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ECD’s standard Micromind µM-65 supplies 8K bytes of memory. Additional 32K byte expansion boards and a mapping option give Micromind expandable access to 64 Megabytes. Utilizing software-controlled I/O channels, Micromind’s advanced encoding techniques load data from ordinary tape recorders at 3200 bits per second.

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What d'ya mean, you can't fix CBs?

Introducing the Sylvania ECG CB10-4 service kit. 26 semiconductors that can turn CB problems into profits.

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Shakespeare's White Knight. The best antenna going. And coming.

Shakespeare comes on strong for the new 40 channel era. With high performance CB antennas that turn on the power on all 23 or 40 channel CB transceivers.

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Ride full tilt into the 40 channel era with the new White Knight CB Antenna. And take the Shakespeare performance route home.

The White Knight Antenna, Style 430/available in a variety of pre-assembled mounting styles. Complete with cables and connectors. Under $25.
ON THE COVER

40-channel CB went into effect January 1, 1977. Right now, all kinds of 40-channel gear is on sale. The 40-channel dial on our cover symbolizes this development. To bring you all the latest information we have produced the special 16-page section starting on page 39 of this issue.

THIS TEST SIGNAL was used to test a new concept in speaker design. Get the whole story. See story starting on page 64.

PITCH BENDER BOARD is important part of keyboard synthesizer. See page 58.

As a service to readers, Radio-Electronics publishes available plans or information relating to newsworthy products, techniques and scientific and technological developments. Because of possible variances in the quality and condition of materials and workmanship used by readers, Radio-Electronics disclaims any responsibility for the safe and proper functioning of reader-built projects based upon or from plans or information published in this magazine.
TED visits United States: The only videodisc system currently on the market anywhere—the TED system, developed by Telefunken of Germany and Decca of Great Britain—was shown recently for the first time in the United States in NTSC color format. The player and a small selection of discs have been on the market in Germany since March 1975 with less than spectacular public acceptance. TED discs measure 8½ inches (21 cm) in diameter, and each thin flexible disc has a playing time of 10 minutes. Recording is done on the hill-and-dale principle and a diamond-tipped "pressure stylus" reads the "bumps" as the disc revolves at 1800 RPM on a cushion of air.

In Germany, the player sells for $500-$600, with discs at $4-$10 each, but Telefunken officials maintain that in large-scale production players could retail as low as $200-$300. A prototype automatic changer that can accommodate up to 12 discs, with a four-second changing time, was also demonstrated. Picture quality and color produced by the player and changer were extremely good during the demonstration.

Despite the slow consumer acceptance in Germany—which Telefunken attributes to low program quality and quantity—the TED player will be manufactured for the NTSC color market by two Japanese firms, Sanyo and General. Two other Japanese companies will manufacture discs, with marketing scheduled to start in Japan in April, well ahead of the estimated December 1977 "test-marketing" date for the Philips-MCA-Magnavox and RCA LP videodisc systems in the U.S. The TED system could show up in this country before the year is over. Sanyo markets television sets here under its own brandname and is also a principal supplier to Sears, Roebuck and Co. General's sets are handled on a private-label basis, and under the Teknika brandname, principally by department stores.

Home VTR suit: A legal challenge to the very concept of home videocassette recording has been filed against Sony by two movie companies. Universal Pictures and Walt Disney Studios jointly sued Sony, its advertising agency, a group of Southern California Sony dealers and one owner of a Betamax home videocassette recorder, charging that Sony is encouraging copyright violations by making VTR's available to consumers to tape copyrighted movies shown over TV. The suit seeks injunctions against manufacture, sale and use of the machines for this purpose and seeks to have the court order all tapes of the studios' copyrighted shows destroyed.

Sony promptly denied that Betamax violates copyrights, calling it a "time-shift machine" that makes it possible for viewers to watch programs at more convenient times. Sony indicated it felt the movie companies were trying to invade consumers' homes and tell them how and when they could see programs broadcast on the publicly owned airwaves. The entire future of home videotape recording of broadcast programs could rest on the outcome of the case.

It's generally agreed that taping a copyrighted show from television and then exhibiting the tape for an admission charge, or selling the tape, would be illegal. But the concept of making tapes from the air for personal use has rarely been challenged in the past. In fact, a 1972 phonograph record anti-piracy law expressly exempts audio recordists who make tapes off the air for their own use from being penalized by law. Whether that precedent extends to video presumably will be decided by the courts.

TV makers' blues: Is the American television manufacturing industry destined to go the way of the domestic radio industry—in other words, out of existence? The substantial increase in total U.S. TV sales in 1976 over 1975 appears almost totally to reflect a substantial increase in imports. In the first nine months of 1976, imports represented 32.8% of the color TV supply, up from 17.1% in the same months of 1975, and 70.5% of black-and-white, as compared with 67.4% a year earlier. Imports' share of the color market in the third quarter rose to a record 43.9%.

One of the casualties of low sales of domestically produced TV sets was Westinghouse's color tube business. The company stopped production last December, leaving only four firms manufacturing color tubes for U.S. set manufacturers—G-E, GTE Sylvania, RCA and Zenith. Meanwhile, some U.S. TV set and component manufacturers, along with labor unions, are seeking government action to trim imports of Japanese color sets on the grounds of "unfair competition" and damage to the domestic industry.

The Elcaset arrives: Cassette convenience with open-reel sound comes at a relatively high price in the new Elcaset decks now being introduced. The Elcaset is about 2½ times as large as a conventional compact cassette and contains ½-inch tape that moves at a speed of 3⅞ IPS. The system is designed for audio automation—special cut-out holes in the case are used to automatically set bias and equalization circuits for any of three different types of tape (low-noise, chrome and Ferrichrome) and switch Dolby circuits in or out. Provision is also made for the use of a special control head in professional models for automatic cueing and program control. Frequency response is claimed to be equal to open-reel tapes.

Elcaset's origin differs from that of a conventional cassette in that the tape is pulled out of its case when inserted in the deck, and three heads can be used (record, playback and erase)—four, with the addition of the control head. The system was developed in Japan by Sony and Teac. Technics by Panasonic will introduce a single-motor Elcaset deck next June at $750, and a four-motor professional model at about $2,000. Superscope and Teac also have scheduled sale of Elcaset decks in the U.S. Sony, Toshiba, Aiwa and JVC have also shown models in Japan.
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Worldwide TV by phone possible with new slow-scan system

It is now possible for televised pictures to be transmitted from one or more CCTV cameras to anywhere in the world by telephone.

THE ROBOT 500 SCAN CONVERTER

The Phoneline Television System uses a scan converter that converts the broad-band 525-line image to a 128-line by 128-element image. The camera "frame-grabs" a complete image every eight seconds and transmits it to the monitor during the period between then and picking up the next image. At the monitor, the picture is displayed eight seconds, then replaced by the next one with a wiping effect. The new image begins at the top of the screen and goes to the bottom.

The manufacturer, Robot Research, San Diego, CA, reports that the system is based on their Robot Series 500 scan converter, which does the frame-grabbing and converting to a signal that can be transmitted at audio frequencies.

"You must raise your prices" RCA official tells technicians

"You should raise your prices," RCA vice president Arnold Valencia told the National Electronics Service Dealers Association at their 1976 convention in San Antonio, TX, "it is important," he said, "that you try to cut your costs by raising efficiency. But I realize that there is a limit to how much you can increase that efficiency. When you reach bottom costs, you have no choice. You have to raise your prices."

He continued: "But I warn you, RCA will challenge you if you raise your charges for doing warranty work on our products. You'll have to justify any request for a price increase by substantiating your costs. Therefore, you'd better understand those costs."

Unfortunately, not all manufacturers were in agreement as to the need for receiving at least cost for doing warranty work. Ray Yeranko, national service manager of Magnavox, said his firm was "a friend of the independent service industry." Yet just the day before, the convention had heard the text of a 7-page telegram that Yeranko had used to defeat a California Senate bill prohibiting below-cost warranty work. Magnavox's reasoning, as expressed in the telegram, was that "Service agencies accept warranty business 'below-cost' only because they can obtain out-of-warranty business as a result of prior contracts." (Apparently the out-of-warranty customers are to be expected to pay part of the manufacturer's cost for in-warranty servicing.) The telegram went on to state that higher prices "could result in a manufacturer's decision to discontinue the use of independent agencies, and to perform the service himself."

The convention was also informed that a General Electric lobbyist, Robert Jordan, had appeared personally before the Finance, Insurance and Commerce committee of California's lower legislative house to defeat the measure.

FCC seizes record quantity of illegal CB radio equipment

More than $65,000 worth of illegally used radio equipment was confiscated last October in a crackdown on illegal radio transmission in Baltimore and the surrounding five counties, according to Jervis S. Finney, US Attorney for Maryland. This was the largest simultaneous execution of search and seizure warrants against illegal radio operators in this country, he said.

The illegal radio stations were reported to have been transmitting on unauthorized frequencies, operating overpowered equipment, invading and overpowered CB channels, and disrupting the communications of legitimate CB users.

No less than 19 search warrants were executed simultaneously by United States Marshals from Baltimore and Washington, accompanied by FCC agents from the Baltimore, Washington, Philadelphia and Norfolk field offices. Operators of the stations, if convicted, face possible fines of up to $500 per day of illegal operation, and a maximum of one year of imprisonment and a $10,000 fine for unlicensed use of CB or amateur equipment.

British cable TV programs transmitted over optical line

Cable TV subscribers in the Hastings area of England are now watching TV programs that have made part of their journey to the viewer through a glass cable instead of a metal conductor. Redifusion Ltd. of London reports that some 34,000 subscribers are now receiving their programs over a line that consists in part of a 1.4-kilometer (about 0.9-mile) length of optical cable.

Electrical signals are converted into light by a Plessey gallium-arsenide light-emitting diode, and back to electrical signals at the other end of the optical line with a photodiode. The cable consists of two strands of extremely pure optical glass, made by Corning in the United States. They are encased in a 7 x 4-millimeter polyethylene sheath with two 1-mm steel wires for mechanical strength. The optical fibers are carried loosely in a rectangular cavity.

Rediffusion reports that information-carrying capacity, compactness and lightness, and eventually lower cost are the three factors leading the company to experiment with optical cable, which can carry 10,000 times as much information per cable as coax.

Greager, Holman and Wiles win Gernsback Memorial Scholarship Awards

This month's winners of the Hugo Gernsback Memorial Scholarship Award, a check for $150 presented annually to an outstanding student in each of eight leading electronics home-study schools, is Alan B. Greager of Nucla, CO.

Mr. Greager completed studies in a technical trades institute, then enrolled with the International Correspondence Schools in their FCC Radiotelephone License program. He states that one of his reasons for taking the course was that he needed additional knowledge about circuits on which he was working.

"From the first lesson," he says, "I have been able to incorporate the knowledge presented by ICS with my job skills. I found that ICS would give me credit for my previous schooling, and offered a toll-free number on which any student can have a direct conversation with instructors. I was also allowed to add to my course lessons that were not required but which I desired.

continued on page 12
now 3-strong

Xcelite® family of attaché tool cases

And here's the newest addition... Model TC-150/ST... containing an intermediate assortment of tools for the technician, serviceman, or field engineer. It contains 52 items in all, including 24 famous Xcelite “Series 99” interchangeable-blade tools, a broad variety of other Xcelite Professional screwdrivers, nutdrivers, pliers, cutters, strippers, measuring tapes, and specialized electronic tools, plus the Weller® Cordless Soldering Iron and charger, an added convenience where outlets aren’t accessible. Tools are mounted in see-thru pockets on removable pallets in a durable, attractive case with Whiskey-tan Marvelon exterior and sun-tan vinyl lining. Plenty of extra space for additional tools, prints and manuals! Solid brass hardware and padded handle are additional quality touches.

It joins the other members of the family...
Model TC-100/ST, the “big daddy” of Xcelite’s cased tool sets, with the greatest variety—a total of 86 types and sizes of drivers, wrenches, pliers, cutters, strippers... and Model TC-200/ST, the 37-piece set that’s unequaled in economy and value.

See the new TC-150/ST at your distributor now. And ask for Xcelite literature, which will give you a detailed listing of the contents of all three Xcelite Attaché Tool Cases.

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Only NRI
25" designed-
DIAGONAL
Color TV
Quadraphonic

Two home training schools give you a hobby-kit Color TV to assemble. Two others give you a commercial set right off the shelf. Neither was designed to teach you how to repair Color TV's. Only NRI invested the time and money to design equipment with learning in mind!

No other home training school gives you both a solid state Color TV and SQ Quadraphonic Receiver complete with four speakers... all in one course. In fact, to even match this kind of thorough training at another school, you'd have to take an extra course costing hundreds of dollars more. And only NRI courses in Color TV/Audio servicing let you learn on equipment designed specifically for training.

It's the only way you can (1) get the feel of typical commercial circuitry, (2) learn bench techniques while building complete units from the "ground" up, (3) perform over 35 "in-set" experiments during construction, and (4) end up with a 25" diagonal solid state Color TV with cabinet and a 4-channel Audio Center.

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NRI engineering eliminates the cost of buying from an outside source. We pay no salesman's commission. Students are enrolled by mail only. The savings are passed on to you in the form of low tuition fees, extras like the TV's console cabinet and the Quadraphonic System; professional test instruments like a 5" triggered sweep oscilloscope, CMOS digital frequency counter, and integrated circuit Color TV pattern generator. You can pay hundreds of dollars more for similar courses and not
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for-learning
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FEBRUARY 1977
Second most outstanding student in the ICS screening is David Holman of Dickson, TN. He writes in part:

"When I started my course in Electronics I was doing manual work in a small repair shop. As I progressed in my studies the job got better and better, and in a few months I was running all the service calls. The shop was then moved to a new location, with me running it. Since then I have left that job and started 'Holman's TV Repair.' In seven months it has grown from a small repair shop to a very nice sales and service center. Without the superior education in electronics I gained from ICS, I would probably still be helping to set TV's up on the bench for someone else."

Mr. Holman receives a model 280 digital multimeter donated for the purpose by B & K.

In third place this month is Terry W. Wiles of Plattsmouth, NE, a graduate of the ICS TV Service Technician program, who receives a WV-529A special service VOM through the generosity of VIZ Manufacturing Co. He writes:

"I started my ICS TV course in late 1973 and completed it in March, 1975. Since then, I have opened a TV-radio service shop and am doing service work in eastern Nebraska and southwestern Iowa. I'm also working with a store in Melvern, IA. I don't believe things would have gone so smoothly if it had not been for the fine training and knowledge I gained through my ICS course. . . ."

OOOOOPS! We blew it! It seems we made an error in reporting the Hugo Gernsback Memorial Scholarship Award winners in the December, 1976, issue. The photographs for the second place winner, Mr Joseph E. Homay, and the third place winner, Mr. Robert Graham, were inadvertently mixed up so that the names appeared under the wrong photographs. To set things straight, we would like to apologize to Mr. Homay and Mr. Graham for the mix up. The photographs, with the correct names, are shown below.

DAVID HOLMAN
from ICS, I would probably still be helping to set TV's up on the bench for someone else."

JOSEPH E. HOMAY
The first place winner Mr. A. H. Christ, was correct as it appeared.

ROBERT GRAHAM

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TERRY W. WILES
radio electronics
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PERIMENTOR 600’s 6/10" center is ideal for microprocessor’s, clock chips, RAM’s, ROM’s, PROM’s, etc. While EXPERIMENTOR 300’s smaller 3/10" center is perfect for smaller DIP’s. Both units, of course, accept transistors, LED’s, resistors, capacitors, pot’s—virtually all types of components with plug-in ease. As well as #22-30 solid hook-up wire for interconnections. Eliminating heat and lead damage to expensive components. And saving you more money, on parts.

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DIGITAL SINEWAVES

Don Lancaster's article entitled "Create Sinewaves Using Digital ICs" that appeared in the November, 1976, issue was both interesting and useful. Using a 4018 IC, we built the five-stage synthesizer shown in Fig. 6 of the article. With 1% summing resistors, the output waveform shown in Fig. 1 was obtained. The scope was set to 1 ms-per-division and 0.5 volts-per-division. The harmonic content of the output signal was analyzed and is listed in Table 1.

TABLE 1—HARMONIC CONTENT of five-stage synthesizer.

<table>
<thead>
<tr>
<th>Harmonic</th>
<th>Frequency (Hz)</th>
<th>Amplitude (dB)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>200</td>
<td>0</td>
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<tr>
<td>3</td>
<td>600</td>
<td>-17</td>
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<td>5</td>
<td>1000</td>
<td>&lt; -60</td>
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<td>7</td>
<td>1400</td>
<td>-14</td>
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<td>&lt; -60</td>
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</tbody>
</table>

In the waveform of Fig. 1, the distance between the bottom level and the first step is larger than the distance between the first step and the second step. A better sinewave approximation would have a larger distance between the 1st and 2nd steps, than between the bottom level and the first step. If the 22.1K and 35.7K resistors are switched, the output waveform shown in Fig. 2 is obtained. The scope settings are the same as before. The harmonic content of this output signal was analyzed and is listed in Table 2.

TABLE 2—HARMONIC CONTENT of modified synthesizer.

<table>
<thead>
<tr>
<th>Harmonic</th>
<th>Frequency (Hz)</th>
<th>Amplitude (dB)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>200</td>
<td>0</td>
</tr>
<tr>
<td>3</td>
<td>600</td>
<td>-57</td>
</tr>
<tr>
<td>5</td>
<td>1000</td>
<td>&lt; -60</td>
</tr>
<tr>
<td>7</td>
<td>1400</td>
<td>-54</td>
</tr>
<tr>
<td>9</td>
<td>1800</td>
<td>-19</td>
</tr>
<tr>
<td>11</td>
<td>2200</td>
<td>-20</td>
</tr>
<tr>
<td>13</td>
<td>2600</td>
<td>-60</td>
</tr>
<tr>
<td>15</td>
<td>3000</td>
<td>&lt; -60</td>
</tr>
<tr>
<td>17</td>
<td>3400</td>
<td>-25</td>
</tr>
<tr>
<td>19</td>
<td>3800</td>
<td>&lt; -60</td>
</tr>
<tr>
<td>21</td>
<td>4200</td>
<td>&lt; -60</td>
</tr>
<tr>
<td>23</td>
<td>4600</td>
<td>&lt; -60</td>
</tr>
<tr>
<td>25</td>
<td>5000</td>
<td>&lt; -60</td>
</tr>
</tbody>
</table>

The modification produces a better approximation that follows the harmonic content specified in the article. The resistor values obtained in Table 1 of the article are also incorrect. The proper values can be determined from the reciprocal of the numbers shown in the parenthesis in Table 1 of the article. For a six stage counter, the resistors should be: 44.2K (1.000), 25.5K (.577), 22.1K (.500), 25.5K (5.77), and 44.2K (1.000).

Thanks again for a useful and interesting article, keep up the good work.

JOHN PEASe

GIL JOHNSON

Marquette Electronics

Milwaukee, WI

OPEN LETTER—AM STEREO

Secretary
Federal Communications Commission
Washington, DC

Dear Sir:

The Federal Communications Commission has announced tender of a petition for rule-making (RM-2717) submitted by Kahn Communications, Inc., of Freeport, New York, I, as an individual, a taxpayer and broadcaster, wish to endorse issuance of a notice of proposed rule-making in this matter.

There are several reasons for this endorsement of the Kahn System:

1) Since I am Program Director at WFBK, I have worked with the Kahn Stereo equipment and I know it works.

continued on page 16
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from mono to stereo and back again (while listening in mono with the tuner in the center position) to see if there was any interference or effect on the regular single radio system. I found there was none. 

5) I checked to see if there was any interference with any near-by station on the dial or any splatter. There was none. 

6) All of the tests that I conducted proved to me that the system is compatible with current broadcast requirements. 

7) In my own mind, I question FCC authority in this matter. A broadcaster is authorized to broadcast material over a commercial frequency. He must meet certain specifications and broadcast with-

in the guidelines set by the Rules and Regulations of the FCC. Since the Kahn System has no effect on these specifications and does not change them in any manner, I see no reason for FCC involve-

ment, thus I see no reason to deny this petition. 

8) Tuning to the sidebands is no problem and takes no special education. It is as simple to do as changing from one radio station to another. 

9) An adaptor could easily be made for stereo tuning. However, the beauty of the Kahn System is that there is no need for the radio listener to run out and buy a new radio. If he wishes to continue listening to a radio station in the mono mode, he may continue to do so. If he has two radios and desires to listen in stereo, he may do so. If he prefers to buy a new receiver with the stereo tuner, and the manufacturers decide to build it, he may do so.

With all of these plus factors, it is hard for me to see why the FCC or any other board would even hesitate to approve use of the Kahn System. It only requires one to snap in the sideband plug for the broadcaster. It is easily available to present equipment. It opens the vista of stereo broadcasting and listening to people who can't afford the luxury of buying new equipment. 

In my opinion it is the duty of the Commission to afford AM broadcasters the right to equally compete with their FM counterparts.

Kahn Communications has followed the proper procedure by placing its request with the FCC rather than an ad-hoc committee.

Delayed action smacks of politics and corporate jealousy. It also deprives the public of present United States technology. It discourages the private individual from making contributions to our society. 

Restrictive and delayed action is never positive action.

NO SYSTEM should be considered unless it is 100 percent compatible with today's AM broadcasting codes.

This letter is being submitted to the Commission on behalf of a radio amateur and citizen as a citizen and taxpayer. I have no financial interests in Kahn Communications, Inc., WFRB, or any manufacturing, broadcasting or media-related industries.

Respectfully submitted,

NORMAN H. BROOKS
Brookandville, MD

HELP!

I am an ardent but occasional reader of your fine magazine. However, it is not regularly available in this country at local stores. Hence I am requesting the help of some of your readers who would not mind to do a good turn to a not so well placed member of their technical fraternity. I would like someone to send me copies of Radio-Electronics magazine after they have finished using them.

I would also like to correspond to some of your readers who are interested in the exchange of small electronic components not readily available in this country, for Indian curios.

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See page 36

NEW—Universal Color TV Game
See pages 3, 65

NEW—“Touch Control” Light Switch
See pages 77, 96

NEW—Two-Way Freezer Alarm
See pages 77, 96

Digital Clocks and Weather Monitors — See pages 2, 8, 9

Amateur Radio Equipment — See pages 54-75

Automotive Accessories — See pages 15-23

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JOHN TITUS, DAVID LARSEN, and PETER RONY*

ONE OF THE MOST IMPORTANT PROGRAMMING characteristics possessed by any digital computer, including a microcomputer, is the ability to make a decision. For a typical microcomputer, we can define a decision as the process of determining further action based upon the logic state of a flag. A flag is a single flip-flop that can be either set or cleared in response to operations occurring within the microcomputer system. A change of state of the flag is usually an indication that either a particular operation has been completed or that a certain condition exists as a result of a microcomputer operation. Flags can be located either internal or external to the microprocessor chip; the ones that we shall discuss in this column are the internal flags that are set or cleared in response to specific types of microprocessor instructions such as arithmetic and logical instructions.

The flags that are located within the microprocessor chip are typically associated with the arithmetic-logic unit (ALU), a region within the chip where all arithmetic and logical operations are performed. In the 8080 microprocessor chip, for example, there exists five flags that indicate the following conditions:

- **zero flag**—If the result of an arithmetic or logical operation is zero, the zero flag is set to logic 1. If non-zero, the zero flag is reset to logic 0.
- **sign flag**—If the result of an arithmetic or logical operation is negative, the sign flag is set to logic 1. If positive, the sign flag is reset to logic 0.
- **parity flag**—If the result of an arithmetic or logical operation has even parity, the parity flag is set to logic 1. If odd parity, the parity flag is reset to logic 0.
- **carry flag**—If the result of an arithmetic or rotate operation has a carry of the most significant bit of the 8-bit result, the carry flag is set to logic 1. If not, the carry flag is reset to logic 0. The carry flag is reset to logic 0 after all logical operations.
- **auxiliary carry flag**—If the result of an arithmetic operation has a carry of bit-3 into bit-4 of the 8-bit result, the auxiliary carry flag is set to logic 1. If not, the auxiliary carry flag is reset to logic 0. The auxiliary carry flag is reset to logic 0 after most logical operations.

Since insufficient space is available in this column to discuss all of the above flags, we shall restrict our attention to the zero flag.

PREVIOUS INSTRUCTION

```plaintext
FLAG = 0
```

**FIG. 1**

Figure 1 shows the traditional flowchart decision symbol. The next instruction that is executed depends upon the logic state of the flag that is associated with this specific decision. For example, consider the JNZ instruction, where JNZ means "Jump If Not Zero:"

```
Instruction code Memonic Description
302 JNZ If the zero flag is at logic 0, jump to the 16-bit memory address given in bytes <B2> and <B3> of this three-byte instruction. If the zero flag is at logic 1, ignore this instruction and proceed to the following instruction.
```

The statement, "Jump If Not Zero," refers to the 8-bit result of a preceding instruction, not the logic state of the zero flag. When this result is zero, the zero flag is set and program control passes to the next instruction.

The JNZ instruction is widely used in the creation of programmed time-delay loops, an continued on page 24
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example of which is provided in Table 1. In this program, both the address and instruction bytes are in octal code: it is assumed that the memory address byte is 000. The program first moves an 8-bit timing byte into register B. This byte, indicated by an asterisk (*) has any value between 000 and 377. The value of the byte will determine the duration of the time delay.

At lo memory address 002, a device-select pulse is generated to set the SN7474 flip-flop shown in Fig. 3. The contents of register B are then decreased by 1. The JNZ instruction immediately tests the logic state of the zero flag. If the contents of register B are not zero, the flag is at logic 0 and a jump occurs back to lo memory address 004. The DCR B and JNZ instructions are executed repeatedly until the contents of register B becomes zero, at which time the zero flag becomes logic 1. The JNZ instruction tests the flag for the last time and shifts program control to the OUT 3 instruction at lo memory address 010. This output instruction generates a device-select pulse that clears the SN7474 flip-flop. Once this has been done, the microcomputer comes to a halt.

