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TV PICTURES ON A DISC

Thin plastic disc rotates at 1500 rpm, floats on air and produces TV pictures. Made by AEG-Telefunken, the unit looks a lot like a conventional record player. The unit shown handles only black-and-white, but color is in the works. For more details see Looking Ahead on page 4.

BATTERY-POWERED CONTINUOUS LASERS

MURRAY HILL, N. J.—A new laser, smaller than a grain of sand, that can be powered by ordinary dry-cell batteries, has been built by Bell Laboratory Scientists. It emits a beam of near-visible infrared light.

Designed by Izuo Hayashi and Morton Panish, the new laser is a semiconductor device that operates continuously at normal room temperature (75-80°F). This is about the average operating temperature of typical transmission equipment used in communications systems.

In the past, the heat generated in semiconductor lasers was too great to permit operation at room temperature for more than a fraction of a second at a time without destroying the semiconductor itself. Even when these lasers were mounted on heat sinks, room-temperature operation was limited to about 1/10,000 second.

By building a new semiconductor laser structure the designers eliminated the need for additional cooling.

The new semiconductor laser consists of four layers of semiconductor materials. Two of the layers, each about 60 millimicrons of an inch thick, confine laser light to a thin central layer of the structure. This layer, called the active region, is about 20 millimicrons of an inch thick.

Laser activity, effectively confined to such a small region, can be produced with much less current, resulting in less heating. This makes the continuous laser operation possible.

Supermagnets To Speed Tomorrow’s Transit

MENLO PARK, CALIF.—A new rapid transit vehicle which can race along an invisible magnetic “track” at over 300 miles-an-hour is under study at Stanford Research Institute. The new transit mode would use the tremendous power of superconducting magnets to lift the weight of the vehicle off its track and propel it to extremely high speeds.

Key to the system is magnetic levitation, a phrase which refers to the use of magnetic repulsion to lift the vehicle free of any physical support, allowing it to achieve very high speeds easily and economically. Pushed by linear induction motors, which also use magnetic forces, the vehicle could travel at high speed in virtual silence.

Operating between downtown urban centers such a vehicle could whisk passengers between New York and Washington in about an hour. The trip between downtown San Francisco and downtown Los Angeles could be made in about 1-1/2 hours. By eliminating the time-consuming auto trip between airport and downtown destinations this new transit system could compete with airliners operating in high-density travel corridors.

SRI researchers reasoned that a transit vehicle carrying powerful superconducting magnets could be levitated as much as a foot above a metal-lined roadway. The same magnet forces could be used to keep the car centered in the roadway.

The suspended vehicle could be propelled by linear-
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Radio-Electronics is indexed in Applied Science & Technology Index (formerly Industrial Arts Index)
For casters.

Video disc system

A new entry in the videoplayer sweepsstakes has some engineers very excited. It's the long-rumored video disc system developed by German manufacturer AEI-Telefunken with British Decca and demonstrated recently in Berlin. The playback unit looks, at first sight, like a conventional phonograph. The grooved discs are extremely thin flexible plastic records measuring eight and 12 inches in diameter, and playing for five and 12 minutes respectively in black-and-white. (See photo in New & Timely, page 2.)

Each disc contains from 120 to 140 grooves per millimeter, and spins at 1,500 revolutions per minute, producing one frame per revolution. On the turntable, the discs ride on a cushion of air, and a new type of pickup responds to the variations in groove elevation. Telefunken is developing longer-playing discs, a disc changer and a system which plays back in color.

American engineers who witnessed the Berlin demonstration say picture quality was very good. Among the claimed advantages of the disc system over other videoplayer techniques: (1) Economy—discs may cost five dollars each, or less; players would be priced somewhere between $140 and $280. (2) Fast duplication—discs may be pressed in a conventional manner, about 1,000 in 12 minutes. (3) Playback ease—the phonograph is still the most familiar type of playback equipment, and the consumer would not have the problem of learning new procedures for playing visual material.

For color consistency

A new series of signals is going out over the air along with current TV broadcasts. It's an experimental system being tested by the television broadcasters as part of an all-out effort to achieve uniformity in color from program-to-program, scene-to-scene and channel-to-channel.

On selected programs of all three networks, a test is being made of the use of part of the "vertical interval" between frames to transmit reference signals for contrast, tint and color intensity. The vertical interval reference signals are inserted in line 20 of field 1. If the tests work out, stations will be supplied with a method for absolute calibration of equipment to produce uniformity in colors. This and other procedures and standards being developed by a top-level engineering committee representing TV broadcasters, set manufacturers, film and tape companies and the telephone company, eventually could lead to color uniformity so good that the tint control could be eliminated from color TV receivers. At least, that's the hope of the engineers.

Standard amplifier measurements

The rule faced by the purchaser of high-fidelity equipment—particularly the unknowledgeable one—has always been "let the buyer beware." Claims for amplifier power, in particular, have been difficult or impossible to compare. There have been several so-called "standard" ways of measuring—the Institute of High Fidelity (IHF) standard, the Electronic Industries Association (EIA) "music power" standard, and something that some manufacturers dreamed up and called "peak music power." Generally, the IHF standard was designed to apply to audio component amplifiers, the EIA and "peak" standards to packaged consoles—and there was no real way for the layman to compare various equipment.

Finally, after some 15 years of discussion and debate, a standard system for describing audio amplifier power appears to be in sight. The EIA has modified its system to the point where it would be compatible with the IHF system. Under the proposal, which had not yet been officially approved by EIA members at press time, all manufacturers would use the same measurements and terminology.

The original EIA "music power output" (MPO) standard, developed in 1960, specified five percent total harmonic distortion. The new proposed standard permits the manufacturer to use any total harmonic distortion level he wishes, but not to exceed five percent. In addition, the EIA has changed test procedures, renamed MPO "momentary power output" and added a continuous power output test and rating procedure.

EIA's proposed new rule also includes definitions and procedures for establishing rated intermodulation distortion, intermodulation distortion-low level, frequency response, channel separation, signal-to-noise ratio, input overload margin, input impedance, output impedance, damping factor and tracking error.

The EIA-IHF efforts to arrive at a common standard were accelerated by a new move by the Federal Trade Commission to step in and police amplifier output claims through a new Trade Regulation Rule. The proposed rule is still in exploratory stages but would require that manufacturers use standard measurements whenever they made claims for amplifier power. EIA and IHF hope that the FTC will accept their proposed joint standard, giving it—in effect—the force of law, and ending at least some of the confusion in the amplifier "horsepower race."

Home videoplayer sweepstakes

Keep your eye on the hot home video-player field. It could end up as biggest boom—or biggest bust—in the history of consumer electronics. It appears that almost everybody has a system, and hardly any one is compatible with any other. In addition to the Telefunken system described above, it's believed that some 20 or more different systems will have been exposed before the year is over. Among non-compatible approaches are RCA's holographic-tape Selecta-Vision, CBS's electronically-recorded EVR film, Sony's Video-cassette magnetic tape system, and other non-compatible tape approaches by Cartridge Television (Cartrivision), Ampex, Arvin, Philips and others. In addition there is the ABO color-encoded film system, plus electronically-scanned Super-8 systems introduced by several manufacturers.

As each new system is introduced, the question of standards becomes more urgent. The promising new field of pre-recorded audio-visual records for the home is in danger of strangling on consumer confusion over the multiplicity of systems. How many years would audio recording have been set back if there had been introduced simultaneously cylinder recordings in eight different speeds, discs which played clockwise and discs which played counterclockwise in five different speeds, magnetic tapes in four widths, grooved tapes recorded by four different non-compatible methods and optical film recordings in several configurations? It's enough to bring back vaudeville, band concerts in the park and other innocent non-electronic diversions.
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CIRCLE 3 ON READER SERVICE CARD
Editorial — TO THE CONSTRUCTOR

FROM DAY TO DAY, OUR PERSONAL EXPERIENCES AND CALLS and letters from you indicate that it is becoming increasingly difficult to obtain all the parts needed to complete any, except the simplest of construction projects. With present-day construction techniques tending toward increased use of semiconductors, printed circuits and miniaturization, it is often impossible to substitute more readily available components.

Take, for example, a magazine article describing a photoelectric control unit with a PC board laid out to accept a particular relay and power transformer. Its cabinet is just large enough to house the electronic circuitry. You, like most readers, want to build the unit exactly as the author described it. The relay, transformer and cabinet — the critical parts in this project — are all standard replacement or consumer-type items listed in manufacturers' catalogs but you can expect some difficulty in obtaining at least one of these parts. This is especially true when you try to buy them locally.

Most of you go to local electronics parts houses for components for your projects. There are three basic types of parts suppliers. One is set up to serve the local TV and radio service industry or possibly the needs of nearby electronics or industrial plants. This type of dealer stocks only those parts experience has shown can be sold fast and with the least difficulty. The variety of components often needed by experimenters and other nonprofessionals are not carried in stock. Experimenters, CB'ers and do-it-yourselfers are not welcome because they often ask for assistance or advice which the dealer is not prepared (or able) to give.

The second type of parts house specializes in hi-fi or CB gear and stocks only those IC's, transistors, tubes, resistors, capacitors and other components that might be required to maintain the equipment it sells. You may find "blister packs" of general-purpose semi-conductors and other parts for beginning students and experimenters but you are not likely to find all the parts needed for an advanced construction project.

Mail-order houses — the third type — are generally the best source of parts for advanced construction projects. Most of these serve all phases of the electronics industry, schools and experimenters and are more likely to have the parts you need in stock. If a component is out of stock, they can supply it without undue delay. Note well that not all mail-order dealers carry all brands of components and some will carry the full line of a particular manufacturer while others do not.

More and more parts manufacturers are gearing their product and sales toward the industrial market and sell only through wholesale (OEM-type) distributors. With the exception of some mail-order houses many distributors are basically wholesalers who have a minimum-order sales policy. That is, they don't accept orders totaling less than a specific amount — usually $8 to 10. RADIO-ELECTRONICS has turned down many articles on equipment requiring a special but inexpensive part obtainable only through sources which have a minimum-order figure many times the value of that part. If a special part for a construction project cannot be readily obtained at reasonable cost, there is no place for that story in RADIO-ELECTRONICS.

If you want to complete your construction projects without the frustration of not being able to locate a critical component, you will have to rely more on mail-order houses. Naturally, you will have to await the arrival of the parts from the mail-order house but this is not nearly as frustrating as waiting for one part ordered through a local dealer.

I have nothing against local dealers. It is just that most are specialists who do not stock the type of parts required for many of your projects.

If you enjoy constructing your own electronic gear from magazine articles, here are a few tips we hope will enhance your enjoyment of future projects.

(1) Before beginning a construction project from a magazine, read the article and parts list again. Then, go over it once more. Often, in your anxiety to get started, you overlook authors' or editors' notes on obtaining a particular part.

(2) Obtain catalogs from at least six mail-order houses. Study them until you are familiar with the types and brands of components listed in each.

(3) When ordering, follow instructions to the letter. List page and stock numbers and a description of each (continued on page 14)

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New&Timely

(continued from page 2)

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tion TV (STV) is the first
system granted technical ap-
proval by the Federal Com-
munications Commission. Jo-
seph S. Wright, Zenith's board
chairman, said, "With FCC
technical approval, Teco, Inc.,
Zenith's licensee for com-
mercial development of Phone-
vision in North America, is in
a position to implement plans
to introduce this new box
office TV service."

Under the FCC rules for
over-the-air STV, it was
noted, no station application
for regular STV broadcasting
may be granted unless the
station has an agree-
ment to use an FCC-approved
system.

Electrons Control
Body Temperature

ERLANGEN, GERMANY—Oper-
a
tions on the heart and aorta
are generally performed with
the aid of a heart-lung ma-
chine. This ensures a proper
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New & Timely

(continued from page 12)

the patient's blood circulation intact. In special cases it is advantageous to lower the temperature of the patient's body during such complicated operations to reduce metabolism as much as possible and thus any post-operative consequences.

Scientists of the Siemens Research Centre in Erlangen (Federal Republic of Germany) have developed a continuous-flow heat exchanger of stainless steel (capacity 190 ml) which is fitted with Peltier cooling blocks (Peltier elements). A Peltier element consists of two semi-conductor legs of which the one is of the n-type and the other of the p-type. The two legs are connected at the face end by a copper strap. If dc flows in a given direction through the Peltier element, the copper strap absorbs thermal energy from the immediate surroundings, and cools off. The thermal energy absorbed by the cold side of the element—boosted by the current input—is given off on the other side of the element. The Peltier element thus "pumps" thermal energy from its cold to its warm side. The direction of the heat transfer can be reversed by reversing the direction of the current. A Sirigor cooling block is made up of several Peltier elements connected meander-fashion in series. Although the elements are connected in series electrically, they are in parallel thermally so that all the cold copper straps are on the one side and all the warm ones on the other side of the cooling block.

Editorial

(continued from page 6)

item when asked for on the order blank. Failure to do this will delay your order. Parts are frequently stocked in bins by catalog number, not by name, brand or part number. If you omit catalog numbers, someone will have to find time to look up the numbers and perhaps rewrite your order in a way that it can be handled by workers in the stockroom. In many cases, the person filling your order may not be able to tell a resistor from an i.f. transformer.

(4) Read new-literature columns in this and other magazines and write for catalogs and data sheets from components manufacturers. Many contain substitution guides, dealer and distributor listings and a wealth of technical information that will prove useful when selecting parts for future projects. You should have catalogs from at least two makers of each type of common electronic component.

(5) Pay particular attention to the parts list when planning a construction project from an old issue. Some parts which were readily available at time of publication may have been discontinued and there may not be an acceptable substitute. If there is any indication of special components in the project, check with the manufacturer or listed source of supply before purchasing any of the more readily available components.

(6) Don't rush headline into a construction project whose operation depends on a special surplus component such as a meter, relay, transformer or filter network. Make sure that the part is available before you begin. The source may have "dried up" and there is no commercial equivalent available.

Construction projects can be fun when you know the ropes. Follow the tips above and then dig in and tackle some of the equipment on the following pages.
Go ahead. Compare them with any others. These new all solid state Precision generators from B&K do more things and do them better. Because B&K makes products that really work.

Take the Model E-200D RF generator, for instance. You get wide-range coverage from 100 KHz to 216 MHz in 5 bands. A completely shielded RF output provides continuously variable monitored attenuation...calibrated to less than a microvolt on fundamentals to 54 MHz. And built-in crystal calibration assures you of frequency accuracy.

You get a giant 5" vernier dial with anti-backlash 12:1 gear ratio. So it's easy to obtain and repeat the results you want.

Or consider the Model E-310B. This new sine/square wave generator gives you step attenuation of both outputs. And a variable control lets you adjust continuously from maximum down to less than 0.25 millivolt.

It also features extended frequency ranges: 20 Hz to 2 MHz for sine wave, 20 Hz to 200 KHz for square wave. Plus constant-voltage output over each band and band-to-band.

Both new generators are based on earlier models that made us famous. Now you can have all the additional benefits only transistors make possible.

See the new solid state generation from B&K at your local distributors. Or write us for complete details.
When your servicemen have to solder, make sure they...

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13126/1H 170
For a long time, old-timers like me have said that a voltage "going up" means it goes more positive, and "going down" means it goes more negative. This worked fine with tubes because plate voltages were always positive and that's what we did most of the work with. Now, we've got to change our "thinking"!

We've got to learn to think "goes more positive" or "goes more negative." This is true even in tube circuits used nowadays, let alone solid-state stuff. And, the way we think about the circuits determines the speed with which we can analyze their operation and find faults in them. You do not do this with voltmeters or any kind of test equipment. You do this with that mass of brain mud between your ears! The proper thinking-methods will speed up the process of getting the information, forming a conclusion, and making the diagnosis.

There are circuits we work with every day that demand this type of thinking. The bias on color picture tubes is a good example. How many times have you been fooled by a "no raster" symptom, only to find out (sometimes very painfully) that there was plenty of high voltage after all? A biased-off picture tube has a dark screen, but lots of high voltage on the ulo to bite unwary fingertips.

All of these things can be traced back to circuit reactions as simple as current flowing through a resistor, Ohm's Law. In the video output stage used in millions of color TV sets, the voltage on the video output tube plate and CRT cathode are determined by the amount of current flowing through the plate load resistor. (Incidentally, we can still say "up" and "down" speaking of current, no matter which way it's flowing!) If the video output tube plate current increases, the voltage at the plate goes more negative. So the cathode of the picture tube becomes more negative since it is directly coupled to the plate.

A negative-going voltage on the cathode of all vacuum tubes is the same as a positive voltage on the grid. (We're assuming now that the grid voltage is fixed.) The important relationship here is the voltage between cathode and grid (not to ground). This determines the conduction of the tube. So if the cathode goes toward a greater negative voltage, the grid-cathode voltage has gone more positive. Result: the raster gets brighter (more beam current).

A positive-going cathode voltage has the opposite effect—the grid has been made more negative with respect to the cathode and beam-current is reduced. A dimmer raster results.

There's a dandy example of the value of thinking of the circuit reactions in terms of "positive-going" or "negative-going" instead of "up" or "down." The terms will remind you of the effect that the voltage-changes have on the circuit operation. You can get the same effects in the CRT grid circuits, of course. A more positive plate voltage on a color reference amplifier tube (directly coupled to one CRT grid) makes that grid more positive. What does a more positive grid voltage do in all tubes? Makes them conduct more current. In this case the raster will be brighter, because the CRT is being biased to draw more beam current.

This kind of thinking can be most helpful in solid-state circuits. Why? Because we can find transistors with "negative-plate voltages!" The collector voltage of a npn transistor in a common-emitter circuit is negative. If this transistor conducts more heavily, it's adding more current through its load resistor, the voltage will go more positive. (Oh yes it will too. Assume that the normal collector voltage was —15 volts, from a —20 volt supply through a load resistor. If this change to —5 volts a —5 volt reading is 10 volts more positive than —15 volts, isn't it?)

In cases like this, you might have a —20 volt supply, with only 5 volts dropped across the load resistor in normal operation. This leaves us —15 volts on the collector. A higher collector current might cause a 15-volt drop across the load resistor, leaving us only —5 volts on the collector. This is a symptom, not a cause. The cause is whatever is making this transistor draw more positive current.

(continued on page 86)
When one gun fades restore color balance with a Perma-Power single-brite COLOR GUN CONTROL

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Correspondence

PROGRESS WITH SCR'S

It is noted that the Noteworthy Circuits item, "Portable SCR Tester," in the March 1970 RADIO-ELECTRONICS (page 94) reproduces, with editorial changes, GE Application Note 201.3.

Our policy is to request that our application notes not be copied without permission. New product introductions constantly require updating of circuits, and when requests are received to publish circuits disclosed in our publications, it affords us an opportunity to inform the author of the most recent advances in the art.

Our suggestion for replacing the C5U SCR described is to employ a C106Q (any type 1-4). For the electronic hobbyist, it may be convenient to pick up a GEMR-5.

CAROL O. THOMAS
Patent Attorney
General Electric Co.

We appreciate the suggested update for the SCR tester, it will aid our readers. Our normal procedure is to credit sources of information that appear in RADIO-ELECTRONICS. The oversight in this instance was due to the absence of any credit or reference to General Electric in our file copy.

FORGOTTEN AMMETER

The auto-transformer story by James Squires (Sept. 1970) was of interest because I don't believe a technician's bench should be without one.

However, after reading the item, there was one item that should have been included in the circuit—an ac ammeter. The ammeter is necessary to know the power drawn by the unit under repair, and quickly lets you know if there's a short in the work.

Use an ac ammeter and voltmeter with a Variac variable line transformer built into my bench, connected to an output plug. By starting the transformer at zero, I've saved many receivers from burning with this rig. I advise anyone building one of these units to wire in the ammeter. I wired a shunted switch across mine so it reads to 2 and 20 amps.

PETER LEGON
Malden, Mass.

