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Amperex
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Electronics and The Worlds Beyond

As long as there's a mountain to climb, a cave to inspect, a territory to map, a sea hole to examine, or a planet to explore—curious and intrepid men will expend their time, their energy, even their lives. The February tragedy at Cape Kennedy has a counterpart in every other field of human curiosity. Undersea, the Thresher tragedy; on Mt. Eiger, John Harlin; at the South Pole, Capt. Robert Scott's party (in 1912); in Africa, David Livingstone. Major accomplishments stand in tribute to those who martyr themselves to knowledge and understanding.

This issue of Radio-Electronics pays special attention to two modern frontiers: space and undersea. All our present technology notwithstanding, to live in either environment requires thinking radically changed from accustomed viewpoints. How much further we'll adjust before we truly unlock the riddles of these other worlds is barely conceivable.

Take the deep ocean. As the science of aquanautics develops, we plumb deeper and deeper into that murky world. Electronic devices we use right now can be adapted to make deep-sea exploration safer and more sure.

Example: A "sea-floor buggy" could crawl around controlled by radio, radar, sonar, hydronics, or some new electronic means from surface ships or shallow-sea substations. The vehicle could carry a TV system as "eyes." Maybe a new hydrosonic or other TV system would have to be developed for depths too dark even for lasers. A video tape recorder could bring back a viewable record.

To record in detail what's down there, laser-holographic equipment could take three-dimensional pictures. Electronic spectrometers, microscopes, and chemistry analyzers could be standard equipment on the sea-floor buggy, with data either recorded in the buggy's own computer storage system or sent back to the tender station by wireless telemetry. If geologic samples were required, remote-controlled laser tools could drill, cut, or chop them loose.

When we send a manned vehicle into the deepest trenches, the communication techniques used for telemetry could keep the aquanaut in touch with his tender station or ship, and could keep them informed of his well-being. Nuclear engines would supply power to drive the buggy and the life-sustaining equipment it has aboard. A large buggy, equipped to support several aquanauts, might be an outpost for smaller-buggy excursions into the so-far tractless depths.

You can surmise how important electronics will be in the watery life below the 3-mile depth.

In space, we have already made strides beyond those we have made under the sea. We have Surveyor sitting on the moon gathering data for the first Apollo landing. The success of Mariner 4, spectacular though it was, only paves the way for other Mariner and Voyager flights to give us more data about our neighbor planets, Venus and Mars.

After moon landings, the obvious followup is a manned flight to one or the other of these nearby planets. Some say by the late 1970's; sooner seems more likely. Despite electronics, the human factor remains vital. No matter how carefully we plan or how fully informed we are of conditions on either planet, the final split-second decision to land must rest in the mind and training of the astronaut commanding the flight.

You can't help thinking: Maybe we could eliminate the chance of life loss by not sending humans. Perhaps TV cameras, remote controls, robots, or other common automated devices could preclude the risk of life on these maiden voyages.

Such thoughts seem futile, at least for now. Consider the communications problem during the last minutes before a landing on Venus or Mars. A radio signal takes an average of 8 or 9 minutes to reach Earth from Venus and 12 to 13 minutes from Mars, one way. If sensors on the spacecraft flashed a warning of imminent danger, even at several hundred miles from touchdown, you can see that last-minute control from Earth would be utterly impossible from this distance.

Either foolproof on-board cybernetics is needed, or the thinking ability of a human. Such high-order automation demands prior knowledge, so it seems for now that a human is necessary on such critical missions. But that human must be supported by some of the most exotic instrumentation ever conceived.

Under the sea or far out in space, electronics builds a stronger role for itself month by month. In both, electronics is only the means used by man to sample, monitor, analyze, adapt to, and control environment. In all our journeys into other realms, the pioneering, fast-thinking and quick-reacting explorer will be the frontiersman—the first line of discovery between our world and the worlds beyond.
Radio-Electronics

May 1967 VOL. XXXVIII No. 5
Over 55 Years of Electronics Publishing

EDITORIAL
2 Electronics and The Worlds Beyond..................Forest H. Belt

SPACE & UNDERSEA ELECTRONICS
37 Build Hydronic-Radiation Transmitter ................Jack Althouse
Is it radio or a new type of communication? Who knows?
39 World's Toughest R/C Job ..................................Allen B. Smith
Remote control by radio at a distance of 240,000 miles
49 Comsat: Communication in the Space Age ..............Ray D.Thrower
Now TV and telephone service around the world by satellite

TELEVISION
22 In the Shop . With Jack ..................Jack Darr
Service Clinic
42 Something New in Color Generators ..................Larry Allen
Makes convergence fast and easy
45 Why Servicers Like Servicing ..........................Dick Glass
Professional technicians explain their reasons for being in the business
60 Troubleshooting Chroma with VTVM ..................Carl H. Babcocke
The lovely voltmeter can tell you a lot, if you know how to use it

GENERAL ELECTRONICS
32 Reaction-Time Testing of Race Drivers ..................Don Davis
Electronics qualifies high-speed wheel jockeys
34 Thinking Computers? Think Small ..................Melvin Whitmer
More than 35,000 of these typewriter-sized problem-solving machines are now in use
53 Selected Circuits from an Experimenter's File ..........Robert F. Scott
Ideas you may never have seen before
57 Imaginary Numbers are a Cinch ..................Norman H. Crowhurst
Operator I—useful math tool, not a telephone girl!
76 Equipment Report: Heathkit GR-180 Color TV Receiver
93 R-E Puzzler
99 R-E Puzzler Answer ..........................Edmund A. Braun

