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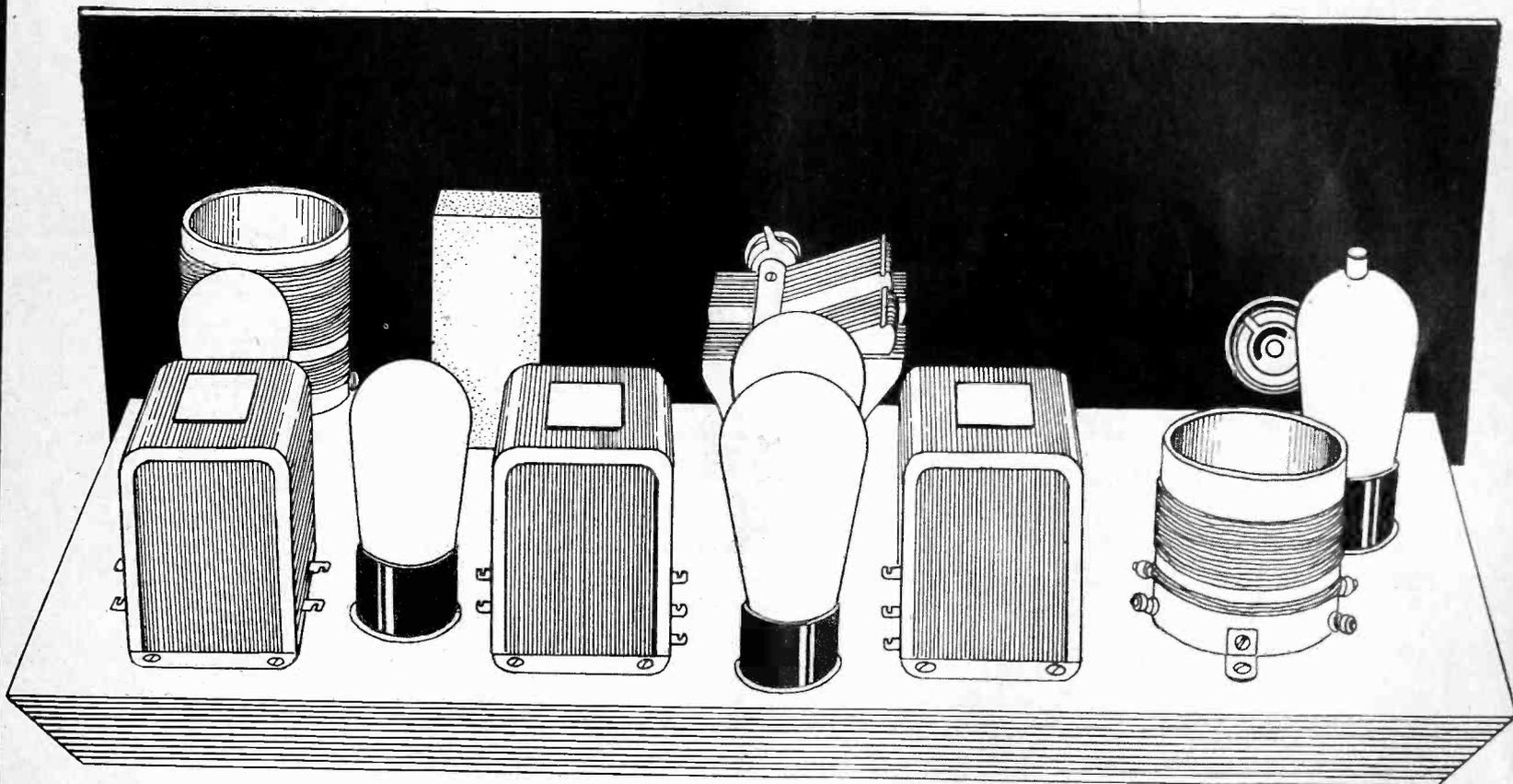
The First and Only National Radio Weekly

376th Consecutive Issue—Eighth Year

AC PUSH-PULL DIAMOND

CONTROL OF VOLUME AND SENSITIVITY

NEW DECIBEL MEASUREMENTS



The AC Screen Grid Diamond of the Air, shown for use with a separate B supply.
See article on page 3.

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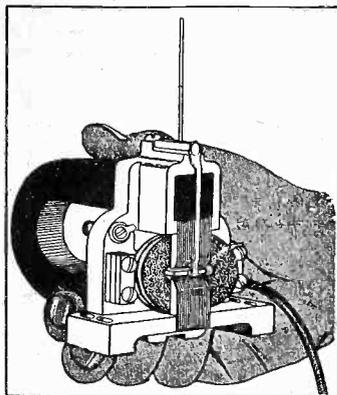
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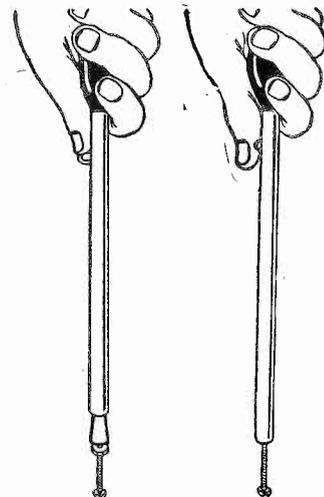
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145 WEST 45TH ST., N. Y. CITY



Vol. XV, No. 12 Whole No. 376
 June 8th, 1929
 15c per Copy, \$6.00 per Year
 [Entered as second-class matter, March, 1922, at the Post Office at New York, N. Y., under Act of March, 1879.]

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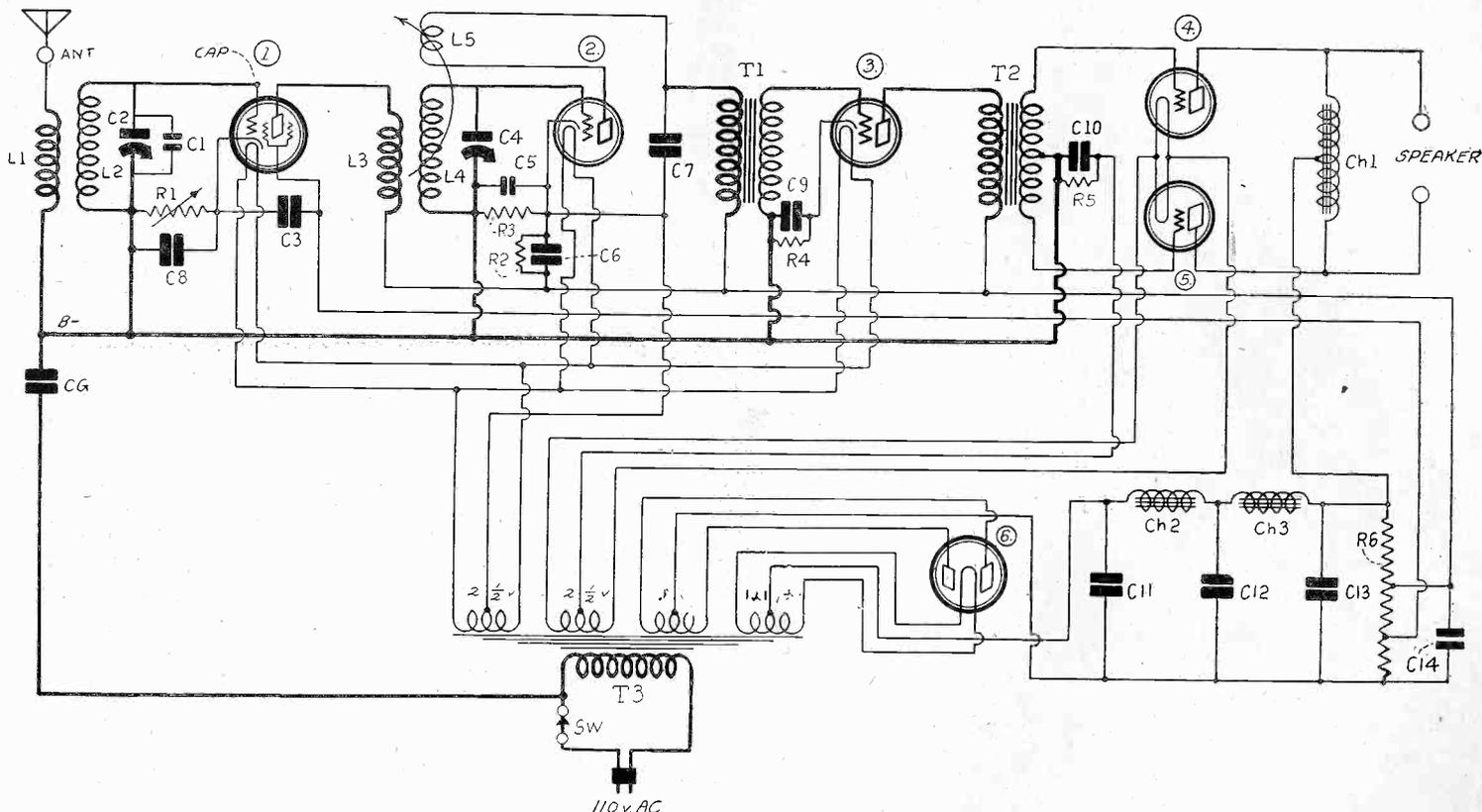
A Weekly Paper published by Hennessy Radio Publication Corporation, from Publication Office, 145 West 45th Street, New York, N. Y.
 Phone BRyant 0558 and 0559 (Just East of Broadway)

The Push-Pull Diamond

AC Model with Power Detector and with 245s in Output

By Herman Bernard

Managing Editor



TWO 245 TUBES IN PUSH-PULL CONSTITUTE THE OUTPUT OF THE 5-TUBE AC DIAMOND OF THE AIR.

TWO 245s have become the most popular pair for push-pull. The Diamond of the Air is shown in a five-tube AC model so as to embody this push-pull output. The design in Fig. 1 is for a com-

plete table model on a 12x20" baseboard or metal subpanel.

If you have a B supply that provides 300 volts maximum you may construct the receiver (include the filament trans-

former), and obtain the B voltages from your present supply. The front cover illustration shows such a model.

The inclusion of push-pull is not for increasing the volume, as this remains

LIST OF PARTS

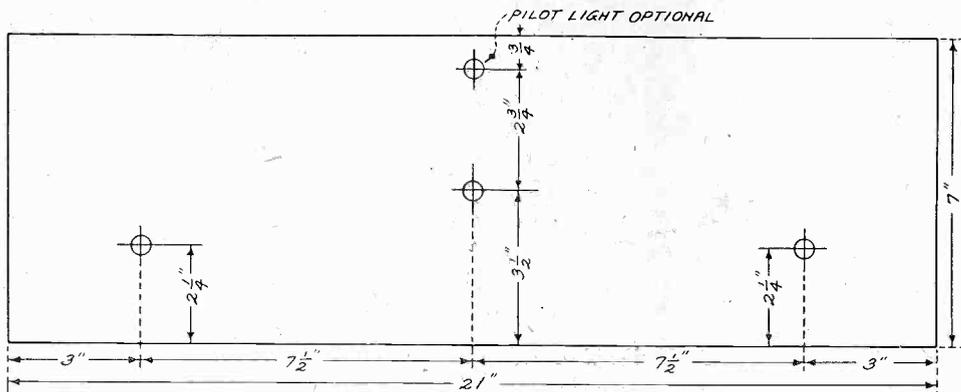
L1, L2—One antenna coil (Cat. AC5).
 L3, L4, L5—One screen grid three-circuit tuner (Cat. SGT5).
 C1—One Hammarlund equalizer, 70 mmfd.
 C2, C4—One Hammarlund 2 gang condenser, each section .0005 mfd. (Cat. MLD23).
 C3, C5, C6, C8, C9—Six Aerovox .02 mfd. fixed condensers.
 C7—One Aerovox .0005 mfd. fixed condenser.
 C10—One Aerovox 4 mfd. bypass condenser.
 C11, C12, C13, C14—Mershon 8-18-18-8.

R1—One Electrad Royalty variable resistor, 5,000 ohms, with 110 volt AC switch.
 R2, R3—One 20,000 ohm Electrad resistor type B (with 3 terminals).
 R4—One 1,000 ohm Electrad resistance strip.
 R5—One 800 ohm Electrad resistor type B.
 R —One Aerovox Pyrohm, type A.
 T1—One National A100 audio transformer.
 T2—One National push-pull input transformer.
 T3—One power transformer (Guaranty Radio Goods Co.).

Ch1—One push-pull output choke (Guaranty Radio Goods Co.).
 Ch2, Ch3—One Silver-Marshall Unichoke 331.
 Ant., Speaker—Three binding posts.
 (1)—One 224 tube; (2), (3), two 227 tubes; (4), (5), two 245 tubes; (6), one 280 tube.
 One 7x21" front panel.
 One 12x20" baseboard with three UY sockets and three UX sockets; or one metal subpanel with sockets affixed.
 One flat type dial.
 Two knobs.
 One roll Corwico Braidite.
 One pilot light, with bracket (optional).

Pointers on Push-Pull

Biasing Analyzed, Impedance Output Explained



DIMENSIONS FOR THE 7 X 21" FRONT PANEL USED FOR THE PUSH-PULL DIAMOND OF THE AIR. THE VOLUME CONTROL R1 AND SWITCH SW, CONSISTING OF A SINGLE COMPOSITE INSTRUMENT, ARE MOUNTED AT LEFT. THE TUNING CONDENSER IS AT CENTER, 3 1/2" UP. THE PILOT LIGHT, IF USED, IS ABOVE THE CONDENSER. AT RIGHT THE THREE CIRCUIT TUNER IS MOUNTED UPSIDE DOWN, THAT IS, WITH TICKLER COIL TOWARD THE SUBPANEL. THREE HOLES EQUIDISTANT AT BOTTOM MAY BE USED FOR ANCHORAGE TO THE SUBPANEL.

(Continued from preceding page)

about the same, but for improving the quality, since push-pull tends to reduce or eliminate extraneous noises by stifling the even order of harmonics.

Power Detection

The tuner is approximately the same as heretofore, except that the volume control is put in the grid biasing circuit of the 224 RF tube, and the power detector's effectiveness is heightened by passing some extra plate current through a lower biasing resistor. The bleeder and the biaser are one resistor, with three terminals, so that one extreme terminal goes to B plus 180 volts, the next one to the cathode and the other extreme terminal to B minus. The third slider is placed near the B minus end. An Electrad wire-wound resistor, type B, 20,000 ohms, is used, and about 2,000 ohms are used for bias and 18,000 ohms for bleeding.

Then the current drawn by the bleeder is passed through the biasing part of the resistor, avoiding the necessity of high reduction of precious plate current in the detector by use of a high value of biasing resistor, which would have to be on the order of 50,000 ohms otherwise.

No ground need be provided externally, as the filament transformer (or, in the case of the complete table model, the power transformer) has a self-grounding effect, and in addition the B minus lead is grounded to the one side of the AC line through a .02 mfd. fixed condenser, CG. The grounding effect through this condenser may not be material unless the plug is inserted in the convenience outlet in the right way. Reverse the plug to determine which connection provides louder signals, then adopt this connection permanently.

Output Impedance

The remaining novelty of the circuit is the push-pull output impedance. This is used instead of an output transformer, because it is more nearly suitable to various types of speakers, whether dynamic or magnetic. The speaker is connected to the two plates, and no current should flow through the speaker, because the flow is through the impedance in opposite directions from its center-tap. The direct current being equal and opposite, it flows

only through a circuit embodying half of the impedance, while none flows across the extreme terminals to which the speaker is connected.

Since magnetic speakers have a relatively high impedance, this output is suitable indeed for them. Also, dynamic speakers have output transformers built in, and these have a high impedance primary but a low impedance secondary, therefore the output serves this purpose well, also.

The biasing resistor in the last stage should be half the resistance of one used for a single-sided output because the plate current passed through it is twice as great. A value of 800 ohms fully takes care of the maximum required bias. But it must be a resistor easily capable of carrying 64 milliamperes. A 5-watt resistor provides an ample margin of safety.

The B Voltages

The B voltages are about 300 maximum, 180 for all plates save the last, and about 75 volts for the screen grid of the 224 tube. This screen grid voltage may be reduced if oscillation trouble arises.

Not all of the 300 volts is effective on the plate of the push-pull output tubes, on account of the effect that the voltage drop in the biasing resistor has on the regulation. As a separate biasing resistor is used for the last stage, the reduction in voltage from the total is not absolute, but relative. However, the voltage actually obtained is never more than a little more than the 250 volts recommended, and any slight excess is tolerable, and is taken care of automatically, by biasing for the higher voltage, since the current is increased through the biasing resistor.

Large capacity condensers of the electrolytic type are used in the B supply. The capacities are, from left to right, 8 mfd., 18 mfd. and 18 mfd., with another 8 mfd. going from the 180-volt tap to B minus. The mica condenser from screen grid to cathode of the 224 tube takes care of the bypassing for the screen grid current. On the other hand, the plate current of three tubes is obtained from the 180-volt post, hence the second 8 mfd. condenser is connected there. The four condensers are in one small copper can. The smaller capacity (8 and 8) are

farther from the periphery than the 18 and 18.

Filament Windings

The filament windings are three: 2 1/2 volts upper, which will stand 9 amperes and to which the tuner tubes and first audio are connected; 5 volts at center for the filament of the 280 rectifier, and 2 1/2 volts, bottom, for the two 245 tubes used in push-pull stage. All windings are center tapped. The top winding's center may be connected to B minus (not shown), to maintain the heater negative in respect to the cathode, and thus prevent the heater from acting as a fifth element.

A separate winding is used for the push-pull tubes' filaments, although the voltage is 2 1/2 also, because a common winding would make the heaters of the three preceding tubes positive in respect to the cathodes, to the extent of the negative bias on the last push-pull pair. Thus the heater might become an electron-attractor, and function as a sort of adventitious plate, which would be objectionable. Hum and instability might arise, also, particularly in respect to the screen grid tube, short life would result.

Absence of Shielding

No shielding is used in the tuner, because only one screen grid stage is incorporated, and the amplification obtained from the tube is maintained within workable limits.

The primary of the three-circuit tuner, nevertheless, has a fairly large number of turns, thus offering a higher impedance than the primary of tuners used with general purpose tubes. It is not possible to obtain the results of high sensitivity unless this primary impedance is larger than the ordinary.

The antenna coil has a small primary, to maintain high selectivity. On a 2 1/2" diameter tubing, 2 1/2" high, wind 6 to 8 turns for L1, and 48 turns for L2, separating the windings by 1/4". For L3 wind 24 turns, leave 1/4", then for L4 wind 48 turns. The diameter is again 2 1/2", but the coil form may be a little higher, around 3". All wire is No. 24 single or double silk covered. The tickler L5 consists of 30 turns on a 1 3/4" diameter, using No. 24 wire or smaller, half the number of turns being placed to one side, half to the other, so that the rotor shaft may penetrate the tickler coil far enough to hold it, and not be obstructed by any of the turns.

Equalized Tuning

The circuits may not tune exactly alike unless a trimmer is placed across the first tuning condenser. Tune in a low wavelength station, say, around 250 meters, with the trimmer set at minimum capacity. This little condenser is of the book type and is adjustable by a screwdriver, but you'd better use a wooden dowel fashioned at the end in screwdriver style, to avoid capacity effects obtained from a metal screwdriver making the adjustment difficult. Turn the trimmer until the station comes in loudest. A slight readjustment of the tuning dial may be necessary. If so, try the trimmer again, always striving for maximum volume. When that is attained, the circuit is set permanently, and resonance will result throughout.