(continued on page 24)
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CORRESPONDENCE
(continued from page 18)

CAR TUNE-UPS

As the article stated, I tried to use a vtvm instead of a vvm ("Tune Up With A Vvm," April 1970 Radio-Electronics). However, my vtvm read in the minus 0 range and, reversing the polarity, in the plus infinity range. What creates this problem?

ROBERT KILL
Worthington, Ohio

The article on tune-up with a vvm was interesting, but I feel that aligning the time marks by pulling on the fan is unsafe. If the blades are bent and then restraightened, metal fatigue could cause the blade to break when the engine is operated.

ROBERT DAVIS
Ottawa, Kan.

My guess is that the vtvm user has a bad diode. The blocking diode, preferably of several hundred PIV, should be checked for a large resistance difference between forward and reverse values. The vtvm voltmeter function will show the polarity of the coil terminal that connects to the points, and the vtvm and diode can be connected in series properly.

When you use the vtvm you will need to modify your calculations for the dwell setting. Vtvm's read down-scale from infinity rather than upscale as vvm's do. Make the calculation for the voltage corresponding to your dwell setting exactly as described on page 24 in the article.

I got a reading of 80 on a 120 volt scale for a 30° dwell. For a vtvm subtract the voltage you get from the full-scale value and use that as the mark to set by. For example, 120 — 80 = 40. With the values given in the article, you'd use the 40V position on the 120V scale with a vtvm, not 80V.

The fan-blade problem is correct in some cases, but most blades I've encountered are heavy enough to pull the engine into place for timing. A blade would be dangerous if restraightened, but common sense should guide you here.

LOUIS E. FRENZEL, JR.
Silver Spring, Md.

HELP!

I have endeavored, thus far without success, to locate a copy of the instruction manual and schematic drawing for a Smith-Lawrence, Inc. Model 851, Potentiometric DC Voltmeter.

I would appreciate any assistance your readers can provide in making this manual and schematic available.

RAYMOND HADLEY
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A. Is the number odd?
B. Is the number in the first or third vertical row?
C. Divide the number by four, is the remainder less than two?
D. Is the square number greater than 3?

As in answer each question, you move the control into the appropriate position ("A", "B", "C", "D"). Each answer is instantly revealed and shown on the board. If you have all the numbers right, you have "Sunk" the Battleship.

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November 1970
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**APPROVED FOR VETERANS**

Accredited Member: National Association of Trade and Technical Schools; National Home Study Council.
To call it "an amplifier" would be like calling a Porsche "Basic transportation."

There is unusual satisfaction that comes from fulfilling a proscac task in a far from prosaic manner.

Hence this amplifying system: the Sony TA-2000 professional preamplifier and the Sony TA-3200F power amplifier. Together, they perform all an amplifier's standard tasks in a satisfyingly impeccable manner; but their 67 levers, switches, meters, knobs and jacks allow you to perform some interesting functions that are anything but standard.

**Dual-purpose meters.**

The two VU meters on the preamplifier front panel, for example, are no more necessary than a tachometer on an automobile. But they do serve the dual purpose of simplifying record-level control when the TA-2000 is used as a dubbing center, and of allowing you to test your system's frequency response and channel separation (as well as those of your phono cartridge), and to adjust the azimuth of your tape heads.

**A broadcast/recording monitor console in miniature.**

The TA-2000 resembles professional sound consoles in more than its VU meters. In addition to the 20 jacks and seven input level controls provided on its rear panel for permanent connections to the rest of your hi-fi system, the TA-2000 boasts a professional patch board in miniature on its front. Thus, you can feed the inputs from microphones, electric guitars, portable recorders or other signal sources into your system without moving the preamplifier or disturbing your normal system connections in the least. And a front-panel Line Out jack feeds signals for dubbing or other purposes into an external amp or tape recorder, with full control of tone and level from the front-panel controls and VU meters.

The tone correction and filtering facilities are also reminiscent of professional practice, allowing a total of 488 precisely repeatable response settings, including one in which all tone controls and filters are removed completely from the circuit.

**The amplifier — no mere "black box."**

A power amplifier can be considered simply as a "black box" with input and output connections, a power cord, and an on/off switch; and such an amplifier can perform as well (or poorly) as the next one. But in designing the TA-3200F, Sony took pains to match the amplifier's facilities to the preamplifier's.

Thus to complement the TA-2000's two pairs of stereo outputs, the TA-3200F has two stereo pairs of inputs, selected by a switch on the front panel. Other front panel controls include independent input level controls for both channels, a speaker selector switch, and a power limiter (in case your present speaker should lack the power-handling capacity of the next one you intend to buy).

**Circuitry unusual, performance more so.**

The single-ended, push-pull output circuitry of the TA-3200F amplifier is supplied with both positive and negative voltages (not just positive and ground) from dual balanced power supplies. This system allows the amplifier to be coupled directly to the speakers with no intervening coupling capacitors to cause phase shift or low-end roll-off (A switch on the rear panel does let you limit the bass response below 30Hz if you should want to, otherwise, it extends all the way down to 10Hz.)

The individual stages within the amplifier are also directly coupled with transformerless complementary-symmetry driver stage, and Darlington type capacitorless coupling between the voltage amplifier stages.

As a result, in part, of this unique approach, the TA-3200F produces 200 watts of continuous (RMS) power at 8 ohms, across the entire frequency range from 20 to 20,000 Hz; IHF Dynamic Power is rated at 320 watts into 8 ohms (and 500 watts into a 4-ohm load). But more important by far is the quality of the sound; intermodulation and harmonic distortion levels are held to a mere 0.1% at full rated output, and 0.03% at the more likely listening level of one-half watt. The signal-to-noise ratio is an incredible 110dB. And the full-damping factor of 170 is maintained down to the lowest, most critical frequencies (another advantage of the capacitorless output circuit).

The companion TA-2000 preamplifier also boasts vanishingly low distortion and a wide signal-to-noise ratio, but this is less unusual in a preamplifier of the TA-2000's quality (and price). What is unusual is the performance of the phono and tape head preamplifier circuits; for though they have sufficient sensitivity (0.06mV), frequency range from 20 to 30Hz (or poorly), the lowest-output cartridges (even without accessory transformers), these preamplifier circuits are virtually immune to overload— even with input signals 80 times greater than normal.

**Their sole vice: they are hardly inexpensive.**

Of course, at a price of $329.50 (suggested list) for the TA-2000 preamplifier, and $349.50 (suggested list) for the TA-3200F power amp, this system cannot be considered other than a luxury. But then, it was intended to be. For there are those to whom fulfillment of prosaic tasks is unsatisfying. And among them are not only many of our customers, but also many of our engineers. Sony Corporation of America, 47-47 Van Dam Street, Long Island City, New York 11101.
Police statistics show that your chances of becoming a burglary victim are growing rapidly. Last year in Los Angeles more than 65,000 burglaries were attempted as compared to only 36,000 in 1960. Across the nation, burglary was the most frequently committed crime. A properly installed burglar alarm will help shift the odds in your favor. This article shows how these systems work and what they can do for you. Turn the page and see...
WHILE CRIME DOESN'T PAY, FEAR OF CRIME DOES. AND while no American likes to admit that he isn't safe in his own home, a great majority of home owners realize they aren't and are installing burglar alarm systems at a rapid rate.

Last year in Los Angeles, the city with the highest crime rate in the nation, more than 65,000 burglaries were attempted as opposed to only 36,000 in 1960. Of the 65,000, over 36,000 were committed against residences and apartments. Across the nation, burglary was the most frequently committed crime.

Statistics such as these are responsible for the boom in burglar or intruder systems. Among the vast array of available devices you're bound to find the one that fits both your pocketbook and needs. Though not all of the available systems will do an adequate job of protecting you, most of them will when properly installed.

What do you really want to protect?

Since the majority of home owners and apartment dwellers insure the contents of their homes against theft, it's really our lives that we wish to protect. This should be kept in mind when considering a system. Protecting life is the single most important function of an alarm system. Unfortunately, not all of the available devices are designed with this in mind. However, a combination of techniques will do the job well.

How sophisticated a system you need depends upon the size and design of your home and how much you are willing to spend. The variety of products on the market range from battery operated single-station units that attach to a door or window, to space-age ultrasonic and microwave detectors similar to the ones used in Vietnam to detect troop movements. They'll do everything from ringing a bell and lighting a light, to dialing the local police station. Costs vary just as widely, ranging from $5.95 to over a thousand dollars.

Unless you want to protect valuable art treasures, jewels, or furs, a self-installed system that will adequately protect your family should run from $100 to $500 for a one-family house and from $100 to $200 for the average four-room apartment.

What you'll get for this amount of money is protection from the amateur or petty thief. Police officials claim that this type of thief scares off quickly when an alarm goes off or a light suddenly flashes on. And if this type of thief knows your home is protected, he may not even attempt to enter it. Why should he take a chance when your neighbor's home is unprotected.

Vandals will also be kept out of your home. And while vandals are not generally after monetary items, they can give you a bigger headache than thieves. A recent vandalism case reported in Long Island, New York resulted in over $10,000 worth of damage to a house, performed over a period of one week while the owners were away on vacation.

Petty thievery and vandalism are often committed by neighborhood teenagers. So upon successful installation of your system, demonstrate it for the neighbors. Word will quickly spread that your house or apartment is protected.

What's Available to do the Job?

Basically there are three different types of systems that are suited to the homeowner. One or a combination of two techniques will afford you adequate protection no matter what the design of your home or apartment. The three systems include electromechanical or perimeter, photo-electric, and ultrasonic and rf systems.

Electromechanical or perimeter systems are the least expensive, but the most difficult to install. Photo-electric devices are seemingly easy to install and are comparable in price to perimeter systems, generally, however, they will only cover a small area. Total protection generally necessitates several units, costing more money.

As might be expected, ultrasonic and rf units are the costliest, yet they can virtually be taken out of their box and plugged in for instant protection. So their high cost is offset by low installation time and costs. The system you select will largely depend upon how much you want to spend and the design of your home.

Fire protection is an added feature of most systems. Low cost sensors ($5 to $10) will convert your intruder system into an effective fire alarm system as well.

Perimeter systems

Though the most difficult to install, electromechanical or perimeter systems offer the greatest degree of protection and flexibility for the money. The reason they are sometimes referred to as perimeter systems is that a properly installed system will surround your home with security. In a complete system all doors and windows are fitted with magnetic or contact switches that trigger the alarm when one is broken. Unless someone saws a hole through your roof or cuts a hole in a sliding glass door, the average thief is detected by such a system.

Since nearly every room is wired, adding fire protection to a perimeter system is easy. Heat or smoke sensors are simply wired into the same circuit the switches are on.

A basic perimeter system consists of a central control box or console (sometimes containing a key switch, batteries and horn) external bell or siren and window and door switches. An advantage of a perimeter system is that it may easily be expanded as desired or required. Some perimeter systems on the market today are single-station devices, that is, they
only protect a single door or window. However, in most cases, they can easily be expanded to afford complete protection.

Low cost perimeter systems are generally battery operated and contain an integral siren or horn, located in the console or control box. Before you purchase any system, ask to hear the alarm. Some units aren't loud enough to wake a sleeping baby.

More expensive and sophisticated perimeter alarms are both ac and dc operated. If there is a power failure or an intentionally cut power line, such units switch over to dc automatically and the system is still much easier.

To provide complete coverage, look for a system mounted on a wall or ceiling. It should contain both an internal and an external sensor. The external signal will alert neighbors and police while you are away. The internal alarm will tell the intruder he's been detected and that it's time to get out—fast! Incidentally, if you plan to go away, instruct a dependable neighbor on the operation of your system, in case it accidentally goes off. And if you add fire protection to your system, use a different sound alarm so you can easily distinguish between a fire and an intruder. If your system does not allow you to hook up an alarm, a simple relay set-up will do the trick.

The actuating key switch is generally located in the control box inside your home or apartment. A delay of from 10 to 30 seconds is built into many systems, allowing you ample time to enter your house and deactivate the system (a feature of some systems is an instant or delay switch for use when you are inside the house). This delay works in reverse when leaving. You'll have 10 to 30 seconds to make your exit. Activation upon closing of a door is another variation in some systems.

If for some reason you prefer to have the actuating key switch outside the house, optional externally mounted key panels are available with some systems. If you intend to eventually install a complete and comprehensive system, look for a manufacturer that offers a complete line of accessories. It will make add-ons much easier.

An important component in any system and especially easy to include in a perimeter system is a series of panic buttons, which upon activation, will sound the signalling device. For the greatest degree of protection, the panic button should sound an alarm even when the system is deactivated. Thus, should a housewife not have the system activated during the day when doors and windows are open, she can sound the alarm should an unwelcome visitor show up. Place panic buttons throughout the house and especially in rooms where a direct view of the entrance foyer is not available.

Windows and doors can be protected by either of three basic means. Magnetic or reed actuated switches have become very popular because of their reliability and ease of installation. They are especially convenient for doors, and cost around $3.50 each. For windows the lever or contact switch is used for its simplicity and low cost (about $1.50). Break-away wire is the most economical “switch” but it can only be used where a particular window or door is virtually unused. Break-away wire is very thin and can not be seen at night. The device you choose will depend upon each situation. More exotic and costly switches which are activated by vibration or pressure are available but they are prone to false alarms. Use them only where such protection is essential.

Naturally, the more you spend the more features you'll get. For instance, some of the deluxe systems have power indicators, battery life testers, etc. The one thing you should add to your system is fire protection. This is done simply by wiring into the same circuit, small thermostats or temperature sensors which activate when a room reaches 130 to 135° F. For warmer climates or rooms, 190° units are also available. Trouble spots like boiler rooms, kitchens and laundry rooms are likely places to install these units.

If you are planning to install a perimeter system in a new, yet to be constructed house, ask the builder if

you can wire the skeleton before the plasterboard is installed. This will greatly reduce the amount of work you'll have to do in cramped attics and crawl spaces.

**Photoelectric systems**

Unlike perimeter systems, photo-electric protection devices are designed to cover a single room or area. To provide the breadth of protection afforded by perimeter systems a photo-electric set-up would require several sets of photo-electric pairs. In addition, the optical alignment of such systems is tricky, and must be doublechecked. There are instances though, where a single photo-electric system strategically placed, will cover the critical portions of a home and at a very nominal cost.
Photo-electric systems usually consist of a light source and photocell sensor. While older systems used easy-to-fool incandescent light sources, modern versions utilize infrared, ultra-violet and modulated light sources which the average thief will find very difficult to "jimmy". A few systems even use relatively new IR (infrared) light emitting diodes as light sources. These gallium-arsenide diodes emit IR in the range of 9100 angstroms and provide the reliability and low power drain of semiconductors.

All photo-electric systems operate on the same basic principle—when a beam is interrupted, the resistance of the photo-cell increases and triggers an alarm. IR, UV (ultraviolet) and modulated beams are used to make the devices tamper proof.

Burglars vs space age electronics

It's doubtful that any thief will find a way to get by the space age protection afforded by ultrasonic, radar and microwave intruder systems. Representing the high technology portion of the burglar business, these devices are difficult to fool, easy to install but relatively costly to buy. Prices range from $75 to about $500.

Each of the three systems operate on the same principle—the Doppler effect. Simply, the Doppler effect dictates that movement within an energy field of an rf or ultrasonic signal will cause a change in the phase, frequency and amplitude of the energy pattern.

Operation of a typical system is illustrated by the diagram shown at right. In this system, an oscillator generates a 20 to 40 KHz signal (depending upon the particular model) which is fed to a piezoelectric transmitter and subsequently radiated throughout a room. Reflected ultrasonic waves from moving and stationary objects are received by a second piezoelectric crystal. Since a moving object causes a change in the characteristics of the original signal, a simple detector/reference circuit detects the difference in the signal and triggers an alarm. Filters help reduce the possibility of false alarms caused by extraneous signals.

Solid state electronics has made it possible to package these systems in a single box, looking something like a small hi-fi speaker. Other systems require the use of detectors (which must be connected to the transmitter) located in various parts of a room. Except for these multi-part setups, all the wiring required for a single-part system are bell, light and panic button circuits similar to those used in perimeter or photo-electric systems. Some of these systems are small enough, and portable enough, to take along on a trip to guard your motel or hotel room.

The basic difference between ultrasonic, radar and microwave systems lies in the frequency range within which they operate. Ultrasonic units operate at about 20 to 40 KHz, radar in the 450 to 500 MHz region, and microwave goes up in frequency where radar leaves off. These different frequency ranges bring about some inherent advantages and disadvantages for each.

Ultrasonic systems seem to be best suited for the average home. Their greatest advantage is that the radiated signal will not penetrate walls and is thus confined to a single room and is not disturbed by outside movement. However, so effective is the protection within a room that air turbulence caused by hot air or air conditioning vents can cause a false alarm under some circumstances. Reducing the sensitivity setting will control this problem. However, to help overcome these unwanted problems, air turbulence compensators and sensitivity adjustments are provided. You'll have to find the right setting for your room.

Radar or higher rf intruder systems have their problems too. Their high frequencies and shorter wave lengths permit the signals to penetrate walls and glass planes, thus posing the problem of outside movements interfering with the system's operation. Also, commercial communications systems such as citizen's band and police radios might possibly trigger an alarm.

Microwave devices operating near and in the Gigahertz region are not at all disturbed by air turbulence or commercial communications systems. However, these units are generally more expensive than ultrasonic units and their radiating pattern is spherical in shape and therefore does not "fill" a room as ultrasonic units do.

Coverage accorded by all three techniques is about the same ranging from 15' x 15' areas for the lower cost units, to 35' diameter areas for the higher power ones. Temperature and humidity changes will cause some variation in their range. Long corridors are relatively easy to protect with some units able to secure a 150-foot long space.

Good protection is afforded by a combination of perimeter and ultrasonic units. There are several ultrasonic units available that have provisions for adding perimeter system for supplementary protection. But no system is better than the installation job. Carefully consider possible entry points and make sure they are covered. Remember, an alarm positioned where it can't detect an intruder is worthless.

For a complete listing of burglar alarm manufacturers turn to pages 83, 94, 95, and 97. If you come up with a manufacturer we haven't listed, let us know. Write Burglar Alarms, c/o Radio-Electronics, 200 Park Ave South, New York, N.Y. 10003.
30 new IC circuits you can use

by R. M. MARSTON

Part II: Versatile four-transistor IC operates from dc to vhf. Build ultra-sensitive switches, ice alarms, smoke alarms, simple robot brain, automatic car light controls, heater and air conditioner controls, and more.

Ultrasensitive switches

The CA3018 is ideally suited for ultra-sensitive switching applications. It can be used as a temperature-operated switch triggered by thermal changes of less than 0.5°C. It can become a light-operated switch that senses changes in light level too small to be detected by the human eye, such as those caused by a puff of cigarette smoke at a range of several feet.

In this section, we present seven ultrasensitive switching applications. All of them use the same basic operating principle.

In the basic circuit of an emitter-coupled dc differential amplifier (Fig. 16-a) Q1 and Q2 share common-emitter resistor R5. Transistor Q1's base is set at 6 volts by R1-R2; Q2's base is variable about 6 volts via R3-R4.

Due to the basic emitter-follower actions of Q1 and Q2, the voltage at the top of R5 is about 0.65 volt less than the larger of the two base voltages, So, if the two bases are at identical potentials, Q1 and Q2 are equally forward-biased and conduct by equal amounts. Under this condition the circuit is balanced.

If, on the other hand, Q2's base is made more positive than that of Q1, Q2 becomes more heavily forward-biased, and turns on harder than Q1. Simultaneously, R5's voltage rises in proportion to Q2's base voltage, reducing the forward bias of Q1.

Transistor Q1 becomes less forward-biased under this condition. Thus, the emitter coupling of the two transistors causes Q2 to conduct more and Q1 to conduct less when Q2 base voltage is increased. The reverse occurs when Q2's base voltage is lower than Q1's.