AUDIO & TEST INSTRUMENTS
47 Remote Control for PA Systems ..................Jack Darr
How to turn the knobs from afar
48 Convert Heath WA-P2 For Transistor Power Amp .......Hank Olson
Lower the output impedance and add a power supply
55 An Engineer Talks About Transistors in Audio ..........Peter E. Sutheim
Solid-state hi-fi tuners can excel their vacuum-tube counterparts
71 A Mechanical Look at Tape Recorder Servicing ..........John Mohan
Pulleys, belts and wheels need service too!
Here's how

THE DEPARTMENTS
16 Correspondence
98 Noteworthy Circuits
99 New Books
99 New Literature
91 New Products
96 New Semiconductors, Microcircuits & Tubes
4 News Briefs

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POSTMASTERS: Notices of undelivered copies (form 3579) to Boulder, Colo. 80302.

COVER FEATURE

p 34—The desk-top computer is the newest arrival in the field of electronic calculating machines. It's compact and can be programmed to perform a series of simultaneous computations.

p 32—Steering a fast racer demands instant response to changing road and traffic conditions. Here's how the professionals' reflexes are measured electronically.

p 60—CHROMA VTVM
p 71—TAPE SERVICE

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BIG BROTHER Listens

The FCC has authorized police agencies to use low-power transmitters on undisclosed frequencies for surveillance, "stake-outs," and raids.

Because normal police frequencies are known and may be monitored by anyone with a suitable receiver, the Commission reasoned that police were hindered in their attempts to capture criminal suspects. For instance, where citizens have been attacked and robbed in a city park, normal police procedure would be to stake out plainclothesmen in the area, equipping them with walkie-talkies. But if they had to use assigned frequencies, criminals might detect their presence and make the attempt useless.

The FCC rule change (Sec. 89.309) provides for the use of any frequency between 40 and 952 MHz which is available in the Police Radio Service, with a maximum of 2 watts' antenna input power. The Commission also emphasized that the rules "...in no sense authorize 'wiretapping' or 'bugging' activities by the police."

R/C Computer

The math instructor enters the classroom at Harvard University in Cambridge, Mass. He assigns students a problem involving the use of a computer. Each works out his idea of how to program and command the problem to the computer. Then the instructor punches the information into the machine. Soon the answer appears on a TV screen. Each student has his chance, in turn, to successfully work the problem.

Simple? Yes—but there's one catch. The computer is over 3,000 miles away—at Santa Barbara, Calif. Telephone and microwave circuits carry coded pulses from the classroom to the computer, and then the answer to the problem is returned from California.

The system could eventually mean that many advanced-level classes could use a single, centrally-located computer to aid them throughout a single day.

TV Service Pricing

In three surveys made last fall, TSA (Television Service Association) of Michigan determined the average service charges their member shops were using.

Average fee for a home call was $5.92 for b-w and 8.05 for color. For servicing a set in the customer's home for half an hour, the average charge was $3.03 (b-w) or 4.56 (color). Thus the total price of a completed service call averaged $8.95 (b-w) or $12.61 (color).

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May 1967

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Most audio engineers are familiar with the basic mechanics of transducer testing using tone burst signal sources. In its most often used form, tone burst testing is used to compare the relative ability of loudspeakers or other transducers to respond to transient audio phenomena.

Generally, however, tone burst testing has been ignored in favor of more traditional testing techniques, such as steady-state sine wave testing, sweep frequency testing, etc., as a means of determining design parameters.

Recently, Electro-Voice instituted a program of design testing using tone burst signals, in association with some conventional techniques, in an effort to develop a correlation between deviations from optimum transient response as displayed in oscilloscope tracings of tone bursts, and data obtained by other techniques.

It was proven that there was indeed a proveable relationship between data displayed and faults determined by more conventional means. For instance, specific peaks and dips in response, shown in steady-state measurements, often were related to poor transient response as shown in tone burst testing. By varying each of the possible contributing causes while observing the oscilloscope tracings, it could be determined which changes improved both frequency response and transient characteristics.

It was also noted that subjective reaction to speaker systems could often be anticipated by careful examination of exhaustive tone burst data. If similar units were compared, trained listeners most often preferred the unit with better transient response as shown in tone burst testing.

Using tone bursts, design parameters such as cone shape and composition, speaker optimum damping, enclosure construction, etc., can be tested with greater precision, and changes in design can be made with greater effectiveness. While no consumer-oriented specification has yet been developed to express the ability of a specific product to respond to such a testing program, it should be noted that several testing organizations use tone burst data in constructing subjective responses to loudspeaker characteristics.

The current testing program at Electro-Voice differs not in kind, but in degree, from previous efforts, using this effective new tool to determine more closely the optimum design parameters of transducers for home and industry.

For technical data on any E-V product, write:

ELECTRO-VOICE, INC. Dept. 573E
613 Cecil St., Buchanan, Michigan 49107

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land despite weather conditions. "Even though visibility was not poor," the Pan Am pilot announced to his 98 passengers afterward, "the landing would have been exactly the same with zero-zero visibility."