The program shown in Table 1 generates a single output-pulse the duration of which can take any value between 0.0125 and 1.925 ms in steps of 0.0075 ms. The calculations associated with the conversion of clock cycles to pulse widths are discussed in the book Microcomputer Interfacing Experiments Using the Mark 80 Microcomputer: an 8080

<table>
<thead>
<tr>
<th>LO memory address</th>
<th>Instruction byte</th>
<th>Mnemonic</th>
<th>Clock cycles</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>000</td>
<td>006</td>
<td>MVI B</td>
<td>7</td>
<td>Move following timing byte into register B</td>
</tr>
<tr>
<td>001</td>
<td>*</td>
<td>—</td>
<td>—</td>
<td>Timing byte for register B</td>
</tr>
<tr>
<td>002</td>
<td>323</td>
<td>OUT 2</td>
<td>10</td>
<td>Generate device select pulse that sets the SN7474 flip-flop</td>
</tr>
<tr>
<td>003</td>
<td>002</td>
<td>—</td>
<td>—</td>
<td>Device code for set input to SN7474 flip-flop</td>
</tr>
<tr>
<td>004</td>
<td>005</td>
<td>DCR B</td>
<td>5</td>
<td>Decrement contents of register B by 1</td>
</tr>
<tr>
<td>005</td>
<td>302</td>
<td>JNZ</td>
<td>10</td>
<td>If zero flag is at logic 0, jump to the memory address given by the following two address bytes; otherwise, ignore this instruction</td>
</tr>
<tr>
<td>006</td>
<td>004</td>
<td>—</td>
<td>—</td>
<td>LO memory address byte</td>
</tr>
<tr>
<td>007</td>
<td>000</td>
<td>—</td>
<td>—</td>
<td>Hi memory address byte</td>
</tr>
<tr>
<td>010</td>
<td>323</td>
<td>OUT 3</td>
<td>10</td>
<td>Generate device select pulse that clears the SN7474 flip-flop</td>
</tr>
<tr>
<td>011</td>
<td>003</td>
<td>—</td>
<td>—</td>
<td>Device code for clear input to SN7474 flip-flop</td>
</tr>
<tr>
<td>012</td>
<td>166</td>
<td>HLT</td>
<td>7</td>
<td>Halt the microcomputer</td>
</tr>
</tbody>
</table>

The number of clock cycles is a measure of the actual time that it takes the microcomputer to execute a single instruction or group of instructions. For a 2 MHz microcomputer, a single clock cycle has a duration of 500 ns.

The program in Table 1 and associated SN7474 flip-flop provide an example of what we mean by "the substitution of hardware by software." In a simple program and a single flip-flop replace a much more complicated hardwired programmable monostable circuit.

**Table 1—Simple Time-Delay Loop**

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The new GR-2001 TV system is the most sophisticated, best performing color TV kit we've ever designed, but it is also much more. It is the heart of a total home viewing system—a Computer TV!

Complete Programming Capability
With its optional Programmer, this Computer TV system allows you to program your set for an entire evening's viewing. The top bank of 8 keys [1] accesses the static NMOS RAM and turns the on-screen clock display into a computer CRT read-out which allows you to see your "program" as you enter it through the bank of 12 keys below the programmer panel. The selected time appears in the first four digits of the clock display, the channel number appears in the last two. First, enter the time at which you want the set to change channels. Next, enter the channel number you want. Then the memory takes over. While you sit back and relax, the Programmer automatically changes to the right channel at the right time. You can program up to 32 channel changes within two 12/24-hour periods!

Those two programming periods add extra versatility. Program the first for your daytime viewing schedule, the second for evening shows. Or, program the first for week nights, the second for weekends. You can even preselect the programs young children can watch—once the programmer is engaged, the manual keyboard is disconnected and can only be reactivated by the remote control or by pressing the correct button on the programming panel. You can even program the set to return to manual operation at a preselected time, then resume automatic operation at another time. When the last program you want to see is over, the set can be programmed to switch to an empty channel. This will cause the screen to go blank and the on-screen readout to flash on and off indicating that it is time to turn the system off with the front panel pushbutton or optional remote control.

Convenient Remote Control
The optional wireless remote control [2] lets you adjust volume, turn the set on or off, adjust tint, activate the digital readout, scan up or down through the pre-selected channels, and turn the optional programmer on and off—all at the touch of a button. This wireless remote control has improved circuitry for greater range and reliability and is the best we've ever offered.

Random Access Tuning
The 3 x 4 keyboard [3] lets you instantly choose any of up to 16 preselected stations—up to 24 with the optional eighth channel accessory. Switch from VHF to UHF, up or down, in any sequence, and be tuned in instantly without switching through empty channels. Up and down buttons on the keyboard also let you scan all the preselected stations.

Automatic Antenna Rotor Control
A Heathkit exclusive! With the optional antenna rotor control [4], you can program the GR-2001 to automatically rotate your outdoor antenna system as it changes from one channel to another, for optimum reception on every channel. No special knobs to turn, no buttons to push. You can select up to eight separate antenna headings with up to three stations per heading. It's perfect for areas where stations are in widely separated locations.

Superb Color and Sound
The TV set itself contains dozens of circuit refinements and improvements designed to give you the best picture and performance you've ever seen. The Automatic Gain Control circuit, for example, has been significantly improved to better resist airplane flutter. And since you build it yourself, you can be assured of a set that is free of mass production "glitches" that show up all too often in other sets now on the market.

Separate Audio IF Stage
The audio circuitry is probably the finest on any commercial set in the world. The sound signal has its own separate IF stage [5] that contributes to real quality audio. Audio and video muting provide smooth, silent between-channel tuning, too. You can hear the difference—especially if you use the output jack to connect the GR-2001 to your stereo system. The built-in wide-range speaker offers excellent fidelity as well. It's one of the first sets ever to give you real hi-fi sound from a TV!

Phase-Locked-Loop Horizontal and Vertical Hold Circuits
New phase-locked-loop horizontal and vertical oscillators [6] "lock-in" on any channel for a picture that's rock-steady and stable. There are no conventional vertical and horizontal hold controls because you never need them! There are no alignment problems either, so you get consistently excellent pictures year after year.

Black-Matrix Picture Tube
The GR-2001's 25" (diagonal) ultra-rectangular picture tube [7] provides one of the brightest, sharpest pictures in the world. The tube is fully shielded to maintain outstanding color purity by eliminating stray magnetic fields.

Easy To Assemble
Though the GR-2001 is one of our more complex kits, the average person shouldn't have any difficulty in assembling it. A step-by-step illustrated manual will lead you through assembly right up to trouble-shooting and testing. A test meter (included) lets you check assembly as you go, and the built-in dot generator aids in setup and service.

GR-2001 Specifications
Antenna Input Impedance: VHF: 300Ω balanced or 75Ω unbalanced. UHF: 300Ω balanced.
Hi-Fi Output: Frequency Response: ±1 dB, 50 Hz to 15 kHz.
Output Voltage: Greater than 1.0 V RMS.
Audio Output: 40 or 8Ω, 2 Watts.
Power Requirement: 110 to 130 Volts AC, 60 Hz, 200 Watts.
Dimensions: 29½" W x 20½" H x 21½" D. GR-2001 TV kit alone (chassis, picture tube and one speaker): 69.95
Heath Company, Dept. 20-261 Benton Harbor, Michigan 49022

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Also pictured above are other units — 5" solid state oscilloscope, vector monitor scope, solid-state stereo AM-FM receiver with twin speakers, digital multimeter, and more. It's the kind of better equipment that gets you better equipped for the electronics industry.

This electronic gear is not only designed for training; it's field-type — like you'll meet on the job, or when you're making service calls. And with NTS easy-to-read, profusely illustrated lessons you learn the theory behind these tools of the trade.

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Get facts on new 2-year extension
B & K Model 2040 CB Signal Generator

For a modest fee, any service shop can now provide a complete laboratory quality check-out of any modern transceiver.

The equipment that does all this is B & K's model 2040 CB Signal Generator that is specifically designed for the test and adjustment of Class-D AM and SSB transceivers. Just looking at the front panel is almost self-explanatory as to features and service conveniences.

In the center of the front panel is the CHANNEL selector for the digitally synthesized oscillator that provides all 40 channels plus all the frequencies in-between authorized channels. For example, the dial is calibrated for Channels 11 and 12. The frequency between the two is not authorized for Class D, yet the signal generator has a dot and frequency readout for Channel 11½, so adjacent channel rejection for channels 11 and 12 can be measured. The two "channels" between Channels 22 and 23 were similarly unassigned to CB. They are now Channels 24 and 25 and are so indicated on the dial.

The CHANNEL selector indicates the frequency and the channel, or a dot if no channel is assigned, for every step of the selector switch.

On the right side of the front panel are the RF output controls and an associated meter calibrated in microvolts and dB. A precision shielded attenuator provides a full-scale RF output of 1 to 100,000 microvolts. An RF level control permits the output to be further reduced to 0.1 µV. For adjacent channel and AGC measurements, the attenuator and meter are calibrated in dB as well as absolute microvolts. The RF output is available at a standard SO-239-type coaxial connector.

For service and troubleshooting there is also a 455-kHz oscillator complete with its own output control and output jacks. This oscillator can be used simultaneously with the RF signal generator.

On the left side of the front panel is a fully metered modulator and incremental frequency control. The modulator provides 400, 1000 and 2500-Hz internal modulation, or external modulation, to full 100%. The modulation amplitude is adjusted by a continuously variable control and the percent modulation is indicated directly on the meter. By "flipping" a meter function switch, the meter is set to indicate the delta frequency (incremental adjustment) of the main oscillator. A control
You get a lot extra with Jerrold’s Xtra Rated Antennas.

Jerrold Super VU-Finder and Paralog Plus Antennas are now Xtra rated to give you:

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Jerrold VU-Finder and Paralog Plus Antennas have long been the performance standard of the industry. Yet, we are constantly working to advance the state-of-the-art and have now added a number of important improvements. Our antenna line is significantly better than any antennas we have ever produced. And that goes for competitive antennas too! The Xtra features found in the Super VU-Finder and Paralog Plus Antennas make them go up faster, perform better and stay up longer.

Here are a few of the features now found in Jerrold Xtra Rated Antennas:

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  - **Serrated Jaw Bites Into Mast**
  - **Evaluating Turning**
  - **Flanged Bracket**
  - **Xtra Rated**

- **Improved Stainless Steel Terminal Screws**
  - **No-Strip Stainless Steel Washers**
  - **Vibration Resistant Lock Nut**

- **FM Breakaway**
  - FM blocking elements reduce received FM up to 12 dB. To receive FM at full gain, break off the element at score mark.

- **Stamped Termination Bracket**
  - The stamped termination bracket provides dc continuity and provides higher front-to-back ratio by using the rear element as a tuned reflector.

Dollar for dollar, and dB for dB, no competitive antennas match the VU-Finder and Paralog Series for ease of installation, performance and reliability. Install a Jerrold Xtra Rated Antenna today and see the difference for yourself or write for complete details.

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external speakers save lost CB sound

Notice what happens to the high frequencies? You lose them mounting the average CB under the dash. The speaker points down into the floor insulation. Sound is lost. With the addition of an acoustically designed "KRIKET®" external speaker, also mounted under the dash but pointing at the driver, the high frequencies come through.

The consonant sounds are in the high frequencies. And they spell the difference between voice intelligibility and just plain noise. That's why you hear remarkably better with a "KRIKET®" external speaker. It's the single best accessory you can add to any CB transceiver—23 or 40 channel—to improve enjoyment of it.

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using American materials and craftsmen.

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channel; the level of the sound carrier, which must be from 13 to 17 dBmV down to the video carrier, and so on and on. These may be checked with a field-strength meter, of course, but with the VSM-5, the operator can get the “whole picture” at once.

The instrument can be used as a field-strength meter, too, for orienting antennas, measuring their gain, and similar tricks. The height of the pipes on the screen is directly proportional to the signal strength. In CATV systems which change channel frequencies, it can be used with an external standard to verify that the new carrier is exactly on frequency.

With a special “slow-sweep” generator, the whole CATV system can be tested for flat response. This can be done with the system in full operation; the subscribers don’t even know it is being done! The sweep-generator is fed in at the head-end, and the VSM-5 is connected to a tap at the end of the line. The slow-sweep generator generates a continuous wave that is swept at such a speed that it passes through active channels without causing any disturbance to the picture. The VSM-5 is set to the same sweep frequency and the resulting trace can be photographed to verify flat response of the entire system.

By adjusting the DISPERSION control, each channel can be checked. The sound and video carriers will show up as sharp pips, and the color carrier as a constantly varying smear. Any sign of intermodulation distortion in the system shows up instantly as a pip where there shouldn’t be one. At a very wide setting of the DISPERSION control, the harmonics of the horizontal sweep show up.

The harmonics are spaced 15.75 kHz apart and any 1M shows up between these.

Hum levels in the system can be checked, and signal-to-noise readings taken. If any interfering signals are spotted, the instrument can be set to a lower resolution so that the exact frequency of the interfering signal can be found. The resolution can be set to 200 kHz, 10 kHz or 0.5 kHz (500 Hz) for this. For verification, markers can be switched in at 1, 10, or 50 MHz. These generate a set of zero-beats or birdies for this test. By counting the number of beats, say 1.0 MHz apart, the signals can be identified.

Beside the variable VHF sweep, three preset bands may be used. These can be tuned to any frequency in the range for tests where the same signal must be checked. For example, we used the markers to check out a set of CB traps by setting up a 27-MHz signal at the left side of the screen, its 2nd harmonic (54 MHz) in the middle, and the 3rd harmonic on the right. By hooking the traps in series with the input and tuning them, their effect could be instantly seen. The same method could be used for setting TV interference traps in a CB (or any) transmitter. All you need is a signal generator that will give you harmonics. I found out, by the way, that my RF signal generator wasn’t all that “pure”: with the output cranked wide open it developed a very perceptible 2nd harmonic! For checking CB transmitters, all you need is an RF pickup that will give you about 1.0 watt into the VSM-5 input. The instrument has pushbutton attenuators on the front panel that are capable of providing a total of 62 dBmV in 7 steps. There are two 20-dBmV attenuators, then 1.2.3.6 and 10 dBmV so that you can get whatever attenuation is needed. The actual signal level is then read from the screen and the in-circuit attenuators added to this.

The VSM-5 is powered by an internal 12-volt battery that is rechargeable or from a 117-230-VAC line. It has battery and may be used “portable” in the field for up to three hours of continuous operation. This makes it ideal for use as a mobile unit for detecting and measuring cable leakage, antenna testing, gain comparisons and your name-it.

This instrument will do a lot more things; its usefulness is limited only by the ingenuity of the operator. (This almost got me at first, but by diligent reading of the instruction manual, I finally made it do everything I wanted it to!) The actual operation is quite simple once you learn how, which really doesn’t take all that long. The suggested retail price of the VSM-5 is $3950.

---

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Unique full-function 8-digit wrist calculator... available only as a kit.

A wrist calculator is the ultimate in common-sense portable calculating power. Even a pocket calculator goes where your pocket goes—take your jacket off, and you're lost!

But a wrist-calculator is only worth having if it offers a genuinely comprehensive range of functions, with a full-size 8-digit display.

This one does. What's more, because it is a kit, supplied direct from the manufacturer, it costs only a very reasonable $19.95. And for that, you get not only a high-calibre calculator, but the fascination of building it yourself.

How to make 10 keys do the work of 27

The Sinclair Instrument wrist calculator offers the full range of arithmetic functions. It uses normal algebraic logic ('enter it as you write it'). But in addition, it offers a % key; plus the convenience functions √x, 1/x, x², plus a full 5-function memory.

All this, from just 10 keys! The secret? An ingenious, simple three-position switch. It works like this:

1. The switch in its normal, central position. With the switch centered, numbers—which make up the vast majority of key-strokes—are tapped in the normal way.
2. Hold the switch to the left to use the functions to the left above the keys...
3. And hold it to the right to use the functions to the right above the keys.

The display uses 8 full-size red LED digits, and the calculator runs on readily-available hearing-aid batteries to give weeks of normal use.

Assembling the Sinclair Instrument wrist calculator

The wrist calculator kit comes to you complete and ready for assembly. All you need is a reasonable degree of skill with a fine-point soldering iron. It takes about three hours to assemble. If anything goes wrong, Sinclair Instrument will replace any damaged components free: we want you to enjoy assembling the kit, and to end up with a valuable and useful calculator.

Contents
Case and display window.
Strap.
Printed circuit board.
Switches.
Special direct-drive chip (no interface chip needed).
Display.
Batteries.

Everything is packaged in a neat plastic box, and accompanied by full instructions.

The only thing you need is a fine-point soldering iron.

All components are fully guaranteed, and any which are damaged during assembly will be replaced free.

The wrist-calculator kit is available only direct from Sinclair Instrument. Take advantage of this 10 day money-back offer.

Send the coupon today.

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Hicksville, N.Y. 11801

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TOTAL $__________

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Address
City
State Zip

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RE-2

FEbruARY 1977

37
Measure $f^\tau$ and a full range of semiconductor parameters for under $300!$

with the B & K-PRECISION Model 530

New from B & K-PRECISION, the Model 530 semiconductor tester can actually perform more tests, on more devices, than any other competitively priced semiconductor tester. In the engineering lab, the 530 provides a full range of accurate data including an exclusive bipolar transistor gain-bandwidth product measurement capability to 1500 MHz.

In-circuit tests include:
- Positive good/bad tests for bipolar transistors, FET's, diodes, and SCR's
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- Identification of all leads of SCR's

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- Ioss and gate leakage of FET's
- Identification of device as bipolar transistor, FET or SCR
- Plus all in-circuit tests and many more parameters

The 530 is not only one of the most complete semiconductor testers available, it's one of the easiest and fastest to use. No charts, tables or calculations are required for basic operation, nor is additional equipment required. The 530 is a direct reading stand-alone instrument. Even unskilled assembly line workers can perform quality control tests with minimal training. Both visual and aural test indications are provided.

The B & K-PRECISION Model 530 is priced at only $295. For additional information or for immediate delivery see your nearest B & K-PRECISION distributor.

*Gain-bandwidth product or unity-gain frequency.
Radio-Electronics

40-CHANNEL CB SPECTACULAR

SPECIAL 16-PAGE SECTION

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The approval of 40-channel mononium among manufacturers to abate. Some new type-accepted shelves and others on the way.

Alpine
Type accepted: BR640. No details.

Audiovox
Several units at FCC of a proposed 10-model line for 1977. Marketing expected to begin in March. Line will consist of six in-dash models, some with FM, and four under-dash units. (150 Marcus Blvd., Hauppauge, NY 11787)

Automatic Radio
Two under-dash models at FCC, prototypes of which had earlier been type-accepted. Expected to be ready for sale on Jan. 3. The sets are: Model CBL2270, list priced at $181.95, with built-in slide-in/out theft-proof bracket, RF gain, Delta tune; and CBH2265, list priced at $202.50, with theft-proof bracket, RF gain, LED readout. (2 Main Street, Melrose, MA 02176)

B & B
At FCC: Five mobiles, one base unit. Prices expected to be "approximately the same price as equivalent 23-channel models when originally introduced." Two models are expected to be ready for January shipment. (185 Park St., Troy, MI 48064)

Benjamin
Type accepted: Model 200. No details.

Bohseii
At FCC: Two mobiles. Expected to be available in March. (7037 Hayvenhurst Ave., Van Nuys, CA 91406)

Boman
Type accepted: Model CBR-9950, not priced, in-dash AM/FM/FM stereo/CB with LED readout, T/R mode indicator light, S/RF meter, ANL, fader control for four speakers.

At FCC: Nine models, consisting of two in-dash, four under-dash, one base station, one modular hideaway (trunk, firewall installation), and one hideaway for attachment to stereo player. Expects to ship CBR-9950 Jan. 1. (9300 Hall Road, Downey, CA 90241)

Browning
Five models at FCC—three AM under-dash, one SSB AM under-dash, one single sideband AM base station. No pricing set.

Company is confident all five will be type accepted, with some product ready for shipment to dealers for Jan. 3 selling, more ready within 10 days. (1269 Union Ave., Laconia, NH 03246)

Channel Master
Type accepted: Under-dash Model CB-5835, featuring PLL, LED readout, Delta tune, ANL, noise blanker. Price not set.

At FCC: One under-dash model. Expects also to submit a base station to
CB unleashed a pandemic that is just starting. Models are on dealer's shelves. Here's what to look for.

FRED PETRAS

a final fix on CB as regards the position of 23-channel versus 40-channel. Several suppliers—some with big, some with small 23-channel inventories—predicted that as the consumer weighs the merits of both types, he will be inclined to lean toward 23's in terms of their initial wide price differential. This, say the producers, may cause some dealers to jack up the prices of 23-channel units in their stocks closer to their earlier prices—but not all the way. For example, units that once sold for $140, later discounted to $110, then to $90, and finally on retail shelves at $70, may be hiked back up to, say, $100 or $110. This will preserve a substantial price differential for the cost-conscious consumer. However, we expect to see the supply of 23-channel equipment dry up by the middle of the year. So if you plan to buy 23, you best buy soon.

Other industry members are predicting that the gap will lessen as 40-channel prices move down to "realistic" levels that will be determined in part by consumer acceptance of, or resistance to, the new products.

While we have scoured the industry to come up with a comprehensive roundup of 40-channel product available for January, 1977, selling, that roundup understandably is not complete in the light of: Early press deadlines common to magazine publishing. The unfinished processing of submitted prototype products by the FCC. Incomplete information on the part of CB manufacturers. We beg your indulgence under these circumstances. We hope to add to the information that follows with a second listing in our next issue.

NOTE: Prices indicated are those given when companies were contacted. Some are approximate, and are so indicated. Some prices may change, in line with production/materials cost vagaries.

FCC for late winter/early spring marketing. (Ellenville, NY 12428)

Clarion

Awaiting approval at FCC are: Model 367, $329, in-dash AM/FM/FM stereo/CB; Model 621, $259, same as 367 plus 8-track player; Model 672, $279, same as 367 but with cassette mechanism; Model RCJ-003, $199, modular system with separate integrated head containing mike and controls, with electronics mountable on firewall, under dash, under seat, or in trunk.

Expected to be in production in January for marketing the latter part of February. (5500 Rosecrans Ave., Lawndale, CA 90260)

Cobra

At FCC: Nine models, most variations on one basic chassis, consisting of six AM mobiles, two SSB mobiles, one base station. Expects to have some units for shipment Jan. 1. Complete 1977 line will consist of 12 models. (6460 W. Cortland Ave., Chicago, IL 60635)

Cobra 21XLR

Colt

Type accepted: Model 290, $200, under-dash, with LED readout, ANL, noise blanker, Delta tune. RF gain, S-meter, PA. Expected to be available at retail mid-January.

At FCC: 450, $220, 8-channel, same features as 290 plus mike control, AM/FM possibility, SWR meter; 480, $320, under-dash, AM/SSB, with same basic features as 450. 800, $300, AM base station, LED readout, walnut cabinet, SWR calibration, Delta tune, ANL, noise blanker. (5725 N. Central Ave., Chicago, IL 60646)

Commando Type accepted: CC-4040, under-dash AM radio, LED readout, PLL. No price set. CC4086, no details. At FCC: CC-4030, under-dash mini model, LED readout, PLL; CC-4055, under-dash full-feature unit with Delta tune, PA, LED readout, PLL; CC-5010, in-dash AM/FM/CB combo with LED readout and PLL.

Units at FCC expected to be approved in time for marketing Jan. 1. (P.O. Box 11071, Chattanooga, TN.)

Craig

At FCC: Model L-101, under-dash, approximate price $170. Features include Day/Night filament channel readout under all illumination conditions, reversible slide-out bracket, switchable ANL, PA, squelch, large S/RF meter, proprietary Craig mike/control unit.

Model L-131, under-dash SSB, with full complement of SSB controls, noise blanker, Day/Night channel indicator. Approximate price, $360.

Model L-231, SSB base station, approximate price $500. Features include digital clock alarm timer circuitry, dual antenna input.

Model L-600, under-dash, approximately $280. Features include 2-speed up/down channel selector on Craig mike and radio itself, provision for feetwitch channel selection, Day/Night channel indicator, headache mike bracket. (921 West Artesia Blvd., Compton, CA 90220)

Fanon/Courier

Fanon division. At FCC: Eight models, consisting of one AM base station, one AM SSB under-dash unit, five
under-dash AM units, and one in-dash combination—AM/FM/FM stereo/CB with Channel 9 priority switch and LED readout/selector built into the mike. One of the five under-dash units also has the latter feature.

Six of the above expected to be ready for sale on Jan. 1. (Subsequently type accepted: Fantaire 100F, no details.)

Courier division: Type accepted: Rebel 40, under-dash, priced at $100.

At FCC: Eleven models, all expected to be type accepted for January 1 selling. (990 South Fair Oaks Ave., Pasadena, CA 91105)

Xtal

At FCC: Six models in the XTAL brand consisting of two under-dash, four modulars for trunk or firewall installation. No price information. (Far Eastern Research Laboratories, 8749 Shirley Ave., Northridge, CA 91324)

Fieldmaster


Both expected to be available at retail on Jan. 3. (21212 Vanowen St., Canoga Park, CA 91303)

Arthur Fulmer

At FCC: Model 15-4035, under-dash AM/FM/FM stereo/CB with automatic warning light and full lineup of features including ANL and noise blanker; Model 16-4040, in-dash AM/FM/FM stereo/CB with LED readout, monitor override, overload circuitry. Former expected to be available “almost immediately” in January, the latter by Feb. 1. No prices set.

Expected for spring sale, a low-priced unit, Model 15-4030, no details. (P.O. Box 117, 260 Monroe, Memphis, TN 38103)

GC

At FCC: Model 9200, under-dash, with full complement of features, including PA and noise blanker. Planning to submit several more for type acceptance. (4000 South Wyman St., Rockford, IL 61101)

Gemtronics

Six models at FCC—five under-dash with “standard” features, plus one base station. Expected to be approved in time for January marketing. (356 South Blvd., Lake City, SC 29560)

General Electric

Type accepted (all under-dash units): Model 3-5801, $135, lighted S/RF meter, quick release mounting system, switchable ANL; Model 3-5811, $160, large meter, PA, switchable ANL; Model 3-5819, $220, with 3-way lighted meter, LED readout, antenna warning light, switchable noise blanker and ANL; Model 3-5821, $240, with channel priority feature. S/RF meter, PA, LED readout. Model 3-5812, $175, with S/RF meter, switchable noise blanker and ANL. Model 3-5825, $330, SSB, with AWM light, S/RF meter, LED readout.

At FCC: Model 3-5871, $250. AC/DC base station with LED readout. S/RF meter, switchable ANL.

