

**EVERYDAY**

**DECEMBER 1992**

**WITH PRACTICAL**

# **ELECTRONICS**

**INCORPORATING ELECTRONICS MONTHLY**

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## **UNIVERSAL IR REMOTE CONTROL**

**Designed to adapt to almost any control requirement**

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INSIDE**

**BULL ELECTRICAL  
64-PAGE CATALOGUE**

**MIDI LEAD TESTER  
COMBINATION SWITCH**

**PLUS**

**TEACH-IN '93 PART II  
OUR GCSE & 'A' LEVEL COURSE**



**THE No. 1 INDEPENDENT MAGAZINE for ELECTRONICS, TECHNOLOGY and COMPUTER PROJECTS**

# EVERYDAY WITH PRACTICAL ELECTRONICS

INCORPORATING ELECTRONICS MONTHLY

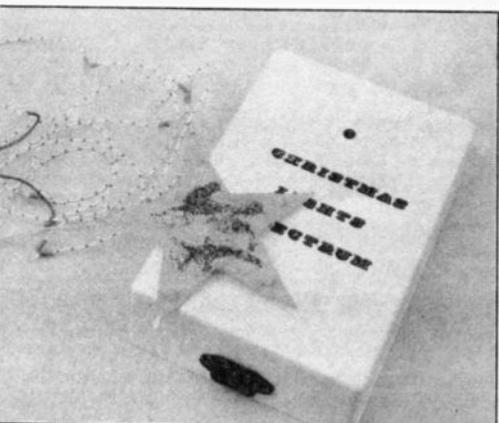
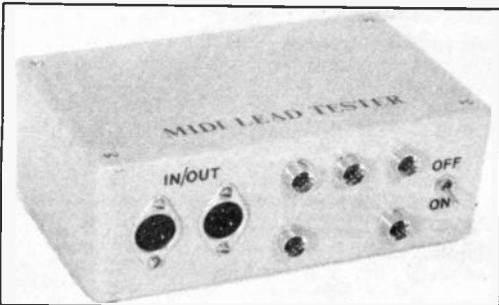
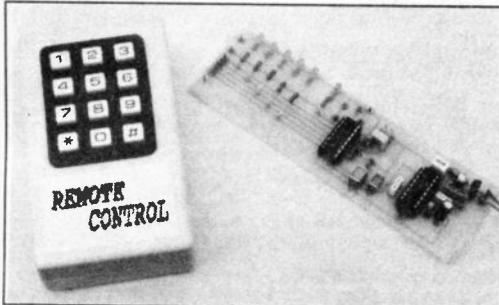
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The No. 1 Independent Magazine for Electronics,  
Technology and Computer Projects

ISSN 0262 3617

PROJECTS... THEORY... NEWS...  
COMMENT... POPULAR FEATURES...



## Projects

- UNIVERSAL INFRA-RED REMOTE CONTROL** 774  
by Chris Walker From one button to 96 combinations this basically simple design has a diverse range of applications
- MIDI LEAD TESTER** by Robert Penfold 792  
Makes checking MIDI leads simple, quick and error free
- MINI LAB** by Alan Winstanley and Keith Dye 810  
We add a L.E.D. Voltmeter to our prototyping board. (Teach-In '93)
- COMBINATION SWITCH** by Max Horsey 814  
Can be used for an alarm or lock, has programmable user password
- CHRISTMAS LIGHTS COLOUR SPECTRUM** 822  
by Mark Daniels Use up to 30 l.e.d.s, which continuously change colour
- AUTO SWITCH ADD-ON** by Ken Taylor 838  
A battery saver that automatically turns off equipment after a pre-set time

## Series

- ALTERNATIVE ENERGY - 4** by T. R. de Vaux Balbirnie 786  
Some further Alternative Energy sources
- TECHNIQUES - ACTUALLY DOING IT** by Robert Penfold 798  
Working with metal and plastic cases for project construction
- TEACH-IN '93 - 1** by Alan Winstanley and Keith Dye 800  
Our series for everyone learning about electronics and particularly GCSE and "A" level students
- INTERFACE** by Robert Penfold 820  
Our page for computer enthusiasts
- CIRCUIT SURGERY** by Mike Tooley 828  
Clinic for constructors - your problems solved
- AMATEUR RADIO** by Tony Smith G4FAI 842  
New Morse Test; Buy But Don't Use; New Book; No Amateur!

## Features

- EDITORIAL** 773
- INNOVATIONS - EVERYDAY NEWS** 782  
News and new products from the world of electronics
- TECHNOLOGY** by Ian Poole 784  
Technological developments that could produce products of the future
- READOUT** A selection of readers letters 795
- HOME BASE** by Terry Pinnell 796  
Jottings of an electronics hobbyist
- SHOPTALK** with David Barrington 797  
Component buying for EPE projects
- FOX REPORT** by Barry Fox 812  
On The Cards; Folklore; Smartcards; Charge It
- ELECTRONICS VIDEOS** 830  
Our range of educational videos
- DIRECT BOOK SERVICE** 831  
Selected technical books by post
- PRINTED CIRCUIT BOARD SERVICE** 835  
PCBs for EPE projects
- INDEX FOR VOLUME 21** 843

**FREE WITH THIS ISSUE - BULL ELECTRICAL  
64-PAGE CATALOGUE** Between pages 804 and 805  
**ADVERTISER'S INDEX** 848

Readers Services • Editorial and Advertisement Departments 773

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Our January '93 Issue will be published on Friday, 4 December 1992. See page 771 for details.

Everyday with Practical Electronics, December 1992

# SURVEILLANCE PROFESSIONAL QUALITY KITS

## No. 1 for Kits

Whether your requirement for surveillance equipment is amateur, professional or you are just fascinated by this unique area of electronics SUMA DESIGNS has a kit to fit the bill. We have been designing electronic surveillance equipment for over 12 years and you can be sure that all of our kits are very well tried, tested and proven and come complete with full instructions, circuit diagrams, assembly details and all high quality components including fibreglass PCB. Unless otherwise stated all transmitters are tuneable and can be received on an ordinary VHF FM radio.

### UTX Ultra-miniature Room Transmitter

Smallest room transmitter kit in the world! Incredible 10mm x 20mm including mic. 3-12V operation. 500m range.....£16.45

### MTX Micro-miniature Room Transmitter

Best-selling micro-miniature Room Transmitter  
Just 17mm x 17mm including mic. 3-12V operation. 1000m range.....£13.45

### STX High-performance Room Transmitter

Hi performance transmitter with a buffered output stage for greater stability and range. Measures 22mm x 22mm including mic. 6-12V operation, 1500m range.....£15.45

### VT500 High-power Room Transmitter

Powerful 250mW output providing excellent range and performance. Size 20mm x 40mm. 9-12V operation. 3000m range.....£16.45

### VXT Voice Activated Transmitter

Triggers only when sounds are detected. Very low standby current. Variable sensitivity and delay with LED indicator. Size 20mm x 67mm. 9V operation. 1000m range...£19.45

### HVX400 Mains Powered Room Transmitter

Connects directly to 240V AC supply for long-term monitoring. Size 30mm x 35mm. 500m range.....£19.45

### SCRX Subcarrier Scrambled Room Transmitter

Scrambled output from this transmitter cannot be monitored without the SCDM decoder connected to the receiver. Size 20mm x 67mm. 9V operation. 1000m range.....£22.95

### SCLX Subcarrier Telephone Transmitter

Connects to telephone line anywhere, requires no batteries. Output scrambled so requires-SCDM connected to receiver. Size 32mm x 37mm. 1000m range.....£23.95

### SCDM Subcarrier Decoder Unit for SCRX

Connects to receiver earphone socket and provides decoded audio output to headphones. Size 32mm x 70mm. 9-12V operation.....£22.95

### ATR2 Micro Size Telephone Recording Interface

Connects between telephone line (anywhere) and cassette recorder. Switches tape automatically as phone is used. All conversations recorded. Size 16mm x 32mm. Powered from line.....£13.45

### UTLX Ultra-miniature Telephone Transmitter

Smallest telephone transmitter kit available. Incredible size of 10mm x 20mm!  
Connects to line (anywhere) and switches on and off with phone use.  
All conversation transmitted. Powered from line. 500m range.....£15.95

### TLX700 Micro-miniature Telephone Transmitter

Best-selling telephone transmitter. Being 20mm x 20mm it is easier to assemble than UTLX. Connects to line (anywhere) and switches on and off with phone use. All conversations transmitted. Powered from line. 1000m range.....£13.45

### STLX High-performance Telephone Transmitter

High performance transmitter with buffered output stage providing excellent stability and performance. Connects to line (anywhere) and switches on and off with phone use. All conversations transmitted. Powered from line. Size 22mm x 22mm. 1500m range.....£16.45

### TKX900 Signalling/Tracking Transmitter

Transmits a continuous stream of audio pulses with variable tone and rate. Ideal for signalling or tracking purposes. High power output giving range up to 3000m. Size 25mm x 63mm. 9V operation.....£22.95

### CD400 Pocket Bug Detector/Locator

LED and piezo bleeper pulse slowly, rate of pulse and pitch of tone increase as you approach signal. Gain control allows pinpointing of source. Size 45mm x 54mm. 9V operation.....£30.95

### CD600 Professional Bug Detector/Locator

Multicolour readout of signal strength with variable rate bleeper and variable sensitivity used to detect and locate hidden transmitters. Switch to AUDIO CONFORM mode to distinguish between localised bug transmission and normal legitimate signals such as pagers, cellular, taxis etc. Size 70mm x 100mm. 9V operation.....£50.95

### QTX180 Crystal Controlled Room Transmitter

Narrow band FM transmitter for the ultimate in privacy. Operates on 180 MHz and requires the use of a scanner receiver or our QRX180 kit (see catalogue). Size 20mm x 67mm. 9V operation. 1000m range.....£40.95

### QLX180 Crystal Controlled Telephone Transmitter

As per QTX180 but connects to telephone line to monitor both sides of conversations. 20mm x 67mm. 9V operation. 1000m range.....£40.95

### QSX180 Line Powered Crystal Controlled Phone Transmitter

As per QLX180 but draws power requirements from line. No batteries required. Size 32mm x 37mm. Range 500m.....£35.95

### QRX180 Crystal Controlled FM Receiver

For monitoring any of the 'Q' range transmitters. High sensitivity unit. All RF section supplied as a pre-built and aligned module ready to connect on board so no difficulty setting up. Outpt to headphones. 60mm x 75mm. 9V operation.....£60.95

### A build-up service is available on all our kits if required.

UK customers please send cheques, POs or registered cash. Please add £1.50 per order for P&P. Goods despatched ASAP allowing for cheque clearance. Overseas customers send sterling bank draft and add £5.00 per order for shipment. Credit card orders welcomed on 0827 714476.

**OUR LATEST CATALOGUE CONTAINING MANY MORE NEW SURVEILLANCE KITS NOW AVAILABLE. SEND TWO FIRST CLASS STAMPS OR OVERSEAS SEND TWO IRCS.**

## ★★★ Specials ★★★

### DLTX/DLRX Radio Control Switch

Remote control anything around your home or garden, outside lights, alarms, paging system etc. System consists of a small VHF transmitter with digital encoder and receiver unit with decoder and relay output, momentary or alternate, 8-way di1 switches on both boards set your own unique security code. TX size 45mm x 45mm. RX size 35mm x 90mm. Both 9V operation. Range up to 200m.

Complete System (2 kits).....£50.95  
Individual Transmitter DLTX.....£19.95  
Individual Receiver DLRX.....£37.95

### MBX-1 HI-FI Micro Broadcaster

Not technically a surveillance device but a great ideal Connects to the headphone output of your Hi-Fi, tape or CD and transmits Hi-Fi quality to a nearby radio. Listen to your favourite music anywhere around the house, garden, in the bath or in the garage and you don't have to put up with the DJ's choice and boring waffle. Size 27mm x 60mm. 9V operation. 250m range.....£20.95

DEPT. EE

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DESIGNS**

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BAXTERLEY, NEAR AATHERSTONE,  
WARWICKSHIRE CV9 2LE



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Surplus always wanted for cash!

## COMPUTER SCOOPS

### PC-AT 386 20-DX FULLY LOADED FOR £499!

- 20 Mhz DX processor
- 2 megs RAM. Exp 10 meg
- 40 meg hard drive
- 1.2 meg 5-1/4" floppy
- 32K cache exp. 64K
- Installed VGA card
- Enhanced 102 key k/board
- Complete with MS-DOS 4.01
- 2 serial 1 parallel ports
- 8 free slots- 6 off 16 bit!



The MP386 quality made by Mitsubishi to last a lifetime! Brand new with all manuals and software plus Super PC-Quick Disk Accelerator for lightning disk access.

**£499 (F)**

### A COMPLETE IBM PC COMPAT SYSTEM FOR ONLY £99!!

Just plug in and go - fully expandable - the Display PC-99! System supplied complete with 12" mono monitor, 84 key keyboard, 360k 5-1/4" floppy disk drive, 128K RAM, 2 serial and 1 parallel port plus DOS with manual. Many other features include 7 slot backplane, all metal case, 150 watt PSU and US made motherboard. In very good used condition with 90 day guarantee. At the unique price of:

Optional FITTED extras: 256K RAM £15; 640K RAM £39. 12" CGA colour monitor with card £89. 2nd 5-1/4" 720K floppy (or 360K if preferred) £29.95. 20 mbyte hard drive £99.

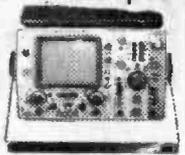
**£99 (E)**



### HP Double Beam 100mhz Scopes

A Major purchase from a liquidated company enables us to offer you this superb Hewlett Packard quality HP1740A oscilloscope in HARDLY USED CONDITION at a fantastic price! Although exceptionally simple to use the full spec. of this scope would take up the whole page, so briefly here are a few of its features:

- 5 mV/div to 20V/div in 12 ranges
  - Large 8 x 10cm CRT with scale illuminated
  - 50ns/div to 2s/div 24 ranges (timebase)
  - Third channel trigger view
  - Dual input impedance 50Ω/1 megΩ
  - TV sync trigger
  - Full delayed time base facilities
  - Light weight fully portable, dual voltage
- Only **£435.00 (F)**  
Fully guaranteed with full user manual.



### COLOUR MONITORS

SAVE £59 - ONLY £100 WHEN BOUGHT WITH THE 386 ABOVE!

14" Philips Model CM8873 VGA multi-sync all the way up to 34KHz with 640 x 480 resolution. This one has everything! Two switches enable you to select CGA,



EGA or VGA and digital/analog. Sound is also provided, with a volume control. There is also a special "Text" switch for word processing, spreadsheets and the like. Compatible with virtually all computers including IBM PC's, Amiga, Atari, BBC, Archimedes etc. Good used condition (possibly minor screen burns) 90 day guarantee. 15" x 14" x 12". Only **£159 (E)**

KME 12" high definition colour monitors. 0.28" dot pitch for superb clarity and modern tone plastic styling. Operates from any 15.625 khz sync RGB video source, with RGB analog and composite sync such as Atari, Commodore Amiga, Acorn Archimedes & BBC. Measures only 13.5" x 12" x 11". Also works as quality TV with our RGB Telebox. Excellent used condition with 90 day guarantee. Beige and brown case. Only **£149 (E)**



Brand new Centronic 14" monitor for IBM PC and compatibles at a lower than ever price! Completely CGA equivalent. Hi-res Mitsubishi 0.42 dot pitch giving 669 x 507 pixels. Big 28 MHz bandwidth. A super monitor in attractive style moulded case. Full 90 day guarantee. Only **£129 (E)**



NEC CGA 12" IBM-PC compatible. High quality ex-equipment fully tested with a 90 day guarantee. In an attractive two tone ribbed grey plastic case measuring 15" x 13" x 12". The front cosmetic bezel has been removed for con- **£79 (E)** tractual reasons. Only **£129 (E)**

20", 22" and 26" AV SPECIALS

Superbly made UK manufacture. PIL all solid state colour monitors, complete with composite video & sound inputs. Attractive teak style case. Perfect for Schools, Shops, Disco, Clubs. In EXCELLENT little used condition with full 90 day guarantee.

20".....£135 22".....£155 26".....£185 (F)

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### Superb Quality 6 foot 40u

### 19" Rack Cabinets

### Massive Reductions

### Virtually New, Ultra Smart!

### Less Than Half Price!

Top quality 19" rack cabinets made in UK by Optima Enclosures Ltd. Units feature designer, smoked acrylic lockable front door, full height lockable half louvered back door and removable side panels. Fully adjustable internal fixing struts, ready punched for any configuration of equipment mounting plus ready mounted integral 12 way 13 amp socket switched mains distribution strip make these racks some of the most versatile we have

ever sold. Racks may be stacked side by side and therefore require only two side panels or stand singly. Overall dimensions are 77-1/2"H x 32-1/2"D x 22"W. Order as:

£ 650 Rack 1 Complete with removable side panels.....£275.00 (G)

£ 375 Rack 2 Less side panels.....£145.00 (G)

£ 950

£ 225

£ 2000

£ 1850

£ 350

£ 139 (D)

£ 149 (D)

£ 39.95 (D)

£ 149 (D)

£ 99 (D)

£ 8.50 (B)

£ 12.95 (B)

£ 9.95 (B)

£ 12.95 (A)

£ 10.95 (B)

£ 10.95 (B)

£ 15.95 (A)

£ 19.95 (A)

£ 14.95 (A)

£ 12.50 (B)

£ 14.50 (B)

### FLOPPY DISK DRIVES

5 1/4" from £22.95 - 3 1/2" from £21.95!

Massive purchases of standard 5 1/4" and 3 1/2" drives enables us to present prime product at industry beating low prices! All units (unless stated) are removed from often brand new equipment and are fully tested, aligned and shipped to you with a 90 day guarantee and operate from standard voltages and are of standard size. All are IBM-PC compatible (if 3 1/2" supported).

3.5" Panasonic JU364 720K\* £21.95 (B)

3.5" Mitsubishi MF353C-L 1.4 Meg. Laptops only £29.95 (B)

3.5" Mitsubishi MF353C-D 1.4 Meg. Non laptop\* £29.95 (B)

5.25" Teac FD-55(36). 360K half height. £22.95 (B)

5.25" Teac FD-55(72). 720K half height. £24.95 (B)

\* Data cable included in price.

Shugart 800/801 SS refurbished & tested £175.00 (E)

Shugart 851 double sided refurbished & tested £275.00 (E)

Mitsubishi M2894-63 double sided switchable hard or soft sectors - BRAND NEW £250.00 (E)

Dual 8" drives with 2 mbyte capacity housed in a smart case with built in power supply! Ideal as exterior drives! £499.00 (F)

End of line purchase scoop! Brand new NEC D2246 8" 85 megabyte of hard disk storage! Full CPU control and industry standard SMD interface. Ultra hi speed transfer and access time leaves the good old ST506 interface standing. In mint condition and comes complete with manual. Only **£299 (E)**

### THE AMAZING TELEBOX!

Converts your colour monitor into a QUALITY COLOUR TV!!



TV SOUND & VIDEO TUNER!

The TELEBOX consists of an attractive fully cased mains powered unit, containing all electronics ready to plug into a host of video monitors made by manufacturers such as MICROVITEC, ATARI, SANYO, COMMODORE, PHILIPS, TUNG, AMSTRAD and many more. The composite video output will also plug directly into most video recorders, allowing reception of TV channels not normally receivable on most television receivers (TELEBOX MB). Push button controls on the front panel allow reception of 8 fully tuneable 'off air' UHF colour television or video channels. TELEBOX MB covers virtually all television frequencies VHF and UHF including the HYPERBAND as used by most cable TV operators. Composite and RGB video outputs are located on the rear panel for direct connection to most makes of monitor. For complete compatibility - even for monitors without sound - an integral 4 watt audio amplifier and low level Hi Fi audio output are provided as standard.

Telebox ST for composite video input monitors £32.95

Telebox STL as ST but with integral speaker £36.50

Telebox MB as ST with Multiband tuner VHF-UHF-Cable, & hyperband For overseas PAL versions state 5.5 or 6mhz sound specification. £69.95

Telebox RGB for analogue RGB monitors (15khz) £69.95

Shipping code on all Teleboxes is (B)

RGB Telebox also suitable for IBM multisync monitors with RGB analog and composite sync. Overseas versions VHF & UHF call. SECAM / NTSC not available.

### No Break Uninterruptable PSU's

Brand new and boxed 230 volts uninterruptable power supplies from Densel. Model MUK 0565-AUAF is 0.5 kva and MUK 1085-AHBH is 1 kva. Both have sealed lead acid batteries. MUD are internal, MUD has them in a matching case. Times from interrupt are 5 and 15 minutes respectively. Complete with full operation manuals.....MUK.....£249 (F) MUD.....£255 (G)

### POWER SUPPLIES

Power One SPL200-5200P 200 watt (250 w peak). Semi open frame giving +5v 35a, -5v 1.5a, +12v 4a (8a peak), -12v 1.5a, +24v 4a (6a peak). All outputs fully regulated with over voltage protection on the +5v output. AC input selectable for 110/240 vac. Dims 13" x 5" x 2.5". Fully guaranteed RFE. £85.00 (B)

Power One SPL130. 130 watts. Selectable for 12v (4A) or 24 v (2A). 5v @ 20A ± 12v @ 1.5A. Switch mode. New. £59.95 (B)

Astec AC-8151 40 watts. Switch mode. +5v @ 2.5a. +12v @ 2a. -12v @ 0.1a. 6-14" x 4" x 1-3/4". New. £22.95 (B)

Greendale 19AB0E 60 watts switch mode. +5v @ 6a ± 12v @ 1a. +15v @ 1A. RFE and fully tested. 11 x 20 x 5.5cms. New. £24.95 (C)

Conver AC130. 130 watt hi-grade VDE spec. Switch mode. +5v @ 15a. -5v @ 1a. ± 12v @ 6a. 27 x 12.5 x 6.5cms. New. £49.95 (C)

Boshert 13090. Switch mode. Ideal for drives & system. +5v @ 6a. +12v @ 2.5a. -12v @ 0.5a. -5v @ 0.5a. £29.95 (B)

Farnell G6/40A. Switch mode. 5v @ 40a. Encased £95.00 (C)

Farnell G24/5S. As above but 24v @ 5a. £65.00 (C)

### BBC Model B APM Board

WIN £100 CASH!

£100 CASH FOR THE MOST NOVEL DEMONSTRABLE APPLICATION!

BBC Model B type computer on a board. A major purchase allows us to offer you the PROFESSIONAL version of the BBC computer at a parts only price. Used as a front end graphics system on large networked systems the architecture of the BBC board has so many similarities to the regular BBC model B that we are sure that with a bit of experimentation and ingenuity many useful applications will be found for this board! It is supplied complete with a connector panel which brings all the I/O to 'D' and BNC type connectors - all you have to do is provide +5 and +12 v DC. The APM consists of a single PCB with most major IC's socketed. The IC's are too numerous to list but include a 6502, RAM and an SAA5050 teletext chip. Three 27128 EPROMS contain the custom operating system on which we have no data. On application of DC power the system boots and provides diagnostic information on the video output. On board DIP switches and jumpers select the ECONET address and enable the four extra EPROM sockets for user software. Appx. dims: main board 13" x 10". I/O board 14" x 3". Supplied tested with circuit diagram, data and competition entry form.

Only **£29.95** or 2 for **£53 (B)**

### SPECIAL INTEREST

Trilo 0-18 vdc bench PSU. 30 amps. New £ 470

Fujitsu M3041 600 LPM band printer £2950

DEC LS/02 CPU board £ 150

Rhode & Schwarz SBUF TV test transmitter 25-1000mhz. Complete with SBTF2 Modulator £6500

Calcomp 1036 large drum 3 pen plotter £ 375

Thurby LA 160B logic analyser £ 950

1.5kw 115v 60hz power source POA

Anton Pillar 400 Hz 3 phase frequency converter 75Kw POA

Newton Derby 400 Hz 70 Kw converter £ 225

ADDS 2020 VDU terminals - brand new £ 2000

Sekonic SD 150H 18 channel Hybrid recorder £ 1850

HP 7580A A1 8 pin high speed drum plotter £ 350

Kenwood DA-3501 CD tester, laser pickup simulator £ 149 (D)

### BRAND NEW PRINTERS

Microline 183. NLQ 17x17 dot matrix. Full width. £ 139 (D)

Hyundai HDP-920. NLQ 24x18 dot matrix full width. £ 149 (D)

Qume LetterPro 20. daisy. Qume QS-3 interface. £ 39.95 (D)

Centronics 152-2-9 x 7 dot matrix. Full width. £ 149 (D)

Centronics 159-4-9 x 7 dot matrix. Serial. 9-1/2" width £ 99 (D)

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## SPARKOMATIC 4 x 150 watt CAR AMPLIFIER



The SA3200 is our top of the line 4 Channel Amplifier which is extremely well specified. It is very powerful and versatile and features separate bass and treble controls which gives the user the possibility of reducing bass response to the front speakers and adding treble for better stereo imaging. The bass response can then be increased to the rear speakers which are usually larger and capable of offering better reproduction. The SA3200 features a bridge operation switch which offers the possibility of using the amplifier in 4, 3 or 2 channel mode. The 3 channel mode is ideal for installations where rear deck speakers are used in combination with a separate subwoofer.

● 4 x 150 Watts max ● 4 x 80 Watts into 4 Ohms at less than 0.5% THD ● 2 x 80 Watts plus 1 x 160 Watts at less than 0.5% THD ● 2 x 160 Watts into 4 Ohms at less than 0.5% THD ● Separate bass and treble controls for front and rear channels ● Separate sensitivity controls for front and rear channels ● 2, 3 or 4 channel operation ● Heavy duty power wires ● Glass blasted aluminium heatsink ● High current capacity

£251.65 plus £7 p&p

## SPARKOMATIC 2 x 150 watt CAR AMPLIFIER

The SA1500 is a very highly specified 2 Channel Amplifier with built-in sub bass crossover. The SA 1500, which is ideal for powering medium sized subwoofers, will also operate in bridge mode as a 150 Watt mono amplifier.

● 2 x 150 Watts max. Into 4 Ohms ● 2 x 70 Watts per channel at 0.5% THD ● Bridge mode operation ● Sensitivity adjustment ranging from 100mV to 1V ● Heavy duty power wires ● Built-in sub bass crossover ● Glass blasted aluminium heatsink ● High current capacity

£117.65 plus £6.50 p&p

## SPARKOMATIC 80 watt CAR POWER AMPLIFIER

The AMP 7000 produces high power at low distortion. The amplifier accommodates low level, high level and high power radio speaker inputs. The response is linear and extends beyond the capability of all music sources. This compact unit mounts easily and its quick connect terminals accept RCA or straight wire input terminals. Power rating 2 x 40 watt per channel. MMP 2 x 20 watt at 10% THD response 20Hz-20kHz. Size 160mm x 130mm x 45mm.

£32.95 plus £3.50 p&p

## 11 BAND COMPONENT GRAPHIC EQUALIZER FOR CARS



This neat unit connects between the line output of your car stereo and your power amplifiers so that you are able to adjust the sound as in a studio compensating for soft furnishing and sound reflections from glass, also it has a sub-woofer output to drive a separate amplifier for that extra deep bass sound. FEATURES: 2 channel inputs 4 channel outputs via phono sockets, CD input via 3.5mm jack 11 band graphic. SPECIFICATION RANGE 20Hz-60kHz THD 0.05%, S/N RATIO 85dB. EQ FREQUENCIES 60Hz, 120Hz, 250Hz, 380Hz, 500Hz, 750Hz, 1kHz, 2kHz, 4kHz, 8kHz, 16kHz (boost cut of ±12dB) SIZE 178mm x 25mm x 140mm.

£32.70 postage £1.80

## EMINENCE 4Ω PROFESSIONAL USA MADE IN CAR CHASSIS SPEAKERS

All units are fitted with big magnets "Nomex" Voice coils NOT ALUMINIUM, "Nomex" is very light and can stand extremely high temperatures, this mixture makes for high efficiency and long lasting quality of sound.

V6 6 1/2" 200W Max	Range 50Hz-3kHz	£34.40
V6 8" 300W Max	Range 45Hz-3kHz	£39.35
V10 10" 400W Max	Range 33Hz-4kHz	£44.45
V12 12" 400W Max	Range 35Hz-3kHz	£45.95
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# RTVC

No.	Qty. per pack	Description	Price
MO20	1	30W dome tweeter by Eagle/Japan Made size 90mm x 66mm	£1
MO21	1	60W HiFi tweeter made for Jamo UK size 90mm sq.	£1
MO22	2	30 watt 8 ohm HiFi chassis speakers. Made for Hitachi UK midi systems, size 125mm sq. with large 70mm magnet	£9.00 + £2.00 p&p
MO23	2	Pod Car Speakers. Moulded in black plastic with 15 watt 10cm Goodmans unit fitted	£4.95 + £2.50 p&p
MO23A	1pr	40 watt Car Speakers made for Roadstar of Switzerland. Fitted with dual polypropylene cone and foam rubber surround. Big 70mm magnet for good base response. Supplied with grills fixing screws and cable. Size 13cm, weight 1.5Kg	£11.70 pair + £3.65 p&p or TWO pairs for £25.00 UK post paid
MO24	2	Audax JBL 40-100watt dome tweeters. High performance 10mm Ferrofluid cooled horn loaded unit for load distortion and high output. Supplied with 1st order crossover, spec. 40 watts at 3kHz, 100 watt at 8kHz; size 51mm x 51mm x 16.5mm. Ideal for car use	£7.50 + £1 p&p
MO25	2	33000µF 10V d.c. can type computer grade quality electrolytic UK made	£1
MO25A	1	47µF 385V d.c. can type electrolytic. Size 350mm x 250mm. UK made by Philips	£1.75
MO26	2	680µF 100V d.c. can type electrolytic size 45mm x 25mm	£1
MO27	3	2200µF 25V d.c. can type electrolytic size 45mm x 25mm	£1
MO28	1	15000µF 40V d.c. can type 23A electrolytic size 113mm x 50mm	£1
MO29	1	33000µF 16V 27A can type electrolytic size 113mm x 50mm	£1
MO30	20	Assorted Variable trimmers	£1
MO31	4	Tuning capacitors 2-gang dielectric type	£1
MO32	2	10k + 10k wirewound precision potentiometer	£1
MO33	8	Rotary potentiometers	£1
MO34	5	100k multivolt Varicap type tuning potentiometer with knob size 45mm x 5mm	£1
MO35	200	Carbon resistors	£1
MO36	2	Large VU meters. Japan Made	£1
MO37	1	Large Tuning meter 125µA-0-125µA size 55mm x 47mm	£1.75
MO38	1	Dual VU meter 280µA f.s.d., size 80mm x 42mm x 15mm	£1.50
MO39	5	Coaxial Aerial Plugs, all metal type	£1
MO40	6	Fuseholders, chassis mounting for 20mm size fuses	£1
MO41	4	Fuseholders, in-line type for 20mm size fuses	£1
MO42	20	5 Pin Din 180° chassis mount sockets	£1
MO43	6	Double phono sockets	£1
MO44	5	6.35mm (1/4") Stereo Jack sockets	£1
MO45	4	6.35 (1/4") Mono Jack Plugs	£1
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MO47	2	Case handles plated U-shape, size 97mm x 50mm	£1
MO48	30	Mixed control knobs	£1
MO49	1	Cassette tape transport mechanism, belt-drive, top loading, six piano key operation with knobs, stereo record/replay erase heads, heavy fly-wheel £5.50 + £2.65 p&p	£5.50 + £2.65 p&p
MO50	1	HiFi stereo pre-amp. module. Input for CD Tuner record player with diagram. Made by Mullard	£1
MO51	2	AM/FM tuner head modules. Made by Mullard	£1
MO52	3	AM I.F. modules. Made by Mullard	£1
MO53	1	FM stereo decoder module with diagram. Made by Mullard	£1
MO54	3	UHF Varicap tuned tuner heads unboxed, untested but complete. Made by Mullard	£1
MO55	1	25V d.c. 150mA Mains adaptor in neat plastic box, size 80mm x 55mm x 47mm	£1
MO55A	1	ETRI Brand new 80mm Cooling Fan. Five bladed A.C. impedance corrected motor on a cast aluminium chassis. Size 80mm x 40mm. Voltage 115V a.c. working. 130mA. Japanese made	£5.95 + £1.40 p&p, TWO for £11.20 UK post paid
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MO57A	1	SRBP Copper Clad Printed Circuit Board. Size 410mm x 360mm x 2mm £3.65 + 75 p&p	£3.65 + 75 p&p
MO58	2	Mono cassette tape heads. Japan Made	£1
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MO65	5	Motor Speed Control i.c.	£1
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# RTVC

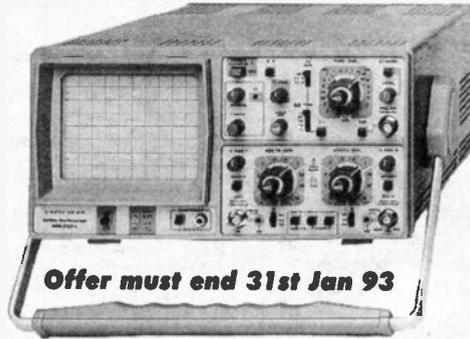
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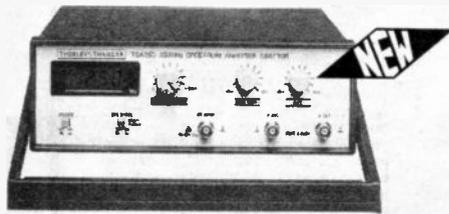
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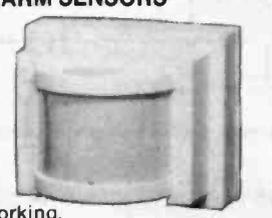
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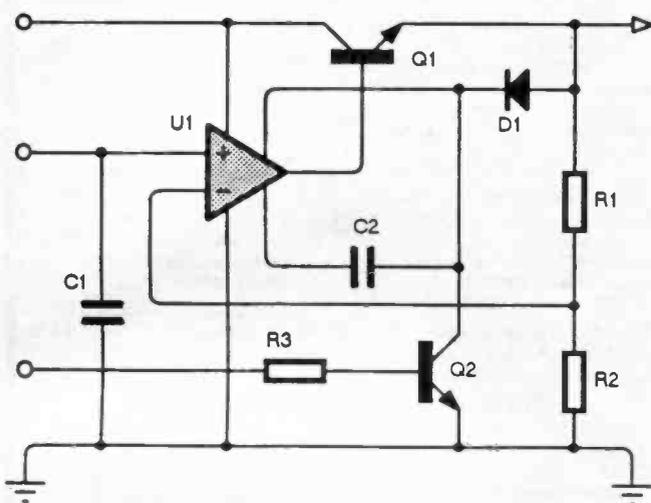
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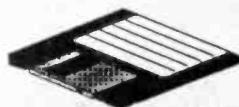
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Our list will give you fuller details of all our Audio Kits, components and special offers.

### AUDIO DESIGN 80 WATT POWER AMPLIFIER.



This fantastic John Linsley Hood designed amplifier is the flagship of our range, and the ideal powerhouse for your ultimate hifi system. This kit is your way to get £K performance for a few tenths of the cost! Featured on the front cover of 'Electronics Today International' this complete stereo power amplifier offers World Class performance allied to the famous HART quality and ease of construction. John Linsley Hood's comments on seeing a complete unit were enthusiastic:- "The external view is that of a thoroughly professional piece of audio gear, neat elegant and functional. This impression is greatly reinforced by the internal appearance, which is redolent of quality, both in components and in layout." Options include a stereo LED power meter and a versatile passive front end giving switched inputs using ALPS precision, low-noise volume and balance controls. A new relay switched front end option also gives a tape input and output facility so that for use with tuners, tape and CD players, or indeed any other 'flat' inputs the power amplifier may be used on its own, without the need for any external signal handling stages. 'Slave' and 'monobloc' versions without the passive input stage and power meter are also available. All versions fit within our standard 420 x 260 x 75mm case to match our 400 Series Tuner range. ALL six power supply rails are fully stabilised, and the complete power supply, using a toroidal transformer, is contained within a heavy gauge aluminium chassis/heat sink fitted with IEC mains input and output sockets. All the circuitry is on professional grade printed circuit boards with roller tinned finish and green solder resist on the component ident side, the power amplifiers feature an advanced double sided layout for maximum performance. All wiring in this kit is pre-terminated, ready for instant use!

RLH11 Reprints of latest articles.....£1.80  
K1100CM HART Construction Manual.....£5.50

### LINSLEY HOOD 1400 SERIES ULTRA HIGH-QUALITY PREAMP

Joining our magnificent 80 Watt power amplifier now is the most advanced preamplifier ever offered on the kit, or indeed made-up marketplace. Facilities include separate tape signal selection to enable you to listen to one programme while recording another, up to 7 inputs, cross recording facilities, class A headphone amplifier, cancellable 3-level tone controls and many other useful functions, all selected by high quality relays. For full details see our list.

Send or phone for your copy of our List (50p) of these and many other Kits & Components. Enquiries from Overseas customers are equally welcome, but PLEASE send 2 IRCs if you want a list sent surface post, or 5 for Airmail.

Ordering is easy. Just write or telephone your requirements to sample the friendly and efficient HART service. Payment by cheque, cash or credit card. A telephoned order with your credit card number will get your order on its way to you THAT DAY.

Please add part cost of carriage and insurance as follows:- INLAND Orders up to £20 - £1.50  
Orders over £20 - £3.50 Express Courier, next working day. £10 (For safety all computer parts are only sent by courier) OVERSEAS - Please see the ordering information with our lists.

### LINSLEY HOOD 'SHUNT FEEDBACK' R.I.A.A. MOVING COIL & MOVING MAGNET PICKUP PREAMPLIFIERS



Modern, ultimate sound systems are evolving towards built-in preamplifiers within or near the turntable unit. This keeps noise pickup and treble loss to a minimum. We now offer two units, both having the sonically preferred shunt feedback configuration to give an accurate and musical sound, and both having the ability to use both moving magnet and moving coil cartridges.

Kit K1500 uses modern integrated circuits to achieve outstanding sound quality at minimal cost. The very low power requirements enable this unit to be operated from dry batteries and the kit comes with very detailed instructions making it ideal for the beginner. K1500 Complete kit with all components, printed circuit board, full instructions and fully finished case.....£67.99  
Instructions only.....£2.80

Kit K1450 is a fully discrete component implementation of the shunt feedback concept and used with the right cartridge offers the discerning user the ultimate in sound quality from vinyl disks. Can be fitted inside our 1400 Preamp, used externally or as a standalone unit. It has a higher power requirement and needs to be powered from our 1400 Series preamplifier or its own dedicated power supply. K1450 Complete kit of board mounting parts for discrete component RIAA preamplifier.....£61.06  
1500/2-8 Case to suit, including Hardware....£39.52  
K1565 Power Supply in matching case. Features shielded toroidal transformer and upgrade path to full preamp power supply.....£79.42

### ALPS PRECISION LOW-NOISE STEREO POTS.



To fulfill the need for higher quality controls we are now importing an exciting new range of precision audio pots in values to cover most quality amplifier applications. All in 2-gang stereo format, with 20mm long 6mm dia. steel shafts. Now you can throw out those noisy ill-matched carbon pots and replace with the real hi-fi components only used selectively in the very top flight of World class amplifiers. The improvement in track accuracy and matching really is incredible giving better tonal balance between channels and rock solid image stability. Motorised versions have 5v DC Drive motor.

2-Gang 100K Lin.....£8.67  
2-Gang 10K & 50K Log.....£9.40  
2-Gang 10K Special Balance, zero crosstalk and zero centre loss.....£10.48  
2-Gang 20K Log (Volume Control) MOTORISED.....£19.20  
2-Gang 10K Special Balance, MOTORISED, zero crosstalk and < 10% centre loss with near Log/Antilog Tracks).....£19.98

### STUART REEL-TO-REEL TAPE RECORDER CIRCUITS

Complete stereo record, replay and bias circuit system for reel-to-reel recorders. These circuits will give studio quality with a good tape deck. Separate sections for record and replay give optimum performance and allows a third head monitoring system to be used where the deck has this fitted. Standard 250mV input and output levels. Ideal for bringing that old valve tape recorder back to life. Suitable stereo heads are in our head list. This basic kit is suitable for advanced constructors only. K900W Stereo Kit with Wound Coils and Twin Meter Drive.....£123.93  
RJS1 Reprints of Original Descriptive Articles.....£3.60

### LINSLEY-HOOD CASSETTE RECORDER CIRCUITS

Complete record and replay circuits for very high quality low noise stereo cassette recorder. Circuits are suitable for use with any high quality cassette deck. Switched bias and equalisation to cater for chrome and ferric tapes. Very versatile, with separate record and play circuits and easy to assemble on plug-in PCBs. Complete with full instructions. Complete Stereo Record/Play Kit.....£62.58  
VU Meters to suit.....(Each) £3.99  
RLH1 & 2 Reprints of original Articles.....£2.70

### HIGH QUALITY REPLACEMENT CASSETTE HEADS



Do your tapes lack treble? A worn head could be the problem. For top performance cassette recorder heads should be replaced every 1,500 hours. Fitting one of our high quality replacement heads could restore performance to better than new! Standard inductances and mountings make fitting easy on nearly all machines (Sony are special dimensions, we do not stock) and our TC1 Test Cassette helps you set the azimuth spot on. As we are the actual importers you get prime parts at lower prices, compare our prices with other suppliers and see All our heads are suitable for use with any Dolby system and are normally available ex stock. We also stock a wide range of special heads for home construction and industrial users. HC80 NEW RANGE High Beta Permalloy Stereo head. Modern space saver design for easy fitting and lower cost. Suitable for chrome metal and ferric tapes, truly a universal replacement head for everything from hi-fi decks to car players and at an incredible price too!.....£8.30  
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HRP373 Downstream Monitor Stereo Combination Head.....£53.90  
HC15 Special Offer of Standard Quality Stereo R/P Head with slight face scratches.....3 for Only £4.80  
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HM120 Standard Mono R/P Head.....£3.44  
H524 Standard Erase Head.....£11.90  
H561 Hi Field Erase Head for METAL Tapes.....£3.49  
SM150 2/2 (Double Mono) DC Erase Head.....£5.20  
HQ751E 4/4 True 4-Track Erase Head.....£57.06

REEL TO REEL HEADS  
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TAPE RECORDER CARE PRODUCTS  
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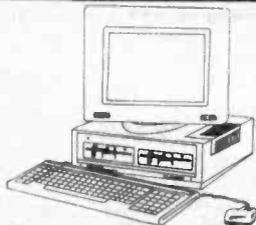
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## PC SOFTWARE AT AMAZING PRICES!!

Words & Figures by Lifetree - combined spreadsheet, word processor and database with graphics. Needs PC with DOS 2+, 256k RAM, mono or colour display. Supports over 200 printers. 546 page manual, spread out menu and disks, all in library case. *Just look at the price!!*

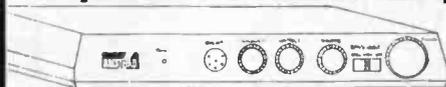
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## VHF/UHF TV RECEIVER/CONVERTER



Z8991 Amstrad  
MP3 computer  
modulator/converter.

A complete, fully tuneable VHF/UHF TV receiver with RGB and composite video out, and sound on the internal speaker. For use with the Amstrad CTM644-2 monitor specifically, or can be used with any colour/mono home computer monitors that have a 15.625kHz line frequency. Grey case 330x250x50mm. Controls: contrast, colour, tuning, volume and band select (VHF-L, VHF-H, UHF) RGB output on 6 pin DIN skt, and composite video. Intended for European market - needs 2 ceramic filters changing. Parts and instructions supplied. Needs stabilized 12V DC, either from monitor or separate power supply (our AL12, £8.14 is ideal)

**ONLY  
£14.95**

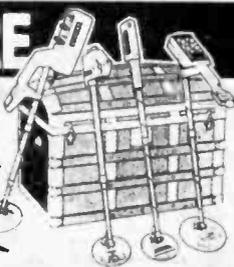
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Inputs for 2 paddles, analogue joysticks, light gun/pen etc. Outputs: UHF, RGB on 8 pin DIN skt, stereo sound 35mm skt. Uses 168 pin dedicated chip, Z80A CPU, AY-3-8912 sound chip, UM1234 UHF modulator

## 3.5" DISK DRIVE



Model FD9. Brand new and boxed, this cased 3.5" 720k unit comes complete with cables and instructions. Can be used with 1x86 & 2x86 "as is", and with 1512/1640 with mod.

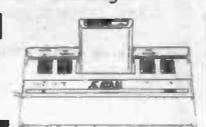
**Superdeal Price £24.95**

## ATARI 2600 GAMES CONSOLE

Complete and boxed with joystick, power supply, TV lead and games cartridge (centipede). Not new, but fully checked and working

Special Price

**£29.95**



**THE 'Greenweld Guardian' IS OUR MONTHLY NEWSLETTER FREE TO ALL BARGAIN LIST SUBSCRIBERS - ONLY £6 A YEAR FOR THE NEXT 12 ISSUES OF ALL OUR LISTS!!**

## CASIO

Casio have just introduced a range of personal stereos and portable audio equipment at very competitive prices:



- W880 Bass boost, auto stop, belt clip £8.99
- AS51R As above but with AM/FM radio £15.99
- AS500R As AS51R, but with auto reverse £23.50
- W120 As AS500R but with graphic equalizer £27.99

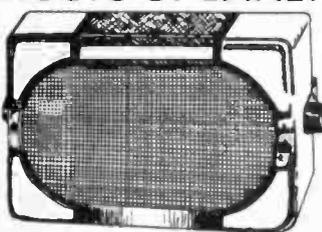


- CP80 Bass boost, AM/FM/LW radio, single cassette, 6W PMPO £32.95
- CP200 Bass boost, AM/FM/LW radio, twin cassette, hi-speed dubbing, graphic equalizer, 6W PMPO £42.95
- CD510 CD player, twin cassette, AM/FM/LW radio + lots of other features, 45W PMPO £159.95

## FOREGROUND MUSIC SPEAKERS

High quality for pubs, clubs etc. Bass unit, mid and 2 tweeters. Moulded cabinet with adjustable bracket. Max power 80W. Size 275x170x125mm. 4R imp.

**Less than half price!  
£39.95 per pr!**



## INSULATION TEST UNIT

Y136C 500V tester that can be used with most digital meters. 2 ranges covering 100k-1999M. Supplied with leads, batts, instructions and carry case. Original trade price £34

**£12.95**

## CAMERA CLEARANCE

110 & 35mm - all are returns, some have small parts missing, but great value for lenses, electronics etc (most have built in flash units)

**5 for £10**

## STEREO AUTO-REVERSE CASSETTE MECHS

Z5405 High quality heavy duty all metal construction stereo cassette player mechanism, probably intended for continuous background music. This is a lovely bit of kit - starts playing as soon as a cassette is inserted. Has fast forward, rewind and eject keys. It's bi-directional, and the sensing circuit automatically reverses the tape at the end. Has a Canon motor and works off 12V DC. Great value at £4.95.

## NEW POWER SUPPLIES

Z5406D High efficiency step down power regulator module by SGS. This is a GSR400 type, as listed by Farnell at £41.11 each. Output is 7V @ 4A from a DC input of 10-46V. Possible uses include battery charger, or put two together and use 24V lorry battery to power car equipment. Our special price - just £5.75 each.

Z5409 Eurocard size - 160x100mm by Protek. 115/230V input. Outputs: +5V @ 3A; +12V @ 2A; -12V @ 0.25A Price £8.95

## TEACH-IN 93 We have good supplies of VEROBLOCK

and can offer 2 for £13.70 + P&P with a FREE (on request) miniature 12/24V ac/dc soldering iron!

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## VERSATILE BBC INTERFACE

A comprehensive interface which allows the BBC computer to be connected safely to a wide range of input and output devices. Two leads connect the interface to the User Port and the Printer port. The interface connects to the 'real world' via standard screw terminal blocks. Up to 16 outputs (all via plug-in single pole change over relays - 8 supplied) and 8 fully protected inputs. L.e.d. status monitoring is provided on all input and output lines. The interface requires an Independent 12 Volt supply.

**KIT 844.....£51.95**

## STEPPING MOTOR DRIVER & INTERFACE

A single board stand-alone stepping motor driver with built in oscillator and speed control circuits. A computer is not required with this board which will drive most unipolar 4 phase motors. Variable Acceleration, Speed, and Direction, may be controlled in HALF STEP, FULL STEP, and ONE PHASE modes. Up to 35V and 1.5A per phase. L.e.d. mimic display. Connector is provided for a computer port. The Kit includes our MD35 motor

**KIT 843 £29.95 - BUILT £44.95**

## DIGITAL LCD THERMOSTAT

A versatile thermostat using a thermistor probe and having an l.c.d. display. MIN/MAX memories, -10 to 110 degrees celsius, or can be set to read in Fahrenheit. Individually settable upper and lower switching temperatures allow close control, or alternatively allow a wide 'dead band' to be set which can result in substantial energy savings when used with domestic hot water systems. Ideal for greenhouse ventilation or heating control, aquaria, home brewing, etc. Mains powered, 10A SPCO relay output. Punched and printed case.

**KIT 841.....£29.95**

## 4 CHANNEL LIGHT CHASER

A 1000W per channel chaser with Zero Volt Switching, Hard Drive, and full inductive load capability. Built-in mic. and sophisticated 'Beat Seeker' circuit - chase steps to music, or auto when silent. Variable speed and mic. sensitivity control. L.e.d. mimic on front panel. Switchable for 3 or 4 channels. P552 output socket. Suits Rope Lights, Pin Spots, Disco, and Display lighting.

**KIT 833.....£32.13**

## SUPERHET LW MW RADIO

At last an easy to build SUPERHET AM radio kit. Covers Long and Medium waves. Built in loudspeaker with 1 Watt output. Excellent sensitivity and selectivity provided by ceramic IF filter. Simple alignment and tuning without special equipment. Supplied with pre-drilled transparent front panel and dial, for interesting see-through appearance.

**KIT 835.....£17.16**

## ACOUSTIC PROBE

A very popular project which picks up vibrations by means of a contact probe and passes them on to a pair of headphones or an amplifier. Sounds from engines, watches, and speech travelling through walls can be amplified and heard clearly. Useful for mechanics, Instrument engineers, and nosy parkers!

**KIT 740.....£19.98**

## PEST SCARER

Produces high power ultrasound pulses. L.e.d. flashes to indicate power output. Battery powered 9 - 12V, or mains adaptor £2.00 EXTRA.

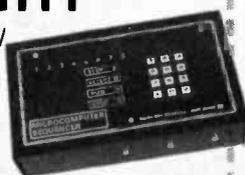
**KIT812.....£14.81**

## KIT HIGHLIGHT

### 8 CHANNEL LIGHT SHOW PROGRAMMABLE SEQUENCER KIT 838

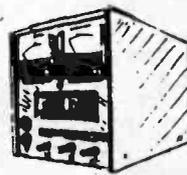
An advanced design using a pre-programmed microcontroller IC to generate over 100 light sequences. Additional battery backed RAM area to store your own sequences. Keypad control allows lamps to be controlled manually, sequences entered and selected, and sequence speed to be increased and decreased. ZERO VOLT SWITCHING. Programs include 3 and 4 channel versions so that existing lights can be used as well as 8 channel arrangements. Special output drive using a two winding transformer ensures foolproof operation with pin-sockets and other difficult loads. This is a superbly finished kit with pre-drilled case and screen printed front panel. Full LED mimic, 2 P552 output sockets, 8 Amp isolated tab triacs with heatsink. Kit includes everything - down to the last nut and bolt. Tremendous Value.

**KIT PRICE  
£64.89**



### MOSFET VARIABLE BENCH POWER SUPPLY 25V 2.5A

Our own high performance design. Variable output Voltage from 0 to 25V and Current limit from 0 to 2.5A. Capable of powering almost anything. Two panel meters Indicate Voltage and Current. Fully protected against short-circuits. The variable Current limit control makes this supply ideal for constant current charging of NICAD cells and batteries. A Power MOSFET handles the output for exceptional ruggedness and reliability. Uses a toroidal mains transformer.



**KIT 769.....£56.82**

### DIGITAL CAPACITANCE METER

Provides clear readings of capacitance values from a few pF up to thousands of µF. Ideal for beginners. It allows obscurely marked components to be identified quickly and easily. Quartz controlled accuracy of 1%, and large, clear 5 digit display. Kit is now supplied with a punched and printed front panel, case, all components and top quality printed circuit board. New low price.

**KIT 493.....£34.95**

### BAT DETECTOR

An excellent circuit which reduces ultrasound frequencies between 20 and 100 kHz to the normal (human) audible range. Operating rather like a radio receiver the circuit allows the listener to tune-in to the ultrasonic frequencies of interest. Listening to Bats is fascinating, and it is possible to identify various different types using this project. Other uses have been found in industry for vibration monitoring etc.

**KIT 814.....£21.44**

### QUICK CAPACITANCE TESTER

A low cost hand-held audio/visual unit which can identify short, open and working capacitors quickly and with a minimum of fuss. Also gives indication of leakage current. An ideal kit for beginners, built on a single printed circuit board which has large copper areas used as test pads. Only a minimum of wiring is needed. 2 l.e.d.s and a piezo transducer provide the output indication.