The system, developed by Boeing and Sperry-Phoenix, includes an autopilot, a flight-director system, a flare computer, an auto-throttle, and a radio altimeter. Complete redundancy is used: every unit has its twin to provide backup in case of equipment failure. Furthermore, the system continuously monitors itself; if it fails during operation, it disconnects itself from the aircraft controls.

Using ALS, the pilot does not relinquish his command, for he can override the computer at any time.

Pan American has two Boeing 727's equipped for automatic landings, and plans to outfit more planes soon.

LATEST ON LASERS
How do you take motion pictures of a rocket sled going 4,000 mph? It's simple, now—you use a laser tracker.

Developed by Sylvania for an Air Force testing program, the laser system is used like radar, bouncing a beam off the moving object. The return beam generates an error signal which controls the mirror angle (see diagram) to keep the target "on camera."

The sleds, which move at twice the speed of a rifle bullet, are impossible to photograph manually. Once the laser system is aimed at the vehicle, however, it tracks with an accuracy of 1/4 feet at 10 miles.

Symmetrical radiation patterns—which make it easier to couple into optical devices—have been obtained for the first time from a laser. The Bell Telephone Labs development is a Gallium Arsenide laser using a p+ layer only 2 microns thick, and a stripe contact on narrow strip of metal, which touches the semiconductor exciting area.

The radiation pattern produced by the device is said to facilitate laser-beam transmission over long distances. A more consumer-oriented use of lasers is a new system which is neither television nor movie. General Telephone & Electronics has built an experimental laser display system which uses phototastic cells for pickup of a live scene. This image is transmitted electronically to a light-beam "steering" device which projects three focused laser beams on a conventional movie screen. Since each beam is a different primary color—red, green and blue—the combination on the screen produces full-color images.

RE: TRANSISTOR TEST SET
Several readers have questioned the transistor socket markings on the panel of the transistor test set on page 57 of the March issue. They assumed that the socket is mounted so the base (B) pin terminal is on the left. The closeup shows that this is not the case.
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impedance mismatch problems?

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NOT ALL BAD

Dear Editor:

Recent correspondence in your magazine berates manufacturers for their negligence in answering correspondence.

I've written to very few, but my results have been most favorable. I was constructing a single-sideband transmitter, and needed a 1,500-kHz discriminator detector transformer which wasn't available locally. A letter to J. W. Miller Co. of Los Angeles not only brought the transformer but also gave me the capacitance change necessary to retune the transformer to the 160-meter amateur band—all for the sale of one transformer.

A similar response was received from their chief competitor regarding a power-type transformer needed.

In both cases, the part number and the proposed application for the part were described to me in detail. I suppose it's from such letters as mine that manufacturers find new uses for parts they advertise.

William Waite
Wellington, Ohio

ENG-A-LAND SWING LIKE...

Dear Editor:

Rereading your October 1966 editorial about a shortage of service technicians, I was again struck by how similar the situation is here. Firms who seek to economize on wages complain of the poor standard of TV mechanics, because anyone worth his salt soon moves on. The poor payers cannot hire an incompetent workman as they find it hard to hire anyone else: even an incompetent is better than no one.

A good bench man backed by comprehensive equipment may seem dear, but it's cheaper than two so-so's without instruments. A shop is only as good as the man wielding the soldering iron.

Here in England we seem to be plagued with incompetent outside men. Not that this worries me unduly. I am an outside man myself, working for a top-paying company with top fringe benefits. I take pride in repairing sets in the house and regard it as a personal defeat to bring the set into the shop. For myself and those like me, this mass of tube pullers makes our limited talents stand out. But they do no good to the industry.

I am in full agreement with your many articles that say: Be efficient, get value for work, give value for money.

P. M. Leyden
Nottingham, England

AND HERE, TOO

Dear Editor:

Regarding the current discussion of the scarcity of good trained technicians, the public is too difficult to please. I gave up the five shops I had 20 years back after I realized that radio and TV repairs were sought only as a necessary evil. Almost everyone would far rather buy a new set than pay for repairs and service.

And their character assassinations are frequently too much for the average stomach of the poor tech or shop owner. After all, who wants to be called a gypsy, crook, etc., for the "mounificent" earnings of $2 to $4 per hour.

It is still hard, even in 1967, to get across to the average set owner that repairs to a circuit cannot guarantee against "failure" in another circuit. Golly, we only repair 'em, not manufacture 'em. Even the service people who make calls for manufacturers get the same abuse.

A. H. Fish
Alexandria, Va.

EASIER SWITCH

Dear Editor:

As always, we enjoy receiving and reading Radio-Electronics. The March 1967 issue was no exception. We were especially interested in the construction article on the "Walkie-Talkie Power Booster" (page 60).

In the text, the author asks the readers to modify the switch. Open-leaf-type switching blades require proper adjustment to give proper "make" and "break" switching. To modify this switch would require skilled hands and a knowledge of how to adjust it.

If your readers have difficulty modifying this switch, they can purchase our FF switch, part No. 1006, This will give them a switch with the two break-make (dpdt). Their wiring, therefore, would be the C circuit and B circuit combination from the second C circuit on the stack.

C. J. Schultz
SwitchCraft
Chicago, Ill.

END
Capacitor stability at bargain prices

Any capacitor changes its microfarad value when temperature varies. And some capacitors change more than others. In some circuits, capacitance drift with temperature can cause real problems.