Units expected to be ready for delivery to dealers on Jan. 1. (Audio Electronics Products Dept., Syracuse, NY 13212)

Hallicrafters (Breaker)


Expected to be type accepted in time for early January sale. (1101 Great Southwest Parkway, Arlington, TX 76011)

Hanabashiya

No products at FCC. Earliest possible 40-channel products in March. (39 West 28th Street, New York, NY 10001)

Handic

Sent 15 models to FCC too late for

Johnson Viking 4740

January selling. They consist of four mobile under-dash units, 1 in-dash AM/FM/FM stereo/CB, 2 base stations, 1 mobile SSB and 1 base SSB, 4 handheld transceivers, 2 limited-channel (3 and 6) mobiles using crystals. Handic is projecting a starting price of $140, with most full-feature models in the $200 to $250 range. (14560 N.W. 60th Ave., Miami Lakes, FL 33014)

Hy-Gain

Type accepted: Model 2701, $139.95, under-dash model with ANL, volume/squelch control, remote speaker jack, illuminated channel selector, TVI filter; Model 2703, no details. Both expected to be available Jan. 1.

At FCC: “About 20” models, consisting of in-dash, under-dash, remote-control and SSB sets, including one AM/FM/FM stereo/CB combo with cartridge playback facilities. (8601 N.E. Highway 6, Lincoln, NE 68505)

I. A. Sales

At FCC: Models TRX-400 and TRX-500, both under-dash, both with full complement of features. Expected to be retail in January. (766 Lakefield Road, Westlake Village, CA 91361)

Inland Dynatronics

No information available. (10 Horizon Blvd., South Hackensack, NJ)

J.L.

At FCC: Model 860, suggested list price $400, in-dash AM/FM/FM stereo/8-track cartridge/CB; Model 615, $420, same as 860 but with cassette mechanism instead of cartridge: Model 853-401, $390, in-dash, AM/FM/8-track/CB; Model 853-201, $690, 120-channel SSB in-dash AM/FM/8-track cartridge facilities. Model 706, $375, in-dash AM/FM/FM stereo/CB with push-button tuning. SSB version of 706, $670. (737 West Artesia, Compton, CA 90220)

Ray Jefferson

At FCC: CB-740, $150, under-dash; CB-845, $200, deluxe under-dash with LED readout, ANL, Delta tune; CB-7120, $300, SSB under-dash; CB-702, $280 (approximate), LED readout; CB-712, $230 (approximate), marine unit.

Expected to be ready for January 1 selling. (Main and Cotton Streets, Philadelphia, PA 19127)

E. F. Johnson

At FCC: Five under-dash units under the Messenger brand name, all with LED readouts, as follows: 4120, suggested list $110: 4140, $180; 4145, $180; 4170, $250; 4175, $250. Messenger Model 4230, $250, AM base station. Viking Model 4740, under-dash SSB priced at $360. (299 10th Ave., SW, Waseca, MN 56093)

Kraco

Type accepted: Model KCB-4003, under-dash leader model, no price set; Model KCB-4030, $220, under-dash with PLL, S/RF meter, squelch, ANL, noise blanker, Delta tune; Model KCB-4020, $180, same features as 4030 but no noise blanker; KCB-4088, no information.

At FCC: Model KCB-4010, $140,
under-dash mini model. PLL, squelch, ANL, S/RF meter. (505 East Euclid Ave. Compton, CA 90224)

**Kris**

Has three under-dash CB radios at FCC. They are: **XL-40** $160, with ANL, PA, RF meter: **XL-45** $210, with ANL and noise blanker, PA. LED readout: **XL-50** $260, deluxe full-feature unit with 3 meters (RF, signal strength, modulation) talk-back feature, ANL and noise blanker, meter dimmer control. Kris expects to have the units type accepted in time for "early January" sale. The company also plans to submit "at least two other models" for eventual marketing later in 1977 (Pioneer Road, Cedarsburg, WI 53012)

**Krypton**

One under-dash model at FCC. Sending one under-dash and one-in-dash to FCC in January. Markets under BETA brand name. (18 Mileday Way, Avenel, NJ 07001)

**Lafayette**

Type accepted: **IB-940** $200, LED readout, SWR calibration control. RF gain, squelch. Delta tune, noise blanker, ANL, S/RF meter. PLL: **IB-740** $160. PLL: Delta tune. ANL, noise blanker: **IB-640** $120. with PLL S/RF meter, transmission indicator, squelch, PA.

At FCC: base station and two SSBs. Type accepted models to be available in all Lafayette company-owned and franchised stores Jan. 3. (111 Jericho Turnpike, Syosset, NY 11791)

**Lake**

No information available. (1948 E. Lehigh Ave., Glenview, IL 60025)

**Maxon**

Type accepted: **Maxon 40**. No details.

**Meishoh**

Type accepted: **ME-402**. No details.

**Meriton**

Type accepted: **RS-5111**. No details. (35 Oxford Drive, Moonachie, NJ 07047)

**Midland**

Type accepted: Four under-dash models all with ANL and Delta tuning, squelch, and illuminated rotary dial tuning, among key features. They are: Models 77-857; 77-882; 77-888; 77-883. No price or shipping date information available at press time. Also type accepted. **Model 63-240**, and 77-853, no details. (P. O. Box 1903, Kansas City, MO. 64171)

**Motorola**

No information available. (1499 E. Algonquin Road, Schaumburg, IL 60196)

**Nuvox**

No 40-channel products planned at this time. Will enter the market when pricing "becomes practical and reasonable" as opposed to the current heavy discounting prevalent in 23-channel CB, said a spokesman. (150 Fifth Avenue, New York, NY 10011)

**Pace**

Type accepted: **Model 8041**. $200, under-dash unit featuring Delta tune, PLL, antenna warning light.

At FCC: Eighteen models consisting of 12 mobiles, three base stations, two in-dash AM/FM/CB combos, one scanner. Combos include cartridge or cassette facilities. 

Pace expects to ship between 15 and 17 models starting Jan. 1. (24049 S. Frampton Ave., Harbor City, CA 90710)

**Palomar**

Six units at FCC. Expects three imported models to be approved in time for February selling, three all-American-made in time for sale by end of first quarter.

Import units for possible sale in February are: Palomar Model 41, listed priced at about $185. Features include LED readout, switchable ANL, squelch, PA. Model 4100, approximately $219, whose features include those of 41 plus RF and microphone controls, snapper styling: **Model 49**, approximate price $249, same features as above two units plus noise blanker. All three are 4-watt, under-dash units.

Three all-American-made models consist of one tube-type base station, one mobile AM under-dash, and one deluxe SSB AM base unit. (665 Upper St., Escondido, CA. 92025)

**Pearce-Simpson**

Type accepted: **Tiger 40**, no price set, under-dash. Delta tune. PA. ANL, noise blanker, S/RF meter, tone, squelch; **Super Cougar 40**, under-dash unit whose features and styling are expected to be revised before marketing, along with a name change.

Submitted "at least 10 pieces to the FCC," in line with plans to have an eight to 10-unit 40-channel line in 1977.

J. C. Penney

Type accepted: **Model 6203**. No details.

**Pioneer**

Type accepted: **GT-6600**, no price set, in-dash AM/FM/CB with ANL, squelch, S meter, five pre-set tuning buttons. Expected to be on sale Jan 1.

At FCC: **Model 1100**, no price, in-dash, AM/FM/CB with ANL, squelch, S meter, manual tuning. (1555 E. Del Amo Blvd., Carson, CA 90746)

Panasonic

**Consumer Electronics Group**: Four under-dash units at FCC, as follows: RJ-3250, $200, with quick-release mounting bracket, Channel 9 priority switch, LED readout; RJ-3450, $230, 2-piece modular system for under-seat, trunk, firewall installation, with built-in scanner system, remote control mike; RJ-3510, $170, and RJ-3030, $130, both with LED readout.

**Auto Products Dept**: Type accepted: Model CR-B4747, in-dash AM/FM/FM stereo/CB radio with push-button tuning; Model CR-6470, mini in-dash AM/FM/FM stereo/CB radio with manual tuning. No prices set.

Three models at FCC. No details.

Type accepted models will "absolutely be ready for January 1 shipping." (One Panasonic Way, Secaucus, NJ 07094)
President

Type accepted: Zachary T., $250, AM base station with ANL, PA, RF gain, LED readout, S/RF meter.

At FCC: Six models—also named after presidents of the United States—expected to be type accepted in time for January 1 retailing. They are: Model John Q., suggested list price, $170, under-dash AM mobile with ANL, PA, LED readout, S/RF meter; Honest Abe, $200, under-dash AM, PA, ANL, Delta tune, S/RF meter, LED readout; Teddy R., $230, under-dash AM, PA, noise blanker, SWR calibration, Delta tune, LED readout, S/RF meter, PA; Dwight D., $330, AM base station, noise blanker, PA, RF gain, tone, SWR calibration, clock, alarm control, Delta tune, PA, on-air light, LED readout, with separate speaker; Grant, $340, SSB AM under-dash, PA, noise blanker, LED readout, S/RF meter, mike gain; Washington, $430, SSB AM base station, PA, noise blanker, mike gain, RF gain, clarifier, PA, LED readout, S/RF meter. (16691 Hale Ave, Irvine, CA 92714)

Radio Shack

Type accepted: TRC-455, AM base station, PLL, LED readout; TRC-466, low-priced "very small" under-dash AM with illuminated channel switching; Mini-40, compact under-dash AM with illuminated channel switching.

Company expects to have 14-unit line of 40-channel CB's for 1977 selling. (2617 West 7th St, Fort Worth, TX 76107)

RCA

Type accepted: Model 14T304, AM mobile unit with LED readout; 14T720, AM mobile unit, 40-detent channel switch. Both models with ANL. No prices established.

At FCC: Eight models, as follows: one SSB, one base station; five mobiles, one AM/FM in-dash combo. These are expected to be accepted in time for sale in January. (Cherry Hill Offices, Bldg. 206-2 Camden, NJ 08101)

Regency

Type accepted: Models CR-430, 485, 486, all foreign-made under-dash units with LED readouts, expected—"Hopefully"—to be available at retail Jan. 3.

At FCC: Model CB-501, American-made under-dash unit, first of what will be a completely American-made line.

Price range of the above four models is from $159.95 to $219.95. (7707 Records St, Indianapolis, IN 46226)

Robyn

Eleven models at FCC, with first seven of those in the following list expected to be approved in time for Jan. 1 marketing, the rest shortly thereafter. All units listed feature PLL. All except T-240-D are under-dash radios. Model WFV-110, suggested list price $150. WFV-110P, same as 110 with antenna and coaxial speaker, $170. LB-120, $160. (Above three units feature back-lighted rotary channel indicator.) DG-130D, $180, with digital readout. SX-401, $190, with rotary back-illuminated channel indicator. SX-402D, $240, with digital readout. 007-140, $200, with antenna, coaxial speaker. TR-210D, $200. GT-410D, $260, with digital readout, turn-on/tun-off switch that automatically returns to Channel 9, scanner switch. GT-440D, $380, SSB with digital readout. T-240D, $300, mobile or base station radio using vacuum tubes, featuring digital readout. PLL. (P. O. Box 478, 10901 Northland Drive, Rockford, MI, 49341)

SBE

Ten under-dash radios at FCC. Most feature LED readouts, all use PLL circuitry.

It is expected that the units will be type accepted in time for Jan. 3 retail sale. (220 Airport Blvd, Watsonville, CA 95076)

Sears Roebuck

Type accepted: Models 28 62674, CM 23785A, CM 6000LB, TA 4501, 28 62676, CM 6000LA, CM 6100S. No details available.

Shakespeare

Type accepted: Model GBS-240, no details.

At FCC: Two mobiles and one base station. The firm expects these to be accepted by Jan. 1, but their actual retailing time will be determined by Shakespeare's Japanese supplier. (P. O. Box 246, Columbia, SC 29202)

Sharp

Type accepted: CB-2260, approximately $140, under-dash, Delta tune, lighted drum channel indicator, flashing Channel 9 indicator, PA, squelch; CB-2460, approximately $150, same as 2260 but with LED readout.

Delivery expected first week of January. (10 Keystone Place, Paramus, NJ 07652)

Siltronix

Late in submitting prototypes to the FCC, this company will not have 40-channel CB's until late February. Awaiting FCC action are three under-dash units, all featuring PLL and LED readouts. (330 Via El Centro Ave, Oceanside, CA 92054)

Sparkomatic

At FCC: Five mobile under-dash and two base stations. Three of the mobiles have touch-bar channel switching. One has "memory" control for automatic tuning of one channel. (Milford, PA 18337)

Standard

Expects to have unit now at FCC approved for sale in January. It is a 40-channel version of the firm's Horizon 29, to be called Horizon 29A, a mobile model featuring ANL, AN blanking, Delta tune, and squelch, among features. It will carry the same list price as the 23-channel model. $229.95. (P. O. Box 92151, Los Angeles, CA 90009)

Surveyor

At FCC: One AM base station; one SSB base; one in-dash AM/FM/CB combo with LED readout, three under-dash (one with LED readout). (7 Electronics Court, Madison Heights, MI 48071)

Teaberry

Type accepted: Model 4002, AM base, dial selector; 4006, AM mobile, LED readout; 4001, SSB mobile, dial selector. All three with PLL.

At FCC: One SSB base unit, one SSB mobile, one base, two AM mobiles.

Pricing to be determined at end of December. At least five models expected to be ready for shipment Jan. 1. (6330 Castleplace Drive, Indianapolis, IN 46250)

Tram/Diamond

Five units at FCC: Model D-42, with ANL, LED readout. PLL: D-12, with PLL (both units under-dash); one SSB

continued on page 50
IT SHOULD COME AS NO SURPRISE THAT CB radio has entered an exciting new stage of growth. The FCC announcement of expanded Class-D service with the addition of seventeen channels was encouraging since it supported the largest of radio services. Unfortunately, the announcement also generated considerable confusion about the status of existing CB products. It also raised the question of just what new products would be made available to fill the needs of the 40-channel market. Now that the dust has settled, what Antenna Specialists has done to meet the need for forty-channel CB antennas should reduce some of the confusion about antenna products and the expansion.

First, it is very important to know that an antenna is not just a piece of cable, a coil, and a steel tip rod. There are many factors that enter into antenna design and affect antenna performance. Since base-station antennas are already "broad band" and will cover the proposed additional channels, they will not be discussed in this article. Mobile antennas provide a whole different set of conditions; however, and those conditions should be understood if proper antennas are to be selected for each application. Some basics of antenna design and history will help to clarify the problem.

Basic antenna design

As mentioned above, an antenna is more than a length of cable, a coil, and a rod. Stated in its most simple terms, an antenna is a device that is capable of sending and receiving radio-frequency energy.

That's simple enough, but simple things are often misleading. For example, each antenna must be constructed to operate over a specific range of frequencies. In addition, each antenna must be tuned to operate with a maximum efficiency and it must function equally well both transmitting and receiving. Under most conditions an antenna works best as a half-wavelength radiator.

On CB frequencies, 27-MHz, one-half wavelength is about 18 feet long. A quarter-wavelength radiator and a quarter-wavelength mirror image or ground plane make up the 18-foot system. In the early days of CB, the most common kind of antenna was the 108-inch whip and spring combination. Whip and spring were the active or radiating half of the system and the vehicle body acted as its "mirror image" or ground plane. In effect, this formed the dipole antenna. The long whip system was broad band and offered little problem in terms of matching the antenna to the transmitter.

The quarter-wavelength whip and spring combination did have some problems in terms of aesthetics. In addition, the quarter-wave whip & spring combination has some mounting difficulties since most people don't like a 9-foot whip on top of their car. The most common place to mount the antenna, bumper or deck, played games with the antenna's radiation pattern. The quarter-wave whip can also be a problem where areas of low clearance exist.

To meet some of the above shortcomings, a loaded antenna was developed. In this way we make antennas mechanically shorter without significantly changing electrical efficiency. The loading coil changes the antenna's voltage and current components and narrows the bandwidth of operation.

RICHARD BITNER
*The Antenna Specialists Co.

THREE BASE-LOADED WHIPS from Antenna Specialists. The models MS178 (left), M-276 (middle) and M-440 right.

FEBRUARY 1977
Operation at or near the resonant frequency is excellent. Using a loading coil permits an antenna of almost any length to be developed but, the shorter the length the more restricted the design and the bandwidth. This is important in considering 40-channel antenna design.

There are three kinds of loaded mobile antennas: base loaded, center loaded, and top loaded. The top loaded antenna, if it is a true top load, is impractical for mobile service since it requires the use of a large capacitance "hat" to couple it to its mirror image. They were not considered for redesign. The most common type of antenna, the base-load design and the center-load design both have advantages and were redesigned for 40-channel coverage.

**Mounting 40-channel antennas**

The mounting location of the antenna also has a direct bearing on its performance. The best place to mount any mobile antenna is still the center of the vehicle roof. Trunk or lid mounting with a center location is second best. Other mounting locations tend to distort the pattern, provide restricted mirror image, or reduce effective bandwidth and matching capability. Site election for 40-channel antennas may be somewhat more critical than for 23-channel applications.

With a multiple of variables and design considerations to take into account, design engineers started to develop broadband CB antennas. Keeping antennas as short as possible was still a main criteria.

**BASE-LOADED WHIP AND SPRING combination. Antenna Specialists model M-125.**

Basically, the base-loaded designs were changed by use of a lower Q coil and a slightly longer tip rod. Q is a measurement of antenna selectivity. In essence, the higher the Q the narrower the bandwidth. Since we are concerned with effective bandwidth, those frequencies on either side of the antenna's resonant frequency, lowering the Q expanded the bandwidth of the antenna. Antennas that are broadband will feature a reworked coil and a longer tip rod. They will physically look the same but have bandwidths that approach 700 kHz at a 2:1 VSWR.

**CENTER-LOADED WHIP with gutter mount. Antenna Specialists model M-489**

Center-load designs involve the different set of design parameters. Broadbanding was accomplished by using a coaxial matching section in series with the feedline. This has caused expanded bandwidth to approach 300 additional kilohertz and makes it possible to cover the entire band. Short center-load antennas, however, will remain narrow-band in configuration.

Since effective bandwidth is a function of VSWR perhaps it would be wise to restate some facts about VSWR. Throughout the past several years, the term VSWR, or more simply SWR, has become greatly overused. There seems to be currently, a feeling that an SWR ratio of 2:1 is unusable. This could not be further from the truth for actual reflected power at a VSWR of 2:1 is only about 11% of the total radiating power.

What this really means is that signal loss with a VSWR of 2:1 is in the neighborhood of 1/2 of a dB. This is undetectable at either end especially during short-range CB communications. Above 2:1, however, the ratio in its effect changes greatly. No one should be overly concerned with a VSWR in any installation of 2.0:1 since it is a nominal figure and sought by commercial installers.

Any installation variable that changes the design parameters of an antenna can reduce bandwidth and cause an increase in VSWR. Good bonding to the vehicle body, proper antenna site selection, and care in tuning are most important. Since some applications will call for specific types of mounts for antennas, then any limitation imposed by that requirement should be understood by the user. The new 40-channel designs will work very well with 23-channel equipment. Also, remember that any current antenna will work with 40-channel equipment over any given segment of frequencies.

Short antenna designs such as rain-gutter mounts contain built-in compromises which make their inclusion in 40-channel expansion programs unpractical. Again, it is up to the user to determine exactly what his specific needs are in terms of installation requirements and select the antenna which will meet his needs.

Since there is a wide variety of antenna products manufactured, all application needs should be able to be met without significant difficulty. Such designs as dual antennas, provided they are not overly short, foldover mast designs and 108-inch whip configurations are inherently broad banded and will function well in the 40-channel expansion program.

**THREE-QUARTER LOADED. Antenna Specialists model M-312**

As we enter this new phase of CB, the product which is currently being manufactured to meet the new expansion requirements will provide those using the service with a means of pursuing it with a minimum of fuss. It will be important for the buyer as well as the seller to know what equipment will meet their specific needs. Personal Communications is an ever growing part of the electronics industry and will continue to grow as the need for CB product grows. The additional seventeen channels authorized by the FCC are just the start and as further changes come about, legitimate manufacturers will meet the changing demands. Personal Communications as a means of communicating is here to stay.
Using PLL for CB Frequency Synthesizers

The PLL, an old concept in frequency synthesis, when applied to 40-channel CB, provides superior accuracy and stability with just one or two crystals. Here's how it works.

ROBERT F. SCOTT
TECHNICAL EDITOR

WHEN FREQUENCY SYNTHESIS WAS FIRST APPLIED TO CB TRANSCEIVER TECHNOLOGY IN 1963, IT WAS IN THE INTEREST OF CONSERVING SPACE AND PROVIDING A HIGHER DEGREE OF ACCURACY AND STABILITY IN THE TRANSMIT AND RECEIVE FREQUENCY GENERATING CIRCUITS. THE TWELVE OR SO CRYSTALS OCCUPIED ABOUT ONE-FOURTH THE SPACE NEEDED FOR DIRECT FREQUENCY CONTROL IN A 23-CHANNEL RIG.

WHEN THE CB BOOM HIT A COUPLE OF YEARS AGO, THE CRYSTAL MANUFACTURERS OF THE WORLD DID NOT HAVE THE CAPACITY TO MAKE CRYSTALS AS FAST AS THEY WERE NEEDED FOR USE IN CB TRANSCEIVERS.

The obvious solution to the problem was to use a frequency synthesis scheme that required a minimum number of crystals. This proved to be the phase-locked loop (PLL) frequency generator that uses from one to three crystals—depending on circuit design.

Just as the PLL frequency synthesizer was gaining a foothold in the top-of-the-line models of some manufacturers, the FCC expanded the Class-D Citizens Radio Service to include seventeen new channels—extending the band up to, and including, 27.405 MHz.

With the crystal shortage still one of the most pressing component procurement problems, the phase-locked loop frequency synthesizer is the only way to go for a manufacturer who wants to remain in the CB transceiver business.

The phase-locked loop

The principle of the phase-locked loop is not new. It was used in some specialized receiver designs developed in the middle 1930's. The most common and most evident application was in the AFC circuits in some of the deluxe AM broadcast receivers of that day. Next came the application of the PLL principle to the synchronization of the vertical and horizontal deflection circuits in TV receivers. Now that we know of two familiar applications of the phase-locked loop, let's see just what it is and how it fits into the CB transceiver design.

Basically, the phase-locked loop is a type of servo-system whose output frequency locks onto and follows, or tracks, an input reference signal. The PLL servo system consists of a phase detector, a low-pass loop filter and a voltage-controlled oscillator or VCO. The function of the circuit is to generate an output signal that is an integral multiple of, and in phase with, the input frequency standard.

The input reference signal and the output of the VCO are locked in phase by comparing the phase of the two signals in a detector that converts any phase difference into an error-correcting voltage or current. The error voltage shifts the frequency—and therefore the phase—of the VCO output signal in the direction that causes it to track the input. The basic PLL system is shown in Fig. 1.

FIG. 1—SIMPLE PLL CIRCUIT produces an output frequency equal to the Input reference frequency.

With no reference voltage applied to the phase detector, there is no error voltage at the phase detector output and the VCO runs at its natural free-running frequency as determined by L-C or R-C constants. When a reference signal is fed into the PLL circuit, the phase detector develops two error signal voltages that are proportional to the difference between the reference and VCO signals.

One of the error voltages is the sum of the two frequencies—let's call them F_r and F_c for oscillator frequency and reference frequency, respectively. The other error voltage is determined by the difference between F_r and F_c. The two error signals are passed through the low-pass filter that eliminates F_r + F_c leaving only F_r − F_c as a DC, or near-DC voltage to control...
the VCO.

The error voltage reduces the phase, and therefore the frequency, difference between \( F_r \) and \( F \), until the frequencies are exactly the same. However, there will be a small but constant phase difference required to generate the error voltage that keeps the VCO and reference signals in lock.

If either the reference or VCO output changes phase, the phase detector and filter produce a DC control voltage that is proportional in magnitude and polarity to the phase change. The error voltage thus produced shifts the phase of the VCO signal by altering its frequency until it again locks onto the reference signal.

About now, you're saying, "So what? If all the VCO does is generate a frequency equal to the reference frequency, what good is it? We would need one precise reference frequency for each of the forty CB channels."

Well, what we have been discussing is the basic PLL frequency generator. In practical multi-channel PLL frequency synthesizers, \( F \) is a very low value and a programmable frequency divider is inserted in the feedback loop as in Fig. 2.

FIG. 2—THE OUTPUT FREQUENCY can be made different from the reference frequency by inserting a divide-by-N counter in the feedback loop. Now the output frequency is equal to \( N \times F_r \).

The divider is arranged so it can be made to divide the VCO output frequency by any desired integer \( N \). Phase lock is obtained by comparing the resultant frequency \( F_r/N \) with the reference frequency \( F_r \). The phase detector output is now proportional to the difference between \( F_r/N \) and \( F_r \).

In most applications, the reference signal is derived from a precision crystal oscillator whose output is fed to a series of frequency dividers. Dividing the crystal frequency by as much as two thousand is not uncommon as the resulting reference frequency is usually made equal to the channel spacing—10 kHz except for five 20-kHz gaps. (There are 20-kHz gaps between Channels 3 and 4, 7 and 8, 11 and 12, 15 and 16, and

19 and 20.)

The block diagram of the PLL synthesizer now appears as in Fig. 3. The crystal oscillator frequency is divided by a fixed factor (M) to develop an \( F_r \) that equals 10 kHz. The VCO must oscillate at \( N \times F_r \) to achieve phase lock. A programmable counter is used to select the desired VCO output that is always a multiple of \( F_r \). The divisor \( N \) can be derived from BCD switches, keyboard logic or from BCD data from a two-digit 7-segment channel indicator.

FIG. 4—PLL CIRCUIT of a single-crystal AM CB transceiver.
The CB channel frequencies are not divisible by 10, so various schemes have been used to, in effect, reduce channel spacing and \( F_0 \) to 5 kHz. In this case (considering circuits that develop the direct signal frequencies for the transmitter) \( N_{max} \) equals \( F_{trans}/F_0 \), and \( F_{max} \) equals \( F_{trans}/F_0 \). Thus, for Channel 1, \( N \) equals 26,965 kHz/5 kHz or 5393. For Channel 40, where \( N \) is maximum, \( N \) equals 27,405 kHz/5 kHz or 5481.

Most of the earlier PLL frequency synthesizers used in CB transceivers employed large-scale integration and/or discrete semiconductor components along with two or more crystals to generate the precise frequencies needed for the transmitter and receiver circuits. For example, in a transceiver with a single-conversion superheterodyne receiver, one crystal develops the reference voltage. The VCO locks onto \( F_0 \), and is used to develop the channel frequency for the transmitter circuits.

A second crystal generates a frequency which, when mixed with the VCO output, is used as the injection oscillator frequency to develop the receiver’s first IF. A third crystal is used in sets that have dual-conversion receivers. A fourth crystal is needed for sideband selection in SSB transceivers.