Note that the degree of conduction of the two transistors is dictated mainly by the difference between the two base voltages, rather than by the absolute base potentials. Both Q1 and Q2 give a current and voltage gain between collector and base. If each transistor has a voltage gain of 100, a differential change of 10 mV on Q2's base causes a 1-volt change on Q2's collector.

Resistors R1 through R4 are in the form of a bridge, so the balance point of the circuit (at which Q1 and Q2's base potentials are equal) is determined solely by the relative values of these resistors and is independent of variations in supply voltage.

Since Q1 and Q2 are connected as a differential amplifier, thermal variations in the forward base-emitter voltages of the two transistors automatically cancel, particularly if Q1 and Q2 are part of an integrated circuit. The balance point of the circuit is thus independent of ambient temperature.

Finally, if R4 is adjusted so the circuit is near balance, Q2's collector voltage can be shifted 1 volt by differentially increasing Q2's base voltage by only (say) 10 mV. Since a standing bias of 6 volts is normally applied to Q2's base via voltage divider R3-R4, this shift in voltage is obtained by changing R4's value by less than 0.2% The circuit balance is thus very sensitive to changes in the bias values of resistors R1 to R4, but is virtually independent of variations in actual supply voltage and ambient temperature. As a result, this circuit can be used as the basis of an exceptionally stable but sensitive resistor-operated electronic switch.

The basic differential circuit can be modified to act as a regenerative switch that drives a relay output (Fig. 16-b). Here, Q2's collector is direct-coupled to the base of npn transistor Q3, and Q3 drives a relay load in its collector. Part of Q3's collector voltage is also fed back to Q2's base via R7 and the R4-R8 junction.

Suppose R4 is adjusted so that Q2
By selecting the value of R8, the triggering backlash of the circuit can be varied to suit individual needs—if R8 is given a very small value, the circuit can be adjusted so the relay turns on when R4 is 10,000 ohms, but does not turn off again until R4 is reduced to (say) 9900 ohms, giving a backlash of 100 ohms.

The overall action of the circuit of Fig. 16-b is fairly simple. R1 through R4 are wired in the form of a bridge which is balanced when R4 equals 10,000 ohms. If R4 is (say) 0.5% or more below 10,000 ohms, the bridge goes out of balance in such a way that Q2 is turned off, so Q3 and the relay are off also.

If, on the other hand, R4's value increases to within less than 0.5% of 10,000 ohms or is made greater than 10,000 ohms, the bridge either balances or goes out of balance and turns Q2 on. Either way, a regenerative action is set up between Q2 and Q3, and the relay is driven sharply on.

The circuit drives the relay on whenever R4's value reaches a pre-selected magnitude. This value can be set with an accuracy of a fraction of a percent. By replacing R4 with a thermistor or a LDR (light dependent resistor), the circuit can be used as an ultrasonic temperature- or light-operated electronic switch.

Fig. 17 shows how to use the CA3018 as a high-performance ice-warning indicator. Transistors Q3-Q4 (within the IC) are Super-Alpha-connected and used as one half of the differential stage (with input to pin 9). Transistors Q2-Q1, also Super-Alpha-connected, are used at the other half, with input to pin 3. Capacitor C1 prevents high-frequency instability of the differential stage, and C2 insures that supply transients do not cause erratic triggering of the regenerative section of the design.

In action, the relay goes on as soon as thermistor temperature falls below its preset value. The triggering temperature can be varied over quite a wide range via R3.

When the unit is used as an ice-warning indicator, R3 is set so the relay goes on as soon as R8 falls below 0°C. The thermal backlash of the circuit is 0.5°C—if the relay goes on when the temperature falls to 0°C, it goes off again when the temperature rises to 0.5°C.
+0.5°C. Backlash can be increased, if required, by increasing R7's resistance.

This circuit is very valuable as an ice-warning indicator for a car. The thermistor is mounted on the front under-side of the vehicle, about 10 inches above the road surface.

Alternatively, the unit can be used at home to turn on a heater when room temperature falls below a preset value.

The circuit can be modified for use as an excess-temperature indicator (Fig. 18). Now the relay goes on as soon as R8's temperature exceeds its preset value.

The circuit can be made to operate at any temperature between about 30°C to 110°C, using the thermistor types shown. Thermal backlash is 1°C at an operating temperature of 100°C. Other operating ranges require different thermistor types. Their resistive values must be between 2000 and 80,000 ohms at the selected trigger temperature.

This circuit can be used to give an indication of engine, gear-box, or brake-drum overheating in a car. It can automatically turn on air conditioning equipment, etc., when temperatures exceed a preset value.

Fig. 19 shows how the circuit can be made to work as an ultrasensitive light-operated switch, which turns the relay on as soon as light level falls below a preset value. The action can be modified so the relay goes on when the light rises above a preset value, by transposing R8 and R1.

Circuit sensitivity is so high that the relay can be operated by changes in light level caused by a puff of cigarette smoke at a range of several feet (changes too small to be detected by the human eye). As a result the unit makes a good ultrasensitive detector of smoke, fog, mist, etc. When used in applications of this type, the brilliance of the source illumination must be stabilized.

The circuit can be modified for use as a lighting slave (Fig. 20), which operates a car's parking lights automatically. It turns them on when the car is driven at night, and keeps them off when the car is used in daylight. It turns lights off automatically when the car is parked at night.

Power is applied to the unit via the car's ignition switch, and the relay contacts are wired in parallel with the vehicle's parking-light switch. R1-R11-R2-R3-D1-C1-C2 and R4 are wired as a dual-time-constant light-level integrating network to control the variable input voltage to the differential stage. The regenerative section of the unit has a moderately large value of backlash (via R10).

Because of the integrating network (when the car is in use) the relay is energized when the light level on R11 falls briefly below a value preset by R2. But the relay releases again only if the light fails to fall below that level again.

**FIG 20 (top)—TURN CAR LIGHTS ON automatically with this simple IC circuit.**

**FIG. 21 (center)—ELECTRONIC LOCK uses variable resistor as its adjustable key.**

**FIG. 22 (bottom)—DIFFERENTIAL BALANCE DETECTOR can be used as a robot brain.**

**NOTE: RY1, RY2—ANY -12V RELAYS, COIL RESISTANCE GREATER THAN 120Ω.**

**NOTE:** The IC's are not circuit sensitive.
NOTE: RY1, RY2 = ANY 12-V RELAYS WITH COIL RESISTANCE GREATER THAN 100Ω.

FIG. 23—HIGH-IMPEDANCE version of the differential balance detector of Fig. 22. Q7 and Zener diode D1 are used as a constant current source in this circuit.

within the following 30 sec; or so. The circuit is unaffected by sudden increases in light level, such as lightning flashes, etc. R2 is set so that the relay is on. It only turns off again when the light level reaching R11 rises to normal dusk or dawn values, and remains at those values for at least 30 seconds.

When the car is used at night, the relay and the lights turn on automatically in response to the low illumination on R11, and are unaffected by sudden transient increases in light level caused by street lights and lightning flashes. When the car is parked, power to the unit is cut off by the car ignition switch, so the lights go off automatically.

When the car is used in daylight, the lights are normally held off via R11. They will turn on automatically, however, if you drive under a bridge or through tunnel in which the light is below the preset level. They turn off again about 30 sec. after the vehicle emerges from the dark area.

Finally, when the vehicle is used on the threshold of dusk, the car lights turn on as soon as R11 illumination falls momentarily to a value perceptibly below that of the mean dusk level (when a shadow caused by a tree, a building or a passing vehicle falls on R11's face). The moderately high backlash of the circuit then insures that the lights do not turn off again until the illumination is held above the mean dusk level for at least 30 sec. (until the following morning). Thus, the car lights operate without the slightest sign of flicker, even on the verge of dusk (or dawn).

When this unit is used in a vehicle, the LDR (light dependent resistor R11) should be mounted on the dash, with its face looking into the car interior, so it is intensive to bright external lights. When setting the trigger levels of the circuit via R2, shade the LDR face so the relay goes on, and then carefully trim R2 so that the relay just turns off again under dusk or dawn illumination conditions. Because of the long time constants of the unit's integrator network, at least 2 min. should be allowed to elapse between each trial adjustment of R2.

The CA3018 can be connected as a precision resistor-operated electronic lock too (Fig. 21). The circuit uses two relays and has a regenerative switching stage wired to each side of the differential amplifier. The two sets of relay contacts are wired in series, so there is an output only when both relays are switched simultaneously. Resistor R2 is the "key" to the lock, and can have any value between 100 and 100,000 ohms. Resistors R1 and R2 must have nearly equal values. R3 balances small differences between R1 and R2.

The two relays go on simultaneously only if the key (R2) is the correct value. The tolerance of the circuit can be adjusted between about ±0% and ±5% via R2. If a wrong key is used, only one of the two relays operates—if R2 is too large. RY1 operates; if R2 is too small, RY2 is energized.

If this circuit is to be used as a practical electronic lock, the output should be taken via an additional 5- or 10-sec. delay relay, so the lock cannot be fiddled open with a variable resistor as the key.

The circuit can be modified for use as a differential balance detector (Fig 22) which drives RY1 and RY2 on simultaneoussly only if R3 and R4 are equal. Balance is independent of the absolute values of R3 and R4 between 470 and 20,000 ohms.

This independence is obtained by using Q3 and Q4 as a constant-current generator in the tail of the Q1-Q2 differential amplifier stage, so voltage gain is independent of pin 3's and pin 6's bias voltages.

This particular circuit is intended for use as a "brain" in a simple robot animal. Simply replace R3 and R4 with a pair of matched transistors or LDR's. In the LDR application, the LDR faces point toward a white line drawn on the ground, and the relay contacts control track-driving motors that move the animal forward. Relay RY1 operates the left-hand track, and RY2 the right-hand track.

When the animal is squarely over the white line, both LDR's are illuminated equally and both motors operate simultaneously to drive the animal forward. If the animal moves to the left of the line, the right-hand LDR is illuminated more than the left-hand one, RY2 drops out, and the left-hand track automatically moves the animal to the right until it is squarely over the white line again.

Similarly, if the animal moves to the right of the line, RY1 drops out, and the right-hand motor drives the animal back over the line. Thus, the animal automatically follows the path of the white line.

If you use thermistors, the animal can be made to automatically move toward or away from sources of heat.

Circuit sensitivity is very high, but can be controlled to a limited degree by R3. Sensitivity can be greatly reduced, if required, by wiring a fixed resistor (between 10 and 470 ohms) in series with the emitter of each of the differential transistors. Q1 and Q2.

Finally, in Fig. 23, there is a high-impedance version of the differential balance detector; here R3 and R4 can have any value between 1000 and 200,000 ohms. Resistor R1 must be 2 × R3, and R2 must be equal to R4.

This circuit is similar to that of Fig. 22, except that the differential stage is Super-Alpha-connected, and Q7 and Zener diode D1 are used as a constant-current source. (continued on page 69)
# NEW R-E EXCLUSIVE

**Kwik-Fix™ picture and waveform charts**

by Forest H. Belt & Associates®

## SCREEN SYMPTOMS AS GUIDES

<table>
<thead>
<tr>
<th>SYMPTOM PIC</th>
<th>DESCRIPTION</th>
<th>VOLTAGE</th>
<th>WAVEFORM</th>
<th>WHERE TO CHECK FIRST</th>
<th>PART</th>
</tr>
</thead>
<tbody>
<tr>
<td><img src="image" alt="No sync, accompanied by no sound" /></td>
<td>No sync, accompanied by no sound</td>
<td>—</td>
<td>—</td>
<td>T1, if in output circuit of a sync-and-sound-i.f. amplifier stage</td>
<td></td>
</tr>
<tr>
<td><img src="image" alt="Sound very weak; buzz and noise with volume control up" /></td>
<td>Sound very weak; buzz and noise with volume control up</td>
<td>—</td>
<td>—</td>
<td>C1 open, C6 shorted, T1 open, C7 open, C8 leaky, C9 open, R3 open</td>
<td></td>
</tr>
<tr>
<td><img src="image" alt="No sound" /></td>
<td>No sound</td>
<td>—</td>
<td>—</td>
<td>C2 open, R6 open, C4 shorted, R9 open, C10 shorted, L1 open, C10 leaky, C11 open, R5 open, R5 high</td>
<td></td>
</tr>
<tr>
<td><img src="image" alt="Sound very weak; may be distorted" /></td>
<td>Sound very weak; may be distorted</td>
<td>—</td>
<td>—</td>
<td>R2 open, R9 low, C10 leaky</td>
<td></td>
</tr>
<tr>
<td><img src="image" alt="No video; no sound" /></td>
<td>No video; no sound</td>
<td>—</td>
<td>—</td>
<td>T1, if in output of video detector or video amp stage</td>
<td></td>
</tr>
<tr>
<td><img src="image" alt="Sound critical at fine tuning; volume about normal" /></td>
<td>Sound critical at fine tuning; volume about normal</td>
<td>—</td>
<td>—</td>
<td>C3 open, C5 open, C6 open, C8 open</td>
<td></td>
</tr>
<tr>
<td><img src="image" alt="Sound critical at fine tuning; volume weak" /></td>
<td>Sound critical at fine tuning; volume weak</td>
<td>—</td>
<td>—</td>
<td>C4 open</td>
<td></td>
</tr>
<tr>
<td><img src="image" alt="Sound critical at fine tuning; volume above normal" /></td>
<td>Sound critical at fine tuning; volume above normal</td>
<td>—</td>
<td>—</td>
<td>C4 leaky</td>
<td></td>
</tr>
</tbody>
</table>
The Stages

Tube sets nowadays, both monochrome and color, use quadrature detectors almost exclusively. The 4.5-MHz sound i.f. is taken off following the video detector diode, or after the video amplifier in some sets. In color chassis, there's often a separate diode to detect sound i.f.

Generally, only one stage of sound i.f. amplification is needed. A sound takeoff coil, tuned to the 4.5-MHz sound i.f., couples the frequency-modulated station sound to an amplifier stage. The amplified signal is fed through another 4.5-MHz tuning network to the quadrature detector.

This special tube, a 6GX6 in the example used in this Kwik-Fix™, has a special grid (pin7) that is tuned 90 degrees away from the primary signal fed to the main control grid (pin1). That's where the term quadrature comes from.

The two influences (control grid and quadrature grid) on the electron stream to the plate make the tube sensitive to phase variances in the FM signal. The result is demodulation. The extracted sound modulation is taken from the plate load.

Sound-signal amplitude from a quadrature detector is enough that the usual stage of audio amplification is not necessary. The output goes to a volume control and then directly to a power audio output stage.

Signal Behavior

The 4.5-MHz sound intermediate frequency is the result of the sound and picture i.f. carriers (41.25 and 45.75 MHz) beating together in a diode (usually the video detector) at the end of the video i.f. stages. Their difference is 4.5 MHz. Takeoff transformer T1, resonant at 4.5 MHz, couples the sound i.f. signal to sound i.f. amplifier V1. C2 is a dc-blocking, signal-coupling capacitor; R1 is the grid load. C3 is the cathode bypass.

The amplified 4.5-MHz is coupled by C4 and tuned coil L1 to grid-pin1 (control grid) of V2. In this version, part of L1 is resonated by capacitor C6. The bottom portion of the winding, below the grounded tap, feeds some of the i.f. signal back to grid-pin2 of V1 through C5. This neutralizes the stage.

The quadrature element, quad-grid-pin-7, is tuned by resonant tank L2. Resistor R7 broadens the response of L2, and damps out any tendency to self-oscillation in that circuit of the tube.

I.f. signals that might remain in the plate circuit are eliminated by C10. The audio signal that results from demodulation is coupled to the volume control and output audio stage by C11.

Screen-pin-6 and cathode-pin-2 are held at signal ground by C8 and C7, respectively. C9 keeps the bottom of L2 at signal ground.

DC Distribution

Plate voltage for both tubes comes from a 240-volt dc source. Voltage for screen-pin-6 of V2 comes from a 140-volt dc source. R3 is the plate supply resistor for V1; R9 is the plate supply resistor for V2. Resistor R6 carries dc to screen-pin-6.

Both tubes are cathode-biased. R2 develops bias for V1 and R5 does it for V2. The bias path to grid of V1 is through R1. The bias path to grid of V2 is through R4 and a portion of the L1 winding.

Quad-grid-pin-7 may have a small contact-type bias developed across R8 by normal signal. With only a sweep signal applied, as when voltages were taken for this Kwik-Fix™, almost no detectable dc voltage develops—unless there's trouble in this circuit.
### DC Voltages as Guides

<table>
<thead>
<tr>
<th>Change to Zero</th>
<th>Very Low</th>
<th>Low</th>
<th>Slightly Low</th>
<th>Slightly High</th>
<th>High</th>
</tr>
</thead>
<tbody>
<tr>
<td>V1-cathode, pin-1 Normal 0.7 V</td>
<td>C3 shorted, C4 shorted, R3 open</td>
<td>C4 leaky, R3 high</td>
<td>C3 shorted, C4 leaky, R2 low, R3 high</td>
<td>R2 open</td>
<td>R2 high, R3 low</td>
</tr>
<tr>
<td>V1-grid, pin-2 Normal 0 V</td>
<td>C4 shorted, C4 leaky, R3 open, R3 high</td>
<td>C4 leaky, R3 high</td>
<td>C3 shorted, C4 leaky, R2 low, R3 high</td>
<td>R2 open</td>
<td>R2 high, R3 low</td>
</tr>
<tr>
<td>V1-plate, pin-3 Normal 75 V</td>
<td>C4 shorted, R3 open</td>
<td>C4 leaky, R3 high</td>
<td>C3 shorted, C4 leaky, R2 low, R3 high</td>
<td>R2 open</td>
<td>R2 high, R3 low</td>
</tr>
<tr>
<td>V2-cathode, pin-2 Normal 3.5 V</td>
<td>C7 shorted, R5 shorted</td>
<td>R6 open</td>
<td>C8 leaky, R6 high, L1 open</td>
<td>C8 open</td>
<td>R5 open</td>
</tr>
<tr>
<td>V2-plate, pin-5 Normal 120 V</td>
<td>C10 shorted, R9 open</td>
<td>C10 leaky, L2 open</td>
<td>C6 shorted, C7 shorted, C8 open, C10 leaky, R5 low, R5 shorted, L2 shorted</td>
<td>C5 open, R2 open</td>
<td>C7 open, R5 open, R6 open, R9 low</td>
</tr>
<tr>
<td>V2-screen, pin-6 Normal 100 V</td>
<td>C8 shorted, R6 open</td>
<td>C8 leaky, R6 very high</td>
<td>C7 shorted, C8 leaky, R5 low, R6 high</td>
<td>C8 open, R6 low, L2 open</td>
<td>R5 open</td>
</tr>
<tr>
<td>V2-quadr-grid, pin-7 Normal 0 V</td>
<td>C7 shorted, C8 open, R5 low</td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

### Notes:
*Goes negative
Use this guide and the Waveforms Guide to help you pinpoint the faulty part.
All voltages are taken with sweep-generator signal applied (see Waveforms Guide notes.)
Measure each of the seven key voltages with a vcm.

### Signal and Control Effects
Station signal doesn't have much effect on dc voltages in these stages, except as already described for the bias at quadrature-coil-pin-7. It becomes more negative with stronger station signals.
Tuning the coils affects voltages only slightly.

### Quick Troubleshooting
As is so common with stages that handle rf and i.f. signals, tuning is frequently the quickest way to spot a trouble. Before you try anything else, adjust the coils by the station-sound method.
Tune in the weakest TV station you get. Leave the antenna lead-in coupled loosely if necessary, but be sure you have a weak and noisy station signal.

For each, move across to the column that describes whatever change you find in that voltage.
Notice which parts might cause that change.
Finally, notice which parts are repeated in the combination of changes you find.
Test these parts individually for the fault described.