Look at circuits where you have fractional microfarad values of paper, film, ceramic or mica capacitors. During warm-up from room temperature to 65°C ambient, a capacitor with a temperature coefficient of, for example, 500 parts per million per degree C will increase capacitance value by 2%. This change is enough to cause troublesome drift in tuned circuits, where inductance also increases with temperature. It can knock the accuracy of a timing circuit off, or mess up the performance of a differentiator network. For these applications, we have a new kind of capacitor that beats anything we've seen in the stability race. It's the new Mallory Polystyrene Capacitor. They're made of stretched polystyrene film and high purity aluminum foil. The assembly is fused into one piece, with the polystyrene forming a solid case of clear plastic that you can look through and see the foil. Their temperature coefficient is less than 150 parts per million per degree C, which is about half that of polyester film capacitors. And the coefficient is negative; capacitance goes down when temperature goes up, compensating for the upward drift of inductance elements in tuned circuits.

Want more? Mallory Polystyrene Capacitors have the lowest dielectric loss... only a small fraction of that of other film capacitors. Their insulation resistance is way above that of mica, film or paper capacitors. And the best part of the whole deal is that they're really low in price!

There's something new from Mallory, too, in stable electrolytic capacitors. It's the molded-case MTA, which has temperature stability that beats most metal case types. It has shown up so well on life test that manufacturers are using it in instruments and computers. And while it's priced down with cardboard-case tubulars, it beats them every way on quality.

You can get these stable Mallory capacitors, and everything else you need for service or experimenting, from your nearby Mallory Distributor. Ask him for a copy of our 1967 General Catalog, or write to Mallory Distributor Products Company, a division of P. R. Mallory & Co. Inc., Indianapolis, Indiana 46206.

DON'T FORGET TO ASK 'EM — "What else needs fixing?"

MAY 1967

Circle 16 on reader's service card
How to get into
One of the hottest money-making fields in electronics today—
servicing two-way radios!

HE'S FLYING HIGH. Before he got his CIE training and FCC License, Ed Dulaney's only professional skill was as a commercial pilot engaged in crop dusting. Today he has his own two-way radio company, with seven full-time employees. "I am much better off financially, and really enjoy my work," he says. Read here how you can break into this profitable field.

More than 5 million two-way transmitters have skyrocketed the demand for service men and field, system, and R&D engineers. Topnotch licensed experts can earn $12,000 a year or more. You can be your own boss, build your own company. And you don't need a college education to break in.

HOW WOULD YOU LIKE to start collecting your share of the big money being made in electronics today? To start earning $5 to $7 an hour... $200 to $300 a week... $10,000 to $15,000 a year?

Your best bet today, especially if you don't have a college education, is probably in the field of two-way radio.

Two-way radio is booming. Today there are more than five million two-way transmitters for police cars, fire department vehicles, taxis, trucks, boats, planes, etc. and Citizen's Band uses—and the number is still growing at the rate of 80,000 new transmitters per month.

This wildfire boom presents a solid gold opportunity for trained two-way radio service experts. Many of them are earning $5,000 to $10,000 a year more than the average radio-TV repair man.

WHY YOU'LL EARN TOP PAY
One reason is that the United States Government doesn't permit anyone to service two-way radio systems unless he is licensed by the Federal Communications Commission. And there simply aren't enough licensed electronics experts to go around.

Circle 13 on reader's service card

RADIO-ELECTRONICS
Another reason two-way radio operators earn so much more than radio-TV service men is that they are needed more often and more desperately. A home radio or television set may need repair only once every year or two, and there's no real emergency when it does. But a two-way radio user must keep those transmitters operating at all times, and must have their frequency modulation and plate power input checked at regular intervals by licensed personnel to meet FCC requirements.

This means that the available licensed experts can "write their own ticket" when it comes to earnings. Some work by the hour and usually charge at least $5.00 per hour, $7.50 on evenings and Sundays, plus travel expenses. A more common arrangement is to be paid a monthly retainer fee by each customer. Although rates vary widely, this fixed charge might be $20 a month for the base station and $7.50 for each mobile station. A survey showed that one man can easily maintain at least 100 stations, averaging 15 base stations and 85 mobiles. This would add up to at least $12,000 a year.

Be Your Own Boss

There are other advantages too. You can become your own boss—work entirely by yourself or gradually build your own fully staffed service company. Instead of being chained to a work bench, machine, or desk all day, you'll move around, see lots of action, rub shoulders with important police and fire officials and business executives who depend on you two-way radio for their daily operations. You may even be tapped for a big job working for one of the two-way radio manufacturers in field service, factory quality control, or laboratory research and development.

How To Get Started

How do you break into the ranks of the big-money earners in two-way radio? This is probably the best way:

1. Without quitting your present job, learn enough about electronics fundamentals to pass the Government FCC Exam and get your Commercial FCC License.

2. Then get a job in a two-way radio service shop and "learn the ropes" of the business.

3. As soon as you've earned a reputation as an expert, there are several ways you can go. You can move out and start signing up and servicing your own customers. You might become a franchised service representative of a big manufacturer and then start getting into two-way radio sales, where one sales contract might net you $5,000. Or you may even be invited to move up into a high-prestige salaried job with one of the major manufacturers either in the plant or out in the field.

The first step—mastering the fundamentals of Electronics in your spare time—and getting your FCC License—can be easier than you think.