**KIT 834.....£10.34**

### IONISER

A highly efficient mains powered Negative Ion Generator that clears the air by neutralising excess positive ions. Many claimed health benefits due to the ioniser removing dust and pollen from the air and clearing smoke particles. Costs virtually nothing to run and is completely safe in operation. Uses five point emitters.

**KIT 707.....£17.75**

### ACTIVE I.R. BURGLAR ALARM

This alarm is useful where ordinary 'passive' (pir) detectors are not suitable. It works by detecting disturbances to its own short wave infra-red beam. Output is via mains rated relay contacts. Built in timer, and mains transformer.

**KIT 700.....£40.74**

## 12V EPROM ERASER

A safe low cost eraser for up to 4 EPROMS at a time in less than 20 minutes. Operates from a 12V supply (400mA). Used extensively for mobile work - updating equipment in the field etc. Also In educational situations where mains supplies are not allowed. Safety Interlock prevents contact with UV.

**KIT 790.....£28.51**

## EE TREASURE HUNTER

Our own widely acclaimed design. This sensitive Pulse Induction metal detector picks up coins and rings etc up to 20cm deep. Negligible 'ground effect' means that the detector can even be used with the head immersed in sea water. Easy to use, circuit requires only a minimum of setting up as a Quartz crystal provides all of the critical timing. Kit includes search-head, handle, case, PCB and all components.

**KIT 815.....£45.95**

## INSULATION TESTER

A reliable and neat electronic tester which checks insulation resistance of wiring and appliances etc., at 500 Volts. The unit is battery powered, simple and safe to operate. Leakage resistance of up to 100 Megohms can be read easily. A very popular college project.

**KIT 444.....£22.37**

## 3 BAND SHORT WAVE RADIO

Covers 1.6 to 30MHz in three bands using modern miniature plug-in coils. Audio output is via a built-in loudspeaker. Advanced stable design gives excellent stability, sensitivity and selectivity. Simple to build battery powered circuit. Receives a vast number of stations at all times of the day.

**KIT 718.....£30.30**

## DIGITAL COMBINATION LOCK

Digital lock with 12 key keypad. Entering a four digit code operates a 250V 16A relay. A special anti-tamper circuit permits the relay board to be mounted remotely. Ideal car immobiliser, operates from 12V. Drilled case, brushed aluminium keypad.

**KIT 840.....£19.86**

## PORTABLE ULTRASONIC PEST SCARER

A powerful 23kHz ultrasound generator in a compact hand-held case. MOSFET output drives a special sealed transducer with intense pulses via a special tuned transformer. Sweeping frequency output is designed to give maximum output without any special setting up.

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## LIGHT RIDER DISCO LIGHTS

A six channel light driver that scans from left to right and back continuously. Variable speed control. Up to 500 watts per channel. Housed in a plastic box for complete safety. Built on a single printed circuit board.

**KIT 560.....£22.41**

## LIGHT RIDER 9-12V CHASER LIGHTS

A low voltage DC powered end-to-end type chaser that can be set for any number of lights between 3 and 16. The kit is supplied with 16 l.e.d.s but by adding power transistors it is possible to drive filament bulbs for a larger brighter display. Very popular with car customisers and modellers. L.e.d.s can be randomly positioned and paired to give twinkling effects.

**KIT 559.....£15.58**

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### HAMEG HM203-7 20 MHz DUAL TRACE OSCILLOSCOPE & COMPONENT TESTER

Western Europe's best selling oscilloscope - It is RELIABLE, HIGH PERFORMANCE, & EASY TO USE.

Sharp bright display on 8 x 10cm screen with internal graticule. A special extra feature is the built-in component tester which allows capacitors, resistors, transistors, diodes and many other components to be checked. The quality of this instrument is outstanding, and is supported by a two year parts and labour warranty.

If you are buying an oscilloscope - this is the one. - It costs a fraction more than some other 20 MHz 'scopes but it is far far superior. Supplied with test probes, mains lead, and manual.

**£338.00 + £59.15 VAT** Includes FREE Next-day delivery

(Cheques must be cleared)

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The classic book by Tom Duncan used throughout schools. Very well illustrated, ideal first book for age 10 on. No soldering. Uses an S.DEC breadboard.

**Book & Components £28.95, Book only £6.25**

#### FUN WITH ELECTRONICS

An Usborne book, wonderfully illustrated in colour. Component pack allows 6 projects to be built and kept. Soldering is necessary. Age 12 on, or younger with adult help. **Book & Components £20.88, Book only £2.95**

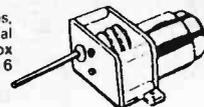
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A more advanced book to follow the others. No soldering. Circuits cover a wide range of interests. **Book & Components £30.69, Book only £2.95**

### DC MOTOR/GEARBOXES

Ideal for robots, buggles, and many other mechanical projects. Min. plastic gearbox with 1.5-4.5V DC motor. 6 ratios can be set up.

**Small type MGS.....£4.77**  
**Large type MGL.....£5.58**



### STEPPING MOTORS

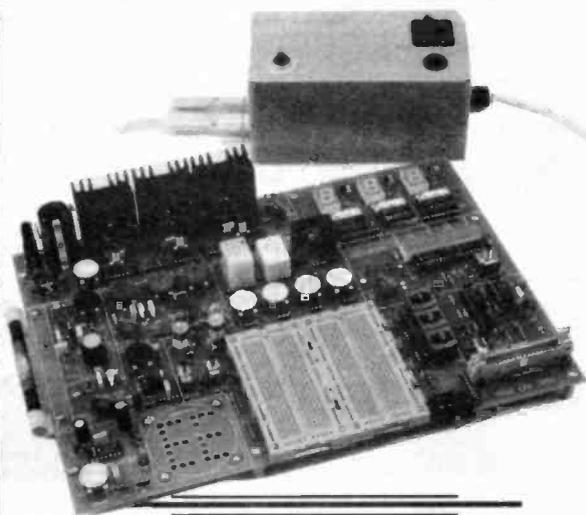
For computer control via standard 4 pole unipolar drivers. **MD35 1/4 - standard 48 steps per rev.....£12.99**

**MD38 - miniature 48 steps per rev.....£9.15**

**MD200 - miniature 200 steps per rev.....£17.10**

# MAGENTA ELECTRONICS LTD

## Teach-In '93



### MINI LAB KITS

ALL COMPONENTS TO ASSEMBLE THE EPE MINI LAB ARE AVAILABLE FROM MAGENTA.

The easy way to buy the correct parts to follow this exciting new educational series.

Components are supplied in packs to keep ordering simple.

A full MINI LAB consists of ML1, ML3, ML5, ML6. These are available at a special combined price of.....**£114.99**  
or less the p.c.b.

ML2, ML3, ML5, ML6 at.....**£104.99**

The transformer unit ML4 is also needed....**£21.45**

KIT ML1	MINI-LAB P.C.B. + all components inclusive of two breadboards for Part 1 (Nov. '92).....	<b>£49.95</b>
KIT ML2	All Components for Part 1 less p.c.b.....	<b>£39.95</b>
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# EVERYDAY WITH PRACTICAL ELECTRONICS

INCORPORATING ELECTRONICS MONTHLY

**VOL. 21 No. 12 DECEMBER '92**

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## SUCCESS

Success breeds success as they say and our first combined issue of *Everyday with Practical Electronics* (last month) has certainly been a success, so much so that this issue is even bigger. In an effort to get everything in we have produced our biggest ever issue. Eighty-eight pages of magazine plus sixty-four pages of catalogue. We even had to ask Bull Electrical to accept thinner than normal paper for their catalogue so that we could bind all the pages together and have a big enough cover to fit around it.

In short, we believe this is the best value ever hobby electronics magazine and we will try to keep it going in that direction. Of course our advertisers have recognised the value of our increased circulation and that also means more of their interesting advertisements and a very wide range of products for you to choose from.

## TEACH-IN '93

Our new *Teach-In '93* has also kicked off to an excellent start and we have had so many requests for the *Mini Lab* p.c.b. that there has been a delay in sending them out. If you had to wait for yours we are sorry but hopefully, by the time you read this, we will have caught up to our normal seven day turnaround.

Teach-In also upset some readers - see the *Readout* letters page - and we apologise for that. No excuses but the offending mnemonic is widely used in education and was suggested by someone in that profession.

## TOPICAL

In view of the recent UK coal mine closures and the subsequent debates and manoeuvres our *Alternative Energy* series makes interesting reading. This month's part puts the whole field of energy production into perspective and also gives some interesting figures on nuclear pollution, etc.

I certainly have gained some knowledge from the series and found it informative reading. I'm only sorry that we have had to omit a few interesting photos from the last two parts due to lack of space. There were, in fact, so many photographs available of wind farms, solar power stations, etc., that we could easily have filled the whole magazine with them. It just shows how much is going on in this field that many of us are unaware of.



## SUBSCRIPTIONS

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## READERS' ENQUIRIES

We are unable to offer any advice on the use, purchase, repair or modification of commercial equipment or the incorporation or modification of designs published in the magazine. We regret that we cannot provide data or answer queries on articles or projects that are more than five years old. Letters requiring a personal reply must be accompanied by a stamped self-addressed envelope or a self addressed envelope and international reply coupons.

All reasonable precautions are taken to ensure that the advice and data given to readers is reliable. We cannot however guarantee it and we cannot accept legal responsibility for it.

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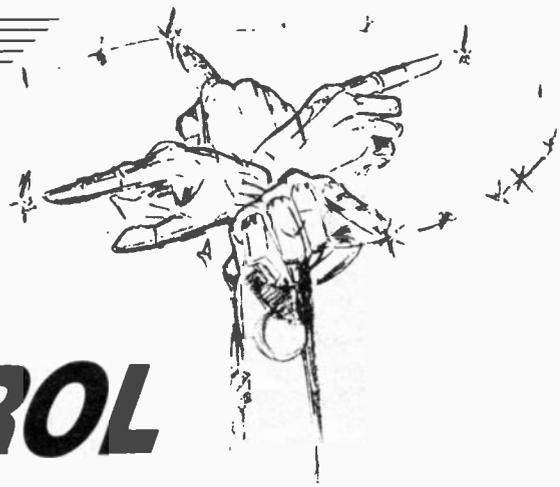
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# UNIVERSAL INFRA-RED REMOTE CONTROL

**CHRIS WALKER**

*One for the experimenter.  
Has been designed to be adaptable  
to almost any control requirement.*



**H**AVING been involved in the design of numerous electronic control and measurement systems, both on a personal level and when assisting Technology students, the author has frequently encountered projects which require the use of a remote control facility.

Some of the applications which spring to mind are: remote control of a free-running robot buggy; security lock systems; remotely controlled garage doors; remote entry of data into a personal computer; remote control of household appliances etc. Needless to say that readers could suggest countless more uses.

Quite often, all the different applications have a common remote control theme running through them, namely the need for a hand-held transmitter unit featuring one or more keys transmitting a short distance (some metres) to a receiver unit which indicates which key on the transmitter has been pressed and activates other parts of the system.

Since the transmitter/receiver remote control sections of the various different systems are all essentially identical, the author thought that it was high time to design a Universal Infra-Red Remote Control unit which could be adapted to a diverse range of uses. This article describes such a "drop-in" unit and applications are limited only by readers' imagination.

## SPECIFICATIONS

The entire Versatile Infra-Red Remote Control system is designed around three integrated circuits produced by GEC Plessey Semiconductors. Use of such advanced i.c.s greatly simplifies construction yet results in a flexible and powerful remote control facility.

The hand-held Transmitter can feature between one and thirty-two keys depending on the application requirement. Data may be transmitted at three different rates and will only be decoded by a receiver set to an identical data rate.

It is, therefore, possible to send 32 different codes to three different receivers in one room, a total of 96 code combinations. This powerful feature will probably far exceed the demands of most applications.

Standby transmitter current consumption, when no keys are pressed, is less than one microamp and the PP3 battery should give an extensive working life.

In the Receiver unit, which is operated from a power supply of between 5V and 25V, the infra-red (IR) data picked up from the Transmitter is amplified using a specially designed integrated circuit. The benefits of using a custom amplifier i.c. for

this job (rather than discrete transistors) are enormous: apart from simplifying construction, the receiver can reject a large amount of electrical and IR noise. Operation is possible in bright sunlight (important for external use) and also in rooms illuminated with artificial lighting.

The latter often causes a problem because mains-operated incandescent lamps (i.e. light bulbs) generate huge amounts of infra-red radiation which flickers at a rate of 100Hz. If this flicker was not rejected by the IR amplifier it would swamp the weak IR pulses from the Transmitter unit.

## IN CODE

The data output from the Receiver is in the form of a 5-bit binary code, see Table 1. Each output line can either be at 0V (logic 0) or +5V (logic 1) and the code on the output signifies which of the 32 possible keys on the Transmitter has been pressed.

This binary-coded output is ideal if the Receiver is to be interfaced to a computer or other digital device. In addition, there is a Data Ready output which can be used to signal to a computer that a valid transmission code is being received. The 5-bit output from the Receiver is tri-state and can be switched to high impedance (switched off) if necessary.

For more basic applications, this article terminates with a selection of circuits which will decode the binary output to provide individual off/on control of up to 32 appliances. The degree of complexity can be tailored to suit the desired application.

## TRANSMITTER CIRCUIT

The circuit diagram for the hand-held infra-red Transmitter is shown in Fig. 1.

The 32-switch keypad is wired as an eight-by-four matrix. Fig. 2 shows an example of how five switches can be wired to send codes 1, 2, 4, 8 and 16 (this is a particularly useful set of codes for simple applications as will be seen later).

When a key is pressed, one row of the matrix is connected to one column. IC1 detects which row/column pair are shorted together and transmits the relevant code.

For example, pressing key "1" connects pin 9 on IC1 to pin 12. Pressing key "16" connects pin 5 on IC1 via resistor R1 to +9V. In order to transmit code "26" pin 3 must be shorted to pin 11, and so on. Constructors can build up their own keypads to transmit a particular set of codes needed for their application.

Table 1: Decmal/Binary Codes

Decimal	Binary
0	00000
1	00001
2	00010
3	00011
4	00100
5	00101
6	00110
7	00111
8	01000
9	01001
10	01010
11	01011
12	01100
13	01101
14	01110
15	01111
16	10000
17	10001
18	10010
19	10011
20	10100
21	10101
22	10110
23	10111
24	11000
25	11001
26	11010
27	11011
28	11100
29	11101
30	11110
31	11111

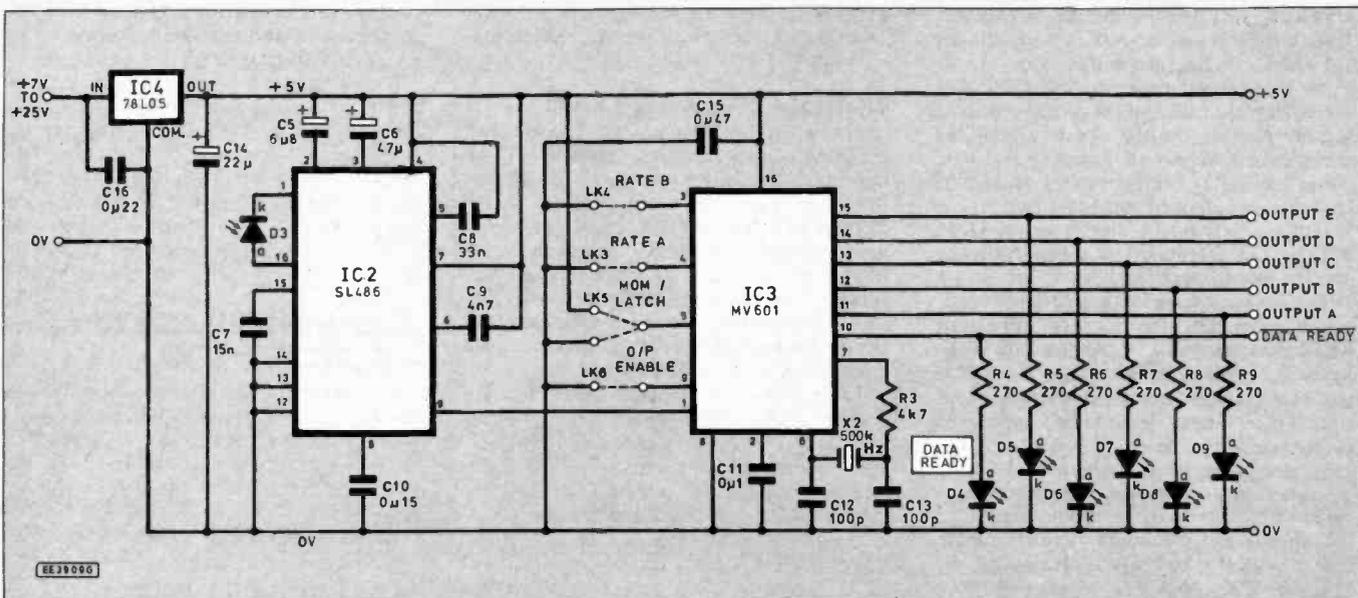


Fig. 4. Complete circuit diagram for the Receiver, including the voltage regulator IC4.

## RECEIVER CIRCUIT

The circuit diagram for the Receiver and decoder is shown in Fig. 4. The IR pulses from the Transmitter are picked up by the photodiode D3 and the resulting current pulses through the diode are processed by amplifier IC2.

Capacitors C5 and C6 control the frequency dependent gain of a gyrator circuit within IC2. The gyrator only significantly amplifies the short pulses of infra-red data but ignores continuous background IR such as sunlight or room lighting.

Following the gyrator are three stages of amplification with decoupling capacitors C7, C8 and C9 which *de-sensitise* the amplifiers low frequency infra-red variation such as 100Hz lamp flicker. Capacitor C10 governs the automatic gain control which assists operation under "noisy" conditions.

The processed data emerges from IC2 at pin 9 and is routed directly to the p.p.m. data input (pin 1) of the decoder, IC3.

In likeness to the Transmitter circuit, ceramic resonator X2 and capacitors C12 and C13 form a 500kHz oscillator. Resistor R3 is used as a "damper" to prevent X2 oscillating at harmonics of its fundamental frequency.

The rate inputs A and B (links LK3 and LK4 respectively) should be set to match the Transmitter. Note that, unlike the Transmitter, leaving a link disconnected on the Receiver sets that rate input to a logic 1. Once again, switches may replace all these links if desired.

Once two valid data words have been received the five-bit data is placed on the five output lines (IC3 pins 11 to 15), Output A (least significant bit) to Output E (most significant bit). After a short delay (about 4µs) the Data Ready line (pin 10) drops to logic 0 and stays low as long as valid data is being received. All outputs are compatible with TTL and CMOS logic levels.

## OPERATING MODE

When data transmission stops, Data Ready goes high again and the five data outputs will behave in one of two ways.

- If link LK5 is set to logic 1 (momentary mode), the outputs will all return to 0V.
- If link LK5 is set to logic 0 (latched mode), the outputs will remain latched to the last valid code received and will not change until a new code is received.

Light emitting diodes D4 and D9 indicate the logic states of the outputs and, although their presence is not essential, it is recommended that they are included as they greatly aid setting-up and fault tracing.

In "latched mode" the data outputs may be reset to 0V at any time by briefly taking IC3 pin 2 to 0V. Capacitor C11 provides a reset pulse during power-up.

The tri-state outputs (A to E) may be put into a high impedance state by taking IC3 pin 9 to a logic 1 level. The ability to isolate the output lines makes it easy to interface them directly to a microprocessor bus where it is important that they do not clash with other devices serving the bus. For "normal" use, link LK6 enables the outputs continuously by tying pin 9 to 0V.

## RECEIVER POWER SUPPLY

The power supply requirements for IC3 calls for a well regulated +5V rail and this is supplied by the voltage regulator IC4. The input voltage to IC4 should lie in the range +7V to +25V. Current consumption of the Receiver circuit rises to about 60mA when *all* l.e.d.'s are switched on.

Capacitors C14 and C15 decouple and stabilise the +5V supply rail and capacitor C16 helps to stabilise IC4. If a clean, regulated 5V source is already available in the main system in which the Receiver is being used then IC4 and C16 may be omitted and the Receiver can be supplied with power directly onto its +5V rail.

## CONSTRUCTION - TRANSMITTER

Thanks to the use of dedicated integrated circuits to perform the complex functions of data encoding and decoding and signal processing, the construction of the Transmitter and Receiver sections is very straightforward.

The Transmitter and Receiver printed circuit board (p.c.b.) component layouts, along with the full size copper track patterns, for the Universal Infra-Red Remote Control are shown in Fig. 5 and Fig. 6. These two boards are obtainable, as a pair, from the *EPE PCB Service*, codes 811T and 811R.

Commence construction with the Transmitter p.c.b. Solder all the components to the board, paying attention to the polarity of the electrolytic capacitor C4. Similarly,



REMOTE CONTROL

transistor TR1 must be inserted the correct way round as must the two infra-red diodes D1 and D2 (details given in Fig. 5).

In the prototype unit these two diodes are soldered directly to the p.c.b. and their lenses protrude through the end panel of the plastic case which houses the entire Transmitter. This anchorage helps to support the circuit board within the case.

Use an 18-pin d.i.l. socket for IC1, do NOT solder the i.c. directly to the board, and do NOT insert the i.c. into its socket until all construction work is complete.

Use p.c.b. terminal pins for all flying lead connections, that is connections to the keypad and to the battery clip for B1. It is also convenient to make the links LK1 and LK2 between terminal pins as they can be changed at a later date.

At this stage, decide which data transmission rate to use (see earlier in text) and solder link LK1 to set the Rate A input of IC1 to either +9V (logic 1) or 0V (logic 0). Do similar with the Rate B input using

link LK2. The links may be replaced with changeover switches if regular alteration of the data rate is anticipated.

## KEYPAD CASE

The actual type and size of case used depends on the number of keys to be crammed onto the front panel. Constructors can create their own keypads from discrete switches wired to give the required set of transmission codes or it may be more convenient to buy one of the numerous types of matrix-wired keypads available, ranging from the "telephone style" pad to the "rubber membrane" type which can be customised with your own logos.

The prototype model uses one of the low-cost telephone style keypads, with numeric keys 0 to 9. It should be noted that, because the keypad is only three columns wide, and since it is pre-wired in a matrix fashion, the transmitted codes do not correspond to the numbers on the keys (see Fig. 2).

The "ON" resistance of the keypad

switches can be anything up to 20 kilohms, so the switches used do not need to be particularly high quality!

If you possess an oscilloscope or a logic probe, you can test the completed Transmitter circuit by connecting the probe between the 0V line and the "drain (d)" terminal of transistor TR1. You should observe a potential of about 6V which, when a keypad switch is pressed, pulses down to almost 0V showing an inverted version of the type of p.p.m. data illustrated in Fig. 3.

## CONSTRUCTION - RECEIVER

The receiver printed circuit board component layout and track pattern is shown in Fig. 6. Once again, assembly of this board is very straightforward and the components can be inserted in any convenient sequence.

Use two 16-pin d.i.l. sockets for IC2 and IC3, the voltage regulator IC4 can be sol-

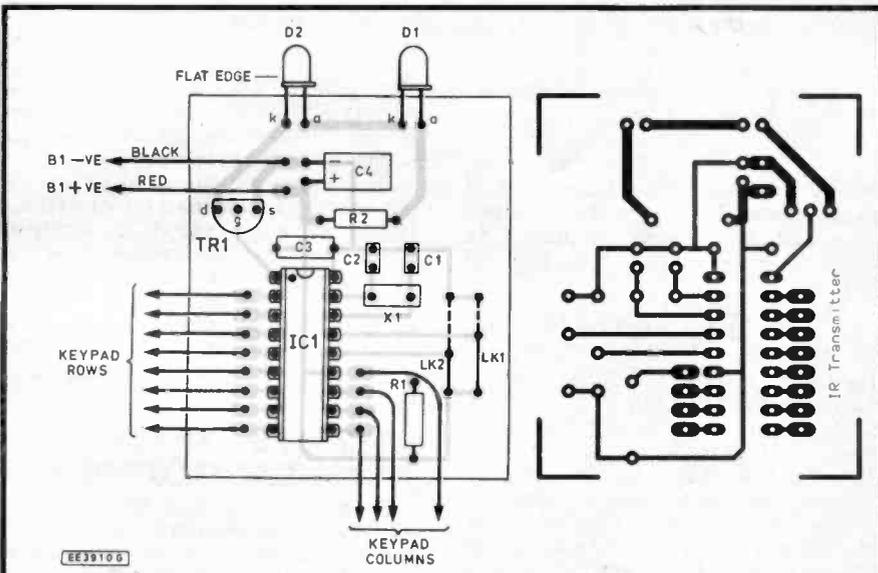
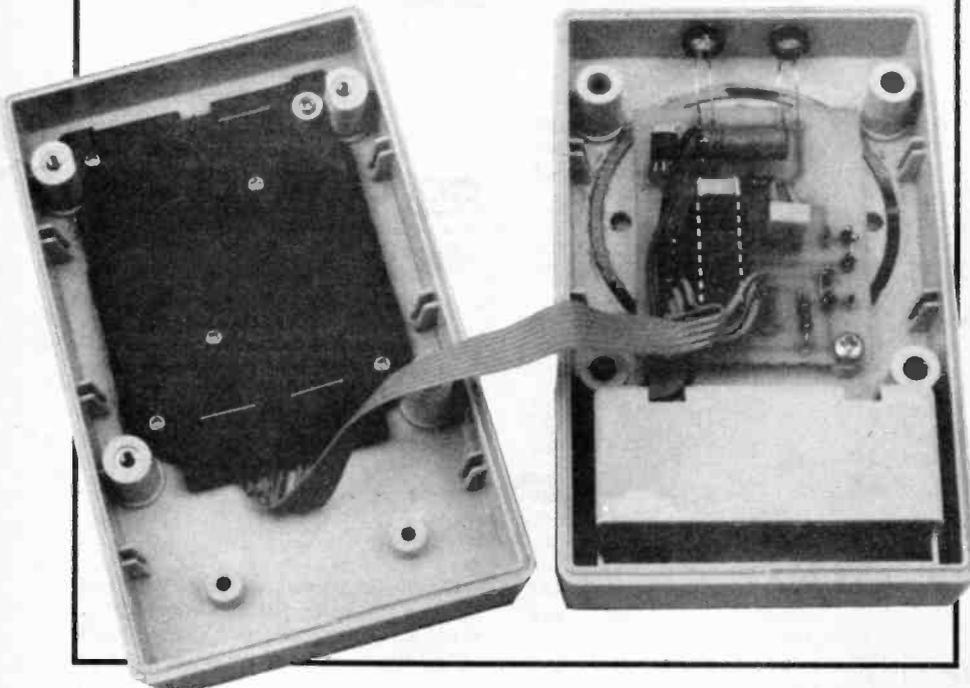


Fig. 5. Printed circuit board component layout, lead-off wiring and full size underside copper foil master pattern for the hand-held transmitter. The completed transmitter, the position of the p.c.b. and the ribbon cable from the board the the keypad can be seen in the photograph below.



## COMPONENTS

### Transmitter & Receiver

#### Resistors

R1	10k
R2	1
R3	4k7
R4 to R9	270 (6 off)
All 0.25W 5% carbon (or better)	

See  
SHOP  
TALK  
Page

#### Capacitors

C1, C2,	
C12, C13	100pF ceramic (4 off)
C3, C11	0μ1 polyester (2 off)
C4	100μ radial elect., 25V
C5	6μ8 tantalum bead, 16V
C6	47μ radial elect., 25V
C7	15n polyester
C8	33n polyester
C9	4n7 polyester
C10	0μ15 polyester
C14	22μ radial elect., 25V
C15	0μ47 polyester
C16	0μ22 polyester

#### Semiconductors

D1, D2	High power infra-red emitters (2 off)
D3	Infra-red sensitive photodiode
D4	Green l.e.d.
D5 to D9	Red l.e.d. (5 off)
TR1	VN10KN n-channel MOSFET
IC1	MV500 transmitter
IC2	SL486 IR amplifier
IC3	MV601 receiver
IC4	78L05 5V 100mA voltage regulator
X1, X2	500kHz ceramic resonator (2 off)

#### Miscellaneous

B1	9V PP3 battery and clip for transmitter
Single-pole push switches for keypad or ready-made membrane keypad etc.; suitably-sized case for transmitter; 16-pin d.i.l. sockets (2 off); 18-pin d.i.l. socket; terminal pins; connecting wire; solder, etc.	

Printed circuit board for Transmitter (811T) and Receiver (811R) available from the EPE PCB Service as a pair.

Approx cost  
guidance only

£32

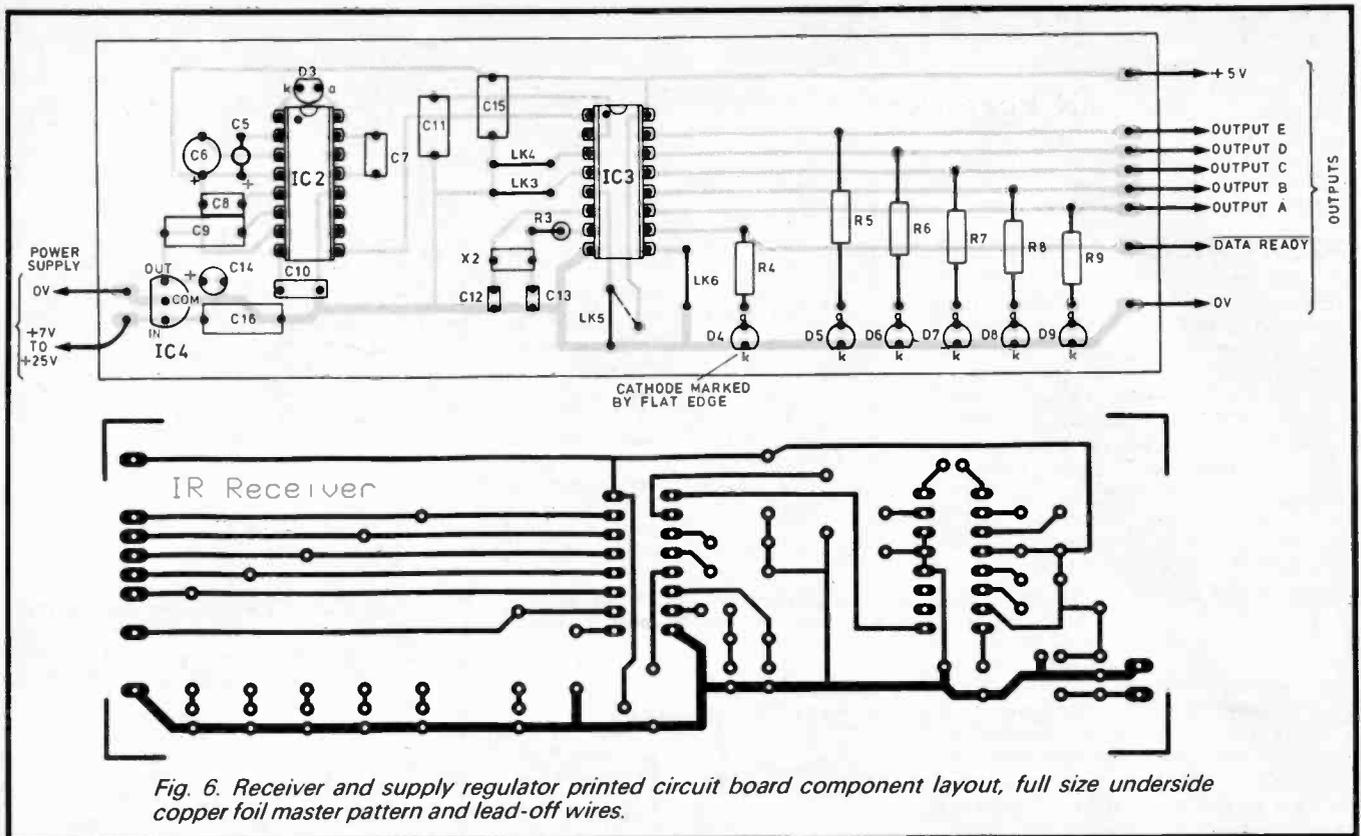


Fig. 6. Receiver and supply regulator printed circuit board component layout, full size underside copper foil master pattern and lead-off wires.

dered directly to the board. The tantalum bead capacitor C5 and the two electrolytic capacitors C6 and C14 must all be inserted the correct way around; pay careful attention to the *plus* or *minus* markings on their cases. Similarly, the six light emitting diodes D4 to D9 have a small flat edge on their case which denotes the cathode (k) lead.

The cathode lead of photodiode D3 needs to be soldered to the track connecting to pin 1 of IC2. Methods of notating the cathode vary on different photodiodes so refer to the manufacturer or supplier for details.

The infra-red amplifier (IC2) is very sensitive and to avoid excessive noise being picked up, the photodiode should be soldered *directly* to the board rather than being linked to the board with flying leads. When used in surroundings which are prone to electrical noise, the entire Receiver p.c.b. may need to be screened inside a metal box containing a small "window" to admit IR light to the photodiode. Screening was not found to be necessary when using the prototype unit.

Insert terminal pins into the p.c.b. for all the off-board connections and for the links LK3 to LK6. The data rate links LK3 and LK4 should be soldered to *match* those set on the Transmitter. On the Receiver p.c.b.,

leaving a data rate link **disconnected** sets that Rate input to logic 1.

Link LK5 should be made between IC3 pin 5 and 0V for latched output mode, or pin 5 and +5V for momentary mode. Unless the tri-state output feature is required, insert link LK6 to enable the outputs continuously.

## TESTING

Before inserting IC2 and IC3 into their sockets, test the voltage regulator on the Receiver p.c.b. by connecting a power supply of between 7V and 25V to the Receiver board. Using a voltmeter, check the p.d. between the 0V and +5V terminal pins adjacent to the Receiver outputs; the reading should be close to 5V.

If the voltage is significantly away from 5V or if voltage regulator IC4 becomes hot, switch off and find your mistake! Do *not* insert the other two i.c.'s until they can be sure of a safe 5V supply.

Remove power whilst inserting IC2 and IC3. When the completed circuit is switched on, only the "Data Ready" i.e.d. (D4) should illuminate. It is worth re-checking the Receiver power supply with the chips now in their sockets; it should still be around 5V.

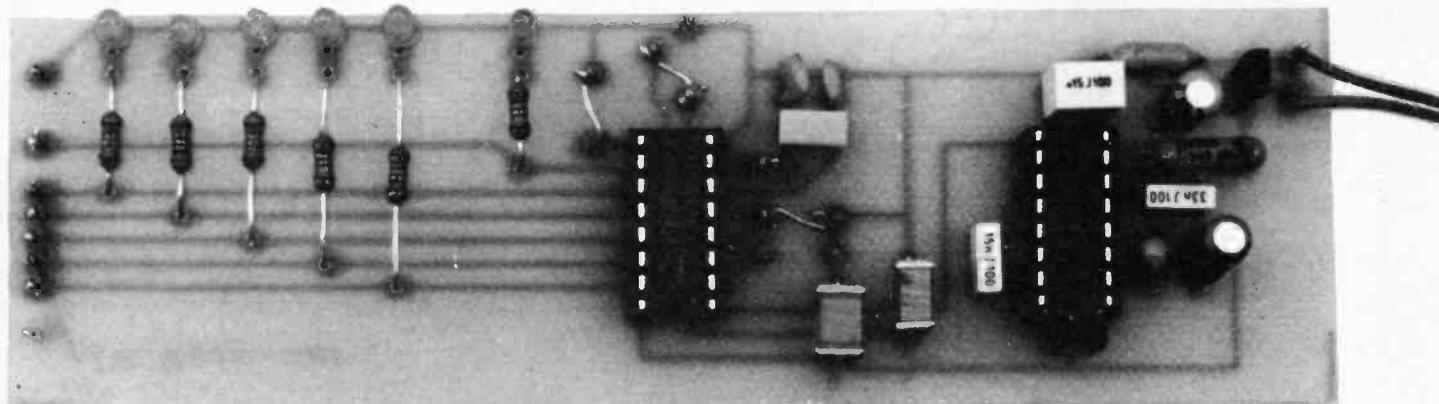
Pressing a key on the Transmitter should cause the Data Ready i.e.d. to extinguish and the five output i.e.d.'s (D5 to D9) to display the binary code of the selected key. Voltage regulator IC4 may become warm when several i.e.d.'s are lit up together; this is normal.

If the Receiver fails to respond to the Transmitter, check for obvious errors such as wrongly inserted i.c.'s or a flat Transmitter battery. Don't forget that the selected data rate on the Transmitter *must match* that on the Receiver. Also, link LK6 must be made for the i.e.d.'s to light.

An oscilloscope connected between the 0V line and IC2 pin 9 should display the received p.p.m. data (see Fig. 3) after amplification. If there appears to be excessive background noise on this signal then screening of the Receiver p.c.b. may be necessary.

## OUTPUT CONSIDERATIONS

This completes assembly of the Universal Infra-Red Remote Control. What now follows are some suggestions for using the 5-bit Receiver output to control appliances in various modes. No construc-



tional details are given, but it is hoped that these circuits will give constructors ideas for their own customised designs.

It is suggested that the required output circuit is initially constructed on a prototype board (breadboard) linked to the IR Receiver. When its satisfactory function is assured, it can be built permanently on p.c.b. or stripboard.

### CURRENT DRIVER

Each output line from the MV601 decoder (IC3) can supply a current of up to 45mA, 10mA is used to illuminate the output l.e.d., leaving 35mA to drive an external device.

To power any respectable-sized output device, some form of current amplifier is needed. The transistor/relay combination shown in Fig. 7 is the easiest and safest method of achieving this.

Two output drivers are shown, but the arrangement may be repeated for the number of outputs required. The 12V power supply is not exceptionally critical and does not need to be regulated.

The contacts of each relay are used as switches to turn on and off the current to each output device. Low voltage devices can be powered from the 12V supply, or mains-operated devices can be switched provided the relay contacts are rated for such work.

*Take great care if you are dealing with mains voltages. Make sure that you fully understand what you are doing; if in doubt, get help from someone with the necessary experience. Mains electricity can kill!*

It is possible to independently control up to five appliances directly from the five output lines, A to E. Pressing key "1" on the Transmitter will make output A go high and turn on the relay connected to this output. Similarly, key "2" will make Output B go high; key "4" activates Output C; key "8" activates output D and key "16" activates output E. Wiring details of the five-button keypad for this application are shown in Fig. 2.

Furthermore, if two or more of these five keys are pressed simultaneously then the appropriate lines will still respond.

Fig. 8. Using two 3-to-8 decoders to give 16 decoded outputs.

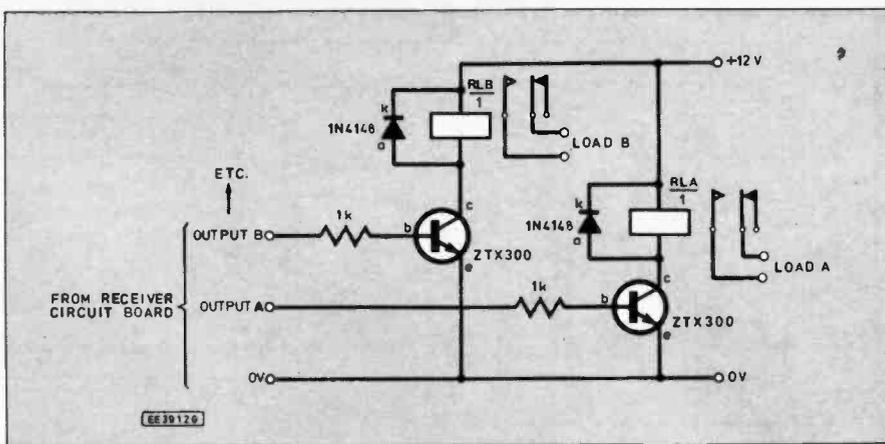
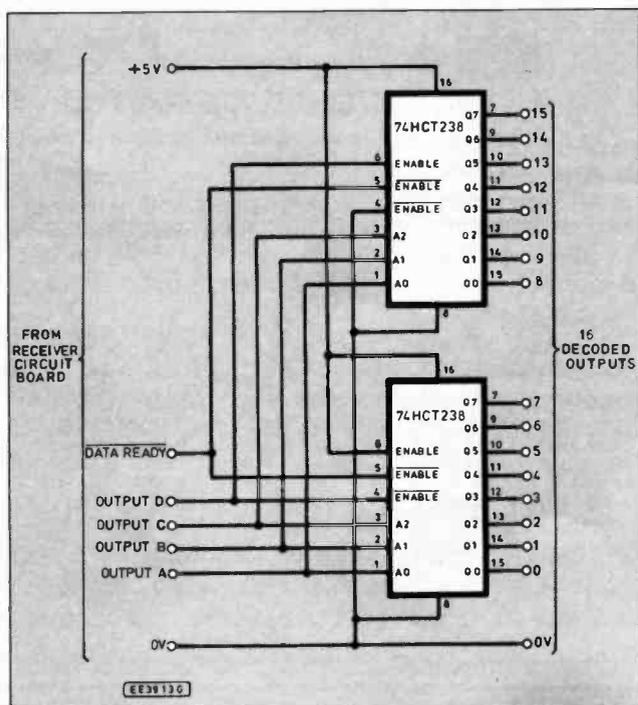


Fig. 7. Suggested transistor/relay output switching circuit diagram.

For example, pressing keys "8" and "2" together will switch on Outputs D and B. Neat, eh!

Depending on the desired application, the outputs may be used in latched or momentary mode.

### OUTPUT DECODING

If more than five separate outputs are required then some degree of decoding is required.

The circuit diagram Fig. 8 shows how two 3-to-8 line decoders can be chained together to give sixteen independent outputs. Only one output will go high at any one time depending on which key, from 0 to 15, had been pressed on the Transmitter.

If eight or less outputs are required then only one 74HCT238 i.c. is required. Furthermore, chaining in further 74HCT238 chips will allow all 32 outputs to be decoded. This feat is left as a reader's exercise!

Every output from the decoder i.c.s will need a transistor/relay current switch as shown in Fig. 7 if they are to control external appliances. The 5V power supply for the decoder i.c.s comes from the IR Receiver board.

### OUTPUT TOGGLING

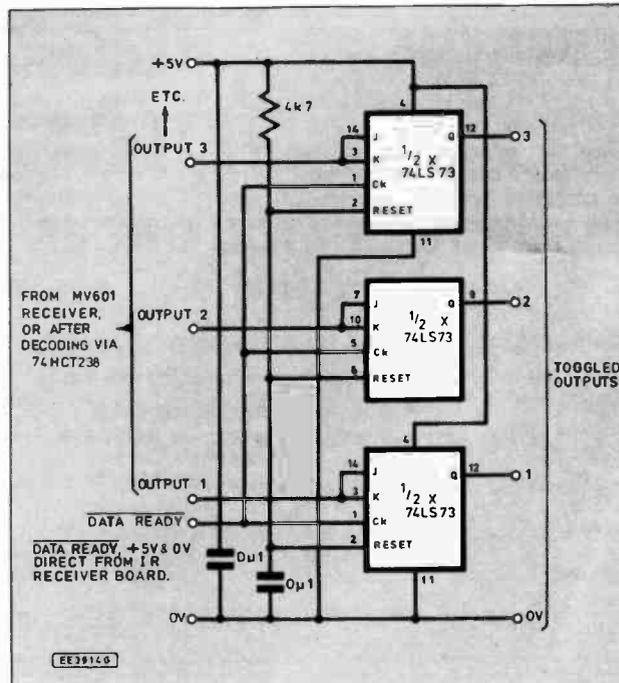
It is often convenient to be able to switch an appliance ON with one press of a Transmitter key, and switch it OFF again with another press of the same key. This on-off-on switching is called "toggling" and the circuit of Fig. 9 shows one way of achieving it.

Each box in the circuit is a J-K flip-flop and there are two such flip-flops in each 74LS73 integrated circuit. A separate flip-flop is required for every output to be toggled. The inputs to each flip-flop come either direct from outputs A to E on the IR Receiver p.c.b. (if five or less outputs are required) or from the 74HCT238 decoder i.c.s (if more than five outputs are needed).

The Reset inputs on all the flip-flops are connected together and the 0μF capacitor and 4k7 resistor provide a reset pulse at switch-on to put all outputs in the low (off) state. All outputs can be reset at any later time by momentarily pulling the Reset line down to 0V.

The 5V power supply originates from the IR Receiver board. For controlling external appliances, each toggle output will need buffering with the transistor/relay arrangement shown in Fig. 7.

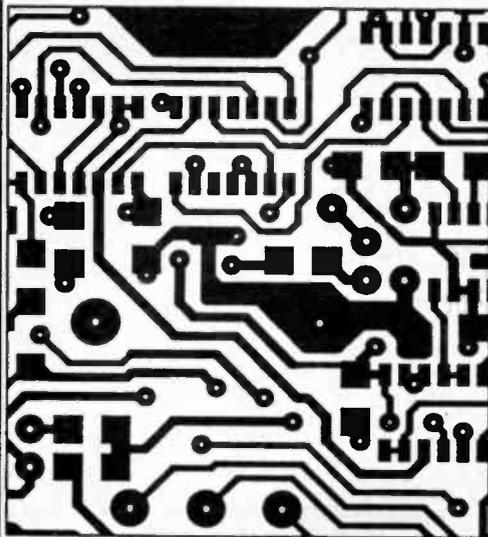
Fig. 9. Circuit diagram to enable on-off "toggling" to be achieved.



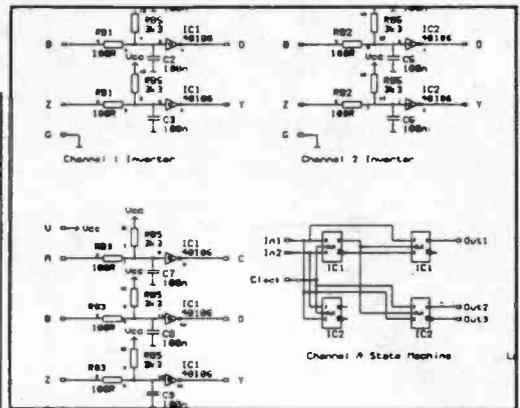


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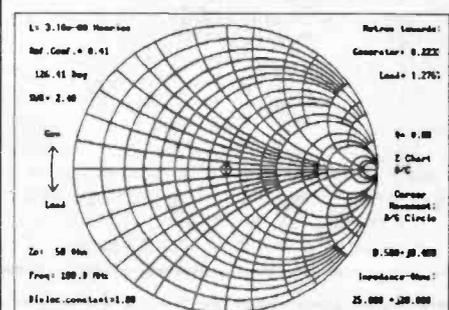
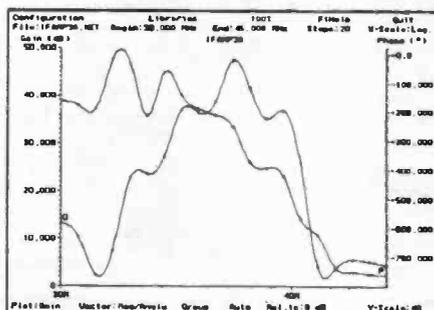
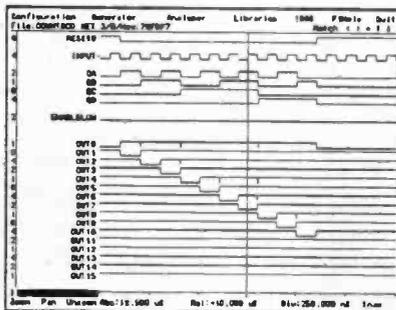
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# Innovations

A roundup of the latest Everyday News from the world of electronics

## Keep Your Distance

**A** SYSTEM that provides an optical or acoustic signal to indicate the distance between vehicles when parking would be a big help, especially to learner drivers. That is only one potential application for the Sharp GP2D02 Distance Sensor. Other conceivable uses include, for example, measurements of fluid levels inside closed vessels, automatic flushing for toilets, control of tools and machining operations for CIM systems, etc.

The Distance Sensor functions as both, a proximity sensor and a distance measurement sensor, and is unaffected by the object colour or reflection ratio. An infra-red l.e.d., a position sensor and a signal-processing circuit are all integrated into a single module which is less than one-eighth the size of conventional components!

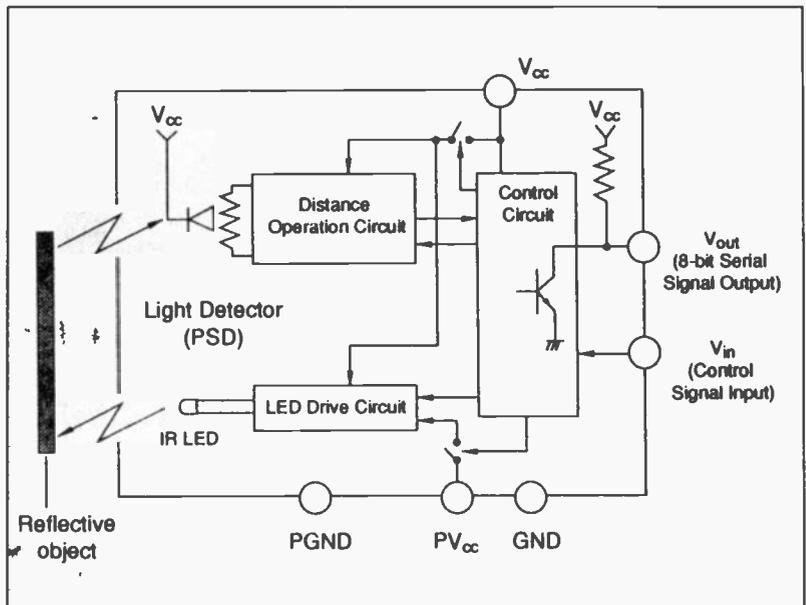
### How It Works

The GP2D02 is based upon the principle of using a light-emitting diode to send out high-speed bursts of infra-red light so that objects lying within the beam of light (angle of dispersion  $\pm 3^\circ$ ) reflect part of the light back to a unique sensor developed by Sharp.

The relationship between two currents in the position-sensitive detector changes as a function of the angle of reflection, which is determined by the distance between the object and the sensor. This change is monitored at various points, and the signal-processing circuitry uses this information to produce an 8-bit value which is also the average of the various input values.

This method permits a claimed accuracy to within  $\pm 10\%$  for distances between 4cm and 30cm (Type 1) or 10cm to 80cm (Type 2). The measurement range can also be varied to meet customer requirements by changing optical components.

The module consumes 7mA in the quick-charge mode and 4.5mA in the standard mode. The sensor switches to a stand-by mode in the absence of reflected light, and current consumption drops to 7 $\mu$ A. The infra-red l.e.d. operates on voltages between 2.7V and 7V and the control circuit on voltages between 2.7V and 3.5V.



## HOME AUTOMATION

HOME AUTOMATION is in its early days – about the same stage as the micro computer in the early '80s. It offers a tremendous opportunity for the enthusiast to contribute to progress in the same way as with micro computers.

The current home environment is a mass of discrete systems, lighting, heating, security, audio/visual, just waiting to be linked to provide convenience, security and energy savings. Future home automation standards are under development (Echelon, CEBus) though almost all of today's systems are based on the well established X-10 power line carrier standard.

Smart House Systems offers free advice and an ever expanding range of products from X-10 lighting control modules at £19.95 to the Ener-

logic home automation controller at £349.95.

Smart House Systems Ltd, Dept EPE, 3 Buckland Street, Largs, Ayrshire KA30 8PP. Tel: 0475 672589

### BUILD YOUR OWN PC

ORGANISERS of *The Christmas Computer Shopper* show, Blenheim, are hosting the UK's first Build Your Own PC competition. The competition will take place during the event from 19 to 22 November.

Schools throughout London have been invited to put together teams for the competition. Each received a copy of the manual in advance and will compete under the guidance of their IT Master to build a PC in the fastest time. The winning team will keep the PC for their school.

## 21 YEARS COMPETITION

THE little free competition we ran in the news pages of the October '92 issue of EE resulted in a truly underwhelming response. Perhaps there are fewer readers that still have the first issue from 21 years ago than we expected. However, the three (yes three!) entrants each get a free Maplin catalogue and just to thank them all for taking the trouble we have also sent them a magazine binder each.

The correct answers were:

The first Teach-in was written by *Mike Hughes*

The first Shop Talk was written by *Mike Kenward*

(Yes indeed the present Editor, though I have had other jobs in the meantime-Ed.)

Our "lucky" prizewinners are: Mr S. E. Lartice from Enfield; Mr S. Keir from Dundee and Mr S. Carter from Portsmouth.

See the Readout page (page 795) for another "little competition" this month!

# Young Radio Amateur of the Year



**T**HE *Young Amateur of the Year* award for 1992 has gone to 17 year old radio buff Martin Saunders (call sign G7JCJ) of Broadstone, Dorset.

Martin was presented with the first prize of £250 by the Radiocommunications Agency's Head of Mobile Radio, Stephen Spivey, at the recent Radio Society of Great Britain's HF Convention at Windsor.

He also received a certificate signed by Michael Heseltine, President of the Board of Trade, to commemorate the presentation, and will be invited to the Agency's Radio Monitoring Centre at Baldock, Hertfordshire, for a conducted tour.

Martin's main interest is in packet radio, which involves transmitting data in packets of information. He assembled his own equipment and has been granted a Notice of Variation to his Amateur Radio Licence to operate a mailbox, a computerised distribution agency which can be accessed by other radio amateurs wishing to hear a message.

He has written a number of articles explaining packet radio and amateur radio generally. He has been appointed Secretary of his local packet group and serves on the forward planning committee of the Flight Refuelling Amateur Radio Society.