Peak T1 and L1 for the loudest sound with least noise; a compromise between noise and sound strength may be necessary at first. Then tune quadrature coil L2 for loudest sound with least distortion. Finally, go back over all three adjustments. Try to end up with loudest sound, least noise, and least distortion.

If any coil doesn't respond, suspect it or the components near it in the circuit. If the coils seem to tune ok, but still don't produce normal sound, make quick voltage checks at V1-plate-pin-3, V2-plate-pin-5, and V2-screen-pin-6. If those tests don't tell you anything, consult the Voltages Guide above and Waveforms Guide on the next two pages.

Sweep-response waveforms are used here because they are more informative of stage operation than any other kind of waveforms would be.

**Turn page for waveform guide**
WAVEFORMS AS GUIDES

WF 1 Normal 0.3 V p-p
Taken at V1-grid-pin-2, this is response curve of the input tuned circuit. As indicated by the markers, this may not be centered at 4.5 MHz, because responses of tuned circuits farther along affect the overall response of the sound i.f. section. Amplitude may vary 30 to 50 percent. Curve shape is what’s important (see Notes).

<table>
<thead>
<tr>
<th>V p-p low</th>
<th>V p-p high</th>
<th>V p-p zero</th>
</tr>
</thead>
<tbody>
<tr>
<td>C1 open</td>
<td>T1 mistuned up</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>T1 mistuned down</td>
<td></td>
</tr>
<tr>
<td>R2 open</td>
<td>0.25 V p-p</td>
<td></td>
</tr>
<tr>
<td>R2 open</td>
<td>0.3 V p-p R6 low</td>
<td></td>
</tr>
<tr>
<td>R2 open</td>
<td>0.2 V p-p C8 open</td>
<td></td>
</tr>
<tr>
<td>C6 open</td>
<td>0.3 V p-p</td>
<td></td>
</tr>
<tr>
<td>C6 open</td>
<td>0.15 V p-p C4 open</td>
<td></td>
</tr>
<tr>
<td>C6 open</td>
<td>0.3 V p-p C5 open</td>
<td></td>
</tr>
<tr>
<td>C6 open</td>
<td>0.2 V p-p</td>
<td></td>
</tr>
</tbody>
</table>

WF 2 Normal 3.0 V p-p
Taken at V1-plate-pin-3, this curve includes reflected response of L1. Also includes effect of feedback from extra turns of L1, through C5 to grid-pin-2. Still is not the shape you might expect, but note that 4.5-MHz (right-hand) marker is near center of irregular curve. The 4.08-MHz marker (left) gives you some reference from which to gauge the width of these response curves.

<table>
<thead>
<tr>
<th>V p-p low</th>
<th>V p-p high</th>
<th>V p-p zero</th>
</tr>
</thead>
<tbody>
<tr>
<td>C1 open</td>
<td>T1 mistuned up</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>T1 mistuned down</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>3.0 V p-p L1 mistuned</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>2.0 V p-p R2 low</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>1.5 V p-p C3 open</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>2.0 V p-p R3 low</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>0.5 V p-p C4 open</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>0.7 V p-p C4 leaky</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>0.4 V p-p C5 open</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>1.0 V p-p C6 open</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>0.5 V p-p R5 low</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>0.2 V p-p C7 open</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>0.3 V p-p R6 high</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>0.3 V p-p C8 open</td>
<td></td>
</tr>
<tr>
<td>C2 open</td>
<td>0.2 V p-p</td>
<td></td>
</tr>
</tbody>
</table>

WF 3 Normal 3.5 V p-p
Taken at V2-grid-pin-1, this curve includes the response with all effects of L1. Shape is more like you might expect, combining the responses of the entire sound i.f. stage (V1). Curve is still slightly to left (down-frequency) of 4.5-MHz center, but that’s normal. Don’t let sweep width (on the generator) be too wide when you look at this waveform, or the curve will take an inaccurate shape.

<table>
<thead>
<tr>
<th>V p-p low</th>
<th>V p-p high</th>
<th>V p-p zero</th>
</tr>
</thead>
<tbody>
<tr>
<td>L1 mistuned</td>
<td>L1 mistuned</td>
<td></td>
</tr>
<tr>
<td>L1 mistuned</td>
<td>3.5 V p-p R2 low</td>
<td></td>
</tr>
<tr>
<td>L1 mistuned</td>
<td>0.5 V p-p C3 open</td>
<td></td>
</tr>
<tr>
<td>L1 mistuned</td>
<td>1.0 V p-p C5 open</td>
<td></td>
</tr>
</tbody>
</table>

V p-p low | V p-p high | V p-p zero |
<table>
<thead>
<tr>
<th></th>
<th></th>
<th></th>
</tr>
</thead>
<tbody>
<tr>
<td>L1 mistuned</td>
<td>L1 mistuned</td>
<td></td>
</tr>
<tr>
<td>L1 mistuned</td>
<td>3.5 V p-p R2 low</td>
<td></td>
</tr>
<tr>
<td>L1 mistuned</td>
<td>0.5 V p-p C3 open</td>
<td></td>
</tr>
<tr>
<td>L1 mistuned</td>
<td>1.0 V p-p C5 open</td>
<td></td>
</tr>
</tbody>
</table>
## WAVEFORMS AS GUIDES

<table>
<thead>
<tr>
<th>V p-p low</th>
<th>V p-p high</th>
<th>V p-p zero</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td>0.7 V p-p C6 open</td>
</tr>
</tbody>
</table>

### WF 4 Normal 3.0 V p-p

This is taken at quad-grid-pin-7. You can use a direct probe, but the shape shown here is with a demodulator probe. The 4.5-MHz marker is in the center of the slanted straight portion of the curve. The top left is rounded slightly, the bottom right is pointed. Patterns below show how mistuning L2 affects this waveform.

<table>
<thead>
<tr>
<th>V p-p low</th>
<th>V p-p high</th>
<th>V p-p zero</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td>C1 open</td>
</tr>
<tr>
<td></td>
<td></td>
<td>L1 mistuned</td>
</tr>
<tr>
<td></td>
<td></td>
<td>1.0 V p-p R3 high</td>
</tr>
<tr>
<td></td>
<td></td>
<td>0.5 V p-p R6 high</td>
</tr>
</tbody>
</table>

### WF 5 Normal 15 V p-p

Taken following the dc-blocking capacitor at V2-plate-pin-5, this is the S-curve response of the quadrature detector. As you can see from the waveform patterns below, the shape and centering of the S-curve depends on all the tuned circuits in both stages. The slanted straight portion of the curve is very steep; the marker spreads itself all up and down it, making the marker somewhat indistinct.

<table>
<thead>
<tr>
<th>V p-p low</th>
<th>V p-p high</th>
<th>V p-p zero</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td></td>
<td>C1 open</td>
</tr>
<tr>
<td></td>
<td></td>
<td>L2 off CCW</td>
</tr>
<tr>
<td></td>
<td></td>
<td>T1 off CCW</td>
</tr>
<tr>
<td></td>
<td></td>
<td>R5 high</td>
</tr>
</tbody>
</table>

### NOTES:
* *from bottom of coil*
* Use this guide and the Voltages Guide to help pin down fault possibilities.*
* Before using this guide, carefully go through preliminary alignment of the sound i.f. section by the station-signal method (see Quick Troubleshooting in text).*
* Check all key waveforms. Readjust the scope input to suit the p-p size of each waveform.*
* Note amplitude. If it's low or high, check the parts listed under those columns.*

---

**NOVEMBER 1970**
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RE-24

Circle 21 on reader service card
by R. W. FOX

LAST MONTH WE INTRODUCED YOU TO the programable unijunction transistor (PUT) and showed you several practical applications of this device. Now let's continue and see how the PUT can be used in a burst oscillator, ring counter and SCR trigger circuits.

In some cases, the PUT is flexible enough to serve both as the timing element and load current device at the same time. One such example is shown in Fig. 12. When the circuit is energized, the capacitor is charged through the A14 diode and the 22K resistor to full supply voltage. When the initiate switch is closed, the low side of the capacitor is suddenly raised to the supply voltage. This raises the diode side of the capacitor to twice the supply voltage. The capacitor immediately begins discharging through the 1 and 3.3-meg resistors in series. Eventually the gate of the D13T1 becomes forward biased and the device turns on and applies power to the relay. The delay (40-60 sec) is virtually independent of supply voltage.

It is often handy to have a burst oscillator available. Figure 13 presents a simple 1-second, 1-kHz burst oscillator. The D13T2 is normally open due to excess holding current through the 100K resistor. When the initiate switch is momentarily closed, the 10μF capacitor charges to 15 volts and the circuit starts oscillating (1.8 meg and 820 pF). The circuit latches when the Zener diode (the emitter-collector section of a 2N2926) conducts again (after the 10μF capacitor has discharged to the Zener voltage).

PUT's can also be used for ordinary thyristor applications. The ring counter in Fig. 14 makes an efficient low-cost circuit featuring automatic resetting through the first stage 3N84, SCS.

The unique characteristics of the complementary SCR allows construction of a ring counter with extremely broad tolerances. The circuit in Fig. 15 puts the load in the gate and triggers at the anode. When power is applied, the first stage is triggered by the transistor-Zener combination when the anode line reaches 7 volts. The first pulse turns off the first stage and when power is restored the second device is triggered by the charge on the coupling capacitor. Note the long discharge time-constant of this capacitor, allowing the use of low-value, low-cost capacitors and long counting pulses. Succeeding stages are similarly triggered. As illustrated, using 24-volt lamps as loads, the circuit operated with supply voltage ranging from 12 volts to the 30-volt maximum of the C13Y, with pulse widths from 3 to 70 microseconds and at repetitive rates to 10 kHz.

The chirp oscillator shown in Fig. 16 is an interesting application of the D13T1 CSCR. Operation commences when the switch opens and lasts 1 to 2 seconds or until the switch is closed. There is no current flow through the anode to cathode terminals while C1 is charging so C2 can be tied to the cathode. When C1 charges to the peak-point voltage, it discharges rapidly through the D13T1 into the storage capacitor, C2. This produces a positive step in voltage across C2 as shown in the accompanying waveforms. The 22-meg resistor is added to point B to prolong the duration of the chirp and discharges C2 towards ground during each period of C1. When C2 reaches a voltage just under the intrinsic stand-
Putter With The Put

built around this fascinating semiconductor device

off voltage, the oscillation stops. The oscillations terminate after approximately 1 second.

The D13T can also be used as a Schmitt trigger, as shown in Fig. 17. The circuit can be considered a npn transistor operated in the common-base mode, driving a common collector npn. With the variable resistor shorted out only linear amplification occurs. As the resistor is increased, regeneration makes the circuit perform as a Schmitt trigger.

The complementary SCR's ability to drive a load in the gate makes anti-coincidence circuit (Fig. 18) easy to implement. This circuit provides ground only to the one load selected first. If other loads are selected, the gates are reverse biased on the associated CSCR's until the first load is released. Note that two loads cannot be driven at one time since any load that is driven deprives all others of triggering voltage. Any number of loads may be operated with this circuit at supply voltages up to 50 volts and load currents to 50 mA. The 1N4156 stabilizer (low-voltage reference diode) and 10K resistor provide gate trigger voltage for the CSCR, while the 1K resistor provides holding current.

A temperature-compensated battery voltage monitor using the CSCR is illustrated in Fig. 19. If battery voltage is low, trigger voltage is not applied to the C13Y gate by the voltage divider and the lamp stays off.

High current SCR triggers

Figures 20, 21 and 22 are also thyristor applications for the D13T1. Generally, it is necessary to trigger a high-current SCR from a pulse transformer secondary.

The circuit in Fig. 20 uses the slow-rising waveform on the transformer secondary to charge the 0.1-µF capacitor through the diode. The capacitor stops charging at the peak of the pulse. As the voltage across the secondary collapses, the D13T1 gate becomes forward biased and the device turns on. This quickly discharges the capacitor into the gate of the SCR. Measurement with a 0.1-µF capacitor and a 20-ohm load resistor (instead of SCR gate) showed rise-times of 50 nsec to 1 amp of pulse current.

The circuit in Fig. 20 requires that the total energy needed to trigger the SCR be delivered by the pulse transformer. Sometimes this is not possible. The circuit in Fig. 21 derives its trigger energy from the power line thereby using the pulse transformer secondary to trigger the D13T1. The voltage across C2 is initially derived through C1 and R1. When the voltage across C2 reaches the Zener breakdown voltage, the rest of the line voltage is absorbed across C1. Dissipation in the Zener diode is minimized by the time constant of R1 and C1. When a trigger pulse is applied to the transformer secondary across the D13T1 gate, the D13T1 triggers thereby discharging C2 rapidly into the SCR.

This same trigger technique can be used to convert a wide, low-amplitude pulse from a dc circuit into a sharp trigger pulse for ordinary SCR's. Figure 22 shows the arrangement necessary. Here the dc circuit is considered as a 10-volt dc supply switched on via Q1 and R1. When Q1 is turned on, the C1, charges via diode D1. During this period, the D13T1 gate is back biased. When Q1 is switched off, the anode gate becomes forward biased via R2 and discharges C through the pulse transformer primary very rapidly. A circuit of this sort could be the interface between low-power integrated circuits such as the PA424 zero-voltage switch and high power SCR's.
CAREERS in ELECTRONICS
blueprint to your future

by GEORGE G. GARMUS

Computer maintenance is really big business. Your electronic background may be all you need to get in on the ground floor.

The role of the computer maintenance specialist is extremely vital. A billion dollar corporation's business activities can come to a standstill when its computer system is "down." The maintenance engineer must be able to think and act coolly under pressure while he's trying to correct a computer system problem. He must have a solid understanding of electronics, computer theory, programming and a high mechanical aptitude. Since computer systems are extremely complex he must be able to think in a logical, analytical manner.

Computer maintenance is at least a $100 million business, ranking on par with manufacturing and sales. Most of the giant computer vendors have recognized this by establishing their computer maintenance operations as separate divisions or departments. In some instances more profit is made from maintenance contracts than from actual sales or rentals of computer systems.

The maintenance of computer systems is truly a "Big Picture" job. The typical field engineer's activities involve electronic circuit analysis, electromechanical adjustments, computer programming and analysis, complex diagnostic systems analysis, customer relations, logistics and business administration. He may, during the course of a month's activities, deal with people in design engineering, manufacturing, sales, training, programming, product planning and several levels of management at the customer's installation.

Computer maintenance is an evolving profession. The growing complexity of computer hardware and software requires special skills and insights to achieve greater operational efficiency and reliability. Maintenance engineers are now considered an integral of the computer systems design team. Engineering courses on systems maintainability are becoming an increasingly important part of an engineering curriculum. Maintainability principles were developed by maintenance engineers. Also, maintenance engineering is now recognized as a professional specialty by engineering and trade associations such as BEMA (Business Equipment Manufacturers Association).

Qualifications for computer maintenance

The typical computer field engineer has at least 2 years of formal electronics training or armed forces training. Some of the best qualified men have had experience in the military as fire control technicians or in the army as missile systems electronic technicians. This type of training involves exposure not only to complex electronic systems but to intricate electromechanical systems as well.

The computer man must be equally skilled in maintaining both electronic and mechanical equipment. High-speed printers, magnetic and paper tape readers, card processors and similar equipment combine electronic, electrical, mechanical and electromechanical elements in their operations.

The candidate must be thoroughly familiar with pulse circuitry theory and be able to use oscilloscopes and voltmeters. He should be able to analyze complex waveforms—quickly and accurately.

When a computer system "goes down" it can be due to a number of reasons—component malfunctions or program errors. Therefore the candidate, in addition to possessing skills in electronics and mechanical aptitudes, must also have a strong talent for logical and analytical thinking. Unless he has this combination of skills he wouldn't be able to isolate the source of a computer trouble. He might spend precious minutes or hours chasing a trouble in an electronic unit—when it was actually a goof in the computer program! Or he might get hung up making adjustments on one of the high-speed printers.

He should have a clear understanding of present electronic theory and be able to read and correctly interpret

THE EXPLOSIVE GROWTH of the computer industry has created numerous career opportunities for those trained in electronics. A new spectrum of technical specialties has been spawned by the computer industry in the past decade. These new specialties range from logic designers, systems engineers, programmers and input-output designers to computer field engineers. Attractive salaries, fringe benefits and job security are a common denominator of these specialties.

One of the most important and interesting jobs in the computer industry is computer systems maintenance. The electronic specialists who maintain and troubleshoot computer systems in the field are usually called field or customer engineers. Some computer manufacturers call their maintenance specialists field service engineers or representatives. However, the important fact is that computer maintenance men are responsible for keeping complex, expensive computer systems "on the air."

Radio-Electronics zeroed-in on the facts about a career in computer maintenance by arranging an interview with an expert who maintained one of the very first commercial digital computer systems and now heads up General Electric's commercial computer maintenance training operations, Kenneth Strandberg. Ken is manager of Technical Operations for G-E's Field Engineering Department headquartered in Phoenix, Arizona.

Maintenance is a business

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He should have a clear understanding of present electronic theory and be able to read and correctly interpret

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schematic diagrams. Equally important is his ability to get along with people effectively. Working under pressure at a customer's facility requires tact, empathy and integrity. Blowing up under pressure could result in having a computer system thrown out!

The computer field engineer should be a "self-starter"—requiring a minimum of supervision. Ideally, he should have many of those talents required to run your own business. That is, he recognizes the importance of time and money in his sphere of activities. He knows how to schedule the use of his time effectively.

Screening and training

At G-E, candidates are screened to determine their electronics knowledge, mechanical aptitudes and customer relations potential. He must present a personal appearance that is above average because of his constant contact with all levels of customer personnel. The candidate must be able to effectively communicate with a wide range of people, in both oral and written communications. He must be in good health and be bondable. The job interview and screening is usually done by a field engineering manager in the city in which the candidate is located. The general policy is hire candidates from cities in which they will work.

The salary ranges from approximately $125 to $300 per week for 40 hours. An inexperienced new trainee generally starts at the low end of the salary range. There is additional compensation at a rate of 1.5 or 2 times straight pay for overtime in excess of the straight 40. The company provides life and medical insurance, sick leave and paid vacations to all field engineer reps.

The objectives of the training program are to provide the field engineer with a thorough educational program covering: data processing equipment installation and servicing; customer relations and site management.

The student's training program begins the day he becomes employed and continues through his career.

The new employee's first phase of training is an orientation period. Initially, he is assigned to work with field engineering personnel in the area where he will be assigned for a period of 3 to 4 weeks. During this time he is familiarized with the computer maintenance business, its operations, maintenance procedures and is acquainted with the men with whom he will work.

G-E has found this is one of the most practical approaches to training computer field engineers—"seeing it the way it is." In the past, many students became confused when they thought what they learned in school didn't seem to jibe with what they saw in the field. Another benefit of this approach is that it instills a higher degree of confidence in the new student in the formal training program.

The second phase of training begins when the student reports to the Phoenix training center. His training at this time is centered around basic mechanics, diagnostic troubleshooting and equipment maintenance procedures. He works on actual computer systems and performs adjustments on electromechanical assemblies.

He is taught basic programming, computer operating methods, and is familiarized with software and typical applications programs. It is extremely important that the field engineer be able to determine why a type of trouble is causing a computing error. Is it a bug in the program? An operator goof? An electronic component malfunction? A mechanical unit out of adjustment? These are the questions that come up when a field engineer is zeroing in on a computer trouble.

The length of the student's training program ranges from 3 to 25 weeks, depending on the specific equipment to which he is initially assigned. There are an average of 12 to 14 students in each class at Phoenix. The school has an international flavor with students attending from Japan, France, Italy, Canada, Australia, Sweden, Holland and Mexico.

Upon completion of his training, the field engineer is returned to his district to assume his duties of equipment maintenance, customer relations and site management.