Recently, several semiconductor manufacturers developed single-IC frequency synthesizers that require only a single crystal to generate all transmit and receive frequencies. In addition, the devices can be programmed by a simple binary-coded channel selector switch, compatible with digital display devices, can be used in both single- and double-conversion receivers and can provide positive lockout against operation on frequencies not authorized as CB channels.

Among the devices available—but not necessarily used in CB production at this time—are Motorola’s XC3390P and the NC6402 by Nitron, a division of McDonnell Douglas Corp. These devices provide all receive and transmit frequencies while positively locking out all other frequencies.

The Nitron synthesizer

The NC6402 is one of a series of PLL-type frequency synthesizer IC’s developed for CB applications. It is a 16-pin dual-in-line device using N-channel MOS integrated circuitry. A block diagram of a single-crystal, 40-channel AM CB transceiver is in Fig. 4. All that is needed (in addition to the NC6402) for a complete CB frequency synthesizer is one crystal reference oscillator, a single-wafer rotary switch, a dual D-type flip-flop and an optional two-digit, seven-segment display.

The NC6402 interfaces with a two-digit, seven-segment encoded channel-selector switch. Five of the seven lines of the one’s digit and three of the seven lines of the ten’s digit are used to generate the address codes for the eight input program lines on the IC. Channels are selected by connecting the appropriate inputs to ground. Figure 5 illustrates switch programming.

The send/receive switch (Fig. 4) develops a logic signal to indicate whether the selected channel is in the transmit or receive mode. A logic “1” for receive and a “0” for transmit. The switch simply makes or breaks a connection to ground.

The VCO output terminal on the device varies between +3 volts and ground during the time that the unknown frequency (\( F_{unk} \)) is being locked in. It goes to a high-impedance state when the unknown frequency is locked in. The output is positive when \( F_{unk} \) is lower than the reference frequency and negative when \( F_{unk} \) is higher than the reference. The output current is sufficient to drive a passive low-pass filter and a varactor-tuned VCO.

While you are absorbing this introduction to PLL frequency synthesis for CB transceivers, we will be preparing to describe the Motorola XC3390P and to take a look at the circuitry used in the General Electric model 3-5800A 23-channel rig.

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**Note:** Switch closure to ground.

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**FIG. 5—SWITCH PROGRAMMING** of the NC6402 shown in Fig. 4.
New FCC Rules

ON JULY 27, 1976, THE FEDERAL COMMUNICATIONS COMMISSION (Charlie) announced changes in Part 95 of the Rules and Regulations authorizing seventeen additional channels for use on the Class D Citizens band, effective January 1, 1977. In addition, the Commission tightened some of the technical standards for type acceptance and amended some of the sections of the Rules and Regulations to clarify the FCC's intent. The sections affected by the changes are 95.3, 95.41, 95.42, 95.49, 95.55 and 95.58. These amended sections are summarized as follows:

95.3 Definitions (b) A Class-D station is now defined as one operated for radiotelephony in the 26.96 to 27.41-MHz band. (The band has been expanded from 27.255 MHz to 27.410 MHz for a bandwidth of 450 kHz. Fifteen of the added channels are spaced 10 kHz apart over the range of 27.255 to 27.405 MHz.)

95.41 Frequencies available. (2) Formerly listed the seven channels used for communications between units of different stations. Now, this section lists the forty channels ranging from 26.965 to 27.405 MHz that may be used for communications between Class D stations, effective January 1, 1977.

The FCC no longer refers to channel numbers—only frequencies. However, the industry has agreed to continue using channel-number designations with channel indicators and selectors noting Channels 1 through 40 as in the table.

(Note that Channels 24 and 25 have been centered in the 30-kHz gap between Channels 22 and 23)

(3) Deleted. Formerly listed Channel 11 (27.085 MHz) for use solely as the calling channel.

95.42 Special provisions. A new section that states that as of September 10, 1976, authorizations for use of frequencies between 26.96 and 27.41 MHz will be issued only to applicants in the Citizens Radio Service. Any license in a radio service other than the Citizens Radio Service authorizing the use of frequencies between 26.96 and 27.41 MHz shall remain in effect until December 31, 1979.

(Since its inception, the Class D CB service has shared the band—without protection against interference—with medical, industrial, public service and scientific users. No new licenses will be issued in this band except to CB users. Other services sharing the band with CB'ers must vacate the band by December 31, 1979.)

95.49 Emission limitations. (d) This section has been amended to require that the harmonic and other spurious outputs of a CB transmitter type accepted after September 1, 1976, be attenuated at least 60 dB below the fundamental power output in watts.

95.55 Acceptability of transmitters for licensing. (c) (4) New section. A transmitter capable of operating on any frequency other Class D CB channels may not be installed or used in any Class D station unless a special license authorizing that transmitter is posted in the CB station. In other words, if you have a ham transmitter, transceiver or linear amplifier at the CB location, you had better have a valid license posted for the non-CB gear.

95.58 Additional requirements for type acceptance. (c) (2) Multi-channel transmitters shall be capable of operating only on those channels authorized for Class-D CB service. All frequency-determining circuitry and components (other than the channel selector) must be inside the equipment cabinet and not accessible from outside the cabinet or from the control panel.

Add-on devices—whether internal or external to the original equipment—whose function is to increase the frequency coverage of a Class D transmitter beyond its original coverage range, shall not be sold, manufactured or attached to any transmitter designed for Class-D CB use. (No matter what you may have heard, add-on converters or adapters that give you 40-channel performance from a 23-channel rig are illegal and are not to be used.)

If you purchase a new CB rig—either 23 or 40 channels—you'll find a current copy of Part 95 of Rules and Regulations, an application blank for a Class-D CB license (FCC Form 505) and a temporary permit (Form 555-B) packed in the carton.

The serial number of every Class-D CB unit sold after January 1, 1977 will be engraved on the unit's chassis. No more gummed stickers or rubber stampings.

Any CB'er causing interference on television Channels 2, 5 or 6—because of spurious emissions from his rig—will be required to install a low-pass filter between the transmitter output and the transmission line that is feeding the antenna.

(All CB'ers would be wise to take a tip from hams who, for years, have used a low-pass filter on their rigs and high-pass filters on their TV sets.)

40-CHANNEL ROUNDUP
continued on page 44

in-dash with LED readout: TD-32 under-dash AM with controls in microphone; TD-52, single sideband under-dash, controls in the microphone. Latter two are full-featured, including PLL, LED readout.

"Two or three" of the above expected to be available for sale toward the end of January, 1977. (P. O. Box 187 Lower Bay Road, Winnisquam, NH 03289)

Tran Sonic
Type accepted: MCH-41. No details.

Toyota Motor
Type accepted: Model 00860-00001. No details.

Well
Type accepted: W-605, W-705.

Windsor
At FCC: Four under-dash units, ranging from $90 to $150. (10 Hub Drive, Melville, NY 11746)

Zodiac
"We missed the FCC deadline for January selling. We're not ready yet. We'll let you know when we have 40-channel CBs," said a spokesman. (626 Chrysler Building, New York, NY 10017)
samples of the following models of Class D, Citizens Radio Service, 40-channel transceivers have been tested by the Commission Laboratory and found to be in compliance with applicable technical requirements for type acceptance and certification.

As stated in earlier releases, grants of certification and type acceptance for these, and for other transceivers yet to be tested which were submitted by November 1, 1976; will be mailed in the latter part of December 1976, and will be effective January 1, 1977. Marketing and importation of these transceivers and others is prohibited prior to the effective dates of the type acceptance and certification grants issued. (See Section 2.803 of the Commission’s Rules.)

"Future lists of transceiver models found in compliance with Part 15 and Part 95 rule requirements by the Commission Laboratory tests will be issued in the coming weeks." We are told that there are approximately 300 waiting to be tested.

**40-Channel Transceiver Directory**

- **Alpine Electronics Co., Ltd.**
  - Model: BR-640

- **Audiovox Corporation**
  - 150 Marcus Blvd.
  - Hauppauge, NY 11787
  - Model: MCB-3000

- **Automatic Radio Mfg. Co.**
  - 2 Main Street
  - Melrose, MA 02176
  - Models: CBH-2265
    - Model: CB 2176

- **B & B Import-Export**
  - 185 Park Street
  - Troy, MI 48064
  - Model: B-4700

- **Benjamin Electronic Sound Co.**
  - 40 Smith St.

- **Boman Industries**
  - 9300 Hall Road
  - Downey, CA 90241
  - Model: CBR-9950
    - CB-950

- **Channel Master**
  - Ellenville, NY 12428
  - Model: CB6835

- **Colt Communications, Inc.**
  - 5725 North Central Ave.
  - Chicago, IL 60646
  - Model: Colt 290

- **Commando Communications Corp.**
  - P.O. Box 11071

- **Craig Corp.**
  - 921 West Artesia Blvd.
  - Compton, CA 90220
  - Models: L101
    - L301

- **Daiei Electric Co. Ltd.**
  - Model: CTI-90

- **Dynascant Corp., Cobra Communications Products**
  - 6460 W. Cortland Avenue
  - Chicago, IL 60635
  - Models: 21XLR
    - 89XLR
    - 138XLR
    - 139XLR

- **Farmingdale, NY 11735**
  - Model: 200

- **Chattanooga, TN 37401**
  - Models: CC4040
    - CC4086
Fanon/Courier Corp.
990 South Fair Oaks Ave.
Pasadena, CA 91105
Models: Fanfare 100F
Rebel 40
Conqueror 40D
Fanfare 880DF

Fukuyama Electronics Co. Ltd.
Model: FKCB001A

Gemtronics
356 South Blvd.
Lake City, SC 29560
Models: GTX-44
GTX-55

General Electric Co.
Electronics Park
Building 5
Syracuse, NY 13201
Models: 3-5801A
3-5811A
3-5811B
3-5812A
3-5819A
3-5821A
3-5825A

General Motors Corp.
Models: GM CBD-10A
GM CBD-20A

Hitachi Sales Corp. of America
401 W. Artesia Blvd.
Compton, CA 90220
Models: CM-2410H
CM-4850H

Hy-Gain Electronics Corp.
8601 Northeast Highway 6
Lincoln, NE 68507
Models: 682A
2700
2702
2703
2705
2679A

Ray Jefferson
Main and Cotton
Philadelphia, PA 19128
Model: CB-740

Kraeco Enterprises, Inc.
505 East Euclid Avenue
Compton, CA 90224
Models: KCB-4003
KCB-4005
KCB-4010
KCB-4020
KCB-4030
KCB-4040
KCB-4088
KCB-4075

Kris, Inc.
N144 W5660 Pioneer Rd.
Cedarburg, WI 53012
Models: XL-45
XL-23A

Kyodo Communications & Electronics
Models: Sawtron 770
Sawtron 790

Lafayette Radio Electronics Corp.
111 Jericho Turnpike
Syosset, NY 11791
Models: HB-640
HB-740
HB-940

Mars Radio Corp.
1420 Rodondo Beach Blvd.
Gardena, CA 90274
Models: CON-400
CON-450
M-368
M-374
M-375
M-379

The Panasonic Company
One Panasonic Way
Secaucus, NJ 07094
Models: CR-B4700EU
CR-B4701EU
CR-B4747EU
CR-B4748EU
CR-B4247EU
CR-B4737EU

Maxon Electronics Co., Ltd.
Sunrise Highway
Great River, NY 11739
Model: Maxon 40

Meishoh Electronics Co., Ltd.
Models: ME-400
ME-402
ME-401A

Meriton Electronics, Inc.
35 Oxford Drive
Moonachie, NJ 07074
Models: RS-5111

Midland International Corp.
P.O. Box 1903
Kansas City, MO 64171
Models: 63-240
77-853
77-857
77-882
77-883
77-888
77-955
77-825

Pathcom, Inc.
Communications Division
24049 S. Frampton Avenue
Harbor City, CA 90710
Models: 1000B
8041
2300CA

Pearce-Simpson, Inc.
4701 N.W. 77th Ave.
Miami, FL 33152
Models: Super Cougar 40
Super Cat 40
Super Tiger 40
Tiger 40

J. C. Penney Co.
Models: 6203
6237

Pioneer Electronics Corp.
1555 East Del Amo Boulevard
Carson, CA 90746
Models: GT-6600
GT-1100

President Electronics, Inc.
16691 Hale Avenue
Irvine, CA 92714
Models: Dwight D
Zachary T
Honest Abe
Teddy R

Radio Shack
2617 West 7th Street
Fort Worth, TX 76107
Models: 21-1520
21-1521
21-1524
21-1526
21-1542
21-1562
21-1580

RCA Corporation
Cherry Hill Offices
Bldg. 206-2
Camden, NJ 08101
Models: 14T270
14T304

Regency Electronics, Inc.
7707 Records Street
Indianapolis, IN 46226
Models: CR-430
CR-485
CR-486

Royce Electronics, Inc.
1746 Levee Rd.
North Kansas City, MO 64116
Models: 1-648
1-675

Sanyo Electric Co. Ltd.
1200 West Artesia Blvd.
Compton, CA 90220
Model: 1A-4000
SBE, Inc.
220 Airport Blvd.
Watsonville, CA 95076
Models: Brute 40
SBE-43CB

M.H. Scott Co. Inc.
Model: DAK MARK 3A

Sears, Roebuck & Co.
Models: 2862674
       2862676
       CM 2378SA
       CM 6000LA
       CM 6000LB
       CM 6100S
       CM 6200S
       TA 4501
       CB-4700S

Shakespeare Company
P.O. Box 246
Columbia, SC 29202
Model: GBS-240

Sharp Electronics Corp.
10 Keystone Place

Paramus, NJ 07652
Models: CB-2260
       CB-2460

Sony Corp. of America
9 West 57th Street
New York, NY 10019
Model: ICB-2500

Standard Communications Corp.
P.O. Box 92151
Los Angeles, CA 90009
Model: Horizon 2900

Superscope, Inc.
8150 Vineland Avenue
Sun Valley, CA 91352
Models: CB-340
       CB-140

Teaberry Electronics Corp.
6330 Castleplace Drive
Indianapolis, IN 46250
Models: 4001
       4002
       4006

Well, Inc.
Models: W-605
       W-705

Western Auto Supply Co.
2107 Grand Avenue
Kansas City, MO 64108
Models: CY14832A-87
       CYJ4854A-87
       CYJ4887A-87
       CYJ4862A-87

Toyota Motor Sales USA, Inc.
Models: 00860-00001
       00860-00020

Tran Sonic Industries, Inc.
P.O. Box 326
Lexington, MA 02173
Model: MCB-41

Uniden Corporation
Model: 805

RCA model 14T270 (below)
KRACO model KCB-4020 (above)

PACE model 8025

TRAM model D62
COURIER Spartan PLL40
New Solderless Coax Connector

FRED SHUNAMAN

IN THE "FIRST SIGNIFICANT IMPROVEMENT IN CB CONNECTORS since the now-standard PL-259 first became available in 1941," Bunker Ramo's RF division has introduced a new, no-solder RG-58 A/U termination that requires no special tools nor training for installation. The new, reusable Amphenol 83-58FCP (field crimp plug) is useful in both base and mobile installations and will be widely used in CB antennas, accessories, and all similar applications that use RG-58 A/U coax.

"Overheating and the damage it causes has been an often-avoidable consequence of soldering," points out Lee E. Eichenseer, vice president of Bunker Ramo's RF division. "Until now, there has not been a fast, simple means of making reliable connector-to-cable connections. Previously-available solderless connectors required special crimping tools and in many cases their performance was unsatisfactory.

"The new Amphenol 83-58FCP is the first connector to solve all these problems. It offers performance equal to that of standard Amphenol 83-ISP connectors, at the same price."

The specs are indeed equivalent to those of older types. The 83-58FCP has a frequency range of 0-300 MHz and a voltage peak of 500. The thermal limits are -67 to +300 degrees F (149 degrees C). The connectors have standard 1/8-24 threads and mate with regular UHF receptacles and adapters. They are not waterproof.

Not only are the new connectors offered for field use, but for manufactured equipment as well. They are being made in OEM (Original Equipment Manufacturer) quantities and, according to Mr. Eichenseer, will soon start appearing on CB products offered by various leading accessory, coaxial line and antenna manufacturers.

STEP 1—SLIDE THE OUTER FERRULE and the coupling nut onto the trimmed cable. Fan the braid out slightly by rotating the insulation. Slide the body assembly over the center conductor of the cable.

STEP 2—PUSH THE BODY ASSEMBLY over the cable so that the barrel goes over the dielectric but under the braid.

STEP 3—POSITION THE BODY so the braid flattens against its rear flange.

STEP 4—SLIDE THE COUPLING NUT up over the body assembly. Grasp the cable with one hand and push the ferrule forward with the other until it snaps into place.

STEP 5—SQUEEZE-CRIMP THE TIP of the center conductor only, with the pliers. If the connector is to be used again, the center contact may be soldered.

STEP 6—TRIM OFF THE TIP of the coax center conductor to flush it with the barrel.
CONTINUOUS PICTURE TUBE DEVELOPMENT. solid-state techniques and the integrated circuit have matured color TV receivers into truly reliable high-performance instruments. Manufacturers are bringing performance enhancements from the research lab into the latest products.

Progress has been made in such areas as electronic tuning with a system that the customer can install, in flesh tone correction by a system that changes only the flesh tones, and in power management by reduction in power consumption to the 100-watt level.

Though the frequency of servicing over a period has been statistically reduced, new problems have been created by the increasingly complex technology. New product innovations are slanted toward making set repair quicker and more systematic.

Broadcasters and set manufacturers are cooperating by recognizing the amazingly few weaknesses in the color broadcasting standards as originally devised. They have come up with an experimental system that subdues one of the deficiencies.

**Quasar**

Servicing solid-state receivers can be a very frustrating experience. Very special parts (triacs, integrated circuits) are encountered more and more. Unless you are a service organization specializing in one or two brands, these parts are probably not on your shelves. The trend makes it more difficult for the small service company to stay alive.

Quasar has made a two-front attack on the related problems of serviceability and reliability by the introduction of the "Super Module" and new integrated circuits.

Modular chassis are becoming universal, but Quasar's new Super Module retains the basic concept while making it more practical for a consumer product. The signal package is designed into a single replaceable panel. It performs most of the low-level signal and video processing in designated 1977 models. One module fits all applicable console, table, and varactor and nonvaractor tuned models. As much as 75 percent of the total chassis circuitry is on the module. It plugs into "Works in a Drawer" consoles, continuing the convenience of servicing from the front of the set. To free the panel from the chassis, remove six Phillips head screws and unplug the connectors.

The concept of replacing a large chunk of the receiver in a couple of minutes, with one large module instead of five or six smaller ones, means many fewer plug-and-socket connections. Each pin eliminated is one less spot where corroded or bent pins can cause trouble. Secondary controls mount directly on the module, reducing the number of connections as well as the wiring to the components.

Five integrated circuits in the IF, AFT, sound video, and color sections reduce the component count and the interconnections, again significantly contributing to reliability.

Figure 1 is the block diagram of the TS958/959 receivers. The first two stages of IF amplification and the AGC functions are contained in IC101. Based on sync amplitude, the IC generates automatic gain control signals for the 1st and 2nd IF stages. Both AGC polarities are available for forward-biased mechanical and reverse-biased varactor tuners.

Video processor IC301 accepts the output of the delay line and amplifies the signal. Brightness and sharpness corrections from the viewer controls are performed by IC301 which then feeds video driver transistor Q303. Noise inverter and sync separator stages in IC301 process the video and produce sync outputs for the horizontal AFC and vertical blocking oscillator circuits.

Color processor IC601 takes the video from emitter follower Q301. ACC (Automatic Color Control) helps maintain the chroma level selected by the color intensity control. Demodulated red, green, and blue color difference signals feed the bases of the video output transistors on the picture tube socket (the emitters of the output transistors are driven with the luminance signal from video driver Q303. Relatively few external components surround IC601.

The fifth monolithic device, IC201, takes the 4.5-MHz output from the sound detector and frequency demodulates the audio information. Two audio preamplifier stages increase the detected signal to the necessary level to drive the audio output transistor Q201.

An outstanding accomplishment of the designers of the two Super Module chassis is the low power consumption. The TS958 25-inch models consume 110 watts and the
TS959 19-inch sets use only 88 watts! The TS958 has a ferroresonant transformer for voltage regulation. Solid-state electronic regulation is featured in the TS959.

RCA receivers

RCA's new models have a number of innovative features. The top of the line direct-access tuning system with a calculator-type remote control keyboard is continued from last year. A dynamic flesh tone correction circuit has all the advantages of the older breed of color correction circuits but does not introduce the color distortion to colors removed in phase from the range of flesh tones. And a new luminance system optimizes peaking adjustment and produces a sharp picture with improved resolution.

Flesh correction circuits generally distort the color signal. Some colors may be shifted more than others and overall phase corrections inadvertently shift correct colors as well as incorrect hues. RCA's new system restricts the correction to colors closest to the +1...
vector near the flesh tones. Pure reds, blues, greens and normal flesh tones are minimally affected. Magenta and yellow-green color components are the "off flesh tones" that are shifted in phase toward the +I direction.

The dynamic correction system amplitude limits the chroma signal and compares it to the color subcarrier with a phase comparator. At this point the corrections exist for all colors not on the I axis. The phase comparator output is gated in a modulator to produce an output that is added to the color subcarrier before it is used by the Q and I demodulators. Figure 2 shows the action of the modulator. An offset bias applied to the modulator blocks the color corrections for colors outside the portion of the spectrum near flesh colors.

The block diagram of the Colortrak luminance system is shown in Fig. 3. A new integrated circuit produces a unique combination of functions including DC controlled phase corrected peaking, nonlinear transient compression in the white direction, and true black level clamping.

Optimum video transient response has one preshoot and a matching overshoot. Until now, the techniques used in TV receivers have not produced this ideal response. The new RCA sets use a transversal filter system that adds the signals from taps on an otherwise conventional delay line. A high-frequency signal is generated by adding the voltages from the taps closest to the ends of the delay line. By adding or subtracting the high-frequency signal from the output of the center tap, a variable peaked signal is formed while maintaining the ideal transient response.

As illustrated in Figure 3, one input to the peaking system is picked off the end of the delay line so only two actual taps are needed.

Zenith

New Zenith engineering advances for 1977 include a 19-inch diagonal Chromacolor 100-degree picture tube with the high-resolution EFL electron gun. The patented EFL gun focuses the electron beam with four instead of two electron lens elements, reducing spot size up to 60% compared with conventional guns. Improved picture sharpness and highlight detail is the result.

Also new is the Color Sentry automatic picture control system that ties together five electronic circuits to adjust the color, level, saturation, tint, color balance, and other parameters automatically. Color Sentry adjusts the picture when changing channels or any other time the program material is altered.

Up to 12 VHF or UHF channels are selected with the touch of a button with Zenith's Touch-Command. Married to the Electronic Video Guard tuning system, light pressure on the selected channel button brings in the viewer's choice without the distraction of passing through other stations. Electronic Video Guard tuning is in 85% of Zenith's color line which is the broadest use of electronic tuning in the industry.

Admiral

The Admiral Group of Rockwell International has a new line with a 100 percent deluxe solid-state television chassis. The ERA II Limited chassis features the industry's first optional electronic remote control that can be installed by the customer. The suggested retail price of the tuning system is $99.95. A select button on the control key keyboard enters the previously keyed-in channel number. Apparently this is Admiral's way of dealing with the problem of differentiating between one and two-digit channel entries.

Remote control can be added to five sets that have the built-in adapter. The remote transmitter is a duplicate of the tuning panel on the set. Installation consists of simply plugging the unit into the adapter after removing the cover plate on the front of the receiver. A 9-volt battery is needed in the transmitter and the system is in operation.

The ERA II Limited remote transmitter has a self-checking system that indicates battery condition. If the battery is strong, pressing the select button causes the least significant digit on the receiver's digital display to flash. When the battery is weak, the lower digit stops flashing.

The Admiral ERA II Limited chassis is offered in certain 23 and 25-inch consoles. They feature a negative black-matrix picture tube and ambient light sensing. Some 25-inch models are equipped with a digital channel read-out display.

All complex parts, including the power supply and the high-voltage deflection system, have been designed into efficient compact units. Each module is computer tested before installation at the factory.

General Electric

G-E also has an industry first for 1977. It is the 25-inch in-line gun picture tube used in the new YM chassis. One of the main features of the in-line picture tube is the simplified convergence procedure that G-E calls the "4 minute half hour."

The big G-E feature for 1977 is the VIR "Broadcast Controlled" color system. Details of the experimental Vertical Interval Reference system are described in the November, 1976, issue of Radio-Electronics. As explained in the article, General Electric uses tint and color controllers to compare the received luminance and color signals with information extracted from the VIR signal. The VIR reference was carefully designed to allow easy correction for transmission distortions. The signal format is such that it is relatively unaffected by the transmission errors that affect the color burst reference phase and amplitude.

Figure 4 is the schematic of the tint controller and demonstrates the principle of both the phase and amplitude correction mechanisms.

The R-Y color difference signal from the chroma/video module feeds the base circuit of emitter follower Q38. Limiting the bandwidth with R37 and C37 reduces noise. Transistor Q38 drives Q41 and Q42 when the signals are allowed through by the intervening diode gates. The lower switching diode-gate (diode D39) is controlled by the chroma reference interval and the upper gate (D40) by the black reference interval keying-pulses developed in the line recognizer portion of the system. The line-recognizer detects the nineteen picture line where the VIR signal is located and extracts the chroma and luminance levels from the appropriate signal segments.

When the cathode of either D39 or D40 is at ground potential, the corresponding anode is one Vx above ground and the bases of the transistors at ground potential are turned off. One Vx is insufficient to forward-bias the two-junction combination of the series diodes D42 or D41 and the base-emitter junction of transistor Q41 or Q42. Positive continued on page 84

FIG. 4—GENERAL ELECTRIC VIR TINT COMPARATOR.
LAST MONTH, PART I OF THIS ARTICLE introduced Oz and presented the schematic, foil patterns and began the construction details.

This month, the article concludes with the rest of the construction details and a description of how to connect and use Oz.

Mount the circuit board, keyboard and front panel in the housing that you've built or purchased. (Another tip: build a box, then cut it in halves or quarters or whatever, to make it open; don't cut the pieces first then try to put them together so that they form a closing box.) Then wire the connections between the circuit board and front panel (see Fig. 6). The caveats that I mentioned for PC board to keyboard wiring apply equally here. Leave enough slack so if something breaks, it can be fixed without dismantling the entire thing—but not so long that radiation problems pop up.