His close runner-up was Neil Mothew, aged 16 (call sign G7NGM) of Ashfields, Loughton, Essex, who was also very strong on home construction of radio equipment. Neil will also be invited to Baldock.

## Breach of Security

A RECENT National Computing Centre (NCC) survey estimated that security breaches will cost Britain £1.1 Billion this year. This means that every IT installation in the UK stands to lose on average £22,000 this year due to loss of security. Even this figure may be too low as many incidents are unreported.

# APPROVAL FOR CIRKIT

CIRKIT are pleased to announce that they have achieved approval to BS5750, Part 2.

The approval has been awarded as part of the company's on-going commitment to quality and service, meeting the exacting standards of BS5750, Part 2 - with the additional stockists supplement and equivalent to European EN29002 and International ISO9002.

As everyone involved in the electronics industry will know, this is not an easy achievement. This coveted award is granted only after a thorough examination of the company's systems and procedures, and reflects the commitment of the entire workforce to a consistently high level of service.



*Cirkit's Quality Manager, Andrea Chapman, responsible for the successful granting of BS5750 approval.*

# THEY'VE GOT THE POWER

A DOUBLING of power is claimed by BK Electronics with the release of their new 1000W MOS-Fet amplifier module. The OMP-MF1000 has taken over from the ever popular OMP/MF450 as the flagship of the OMP module range.

Delivering 1000W r.m.s. into a two ohm load, the OMP/MF1000 is claimed to provide a no compromise on price/power/performance ratio and, as with all of the OMP range, its advanced MOS-Fet circuitry achieves superb sonic clarity with enduring reliability.

The OMP/MF1000 is priced at £259, including VAT, plus £12.50 delivery, and

is available from BK Electronics, Dept EPE, Units 1&5 Comet Way, Southend-on-Sea, Essex SS2 6TR. Tel: 0702 527572



## AMS6 & Radio Rally

THE All Micro Show and Radio Rally will be held on Saturday, 14 November at Bingley Hall, Staffordshire Show Centre, Stafford.

This is the fourth consecutive year at the Staffordshire Show Centre and the All Micro Show is now established as a major computer and electronics show in the Midlands, ideally timed for the Christmas trade.

Previous shows have attracted a diverse mix of exhibitors ranging across the computing and electronics spectrum, including national PC suppliers, individual user groups, electronics component suppliers, TV and Video sales and spares, Ham Radio, Computer Media etc.

## Impact On Growth

CLAIMING to be one of the worlds largest dot-matrix printhead repair companies, the American based Impact Printhead Services, opened its UK offices in January 1991.

It has obviously struck a niche in the market. Sales in 1992 have increased by 344 per cent compared with the same period last year.

"We have been staggered by our growth" said Andrew Gadd, UK General Manager, "but when we can save a customer over half the price of a new printhead, and give a one year warranty, it's not really that surprising".

Impact can be contacted at The City Business Centre, Dept EPE, 10 City Business Centre, Brighton Road, Horsham, West Sussex RH13 5BA. Tel: 0403 272627.

## European Customer Service

AN agreement between BT and Olivetti has been signed to ensure round-the-clock maintenance and support for the telecommunications giant's growing number of customers in Europe.

The agreement gives BT's customer support organisation access to Olivetti's established customer service infrastructure *Oliservice* throughout mainland Europe and means that customers can be offered service level agreements which guarantee response times of as little as two hours. The European Customer Service Centre (CSC) in BT's Paris Eurocentre provides multi-lingual support to customers. The CSC, which also houses a network control centre, will now be able to call upon trained Olivetti staff in addition to its present resources.

The European Customer Service Centre is also linked to network management centres in London and the US, providing comprehensive service and support.

# New Technology Update

*Ian Poole reports on techniques for reducing i.c. size, high output l.e.d.s and low voltage logic*

ONE of the major thrusts of i.c. technology development is to produce ever smaller structures for use in chips. By doing this more can be packed onto a single i.c. and this in itself brings many benefits. Increased speed and reliability together with reduced costs are just three.

This pressure for size reduction is being applied to the reduction of dynamic RAM sizes as much as anywhere else. However one of the major size restraints is imposed by the small capacitors which are used to store the data in these chips. For the individual cells to be able to reliably store data they must have sufficient capacitance. If the capacitor is too small then insufficient charge will be stored for the data to be detected at a later time.

## Capacitors

The main problem for RAM designers is that the amount of capacitance required governs the physical size of the capacitor. To try to overcome this capacitors using a V or trench structure have been used. This gives a larger area for the capacitor whilst taking up no more surface area on the chip itself. Whilst this method is reasonably successful it only provides a limited improvement.

A much more promising approach is now being developed. The key to it has been made possible because researchers have been able to deposit very thin layers of titanium oxide onto the i.c.s. This material has about 20 times the dielectric constant of silicon oxide/silicon nitride which is normally used. Consequently it means that the size of the capacitors can be reduced by about the same ratio.

Apart from the reduction in size, titanium oxide offers a number of other advantages. The most important is that capacitors made with it do not suffer from the same loss of charge that ones made from silicon oxide do. This means that these i.c.s will be more reliable than their silicon oxide counterparts.

Whilst there are several advantages to using titanium oxide there are also some disadvantages. The major one is that the processes needed for handling titanium oxide need to be controlled very stringently. Even so, they are now well advanced and i.c.s using the new process should be seen on the market within the next year.

## Bright Lights

Traditionally l.e.d.s have only been used as small indicators. They have not had the power to compete with tungsten lamps.

Whilst some high power red and orange l.e.d.s have been manufactured no yellow ones have been available. This is set to change. Toshiba have recently launched a range of high output yellow l.e.d.s which can be used in a number of areas where only tungsten lamps have been used.

To obtain the high light output a totally new structure has been used. This was developed because the l.e.d. itself absorbs a large amount of light, reducing the brightness quite considerably.

To overcome this absorption the new structure uses an indium gallium aluminium phosphide active layer. This layer also includes a reflector consisting of a multi-layer structure to cut the light loss.

By using this structure the light output of the l.e.d.s has been raised to a level which is typically ten times as much as standard yellow l.e.d.s. A figure of six candela at a wavelength of 590nm is quoted. This means that these l.e.d.s will be able to be used in a number of outside applications including roadside displays where previously only the relatively inefficient tungsten lamps have been used.

## Low Voltage Logic

Five volts has been a standard voltage rail for digital circuits for as long as most people can remember. It has served well as the supply voltage for virtually all the families of i.c.s since before the power hungry standard 74 family of i.c.s was introduced.

Now technology is advancing and more requests for very low power systems are being placed before system designers. Laptop and notebook computers are two prime examples of the need for power reductions.

By reducing power consumption battery lifetime can be increased in addition to reducing the size of batteries. This is a great advantage because the two main complaints about laptop computers are the short time which these computers can be used on their batteries between charge and the size and weight of the batteries.

To achieve power reduction designers are turning to lower voltage logic as a means of achieving a solution. As a result it now appears that 3.3 volts is becoming an alternative standard to 5 volts for these applications.

A surprising amount of progress is being made. All the major manufacturers of 286, 386, and 486 processor chips are now developing or actually marketing 3.3 volt

versions of these chips. In addition to this most of the peripheral chips are now available to operate along side the processors.

For the i.c. manufacturers there are a number of advantages to using 3.3 volts. Having lower supply rails means that some of the internal dimensions of components within the chips can be reduced. The most obvious is that the thickness of the gate insulation layer in the f.e.t.s can be reduced from about 150Å to just 100Å. Gate sizes can also be reduced. Consequently the component sizes in the chip can fall quite significantly.

## Redesign

The modifications to enable the i.c.s to operate at lower voltages require some major changes. A full process redesign is needed in addition to any small dimensional reductions.

Many of the circuits have to be changed. This can be very expensive but it has to be undertaken to ensure that the circuit operates correctly and efficiently at the new voltage. Transistor thresholds need to be altered. Also inefficient logic structures need to be replaced by new ones which operate more satisfactorily at the new levels.

Standard techniques do not always provide the best solution. Often new techniques have to be used. One such example is a new process called SIMOX (separation by implantation of oxygen) which can give enormous improvements.

It has been found that transistors fabricated using this process have a much higher drive capability. On top of this they can run 50 per cent faster than transistors fabricated in the normal way. This means that the new transistors can operate just as fast at 3 volts as standard ones do at 5 volts.

With the reduction in size and the use of SIMOX designers now have several options open to them. If all the parameters are optimised for speed then the new 3.3 volt chips can operate faster than their 5 volt counterparts. Alternatively the chip could be designed to give the lowest power consumption, or the smallest chip size whilst keeping the performance the same.

Whatever option is taken by the i.c. designers the new 3.3 volt logic is certainly an exciting development which is likely to have a large presence in the semiconductor market of the near future.

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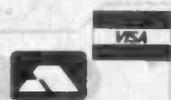
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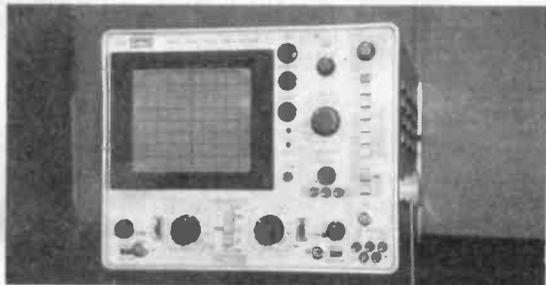
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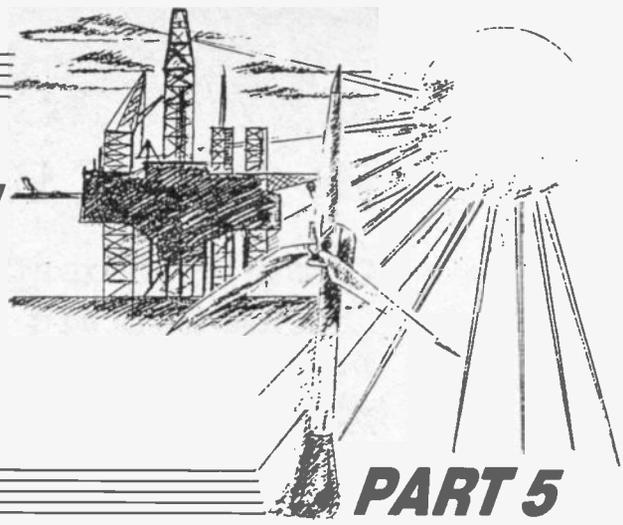
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# ALTERNATIVE ENERGY



T. R. de VAUX BALBIRNIE PART 5

Some further Alternative Energy sources.

**T**HIS is the last in a short series about Renewable Energy – that is, sources of power which will last for ever unlike those derived from fossil fuels which will eventually run out. This month we shall look at some remaining renewable energy sources which have the potential for large-scale exploitation in the UK.

These are: energy produced by burning household and industrial waste, landfill gas, biomass and geothermal energy. These methods have been grouped together because they all involve the use of direct or indirect heat.

We shall also take the opportunity to look more closely at conventional nuclear power - that is, nuclear fission. Although not itself a source of renewable energy, it also produces heat and will help to put the whole topic of alternative energy into perspective.

## WASTE NOT WANT NOT

Currently, 30 million tonnes of waste material per year are disposed of in the UK – that is, around half a tonne for every member of the population. Some 90 per cent of this is disposed of in the 5,000 landfill sites which are scattered around Britain. Using waste to provide energy seems a very attractive idea and one in which the Department of Energy is actively involved through a programme of funding research and development.

Waste material suitable for providing a useful amount of energy may be divided into two groups – *dry organic waste* and *wet waste*. Dry waste may, after suitable preparation, be burned directly. The fairly low technology required and the simplicity of the method are its greatest strengths. A further advantage is that it can reduce vast quantities of waste into a small volume and so simplify its ultimate permanent disposal.

## DRY WASTE

Burning dry household and industrial waste involves certain problems. Firstly, the energy content is likely to be much smaller than in a comparable quantity of fossil fuel. Also, there will be a large amount of incombustible matter mixed with it.

Removal of certain types of material is possible such as the ferrous metals using an overhead magnet. However, sorting some of the waste at source would help. Sensible waste management involves *recycling*, whereby the producer of the waste removes certain combustible and non-combustible

material such as paper, glass and metals for re-use.

Obtaining heat by burning dry waste in incinerators may be used for space heating, industrial processes close to the site or to produce electricity. This process already saves the equivalent of 100,000 tonnes of coal per year and there is the potential to improve on this. As well as household waste, it is possible to burn certain other types of scrap industrial and agricultural material such as wood.

## BURNING RUBBER

One form of dry waste which deserves special mention is scrap tyres. Currently we in the UK dispose of old tyres with an energy content equivalent to 200,000 tonnes of coal each year. Disposal of tyres in landfill sites causes instability so they are often refused. The cost of disposal by other means is then considerable. Burning them under controlled conditions seems an attractive idea and could present a worthwhile energy source in the future.

A further idea is to burn hospital waste. Some is harmless general combustible waste but there is other material which could be dangerous if not disposed of by high-temperature incineration – the so-called *clinical waste*. Separating the two types is difficult and, in practice, the health authority will burn all of it. This involves the use of fuel to reduce it to ash – an energy-hungry process.

The Yorkshire Regional Health Authority alone currently produces 12,000 tonnes of waste per year. Nationwide, Britain produces hospital and similar waste (i.e. from homes for the elderly, etc) to the equivalent energy value of over 400,000 tonnes of coal per year. Several health authorities are now considering self-fuelled on-site incinerators with energy from the waste itself being used for central heating coupled with electricity generators.

In some schemes, the waste is used to produce bulk background power, with conventional (fossil) fuels being used to top-up the output as necessary – in particularly cold weather, for example.

## WET WASTE

The use of wet waste and landfill material to generate gas involves a process of *anaerobic digestion*. Here, bacteria decompose organic material in the absence of oxygen to produce *methane* (see Fig. 1). Methane is the same as natural gas obtained from the North Sea reserves and is an excellent fuel.

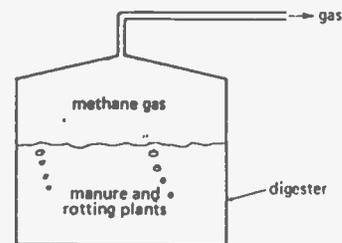


Fig. 1. Production of methane from anaerobic digestion.

On a small, local scale – for example on a farm – the process may be carried out in a large vessel called a *digester*. Any animal or plant waste (such as chicken-droppings), sewage and certain other material may be treated by simply allowing it to decay in the digester and piping off the gas. This process can yield approximately 60 per cent methane (but usually nearer 50 per cent) the rest being *carbon dioxide*. This gas mixture may be used for heating or cooking. There may be a future for large-scale digesters for bulk treatment of household waste to produce electricity.

## LANDFILL GAS

The UK has been using landfill gas (LFG) since the early 1980's and, together with the United States, is a leader in this technology. There are currently more than 33 large-scale landfill gas extraction schemes in operation in the UK. Burning the gas and using the heat to generate electricity produces some 18MW of power with a total coal saving of 150,000 tonnes per year.

So attractive is this method that it is thought that by the end of the century, the energy contribution from LFG could be equivalent to one million tonnes of coal per year. Apart from the benefits to be gained by using less fossil fuel, it also means that the gas is removed from the site instead of being left to find its way into the atmosphere. This could cause offensive odour blown by the wind over a wide area. Also, since methane is a very flammable gas, there is a danger of explosion if it is allowed to build up in confined spaces.

The rate at which landfill material produces gas depends on several factors such as the exact nature of the waste, the temperature and its moisture content. Typically, on a new site, production begins within a few months of deposition then builds up and maintains a useful output for 15 years or more.



*A digester producing methane on a farm.  
Crown copyright. Reproduced by permission of the Controller of HMSO.*

One tonne of refuse may be expected to yield 135 cubic metres of gas with present technology. However, the theoretical yield is much greater and will, no doubt, increase as the technique for extraction improves. At the moment, the gas is removed by a series of wells drilled into the waste and connected to a pumping system.

Landfill gas may be used simply for local heating or to fuel some industrial process close to the site – to fire brick kilns, for example. Alternatively, the methane may be used to power a *gas turbine* and generator to make electricity. A gas turbine is like a jet aircraft engine – the fuel is burned and the hot gasses produced turn the turbine blades.

In a jet aircraft, *thrust* is produced but here the power is taken from the rotating turbine shaft to turn the generator. The Centrax gas turbine at Meriden fuelled by LFG has a rated output of almost 4MW. The electricity is then sold to the Midland Electricity Board under contract.

It is thought that by the end of the century, the electrical power generated by landfill gas could exceed 175MW (the output of a small coal-fired power station). Research is being funded by the Department of Energy with a view to exploiting this source of power at more sites. Research includes the study of the microbiological process by which the methane is produced. This is because, at the moment, they are not particularly well understood and such knowledge could help to improve the yield.

## **BIOMASS- ENERGY FOREST**

Growing crops for fuel rather than food or manufacture is called *biomass*. In Part 1 of this series (August, 1992 issue), we saw how the Brazilians have turned this idea to good use by making alcohol from sugar. This may then be used as a petrol substitute and used as a fuel for cars. This relieves the pressure on oil from which petrol is obtained. Up to five per cent of alcohol may be blended with petrol with no detrimental effect but more may be used providing certain modifications are made to the engine.

It is interesting to note the large area of land in the UK which over-produces food – currently, one million hectares approximately (almost 2.5 million acres). This land is therefore available for other purposes such as for the production of biofuels. Trees may be grown simply for their fuel value and the Department of Energy is investigating various *energy forestry* schemes in the UK –

in particular, in Northern Ireland, Scotland and southern England.

Wood can be used as an industrial fuel and is a renewable energy source so long as the trees are grown at least at the same rate as they are being burned. It is anticipated that by the end of the century, an equivalent saving of one million tonnes of coal will be made using existing sources alone.

By growing wood specifically for the purpose, the saving could be equivalent to at least eight million tonnes of coal per year. Most of this would be used directly for heating purposes simply as a substitute for coal or oil. However, it is also possible to burn it in a more-or-less conventional power station to generate electricity.

Using wood as a fuel has the advantage that sulphur dioxide emission and hence acid rain is much less of a problem than with coal. A further advantage is that much of the technology needed for energy forestry is familiar, well-developed and relatively inexpensive. This, in practice, means that the method is a very attractive one.

## **GRASPING AT STRAWS**

At present, some eight million tonnes of straw are either burnt or ploughed back (incorporated) into the land each year. Straw does have some industrial applications – in the manufacture of fibreboard, for example. It also has some uses on the farm. However, an enormous surplus remains for which there is no particular use. Up to the present time, much of this excess straw has been simply burned in the field. However, this practice was prohibited after the 1992 harvest.

One way of using it is to burn the straw as a fuel under controlled conditions. Some 200,000 tonnes of straw are already used in this way and this is expected to rise to 800,000 tonnes by the end of the century. Unfortunately, straw can cause problems with dust. Also, it is an abrasive material which tends to wear the handling machinery.

Straw must be baled and stored (because it needs to be kept dry). Unfortunately, it has a low density – that is, it has a large volume for a given weight – so storage is expensive. Straw is also expensive to transport again, because of its low density. It may therefore be uneconomic to take it to a distant incinerator. A good way of using it is to burn the straw on the farm where it is produced using whole-bale burners to produce hot water.

Much of the heat may be used locally for domestic heating and for certain other jobs – grain drying and greenhouse heating, for example. Any excess hot water may be sold to local houses, factories, schools, etc. for central heating. The hot water could be sent the relatively short distances involved using feed pipes well-insulated to minimise heat loss.

Woburn Abbey already has a straw-burning scheme in operation which supplies heat for the entire historic home. The aim was to save £20,000 per year on oil, gas and electricity charges. The cost of the straw-burning installation was £150,000 but three boilers needed to be replaced anyway at a cost of £60,000. Also, it was found that further areas of heating could be added to the system.

The straw is handled in 1.2 x 1.2m round bales and delivered to the boiler twice per day from the nearby farm. A straw grater teases the straw out of the bale whereupon it is fed into the boiler. Straw burning is notoriously unpredictable since the material is not one definite substance but varies in composition. In practice, this means that although the plant is under automatic control, a boilerman is needed to solve minor problems as they arise.

## **COMPACT STRAW**

There is some research being carried out with a view to compacting straw into high-density blocks or pellets. This could make transportation to large straw-burning power stations much cheaper. The use of straw as a fuel promises to make a significant contribution to our total energy requirements in the future and the methods will become even more attractive as fuel prices in general rise.

As well as being prominent in the field of wind energy (as we saw in Part 3 of this series – October, 1992 issue) Denmark also leads the way in straw-burning technology. A large number of projects have been set up there over the past few years to heat water and pipe it to private houses and public buildings for use in central heating.

Schemes are also in operation to generate electricity using special furnaces and conventional boilers and turbogenerators. One particular 30MW plant burns 25,000 tonnes of straw per year. Yet other power stations use straw enhanced by coal.

It is important to note that Biomass techniques are *carbon dioxide neutral* – that is, they do not contribute to the greenhouse effect because the carbon dioxide released on burning the crop is the same as the amount absorbed from the atmosphere to perform photosynthesis.

## **GEO THERMAL ENERGY**

Geothermal energy – often called the energy of *hot rocks* – is heat derived from the interior of the earth. Strictly speaking, geothermal energy is not a form of renewable energy since the heat of the earth is a finite resource – that is, the rocks gradually cool down as heat is extracted from them. However, the reserve of energy is so great that it may be regarded as renewable.

Some estimates put the equivalent heat content of all known hot rocks in the world as high as 50,000,000,000,000,000 tonnes of coal! This could, technically, provide the world with all its energy requirements for several million years.

It seems that there are two reasons why hot rocks exist below the earth's surface. Most of the heat probably derives from the heat of the Earth's core. Also, some of it

may come about from the decay of natural radioactive material such as uranium.

Certain forms of geothermal energy are familiar and include volcanoes, hot springs and geysers. It is thought that the energy of volcanoes could not be exploited by present technology – not economically at any rate. However, geysers and hot springs can yield a significant amount of energy.

Geysers work intermittently as cold water flowing deep in the earth's crust becomes suddenly heated by the hot material around it. The water boils violently and bursts out through a crack in the ground. The gushing water and steam may then reach a height of up to 70m. The *Geysir* in Iceland (from which all others are named) sometimes reaches 55m.

Geysers and hot springs are chiefly found in volcanic regions such as Iceland. Here, the water may be used direct for central heating or for warming swimming pools, etc. The pavements of Reykjavic are heated by water derived from hot springs and this keeps them clear of ice.

## AQUIFIERS

Aquifers are deep underground rocks which heat water naturally. In Britain, aquifers occur some 2km below the surface. These usually provide water at between 60 and 80 degrees (i.e. hot but not boiling). The hot springs at Bath emerging at about 49 degrees C are a good example of such hot water leaking to the surface.

Another example is the Southampton City borehole project which is a joint private sector/local authority district heating scheme. However, there is little prospect of really large-scale development of aquifers in the UK and the research programme funded by the Department of Energy has been curtailed. Even so, the Department is continuing to fund the monitoring of the Southampton bore-hole project.

## HDR TECHNOLOGY

Although Britain is not ideally served, we do have some sites especially in Cornwall where hot granite rocks exist deep below the surface and which may be used as a source of energy. It is thought that a significant proportion of Britain's electricity could be generated this way.

*A view of Rosemanowes HRD site showing the well-head structures and the heat exchanger in the background.* Crown copyright. Reproduced by permission of the Controller of HMSO.



To extract the energy from the hot dry rocks (HDR technology), water is sent down a borehole or natural fissure increased in size by hydraulic fracture – the *injection well*. The water reaches the rocks, boils, and the steam is allowed to return to the surface through the *production well*. This may then be used for simple heating in domestic or industrial processes or used to operate conventional turbogenerators and produce electricity.

To boil water needs a temperature of at least 100 degrees C but, in practice, 200 degrees C is needed to superheat the steam and produce electricity efficiently. Drilling the boreholes is quite difficult due to the depth required – some 6km – and it may involve drilling through some very hard rocks on the way.

Less geologically stable areas than Britain are better suited to exploiting the energy of hot rocks. As well as Iceland already mentioned, countries with better resources include parts of Italy, New Zealand, Japan, Mexico and the United States. These countries all have large-scale geothermal energy programmes already in operation.

## THE NUCLEAR ROAD

Nuclear *fission* – that is, the energy technology widely used at present is, like geothermal energy, not strictly a renewable energy source. This is because the process relies on *uranium* and this will eventually run out like the fossil fuels. However, the time scale is much longer.

It is thought that there are sufficient uranium reserves in the world – chiefly in the USA, Canada, Australia, Africa and France – to provide electricity for 1,000 years or more. The fission process produces no carbon dioxide so does not directly contribute to the greenhouse effect and possible global warming. However, there is some *indirect* use of fossil fuels.

Some fossil fuel is needed to build the nuclear plant including the power stations themselves, the re-processing works and storage sites. Some is needed for waste disposal. Yet more fossil fuel energy will be needed for decommissioning purposes when the useful life of the power station is over.

On the other hand, there are some positive reasons for using uranium rather than the fossil fuels apart from the fact that the latter will eventually run out. The fossil fuels have great importance in their own right as chemicals for the manufacturing industry whereas uranium is a fairly useless material apart from its use in nuclear reactors. Also, uranium is a very concentrated source of energy compared with the others – one tonne providing the same amount of energy as 20,000 tonnes of coal.

Several countries have gone down the nuclear road to generate very significant amounts of their total energy needs this way. Table 1 shows, in descending order, the percentage of total electricity requirement generated using nuclear energy in several countries. The overall figure for Europe is about 30 per cent.

Table 1

Country	% of nuclear energy
France	75
Belgium	61
Sweden	45
Switzerland	42
Spain	38
Finland	35
Germany	34
Japan	28
UK	22
USA	19
Canada	16

Nuclear power production only differs from conventional electricity generation in the source of heat used. This is provided by the fission process carried out in the core of the nuclear reactor (see diagram opposite). There are several different types of reactor used in Britain.

The original design is called a *Magnox* reactor – the first practical one in the world was Calder Hall at Sellafield which was put on stream in 1956 and is generating to this day. The *Advanced Gas-Cooled Reactor* (ARG) followed. The most widely-used type of reactor in the world is in *Pressurised Water Reactor* (PWR) – see opposite. One of these is currently under construction at Sizewell (Sizewell B) in Suffolk and will operate alongside the existing (Sizewell A) *Magnox* power station.

Whatever type of reactor is used, they all behave in the same basic way – nuclei of uranium atoms are split by bombardment with sub-atomic particles called *neutrons*. This, in turn, releases more neutrons which can cause more splitting of uranium nuclei and a *chain reaction* results (see Fig. 2).

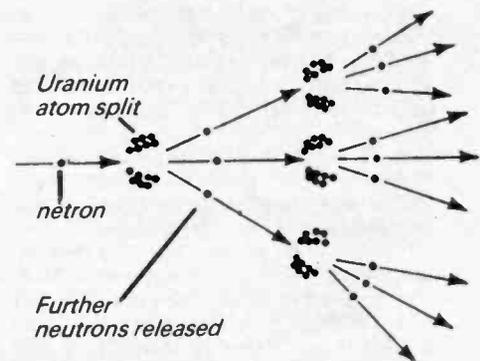
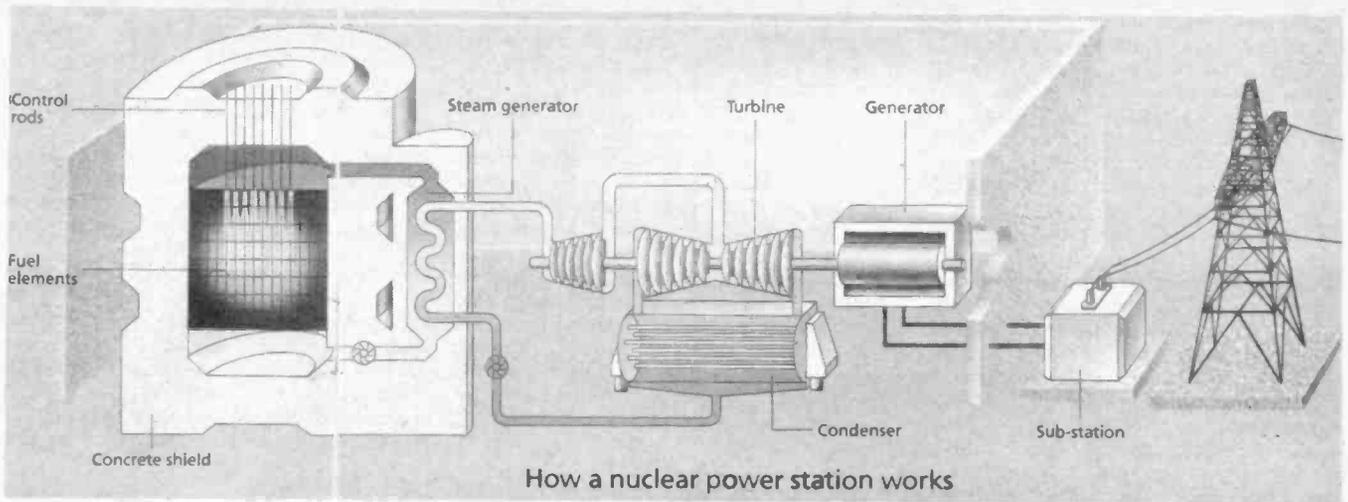


Fig. 2. The fission process.



This process produces a colossal amount of heat which is transferred to and carried away by a coolant such as water or carbon dioxide gas. This coolant then transfers its heat to water in a heat exchanger and returns for re-heating. The steam produced operates a turbogenerator and hence generates electricity.

Clearly, there must be a way of preventing excessive splitting of uranium nuclei with the ever increasing number of neutrons released. This would cause the system to go out of control. Excess neutrons are therefore absorbed by *control rods* to leave just enough to maintain a chain reaction and provide the amount of heat needed. Control systems ensure that the reactor shuts down completely in the event of any fault developing such as overheating.

## WASTE

The nuclear fuel in the reactor gradually becomes less efficient and has to be replaced. In fact, only three per cent of the weight of this material is "waste" – the rest is unused uranium and approximately one per cent of a new material formed in the process called *plutonium* – a decidedly nasty substance both intensely radioactive and highly toxic.

The fuel may be re-processed to remove at least 96 per cent of the uranium and the plutonium. This leaves three per cent

of the original material which constitutes highly radioactive liquid waste which has to be stored permanently. The volume of the high level waste produced by the British nuclear industry over the last 30 years or so is about the size of two semi-detached houses. Although this is a very small volume compared with the amount of waste generated by other industries, this material is very dangerous and must be stored carefully.

At present, the storage method used is to keep it in double-walled stainless steel tanks. A more recently-developed method is to reduce it to powder and then melt it permanently into blocks of glass – *vitrification*. This takes up only one-third of the volume of liquid waste and is much easier to store.

The process produces much larger quantities of intermediate-level and low-level waste – liquid and solid. The intermediate-level material is much less dangerous than the high-level variety and may be stored by encasing it in cement inside steel drums. The low-level material is only slightly radioactive and, after some treatment to make it still weaker, the liquid is disposed of in the sea and the solid stored in enormous vaults.

## FAST BREEDER

The plutonium obtained from spent uranium fuel rods is itself used as a fuel in a certain type of reactor called a *fast breeder reactor*. This technique increases the energy

obtained from the original uranium by a factor of 60 approximately.

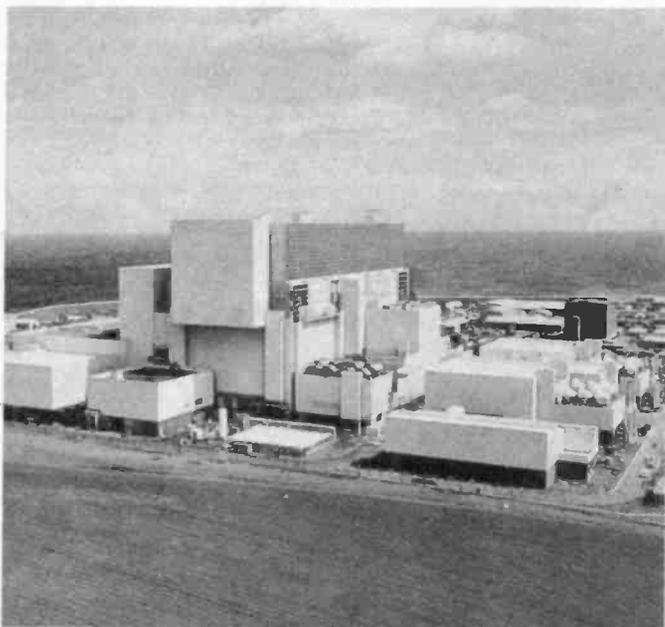
There is public concern about certain safety aspects of nuclear power. One fear – that of danger from direct radiation is unfounded – 87 per cent of the average dose of radiation for a person living in Britain comes from natural sources. Medical uses (such as X-rays) contribute about 12 per cent. Even flying in aircraft increases the radiation dose due to cosmic rays entering the earth's atmosphere from space – see Fig. 3. Less than 0.1 per cent comes from the nuclear industry.

Other concerns come from the possibility of an accident resulting in the discharge of radioactive material into the atmosphere such as from the catastrophic event at Chernobyl in the former Soviet Union in 1986. However, this type of reactor was known to have serious design faults and would not be allowed in Britain. Safety systems were turned off by the staff in a way which could not happen here.

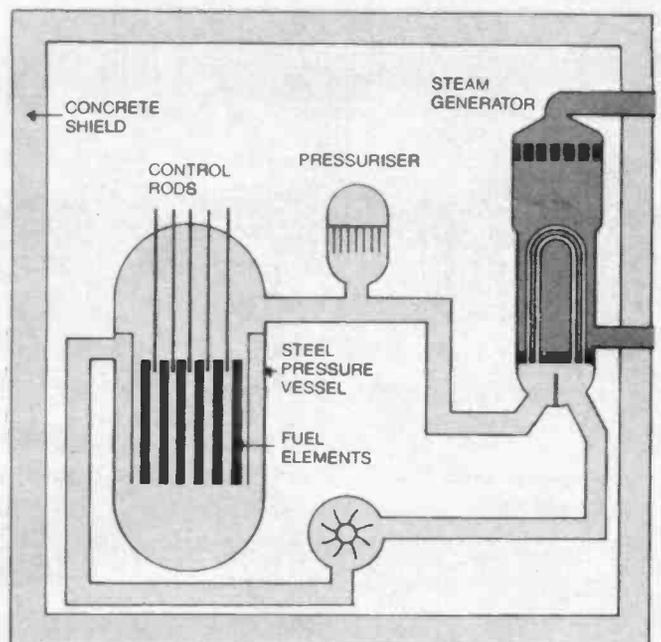
The other concern is that of nuclear waste disposal and the possibility of it entering the environment – perhaps by entering water supplies. British Nuclear Fuels, who manage the waste, go to great lengths to ensure the safety of their waste storage and disposal programme.

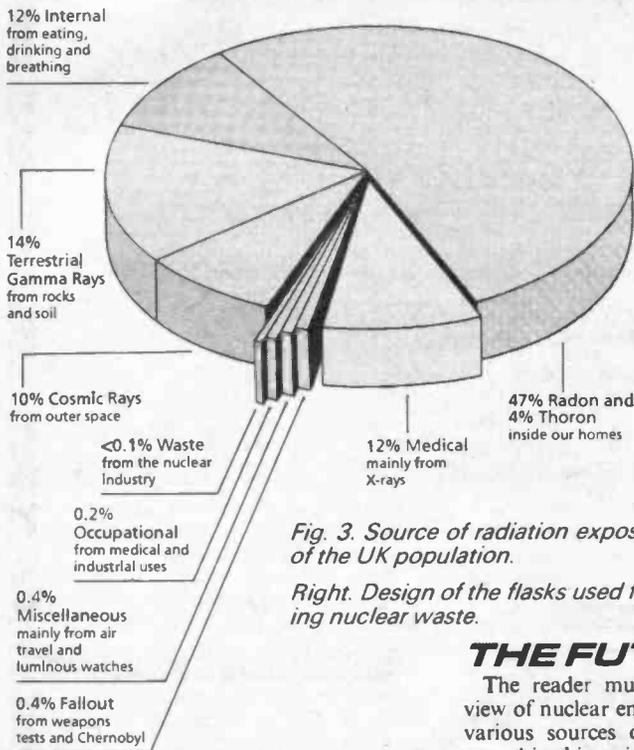
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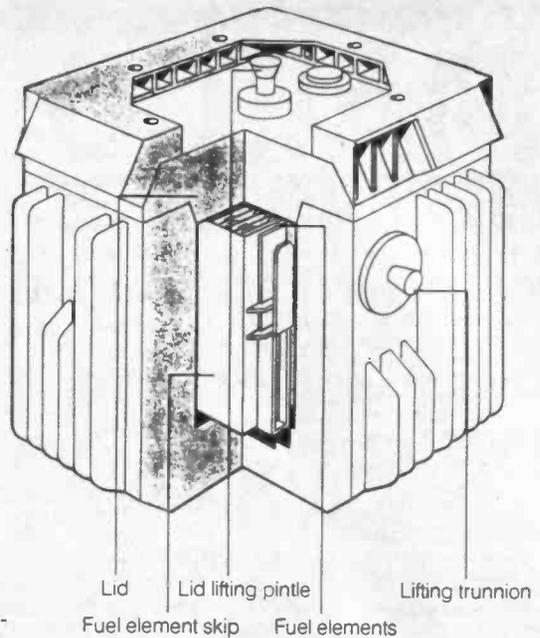




Source: National Radiological Protection Board

Fig. 3. Source of radiation exposure of the UK population.

Right. Design of the flasks used for transporting nuclear waste.



It seems that man will have the energy he requires independently of fossil fuels into the foreseeable future. We will not have to go back to the days when he had to do all the work for himself or use animal power or slaves.

Many familiar techniques which have been developed for the old technology will continue. Perhaps the prophets of doom and gloom who foresaw unspeakable calamity – even the destruction of the Earth itself will be proved wrong after all. With good housekeeping the future is bright. □

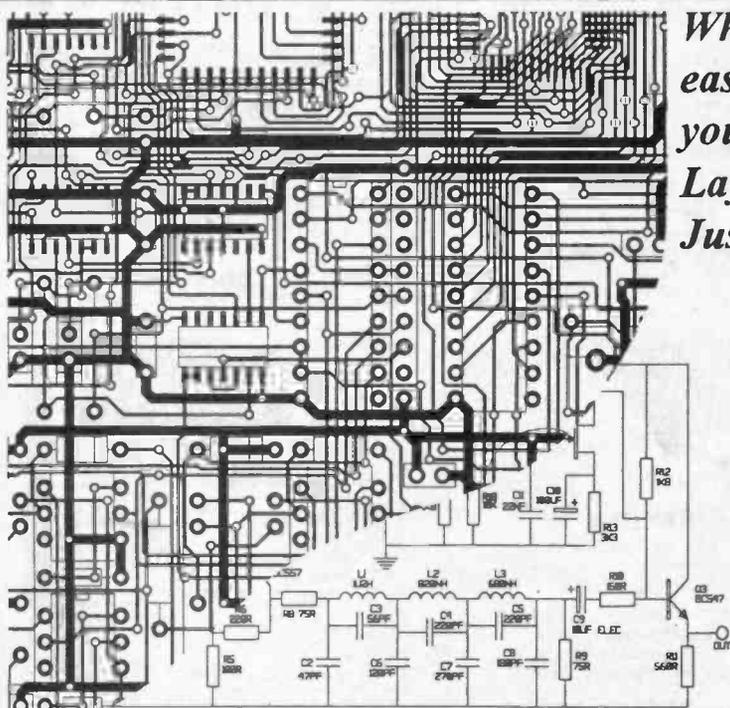
We would like to thank British Nuclear Fuels PLC for their permission to reproduce many of the photographs and diagrams used in this article.

### THE FUTURE

The reader must form his or her own view of nuclear energy balanced against the various sources of renewable energy discussed in this series. There is no doubt that nuclear power will continue to be a large-scale supplier of world electricity needs.

It does seem that several technologies will operate side by side with each making its own contribution to the total energy requirement. It may be that *nuclear fusion* will be the eventual single solution to provide our total energy needs – this topic was discussed in Part 2 of this series (Sept., 1992 issue).

it is carried in steel or steel and lead flasks each weighing up to 110 tonnes (see above). The flasks have undergone a full-scale test involving a collision with a 140 tonne train travelling at 100mph. The locomotive was destroyed but the flask, which was placed in the path of the train in its most vulnerable position, withstood the impact without fracturing.



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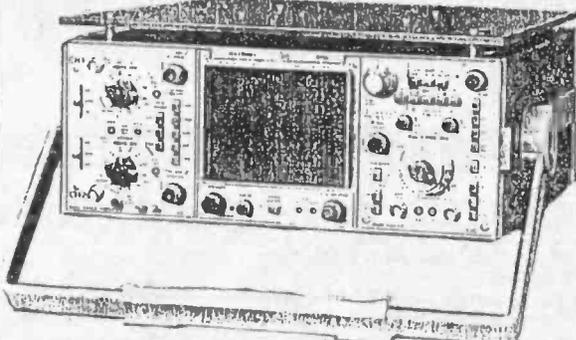
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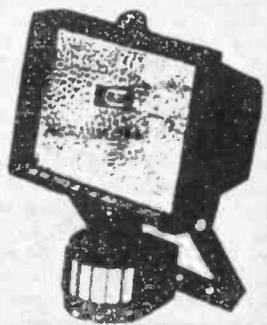
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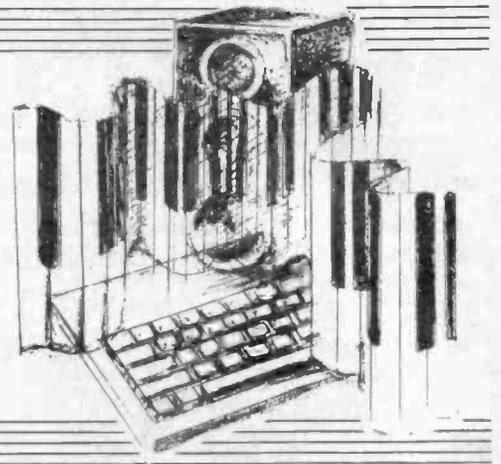
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# MIDI LEAD TESTER

**ROBERT PENFOLD**

*Now you can check-out your MIDI leads as quick as a flash with this easy-build tester.*



**M**IDI is quite a 'high-tech' means of communicating between units in an electronic music system, but it relies on ordinary plugs, sockets and leads to actually carry the signals. These will function well enough for long periods if left undisturbed. However, in practical systems there is often a lot of plugging-in and unplugging. Also, leads do tend to get kicked and tripped over.

When a MIDI system fails to work properly, unless something in the system is obviously malfunctioning, there are two likely causes. One is that something in the system is simply not set up correctly. The other is that there is a faulty MIDI cable somewhere in the system. Checking for a damaged lead is probably the best starting point.

This unit enables MIDI connecting leads to be quickly checked, and it will show up broken wires/connections or short circuits from one lead to another. In use it is much quicker and easier than using a multimeter or continuity checker. Checking leads with ordinary test equipment is easy enough provided you have four hands (two for the test leads and two for the plugs)!

It should be possible to build this tester for less than the cost of ready made units, some of which seem to be rather crude in comparison.

It can help to greatly speed-up checks on a faulty MIDI system. It can also save a lot of time if non-technical friends with MIDI equipment tend to present you with sets of MIDI leads for testing on your multimeter.

## OPERATING METHOD

On the face of it, all the tester has to do is feed five signals down the conductors of the cable, and have five indicators to show whether or not these signals make it to the other end of the cable. This method will indicate any breaks in the conductors, but it will not show any short circuits between conductors. With leads that use moulded plugs it is unusual for short circuits to occur, but with ordinary DIN plugs these can easily arise where multi-strand wires have partially come adrift and frayed.

The tester to be described here has five l.e.d. indicators, one for each pin of the 5-way DIN connectors. These are driven one at a time, so that any short circuits will

be indicated by two or more l.e.d.s coming on simultaneously. A break in the cable will, of course, be indicated by one or more l.e.d.s failing to switch on.

## MIDI

The block diagram of the MIDI Lead Tester, Fig. 1, shows how this simple scheme of things is implemented in practice. The sequential outputs are provided by a one of five decoder, which is driven by a low frequency clock oscillator.

The decoder has five outputs ("0" to "4") which go high, in sequence, for one clock cycle per output. The decoder cycles continuously, with output "0" going high for one clock cycle after output "4" has had its turn.

Each clock cycle lasts about one fifth of a second, so that things progress at a slow enough rate for the testing of each connector to be clearly seen. On the other hand, the decoder is cycled at a fast enough rate to permit leads to be thoroughly checked in just a few seconds.

The suspect MIDI lead is plugged into the tester, and it should carry the signals to a five l.e.d. display. A faulty lead will scramble the signals in some way, which will be clearly indicated by the l.e.d. display.

In order to use the unit properly you need to know a little about MIDI leads, and the methods of interconnection that they use. This is something that is fully explained at the end of this article.

## CIRCUIT OPERATION

The full circuit diagram for the MIDI Lead Tester is shown in Fig. 2. The clock oscillator is based on a low power 555

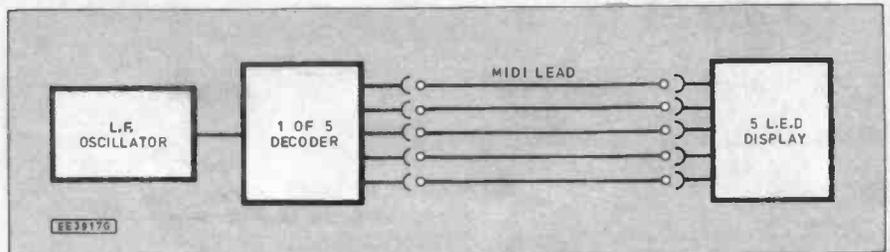
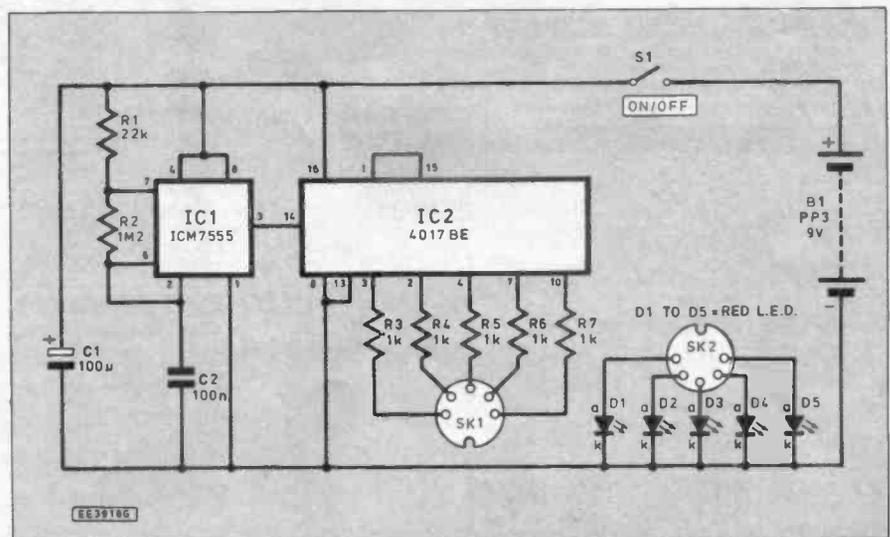
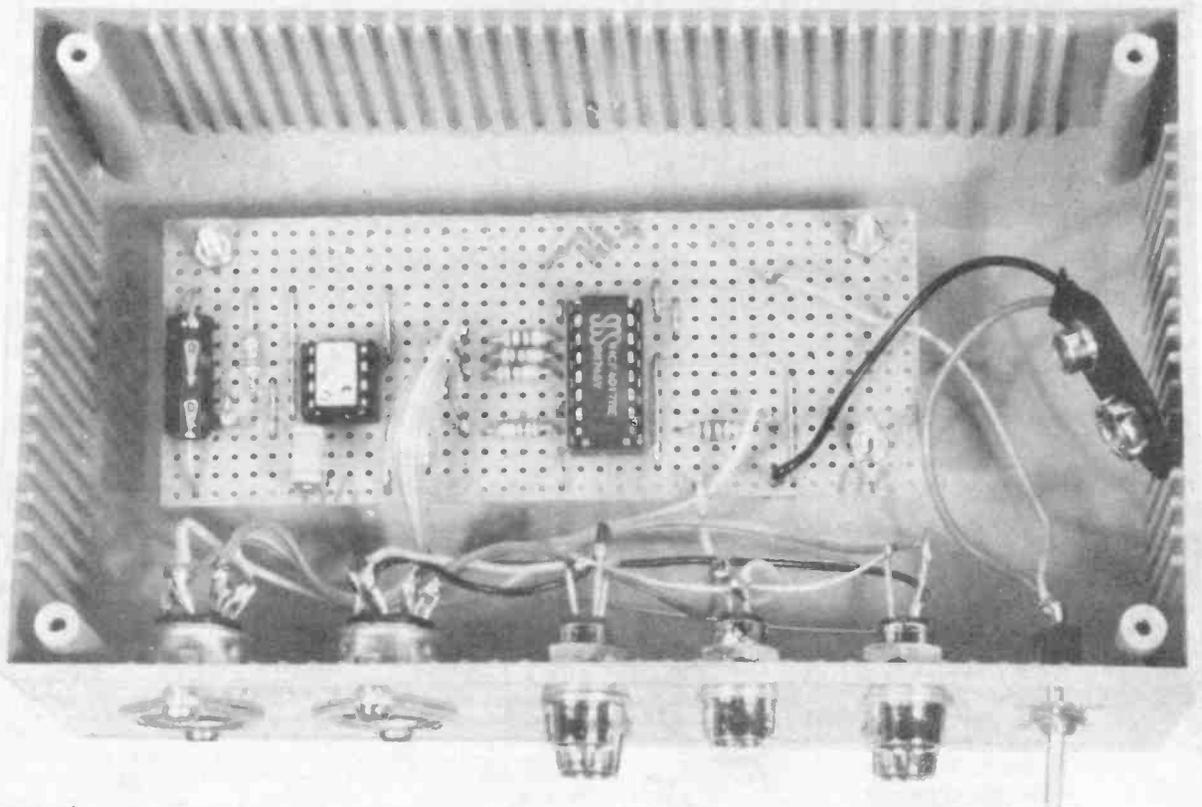


Fig. 1. The MIDI Lead Tester block diagram. To avoid misleading results, the connectors are tested one at a time.

Fig. 2. Complete circuit diagram for the MIDI Lead Tester.







Layout of component and wiring inside the case. The battery sits on the bottom of the case to the right of the circuit board.

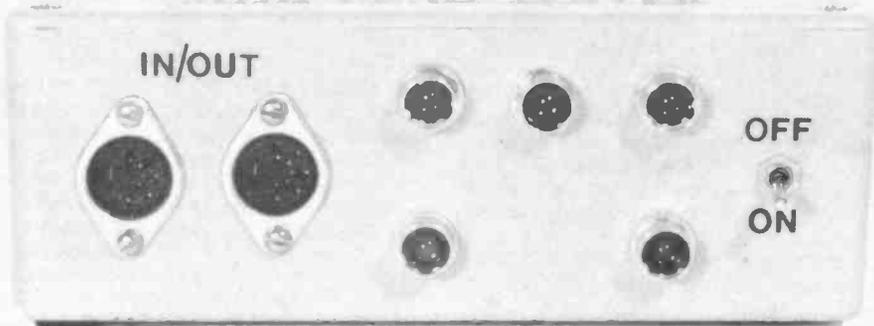
edges, but these are easily filed flat using a small file. Stripboard is not particularly tough, and it must therefore be worked with reasonable care.

Next drill the three mounting holes (refer to Fig. 3 for details of the board). Use 6BA or M3 fixings which require 3.3 millimetre diameter mounting holes. If you use plastic stand-offs, the size of the mounting holes must be selected to suit the particular stand-offs used. The 25 breaks in the copper strips on the underside of the board should then be made. A special tool for cutting the strips is available, but a hand-held drill bit of about 5 millimetres in diameter will do the job quite well.

The board is now ready for the components and link wires to be fitted. IC2 is a CMOS device which consequently requires the usual anti-static handling precautions. The most important of these is to fit IC2 onto the board via a holder instead of soldering it directly onto the board.

Do not remove IC2 from its anti-static packing and plug it into place until the unit is, in all other respects, finished. Try to handle IC2 as little as possible once it has been removed from its packing.

The completed MIDI Lead Tester showing layout of components on the front panel.



The timer i.e., IC1 does not require any special handling precautions, but the use of a holder for this component is still recommended. Note that IC1 has the opposite orientation to IC2.

Fitting the other components and the link wires on the board is quite straightforward. However, be careful to fit the electrolytic capacitor C1 the right way round. Capacitor C2 must be a miniature printed circuit mounting (7.5mm pitch) type if it is to fit into this layout easily.

The link wires can be made from 22s.w.g. tinned copper wire, or the trimmings from the resistor leadout wires will probably be suitable. Fit single-sided solder pins to the board at the points where connections to off-board components will be made. Tin these pins with a generous amount of solder. It should then be easy to make good connections to them.