Continuous training and advancement

Computer technology is probably one of the fastest changing fields in engineering. Therefore, the field engineer must constantly keep his professional knowledge up to date. The experienced field engineer receives additional instruction through a comprehensive field training program. Field training is conducted at sites especially designed for the use of audio-visual training techniques. This (turn page)

AT GE'S PHOENIX CENTER (top left) students study actual equipment to gain mastery of computer maintenance. SCOPE AND VOLTMETER (top right) combine with logical analysis to track down troubles. AUDIO AND VISUAL aids are used (2nd photo from top) to introduce actual equipment and techniques. TEAMWORK IS THE KEY (2nd from bottom) to maintaining the more complex equipment. GOOD MANUAL DEXTERITY (bottom) and a semiconductor background are prerequisites. Photos by G-E.
portion of his training is presented on audio tape with instruction books and associated materials. He is assisted in this period by a qualified tutor (another field engineer) who has had special training and is experienced on the equipment the student is studying.

Students return to the Phoenix training center for additional training on an average of once every 2.5 years. Special audio-visual training classes are also conducted at Phoenix for experienced field engineers. The teaching system is essentially the same used in the field training programs.

Training programs in the form of formal classes, seminars and workshops are conducted regularly in the field. "In a business such as ours where the technology is constantly changing," says Strandberg, "continued training is a must to maintain an organization of competent professionals."

There is only one criterion for advancement and this is performance. The prime factor in selecting the field engineer for promotion is how he puts his technical knowledge to work. Basically, there are two paths of advancement open to the field engineer. That of the specialist and that of management.

The field engineer progresses to specialist because of demonstrated technical, customer relations and administrative abilities. The specialist may then join the headquarters staff in technical operations where he will assist in the design and development of new equipment from the maintainability standpoint, or he may enter training operations where he will contribute to the development of new training programs.

The field engineer may proceed along the path toward a management goal, seeking a position as a district, regional or departmental manager. The path is of his own choosing and how he travels along this path depends on his ability and initiative.

The plusses for a career in computer maintenance are that it is a job that uses a wide range of an individual's talents and presents an everlasting challenge to the "problem solver" who also likes to work with his hands. It ties in that rare combination of dealing with very complex equipment and customer relations.

The challenge and variety of a field engineer's duties are hard to beat. Dealing with programming, sales, engineering, logistics, administrative and customer relations problems makes a field engineer's life very interesting. The status of the field engineer ranks high in the computer field and is continually improving. Good salaries and job security are another plus for a career in field engineering.

Like any profession, field engineering has its negative aspects. Working on rotating shifts is probably the most negative aspect of computer maintenance work. However, some field personnel find it to their liking. Being able to go golfing or fishing while everyone is busy in the office can be a plus to many outdoor enthusiasts. Another consideration is periodic relocation for those moving up the promotion ladder.

Computer maintenance outlook

The phenomenal growth of field engineering has increased the demand for qualified men. In 1960, there were approximately 10,000 field engineers in commercial computer systems maintenance in the entire industry. The estimated total in 1970 is about 25,000! The industry estimates that there will be a rock-bottom minimum of 50,000 required in 1980. If anything, most past estimates were entirely too light.

With the growing complexity of computer software, the use of minicomputerization techniques and solid-state technology, the professional requirements for field engineers will be raised to higher levels of technical competence. In all likelihood, we'll soon wit-

COMPLEX COMPUTER CIRCUITS take trained troubleshooting skills to repair.
POWER TOOLS

by THOMAS R. HASKETT

Unless your interest or work in electronics is very limited, you probably use power tools. They enable you to drill holes, cut a chassis or speaker-mounting hole, and sand a speaker enclosure.

Electric drills are versatile

Almost indispensable for cutting holes in metal, the common electric drill consists of a motor in a pistol-grip housing (Fig. 1). The motor shaft is geared down to drive the chuck and the drill bit, which does the cutting. Most drills today have keyed chucks, which you turn with a key (furnished with the drill), causing the chuck to grip the bit. Some inexpensive drills use a knurled chuck which you twist with your hand. Such a chuck is not desirable.

Drills are rated by their chuck capacity and class of service. The three standard chuck sizes are 1/4", 3/8", and 1/2". Larger drills usually turn more slowly, since there is more work to be done. The average speed of a 1/4" drill is 2200 rpm; a 3/8" drill, 1000 rpm; and a 1/2" drill, 600 rpm. Of course, the larger drill has a larger horse-power rating, too (1/7 HP for 1/4" vs 2/5 HP for 1/2").

A standard or light-duty drill will do for general work, but a heavy-duty drill (in any of the above sizes) has a larger-horsepower motor and somewhat slower bit speed. Light-duty drills aren't intended for contin-

kit containing one each of the fractional sizes over a certain range. It's more useful to have several of the same often-used sizes, in case you break one when you can't easily obtain another.

Twist drills are made of one of two types of metal: Carbon steel, or high-speed steel. Carbon-steel (or butter) drills are useful only in soft materials such as wood or plastic, and don't work well in metal. High-speed twist drills, although more expensive, can be used in metal, wood, or plastic. They are marked HS or HSS.

Most drills come in standard lengths, and the length increases as the diameter of the drill does. Extra-long (jobbers) sizes are available, but are useful only when you have to cut a deep hole. Short shanks are better for close, precise work, as the drill is working closer in.

Normally, straight-shank drills are most useful, but to drill a hole larger than the chuck capacity of your drill, you may have to use a reduced-shank twist drill or bit. You don't get something for nothing, however; a reduced-shank type won't drill through as hard a material as a straight-shank model. Also, the tip tends to wobble and may cut a ragged hole. Also available are chuck adapters.

Many bits come with a quick-cut or split point; such a drill starts its own hole without walking around on the work.

The most convenient way of carrying and keeping bits is the kit mentioned previously. It has spaces for drills from perhaps 1/64" to 1/4" or 1/2" Cases are available for various size ranges.

To drill concrete or masonry, you must use a twist drill tipped with an extra-hard metal such as carbide or Carboloy.

Notes on drilling

When you're drilling holes in a painted surface, cover it with masking tape or paper and lay out the hole centers with a pencil. Use a center punch first to indent the chassis; then drill through the paper and surface. Don't remove the paper until you've drilled all the holes. That way you're not as likely to mar the painted surface.

If you're drilling a hole in existing equipment, remove all glass ahead of time—vacuum tubes, meter movement, glass dial faces, etc. To prevent metal chips from clogging tube sockets and other receptacles, cover them with masking tape.

Drill holes larger than 1/4" in two or more steps. Drill a pilot hole first—say 1/8". Then enlarge it with the larger drill or reamer. For precision, cut larger holes (3/4" diameter and above) with chassis punches.
uous operation over long periods of time, and they can't be used for working thick, heavy metals without running the risk of overheating and damaging the motor. Heavy-duty drills can be used all day, and will do large jobs with ease, but they are a little bulky for small work and are more expensive than those rated for light duty.

For most chassis work, a 1/4" standard drill is adequate. A 1/2" drill is useful in construction and electrical work, and for industrial and broadcast electronics. It has a spade grip and crossbar handle for working with both hands. The 3/8" drill is something of a compromise between the other two, and usually has an auxiliary handle on the side of the case.

Some drills have a built-in variable-speed control, which lets you change speed to fit the hardness of the work. Some drills are reversible, a valuable feature when you have to back a drill out of a deep cut.

Most drills have universal motors, which will run on 117 Vac or dc. Another feature of interest is the position of the chuck key on the power cord. It's usually attached at the plug end of the cord. It should remind you to always pull the plug before changing bits. If the plug is in, a short at the trigger switch could start the motor and injure your hand.

At least one manufacturer—Black and Decker—offers a cordless or battery-powered 1/4" drill (Fig. 2). The batteries are rechargeable.

Drill bits aren't all alike

Also called twist drills, drill bits are available to cut holes from .0135" to 1". The most useful sizes for electronics work at No. 33 (to pass 4-40 screws), No. 28 (to pass 6-32 screws), No. 18 (to pass 8-32 screws), No. 11 (to pass 10-24 screws) and 1/4". Many people buy a

When drilling aluminum or brass, you can remove the burrs of these soft metals with a sharp knife. To remove burrs from a steel hole, use a countersink or a cold chisel and hammer.

Only sharp drill bits do a good job. It's not easy to sharpen bits, as the tip angle is critical. Kits are available to hold the bit at the correct angle for a grinder, but many people simply replace small bits when they become dull, and have the larger ones sharpened at a hardware store.

Drill accessories are numerous

You can extend the usefulness of any drill by acquiring special attachments designed for specific jobs. You can also buy a drill and several attachments and bits as a kit. A wire wheel or cup brush (Fig. 3) is handy for removing rust and scale, or for rough-polishing metal.

Fig. 3—Wire cup brushes are handy for removing rust and scale. Unit at left is made by Coastal Abrasive. At right are brushes by Black & Decker.

Sandpaper discs and a lamb's wool buffing wheel are used with a rubber disc for backing. Both are handy for smoothing and polishing. A grinding wheel (Fig. 4) is made of abrasive stone and can be used to sharpen edged tools. The easiest way to perform this job is to clamp the drill body in an accessory horizontal drill stand (Fig. 5) and the stand in a vise. The countersink (Fig. 6) was mentioned above, and is used for putting a dimple in the work to take a flathead screw, as well as for cleaning the edges of a rough hole.
To cut a hole larger than the largest drill bit you have, you can use a hole saw (Fig. 7), which comes in various sizes and uses serrated teeth to saw out the circle. It comes with an arbor and a pilot drill to start the hole. The same job can also be done with a fly or circle cutter (Fig. 8), which spins a cutting bit in a circle around the pilot drill. Unfortunately, a fly cutter is prone to wobble and cut a ragged hole.

Fig. 7—Hole saws are a must when a large-diameter hole is to be made. At left is Stanley unit for metals. Center above are two Black & Decker units. Fig. 8—(right) Adjustable dial saw is another approach to cutting large-diameter holes with an electric drill.

Screwdriver bits (Fig. 9) are useful for extracting and installing tight screws when used with a special adapter.

One of the most recent additions to drill kits is the SCR speed

Fig. 5—Horizontal drill stands convert hand drill to a bench unit. Two units are shown above. At far left is Costal Abrasive stand. In the center is the Skill 45192 horizontal drill stand. Fig. 6—(above left) Costal countersinks.

Less expensive saws run at a fixed speed, although some let you switch between two speeds. But it's sometimes useful to vary blade speed with work thickness, and the variable-speed saw does this job nicely.

Often a saber saw is used for cutting small curved areas and you have to turn the drill as you curve the cut. A few models are available which allow you to rotate the blade (Fig. 12) for easier curve cutting.

Fig. 12—Tilt while cutting circles makes the job a bit easier. Stanley 90486 speed-control saw is shown here.

**Blades for your saber saw**

A variety of blade types are made for saber saws. For scroll (curved) cutting of soft wood and plastic, thin, narrow blades are best. For straight cuts through plywood and hardwood, wider blades are used. Tougher blades with different teeth are recommended for hard, green, or wet wood. You have to use a hacksaw blade, of course, for metal. A knife blade is useful for cutting rubber, leather, tile, etc.

In any case, a coarse-toothed blade (less teeth per inch) is used for rough work, while a fine-toothed blade (more teeth per inch) is used for finish work. Most blades are about 3" long, but some are 4" or even 6", for deep work. In Fig. 13 you see the typical blade compared to the flush-cut blade. To use this type, you remove the guard at the front of the saw (it screws on) and insert the flush-cut blade in the saw. Then

Fig. 13—Sabre-saw blades come in wide variety of sizes, shapes and types. There are blades for wood and metal, flush cut and standard. So pick the blade that meets your requirements. Units at left are made by Skill. Blades at right are P.A. blades.
control (Fig. 10), which gives you continuously variable speed, letting you match bit rpm to the hardness and thickness of the work. Some drill bodies are available with SCR speed control, while accessory units are made for drills not already equipped.

Fig. 10—Add on speed control turns most ac-powered tools into variable-speed units. Just insert control box between tool and wall outlet. Unit shown is made by Chicago Wheel.

Electric saber saw

If you do any cabinet, carpentry, or chassis work, a very useful tool is the saber saw (Fig. 11). Also called a portable jig saw, it contains an electric motor which drives a reciprocating cam that moves a saw blade back and forth across a limited area. The best saws have universal motors which work on both ac and dc. The blade is in the front of the saw, on the bottom. Some models have a light above the blade so you can see where you're going.

Fig. 11—The electric saber saw is really a portable jig saw, and is extremely versatile. Top left is Murphy. Center above is RAM-146. At right is Skill 2-speed model 514. At left is Rockwell unit.

you can cut up against a wall or corner.

Most saws run at a fixed speed, but variable-speed models are on the market. Of course, you can use most saber saws with accessory SCR speed controls.

Other power tools for you

For woodwork, cabinetry and carpentry, a circular saw (Fig. 14) is essential. It guides itself through a straight cut, can make cuts at a varying vertical angle, and cuts fast. Several blade types are used—rip, crosscut, combination—depending on whether you're making a rough or finish cut with or against the grain of the wood.

An electric hand grinder is useful for chassis and hobby work where you wish to do grinding, buffing, and shaping. The kit usually includes wheels and bits of several shapes and grits. You can deburr, carve, rout, and sand small areas with this tool.

A bench grinder is simply a bench-mounted grinding wheel about 6" in diameter for keeping edged tools sharp. Unless you have frequent need for this tool, you can probably get by with a grinding wheel and your hand drill.

It's a good idea to mark your tools for identification to prevent mistakes when lending or borrowing them. Using a sharp awl or pick is tedious; much easier is Dremel's Electric Engraver. This handy little tool is a small vibrator with a hard point. A variable control lets you adjust stroke length for the depth of cut you want and the hardness of the material. The tool cuts metal, ceramic, and glass.

More specialized tools exist—such as routers—but are seldom used in electronics.
New FM tuner
One IC replaces entire I.F. strip

- From rf to audio
- No tuned circuits
- Narrow bandpass
- High linearity
- Low distortion
- Phase-locked loop

An old circuit in a new IC may spell revolution in many radio circuits

by LARRY STECKLER
MANAGING EDITOR

A new kind of IC, a phase-locked-loop (PLL) has been developed by Signetics and Motorola. Other units by other manufacturers are soon to follow. This new unit makes it possible to replace tuned circuits in i.f. strips, demodulators, and other inductor-dependent circuits with an IC arrangement that completely eliminates the need for coils or transformers.

Shown on our cover this month is an FM tuner built around one of the Signetics IC's. It consists of a conventional FM front end of the type used in stereo FM tuners, which is coupled to a PLL IC. This IC takes the 10.7-MHZ output from the front end and converts it to audio. While the quality of this circuit is not equal to the performance of a conventional stereo tuner, it opens the door to a new approach to FM.

The PLL will likely find its way into many types of electronic equipment such as radio receivers, appliances, TV sets, tape recorders, Commercial uses include signal generators, detecting and decoding of tones, etc.

One major advantage of PLL in FM tuners is that mistuning and drift are eliminated. The PLL locks onto a station and its built-in VCO (voltage-controlled oscillator) keeps it on station.

PLL in itself, is not a new idea. It has been used for several years to track satellites, stabilize the frequency of klystrons, and filter information out of noise. Unfortunately, in its discrete-component form the circuit is too costly and complicated to be used in consumer electronics.

An as IC, however, even the cost of this complex circuit is low, and we are likely to find the PLL appearing in many new kinds of equipment.

As shown in Fig. 1, the PLL is made up of four basic sections—the phase comparator, the low-pass filter, the amplifier, and a voltage controlled oscillator (VCO).

The phase comparator is a multiplier that mixes the input signal with the VCO signal to produce sum and difference frequencies f₀ ± f. When the loop is in lock, the VCO duplicates the input frequency and the difference frequency (f. — f.) is zero. At this time, the output of the phase comparator contains a dc component.

The output of the comparator is filtered to remove the sum frequency component (f. — f.), while the remaining dc component is amplified and fed back to the VCO. It is this error voltage that drives the VCO keep f. equal to f. when the loop is in lock. The size of this error voltage is a direct function of the phase relationship between the VCO and the input signals.

If you want to see how error voltage is affected by the phase difference between the two signals, take a look at Fig. 2. Here the VCO signal is a square wave. When the loop is out of lock, and the differ-
ence between f. and f. is large, both sum and difference signals are filtered by the low-pass filter and the VCO runs at its preset frequency.

But as the difference between f. and f. approaches the cutoff frequency of the low-pass filter, a beat note forms which tends to force the VCO frequency toward the input frequency. When this dc component of the beat note grows large enough, the VCO frequency reaches the input frequency and the loop locks. Therefore, lock sensitivity is a function of the low-pass filter. Once in lock, the loop will track signals over its entire lock range.

PLL applications
Some of the expected applications for PLL IC's include FM demodulation in FM tuners, FM telemetry and telephony, frequency-shift keying (FSK), and store-casting music (SCA) detection.

Other potential uses are in signal conditioning where the PLL can be used to filter unwanted sidebands from the input signal. The useful PLL output in this kind of application is from the VCO output signal.

Frequency multiplication is another application. For low orders of multiplication with small FM deviations the loop can be simply locked to a harmonic of the input signal. For very wide FM deviations or higher multiplication, the loop can be broken between the VCO and the phase comparator, and a frequency divider inserted. Now, when the circuit is in lock, the output of the divider is at the same frequency as the input signal and the VCO runs at a multiple of this frequency, thus forming a precise frequency multiplier which tracks the input signal over a wide range.

Some circuits to try
Now let's take a look as some of the practical ways we can expect to use a PLL right now. In each of the four examples that follow, a Signetics SE/NE565 IC is used.

FM demodulation
The 565IC phase Locked Loop is a general-purpose circuit designed for high-linearity FM demodulation. During lock, the average dc level of the phase comparator output signal is directly proportional to the frequency of the input signal. As the input frequency shifts, it is this output signal which causes the VCO to shift its frequency to match that of the input. Consequently, the linearity of the phase comparator output with frequency is determined by the voltage-to-frequency function of the VCO.

Because of its unique and highly linear VCO, the PLL can lock to and track an input signal over a very wide range (typically ±60%) with very high linearity (typically within 0.5%).

A typical FM demodulator circuit is shown in Fig. 3. The VCO free-running frequency is represented approximately by f. = 1/4R1C1, and should be adjusted to be at the center of the input signal frequency range. Capacitor C1 can be any value, but resistor R1 should be within the range of 2000 to 20,000 ohms with an optimum value in the order of 4000 ohms. The source can be direct coupled if the dc resistances seen from pins 2 and 3 are equal and there is no dc voltage difference between the pins.

A short between pins 4 and 5 connects the VCO to the phase comparator. Pin 6 provides a dc reference voltage that is close to the dc potential of the demodulated output (pin 7). Thus, if a resistance is connected between pins 6 and 7, the gain of the output stage can be reduced with little change in the dc voltage level at the output. This allows the lock range to be decreased with little change in the free-running frequency.

In this manner the lock range can be decreased from ±20% of f. (at ±6V).

A small capacitor (typically 0.0001µF) should be connected between pins 7 and 8 to eliminate possible oscillation in the control current source.

A single-pole loop filter is formed by capacitor C2, connected between pin 7 and positive supply, and resistance inside the IC of approximately 3600 ohms.

Frequency shift keying
FSK (frequency-shift keying) refers to data transmission by means of a carrier which is shifted between two preset frequencies. This frequency shift is usually handled by driving a VCO with the binary data signal so that the two resulting frequencies correspond to the "0" and "1" states (commonly called space and mark) of the binary data signal.

A simple circuit using the PLL to receive FSK signals of 1070 Hz and 1270 Hz is shown in Fig. 4. As the signal appears at the input, the loop locks to the input frequency and tracks it between the two frequencies with a corresponding dc shift at the output.