That makes twice that I've mentioned "radiation problems." Conceivably some of you are wondering what that means. It means this: as we've already noted, the output of the top-octave IC is a bunch of squarewaves. I might now mention that these squarewaves have very respectable rise and fall times. When we have fast rise and fall times, Mr. Fourier tells us that we will have high-frequency components. Mr. Maxwell tells us that high frequencies have a great ability for launching themselves into space in the form of electromagnetic radiation. The longer the wire that these high frequencies have to pass through, the greater will be their ability to leave the conductor entirely.

Similarly, long wires connected to the input of an amplifier will act as antennas that pick up the radiation and amplify it just as it would any other signal. The net result, in this case, is that notes work their way from the tone-generating circuitry to the amplifier and speaker without bothering to go through the keyboard. It becomes our task, then, either to prevent the radiation to whatever extent possible (short wires): or, failing that, to at least contain the radiation (shielding). Besides short wires there are a couple of other things that you can do to minimize—and probably eliminate entirely—this problem.

Use coax to make the connections indicated to be coax on the schematic and in the illustrations: specifically, the signal path from keyboard to front panel and back to the circuit board. It is not practical to use coax to make all the connections between the circuit board and keyboard; but fortunately it's not necessary either. Most keyboards have a solder lug attached to their frame somewhere. It's there for a purpose—the purpose being that if you ground it, the whole frame of the keyboard becomes one big shield to prevent radiation spilling out into places it shouldn't. Point "BB" on the PC board is ground and has been provided for just this purpose.

Just a little more, we're almost done. Special note also needs to be given to the pitch-bender panel. If you really want to, you can eliminate it entirely simply by not wiring it in (in which case it would not hurt to parallel an additional 47 pf across C20). But the ability to tremolo and "glide" single notes and chords is to a great extent what Oz is all about, so I will assume you are going to use it. Connect the pitch bender panel to the rest of the circuitry with small diameter coax, like RG-174/U. Coax is important here because we're working with an element that is capacitance sensitive and we want to minimize the effect that grabbing the connecting cord with your hand will have. The most important part of using coax is to make sure that the shield goes to the driving terminal of the clock circuit (the output of IC6-b—point "L" on the circuit board). Figure 7 shows the assembly of
the pitch bender panel.

You're finished. Now, before you hook up the batteries, why don't you check it through one more time? I always do (well, almost always).

OZ requires a 12-volt power source for proper operation. This can be a line-operated supply if you like, but be aware that a portion of the audio bus is connected to the supply lines, so power supply ripple rejection at the output is low. Make sure the supply is well filtered.

The best bet is batteries. "AA" size cells work well and even taking into account OZ's 60-mA typical current drain, a fresh set of quality batteries should last for a month or more of daily intermittent operation.

Testing

Turn the power switch on and observe that one and only one of the range indicating LED's lights. Change the octave switch and observe that the LED's follow the action of the switch.

Depress one of the keys and advance the level control until a tone is heard from the speaker. Test all the keys to make sure they produce a tone and that the pitch of the tone ascends as you work your way up the keyboard. Confirm that all positions of the octave switch work and that as the knob is advanced in a clockwise direction the pitch produced changes by octave steps.

While pressing keys, check the operation of the trigger status LED. With the trigger select switch in the step position, the LED should come on and stay on as long as any keys are down. In the pulse position of this switch, the LED should wink briefly (very briefly) every time a new key is pressed down.

Try connecting the tone-generation circuitry to an external amplifier by connecting the out jack J2 into the input of the outboard amplifier. Under these conditions make sure that signal is being supplied to the external amplifier and that no sound is coming from OZ's speaker.

To test OZ's input jack, you can either run an external program source into J1 (IN) or you can just jumper from OZ's output to its input. If, under these conditions, you still hear sound from the speaker, then the in jack is working.

Using OZ

There is little to be said about using OZ by itself. You simply turn it on, select the octave that you want to play in, set the level and tone controls to your taste and wait!

Interfacing to full synthesizer systems is only slightly more complicated, and more than anything else, requires that you have a firm grasp of exactly what OZ is and what it is supposed to do. Let's pin that down right now: OZ is a combination controller and polyphonic pitch source. Nothing more (it doesn't need to be anything more). In most synthesizer systems, it will be used in place of—or possibly in conjunction with—a keyboard/voltage controlled oscillator combination.

For example, Fig. 8 shows a "classic" synthesizer configuration. In a patching arrangement like this, there are really two different circuits with which we need to be concerned; the control circuit (triggers, control voltages, etc.) and the audio circuit (outputs of oscillators, inputs and outputs of voltage controlled filters and amplifiers, etc.). In the scheme illustrated, the control circuit can also be broken down into control
voltages and triggers. When a key is depressed on the keyboard, a voltage that tunes the VCO appears at the control-voltage output of the keyboard while a trigger appears at the trigger outputs. The trigger activates a function generator that produces a precisely preset time-varying voltage which in this case takes care of controlling the VCA to produce varying attack and decay times. The filter is stuck in there for harmonic control and it can be driven by control voltages from a variety of sources that we're not really concerned about.

The audio circuit is from the output of the VCO, through the VCF, through the VCA and finally to some sort of amplifier or recording device.

The part of this circuit that OZ replaces is only the keyboard and VCO and the control voltage path between the two. The rest of the patching configuration remains unchanged as shown in Fig. 9. The biggest difference is that by using OZ in place of the keyboard/VCO, you can now play full chords instead of the one-note-at-a-time restriction you previously had. Notice that the processed output can, if you wish, be routed back to OZ's amplifier. You do have one drawback—OZ's output is basically a squarewave rather than the multiplicity of waveforms that you have available from most VCO's, but, with appropriate filtering (at the VCF), that is not really as much of a pain as you would first think.

Many of you will (I hope) be using OZ with the Gnome Micro-Synthesizer (See Radio-Electronics, November and December, 1975, and January, 1976. issues.) That's great; that really is what it's designed for. But, you're going to have to make some very light-weight changes to the Gnome before OZ and the Gnome are compatible. Specifically, you're going to have to add an external audio input to the Gnome audio bus.

Figure 10 shows the audio bus of the Gnome as it was illustrated in the series of articles, and you can see that we have added a new resistor that connects to this bus.

How this new input is made available to OZ is largely a matter of personal preference. In mine, the other end of the new resistor connects direct to a piece of co-ax which then drops out of the seam at the back of the Gnome case and terminates in a miniature phone plug. (In fact, the resistor is inside the plug that terminates the coax—but I'm particularly lazy.) A word of advice—Do Not connect both ends of the shield of the coax—there should be one and only one ground connection between OZ and the Gnome and that should be part of the line that runs between the Gnome's output and OZ's input. Multiple grounds in different locations are like an insurance policy guaranteeing hum problems. Ground the shield of the coax mentioned above at the plug that terminates the line only.

From there, the interconnections are simplicity personified. Your "new" audio input cable connects to OZ's output. OZ's input goes to the Gnome output, and the trigger jacks of both tie together. These connections are shown in Fig. 11.

Playing the Gnome/OZ combination is not very different from playing either one separately—though they are very definitely a synergistic pair. As a matter of fact, only one point need really be brought to your attention. It's this: leave the Gnome's VCA sustain switch in the sustain position. If you don't, the Gnome's percussion "mute" function will drive you bananas.
Digital Clock For Your Car

Part II—Construction details on a useful automotive accessory that is rarely available as original equipment. A valuable aid for the trucker and road rally enthusiast that is simple and easy to build.

ROBERT C. ARP*

AN INTRODUCTION TO THIS USEFUL PROJECT appeared last month along with a detailed description of how the circuit works.

This month, the article concludes with the construction and calibration details, and foil patterns.

Construction

So that either type of display may be used with the clock, foil patterns for two double-sided display boards were prepared, as shown in Fig. 7. The two display boards are directly interchangeable with the main board. However, if incandescent displays are used, the copper trace on the main board that connects the emitter of Q3 to the 5-volt bus must be cut. Then, a jumper wire should be connected from the emitter of Q3 to V
t. The component layout for each display board is shown in Fig. 8.

Because the main board is separated from the display board, a variety of mounting schemes are possible: the display board may be dash mounted and connected to the remotely located main board with an edge connector, a display socket may be used to house the display (such as the Allied Electronics Cat. No. 658-1130) or the display board may be plugged into the main board, using Calectro Digi-Klips (Cat. No. J4-645).

Digi-Klips are available at local Calectro dealers and are manufactured by GC Electronics, Div. of Hydrometals Inc., 400 South Wyman St., Rockford, IL 61101.

If you decide to use Digi-Klips, they are available in packages of 24 pins. Although 26 connections are required between the main board and the display board, a package of 24 pins will do the job if two connections are made with wire soldered to each board. The Digi-Klips allow a rigid perpendicular connection between the display board and the main board. They may, however, be placed in the main board with one pin higher than the other to provide a tilted display. If you elect to use the Digi-Klips, a drilling guide is provided in Fig. 9.

Caution: When mounting the Digi-Klips, do not remove them from their plastic carriers; insert the entire row into the holes in the board at the desired angle, and solder each Digi-Klip to its bottom edge-connector pad provided on the board. Then remove the plastic carrier with a small screwdriver and solder each Digi-Klip to its edge connector pad provided on top of the board.

The numeric displays should not be soldered directly to the display board. Mount Calectro Cat. No. J4-635 IC terminal sockets or Molex IC Terminals to the board, in strips of seven or eight, by soldering them both to the top and bottom in-line pads. The two tabs to which the strips of terminals are connected should be closer to the center line of the in-line pads than should be the openings into which the display pins will be inserted. The tabs may be removed by bending them from side to side with long-nose pliers until they snap from the terminals. The displays are inserted into the terminals the same way they are inserted into an IC socket. IC sockets may be used on the main and display boards if the leads are long enough so the sockets can be soldered to all pads on the top of the main board and on the front of the display board that have copper traces running to them. In either case, the terminals or sockets that have copper traces running to them on the top and bottom of the board, must be soldered on both sides of the board.

If the incandescent displays are used, and if a viewing screen is used in front of them, the Radio Shack Cat. No. 276-116 photoresistor must be used (see parts list). It must be mounted to the display board so that the black chimney is in direct, flush contact with the viewing screen. This will prevent light from incandescent displays from reaching the photoresistor by reflection from the viewing screen.

If the LED displays are used, mount the photoresistor to the display board so that it stands as far away from the board as do the displays; the ambient light reaching the
photoresistor should be equal to the ambient light reaching the displays.

Display bezels may be purchased from Allied Electronics or Tracy Design Corp. The Allied No. 658-1240 bezel assembly has a red circularly polarized viewing screen and two mounting screws that can be used to mount it to the display board. The bezel may be separated from the display board with metal or fiber spacers. The No. 658-1240 bezel is actually made for five LED displays; the extra length is specified so that the bezel will accommodate the four displays plus a Clairex

The main printed-circuit board layout demands that these two integrated circuits be mounted from the bottom of the printed-circuit board, or that their pins be bent back 180 degrees before they are mounted from the top of the board. This form of mounting is necessary because of the layout of the MM5385 pins.

There need be no fear of bending back the pins of an integrated circuit as long as the bending is done correctly, with the proper tool. Use a pair of tweezers with sharp, rather than blunt, blades to grasp each pin exactly at the point it enters the package. The pins must be bent precisely at the package exit point and the edge of the tweezers must be held as close to the package as possible during the entire bending operation. This will prevent the pin from bending and breaking at the shoulder.

Bend the pins the full 180 degrees, one at a time, starting with a pin in the upper right hand corner of the IC and proceeding from pin-to-pin in a clockwise direction. You can practice on a bad IC, if one is at hand, or rejected IC's may be purchased at very little expense. Of course, a jig could be set up to bend all the pins on one side of an IC simultaneously.

After all the pins have been bent: check them for alignment before the IC is inserted into the terminals. The pins of all IC's should be carefully observed while the IC's are being inserted to make sure that all pins are actually inserted into the socket.

Next, without removing them from their plastic carriers, insert the first row of Digi-Klips from the top into the holes prepared in the main printed circuit board. Start at the end of the board where the heat sink is mounted. Look through both ends of the carrier and make sure that all center flats are touching the board. If they aren't, press them down with a very small screwdriver or other thin tool inserted into the appropriate end of

FIG. 7—DISPLAY BOARD foil-patterns shown half-size, a—front of incandescent display board; b—rear of same board; c—front of LED board; d—back of the LED board.

FIG. 8—DISPLAY BOARD COMPONENTS LAYOUT. Front view. Bezel mounting holes are centered by drawing a line across board, through the center of the displays. Inner bezel holes accommodate Allied 658-1240 bezel; outer ones the 658-1260. The photoresistor mounting holes are for Archer 276-116 or Clairex CL704L. Photoresistor solder pads are large enough to accommodate Clairex CLM54L. If Allied 658-1260 bezel is used.

FIG. 9—GUIDE FOR DIGI-KLIPS

CL704L or a Radio-Shack No. 276-116 photoresistor. The fit is tight; width of the bezel is 3.2 inches.

If a greater bezel width can be tolerated, the Allied No. 658-1260, with a bezel width of 3.7 inches, may be used. This bezel will allow the use of a Clairex CLM54L photoresistor. All three of the photoresistors mentioned above were used as intensity controls; the Clairex CLM54L gives the best results.

The foil pattern for the double-sided main board is shown in Fig. 10 and the component layout diagram is shown in Fig. 11. Sockets should be mounted to the main board for the integrated circuits and a socket should be mounted for the opto-isolator/coupler in the same manner as they are installed on the display board. However, a decision must be reached about the MM5385 and MM74C221 installation before the terminals are mounted.

FIG. 10—THE MAIN BOARD foil-patterns shown half-size, a shows component side; b shows bottom. If board is made a little larger, corner holes may be moved out slightly; they are now positioned for minimum-sized board.

FIG. 11—COMPONENTS LAYOUT, MAIN CIRCUIT BOARD. Top view. Capacitor C2 mounts to bottom of board. Unless pins of MM5385 and 74C221 are bent back 180 degrees (see text) they must also be mounted to bottom of board.
the carrier.

Turn the board over, and, while holding the carrier in place, solder the Klips to the bottom. Remove the holder with a screwdriver and solder the Klips to the top of the board. Do not solder to block the board opening in the center of the Digi-Klips.

Mount another row of Digi-Klips to the main printed circuit board in the same manner, then remove a Digi-Klip from a row in a fresh carrier. After mounting this Digi-Klip there will be an empty Digi-Klip hole at the end of the row of which the Q2 and Q3 are to be mounted. (A connection must be made to this Digi-Klip hole, through an ammeter, from the display board power bus during calibration.)

Enough room on the main board has been provided for at least two types of crystals: the cylindrical and flat-type crystals up to 1 inch in diameter.

The heat developed by the LM320K-5 voltage regulator during normal operation of the clock may be dissipated in three ways:

1. Because the input to the LM320K is negative to the center of the battery, the voltage regulator may be mounted directly to the metal body of cars with negative ground electrical systems. Choose a level spot, make a good electrical connection, and use any silicone-type of heat sink compound between the regulator and the body of the car.

2. If the clock is to be housed in a metal cabinet, the voltage regulator may be mounted to the cabinet for heat dissipation as long as the total outer metallic surface area of the cabinet is equal to or greater than 42 square inches.

3. A heat sink equivalent to the Wakefield 680-1.25-A (see parts list) may be mounted to the printed circuit board and used to cool the regulator. The size of this heat sink is the minimum that can be used. All heat sink areas and sizes discussed assumes the use of a silicone heat sink compound between the regulator and the heat sink. Separate the heat sink from the board about .05 inch. The heat sink must also be electrically insulated from the circuit board.

After mounting the LM320K-5 voltage regulator, a jumper must be soldered from the top of the main board at the LM320K-5 pin-2 hole to the bottom of the board at the same hole. This connects the 5-volt bus that runs along the top of the board to the output of the voltage regulator.

A heat sink must also be provided for Q2 and Q3. The Allied Electronics No. 957-2670, or a similar heat sink sold at Radio Shack stores as part of an assortment (Cat. No. 276-003), gives excellent results. The height of the heat sink is such that the top of the seven fins may be cut from the Radio Shack heat-sink. Use the three-fin sink for Q2 and the four-fin sink for Q3. Many similar heat sinks may be used; the thermal resistance should be about 24° centigrade-per-watt.

Switched functions may be mounted in any convenient location. Wires from the power supply and switches must be soldered to the appropriate pads. These pads are labeled A through M (see Fig. 11). Solder the wires necessary to provide the desired functions to their pads, and solder one in-line fuse from the accessory terminal of the ignition switch to pad “C” and an in-line fuse from the battery terminal of the ignition switch to pad “A.” A wire must also connect from pad “B” to the negative terminal of the battery. It may be fastened to any part of the car chassis from which paint and/or rust have been removed.

The display board may be mounted to the main board by placing the main board on a sheet of ½-inch styrofoam, then, by pushing the display board into the Digi-Klips. The Digi-Klips should be centered on the connector pads of the display board.

If possible, the main board-display board assembly should be mounted in its permanent location before calibration. (This may not be feasible unless it is to be mounted in a metal cabinet.) Some suggestions for dash mounts have been: in a panel that has been cut to fit an original automobile clock, in seat-belt control cutouts provided in certain models, or in any convenient location of the dash. After calibration, remove the display board by prying it out of the Digi-Klips with a screwdriver so that the last Digi-Klip may be mounted to the main board. This Digi-Klip connects the display board power bus.

Frequency calibration

The calibration instructions are the same for all supply voltage levels. However, with 6-volt systems the intensity of incandescent displays will be lower than that specified for 12-volt systems. Furthermore, the instructions for calibrating the display current are different for each type of display.

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Before applying voltage to the clock, set both the DARK ADJUST potentiometer (R9) and the BIAS potentiometer (R10) for minimum resistance (full counterclockwise). Set R10 for minimum resistance at initial turn-on to prevent the flow of excessive display current that could destroy the displays.

Connect both the battery supply lead and the accessory supply lead to the positive side of the automotive voltage supply (see Fig. 12) with the engine running and the alternator delivering the rated voltage to the electrical system. Insert a 0-500-mA meter between the display power Digi-Klip located at the extreme right on the main board and the display power edge connector located at the extreme right on the display board.

Connect the Vaux supply lead to the negative side of the automotive voltage supply. During the calibration, a cigarette lighter power plug may be used to interface the clock to the vehicle’s electrical system. If the clock is to be mounted in a truck, use the 12-volt supply rather than the 24-volt supply.

If the clock is to be mounted in a truck that does not have a 12-volt supply, change D1 to a 24-volt Zener diode. Also, lower the 24-volt supply to 12 or 15 volts using an LM340K-12 or LM340K-15 and appropriate heat sink. (If a much larger heat sink is used for the LM320K-5, a prerogulator may not be necessary.)

Depress the SLOW SET or the FAST SET switch momentarily, to reset the power failure indicator. LED1, on the main circuit board should begin flashing at some rate between 36 and 180 flashes-per-minute. This is because the input to the clock will be between 30 and 150 Hz.

The 50 Hz input to the MM5385 may be adjusted precisely by connecting a digital counter with a resolution of 0.1 Hz between the collector of Q1 and VSS. C4 may then be adjusted with a non-metallic tuning tool until the counter reads 50.0 Hz.

If a counter with at least this much accuracy cannot be used, the 50-Hz input to the clock can be adjusted by tuning C4 while observing the flashing rate of LED1. Because of the oscillator configuration, LED1 will flash only at rates between 36 and 180 per minute. This may be observed by rotating C4 counterclockwise 360 degrees while observing LED1. At some setting of C4, LED1 will be flashing at 180 flashes-per-minute. As the variable plate of C4 is turned counterclockwise past this setting, a point will be reached at which LED1 stops flashing at 180 flashes-per-minute, and, precisely at that point, begins to flash at a 60 flashes-per-minute rate. At this point, the input to the MM5385

FIG. 12—CALIBRATION SETUP. Install both fuses before turning on power; set both pots to minimum resistance. If displays show an illegal readout after clock has been reset at initial power turn-on, disconnect power and troubleshoot the display board and displays for shorts or opens. Voltages shown are for a 12-volt system, engine running.
Amplifier/Speaker Interface—a new concept

Until now, matching the performance of an amplifier to a speaker has been a hit or miss affair. A new speaker system handles this problem in a unique way.

LEN FELDMAN
CONTRIBUTING HI-FI EDITOR

The stringent rules regarding audio amplifier power ratings imposed nearly two years ago by the Federal Trade Commission has done much to standardize the way that hi-fi amplifier and receiver manufacturers advertise the power capabilities of their products. This standardization makes it easier to compare power specifications of different competing products. As most readers are aware, the law requires that amplifiers be rated on the basis of their ability to deliver continuous sinewave power into a specified resistive load. The amplifier does this over a specified band of audio frequencies at some specified value of total harmonic distortion.

That’s all well and good as far as amplifiers are concerned, but in some ways it has made the problem of interfacing amplifiers with loudspeakers even more difficult than before. For one thing, we do not normally listen to continuous sinewaves and the complex waveforms produced by program sources may be handled quite differently by an amplifier compared to the way that it handles continuous sinewave signals.

Generally, amplifiers can deliver greater short-term power when fed with transient, musical signals than it can when dealing with continuous sinewave signals. Also, since music encompasses a wide dynamic-range of signal amplitudes, it is quite possible to listen to a system at what seems like reasonable loudness levels only to have sudden peaks (that may be greater in amplitude by as much as 10 dB, 20 dB or even more) drive that amplifier well into clipping and audible distortion.

Conversely, a 50 watt-per-channel stereo amplifier, connected to a speaker system that is nominally rated to handle 50 watts, may well be driven to beyond its power handling capability by transient signal peaks. These transient peaks, though instantaneously well in excess of 50 watts, are not causing amplifier clipping because they are of such short duration. However, such power levels fed to the speaker, even intermittently, have been known to damage speakers or at least to drive their cones into non-linear operating regions with attendant distortion produced this time from the speakers themselves.

Finally, the so-called nominal power handling rating of the speaker may be based upon the capabilities of the more rugged woofer and not on the lower power-handling capacity of the midrange or high-frequency drivers in the system. Inordinately high levels of high-frequency energy fed to such speaker systems (either because of super-audible amplifier oscillation, non-typical frequency distribution of musical energy commonly encountered with some forms of electronic music or from amplifier-generated harmonics due to inadvertent amplifier clipping) have been known to “blow” many an expensive mid-range driver or tweeter while the woofer remains intact.

A New Speaker System

A new speaker system, developed by British Industries Company, includes many innovative design features that are intended to alert the user of these dangers and even to protect against their consequences. The B.I.C. Venturi Formula-7 Monitor Series speaker, shown in Fig. 1, is one of two new speakers that the company describes as a “speaker systems that think”. It is our purpose here not to evaluate the sound reproducing qualities of the Formula 7 (preferences in speaker sound remain, as always, largely subjective), but rather to discuss the electronic and indicator circuits that have been incorporated in this unusual system.

The upper section of the speaker system contains a control and indicator panel that is shown in Fig. 2. On the front panel to the left of the speaker system is an LED indicator light and just above it, on the sloped section of the front panel, is an AMPLIFIER CLIPPING control. A special test record is supplied with the Formula 7. It contains a continuous signal made up of repeating bursts of 300 Hz, three cycles of which repeats every 100 milliseconds, as shown in Fig. 3. Thus, the duty cycle of this signal is 10%, not unlike the duty cycle of the peak signals in typical music.

In actual use, the listener plays this record through his amplifier and the Formula 7, increasing the volume setting until a distinct change is heard in the character of the pulse-like sound. This point corresponds to amplifier clipping (assuming the amplifier is the limiting factor of the combination). The user then rotates the AMPLIFIER CLIPPING control until the indicator light just flashes. Once the setting is made, the LED will light whenever the amplifier is driven to clipping levels by other musical program sources. The nice thing
FIG. 2—CONTROL PANEL of the Formula-7 speaker system.

FIG. 3—TEST RECORD supplied with Formula-7 provides three cycles of 300-Hz signal every 100 milliseconds.

FIG. 4—INTERFACE AND CROSSOVER circuitry of the Formula-7 speaker system.

stood by referring to the complete schematic diagram of the Formula 7 shown in Fig. 4. Audio signals delivered by the power amplifier to J1-1 and J1-2 are fed to the usual speaker drivers and crossover network components, with which we will deal shortly. In addition, the audio signal is rectified by diode D2 and applied to the positive input of IC3-a. The trigger point of LED9 is determined by picking off a portion of the fixed 5-volts DC from a built-in power supply via R23 and applying it to the negative input of IC3-a. Thus, the trigger point of LED9 can be adjusted by means of R23 (the front panel AMPLIFIER CLIPPING control previously referred to) over a wide range corresponding to amplifier outputs up to around 125 watts-per-channel.

Now, suppose the user owns an amplifier that exceeds the safe power-handling capability of the speaker system. In that case, auditioning of the test record might result in speaker distortion or damage well before the amplifier begins to clip. To prevent overdriving of the speakers in this manner, additional indicator and protection devices have been incorporated.

Referring now to the portion of Fig. 4 that shows the four drivers you will note that there are three circuit breakers (CB1, CB2 and CB3) wired in series...
with each driver. Each circuit breaker is rated at an appropriate value from 2.0 amperes for CB1 to 0.6 amperes for CB3. When the current exceeds the circuit-breaker rating, the appropriate circuit breaker opens and an indicator light wired across that circuit-breaker flashes in accordance to the signal current that then flows through it.

The three circuit breakers are located on the sloped portion of the front panel and are manually resettable. Dividing up this indicating system instead of having a single circuit-breaker permits the user to determine which driver elements of the unit are being over-driven and even warns of trouble in the event of super-audible high-frequency oscillation that, while inaudible, might easily destroy the high frequency driver elements.

One other aspect of the Formula 7 is worth mentioning before we leave the speaker-driver section of the schematic. Thermistor R37 is the key component in what B.I.C. calls its "dynamic tonal compensation" system. When front-panel mounted potentiometer R38 (which varies the signal level fed to midrange and tweeter) is rotated fully counterclockwise, switch S2 is thrown to the position shown in Fig. 4. The value of thermistor R37 varies in accordance with the average current flowing through it. R37 is chosen so that when loud levels are reproduced, its resistance is very low (due to the heating effect of the thermistor) while at lower listening levels, its resistance increases and reduces the mid- and upper-mid frequency sound levels. This amounts to the same thing as boosting (relatively) the bass response and is designed to compensate for the well known Fletcher-Munson loudness effect in human hearing. This effect is often compensated for by the so-called loudness controls on amplifiers and receivers. The advantage in this approach is that the resultant response can be directly related to actual sound-pressure levels and not to arbitrary settings of a master volume control that cannot take into account such variables as program source signal levels, speaker efficiency or amplifier gain.