Cleveland Institute of Electronics has been successfully teaching electronics by mail for over thirty years. Right at home, in your spare time, you learn electronics step by step. Our Auto-Programmed lessons and coaching by expert instructors make everything clear and easy, even for men who thought they were "poor learners." You'll learn not only the fundamentals that apply to all electronics design and servicing, but also the specific procedures for installing, troubleshooting, and maintaining two-way mobile equipment.

Get Your FCC License... or Your Money Back!

By the time you've finished your CIE course, you'll be able to pass the FCC License Exam with ease. Better than nine out of ten CIE-trained men pass the FCC Exam the first time they try, even though two out of three non-CIE men fail. This startling record of achievement makes possible the famous CIE warranty: you'll pass the FCC Exam upon completion of your course or your tuition will be refunded in full.

Ed Dulaney is an outstanding example of the success possible through CIE training. Before he studied with CIE, Dulaney was a crop duster. Today he owns the Dulaney Communications Service, with seven people working for him repairing and manufacturing two-way equipment. Says Dulaney: "I found the CIE training thorough and the lessons easy to understand. No question about it—the CIE course was the best investment I ever made."

Find out more about how to get ahead in all fields of electronics, including two-way radio. Mail the bound-in postpaid reply card for two FREE books, "How To Get A Commercial FCC License" and "How To Succeed In Electronics." If card has been removed, just send us your name and address on a postcard.

ENROLL UNDER NEW G.I. BILL
All CIE courses are available under the new G.I. Bill. If you served on active duty since January 31, 1955, or are in service now, check box on reply card for G.I. Bill information.
In the Shop... With Jack

By JACK DARR

FOR A LONG TIME, ONE OF THE MOST frequent questions in my mailbag was "How can I change my old Gizmo color TV set to use the new sulphide tubes?" Then, after a while, it was "How can I change my old obsolete sulphide-tube color set to one of the new rare-earth type tubes?"

About all I could give was sympathy. Conversion of a h-w TV set from 70° to 90° tubes was bad enough, but conversion from one kind of color tube to another was worse. None of the set manufacturers would even recommend it. Since it was such an expensive deal, with such a big, fat chance for errors, I didn't dare recommend any rash experimenting!

However, things have changed (as usual). Now, the "converters" can upgrade a lot of TV sets. To be honest about it, the process won't make a brand-new set out of an old CTC4, for instance, but it will produce slightly better pictures if the chassis is in good shape. RCA has brought out a little leaflet giving instructions; they also have conversion kits for replacing the old metal tubes with new glass ones.

The leaflet covers sets like the 21CT55, 21CT660 (up through the 700 and 800 series), CTC5 chassis up through the -N and -W, as well as the -F and -H series; roughly: through 1957. The replacement tube, a glass-bulb 21CYP22, is electrically identical to the original 21AXP22. This means there won't be any great bother, outside of setting up the new tube with the regular screen and background controls.

The major part of this conversion is "mechanical"—changing the original suspension of the metal-cone tube to hold the glass bulb. RCA's 12B101 conversion kit has all the necessary hardware to do the job. Incidentally, since a great many color TV sets built during that period used RCA chassies, this conversion kit will apply to many other brands as well as RCA.

The 21CYP22 tubes have two HV connectors instead of one. One is a "blank" used to hold one end of a 56K current-limiting resistor. The regular HV lead goes to the blank terminal, with one end of the tension spring (see Fig. 1). This is the lower terminal, with the set in normal position.

The heavy insulator around the metal cone of the old CRT is discarded; a new plastic cover is provided for the bell of the tube. You'll notice that instructions emphasize the use of ground straps, springs, etc., on several connections. Be sure that these are installed, and in the right places, so that the conductive coating on the outside of the bulb will be well grounded. If your tube had any doubts, add a couple more grounds!

In later chassis, which were designed with the glass 21CYP22's from the CTC7 on up, you can convert to the rare-earth tubes; late-production 21CYP22's were sulphide-phosphor types. Mounting changes will be minor, since these chassis are made to take the glass-bulb tubes. The 21FBP22 is the recommended replacement. This tube doesn't have the current-limiting resistor connection, as in the -CYP22, so this is taken off and discarded.

Some electrical changes will have to be made in the chassis, if the full benefit of the more efficient rare-earth phosphor is to be realized. The red is much better. (Remember, a long time ago, when a picture tube went out, it was always the red gun? Or, it seemed that way.)

In this conversion, you may have to make some alterations in the shielding, hardware and so on, to avoid any chance of arc-overs to grounded objects. Magnetic shielding can be notched out to keep it away from the HV connector. Keep any grounded object at least 6

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Fig 1—Current-limiting HV resistor is used with the 21CYP22 color CRT.

This column is for your service problems—TV, radio, audio or general and industrial electronics. We answer all questions individually by mail, free of charge, and the more interesting ones will be printed here.

If you're really stuck, write us. We'll do our best to help you. Don't forget to enclose a stamped, self-addressed envelope. Write: Service Editor, Radio-Electronics, 154 West 14th Street, New York 10011.
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what else would you want from a tube tester?

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We know that swift and sure tube testing makes servicing more profitable. So we created the Dyna-Jet 606 Tube Tester, the professional portable loaded with the most-wanted features. Multiple 13-socket design means only 4 test settings, yet it tests the latest miniature and color receiving tubes as well as older types.

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Few test instruments pack the profit-per-square-inch as does the Dyna-Jet 606. It's another product of B & K electronic innovation . . . of B & K's policy to provide maximum value and maximum quality. And the B & K Professional Servicing Equipment emblem assures you . . . and your customers . . . that you use the finest equipment available.