Loop filter capacitor C2 is chosen smaller than usual to eliminate overshoot on the output pulse, and a three-stage RC ladder filter is used to remove the carrier component from the output. The band

![FIG. 1—PLL BLOCK DIAGRAM (below) illustrates basic operation of the phase-lock-loop.](image1)

![FIG. 2—CORRECTION SIGNALS (left) are developed as shown here.](image2)

![FIG. 3—FM DEMODULATOR using the PLL IC. Experimenters will enjoy this one.](image3)
edge of the ladder filter is set approximately half way between the maximum keying rate (in this case 150 Hz) and twice the input frequency (approximately 2200 Hz). The output signal can now be made logic compatible by connecting a voltage comparator between the output and pin 6 of the loop. The free-running frequency is adjusted with R1 to result in a slightly-positive voltage at the output when \( f_0 = 1070 \text{Hz} \).

The input connection is typical for cases where a dc voltage is present at the source and therefore a direct connection is not desirable. Both input terminals are returned to ground with identical resistors (in this case, the values are chosen to effect a 600 ohm input impedance).

**Frequency multiplication**

There are two methods of doing this using the 565:

1. Locking to a harmonic of the input signal.
2. Including a digital frequency divider or counter in the loop between the VCO and phase comparator.

The first method is the simplest, and is used by setting the free-running frequency of the VCO to a multiple of the input frequency. A limitation to this scheme is that the lock range decreases as successively higher and weaker harmonics are used for locking. If the input frequency is to be constant with little tracking required, the loop can generally be locked to any one of the first 5 harmonics. For higher orders of multiplication, or for cases where a large lock range is desired, the second scheme is more desirable. An example of this might be where the input signal varies over a wide frequency range and a large multiple of the input frequency is required.

A block diagram of the second scheme is shown in Fig 5. Here the loop is broken between the VCO and the phase comparator, and a frequency divider is inserted. The fundamental of the divided VCO frequency is locked to the input frequency in this case, so that the VCO is actually running at a multiple of the input frequency. The amount of multiplication is determined by the frequency divider.

A typical circuit is shown in Fig. 6. To set up the circuit, the frequency limits of the input signal must be determined. The free-running frequency of the VCO is then adjusted by means of RI and CI (as discussed under FM demodulation) so the output frequency of the divider is midway between the input frequency limits. Filter capacitor C2 can be neglected, in this case, since it is generally desirable to have the capture and lock ranges as large as possible. The output can now be taken as the VCO square-wave output, and its fundamental will be the desired multiple of the input frequency as long as the loop is in lock.

**SCA decoder**

Some FM stations are authorized by the FCC to broadcast uninterrupted background music for commercial use. To do this, a frequency modulated 67-kHz subcarrier is used.

The SCA signal can be filtered out and demodulated with the NE565 PLL without the use of any resonant circuits. A circuit for this application is in Fig. 7. This circuit also serves as an example of operation from a single power supply.

A resistive voltage divider is used to establish a bias voltage for the input (pins 2 and 3). The demodulated (multiplex) FM signal is fed to the input through a two-stage high-pass filter, both to effect capacitive coupling and to attenuate the strong signal of the regular channel. A total signal amplitude, between 80 mV and 300 mV, is required at the input. Its source should have an impedance of less than 10,000 ohms.

The PLL is tuned to 67 kHz with a 5000 ohm potentiometer. Only approximate tuning is required, since the loop will seek the signal.

The demodulated output (pin 7) passes through a three-stage low-pass filter to provide de-emphasis and attenuate the high-frequency noise which often accompanies SCA transmission. Note that no capacitor is provided directly at pin 7. Thus, the circuit is operating as a first-order loop. The demodulated output signal is about 50 mV and the frequency response extends to 7 kHz.

![Diagram](image-url)
Build with IC's

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For about $40.00 this wide-range function generator will prove a most valuable addition to the work bench of the electronic experimenter or audiophile. It provides sine waves, triangle waves, and square waves simultaneously, from 1 to 110,000 Hz. Each output amplitude is constant regardless of frequency, and adjustable from zero to twenty volts P-P or more. Tuning is variable over one decade, with ten percent overlap on each range. A range selector provides five decade multiples of the vernier tuning.

The heart of the generator is three Motorola MC1709CG integrated-circuit operational amplifiers. Each contains the equivalent of 13 transistors, 2 diodes, and 15 resistors in a standard TO-5 package.

Theory of operation

The unit (Fig. 1) can be divided into three major parts. The oscillator sine-wave converter, and output amplifiers. The oscillator is a form of relaxation circuit called a hysteresis oscillator. It is composed of two integrated-circuit operational amplifiers IC1 and IC2.

Integrated circuit IC2 is connected as a voltage comparator with hysteresis feedback. Because of the high gain of the amplifier, if the non-inverting (+) terminal is slightly more positive than the inverting (-) terminal, the output will be the positive saturation level of +14 volts. Conversely, if the non-inverting input is more negative, the output will be the negative saturation level, -14 volts.

Assuming the output is +14 volts, positive feedback applied to the non-inverting input via R9, keeps the amplifier latched in the positive saturation state. An increasingly negative voltage is applied to the input resistor R7. The ratio of R7 and R9 is such, that when the input reaches -10 volts, it cancels the +14 volt feedback. As the input goes slightly more negative, the amplifier switches to the negative saturation level. It remains in that state until the input exceeds +10 volts when a similar, but opposite reaction occurs causing the amplifier to latch again at the positive level.

Integrated circuit IC1 is connected as an integrator. If a positive voltage (Ei) is applied to the input resistor, R5, the output will try to swing negative. The voltage across the timing capacitor C, (C1-C5) cannot change instantaneously, however, and the negative output thru the capacitor cancels out the positive input in an attempt to maintain the inverting terminal at zero volts. The cancelling current charges the capacitor at a linear rate toward -14 volts at Ei/R5C, volts per second. The opposite effect is true of negative inputs.

If comparator IC2 is latched in the +14-volt state, and some portion of the +14 volts is applied to the integrator input resistor, the integrator output will run down linearly toward -14 volts. Since the integrator output is connected to the comparator input, the comparator will switch to -14 volts when the integrator output equals -10 volts. This applies the opposite polarity to the integrator, and its output runs up linearly toward +14 volts. At +10 volts the comparator switches back to +14 volts completing one cycle. The cycle then repeats and the circuit oscillates at a constant rate.

The frequency of oscillation is a function of the applied voltage to the integrator (Ei), the integrator input resistor (R5), the feedback capacitor (C), and the trigger levels of the comparator (E, max) according to F = Ei/4E, max x R5 x C, E, max is set at 10 volts, R5 is 25,000 ohms, therefore F = E,/C, where C, is in μF. Resistors R1, R2 and R3 are chosen to provide Ei, from ± to ±11 volts, and C, is switched from 1 μF to 100 PF providing frequencies from 1 Hz to 110,000 Hz.

The output from the comparator is a 28-volt p-p square wave while the output of the integrator is a triangle of 20 volts p-p. The triangle wave output of the integrator is converted to a sine wave by the non-linear resistor/diode network R15-R30 and D7-D14.

Assuming the input to R29 and output across R30 are both zero initially, all the diodes are reverse biased by the bias voltage divider R15-R24. If the triangle wave is rising linearly toward +10 volts, its slope is reduced at the output by the voltage dividing action of R29 and R30. When the output reaches the bias level of D10, R25 is switched in parallel with R30, further reducing the slope. As the output increases still more, diodes D9, D8 and D7 conduct in turn (Fig. 2), each reducing the output slope by switching in R26, R27 and R28. The input triangle is therefore "distorted" into a quarter sine wave. As the triangle voltage decreases back toward zero, each diode is cutoff in reverse order to produce the second quarter sine wave. For negative half-cycles,
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**PARTS LIST**

All resistors 1/2 watt, 10% unless noted.

- R1—750 ohms
- R2—potentiometer, 2500 ohms, linear taper
- R3—270 ohms
- R4—22,000 ohms
- R5—27,000 ohms
- R6, R34—1500 ohms
- R7—10,000 ohms
- R8—5600 ohms
- R9—18,000 ohms
- R10, R35—potentiometer, 10,000 ohms, linear taper
- R11, R12, R31, R37, R40, R41—100,000 ohms
- R13, R14, R38, R29, R42, R43—4700 ohms
- R15, R24—1000 ohms
- R16, R23—39 ohms
- R17, R22—68 ohms
- R18, R21—120 ohms
- R19, R20—220 ohms
- R25—47,000 ohms
- R26—18,000 ohms
- R27—10,000 ohms
- R28—510 ohms
- R29—27,000 ohms
- R30—potentiometer, 25,000 ohms, linear tap
- R32—68,000 ohms
- R33—82,000 ohms
- R44, R45—51 ohms, 2 ohms, 5%
- R46, R47—3.9 ohms, 2 watts
- C1—100 pF
- C6, C9—10 pF
- C2—0.01 µF
- C3—0.01 µF
- C4—0.1 µF
- C5—1.0 µF
- C7, C8—15 pF
- C10, C11—1000 pH, 30 volts, electrolytic
- C12—0.0056 µF
- J1, J2, J3, J4—way jacks
- T1—transformer, 11.8 volt primary, 40 volt center-tapped secondary, 1 amp
- BR1—full wave bridge rectifier, 100 pF, 1A or four 1N4004 or similar diodes
- D1 to D14—1N914
- Q1, Q3, Q5—2N423
- Q2, Q4, Q6—2N425
- IC1, IC2, IC3—MC1709CG IC op amp

Miscellaneous: 6" x 4" x 3" utility box, knobs, stand-off terminals, hookup wire.

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(continued on page 68)
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There are three output drives, one for each waveform. They consist of complementary emitter followers. Diodes such as D1 and D2, compensate for the base-to-emitter drop of the transistors. For positive inputs, the npn transistor supplies the current to the load, and for negative inputs the pnp supplies the current. The 470-ohm resistors limit the current to prevent damage to the transistors, as well as provide an output impedance of about 500 ohms.

Construction techniques

A 3 x 4 x 6-inch chassis, such as a Bud AC-430 with BPA-1505 bottom plate makes an economical, compact cabinet. Mount power supply components on the inside top of the chassis box using insulated standoff terminals (Grayhill type 18-1 or similar). Mount all controls and binding posts symmetrically on one 3 x 6-inch face. After all components have been mounted in the chassis and on the PC board, connect the two thru twisted pairs of wires as per the wiring list.

Resistor R1 is connected from the top of R10 to the top of R2, and resistor R3 is connected from the bottom of R2 to the ground post.

The completed PC board is attached to the center of the bottom plate with 1/4-inch spacers. The completed unit should be checked for proper operation. Notice there are no adjustments to be made. If all components are good, and the wiring is correct, the unit will function immediately.

After checkout is completed, fold the PC board up into the chassis in a manner similar to closing a book. Attach the bottom plate with four 1/4-inch No. 6 self-tapping screws. R-E
The close parameter matching and tight thermal coupling of the CA3018 transistors result in exceptionally stable operation when they are used in the dc differential mode as a meter amplifier. The device outperforms conventional units in this type of application.

Fig. 24 shows how to connect the IC so it effectively increases the sensitivity of a 100-µA meter to 1 µA. Transistors Q3-Q4 and Q1-Q2 are wired in the differential mode as Super-Alpha emitter followers, with emitter loads R6 and R7 and with bias set by bridge network R1-R2-R3-R4. Resistors R5 and R9 enable the bridge bias network to be precisely balanced. Resistor R8 permits balancing the two emitter loads so the meter (connected between the emitters of the two emitter followers) reads zero when zero input is connected between the input terminals (pins 9 and 6) of the unit.

The emitter-follower actions of the circuit cause the design to deliver high current gain, but zero voltage gain, between the input terminals and the actual meter. The basic circuit gives a current gain of approximately 1000 times. This gain is subject to some variation with changes in ambient temperature.

In the interest of stability the actual gain of the circuit is reduced to only 100 by connecting SENSITIVITY control R10 across the input terminals.

To set up the circuit:
1. Set R10 (SENSITIVITY) to maximum. Short the input terminals, and adjust R8 for a zero reading on the meter.
2. Remove the short from the input terminals, and adjust R5 for a zero reading on the meter. Finalize the zero reading with R9.
3. Apply a 1 µA current between the circuit's input terminals. Then adjust R10 for full-scale deflection on the meter. The circuit is now ready for use. Fig. 25 shows how to wire the IC as a dc millivoltmeter with a sensitivity of 200,000 ohms/volt, using a 100-µA meter. Transistors Q1 and Q2 are used as a differential amplifier, with the meter connected between their collectors. Transistors Q3-Q4 are used as a constant-current generator in the differential amplifier's emitter line.

The circuit delivers both voltage and current gain between the input terminals (pins 6 and 3) and the meter. Resistor R. acts as a range multiplier, and has a value of 200,000 ohms/volt—if the meter is required to read 10 volts full scale: \( R = 2 \text{ megarhms} \).

To set up the circuit:
1. Short pins 3 and 6, and adjust R9 for zero meter reading.
2. Remove the short from pins 3 and 6, and with no input connected, adjust R5 for zero meter reading.
3. Connect the calculated value of R. (depending on full-scale deflection voltage required), and apply the required voltage to the input terminals. Adjust R10 for full-scale meter deflection.
4. Recheck steps 1 and 2, and readjust if necessary.

Finally, Fig. 26 shows how to wire the IC as a dc millivoltmeter with a 2-megohm voltage sensitivity. The circuit is basically similar to Fig. 25, except that Q3-Q4 and Q1-Q2 are Super-Alpha-connected in the differential mode, and Q5 and D1 are used as the constant-current generator. Resistor R. is given a value of 2 megarhms/volt, depending on the full-scale deflection voltage required.

To set up the circuit:
1. Short pins 9 and 6, and adjust R7 for zero meter reading.
2. Remove the short from pins 9 and 6, and adjust R9 and R10 for zero meter reading.
3. Connect the calculated value of R., and apply the required full-scale deflection voltage to the input terminals. Adjust R8 for full-scale deflection.
4. Repeat steps 1, 2, and 3, and trim if necessary.

Since this circuit has very high sensitivity some care is needed in the construction to insure stable operation. The circuit must be well screened, to keep stray pickup away from the input network. Resistors R1-R3 and R2-R4 must be mounted close together, so they operate at similar temperatures. All must be 5% or better, high-stability types.

This circuit, and those of Figs. 24 and 25, can be used as multirange meters, by switch-selecting R., values.

Schmitt trigger circuits

The high package density and gain-bandwidth products of the CA3018 make the IC ideal for compact, high-performance Schmitt trigger projects. Fig. 27 shows how to connect the IC as a simple Schmitt trigger with dc input and emitter-follower output. Transistors Q1-Q2 form the Schmitt, and Q4 acts as the emitter-follower output stage.

With zero input connected to the unit, the output is close to zero volts. When the input is raised to roughly \( V_{app} \approx 0.22 \), the Schmitt changes state and the output jumps up to a positive voltage. The trigger point of the circuit can be changed, if required, by using different R4 values. The circuit acts as a useful dc level sensor.

Fig. 28 shows how the circuit can be used as a simple sine/square converter. Here Q1 has a fixed bias applied via R1-R2 and R3, and the input signal is applied to the Schmitt via C1. The circuit needs a sine-wave input greater than 1 volt peak to peak for correct sine/square conversion, and gives a useful performance up to several hundred
kHz. Capacitor C2 is a speedup unit that reduces the output rise time.

Fig. 29 shows how to wire the IC so that it acts as a sensitive sine/square converter. Input impedance is 100,000 ohms and an input of only 50 mV peak to peak is needed for correct sine/square conversion.

Transistors Q3-Q4 are used as a Super-Alpha-connected common-emitter preamp with input connected to pin 9 and output direct-coupled to the input of the Q1-Q2 Schmitt trigger stage. The preamp is operated near saturation. The square-wave output of the circuit is taken from Q2's collector via C2, and the waveform is useful up to a couple of hundred kHz.

**Multivibrators to try**

The high package density and very high available current gain of the CA3018 also make the IC ideal for multivibrator projects. Fig. 30, for example, shows how the IC can be set up as a long-period stable multivibrator that operates at a frequency of 0.2 Hz—at 5 sec per cycle.

Transistors Q3-Q4 are wired in the Super-Alpha mode to form one half of the multivibrator. Transistors Q1-Q2 are wired in the Super-Alpha mode to form the other half. Because of the very high current gains of the Super-Alpha-connected transistors, the circuit can be given very high values of time-constant resistance (R1 and R2).

Two outputs are available from the circuit (one at pins 11 and 12, and the other at pins 5 and 9), and are in antiphase. The operating period of the circuit can be increased, if required, by increasing the values of C1 and C2.

---

**Fig. 27**—Schmitt trigger (right) is versatile pulse generator. Fig. 28 (below)—Schmitt trigger converts input sine wave to a square wave at the output.

**Fig. 29** (left), 30 (right) and 31 (below)—Sensitive sine/square converter, low-frequency oscillator and a long-period one-shot multivibrator, respectively.

---

**Fig. 31** shows how to connect the IC as a long period one-shot (or monostable) multivibrator, with emitter-follower output. Transistors Q3-Q4 are again wired in the Super-Alpha mode, and form one half of the multivibrator. Transistor Q2 forms the other half and Q1 is used as the emitter-follower output. The output of the circuit (taken from R7's slider) is normally at zero volts, but rises to positive for a fixed period whenever the monostable is triggered. The period has a duration of approximately 2.5 sec/µF of C1's value—of C1 is 10 µF, the period is 25 seconds. The circuit can be triggered manually by momentarily operating S1, or electronically by a positive pulse on pin 3.

(continued on page 101)
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NEW PRODUCTS

More information on new products is available from the manufacturers of items identified by a Reader Service number. Use the Reader Service Card on page 103 and circle the numbers of the new products on which you would like further information. Detach and mail the postage-paid card.

SOCKET WRENCH SET, Model No. 1001. This 14 piece, 3/4" square drive set can be used in seven different ways to drive both hex and square fasteners. Reach varies form 13/16" to 9/16" depending on combination. Includes reversible ratcheting handle with a short swing for close-quarter operation; 5/8" spin-nerextension with a drive socket insert permits use as a ratchet extension in addition to a regular nutdriver; nine sockets in hex sizes from 3/8" through 3/8"; two dual-purpose sockets with openings of 3/4" and 9/16" for square and hex nuts. Set is packaged in snap-lock case with a free set of personalized initials—Xcelite, Inc. Orchard Park, New York 14127.

Circle 31 on reader service card

CHILLING SPRAY. Blue Frost, tests entire assemblies for cold weather and high altitude operation. Troubleshoots intermittent heating circuits by spraying each suspected component for a few seconds. Healthy components will not be affected, but faulty ones will act up. To test entire units, place unit in cardboard box and flood with Blue Frost. Tight fitting metal parts that are difficult to mate can easily be pressed together if the inner part is first shrunk with the spray. Recommended for industrial or professional use only. Available in 8 oz. and 20 oz. cans—Tech Spray, P.O. Box 949, Amarillo, Texas 79165.

Circle 32 on reader service card

HI-FI SPEAKERS. Model 10, delivers response from under 200 Hz to over 13 kHz with a single wide-range driver. This high flux drive unit has a 3½-inch long throw cone with a 5-pound ferrite magnet. The driver is mounted in a sub-enclosure that performs as a miniature air suspension unit. The dome center produces even dispersion without a whizzer cone. Response of the woofer is above 200 Hz to eliminate boxiness or hints of a nasal effect. The free air cone resonance is just over 20 Hz. 8-ohm unit can handle 10-watts of audio. Enclosure measures 24x14x10½ inches . . . 369.95, Audionics, 9701 S.E. Mill, Portland, Oregon 97216.