### CASE

The prototype is housed in a medium sized plastic box, but this is somewhat larger than is really necessary. The unit should fit into quite a small case without too much difficulty, provided you can find

enough front panel space to accommodate the on/off switch, the five l.e.d.s, and the two sockets. From the electronics point of view the layout of the unit is not critical, but try to lay things out sensibly so that the unit is easy to use.

There is a fair amount of point-to-point wiring, and this is shown in Fig. 4. This diagram should be used in conjunction with the circuit board layout, Fig. 3.

For example, point "A" in Fig. 3 connects to point "A" in Fig. 4. The cathode (k) terminal of a l.e.d. is normally indicated by this lead being shorter than the anode (a) lead. There are a few exceptions though, and it might be worthwhile checking this point in the retailer's catalogue before wiring up the l.e.d.s.

### IN USE

With the Midi Lead Tester unit switched on and no MIDI lead connected, none of the l.e.d.s should come on. If a piece of wire is used to temporarily connect a pin of one socket to the equivalent pin of the other socket, the corresponding panel l.e.d. should flash about once per second. It is a good idea to check all five pairs of pins in this way to check that the unit is fully operational.

A standard MIDI lead has the two pin-2's (the middle pins) interconnected by the screen of the cable. Pins 4 and 5 on one plug are connected to pins 4 and 5 of the other plug via two inner conductors of the cable. Pins 4 and 5 are the ones on either side of pin 2 incidentally. Pins 2 and 3, which are the ones at opposite ends of the arc of pins, should not be connected.

This means that with a standard MIDI lead connected to the unit, only the middle three l.e.d.s of the display should operate. In practice many MIDI leads seem to have all five pins interconnected. One reason for this is that many of the leads which are sold as MIDI types are actually audio cables.

Note that some five-way DIN audio cables have the plugs cross-connected, and are therefore unsuitable for MIDI use.

These can easily be detected using this tester, as with one of these leads the display l.e.d.s will "move" in the wrong direction.

Some MIDI leads supplied with equipment from respectable manufacturers seem to have pins 2 and 3 interconnected. Presumably these leads are purpose made MIDI types, but have pins 2 and 3 connected in case they are utilized in some later MIDI implementation.

Anyway, if the middle three l.e.d.s operate properly, or all five l.e.d.s operate in the correct order, the lead should be usable. It is only if one or more of the middle three l.e.d.s fails to operate, or more than one l.e.d. at a time switches on, that the lead is faulty.

## BE FLEXIBLE

If the middle l.e.d. fails to operate, the lead will still carry MIDI information correctly. However, the screening may not be earthed properly, and strong radio interference might be radiated by the cable.

When checking any cables it is always a good idea to flex the leads slightly, and to gently pull the cable away from each plug. A fair percentage of cable problems involve intermittent faults. You can find that a lead passes a test, but fails to operate when it is put back into the system. Flexing and gently tugging the lead will encourage any intermittent problem to occur while the lead is being checked.

The popular Atari ST computer has a non-standard implementation of the MIDI OUT and THRU sockets. These should be separate sockets with the signals carried by pins 4 and 5. Instead, these two functions are combined in a single socket with pins 4 and 5 carrying the OUT signal. Pins 2 and 3 carry the THRU signal.

An ST combined THRU and OUT lead is therefore a twin cable which has a single plug at the ST end. This couples through to the OUT plug in the usual way. However, pins 2 and 3 couple to pins 4 and 5 of the THRU plug. Obviously this tester can check these leads, one section at a time, and can show which is the OUT plug and which is the THRU plug. □

# READOUT

## OFFENSIVE MNEMONIC

Dear Ed.,

One of my constituents has drawn my attention to the *Teach-In '93* article in the November edition. The article deals with a resistor colour code and suggests a mnemonic to remember the colours.

My constituent found this mnemonic offensive and so do I. I think that, in this day and age, it is totally inappropriate that such mnemonics should be suggested particularly in magazines that will find themselves in schools and colleges.

As a member of the Education Committee of our Regional Council I will be taking this matter further if I do not receive a satisfactory explanation from you. I think it shows a distinct lack of imagination that a mnemonic could not be devised that was unoffensive.

*Councillor Des Loughney*  
Lothian Regional Council

*I am sorry that we have offended you with the mnemonic. Your point is taken and on reflection I agree that we should have come up with something different. The published version has been in use in education and the electronics industry for more than 30 years and I must admit we published it without due consideration.*

*We would like to offer a years subscription to EPE plus a binder (worth £25.95 in total) for the best alternative colour code mnemonic sent in by a reader before December 31st 1992. The winner will be published in EPE.*

## P.C.B. OR NOT P.C.B.

Recently you published a series of articles on printed circuit boards and it was this that prompted me to write in.

I have been an electronics hobbyist for many years. For a long time, I believed that the p.c.b. was the "best" way of making circuits. Many, if not most, of your readers believe the same. Well, I now think this idea is seriously misguided. Let me explain.

When you see just how widespread the printed circuit is it is easy to imagine that there is some natural law of electronics which demands its use, but this isn't so. In fact they are a recent development, not appearing until the 1940's, and despite their seductive good looks they have a number of

serious disadvantages. They are difficult to design, time-consuming to make and impossible to repair (ever *tried* removing a component from a board)?

If all this is so, then why are they used so much? Because they have one big advantage: reliability. For a manufacturer this feature of reliability completely outweighs the disadvantages and this explains how p.c.b.s have taken over.

However, it is important to see this in perspective. The concept of statistical reliability has no meaning for the hobbyist who makes a single copy of a circuit and can fix it if it fails. For him, the disadvantages of the p.c. method, outlined above, can make it a poor choice.

I take the view that for the hobbyist it is a better bet to use matrix board. The recent introduction of wiring pens has made this method of construction rather easy. Additionally, by using the new insulated wires it is possible to implement complex digital circuits where the wiring looks like spaghetti and where using a p.c.b. is out of the question for an amateur.

There is nothing inherently right or wrong about the p.c., only its application. I feel many are using it inappropriately. My advice is just to think before you etch.

*B. Clothier*  
Wallington

*We agree with much of what you say, however the biggest advantage of the p.c.b. to the hobbyist, when made from a published design, is the virtual elimination of wiring errors. In our opinion this generally outweighs the cost and/or production time etc. - Ed.*

## MICROSOFT OK?

Dear Ed.,

I am writing about the item by Barry Fox in *Everyday Electronics* September 1992. Although I have no connection with Microsoft, or to Bill Gates, I feel that Barry has been unfair to them, and showed himself to have a poor grasp of the history of the PC.

Surely the present growth in the marketing of DIY PCs - where you buy components, a motherboard, hard disk, floppy drives, VGA card etc. - shows the (almost) true compatibility standard set by the IBM PC. Unfortunately, the standard is now old technology, with local bus and EISA techniques trying to increase speed and performance while still maintaining compatibility with the early machines.

Bill Gates business acumen aside, Microsoft did not 'condemn some 70 million people (to using MS-DOS)'. When the original IBM PC was made it

was designed to run CP/M-80 - a now seriously outdated operating system. MS-DOS was released as another option which the market chose. It is a consequence of the market's decision *at the time* that 70 million people now use DOS.

The limit of 8-character file names is again a hangover from CP/M. By the way - I work in the real world, with many other people, all of whom use text files, and no one I have asked considers the fact that the operating system places limits on filenames remotely daft.

As for your scornful opinion of Windows 1.0, any software package evolves from the original, through (often) several stages towards the present "best-ever" version. So what if windows is now on 3.1, Wordstar is now in version 71 and DR-DOS a non-Microsoft operating system is on version 6, 1 ahead of MS-DOS. When DOS was originally released, PCs were mainly used by people who knew what they were doing, or at least, knew the dangers involved in delete or format situations. Of course safe format and unerase commands should have been implemented earlier. I for one would have used them many times! However, the extra steps involved in the execution of such commands, or rather in the delete and format commands have in the past added extra time to the execution, and hence were 'undesirable' until of course you needed the files back!

Half way through your column we come to what could be the reason for your violent dislike of Bill Gates and Microsoft - he wouldn't give you an interview. I'd call that a childish attitude.

I use Windows 3.0 at work. I don't like it, but I use it. Although I can see your point about taking quite a while to install Windows 3 itself. Installing applications afterwards is child's play, and while I would also query the times taken to master windows software, certainly many people could learn enough in two and a half hours to use any windows software.

Most windows software packages these days do actually have the necessary PIF files to enable them to run properly.

If I was the ASA I would not take your side, especially since many PC's these days come with windows pre-loaded.

In short, your column was not what I would have expected from a professional electronics and computer journalist. It was inaccurate, ill-informed and biased.

*James Ogden*

Hemel Hempstead

*Everyone is entitled to their opinion - the ASA upheld Barry's complaint - Ed.*



them. Even for someone as clumsy as me, this is fairly easy because the sort of coils I mean have four legs, making convenient anchor points for the wire. I have dozens of formers from one of my "bargain pack" or surplus purchases. Alternatively a less neat transformer can be wound on a 13mm piece of 6mm diameter ferrite rod.

Together with the 22nF capacitor (C2) the choke (L1) forms a filter to minimise r.f. noise. It can be made by winding about 50 turns of 20 s.w.g. enamelled wire onto a

20mm length of 10mm diameter ferrite rod.

The triac should suit your needs. I like a lot of illumination when I'm groping around in our loft, and my string of large bulbs now numbers seven. I therefore used an 8A triac, and chose one with an isolated tab to avoid worrying too much about insulation.

From a tin can that probably started life looking after baked beans, a strip was cut about 150mm long and 20mm wide, doubled up on itself, drilled with a neat

hole near one end, then folded into a U-shape and used as a small heatsink for the triac. (Just in case you are wrong about your triac's tab being isolated, if you are curious about how warm the heatsink gets, unplug the mains before touching it!)

The circuit was cased in a tobacco tin, suitably earthed of course and with a 5A mains fuse in the supply. By the way, please don't build this circuit if you are unfamiliar with mains wiring - we would hate to lose any readers.

# SHOP TALK

with David Barrington

## Mini Lab (Teach-In '93)

Before we tackle the components for this month's *Mini Lab* project, we have some welcome news from two of our advertisers.

The first concerns the specified Verobloc. We were surprised to learn that Maplin no longer carry this item. However, thanks to the good offices of Greenweld Electronics, who do carry stocks, they are making a special offer to readers.

If any reader orders the two "breadboards" from them (£13.70 plus £2.75 p&p), Greenweld will give them a *Free* low voltage 12V to 24V a.c./d.c. 6W soldering iron whilst stocks last (see their advertisement). *You must request the iron when ordering.* Having seen the iron, its tempting to add a car cigar lighter plug to the leads and keep it in the car tool box for emergency breakdown repairs.

We have received a few comments from readers that ordering the correct components from all the various sources mentioned last month "adds considerably to the cost when each postal charge is taken into account." This point has been taken up and we are pleased to report that Magenta Electronics have put together a selection of kits, including a complete package, at prices to suit differing requirements (see their advertisement). The complete kit does not, of course, include the *Micro Lab* demonstration package to be described towards the end of the series.

Back to this month's L.E.D. Voltmeter project. Apart from the toggle switch (covered in last month's roundup) and the

10-segment l.e.d. bargraph module, all components should be fairly readily available.

The bargraph module is available from Maplin, code BY65V; Electromail, code 588-027 and Farnell, code MV57164. You can, of course, use individual rectangular l.e.d.s.

The Mini Lab p.c.b. is only available through the EPE PCB Service see page 835. Please remember that in consideration of the effort by both Authors which went into the Mini Lab design, the copper track layout remains Copyright Dytronics 1992.

## Universal Infra-red Remote Control

If any constructors have difficulties finding a local source for the transmitter, infra-red amplifier and receiver chips, they are currently listed by Electromail (☎ 0536 204555), codes 658-491, 301-527 and 658-508 respectively.

If you wish to take advantage of the 32-way key switching possibility you will have to use two 16-way membrane keypads. These are available from Maplin, code JYO4E. They also supplied the small keypad shown on the model, JMO9K.

The printed circuit boards are available as a pair from our PCB Service, code 811T/R.

## Combination Switch

Finding the combination switch i.c. for the *Combination switch* project was quite a task. The UA3730 switch i.c. is available from Maplin, code UM98G, but was very

difficult to find in their "semiconductor listing". The keypad (JMO9K) and other semiconductor devices were obtained from the same source.

The printed circuit board is available from the PCB Service, code 812 (see page 835).

## Christmas Lights Colour Spectrum

We cannot foresee any component buying problems for anyone wishing to build the *Christmas Lights Colour Spectrum* project. However, you *must* take extreme care when testing the unit as mains voltages are present.

The p.c.b. can be purchased through our PCB Service, code 813.

## Auto-Switch Add-On

The twin coil d.i.l. relay used in the *Auto-Switch Add-On* prototype was the 5V version from Electromail (☎ 0536 204555) code 351-689 (12V version 351-695). A suitable ultra-miniature switch for the small "pill" box version would be one of the "tactile click effect" ones from Maplin, JR89W.

## Midi Lead Tester

We do not expect any component buying problems to be encountered when shopping for parts for the *Midi Lead Tester*. The l.e.d.s are widely available in small panel mounting housings.

## PLEASE TAKE NOTE Battery/Mains Inverter - Daughter Board

Please note the annotations in "brackets" on the two i.c.s. in Fig. 2 of the *Battery/Mains Inverter-Daughter Board* have been transposed.

As the two i.c.s. have different pin counts, they can only be inserted on the printed circuit board in one position and should not, we hope, have caused any problems.



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# Techniques

## ACTUALLY DOING IT!

by Robert Penfold

I SUPPOSE that with any hobby there will be some tasks that you enjoy more than others. Probably in common with many electronic project builders, I find the electronics side of the hobby more absorbing and rewarding than the purely mechanical aspects.

One facet of project construction that I definitely do not like much at all is making large cutouts. If you happen to have the perfect equipment for this type of thing, then it is probably enjoyable.

Unfortunately, the perfect equipment is a power fretsaw fitted in a sort of combined stand and work-top. These tend to cost rather more than most electronics hobbyists' annual expenditure on components and tools.

### COPING WITH CUTOUTS

There are alternatives to an electric fretsaw, and the obvious one is an ordinary hand fretsaw. One of these fitted with practically any fairly fine-toothed blade will cut quite well through an aluminium or plastic panel. Whatever tool you use, cutting through steel panels is likely to be slow work. Unless you have tools that are really up to this type of thing I would strongly recommend that steel cases are avoided if any large cutouts will be needed.

A fretsaw enables cutouts of any shape to be produced, at the edge and within a panel. In the case of cutouts within a panel a small starting hole must first be drilled somewhere within the perimeter of the required cutout. The blade is then threaded through the hole and fixed into the saw.

Fretsaws are rather unwieldy due to the largeness of their frames. Coping saws have substantially smaller frames and are much easier to use, but cannot reach into the middle of large panels.

*Making a ring of holes and then joining them using a needle file is quicker than making the cutout using the file alone.*

Where a coping saw has sufficient reach to make the required cutout, it is a better choice.

With either type of saw it is important that the panel is firmly clamped in place, and that you take your time. Rushing things will almost certainly produce poor cutting accuracy and a lot of broken blades.

Even if things are taken slowly and carefully, one or two blades are likely to be broken while making a large cutout. Presumably this is why fretsaw blades are usually sold in packs of a dozen or more!

### GETTING THE NEEDLE

Miniature round files (also known as "needle" files) offer a good alternative to coping and fret saws. They do not provide quite the same degree of precision, but they are adequate for the vast majority of cutouts.

Some of these files fit into a combined handle and frame which is very much like a coping saw. These are quite fast and easy to use, but like a coping saw they cannot reach the parts of a panel that fretsaws and ordinary needle files can reach.

Ordinary needle files are basically just scaled down versions of the full-size products. Unfortunately, in most cases the handles and not just the files themselves are scaled down. This makes them difficult to hold, so that producing large cutouts with one of these files can be a bit awkward and time consuming.

Take things slowly though, or the file may slip out and smash into the panel. This will badly scratch the panel, and could well do the same to you. A needle file probably represents the cheapest means of making cutouts that is reasonably practical, but you need patience when using one of these tools.

One way of speeding things up when using a needle file for large cutouts is to drill a series of small and closely spaced holes just inside the perimeter of the required hole – see photograph. The file is then used to join up all the holes, leaving what will be a rather rough cutout.

However, the rough edges can soon be filed out to a neat finish, and there is not much point in trying to make the initial cutout at all neat or tidy. It is really just a matter of making sure that the initial cutout is kept within the perimeter line.

### JIGSAWS

These days many household tool kits are equipped with a power jigsaw. For most work in the current context these are too large and cumbersome, although I suppose they may be satisfactory for some project work.

Miniature or "precision" jigsaws are more suitable, and will permit many cutouts to be made quite quickly and easily. Miniature jigsaws do not permit intricate shapes to be cut with quite the same degree of precision as with a fretsaw, but they are adequate for the vast majority of cutouts.

If you have a miniature electric drill for making holes in printed circuit boards, there may be a precision jigsaw attachment available for this drill. An attachment of this type is substantially cheaper than a precision jigsaw, but should be adequate for cutting through thin aluminium and plastic panels.

### QUICK NIBBLE

A hand nibbler tool almost certainly represents the quickest means of producing cutouts. These tools vary somewhat in their appearance and design, but they all operate in basically the same fashion. This is a vaguely scissor-like tool which literally nibbles out little rectangles of material.

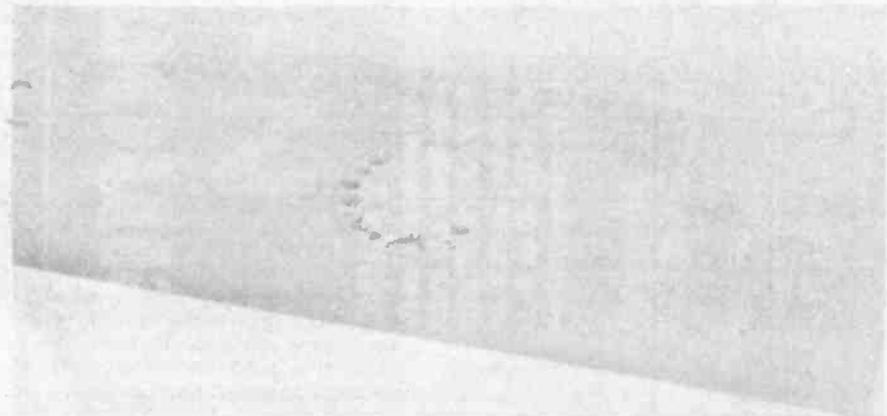
Results with aluminium are good using one of the better quality nibblers, but I could not recommend them for use with steel. They are usable with some plastics, but will simply shatter *hard* plastics.

There is a knack to using a nibbler, but quite accurate results can be produced after some practise has been gained on some scraps of aluminium sheet. It is not possible to make small and intricate cutouts using these tools as they make much wider cuts than a saw blade. The cut is usually about three or four millimetres wide. They can be used for quite small rectangular holes and the like, but are at their best when large irregular shaped cutouts are needed.

Obviously it is not possible to make cuts with curved edges when using a nibbler, since it has to make a series of short but straight cuts. You can make a pseudo curved cut just within the perimeter of the cutout, and then carefully file this out to precisely the required shape.

Unless you are very skilful, it is best to use a similar approach when using any of the tools mentioned so far. Otherwise it is very easy to stray onto the wrong side of the cutting line here and there, which in some cases will result in an unusable end product.

Nibbler attachments for electric drills are now available. These are relatively expensive though, and are probably a bit "over the top" for most electronic project work. It is probably fair to say that most of the attachments for full-size electric



drills are of little use for project construction, where you normally require precision rather than the ability to zip through large work pieces at great speed.

## MAKING YOUR MARK

When making any cutout it is important that the perimeter of the required cutout is very clearly marked, so that it can be clearly seen while you are working on the panel. Making a short cut, looking to see how accurate it was, then making another short cut, looking again, and so on, is a sure way to produce some poor quality results. It also gives very slow progress.

Marking clearly visible lines on most aluminium and plastic panels is not easy. Faint pencil marks or scribed lines may be satisfactory if you have very good eyesight, but are inadequate for many people to see clearly when working on a panel. Some fibre-tipped pens which have spirit based inks will give more distinct lines, but most of these produce quite broad lines which will not give sufficient accuracy.

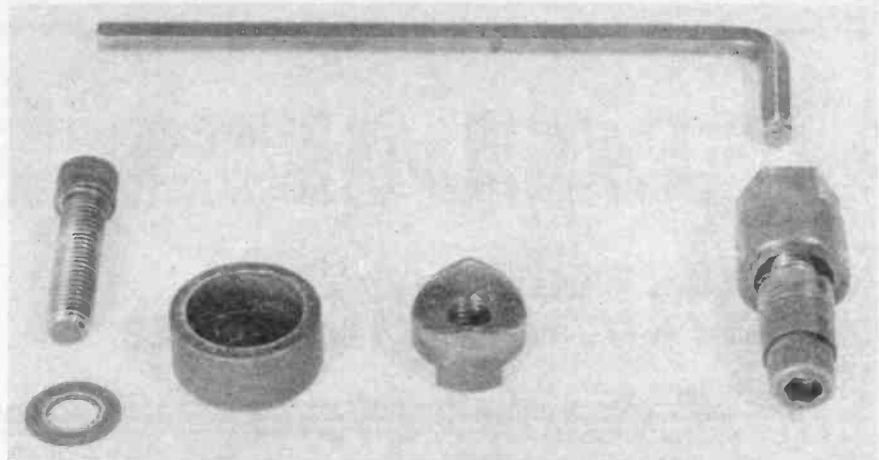
A commonly used ploy is to fix some paper or thin card over the panel, and to then mark the lines onto this using any pen or pencil that gives a fine dark line. The paper or card needs to be easily removed once the drilling and cutting has been completed.

A few pieces of double-sided adhesive tape are probably the best method of fixing the card in position on the panel. An advantage of this method is that it gives the panel some protection against accidental chips and scratches, particularly if thin card is used.

It is very important to check carefully that the marking-out of the panel has been done correctly, particularly with larger projects where mistakes are more easily made (and potentially more expensive to fix). Placing each front panel component over its respective place on the panel *before* starting any drilling or cutting should highlight any discrepancies. If there is a hole too many, a hole missing, a cutout marked too large, not enough space reserved for a component, or anything of this type, the mistake will become obvious once everything is placed in position on the panel.

Once a mistake has been made it can be very difficult to correct it. If a hole has been omitted, then it will probably be possible to accommodate it somewhere on the panel. The panel layout might look a bit odd though, and the finished unit might not be very easy to use.

*A nibbler provides a quick means of making large irregular shaped cutouts.*



*The constituent parts of a chassis punch plus an allen key (top) and assembled punch (right).*

Holes or cutouts that are made too small can be enlarged fairly easily, although with a large cutout a lot of time consuming filing may be required. An extra hole or accidental damage to a panel can be covered over with a veneer, such as a self-adhesive brush aluminium effect veneer.

With cutouts that have been made too large there is no easy solution. It might be possible to make a cutout of the right size in a small aluminium panel and then glue this in place behind the incorrect cutout.

However, if a major mistake is made it is best to discard the panel and start again from scratch. It is much better to proceed carefully and get it right first time.

## ROUNDING OFF

Where large round holes are required there are some quite quick ways of producing them. Chassis punches have been mentioned more than once before in *"Actually Doing It"* over the years, but it is worthwhile mentioning them again.

This seems to be a type of tool which is largely unknown outside the field of electronics, but a chassis punch is useful whenever a large (up to about 30 to 35mm in diameter) circular hole is needed in thin sheet metal. They also work quite well with softish plastics incidentally.

A chassis punch is basically a large nut and bolt – see photograph. A hole for the bolt must first be drilled, and the nut and bolt are then fixed in place.

The "nut" is actually a circular cutting blade of the appropriate diameter, and an allen key is used to tighten the bolt to the point where the blade is forced right

through the panel. There is a large cap beneath the head of the bolt, and the blade goes up into this, together with the washer-like piece of metal that is punched out of the panel.

This may seem like a rather crude way of making holes, but it is quite fast and produces far neater results than any other method I have encountered. The holes produced are invariably clean and burr-free. It is a system which is not restricted to round holes, and practically any shape could be punched out using an apposite cutting blade and cap.

In practice these tools are only widely available in the standard round form, although there are rectangular types available from a few sources (e.g. RS outlets). These are designed to make the cutouts for particular switches, etc.

The main drawback of chassis punches is that good quality ones are quite expensive. I find that 12.5mm and 16mm diameter ones are very useful for making mounting holes for jack sockets, DIN sockets, toggle switches, panel neons, etc. It is only worthwhile buying a larger size if you have a specific purpose for it.

The cheap way of making large round holes is to use a reamer. This is a long tapered tool having about five cutting blades along its full length. There is a hole drilled through the top (thick) end of the tool, and a metal bar which acts as a handle is placed through this.

The basic idea is to first drill a relatively small hole, and to then widen this out to the required size using the reamer. Simply place the reamer in the hole and rotate it until the hole has been increased to the appropriate size.

Some reamers are for small holes from around 3mm to 12mm diameter. The ones which cover larger sizes from about 12mm to 30mm are probably the more useful though.

Enlarging holes to diameters of 20mm or more can take a fair amount of effort, but reamers are very much an economy way of doing things. You need to proceed carefully with these tools as it is easy to slightly overdo things.

Reamers produce holes that are far from burr-free, and you need to allow for the fact that the de-burring process will slightly enlarge the hole. There are special de-burring tools available, but when working on aluminium and plastics simply trimming away the burrs with the blade of a penknife seems to be the quickest and easiest method.

# Teach-In '93

with Alan Winstanley  
and Keith Dye B.Eng(Tech)AMIEE

Part 2

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*Teach-In '93 continues a tradition of offering an interesting and thorough tutorial series aimed specifically at the novice or complete beginner in electronics. The series is designed to support those undertaking either GCSE Electronics or GCE Advanced Levels, and starts with fundamental principles to give the student a solid foundation before proceeding onto further topics.*

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IN ACCORDANCE with the recommendations contained in the various GCSE Electronics Syllabuses, which themselves comply with the National Criteria for Science, no attempt is made to explain the "physics" behind any electronic components. Instead, we are much more interested in what they look like and *how to use them*, rather than what makes them work in the way they do. Many text books on electronics start with the very dry theory of atoms, electrons and semiconductor physics, much of which in our opinion may deter the less able candidate:

IN PART ONE we introduced the fundamental rules which concern voltage, current and resistance, and we performed several demonstrations on the *Mini Lab* which for now, continues to be powered by batteries. We start this month with more basic theory and then we will use the *Mini Lab* to help demonstrate more electronic principles.

We now know that electric current flows from the most positive to the most negative (or least positive) voltage in a circuit. When we

use a 6V battery pack, *conventional* current flows out of the positive terminal, through the circuit and back into the negative terminal of the battery. Remember also that it's more accurate to talk of the "0V" terminal of a battery rather than the -6V terminal.

If we were to draw a graph of the output of our 6V battery pack, perhaps connected to a simple load like a filament bulb or a resistor, when we plot the output voltage  $V_o$  against time, it might look like Fig. 2.1. Ignoring the fact that the battery voltage will fall away as

apart from being of academic interest only, it can also be a bit boring!

We invited a highly-qualified and experienced Moderator and GCE A Level Examiner to join us and the text incorporates his very valuable suggestions to enable *Teach-In* to have maximum appeal to those candidates undertaking GCSE or GCE examinations and coursework. *Teach-In* we hope will also appeal to the experienced *Everyday with Practical Electronics* reader who might like to brush up on his or her theory.

the battery starts to age or flatten, the voltage level is reasonably constant with time.

However, many types of electrical signals actually vary with time. Fig. 2.2 is a very simple system consisting of a battery pack and a push switch in series with a bulb. The adjacent graph plots the possible states of the output voltage and clearly there are two possible conditions: *on* or *off*. The graph actually shows a *waveform* called a *square wave*, although the "squares" could be fairly brief "blips" or on the other hand, very long pulses. The time period "t" is the time taken for one complete cycle to occur.

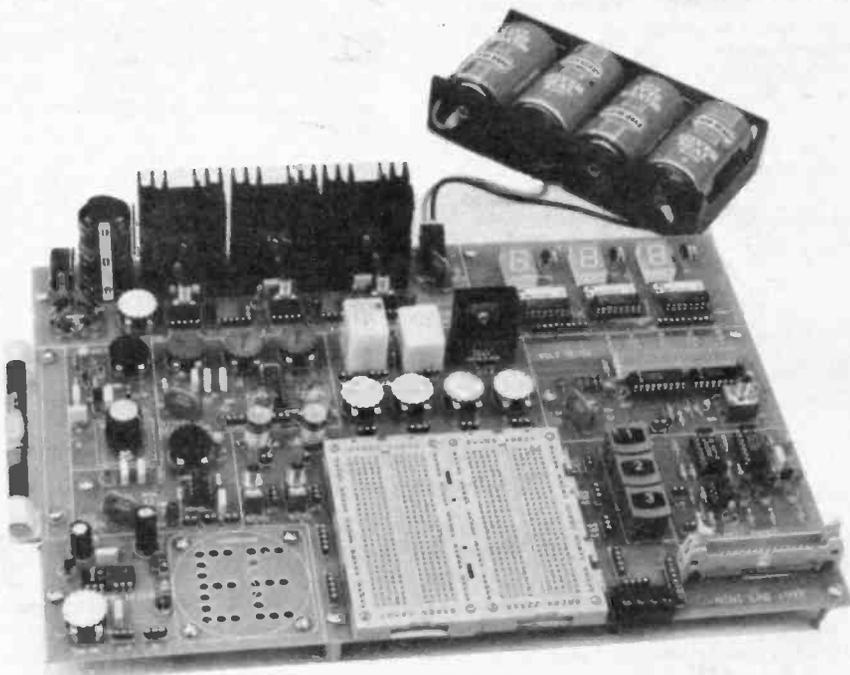
In fact, you've just seen your first *digital* signal. These have only two states, being on and off or "high" and "low" or "one" and "zero". Digital signals are the language of "logic" circuits and microprocessors and we investigate them in great depth in Part Six.

Fig. 2.3 shows a potentiometer (pot for short) connected across a 6V battery and the output voltage is present at the wiper of the pot. By rotating the spindle of the pot, you can vary the output voltage  $V_o$  accordingly. The resultant graph plot would be a wavy line, and how large the peaks and dips are depends on how much you rotate the spindle of the variable resistor.

Clearly there is no consistency in the waveform, and its shape can vary considerably. It's certainly nothing like the distinctive on/off digital waveform we saw earlier. This type of varying waveform is called an analogue signal and they are simply wavy lines which have varying states.

## ALTERNATING CURRENT

We mentioned last month how the current which flows out of a battery is classed as



"direct current" (or d.c. for short) because it can only flow one way. Alternating current (a.c.) signals are different because they flow forwards and backwards and not simply in one direction. This could be simulated as in the circuit diagram and waveform shown in Fig. 2.4.

This time we have added a second 6V

battery B2 in series to form 12V d.c. total, and this is connected across a potentiometer. However, we have made a connection in between the batteries, which forms a "centre tap". Consider this "tap" as 0V (zero volts), and therefore think of the battery terminal voltages as shown, with +6V wired to one end of the pot. and therefore -6V con-

nected to the other end. The p.d. across each battery is still 6V as normal, and the overall p.d. across the potentiometer is still 12 volts (the difference between +6V and -6V).

Now, if we consider the output voltage between 0V and the potentiometer wiper, we have a situation where with the wiper at "A", the potential at the output with respect to 0V will be +6 Volts.

If we move the wiper to the other end of the track ("C"), the output voltage with respect to 0V becomes -6V. In other words, you can actually have an output voltage which is below 0V. With the wiper at the centre "B", the output voltage will be balanced at 0V, because the 12V p.d. of the batteries is divided in two by the potentiometer, i.e. "A" will be at +6V, "B" will be at 0V and "C" will be at -6V.

It's difficult to measure the output voltage at the pot. wiper (with respect to our 0V centre tap) with a test meter unless you have a special centre-zero voltmeter which can swing between plus and minus, with zero in the middle: you would then see the output action clearly. Alternatively, a digital multimeter will automatically change polarity to indicate both positive and negative voltages on its digital display, so you could see the output voltage swing above or below 0V.

By rotating the spindle of the potentiometer, we could generate a varying analogue signal which could be positive or negative of 0V. A graph of the results might look something like that shown in Fig. 2.4. Unlike the waveforms shown earlier, notice this time how it is possible for the waveform to dip into the negative region, below 0V, depending on how you move the control.

It's not unusual for a circuit to require both positive and negative supply "rails" - an arrangement called a "split supply" - though the vast majority of the *Teach-In* demonstration circuits will use a straightforward positive rail together with a 0V rail. We will look at power supplies in more detail next month.

## SINE WAVES

If you were able to rotate the potentiometer spindle at a perfectly timed regular interval from one limit to the other, the output voltage would appear like the undulating waveform of Fig. 2.5 which depicts a sine wave.

We can usually measure a d.c. voltage simply in terms of its voltage with reference to the 0V rail, but how can we define a waveform like a sine wave? And how can you actually measure a waveform like a sine wave?

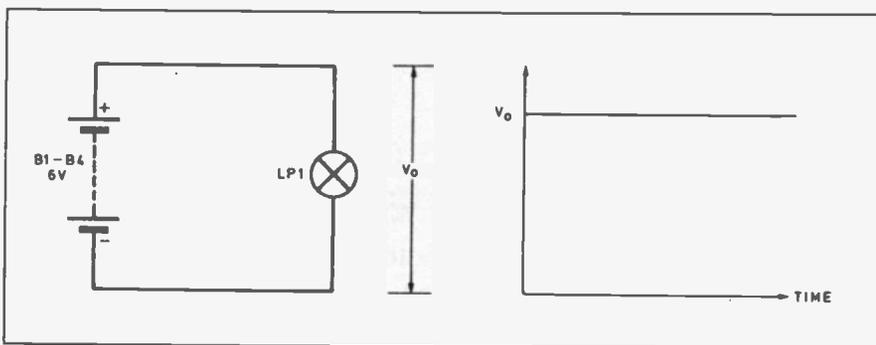


Fig. 2.1. Graph showing the output voltage versus time of a simple circuit.  $V_0$  remains constant with time.

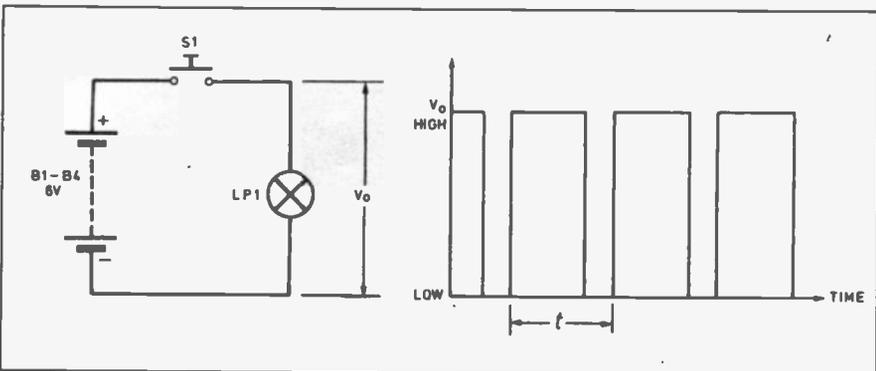


Fig. 2.2. A simple switching system producing a square wave consisting of two states - "high" and "low" or "on" and "off".

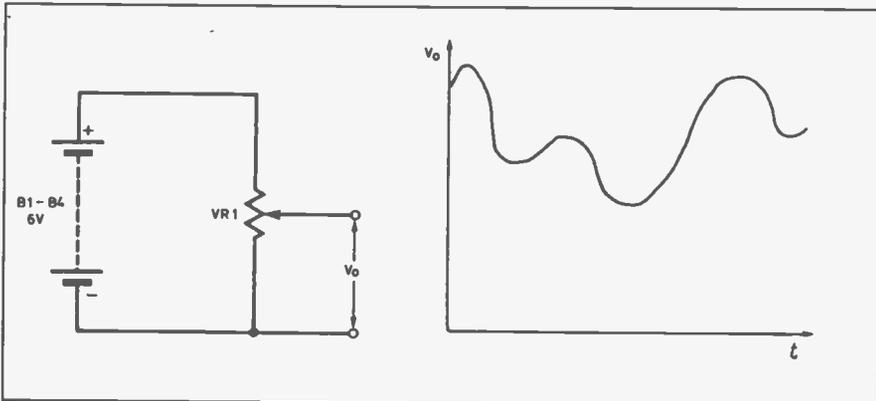


Fig. 2.3. This simple potentiometer circuit will produce an analogue waveform such as that shown in the graph.

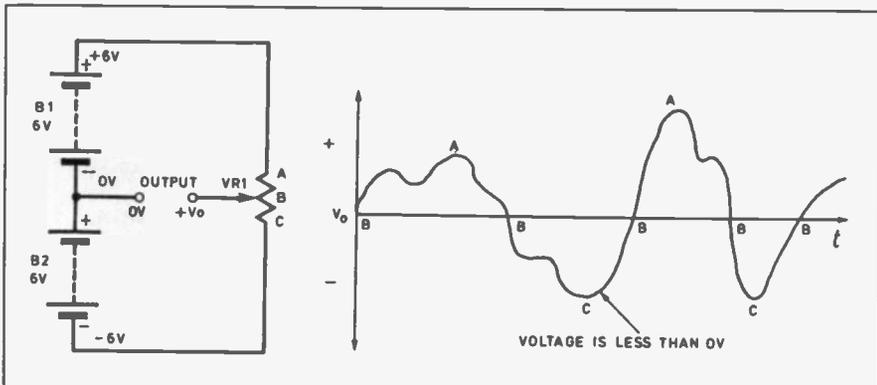


Fig. 2.4. Simulation of an alternating waveform. The output voltage  $V_0$  could be considered as swinging between +6V and -6V with respect to 0V.

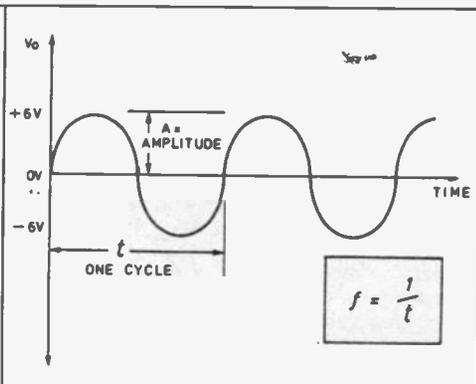


Fig. 2.5. A sine wave. The frequency of the waveform is given by the formula  $f = 1/t$  where  $t$  is the time for a complete cycle, in seconds. Frequency is measured in Hertz.

Several parameters are shown in the sine wave of Fig. 2.5 which will help us to define its characteristics. The first is how high the peaks rise with respect to 0V: this is termed the **amplitude** of the sinewave. In our simple example, the amplitude  $A$  is 6 volts.

Another important parameter is how frequently the sine waves repeat themselves. The frequency would be determined by how quickly you can rotate the spindle back and forth between the two end stops of the potentiometer: the quicker you can do this every second, the higher the frequency.

The unit of measurement of frequency is the **Hertz** (abbreviated simply to "Hz."). It's defined as the number of complete cycles which occur per second. One cycle is equivalent to one complete sine wave in this example, as shown in the diagram. The formula used to calculate frequency is:-

$$f = 1/t \text{ Hertz}$$

where  $f$  is the frequency (cycles per second) in Hertz, and  $t$  is the period, or time taken in seconds for one complete cycle.

So for example if one complete sine wave cycle takes two seconds, the waveform frequency is 0.5Hz, i.e. half a cycle every second. If the period is 25mS (0.025S), the frequency is 40Hz. Similarly a more rapid sine wave might have a frequency of say 10Hz and working backwards from the above formula, we could state if we needed to that the time period  $t$  for the wave is 0.1 seconds.

The Hertz can be quite a cumbersome unit to deal with and, like the Ohm, we often add a prefix to save writing many zeros when dealing with high frequencies.

The symbol "kHz" is the symbol for kilohertz, or thousands of cycles per second. 25kHz equals 25,000 cycles per second. MHz is the abbreviation for megahertz, or millions of cycles per second. If you own a hi-fi system or car radio, you might have a digital tuner which informs you of the frequency of the radio station you're tuned to. Look at the digital display for the "kHz" and "MHz" symbols on the a.m. and f.m. wavebands.

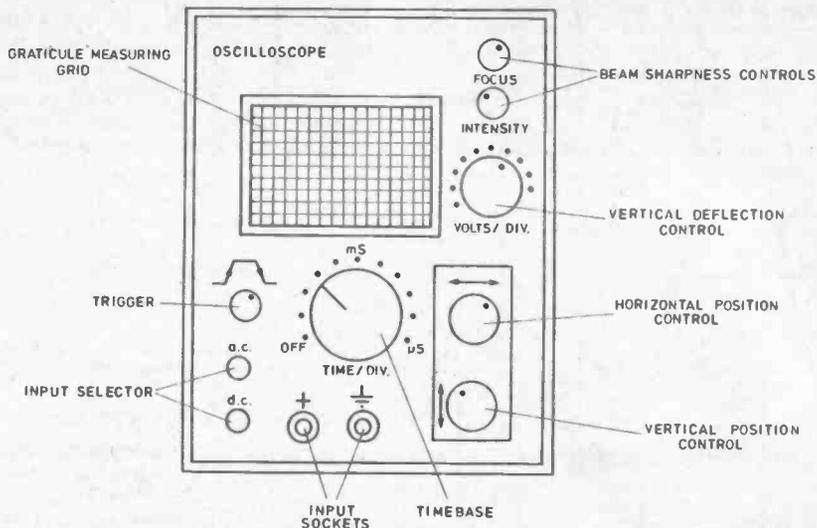
The formula for frequency holds true for any repetitive waveform such as a square wave: if we know the time taken for one complete cycle ("t" in Fig. 2.2) then we can work out the frequency of the waveform.

## THE OSCILLOSCOPE

The cathode-ray oscilloscope (C.R.O.) is one of the most versatile pieces of test equipment you can buy. The basic operation of the C.R.O. is described in a separate section; even if you don't possess one, you should still be familiar with their capabilities and basic controls.

An oscilloscope actually plots a graph of voltage against time, just like the graphs of the waveforms we saw earlier. The graph takes the form of a trace on the screen which mimics what is happening in the circuit under test. If we know the *time-base* and *vertical deflection* settings (q.v.) of the C.R.O., then we can make some meaningful measurements on the waveform which appears on the display. Fig. 2.6 shows a real sine wave which we observed with a C.R.O. and then plotted on a computer printer. The *graticule* grid is also shown. (The sine wave is actually taken from the *Mini Lab* Signal Generator described in future parts.)

The **amplitude** of the sine wave is easily calculated. The vertical deflection has been set at 500mV per division, and we can see that the measurement "A" is 2.5 divisions.



## THE OSCILLOSCOPE

In order to actually "see" what is going on in a circuit, a *cathode ray oscilloscope* (C.R.O.) is invaluable, though they are expensive pieces of equipment which only the keen enthusiast might buy. The diagram above illustrates the controls of a simple C.R.O.

An oscilloscope contains a *cathode ray tube* like a small round television tube. A fine beam of electrons strikes the phosphors coated inside the face of the tube making them glow temporarily, perhaps bright blue or green. Two factors control the path of the beam: the *vertical deflection* on the screen which is directly proportional to the voltage connected to the C.R.O.'s input terminals; secondly, a *time-base* which causes the beam to repeatedly traverse from left to right at an accurate rate.

Superimposed on the screen is a precise measuring grid called a *graticule* which may be calibrated in centimetres or simply in "divisions". If the vertical deflection switch is set to, say, two volts per division then a 6V d.c. input will cause the beam to deflect upwards by three divisions on the graticule.

The time-base control is calibrated in fractions of a second per division, and if the time-base was set at, say, 25mS (25 milliseconds) per division then the beam will traverse the screen left to right at precisely that rate. A timebase can easily operate at microseconds (millionths of a second) per division if necessary.

A *trigger* control locks a waveform on the screen so that the waveform shape "freezes" to enable us to observe it. The trigger control determines the level at which the beam will start its sweep, and whether it's triggered on the positive-going or negative-going slope of a waveform.

Also, it's possible to use a switch to select either d.c. or a.c. inputs: In a.c. mode, a small capacitor is inserted in series with the input signal and it removes any d.c. voltages superimposed on the signal, so that we can simply observe the a.c. "wobble". In d.c. mode, the input signal passes straight through to the C.R.O. (This

topic is discussed in Part Four when we look at a.c. coupling.)

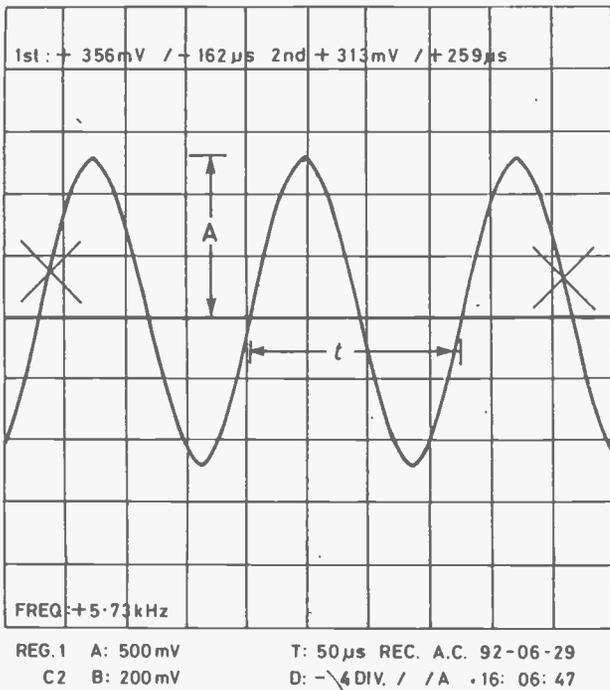
### SETTING UP

After switching on and letting the instrument warm up, firstly centralise the electron beam on the screen with no input voltage present: do this by turning off the time-base completely and adjust the *horizontal and vertical position controls* to manually move the dot of light to the centre of the graticule. Adjust the *focus* and *intensity* controls to obtain a pin-sharp dot of light.

Then turn on the time base and adjust it to make the beam sweep across the screen. Because no input voltage is present, the line now being traced is effectively a 0V reference line. Fig. 2.3 showed how you could obtain a varying voltage from a 6V battery - a circuit like this could be connected to the input of the 'scope. Note the beam's deflection with reference to the graticule so that you can measure the voltage on the screen: it's the distance between your 0V reference line and the beam's new position with the input signal connected. In effect you are using the C.R.O. as a simple voltmeter.

Unlike moving coil multimeters, oscilloscopes don't appreciably load the circuit under test, so they are very accurate. However the major advantage of a C.R.O. over a multimeter is that we are actually able to see the *shapes* of waveforms on the cathode-ray tube. An example is given in the main text of how to utilise a C.R.O. to measure the frequency and amplitude of a waveform like a sinewave.

The *Mini Lab* will in due course sprout a compact waveform generator capable of producing three different types of waveform. This will be of especial interest to those readers who own, or have access to, an oscilloscope. However, we continue on the basis that it is *not* compulsory to possess an oscilloscope in order to follow *Teach-In*. For the purposes of GCSE/ A Level Electronics, though, you should be familiar with the basic operation of the C.R.O. - so try to have a go on one if you can.



The amplitude is therefore 1.25 volts. The peak-to-peak amplitude is equivalent to 2A, i.e. 2.5 volts.

The frequency can be calculated by measuring the time taken for one complete cycle of the sine wave to occur, and then applying our formula  $f = 1/t$ . We set the time-base at  $50\mu\text{s}$  per division. One complete cycle covers 3.5 squares as represented by "t" on the diagram, which is equivalent to a period of  $175\mu\text{s}$ . Hence, the frequency of this waveform is  $1/175\mu\text{s}$  or  $5.7\text{kHz}$ . (Indeed, the built-in digital frequency meter display on our C.R.O. screen confirmed this as you can see.)

We now analyse sine waves in a little more detail. Again, we deliberately avoid any intensive mathematics as *Teach-In* is more interested in the results rather than how we arrived at them. Also, for the next section, it's not necessary to own an oscilloscope in order to progress through the text.

## A.C. AND POWER

We already know from Part One that a resistor will dissipate power when a there is a voltage drop across it. When a simple d.c. voltage is applied, the resistor will become warmer, though you might not physically notice it, depending on circuit values.

Fig. 2.7 shows an a.c. voltage source applied across a resistor. Unlike a d.c. voltage supply, there is no longer a constant voltage across the resistor. Instead, a peak voltage is applied which actually drops to zero before the current switches the other way through the resistor, which heats up at a varying rate as the voltage across it rises and falls.

In fact, the resistor averages out the power dissipation when an a.c. current flows through it. If the a.c. has a peak voltage of, say, 10V, this cannot produce as great a heating effect in the resistor as a d.c. voltage of 10V would: the a.c. peak voltage of 10V only appears across the resistor for a fraction of the time (twice per cycle) – for the rest of the period, it's less than 10V. Compare this against 10V d.c. which would appear across the resistor *all the time*.

It's actually possible to calculate *effective values* for a.c. sine waves which would give the same heating effect in the resistor as a particular d.c. value. We can do this by considering the root-mean-square or r.m.s. values of the sine wave. Without detailing the

$$f = \frac{1}{t} = \frac{1}{175\mu\text{s}} = 5.7\text{ kHz}$$

Fig. 2.6 (left). A real sine wave shown on an oscilloscope graticule. The frequency can be calculated as shown: compare against the actual digital frequency readout on this oscilloscope.

Fig. 2.7 (above right). An alternating waveform connected across a resistor. The voltage has a peak value of 10V, and swings back to 0V before reversing the other way.

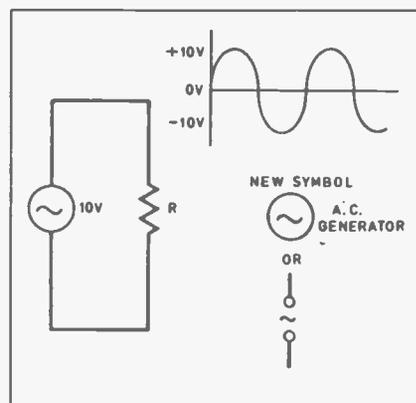
mathematics behind it to any extent, the formula:

r.m.s. equivalent = peak value  $\div$  1.414  
 $(1.414 = \sqrt{2} = \text{the square root of } 2)$   
 can be applied to both the a.c. peak voltage and peak current. Multiplying V r.m.s. and I r.m.s. together (because  $P = I \times V$ ) in an a.c. circuit would give the same value of power dissipation in Watts as multiplying V d.c. and I d.c. would, in an equivalent d.c. circuit.

In Fig. 2.8, a sine wave of 10V *peak value* is plotted – this has an equivalent value of 7.07V (i.e.  $10\text{V a.c.} / 1.414$ ). Hence 7.07V a.c. r.m.s. would cause the same heating effect in a resistor as a d.c. voltage of 7.07V.

When we talk about a.c. values, we nearly always mean the r.m.s. equivalent, not the peak value. The peak value has little use except when we specify the ratings of some components, when we might be interested in how high a voltage or current is likely to be applied. We then have to ensure that the component will be able to withstand these peak values. Note that the formula given above for calculating r.m.s. values applies *only* to sine waves – not a square wave for instance.

In the domestic mains power supply, the output is a sine wave which is stated as being 240V a.c. In actual fact, it's 240V r.m.s. and using the above formula in



reverse, the peak value of the sine wave is 339V ( $240\text{V} \times 1.414$ ). Therefore, the 240V r.m.s. a.c. voltage from a mains outlet gives the same heating effect in a resistor (like an electric fire element) as a d.c. – and equally lethal – level of 240V d.c. would. But the insulation still has to withstand the peak value of 339 volts.

We can always apply the power dissipation formula ( $P = I \times V$ ) and also Ohm's Law ( $R = V/I$ ) in a.c. circuits provided that we are referring to r.m.s. values, not peak or peak-to-peak figures. If your multimeter (either digital or moving-coil) has an a.c. voltage scale, this will actually read in r.m.s. volts, not the peak voltage.

## THE DIODE

Let's turn to the *Mini Lab* for some practical work. We continue to use batteries as our source of power, so ensure that the "EXT BATT" option is selected in the power supply area of your *Mini Lab*.