[Other illustrations on front cover]

(Part II next week, issue of June 15th)

Sensitivity and Volume

How Both May Be Well Controlled at the Same Time

By Bruce Deeming

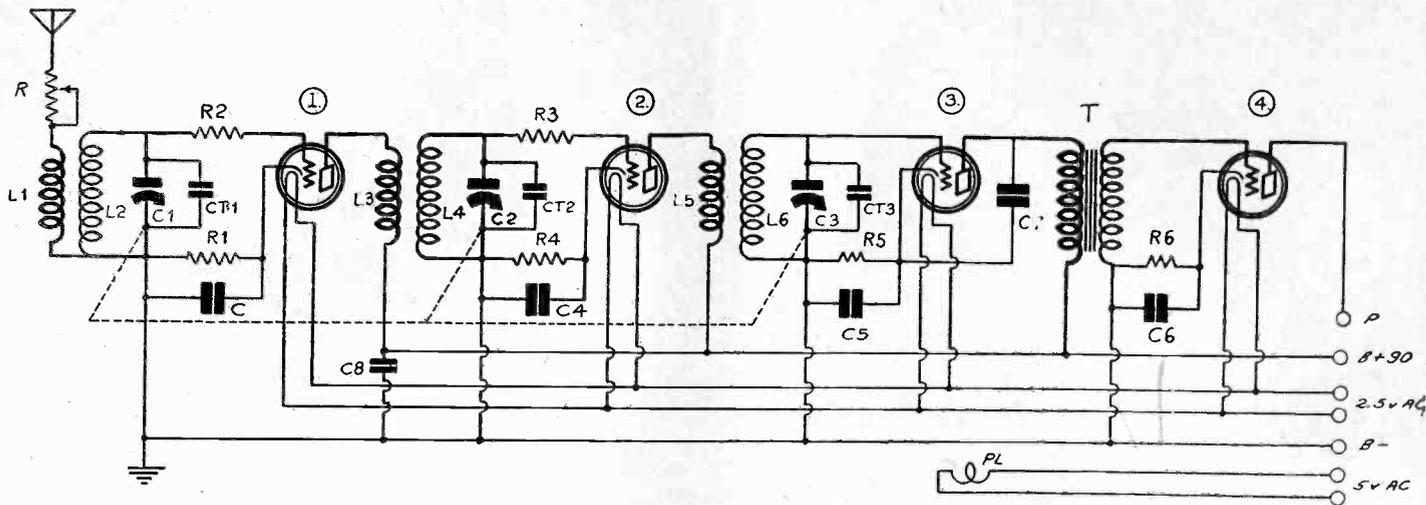


FIG. 1

RECEIVERS CONTAINING HEATER TYPE TUBES CAN BEST BE CONTROLLED AS TO SENSITIVITY AND VOLUME BY MEANS OF VARIABLE GRID BIAS RESISTORS, PLACED IN THE POSITIONS R1 AND R4. THIS IS A LITTLE BETTER THAN THE AERIAL RESISTOR METHOD SHOWN AS R.

STABILIZATION of radio frequency amplifiers and control of sensitivity are two questions which are continually in the mind of radio fans. These questions do no longer present a problem in battery sets, for the filament current in the radio frequency tubes can be changed over any desired range with instantaneous response.

When heater type tubes are used they are problems because the filament current cannot be changed readily, and if it is changed the response is not immediate. For example, suppose the volume is too great so that it becomes necessary to reduce it, that is, to reduce the sensitivity of the receiver. The filament current is reduced by a certain amount. It makes no difference by what means it is reduced—reduction in the primary voltage supplying the heating transformer, or reducing the secondary voltage by means of a resistor. Whatever be the method used, the response is slow. It may take three-quarters of a minute before any change is noticed.

Overshooting the Mark

The amount by which the filament current is changed is pure guess. When a certain change has been made, say downward, the volume will gradually reduce. When a stable state has been reached, the volume may be entirely too low. The guess was a poor one. So it becomes necessary to increase the volume. There is another wait and then the volume may be too loud. Even if the correct guess is made the second time, the signal strength may have changed during the process of adjustment, so that it becomes necessary to start all over again.

Changing the filament current in a set embodying heater tubes is obviously not a satisfactory method of controlling the volume.

One method of controlling the volume, one which responds instantly, is to use a variable resistor in the antenna circuit, as illustrated in Fig. 1. R is a variable resistor which can be changed over a range of from 0 to 5,000 ohms. If the required sensitivity range of the receiver

is not wide, this method is quite satisfactory, and it is convenient.

Insufficient Range

But if the set is very sensitive when adjusted to maximum sensitivity, this method is not quite sufficient. Even when all the resistance is inserted in the antenna circuit there remains a good deal of pick-up. Some of this is through the antenna and some through the coils in the set. A more effective method of controlling the volume is necessary when the radio frequency amplification is high.

In nearly all AC sets the grid bias is provided by means of a bias resistor, often a separate resistor for each tube. Fig. 1 illustrates this case. In this circuit there are four bias resistors R1, R4, R5 and R6. The first two of these, R1 and R4, are in the grid circuits of the radio frequency amplifiers, and for that reason they can be manipulated without effecting the quality of the output.

As is well known, the amplification of the tube depends on the value of the grid bias. If the bias is increased the amplification decreases, if the bias is decreased, up to a certain point, the amplification increases.

Now the bias given to a tube by a resistor is directly proportional to the value of the resistance, the current through it remaining constant. Hence the sensitivity of the receiver can be changed by merely increasing the value of the bias resistor. The effect is instantaneous no matter what the type of tube that is used.

Decrease in Plate Current

But the current through the resistor does not remain constant. It decreases as the resistance is increased. It would seem that this would defeat the purpose of the increase of the resistor, but that is not so. As the bias is increased by the resistor method, the plate voltage decreases in direct proportion. Hence the need for a high bias is lessened. The two effects work together so that the effect on the sensitivity is about the same as if the grid bias actually increased in proportion to the resistance value.

This method of controlling the sensitivity of a receiver, and hence its amplifi-

cation and output volume, is one of the best yet devised, at least for circuits incorporating heater type tubes.

While it should not be necessary to vary more than one of the grid bias resistors, in some instances it may be desirable to vary both R1 and R4. The sensitivity can be varied over a much wider range if both are made variable.

R1 and R4 may even be combined into one resistor in some instances. This is possible when the plate voltage applied to the two tubes involved is the same, as it is in the circuit represented by Fig. 1. When this is done the maximum value of the resistor should be one half as great as the value of a resistor used for a single tube. A suitable value for the resistor for one tube is 20,000 ohms and for one used for two tubes 10,000 ohms.

By-pass Condensers Indispensable

The by-pass condensers C and C4 across the grid bias resistors are indispensable, for without them the maximum sensitivity of the receiver will be very low and there will be feedback from one tube to the preceding.

By-pass condensers C5 and C6 in the detector and audio stages are even more important, for if they are not used there will not only be a reduction in the amplification but also frequency distortion, the bass notes being suppressed. And these condensers must be as large as practical. Condensers of 1 mfd. will help but condensers of 4 mfd. will be much more effective. The radio frequency condensers C and C4 need not be so large, values larger than .1 mfd. being unnecessary.

Stabilizers Employed

In radio frequency amplifiers designed for high gain, or high sensitivity, there is usually oscillation on the higher frequencies in the broadcast band. To stop this oscillation without at the same time reducing the sensitivity greatly, grid resistors R2 and R3 are inserted in the leads to the grids. The value of these required in any circuit is not a definite quantity, depending on the tubes as well as on the circuit associated with them. About 900 ohms has been found to be a good average value.

New Decibel System

Relative Amplification Values Measured Logarithmically

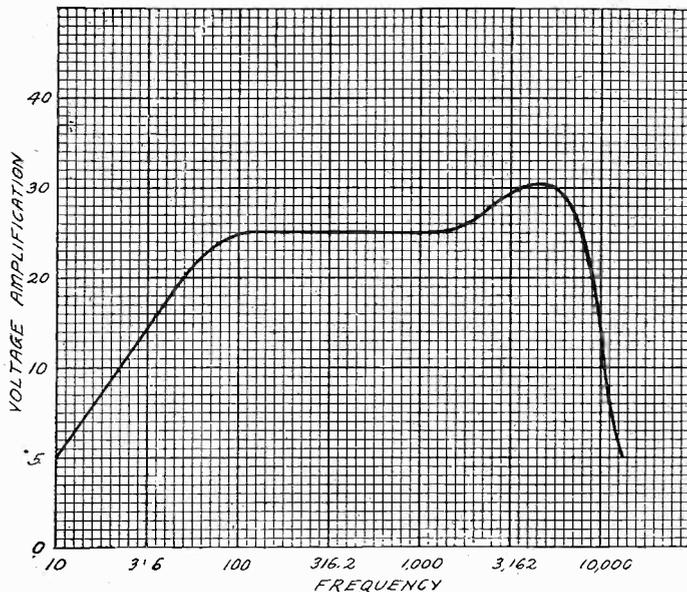


FIG. 1
A TYPICAL RESPONSE CURVE OF A STAGE OF TRANSFORMER-COUPLED AMPLIFICATION WITH FREQUENCIES LAID OUT ON THE FAMILIAR LOGARITHMIC BASIS.

IN plotting response curves it has been customary to plot either frequency or octaves against the amplification. The resulting curves are familiar to most readers of radio articles. Fig. 1 gives the typical form of such a curve. In this the ordinates give the amplification and the abscissas the frequency in octaves, or a scale which amounts to the same thing as octaves. The gradual slope of this curve from 10 cycles to 100, the level region between 100 cycles to a little over 1,000, the rise to a maximum at about 5,000 cycles, and the sudden drop beyond that point are familiar characteristics of amplifiers containing audio transformers.

Plotting in Terms of Decibels

In recent articles the term decibel has appeared as a measure of relative amplification. This is not yet universally known, nor are the amplification curves plotted on this basis familiar.

The term decibel, named in honor of Alexander Graham Bell, the inventor of the telephone, has been adopted the world over in place of the unit formerly known as the transmission unit. It is defined in terms of the common logarithm of the ratio between two similar quantities which it is desired to compare. Stated mathematically, if D expresses the number of decibels by which two quantities V_0 and V_1 differ, then $D = 10 \log (V_0/V_1)$, the common logarithm being understood. If V_0 is larger than V_1 , D is positive, and conversely, if V_1 is larger than V_0 , D is negative.

One of the quantities compared, for example, V_0 , is taken as the zero level, which means simply the basis of comparison. When D is negative, V_1 is "down" with respect to V_0 , or the zero level, and when D is positive, V_1 is "up" with respect to the zero level. The amount by which V_1 is "up" or "down" is expressed by the number of decibels.

Decibel Measures Ratio Only

The decibel tells nothing about the absolute values of the quantities involved, only relative values. Two quantities

measured in kilovolts may differ by the same number of decibels as two quantities measured in microvolts. Often the absolute values of the quantities are of little importance, but the relative values may be of prime importance. This is true especially when response curves are involved. For example, it is of practically no importance what the absolute values of the signal voltages on a certain grid of an amplifier are at two different frequencies, but it is of first importance what the relative values are. The quality of the output with respect to frequency distortion depends on the relative values.

In order to familiarize the reader with the two types of curves plotted in the old way and according to the decibel system, the characteristic shown in Fig. 1 has been replotted in Fig. 2. The same data were used in plotting both. But the curves differ widely in shape.

The zero level was selected as the amplification at 400 cycles, because this is usually employed as a standard for comparison. The amplification at any other frequency is then given as the number of decibels above or below the amplification at 400 cycles.

Level Region

Naturally, the level region in the two curves coincides as far as the frequency axis is concerned, but on the decibel curve this level region coincides with the zero level line, while in the other it falls on the 25 amplification line.

Below 100 cycles the curve in Fig. 2 falls below the zero level line, meaning that the amplification for frequencies below 100 cycles is less than that at 400 cycles. At about 5,000 cycles there is a peak in the curve, and for a certain region about this peak the curve lies above the zero level. That means that the amplification in this region is higher than that at 400 cycles. Above 5,000 cycles the curve drops rapidly so that at 10,000 cycles it is again far below the zero level. The amplification is therefore much less in this region than it is in the middle register.

It has been found experimentally that

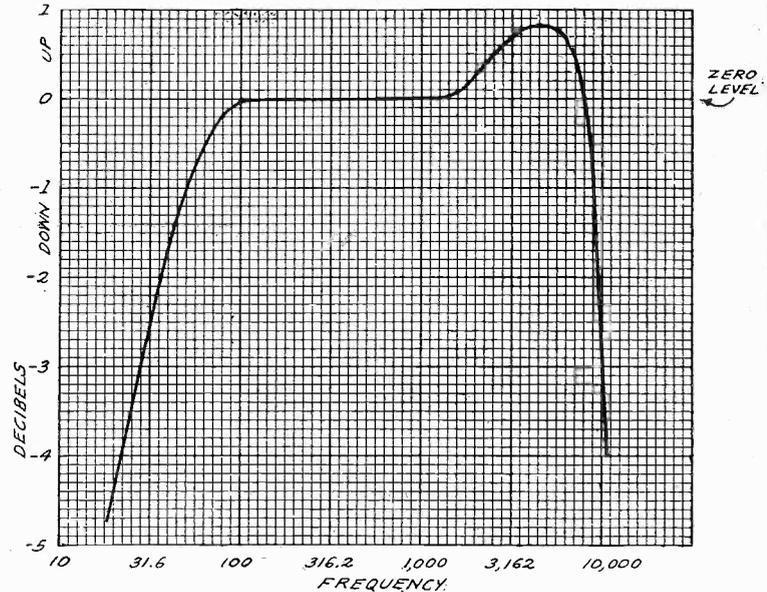


FIG. 2
THE SAME RESPONSE CURVE ON THE SAME FREQUENCY BASIS BUT PLOTTED VERTICALLY ACCORDING TO THE NUMBER OF DECIBELS UP OR DOWN, COMPARED WITH THE AMPLIFICATION AT 400 CYCLES.

the ear appreciates intensity differences according to a logarithmic law, that is, according to the number of decibels by which they differ. It makes little difference what the absolute intensities may be, as long as they lie within the limits of hearing. It is for this reason, mainly, that the logarithmic definition has been laid down.

One Decibel Is Minimum

It so happens that the smallest intensity difference that the human ear can detect is that represented by one decibel, and this is one reason why the decibel, rather than the bel, was chosen as the practical unit. The bel is ten times as large as the decibel, and is too large to be made a practical unit.

A response curve plotted in terms of the decibel is very convenient in judging the quality of an amplifier or of a complete receiver. If the curve does not deviate more than one decibel from the zero level in the essential frequency band, it is known that the quality will be practically perfect, for at no frequency will it rise or fall by an appreciable amount. Wherever it rises more than one decibel the response will be noticeably high, and the result may be an unpleasant blast. If it falls more than one decibel in any region, the response will be appreciably weak in that region. If it is in the bass, the quality will be thin and lacking in body. If it is in the treble and above, the quality will be boomy and lacking in crispness and articulation.

Comparing Powers

When comparing power outputs and different frequencies, or at the same frequency on different sets, the definition of the decibel is changed slightly, but only when using voltages and currents for the comparison of the powers. Suppose two voltage intensities V_0 and V_1 are compared and it is desired to compare also the corresponding power outputs. The definition then becomes $20 \log (V_0/V_1)$. The reason for doubling the constant associated with the logarithm is that the

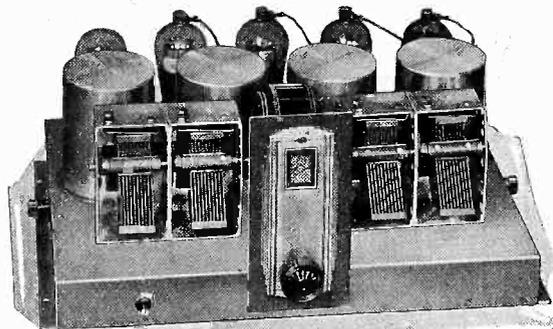
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Why the MB-29 is OK

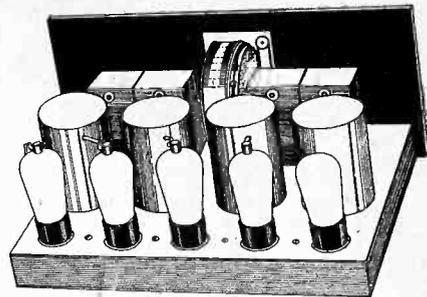
Each Stage Carefully Shielded and Filtered, Preventing Feedback

By J. E. Anderson

Technical Editor



EACH STAGE OF THE MB-29 IS SEPARATELY SHIELDED. THE CYLINDRICAL CONTAINERS ARE THE SHIELDS. THEY ARE ALUMINUM. INSIDE EACH SHIELD ARE THE FILTERS FOR EACH STAGE. NOTE THE NEW NATIONAL MODERNISTIC DIAL IN THE ILLUSTRATION AT LEFT. THE REAR VIEW OF THE MB-29 IS SHOWN AT RIGHT.



[In the issue of May 18th were published the first details of the MB-29 to be presented anywhere. This is a 5-tube tuner, using four of the new AC tubes (224) and a power detector (227). In the May 25th and June 1st issues the intimate technical discussion was carried forward with extreme thoroughness. This week the reasons for the soundness of the circuit—despite some contrary dictum—are set forth. Next week other features of this circuit will be published in the June 15th issue. This is the circuit that brought in nine coast-to-coast stations in one evening on the speaker. —EDITOR.]

* * *

SOME manufacturers of AC screen grid tubes have issued specific recommendations that no more than three of these tubes be used in the same receiver in the same frequency level. On the strength of these recommendations, we have been requested to correct the design of the MB-29, or to explain the discrepancy.

What's To Be Corrected

It would indeed be presumptuous on our part to correct the design of two such eminent engineers as James Millen and Glenn H. Browning, especially a design which these two engineers have worked out with utmost care and precision, a design based not only on theoretic soundness, but on unchallengeable laboratory results. It is not the circuit that needs to be corrected, rather the idea entertained by some that all knowledge of the application of tubes is concentrated in

the minds of engineers of some tube manufacturers.

But the discrepancy can be explained. When two or more screen grid tubes are used in hastily assembled receivers there is likely to be a great deal of feedback that will produce oscillation. This is difficult to control. To stop the feedback, and thus to stop the oscillation, it is necessary to shield the separate stages individually and to insert filters in the voltage supply leads. If this is not done, and done well, even a circuit embodying two of the tubes is almost uncontrollable. One having three of the tubes is then quite unmanageable, and one having more than three of the tubes is then hopeless.

MB-29 Is Stable

But this applies only to circuits in which utmost precautions against feedback have not been taken. It does not apply to the MB-29, for shielding and filtering in that circuit have been done with thoroughness. Each tube in that circuit feeds only the one following, and not any one preceding it.

Naturally, the amplification in a circuit containing four AC screen grid tubes is exceedingly high. There will be some stations which will overload even the second tube, provided the circuit is adjusted to maximum sensitivity. If it did not, it would not bring in distant stations with such good volume. Every circuit designed for very high sensitivity and distance getting ability must necessarily be such as to overload on local stations, when it is adjusted to maximum sensitivity. That is axiomatic.

(Continued from preceding page)
power is proportional to the square of the voltage. Hence the ratio of voltages must be squared before the logarithm is taken. The logarithm of the square of the quantity is equal to twice the logarithm of the quantity itself. Hence the factor 20 is used instead of the factor 10.

If two powers are compared directly, of course, the factor 10 is used, for the factor 2 was used in deriving the powers.

We might illustrate the comparison of powers. Suppose one amplifier will develop a voltage of 100 across a loudspeaker and another amplifier only a voltage of 50 across the same speaker and at the same frequency. The ratio of these voltages is 2, the common logarithm of which is .301. The voltage output of one speaker is then up 3.01 decibels as compared with the other. But the

louder speaker is up 6.02 decibels when the powers are compared.

That this gives the same results as if the powers directly were compared, using the factor 10, can be demonstrated with actual values. Suppose that the resistance of the speaker is 4,000 ohms. Since the voltage in one instance is 100 volts, the current is .025 milliamperes. And the power, being the product of the voltage and the current, is 2.5 watts. Using the 50 volts, the current is .0125 milliamperes and the power is .625 watt. The ratio of 2.5 and .625 is 4, the common logarithm of which is .602. Hence the power in one instance is up 6.02 decibels as compared with the power in the other instance.