Circle 33 on reader service card

COMP-UKIT LOGIC LAB I, consists of 10 assembled computer circuits, 20 patch cords and an instruction book. Dozens of games, experiments and puzzles can be assembled. No previous knowledge of computers or electronics required. It uses the same 54/74 TTL circuits used in many modern computers. Experiment with circuits that include a two-stage binary counter, a three-stage shift register, or a counter with decoding. Modules can be added one at a time for advanced logic. $39.00—Scientific Measurements Co., 9701 N. Kenton Ave., Skokie, Ill. 60076.

Circle 34 on reader service card

CASSETTE TEST TAPE, A-87 Tor-gette, is a calibrated torque indicator.
Our hot ones are the last to go.

The last thing you need is to be called back a day or two after you've replaced the sweep or high voltage tubes in somebody's color TV.
But, they're usually the first to go.
Because they get so hot.
So we figured out how to cool them.
Now, they last a lot longer.

Take our 6JE6C/6LQ6, for example. It's the horizontal deflection tube that takes such a beating when the set gets hot.

Well, we've given it special patented radiator fins that first absorb the heat and then radiate it out of the tube.
Now it runs cooler and lasts longer.
Same for our 6JS6C.
Or take our 6B4K4C/6EL4A. That's the shunt regulator that eliminates runaway high voltage. We gave this one a whole new anode and shield design to improve heat transfer and stability.

Now it also runs cooler and lasts longer.
Or take our 3A3B high voltage rectifier. This one's got leaded glass for added protection. And it lasts longer too.
So next time you have to replace any of the hot ones, just cool it.
You'll both last longer.

SYLVANIA
GENERAL TELEPHONE & ELECTRONICS
User inserts Torquette in tape deck, and the tape take-up torque in grams per centimeter is read through the window by pressing the play button. Indicates condition of the clutch assembly, has high degree of repeatability, is direct reading and very accurate. Eliminates guess work for all technicians. $6.90—Hartak Instrument Co., P. O. Box 1, Frazer, Colorado 80442.

Circle 35 on reader service card

TRIGGERED OSCILLOSCOPE, Model 555. This 5” scope has a broad bandwidth (10-MHz, 5x expansion), sensitivity of 20 mV/cm. Solid-state circuitry, 9-position attenuator, and a built-in square-wave oscillator for calibration functions. A wide sweep range of 1 µsec/cm to 1 sec/cm handles both low frequency and transient signals. Weighs 24 lbs., measures 8” X 11” X 17”. $346. —Kikusui Electronics, c/o Marubeni-Ida, Inc., 200 Park Ave., New York, N.Y. 10017.

Circle 36 on reader service card

CB BASE STATION, Model Titan III, Transmits and receives on either the upper or lower sideband of each channel providing 46 options. It also operates as conventional unit. Single sideband eliminates normal AM carrier and separates each channel into its upper and lower sidebands. The unit provides the F.C.C. maximum of 15 watts P.E.P. power input. While transmitting full carrier AM, output is 3.5 watts minimum. Unit is equipped with 23 crystals with a tolerance of ±0.015%. $598.—Tram Corp., Lower Bay Road, Box 187, Winnisquam, New Hampshire 03289.

Circle 37 on reader service card

RAYGUARD, is a device to warn of potentially dangerous X-radiation from color television sets. It monitors high voltage in the set, and warns the owner, on a small meter, when the voltage exceeds levels set by the set manufacturer. Consists of a sensor unit which is installed in the set, and a meter placed in a bracket in the rear of the set. A needle on the meter enters a danger area when voltage exceeds limits.—Audio Equipment Co., Div. of Walter Kidde & Co., Clifton, New Jersey 07012

Circle 38 on reader service card

AC ADAPTER, Model BC9, for portable radios and cassette recorders. Has reverse polarity plug to adapt for both positive- and negative-ground equipment. Also has a four-way connector that matches jacks on most equipment—3.5 mm plug, 3.5-mm jack, 2.1-mm coax power plug, and 2.5-mm miniature plug. In addition, the adapter has a switch to select 6- or 9-volt outputs.—Workman Electronic Prods., P.O. Box 3828, Sarasota, Fla. 33575.

Circle 39 on reader service card

(continued on page 85)

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LUBRA CLEAN

...A T.V. TUNER CLEANER THAT REALLY WORKS!

MADE ESPECIALLY FOR TELEVISION TUNERS

Finally, a tuner cleaner that really works. Apply LUBRA CLEAN and let it do the work for you. Physical cleaning and polishing contacts is no longer necessary. Simply apply—and LUBRA CLEAN continues to polish and lubricate tuner contacts for months after all other cleaners and lubricants have failed. Sold at leading dealers and distributors for $2.98 per tube.

LUBRA CLEAN COMPANY

P. O. BOX 926 - MARTINSVILLE, VA. 24112 - PHONE (703) 638-1211

Circle 61 on reader service card
The new Permacolor antenna from RCA has:

"Long-lasting" element/ feed line connections.

Waterproof, polypropylene insulators that pivot.

Integrated UHF bow tie.

Rigid, square boom.

Tough vinyl finish.

That ought to prove RCA is serious about the antenna business.

So serious we've set up a new production facility at our Memphis plant. But before we manufactured a single antenna, our engineers literally started designing from scratch. The result? Permacolor.

Here are just a few reasons why RCA Permacolor antennas are different, and are an improvement over what you're now selling.

1. Durable connections/Perma-tuned circuits. All active elements (many perform more than one function) are solidly connected to a symmetrical aluminum feed line by riveted straps. This reduces reception failure due to flimsy or intermittent connections.

2. "Single unit" insulator/element. Insulators are polypropylene. Elements extend 5½ inches into the insulators. Elements and insulators pivot as a unit—lock in place—have no loose connections. No high stress points.

3. Bow tie and 110° corner reflector UHF Section integrated into a single downlead for better UHF reception.

4. Easy installation. Simply unfold and install. Permacolor antennas go up in one piece. No bag of bolts. Nothing to take apart and reassemble. Snap-off elements let you quickly tailor the antenna to reception requirements.

5. Tough, handsome blue and gold vinyl coating protects against weather and airborne chemicals.

The new RCA Permacolor Antenna is the antenna you can put up for good. See it now at your RCA Parts and Accessories distributor.

Parts and Accessories, Deptford, N.J.
New from Heath...in time

New Heathkit® solid-state modular color TV

The result of over five years in research and development, these sets represent one of today's greatest color TV values. Here's why: a total of 45 transistors, 55 diodes, 2 silicon controlled rectifiers, 4 IC's containing another 46 transistors and 21 diodes plus 2 tubes (picture and high voltage rectifier) combine to deliver performance and reliability unequalled by any conventional tube set. Other features include: MOSFET VHF tuner; high-gain 3-stage solid-state IF; emitter-follower output; automatic fine tuning; VHF power tuning; built-in degaussing plus manual degaussing coil; automatic chroma control; adjustable noise limiting and gated AGC; "instant-on"... sound instantly, picture in seconds; bonded-face, etched glass picture tubes; adjustable tone control; exclusive hi-fi outputs; and 48-hour factory service facility for modules. The sets are designed to be owner-serviced... the only sets on the market with this exclusive feature. A build-in dot generator, volt-ohm meter, and modular snap-out epoxy circuit boards make routine adjustments and service a snap... virtually eliminating service calls and offering significant savings over the life of the set. It all adds up to the color TV buy of a lifetime in the GR-270 and GR-370... ready now for Christmas giving!

Kit GR-270, 227" 20v tube, 114 lbs. .................................. $489.95*
Kit GR-370, 295" 23v tube, 127 lbs. .................................. $559.95*
Kit GR-370MX, GR-370 with RCA matrix tube, 127 lbs. ........ $569.95*

Exclusive modular design

1 Exclusive check-out meter
2 Tilt-out convergence/secondary control panel
3 Gun shorting switches
4 3-stage IF assembly
5 Plug-in AGC/Sync circuit board
6 Plug-in 3.58 MHz oscillator circuit board
7 Plug-in Chroma circuit board
8 Plug-in Luminance circuit board
9 Service and Dots switches
10 Plug-in Video Output circuit board
11 High voltage power supply
12 Plug-in Vertical Oscillator circuit board
13 Plug-in Horizontal Oscillator circuit board
14 Plug-in Pincushion circuit board
15 Conservatively-rated power supply components
16 Circuit breaker protection
17 Plug-in Sound circuit board
18 Master control panel
19 Hi-fi sound output
20 Plug-in wiring harnesses and connectors for faster assembly

Choice of factory-assembled cabinets

3 models in 295 sq. in.

Lustrous Mediterranean Cabinet...factory assembled of fine furniture grade hardwoods and finished in a flawless Mediterranean pecan. Statuary brome trim handle, 30½" H x 47" W x 17¾" D. Assembled GRA-204-23, 78 lbs. ......... $129.95*

Deluxe Early American Cabinet...factory assembled of hardwoods & veneers and finished in classic Salem Maple. 29½" H x 37¾" W x 19¼" D. Assembled GBA-303-23, 73 lbs. ........... $114.95*

Contemporary Walnut Cabinet and Base Combination. Handsome walnut-finished cabinet sits on a matching walnut base. Cabinet dimensions 20½" H x 31½" W x 19¼" D. Base dimensions 24½" H x 27¼" W x 18½" D. Assembled GRA-203-20 Cabinet, 46 lbs. $49.95* GRS-203-6 above cab. w/matching base, 59 lbs. .......... $59.95*

Handy Roll-Around Cart and Cabinet Combination. Features the GRA-203-20 walnut cabinet plus a walnut-trimmed wheeled cart with storage shelf. Assembled GBA-203-20 Cabinet, 46 lbs. $49.95* GRA-204-20 Roll-Around Cart, 19 lbs. ... $19.95* GRS-203-5 Cart & Cabinet Combo, 65 lbs. ......... $59.95*

3 models in 227 sq. in.

Exciting Mediterranean Cabinet...factory assembled using fine furniture techniques and finished in stylish Mediterranean pecan. Accented with statuary brome handle, 27½" H x 41½" W x 19½" D. Assembled GBA-202-20, 85 lbs. ............. $114.95*

Contemporary Walnut Cabinet and Base Combination. Assembled GBA-203-20 Cabinet, 46 lbs. $49.95* GRS-203-6 above cab. w/matching base, 59 lbs. ......... $59.95*

RADIO-ELECTRONICS
Imagine a baked potato in 4 minutes; baked beans in a little over 6 minutes; a five-pound roast in 45 minutes. This is the miracle of microwave cooking. And now Heath brings you this modern miracle for the first time in money-saving, easy-to-assemble kit form. For busy families on the go, meal preparation is a matter of minutes. You can cook on china, glass, or even paper dishes since only the food becomes hot. Your cooking dish can be your serving dish. Frozen foods can be defrosted in minutes for quick spur-of-the-moment frozen meals cooked right in their own containers. And there is not the slightest cause for concern about the safety of your Heathkit electronic oven. Exclusive door design prevents microwave leakage from the oven cavity. And with a SAFETY INTERLOCK SYSTEM UNIQUE IN THE INDUSTRY, not only does the oven stop cooking if the door is opened, but the door can't be opened unless the interlock is operating properly. A second independent door interlock is also provided for maximum protection. And all interlock mechanisms are tamperproof. Assembled in accordance with the manual, the GD-29 meets all the new federal standards for safety and radio interference. No special precautions are required when operating. The Heathkit electronic oven is as safe as your conventional oven! Quality components are used throughout: magnetron tube by Litton, the uncontested leader in the field; avalanche diode circuitry for longer tube life; simplified wiring harness with push-on quick-connectors for reliability and ease of assembly. GD-29 prototypes endured grueling "life-tests" equivalent to over 60 years of continuous service... further assurance of uncompromised reliability. Another feature is portability: the Heathkit electronic oven operates on regular household current. Plug it in anywhere... on a countertop, a wheeled cart, in the kitchen, on the patio, at the cottage... anywhere... a grounded 120V AC power outlet is available. Make this a Christmas to remember by putting a Heathkit electronic oven under the tree. It's a gift your wife will thrill to... and a present the whole family will enjoy... meal after meal after meal.

Kit GD-29, 80 lbs. ........................................................................... $399.95*
New Heathkit® AJ-29
AM-FM-FM stereo tuner

This is the feature-packed tuner section of the famous Heathkit AR-29 stereo receiver... now available as a stereo "separate." The pre-assembled, factory-aligned FM tuner boasts 1.8 uV sensitivity for WHOPPING station pulling power using FET design for superior overload characteristics. Three IC's in the IF section offer superior AM rejection, hard limiting, temperature stability, and outstanding reliability. Other features include a computer-designed 9-pole L-C filter for greater than 70 dB selectivity, new "blend" and "mute" functions, and a built-in AM rod antenna that swivels for best reception.

Kit AJ-29, 19 lbs., less cabinet.......................... $169.95*
Assembled AE-19, oiled pecan cab., 9 lbs. .............. $19.95*

New Heathkit® AA-29
100-watt stereo amplifier

Power-packed amplifier section of the Heathkit AR-29, the AA-29 stereo "separate" marks another milestone in superior Heathkit amplifier design. Its 70-watts of continuous power is more than enough to drive even the most inefficient speaker systems. A massive, fully-regulated and filtered power supply, 4 conservatively heat sunked output transistors and the best IM and harmonic distortion specifications in the industry add up to sound fidelity you never expected to hear outside the theater. Modular plug-in circuit boards make assembly easier... snap out in seconds for future servicing.

Kit AA-29, 27 lbs., less cabinet.......................... $149.95*
Assembled AE-19, oiled pecan cab., 9 lbs. .............. $19.95*

New Heathkit® floor model speaker systems

In the new Heathkit AS-101 and AS-102 speaker systems, Heath engineers have combined the best of both worlds of sound and beauty. The AS-101 Heath/Altec-Lansing 2-way system features a 15" woofer and sectoral horn delivering from 35 to 22,000 Hz with uncompromising accuracy. The AS-102 Heath/Bozak 3-way system uses a 12" woofer, 6" mid-range, and two 2½" tweeters in an infinite baffle design to produce clean natural reproduction from 40 to 20,000 Hz. Both systems are housed in assembled Mediterranean pecan cabinets, 29½" H x 27 ¼" W x 19½" D.

Kit AS-101, 53 lbs. ........................................ $259.95*
Kit AS-102, 39 lbs. ........................................ $259.95*

New Heath stereo equipment credenza

Romantic Mediterranean styling in wife-pleasing one-piece console design... yet with plenty of room for your favorite separate stereo components. Six-and-a-half feet of solid craftsmanship executed in North American Hickory veneers and solid oak trim, finished in oiled pecan. Completely assembled and finished, ready for installation of Heath or other components. Speaker enclosures are dUCED port reflex design, pre-cut for 12" speakers. An adjustable shelf has room for stereo receiver, cartridge or cassette tape player or separate tuner and amplifier. Below the shelf is room for your turntable and record storage. Accessory matching drawers on ball bearing slides are available for turntable and tape player.

Model AE-101, 90 lbs. ....................................... $189.95*
New Heathkit® IC15 MHz frequency counter...$199.95*

A highly accurate, low cost frequency meter for anyone requiring accurate frequency measurements. Compare these features to counters selling for over twice this low price: accurate counting, 1 Hz to over 15 MHz; integrated circuitry; automatic trigger level for wide range input without adjustment; five digit readout with Hz/KHz ranges and overrange indicators for eight digit capability; high input impedance; storage circuitry for non-blinking, no-count-up readout; computer-type circuitry, no divider chain adjustment, temperature-compensated crystal time base oscillator; BNC input with cable; double-sided, plated-thru circuit board with sockets; three-wire, removable line cord; heavy-duty aluminum case handle/tilt stand and die cast zinc front panel; no special instruments required for accurate calibration.

Kit IB-101, 7 lbs ..................................................$199.95*

See these and 300 other Heath-gift suggestions at one of the following Heathkit Electronic Centers:

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Chicago Area
Downers Grove, Ill. 60515
224 Ogden Avenue

Cleveland, Ohio 44129
5444 Pearl Road

Dallas, Texas 75201
2715 Ross Avenue

Denver, Colorado 80212
5940 W. 38th Ave.

Detroit, Michigan 48219
18645 W. 8 Mile Road

Fair Lawn, N. J. 07410
35-07 Broadway (Rt. 4)

Houston, Texas 77027
3705 Westheimer

Los Angeles, Calif. 90007
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Milwaukee, Wisc. 53216
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Hopkins, Minn. 55343
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New York, N.Y. 10036
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Philadelphia, Pa. 19149
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Pittsburgh, Pa. 15235
3482 William Penn Highway

St. Louis, Mo. 63123
9296 Gravois Ave.

San Diego Area
LaMesa, Calif. 92041
8363 Center Drive

San Francisco Area
Redwood City, Calif. 94063
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Washington, D. C. Area
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Healthkit Electronic Center Prices Slightly Higher.

During 1971, consult Heathkit Catalog Supplements and local newspapers for announcements of new Heathkit Electronic Centers opening in these places:

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El Cerrito, California

Cincinnati, Ohio

Los Angeles Area
Woodland Hills, California

Atlanta, Georgia

Prices listed are factory mail order. Retail prices are slightly higher.

...or send for your FREE factory mail order catalog

Circle 62 on reader service card

New Heathkit® HW-101 SSB transceiver

The Hams at Heath have done it again . . . with an uprated version of the Heathkit HW-100, one of the most popular pieces of gear on the market! The HW-101 features improved receiver circuitry resulting in better than 0.35 µV sensitivity for 10 dB S+N/N. Image and IF rejection are better than 50 dB. Other improvements are a new 16-1 ball-bearing dial drive; new selectable SSB or CW filters and attractive new front-panel styling.

Kit HW-101, 23 lbs .............................................$249.95*

New Heathkit wattmeter/ SWR bridge

Two switch-selected ranges allow measurement of RF output from 10-200 W and 100-2000 W. Built-in calibrator permits 10% accuracy throughout the 80-10 M ham bands.

Kit HM-102, 3 lbs .............................................$29.95*

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[Credit Application Form]

*Mail order prices: F.O.B. factory

Prices & specifications subject to change without notice.

CL-394
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ITS A TRF AM RECEIVER
An AM Voice Transmitter, a Modulated Code Oscillator contains:

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- RF Oscillator
- Ferrite Coil
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Complete with circuit diagram.

Honeywell Boards 3/4" x 6". Loaded with later no. transistors, diodes, resistors and capacitors. 2 different boards $1.00. 2 lb. Stock No. 9094

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Circle 65 on reader service card

EQUIPMENT REPORT

GBC CLOSED-CIRCUIT TV SYSTEM

For manufacturer's literature, circle No. 96 on Reader Service Card.

A CLOSED CIRCUIT TELEVISION SYSTEM has many everyday applications. Some of them include seeing who's at the front door before opening it and keeping an eye on poolside activities from your kitchen. Combined with a video tape recorder, it makes a complete video tape system. Use the camera from the CCTV system to take the pictures you want to tape, the recorder to tape them, and the monitor to view what you taped.

CCTV systems come in many combinations. A basic system consists of a camera and monitor. A better system, like the GBC MC-15 described here, has a built-in audio channel.

The camera in the GBC system is light-weight and readily portable. It has only one cable dangling behind it. This links the video and audio signals from the camera to the monitor and carries operating power to the camera. The camera has several interesting features. Its lens automatically adjusts to existing light levels to produce a proper picture. A microphone is built into the camera.

The only controls on the camera are focus and an on-off switch. This makes operation simple. An indicator lamp lets you know when the power is on. There is a mount for attaching the camera to a tripod.

The operating controls are on the monitor. There's the inevitable on-off switch, brightness and contrast controls, a volume control for the audio plus a push-to-talk button (the camera microphone becomes a speaker) and a focus adjustment that permits limited variation of focus from the monitor. The only other control is a standby switch.

After using the system I found picture quality excellent, the camera and monitor easy to use. The 2-way audio takes a little getting used to, but as soon as you remember not to talk while the other party does, you're in business. All in all, a system worth owning.—Warren Roy
NEW BOOKS

Cbers' HOW-TO BOOK, by Leo G. Sandi. Hayden Book Co., Inc. 116 W. 14 St., New York, N.Y. 10011. 8% x 5%", 120 pp. Softcover, $3.50.