Fig. 2.9(a) is a simple circuit diagram which shows a d.c. voltage supply and a filament bulb, together with a new symbol. D1 is a component in the class of *semiconductors* and is called a diode, and to demonstrate this device you will require a 1N4001 diode from your supplier. This is a simple two-wire device, generally with one end of the body painted with a white band. Build this circuit as shown in Fig. 2.9(b), using the breadboard section to connect the diode D1, and note the condition of the bulb below:

### CONDITION OF THE BULB:

DIODE CONNECTED AS  
IN FIG. 2.9(b)

DIODE REVERSED

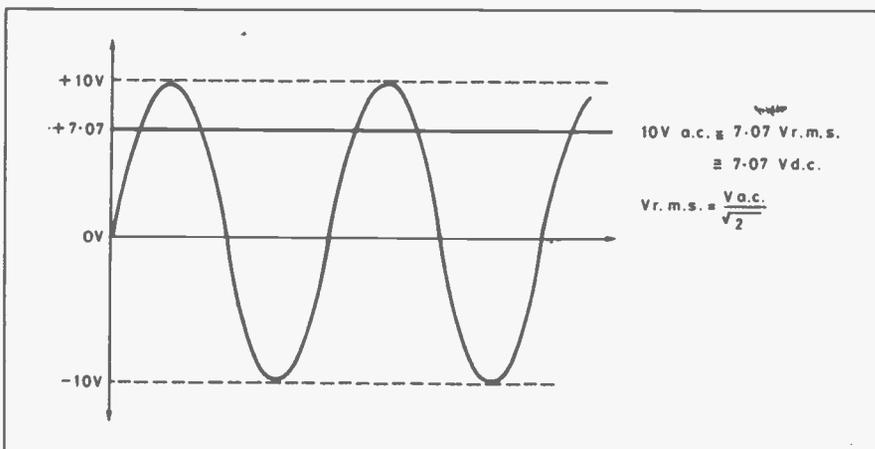


Fig. 2.8. A 10V a.c. (peak value) waveform would have an effective value of 7.07V in d.c. terms, or 7.07V r.m.s. When we refer to a.c. values, we nearly always mean the root-mean-square value rather than the peak figure.

# GCSE QUESTION

This month's question is taken from the Midland Examining Group GCSE Electronics (1751) 1989 Paper 1. We acknowledge their co-operation with thanks. Here's an opportunity to apply your knowledge of light-emitting diodes and series resistors in this straightforward question.

The answers are the work of the authors and do not necessarily constitute the only correct solution.

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- 1 The circuit shown in Fig. 1 has an LED connected in parallel with a buzzer to a 9 V power supply. A 1.2 kΩ resistor is connected in series with the LED. A switch is connected in series with the buzzer.

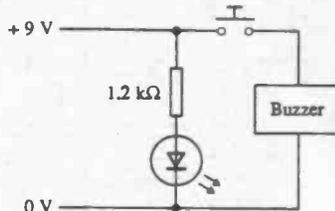


Fig. 1

- (a) Fill in this Table. It shows what the LED and buzzer do when the switch is open and closed. [4]

Switch	LED (dark or light)	Buzzer (silent or buzzing)
Open		
Closed		

- (b) Fig. 2 shows the coloured bands on the 1.2 kΩ resistor. If the resistor has a tolerance of 10%, name the colours of the bands on Fig. 2. [4]

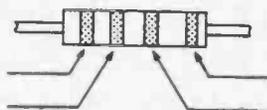


Fig. 2

- (c) The potential difference across the LED is 1.8 V. Calculate the potential difference across the 1.2 kΩ resistor. [2]

..... V

- (d) Calculate the current going through the LED. [3]

.....

What happens when you reverse the diode? Connect it the other way round again and confirm. The bulb will only illuminate when the diode is connected one way round. Which way round? The diode is a "uni-directional" device, and will only permit current to flow one way. It blocks current in the other direction, as its circuit symbol indeed suggests.

(Note: some Examinations Boards have slightly differing but equally correct symbols from ours for diodes and also a few other components. You may lose marks for using a symbol which is not approved by your Examining Board, whilst other Authorities may not object provided you are consistent throughout your work. Please consult your tutor if you have any doubts in this respect.)

With the bulb illuminated, use your multi-meter to measure the voltage across the diode (meter + terminal connected to +6V). Your meter should be set to a low d.c. volts range, say 1V f.s.d.; record your measurement below.

## VOLTAGE ACROSS THE CONDUCTING DIODE: Volts

Your answer will be in the region of 0.6V to 0.7V. When the diode permits current to flow (the bulb lights up), it is said to be forward biased. Conversely, when it blocks current (the bulb is extinguished) it is reverse biased. A voltage of about 0.7V always appears across a silicon diode (like the very common 1N4001) when it is forward biased, and this is called the forward voltage.

In fact the anode of the diode must be 0.7V more positive than the cathode before the diode can become forward biased and allow current to flow through it. See Fig. 2.10 which summarises the function of the diode.

Fig. 2.11 shows the characteristic curve for a typical silicon diode. The plotted line shows how the forward current through a diode can only rise once the forward voltage of the diode exceeds 0.7V. When it is reverse biased, hardly any current flows at all so it doesn't conduct.

## DIODE SPECIFICATIONS

Diodes have electrical ratings, and the main parameters we are normally interested in are these:-

$V_f$  Forward Voltage,

$I_f$  Forward Current, the maximum current which can flow through the diode when it is conducting

PIV Peak Inverse Voltage, the maximum reverse voltage allowed before the device is damaged

The data for our humble 1N4001 diode states that it's a 50PIV 1A device. A forward voltage of 0.7V roughly is fairly standard amongst all silicon diodes and therefore you won't often see it quoted in any published data. You might also see the

symbol  $V_{rrm}$  instead of PIV. Have a look in some suppliers' catalogues to see if you can spot the ratings of diodes. Also check out their shapes and sizes, and compare the cost. Note how the prices increase for diodes which can carry more current (up to

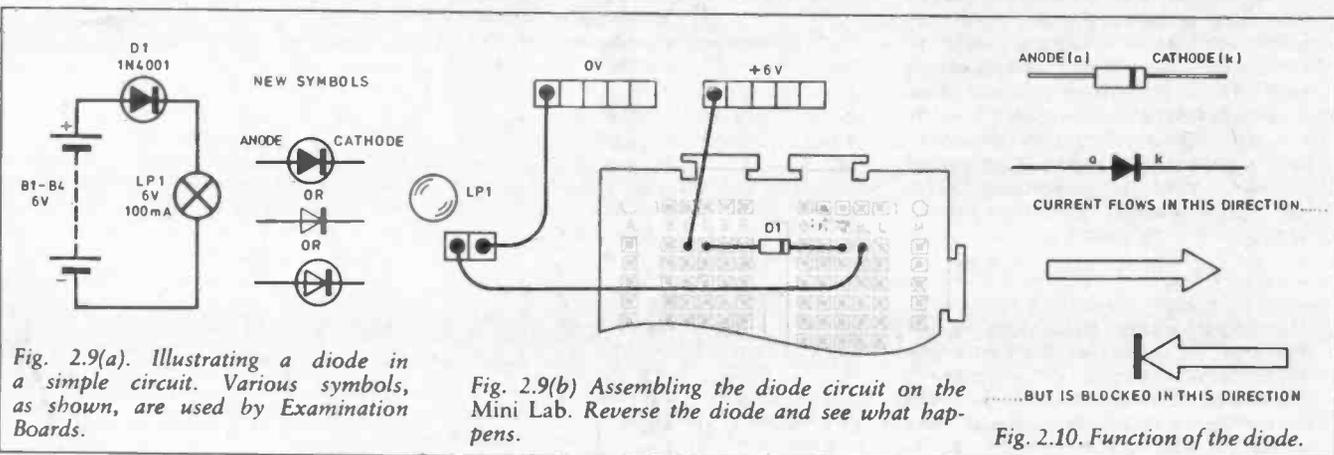


Fig. 2.9(a). Illustrating a diode in a simple circuit. Various symbols, as shown, are used by Examination Boards.

Fig. 2.9(b) Assembling the diode circuit on the Mini Lab. Reverse the diode and see what happens.

Fig. 2.10. Function of the diode.

(e) The graph of Fig. 3 shows how the current through the buzzer depends on the potential difference across it. Use it to find out how much current goes through the buzzer when the switch is pressed. [2]

The current through the buzzer is .....

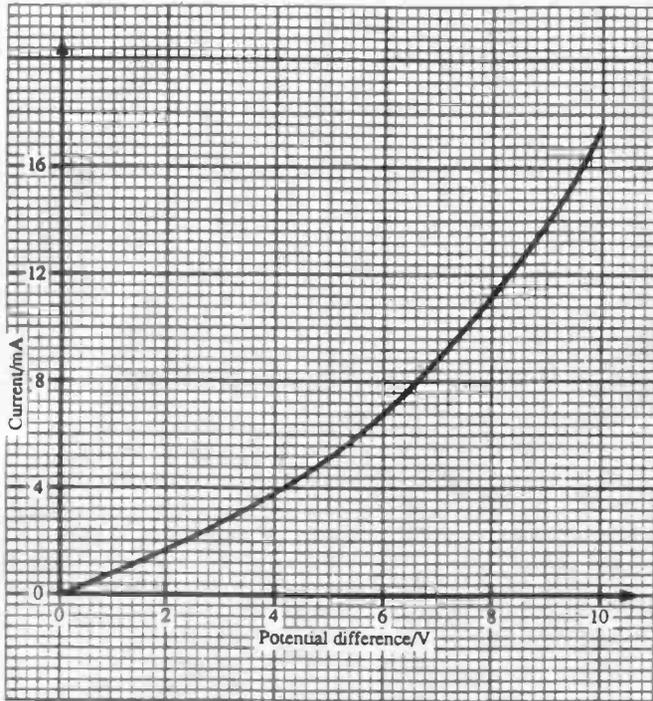


Fig. 3

(f) Sketch on the graph of Fig. 3 how you would expect the current through the LED to change with the potential difference across it. Assume that the LED is forward biased. [3]

(g) The 1.2 kΩ resistor in Fig. 1 is replaced with a 4.7 kΩ resistor. Describe and explain what effect this would have on (i) the LED, and (ii) the buzzer. [4]

- (i) the LED: change .....
- reason .....
- (ii) the buzzer: change .....
- reason .....

many tens of amps) or which have a higher PIV.

Apart from silicon diodes you may come across devices fabricated from another material called "germanium" (not to be confused with "geranium" which is a flower!). These differ from silicon types in several respects: they have a forward voltage of typically 0.2V, and their current-carrying capabilities are vastly inferior to silicon types. A typical germanium device such as the OA91 may have an  $I_f$  rating of only 50mA. These sensitive low-power devices are classed as "signal" diodes in radio circuits for instance, where they work with audio or radio-frequency waveforms.

### DIODES AND RELAYS

One major headache of using relays is the fact that they are based around an electromagnet, which is basically a coil wound onto an iron core. Coils in electronic circuits can have a major drawback: when you switch a coil off, the magnetic field around the coil collapses back into it,

causing a large back e.m.f. to be generated. This takes the form of an often massive reverse electric pulse which can cause havoc with delicate electronic components such as transistors.

The *Mini Lab* incorporates a relay on the circuit board, and Fig. 2.12(a) illustrates a basic switching system where a push-operated switch S1 supplies power to the coil of the relay RLA. Closing S1 will supply power to the relay which operates, and in turn illuminates the bulb LP1. At this time, point "A" is actually at +6V with respect to point "B" which is at 0V.

It's similar to the circuits which were demonstrated last month when we discussed the relay. However there is one important difference, which is the inclusion of a diode D1 across the relay coil. When S1 is closed, D1 is

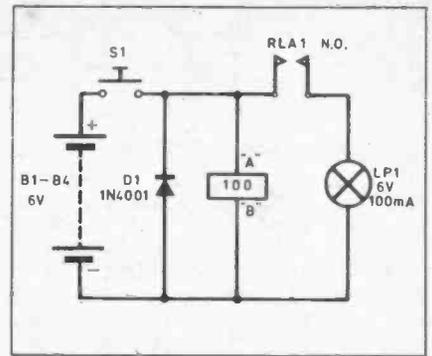


Fig. 2.12(a). A simple relay circuit which will drive LP1 once the switch S1 is closed. The diode D1 is used to short out any reverse-voltage pulse ("back e.m.f.") which is generated when the relay is de-energised.

reverse-biased because its cathode is at a voltage of +6V whilst the anode is at 0V.

Therefore under normal circumstances, D1 does nothing at all because it cannot conduct. However, when S1 is opened again, a reverse voltage (back e.m.f.) appears temporarily across its coil, such that point "B" is very much more positive than point "A". Now, the diode is instantly forward biased. It thus conducts and shorts out the back e.m.f. pulse generated by the coil.

Without the diode, the back e.m.f. which is created by the "implosion" of the relay's magnetic field will probably damage other components in the circuit. You will always find a reverse-connected diode across relay coils when they are used in electronic circuits. It's essential!

Fig. 2.12(b) shows an oscillogram of this actual effect which we captured using advanced test equipment. We connected a relay identical to that on the *Mini Lab* in an arrangement similar to that in Fig. 2.12(a),

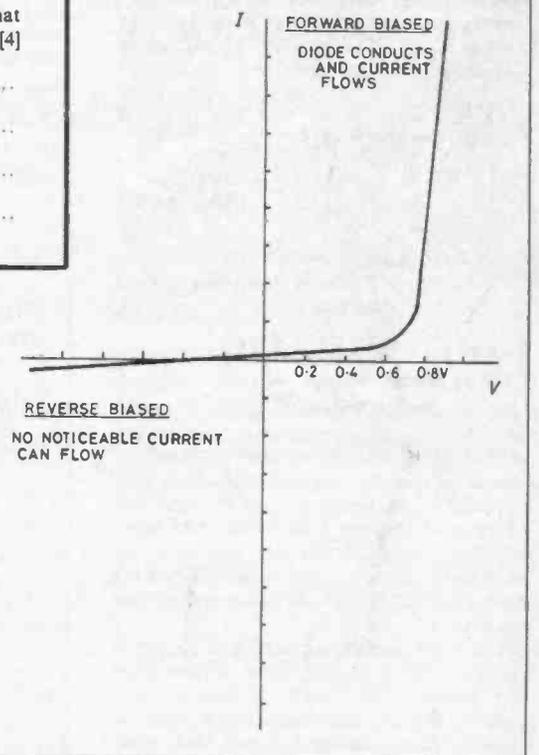


Fig. 2.11. A typical "characteristic curve" of a silicon diode, which will only start to conduct once the voltage across it exceeds approximately 0.6V – the anode must be 0.6V more positive than the cathode. In the opposite direction, no current can flow (apart from a tiny "leakage" current).

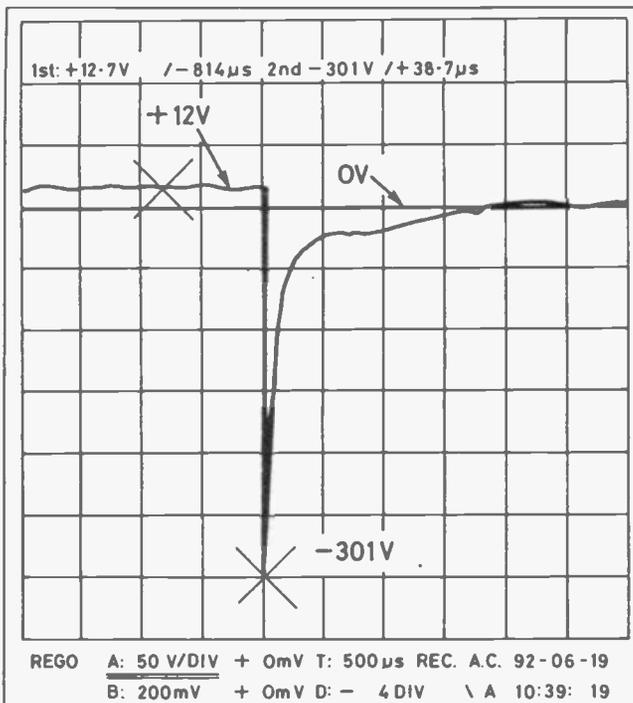


Fig. 2.12(b). Oscillograph of the back e.m.f. generated by a relay coil.

both with and without the diode. Our oscilloscope recorded the back e.m.f. pulse which was generated when the relay was switched off: as you can see, the peak reverse voltage was measured at an incredible -301 volts! Fig. 2.12(c) shows the result of including a diode to "shunt" the back e.m.f. which is now all but eliminated.

There is a location on the *Mini Lab*, adjacent to the general-purpose relay, where a suitable diode (e.g. 1N4001) can be permanently soldered into position to protect your test circuits from back e.m.f. You can solder it into place now, if you wish, taking care not to overheat it as semiconductors can be damaged by excessive soldering times. Once the diode is in place, it becomes imperative that you connect the relay coil correctly so that the diode is normally reverse biased.

Other important applications of diodes are shown next month when we utilise the unidirectional effect in power supplies to form a "rectifier". We also look at an interesting version of the rectifier called a "thyristor" which only conducts when you input a suitable control signal.

## LIGHT-EMITTING DIODES

Diodes contain a special "junction" manufactured from either silicon or germanium, which has been carefully treated to transform it into a semi-conducting material with certain operating characteristics. However, manufacturers have created a whole class of light-emitting diodes which are invaluable as solid-state indicators in electronics. They don't have a filament like a bulb, and in normal use last forever.

A typical light-emitting diode (l.e.d.) is shown in Fig. 2.13, and these devices have two lead-outs designated anode (a) and cathode (k) like a normal diode, and by forward-biasing the l.e.d., it will emit light. You can buy them in red (cheapest and most common), green, yellow, orange and even blue (unusual and very expensive). A whole host of shapes and sizes is available - look through the "Opto-Electronics" section of a good catalogue to see.

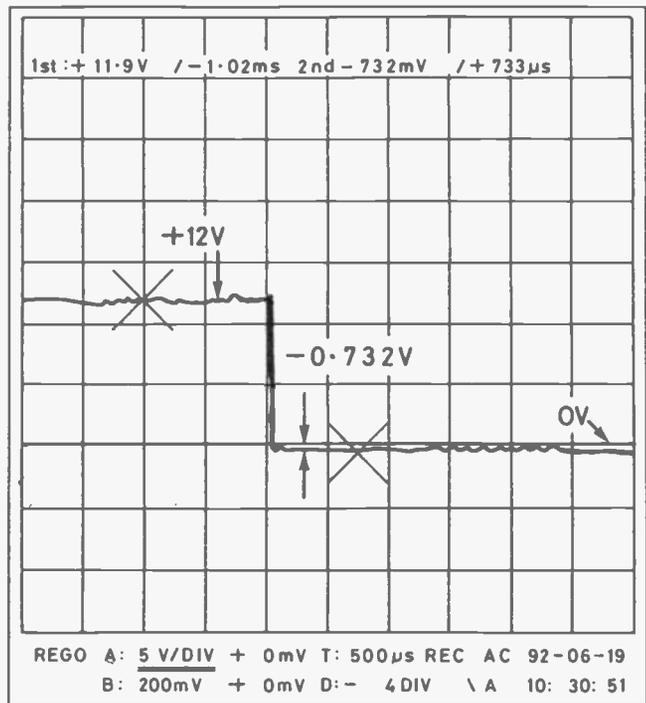


Fig. 2.12(c). The result of including a reverse biased diode across the relay coil.

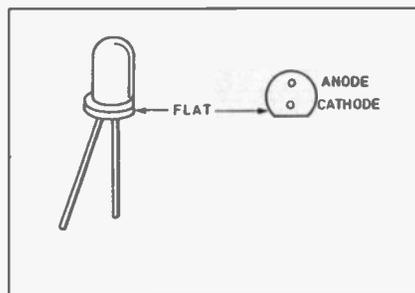


Fig. 2.13. A typical light-emitting diode, with pin-out connections. Usually, the flat notch is nearest the cathode. Sometimes, one lead is longer than the other, in order to identify anode and cathode - check the manufacturer's data.

There are some important operating conditions to observe, or an l.e.d. could be damaged. Firstly, they hate reverse voltages, and are likely to be destroyed by anything over 5V PIV or even less - so connect them the right way round, forward biased! Secondly, they have a typical forward voltage of about 1.8V. It's essential that you don't test them by connecting them directly across a 6V battery (for example), because you will exceed their  $V_f$  rating and destroy them.

Light-emitting diodes always have their lead-outs identified somehow so that you know which is which. Generally you see a flat mark on the base of the resin body adjacent to the cathode. Other times, the anode and

## GCSE QUESTION (see previous page)

### ANSWERS

- With the switch open, 9V is applied to the l.e.d. and series resistor but not the buzzer. Hence the l.e.d. is alight and the buzzer is silent. With the switch closed, 9V is now applied to the buzzer, so the l.e.d. is alight and the buzzer sounds. The switch has nothing to do with the state of the light-emitting diode - it's a red (l.e.d.) herring!
- From the Resistor Colour Code, the digits required are 12 and the multiplier to obtain 1.2k (1200 ohms) is  $\times 100$ . Hence the coloured bands are Brown (1), Red (2), Red ( $\times 100$ ) and finally Silver (10%). The Resistor Code, by the way, is given with the Paper.
- If 1.8V appears across the l.e.d. (forward voltage), the rest of the supply voltage appears across the resistor, so the answer is  $9V - 1.8V = 7.2V$ .
- Hmmm ... the current flowing through the l.e.d. must be the same as that flowing through the series limiting resistor: use Ohm's Law  $I = V/R$  so the current is  $7.2 / 1200 = 6mA$ . No marks were lost if you wrote 0.006A instead, or even 6/1000 Amps.
- The p.d. across the buzzer when the switch is closed is 9V. Therefore, read from 9V on the Voltage scale upwards to intersect the curved line: read across to obtain an answer of 14mA. Interestingly, the Examiners commented that many candidates could not read the graph, a fundamental skill required in electronics.
- A constant 1.8V will appear across the l.e.d. once it's forward biased. Until then, no current can flow. The graph rises steeply when the l.e.d. starts to conduct, when we know from (e) that 14mA will flow through it with 1.8V across it.
- (i) The l.e.d. will dim. Increasing the resistance reduces the current flowing through the l.e.d. to  $(I = V/R)$  1.5mA.  
(ii) There is no change with the buzzer, because its supply voltage remains at 9V. Its operation has nothing whatsoever to do with the l.e.d. series resistor - another red herring to test you!

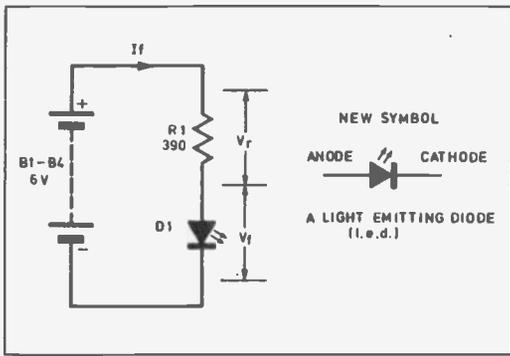


Fig. 2.14(a). A light-emitting diode (I.e.d.) operated in series with a limiting resistor R1.

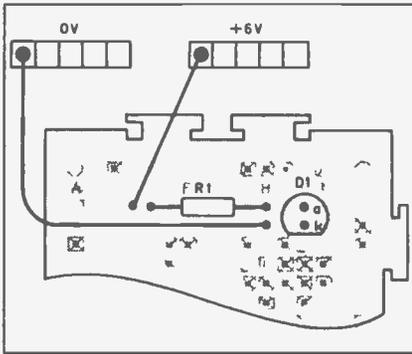


Fig. 2.14(b). Connecting an I.e.d. with a series resistor to the 6V Mini Lab supply.

cathode are denote by different lengths of the lead: check the supplier's data if in doubt.

Nearly always you will want to drive an I.e.d. from a voltage higher than the maximum permissible forward voltage, such as the +5V rail which is extremely common in digital circuits. You have to ensure that when it's illuminated, only the forward voltage value appears across the I.e.d., nothing greater. Anything less, and the I.e.d. simply doesn't light; too much, and it spells death to the I.e.d.!

It is also necessary to consider the *current flowing through the device*. Light-emitting diodes have a maximum forward current ( $I_f$ ) rating which is usually expressed in milliamps. We have to limit the current so that this current is not exceeded. 50mA might be typical, but you can obtain a respectable light output from an I.e.d. when as little as 10-20mA, or even less, flows through it.

Thus when operating a light-emitting diode from a (d.c.) voltage, we generally have to reduce this voltage so that only the forward voltage of the I.e.d. appears across it – the rest of the voltage has to be wasted somewhere – and we have to limit the current flowing through the I.e.d. as well.

Fig. 2.14(a) is a simple circuit to demonstrate the light-emitting diode. For this section, you will need any general-purpose I.e.d., say 5mm diameter so that it isn't too fiddly to handle: they cost about 20p.

The basic specification for a typical I.e.d. is as follows:-

Type: RS 590-446  
 Colour: red  
 Diameter: 5mm  
 $I_f$  (maximum): 30mA  
 $I_f$  (typical): 10mA  
 $V_f$ : 2V typical  
 $V_{rev}$  max: 3V  
 Viewing angle: 30 degrees.

The values shown are typical of other types of I.e.d.

For our demonstration we will operate the I.e.d. from our 6V battery pack, and because the typical forward voltage for this I.e.d. is

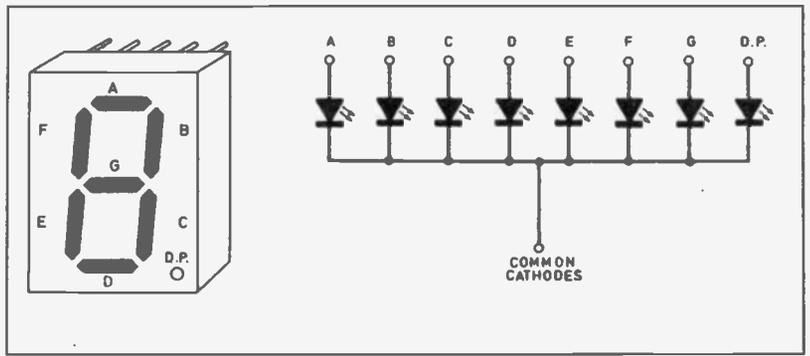


Fig. 2.15. A seven-segment light-emitting diode digital display. Either common-anode or common-cathode types are available. CC-type is shown (right).

2V, we have to lose 4V before we can drive the I.e.d. safely. Additionally, we have to ensure that the forward current ( $I_f$ ) is limited to no more than 30mA – though let's choose 10mA which from experience is adequate to give a reasonable output of light. A lower current simply reduces the brightness.

## SERIES RESISTOR

For the I.e.d. to operate reliably, we have to employ a series limiting resistor so that 4V will appear across the resistor, and the current flowing through the I.e.d. (and therefore through the series resistor) will be limited to 10mA. It's really easy to calculate the resistor value using Ohm's Law (see Part One):

$$R = V/I = 4/0.01 = 400 \text{ ohms.}$$

Remembering that resistors are produced in a range of preferred values, the one nearest to 400 ohms is actually 390 ohms. To double check the current which will flow using a 390 ohm type, use  $I = V/R$  to obtain an answer of 10.2mA – quite acceptable.

The final consideration is the power dissipated by the resistor. You can use either  $P = V^2/R$  or  $P = I^2R$  which will give about 0.04 Watts (40mW), so a standard 0.25W 5% resistor will easily be more than adequate.

The power dissipated by the I.e.d., incidentally, will be a mere 20mW ( $P = IV$ ). Thus they are very energy-efficient devices and ideal for operating with low-power circuits – compare against our 6V filament bulb used in Part One, which draws some 100mA of current and dissipates 600mW of power, though the bulb is of course much brighter.

Now use a 390 ohm resistor and an I.e.d. to build the circuit of Fig 2.14 on the Mini Lab, as shown in Fig. 2.14(b). Note particularly the polarity of the I.e.d. which in accordance with our standard practice is shown from an aerial (bird's eye) viewpoint in this diagram.

You may wish to try other colours in this circuit to assess the effect, and the red device can be exchanged for any other colour, size or shape of I.e.d. if you have any. Simply remember to observe the polarity.

## OTHER TYPES

Check out a catalogue for *bi-colour* types. These glow either red or green, depending which way round you connect them. They're great fun to play with. Also look for *flashing* types, which are designed to run from a stable +5V supply and incorporate a miniaturised circuit to make the I.e.d. flash continually. You might see *constant-current* light-emitting diodes which don't need a series resistor: they adjust automatically to ensure that a safe forward current flows. They're ideal for use with varying voltages.

Our final look at light-emitting diodes is the *seven-segment display*, see Fig. 2.15. The bars of the display are actually elon-

gated light-emitting diodes; and by activating various segments, different numbers can be displayed. Each segment is denoted by a letter A-G as shown, and a decimal point is generally included too.

Rather than have seven cathodes and seven anodes – 14 connections in all – it's normal to connect either all the anodes or all the cathodes together and connect them to one voltage (e.g. the supply rail), and activate the other ends of the segments as needed to form a display. Hence, you can buy either "common anode" or "common cathode" devices to specify which segment terminals are all connected together. The Mini Lab will require three seven-segment displays in due course.

The very first digital watches introduced in the mid 1970's used seven-segment light-emitting diode displays under a magnifying lens; they didn't display the time constantly (you had to press a button) and compared against modern liquid-crystal displays (I.c.d.s) they consumed batteries at a frightening rate!

## CAPACITANCE

The following topic utilises the Mini Lab L.E.D. Voltmeter, which is the first of the modules to be assembled on the Mini Lab printed circuit board. The constructional details for the L.E.D. Voltmeter are given separately. Don't worry if you don't fully understand how the module works: the main thing for now is to assemble it!

Capacitors are components which have the "capacity" (literally) to store electric charge. The circuit symbols for three common types of capacitor are shown in Fig. 2.16 – we will look at each type in due course. The symbol is actually a good representation of a capacitor's construction,

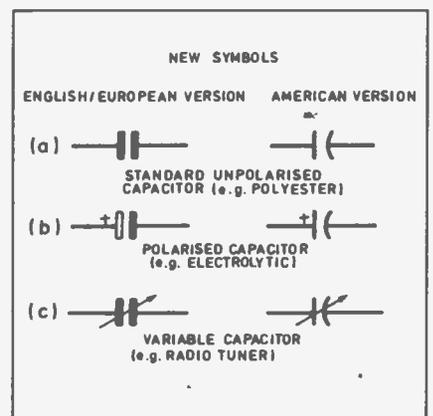


Fig. 2.16. Schematic symbols for capacitors. The American versions are also shown, as these often appear in component manufacturers' data sheets for American-made semiconductors.

# ADVANCED LEVEL

In order not to encroach upon the Physics syllabus, our treatment of Capacitance is deliberately simplified, and only outlines the principles which are of interest to the electronics enthusiast. Physics students must remember that in electronics, by convention we treat charge as positive but in real life, electrons have a *negative* charge.

The basic property of a capacitor is described by the formula  $C = Q/V$  where  $C$  is the capacitance in Farads,  $Q$  is the charge in **Coulombs** stored on the capacitor and  $V$  is the potential difference created across the device. In an ideal capacitor, when a voltage is applied to it a charge of  $+Q$  Coulombs is gained by one plate and so  $-Q$  is lost by the other.

The *net* charge is zero because in mathematical terms  $+Q$  and  $-Q$  cancel out, yet the capacitor stores *electrical energy* in the form of an electric field between the plates. In electronics, when we talk of the charge on a capacitor, we actually mean the charge of magnitude  $Q$  on just one of the plates.

When capacitors are placed in parallel, the p.d. across them remains the same (just like it does with two resistors in parallel) but the total charge is increased. If

both capacitors were identical, the total charge would be doubled. The formula for any number of capacitors in parallel is:

$$C_{\text{total}} = C1 + C2 \dots \text{etc}$$

For capacitors placed in series across a potential difference, the applied voltage is now split up across the capacitors. If two identical capacitors are placed in series across a potential difference, half of the voltage appears across each capacitor. The formula for capacitors in series is:

$$1/C_{\text{total}} = 1/C1 + 1/C2 \dots \text{etc}$$

You might have spotted that these formulae are the opposite of those for resistors in series and parallel.

Apart from their use as energy-storage devices, capacitors have another important characteristic when used with a.c. waveforms: their "resistance" can vary depending on the frequency of the applied waveform. The term "impedance" is used when we mean "a.c. resistance" and the impedance  $Z$  [in Ohms] of a capacitor of  $C$  Farads at a particular frequency  $f$  is:

$$Z = 1/(2 \pi f.C)$$

Their properties make them useful as frequency-dependent *filters* when we want to reduce (or "attenuate") a particular fre-

charge on plate "A" and a negative charge on "B" so a potential difference equal to  $V_s$  is then present across the plates. Whilst a current has not passed through the capacitor itself, there has been a flow of electrons on each side of the component, which often gives the false impression in a circuit that a current did indeed flow through the dielectric. Refer to the "Advanced Level" section for further information on capacitance.

The unit of measurement of capacitance is the Farad (symbol F). Very large capacitors (say 1 to 3F) are used as memory back-up devices in car radios or hi-fi digital tuners, but in general terms the Farad is such a large unit that we mainly use the following:

MicroFarad	$1\mu\text{F} = 10^{-6}\text{F}$
NanoFarad	$1\text{nF} = 10^{-9}\text{F}$
PicoFarad	$1\text{pF} = 10^{-12}\text{F}$

Different types of capacitor are available which utilise different dielectric materials. We generally classify a capacitor by what the dielectric is made of; probably the most common type suitable for general purpose use is the plastic *polyester* capacitor which ranges from say 10nF to 0.47 $\mu\text{F}$  or more. Tiny capacitances of, say 220pF or less might have a *polystyrene* dielectric.

For larger capacitances, e.g. 1 $\mu\text{F}$  to 4,700 $\mu\text{F}$  or more, it's generally necessary to utilise a *polarised* device called an *electrolytic capacitor*. The symbol for this is shown in Fig. 2.16(b) and it will be seen that it has a positive symbol. The real problem with electrolytics is that they have an extremely large manufacturing tolerance, typically  $-20$  per cent to  $+50$  per cent of their stated capacitance value.

## WARNING

Because of their construction and contents, it is essential that electrolytic capacitors are always connected the right way round in a circuit, otherwise damage or personal injury could result. Therefore, *always* connect the  $+$  terminal to the most positive voltage in the circuit. Never ever mess about with electrolytic capacitors.

In theory the capacitor in Fig. 2.17 would retain its charge indefinitely, once the battery was removed. However electrolytics are notoriously bad for having an exceptionally high *leakage current*, because the dielectric they use is far from being a perfect insulator. Some of the charge leaks away through the component - smaller tantalum bead capacitors are an improvement in terms of size and performance (they are also polarised like electrolytics), and for smaller capacitances polyester types are better still in terms of tolerance, leakage and physical size.

Capacitors also have a *voltage rating* and the user must always ensure that the voltage across the capacitor does not exceed this value - especially electrolytic types. It is however permissible to use a component with a *higher* rating than one specified in a circuit.

Fig. 2.18 depicts some common capacitors. Note that an electrolytic type generally has a ring crimped into its body: this signifies the positive (+) end: the "can" is always connected to the negative terminal. Again, have a look through a mail order catalogue and compare size, type and price.

## TIME CONSTANT

Fig. 2.19(a) shows a 1000 $\mu\text{F}$  16V electrolytic capacitor in series with a 27k resistor, connected across the 6V battery pack. The battery also powers the L.E.D.

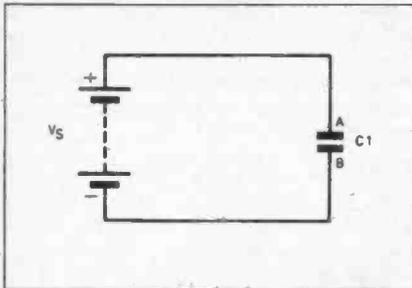
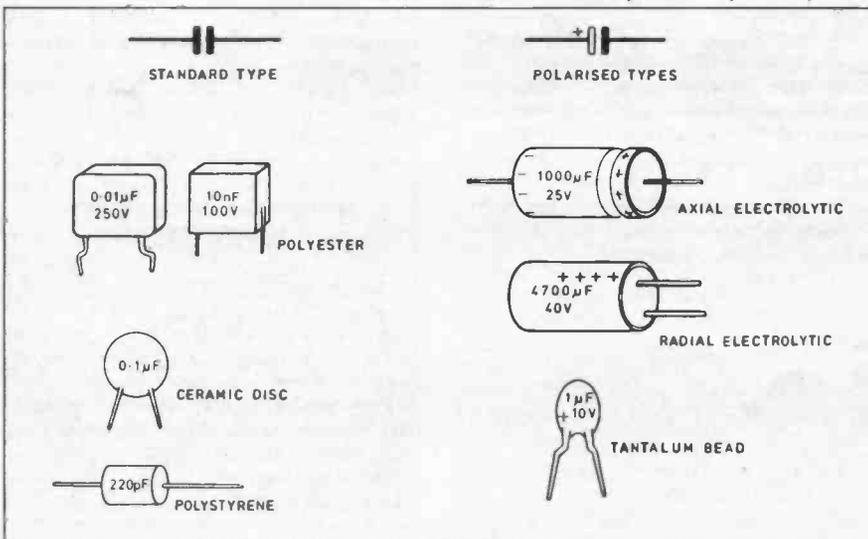


Fig. 2.17. A capacitor  $C1$  connected to a hypothetical d.c. voltage  $V_s$ .

Fig. 2.18 (below). Some typical capacitor types. It is VITAL that polarised types, especially electrolytics, are connected the right way round. Capacitors are classified by the dielectric material.



Voltmeter through a ready-made internal connection on the *Mini Lab*. Build this circuit on the Veroblock as shown in Fig. 2.19(b), ensuring the electrolytic capacitor is connected the right way round. Use the largest electrolytic capacitor you have available – certainly 470 $\mu$ F or more. The resistor should be roughly the value shown.

Before switching on the battery pack, set the range of the L.E.D. Voltmeter to 5V d.c. using its on-board selector plug. The meter is now going to be used to display the charging up action of a capacitor: we have included the 27k resistor in series to "slow down" the capacitor charging current to enable us to monitor the charging time. As the voltage across the capacitor increases due to its charging action, the L.E.D. Voltmeter display will gradually advance. You will find it best to use the "DOT" mode.

You need a stopwatch or a clock with a seconds hand. As soon as you switch on the 6V supply, start the clock and watch the display of the voltmeter: every time the l.e.d. display advances, try to note down the elapsed time in seconds. Continue until the display is static.

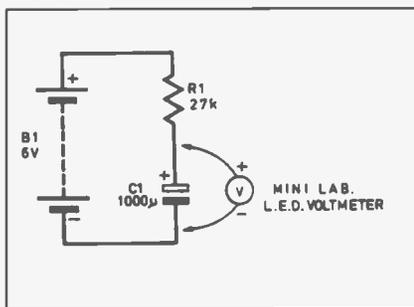


Fig. 2.19(a). A large electrolytic capacitor is charged through a resistor R1. The L.E.D. Voltmeter shows the gradual increase of potential across C1 as it charges.

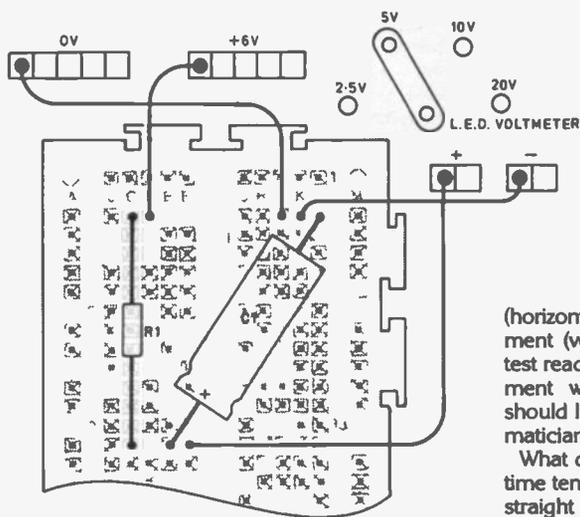


Fig. 2.19(b). Utilising the Mini Lab L.E.D. Voltmeter to monitor the rising voltage of a charging capacitor. Use any electrolytic capacitor you have available, preferably 470 $\mu$ F or more.

A great alternative suggested by our G.C.S.E. Advisor is to use a cassette recorder to tape your test readings as follows: set the tape machine recording and start the experiment. Either speak the time in seconds, or say "now" every time an l.e.d. lights, until no further change is noticed on the l.e.d. scale. Playback the tape and using the stopwatch, note down the elapsed time.

You will finish with a string of numbers,

each one representing the elapsed time when the next l.e.d. in the display lights. With the L.E.D. Voltmeter module set to 5V, each l.e.d. step is equivalent to 250mV (5V divided by 20 l.e.d.s), so the time taken for the next l.e.d. to light represents an increase of 250mV across the charging capacitor.

If you want to start the experiment again, you must ensure that the capacitor is fully discharged. In this instance it's acceptable to short it out with a temporary link wire – or why not use one of the convenient n/o push switches? In general you should not do this with larger capacitors or when higher voltages exist across a capacitor, as electrolytics are high-energy devices.

Using your results, plot a simple graph of voltage against elapsed time for your charging capacitor. Just plot the elapsed time

to fully charge or discharge a capacitor through a resistor.

Whenever you see "RC" mentioned in a formula or circuit containing a resistor and capacitor, you know that somewhere the capacitor is charging and/or discharging through that resistor. It's actually possible to make a capacitor charge up linearly through a "constant current source" – useful if you want to generate a uniform "ramp" waveform instead of a curve.

You can also experiment by discharging a capacitor through a resistor. Use one of the push switches to charge up an electrolytic capacitor from the 6V battery, then release the button to start the experiment and discharge it through the resistor, monitoring the voltage across the RC combination with the L.E.D. Voltmeter. Fig. 2.21 shows a

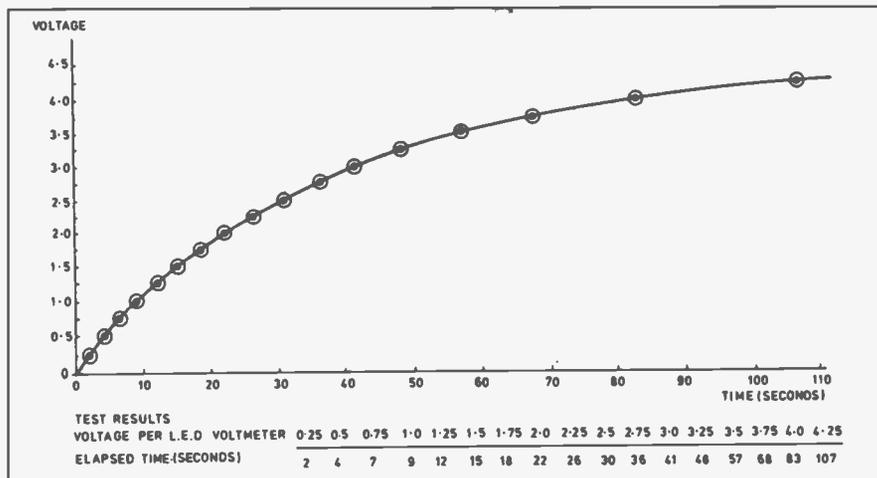


Fig. 2.20. Plot of an exponential charging curve, using the results obtained from the circuit of Fig. 2.19.

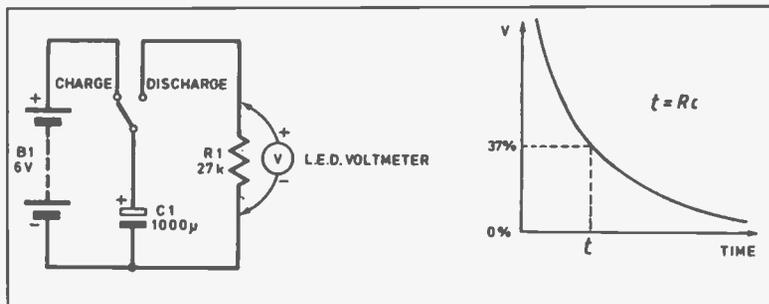


Fig. 2.21. Suggested circuit to observe the discharge action of a capacitor, use the L.E.D. Voltmeter to obtain results as before, then plot a discharge curve.

(horizontally) against each 250mV increment (vertically). Fig. 2.20 shows our own test readings when we performed the experiment with our *Mini Lab* – your result should look roughly the same, what mathematicians call an "exponential curve".

What does the graph show? The charging time tends to tail off in a curve rather than a straight line. It takes progressively longer for each 250mV step to be reached, so a capacitor does not charge in a linear fashion through a resistor. The Time Constant of a resistor/capacitor network (like the one we've just used), is equal to:

$$\text{RC Time Constant (seconds)} = R (\text{ohms}) \times C (\text{Farads})$$

The RC time constant for the 27k/1000 $\mu$ F network works out as 27 seconds. Introducing some useful rules of thumb:-

RC = time taken (seconds) for a capacitor to charge up to 63 per cent of its final value, or discharge down through a resistor to 37 per cent of its final value.

5RC = approximately the time it takes

simple circuit to do this – see if you can work out the intertwining yourself, then take your results (the l.e.d. display moves backwards this time) and plot a discharge graph. Experiment with different values of resistors and capacitors.

Next month, we take a look at power supplies where your *Mini Lab* becomes mains powered – it will then really look like it means business! We also examine the extremely important area of safety in electronics, so there's something for everyone!

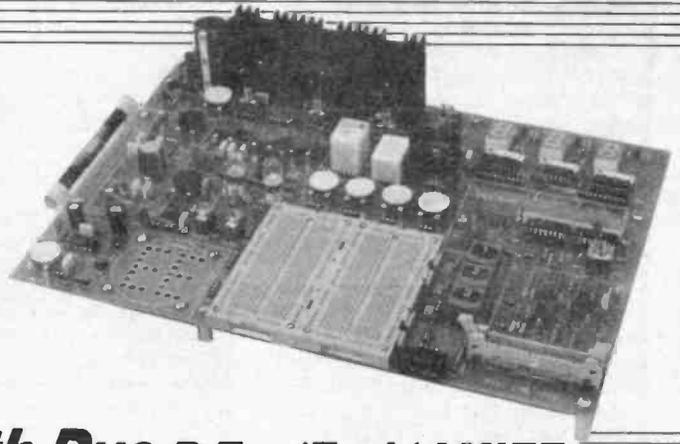
### MINI LAB KITS

Magenta Electronics have advised us that they are now supplying kits of parts for the *Teach-In '93* series. They can supply kits for each part (part 1 with the p.c.b.) or a complete kit for the series.

Greenwell Electronics are also offering the two breadboards for the *Mini Lab* at a special price with a free soldering iron.

See the relevant advertisement and *Shop Talk* for more details.

## MINI LAB



**Alan Winstanley & Keith Dye B.Eng(Tech)AMIEE**

*The Everyday with Practical Electronics Mini Lab has been created to accompany Teach-In '93, and enables the reader to assemble demonstration circuits by following the clear instructions and diagrams contained in the main text, with every chance of it working first time. The Mini Lab is an exciting learning aid which brings electronics to life in an enjoyable and interesting way: you will both see, and hear, the electron in action.*

**T**HE first module to be constructed on your *Mini Lab* printed circuit board (described in detail last month) is a complete voltmeter which uses a row of 20 light-emitting diodes (l.e.d.s) for a scale. It reads up to 2.5V, 5V, 10V or 20V full-scale deflection (f.s.d.). The circuit diagram is shown in Fig. 1: if you are a novice in electronics, don't worry if the circuit diagram doesn't make too much sense – the important thing for now is to build the module successfully.

Space does not permit us to examine the mode of operation in great detail, though the components used will be familiar to experienced readers. Two LM3914N integrated circuits are used, and are connected to form a 20-step l.e.d. bargraph. The 1.25V internal reference voltages of both devices are connected in series to form an accurate 2.5V reference. The LM3914 i.c.s then compare the input voltage against this and illuminate (or "enable") the appropriate number of light-emitting diodes.

An attenuator system of close-tolerance resistors (R1 to R6) is selectable so that the f.s.d. of the module can be determined by linking with a shorting plug the appropriate sockets mounted on the *Mini Lab*. The values have been selected so that when the maximum f.s.d. voltage is presented to the L.E.D. Voltmeter, only 2.5V maximum actually appears at the inputs of IC1 and IC2. On the 2.5V setting, the input voltage is applied directly to the inputs of the l.e.d. drivers.

The meter has a sensitivity of 80,000 o.p.v. (ohms per volt) which is much better than many moving coil multimeters. The i.c.s are protected against overloads up to 35V and reverse input voltages down to -35V d.c. For further information, the manufacturers (National Semiconductor) of the LM3914 produce an Informative Data Sheet which should be consulted.

S1 is a p.c.b.-mounted switch which controls the "mode" pins of the i.c.s and selects either moving dot or bargraph mode – the former is preferable for battery operation in order to minimise battery drain.

The design will be more than accurate

enough for our purposes, and it also adds interest to demonstrations whilst freeing your multimeter to make other measurements. A printed scale is incorporated in the silk-screen printing of the printed circuit board, so you will have no problems interpreting readings.

### CONSTRUCTION

The L.E.D. Voltmeter is assembled on the relevant section of the *Mini Lab* printed circuit board, which is fully silk-screen printed to help with component positioning. Solder in all parts as shown in Fig. 2 starting with the

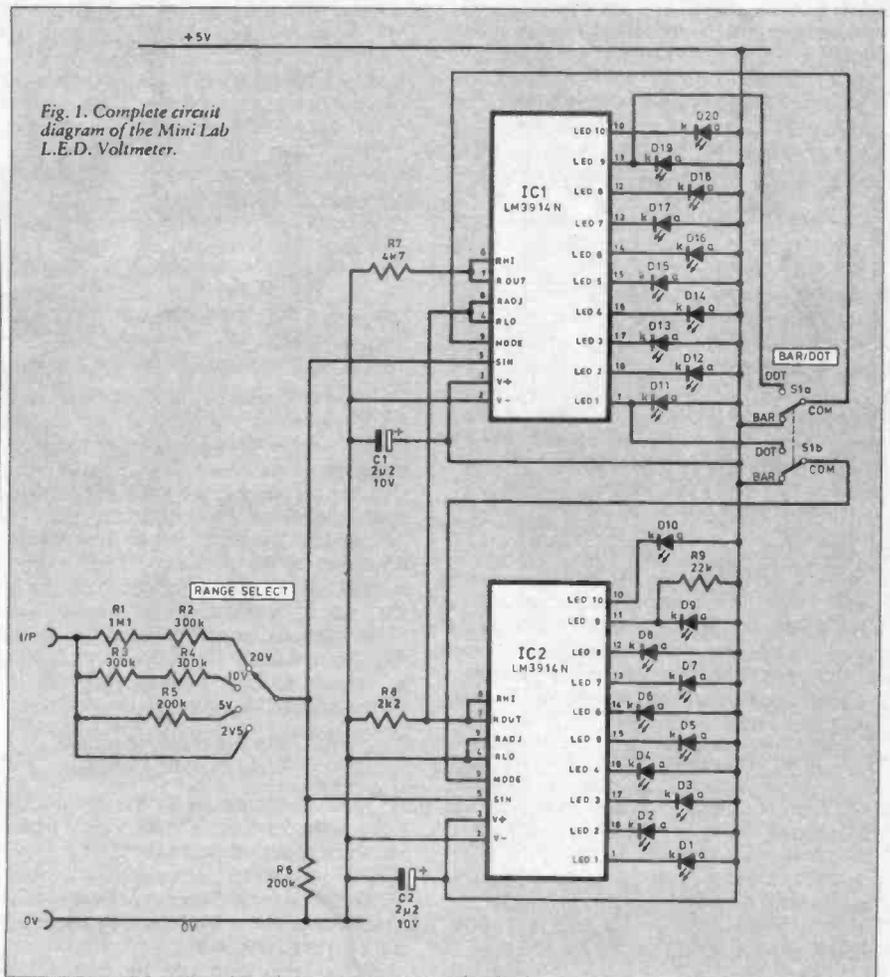


Fig. 1. Complete circuit diagram of the Mini Lab L.E.D. Voltmeter.