Obtaining the Common Logarithm

The common logarithm can be obtained in almost any book on trigonometry or

Hence every super-sensitive receiver must have an effective sensitivity control by means of which it can be adjusted to suit the requirements. When the receiver is tuned to a local, high powered station, the sensitivity control is set so that each tube contributes only a small fraction of its possible amplification power. When the circuit is tuned to a weak distant station the control is set so that each tube contributes maximum amplification. Without such a sensitivity control the best receiver is useless.

The Sensitivity Control

As is well known, the amplification given by a screen grid tube depends largely on the voltage applied to the screen grids. By varying this voltage over a certain range, the amplification of the tube can be changed from zero to maximum possible with the particular circuit arrangement used. This fact has been made use of in the MB-29 circuit. The voltage applied to the screen grids has been made variable from zero to 67 volts, or to a higher value if desired.

The variation of the screen grid voltage is effected by means of the potentiometer R5, which has a value of 50,000 ohms. The range of voltage variation depends only on the voltage that is impressed across this potentiometer.

Since all the screen grids are connected to the same sliding point on the potentiometer, the screen grid voltage on all the tubes is the same at all times, and the amplification by each tube is varied in the same manner. It is thus apparent that the control is exceptionally effective.

in books especially devoted to mathematical tables. It can also be obtained from any slide rule. Most computations for plotting curves of the character in Fig. 2 are made on slide rules, and the logarithm of the number desired is obtained as quickly as the number itself. Indeed, it is not necessary to read off the number at all, only the logarithm of it. Tables or the rule give only the fractional part of the logarithm. The whole number, if any, is obtained by counting the number of figures on the left of the decimal point of the number of which the logarithm is desired, and by subtracting one from the number of figures. For example, for the number 25.6 the table or the rule gives .408. Since there are two figures to the left of the decimal point, the logarithm is 1.408.

—J. E. ANDERSON.

Impedance Audio

Gives Good Reproduction with Little Tendency Toward Trouble

By Malcolm Maxim

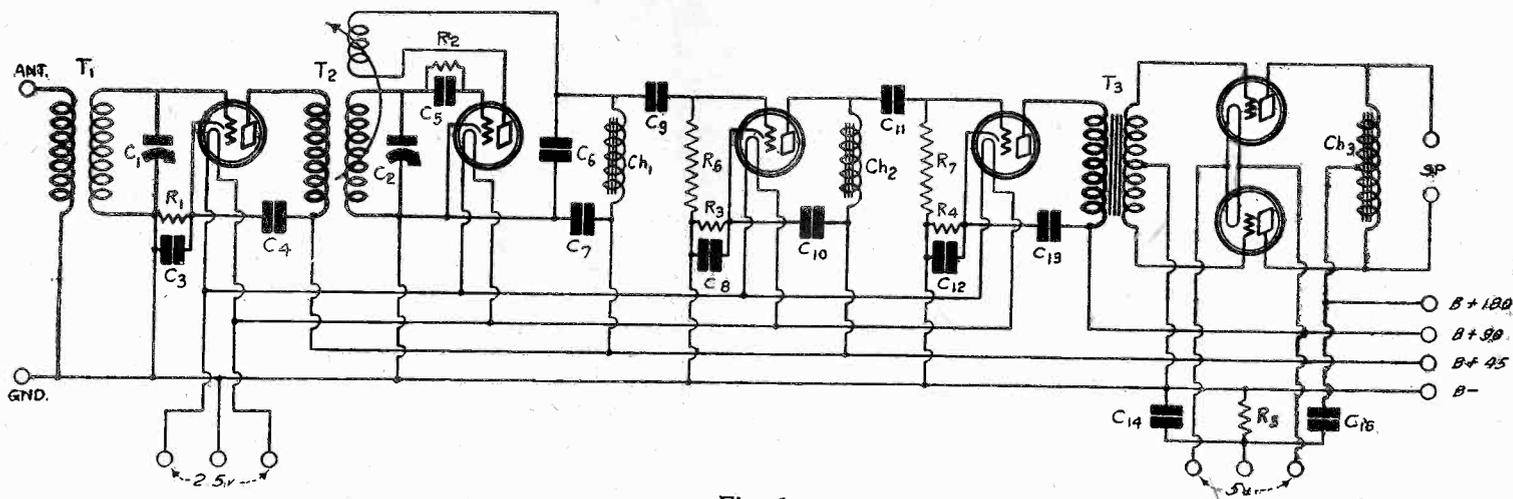


Fig. 1

A QUALITY RECEIVER IN WHICH IMPEDANCE-RESISTANCE COUPLING IS USED IN TWO STAGES AND TRANSFORMER FOR THE PUSH-PULL STAGE

THE impedance coupled amplifier has not been received by the radio fans as well as its virtues merit. It has many distinct advantages over other forms of coupling which are not appreciated. First it has an amplification characteristic practically as flat as that of resistance coupling, and it has a per-stage amplification, with the same tubes, which is slightly greater. It does not motorboat as readily as the resistance coupled circuit, and if it does, it oscillates at a frequency at which the oscillation may be stopped easily with condensers across the plate voltage supply. It does not amplify effectively the sub-audible frequencies and therefore motorboating at these frequencies will not occur in it. It is these frequencies which give the greatest trouble in direct coupled amplifiers.

One reason why the impedance coupled circuit is relatively stable is that the coupling impedances introduce a lag in each of the plate signal currents so that it cannot produce oscillation-inducing feedback as readily as when there is no phase displacement of the currents.

Point in Its Favor

One point in favor of the impedance coupled circuit is that fine quality can be obtained with relatively inexpensive parts. For example, the secondary of a transformer which is unsuited for use in a transformer coupled circuit can be used with the assurance that the quality will be much better than the quality of the average set at the present time, and it must be admitted that the quality of the average set of today is pretty good.

One of the objections against the impedance coupled amplifier is that it cannot be used in push-pull, and since transformers have to be used for push-pull anyway, they might as well be used throughout the set. That is not a good reason. If first rate quality can be obtained with impedance coupling inexpensively, it seems good practice, not only from an engineering point of view but also from an economical point of view, to use impedance coupling in the first stages and push-pull transformer coupling for the final stage, if push-pull output is desired.

A combination impedance and push-pull transformer coupled receiver is shown in

Fig. 1. This circuit contains heater type tubes throughout with the exception of the last stage, which contains a couple of filament type power tubes. They may be of any type provided that the various voltages applied are suited to the tubes used. It will be assumed here that the tubes are of the 171A type, because these tubes are admirable output tubes in a push-pull amplifier for home use. More volume can be obtained from them than will ever be tolerable in a home, even if a rather insensitive loudspeaker is used.

Adequate By-Passing Necessary

In order to get the best quality out of any receiver, regardless of coupling, it is necessary to by-pass adequately all the plate and grid leads. The circuit in Fig. 1 exemplifies a case of thorough by-passing.

Let us analyze the by-passing both in the radio and audio frequency portions of the circuit. First there is C3 which serves the grid bias resistor of the first tube. It aids greatly both to increase the sensitivity and the stability of the receiver. The value of this condenser might be .1 mfd., although a somewhat lower value can be used with satisfactory results. Then there is C4, which by-passes the plate circuit of the first tube. The value of this condenser may be the same as that of C3 because it operates at radio frequency. Note that both these condensers are connected to the cathode of the tube. It is always a good policy to by-pass direct to the cathode rather than to lead the signal currents to that point by a round-about route. This method of by-passing is used in all the succeeding tubes, changed only as local conditions may demand.

Plate Circuit By-Pass

No condenser is necessary in the grid circuit of the detector because the grid return is connected to the cathode. There can be no more direct route than that. But we do find a couple of by-pass condensers in the plate circuit of the detector. C6 is connected across the line so that its value must not be larger than is absolutely necessary. Neither is it necessary to use a large condenser here for it works at radio frequency, and also there is some distributed capacity in the coupling choke Ch1, which adds to the value of C6.

The circuit will not oscillate well if the total by-pass capacity is small, neither will it detect well. It has been found by experiment that .0005 mfd. is a very satisfactory compromise. Actually it may not be necessary to use more than .00025 mfd. because of the effect of the distributed capacity in the choke coil.

C7 works mainly at audio frequency and therefore this condenser should be large. This condenser is also in a critical position because if it permits any feedback, this is amplified by all the succeeding tubes and the effect of this feedback may be very great, causing distortion or actual oscillation at some frequency. A value of 2 mfd. has a very beneficial influence on the stability, but one of 4 mfd. is proportionately more effective. Since this condenser need not be rated at a high voltage there is no good reason why a large value condenser should not be used. One microfarad of prevention here is as good as 10 microfarads placed later in the circuit.

In the next two stages there are two grid by-pass condensers C8 and C9. They serve the same purpose as C3, except that they are functioning at audio frequency, and for that reason must have large values. The lowest value for each of these that will help much is 1 mfd., while 2 mfd. should be the lowest that should be used. These operate at very low voltage so that the cheapest condensers made may be used.

Plate By-passing

In the plate circuits of these two tubes are two condensers C10 and C13. These work at a slightly higher voltage than the grid condensers, but even so, the lowest voltage-rated condensers regularly manufactured can be used with safety. The size of either of these condensers is the same as the corresponding grid condenser.

There is no imperative reason why by-pass condensers should be used in the grid and plate circuits of a push-pull amplifier stage, provided the circuit is balanced. But perfect balance is very rare. Hence it is advisable to use the condensers. The connections are exactly the same as those for the single tube stages, except that allowance has been made for the fact that now the mid-tap of the filament transformer takes the place of the cathodes in the heated tubes.

Better Reception Looms

Synchronization Successful in Isolated Cases and Promises Much

By Dr. John H. Dellinger

CHIEF OF RADIO LABORATORY, BUREAU OF STANDARDS

The following is an excerpt from an address delivered by Dr. Dellinger before the recent convention of the Institute of Radio Engineers, at Washington, D. C.

THE broadcasting band, which has hitherto occupied the chief attention of the Federal Radio Commission, exhibits a number of very special engineering problems. Here the width of each channel in 10 kilocycles, which is necessary for satisfactory musical reception. Even this is not sufficient for musical reproduction of the highest quality. With only 90 such channels, and more than 600 stations on them, there was naturally, very great interference prior to the November, 1928, allocation. To remedy this the Commission had to choose among various alternatives.

It decided as a matter of policy not to reduce radically the total number of stations. It was then necessary to (a) limit the simultaneous operation of an excessive number of stations by making many of them divide time; (b) assign frequencies carefully selected with regard to geographical separation of stations, to reduce inter-channel interferences (i. e., disturbance of reception of a station on one frequency by other stations on adjacent frequencies); and (c) limit the power of stations so they would not cause interferences to other stations on the same frequency.

Calls Results Accomplished

The accomplishing of this constituted the allocation of broadcasting stations which the Commission put into effect on November 11th, 1928.

The most striking of the problems involved in the allocation was the carrying out of requirement (c), which determines power limitations. Stations assigned to the same frequency have not, up to the present, been able to maintain their frequencies with sufficient accuracy to prevent the existence of a slight difference (or beat) frequency, producing what is commonly known as heterodyne interference, or whistles.

Unfortunately the heterodyne interference reaches out to enormously greater distances from a station than the program. Consequently the operation of two or more stations on a channel results in an area of destructive interference far in excess of the area, in which program service is provided.

For instance, the program of a 5-kw station can be heard with fair intensity under good conditions at 100 miles, while the heterodyne interference from two such stations is heard at 3,000 miles. Two stations of 5 kw or more therefore cannot be assigned the same frequency in the United States.

It is possible, on the average, to put two or more 1-kw stations on the same frequency if they are at least 1,800 miles apart, and two or more ½-kw stations if they are at least 1,200 miles apart.

All stations subject to these restrictions have only a small service area, and give no service to remote rural areas. Such distant service is given only by stations having exclusive use of the channels to which they are assigned.

In order to provide rural service, 40 channels are each used by one station exclusively. The stations on the exclu-

sive channels not only serve very great areas, but deliver a more satisfactory intensity at every point within those areas.

Their service is better for all concerned, the greater the power they use. This fact is not commonly understood by others than radio engineers. It is clear when the distinction between the exclusive and the other channels is comprehended. Service on the non-exclusive channels would be utterly ruined if the power limits fixed by the facts of heterodyne interference should be exceeded, and in consequence such stations cannot in general use more than 1 kw. But on the exclusive channels the service is better the higher the power level, and indeed such stations will not be serving the public most effectively until the level reaches hundreds of kilowatts.

Increased Service Area

There is some hope that the limitation of power and service on the non-exclusive channels may be overcome. If the frequencies of stations on the same channel are maintained to a certain high accuracy, the heterodyne or whistle becomes inaudible. The technique of frequency control is fast approaching this goal, and success has been attained in isolated instances.

The satisfactory service area of such "synchronized" stations is not yet known. It is likely that some additional fading and interference effects will be introduced but it is believed that the net service area will be substantially greater than when heterodyne interference exists.

The significance of this is that the present power limits for stations, on shared channels can be raised, better service given, and wider areas served. Synchronization is therefore looked for as the next great advance in broadcasting.

Different in Day Time

This discussion of broadcasting has been largely with reference to night conditions. Broadcast transmission is entirely different in the daytime. Transmission distances are much less, and somewhat greater power can be allowed the stations. Furthermore, additional problems are not as acute as are those of night-time transmission, they are being handled with care so that day-time broadcasting may be developed as a valuable service.

The difference between day and night transmission conditions raised one technical problem of considerable moment, viz., determination of the time when day ends and night begins, and thus at what hour daytime stations should close.

Investigation had revealed that the change from day to night radio conditions extends over a period of something more than an hour and a half, beginning about a half hour before sunset and closing a half hour after sunset.

The most reasonable time to choose as the transition point is the moment of sunset, and this was done in the Commission's General Order No. 41.

Most of the regulations of the Com-

mission involve engineering problems. Examples include the General Orders on chain program limitation, visual broadcasting (television), increased power in daytime, and prohibition of damped waves.

The Broadcasting Committee of the Institute of Radio Engineers has been of great assistance to the Commission by the studies it has made of certain subjects leading to new regulations.

These are on such subjects as: The requirement of an artificial antenna in broadcasting stations for use during warming-up periods, etc.; the location of high-power stations with respect to populous areas; requirement of highly accurate frequency control; allowable ratio of day to night power; permissible intensity of harmonics; per centage modulation; and fidelity of transmission.

Complex Situation

All of the engineering work involved in Federal radio regulation has the peculiar difficulty that the facts dealt with are extremely complex. They are indeed rapidly shifting. Not only must allowance be constantly made for the flux of changes inherent in a rapidly developing art, but radio waves themselves exhibit extraordinary vagaries.

Orderly radio regulation must proceed on a consideration of the distances at which the waves are received. But distances vary enormously between day and night, from season to season, even from night to night, and are different over different kinds of terrain.

Knowing this is not counsel despair. These vagaries have, after all, certain discernible laws, becoming more and more calculable as the results of scientific investigations accumulate.

It is not necessary to throw up our hands and say that the whole situation is chaotic. In spite of their vagaries, radio phenomena are subject to known engineering principles. An engineering principle is nothing but an organized body of facts affecting a practical situation. Violation of such engineering principles in radio regulation would sooner or later reduce the service of radio to the public.

Summarizing, the Federal regulation of radio involves extensive and difficult engineering problems. These are characterized by certain outstanding facts or principles.

First, radio waves spread out everywhere and potentially interfere with one another.

Secondly, at any given stage of radio technique, the available number of communication channels is definitely limited.

Another controlling principle, as the art stands today, is that heterodyne interference sharply limits the power that may be permitted any two or more broadcasting stations on the same channel.

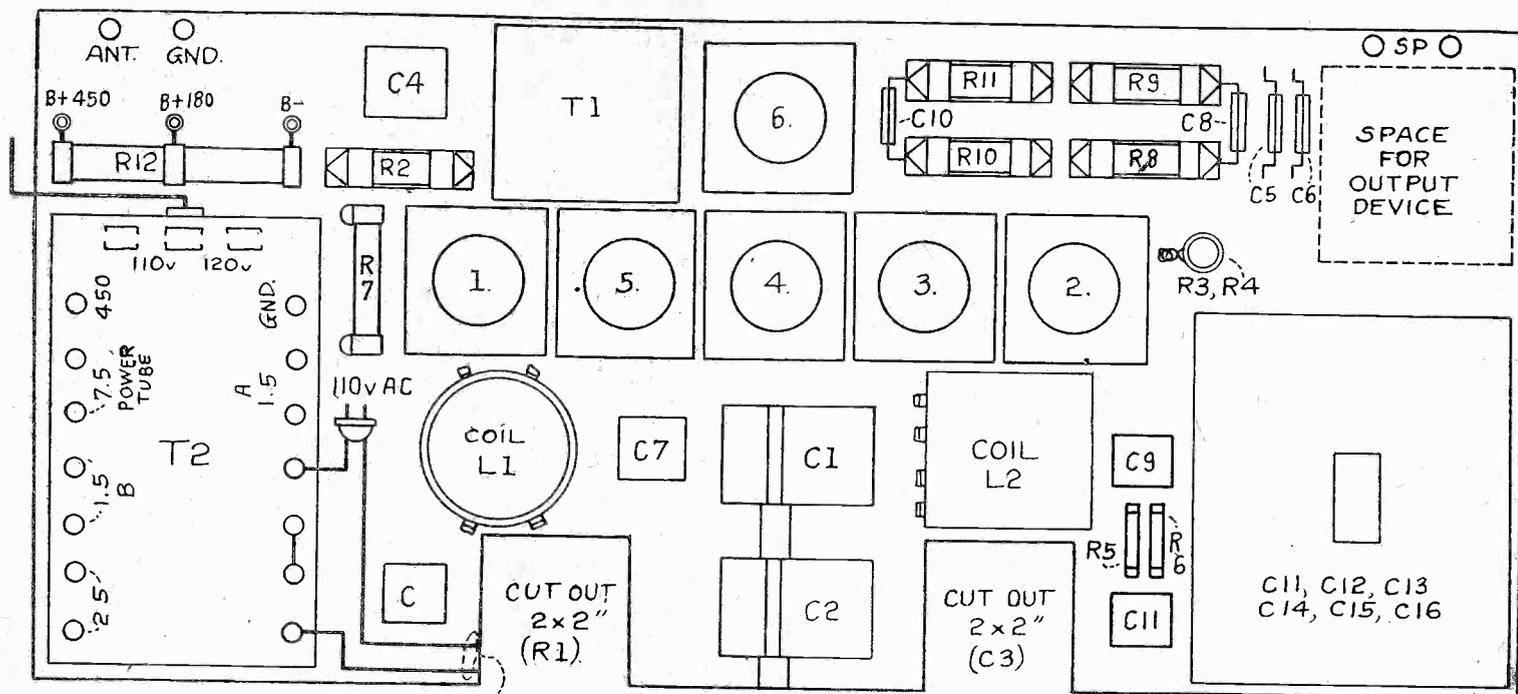
Finally, radio wave transmission is characterized by extreme vagaries. The facts and implications of each of these principles are subject to constant revision as radio progresses. Such facts constitute the natural limitations of radio. They are inescapable conditions of its Government regulation.

The 250 Universal

Complete AC Receiver on 8 1-2x20 Baseboard

By Capt. Peter V. O'Rourke

Contributing Editor.



LAYOUT OF PARTS OF THE RECEIVER. NOTE THE TWO CUTOUTS

[The second and final instalment of this article on a compact receiver using six tubes, including rectifier, is published herewith. Part I was published last week, June 1.]

S grid voltage changes produce corresponding plate current changes, the plate current must accommodate this pulsation, otherwise tube distortion arises, as evidenced by the wobbling needle of a microammeter placed in series with the plate circuit. Without such precaution power detection would be more of a name than an achievement.