**SPL indication**

The audio signal from the amplifier is rectified and filtered by capacitor C3 and diode D1. (See Fig. 4.) This varying DC voltage is applied to the negative inputs of eleven comparator circuits (four in IC1, four in IC2 and the three remaining in IC3). The positive inputs of these comparators are supplied with progressively lower and lower DC voltages that are determined by the string of series resistors R4 through R18. Thus, when the lowest-level signals are applied, LED2 will light, while increasingly stronger input signals will trigger LED1, LED10 and so forth up to LED1 which flashes when the greatest audio signal levels are applied to the system.

The row of LED's on the front panel are calibrated in approximately 4-DB increments all the way from 75-dB SPL (Sound Pressure Level) to 117-dB SPL. A chart inscribed right on the system's front panel (see Fig. 5) permits the user to translate the LED indications into actual sound pressure levels, measured at specific distances on-axis from the speaker system. Alternatively, the user can read the average sound pressure level attained in listening rooms varying in cubic volume from 1000 cubic feet to several others inscribed over the various LED's below the chart. Observing the flashing lights while listening to music for extended periods may appeal to some hi-fi fans while others may wish to just listen and not look, having established desired levels. For this reason, the Formula 7 has a switch (SW2 in the schematic diagram) that permits you to turn off the sound pressure level indicators at will.

Much of what B.I.C. has built into the Formula 7 is fairly simple in terms of circuit complexity, and the information it provides could be obtained if you owned a good SPL meter and an accurate audio power meter. But few...
Radio-Electronics Tests
Soundcraftsmen PE2217 Equalizer

LEN FELDMAN
CONTRIBUTING HI-FI EDITOR

LAST MONTH WE EXAMINED A RELATIVELY INEXPENSIVE FIVE-BAND ADD-ON GRAPHIC EQUALIZER SUPPLIED IN KIT-FORM BY Heath. The add-on type of graphic equalizer lends itself particularly well to application in Hi-Fi systems that have already been assembled and which may contain an all-in-one receiver, or an integrated amplifier, or even a separate preamplifier and basic power amplifier. The audio enthusiast who chooses to use completely separate components may find it a bit cumbersome to add yet a fourth major component such as an equalizer to what is already a growing number of individual components. After all, one's shelf or cabinet is only so large. Accordingly, Soundcraftsmen decided that it would be a good idea to combine one of their better graphic equalizer circuits with a highly flexible and well designed preamplifier/control unit, and that is just what their PE2217 adds up to. The front panel, shown in Fig. 1, looks very much like the front panel of that company's model RP2212 equalizer that sells for around $370.00 alone. So, if you are in the market for a preamp and a graphic equalizer you might say that the preamp section of the PE2217 only adds around $160.00 or so above the cost of a 10-band graphic equalizer. Viewed from that point, that's quite a bargain, especially when you consider what the preamp section of the PE2217 can do all by itself.

Preamplifier related controls are located along the bottom of the front panel and include a POWER ON/OFF pushbutton with an indicator light just above it, a MASTER VOLUME control, a BALANCE control, two STEREO HEADPHONE jacks (one disables the line outputs at the rear when a plug is inserted, the other does not) and a TAPE 2 input and output jack pair that provides front-panel connection of a second tape deck and are wired in parallel with similarly labeled phono-tip jacks at the rear. The rightmost bank of pushbuttons take care of program source selection. Six more pushbuttons at the center handle the Tape 1 and Tape 2 monitor circuits, as well as the dubbing options from Tape 1 to Tape 2 or vice versa. The two remaining buttons in this bank permit instant comparison of equalized or unequalized sound and also enable you to apply pre-equalized signals to your tape deck inputs—a very handy feature for adventurist recordists, and one generally not found on preamps or even on separate equalizers.

The upper section of the panel resembles the front of the previously referred to RP2212 equalizer. Two identically calibrated groups of ten slide controls handle octave-by-octave response adjustments for left and right stereo channels. Between these banks of slide controls are two additional gain controls, flanked by upper and lower LED indicator lights for each channel. These controls are active only when the equalization circuits are selected and are used to re-establish identical overall sound levels with those of the unequalized signals, when comparisons are to be made. This is necessary because even if a few of the slide controls are in the “boost” region, overall apparent reproduced levels will be louder than would be the case if all sliders were set to their mid-positions, where the response of the system is flat. The indicator lights flash when program material is channeled through the equalizer, and equal flashing of upper and lower lights indicates correct setting of the gain adjust controls.

The rear panel of the PE2217 is shown in Fig. 2. The two pairs of low-level phono inputs at the left are isolated from the high level inputs and tape output at center, a layout detail that permits the most direct and shortest wiring from these inputs to the critical, low-level phono preamp section visible at the rear right of the chassis in Fig. 3. A chassis ground terminal is located near the phono input jacks, and, at the extreme right of the rear panel, there are six convenience AC outlets (four switched, two unswitched) for connection of other components. A line fuseholder is located at the lower right, adjacent to the power cord.

Circuit layout
In examining the layout and construction of the PE2217 (Fig. 3), we noted that two identical circuit boards are used for the two channels of equalization. LC filters, using discrete inductances, are employed for each octave filter circuit and, if like us, you can only count six “cans” on each board, rest assured that there are four more per board, resting securely on the surface of the PC boards since they are the ones that handle the lower octaves and are therefore too large in size for standard encapsulation. The power supply at the left-rear is in its own fully shielded compartment, as is the phono preamp section at the right rear. Separate power-supply voltages are provided for the preamplifier circuits and the equalizer circuits in this design. All switches (obscured by the equalizer boards) have gold plated contacts for lowest possible resistance, and the PC boards themselves are of military-grade G-10 glass-epoxy construction. In examining the circuit boards closely, we noted the use of several low-noise carbon-film resistors in critical low-level circuits. An idea of the flexibility of this combination unit can be gained by examining the

MANUFACTURER'S SPECIFICATIONS:
Frequency Response: High Level Inputs: 5 Hz to 100 kHz, ± 0.25 dB; Phono Inputs: RIAA ± 0.5 dB. Harmonic Distortion: 0.05% at 1 volt. Signal-to-Noise: High Level Inputs: 100 dB; Phono Inputs: 84 dB (ref. 10-mV input); Equalizer Section: 90 dB below 1 volt Overall Maximum Gain: High Level Inputs: 21 dB; Phono Inputs: 63 dB. Phono Overload Level: 105 mV. Output Impedance: 600 ohms. Octave Control Range: ± 12 dB minimum, Octave Centers: 30 Hz, 60 Hz, 120 Hz, 240 Hz, 480 Hz, 960 Hz, 1920 Hz, 3840 Hz, 7680 Hz, 15,360 Hz. Dimensions: (In supplied wood cabinet): 20 wide by 7¼ high by 11½ inches deep. Panel Dimensions: 5¼ high by 18 inches wide. Also available for rack-mounting, less cabinet at same price. Suggested Retail Price: $529.50 (cabinet or rack panel mount).
signal flow. The master volume control is located just ahead of the line amplifier so that best signal-to-noise ratios are maintained. The equalizer section contains those level-matching gain controls that were mentioned earlier and insure against possible overload or distortion within the equalizer section itself. The position of the equalizer section and the switching arrangement permits the user to equalize both the signals at the tape-output jacks and at the line-output jacks, and either of these outputs can be fed with an unequalized signal while the other delivers an equalized signal. Tape dubbing can be accomplished even while listening to another program source such as phone or tuner.

Laboratory measurements

Our measurements of the PE2217 necessarily consisted of two separate groups. First, we checked the action of the equalizer controls themselves. We did this by "plotting" the response obtained at the maximum boost and maximum-cut position of each of the ten octave-controls of a single channel and storing the combined results on the storage-scope of our spectrum analyzer. The spectrum analyzer had been swept from 20 Hz to 20 kHz for each of the twenty-one (including flat response) response curves. The results are shown in the photo of Fig. 4. Maximum boost range to maximum-cut of each control measured approximately 28 dB, or ±14 dB from the mid-position of each slide control. Peaks are quite evenly spaced which, in this logarithmically plotted sweep means that they were exactly an octave apart, as claimed.

To check the interaction of adjacent controls, we set up an arbitrary response curve that might typically be required in a less-than-perfect component/acoustic situation. The ten control knobs were set as shown in the close-up photo of Fig. 5 and a response curve was plotted for these settings. The response obtained is shown in Fig. 6. It corresponds quite closely with the settings of the ten knobs, with the exception of the region of the eighth-octave control that had called for a slight rise in frequency and was offset by the dip position of the ninth-octave adjacent control and the flat setting of the adjacent seventh-octave slider. The likelihood of ever needing such a narrow peak in the response of a real system is very slim.

Measurements related strictly to preamplifier performance of the PE2217 are listed in Table 1 and generally ranged from good to excellent. The sole exception was the action of the slide controls themselves, and our criticism of these will be found in our summary comments following our overall product analysis in Table II.

Use and listening tests

In general, the PE2217 handles well. Program-selection pushbuttons and the six centrally located pushbuttons that handle the tape functions and equalization selection are all interlocked within each grouping to prevent accidental pushing of two buttons at once. The preamplifier's output was at all times adequate for driving even the most insensitive (low gain) of basic power amplifiers.

We found that it is extremely important to be guided by those flashing gain-adjust test lights when the equalizer is "in-circuit", for to ignore their warning is to invite higher levels of overload and distortion. While it is possible to disable the indicator lights at the push of a button (the presumption being that you will tire of watching them flash after proper levels have been set up) we would suggest that they be allowed to do their job at all times, or at least when you switch from one program source to another.

While much of the flexibility of the PE2217 is related to tape recording activities, we feel that even the audio enthusiasts who does not own a tape deck but wants to own a good preamp and an effective graphic equalizer will find that Soundcraftsmen's PE2217 merits consideration on both counts.

Accessories

Although, in theory, one setting of the equalizer controls should suffice for all times (as far as one's system and room acoustics are concerned), there are times when particular program sources may call for different settings of the twenty octave controls. After all, recording engineers apply "equalization" at the studio, too, and you may not always agree with their ideas of tonal balance. For this reason, and because some over-enthusiastic visiting sound-buff friend may not be able to resist the temptation to try his or her
TABLE II
RADIO-ELECTRONICS PRODUCT TEST REPORT
Manufacturer: Soundcraftsmen  Model: PE-2217

OVERALL PRODUCT ANALYSIS

<table>
<thead>
<tr>
<th>Feature</th>
<th>Rating</th>
</tr>
</thead>
<tbody>
<tr>
<td>Retail price</td>
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</tr>
<tr>
<td>Price category</td>
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<td>Price/performance ratio</td>
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</tr>
<tr>
<td>Styling and appearance</td>
<td>Excellent</td>
</tr>
<tr>
<td>Sound quality</td>
<td>Good</td>
</tr>
<tr>
<td>Mechanical performance</td>
<td>Good</td>
</tr>
</tbody>
</table>

Comments: Combining the functions of a ten-band graphic equalizer and a preamplifier/control unit in a single chassis makes a lot of sense to us, particularly since the increased popularity of graphic equalization, or "room voicing," is primarily one that has arisen from audio enthusiasts who choose to assemble their systems from separate components. Octave-by-octave tone adjustment is, in our opinion, the minimum segmentation needed to do a precise compensating job and, while not as precise as third-octave equalization, it is more effective than what can be done with the lower-cost five-segment units that are available. Although Soundcraftsmen suggests that a good job of equalization of one's system can be done using their supplied test record and one's own ears, we found that a much better job can be done using a sound-level meter—even if it has to be one of the very inexpensive ones that are now available. The ear is too easily deceived, even when pink noise bands are used as the test signal.

As for the preamplifier/control sections of the PE2217, they measured right up there with the best of the separate preamps around. Audible distortion is not likely to be a problem, unless generated by other components in your system, and the flexibility of the switching facilities (especially those related to tape recording, including the pre-equalization possibilities) make the unit a self-contained "patch panel" that will be appreciated by those who own a pair of tape decks and are constantly finding it necessary to plug and unplug cables to achieve required results.

We wish that Soundcraftsmen had supplied detents, or click-stops for the flat settings of each octave control, and that the overload characteristics of the unit (when in the equalization mode) were not so critically dependent upon correct "zero gain" adjustment of those gain controls, but when care is taken, the preamp behaves like a good preamp should, and the equalizer does its expected job much as a separate one would.

Bigston BSD-300 Cassette Deck

WHEN FRONT-LOADING CASSETTE DECKS WERE first introduced for home high-fidelity systems, they were fairly expensive gadgets selling for $300.00 or more if they included Dolby noise reduction circuitry. The public's acceptance of the front-loading cassette configuration has prompted a great number of manufacturers to produce lower-cost decks that offer the same advantages of "stackability" with other high-fidelity components. One of the lowest priced of these that we have had an opportunity to test and evaluate is the Bigston BSD-300. Bigston Corporation is perhaps not too well known here for its own products but, according to information we obtained from the company's U.S. headquarters, that company has been and continues to be a large supplier of "private label" or custom manufactured tape recording products for retail chains.

As for the BSD-300 itself, the photo of Fig. 1 shows its front panel layout. Cassette are inserted, tape forward, into the sloped compartment at the left. This compartment is illuminated for better visibility inside when pressing the single RECORD button. Most others require simultaneous hold-down of both the RECORD and PLAY buttons. The dual-button arrangement is normally a precautionary measure to prevent accidentally erasing a precious recording and is a carry-over from the early days of cassette decks. We have often thought that the bother of having to carefully depress two buttons at once to get into the record mode more than offsets the safety feature and apparently Bigston thought so too.

The REWIND/REVIEW and FAST FORWARD/CUE levers serve a double purpose. Used alone, they perform the normal fast rewind or fast forward functions. Used when the PLAY button is depressed, they permit fast wind in either direction as long as the button is held down manually while some sound is heard through your system, thereby enabling rapid location of a specific point in a given cassette tape. The only other units we know of that have this capability (aside from low-fi portable dictating cassette decks) are those offered by Kenwood and sell for quite a bit more than this Bigston model.

The right section of the panel is equipped with a pair of small VU level meters, calibrated from -20 dB to +5 dB and with a Dolby level mark. A RECORD indicator lamp is located between the meters. Below, are three buttons that select tape bias, equalization and Dolby circuitry. The bias and EQ buttons work independently, thereby afford-

FOR MANUFACTURER'S LITERATURE, CIRCLE 88 ON FREE INFORMATION CARD

POWER: 110 Volts, 50/60 Hz, 60 W

MANUFACTURER'S PUBLISHED SPECIFICATIONS:

Frequency Response: Standard Tape: 30 Hz to 13,000 Hz; CrO2 Tape: 30 Hz to 15,000 Hz. Wow-And-Flutter: 0.08% W/RMS. S/N Ratio: Without Dolby: 48 dB; With Dolby: 56 dBA. Output Level: Line: 0.58 volts; Phones: 0.3 mV/8 ohms. Power Consumption: Approx. 10 watts. Dimensions: 14¾ inches wide x 5½ inches high x 10½ inches deep. Weight: 11¾ lbs. Suggested Retail Price: $169.95

FEBRUARY 1977

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Advanced Electronics

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TABLE I
RADIO-ELECTRONICS PRODUCT TEST REPORT

Manufacturer: Bigston Model: BSD-300

CASSETTE TAPE DECK MEASUREMENTS

<table>
<thead>
<tr>
<th>FREQUENCY RESPONSE MEASUREMENTS</th>
<th>R-E Measurements</th>
<th>R-E Evaluation</th>
</tr>
</thead>
<tbody>
<tr>
<td>Frequency response, standard tape (kHz-kHz = dB)</td>
<td>36-10, ±3 dB</td>
<td>Poor</td>
</tr>
<tr>
<td>Frequency response, CRO2 tape (kHz-kHz = dB)</td>
<td>32-8, ±3 dB</td>
<td>Poor</td>
</tr>
</tbody>
</table>

DISTORTION MEASUREMENTS (RECORD/PLAY)

<table>
<thead>
<tr>
<th>Standard tape, Dolby off (dB)</th>
<th>Excellent</th>
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</thead>
<tbody>
<tr>
<td>CRO2 tape, Dolby off (dB)</td>
<td>Excellent</td>
</tr>
<tr>
<td>CRO2 tape, Dolby on (dB)</td>
<td>Excellent</td>
</tr>
</tbody>
</table>

MECHANICAL PERFORMANCE MEASUREMENTS

<table>
<thead>
<tr>
<th>Wow and flutter (%)</th>
<th>Wind and rewind time, C-60 (seconds)</th>
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</thead>
<tbody>
<tr>
<td>0.06 (0.09 RMS)</td>
<td>75</td>
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COMPONENT MATCHING CHARACTERISTICS

<table>
<thead>
<tr>
<th>Microphone input sensitivity (mV)</th>
<th>Line input sensitivity (mV)</th>
<th>Line output level (mV)</th>
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<tbody>
<tr>
<td>0.17</td>
<td>120</td>
<td>500</td>
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TRANSPORT MECHANISM EVALUATION

<table>
<thead>
<tr>
<th>Action of transport controls</th>
<th>Absence of mechanical noise</th>
<th>Tape head accessibility</th>
<th>Construction and internal layout</th>
<th>Evaluation of extra features, if any</th>
</tr>
</thead>
<tbody>
<tr>
<td>Very good</td>
<td>Good</td>
<td>Average</td>
<td>Good</td>
<td>Very good</td>
</tr>
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CONTROL EVALUATION

<table>
<thead>
<tr>
<th>Level indicator(s)</th>
<th>Level control action</th>
<th>Adequacy of controls</th>
<th>Evaluation of extra controls</th>
</tr>
</thead>
<tbody>
<tr>
<td>Fair</td>
<td>Good</td>
<td>Good</td>
<td></td>
</tr>
</tbody>
</table>

OVERALL TAPE DECK PERFORMANCE RATING

Good

TABLE II
RADIO-ELECTRONICS PRODUCT TEST REPORT

Manufacturer: Bigston Model: BSD-300

OVERALL PRODUCT ANALYSIS

<table>
<thead>
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<th>Retail price</th>
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<td>Price category</td>
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<tr>
<td>Price/performance ratio</td>
<td>Good</td>
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<tr>
<td>Styling and appearance</td>
<td>Excellent</td>
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<tr>
<td>Sound quality</td>
<td>Good</td>
</tr>
<tr>
<td>Mechanical performance</td>
<td>Excellent</td>
</tr>
</tbody>
</table>

Comments: If our overall evaluation of this low-cost deck seems particularly negative (considering its low price and the fact that it does incorporate Dolby noise reduction at that price) it is because there is so much that is good about this unit that we were particularly disappointed at its very poor record/playback frequency response. If that response could be tweaked up to, say, 14 kHz or 15 kHz when used with top-grade tapes (and we have a feeling that it could be, very easily), the other parameters such as signal-to-noise, headroom, harmonic distortion and wow-and-flutter are all so superior to anything found in this price category that the unit would be a most outstanding buy. We suspect that in our sample, the cue/review feature was not operating properly for, instead of rewinding when this button was depressed (with play button depressed too), the tape simply came to a standstill. The fast forward cueing feature worked well, however, and is a handy added feature that permits one to reach a desired point in the tape without having to go through the stop mode. Record-level meters are rather on the small side and in our sample, there was a 1.5 dB discrepancy between left and right meters when input levels to both channels were identical and level controls were at maximum. The manner in which the front-loading cassette compartment is arranged makes for fool-proof insertion of cassettes and a ledge up front forestalls cassettes from popping out on to the floor when the eject button is pushed (a failing on some other front-loaders we have seen). In summary, then, the Bigston BSD-300 is in many ways equal to cassette decks costing much more—but that limited frequency response range prevents us from concluding with overall raves for this product.
Last month, in Part I of this article, we discussed the different types of analog voltmeters that are currently available.

This month we will cover the various features of these voltmeters and the meaning of their specifications.

The ammeter circuit

The most common form of ammeter on TVM's is the simple shunt. The TVM offers the advantage of high gain and low insertion loss. Since the VTVM has an insertion loss of 1 to 1.5 volt for a shunt, ammeters are not normally included in VTVM's. When shunt ammeters are used, the DC and AC ammeters are very similar. The change from AC to DC consist of selecting the correct shunt to be placed across the voltmeter.

Even though the shunt insertion loss can be kept low with a meter having sensitive ranges (0.1 volt, for example), the shunt resistance becomes large for the sensitive current ranges. For example, a meter with 0.1 volt sensitivity has 10,000 ohms insertion resistance for the 10-μA full-scale range. On the same meter, the 1-ampere range has a 0.1-ohm shunt. Due to other series resistances, the 1-ampere range might have an insertion resistance of 0.2 ohms or more.

The TVM ammeters appear in two other, less frequently used forms. First, the scheme shown in Fig. 9-a. In this circuit, an operational amplifier is used as a current-to-voltage converter. The unknown current is introduced to the summing junction of the operational amplifier. This current is nullled by an equal amplitude but opposite polarity current developed by the output voltage of the operational amplifier across the feedback resistor. As this current is flowing through the feedback resistor, the voltage drop across the resistor is directly proportional to the resistance and the unknown current. The output voltage of the operational amplifier (the voltage across the resistor) is then measured with the TVM voltmeter. Since there is no voltage drop at the input to the null amplifier, this technique offers the advantage of essentially zero input impedance.

Figure 9-b shows schematically the technique developed and patented by Weston for current measurement in their model 670 TVM. This TVM features in-circuit current measurements. Measuring the current does not require opening the circuit, a considerable inconvenience with other ammeters. The circuit relies on the inherent resistance in the circuit conductors. The voltage drop across the circuit conductors is reduced to zero by a current equal in amplitude and opposite in polarity to the unknown one. This “bucking current” is maintained at exactly the correct value by the null amplifier.

This technique requires a special pair of two-terminal probes and is especially useful for measuring currents in printed circuit board foils. Both these latter techniques are applicable only where the currents being measured are low enough so the measuring circuit can produce an equal and opposite current, and where DC currents are being measured.

The TVM ohmmeter circuit

The simplest and most frequently used ohmmeter circuit for TVM's is identical to the series-resistance type of the VTVM. The newer and more elegant TVMs fea-
ature a dual-range ohmmeter circuit which has a choice of two open-circuit voltages. A low-voltage range is provided so semiconductor junctions (both germanium and silicon) are not forward-biased. The maximum open circuit voltage of such ohmmeters is about 0.085. The high-voltage range of these ohmmeters is usually 1.5 volts and can easily forward-bias a semiconductor junction.

These two circuits are constructed in two different ways. First and simplest, the low voltage is obtained by reducing the 1.5-volt ohmmeter battery to the low voltage by a voltagedivider circuit. This effectiveld yields a low-voltage source with a 10-ohm series impedance. Alternatively, the test signal is derived by applying a constant current source to a resistor as shown in Fig. 10. In this circuit, the constant current source is applied to the basic range resistor (say 10 ohms). The value of the current flowing through this resistor determines the maximum open circuit voltage of the ohmmeter. If the 10-ohm resistance is shunted with an unknown resistance, the voltage drop across the resistor is determined by the parallel value of the two resistors. A two-range constant-current source is used for high/low measurements. The current is the same as the maximum open circuit output voltage of the ohmmeter.

The AC voltmeter

A specialized form of electronic analog voltmeter amplifies and measures only AC voltages. The AC voltmeter finds special application in laboratory work. It is also part of the standard test equipment used in audio service work. The AC voltmeter is characterized by high sensitivity (compared to the ordinary VTVM or TVM), a voltage-only function, extended frequency response, alternate calibration in dB (usually dBm) and of course AC-only response.

The major difference between the AC voltmeter and the other types discussed lies in the amplifier. As DC voltages are not being measured, they need not be amplified. This allows an amplifier design with considerably more gain but without the worries of DC drift which accompanies a high gain DC amplifier.

The input attenuator of the AC voltmeter is similar to the attenuator described for that AC portion of the TVM. On AC voltmeters with extended frequency response, the input attenuator is often divided into two portions. The first portion is a compensated attenuator providing decade attenuation. Finer attenuation follows an input buffer stage. This is a low-impedance attenuator and does not need compensation. A specialized form of AC voltmeter offers linear response to logarithmically changing signals. That is, the voltmeter is calibrated linearly in dB and logarithmically in voltage. Such a meter has an additional logarithmic converter placed between the rectifier and the meter circuit. The whole amplifier design is required for the amplifiers used for the AC voltmeter.

VTVM AC and DC voltage ranges

The voltage ranges of the VTVM are most common in 1.5—5—15 sequence. This is a modification of the 1—3—10 sequence and is chosen because the steps represent 10 dB voltage changes. This is convenient when working with AC signals, and most VTVM's have a dB scale for this purpose. The 0 dB level is 0.774 volts (1 mW on 600 ohms). Figure 11 shows a typical VTVM scale. Occasionally VTVM's are found with a 1—3—10 sequence.

Normally the most sensitive range of the VTVM is 1.5 volts. The 1.5-volt AC range may require special calibration to correct for the nonlinearities of the rectifier when operated at this low input level. The 1.5—5—15 sequence also permits the user to make all measurements in the upper two-thirds of the scale, where the meter is most accurate.

The full scale of 150 volts AC is also handy in countries where a nominal 115 to 120 VAC power-line voltage is used, as measurements of the line voltage are made in the upper third of the meter scale. Needless to say, this advantage disappears when the common line voltage is 200 to 240 VAC. VTVM’s have an upper voltage limit of 1,000 to 3,000.

Input impedance

Input impedance for the VTVM is generally specified separately for the AC and DC modes. The general purpose VTVM has an 11-megohm input impedance. Although the 11-megohm input is the most common, other input impedances are used. There is little or no significant difference between the VTVM with 11 megohms and the VTVM with 13-megohms input impedance. There is, however, a significant advantage with the VTVM that offers an input impedance in the hundreds of megohms. Such a high input impedance costs money!

If the VTVM is to be used with a high-voltage multiplier probe, then its input impedance, with probes disconnected, must be some standard value. The accuracy of the input impedance is of no consequence unless a multiplier probe is being used. Often the accuracy of input impedance is not specified, or a simple note stating the attenuator is composed of 1% tolerance resistors is given. Remember, a VTVM which has 11 megohms input impedance with one megohm in the probe must be calibrated for that series impedance. The direct input to the VTVM without the probe finds the calibration at these points sensitive by an extra 10%.