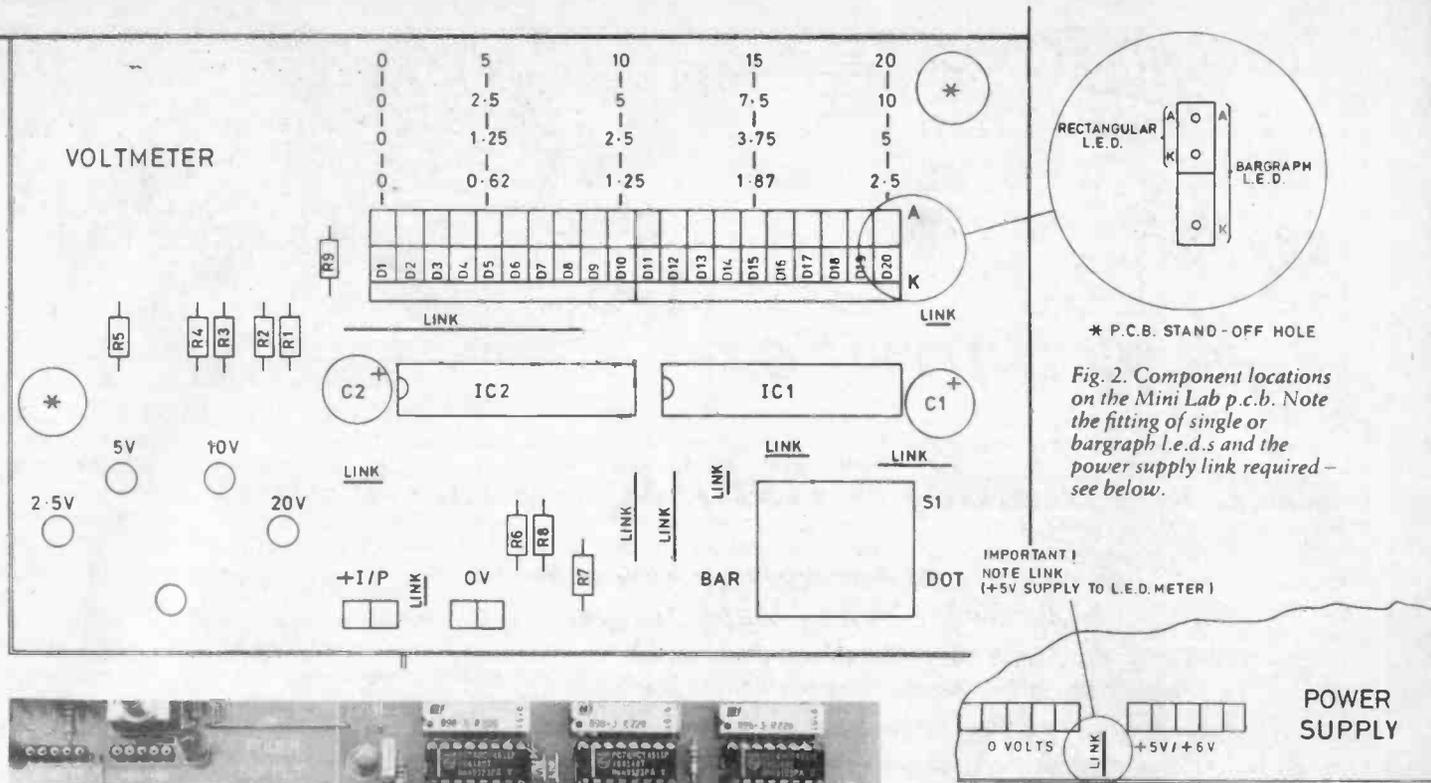
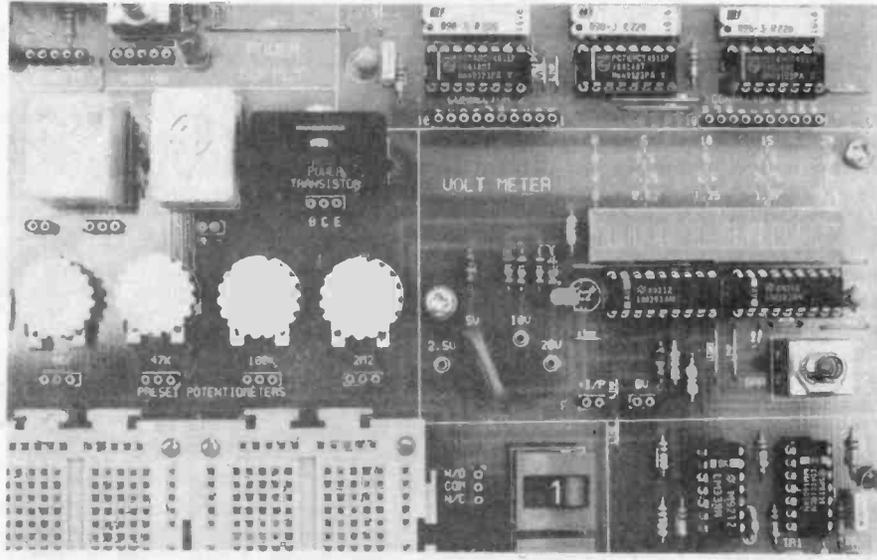
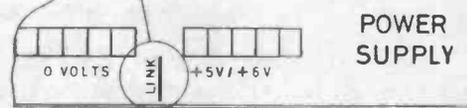


Fig. 2. Component locations on the Mini Lab p.c.b. Note the fitting of single or bargraph i.e.d.s and the power supply link required - see below.

IMPORTANT!  
NOTE LINK (+5V SUPPLY TO L.E.D. METER)



To test your completed L.E.D. Voltmeter module, choose 10V f.s.d. and switch on the 6V battery pack. Then use one of the on-board preset resistors (e.g. 47k) to apply a varying voltage to the input terminals of the Voltmeter (see Fig. 2.3 in the main tutorial). The relevant number of light-emitting diodes will illuminate when you rotate the preset control. Also test the Dot/ Bar selector, after which your L.E.D. Voltmeter is ready for use.

Next Month: Power supply.

## Mini Lab Components

- Resistors**
- R1 1M1 1%
  - R2 to R4 300k 1% (3 off)
  - R5, R6 200k 1% (2 off)
  - R7 4k7 -
  - R8 2k2 -
  - R9 22k ✓
- All 0.25W 5% except where shown

See  
**SHOP  
TALK**  
Page

- Capacitors**
- C1, C2 2μ2 35V tantalum bead (2 off)

- Semiconductors**
- IC1, IC2 LM3914N i.e.d. bargraph driver i.c. (2 off)
  - D1 to D20 EITHER: 10 segment i.e.d. bargraph module (2 off),  
OR: D1 to D4, D6 to D9, D11 to D14, D16 to D19: Red rectangular i.e.d. (16 off)  
D5, D10, D15, D20: Green rectangular i.e.d. (4 off)

- Miscellaneous**
- S1 d.p.d.t. toggle switch (C&K Z200 series)
- P.C.B. mounting "jacks" (5 off); p.c.b. 0.5 inch insulated shorting link plug; turned pin s.i.l. sockets (4 off); 18-pln d.i.l. i.c. sockets (2 off); s.i.l. sockets for i.e.d. arrays, if required (4 off); wire, solder etc.
- Note** as detailed last month the Mini Lab p.c.b. is available from the EPE PCB Service, order code MINI LAB.

## IN USE

The Voltmeter is powered by the 6V battery pack during the early stages of *Teach-In* after which mains operation is introduced and it is then powered from the 5V rail of the Mains Power Supply. External battery or Mains operation is selected with an on-board selector plug near the Power Supply section of the p.c.b., with the L.E.D. Voltmeter being powered from the appropriate source.

A set of turned-pin sockets are used for the + and - inputs to the L.E.D. Voltmeter, and these can quickly be connected using jumper wires to the Veroblock breadboard when required, in order to make a measurement. The 0V rail connects by the p.c.b. to the other modules (e.g. the Mains Power Supply), so often it will not be necessary to link a Jumper wire to the 0V input socket - this connection is ready-made by the board.

Choose the required range of the Voltmeter using the selector plug to link the two appropriate sockets on the printed circuit board. In the demonstration circuits of *Teach-In*, you are always advised which f.s.d. is required for any experiments. Generally, you can also select either dot or bargraph mode, whichever you find easiest to read.

jumper link wires which can be made from tinned copper wire.

It is best to use i.c. sockets if you are unaccustomed to soldering integrated circuits, so that thermal damage is avoided. In any event, do not overheat any joints or you may damage the p.c.b. and lift off the copper track. A fine-tipped soldering iron is best - take your time! Then insert the two integrated circuits carefully into their sockets, observing polarity or you will damage the devices.

Separate i.e.d.s can be used for the indicators, and to help when reading the scale, each 5th i.e.d. can be green coloured, the remainder being standard red devices. Rectangular types are best, and you must ensure that the i.e.d.s are soldered in the right way round, fitting them so that they are flush against the board.

Alternatively, a much neater effect is produced by utilising two 10-segment i.e.d. bargraph modules, although unlike separate i.e.d.s (the p.c.b. is designed to accept either) it will not be possible to include different coloured bars as "milestone" indicators. Either solder the bargraphs directly to the board or use dual-in-line sockets to carry them.

**Price** **£13**  
approx

# FOX REPORT

by Barry Fox



## On The Cards

Looking at the piles of discarded payphone cards littering the street near a public call box recently, someone asked why the cards cannot be re-cycled and whether any other country re-cycles them.

British Telecom now has 100,000 payphones, of which 21,000 are card phones. BT sells 25 million cards a year and they cannot be re-cycled because the credit units are stored in the card as an optical hologram. The card reader literally burns along a track as it uses up the credits.

If you look carefully at the top (green) surface of a BT phone card you will see a strip, stretching the length of the card with parallel markings, like rungs of a ladder. These are the credit units.

The object of using this type of card (made by Continental company Landis and Gyr) is to prevent fraud. The disadvantage of BT's system is that if the card surface becomes scuffed, the optical reader cannot read the track. Tip: Keep BT cards in a plastic holder.

## Folklore

There have been all kinds of rumours and folklore stories about clever ways of rejuvenating BT cards. One theory was that you could put the cards in a freezer overnight to restore their value. Another was that you could cover the card surface with nail varnish to stop the laser destroying the track.

Needless to say British Telecom describes both tricks as "complete nonsense". But I have not yet met anyone with first-hand experience of reviving a card and I am not going to risk trying myself because it would of course be fraud and illegal to do so. Common sense suggests it is indeed impossible.

In theory cards, like credit cards, which have a magnetic strip can be recycled. But in practice the strips are unlikely to be sufficiently robust for reused cards to be reliable.

In any case the criminal fraternity has come up with several ways of copying the information on a magnetic strip card, either by heat transfer or the kind of card readers and encoders which large companies use for their own security cards. This is why magnetic cash cards will only work when the user keys in a secret PIN, or personal identification number.

It thus comes as quite a shock to find that Mercury, which went into card phones later than BT, adopted a magnetic card. Although there is no obvious magnetic strip on the surface, magnetic material hidden under the surface works just like the surface strip on a credit card.

How then does Mercury stop people cloning cards? No company wants to talk about security matters, but those in the know believe that Mercury uses a central monitoring system.

All the payphones are connected to a computer, which cross checks the serial number of the payphone card with a record of its last use. This is what causes the delay when some calls are made through a Mercury payphone, even though dialling is by speedy touch tone. The computer is checking its records.

If the computer finds that the card has suddenly accumulated value, instead of lost value, then the system rejects it.

The disadvantage of any magnetic system is that data can be erased if

the card is brought near a powerful magnetic field. There was for years a folklore story that sealskin wallets erased credit cards. In reality it was the magnetic catch on the wallet which was erasing the data.

## Smartcards

In France telephone cards are Smartcards, credit cards with a built-in computer chip and memory. Technically this is expensive overkill, but France has for years been subsidizing the use of Smartcards in an effort to make the French invention a way of life. Although it would be easy to re-cycle Smartcards, and make good financial sense, I checked with France Telecom, and they tell me that cards are not re-cycled.

The Smartcard system may be hi-tech, but it falls flat on its face in practice. Recently I tried to make a call from a French airport. I had no card and there was nowhere to buy one. I tried for half an hour to make a reverse-charges "collect" call, but never did get to speak to a human operator.

## Charge It

For the first time British Telecom is feeling the real bite of competition from Mercury. If you are moving house, or planning to install a second phone line, or have a fault on your line, it pays to know how to turn this to your advantage.

Recently I needed a second line installed in a London flat. BT charged £163.75 (including VAT) to do the work, even though there were plenty of BT lines already running into the block.

Just after I paid, I received a press release from British Telecom quoting Michael Hopher, BT's Group Managing Director. He was puffing about BT "making it easier to get on the phone". Hopher quoted BT's line connection charge of £163.75 as a "maximum". "Very few customers are asked to pay the maximum connection charge" he went on.

I asked BT why I had been charged the "maximum" when there were already plenty of spare lines and the extra work done was minimal. Various BT people then came up with various excuses which boiled down to one practical fact. You only avoid paying BT's "maximum" connection charge if a full telephone service

has previously been installed and disconnected, so that the only work involved is re-connection at the exchange.

While I was wasting time and money writing letters to BT, something else happened. The local cable TV service "passed" the premises, and offered me either a cable TV or Mercury telephone line connection, or both, for a flat fee of £30. Also, the cable company reminded, long distance calls on Mercury are cheaper.

## Sunday Special

Incidentally this is doubtless why BT is now advertising the "Sunday Special", long distance calls at local rate on Sunday afternoons in November and December. Mercury and the cable companies now have to change their publicity adverts to "cheaper calls except on Sundays during November and December", which is hardly a catchy slogan.

BT's "present" also hits everyone with one of Mercury's "smart boxes", which supposedly route calls by the cheapest route. On Sundays this winter they no longer will. If you have one of these boxes you may well save money by disconnecting it on Sundays.

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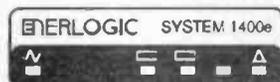
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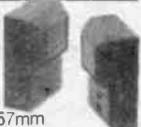
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# COMBINATION SWITCH

MAX HORSEY

A versatile project which can be used as a home security lock or alarm control switch. Personal "password" is held in memory and may be reprogrammed at any time

**T**HE Combination Switch or Combination Lock is a versatile project which may be used to unlock a door or switch off an alarm etc. The user's password is held in memory, which is protected by a back-up battery in case of power cuts.

The password may be re-programmed by the user at any time, and up to 12 digits can be stored. If the wrong password is entered more than 3 times, a siren output is activated for a minute or so.

There is also provision for a switch input so that the siren sounds if the switch is closed before the correct password has been entered. This may be used to detect a door or window being forced. The project may be fully integrated with an alarm system if required.

The outputs are via transistors and a relay as follows:

**Solenoid Output** (up to 3A) – to unlock a door.

**Relay** – to turn on or off an alarm or any other equipment.

**Siren** (up to 3A) – which sounds for a limited time if the wrong password is entered three times, or if the door is forced.

**Small buzzer** – to indicate correct operation.

The constructor may select one or more of these output options.

## OPTIONAL BACK-UP BATTERY

A power cut would cause the loss of the user's password in memory, resulting in the entry password defaulting to 0. This problem may be overcome by the use of a back-up battery. It is not intended that this battery should allow the project to function normally (a much larger 12V battery across the whole circuit would be needed), but the battery suggested will provide enough power to keep the memory of the i.c. active.

The supply suggested is three AA cells, which may be non-rechargeable (dry) cells, or re-chargeable Ni-Cad cells. In the latter case, resistor  $R_X$  should be fitted to trickle charge the cells.

**Warning:** Do not fit  $R_X$  if non-rechargeable cells are employed.

## HOW IT WORKS

The essential features of the Combination Switch are shown in Fig. 1. At the heart of the circuit is IC1, a coded "security lock" chip type UA3730. Data is entered with an inexpensive matrix keypad, and all the complex functions are performed by IC1. A 5V supply is required, but since most solenoids and sirens require 12V, a regulator (IC2) is used to power IC1.

The voltage at pin 14 allows a small piezo buzzer to sound each time a key is pressed, and to indicate other functions. When the correct password is entered pin 17 goes low (i.e. 0V) for two seconds. The current available is amplified by a Darlington pair.

Also, each time the correct password is entered pin 16 toggles – in other words it changes from low to high, or high to low. A single transistor and relay are used here to switch on or off an alarm etc.

The output at pin 15 goes low for one minute to activate the siren via a Darlington amplifier, if the wrong password is entered more than three times.

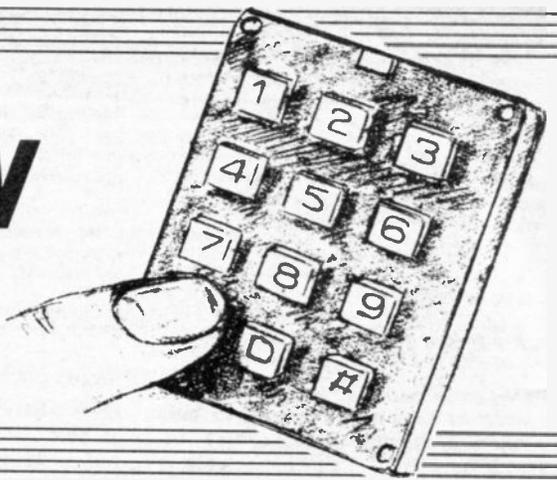
## OPERATING THE SYSTEM

Each time a key is pressed, the small buzzer WD1 will sound. The password is entered by pressing the correct keys, followed by the button marked "#". When first switched on the password is 0.

The password may be re-programmed as follows:

- 1 Enter the existing password. (Do not press "#")
- 2 Press \*
- 3 Key in the new password – any combination up to 12 digits.
- 4 Press \*

The "security" IC1 is now programmed with your new password.



## CIRCUIT DESCRIPTION

The complete circuit diagram for the Combination Switch is shown in Fig. 2. The 12V power supply is delivered via diode D9 which prevents damage if the supply is accidentally connected the wrong way round. Capacitor C4 decouples the supply, and 12V is available for the transistors, solenoid, relay and siren.

The 12V supply is reduced to 5V by the regulator IC2. This is a 78L05 i.c. which provides a very stable 5V output at up to 100 mA. Diode D2 prevents current returning to the regulator from the back-up battery in the event of a power failure, and capacitor C3 provides further supply decoupling.

The 5V supply from IC2 is applied to pin 9 of IC1. Diode D1 allows current to flow from the back-up battery to IC1 during a power failure, but prevents current flowing into the battery from the circuit during normal operation.

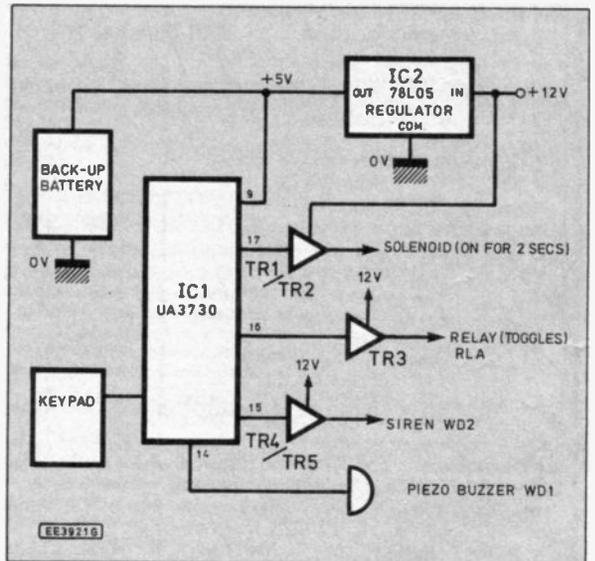


Fig. 1. Block diagram for the Combination Switch.

Optional resistor  $R_X$  should be fitted *only* if re-chargeable cells are employed.  $R_X$  allows a trickle charging current to flow when power is restored.

The suggest value of 100 ohms for  $R_X$  provides a trickle charge current of about 3mA. If the power supply is frequently disconnected,  $R_X$  could be reduced to 56 ohms. Note that  $R_X$  **MUST NOT** be included if dry-cells are used, because there is a risk of the cells exploding.

## INPUTS

The keypad is connected to IC1 via pins 5, 6, 7, 8, 10 and 11. The matrix arrangement of the switch contacts reduces the number of connecting wires and pins required.

Switch S1 is optional, and when closed, causes the siren WD2 to be triggered. S1 could be used on a door or window or as an under-carpet pressure mat (or all three if wired in parallel), to trigger the siren if entry is forced without first entering the correct password.

## OUTPUTS

Output pin 17 is "active low". In other words the pin is normally open circuit, but after the correct password is keyed in, pin 17 is internally connected to 0V. Pin 17 is connected via current limiting resistor R3 to the base of TR1.

Transistors TR1 and TR2 form a *pn*p Darlington pair. When pin 17 is open circuit, no current flows from the base of TR1, and both transistors are turned off.

When pin 17 switches to 0V, current flows from the base of TR1, and both TR1 and TR2 turn on.

Assuming that the gain of TR1 is about 150, and the gain of the power transistor TR2 is 50, the total gain will be  $150 \times 50$ , i.e. 7500. Thus a current of 3A supplied to the solenoid requires only 0.4mA from the base of TR1. The circuit more than meets this requirement.

As mentioned earlier, output pin 16 toggles whenever the correct password is entered. In other words pin 16 changes from open circuit to low, then changes back again at the next correct entry.

The use of a relay here ensures that the project can be used to control any type of device from an alarm system to unauthorised use of a Hi-Fi unit! Transistor

TR3 is employed to increase the current available from pin 16 to that required by the relay coil.

The output at pin 15 switches from open circuit to low when an incorrect password is entered more than three times. It is used to drive a siren via the Darlington pair TR4 and TR5. This operates in exactly the same way as with pin 17.

Diodes D6, D7 and D8 remove any high voltage spikes which tend to be produced by electro-magnetic devices such as solenoids and relays.



## COMPONENTS

### Resistors

R1	12k
R2	4k7
R3-R5	2k2 (3 off)
Rx	100 (see text)
All 0.25W 5% carbon film	

See  
**SHOP  
TALK**  
Page

### Capacitors

C1	270p ceramic plate
C2	1µ radial elect., 63V
C3	0µ1, disc ceramic
C4	1000µ, axial electrolytic 16V or more

### Semiconductors

D1, D2,	
D9	1N4001 1A 50V rec. diode (3 off)
D3-D8	1N4148 signal diode (6 off)
IC1	UA3730 combination switch i.c.
IC2	78L05 5V 100mA voltage regulator
TR1, TR3,	
TR4	BC214L silicon <i>pn</i> p transistor (3 off)
TR2, TR5	TIP42A silicon <i>pn</i> p power transistor (2 off)

### Miscellaneous

WD1	Piezoelectric sounder
WD2	12V high power siren
RLA	12V 180 ohm coil or more, with changeover contacts (see text)
S1	Door or window micro- switch (see text)

Keypad, 12-key 4 by 3 X/Y matrix; plastics case (main unit), size 150mm x 100mm x 60mm; small plastic case (relay housing), size 76mm x 50mm x 38mm; 12V solenoid or lock mechanism (see text); 18-pin d.i.l. socket; self-adhesive p.c.b. stand-off (2 off); back-up battery holder (for 3 AA cells), with battery clip; dry or rechargeable AA battery (3 off) - see text; multi-coloured connecting wire; 7-way "rainbow" ribbon cable (see text); double-sided sticky pads; solder, etc.

Printed circuit board available from PCB Service, code 812.

Note: you may not require all these parts - see "Options" section.

Approx cost  
guidance only

**£32**

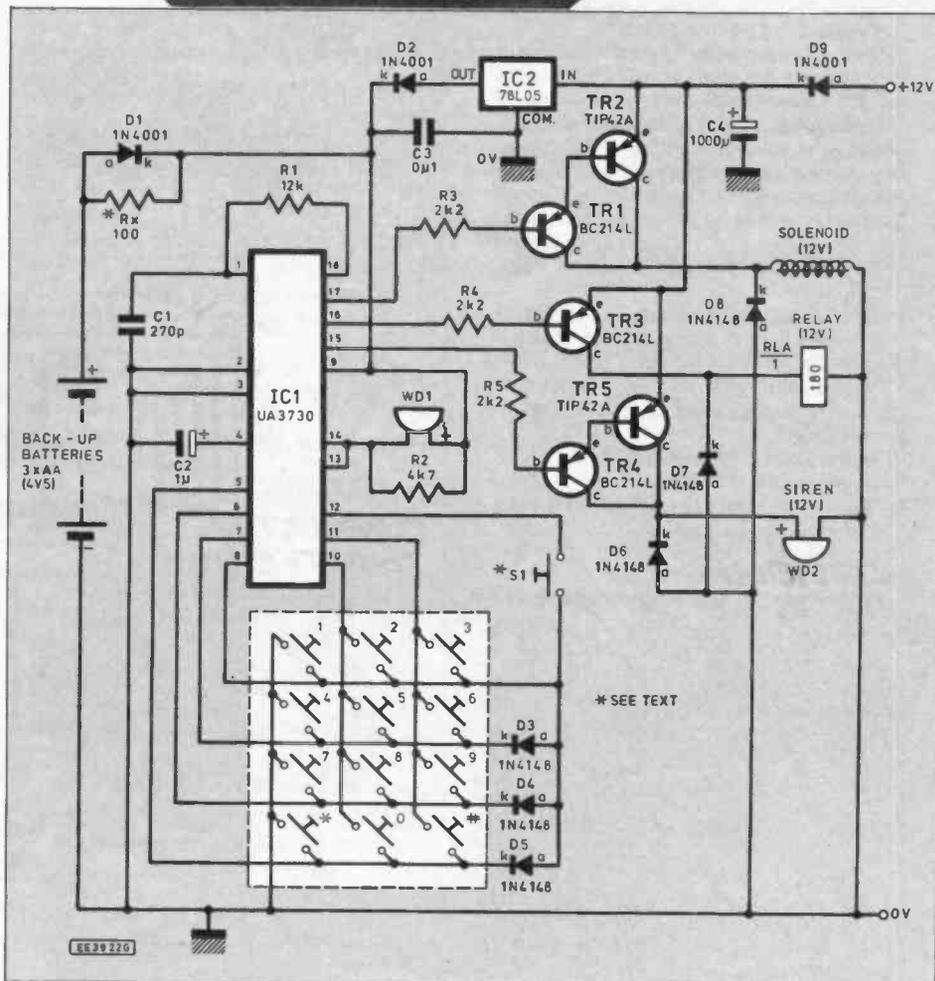


Fig. 2. Complete circuit diagram for the Combination Switch.

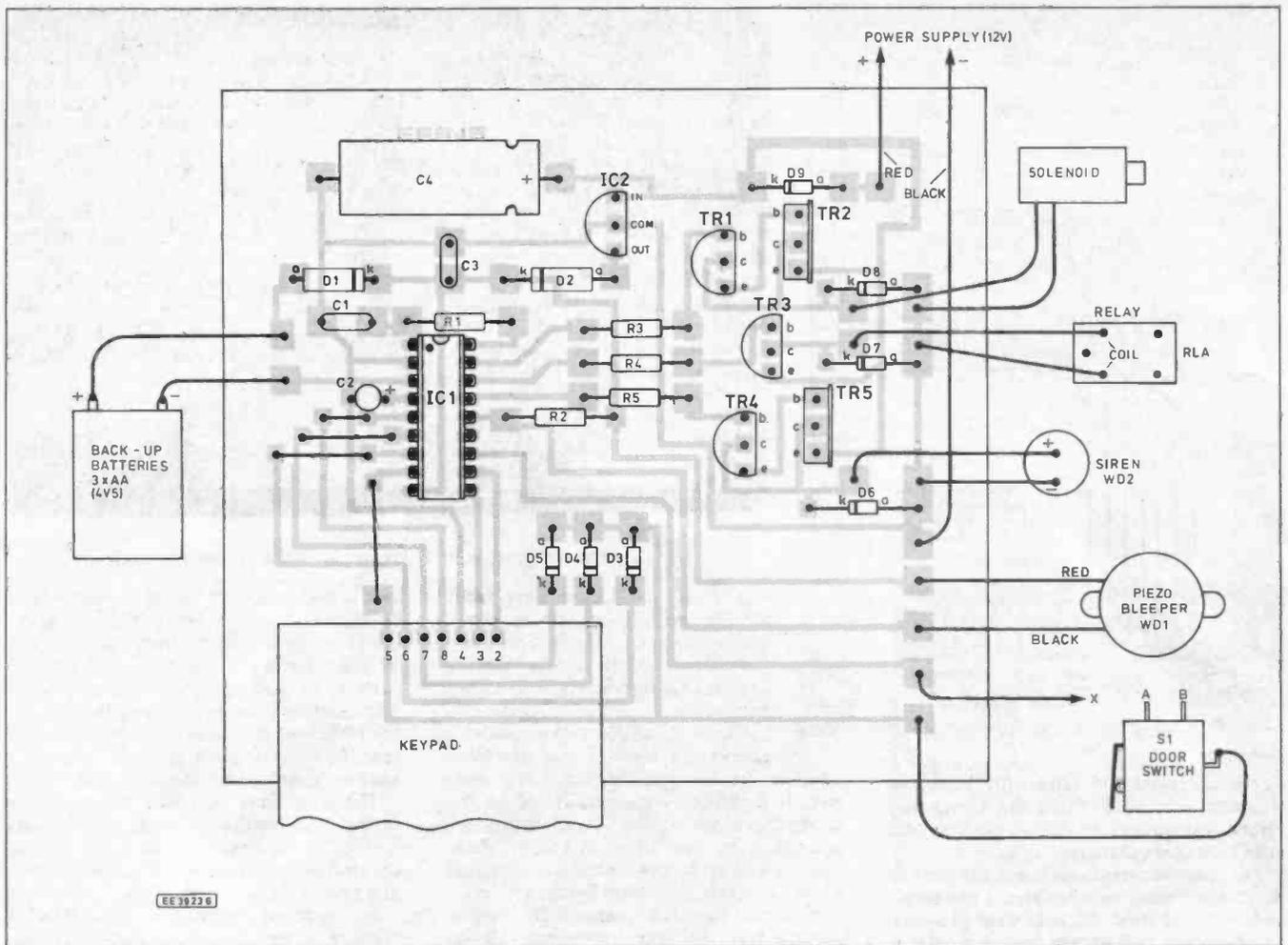


Fig. 3a. Printed circuit board component layout and wiring to off-board components. Wiring from the keypad is shown in Fig. 4.

## OPTIONS

The solenoid, relay and siren are all optional, and you may not require all three. If you omit any of these devices, certain other components can also be left out as follows:

- Solenoid not fitted: Omit R3, TR1, TR2, D8
- Relay not fitted: Omit R4, TR3, D7
- Siren not fitted: Omit R5, TR4, TR5, D6

## CONSTRUCTION

All the discrete components for the Combination Switch are mounted on a small printed circuit board (p.c.b.) which is fixed to the rear of the case lid. The memory back-up battery, sounders and relay are mounted on the base of the case.

The printed circuit board component layout and full size copper foil master pattern is shown in Fig. 3. This board is available from the PCB Service, code 812.

Before soldering any components on the p.c.b., the final layout inside the case must be decided on. Remember, the p.c.b. is mounted on the back of the case lid and should be taken into account when setting out the off-board components, making sure they do not "short-out" on the p.c.b.

The best positioning of components can be found by laying the battery back-up holder in one corner of the case, the keyboard buzzer and siren down one side and the relay in the remaining corner - see photograph. If the mains supply is to be controlled by the relay, it should be covered or housed inside an additional

small plastic case (inside the main case) for added safety.

Begin construction of the board with the i.c. socket, checking that the notch is towards the top of the p.c.b. Do not fit IC1 into the socket at this stage.

Next solder in the wire links using bare solid wire. Fit the resistors, with a careful check on their values. Diodes D1 to D9 may now be fitted, checking whether they are 1N4001 or 1N4148 types, and also checking that the black or silver cathode

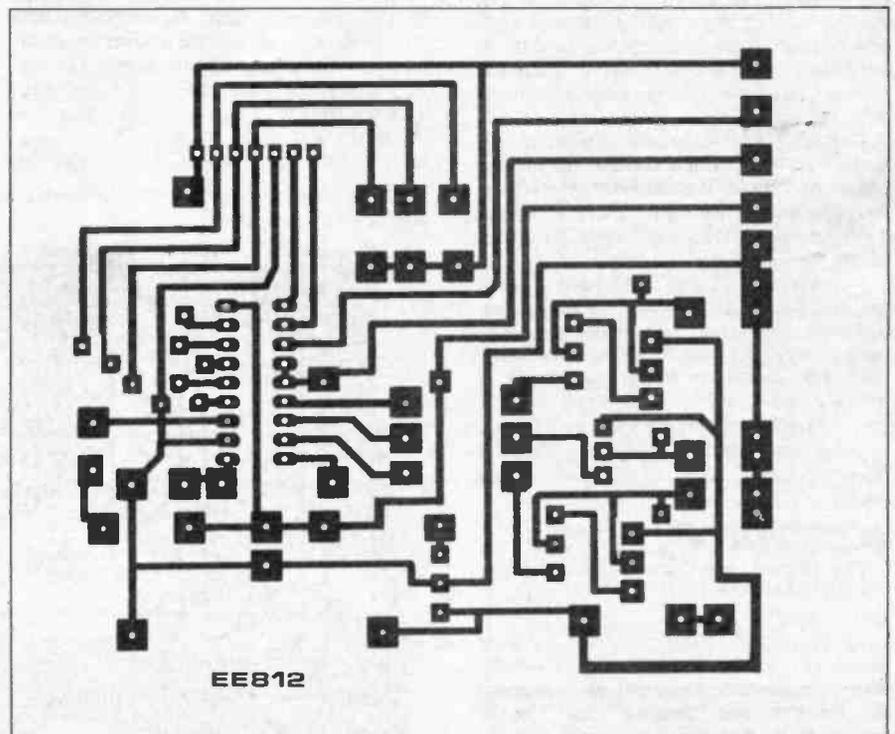


Fig. 3b. Full size copper foil master pattern for the Combination Switch.

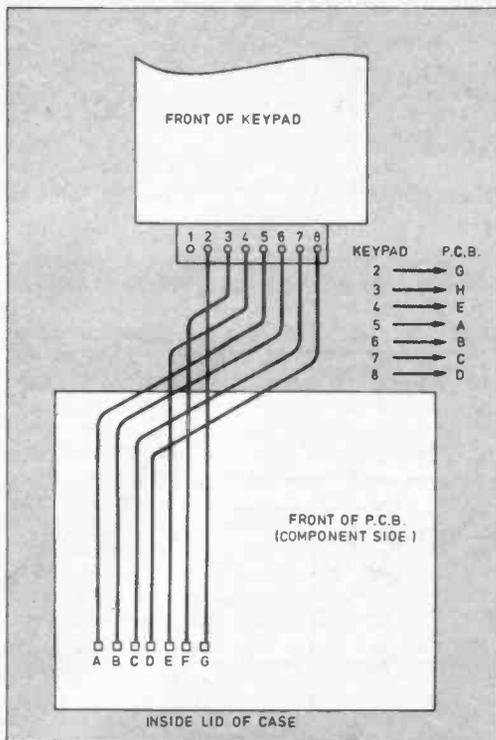


Fig. 4. Wiring from the keypad to the circuit board.

(k) bands at one end agree with the layout diagram. Any diode fitted the wrong way round will prevent the circuit working and may even cause damage.

No provision has been made for resistor  $R_X$ , since it must *only* be fitted if re-chargeable cells are used. If pads were provided for  $R_X$ , there is a danger that it would be fitted accidentally, and if dry cells are then employed there is a risk of the cells exploding. Consequently,  $R_X$  must be soldered to the pads used for diode D1 on the *copper* side of the p.c.b. if re-chargeable cells are to be used.

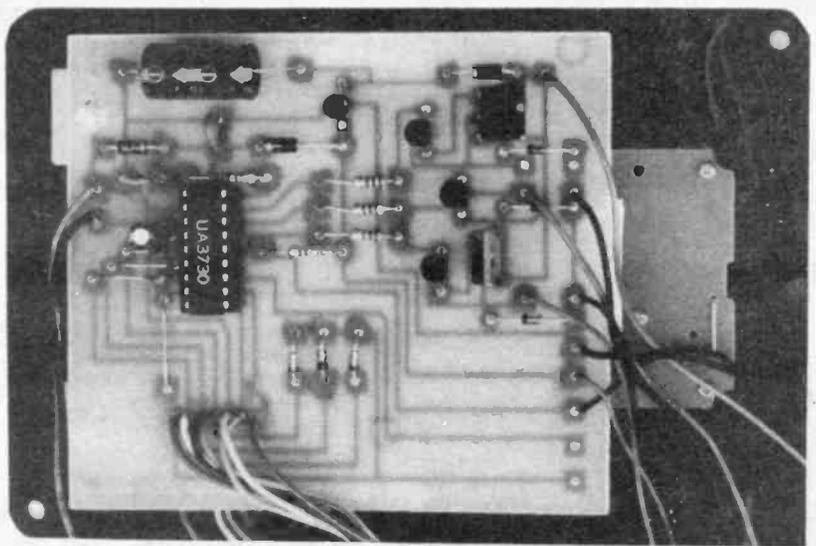
The small capacitors, C1 and C3, may be fitted either way round, but the larger electrolytic capacitors must be fitted with the correct polarity, noting carefully that the negative side is usually marked with small minus signs and arrows. Just to be confusing, the *positive* lead of axial (lie down) capacitors nearly always emerges from the *black* end.

Regulator IC2 looks like a transistor, so check the code, which should include the characters 78L05. It must be fitted with the flat side facing the right. TR1, TR3 and TR4 are also fitted with their flat sides facing the right.

Transistors TR2 and TR5 have a metal tab, and, looking from above, this tab must be on the right-hand side. A heatsink may be attached to these tabs if required; however, this was not found necessary in the prototype circuit. The transistor body can become surprisingly hot before any damage results, and in this circuit both the current required and the "on time" is quite small.

## KEYPAD

The keypad may now be fitted on the outside of the case lid. But before doing this, suitable lengths of multi-coloured leads or 7-way "rainbow" ribbon cable should be soldered to the key contacts. These leads/cable should be threaded through a suitable "window" cutout in the lid so that the opposite ends can be soldered to the board, as shown in Fig. 4.



The completed circuit board mounted on the rear of the case lid.

If using leads, use long wires (which can be bunched up later), and a variety of colours if possible. Note that from the underside of the keypad, the numbers 1, 2, etc. read backwards.

Solder in all the other insulated connecting leads noting the convention: red for positive, black for negative. It is difficult to generalise about the connections for the solenoid etc. since individual circumstances will vary. These devices could be connected via terminal blocks to retain maximum flexibility.

Note that the small buzzer WD1 *must* be a *piezo* type; an ordinary buzzer will not work. Wires may be connected to the back-up battery holder, but do not load the cells at this stage.

Finally, IC1 may now be inserted into its socket, checking that the notch or dot is towards the top of the p.c.b., regardless of which way round you fitted the socket.

## CASE

Assuming that the circuit is to be housed in a conventional case, the keypad is best connected to the p.c.b. via flexible insulated leads. The photographs show how the battery box, relay (and small plastic case if required), siren and buzzer are positioned in the main part of the case, with the p.c.b. mounted on the inside of the lid. The key

pad is fixed with self-adhesive sticky pads to the outside of the lid or if the cutout in the case matches the keypad exactly, it may be glued in place.

Begin by drilling all the necessary holes, and a slot for the keypad connecting pins. In the prototype the siren was fitted inside the case, but in practice it might be connected via long leads to the side of a wall etc.

The relay may be glued into position. If the relay switch contacts are to control mains equipment, take great care to prevent the mains coming into contact with any part of the low voltage wiring.

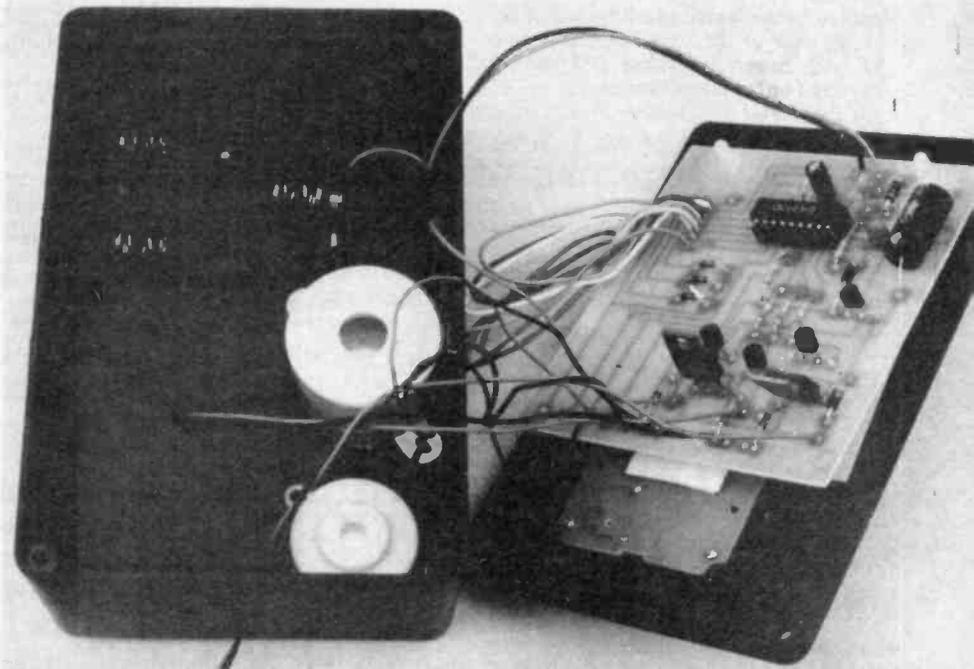
A wise safety precaution is to house the relay in a separate small plastic box inside the main case. This guards against electric shock in the event of internal parts such as batteries becoming loose and falling against live contacts.

The battery box, siren and buzzer may be secured with glue or double sided sticky pads. The p.c.b. may be fixed to the lid of the case using self-adhesive p.c.b. supports, taking care to position it in such a way that the lid will close easily.

## TESTING

When the main power supply has been connected, press a key, and check that the piezo buzzer WD1 sounds. If it does not, disconnect the power supply and check for obvious faults such as either IC1, WD1 or

The back-up battery holder, keypad piezo buzzer, siren and relay box mounted in the base of the case.





The completed unit showing position of the keypad.

any diode the wrong way round. Double check the connections between the keypad and the circuit particularly if flexible wires have been used.

Re-connect the power supply and if WD1 still fails to sound, using a voltmeter (multimeter set to d.c. volts) attach the negative lead to 0V on the board (the negative end of C4 is a convenient place to clip a lead), and use the positive voltmeter lead as a probe. Touch the positive end of C4 with the probe; a reading of between 11V and 12V should result.

Check the following results:

Voltage regulator IC2  
Input = 12V (approx)  
Common = 0V  
Output = 5V

Combination switch IC1  
Pin 2 = 0V  
Pin 3 = 0V  
Pin 9 = 4V to 5V  
Pin 14 = 4V to 5V, dropping slightly if any key is pressed.  
Pin 13 = as pin 14

If the pin 14 test is satisfactory, the buzzer WD1 must be at fault.

If WD1 is operating correctly, try inputting the password 0, followed by the button marked \*. The solenoid should operate for two seconds and the relay should toggle. Try entering the wrong password three times. The siren should operate.

If any or all of these tests fail, establish whether the i.c. or the transistors are at fault, by taking voltmeter readings as follows:

Connect the positive lead of the voltmeter to 12V positive (the positive end of C4 is a handy connecting point), and use the negative voltmeter lead as the probe.

The voltmeter readings (not the actual voltages relative to 0V), should be as follows:

IC1 pin 17 changes from 0V approx. to 12V approx. for two seconds on correct password.

Pin 16 toggles between 0V and 12V approx. at each correct password.

Pin 15 changes from 0V approx. to 12V approx. for one minute if three wrong passwords are entered.

Failure to obtain these readings indicates

that IC1 is not working correctly, or not connected properly, otherwise one or more of the transistors or associated components must be at fault.

## FINAL SETTING UP

If all is well and the circuit is working correctly, try changing the password as described at the beginning of the article.

The back-up battery may now be connected. If the 12V power supply is disconnected the password should still be retained by IC1.

The optional door switch S1 must be normally open for correct operation of the circuit. When the contacts of S1 close, the siren is triggered for one minute. This trigger could be used to set off a conventional alarm system if preferred.

A number of options are possible; for example, switch S1 could be an under-carpet pressure mat, or a lever-operated microswitch (as shown in Fig. 3).

If a microswitch is used, then the lead marked X (see Fig. 3) must be connected to contact A for correct operation if the switch is positioned such that it is pressed when the door or window is closed. S1 could also be used in this way to detect somebody tampering with the case of the project.

## SOLENOID

An inexpensive solenoid will provide a small amount of movement, and a degree of mechanical skill is required to enable it to operate a lock successfully. For something over £10 an electric door catch release can be purchased. The solenoid only operates for two seconds since its purpose is to release the door, not open it!

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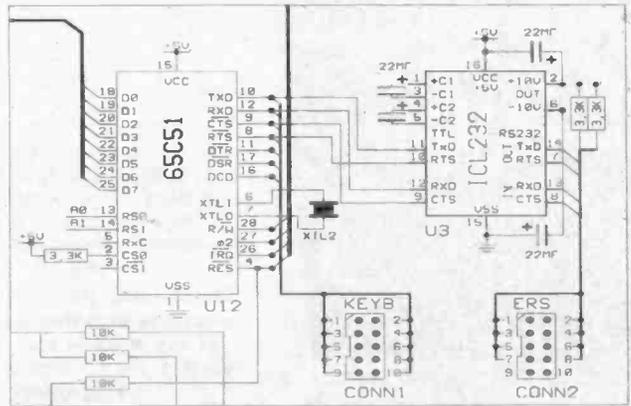
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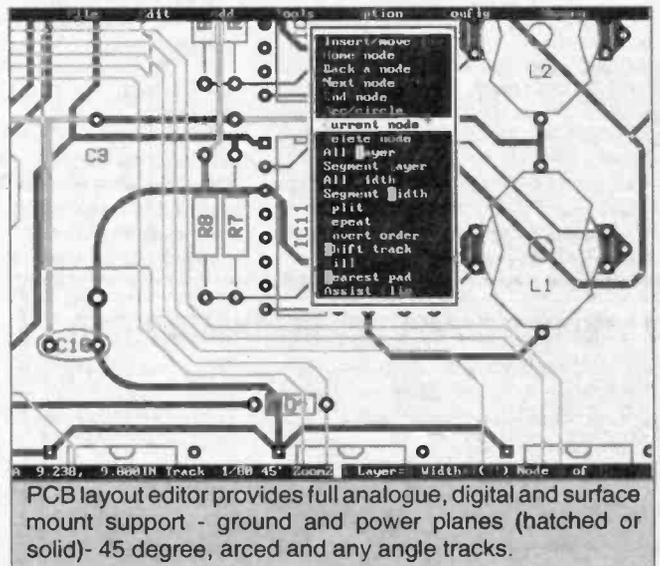
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# INTERFACE

**Robert Penfold**



**A**S POINTED out in last month's *Interface* article, it is not essential to have a special driver circuit in order to control a stepper motor. Control can be provided by driving the four solenoids from four latching digital outputs of a parallel port, via relay driver style circuits. This table shows the output states needed in order to drive a four phase stepper motor correctly.

Step Number	Q1	Q2	Q3	Q4
1	L	H	L	H
2	H	L	L	H
3	H	L	H	L
4	L	H	H	L
5	L	H	L	H

Of course, step 5 just takes things back to the beginning of the sequence, and for continuous rotation steps 1 to 4 are just repeated over and over again. For continuous rotation in the opposite direction the sequence is reversed (i.e. repeat steps 5 to 2 over and over again).

You will notice that Q1 is always at the opposite state to Q2, and that Q3 is always at the opposite state to Q4. This permits simplified control with each pair of solenoids being driven from one digital output line. One solenoid in each pair is driven in-phase, while the other is driven out-of-phase via an inverter stage.

The basic scheme of things is shown in Fig. 1. An identical circuit (controlled by a different output line) is needed to drive the motor's other two solenoids. The inverter can be any TTL or CMOS type powered from a 5 volt supply, or a 4000

series CMOS inverter powered from the 12 volt motor supply. Alternate toggling of the two digital outputs steps the motor to give continuous rotation.

## SAA1027

Although driving a stepper motor without a special interface device is clearly not particularly difficult, most users seem to prefer the convenience of a proper interface circuit. The hardware is much more costly, but the software side of things is more straightforward.

The standard stepper motor driver chip is the SAA1027, which is not a particularly modern device, and it is out-performed by some more recent chips. For the amateur user it is still probably the best choice though, as it is relatively cheap, has a perfectly adequate level of performance, and it is readily available. It is very easy to use, as can be seen from the basic SAA1027 stepper motor driver circuit of Fig. 2.

On the output side of the circuit there is no need for discrete protection diodes as protection circuitry is incorporated in the SAA1027. The supply

voltage can be anything from 9.5 to 18 volts, but most small stepper motors seem to be designed for 12 volt operation. The maximum permissible output current is 500 milliamps, which means that the resistance of each solenoid should be 24 ohms or more for operation on a 12 volt supply. In practice most stepper motors seem to have coil resistances of a round 100 ohms or more, and are therefore perfectly suitable for use with this circuit.

The circuit is controlled via three input lines, one of which is the "Clock" input. Each input pulse on this line shifts

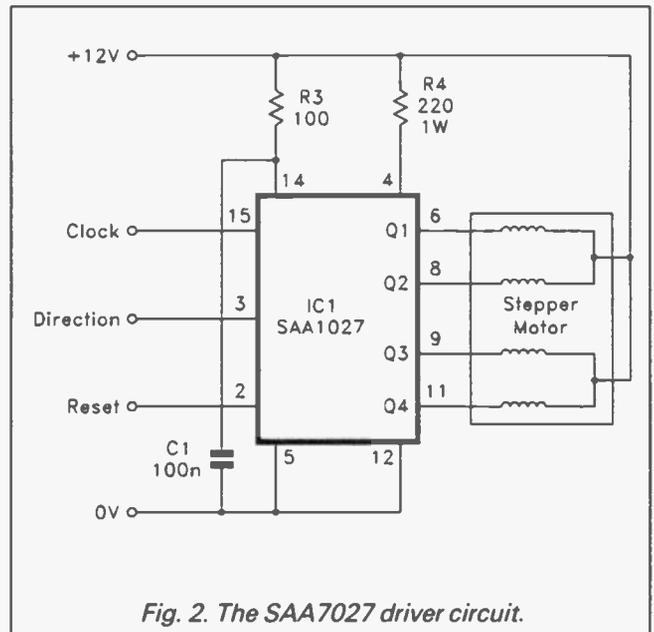


Fig. 2. The SAA1027 driver circuit.

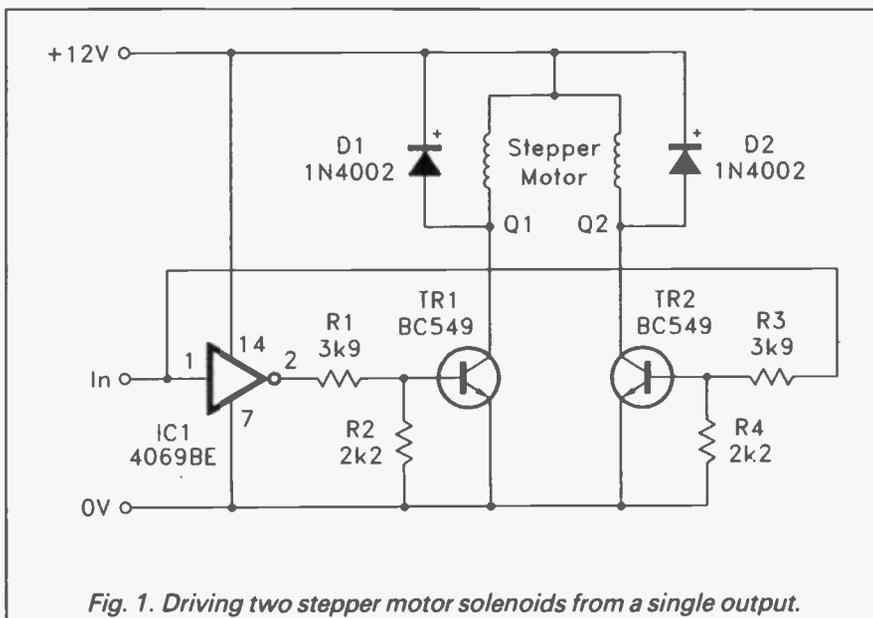


Fig. 1. Driving two stepper motor solenoids from a single output.

the motor one step, and it is on high to low transitions that the outputs of IC1 change state. With the motor connected to the circuit correctly, taking the "Direction" input low gives clockwise rotation of the motor - taking it high gives counter clockwise rotation.

Pulsing the "Reset" input low simply takes the outputs back to their initial state (i.e. step 1/5 in the table provided previously). This does not mean that the motor will be taken back to its starting position. To permit this the software must include a counter which is incremented each time the motor is stepped in one direction, and decremented whenever it is stepped in the opposite direction.

The counter will then indicate how many steps the motor must be moved, and in which direction, in order to return it to the starting position. It is then quite easy to implement a software routine which can read the counter and step the motor back

to its starting point. In some applications the "Reset" input will not be needed, and it is then simply connected to the +12 volt supply line.

### Level Shifting

One slight problem when using the SAA1027 is that its inputs are not designed to operate using normal 5 volt logic levels. Logic 1 is at a potential of around 12 volts, not 5 volts. There should be no difficulty in controlling the SAA1027 from a CMOS control circuit having a 12 volt supply, but with 5 volt logic control circuits a level shifter will be needed at each input.

There are several ways of providing this simple level shifting, including circuits based on operational amplifiers and open collector TTL chips. In this case a very high operating speed is not essential, and simple transistor level shifters are perfectly adequate. Fig. 3 shows the circuit diagram for a simple level shifter of this type. Of course, one of these circuits is needed at each input of the SAA1027.

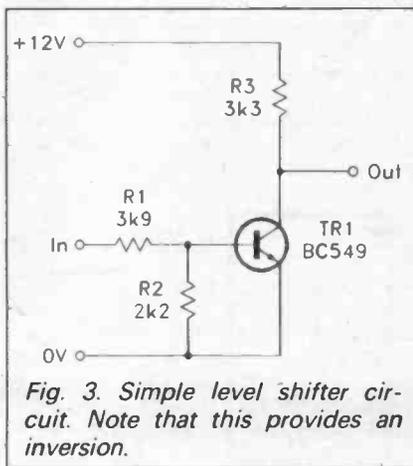


Fig. 3. Simple level shifter circuit. Note that this provides an inversion.

Note that there is an inversion through each level shifter. Therefore, with these circuits added a high level on the "Direction" input gives clockwise rotation, pulsing the "Reset" input high resets the circuit, and the motor is stepped as the "Clock" input is taken through a low to high transition.

### Comio V.1.00

Comio is a set of C callable functions which form a buffered serial i/o communications library. It is an interesting product, both for what it does and how it is presented.

Intended for IBM PC computers and compatibles, Comio is supplied as a set of source code files in C and 8086 assembly language. No object files are supplied, so you need both a C compiler and an assembler to use it. It was written using Microsoft's Quick C with Quick Assembler package (something the author has used extensively), but it should also be usable with the full professional Microsoft C and Macro Assembler (MASM).