RF Channel

The radio part of the installation is sufficiently sensitive for most modern needs, with the aid of regeneration provided by the small condenser C3 connected from plate of the detector tube to midtap of the interstage coupler's primary. The regeneration will fail unless the connections of respective terminals of the primary of L2 are reversed in comparison with the connections of the terminals of the secondary of L2. But if regeneration fails, remove the plate and B plus connections of the primary of L2 and replace them to the opposite points to the one to which they formerly were connected.

Volume control is achieved by one of the best methods for AC receivers, by varying the bias on the radio frequency amplifier. This tube is the new 224 screen grid AC tube. The biasing resistor R1 is 0-5,000 ohms, but the zero position is never used, as that would stop the plate current in the tube. The bias may be varied from a volt or so negative to 25 volts negative, which brings the RF tube even beyond the best detecting point. Any preliminary detection is lost, as the succeeding coil will pass only radio frequencies, and these would be merely stray components when detection takes

place, and would constitute a feeble input to the detector. Hence the volume is variable from the conventional shout to the still more conventional whisper. The resistor is bypassed to sustain the high amplification of which the tube is capable.

Source of Voltages

The voltages at the radio frequency side are derived as follows:

The RF tube and the detector both get their heater voltage from the same winding of the transformer, 2.5 volts (actually 2.15 volts). This is not the filament, because both the 224 screen grid AC tube and the 227 AC tube are of the heater type. The heater is connected to the power transformer and gets hot. This heat is communicated by the heater to the cathode, or functioning filament, by thermal radiation. The cathode emits the electrons.

The sum of the plate and grid voltages is 180 to 190, obtained by connecting to the midtap of a 9,000 ohm resistor that also bleeds current from the rectifier output. Besides this stabilizing bleeder current through all of R12 there is the plate current of all tubes save the last, through the midtap-to-B-minus half of R12.

The 224 tube takes the voltage from this midtap for its plate. Then another resistor, R2, which is .1 meg., drops the 180 to 75 to 80 volts, which voltage is applied to the screen grid or G post of the 224 tube. C4 bypasses RF currents to B minus, while C13 bypasses plate the lower half of the midtapped resistor R12 to the same point. C13 also serves to complete the circuit for C2 to the end of the plate winding or primary of L2.

The same plate voltage is applied to the detector as to the RF tube, but of course the current flow is much less, due to the drop in the load resistor R8 in the plate circuit.

As for the audio plate voltages, the first

stage (tube 3) gets 180 ahead of the load resistor, but actually this is dropped to around 60 volts at the plate, whereas the same 180 volts applied through the primary of the transformer T1 affords about 120 volts on the plate.

Facts on Power Transformer

The power transformer has provisions for grounding the center taps of all the filament windings, except the power rectifier tubes. Thus the tuned circuits are grounded through the biasing resistors, the bypass condensers across these resistors, and the extra bypass condenser C13. The two fuse clips are marked 110 volts and 120 volts, and the fuse is to be put in the correct clip, as determined by your AC line voltage, or, you have no meter that reads that high, you may put the clip in the fuse in the clip that affords 1.4 volts for the 226 tubes or 2.15 volts or less for the heater tubes. The transformer is purposely designed to keep the heater voltages well under 2.25 volts even if the line is 120 volts and no means of reading any voltages is at hand.

In this transformer is the choke Ch, a husky one, so that only one choke is necessary for the filter circuit. The filter condensers are C14, C15 and C16, which are 2 mfd. each, the first being of 1,000 volts AC test, the two others of 800 volts AC test. These three, and C11, C12 and C13 are in the capacity bank. All other condensers are external.

The circuit diagram shows no output transformer, as it is assumed a dynamic speaker will be used. However, room is left on the baseboard for the inclusion of such a transformer, or of a condenser-choke output filter, should one use a magnetic speaker. Be sure that the output transformer or filter is of the 250 type. An ordinary one will not stand the 50 milliamperes.

Chain Restriction Due to Be Lifted

Washington. The order of the Federal Radio Commission, prohibiting stations, 300 miles or less apart, from broadcasting the same chain program, which rule was never put into effect, is due to be abrogated entirely. It was meant to apply only to stations in cleared channels, but the effective date was advanced three times, due to serious objections from stations and some listeners.

It would finally become effective June 1st, but a prior abrogation order is expected. At least another postponement is promised.

Users Find Ce-Lec-Tor, Excellent Filter Trap

Commendations upon the working efficiency of the Ce-Lec-Tor, the new band-pass filter trap, are being received from all over the United States, according to Julien J. Proskauer, head of the Trutone Radio Sales Co., 114 Worth Street, New York City.

A fan from San Francisco wrote that he had no trouble cutting out such powerful stations as KGO and receiving distant stations from the middle Western and the Eastern states.

The Ce-Lec-Tor is sturdily built and compact. It is cased in black bakelite having the interior condensers and coils immersed in a special insulating wax which renders them impervious to moisture. Full information and details may be had by addressing this concern. Mention RADIO WORLD.—J. H. C.

World Wave Parley Needed, Says Craven

Washington.

Lieutenant Commander T. A. M. Craven, naval radio specialist, before the Senate Committee on Interstate Commerce recently predicted that it would be necessary soon to call an international conference to adjust the world radio situation, because the demands of nations for radio channels are conflicting. The naval officer testified in connection with the Couzens Bill for creating a Federal communications commission with full jurisdiction wire and radio services.

Emphasizing that the number of channels in the spectrum is definitely limited, he said that "engineers are having some difficulty in finding out how to use the channels in the most effective manner." He recommended that the Federal licensing authority should safeguard radio communication by assigning particular blocks of channels to particular services. Stations of different character and function, he said, cannot operate effectively on the same or adjacent channels if the channels are to be employed to their fullest extent.

A THOUGHT FOR THE WEEK

CENSORSHIP is un-American, undemocratic, non-republican, anti-a-lot-of-things, and a confounded nuisance generally. Careless broadcasters should remember that and forego sending over the air those things which will give the enemy an excuse for insisting on the opportunity of a radio censor. An ounce of commonsense is worth a ton of protest that's too late.

Right or Wrong?

(Answers below)

1. The shielding effect of a sheet of metal is greater the lower the specific resistance of that metal. It is also greater the higher the permeability of the metal.
2. Shielding is due to a reflection of the radio wave and the action of the shield is similar to the action of a mirror to light.
3. The value of a dry cell used as a grid battery is truly indicated by a voltmeter, even if the dry cell is practically exhausted.
4. There are only two ways of producing an electrical difference of potential, or an emf, namely, by a generator and by chemical means.
5. It is not possible to get a higher amplification per stage than the amplification constant of the tube.
6. A wire-wound resistor cannot be used in resistance coupled amplifiers because it has a high inductance which would introduce frequency distortion.
7. The only possible arrangement of two inductance coils which will result in no mutual inductance, or coupling, is that which makes the axes of the two coils at right angles.
8. More than three inductance coils can be placed so that there is no mutual inductance between any two.
9. Mutual inductance is a purely fictitious quantity. Two inductance coils placed close together have no inductance in common but they merely act as if they had.
10. Inductance of an air-core coil depends only on the geometry of it.

Answers

1. Right. Silver is the best shield material of the non-magnetic metals because it has a lower specific resistance than any other. Copper comes next, being only slightly inferior. Iron is the best of all due to its very high permeability.
2. Right. The waves cannot penetrate because eddy currents are induced in the metal, and these send out waves in the opposite direction to the incident waves. A light wave is of the same nature as a radio wave and responds essentially the same way.
3. Wrong. The value of the cell for bias is determined by the emf of the cell. A voltmeter of the ordinary type always reads less than the true emf.
4. Wrong. An emf can also be produced by heat, as in the thermo-couple.
5. Wrong. The statement holds true for resistance and impedance coupled circuits, but not for transformer coupled amplifiers. The limiting amplification with transformer coupling is the amplification factor of the tube times the step-up ratio of the transformer.
6. Wrong. The inductance of the coil is very small and does not affect the amplification adversely. A resistor may be wound so that its inductance is negligible.
7. Wrong. It is possible to place them in any angular relationship and still have the coupling zero provided that one coil is not inside the other.
8. Wrong. There are only three dimensions and therefore only three coils can be placed mutually at right angles.
9. Wrong. Mutual inductance is as real as self-inductance. When two coils are placed together, either lies in the induction field of the other, and the magnetic flux from one threads the other coil.
10. Right. In the electromagnetic system of measuring inductance is a length and is measured in linear units. The unit is the centimeter.

Hotel Wins Suit Over Copyright

Kansas City, Mo.

In a suit filed here by the American Society of Composers, Authors and Publishers against the La Salle Hotel of this city, Federal Judge Merrill E. Otis upheld the right of hotels to receive and transmit to guest rooms copyright music broadcast by radio stations. The decision is said to be the first of its kind.

Judge Otis held that in receiving the radio programs and transmitting them to the rooms the hotel owners did not "perform" the compositions, but merely "provided the means" for hearing the music.

The decision differentiated between radio programs and the playing of phonograph records and awarded the plaintiffs \$250 damages for alleged playing of a record by the hotel. The playing of a record was held to be a "performance," as the music was permanently recorded and the rendition was made at a time that suited the hotel's purpose and convenience.

Chain of Radio Schools Planned

Plans for a nationwide radio chain to broadcast educational programs to all sections of the country were announced by Headmaster Ellsworth Tompkins of the WODA Free Grammar and High School of the Air, Paterson, N. J.

A high frequency relay system, transmitting complete courses in six academic subjects, will be used by the station to link together the stations of the chain. Application for the short wave length is now before the Federal Radio Commission. Tentative plans call for two stations in the South and several stations throughout the Middle West.

A complete grammar and high school course will be available to every radio listener when the chain begins operation. According to Commissioner of Education Richard E. O'Dea of Paterson, owner of WODA, the courses will be broadcast as good will programs, free from commercialism. None of the chain stations will be allowed to commercialize the school.

"The success of the WODA Free Grammar and High School, now in its second year with an enrollment of 1,000 pupils," said Commissioner O'Dea, "proves conclusively that there is a need for complete academic courses on the radio. It was with this thought in mind that the school alumni, composed of prominent educators in New Jersey, arranged courses so they are equally valuable to listener-students all over the country."

Broadcasting of the school will begin on an experimental basis during the summer. Several weeks will be given to preliminary tests to enable engineers of the station to work out signal strength charts and gather other important technical data. A 250-watt transmitter, similar to transmitters used for transcontinental and transoceanic work, has already been completed by engineers of WODA.

The school's curriculum would include English, grammar, arithmetic, algebra, history, Americanization, general science, social science, nature study and home economics.

POWER AMPLIFIERS

By J. E. Anderson and Herman Bernard

[Here is the second instalment of the noteworthy contribution of Anderson and Bernard to the all-too-scarce literature of power amplifiers. The series began last week, issue of June 1st. There will be an instalment each week for several months. Don't miss a single one.—Editor.]

Loudspeaker Coupling Devices

There are three principal methods of coupling the loudspeaker to the power amplifier: direct, output filter, and transformer.

The direct method of coupling, shown in Fig. 9A, is used when the power tube delivers only a small plate current, or when the loudspeaker winding is such as to withstand the heating caused by considerable current. Ordinarily, this method of coupling is not used with tubes larger than the 112A. It has one advantage not possessed by any other, and that is it avoids the addition of all reactive devices which always introduce some frequency distortion. If the loudspeaker is wound to match the tube used, and if the plate current is small so that the current will not damage the winding, the connection gives the best quality.

Unfortunately, it is not practical to make speakers so as to carry heavy direct current, so that when larger tubes are used some output device is necessary. One of the most popular is the output filter, shown in Fig. 9B. The plate current flows through a high inductance choke Ch and the signal current through a stopping condenser C2 and the speaker, SP. This method of coupling requires that the speaker impedance be matched to that of the tube, because for AC this connection is essentially the same as that in Fig. 9A. The matching of impedances does not mean they should be made equal.

The choke coil Ch must have a high inductance and a low direct current resistance, and condenser C2 must have a large capacity. If the inductance is not high, part of the signal current will flow through the coil, especially the low frequencies, and if the condenser C2 is not large the low frequencies cannot flow through the speaker circuit. Hence the conditions imposed on the choke coil and the condenser are mainly for the purpose of allowing the low notes to reach the speaker. The DC resistance of the choke should be low so that the voltage drop in the coil be as low as possible.

Another popular output device, the transformer, is shown in Fig. 9C. This has the advantage, in common with the output filter, of preventing the direct current from reaching the speaker. Another point in favor of the transformer is that it can be used to match the tube and speaker impedances so as to get the maximum undistorted output. An output transformer is built into all dynamic speakers.

The output transformer used with push-pull amplifiers, shown in Fig. 9 D, shows an alternative connection. The loudspeaker

may be connected directly between the two plates of the amplifier, as indicated by the dotted lines. This connection may be used when the sum of the plate impedances of the two tubes is comparable with the impedance of the speaker. For example, a high impedance magnetic speaker might be connected to the output of a power stage using two 245s. The direct plate current will not flow through the speaker.

E, F, and G in Fig 9 show the connections when AC tubes are used. In E and F the output filter has been used for illustration. If the connection A is desired the speaker is connected where Ch is and the condenser C2 is omitted. If the condenser in C is desired the primary of the transformer is connected where the choke is, the condenser C2 again being omitted.

It should be noted that in each of the circuits B, E and F the loudspeaker returns to filament or cathode. In B it returns to minus A, in E it goes to the cathode, and in F, to the mid-tap of the filament transformer. These are the best returns possible for the different types of tubes. The object of the direct filament, or cathode, return is to minimize feedback through the plate voltage supply.

It should also be observed that in each of the circuits in Fig. 9 there is a condenser C connected from B plus to the filament or cathode. While this condenser is not absolutely required in every instance, it always serves a good purpose. The object of this condenser is also to eliminate feedback. If the condenser is used at all it should be at least 1 microfarad. The larger it is the more effectively will it prevent feedback.

Battery Operated Amplifiers

Although battery operated amplifiers can be called power amplifiers only by courtesy since the undistorted power output is not considerable, they fall, within the bare technical classification, and formed the basis for the early lore of audio amplification.

Three types will be described, namely, two-stage transformer coupled, three-stage resistance coupled, and two-stage transformer coupled with the output stage to push-pull.

The most common in battery receivers is the two-stage transformer coupled, one form of which is shown in Fig. 10. This circuit comprises two audio transformers, T1 and T2, two amplifier tubes, (1), and (2), and output filter ChC1. In addition to the main elements it contains two Amperites (filament ballast resistors), R1 and R2, and a filament switch Sw. Filament, plate and grid batteries are supposed to be used with it.

The requirements of the audio transformers are the same as those of transformers used in any other similar circuit, regardless of the power supply, provided similar tubes are used. Since nearly all general purpose tubes are designed to have nearly the same output characteristics, most audio transformers designed for quality reception are suitable. Both transformers may be of the same design, but if one of two available transformers has a higher primary impedance, which usually means a lower step-up ratio, that should be used in the first stage, assuming the amplifier is to be connected to a detector tube,

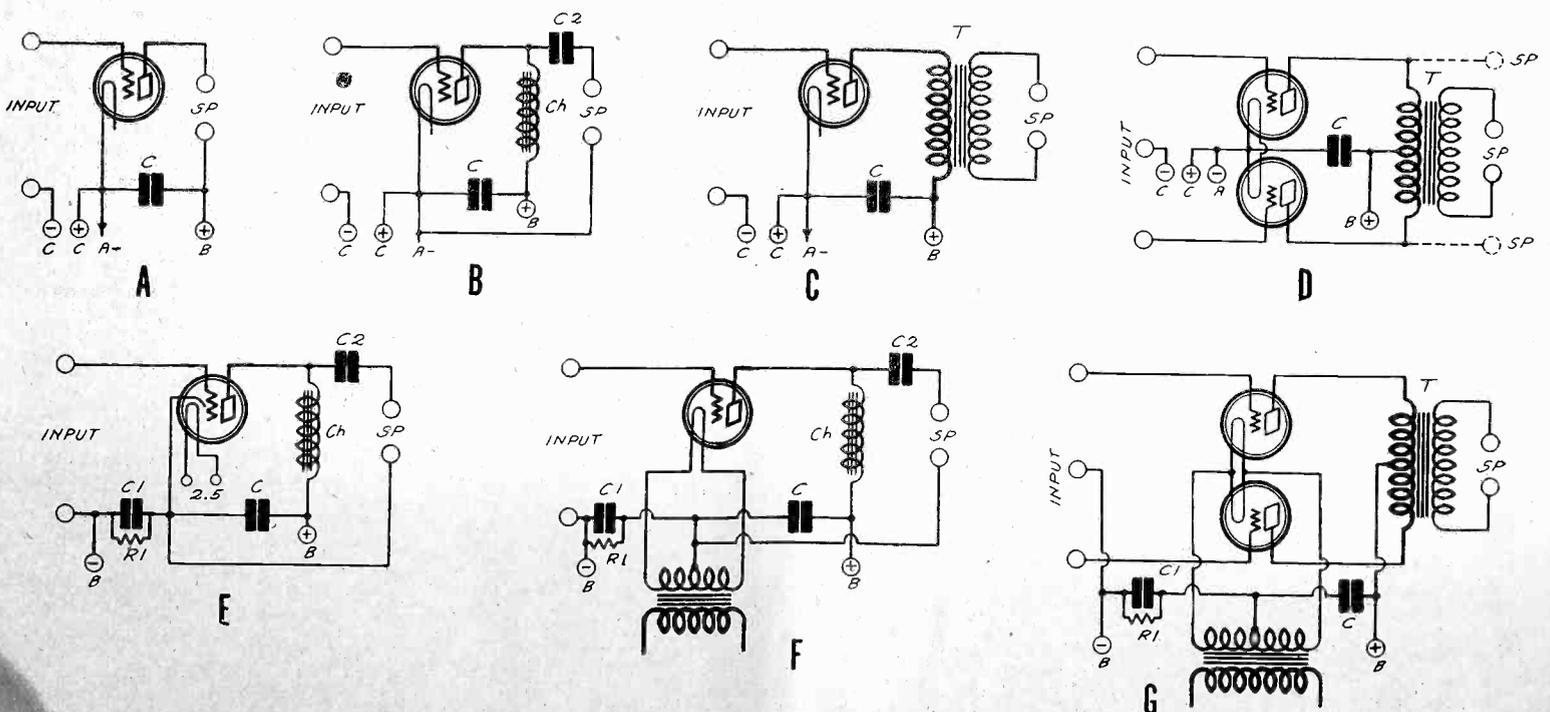


FIG. 9. THE THREE TYPES OF COUPLING BETWEEN THE POWER TUBE AND THE LOUDSPEAKER AS APPLIED TO DC AND AC TUBES. A—SPEAKER CONNECTED DIRECTLY IN THE PLATE CIRCUIT; B—THE OUTPUT FILTER WITH DC TUBE; C—OUTPUT TRANSFORMER WITH DC TUBE; D—PUSH-PULL OUTPUT TRANSFORMER WITH AC TUBES, AND THE PLATE-TO-PLATE CONNECTION; E—OUTPUT FILTER WITH HEATER TUBE; F—OUTPUT FILTER WITH AC FILAMENT TUBE; G—PUSH-PULL WITH AC FILAMENT TUBES.

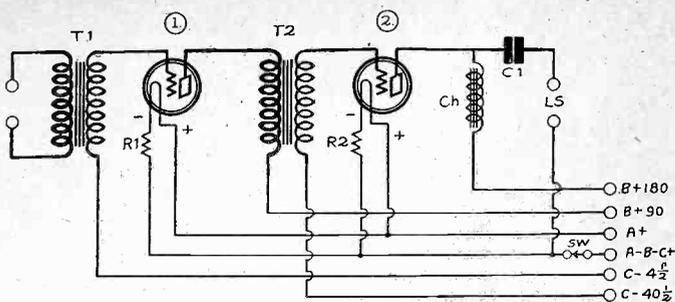


FIG. 10

TWO-STAGE, TRANSFORMER COUPLED, BATTERY OPERATED AUDIO FREQUENCY AMPLIFIER.

since the impedance of the detector tube to audio frequencies is very high.