The VTVM AC mode input impedance is one megohm, except on the expensive VTVM’s, which use the same attenuator for the AC and DC modes. The AC input specifications also include the value of shunt capacitance across the input terminals. This figure does not include the probe capacitance, and therefore is simply an indication of the lowest capacitance achievable with special care.

Accuracy

The accuracy specifications of both the AC and the DC modes of the VTVM are given as a percentage of full scale. This is a good reason to keep the measurements as close to full scale as possible. Accuracies of ±1% to ±3% are common for the DC ranges and ±2% to ±5% are common for the AC ranges. The lowest AC range may have a separate error specification due to the nonlinearity of the rectifier at low forward voltages. Accuracy of course depends to a great extent on the price paid for the VTVM. Accuracy also diminishes with time and environmental conditions such as temperature and humidity.

Accuracy specifications of the ohms function are given in one of two ways. On the lower-cost VTVM’s, the ohmmeter
Accuracy is indicated by noting the accuracy of the resistors used in the measurement circuits. This indicates an error about the center scale value of the ohmmeter. The error at other points must be calculated. The lowest ohmmeter range does not have the indicated accuracy, due to variations in the battery internal resistance. Accuracy also depends on the preciseness of the ohmmeter scale calibration.

The alternate mode of accuracy specification notes the accuracy as a percentage of the mid-scale reading. VTVMs with such a specification have different specification values for the first range and other ranges of the ohmmeter. A 2 to 5% accuracy figure is common.

Frequency response
Although the accuracy of the AC mode of the VTVM is given as a percentage of full scale, this specification must be further qualified by a frequency range. The most common method of specifying frequency response is to indicate the frequency range over which the response will not vary more than ±1 dB when referred to a low-impedance 60 Hz source. The common frequency range for the low-cost VTVM is 25 Hz to 1 MHz. Higher upper frequency limits can be had for additional cost. The Hewlett Packard 410B has a 20 Hz to 700 MHz specification, for example, and costs $275.

Ohmmeter ranges
As noted earlier, ohmmeters conventionally used in the VTVM are specified with a center-scale value, usually 10 ohms. The ohmmeter is ranged in decades steps from this point up, R x 1 megohm being the upper limit. This gives usable measurements in the vicinity of 100 megohms.

Probes
The VTVM is normally supplied with some form of probe. In many cases these probes are either attached to the VTVM or have connectors or characteristics of such a special nature they must be used only with the original model of VTVM. As noted before, the DC probe often has an internal 1-megohm resistance at the tip. If one probe is used for all modes of the VTVM, the probe is switched at the tip.

Alternatively, separate leads may be supplied for the DC and AC/ohms functions. There is always a separate lead for the common or return path. Some specialized VTVM’s offer an additional probe for extended high-frequency operation. The mechanical ease and flexibility of the probe are important to consider when purchasing the VTVM.

Controls
The VTVM has two front-panel controls in addition to those for selecting range and function. These are the zero control and the ohmmeter adjustment. The zero control balances the meter electrically when the test leads are shorted together. This control can also move the meter from left zero to center zero, permitting readings of a signal that passes through zero volts. This feature is particularly handy for null detection.

The ohmmeter adjustment sets the full-scale sensitivity of the voltmeter when used in the ohmmeter function. This feature is necessitated by the variations in the battery used for the ohmmeter voltage source.

The function switch selects AC, ±DC, ±AC, and ohms. The ± selection of the DC mode is useful as it eliminates the need to reverse the test leads. Often reversing the test leads is not practical as it is with the conventional VOM, as the common test lead may be connected to earth ground. If the common lead is not connected direct to earth ground, the impedance to ground (especially the capacitive component) may be quite low. Either situation will cause considerable disturbance to the circuit under test if the voltage source in the circuit is connected to the common lead.

The TVM—voltage ranges
The typical scales for a VTVM are shown in Fig. 11. Decibel, peak-to-peak, and zero-center scales are given for special applications. Often meter scales are color-coded for ease of interpretation.

Some different range sequences are commonly used for the TVM. The 1-3-10 sequence is most common at present. These meters start at 0.1, 0.3 or 1 volt and extend to 1,000 or 3,000 volts. A few TVMs use the 1.5-5-15 sequence, and cover approximately the same voltage range. When a very low cost TVM is designed, the scales may run in a 1-10-100 sequence. Such units start with a 1-volt range and go to 1,000 volts. The 1-3-10 sequence is the most practical when working with voltages found in solid-state circuits.

Accuracies
Accuracy for the TVM, like accuracy for the VTVM, is given as a percentage of full scale, and is given separately for the AC and DC voltage modes. DC accuracies are in the ±2% to ±3% area. The accuracy of the DC voltmeter is further limited by environmental factors and possibly the position of the meter.

The AC accuracy lies in the ±2% to ±5% of full scale area, with accuracy being more proportional to price. Accuracy is either specified at a particular frequency or specified over a limited band of frequencies.

Ohmmeter accuracy is much the same as the VTVM ohmmeter accuracy. That is to say, it has a specification indicating a percentage of the mid-scale figure. Accuracies are in the area of 5% for TVM ohmmeters. An alternate method of ohmmeter specification indicates the accuracy in degrees of arc of meter swing. This method of ohmmeter specification is one of the most descriptive. An ohmmeter specification of ±3 degrees of arc is common, on a meter with 100 degrees of swing. Further limitations are given for very low values of resistance.

Frequency response
The frequency response of the AC voltmeter portion of the TVM may be given in a number of ways. The specification may be identical to the AC specification of the VTVM, with AC accuracy specified at a particular frequency (and usually with a low driving impedance) and a frequency response characteristic of plus/minus a number of dB given for a range of frequencies.

Response limits of ±1 dB or ±2 dB are common for TVM specifications; ±1 dB is approximately ±10% error. The specification may be given as error due to the influence of frequency (for example ±5%, 40 Hz to 10 kHz), or the frequency response may be given graphically. The graph plots dB error vs. log frequency.

When limited specifications, such as those for distributors catalogs, are given, the actual numerical values of the frequency response may be omitted. In this case, the meter may have only a frequency range specified, or perhaps a simple error specification on the AC ranges. In the latter case, this error figure should not be taken as the error figure including frequency response, but more as an idea of the error at some fixed frequency such as 60 Hz. Note that the frequency response of the TVM is much more limited than the response we have come to expect from the VTVM. Where the VTVM is specified to one or two megahertz, the upper limit of all but a few of the TVMs is 10 to 100 kHz. When buying the TVM, pay close attention to this upper specification of frequency response if you need accurate measurements at the upper end of the audio spectrum. A meter that stops at 10 kHz may not suffice.

continued next month
Screen controls

Obscure reactions

COLOR TV SETS ARE FULL OF JOLLY LITTLE surprises for us. Some of them are really obscure and some of them are so obvious that we overlook them. We rely heavily on symptoms and reactions for our diagnoses. The picture-tube section of the set has some really good ones. Here's one of the most popular (?) (This column was prompted by a surprisingly large amount of letters on just these symptoms and reactions!)

In the stock tube circuit, the picture-tube cathode voltages are determined by the plate voltage and current of the video-output tube. If we pull this tube, or if the tube is dead, no plate current can flow. Without a load the plate voltage promptly rises to the supply level, which is usually about +400 volts. (See Fig. 1.) Since this point is directly connected to the picture-tube cathodes, their voltage goes more positive also.

Since the picture-tube grids have presumably been unchanged, the cathode voltage going up makes the grids far more negative. Normal ballpark DC voltages are: +300 volts for the cathodes and +200 volts for the grids. This results in a net bias of -100 volts. If the cathode voltage increases to +400, we now have a bias of -200 volts. The raster promptly blacks out because the picture tube is completely cut off.

Or does it? What if we pull the video-output tube or it goes dead and we still have a raster? This is a reverse reaction to what we ought to see; it gives us a very useful clue. If the raster doesn't go out, this is telling us that the picture tube is not cut off, and it ought to be. Something has changed the cutoff voltage. (Have you got it yet? What affects the cutoff voltages of the picture tube?)

The answer in almost all cases is a simple one. One other voltage has a great effect on the cutoff point of the picture tube, and this is the screen grid voltages. If these are set too high, you will still see a raster. One more good clue here; you will not see any video though you may have sound. If the screen voltages are raised to their maximum level, somewhere around 900-1,000 volts, the cutoff point of the tube will be raised to the point where even the -200-volt bias won't kill the raster.

Why? Very simple. Someone has turned all three of the screen controls wide open! This first happened to me a long time ago. The set came in with the symptoms just described—zero plate current in the video-output tube, no video and a raster. The no-current problem turned out to be a contrast control that was wide open. Since this control is the cathode resistor of the video-output stage, this stopped things dead. Further checking (and some little elapsed time!) showed that all three screen controls had been turned full open.

This can happen when a novice technician can't think of anything else, so he turns the screen controls full on. It can also happen if a child gets behind the set and turns the pretty little red, blue and green controls. (Age anywhere between about 10 and 60 or so.) So; when you get this kind of reaction, don't go looking for complicated things yet. Check the setting of the screen controls. Fix the problem in the video stage, run a grey-scale setup adjustment and you're back in business.

My own pet method of setting these controls is to turn each one till I just see a line. Then back up till it has just gone out. This gives you a little better equalization of the three screens. In the sets that do not have a service switch, you can use the same method. Turn the brightness to about 25% on and bring each screen control up till you can barely see a raster, then back it off till it...

This column is for the service technician's problems—TV, radio, audio or industrial electronics. We answer all questions submitted by service technicians on their letterheads individually, by mail, and the more interesting ones will be printed here.

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disappear. This gives the same results. From here on, setup is the same as before. If dark areas of the picture show color, back off the screen control for that color. If the lighter areas (whites) show color, adjust the drive controls until these are a good white.

Other reactions

Excessively high screen-voltages can cause other problems that are not so obvious. Among these are retrace lines that won't go away as they should with brightness and contrast properly set up. In some cases, excess blooming can be caused by a high screen voltage. Before going through a lot of other and more difficult tests, run a grey-scale setup adjustment and see if this won't solve the problem. This can save a lot of time.

This can work the other way, of course. If none of the screen controls will give you a line or raster on setup and yet you get a dim raster with the Service Switch in the Normal position, check to see if the screen-grid voltages are up to normal. Read them on the sliders of the screen controls or on the picture-tube socket. In the stock circuit, all three controls will have the boosted-boost (B+ +) of 1,000–1,200 volts on one end and B+ of about +400 volts on the other. If the B+ + rectifier that is almost always (or always), I'm pretty sure a solid-state diode is down a little, this will reduce the B+ + voltage and the screens won't be able to reach the correct voltage. Quick-check: see if B+ and the raw boost are normal. The boost should be about +850 volts. If these are OK but B+ + is low, try a new diode unit. This is practically a dry circuit—there is hardly any measurable current flow aside from the bleeder current through the high-resistance screen controls and these go back to B+ . Large resistors can be used. If you find one of these that is too hot, look out for leakage somewhere.

Another cute one is the bypass capacitors that are usually connected from the sliders of the screen controls back to the low end. If one of these is leaky or shorted, you won't be able to get that screen up to normal voltage. This is one possible cause for the "two colors make lines on setup but the other one won't" symptom.

Ballpark figures for a typical 21–25-inch picture tube will be +300 volts on the cathodes, +200 volts on the control grids and about +700 volts on the screens. You will find sets with the same tubes using lower DC voltages! So, check the schematic before you start tearing things up!

It's always a good idea to check the bias voltages on the picture tube when you run into unusual reactions involving low high-voltage, dim raster and similar symptoms. The typical maximum beam-current for the picture tube runs somewhere around 1.3–1.4 mA. So, it doesn't take much of an amp in bias to overload the high-voltage supply and cause problems.

**reader questions**

**SCHEMATIC FOR COLOR BAR**

I've a problem of horizontal instability in an Amphenol model 860 color-bar generator. I've checked everything I can with no results. I don't have a schematic. Do you?—P.C., New York, NY.

**DEAD HORIZONTAL OSCILLATOR**

No high-voltage, no bias on horizontal output! This is a Wards GCI-17241A. I know the horizontal oscillator isn't running! I've changed the sinewave coil and the horizontal oscillator coil. No go. Help, help! Customer mad also!—L.H., Detroit, MI.

Don't get mad at it; it's just a machine! Go back over all of the things you've tried and recheck. Also, check...
for signal on the DC voltage supply. A feedback loop here, due to an open filter, can upset the oscillator. (Feedback: “Better glad than mad! I had replaced C521. Turned out that the one I replaced it with had died on the shelf! Thanks for the help.” Note: C521 was not an electrolytic, as I assumed at first. It was the 390-pF coupling capacitor in the horizontal oscillator circuit! These can go bad, too.)

HORIZONTAL OSCILLATOR SUPPLY?

I can’t see where the plate voltage comes from, for the horizontal oscillator, in circuits like a Motorola TS-921. It is fed from the boost; there isn’t any boost until the oscillator starts. Is there? So, this is a chicken-and-the-egg question to me. Which comes first and how does it get there?—M.M., Madras, OR.

This causes a lot of confusion among technicians (including me, for far too long). The secret lies in the fact that there is DC voltage on the boost line, even though the horizontal oscillator and output are not running.

This voltage comes “through” the damper tube. Its plate is tied to B+. As soon as this tube warms up and starts to conduct current, DC appears on the flyback, and also on the plate of the horizontal oscillator. This voltage must be present at normal value on the damper cathode or nothing can work. It will be high enough to start the oscillator and let the output tube conduct.

COLOR PROBLEMS

This Philco-Ford 20KT40B chassis has had repeated loss of color problems for quite a while. It seems to be in the 3.58-MHz oscillator IC. The DC voltages on pins 1 and 7 are high. The supply voltage is a little high, too. 23 volts instead of 20.0. What do you think?—D.N., Baton Rouge, LA.

I think two things right at the moment. One, I think this IC is shorted or damaged internally. The DC voltages on pins 1 and 7 come “from inside the IC”. Whenever any of these are quite a bit off, as they are here, this indicates internal problems. Second, I think the supply-voltage situation should be checked. This voltage should be brought down to the rated 20.0 volts. Over-voltage operation could be the cause of the repeated failures in this circuit.

(Feedback: reader says “Bingo!” That did it.)

ROLL AND DIM-OUT

The raster shrinks, then the picture starts rolling, then it gets dimmer and goes dark; this is in an RCA CTC-53. At the shop, I had to raise the AC line voltage by 10 volts to make it show up. What could be affecting all of these things at once?—J.M., Evanston, IL.

There is one thing that could affect the vertical size, the sync and the brightness all at once in this chassis. This is the boost voltage. It affects the picture tube screen, and also feeds the input half of the vertical oscillator. So, if the boost is dropping, this is the kind of symptom you would see. (Just for luck, check the 31LZ6 horizontal output tube. We have found some of these with gas or grid emission, etc.)

TRANSISTORS OVERHEATING

While testing an RCA Stereo, a YVD-994, on my bench speakers (4-ohm), I noticed that the output transistors got very hot. The emitter-collector voltage across the transistors would drop to less than 0.5 volt at high volume. What is causing this?—S.R., Macogoches, TX.

This overheating at full volume under these conditions is normal. The speaker rating on this model is 8 ohms. If you try to run the amplifier into a 4-ohm load, it will try to develop the same amount of power. This means that the collector current will be double the normal amount. You’re “bottoming” the power transistors. Output-transformerless (OTL) stages in class B or AB, which these are, draw much more current with
a high signal level than they do at rest or no-signal.
You can always use a greater load impedance, but don't go below rated load or short the outputs. This can damage the output transistors.

**REPLACEMENT TRANSISTOR TRICK**

I'd like to know what transistors will replace these: A634 and C1096. That's all the markings; they were in a little orphan stereo.—C.W., Graham, WA.

There's a little trick you can use with some of these import transistors when seeking replacements in Substitution Guides. If you look under "A634" and get zilch, try adding "2S". Looking under 2SA634, you'll come out with something like a Sylvania ECG-187. These seem to be complementary-symmetry, so the other one would be an ECG-186. In RCA, SK-3083/SK-3054.

**HORIZONTAL OUTPUT TUBE GLOWS RED**

The horizontal output tube in this G-E G1 chassis glows red, and the fuse blows. I can get high voltage by driving the plate lead to the flyback with an Analyst. Doesn't this clear the flyback and other circuits?—J.M., Nashville, TN.

Yes! From the symptoms, you've lost the grid drive to the horizontal output tube. Check to see if you get the normal -60 volts on the 21LG6 grid. If not, check the horizontal oscillator. We have had some troubles with intermittent-starting oscillators. Try replacing the capacitors across the oscillator coil. Use exact replacements.

(Feedback: .0033- and .0068-µF capacitors replaced and she plays pretty! Thanks.)

**NO COLOR**

Here's one for you: An Admiral 2K2084 with no color. All color circuits seemed to check out OK. Color very pale at maximum-on color control: some slight barber-pole effect. Finally checked the tint control and found it open. When the slider hit the bad spot, it bridged it and the color came back. New control fixed the whole thing. Have you ever run into this one?—L.S., Brunswick, OH.

Not exactly, but I'm glad to get the data for future use.

**REPEATED DAMPER BLOWING**

This Magnavox T940 chassis has a bad habit; it blows the damper tubes at assorted intervals. This is getting monotonous. What can I do? I noticed that the last tube that blew had a small blob on the end of the cathode ribbon. Does this tell us anything?—W.M., Phoenix, AZ.

Yes, indeed! The blob on the end of the cathode ribbon means that this was blown out by a sudden heavy pulse of current. This is NOT likely to come from something inside the damper tube itself. Look for something that would indicate an anode; maybe in the 6JE6 tube boost, or anywhere in the high-voltage circuitry. Try tapping things.

(Feedback: It worked! I put a 1/2-ampere fuse in the damper plate and then started tapping things. At first I got nothing, then after the set had warmed up a little more, I tapped the 6JE6! There was a flash inside it and the fuse blew. It's been working ever since with a new 6JE6! Thanks.)

**BULLSEYE!**

You told me to look for an intermittent connection around the horizontal oscillator circuit, when I asked about reasons for the damper tube in this set turning red hot. Bullseye! The oscillator coil had a very intermittent connection due to a crack in the PC board. The customer told me why. He and a friend had the set out on the floor, trying to test tubes, and his 4 year old son dropped a pop bottle in it! What next?—P.T., Chicago, IL.

What next? Goodness only knows!

**SERIES-HEATER BREAKDOWN**

A Magnavox T931 keeps blowing 12AT7 tubes! This shuts down the whole

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thing since it's a series string. I replace it and the thing works for a while, then pow; same 12AT7 out. I found a couple of bad capacitors, but this didn't help.—G.B., Seattle, WA.

This used to be quite a common problem in series-heater string sets. The most likely cause is a heater-cathode short in the next tube in the circuit! This will put a much higher AC voltage across the tubes remaining between this point and the input. In this circuit, it would be V1, the 4EH7 1st IF tube.

(Feedback: Right on! That was it.)

VERTICAL RETRACE
I can't get rid of the vertical retrace

lines on this Magnavox T308 black-and-white set. The blanking-gate Zener diode had been broken and replaced by someone else. It connects from the collector of the vertical output to the +420-volt boost. All of the DC voltages seem to be a little low, too. The picture tube heater is only 5.1 volts instead of 6.3. Is this significant?—G.R., Pine Bush, NY.

You seem to have two problems! The low voltage isn't significant; probably due to low line-voltage since even the AC voltage to the picture tube heater is down. Not too bad anyway.

The other one is. The vertical blanking gate Zener does NOT connect from the vertical output collector to the +420-volt boost! I'll admit that if you do not examine the schematic very closely, it LOOKS as if it does! The anode of this Zener goes to the vertical output collector, and the cathode goes to the emitter of the video output, through a 15K resistor. Change it; probably blown up.

NO SOUND FROM AMP WITH GUITAR

My son is trying to use an electric guitar he just got on my stereo amplifier. However, we get no sound from the Auxiliary or Tape Inputs. The guitar is OK; we checked it on a guitar amplifier. Why can't we get sound from the stereo?—E.P., Philadelphia, PA.

For a ballpark guess, the guitar pickup doesn't have quite enough output to drive the stereo amplifier input. The voltage output of these is very low. You can probably get better results by adding a small preamp stage. Your stereo amplifier has plenty of power output, but may not have enough voltage gain in the early stages. Phono pickups and tape decks usually have a fairly high signal output, and these will drive it.

Raster with Dead Damper Tube?

Here's one for you, Swami! In an old G-E, the raster was about half width and folded over very badly in the middle. Looking around, I found the damper tube completely dead! New damper tube fixed it. Here's the question: How can this happen?—L.C., Mena, AR.

That's a good question and I wish you hadn't asked. The only thing I can think of is that the damper tube had a dead heater, but also had a very good plate-to-cathode short! This would be necessary so that you could at least get B+ voltage on to the horizontal-output tube. The loss of any "damping" action would cause the severe foldover, and the loss of boost would reduce the width. (If you want the truth, I have never seen this one at all, but I'll believe anything.)

Loss of High Channels

I've a peculiar condition in an Admiral tuner. We get Channels 2, 4, 5 and 6 on the low band, and 7 and 11 on the high band. Low-band stations are pretty good, but the high-band stations are weak; in fact, it won't even get Channel 11. I tried a new 4GS7 mixer tube, but it didn't help. Tried a 5J8 which was given as a substitute, same thing. What is it?—R.M., Houston, TX.

The most likely thing for this is a load resistor on the oscillator plate, which has gone away in value. A previous mixer tube with a short can cause this. Take a resistance reading from the B+ terminal on the tuner to the oscillator.
plate pin on the tube socket; you can do this from the top. If this resistor is far higher than it should be, replace it. Hint: you can usually clip the leads and tack the new one in. This is much easier than trying to get the ends out of the switch lugs!

**WHAT'S THE CHASSIS NUMBER?**

*I need a schematic on this Emerson. All of the labels and markings have disappeared.*—M.H. Maple Shade, NJ.

*Service Ed.:* "Send me a tube layout and any part numbers you can find"—M.H.; OK, here. Only part number I can find is 294-6631 on the yoke; also "70841" but I believe it's a patent number or something.

*Service Ed.:* Nope: that's the part number for the deflection yoke! We're so lucky we stink. Emerson used this yoke in only one model, which is a Chassis 120837/839/840, etc., etc., in Sams Photofact Folder 838-2! Tube layout, etc... all agrees with your sketch.

**BALLAST TUBE FOR OLD RADIO**

*I've got an old radio I want to fix. Uses 25L6-6SK7-6Q5-25Z6 and a BL42D. What's the heck is a BL42D? It is an odd looking thing, all metal.*—B.F., Chicago, IL.

A BL42D is a "ballast tube"; it is just a heavy duty wirewound resistor mounted in a can on a tube base. The "42" in the type number stands for the voltage drop across it. Add up the heater voltages of the tubes, and subtract it from 110 (which was the line voltage used at that time) and you'll get 42 volts.

This comes out as something like 140 ohms. Your heater current is 300 mA. Use at least a 10- or 20-watt resistor. 140 ohms is an oddball size so use 150 ohms. You can use a little more resistance since line voltages are higher now.

**INTERMITTENT VIDEO**

Thanks for your suggestions on my problems with a CTC-25 RCA. The video was intermittent; horizontal lines, flashes complete picture fade and so on. You said "Check the ground points on the PC board." I looked at all of them and resoldered the ones that didn't look good. No help. So, I got out some solder lugs, nuts and bolts, and grounded those points!

That did it. The problems all went away and haven't come back!

Persistence and perspiration all help.

Thanks to Wes Terry, Carlsbad, CA.)

**INTERMITTENT WIPEOUT OF 6JE6'S**

This RCA CTC-38 chassis has a bad habit of wiping out the 6JE6 horizontal-output tube at odd intervals! Most of the time it comes on and plays perfectly, but every so often it will blow the 6JE6 tube! I've been all over it and I can't find what's doing this. No smoke, no smell, no nothing.—A.S., Gardena, CA.

This evidently is not an intermittent short. The most likely cause would be a horizontal oscillator stage that is intermittent. If the oscillator doesn't start, you will lose the 6JE6. Try tapping and bending the PC board around this stage; bad solder joints are a common cause.

(Feedback: That was it! Bad solder joint on the coupling capacitor from horizontal oscillator to horizontal output stage!)

**CONVERT TV TO SCOPE?**

*I've got a curve tracer and I need a scope to work with it. Can I convert a portable TV to do this?—J.M., Bristol, TN.*

There is a way of converting old TV's for scopes, but you'd probably be better off to use a regular type! The scope bandwidth has a good deal to do with the curve-tracer patterns! Look for a fairly good "previously-owned" scope, and this might do very well. Check a few known good transistors and you'll get an idea of what the patterns should look like.

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</thead>
<tbody>
<tr>
<td>Sensitivity</td>
<td>2.0 Mv. for 1.0 Volt Output</td>
</tr>
<tr>
<td>Maximum Output</td>
<td>5.0 Volts RMS</td>
</tr>
<tr>
<td>Freq. Response</td>
<td>± 1 Db 10 Hz to 100K Hz.</td>
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<tr>
<td>Distortion</td>
<td>Less than 0.05% Harmonic or IM</td>
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<tr>
<td>Noise and Hum</td>
<td>More than 65 Db down</td>
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R-E TESTS BIGSTON
continued from page 74

cassette decks on which we have used it as a reference tape.

Judging from the very low distortion figures obtained with both varieties of tape, we suspect that the people at Bigston have set their bias too high for both standard and CrO$_2$ tapes. A recording level of +8 dB was reached before 3% total harmonic distortion was recorded during playback measurements and, while we certainly are in favor of good "headroom" in a cassette deck, we suspect that backing off on the bias a bit would improve frequency response. Even if the bias reduction resulted in a reduced THD point for this deck to a record level of say +4 or +5 dB, we would have had much higher praise for the deck. All of which goes to prove once more how inter-related the three important tape specifications of THD, frequency response and signal-to-noise level are. If the settings of bias on this deck are typical of all production units of this model (and they well may not be), we would urge Bigston's engineers to reconsider those settings at once.

Use and listening tests

Transport action and control functions were amazingly smooth for a deck priced this low and after a few moments of use we were thoroughly familiar with the control layout. Not mentioned earlier is another innovative feature not found on decks in this class. If both the REWIND and PLAY button are depressed simultaneously, a given tape being played will come to its end and then the PLAY button will automatically spring up (thanks to the automatic end-of-play stop feature), at which time the REWIND button takes over and automatically rewinds the tape at high speed to its starting point. With so many fine features going for it, we were disturbed at the absence of highs during playback of sample musical material that we recorded. Our product analysis and summary will be found in Table II.