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Circle 17 on reader's service card

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For precise speed-regulation, Sony employs the first precision servo-controlled motor ever used in a turntable. No rumble, no noise, precise speed, beautiful music from the Sony TTS-3000. $149.50 (base extra).

For the finest playback system, add the Sony VC-8E moving coil cartridge, $65 and the PUA-237 professional arm, $85. Prices, suggested list. At your Sony high fidelity dealer. Sony Corporation of America, Dept. H, 47-47 Van Dam Street, Long Island City, N.Y. 11101
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*Optional Distributor resale price. All prices subject to change without notice. Price may be slightly higher in Alaska, Hawaii, and the West.

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The Most Trusted Name in Electronics

MAY 1967

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In the Shop... With Jack

Fig. 4—This modification will prevent damage if new CRT should be over.

Capacitor. The 0.15 μF capacitor (.047 μF in CTC7's) is disconnected from the R—Y amplifier plate, and hooked to the junction of the two new resistors.

In all these conversions, a 0.1-μF 600-volt bypass capacitor should be added from pin 7 of the 6BK4 high-voltage regulator to ground. The picture-tube heater is connected to the same winding to protect it against damage from possible flashovers, inside the tube neck. Add a 1,000-ohm resistor from the 6BK4 cathode to B+; same purpose. These changes are shown in Fig. 4.

After the mechanical part of the job is done, and the set fired up, check the color temperatures, brightness tracking, and so on. If you have trouble getting highlights to track, take the red cathode lead off, at the chassis tie point, and exchange it with either the blue or green cathode, whichever gives the best results.

There is one thing you positively cannot do! It's bad enough in b-w sets, but in color it's a sheer impossibility. This is to swap a 70° round tube for a 90° rectangular tube! A 25GP22A is not a replacement for a 21AXF22, that's all there is to it!

Why? Well, in b-w sets, our problem in such swaps is to get enough sweep power. It takes 25% more horizontal sweep power and 50% more vertical, to change from a 70° to a 90° tube: 110° tubes I'd rather not even think of! In a color set, we'd have all of these problems, plus the convergence!

Besides, a rectangular b-w tube can be corrected for pincushioning with a pair of simple PM magnets. In a rectangular color tube, the same job requires a very complex built-in pincushion corrector, involving the yoke, vertical output transformer, and so on! We would probably wind up putting on a new yoke, flyback, and a few other goodies.

At any rate, there isn't much difference in cost between the newer tubes and the originals, and the conversion kits aren't expensive. So, if you've got to replace the color picture tube anyhow, you might as well go first class.

---

**Centralab's full line of color TV capacitors includes:**

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Circle 22 on reader's service card

MAY 1967
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### Reaction-Time Testing of Race Drivers

**High-speed competition driving requires fast reflexes—abilities that can be measured only by electronics**

By DON DAVIS

Automobile racing drivers, as viewed by an enthusiast who also is an audio engineer, have interesting characteristics that demand analysis. And, surprisingly, such an analysis reveals patterns similar to those commonly encountered in evaluating audio and acoustical phenomena.

As an example, consider the two weekends in May during which drivers and cars compete to enter the world-famous Indianapolis 500-mile race. Only 33 starting positions are available, yet as many as 70 cars and drivers arrive at the track to prove themselves fast enough and talented enough to race on Memorial Day. Each aspirant runs 4 laps (21 1/2 miles a lap) at the fastest speed he can maintain for the 10-mile test. The 33 fastest qualifiers are in: the others are out. The fastest and the slowest cars in the lineup are separated by fewer than 3 seconds a lap.

If the speed of 33 racing cars varies less than 3 seconds a lap from the slowest to the fastest, it would appear that the cars and drivers are almost equal. Fig. 1-a, however, shows this assumption to be incorrect. Plotted on this first chart are the 1964 qualifying speeds in 1-mph increments versus the number of drivers who qualified at each speed. When rough curve A is smoothed to form curve B, it assumes the classical Gaussian or Laplacian distribution pattern. Fig. 1-b (1965) shows clearly what happened when the very best European drivers were in competition with the best US drivers. Those who were below norm (and the norm at Indianapolis is extraordinary compared to the norm of the man on the street) have been eliminated.

Distribution of the sharpness of the five physical senses in any group of people would follow a similar pattern. If, for example, you were to test the visual perception of 1,000 people selected at random, your measurements would produce a curve like that shown in Fig. 1-a. The curve would plot visual acuity (sharpness) against the percentage of the total sample taken. Similar curves would apply for tests of hearing, balance, touch, smell, and reaction time. One can therefore conclude that what happens out on the race track depends greatly on a driver’s physical abilities. Some drivers are going to have better basic abilities than others.

The literature (sparse), observations (emotionally tainted) and speculations (rampant) all suggest that supernormal eyesight and acute reaction time are the two most important abilities a race driver requires as natural endowments. Because eyesight can be measured and evaluated very accurately, Dr. Thomas Hanna, medical director of the Indianapolis Motor Speedway, has been able to confirm that the eyesight of competition drivers tends to be exceptional, particularly in regard to equal abilities with both eyes.