The assembly language file is supplied in two versions, one of which uses the simplified segment directives (.MODEL, .CODE etc.) and the macro features of Quick Assembler and MASM 5.1 and 6.0.

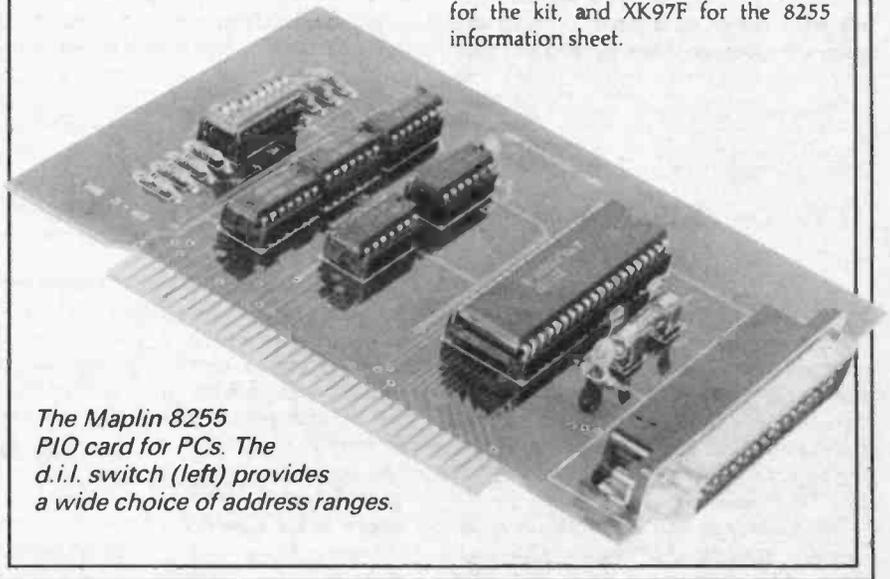
## Inexpensive PIO

On more than one occasion I have bemoaned the lack of low cost prototype cards etc. for the PCs. The situation is now slightly better than it was two or three years ago. Maplin sell three useful prototyping cards at reasonable prices, and there are some useful products available from R. Bartlett which are often advertised in *Everyday with Practical Electronics*. Recently I have been trying out the Maplin 8255 PIO card for the PCs.

This is a kit for a straightforward 8255 PIO card which gives 24 input/output lines. These, plus the

supply lines, are made available on a 37 way D connector. The address range occupied by the card can be altered via a bank of d.i.p. switches so that conflicts with other hardware can be avoided. The printed circuit board is a good quality through-plated type. Everything fits together properly and the finished unit worked first time.

Even with the recently increased price of £21.95 (plus 80p if you require the 8255 data sheet) this kit still represents very good value for money, and is one of the cheapest ways of connecting a PC to relay drivers, D/A converters, etc. The Maplin order codes are LP12N for the kit, and XK97F for the 8255 information sheet.



The Maplin 8255 PIO card for PCs. The d.i.l. switch (left) provides a wide choice of address ranges.

The other is plain vanilla 8086 assembly language, and should be usable with other assemblers. The C code is ANSI compatible. Most current C compilers meet this standard.

The product implements a fully buffered, instant return communications system. When a program written using Comio sends data to a port, the functions return immediately, so the program can continue operation while the Comio functions deal with transmitting the data in the background. Buffered read functions are also provided.

### Functions

The full set of functions includes opening and closing routines, functions to read and write single bytes, and to read and write multiple bytes (block or null-terminated string), functions to check for completion of read write operations, and error handling functions. There are also functions to check the validity of ports and file pointers, and a number of other "utility" functions. In all, 23 functions are provided.

The ports can be fully configured for Baud rate, parity etc. in the opening function, and the closing function makes sure your program tidies up after itself. The memory for the buffers is not allocated until communications are opened, and it is

freed on closure. This means memory is not occupied if the communications facilities in a program are not used. The buffer sizes can be from 19 to 66519 bytes. The two serial ports, COM1 and COM2, can be used concurrently for read or write.

The manual is extensive. It runs to 77 A4-size pages in a channel binder, is well-written, and includes a full description of all the functions, plus background information and descriptions of the example programs. You should not find yourself foundering, short of some vital information, with this manual. The example programs are interesting, but you will, of course, need something for them to communicate with to explore them fully.

The most obvious uses for this product are perhaps to write a printer driver or a communications system between two computers, but other uses are possible, including control and robotics. The serial ports have, perhaps, been under-used in these areas, but Comio could help to change this. Being supplied in source code, it may also interest students of programming, who may be interested to see how a device driver is implemented.

Comio is remarkably inexpensive at £34.00 including VAT. It is supplied direct by Micro SciTech Ltd., Dept EPE, 2 Coxford Drive, Maybush, Southampton, SO1 6FD, Telephone/Fax 0703 784578.

# CHRISTMAS LIGHTS COLOUR SPECTRUM



BY MARK DANIELS

*Drives up to 30 l.e.d.s and produces a steadily changing colour sequence from red to orange, yellow and green to yellow, etc.*

**T**HERE have been many designs published for Christmas lighting effects units, a number of which have been add on units for conventional filament lamp sets, producing various flashing, sequencing or shimmering effects. Other designs have made use of more modern, semi conductor lamps, or l.e.d.s as they are better known.

Using single colour l.e.d.s a conventional string of lights can be made which may then be driven by suitably designed electronics to produce one or more of the effects outlined above. If, however, two l.e.d. dies of different colours are encapsulated in a single colourless, translucent package a broad range of the light spectrum may be produced simply by driving the two dies at various current ratios.

The two colours best used for this are Red and Green. Being two of the primary

colours they can, between them, produce a wide variety of other colours including yellow and orange of various shades. The third primary colour is, of course, blue, but is not readily available in a bi-colour package.

## OBTAINING THE SPECTRUM

There are two possible ways of connecting a red and a green l.e.d. in a single package, each with its own advantages and disadvantages. The simplest way, as depicted in Fig. 1a, is to connect them in a common cathode (k) (or common anode (a)) configuration, with three connection leads. This has the advantage of allowing both l.e.d.s to be lit simultaneously providing a continuous spectrum by driving them over a range of current ratios.

The alternative is to connect them in inverse parallel (Fig.1b) such that connection to a d.c. supply will cause either one or the other of the diodes to emit light (but not both together) depending upon the polarity of the supply. This has the advantage of requiring only two connecting leads.

It would appear at first glance that the only way a wide colour spectrum may be obtained is by using the "tri-colour" l.e.d. with its three connecting leads. This is, indeed, a possibility and may be realised with fairly simple drive circuitry.

Further investigation reveals that parallel connection of the devices using three-core connecting wire is necessary to produce a chain of lamps and that each individual l.e.d. requires its own current limiting resistor. This does not make it easy to produce an attractive chain of Christmas lights.

The two pin "bi-colour" device appears, at first glance, to be even more awkward to use in this application than its multi-pin cousin, since it is not possible to light both the green and red l.e.d.s simultaneously due to their conflicting polarity requirements. It does however eliminate the above connection problems, as simple series connection of the devices is now possible.

The fact that the red and green l.e.d.s may not be on together is not a particularly serious problem as the response of the human eye is fairly slow. If we

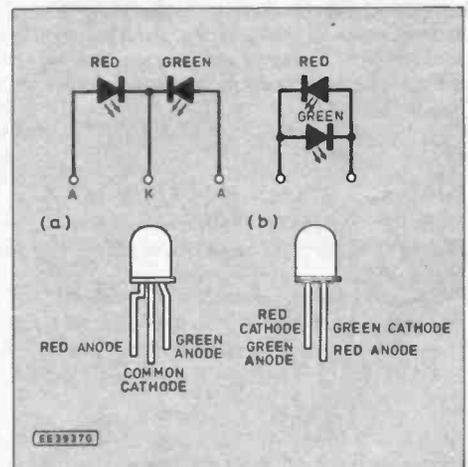
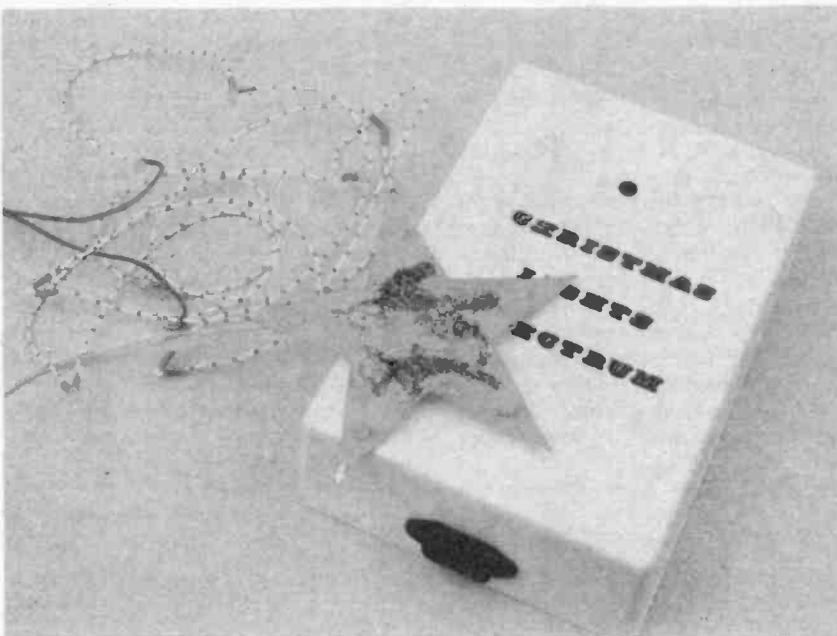


Fig. 1. (a) "Tri-colour" l.e.d. diagram and encapsulation and (b) "bi-colour" diagrams.

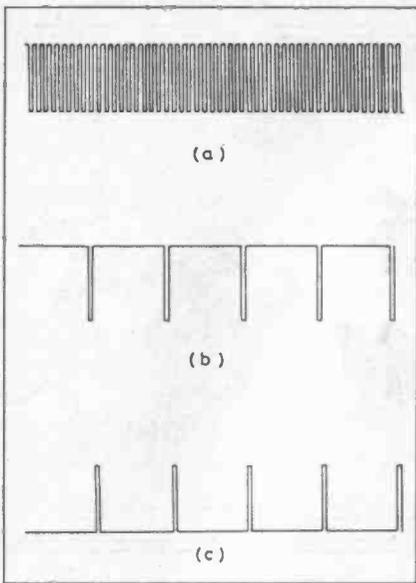


Fig. 2. Output waveform produced by the unit.

switch between the two colours sufficiently rapidly (more than about 25 times per second) the brain will perceive the result of the two colours being mixed together to produce yellow or orange, depending on the intensity ratio.

This allows us to use an alternating current supply which lights alternately one, then the other of the two colours. A symmetrical square wave as depicted in Fig. 2a will light the red and green for equal lengths of time in each cycle producing yellow light. Using the asymmetrical waveforms of Fig. 2b and Fig. 2c will produce nearly pure red or pure green light.

### CIRCUIT DESCRIPTION

The complete circuit diagram for the control board of the Christmas Lights Colour Spectrum is given in Fig. 3. The circuit is powered from the mains via a conventional stabilised power supply comprising transformer, T1, rectifier diodes, D1 to D4 and electrolytic smoothing capacitor, C1. Zener diodes, D5 and D6 with current limiting resistor, R1 regulate the supply providing 15 volt and 30 volt supply rails.

A 555 timer, IC1, is employed here in its astable mode to provide the low frequency timing waveform, which has a period of approximately ten seconds. The non-linear triangular waveform produced across capacitor C4 by IC1 (see Fig. 4) is buffered by transistor TR1 and provides a modulating input to the control voltage pin (9) of IC2, a phase locked loop.

### V.C.O.

In this application only the voltage controlled oscillator section (v.c.o.) of the 4046 is used. A 5.4V Zener diode is connected internally across pins 8 and 15 of IC2 which is used in conjunction with Zener diode D8 and resistor R6 to provide a stabilised supply above the normal zero volt rail. This allows the triangular waveform produced by IC1 to sweep the complete control voltage range of the v.c.o.

The variable frequency output of IC2 is used to repeatedly trigger a short period monostable, IC3, the pulse length of which may be adjusted by preset VR1. At the highest output frequency of IC2, each pulse produced by IC3 is very nearly equal to one output cycle of the v.c.o. and thus produces an almost continuous positive voltage, as in Fig. 2b.

As the frequency of the v.c.o. reduces, the time interval between the pulses from IC3 increases and a point is reached when the time between pulses is equal to the duration of the pulses, shown in Fig. 2a. At this point the red and green l.e.d.s are on for equal time intervals and will produce light of a yellowish colour. Reducing the frequency of the v.c.o. still further, to its minimum, will give the longest intervals between pulses and the effect of the pulses is almost negligible.

### BRIDGE AMPLIFIER

The output at pin 10 of IC3 is a low level signal of approximately five volts peak and capable of providing only a few milliamps of current and thus requires amplification to provide a useful output. A dual op-amp IC4 is connected as a bridge amplifier to provide dual complementary outputs of 60 volts peak-to-peak from the single input.

The high current output stage comprises two complementary followers, consisting of pairs of npn/pnp transistors, TR2 to TR5, connected in a bridge arrangement. This gives an output current capability approaching 200mA with the specified transistors, allowing at least three chains of l.e.d.s to be driven.

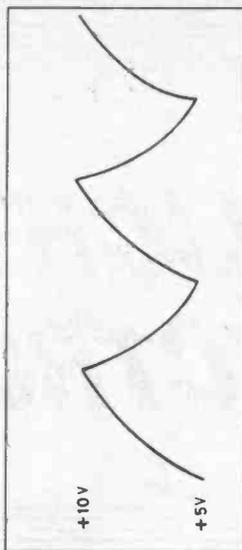


Fig. 4. Non-linear triangular waveshape produced across C4 by IC1.

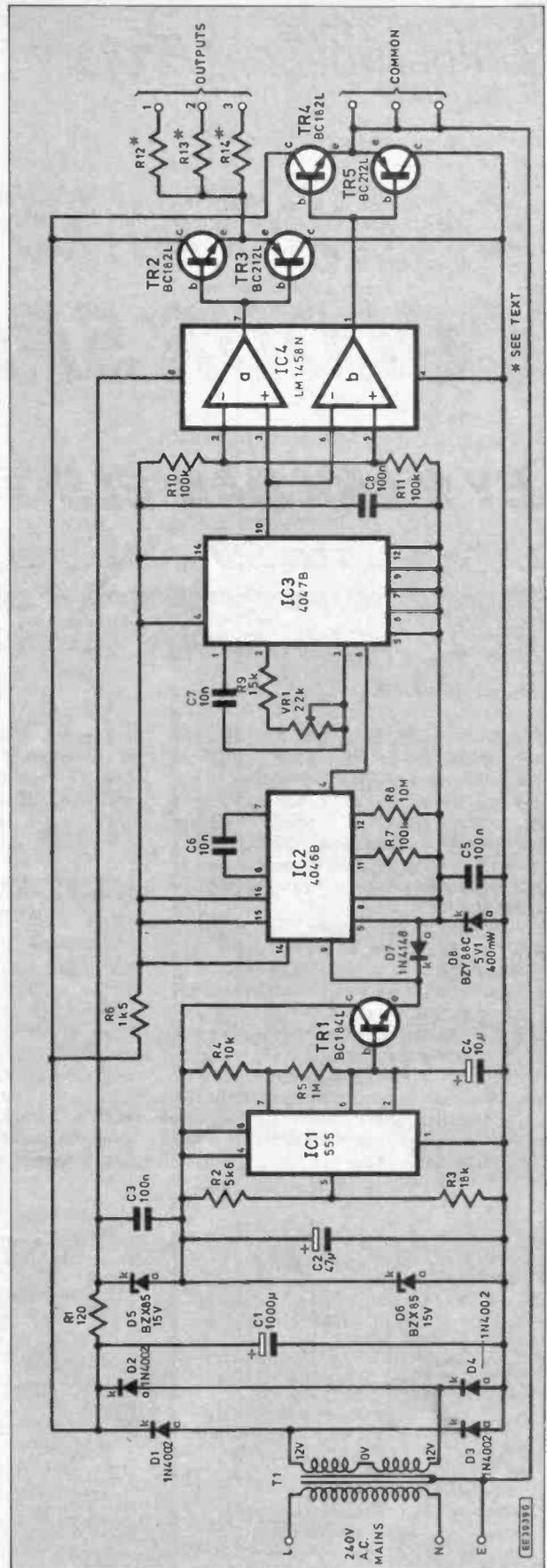


Fig. 3. Complete circuit diagram of the Christmas Lights Colour Spectrum.

Resistors, R12 to R14 are the current limiting resistors for the l.e.d.s and are selected to give a current through the chains of l.e.d.s of around 20mA. Table I gives suitable values of resistor for up to ten l.e.d.s per chain, which is considered to be a sensible limit, although under ideal conditions twelve is possible, but the resistor will have to be selected by trial and error.

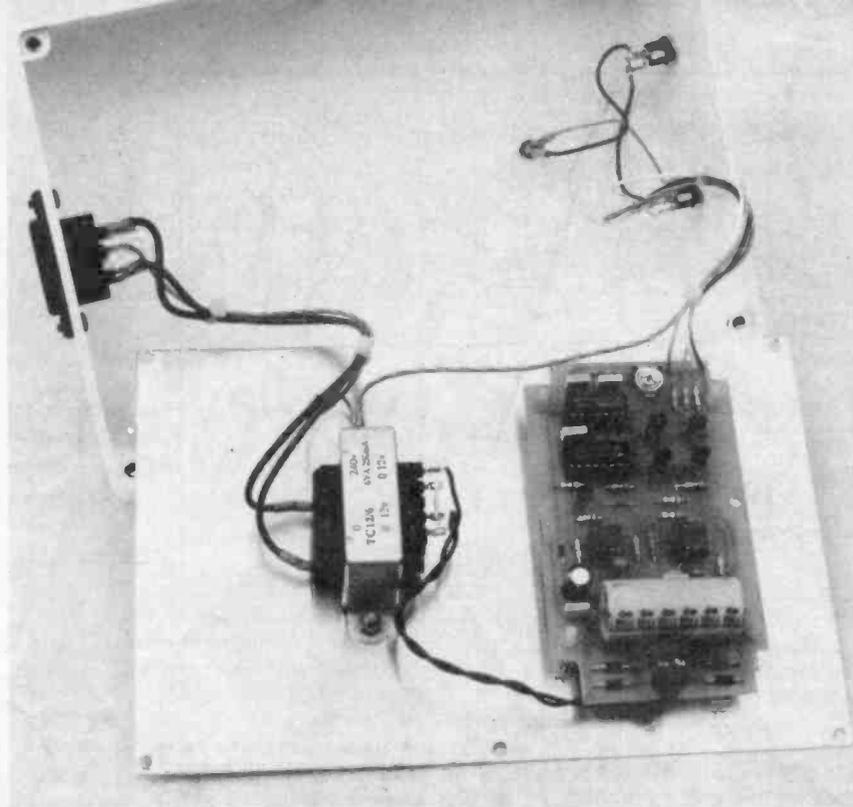
## ALTERNATIVE CURRENT LIMITING

With large numbers of l.e.d.s in a chain (see Fig. 5), the current difference between the red and green dies can become very significant due to the somewhat higher forward volt drop of the green die and the low voltage dropped across the current limiting resistor. This can be overcome to a certain extent by shunting the resistor,  $R_L$  with a diode and second resistor,  $R_G$  in series (see Fig. 6) to allow the current through the green dies to be increased without increasing the current through the red ones.

The value of resistor,  $R_G$  will have to be determined by experiment. A suitable starting point would be to make  $R_G$  equal to  $R_L$ .

If the red brightness increases (rather than the green) when this circuit is connected, merely turn the diode round to obtain the correct result.

The circuit of Fig. 6 replaces one of the



## COMPONENTS

### Resistors

R1	120
R2	5k6 1% metal film
R3	18k 1% metal film
R4	10k
R5	1M
R6	1k5
R7, R10,	
R11	100k (3 off)
R8	10M
R9	15k
R12, R13,	
R14	See Text and Table 1

0.25W 5% carbon film, unless stated

See  
**SHOP**  
**TALK**  
Page

### Potentiometer

VR1 22k sub-min horz. preset, lin

### Capacitors

C1	1000 $\mu$ axial elect., 50V
C2	47 $\mu$ radial elect., 25V
C3	100n polyester 5mm
C4	10 $\mu$ tantalum bead, 25V
C5, C8	100n polyester (2 off)
C6, C7	10n polyester (2 off)

### Semiconductors

D1 to D4	1N4002 1A 100V rec. diode (4 off)
D5, D6	BZX85 15V 1.3W Zener (2 off)
D7	1N4148 signal diode
D8	BZY88C5V1 400mW Zener
D9, D10	Bi-colour two pin l.e.d.s, as etc many off as required
TR1	BC184L npn silicon
TR2, TR4	BC182L npn silicon (2 off)
TR3, TR5	BC212L pnp silicon (2 off)
IC1	555 timer i.c.
IC2	4046B phase-locked loop
IC3	4047B CMOS multivibrator
IC4	LM1458N dual op. amp

### Miscellaneous

T1	6VA mains transformer: 240V a.c. primary; 12V-0V-12V secondary
----	--

Plastic case, size 220mm x 150mm x 65mm; 8-pin d.i.l. socket (2 off); 14-pin d.i.l. socket; 16-pin d.i.l. socket; miniature d.c. power-in connectors (3 off); 3-core mains cable; connecting wire; rubber grommet; solder tag; M3 x 12mm screws and nuts; solder, etc.

Approx cost  
guidance only

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inc. 10 l.e.d.s.

Table 1. Suggested Current Limiting Resistor Values

No. of l.e.d.s	Approx value for R12, R13 or R14	Red l.e.d. Current (mA)	Green l.e.d. Current (mA)
1	1k2	21.7	21.5
2	1k2	20.0	19.7
3	1k	22.0	21.4
4	1k	20.0	19.2
5	820	21.9	20.7
6	820	19.5	18.0
7	680	20.6	18.5
8	560	21.4	18.6
9	470	21.3	17.4
10	330	24.2	18.2

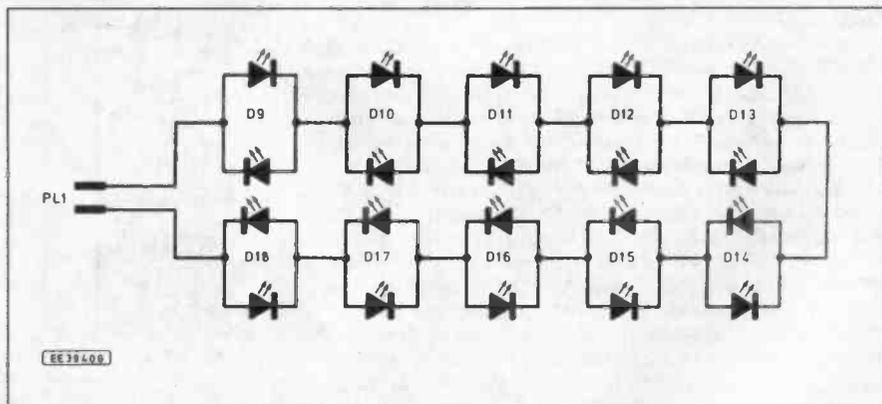


Fig. 5. Circuit diagram of a 10 l.e.d. light chain.

resistors, R12 to R14 in the main circuit of Fig. 3 and may be repeated for each chain where a relative brightness problem is apparent.

## CONSTRUCTION

The majority of the components are assembled on a single sided glass-fibre printed circuit board. This board is available from the *EPE PCB Service*, code 813. The component layout and full size copper foil master pattern is shown in Fig. 7.

Before commencing assembly of the printed circuit board decide how many l.e.d.s are going to be used in each chain and then, from Table 1, select the appropriate value of resistor for each of R12 to R14. If it is intended that a large number of l.e.d.s (e.g. eight or more) are to be used in one or more of the chains then it may be desirable to use the alternative method of current limiting, described above, for each such chain.

For ease of assembly it is recommended

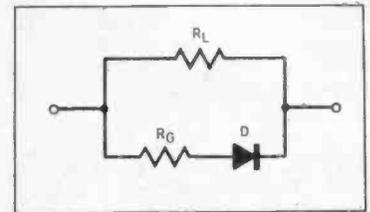
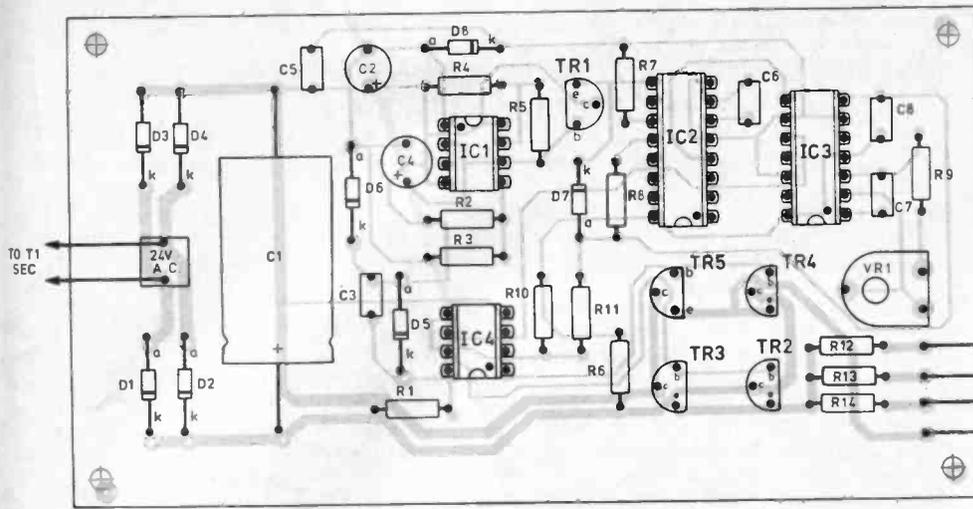
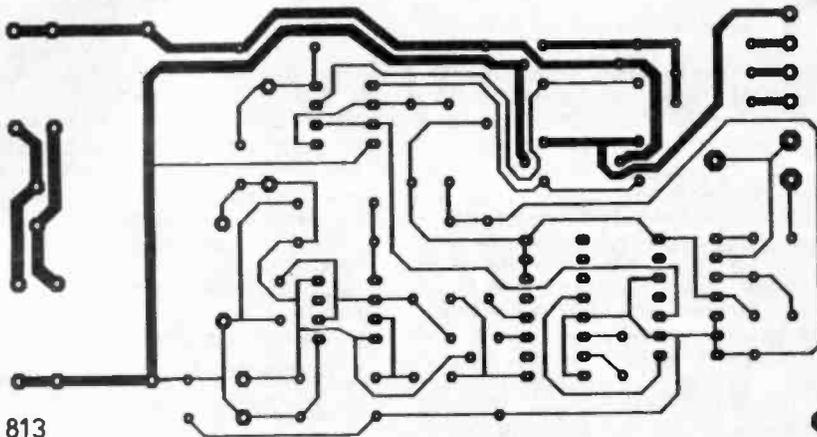
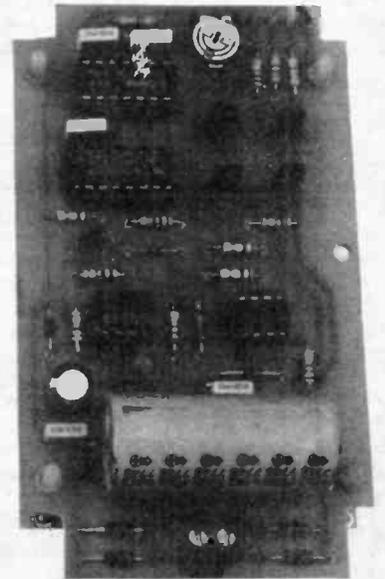


Fig. 6 (above). Alternative "biased" current limiting circuit.

Fig. 7 (left). P.C.B. layout and wiring for the Christmas Lights Colour Spectrum.



that the components are fitted in the following order: i.c. sockets, resistors, small capacitors (other than the i.c.s) and finally the large power supply smoothing capacitor, C1. Check carefully the orientation of semiconductors and the tantalum and electrolytic capacitors, as a mistake here could result in the affected components being destroyed or the completed circuit malfunctioning. A p.c.b. mounting screw terminal block was used on the prototype for the a.c. input and, if used, should also be fitted at this stage.

It is suggested that the i.c.s be fitted in their sockets after the board has been mounted in the case. Care should be taken in handling IC2 and IC3 since these are both CMOS devices and may easily be damaged by exposure to static electricity.

### MAIN UNIT ASSEMBLY

The assembled printed circuit board (p.c.b.), mains transformer and output sockets are all fitted in a plastic case for safety. A suggested case layout is shown in Fig. 8, with the majority of the parts fitted to the case lid. If the recommended case is used there will be plenty of room to get everything in without it being cramped.

Using the p.c.b. as a template, drill four 3mm mounting holes in the case lid and fit a plastic p.c.b. support pillar to each hole using suitable size self-tapping screws. The board can now be mounted on top of these

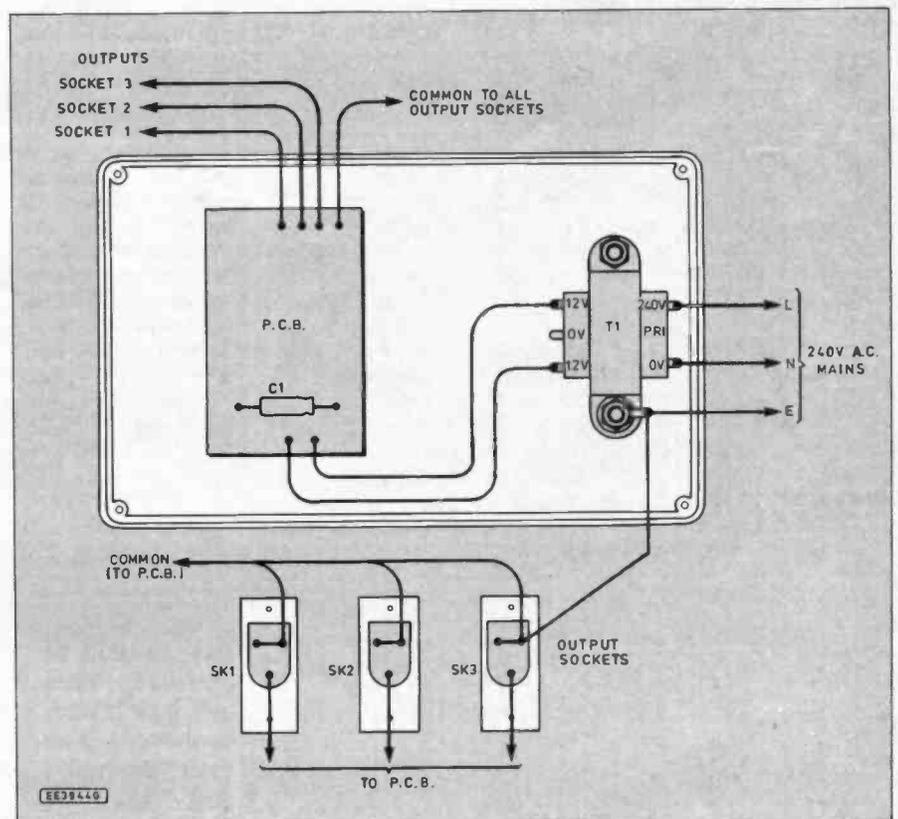


Fig. 8. Case layout and interwiring of T1 and the sockets.

pillars with four more screws after all external connections have been made to it, ensuring that the pillars are long enough not to allow the points of the screws to contact each other.

The mains transformer is mounted adjacent to the p.c.b. using two M3 × 12mm machine screws and nuts, an "Earth" solder tag being placed under one of the nuts for the mains earth (E) wire connection. Make the a.c. connections from the two ends of the transformer secondary winding to the p.c.b. using 7/0.2 wire. No connection is made to the centre tap of the winding.

If the transformer used has two separate 12 volt secondaries then these will need to be connected in series in accordance with any instructions supplied with the device. In the absence of any data determine which connections go to each secondary by using a meter or a lamp and then connect them in series (i.e. one end of the first winding to one end of the second winding) and check for voltage across the two free ends. If there is no output then reverse the connections to only one of the windings.

Decide how many chains of lights are going to be connected to the unit and fit a d.c. power socket to the main part of the box for each light chain. Up to three may be fitted without modifying the circuit. On the prototype unit only two sockets were fitted and the third output was used to drive a bi-colour indicator l.e.d. fitted to the top of the case to show when the unit is operational.

Drill a hole in one end of the box and fit it with a grommet for the mains lead. Thread a length of three core mains flex through the grommet and secure it internally with a cable tie. Make the mains connections directly to transformer T1 primary and connect the Earth lead to the solder tag fastened under the transformer mounting screw.

## LIGHT CHAINS

It is important to ensure that all the l.e.d.s in a single set are from the same manufacturing batch to guarantee inten-

sity and colour matching of the devices. The l.e.d.s may be made up into conventional strings or mounted on decorative card cut into stars or other traditional Christmas designs.

The connection details for a string of ten l.e.d.s is shown in Fig. 9a. Note that in this diagram all the l.e.d.s are connected with the same polarity and will therefore follow the same colour sequence as each other. Connecting alternate devices the opposite way round, as in Fig. 9b, will cause every other l.e.d. to follow the reverse colour sequence to its neighbours.

Use thin single core stranded connecting wire, of an appropriate colour, e.g. green to match the Christmas tree, for making up the light chains. Cut the leads on each l.e.d. short and solder the wires to the l.e.d.s to connect them in series.

The connections on the finished set may be given some protection by coating them with "Araldite" or hot glue from a glue gun. Finally run a long wire from each end of the chain and connect to a miniature d.c. power plug to fit the socket(s) on the main unit

An alternative method is to fix the l.e.d.s to a card star or other decoration, fold the pins over to secure them in place and make connections on the back of the card where they will not be seen and then glue a piece of paper over the wiring.

## TESTING

Fit a mains plug, fused at 3 amps or less, to the end of the mains cable and give all the wiring a final check. If the i.c.s have not already been fitted now is the time to fit them, observing the handling instructions given earlier.

Temporarily assemble the case (closing the lid), connect a string of lights to one of the output sockets and plug the unit into the mains. If all is well the lights should come on and, after a few seconds, start changing colour, eventually reaching the opposite primary colour to which they started at. The sequence then reverses and is repeated indefinitely.

If all appears o.k. preset VR1 may be

adjusted to obtain the best result. This adjustment allows the widest range of colours to be selected and also enables any "jittering" at the ends of the spectrum to be smoothed out.

The "jittering" is caused by the monostable, IC3 failing to retrigger, due to it not having finished its current timing cycle before the next trigger pulse arrives. This does prevent absolutely pure red and green from being obtained, but in practice the differences are virtually unnoticeable.

## FAULT FINDING

If the Christmas Lights Colour Spectrum fails to work immediately at switch-on do not despair as the fault may turn out to be something quite simple. Check first for anything obvious such as a failed fuse in the plug or bad connections, dry joints, solder splashes bridging copper tracks, or semiconductors inserted the wrong way round, etc.

If the unit still does not work then the fault may lie in malfunctioning components. With the circuit diagram at hand check power supply voltages first, across transformer T1 secondary, across capacitor C1 and across the Zener diodes. Then check for power to all the i.c.s and the collectors of the transistors.

The voltage across TR1 emitter (e) and pin 1 of IC1 should be steadily changing over a period of several seconds, between approximately 5V and 10V or 11V (see Fig. 4). If it remains fixed at a steady potential try replacing IC1 with a known working device (this is made much simpler if i.c. sockets were used when constructing the board). If this does not cure the fault then replace transistor TR1.

## CHECKING THE V.C.O.

With the control voltage on pin 9 of IC2 sweeping up and down in the correct manner a variable frequency square wave should be present at pin 4 of the i.c. Check for this with an oscilloscope, if one is available or use a loudspeaker with a one kilohm resistor in series with it across pins 9 and 10 of IC2 and listen for a note which changes pitch over a few seconds. The note will become inaudible as the frequency falls below the lower frequency response of the ear.

If IC2 fails to produce the correct output check the values and location of C6, R7 and R8 before replacing it. Also check the orientation of diode D7, as this protects the i.c. from negative input voltages as capacitor C4 initially charges from zero volts on the first cycle of IC1.

The output from IC3 is best checked by connecting a bi-colour l.e.d. and a one kilohm series resistor across pin 10 and pin 11 of the i.c. (pin 11 is the complementary output to pin 10). The result if the unit is functioning correctly up to this stage will be the same as the expected output from a fully functional unit. C7, VR1 and R9 are the components associated with IC3 and should be checked before suspecting the i.c.

The dual op.amp IC4 should only really need checking for its power supply rails (30 volts) and outputs across pin 1 and pin 7, which may be checked in a similar manner to the outputs of IC3, but increase the resistor to 4k7.

If everything is working up to this stage and the unit is still causing problems the fault will most likely be in one or more of the output transistors, TR2 to TR5. □

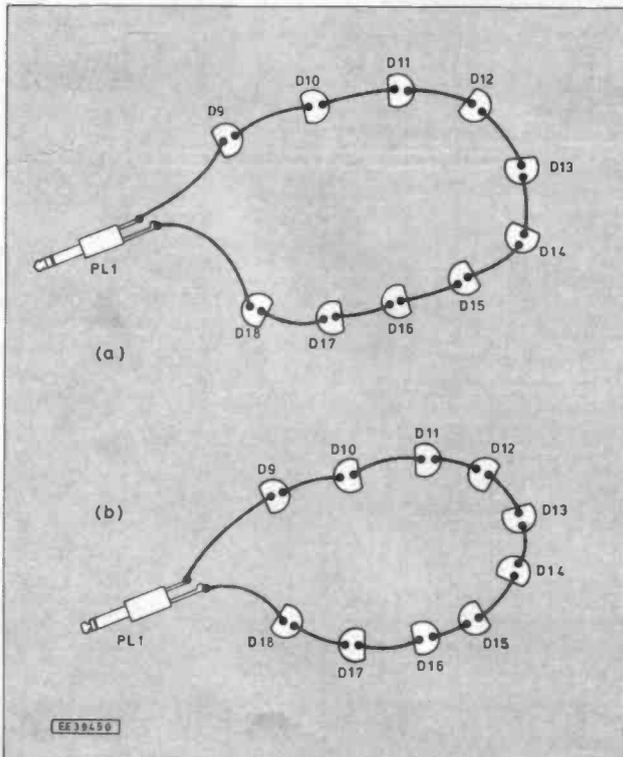


Fig. 9. (a) Connection details for an l.e.d. where all "lamps" follow the same colour pattern. (b) Alternative connection where alternate l.e.d.s follow an opposite colour pattern. The alternative current limiting circuit is not required with this type of display.

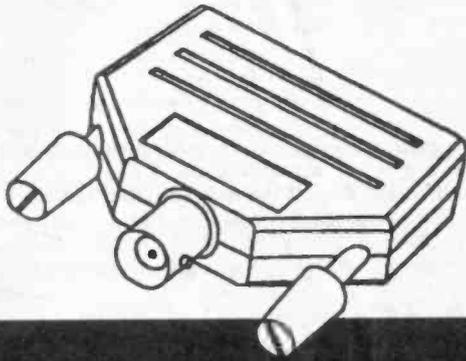
## WARNING

Due to mains voltage being present, extreme care should be exercised when carrying out work on the unit. In all cases the mains plug must be removed

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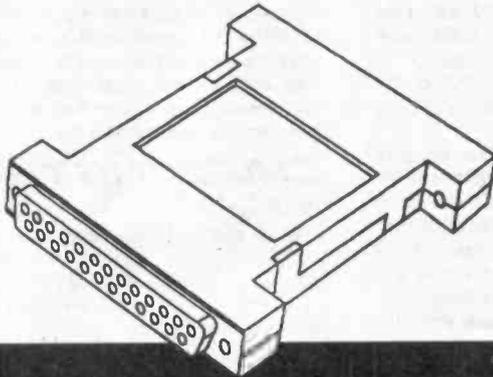
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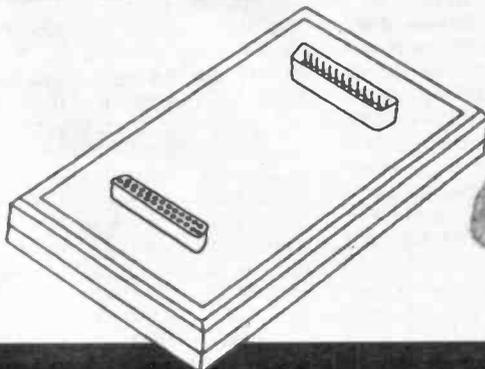
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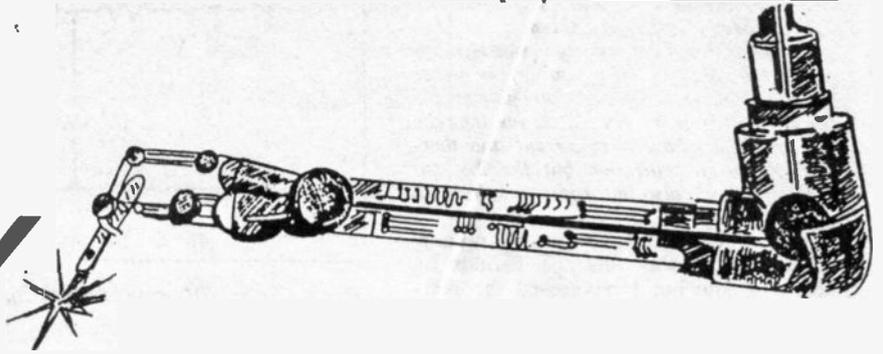


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# CIRCUIT SURGERY



MIKE TOOLEY B.A.

Once again, welcome to *Circuit Surgery*, our regular clinic for readers' problems. In this month's *Surgery* we take a look at reliable regulator design and attempt to explain some of the mysteries of "safe area protection". For the test equipment enthusiast, we describe several probes which can be used to extend the performance of signal tracing equipment. We end this month's instalment with a "round-up" of hints and tips sent in by readers.

## Probing around

A letter from *J. Ryan* has prompted me to give some thought to the use of various types of signal-tracing probes. Mr Ryan has asked for some clarification concerning the probe which was designed for use with the *Audiotest* signal tracer which appeared in *Everyday Electronics*, March 1978 (issue NOT available).

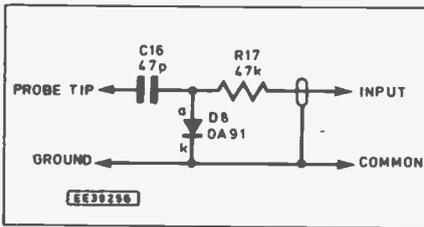


Fig. 1 Audiotest r.f. probe

The *Audiotest Probe* (see Fig. 1) provides a simple means of demodulating amplitude modulated (a.m.) radio signals, allowing the audio frequency signal information to be recovered and then amplified within the *Audiotest* itself. The probe was linked to the *Audiotest* by means of a short length of single-core coaxial cable (the outer braid of the cable being connected to earth/common).

I must confess to using an almost identical home-constructed probe in my own workshop and have found it to be

invaluable when fault-finding on radio equipment. Other signal tracing probes which I have found useful include an "infinite impedance" probe (Fig. 2) and a "wideband" probe (Fig. 3). Both can be constructed on a small piece of matrix board and housed, together with a PP3 battery, within a small probe case (e.g. Maplin JX57M).

## Taken to task

*E. Fielder* of Southampton has taken me to task over the position of the fuses in Mr Taylor's high current power supply (*Circuit Surgery*, October 1992). For those who may have missed it, Mr Taylor's power supply had fuses fitted in both the line (L) and neutral (N) supply connections.

As Mr Fielder rightly surmises, if the fuse in the neutral rail was to become open-circuit the remainder of the mains input circuit although apparently dead would still be live. This is a point worth noting and, if nothing else, should re-inforce the importance of disconnecting *any* item of mains powered equipment *before* delving inside.

There is, however, some justification for Mr Taylor's use of a fuse in both line and neutral. Firstly, one cannot always rely on the line and neutral being correctly wired and secondly the neutral fuse could provide protection against the excessive current

which might flow if the neutral end of the mains transformer primary was to become shorted to the transformer core (presumably earthed).

Depending upon the *actual* potential of the mains neutral connection (rarely is this exactly 0V) a considerable current may circulate when the neutral and supply earth connections are shorted together. Clearly the correct position of mains fuses (not to mention switches) is a topic for some debate and other readers' views and experiences on this topic would be most welcome.

## Safe power supplies

Regular reader *Kevin Davies* writes:

"I have a worrying problem concerning three-terminal voltage regulators. A friend recently asked if I could run his 6V camcorder from a 12V battery temporarily until a purpose designed power supply was available.

I suggested a 7805 type voltage regulator, voltage boosted to 6.2V with two diodes in the common terminal. However, when I went to buy one from a local TV repair shop, the engineer suggested that it was unwise to use this method without over-voltage protection since these regulators commonly fail and go short-circuit. This would clearly damage the camcorder by applying the full 12V.

Fig. 2 (below). Infinite impedance probe

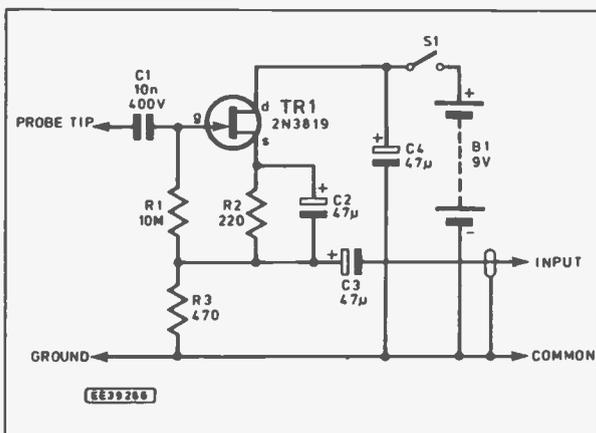
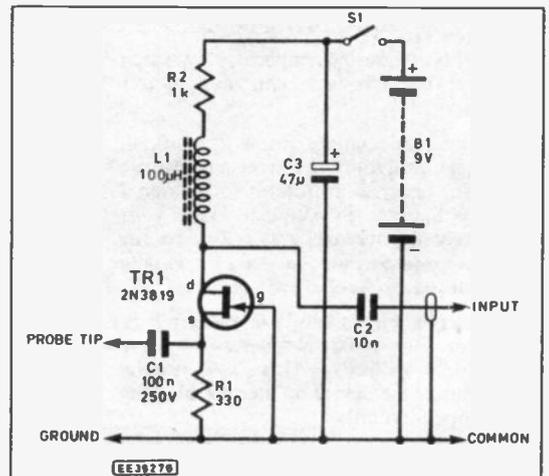


Fig. 3 (right). Wideband (low-impedance) probe



This is both puzzling and worrying since many other items of expensive equipment seems to use these regulators without any special over-voltage protection. Is it true that these regulators do go short-circuit when they fail? Also would you please explain what Safe Area Protection means since I notice in the RS Catalogue that the 1A regulators have over-current and thermal shutdown protection but the 2A and 3A regulators also incorporate safe area protection."

Well, Kevin, it really is difficult to know where to start with this one. Perhaps its tempting fate but I have used three-terminal voltage regulators in numerous circuits and not had even a single failure (but perhaps I have been lucky).

In any event, I decided to find out just what it takes to make one of these regulators go short-circuit by applying the output of a high-current 13V d.c. supply (via a 5A fuse for protection) to a 7805 mounted on a 4°C/W heatsink. The camcorder load was simulated using a 20 ohm 5W resistor. An intermittent short-circuit applied to the output did not produce any adverse effects; the output voltage merely fell to 0V when the short-circuit was applied and then immediately returned to 6.2V when the short was removed.

Heartened by this, I applied a short-circuit for 60 seconds. The output voltage fell to 0V and the regulator quickly became very hot. Happily, the output voltage rose again (a little more slowly than before) to 6.2V when the short-circuit was removed. This test was repeated several times, again with no adverse effects.

I then resolved to leave the short-circuit in place permanently (or for as long as was necessary to cause the regulator to fail). After approximately ten minutes (with the heatsink running extremely hot), the output voltage suddenly rose to 9.8V signifying the point at which the regulator failed. Even so, it is worth noting that it did not go completely short-circuit.

I am not sure what this proves other than demonstrating that three-terminal regulators can (and do) fail internally and produce excessive output voltages. That said, it is worth emphasising that prudent design can be instrumental in significantly reducing the chance of regulator failure.

Points to bear in mind when designing a "safe" supply include:

1. Check that the regulator is correctly rated.
2. Keep regulator dissipation to an acceptable level by avoiding an excessive difference between input and output voltages. For most regulators a difference of 3V to 5V is all that is required. A series-connected low-value wirewound resistor can help in this respect.
3. Fit an adequately rated heatsink in order to avoid excessive temperature rise. As a rule of thumb (or should I say finger?), if you can place your finger on the heatsink close to the device and count to ten this usually indicates that all is well.
4. Insert a silicon diode in the input in order to avoid inadvertent reverse supply polarity. This will usually destroy the device by literally blowing it open-circuit!
5. Fit electrolytic capacitors of 47µF to 100µF with adequate voltage ratings

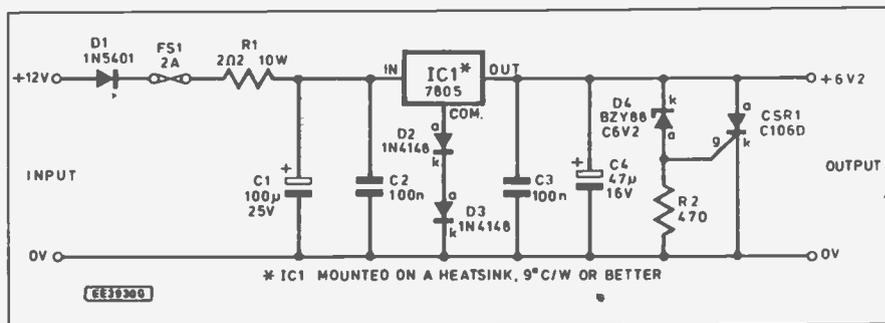


Fig. 4 "Safe" 6V power supply suitable for a camcorder

at the input and output. These can help to prevent momentary voltage surges.

6. Fit ceramic 100nF capacitors at the input and output to prevent high-frequency instability and noise spikes.
7. In exceptional cases, consider fitting an external "crow-bar" and input fuse to provide "last-ditch" protection.

A representative circuit incorporating all of the above features is shown in Fig. 4.

Kevin's second point relating to "Safe Operating Area Protection" is somewhat more general. One can reasonably expect a semiconductor device to cope with either a high current or a high voltage. However, some devices (including power transistors, regulators and power hybrids) are required to withstand high current and high voltage simultaneously during normal operation.

The ability of a device to tolerate such conditions is normally shown by means of a graph showing the region in which the manufacturer considers that the device can be used reliably. It is worth noting that the situation is a little more complicated when the device in question is intended for use in pulsed or switching applications. Here, there may be several graphs relating to different conditions of duty cycle or pulse duration.

A typical safe operating area graph for a power transistor is shown in Fig. 5. Modern power integrated circuits can in-

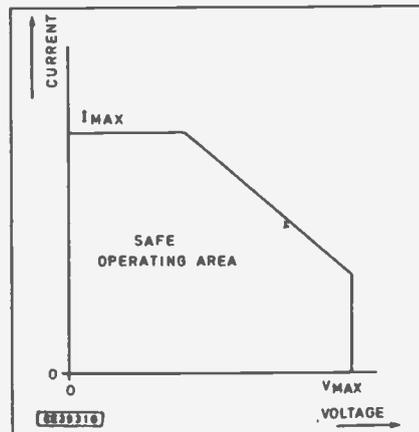


Fig. 5 Typical safe operating area graph for a power semiconductor

corporate voltage and current sensing circuitry which ensures that the device in question operates within the safe area.

Next month: We shall be describing a range extender which will permit low-level audio measurements on a basic multi-range d.c. meter. In the meantime, if you have any comments or suggestions for inclusion in *Circuit Surgery*, please drop me a line at: Faculty of Technology, Brooklands College, Heath Road, Weybridge, Surrey, KT13 8TT. Please note that I cannot undertake to reply to individual queries from readers however I will do my best to answer all questions from readers through the medium of this column.

## Hints and tips

One of the nice things about writing this column is that readers often include hints and tips in their letters. These are usually borne out of considerable experience and many are worthy of a wider audience. Hence, we end this instalment of *Circuit Surgery* with a quick round-up of those which have arrived in my post bag during the last six months:

**J. Holt** recommends sorting unmarked transistors into *npn* and *pnp* groups and marking the TO18 and TO5 cases with a small spot of enamel paint; red for *pnp* and blue for *npn* (does anyone else remember the "red spot" transistors of the 1950's?).

**Andrew Wilkinson** suggests that empty egg boxes make excellent storage "bins" for small components when building projects. Andrew also suggests storing nuts and bolts in see-through glass jars (rather than in drawers where they can't be seen).

Finally, **Jeremy Field** has found a novel use for unmarked plastic transistors, the base-emitter junction of which can be used as a Zener diode. Transistors of the "BC548/BC212 variety" can be used (whether they be *pnp* or *npn*) as shown in Fig. 6. The "Zener voltage" produced will normally be in the range 7V to 10V (around 8.5V being the most common).