COLOR TV '77
continued from page 57

pulses on the diode cathodes back-bias their junctions, allowing their anodes as well as the base and emitter of the corresponding transistor to rise.

Transistors Q41 and Q42 act as peak detectors, since they are emitter followers with high current capability in one direction. Emitter-connected capacitors C43 and C44 are the detector storage elements.

Capacitor C42 charges to a voltage dependent on the amplitude of the chroma
signal during the 35-microsecond blank reference interval. Capacitor C41 charges during the 35-microsecond chromo reference pulse. One of the four op-amps in K60, an LM3900, is connected differentially to the two detector outputs. The output of the amplifier is filtered by R48, C48 and R49, and is proportional to the error between the two detected signals.

Selecting the VIR feature with the front-panel control substitutes the filtered output of the op-amp for the DC voltage on the slider of the manual tint control. Variations between the chroma tint and what the VIR signal says it should be is automatically corrected in closed-loop fashion.

**GTE Sylvania**

The 1977 Sylvania color line includes the carriage GT-Matic II E40 chassis in 21- and 25-inch models, the varactor-tuned E41 and E42 19- and 25-inch sets, and the E20 and E21 17- and 19-inch models. Continuation of the countdown vertical synchronization system remains unique to the GT-Matic receivers.

Seven IC's are used in the high-end GT-Matic II's. They feature an electronically regulated power-supply, low B+ short protection and horizontal shutdown circuits. Three circuit modules plug into the chassis. The IF module contains the power supply minus the transformer, the video IF and detector, the tuner AFC, and the sound IF, audio detector and power amplifier.

The video chroma module contains the chroma processor, the 358-MHz CW regenerator and control, video processing and blanking, and the color drive circuits. Vertical and horizontal drive and sweep systems, the pincushion circuitry and the Boost B+ power-supply are all on the detection module. The remaining heavy parts, including the power transformer and filter, horizontal transformer, and tripler, and heat sinking for the B+ regulator and horizontal output transistor, mount on the main chassis.

All IC's and transistors plug in for serviceability. The receiver is loaded with safety features for protection against X-ray radiation and fire. Contributing to fire protection are the power-supply short circuit sensing circuits, the horizontal shutdown system, UL-rated self-extinguishing wire, and a mylar insulated and epoxy encapsulated high-voltage transformer. X-ray protection components are the horizontal shutdown device, the high-voltage tripler and the strontium-90 softened CRT glass.

The varactor E41 and E42 receivers use upgraded E11 and E12 chassis. Single-knob control of the varactor tuners makes VHF, UHF, and remote control tuning equal. Mechanical memory tuning centers around a 21-pot turret tuner that uses a regulated 33-volt Zener supply as the reference supply. Bandswitching applies power only to the VHF or UHF tuner at any time. The slider on the potentiometers feed a PNP-NPN dual emitter-follower circuit that buffers the tuner from the pots. The voltages of the two oppositely poled base-to-emitter junctions tend to cancel.

These sets use a constant voltage ferroresonant transformer for supply regulation.

A special retrace capacitor design prevents an increase in high voltage due to opens or shorts in the capacitor. Normally either failure would shorten the flyback time and increase the 2nd anode voltage. Figure 5 shows the four-legged arrangement. Each of the two capacitor foils acts as jumpers, connecting the horizontal output transistor to the output transformer and to ground. If any connection opens, the output transistor is disconnected and the circuit is disabled. Shorts in the capacitor divert the current around the output transistor which shuts everything down.

The regulated supply in the E20 and E21 chassis uses a current sensing resistor to turn on a current limit transistor when currents exceed 1.5 A. An SCR shuts off the drive pulses to the horizontal driver and holds the horizontal output transistor off.

Due to differences in the chromaticity coordinates of today's picture tubes and the NTSC chromaticity specifications, color errors result in straight inverse NTSC decoding. Decoding in the E20 and E21 have been modified to minimize these errors.

---

**FIG. 5—SAFETY RETRACE CAPACITOR acts as fuse in Sylvania E41, E42 chassis.**

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**FEBRUARY 1977**
CAR CLOCK
continued from page 63

reaching the setting that causes it to flash at 60 flashes-per-minute.
If C4 is adjusted to provide the 50-Hz input to the clock up to 30 Hz, the oscillator may not begin oscillations at 100 kHz whenever power is initially applied to the clock. This is true even if a counter is used to adjust the frequency.

Display current calibration
During display-current calibration, the time should be set to 10:08 or 10:09 for optimum results. This can be accomplished by manipulating the FAST SET and SLOW SET switches.

The BIAS potentiometer (R10) and the DARK ADJUST potentiometer (R9) must be set to values that allow maximum safe current to flow through the display when maximum ambient light is striking its face. In addition, the display current must fall to a value that will yield the desired display intensity for night-time driving. Therefore, the display current must be monitored with an ammeter between the display power Digi-Klip located on the main board and the display power edge connector located on the display board.

During LED display calibration, the nuts that hold the bezel to the display board should be loose enough to allow easy access to the photoresistor. However, during the incandescent display calibration, the bezel must be secure so that the chimney of the photoresistor is flush against the viewing screen: this will eliminate the problem of light reaching the photoresistor via the display itself.

For this reason the calibration procedures for the two types of displays are different. Calibration for the LED display will be discussed first.

LED display calibration
With R9 and R10 set at minimum resistance, and the viewing screen in place, allow the maximum expected ambient light to strike the display and the photoresistor through the viewing screen.

Increase the resistance of R10 until the display current, as read on the ammeter, is about 210-220 mA. The display should be visible, but not flashing to indicate a power failure.

After a 5-minute warm-up period, adjust R10 for a display current of 300-310 mA. Allow another 2-minute warm-up period; adjust the potentiometer, if necessary, to keep the display current at or below 310 mA. Set the time to 10:08 or 10:09 and reset R10 to 300-310 mA, if necessary.

Remove the viewing screen or bezel, and cover the photoresistor with black electrical tape. With the time set to 10:08 or 10:09, increase the resistance of R9 until the display current is about 20 mA. In bright light, the display will not seem visible; however, it will be about right for nighttime viewing.

Remove the black tape from the photoresistor, replace the viewing screen, and apply maximum expected ambient light to the display. Readjust R10 for a display current of 300-310 mA with the time set to 10:08.

Repeat the dark adjustment and the light adjustment until the display current is about 20 mA when the photoresistor is covered with black tape, and 300-310 mA with the black tape removed, the viewing screen in place, and maximum expected ambient light striking the display.

The 20-mA dark display current is an approximation only; the display should be adjusted for final desired intensity under actual conditions. During all display current adjustments, the time should be set to 10:08, if possible. After adjustments are completed, the display brightness should be sensitive to a hand passed in front of the photoresistor. If absolutely necessary, the maximum permissible display current is 315 mA. However, some allowance should be made for seasonal changes in the weather. With the display disabled, maximum current drawn by the clock will be about 15 mA: with displays enabled, about 325 mA.

Incandescent display calibration
With R9 and R10 set at minimum resistance, the viewing screen in place, and the chimney of the photoresistor flush against the viewing screen, allow the maximum expected ambient light to strike the display and the photoresistor through the viewing screen.

Increase the resistance of R10 until the display current is about 200 mA. Allow a 2-minute warm-up period, then set the time to 10:08 or 10:09.

Readjust R10 for a display current of about 290 mA. Remove the viewing screen and cover the photoresistor with black tape. Then increase the resistance of R9 until the display current is about 100 mA.

Remove the black tape from the photoresistor, replace the viewing screen, and allow maximum expected ambient light to strike the display. The display current should rise to about 290 mA if the time is set for 10:08.

Once again, the viewing screen should be removed and the photoresistor covered with black tape to make sure that the display current will fall to about 100 mA when the photoresistor is fully dark.

If this operation is not correct repeat the above adjustments, but adjust R10 to about 265 mA with the time set at 10:08. Continue in this manner, reducing the initial light current in steps of 5 mA, if necessary, until the maximum display current (not to exceed 290 mA) can be obtained under maximum ambient light. However, the display current should fall to about 100 mA when the photoresistor is covered with black tape.

With the incandescent displays, only one adjustment of the BIAS and DARK ADJUST potentiometers should be necessary; the additional steps are simply given in case they are needed.

With the display disabled, maximum current drawn will be about 15 mA: with displays enabled, about 305 mA.

R-E JONES
TV REPAIR
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"I hardly think it's proper, to have my alma mater, scavenging upon my clientele."
RESISTOR BURNUP

I've been very confused trying to get this Kenwood amplifier straightened out. There's a 68-ohm resistor in the left channel output that keeps burning up. What can I do to prevent this?—W.Q., Granada Hills, CA.

For one thing, check the resting current in both channels. This is the output stage current with no signal at all. It should be the same in both channels and should be fairly small, say about 20 mA ballpark. It will increase as the volume is turned up but should still be about the same in both.

(Feedback: That did it! One channel read about 15 mA, the other read over 100 mA resting current! This was the one that was blowing the 68-ohm resistor. Further checking showed up a couple of resistors in the bias network that had changed in value. I cross-checked against the same resistors in the other channel. Thanks.)

WHY DOESN'T THE RASTER GO OUT?

There is raster and sound on this CTC-39XA RCA but no picture. The plate voltage is high on the 12HG7 video-output tube but the cathode voltage is zero. Screen voltage is high too. In fact, I can pull this tube with no effect. Why does the raster stay on? I thought it should go out—T.H., Pontiac, MI.

It ought to. In either case (zero cathode voltage indicating no plate current, or tube out), the plate voltage should rise to the supply level. This bias change on the picture tube should cut it off. Now: let us heat up the analyzing iron and see.

One: It sounds to me as if your controls are all hooked up! See if all three screen controls on the picture tube aren't turned full on. This will raise the cutoff so that you can get a raster.

Two: Check continuity in the 12HG7 cathode circuit; the contrast control may be open. Also, look to see if this tube is LIT! This booby-trapped me in a case with the same symp-
toms!

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That spelled "opportunity" to Herb. Plus he liked the idea of having a set of skills that might lead to jobs in places as different as a TV station... a hospital... an airport... a petroleum refinery.

But what Herb liked _most_ about electronics is that it's just plain _interesting_. Even though it takes time and effort to learn, the subject is so fascinating it _almost_ doesn't seem like "studying" at all!

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S/RF METERS. Signal strength and relative power output are easily read with these units. The base version is in a 3 x 7 x 4-inch case; the mobile version in a 3 x 4 x 5-inch case. Both units come with a 5-foot shielded cable and a standard phone plug—Kits, Inc., Pioneer Rd., Cedarburg, WI 53012

CIRCLE 101 ON FREE INFORMATION CARD

CB HEADSET ADAPTER. Model CB-88 consists of an eyeglass adapter and a CB-88 lightweight headset. The adapter is an eyeglass clip onto which either of the eyeglass temples is inserted. The adapter is held firmly in place with a setscrew. The headset weighs less than 3 ounces. When receiving and transmitting ele-
ments are used with the eyeglass, making use of the headband unnecessary, the weight is further reduced. The unit includes a noise-canceling power mike and is mounted on a pivoting boom. The push-to-talk switch is equipped with a clothing clip. The unit is priced for less than $70.00—Telex Communications, Inc., Minneapolis, MN 55420

CIRCLE 103 ON FREE INFORMATION CARD

CB ANTENNA DISGUISE. The Little Fooler AM/FM/CT antenna deters theft simply because it doesn't look like a CB antenna. It looks like an ordinary three-section telescopic antenna.

The antenna transmits and receives CB simultaneously with AM and FM. It covers all CB channels. VSWR of 1.2:1 is claimed on any of the 40 channels—Anixter Bros., Inc., 4711 Golf Road, One Concourse Plaza, Skokie, IL 60076

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CB ANTENNAS. Four styles of fiberglass antennas offering a complete selection of top-center- and base-loaded models. They are available individually, in kits with mounts, or twin antenna kits with harnesses featuring GC Co.

Phase Mixer-Balun Circuity—GC Electronics, Div. of Hydrometals, Inc., 400 S. Wyman, Rockford, IL

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CB ANALYZER, model GN-1375B is AC/battery-powered portable instrument for testing performance and troubleshooting CB transceivers and other communication products. Comes complete with test cables, probe, battery pack module, regulated power supply-charger. AC cord and instruction manual. The unit uses crystal-controlled RF oscillators to provide laboratory-grade performance of high accuracy, stability, and reliability. Also, it is equipped with an RF/AF signal tracer, and RF wattmeter, dummy load, SWR meter, relative field-strength monitor and % of modulation meter.—Nikoltronix Electronic Engineering Co., 2437 W. Peterson Ave., Chicago, IL 60659

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Kits are complete (less cabinet) including PC boards, power supply, IC socket, 9 switches, 16 transistors and all parts required for above features and options. [All #7001 Kits Will Cabinet] PRINTED CIRCUIT BOARDS for CT-7001 Kits sold separately with assembly info. PC boards are drilled Fiber glass, solder plated and screened with component layout. Scratch for #7001B or #7001C (Set of 2) $7.95.

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Printed Circuit Board for kit # 850-4 (etched & drilled fiberglass) $2.95 Mini-Brite Red Led’s (for col in clock display) pkg of 5 $1.00. Molex pins - 24-Molex pins for IC socket.

NOTE: Entire Clock may be assembled on one PC Board or Board may be cut to remote display. Kit # 850-4 will fit Plexiglas Cabinet II.

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NOTE: Entire Clock may be assembled on one PC Board or Board may be cut to remote display. Kit # 850-4 will fit Plexiglas Cabinet II.
HI-FIDELITY SPEAKER SYSTEMS. The model SP-325 is a three-way acoustic suspension speaker system featuring an 8-inch woofer with a 1/2-pound magnet and a 11/4-inch long-throw voice coil. The SP-325 also has a 3-inch mid-range speaker and a 3-inch super tweeter. Crossover frequencies at 6,000 and 9,000 Hz, frequency response is 50 to 20,000 Hz at 8 ohms, power handling capacity is 20 watts RMS with peak power 30 watts. This speaker system also features a cabinet structure, 10 1/8 by 16 1/2 by 7 1/8 inches, of oiled walnut with a laminated finish over 1/2-inch composition board. The removable grille is contrasting brown grille cloth fabric, stretched over plastic molded perforated base. The model weighs 15 lbs. and retails for $40.00 — Olson Electronics, 260 Forge St., Akron, OH 44327.

CIRCUIT 87 ON FREE INFORMATION CARD

INTERFERENCE FILTERS, models AV-800, AV-811 and AV-820. If the CB transceiver is radiating harmonics of the same frequency assigned to one or more of the local TV channels, installation of the model AV-800 low-pass filter (shown) on the transceiver should clear up the problem if the problem is at the TV receiver due to front-end overloading, the model AV-811 filter on the TV lead-in should solve it. The filter lets the TV signals come through unhindered, while choking off the incoming CB signal. A third filter, model AV-820, used at the TV set, prevents outside CB signals from entering the TV through the AC power line. — Avanti Research & Development, Inc., 340 Stewart Ave., Addison, IL 60101.

CIRCLE 88 ON FREE INFORMATION CARD

CB METER, Model 10043, is a combination power-meter and voltage standing-wave ratio bridge. The new meter measures RF output power up to 10 watts as well as providing VSWR measurements. An important feature of the meter is that it does not require a perfect impedance match to read accurately. It will read ±5% regardless of the input-output impedance match.

The meter is housed in a high-impact black plastic case. The suggested retail price is $88.00 — Antenna Incorporated, 23850 Commerce Park Road, Cleveland, OH 44122.

CIRCLE 89 ON FREE INFORMATION CARD

CB MONITOR, model CB 23, converts any AM car radio to a Citizen-band receiver. Includes a built-in noise suppression circuit using 4 transistors and 3 diodes to minimize ignition noise.

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The booklet describes cassette tapes and explains in detail the steps to take before recording. Includes a section on the language of cassette tapes and converting engineering terms into plain English. Send SASE to Audio Tape Division Fuji Photo Film U.S.A., Inc., Empire State Building, New York, NY 10001.

FLAT CABLE AND CONNECTOR CATALOG, No. FC-2, is a 20-page, 2-color catalog. The catalog covers a complete line of flat cable and connectors used for multiple termination in electronic circuits.

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KARL SAVON
SEMICONDUCTOR EDITOR

SONAR AND SODAR TECHNOLOGIES ARE THE BASIS of a variety of fish finder, burglar alarm, and collision avoidance products. A new integrated circuit from National Semiconductor reduces the active circuitry for this useful class of equipment to little more than the single chip and an ultrasonic transducer. The medium-sized 80 x 93-mil chip contains a complete transmitter-receiver.

Choice of two frequencies adapts the equipment to both water-confined Sonar (Sound Navigation and Ranging) and air-operative Sodar (Sound Detection and Ranging) systems.

Figure 1 is the functional schematic of National's LM1812 ultrasonic transceiver. The single transducer is time-shared by the receiver and transmitter circuits. One-millisecond transmit intervals are interleaved with longer, range-dependent intervals to listen for the reflected echo.

Modulation input pulses at pin 8 proportion the two time periods. Some applications generate the modulation pulse with a permanent magnet attached to an electromechanical display that excites an inductive pickup. Under the control of the pin-8 pulse, Q6 is turned on, enabling the gated oscillator tuned by the L1-C3 tank. The tank is used by both transmitter and receiver, ensuring that they are always at identical frequencies.

The transmitter

Multi-collector transistor Q7 couples the oscillator signal to the 1-μs one-shot. Having been enabled by the modulating pulse on pin 8, the monostable produces a pulse train of 1-μs positive-going pulses with a total 5-μs cycle time defined by the reciprocal of the 200-KHz oscillator frequency. Incidentally, 200 KHz is the water sonar application frequency. Because of the attenuation of the higher frequencies in air, a lower 40-KHz rate is used in sodar.

The 20 percent duty-cycle monostable output feeds the three-transistor transmitter output stage. One-ampere transistor current pulses multiplied by the 12-volt supply is 12 watts of peak power, yet no special heatsinking is needed. The power is dealt with by switching the output transistors on fully during the shorter 1-μs positive pulse.

Transistor Q14 is a buffer amplifier that turns on during the 1-μs period. When it is conducting, its collector is pulled low and an inverted modulation pulse appears at pin 7. Transistors Q15 and Q16 are a high-current-
gain Darlington output stage. The output voltage on pin 6 is limited to the base-emitter voltage drop of Q16 because as Q15 becomes saturated, the base and collector of Q16 are effectively tied together and Q16 looks like a diode. The base voltage of Q16 is the output voltage on pin 6. With the output transistors off, no power is dissipated simply because there is no current. Switched on, the power dissipation of the output device is limited to the saturation $V_{BE}$ output times the output current. Averaging further limits the power dissipation to $\frac{1}{2}$ the peak power since the stage is only active for that portion of the total duty-cycle.

Technically, the output operation falls into the Class-C category since the stage conducts for less than half a cycle. The square-wave output is filtered by the low-Q tuned transformer network. Tuned to resonate with the combined cable and transducer capacitance, the transformer coupling tolerates field transducer replacements without retuning.

In some applications such as hydroacoustic communication systems, either AM, FM, or pulse modulation can be used at reduced power levels.

The ultrasonic receiver

When the input swings negative, capacitor C1 charges more positive at the base of Q1 due to the conduction of D1. Positive input signals then drive Q1 into conduction. Q2 is an emitter-follower that buffers the high-gain first RF stage from the gain control pot. The pot slider is coupled to the second RF stage that is biased by the same diode-capacitor method. Switching off the cascode transistor Q4 disables the second RF stage during the transmit mode.

During the longer receive periods, the base of Q4 is biased at two $V_{BE}$'s above ground by D3 and D4 fed from a current source. The second stage of the receiver is disabled by duty-cycle control transistor Q5 to Q6. Turning on Q5 grounds the base of Q4 directly. When Q6 goes on, the subtractive effect of D5 and D6 also grounds Q4, shutting down the receiver.

L1–C3 acts as a selective filter tuned to either 200 or 40 kHz in the receive portion of the sequence. Transistor Q7 sends the received signal to pulse-train detector Q8. Its purpose is to discriminate against noise. Two or three missing pulses allows C5 to charge sufficiently to dump the charge from integration capacitor C4. Noise reject pot R2 and series resistor R1 in combination with C4 provide an adjustable noise-threshold. After a number of cycles have been received, indicating a valid return echo, display driver Q10 turns on, which in turn drives drives display transistors Q11, Q12, and Q13. Like the transmitter stage, it can deliver ten watts of peak display drive.

The diagram shows an LED display although neon, digital, and CRT readouts can also be used. Neon displays common in depth measuring and fish finding equipment can also be used. Neon displays common in depth measuring and fish finding equipment use a voltage step-up transformer. Various methods of chip protection are possible to prevent excessive current buildup in the transformer primary. Returning the transformer winding to an RC filter gives the peak power delivery of the capacitor and the long-

![Figure 1: National Semiconductor LM1812 Ultrasonic Transceiver](image-url)
CMP1802 microprocessor

Microprocessors are available in different technologies—PMOS, NMOS, bipolar, and CMOS. CMOS has the distinct advantages of low power dissipation and wide-temperature operation. RCA's 1801 microprocessor was a two-unit deal that was relatively slow and expensive. Started out at $200, the price of the pair dropped into the still unattractive $50 range.

But they haven't been sleeping! Just announced is a single-package CMP1802 COSMAC microprocessor that is a dramatic improvement over the earlier version. New instruction repertoire has been added, cost is competitive, down to under $30 levels, and the execution speed of single-cycle instructions is down to 2.5 microseconds. Standard aluminum gate technology has been supplemented with a self-aligned silicon-gate process that cuts down chip area and increases yield.

RCA's new 230 × 180-mil µP retains the unique architecture of the older 1801 design. Programs written for the earlier circuit will run without change on the new one. It is built with a new CMOS closed-COS/MOS logic. Source connections are common and separate contacts are not needed. Guardbands to prevent parasitic action are not needed, yet the CMP1802D can operate at full supply range of 3–12 volts.

The CMP1802 block diagram in Fig. 2 shows an 8-bit address bus MAO through MA7. The bus is time-multiplexed to 16 bits. Storing the first 8 bits in an external latch gives 65K of memory addressing capability.

Sixteen general-purpose registers distinguish the RCA approach from contemporary processors. Each register is 16 bits long, which adds up to 32 8-bit words or a total of 256 read-write scratchpad bits. By loading these registers with frequently used addresses and data, efficient programming code can be written. Single-byte instructions replace multiple-byte main memory addressing operations in other schemes.

Any register can be designated as the program counter or data pointer. One of these 4-bit registers selects the particular one of the 256 registers.

Many of the memory and register instructions include a 4-bit operand that is stored in the N register. The most significant 4 bits of the instruction is the operation code and is stored in the 4 bit I register. For example, the instruction format 1N is an increment register operation. To increment register 5 by one, the instructions would be 15 in hex or 00010101 in binary. Binary number 0001 would be stored in the I register and 0101 in the N register. The control logic decodes the opcode and sends out signals to direct the N register to select register 5 in the array. The control logic then causes the addressed register to be incremented by one by the increment block in Fig. 2.
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The 4-bit X register picks the register to be referenced by arithmetic and logic oper-
ations.

Interruptions in program executions by peripherals such as terminals and disk memories are standard computer procedures. The 8-bit T register is used to hold the X and P register contents and store them in a single memory-location. After servicing the inter-
rupting device with a special program routine the T register is used to reset X and P to their original values so the main program can continue from where it left off.

The 8-bit D register is the \( \mu P \) accumulator. Data is transferred in either direction between the D and stack registers. Since the general registers are twice the length of the D register, either the low or high byte can be loaded by separate instruc-
tions.

Data transferred to and from memory must pass through the accumulator and is handled by a series of eight memory refer-
ence instructions. The instructions use either the X or P registers as operands. These pointers will either stay fixed or will be automatically incremented by the choice of instruction.

Arithmetic and logic operations take place between memory and the D accumulator, governed by an extensive series of instruc-
tions. The 1802 has a number of new imme-
diate instructions that are two bytes in length. These load constants or use them as other operands. Immediate means the constant is stored in the second byte of the instruction. This is very useful when setting up indexes or initial memory pointers, which can then be incremented or decremented by the pro-
gram.

At the heart of many machine-language routines are the branch instructions, and a large assortment will save programming code. The 1802 selection includes two-byte short branches where the second byte is the address that replaces the low program counter byte. Jumps over a 2-256 word total range on the same memory page use these instructions. New to the 1802 are long branches that are three-bytes and allow jumps to any location in memory. The high and low address bytes to be inserted into the program counter are in the second and third instruction words. Except for the uncondi-
tional branch, the decision to jump or not is based on the state of the D register, and the DF, Q, and EF flags. DF, the data flag, is a one-bit ALU carry flip-flop. Q is a program controlled flag, and EF1 through EF4 are a group of flags controlled by peripherals.

Short and long skip instructions are similar to the branches in that they are executed in response to the same flag conditions, but their action is limited to the skipping of either one or two steps in the program. They are single-byte instructions that are very efficient in terms of memory usage.

The ten control instructions include the stack return and idle operations. The remain-
ging group are the output instructions that route data from the registers pointed to by X to the output data bus or from the input data bus. During input-output operations, the 4-
bit N register is set to a 10 value between 1 and 7 and used to select the peripheral device.

The unit price of the 4-6-volt 5-\( \mu P \) CPD1802D is $29.50. The 3-12 volt full speed CPD1802CD is $43.50. Information is available from RCA Solid State Division. Route 202, Somerville, NJ 08876.
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Cobra has a reputation for punching through loud and clear. The new Cobra 32XLR, of course, continues the reputation. And creates another—for innovative design, superb engineering and technical superiority.

Start with the illuminated 4-in-1 meter. It tells you exactly how much power you're pushing out and pulling in. As well as monitoring your modulation in precise percentages. And measuring your punch with an SWR check. In short, the 32 XLR lets you keep an eye on your ears.

ScanAlert, Cobra's unique scanning system, continually monitors Channel 9 when you're on another channel. If an emergency comes up, the ScanAlert light goes on. And the incoming message automatically locks the receiver on the active channel.

The 32XLR's Digital Channel Selector is the very latest. With large LED numerals—for a read-out that registers clearly and quickly. Plus switchable "pulse block" noise blanking that rejects short-pulse noise not normally blocked by other systems. Which makes it the most effective in the business. Finally, add automatic noise limiting, Dynamike Plus (with built-in power mike) and Delta Tuning.

The new Cobra 32XLR. It has virtually everything. And it has everything to do just one thing. Punch through loud and clear.

For information on our complete line write for brochure #CB-2

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