There is, however, little data available on how to measure reaction time. How this all-important physical characteristic affects the race driver can best be illustrated by a brief look at the speeds racing cars travel. On the main straightaway at the Indy speedway, they reach speeds of 200 mph, or 293 ft/sec. (mph x 1.466 = ft/sec.). At this speed, a misjudgment as small as 0.2 sec would carry the car almost 60 extra feet (a disastrous mistake on the walled corners at Indianapolis). It’s ap-

**Fig. 1—Qualifying speeds at the 500-mile race. a—Nearly all drivers were from US in 1964. b—European drivers raised the speed norm in 1965, eliminating some low scorers.**

reaction time of a top professional driver like 1965 Indianapolis champion Mario Andretti is far faster than that of the noncompetition driver. Slow reflexes in a race can mean disaster or a poor showing.
The oscilloscope provides a readily available, accurate and standard instrument easily adapted to reaction-time measurements. Fig. 2 shows a schematic for a reaction-time tester using an oscilloscope indicator and a camera as the recording instrument.

A simple procedure for using the circuit shown is: The panel-mounted TRIGGER switch (Fig. 3) is set so the battery is connected to the green indicator light which won't light until the accelerator pedal on the subjects console is pushed down. The driver being tested faces a panel on which there are four lights—green, yellow, red and white—plus an AUDIO pushbutton. Just below the driver's panel on the floor are the brake and accelerator pedals. The driver sits on a chair in front of the panel and grips two handles at the side of the chair (to keep his hands away from the panel).

The operator's panel also has four lights wired in parallel with the subject's console lights. In addition, the operator has a TRIGGER switch. When this switch is set at READY, the green light will glow as soon as the driver presses the accelerator. A mode-selector switch and an audio switch also are mounted on the operator's panel.

With the driver in position and the accelerator depressed, the operator may choose which driver reactions he will test. He can, by choosing the proper position on the selector switch, (1) measure how long it takes for the driver's foot to come off of the accelerator; (2) measure how long it takes the driver to lift his foot from the accelerator and hit the brake, and (3) measure the time required by the driver to let go of the seat handles and depress the audio pushbutton on his panel after hearing an audible signal.

A simple “light on” push-the-button test was not used, because in actual racing situations, not only is a reaction required, but a sound decision must precede it. Therefore, the testing panel was designed to force a decision and then the reaction.

To perform a test, the operator turns the SELECTOR switch to ACCELERATOR. When he throws the TRIGGER switch from READY to TEST, the green light goes out, the yellow light comes on, and a small dc voltage is fed to the oscilloscope variable trigger. This signal starts a sweep across the scope. Sweep time on the scope can be varied, but 0.1 sec/cm allows a full second of time to be displayed in large 0.1-sec increments. The driver being tested sees the yellow light come on and, if he makes the correct decision, lifts his foot from the accelerator. As soon as his foot lifts, the white RESPONSE light comes on and a small small dc voltage is fed to the vertical input of the oscilloscope. This causes a “blip” to appear on the scope tube. The distance from the beginning of the sweep at the left-hand marker on the scope face to the “blip” is elapsed reaction time.

If the driver makes a wrong decision when the yellow light flashes and hits the brake instead, no voltage can get to the white light or to the oscilloscope. The selector switch not only selects the test but blocks any response other than the correct one.

Interesting data will be taken using this equipment this May at the Indianapolis track. Once enough data are gathered, meaningful criteria can be established for use in the demanding profession of driving racing cars.

![Fig. 2-The reaction-time tester displays a scope trace which denotes the speed and dexterity of the driver being tested. Lights tell driver what he is expected to do.](image)

![Fig. 3-Pilot lamps on the operator's panel monitor operational modes. The driver's panel contains similar pilots wired in parallel to the operator's, along with a speaker. On the floor are accelerator and brake pedals which operate switches and trigger the scope.](image)
Thinking Computers? Think Small

By MELVIN WHITMER

As recently as 15 years ago, according to the American Federation of Information Processing Societies (AFIPS), there were fewer than 25 computers at work in all of the US. That number has grown today to well over 35,000, and the AFIPS predicts that by 1975 there will be more than 85,000—representing an annual investment of $30 billion.

Understandably, the greatest increase—because of their lower initial cost—will come in the area of compact and desk-type computers. Though physically small, many of this new generation of time-savers are capable of a wider range of computations than some of the huge multi-rack installations of a decade or so ago.

Strictly speaking, the subject of this article is a hybrid. An electromechanical calculator, or adding machine, can perform only simple operations. A true computer has large-capacity memory banks and can be programed, or instructed, to “learn” a long and intricate series of computations. A desk-top computer can do more than an adding machine but less than a true computer. It is becoming a popular item with many businesses, and will be used so much in the future it will open many job opportunities. Electronics technicians will be needed to service these computers.

In all but a few isolated instances, factory-service opportunities include extensive training programs for technicians with a good technical background. Factory-trained technicians service and repair computer hardware directly for the factory or for local franchised distributors of the equipment.

Although service during warranty will be performed by the factory or by its authorized distributors, most manufacturers later will encourage in-plant electronics personnel to perform all routine computer maintenance. Service information and short training courses will be made available to the personnel of qualified plants. Many users obviously will have to add technicians to their existing staffs when complex computer equipment becomes a significant part of the company’s office equipment.

Manufacturers of desk-top computers agree, however, that independent industrial-service organizations will be unable to enter the field of desk-top computers. Few, indeed, would want to. Sales and service functions of this magnitude are best performed under factory direction. Service information, therefore, seldom will be available to independents.