The choice of tubes for this amplifier depends on the purpose for which the circuit is to be used. If it is to be a portable amplifier, the first tube might be a 199. In that case the last tube should be a 120, the plate voltage being 135 volts instead of 180 volts, and the negative grid bias $22\frac{1}{2}$ volts instead of 40.5 volts. The filament voltages must be adjusted to suit the tubes. Also, if the 120 tube is used it is not necessary to use an output filter, but the loudspeaker, or earphones, may be connected directly in the plate circuit of the second tube.

For broadcast reception in a home a better choice of tubes is a 201A for the first tube and a 171A for the second. An output filter is necessary with this combination of tubes, for the 171A requires a high plate current. The plate and grid voltage values given in Fig 10 assume that this combination of tubes is used. The plate voltage on the first tube is 90 volts and that on the second is 180 volts. The bias on the grid of the first tube is 4.5 volts and that on the second is 40.5 volts. These are standard values and easily obtainable.

The filament voltage source, of course, is six volts, obtained from a storage battery. The filament voltage is cut down to 5 volts, at .25 ampere, by means of 1A Amperites.

The choke Ch should be preferably a 100 henry coil, measured with no DC in the winding, and it should be large enough in physical dimensions to avoid undue drop in the inductance with the flow of direct current. Certainly it should not be less than 40 henries when 20 milliamperes are flowing.

The stopping condenser C1 should be not less than 4 mfd. The value depends somewhat on the speaker, but the capacity could not be too large in any case.

Since the amplifier is supposed to be used with plate and grid batteries, by-pass condensers have not been included in the circuit. If the plate battery is of the storage variety no by-pass condensers will be required at any time, but if dry cell plate batteries are used, the familiar B batteries, and if they are allowed to become exhausted, distortion or even oscillation might result without condensers. However, when distortion due to the battery resistance becomes appreciable, there is so little life left in the battery that it is not worth while to employ condensers to attempt to prolong it. Noticeable distortion of this enlarging type is one warning to replace the battery. The grid battery usually will last a year or more without replacement, no current is drawn from it, and its life in such use is about equal to its shelf life. Even if a voltmeter indicates a 50 per cent reduction in the voltage, the grid bias value of the batteries is practically the same as that of a new battery. This does not apply to all circuits.

It is of utmost importance to keep the filament storage battery fully charged. The amplifier will not work well when the voltage is below six volts. If the battery is permitted to discharge completely periodically, it will not last long. A hydrometer should be kept handy at all times for testing the charge.

Resistance Coupled Amplifier

A resistance coupled amplifier such as that shown in Fig. 11 is particularly suited to battery operation. Indeed, only in a battery operated circuit can this type of amplifier be used in its simple form. If any other source of plate voltage is used it is necessary to take utmost precautions against audio frequency oscillation. With batteries it gives little trouble, except when the batteries are old and nearly exhausted. When the circuit starts to misbehave, just as in the case of the transformer coupled circuit, it is time to replace the battery.

If a storage battery is used to supply the plate voltage there will be no trouble as long as the battery is good.

The first two tubes in this circuit may well be of the high gain type. A special purpose tube, the 240, with an amplification factor of 30 has been especially designed for the circuit. With two such tubes a total voltage step-up of 500 can be obtained between the grid of the first and the grid of the third tube. General purpose tubes also may be used, but with a lower gain. Tubes like 201A with an amplification fac-

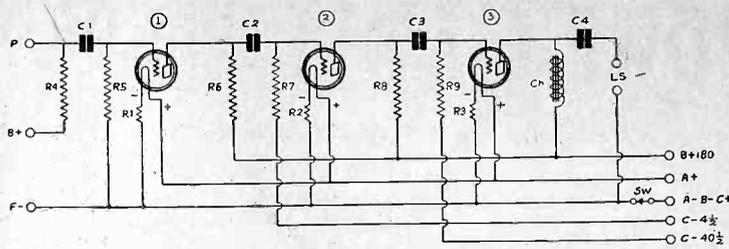


FIG. 11

THREE-STAGE, RESISTANCE COUPLED, BATTERY OPERATED AUDIO FREQUENCY AMPLIFIER.

tor of 8 might yield a gain of between 50 and 60, depending on the values of the coupling impedances.

The higher the coupling impedances the higher will be the amplification.

The lowest values for the resistors R4, R6, and R8 should be 100,000 ohms. If they are increased to 250,000 ohms the amplification will be considerably greater. For higher values the increase in the amplification will not be great.

But it is of no value to increase the coupling resistors if the grid bias resistors R5, R7 and R9 do not remain considerably larger than their corresponding plate resistors. The first two, R5 and R7, should be at least 2 megohms. The last, R9, should be equal to this if the circuit will operate without audio oscillation when such a value is used. If it must be lowered it should be kept as large as it can be without losing stability.

The values of the stopping condensers, in general, should be as large as practicable, but it is of doubtful value to increase the size if that requires using condensers with paper dielectric. The best condensers to use are those with mica dielectric, standard values of which come as high as .02 mfd. This is large enough provided the grid leak associated with any condenser is not less than 1 meg.

A high grid leak resistance and a large capacity condenser will permit full amplification of all the low frequencies. A low value leak and a low value stopping condenser will suppress the amplification of the low notes. The product of the capacity of the condenser in microfarads and the resistance of the leak in megohms should not be smaller than .02, unless, for any reason, it is desirable to suppress the low notes. This product of .02 holds no matter what is the capacity of the stopping condenser.

Satisfactory operation of battery amplifiers of this type have been obtained, with two high mu tubes, with 250,000 ohms for the plate resistors, .02 mfd. for the stopping condensers, and 2 megohms for the grid leaks. In some instances R9 has to be reduced to as low as 250,000 ohms.

There is another reason besides instability for reducing the value of the last grid leak. Leakage through and over the insulation of the grid often makes the grid positive notwithstanding the high value of applied grid bias. A low value of grid leak will maintain the potential of the grid at the correct operating value.

The amplifier shown in Fig. 11 is more stable than one containing only two tubes, in resistance coupling, provided that all the tubes, including the detector, receive their plate voltage from the same source. The use of high values of plate resistors increases the stability, and the return of the speaker to the A battery increases it still further.

There are two frequency regions in which instability might occur when the plate battery has a high resistance, at very low frequencies and at very high. The high frequency instability can be remedied by connecting a condenser, about 1 mfd., across the battery. The low frequency instability can be remedied by reducing R9, or by replacing a deflated B battery with a fresh one.

When this amplifier is connected to a detector tube the three terminals at the left should be connected to the tube as indicated. If the same storage battery is used for both the detector and the amplifier, the F connection should be omitted, for it is made automatically through the B battery. The B plus terminal may be connected to the 180 volt tap on the plate battery, or to any lower tap.

When the amplifier is connected to a phonograph pick-up unit, the P and the F terminals should be used. B plus detector then should not be connected to anything. The pick-up unit may also be connected across R5, or in place of R5.

The final tube in this amplifier is supposed to be a 171A, and for that reason the output filter ChC4 is employed. The specifications for this filter are the same as those given for the transformer coupled receiver in Fig. 10.

Since all the tubes specified are five volt tubes, the source of filament current should be a six volt storage battery. The three ballast resistors R1, R2, and R3 should be 1A Amperites.

The resistance coupled amplifier is capable of excellent quality provided that the plate voltage battery has a negligible internal resistance, or provided that special precautions be taken to nullify the feedback through any resistance that may exist.

* * *

[Part III of "Power Amplifiers" next week]

Radio University

QUESTION and Answer Department conducted by RADIO WORLD, by its staff of experts, for University members only.

When writing for information give your Radio University subscription number.

IS IT REALLY necessary to use a baffle board with a dynamic speaker to get the low notes? I have such a speaker and I can hear the low notes very well.

(2)—If it is necessary to use a baffle board with a dynamic, why is it not equally necessary to use one with magnetic speakers having a similar cone structure?

(3)—Can you suggest a simple way of demonstrating the effect of a baffle board, one which shows the difference convincingly?

(4)—If you recommend baffle boards, what do you consider the best size and best material for them?

DON McLEOD,
Toronto, Canada.

(1)—It is not necessary, but it is highly desirable to use a baffle with speakers having small cones or any small sound radiating surface. The low frequency cut-off due to the absence of a baffle board is not very sharp. The low notes come through, but with greatly reduced volume.

(2)—It is even more desirable to use a large baffle with a magnetic speaker having a small cone.

(3)—Get a baffle board of fairly large dimensions. Put the cone up against the aperture. Note the quality of the sound. Then tilt the speaker away from the aperture and note the striking difference. This is the simplest and the most convincing demonstration of the effect of the baffle board.

(4)—The larger the baffle board the better. A compromise is often struck on 24 in. or 36 in. square. There is no one material that is best. The material must be rigid so that it will not vibrate.

* * *

I RECENTLY got a severe shock from my power pack after I had turned off the power. Where did the voltage come from?

(2)—What precautions would you recommend to prevent such shocks in the future?

(3)—Can static-like noises be introduced by the ground used with the receiver? I have a set which I can use both with a loop and an antenna-ground, and there is not nearly so much noise in the output when I use the loop as when I use the antenna.

BORIS LENSKY,
Toledo, Ohio.

(1)—It came from the condensers in the filter.

(2)—Keep your fingers away from the power pack until the charges on the condensers have had a chance to leak off.

(3)—The ground is often a prolific cause of noise. But the loop does not necessarily demonstrate the fact, for the loop always picks up less noise than an antenna and ground. Install a good counterpoise and switch from the ground to the counterpoise, using the same antenna, and then compare the noise.

* * *

IS IT POSSIBLE to improvise high resistance voltmeter by using a 0-1 milliammeter in series with a high resistance?

(2)—If it is, please tell how to figure the resistance to be connected in series.

(3)—Is it also possible to use an ammeter to improvise a voltmeter?

WILLIAM H. FRANCIS,
Jefferson City, Mo.

(1)—That is how every 1,000-ohms-per-volt voltmeter is made.

(2)—Figure on 1,000 ohms for every volt you want to measure, full scale. If the range of the voltmeter is to be 0-1

volt, use 1,000 ohms. If you want the range to be 0-100 volts, use 100,000 ohms. If you want to make the scale 0-500 volts, use half megohm. Readings below the full scale are proportional to the scale. Use accurate resistors, for the accuracy of the readings will depend on this.

(3)—Not possible, except, perhaps, to measure the voltage of a storage A battery.

* * *

I FREQUENTLY read in technical articles and books that the time constant of the grid leak and grid condenser in a direct coupled circuit should be small to get full response on the high notes, but I have also seen statements in Radio World that it should be as large as possible. Which is correct?

(2)—Under what conditions should the time constant be small in order that the circuit respond quickly to rapidly changing voltages such as those occurring in high frequency signals?

(3)—How is the time constant measured and what is its meaning?

CHARLES H. BURT,
Baltimore, Md.

(1)—Somebody once made a mistake by saying the time constant should be small and some writers have carried on the error. The time constant should be large.

(2)—If the condenser and the resistance are in parallel and so placed in the circuit that the condenser must be charged up before the full voltage is attained, then the time constant should be small. The stopping condenser in a resistance or impedance coupled circuit is in series with the grid leak, and the only condition is that it be large so that it does not offer any impedance to the low frequencies.

(3)—The time constant is measured by the product of the resistance and the capacity, ohms and farads being the units. The results is measured in seconds. The meaning of the time constant might be explained as follows: Suppose a condenser of capacity C be charged to a voltage V and that a resistance of R ohms be connected across the charged condenser, then the time constant RC is the number of seconds that will elapse before the voltage across the condenser has been reduced to 1/2.718 of its original

value. The number 2.718 is the base of natural logarithms.

* * *

THERE IS a terrific hum in my Screen Grid Universal receiver, which is operated by A and B battery eliminators. What is the cause of this hum?

(2)—The volume of this set is terrific and it cannot be controlled with the rheostat in the filament circuit of the screen grid tube, which has a resistance of 20 ohms. What would you suggest for controlling the volume?

(3)—Is it possible to use a 171A power tube in the last stage?

SIGWALD ORRE,
St. Paul, Minn.

(1)—The hum is due to a combination of two things, regeneration and residual hum in the eliminators. Control the regeneration and the hum will disappear, unless the eliminators are hopelessly overloaded.

(2)—Use the rheostat recommended—50 ohms.

(3)—Yes, it is possible if you apply the proper plate and grid potentials to the power tube. You should also use an output filter or transformer to protect the loudspeaker.

* * *

WHAT IS THE OBJECT of decreasing the filament voltage on AC tubes as compared with DC tubes? I refer especially to the reduction in the filament voltage in the design of the 245 tube.

(2)—Is it practical to heat the filaments of a 245 and one or more 227s by the same 2.5 volt winding?

(3)—Is there any marked advantage of using a 250 tube over a 245 in a regular home receiver? If so, what?

LOUIS G. WILLIAMS,
Paterson, N. J.

(1)—There are two objects. One is that if the filament voltage is low the hum produced in the tube due to the AC will be less the lower the filament voltage. The other is that it simplifies the filament transformer that may be used.

(2)—It is practical in most instances. There are circuit arrangements in which it is not advisable to use the same winding but in most circuits there is no reason why the same winding should not be used.

(3)—The 245 tube in conjunction with a good speaker will give enough undistorted volume for any home. The 250 will give much more but it is not needed. Of course, for a given output that from the 250 will contain less harmonic distortion than that from the 245, but the distortion cannot be detected by the ear until the volume is so loud as to be unpleasant.

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- A. E. Averett, 16 Bedford Rd., Schenectady, N. Y.
- Arthur H. Pullon, 17 Forest Ave., Ansonia, Conn.
- W. J. Lyons, Blausers Elec. Co., 25 Main St., Oil City, Pa.
- Harry L. Evans, Binghamton, N. Y.
- Oliver Brothers, Inc., 71 Murray St., New York City.
- Dr. C. L. Zimmerman, 412 Cumberland St., Lebanon, Pa.
- Ernest E. Palmer, Box 122, No. Stonington, Conn.
- W. J. Kulabonish, 809 Washington St., Martins Ferry, Ohio.
- John T. Hosey, M.D., Palatka, Fla.
- L. S. Forsyth, 515 5th St., Portsmouth, Va.
- G. F. Schaaf, 430 W. Grand Ave., Lima, Ohio.

- W. C. Wells, P. O. Box 230, Brantford, N. Dak.
- E. Stead, 138 Eastwood Ave., Toronto, 13, Ont., Can.
- Edward Morrill, 16 Harding St., Worcester, Mass.
- Robert L. Potter, care Yale & Towne Mfg. Co., Stamford, Conn.
- Geo. Anglim, P. O. Box 13, Duncan, B. C., Can.
- John Howard, 5 Willow Ave., Towson, Md.
- Wm. R. Scott, Box 125, Rt. 2, Holmes Rd., Pittsfield, Mass.
- H. de Long, 41 R. St., N. E., Washington, D. C.
- A. J. McNally, N. E. Cor. 40th & Woodland Ave., Philadelphia, Pa.
- Avery Macklem, 1700 Atkinson Ave., Detroit, Mich.
- Robt. J. Orr, 3503-A Nat. Bridge, St. Louis, Mo.
- Community Garage, 388 Belleville Ave., Bloomfield, N. J.
- Andrew Feilsch, Sr., 2720 Van St., Cincinnati, O.
- George Weller, 5233 Warnock St., Phila., Pa.
- L. E. Funk, 637 Linden St., Bethlehem, Pa.
- M. J. Loir, 72 Walden St., No. Cambridge, Mass.
- Edgar Van Gilder, 2407 W. 32d Ave., Denver, Col.
- Wm. T. Brooks, P. O. Box 246, Herkimer, N. Y.
- Oliver A. Ayers, Mohawk Lake, N. Y.
- Joseph Turner, 3104 No. Hancock St., Phila., Pa.
- G. W. Hill, 925 Whitney Bldg., New Orleans, La.
- Geo. E. Harbour, 342 Pearl St., Cambridge, Mass.
- Peter L. Czajkowski, 202 E. Pine St., Olean, N. Y.
- Johnston's Radio Shop, Mr. K. S. Johnston, 9008 Jamaica Ave., Woodhaven, N. Y.
- Frank Idner, 309 Okeechobee Road, West Palm Beach, Fla.
- E. Langworthy, Box 104, Matteson, Ill.
- N. A. Sell, City Produce Co., Union St. & B. & O. R. R., Cumberland, Md.
- James Brennan, 73 Poulton Ave., Toronto, Ont., Can.
- C. J. Werila, 622 W. Gandy St., Denison, Tex.
- W. E. Barnes, 4119 N. Broad St., Phila., Pa.
- Grant Irwin, Sulphur, Akla.
- M. A. Wesling, 1901 Denver St., Covington, Ky.
- Dr. Frank C. Kinsey, 315 Powers Theatre Bldg., Grand Rapids, Mich.
- Chas. A. Winsor, Evansville, Wis.
- Geo. G. Mitchell, Jr., P. O. Box 55, Hacketts-town, N. J.
- C. F. Sleeman, Dixie, Ont., Can.
- F. L. Pittman, 108 W. Patent Office, Washington, D. C.
- John G. Rieger, Radio Editor, Buffalo Evening News, 216-218 Main St., Buffalo, N. Y.
- S. J. Ceranski, 6043 Harrell Ave., Detroit, Mich.
- Geo. E. Grosvenor, R. F. D. No. 1, Olyphant, Pa.
- T. A. Jennings, 209 W. 12th St., Cincinnati, Ohio.
- John R. Purcell, Port Jervis Radio Co., Port Jervis, N. Y.
- Gordon W. Brown, 38 Elgin St., Lindsay, Ont., Can.
- Harry M. Wise, 122 E. Poplar Ave., Wildwood, N. J.
- Fred Breidenbach, 411 N. Bancroft St., Indianapolis, Ind.
- H. Holbrook, 19 4th St., Lakewood, N. J.
- W. A. Spangler, 41 Burns Ave., Lokland, Ohio.
- Alfred Barcklow, 270 W. 2nd St., Moorestown, N. J.
- Ernest G. Keler, 351 E. Lincoln Ave., McDonald, Pa.
- Robt. B. Pfancook, 6 E. Diamond Ave., Hazleton, Pa.
- W. A. Elmore, 113 Fairley St., Laurinburg, N. C.
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- Mrs. W. E. Paine, Zealandia, Sask., Can.
- Geo. H. Keller, 1048 Lorena St., Los Angeles, Cal.
- J. W. Newson, Medina, Tenn.
- Robt. C. Hopkins, 5205 14th Ave., Brooklyn, N. Y.
- Geo. G. Steele, 8113 Maple St., New Orleans, La.
- M. H. Kennedy, Art Dept., "The Sun-Telegraph," Pittsburgh, Pa.
- Stanley Glenn, Perry, Mich.
- Bert P. Hanavan, 1658 Farwell Ave., Chicago, Ill.
- Waechter Lumber Co., Harper, Iowa.
- John C. Hoffmann, 511 Lexington Ave., Brooklyn, N. Y.
- S. A. Vatter, Directory Section, Main P. O., Grand Rapids, Mich.
- James Chapman, 2847 N. Marvine St., Phila., Pa.