Explains how to operate CB gear and includes tables for all CB codes and channels. Tells how to select and install the right unit, with analysis of mobile unit and base-station features. Walkie-talkies and phone patches also discussed. Covers test and repair techniques and modifications to improve operation.

1970 POPULAR TUBE/TRANSISTOR SUBSTITUTION GUIDE, by TABB editorial staff. Tab Books, Blue Ridge Summit, Pa. 17721. 224 pages, 5% x 8% in. Leatherette, $4.95, paper, $2.95.

First four sections cross reference popular American receiving tubes; commercial and industrial tubes are also covered. Third and fourth sections show substitutes for foreign tubes and base diagrams. Final four sections list transistors in a similar format.


Early chapters describe analog and digital computers and basic logic circuits. Other topics discussed are: information theory, automation deductive logic, inductive logic; later chapters discuss possibilities that a computer might show true intelligence.


Supplements operating instructions of a scope by describing how to use them, for everything from finding a diode's switching time to the hysteresis factor of a memory core. After scope basics, chapters on voltage, current, time, frequency and phase measurements are described. Other chapters explain scope use with sweep generators, checking components, amplifiers, communications gear, TV and industrial items.


Presents circuits for color TV reception from antenna input through to a decoder. Also introduces decoder design for color systems in use in different countries. Text will serve as an introduction to assessment of TV circuit performance in terms of waveform design. Covers signal-to-noise problems of vhf and uhf TV reception. Includes receiver input circuit, designs for low distortion in the luminance and chrominance channels. NTSC decoder, PAL decoder, SECAM decoder and NR decoders, applications of waveform measurement techniques.

ALARM SYSTEM DIRECTORY

If you can't come to the world's newest, largest and most exciting electronics department store, well mail the store to you!

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Allied TD-1099. 3-Head Stereo Tape Deck. One-piece head makes tape threading a snap. Walnut case. $179.95

Realistic STA-120. Wideband AM, FET-FM Stereo Receiver, 140 Watts. With case. $269.95

1971 Allied Radio Shack Catalog

Mail Coupon To Address Below or bring to Allied Radio Shack store for new 1971 catalog!

ALARM SYSTEM DIRECTORY

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Adcor Electronics, 349 Peachtree Hills Ave. Atlanta, Georgia 30305

Advanced Devices Laboratory 701 Kings Row San Jose, Calif. 95112

Aerolarm P.O. Box 806 Butler, N.J. 07405

Alarm Device Mfg. Co. 165 Eileen Way Syosset, N.Y. 11791

Alarm Prods., International, Inc. 24-02 40th Ave. L.I.C. New York 11101

Alarm Systems Co. 31807 Uteca Rd. Fraser, Mich. 48026

Alarmtronics Engineering, Inc. 154 California St. Newton, Mass. 02195

Almact Industries, Inc. 5201 N.E. 12th Ave. Ft. Lauderdale, Fla. 33308

Allied Radio Corp. 100 N. Western Ave. Chicago, Ill. 60680

American District Telegraph Co. 155 Sixth Ave. New York, N.Y. 10013

Ampex Corporation 401 Broadway Redwood City, Calif.

(continued on page 94)
Compare these features:

- Solid state circuitry
- High sensitivity — 10MV/cm
- Low heat dissipation
- Extra wide bandwidth — 7 MHz
- Lightweight — only 8½" high
- Simplified controls makes for easy operation
- Extremely reliable performance
- Built-in square wave oscillator for calibration functions
- Vertical channels
- AC/DC inputs
- FET protected input
- Wide sweep range — 1 μs/cm

ORDER ONE TODAY!
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Circle 67 on reader service card

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Circle 68 on reader service card

COMING NEXT MONTH

DECEMBER 1970

- Build A Computer Lab ICs, patchboards, and logic. It all adds up to a convenient way to learn how computers work by setting up actual computer logic circuits

- Designing Semiconductor Hi-Fi Stereo Amplifiers
  First in a series of articles that shows how to custom-design the features you want into a solid-state amplifier

- Darkroom Temperature Monitor
  Know what your solution temperatures are instantly and precisely

- Building Solid-State Power Supplies
  Matt Mandl tells how to make the power supply you need for that special application

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Jack Darr’s Service Clinic
Your Career In Electronics
Kwik-Fix Troubleshooting Charts
NEW LITERATURE

All booklets, catalogs, charts, data sheets and other literature listed here with a Reader's Service number are free for the asking. Turn to the Reader's Service Card facing page 102 and circle the numbers of the items you want. Then detach and mail the card. No postage required!

**NEW PRODUCTS**

(continued from page 74)

CONTACT OVERHAUL KIT, for TV tuners, contains cleaning, polishing, and lubricating materials for switch contacts in all color and black & white TV tuners. Includes aerosol spray of degreaser/cleaner for high pressure removal of "muck" from the surfaces with minimum of frost. An aerosol spray of lubricant completes the kit. Comes with instructions...

$5.50—Castle TV Tuner Service, 5710 N. Western Av., Chicago, Ill. Circle 40 on reader service card

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Circle 43 on reader service card

**Electronic Kits Catalog, EICO 1970**

Features do-it-yourself kits and factory-assembled equipment. Lists color organs, strobes, transistors, solid-state test instruments, electronic project kits, amateur shortwave equipment, Cortina stereo components, oscilloscopes, generators, power supplies, multimeters, probes and many others. Also includes an index of authorized dealers—Eico, 283 Malta St., Biklyn, New York 11207

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**ELECTRONIC KIT CATALOG**

EICO 1970. Features do-it-yourself kits and factory-assembled electronic equipment. Lists color organs, strobes, transistors, solid-state test instruments, electronic project kits, amateur shortwave equipment, Cortina stereo components, oscilloscopes, generators, power supplies, multimeters, probes and many others. Also includes an index of authorized dealers—Eico, 283 Malta St., Biklyn, New York 11207

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Only Injectorall's new BLUE LUBE foam spray contains Oxidant — an anti-oxidation chemical that cleans contacts, cleans them permanently clean and prevents callbacks. Eliminates picture streaking and intermit-tents with one application.

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Only Injectorall's new BLUE LUBE foam spray contains Oxidant — an anti-oxidation chemical that cleans contacts, cleans them permanently clean and prevents callbacks. Eliminates picture streaking and intermit-tents with one application.

Circle 72 on reader service card

IN THE SHOP
(continued from page 17)
times the normal current. Likely sus-pects, an internal leakage in the transis-tor or incorrect bias voltage on the base. So that's where we should start looking for the trouble.

We must learn to think in terms of "What is the normal reaction in this cir-cuit?" In tube circuits they were all the same — more negative grid voltage, less plate current. In transistor circuits, a more-negative base voltage means less collector current only if the transistor is an npn! A negative voltage is a reverse bias (in the direction which reduces col-lector current) for nnp's. For pnp's, a more-positive voltage reduces the collector current, since a positive voltage is reverse-bias for these.

I know that this is going to be hard to do, for some of us "set in our ways" old-timers. In fact, every now and then you'll probably find "up" or "down" sneaking into this space! The habits of a lifetime aren't easy to break; I write "Hertz" but I still say "cycles!" How-ever, if you can discipline your thinking to use these terms, it will be a big help in dealing with circuit reactions in mod-ern electronic equipment. And, frankly, the way things are now, we can use all the help we can get!

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Service Clinic

By JACK DARR

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This column is for your service problems—TV, radio, audio or gen-eral and industrial electronics. We answer all questions individually by mail, free of charge, and the more in teresting ones will be printed here.

If you're really stuck, write us. We'll do our best to help you. Don't forget to enclose a stamped, self-addressed envelope. Write: Service Editor, Radio-Electronics, 200 Park Ave., South, New York 10003.

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Negative Pix, Old Conar TV

I've got a very weak negative picture on an old Conar 70 TV that I'm fixing up...J. B., Dalmays, Va.

Check the 6A58 used as the 3rd video i.f. If the diode (video detector), has mount-leakage or is shorted that'll do it. Also see if the video out put amplifier plate load resistor hasn't gone away up in value. This can cause the same symptoms too.

Intermittent color flashes

This color set works perfectly most of the time; every now and then the screen suddenly flashes a bright...
blue. Sometimes, I can tap the chassis and make it flash. Replaced tubes, checked picture tube, etc. Do you think this could be an intermittent heater-cathode short in the CRT? — R. G., Chickasha, Okla.

Possible, but let's exhaust all of the cheaper alternatives first! For one common cause of this symptom, check the plate connections to the B-Y color difference amplifier tube. If, as usual, there's a tiny rf choke in there, this can be intermittently opening. This lets the plate voltage jump away positive, and turns the blue gun hard on, causing the flash. Intermittent solier joints on the PC board in the same circuit can also cause identical symptoms.

This can affect any color, of course, not blue alone.

Fanout of R, G, B lines

The red, green, and blue vertical lines fan out like a palm tree in this Silvertone color TV chassis. Controls don't seem to have a lot of effect on it.

Top and bottom of the screen. What's going on?— A. M., Lake Charles, La.

This has been known to happen in these sets when the vertical centering control opens. This upsets the vertical convergence waveforms. Easy to check; just see if the vertical centering control will move the picture up and down. If it won't, there you are.

Yoke replacement

I have a Sharp TSP109A, with a bad yoke. Can't find a replacement listed for it, and can't get one from the manufacturer. Do you know of a U.S. manufacturer who makes one?— D. K., New York.

The Merit MDF-127 will replace this one. Your distributor may have to order it specially, but it'll fit.

Faint vertical bars

I've got faint vertical bars all the way across the screen, on a Sylvania 912 color chassis. These seem to show up mostly on black and white pictures. Checked tubes, etc. No help.— D. Q., Bangor, Maine.

Check the horizontal blanking amplifier transistor, Q604 (Q14 in 'photofacts). If you find the same pulse waveform on both base and emitter, there you are. This transistor is leaky. Replace it.

Hot horizontal output transistor

I've got a mysterious ailment in a title transistor TV. The horizontal output transistor gets very hot, I have a high voltage, and I get too much screen. Everything else seems OK; sets in the flyback circuits, and supply voltage is a little low. What's going on here? Tried a new output transistor, and it gets hot too.— P. D., Tucson, Ariz.

Check the frequency of your horizontal oscillator. The drive pulses in transistor horizontal output stages must be of the right width, to get the correct ON time for the output transistor. They must also be at the right frequency. If the oscillator is running much too fast, you'll have pulses at the input, but there will be too many of them. This has the same effect as a too-wide pulse: it keeps the output transistor turned on too long, thus drawing too much current. Check with scope, comparing to video signal.

One line or three?

When I throw the switch to the service position in color sets, I often get three separate lines. I thought you were supposed to get a single white line when you had the screens set correctly. How come?— R. G., Altoona, Penna.

When you set the switch to 'service,' you kill the vertical output stage. So, you lose your vertical dynamic convergence waveforms. Therefore, the three lines may or may not cover each other. Pay no attention to this; just set the three lines to equal intensity and turn it back to normal. Personally, I like to set screens so that the lines have just gone. I think you get a more accurate adjustment by doing this. Others differ and have their own favorite adjustment.

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Circle 74 on reader service card

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**LOGIC DESIGN COURSE**

Recently introduced by the Center for Technical Development, a new Ohio-based company, the Logic Design Program DDP-1 consists of a 6 x 10-inch PC board, a 13-lesson loose-leaf notebook along with 13 programmed learning tests and 10 breadboard project instruction sheets. All components for the lessons are provided with the kit, and the parts may be removed from the push-in terminals for repeated use.

By lesson three, you've derived the alpha and beta formulations for transistors, covered the forward and reverse bias characteristics of diodes, and have been introduced to transistor load curves. Lesson four looks at design procedures for simple inverters, showing how to find IӨ, RӨ, and Rө. Five is a study of emitter follower operation. Starting with lesson four, separate breadboard lesson sheets tell you how to install components on the PC board to measure (a voltmeter is needed) and observe what you've been studying.

The remaining lessons (6-14) describe logic circuits, starting with diode gates and truth tables. Diode transistor logic (DTL) and resistor transistor logic (RTL) are covered in seven and eight—with corresponding breadboard projects. The on or off states of the circuits is indicated by small lamps, operated with two switches premounted on the board.

Next comes NAND/NOR gate design and set/reset flip-flops. Here's how a typical design example goes: "Design an SK flip-flop, using NOR gates capable of driving 6-volt, 150-ohm lamps. Use germanium transistors type 2N404. Make Vө = +3V and Rө = Rө. The lesson shows how to complete this design. At this stage all four of the premounted transistors are used in the breadboard projects.

The final four lessons (11-14) are on bistable and monostable flip-flops and Schmitt trigger design. Also covered are differentiation networks and commutating capacitors. A scope is useful—but not necessary—with these lessons to observe waveforms.

The course packs a lot of basic solid-state design theory into each of the brief lessons (each one is about 6 pages long). Examples and solutions for important design principles are provided in the lesson and tested in the sheets of the programmed course.

The breadboard projects provide practice at wiring up components from simple schematics, and classroom instructors should find the non-permanent feature useful. Cost of the course is $99. —John Williams

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Circle 76 on reader service card
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The A297 is a new neon lamp specifically designed for long-life use in a seven-bar numeric display. It is constructed with 7-mm electrodes in a glass envelope considerably longer than is necessary for an ordinary neon lamp. The longer glass envelope and shorter electrodes eliminates the possibility of characteristics changes due to sputtering. As a result, the tip remains clear and its effective life, measured by light output, is considerably longer than a device with normal length electrodes in the envelope. Light is taken from the end of the envelope and applied to its segment of the seven-bar display.

The Signalite A297 has a breakdown voltage rating of 215 volts max and a maintaining voltage rating of 128 to 148 volts. Design current is 5 mA. The lamp is in a T2-size envelope that is 27/32 inch long and 0.244 inch in diameter.

For information, contact Signalite Inc., Neptune, N.J.

EPOXY SCR'S

The 2N5787 through 2N5792 are six new low-cost Fairchild SCR's with forward and reverse voltage ratings from 30 to 400 volts. They are electrical equivalents of the 2N5060 through 2N5063 and the Texas Instruments TIC-44 through TIC-47. Gate sensitivity is 10 mA (typical) and on-state voltage drop is 1.0 volt (typical at 200 mA). All have a maximum current rating of 500 mA. Housing is Fairchild's TO-106 package, which has the same circular pin configuration as the TO-18 metal can. Guaranteed to operate at junction temperatures of -65° to +125°C. Ratings and prices (1 to 99) of these Fairchild devices are: 2N5787, 30 V, $0.77; 2N5788, 60 V, $0.83; 2N5789, 100 V, $0.96; 2N5790, 200 V, $1.09; 2N5791, 300 V, $1.61 and 2N5792, 400 V, $2.00.

EPDXY SCR'S

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NOVEMBER 1970
TRY THIS ONE

DEMANDIZER

A simple tape head demagnetizer can quickly be made by attaching No. 14 household wire formed into a two- or three-inch coil to your pistol grip soldering gun. A reminder for this as well as all demagnetizers, always withdraw the field from the head before turning power off. — John F. Agee

REMOVING PC COMPONENTS

When you have to remove multiple-terminal components like tube sockets; if transformers, etc. from printed circuit boards use this trick. Rather than heating each terminal successively and hoping that the solder remains molten until all the terminals are heated, form a piece of No. 12 or No. 14 wire to fit down on the terminals of the part to be removed. Tin the wire and make sure it touches all the terminals. Then apply heat to two or three places along this "heat sink" with your soldering gun. The part may then be removed with a minimum of heat—avoiding needless damage to circuit board or component. — H. Joseph

FUSE SAVER

On some color TV's you find circuit breakers, and on others plug-in fuses. But when we make a service call and find a set with a blown fuse, we don't plug in a new fuse. Instead we made up a circuit breaker by taking the old plug-in fuse pins and soldering them to a breaker. Then we plug in the breaker until we find what is causing the fuse to blow. When repaired, we plug in the regular fuse. This way you can save a lot of fuses. — Harry J. Miller

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(continued from page 83)

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Circle 83 on reader service card

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May be used with electrostatic speakers.

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Z-30 $15.95

Z-50 $19.85

The Stereo 60 module is a matching unit in performance for use with the Z-30/50 modules but may be used with other amplifiers. The Stereo 60 features low distortion (less than 0.05%), three input tape output, a frequency response from 100 Hz to 10,000 Hz, and a channel separation of better than 70 db. The active FILTER UNIT may be used with the Project 60 systems plus specialized applications requiring a continuously variable rumble and scratch filtering system. The AFU uses Sallen & Key active filter stages to provide rapid rejection of unwanted frequencies (12 db per octave).

H.F. FILTER: Variable 5 KHz to 28 KHz.

L.F. FILTER: Variable 25 Hz to 100 Hz.

SINCLAIR ACTIVE FILTER UNIT $20.95

POWER SUPPLIES:
PZ-5: A standard 25-30 Vdc supply for use with the Project 60 modules $13.95

PZ-6: A regulated supply adjustable from 15-45 Vdc; ½ amp at 35 Vdc. For use with 25-30v when maximum power is required $23.95

PZ-8: A regulated supply adjustable from 35-50 Vdc providing over 3-amps of current. For use with 25-50 when maximum power is required $27.95

SINCLAIR PROJECT 60 SYSTEMS are available from many dealers. If your dealer does not yet stock Sinclair products, order direct from Audionics.

AUDIONICS, INC.
9701 S.E. Mill Street, Portland, Oregon 97216

Please ship the following:

#1 Two Z-30's, PZ-5 and Stereo 60 $74.95

#2 Two Z-30's, PZ-6 and Stereo 60 $84.95

#3 Two Z-50's and Stereo 60 $90.95

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CB Troubleshooter's Casebook
Compiled by Andrew J. Mueller

Case 1: Radio is dead. New batteries go dead when the switch is on for 30 seconds.

Common to: Midland 13-775B

Remedy: Replace audio transformer, T1.
Reasoning: The audio output transformer primary has shorted to the core. This connects battery voltage directly to ground. Replacing T1 will restore normal operation.

Case 3: Radio receives poorly. Transmit is ok.

Common to: Pearce-Simpson Sentry

Remedy: Exchange C1 and C2
Reasoning: C1 and C2 influence the local oscillator injection voltage. From the factory these parts were reversed, reducing receiver sensitivity. When they are properly installed, sensitivity will return to normal. The schematic shows their proper locations.
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Finally, Fig. 32 shows how the IC can be wired as a r/c/dc oscilloscope-calibrator. In this circuit, Q1 and Q2 are wired as a free-running (astable) multivibrator, operating at a frequency of about 1 KHz. Diodes D1 and D2 protect the transistors against reverse base-emitter breakdown, and R5 allows presetting the mark-space ratio.

Output is clipped at about 6.5 volts by Zener-connected Q3, and is passed on to the R7-R10 attenuator network via emitter follower Q4. In use, R7 is adjusted so that a signal of 5 volts peak appears at the R7-R8 junction. Calibration voltages of 50 mV, 500 mV and 5 volts are then available at the outputs. When S1 is set to the dc position, the multivibrator is switched out of circuit, but Q3, Q4 and the attenuator network are still operating. Dc calibration voltages of 50 mV, 500 mV and 5 volts are now available at the attenuator outputs. Due to Zener stabilization by Q3, these calibration voltages are virtually independent of supply line variations. The unit can be calibrated initially by turning S1 to the dc position and connecting a sensitive (20,000 ohms/volt or greater) meter between the 5-volt terminal and ground. Then adjust R7 for precisely 5 volts output and the unit is ready to use as a scope calibrator. R-E

Fig. 32—OSCILLOSCOPE CALIBRATOR provides dc and peak-to-peak outputs.

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