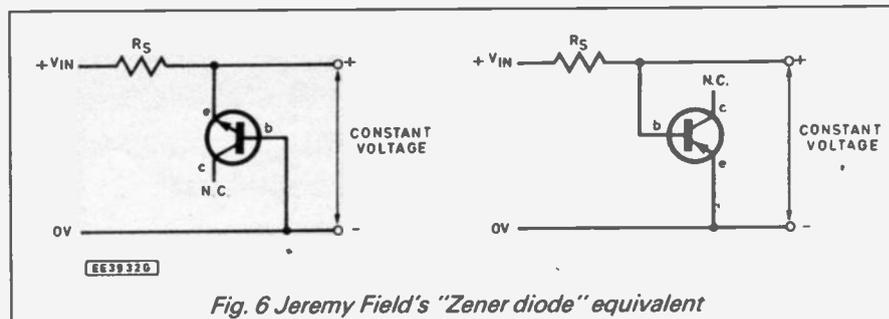


Fig. 6 Jeremy Field's "Zener diode" equivalent

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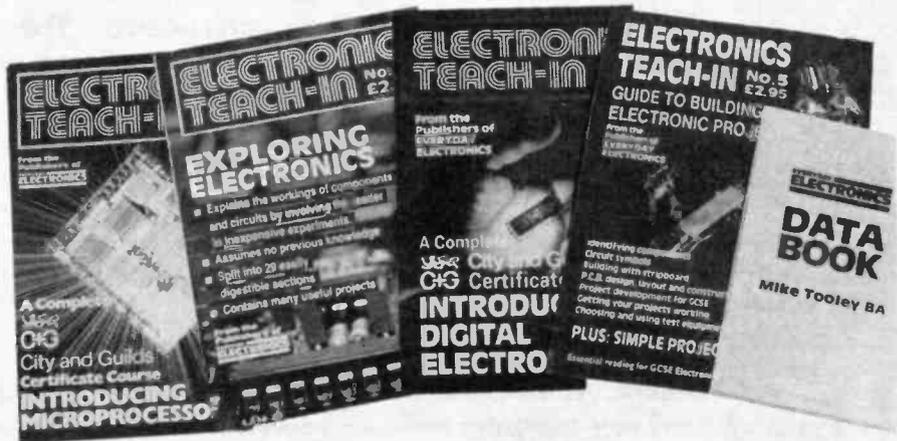
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 If you are more used to the black and white keys of a synth keyboard than the QWERTY keyboard of a computer, you may be understandably confused by the jargon and terminology bandied about by computer buffs. But fear not, setting up and using a computer-based music making system is not as difficult as you might think.  
 This book will help you learn the basics of computing, running applications programs, wiring up a MIDI system and using the system to good effect, in fact just about everything you need to know about hardware and the programs, with no previous knowledge of computing needed or assumed. This book will help you to choose the right components for a system to suit your personal needs, and equip you to exploit that system fully.  
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## COMPUTERS AND MUSIC.

See Computers section

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## GUIDE TO BUILDING ELECTRONIC PROJECTS

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See the first page of books - ELECTRONICS TEACH-IN No.5 - for full details.

## ELECTRONICS PROJECT BOOK

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See the first page of books for full details.

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Contains: Amplifiers - low level discrete and op-amp circuits, voltage and buffer amplifiers including d.c. types. Also low-noise audio and voltage controller amplifiers. Filters - high-pass, low-pass, 6, 12, and 24dB per octave types. Miscellaneous - i.c. power amplifiers, mixers, voltage and current regulators, etc.

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Filters play a vital part in almost all electronic circuits, yet many people believe that they are difficult to understand. This is probably because so many of the books on this topic are extremely mathematical. By contrast, this book deals with the subject in a non-mathematical way. It reviews the main types of filter, explaining in simple terms how each type works and how it is used.

The book also presents a dozen filter-based projects with applications in and around the home or in the constructor's workshop. These include a number of audio projects such as a rhythm sequencer and a multi-voiced electronic organ. Project descriptions include circuit diagrams, explanations of their operation, and detailed instructions for building them. A number of the projects are suitable to the beginner while others will be of interest to the more advanced constructor.

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Topics such as Boolean algebra and Karnaugh mapping are explained, demonstrated and used extensively, and more attention is paid to the subject of synchronous counters than to the simple but less important ripple counters.

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The main stumbling block for most would-be robot builders is the electronics to interface the computer to the motors, and the sensors which provide feedback from the robot to the computer. The purpose of this book is to explain and provide some relatively simple electronic circuits which bridge this gap.

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Ian R. Sinclair

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The devices, their operating principles and typical cir-

cuits are all dealt with in detail. The action of rectifiers and the reservoir capacitor is emphasised, and the subject of stabilisation is covered. The book includes some useful formulae for assessing the likely hum level of a conventional rectifier reservoir supply.

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## PRACTICAL ELECTRONIC DESIGN DATA

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A helpful major section covers the main kinds of components, including surface-mount devices. For each sort, it lists the most useful and readily available types complete with details of their electronic characteristics, pin-outs and other essential information. A special feature of this section are the easily followed charts and tables which advise the reader on how to select the best type of components for any particular purpose.

Basic electronic units are defined, backed up by a compendium of the most often required formulae.

There are five more extensive sections devoted to circuit design, covering analogue, digital, radio, display, and power supply circuits. Over 150 practical circuit diagrams cover a broad range of functions. The reader is shown how to adapt these basic designs to a variety of applications.

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AND S.W.L.S.

R. A. Penfold

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The circuits covered include: An aerial tuning unit; A simple active aerial; An add-on b.f.o. for portable sets; A wavetrapp for combat signals on spurious responses; An audio notch filter; A parametric equaliser; C.W. and S.S.B. audio filters; Simple noise limiters; A speech processor; A volume expander.

Other useful circuits include a crystal oscillator, and RTTY/C.W. tone decoder, and a RTTY serial to parallel converter. A full range of interesting and useful circuits for short wave enthusiasts.

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## AN INTRODUCTION TO AMATEUR RADIO

I. D. Poole

Amateur radio is a unique and fascinating hobby which has attracted thousands of people since it began at the turn of the century.

This book gives the newcomer a comprehensive and easy to understand guide through the subject so that the reader can gain the most from the hobby. It then remains an essential reference volume to be used time and again. Topics covered include the basic aspects of the hobby, such as operating procedures, jargon and setting up a station. Technical topics covered include propagation, receivers, transmitters and aerials etc.

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## SIMPLE SHORT WAVE RECEIVER CONSTRUCTION

R. A. Penfold

Short wave radio is a fascinating hobby, but one that seems to be regarded by many as an expensive pastime these days. In fact it is possible to pursue this hobby for a minimal monetary outlay if you are prepared to undertake a bit of d.i.y., and the receivers described in this book can all be built at low cost. All the sets are easy to construct, full wiring diagrams etc. are provided, and they are suitable for complete beginners. The receivers only require simple aerials, and do not need any complex alignment or other difficult setting up procedures.

The topics covered in this book include: The broadcast bands and their characteristics; The amateur bands and their characteristics; The propagation of radio signals; Simple aerials; Making an earth connection; Short wave crystal set; Simple t.r.f. receivers; Single sideband reception; Direct conversion receiver.

Contains everything you need to know in order to get started in this absorbing hobby.

88 pages **Order code BP275** £3.95

## AN INTRODUCTION TO SATELLITE TELEVISION

F. A. Wilson

As a definitive introduction to the subject this book is presented on two levels. For the absolute beginner or anyone thinking about purchasing or hiring a satellite TV system, the story is told as simply as such a complex one can be in the main text.

For the professional engineer, electronics enthusiast, student or others with technical backgrounds, there are numerous appendices backing up the main text with additional technical and scientific detail formulae, calculations, tables etc. There is also plenty for the DIY enthusiast with practical advice on choosing and installing the most problematic part of the system - the dish antenna.

104 pages **Order code BP195** £6.95

## AN INTRODUCTION TO AMATEUR COMMUNICATIONS SATELLITES

A. Pickford

Communications and broadcast satellites are normally inaccessible to individuals unless they are actively involved in their technicalities by working for organisations such as British Telecom, the various space agencies or military bodies, even those who possess a satellite television receiver system do not participate in the technical aspects of these highly technological systems.

There are a large number of amateur communications satellites in orbit around the world, traversing the globe continuously and they can be tracked and their signals received with relatively inexpensive equipment. This equipment can be connected to a home computer such as the BBC Micro or IBM compatible PCs, for the decoding of received signals.

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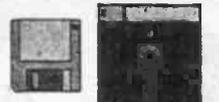
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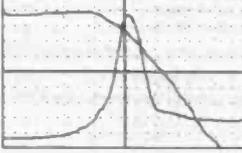
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# AUTO-SWITCH ADD-ON

**KEN TAYLOR**

*Prolong the operating life of your battery operated equipment with one of these two low-cost add-ons*

**H**ERE'S a simple device that should reduce your battery costs if you – or your family – often forget to switch off toys, radios, test equipment, or other battery operated items. It will do as much for saving PP3 type battery costs as the *Dry Cell Recharger* (Sept '91 – for back issues) did for single cells.

You may remember that the one common battery type that couldn't be recharged was the PP3 and it was advised that devices using 9 volt batteries shouldn't be purchased. But this battery is being used in more and more toys and equipment and if you want one of these there is little you can do.

This problem came to prominence when the author decided to purchase one of the small Digital Volt Meters – DVM's – on the market. Some have an *auto switch-off* facility but the one selected didn't have this. With the most conscientious intentions it was found difficult to remember to always return the rotary switch to the OFF position after each use and some hours or even days later the meter would be found

to be still switched on. Something was needed urgently.

## REQUIREMENT

For the DVM, a unit was required which was small enough that it could easily be attached to the meter, take a negligible extra current to the five milliamps taken by the meter and provide times of about two to 10 minutes. When it switched the unit off it should ideally also switch itself off and cease to take any current.

For radios, toys and other battery operated equipment, the switching current might be as high as 500 milliamps and the maximum time up to one hour. The voltages of these devices will also vary and the unit should cater for a range of about 5V to 12V. The unit should be small enough to be built into these larger devices.

## SIMPLE VERSION

The simplest version of the Auto-Switch Add-On Unit circuit, designed to operate the DVM, is shown in Fig. 1. Because the meter needs such a small current it was decided that a circuit using just a CMOS timer might meet the requirements.

However, a look at the timer specification revealed the first problem. The timer output in common with many i.c.s is capable of sinking a much higher current than it can source, therefore the circuit should connect the negative lead of the

DVM to the timer so that it will be switched on when the timer output is low. But in the monostable mode the timer output is normally low and goes high when timed out.

This circuit reverses the sequence being low whilst timing and going high when it has timed-out. So you will find the circuit slightly different to the one you would normally expect. Also there is a single push-button switch S1 to both reset and start the timing cycle.

There is no doubt that this circuit fulfils some of the requirements and would certainly pay for itself in battery saving for *low current* devices; but it also has some serious failings. Although while timing it only takes 170 microamps, it continues to take this current after it has switched off

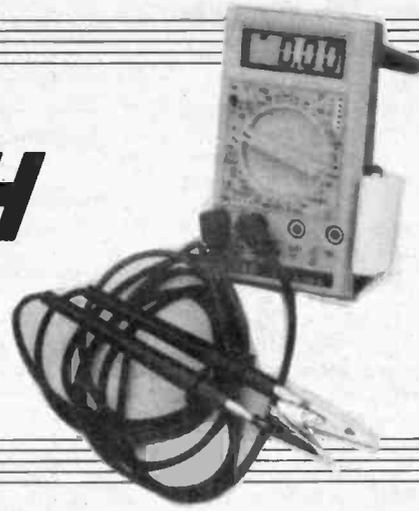
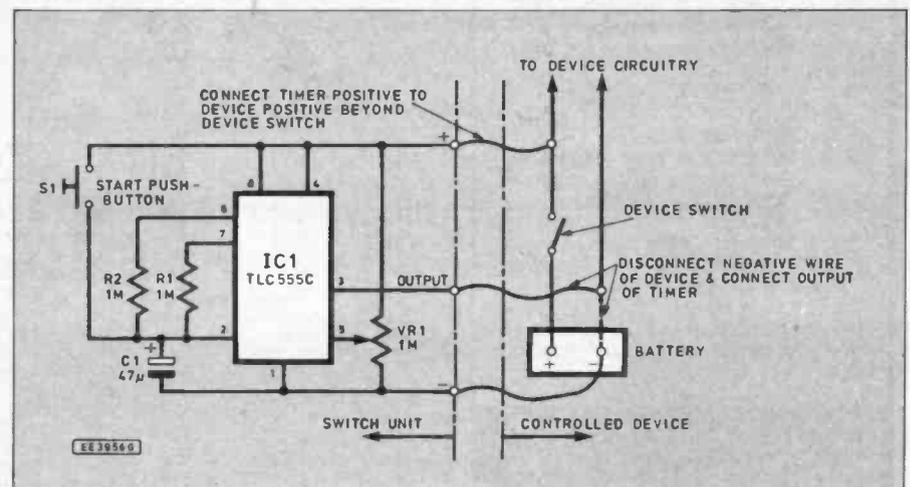


Fig. 1. Circuit diagram for the Simple Auto-Switch Add-On for low current devices.



## COMPONENTS

### Simple Version (Fig. 1)

#### Resistors

R1, R2 1M (2 off)  
All 0.25W 5% carbon film

See  
**SHOP  
TALK**  
Page

#### Potentiometer

VR1 1M enclosed vertical min.  
preset, lin.

#### Capacitor

C1 47µ sub-min. radial elect.,  
16V or more

#### Semiconductor

IC1 TLC555C CMOS timer i.c.

#### Miscellaneous

Miniature pushbutton switch; 8-pin d.i.l. socket; stripboard, 0.1in. matrix (size as required); connecting wire; solder, etc.

Approx cost  
guidance only

**£3**

the device. This current drain will continue until the device itself is switched off.

Another undesirable feature of this circuit is the internal resistance of the timer. At five milliamps there is a voltage drop of 0.072V (see Table 1) which doesn't look too bad. But this gives an internal resistance of 14.4 ohms that at 100 milliamps would drop 1.4V.

This drop on a supply of 9V could be quite serious and at lower voltages would be unacceptable. It could also give rise to distortion in an audio device unless decoupling measures were taken.

So although the circuit Fig. 1 could be useful in certain circumstances further improvements were necessary. These would keep all the good points, such as the low operating current, of the simple circuit but removing the undesirable features of internal resistance and lack of switch-off.

## IMPROVED CIRCUIT

One method of solving the problem with the first circuit would be to use a relay. But a normal relay requires too much current even if the "holding" current is reduced to a minimum.

A while ago the author produced a reed relay which by mounting a permanent magnet in a suitable position would latch ON when the coil current assisted the magnet and unlatch when the direction of the current was reversed. This type of relay needs no holding current and therefore would be ideally suitable for this application.

However, it was found to be very difficult to achieve reliable operation and so enquiries were made of the major suppliers to find a commercial equivalent. The relays that were obtained were found to operate on the same reed switch principle but they

Table 1: Operating Data for the Low Current Timers

		ICM 7555	TLC 555C
Operating voltage	Max	18 volts	18 volts
	Min	2 volts	2 volts
Operating current at 9 volts (in Fig. 1 Circuit)		0.12mA	0.17mA
Output current (Sink)	Max	100mA	100mA
Volt drop at 5mA (Sink)		0.117 volts	0.072 volts
Internal resistance		23 ohms	14.4 ohms

also had an improvement. They had two coils one for on and one for off and this greatly simplifies the switching.

The final circuit diagram for the Auto-Switch Add-On is shown in Fig. 2. Because the relay unlatches at the end of the timing period when the timer output goes low, the circuit looks more like the one you would expect for a 555 timer.

The timer IC1 is initiated by the Start pushbutton switch S1 which latches the relay by energising the start coil winding across pin 8 and pin 9. This closes the two relay contacts pins 12/16 and pins 1/5. The closing of the relay contacts at pin 12 and pin 16 completes the battery positive supply to the "equipment" or device switch.

Providing the device switch is ON the timer output goes high and the timer starts to time-out. At the end of this period the output goes low and energises the unlatch winding across pin 7 and pin 10. This opens both sets of contacts and the device being powered and the timer are switched off. Similarly all current ceases if the device switch is turned off; but then at switch-on the timer will restart without the button being pressed.

According to the specification for the timers no supply decoupling capacitor is needed. Despite this it has sometimes been found necessary when operating at five volts to fit one, shown dotted (C3) in Fig. 2, otherwise the relay drops out when the pressbutton switch is released.

There are 5V and 12V versions of the twin coil, latched relays available, covering the range 4V to 18V. This is virtually the same as a low current timer and defines the operating range of the unit.

The five volt relay should be used in a circuit with a voltage of 4V to 8V and the

12 volt relay for voltages of from 8V to 18V. While the lower voltage relay does not seem reliable below 4V the 12V one has worked reliably down to six volts this is well below the voltage at which a 9V battery would be replaced. The maximum switch rating for both of these relays is 1A which should cater for most requirements.

## MAKING TIME

There are two readily obtainable low current timers, ICM7555 and TLC555C. Both give the same time delay but differ slightly in some of the other parameters.

Table 1 compares these two types but basically the TLC555 is best for the simple circuit because it has the lowest internal resistance and the ICM7555 is best for the relay circuit, having a slightly lower operating current. There is very little difference in the price and both have the same operating voltage range.

Completed relay version (left) and a smaller version (right).

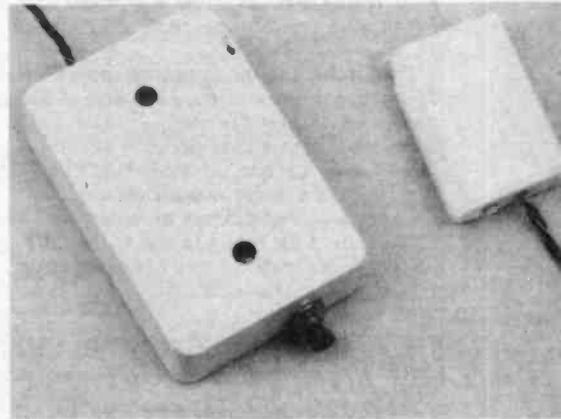
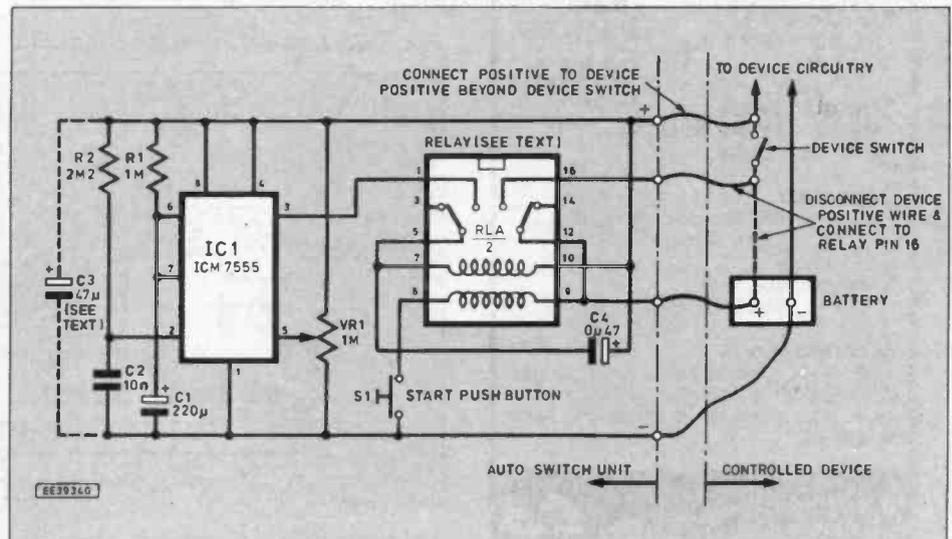


Fig. 2. Circuit diagram for the improved relay version of the Auto-Switch Add-On.



## COMPONENTS

### Relay Version (Fig. 2)

#### Resistors

- R1 1M
- R2 2M2
- All 0.25W 5% carbon film

See  
**SHOP  
TALK**  
Page

#### Potentiometer

- VR1 1M sub-min. enclosed horizontal preset, lin.

#### Capacitors

- C1 220µ sub-min. radial elect., 16V or more
- C2 10n ceramic
- C3 47µ sub-min. radial elect., 16V or more
- C4 0µ47 sub-min. radial elect., 16V or more

#### Semiconductor

- IC1 ICM7555 CMOS timer i.c.

#### Miscellaneous

- RLA 5V or 12V, twin coil (167 or 960 ohm), d.i.l. latching relay with double-pole changeover contacts

- S1 Min. pushbutton switch, press-to-make non-locking

Plastics case, size 75mm x 50mm x 25mm; stripboard 0.1in. matrix, size 18 strips x 24 holes; 8-pin d.i.l. socket; 16-pin d.i.l. socket; connecting wire; solder, etc.

Approx cost  
guidance only

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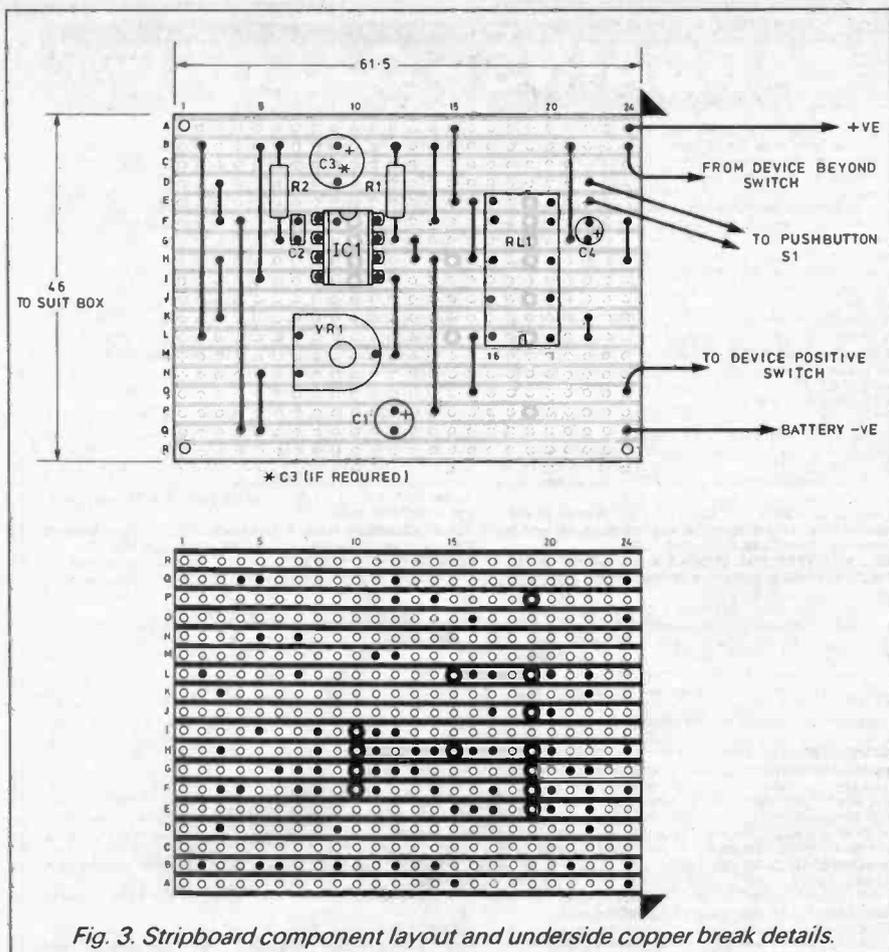


Fig. 3. Stripboard component layout and underside copper break details.

## CONSTRUCTION

No construction details or layout is given for the Simple Auto-Switch Add-On circuit (Fig. 1) as this unit will probably be tailored to suit the device to be powered. The problem with the relay version was to find a suitable container.

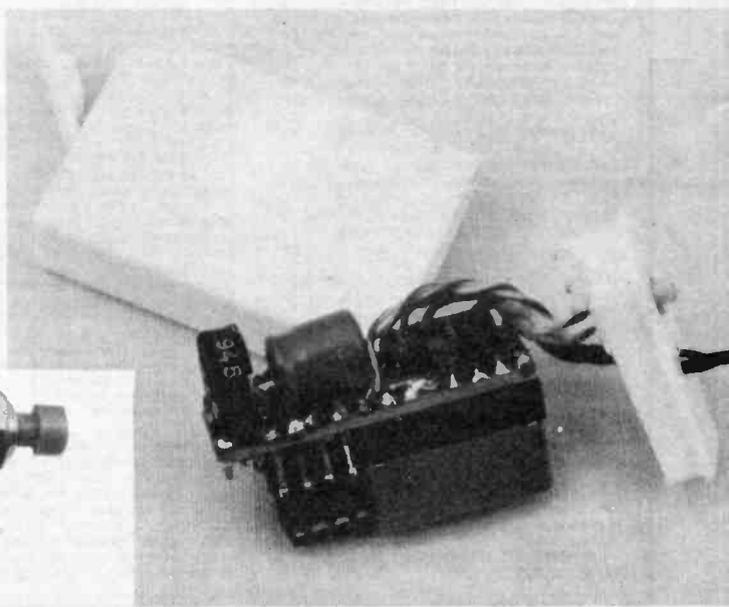
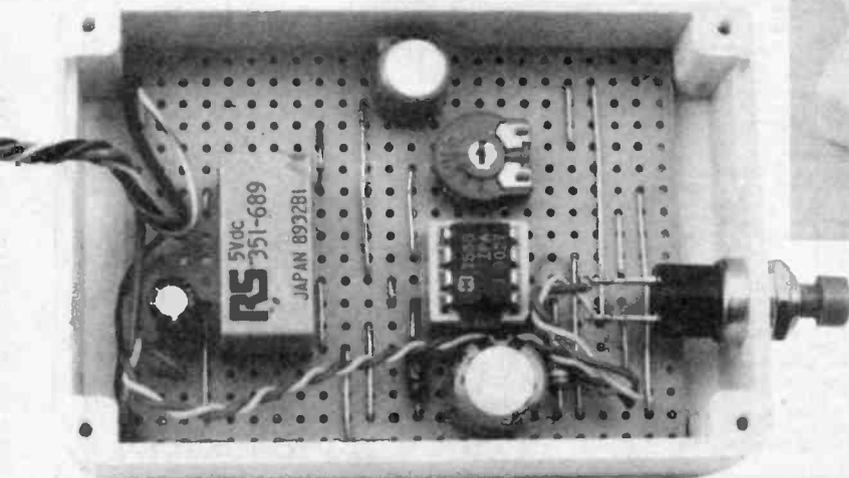
If there is plenty of room inside the case of the device to be controlled, or if a box is acceptable, the Relay version can be constructed on a piece of stripboard, size 18 strips x 24 holes. The stripboard component layout and details of the breaks in the underside copper tracks is shown in Fig. 3.

The unit which was designed for operation with the DVM was also built on a small piece of off-cut stripboard, size 14 strips by 4

holes, and was designed to fit into a "Hermesetas" sweetener box - see photographs. This was ideally suited as it was just large enough and has an opening gap at the front providing access to the time adjustment potentiometer. The copper strips on the board are cut *between* the two centre holes, along its length, to form a miniature "tag-board".

This is an example of what can be done to pack the components into the space available. Most of the components are small and in this unit the timer i.c. was fitted into a wire-wrap type socket which was soldered well clear of the stripboard to provide space underneath for the resistors and the 10nF capacitor C2.

*Two versions of the Auto-Switch Add-On. To enable the smaller version to fit into the tiny box, some of the components are mounted underneath a wire-wrap d.i.l. socket.*



At the time it was constructed a push-button switch small enough to fit the box couldn't be found and one salvaged from the front p.c.b. of an old Betamax video was used. However, this type of sub-miniature switch is now available if you need it. (See *Shop Talk*). This unit is attached to the DVM with double-sided adhesive tape.

While it is not essential, both the timer IC1 and the relay RL1 are fitted into d.i.l. sockets, enabling the soldering to be finished before the main components are fitted. This also enables the relay to be changed very easily if the unit is required to operate at a different voltage.

## IN-CIRCUIT

To fit the Auto-Switch to a piece of equipment it is only necessary to disconnect the equipment positive wire from the battery and connect it to board location O24 (pin 16 of the relay RL1). The positive (relay pins 9 and 12) and "ground" (-V) leads from stripboard points A24 and Q24 are then connected directly to the equipment battery.

A connection is made from the *circuit side* of the device/equipment On/Off switch to pin 10 of the relay at board point B24. This is not usually a problem. It is only necessary to search the device p.c.b. with a voltmeter until a contact is found where the full battery voltage is present, checking that this voltage is controlled by the device switch.

Often this will be pin 14 or pin 16 of an i.c. or the positive connection of an electrolytic capacitor. However, if it is not possible to make this connection, relay pin 16 can be joined to pin 10 and the timer will still work satisfactorily. The advantage of wiring the timer through the device switch is that when the device is switched off all current ceases: even the small current drain taken by the timer.

The component values shown in the circuits of Fig. 1 and Fig. 2 give approximate times of one to ten minutes. This is a useful range for equipment such as the DVM which in use has been found to be best with a time of about four minutes.

Those who have used the 555 timers before, will know that the timing is directly proportional to the value of the capacitor C1 and/or resistor R1. Therefore to double the time either the resistor value or the capacitor value should be doubled. □

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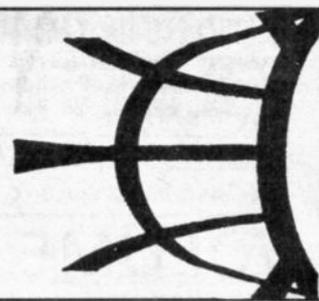
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# REPORTING

# AMATEUR RADIO

Tony Smith G4FAI



## NEW MORSE TEST

To obtain an amateur "A" licence one needs to pass a 12 words per minute Morse test, as well as the *Radio Amateurs' Examination*, and for many years the test has simply required candidates to send and receive plain language text and, separately, to send and receive groups of figures.

When the new Novice licence was introduced last year, with an "A" version requiring a 5 w.p.m. Morse ability, the Novice test had a new format. This required candidates to send and receive typical exchanges between amateurs, as in real-life contacts, including mixed text and figures and the use of regularly used abbreviations, Q-codes and procedural signals.

This style of test has proved so successful in preparing Novices to go on the air that it will also be used for the full 12 w.p.m. test as from 1 January 1993. Those who have been studying for the old test can still take it until 31 March 1993 but after that the new test will become compulsory.

Under the old system, many successful candidates were hardly better prepared to use Morse on the air than when they began to learn the code and, not surprisingly, most picked up their microphones and forgot all about Morse as soon as they got their licence. It is hoped that the new-style test will encourage more newcomers to try the code since they will at least be familiar with the basics of CW communication when they first get their licence.

As reported previously, it is not absolutely necessary to learn Morse to become a radio amateur. The "B" licence, whether Full or Novice, permits operation on the v.h.f. and u.h.f. bands without a Morse test and this is a popular way of coming into the hobby. This approach leaves open the option of taking the test later to extend operation into the h.f. bands.

## BUY BUT DON'T USE!

Radio scanners are increasing in popularity, providing reception over v.h.f. and u.h.f. frequencies together with control facilities well beyond those of older-style shortwave receivers. World band radios have some of the facilities of scanners and between them these two types of microprocessor equipped receiver provide an impressive means of listening to the entire spectrum of radio transmissions while some scanners are now available which cover the whole range in one set.

However, while such sets may be legally purchased by the general public, it is *not legal* to listen to *everything* that can be heard on them. Anyone can listen to

licensed radio broadcast stations, to radio amateurs and standard frequency transmissions but licenses, mostly not available to the public, are required to listen to other transmissions.

It seems to be a feature of the law in this country that while certain things may not be done legally, it is perfectly legal to obtain the means of doing them. Before CB was legalised there was no problem in obtaining illegal rigs.

Surveillance bugs are widely advertised today even though licences cannot be obtained for them. Non-BABT approved modems and other telephone equipment can be bought although it is not legal to use them, and there are many situations where the good citizen can be quite unaware that the *use* of the latest "goodie" he has just purchased could put him on the wrong side of the law.

## NEW BOOK

These musings have been brought on by the publication of a new book by the prolific Ian Poole who has produced an excellent introduction to yet another aspect of hobby radio. *An Introduction to Scanners and Scanning* covers the whole gamut of this activity although the author does make a point of referring to the legal position in the first chapter, including the fact that prosecutions have taken place for illegal listening.

For those thinking of obtaining a scanner or a world band radio this handy book (152 pages, 178mm + 108mm) will tell them about radio waves and how they travel; the types of transmissions they will be able to hear legally, and what they may not listen to without a licence; what facilities they will find on a typical scanner; what specifications to look for when choosing a set; and aerials suitable for different ranges of frequency.

There are some useful notes on listening to broadcast stations on the long, medium, shortwave and v.h.f. bands. A chapter on amateur radio provides a helpful summary of amateur activities from h.f. to u.h.f. plus some notes on procedure and jargon which will help a newcomer make sense of what he hears.

A similar chapter provides information about Citizen's Band Radio, including its background, informal procedures, and frequency bands. I was surprised to note that a CB licence is required to *listen* to CB transmissions. Three useful appendices provide a glossary of receiver terminology; a list of ITU callsign prefix allocations; and an amateur prefix list.

Covering a wide range of subjects and containing a lot of factual information, this book provides a helpful introduction to both world band listening and scanning over the v.h.f. and

u.h.f. range. Its style makes it particularly suitable for the beginner and it will also provide a useful reference facility for the more experienced. Written by I.D. Poole, G3YWX, the book is published by Bernard Babani (publishing) Ltd at £4.95, and is obtainable from the *EPE Direct Book Service*, code BP311.

Incidentally, a Radiocommunications Agency Information Sheet, *RA 169 Receive Only - Scanners etc.*, provides official advice on scanner use. This can be obtained free of charge by telephoning 071-215 2072 (24-hour answer-phone service).

## NO AMATEUR!

A specific example of illegal listening on legal equipment occurred recently in the case of the "radio ham" who recorded a cellular telephone conversation allegedly involving a member of the Royal Family, and sold the recording to a popular newspaper.

This was another of those instances, which I mentioned last month, when the term "ham" is applied indiscriminately by the press to anyone interested in radio whether they are licensed radio amateurs or not; and on this occasion there were some very critical remarks about radio amateurs in some of the papers together with ill-informed assumptions about activities outside the terms of the amateur licence.

It appears however that the person who made the recording was not a licensed amateur at all but a shortwave listener who at one time operated on CB before giving it up because he didn't like the language he heard.

The *W5YI Report*, an amateur radio newsletter published in the USA, telephoned this person who, according to their report, acknowledged that he had made the recording on January 4, 1990. He said that the 23 minute tape, subsequently heard by thousands of callers to a British Telecom 0898 number, was really two tapes spliced together to make one conversation. The other part was from a phone call intercepted by "a lady" on New Year's Day 1989. He knew that the recording was illegal and the paper had gone back on their agreement with him not to release his name.

It seems a pity that amateur radio, with its great tradition and background of public service, should have become involved in all this. At a recent Strategy Conference the Radio Society of Great Britain received a recommendation that it should adopt a professional approach to public relations. If it did that, perhaps there would be fewer inaccurate and harmful references in the media to amateur radio and to what it actually does.

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Pages	Issue	Pages	Issue
1-72	January	401-472	July
73-128	February	473-544	August
129-192	March	545-608	September
193-256	April	609-680	October
257-328	May	681-760	November
329-400	June	761-848	December

The No 1 Magazine for Electronic & Computer Projects

**VOLUME 21 INDEX**

**CONSTRUCTIONAL PROJECTS**

**NEXT MONTH  
 PE INDEX VOL. 28**

ALARM, GAS	484	LAPEL BADGE	86
ALARM, MICRO-SENSE	18	LCD ULTRASONIC TAPE MEASURE <i>by John Becker</i>	556
ALARM, VEHICLE WATCHDOG	342	LEAD TESTER, MIDI	792
ALARM, VIBRATION	742	L.E.D. TIE PULSER	370
ALPHA WAVE PROGRAMMER	42	L.E.D. VOLTMETER	810
ALTIMETER, EPE <i>by John Becker</i>	694	LIGHT, AUTO GARAGE	154
AMPLIFIER, CAMCORDER HEADPHONE	272	LIGHT BACKUP, BICYCLE	170
AMPLIFIER, CLASS-A HEADPHONE	412	LIGHT BOX, ARTWORK	532
AMPLIFIER, PERSONAL STEREO	720	LIGHT, EMERGENCY PLUG	138
AMPLIFIER, VERSATILE AUDIO	232	LIGHTHOUSE, BISHOP ROCK MODEL	55
ARTWORK LIGHT BOX <i>by Alan Winstanley</i>	532	LIGHTING UNIT, SOLAR POWERED	510
AUDIO AMPLIFIER <i>by Paul Henderson</i>	232	LIGHTS COLOUR SPECTRUM, CHRISTMAS	822
AUDIO TELESCOPE <i>by Robert Penfold</i>	202	LIGHTS, MODEL TRAFFIC	636
AUTO GARAGE LIGHT <i>by A. R. Winstanley</i>	154	LIGHTS-ON WARNING <i>by T. R. de Vaux-Balbirnie</i>	662
AUTO MELODY <i>by M. P. Horsey</i>	116	LOUDSPEAKER ENCLOSURE, BASS SUB-WOOFER	522
AUTO-SWITCH ADD-ON <i>by Key Taylor</i>	838	MAINS SWITCH	216
BADGE, LAPEL	86	MELODY MAKER, AUTO	116
BARGRAPH VOLTMETER	108	METRONOME, DUAL	492, 582
BASS SUB-WOOFER	522	MICRO-SENSE ALARM <i>by Jason Sharpe</i>	18
BATTERY MONITOR	108	MIDI LEAD TESTER <i>by Robert Penfold</i>	792
BATTERY-TO-MAINS INVERTER DAUGHTER BOARD <i>by Mark Daniels</i>	718, 797	MIND MACHINE PROGRAMMER	42
BBC INTERFACE <i>Mark Stuart</i>	176	MINI LAB <i>by Alan Winstanley &amp; Keith Dye</i>	736, 810
BIOFEEDBACK PROGRAMMER	42	MODEL RAILWAY AUTOMATIC SIGNAL <i>by Stuart Dando</i>	558
BISHOP ROCK LIGHTHOUSE, MODEL	55	MODEL SERIES, SIMPLE	55
CAMCORDER HEADPHONE AMPLIFIER <i>Chris Brown</i>	272	MOTOR DRIVER, STEPPER/INTERFACE	34
CAPACITANCE METER	620	NIGHTLIGHT, SOUND OPERATED	568
CAR BATTERY MONITOR	108	PERSONAL STEREO AMPLIFIER <i>by I. A. Duncombe</i>	720
CAR DUMMY ALARM	342	PHONE RINGER, STAGE	239
CAR LIGHTS-ON WARNING	662	PLUG LIGHT, EMERGENCY	138
CAR SCREENWASHER BOTTLE MONITOR	580	POWER SUPPLY, 5V to 15V SWITCH-MODE	355
CCD REVERB UNIT <i>by Robert Penfold</i>	338	PROGRAMMABLE TIMER <i>Stephen Holland</i>	100, 236, 288
CHARGE STATE MONITOR <i>by T. R. de Vaux-Balbirnie</i>	108	PROGRAMMER, MIND MACHINE	42
CHECKER, TRANSISTOR	12	PROGRAMMER, QUICK TEST	418
CHRISTMAS LIGHTS COLOUR SPECTRUM <i>by Mark Daniels</i>	822	QUICK PROM <i>by John Becker</i>	418
CLASS-A HEADPHONE AMPLIFIER <i>by P. Henderson</i>	412	QUICKTEST <i>by Mark Daniels</i>	596, 724
COMBINATION SWITCH <i>by Max Horsey</i>	814	RAILWAY AUTOMATIC SIGNALLING, MODEL	588
CONTINUITY TESTER	596, 724	REACTION TIMER <i>by T. R. de Vaux-Balbirnie</i>	710
CONTINUITY TESTER, SONIC	206	RECHARGEABLE TORCH	138
CRICKET GAME <i>by Steve Knight</i>	443	REMOTE CONTROL, UNIVERSAL INFRA-RED	774
CYCLE LIGHT BACK-UP <i>by T. R. de Vaux-Balbirnie</i>	170	REVERBATION UNIT, CCD	338
DEMONSTRATION BOARD	736, 810	ROBOT SERVO INTERFACE	382
DRILL CHARGER/PSU, 12V	314	SERVO INTERFACE, DIGITAL	382
DUAL METRONOME <i>by R. M. Worthington</i>	492, 582	SOLAR POWERED LIGHTING UNIT <i>by T. R. de Vaux-Balbirnie</i>	510
EASY SWITCH <i>by T. R. de Vaux-Balbirnie</i>	216	SONIC CONTINUITY TESTER <i>by Mark Daniels</i>	206
ECHO UNIT	338	SOUND OPERATED NIGHTLIGHT <i>by Eleanor Maitland</i>	568
ECONOMY SEVEN TIMER <i>by Chris Brown</i>	142, 236	SPECTRA-LITE <i>by Chris Walker</i>	86
EMERGENCY PLUG-LIGHT <i>by Ivor Sadler</i>	118	SPINNING HEART <i>by Rafe Aldridge</i>	98
EXPERIMENTAL WEIGHING SCALE <i>by Chris Walker</i>	266	STAGE TELEPHONE RINGER	239
EXTENDED RANGE CAPACITANCE METER <i>by Steve Knight</i>	620	STEPPER MOTOR DRIVER/INTERFACE <i>by Mark Stuart</i>	34
GAME, CRICKET	443	STEREO AMPLIFIER, PERSONAL	720
GAME, IT'S A KNOCKOUT	280	STRAIN GAUGE WEIGHING SCALE	266
GARAGE LIGHT, AUTO	152	SUB-WOOFER <i>by Paul Henderson</i>	522
GARDEN HOSEPIPE CONTROLLER <i>by T. R. de Vaux-Balbirnie</i>	154	SWITCH, ADD-ON, AUTO	838
GAS ALARM <i>by Robert Penfold</i>	484	SWITCH, COMBINATION	814
HEADPHONE CLASS-A AMPLIFIER	412	SWITCH, EASY	216
HEADPHONE, STEREO AMPLIFIER	272	SWITCH-MODE POWER SUPPLY <i>by Steve Knight</i>	355
INFRA-RED REMOTE CONTROL, UNIVERSAL	774	SWITCH, WHISTLE	640
INTERCOM, VERSATILE	646	TAPE MEASURE, LCD ULTRASONIC	556
INTERFACE, BBC	176	TEACH-IN '93 MINI LAB	736, 810
INTERFACE, DIGITAL SERVO	382	TELEPHONE RINGER <i>by Chris Walker</i>	239
INTERFACE, QUICK PROM	418	TELESCOPE, AUDIO	202
INTERFACE/STEPPER MOTOR DRIVER	34	TELESOUND <i>by Gary Calland</i>	82
IT'S A KNOCKOUT GAME <i>by Roy Bebbington</i>	280	TESTER, MIDI LEAD	792
INVERTER DAUGHTER BOARD, BATTERY-TO-MAINS	718, 797	TESTER, SONIC CONTINUITY	206
		TIE PULSER <i>Max Horsey</i>	370
		TIMER, ECONOMY SEVEN	142, 236

TIMER, PROGRAMMABLE	100, 236, 288	VERSATILE INTERCOM <i>by I. A. Duncombe</i>	646
TIMER, REACTION	710	VIBRATION ALARM <i>by M. G. Argent</i>	742
TIMER, UV EXPOSURE	437	VOLTMETER, BARGRAPH	108
TRAFFIC LIGHT SYSTEM <i>by J. Hewes</i>	636	VOLTMETER, L.E.D.	810
TRANSISTOR CHECKER <i>by Steve Knight</i>	12	WASHER BOTTLE <i>by Alan Winstanley</i>	580
UNIVERSAL INFRA-RED REMOTE CONTROL <i>by Chris Walker</i>	774	WEIGHING SCALE, EXPERIMENTAL	266
ULTRASONIC TAPE MEASURE, LCD	556	WHISTLE SWITCH <i>by Steven Holland</i>	640
UV EXPOSURE TIMER <i>by Alan Winstanley</i>	437	5V to 15V SWITCH-MODE POWER SUPPLY	355
VALENTINE'S DAY HEART	98	8-CHANNEL SERIAL PORT INTERFACE	382
VEHICLE WATCHDOG <i>by Chris Walker</i>	342	12V DRILL CHARGER/PSU <i>by Mark Daniels</i>	314
		80 WATT AUDIO AMPLIFIER	232

### SPECIAL SERIES

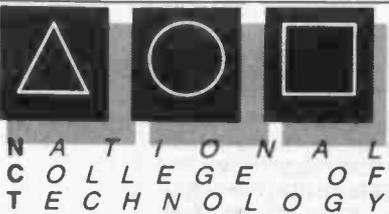
ACTUALLY DOING IT <i>by Robert Penfold</i>	106, 174, 244, 306, 368, 450, 500, 584, 634, 740, 798	Telesound Baby Monitor	286
ALTERNATIVE ENERGY <i>by T. R. de Vaux-Balbirnie</i>	504, 562, 628, 714, 786	Timer Software	627
1 - Introduction	504	Versatile Power Supply	423
2 - Fossil Fuels; Wind Power	562	Video Signal Booster	627
3 - More on Wind Power	628	30V Power Supply	744
4 - Power from Water	714	INFORMATION TECHNOLOGY AND THE NATIONAL CURRICULUM <i>by T. R. de Vaux-Balbirnie</i>	22, 92, 146, 210, 274, 426, 516, 574, 628, 656
5 - Alternative Energy Sources	786	3 - Information Storage	22
AMATEUR RADIO <i>by Tony Smith G4FAI</i>	61, 119, 168, 238, 324, 396, 468, 540, 604, 676, 756, 842	4 - Computers Past and Present; Microelectronics	92
CIRCUIT SURGERY <i>by Mike Tooley</i>	228, 286, 346, 422, 572, 626, 744, 828	5 - Microelectronic Experiments	146
Analogue-to-Digital Conversion	423	6 - Circuit Symbols and Logic	210
Audio Compressor	744	7 - Analogue and Digital Conversion	274
Bargraph Display	573	8 - Telecommunications	374
Basic Logic Functions	424	9 - Logic Systems	426
Capacitance Adapter	346	10 - Computer Monitoring and Communications Systems	516
Constant Level	573	11 - Problem solving and Feedback	574
Digital Audio	745	12 - Ohms law and Power Calculations	656
Fuses	626, 828	INTERFACE <i>by Robert Penfold</i>	50, 90, 160, 224, 312, 350, 458, 528, 592, 666, 747, 820
Go/No-Go Transistor Tester	228, 346	MAGNETIC RECORDING <i>by Vivian Capel</i>	52, 112, 162
Hints and Tips	828	4 - Head Driver Circuits	52
Linear Scale Resistance Measurement	286	5 - Recording Tape	112
LM380	287	6 - Recording Tape	162
Logical Choice	287	MAKING YOUR OWN PCBs <i>by Alan Winstanley</i>	289, 363, 433
Low Battery Indicator	347	1 - Introducing the P.C.B. (Supplement)	289
Modern Connections	745	2 - Ultra-Violet Processing Techniques	363
Pin Connections	626	3 - Originating Your Own Artwork	433
Resistance range Extender	287	SIMPLE MODEL SERIES <i>by Owen Bishop</i>	55
Safe Power Supplies	828	TEACH-IN '93 <i>by Alan Winstanley &amp; Keith Dye</i>	726, 800
SCART Connector	572	1 - Introduction and the Resistor	726
Signal-tracing Probes	828	2 - Alternating Current and the Diode	800
Simple A.M. Tuner	423		
Soldering Iron Standby Circuit	228		
Synchronising Engine Speed	422		

### GENERAL FEATURES

ABC OF R.I.A.A. <i>by W. F. Harm</i>	380	NEW TECHNOLOGY <i>by Ian Poole</i>	784
ANIMATED CIRCUITS FOR ELECTRONICS <i>by Mike Tooley</i>	166	NEWS	40, 96, 152, 230, 310, 352, 440, 514, 583, 643, 702, 782
CROSSWORD/WORDSEARCH <i>by Malcolm Argent</i>	32	NEXT 21 YEARS <i>by Barry Fox</i>	704
DOWN TO EARTH <i>by George Hylton</i>	182, 246, 388, 461, 590, 668-	PLEASE TAKE NOTE	181, 236, 582, 724, 797
EDITORIAL	11, 81, 137, 201, 265, 337, 411, 483, 555, 619, 693	Dual Metronome	582
EUROPEAN SHOW REPORT <i>by Barry Fox</i>	28	Economy Seven Timer	236
FOR YOUR ENTERTAINMENT <i>by Barry Fox</i>	104, 144, 226, 308, 348, 416, 502, 594, 645	MARC System Chip-Encoder	181
FOX REPORT <i>by Barry Fox</i>	812	Programmable Timer	236, 288
HOME BASE <i>by Terry Pinnell</i>	796	Quicktest	724
INSIDE THE MINI DISC <i>by Ian Graham</i>	498	Inverter - Daughter Board	797
LETTERS	223, 309, 389, 462, 490, 567, 669	SHOPTALK <i>by David Barrington</i>	40, 103, 181, 236, 288, 351, 459, 501, 566, 660, 724, 797
		STRAIN GAUGES <i>by Chris Walker</i>	220
		TOMORROW'S TECHNOLOGY <i>by Ian Poole</i>	705
		WORDSEARCH/CROSSWORD	32

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COMPONENTS CATALOGUE (Cricklewood '92)	banded with JAN '92	MAKING YOUR OWN PCBs (Supplement) <i>by Alan Winstanley</i>	289
COMPONENTS CATALOGUE (32-page Greenweld Spring '92)	between 224/225	* PRINTED CIRCUIT BOARD SERVICE	66, 124, 188, 252, 322, 394, 466, 538, 602, 674, 745, 835
COMPONENTS CATALOGUE (32-page Greenweld Summer '92)	between 436/437	ELECTRONICS VIDEOS	390, 442, 509, 579, 670, 750, 830



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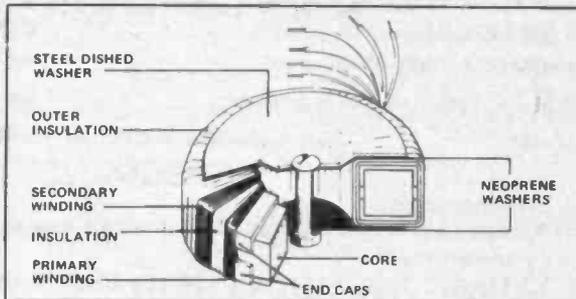


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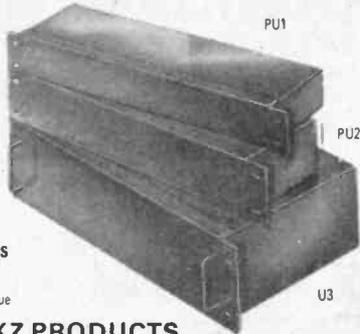
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## ADVERTISERS INDEX

AKULA SYSTEMS.....	834	MAGENTA ELECTRONICS.....	770/771
AUTONA.....	813	MAILTECH.....	766
N. R. BARDWELL.....	846	MARLIN ELECTRONICS Cover (iv)	647
R. BARTLETT.....	818	MARAPET.....	791
BK ELECTRONICS.....	Cover (iii)	MARCO TRADING.....	846
BLB ELECTRONICS.....	818	MAURITRON.....	846
BRIAN J. REED.....	848	M&B ELECT. SUPPLIES.....	841
BULL ELECTRICAL.....	Cover (ii)	MODERN AMPS.....	834
CAMBRIDGE COMP.....	818	MODERN ELECTRONICS.....	836/837
SCIENCE.....	765	NATIONAL COLLEGE OF TECH.....	845
CIRKIT DISTRIBUTION.....	848	NUMBER ONE SYSTEMS.....	781
COMPELEC.....	797	OMNI ELECTRONICS.....	845
CRICKLEWOOD ELECT.....	785	PENTAGRAM ELECT. DESIGNS.....	790
CROTECH INSTRUMENTS.....	848	PICO TECHNOLOGY.....	827
CR SUPPLY COMPANY.....	763	RACKZ PRODUCTS.....	848
DISPLAY ELECTRONICS.....	813	RADIO & TV COMPONENTS.....	764
ELECTRONIC DESIGN.....	772	SETRAX CAE.....	785
ESR ELECTRONIC COMP.....	769	SERVICE TRADING CO.....	845
GREENWELD ELECTRONICS.....	768	SHERWOOD ELECTRONICS.....	848
HART ELECTRONIC KITS.....	834	SMART HOUSE SYSTEMS.....	813
HENRY'S AUDIO ELECTRONICS.....	847	STEWART OF READING.....	834
HESING TECHNOLOGY.....	847	SUMA DESIGNS.....	762
ICS.....	847	TECHNICAL INFO. SERVICES.....	847
JAYTEE ELECTRONIC SERVICES.....	845,847	TSIEN.....	819
JPG ELECTRONICS.....	847		
LABCENTER ELECT.....	767		

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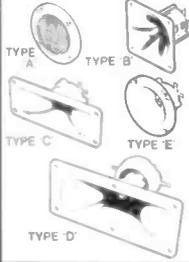


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