- Jos. D. McGrath, 302 W. Jefferson St., Ft. Wayne, Ind.
- F. K. Hodges, Clark Hotel, Roanoke, Va.
- Fred F. Norris, 604 S. Spring St., Nevada, Mo.
- C. B. Smith, The Park Club, Pittsfield, Mass.
- Harry A. Strong, 2520 10th St., N. E., Washington, D. C.
- Fred L. Cutler, 720 Tonawanda St., Buffalo, N. Y.
- W. H. Collins, Jr., Bowen High School, 8860 Manistee Ave., Chicago, Ill.
- A. S. Bailey, The Radio Shop, 2114 Gentry St., Houston, Tex.
- Albert Stone, 1444 E. 10th St., Brooklyn, N. Y.
- Clarence A. Brodt, Columbine, Ohio.
- E. B. Carpenter, 106 Poultney Ave., Buffalo, N. Y.
- W. J. Deane, 13th Floor, Cosden Bldg., Tulsa, Okla.
- Adolph F. Fey, 4443 No. Chadwick St., Phila., Pa.
- Clement Preller, 2025 T St., N. E., Washington, D. C.
- F. H. Merriam, M. D., 1000 Washington St., So. Braintree, Mass.
- Archie Pinkley, 5161 Holcomb, Detroit, Mich.
- F. C. Rogers, Alton, Wayne Co., N. Y.
- A. P. Seidel, 2208 Pasadena Ave., Los Angeles, Calif.
- W. G. Yarrow, 1908 W. 15th St., Davenport, Iowa.
- Georges Payen, Ste-Theole, co. Champlain, P. Q., Can.
- Clarence D. Bussars, 104 17th St., S. E., Washington, D. C.
- A. J. Anderson, Angola, N. Y.
- G. E. Helbeit, 6761 4th Ave., Los Angeles, Calif.
- Robert H. Roth, Sidney, Ohio.
- L. A. Morrow, care The Dow Drug Co., Springfield, Ohio.
- J. E. Tracy, 220 Victoria St., Jacksonville, Fla.
- R. L. Watson, 2050 John Ave., Butte, Mont.
- Wm. A. Harding, 2013 No. 29th St., Phila., Pa.
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- B. C. Coeman, Box 1742, Yardley, Wash.
- A. B. Erickson, 7935 Kimbark Ave., Chicago, Ill.
- Ernest F. Wright, P. O. Box 273, Lansford, Pa.
- Arthur Vinton, P. O. Box 83, Southbridge, Mass.
- Chas. M. Ross, 403 King St., E., Hamilton, Can.
- Bob Stutz, 1128 Mackow Dr., Toledo, Ohio.
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- J. C. Doss, 1026 Hanover Ave., Roanoke, Va.
- Arthur Walker, 155 Abbott St., Lawrence, Mass.
- Joe Rauh, R. F. D. No. 1, Kent, Iowa.
- W. Baker, 611 Galt Ave., Verdun, Montreal, P. Q., Canada.
- Edwin Anderson, 1068 71st St., Brooklyn, N. Y.
- C. Wills, 132 Queen St. (East), (T2), Toronto, Can.
- John McKenzie, 54 12th St., Norwich, Conn.
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- Howard B. Stevens, 49 High Street, Waltham, Mass.
- H. E. Beavens, 216 10th St., S. E., Washington, D. C.
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- L. C. Cover, 170 E. 2nd St., Peru, Ind.
- Wm. C. MacKay, 271 Washington St., Newtonville, Mass.
- Sam H. Dooley, 717 Glenn St., Cumberland, Md.
- P. J. Kuprion, 600 Barbee Ave., Louisville, Ky.
- Robert E. Beaupre, 15 Allen St., Burlington, Vt.
- Television Radio Shop, 115 Winter St., Haverhill, Mass.
- Theodore C. Janes, 36 So. Union St., Burlington, Vt.
- James Watkins, P. O. Box 72, Russell, Ky.
- D. M. LeBlanc, 65 Richmond Hill Ave., Stamford, Conn.
- J. E. Bures, P. O. Box 7117, Mexico, D. F., Mex.
- Ray Hoffman, 364 Williams Ave., Hackensack, N. J.
- Ivan F. Pierson, 736 Evergreen, Bremerton, Wash.
- L. R. Hutchison, 120 Patterson Ave., Butler, Penna.

Choose Your Speaker from This Complete Array!

EXPONENTIAL TYPE HORNS

Modern acoustical science is striving to equal the performance of a large air column horn with powerful unit, while the horn enjoys its rightful popularity with trained experts. The larger the horn, the better, hence we offer two models: one with 7 1/2 ft. tone travel, the other (where space permits) with 10 ft. tone travel. The material used is patented Racon. Nozzle is standard size.



Cat. 200
\$7.50 Net

This horn has a 92-inch air column. No resonance peaks. Front, 18"x18". Depth, 18 1/2". Weight, 5 lbs.



Driving motor, the unit needed to work the air column horns. Standard size thread. Cat. 303. Price, \$3.50 net.



Cat. 300
\$10.50 Net

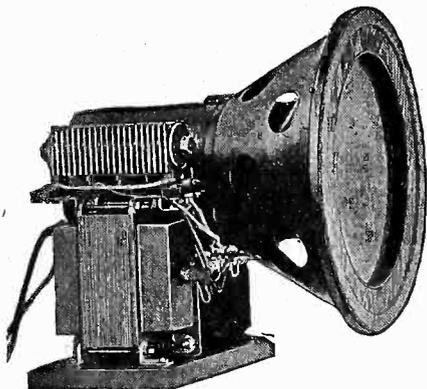
The larger horn is preferable, where space permits. Air column, 120". Front, 18"x18". Depth, 18". Weight, 7 lbs.

DYNAMIC CHASSES and Baffle

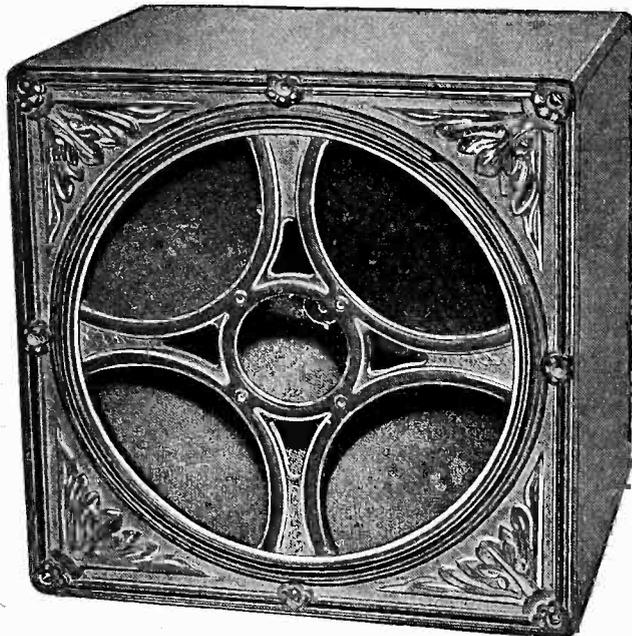
The dynamic speaker is the most popular one by far, and here is your opportunity to get a real fine chassis at a low price. Cat. 110 A.C. operates directly from the 110-volt A.C. (alternating current) lamp socket, to which built-in plug is connected, while the tipped cords go to your receiver output. Dry rectifier and output transformer built in this model.

Those whose place is wired with 110-volt D.C. (direct current) should use Cat. 110 D.C. @ \$17.50 net. Those who have no electricity should use the model that works from a 6-volt storage battery. Cat. 6 D.C. @ \$14.75 net.

At left is illustrated an 18"x18" baffle, Cat. 111, with cane sides and top, for any dynamic speaker. Specify speaker. Walnut 5 ply veneer. Price \$11.00 net.



Cat. 110 A.C.; Price, \$20.50 Net



Cat. 113
Price, \$13.50 Net

New Model Polo Speaker, with five-ply veneer walnut housing, moulded, decorated metal front piece, and containing Pole Twin Magnet Unit and Textile Cone. All ready to play. Stands 150 volts without filtering. Will work fine from any output tube from 201A to a pair of push-pull 250s without rattling.



Cat. 111; Price, 11.00 Net

Cat. 110 A.C., shown inside, \$20.50 extra.

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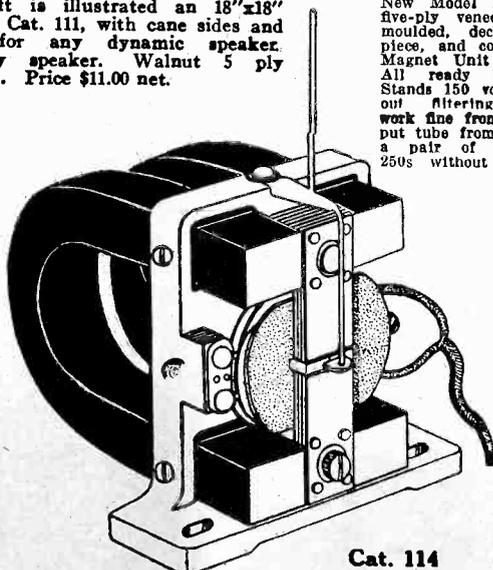
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| <input type="checkbox"/> Cat. No. 200 | <input type="checkbox"/> Cat. No. 111 |
| <input type="checkbox"/> Cat. 300 | <input type="checkbox"/> Cat. No. 113 |
| <input type="checkbox"/> Cat. No. 110 A.C. | <input type="checkbox"/> Cat. No. 114 |
| <input type="checkbox"/> Cat. No. 110 D.C. | <input type="checkbox"/> Cat. 114A |
| <input type="checkbox"/> Cat. No. 6 D.C. | <input type="checkbox"/> Cat. 115 |
| <input type="checkbox"/> Cat. No. 300 | <input type="checkbox"/> Cat. 116 |
| <input type="checkbox"/> Please send C.O.D. | <input type="checkbox"/> Cat. No. 303 |
| <input type="checkbox"/> Remittance enclosed. | <input type="checkbox"/> Please send prepaid. |

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Address

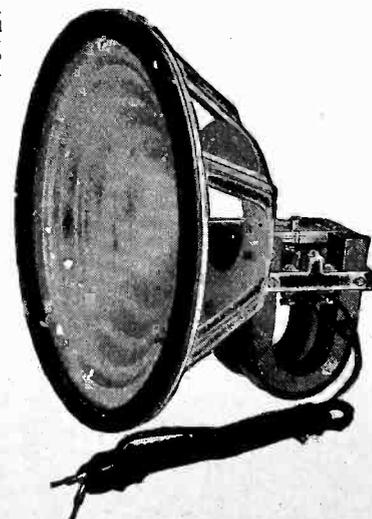
City State

5-DAY MONEY-BACK GUARANTEE



Cat. 114
Price, \$9.25 Net

Pole Twin Magnet Unit—weight, 3 1/4 lbs., or twice as heavy as ordinary unit. Twin magnets double sensitivity. This unit gives more volume, clearer tone, and stands the gaff. Supplied with 10-ft. cord. Cat. 114. Tri-foot moulded unbreakable metal mounting bracket and apex constitute Cat. 114A @ \$0.75.



Cat. 115; Price, \$11.50 Net

Molded 9" spider, unbreakable metal, with Textile cone and felt ring and apex, and Pole Unit mounted on the assembly, which stands on own feet. Cat. 115.

LABOR STATION LOSES ITS PLEA FOR FULL TIME

Washington.

The application made by the Chicago Federation of Labor for full time and 50,000 watts on a cleared channel for its station, WCFL, has been denied by the Federal Radio Commission. The station's frequency is 770 kc.

After the application was filed the Commission notified the station that it was not satisfied that public interest, convenience, or necessity would be served by the grant, and the station was notified that it would be given a hearing on its application. Such hearing was held. The applicant was permitted to offer testimony to show that public interest would be served by the granting of the license requested.

WBBM, Glenview, Ill., and KFAB, Lincoln, Nebr., were made respondents because they operate on the same frequency with WCFL. The two stations were represented at the hearing by speakers who opposed the application and offered testimony to show that the granting of the application would not be in the public interest.

Super Power Clarostat Rated at 250 Watts

The newest addition to the Clarostat line is the Super-Power Clarostat, a heavy-duty adjustable resistor. It is a Clarostat compression device for obtaining stepless and noiseless resistance that remains set at any value desired. The Super-Power Clarostat is furnished in three resistance ranges.

Filament range .. ¼—10 ohms
Low range 25—500 ohms
Universal range .. 100—100,000 ohms

It is built for heavy-duty service and is capable of withstanding high temperatures when dissipating up to its maximum capacity of 250 watts. It is provided with a long shaft and special bracket, enabling it to be mounted clear of the panel for proper air circulation, or, again, for mounting on thick slate panel. It is intended as a heavy-duty line control, variable speed motor control, plate voltage control for transmitters, field control for shunt type generators, etc. Full information may be had from the Clarostat Manufacturing Co., Inc., 291 North Sixth Street, Brooklyn, N. Y. Mention Radio World.

Crosley Sends Programs via W8XAL on 49.5 m.

Cincinnati

Broadcasting the programs of WLW, the Crosley Radio Corporation's short-wave transmitter, W8XAL, is on the air once more after several months of silence during which it has been completely rebuilt and moved from the old WLW site at Harrison, O., to the new Crosley transmitter building at Mason, O.

W8XAL is licensed to operate with 250 watts power on a frequency of 6060 kilocycles (49.5 meters). One hundred per cent modulation is claimed.

The short-wave broadcasts are from 6:15 A. M. until 11:00 A. M.; from 1:30 P. M. until 2:30 P. M., and from 6:00 P. M. until the long-wave station signs off at night.

Chain Limitation Again Deferred

Washington.

Another postponement of the order of the Federal Radio Commission, adopted last November, proposing to limit duplication of chain broadcasting programs on cleared channels to stations separated by more than 300 miles, was announced by the Federal Radio Commission.

In a new general order (No. 63), the Commission announced that General Order No. 43, which provided for the restriction, has been deferred from June 1st to October 1st. This is the fourth time the order has been deferred, it having been originally promulgated to become effective with the reallocation of broadcasting stations of November 11th.

It was explained at the Commission that following the issuance of the order, a number of broadcasting stations complained vigorously, and the order was deferred pending further investigation.

Kolster Plans to Sue Nineteen Set Makers

Nineteen set manufacturers have been notified by Kolster Radio Corporation, through a subsidiary, the Federal Telegraph Company of California, of claims for infringement of four patents on gang condensers and vertical drum dials. Suit is threatened.

A test case is to be made and large damages demanded. Among the companies notified are Grigsby-Grunow Company, manufacturers of Majestic sets; National Carbon Company, makers of Everready sets; Crosley, Zenith, Sparks-Withington, makers of Sparton sets; Steinite, A. C. Dayton, Stewart-Warner, Silver-Marshall, Philco, All American Mohawk, Gilfillan and Sonora.

CeCo Licensed Under RCA Tube Patents

The CeCo Manufacturing Company, Inc., of Providence, R. I., third largest tube manufacturer, with a capacity of 45,000 tubes daily, has taken out a license from the Radio Corporation of America to manufacture under RCA tube licenses. CeCo is the second independent to obtain a license. The first was the Raytheon Manufacturing Company, the sale of whose tubes has since passed to exclusive control of the National Carbon Company, which is backed by General Electric.

RCA and Cunningham, the latter owned by RCA up to 51% of the stock, are first and second in tube production and sales.

Schnell Heads Staff Of Aero Engineers

Lt. Commander F. H. Schnell, U. S. Naval Reserves, for six years traffic manager of the American Radio Relay League, and more recently with the engineering and research laboratory of the Burgess Battery Co., has become chief radio engineer of Aero Products, Inc. of Chicago, specialists and manufacturers of short-wave radio equipment.

Commander Schnell is now working on some new developments in short-wave, which will shortly be announced from the Aero Products laboratories. Commander Schnell's experience in short-wave radio development covers a wide range of activities.

LAW IS ASKED FOR LIGHTS ON STATION MASTS

Washington.

The Federal Radio Commission is without authority to require radio stations to paint and illuminate their towers near commercial airways as a safeguard for aviation, nor has the Secretary of Commerce the power under the Air Commerce Act to make such a requirement, according to the opinion of the general counsel for the Commission, Bethuel M. Webster, Jr. Mr. Webster's opinion has been placed in the record of the hearings of the Senate Committee on Interstate Commerce on the Couzens bill (S. 6) to establish a Commission on Communications.

The matter was brought up at the hearings of the committee by the statement of Ira E. Robinson, a member of the Radio Commission, that the Assistant Secretary of Commerce, William P. MacCracken, Jr., had called the attention of the Commission to the failure of the Radio Corporation of America to light towers in Brunswick Meadows, N. J., and the taking off of the lights on the towers at Cleveland, Ohio. The opinion of Mr. Webster suggested additional legislation to cope with the situation.

Intelligence Confined to Earth, Says Maxim

According to Hiram Percy Maxim, the inventor, all intelligence is confined to the earth. Speaking at a dinner concluding the convention of the Hudson Division of the American Radio Relay League, a national organization of radio amateurs, he said:

"We have received nearly every frequency in the spectrum from the lowest relation to the highest, which is known as the Milliken Cosmic Ray.

"Is it not odd that of all these radiations from inter-stellar space there is not a scintilla of evidence that any of these radiations were sent out by an intelligent being. So may we not judge from that that on this little earth of ours resides the only intelligence of all cosmos?"

Transformer Leads for "Universal 250"

The connections for the power transformer used in the Universal 250, begun in last week's issue, and concluded in the present issue, are: red, to plate of 281 rectifier; brown, midtap of 7.5 volt power tube filament winding; blue, midtap of 1½ volts that are next to the 450-volt post; yellow, B—; green pair, 7.5 volts for filament of rectifier; black, to ground.

BID ON LEAGUE STATION

Geneva

Two important European radio companies, the Compagnie Generale de Telegraphie of Paris and the Telefunken Company of Berlin, have made a joint proposal to build the radio station which the League of Nations is considering erecting.

A THOUGHT FOR THE WEEK

A COMMERCIAL check-up of different businesses shows that there are nearly five hundred radio stores in Greater New York—and that a few of them are not in the Cortlandt street district.

**For Those
Screen Grid Tubes—**

Don't guess at grid bias voltages! Screen grid tubes are precision devices that do not react well to blacksmith methods. For the D.C. screen grid tube, try the 20-ohm HUNDINGER as filament resistance and adjustable grid bias. For A.C. screen grid tube, try the VOLUME CONTROL CLAROSTAT.

Your dealer can show you both these devices, as well as other items of the CLAROSTAT line. And we'll gladly send you literature.

CLAROSTAT MFG. CO., Inc.
Specialists in Radio Aids
291 N. 6th St., Brooklyn, N. Y.

CLAROSTAT

**FAMOUS
NATIONAL**

Velvet B Eliminator \$16.13
180 Volts (280 Tube Free)



Latest Model National Velvet-B, Type 3580, in handsome crackle finish black metal casing, for use with sets up to and including six tubes. Input 105-120 volts AC, 50 to 60 cycles. Output, 180 volts maximum at 35 milliamperes. Three variable output intermediate voltages. (Det., RF, AF). Eliminator has excellent filter system to eliminate hum, including 30 henry choke and 18 mfd. Mershon condenser. No motorboating! (Eliminator Licensed under patents of the Radio Corporation of America and associated companies.)

Guaranty Radio Goods Co.
145 W. 45TH STREET
(Just East of Broadway)
NEW YORK CITY

GUARANTY RADIO GOODS CO.,
145 W. 45th St., N. Y. City

Please send C. O. D. on 5-day money-back guaranty, one National Velvet B (180 volts maximum with three variable intermediate voltages), 280 tube free, at \$16.13.
 Enclose find \$16.13. Please send above. You are to pay cartage.

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City..... State.....

5-DAY MONEY-BACK GUARANTY

LYNCH METALLIZED

Standohm—The "Self-supporting" Register. A big improvement in a well-known, standardized QUALITY resistance unit. Moderate in price. Efficient in operation.

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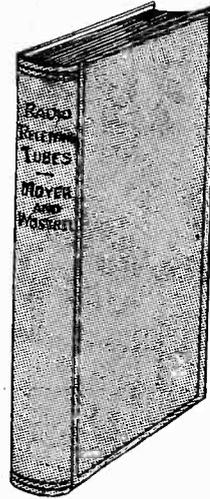
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**NEWEST BOOK
ON VACUUM TUBES**



"RADIO RECEIVING TUBES," by James A. Moyer and John F. Westrel, first edition just off the press. No radio service man, experimenter or student of radio should be without this authoritative book on the principles and applications of vacuum tubes. It answers all your questions relating to receiving, amplifying and rectifying tubes. It is a complete discussion of tube principles, functions and uses, thoroughly up-to-date.

In this book the essential principles underlying the operation of vacuum tubes are explained in as non-technical a manner as is consistent with accuracy. The book covers the construction, action, reactivation, testing and use of vacuum tubes as well as specifications for vacuum tubes and applications for distant control of industrial processes and precision measurements.

Price \$2.50

RADIO WORLD
145 West 45th Street, New York City
(Just East of Broadway)

Gothic Polo Speaker, \$15.00

Housed in a beautiful Gothic structure of genuine walnut; hand-rubbed to an attractive finish, the Polo driving mechanism and cone combine best quality reproduction with finest appearance. The grille is specially constructed for two-tone effect, so popular in walnut these days. The Polo Speaker in the Gothic housing is an adornment, besides being an outstanding speaker in performance. The design of the cabinet is exclusive. The height is 12 1/2". Shipping weight, 10 lbs. Cat. No. T.M.P.G.....



Guaranty Radio Goods Co.
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FREE!

Your Choice of Any One of These Nine Meters FREE!

- 0-6 Voltmeter D.C. No. 326
- 0-50 Voltmeter D.C. No. 337
- 6-Volt Charge Tester D.C. No. 23
- 0-10 Amperes D.C. No. 338
- 0-25 Milliamperes D.C. No. 325
- 0-50 Milliamperes D.C. No. 350
- 0-100 Milliamperes D.C. No. 390
- 0-300 Milliamperes D.C. No. 399
- 0-400 Milliamperes D.C. No. 394

**Equip Yourself Now
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RADIO WORLD
145 WEST 45TH ST., N. Y. CITY

Alphabetical List of Stations by Call Letters; Location and Frequency

[FROM LIST REVISED AND CORRECTED BY FEDERAL RADIO COMMISSION AND RELEASED ON MAY 28TH.]

Station	Location	Frequency	Station	Location	Frequency	Station	Location	Frequency	Station	Location	Frequency	
WAAF	Chicago, Ill.	920	WGBS	New York City, 1180	WMAF	Dartmouth, Mass., 1360	WTAR	WPOR—Norfolk, Va., 780	KGHL	Billings, Mont., 950		
WAAM	Newark, N. J., 1250	WGCM	Gulfport, Miss., 1210	WMAK	Buffalo, N. Y., 900	WTAW	College Station, 1120	KGHX	Richmond, Tex., 1500			
WAAT	Jersey City, N. J., 1070	WGCP	Newark, N. J., 1250	WMAL	Washington, D. C., 630	WTAX	Streator, Ill., 1210	KGJF	Twin Falls, Idaho, 1320			
WAAW	Omaha, Nebr., 660	WGES	Chicago, Ill., 1360	WMAN	Columbus, Ohio, 1210	WTBO	Cumberland, Md., 1420	KGIR	Butte, Mont., 1360			
WABC	WBOQ—N.Y. City, 860	WGHP	Newport News, Va., 1310	WMAQ	Chicago, Ill., 670	WTFI	Toccoa, Ga., 1450	KGIW	Trinidad, Colo., 1420			
WABI	Bangor, Me., 1200	WGLP	Detroit, Mich., 1240	WMAZ	Macon, Ga., 890	WTIC	Hartford, Ct., 600, 1060	KGIX	Las Vegas, Nev., 1420			
WABZ	New Orleans, La., 1200	WGLT	Ft. Wayne, Ind., 1370	WMBA	Newport, R. I., 1500	WTMJ	Milwaukee, Wis., 620	KGJF	Little Rock, Ark., 890			
WADC	Akron, O., 1320	WGMS	See WLB-WGMS	WMBC	Detroit, Mich., 1420	WWAE	Hammond, Ind., 1200	KGKB	Brownwood, Tex., 1500			
WAFD	Detroit, Mich., 1500	WGN	WLIB—Elgin, Ill., 720	WMBD	Peoria Hts., Ill., 1440	WWJ	Detroit, Mich., 920	KGKL	San Angelo, Tex., 1370			
WAGM	Royal Oak, Mich., 1310	WGR	Buffalo, N. Y., 550	WMBG	Richmond, Va., 1210	WWL	New Orleans, La., 850	KGKO	Wichita, Falls, Tex., 570			
WAIU	Columbus, O., 640	WGST	Atlanta, Ga., 890	WMBH	Joplin, Mo., 1420	WWNC	Asheville, N. C., 570	KGKX	San Point, Idaho, 1420			
WAPI	Birmingham, Ala., 1140	WGY	Schenectady, N. Y., 790	WMBI	Addison, Ill., 1080	WWRL	Woodside, N.Y., 1500	KGO	Oakland, Calif., 790			
WASH	Gd. Rapids, Mich., 1270	WHA	Madison, Wis., 940	WMBP	Pittsburgh, Pa., 1500	WWVA	Wheeling, W.Va., 1160	KGRC	San Antonio, Tex., 1370			
WBAK	Harrisburg, Pa., 1430	WHAD	Milwaukee, Wis., 1120	WMBL	Lakeland, Fla., 1310	KCRC	Enid, Okla., 1370	KGRS	Amarillo, Tex., 1410			
WBAL	Baltimore, Md., 1060	WHAM	Rochester, N. Y., 1150	WMBO	Auburn, N. Y., 1370	KDB	Santa Barbara, Cal., 1500	KGU	Honolulu, Hawaii, 940			
WBAP	Fort Worth, Tex., 800	WHAP	N. Y. City, 1300	WMBQ	Brooklyn, N. Y., 1500	KDKA	Pittsburgh, Pa., 980	KGW	Portland, Ore., 620			
WBAW	Nashville, Tenn., 1490	WHAS	Louisville, Ky., 820	WMBR	Tampa, Fla., 1210	KDLR	Devils Lake, N.D., 1210	KGW	Lacey, Wash., 1200			
WBAX	Wilkes-Barre, Pa., 1210	WHAZ	Troy, N. Y., 1300	WMBT	Memphis, Tenn., 570	KDYL	Salt Lake, Utah, 1290	KHI	Los Angeles, Calif., 900			
WBBC	Brooklyn, N. Y., 1400	WHBC	Canton, Ohio, 1200	WMC	New York, N.Y., 570	KEJK	Beverly Hills, Calif., 1170	KHO	Spokane, Wash., 590			
WBBL	Richton, Va., 1370	WHBD	Bellefontaine, O., 1370	WMES	Boston, Mass., 1500	KEK	Portland, Ore., 1180	KICK	Red Oak, Iowa, 1420			
WBMM	WJBT—Chicago, Ill., 770	WHBF	Rock Island, Ill., 1210	WMMN	Fairmont, W. Va., 890	KEX	Portland, Ore., 1180	KID	Idaho Falls, Idaho, 1320			
WBRR	Rossville, N. Y., 1300	WHBL	Sheboygan, Wis., 1410	WMPC	Lapeer, Mich., 1500	KFAB	Lincoln, Nebr., 770	KID	Boise, Idaho, 1250			
WBYY	Charleston, S. C., 1200	WHBP	Johnstown, Pa., 1310	WMRJ	Jamaica, N. Y., 1420	KFAD	Phoenix, Ariz., 620	KIT	Yakima, Wash., 1370			
WBBZ	Ponca City, Okla., 1200	WHBO	Memphis, Tenn., 1370	WMSG	New York, N. Y., 1350	KFBB	Great Falls, Mont., 1360	KJBS	San Francisco, Cal., 1070			
WBGM	Bay City, Mich., 1410	WHBU	Anderson, Ind., 1210	WMT	Waterloo, Iowa, 1200	KFBK	Sacramento, Calif., 1310	KJR	Seattle, Wash., 970			
WBIS	See WNAC	WHBW	Philadelphia, Pa., 1500	WMT	Waterloo, Iowa, 1200	KFB	Everett, Wash., 1370	KJL	Blytheville, Ark., 1290			
WBMS	Fort, Lee, N. J., 1450	WHBY	W. De Pere, Wis., 1200	WNAC	WBIS—Boston, 1230	KFDM	Beaumont, Tex., 560	KLO	Ogden, Utah, 1370			
WBNY	New York, N. Y., 1350	WHDF	Calumet, Mich., 1370	WNAD	Norman, Okla., 1010	KFDY	Brookings, S.Dak., 550	KLRA	Little Rock, Ark., 1390			
WBOO	See WABC	WHDH	Gloucester, Mass., 830	WNAX	Yankton, S. D., 570	KFEL	Denver, Colo., 940	KLS	Oakland, Calif., 1440			
WBOW	Terre Haute, Ind., 1310	WHDI	Minneapolis, Minn., 1180	WNBF	Binghamton, N.Y., 1500	WFGQ	St. Joseph, Mo., 560	KLX	Oakland, Calif., 880			
WBRC	Birmingham, Ala., 930	WHDL	Tupper Lake, N.Y., 1420	WNBH	N'w Bedford, Mass., 1310	WFOQ	Boone, Iowa, 1310	KLZ	Dupont, Colo., 560			
WBRE	Wilkes-Barre, Pa., 1310	WHDC	WABO—Rochester, N.Y., 1440	WNBI	Knoxville, Tenn., 1310	KFHA	Wichita, Kans., 1300	KMA	Shenandoah, Iowa, 930			
WBRL	Tilton, N. H., 1430	WHFC	Cicero, Ill., 1310	WNBO	Washington, Pa., 1200	KFLA	Gunnison, Colo., 1200	KMB	Independence, Mo., 950			
WBSE	Wellesley H., Mass., 780	WHIS	Bluefield, W. Va., 1420	WNBW	Memphis, Tenn., 1430	KFL	Los Angeles, Calif., 640	KMD	Medford, Ore., 1310			
WBT	Charlotte, N. C., 1080	WHK	Cleveland, Ohio, 1390	WNBW	Carbondale, Pa., 1200	KFLF	Portland, Ore., 1420	KMJ	Fresno, Calif., 1200			
WBZ	Springfield, Mass., 990	WHN	New York, N.Y., 1010	WNBZ	Springfield, Vt., 1200	KFIO	Spokane, Wash., 1230	KMMJ	Clay Center, Nebr., 740			
WBZA	Boston, Mass., 990	WHO	Des Moines, Ia., 1000	WNBZ	Saranac Lk., N.Y., 1290	KFIZ	Fond du Lac, Wis., 1420	KMO	Tacoma, Wash., 1340			
WBZC	Storrs, Conn., 600	WHP	Harrisburg, Pa., 1430	WNY	Newark, N. J., 1450	KFJB	Marshalltown, Ia., 1200	KMOX	KFOA—St. Louis, 1090			
WBZD	Canton, N. Y., 1220	WIAS	Ottumwa, Iowa, 1420	WNOX	Knoxville, Tenn., 560	KFJF	Oklahoma City, Okla., 1470	KMTR	Hollywood, Calif., 570			
WBZL	Pittsburgh, Pa., 1220	WIBA	Madison, Wis., 1210	WNRC	Greensboro, N.C., 1440	KFJJ	Astoria, Ore., 1370	KNX	Hollywood, Calif., 1050			
WBZM	Lincoln, Nebr., 590	WIBG	Elkins, Park, Pa., 930	WNYC	New York, N.Y., 570	KFJM	Gd. Forks, N.D., 1370	KOA	Denver, Colo., 830			
WBZP	Northfield, Minn., 1250	WIBM	Jackson, Mich., 1370	WOAI	San Antonio, Tex., 1190	KFJR	Portland, Ore., 1300	KOAC	Corvallis, Ore., 560			
WBZQ	Camden, N. J., 1280	WIBO	Chicago, Ill., 570	WOAN	Lawrenceburg, Tenn., 600	KFJY	Fort Dodge, Iowa, 1310	KOB	State College, N.M., 1180			
WBZR	Baltimore, Md., 600	WIBR	Steubenville, O., 1420	WOAX	Trenton, N. J., 1280	KFJZ	Fort Worth, Tex., 1370	KOCW	Chickasha, Okla., 1400			
WBZS	Asbury Pk., N. J., 1280	WIBS	Elizabeth, N. J., 1450	WOBT	Union City, Tenn., 1310	KFKA	Greeley, Colo., 880	KOH	Reno, Nev., 1370			
WBZT	Rapid City, S.D., 1200	WIBU	Poynette, Wis., 1310	WOB	Charlestown, W.Va., 580	KFKB	Milford, Kans., 1050	KOIL	Council Bluffs, Ia., 1260			
WBZU	Philadelphia, Pa., 1170	WIBW	Topeka, Kan., 1300	WOC	Davenport, Ia., 1000	KFKC	Lawrence, Kans., 1220	KOIN	Portland, Ore., 940			
WBZV	Carthage, Ill., 1070	WIBX	Utica, N. Y., 1200	WOC	Jamestown, N. Y., 1210	KFKX	KYW—See KYW	KOL	Seattle, Wash., 1270			
WBZC	Allentown, Pa., 1440	WICC	Bridgeport, Conn., 1190	WODA	Paterson, N. J., 1250	KFKZ	Kirkville, Mo., 1200	KOMO	Seattle, Wash., 920			
WBZD	Zion, Ill., 1080	WIL	St. Louis, Mo., 1200	WOF	Ames, Iowa, 560	KFLV	Rockford, Ill., 1410	KOOS	Marshallfield, Ore., 1370			
WBZM	Baltimore, Md., 1370	WILL	Urbana, Ill., 890	WOK	WMBB—See WMBB	KFLX	Galveston, Tex., 1370	KORE	Eugene, Ore., 1420			
WBZP	Springfield, Ill., 1210	WILM	Wilmington, Del., 1420	WOKO	Poughkeepsie, N.Y., 1440	KFMX	Northfield, Minn., 1250	KOY	Phoenix, Ariz., 1390			
WBZQ	Minneapolis, Minn., 810	WINR	Bay Shore, N. Y., 1210	WOL	Washington, D.C., 1310	KFNH	Shenandoah, Iowa, 890	KPCB	Seattle, Wash., 1210			
WBZL	Cliffside, Pk., N.J., 1350	WIOD	Miami Beach, Fla., 560	WOMT	Manitowoc, Wis., 1210	KFOR	Lincoln, Nebr., 1210	KPJM	Prescott, Ariz., 1500			
WBZM	Chicago, Ill., 970	WIP	Philadelphia, Pa., 610	WOOD	Gd. Rapids, Mich., 1270	KFOX	Long Beach, Calif., 1250	KPLA	Los Angeles, Calif., 1000			
WBZN	Coney Island, N.Y., 1400	WISN	Milwaukee, Wis., 1120	WOQ	Kansas City, Mo., 610	KFPL	Dublin, Tex., 1310	KPO	San Francisco, Cal., 680			
WBZY	Harrison, O., 1480	WIAD	Waco, Texas, 1240	WOR	Newark, N. J., 710	KFPM	Greenville, Tex., 1310	KPOF	Denver, Colo., 880			
WBZC	Lg. Beach, N.Y., 1500	WIAG	Norfolk, Nebr., 1060	WORD	Chicago, Ill., 1480	KFPY	Siloam Spgs., Ark., 1340	KPPC	Pasadena, Calif., 1200			
WBZD	Kenosha, Wis., 1200	WIJK	Kokomo, Ind., 1310	WOS	Jefferson City, Mo., 630	KFPY	Spokane, Wash., 1390	KPQ	Seattle, Wash., 1210			
WBZL	Joliet, Ill., 1310	WIAR	Providence, R.I., 890	WOW	New York, N.Y., 1130	KFOA	KMOX—See KMOX	KPRC	Houston, Tex., 920			
WBZM	Culver, Ind., 1400	WIAX	Jacksonville, Fla., 1260	WOW	Omaha, Nebr., 590	KFOD	Anchorage, Alaska, 1230	KPSN	Pasadena, Calif., 950			
WBZP	Pensacola, Fla., 1120	WIAY	Cleveland, Ohio, 620	WOWO	Ft. Wayne, Ind., 1160	KFOU	Holy City, Calif., 1420	KPWF	Westminster, Calif., 1490			
WBZQ	Columbus, Miss., 880	WIJZ	Chicago, Ill., 1480	WPAP	WQAO—See WQAO	KFOW	Seattle, Wash., 1420	KQV	Pittsburgh, Pa., 1380			
WBZR	Yonkers, N.Y., 1210	WIJZ	Chicago, Ill., 1480	WPAP	Pawtucket, R.I., 1210	KFQZ	Hollywood, Calif., 860	KQW	San Jose, Calif., 1010			
WBZS	Chicago, Ill., 1210	WIJC	La Salle, Ill., 1200	WPCC	Chicago, Ill., 570	KFR	San Francisco, Cal., 610	KRE	Berkeley, Calif., 1370			
WBZT	Portland, Maine, 940	WJBI	Red Bank, N. J., 1210	WPCH	New York, N.Y., 810	KFRU	Columbia, Mo., 630	KRGV	Harlingen, Tex., 1260			
WBZU	Springfield, O., 1380	WJBC	Ypsilanti, Mich., 1370	WPG	Atlantic City, N.J., 1100	KFSD	San Diego, Calif., 600	KRLD	Dallas, Tex., 1040			
WBZV	Tampa, Fla., 620	WJBL	Decatur, Ill., 1200	WPOE	Patchogue, N. Y., 1420	KFSG	Los Angeles, Calif., 1120	KRMD	Shreveport, La., 1310			
WBZW	Kansas City, Mo., 610	WJBO	New Orleans, La., 1370	WPOR	WTAR—See WTAR	KFUL	Galveston, Tex., 1290	KRSC	Seattle, Wash., 1120			
WBZX	Amarillo, Tex., 1410	WJBU	Lewisburg, Pa., 1210	WPRC	Harrisburg, Pa., 1200	KFUM	Colo. Springs, Colo., 1270	KSAC	Manhattan, Kans., 580			
WBZY	El Paso, Tex., 1310	WJBW	New Orleans, La., 1200	WPS	State College, Pa., 1230	KFUP	Clayton, Mo., 550	KSCJ	Sioux City, Ia., 1330			
WBZC	Fargo, N. D., 1280	WJBY	Gadsden, Ala., 1210	WPSW	Philadelphia, Pa., 1500	KFUP	Denver, Colo., 1310	KSD	St. Louis, Mo., 550			
WBZD	Roanoke, Va., 930	WJID	Woohearth, Ill., 1130	WPTF	Raleigh, N. C., 680	KFVD	Culver City, Calif., 710	KSEI	Pocatello, Idaho, 900			
WBZE	Orlando, Fla., 620	WJIS	Gary, Ind., 1360	WQAM	Miami, Fla., 1240	KFVS	Cape Girardeau, Mo., 1210	KSL	Salt Lake City, U., 1130			
WBZF	Wilmington, Ind., 1120	WJL	Detroit, Mich., 750	WQAO	WPAP—N.Y.C., 1010	KFWB	Hollywood, Calif., 950	KSMR	Santa Maria, Calif., 1200			
WBZG	Minneapolis, Minn., 1180	WJW	Mt. Vernon Hills, Va., 1460	WQBC	Utica, Miss., 1360	KFWC	S. Pomona, Calif., 1200	KSO	Clarinda, Iowa, 1380			
WBZH	Chattanooga, Tenn., 1280	WJZ	New York, N. Y., 760	WQZ	Weirton, W.V., 1420	KFWF	St. Louis, Mo., 1200	KSOO	Sioux Falls, S. D., 1110			
WBZL	New Haven, Conn., 1330	WKAO	San Juan, P. R., 890	WRAF	LaPorte, Ind., 1200	KFWI	San Francisco, Calif., 930	KSTP	St. Paul, Minn., 1460			
WBZM	New Orleans, La., 1270	WKAR	E. Lansing, Mich., 1040	WRAC	Eric, Pa., 1370	KFWM	Oakland, Calif., 930	KTAB	Oakland, Calif., 550			
WBZN	WLSI—C'nst'n, N.J., 1210	WKAW	Laconia, N. H., 1310	WRAP	Reading, Pa., 1310	KFXD	Jerome, Idaho, 1420	KTAP	San Antonio, Tex., 1420			
WBZP	Tuscola, Ill., 1070	WKBB	Joliet, Ill., 1310	WRAX	Philadelphia, Pa., 1020	KFXF	Denver, Colo., 940	KTAT	Ft. Worth, Tex., 1240			
WBZQ	New York, N.Y., 660	WKBC	Birmingham, Ala., 1310	WRBC	Valparaiso, Ind., 1240	KFXJ	Edgewater, Colo., 1310	KTBI	Los Angeles, Calif., 1300			
WBZR	Providence, R.I., 550	WKBE	Webster, Mass., 1200	WRBI	Tifton, Ga., 1310	KFXK	Oklahoma City, Okla., 1310	KTBR	Portland, Ore., 1300			
WBZS	Columbus, O., 550	WKBF	Indianapolis, Ind., 1400	WRBJ	Hattiesburg, Miss., 1500	KFXV	Flagstaff, Ariz., 1420	KTBS	Shreveport, La., 1450			

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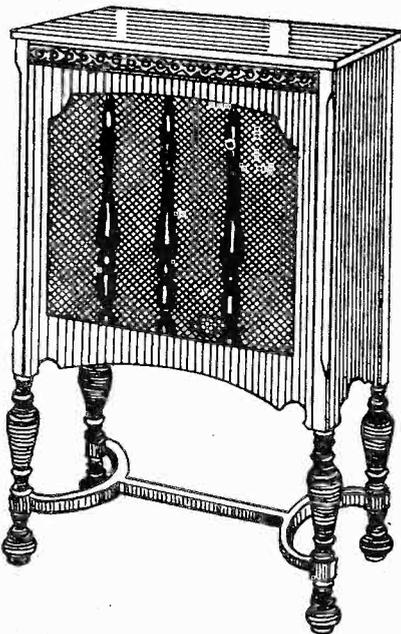
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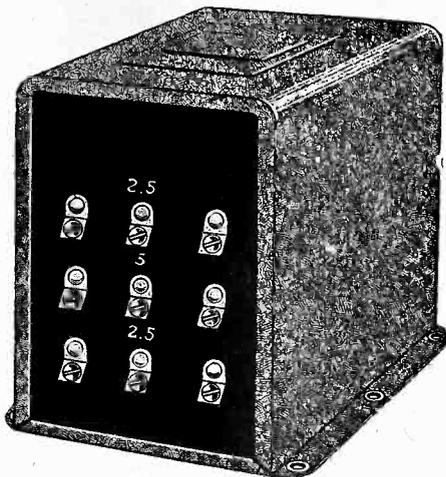
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All three windings are tapped at the exact
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The heater type tube is represented by the 227,
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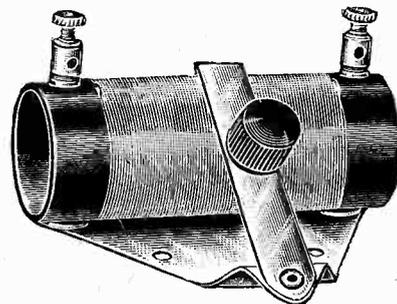
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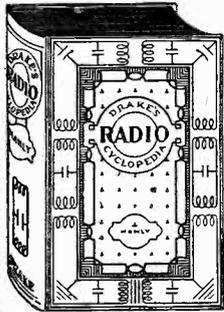
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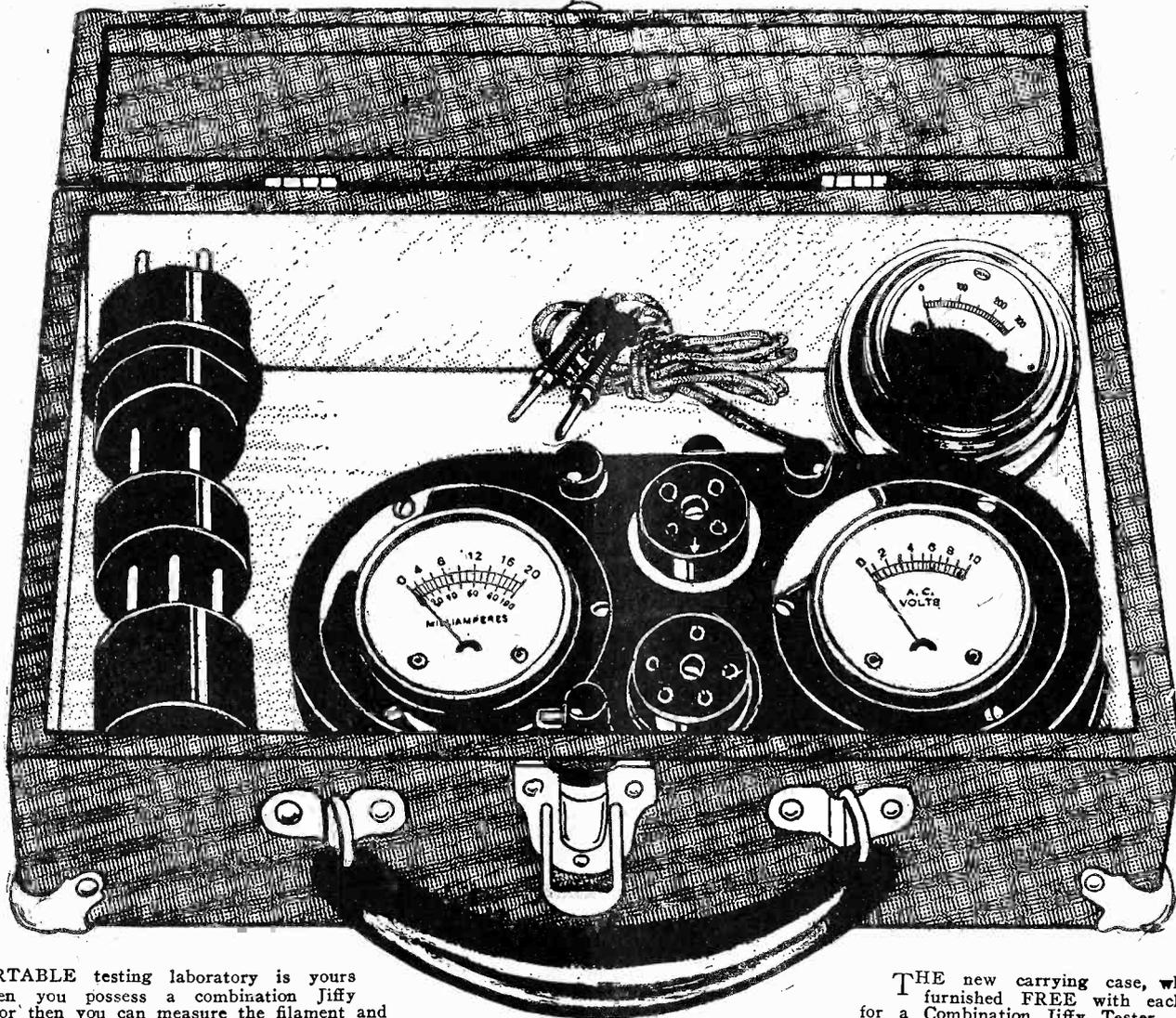
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THIS OFFER EXPIRES AT NOON JULY 15TH, 1929

New Style DeLuxe Leatherette Carrying Case FREE with each Jiffy Tester!

This combination of meters tests all standard tubes, including the new AC screen grid tubes and the new 245 tube, making thirteen tests in 4½ minutes! Instruction sheet gives these tests in detail.



A PORTABLE testing laboratory is yours when you possess a combination Jiffy Tester, for then you can measure the filament and plate voltages of all standard tubes, including AC tubes, and all standard battery-operated or AC screen grid tubes; also plate voltages up to 500 volts on a high resistance meter that is 99% accurate; also plate current.

The Jiffy Tester consists of a 0-20, 0-100 milliammeter, with change-over switch and a 0-10 volt AC and DC voltmeter (same meter reads both), with two sockets, one for 5-prong, the other for 4-prong tubes; a grid bias switch and two binding posts to which are attached the cords of the high resistance voltmeter; also built-in cable with 5-prong plug and 4-prong adapter, so that connections in a receiver are transferred to the Tester automatically. Not only can you test tubes, but also opens or shorts in a receiver, continuity, bias, oscillation, etc. The instruction sheet tells all about these tests.

In addition you can test screen grid tubes by connecting a special cable, with clip to control grid (cap of tube) and other end of special cable to the clip in the set that went to the cap before the tube was transferred to the tester.

THE new carrying case, which is furnished FREE with each order for a Combination Jiffy Tester, contains the entire outfit, including the three meters, cable and plug, and three adapters (one for 4-prong tubes, two for 199 tubes). This case is 10½ x 7¼ x 3½" and has nickel corner pieces and protective snap-lock. The case is made of strong wood, with black leatherette overlay.

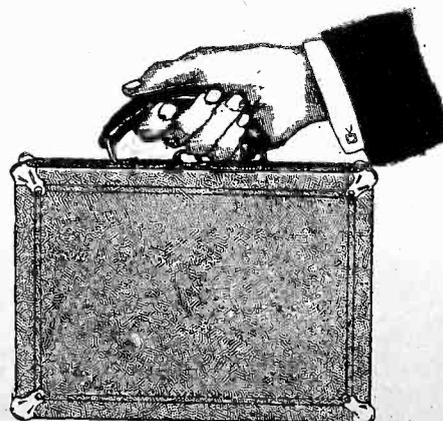
To operate, remove a tube from the receiver, place the cable plug in the vacant receiver socket, put the tube in the proper socket of the Tester, connect the high resistance meter to the two binding posts, and you're all set to make the thirteen vital tests in 4½ minutes!

The Combination Jiffy Tester is just the thing for service men, custom set builders, experimenters, students, teachers and factories. Order "Jiffy 500." The price is only \$14.50.

If a 0-600 AC and DC high resistance meter (99% accurate) is desired, so house electricity line voltage and power transformer voltages can be measured, as well as plate voltage, instead of the 0-500 DC voltmeter, order "Jiffy 600" at \$15.50.

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- (4) One 5-prong plug with 30' cord for AC detector tubes, etc., and one 4-prong adapter for other tubes.
- (5) One grid switch to change bias.
- (6) One 5-prong socket.
- (7) One 4-prong socket.
- (8) Two binding posts.
- If 0-300 DC high resistance 99% accurate voltmeter is preferred to 0-500, put check here. Price is same, \$14.50.
- Same as above, except substitute a 0-600-volt AC and DC high resistance 99% accurate voltmeter (same meter reads both) for the 0-500 DC meter. Price \$15.50.
- (9) One handsome moire metal case.
- (10) One instruction sheet.
- (11) One de luxe carrying case.
- (12) One screen grid special cable.



The new de luxe leatherette carrying case is compact and handy. Size 10½" long, 7¼" wide, 3½" deep.

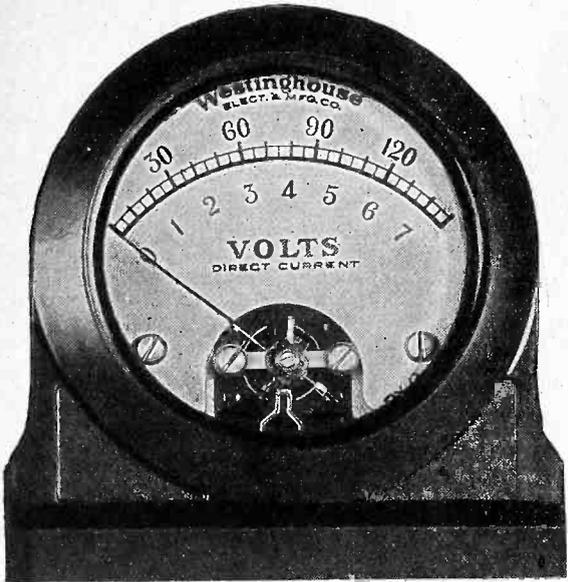
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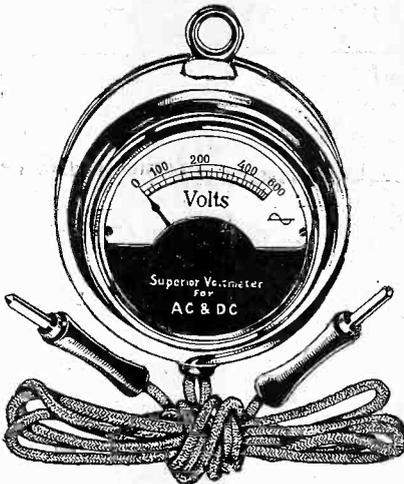
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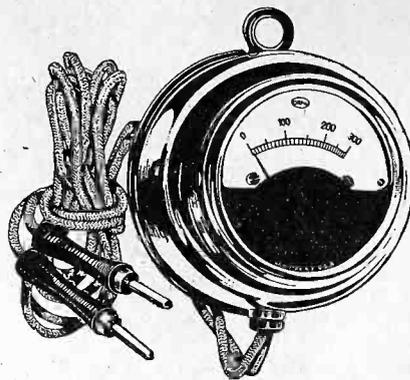


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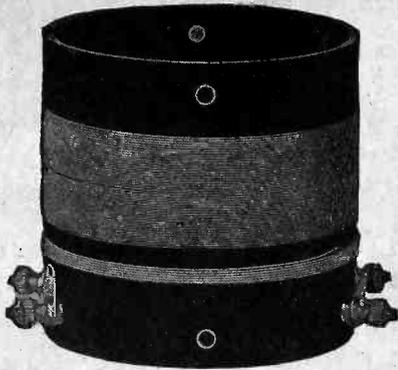
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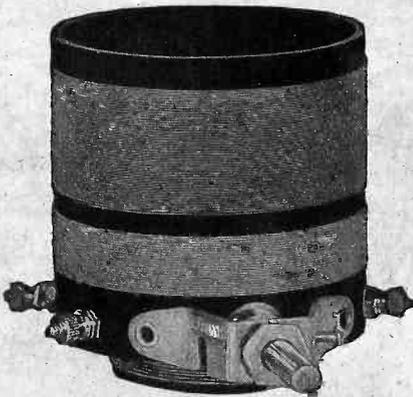
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Highly selective antenna coil for any circuit, and interstage coil for AC circuits. Step-up ratio, 1-to-8. Tunes with .0003 mfd. Model AC3, for .00035 mfd. \$1.75



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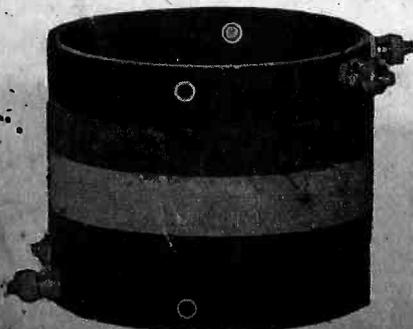
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TP5 \$3.00

Interstage coupler to work out of a screen grid tube, where the primary in the plate circuit is tuned, the secondary, in the next grid circuit, untuned. Tunes with .0005 mfd. Model TP3, for .00035 mfd. \$3.25

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Conductively coupled antenna coil, for maximum pickup, where selectivity is not the main consideration. Continuous winding in two colors. Tunes with .0005 mfd. Model A3, for .00035 mfd. \$2.00

Screen Grid Co., 145 W. 45th St., N. Y. City

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The best appearance of the New Diamond of the Air results from using the official aluminum subpanel, 10 x 20 inches, with the four sockets built in, and with self-bracketing front. Hardware and insulating washers supplied with each sub-panel. The aluminum sub-panel is exactly the same as the one used in the laboratory models of the battery operated and the AC Screen Grid Diamonds. Holes are drilled for mounting parts, but as this aluminum drills like bakelite you can drill any holes you want.

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Enclosed please find \$5.00 for both the aluminum subpanel, etc., and the drilled Bakelite front panel of the battery model.

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Write Dept. RW3. for useful data

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CA, CG, C7—Three Aerovox .0005 mfd. mica fixed condensers, moulded, @ .25	\$0.75
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C6, C9—Two Aerovox .02 mfd. mica fixed condensers, moulded, @ \$1.00	2.00
C10—One Aerovox 4 mfd. condenser	2.50
C1—One Hammarlund Equalizer, 70 mfd.	.40
C2, C5—One Hammarlund Midline double condenser, each section .0005 mfd. (MLD23)	0.00
R1, SW—One Electrad Royalty volume control (0-5,000 ohms) with 110-volt Hart & Hegeman AC switch built in	1.50
R2—One Electrad 900-ohm resistance strip (grid suppressor type)	.20
R3—One Lynch 50,000-ohm resistor, with clips	.20
R4—One Electrad 2,000-ohm resistance stub	.20
R5—One Electrad 2,000-ohm type B resistor (B20)	.20
T1, T2—Two National A100 audio frequency transformers @ \$5.70	11.40
T3—One filament transformer; one winding 2.5 volts at 9 amperes or more, one winding 2.5 volts at 3.5 amperes or more, one winding 5 volts at 2 amperes or more (merchandise by Guaranty Radio Goods Co.)	0.00
PL—One Yaxley pilot light bracket with green jewel and lamp	.55
Ant., ground, speaker —, speaker +, four blinding posts @ .10	.40
One roll Braidite wire	.30
One 7 x 21-inch front panel	2.35
One 10 x 20-inch official AC Diamond aluminum subpanel, self-bracketing, with three five-prong sockets and one four-prong socket built in	3.25
Front and subpanel together	5.25
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One dial	.94
Two knobs (one for tickler, the other for volume control) @ .20	.40
Note: The optional condenser, CX, is .006 mfd. @ .50	

PARTS FOR BATTERY MODEL DIAMOND

C1—Aerovox .0005 moulded fixed	.25
C2—Hammarlund .0005 mfd. Midline	3.30
C3, C4—Two Aerovox moulded .006 mfd. @ .50 ea.	1.00
C5—Hammarlund .0005 mfd. Midline	3.30
C6—Aerovox .00025 moulded fixed with clips	.30
C7—Aerovox .0005 moulded fixed	.25
A1—322 Amperite with mount	.85
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R1—50-ohm Frost rheostat	1.00
R2—5-meg Lynch metallized leak	.40
T1, T2—Two National A100 audios @ \$5.70 ea.	11.40
Ant., Gnd., Sp., Sp., + posts @ .10 ea.	.40
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10 x 20-inch official battery Diamond subpanel, self-bracketing, with four sockets affixed; subpanel hardware, insulated bushings, washers	3.00
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Four Kelly tubes: one 222, two 201A, one 112A or 171A	7.00

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NEW YORK CITY

Blueprint of the AC Diamond

BUILD this 4-tube receiver, using one 222 tube, two 227 and one 112A (or 171A), and enjoy tone quality, selectivity and ease of control. The official blueprint gives the picture diagram life size, both top and bottom views; also schematic diagram and list of parts. You can use your present B eliminator externally, but the filament transformer is a part of the circuit.

Enjoy the convenience of AC operation, and still have just as selective and sensitive receiver, by building the AC Diamond. If you have 110-volt, 50 to 60 cycle AC house current, then this is the circuit for you. Fine performance. No hum.

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 30c for the March 23d and 30th issues (1929) describing this circuit.
 \$3.00 for 6 months subscription for Radio World. Send blueprint and two AC Diamond issues FREE.

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Most Selective Diamond Follow Blueprint THIS IS THE BATTERY MODEL

Here is the circuit of circuits—the design that makes a neighboring cleared-channel, high-power broadcaster snap out of audibility at a slight turn of the dial.

No need to worry about the selectivity requirements imposed on receivers by the reallocation.

Volume "to fill the house"—even on distance. Tone quality excellent.

Get the official blueprint of the laboratory model of the new SG 4-tube Diamond, exactly as built by Herman Bernard, the designer.

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