Personnel requirements

It’s hard to generalize about what talents and training are required for factory-service work. Each manufacturer has his own employment standards. Very few require applicants to hold degrees for field and shop work—high school and an interest in electronics generally are minimum requirements. Technical-school graduates should have no trouble being accepted as trainees in any manufacturer’s sales and service department. Experience in office-machine servicing is most desirable, but an electronics technician with this type of background is rare indeed. The market is wide open, and most companies are expanding their sales and service staffs for the time when their electronic calculators become part of every corporation’s office-machine budget.

Integrated circuitry is largely responsible for stuffing a useful measure of large-computer capability into a small package. And some manufacturers use interchangeable printed-circuit cards to hold several IC’s at one time. These modules are stacked compactly in a holder or “bin.”

The capabilities of a typical desktop computer reach far beyond those of mechanical calculators or adding machines. Operations include adding, subtracting, multiplying, dividing, working with powers and roots, logarithms, factorials, and trigonometric functions. These and other capabilities of small computers will add efficiency to business offices and engineering departments. Full-scale data-handling systems won’t become obsolete, of course, but only those problems requiring very complex processing will be sent to the central computer.

Simple problems, or preliminary steps in a more complex program, will be

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Sage 1 (Dero Research & Development) has 20-digit register. IME-86 has 7 registers to store various parts of a problem.
accomplished with the desk-top computer. This elimination of trivial tasks will greatly increase the usefulness of the central computer. Since operating a small computer is similar to operating a mechanical calculator, it's a job easily assigned to non-technically trained employees.

An integrated circuit deposited and etched on a chip of silicon or other semiconductor material performs the same function as a conventional multicomponent stage. Resistors, diodes, capacitors and transistors that normally would be used to build a typical transistorized stage are replaced by small areas of deposited metallic oxides.

The circuit shown in Fig. 1-a is the equivalent of that constructed on a single integrated-circuit chip. In operation the circuit establishes two stable conditions: The first occurs when Q1 and Q2 are conducting; the second, when Q3 and Q4 are conducting. Input pulses are applied to S1 or S2. Clock pulses—which synchronize the advance of data through the computer—are applied at connector C. Outputs are obtained from R1, Q, S1, or Q and differ primarily in current-output characteristics. Outputs R1 and S1 vary about 1.5 volts, based upon the stable condition. Outputs Q and Q vary about 3.5 volts, at a higher impedance than that available at R1 and S1.

If a set pulse is applied to Re, then Q1 and Q2 are driven into conduction by resultant decrease in base current for Q3 and Q4. The first input pulse at Sc reverses the stable state, cutting off Q1 and Q2 and driving Q3 and Q4 into conduction. A clock pulse will then produce outputs (negative pulses) at Sc and Qc.

Microelectronic circuits formed in metal-oxide-semiconductor (MOS) chips often replace as many as 200 standard components; a complete 10-digit counter, for example, can be built into an MOS chip. This greatly reduces the size of the computer without sacrificing problem-solving ability.

Servicing MOS circuitry requires some understanding of the computer layout, a visual inspection for open or shorted printed-circuit paths, voltage tests and finally, replacement of the defective unit. Most computer manufacturers use some form of modular design to facilitate plug-in repairs. The defective plug-in board module may then be returned to the manufacturer's central-repair station for detailed analysis and for repair, when that is possible. Alternatively, the digits may appear on Nixie tubes or a CRT.

The block diagram of a typical small computer is shown in Fig. 2. Keyboard output is fed to the control unit. From the control unit, input data pass to the "learn" or storage sections. The mathematical unit may accept data from the storage register or from the keyboard input (through control unit). Steps, operators and solutions are fed to the printer control circuits, then imprinted on the tape.

A computer like this may "learn" 24 or 48 steps, depending upon the model selected, and there are eight storage registers. The learn feature is the next best thing in computer capability to a completely stored program. Several steps can be stored in a program memory when a problem is repetitive. Working with progressions (1, 7, 13, 19, etc.) and factorials (17 x 16 x 15 x—etc.) could be tedious if done manually. The program memory learns a few basic steps and the operator then activates an automatic sequence.

### Desk-top computers defined

All categories of computers are achieving a far greater component den-

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**Fig. 1**—The integrated circuit has made the small computer possible. The entire circuit at left is manufactured on the 1/4 x 3/4-in plug-in module shown above.
Thinking Computers? Think Small

Sity through microcircuit techniques. So, the distinction between desk-top and central computers has to be drawn not only on physical size but on capability as well. A desk-top computer is generally considered to be one that requires no formally prepared stored program entered in a special language. Most desk-top computers are operated like adding machines or mechanical calculators. Problem information is inserted through a digital keyboard or typewriter in much the same sequence as it would be written by hand. When a special program language is required, the computer is properly called a general-purpose digital computer, even though the keyboard or typewriter can be placed on a desk.

Physically, desk-top calculators are either completely self contained or have only a small auxiliary unit that can be mounted under the desk. When the electronic package is built into a desk or console, the computer's computational capacity is usually so large that some form of preprogramming is necessary. Typewriter, teletype, punched tape or card-reader inputs usually put a computer out of the desk-top class. The exception, of course, would be the addition of one of the above input units to a keyboard-input computer to extend its versatility.

Operation

Since a desk-top calculator seldom requires a stored program, steps to be performed are entered as the computer works the problem. Most small units have storage registers for holding partial solutions or constants. Thus, solving a problem involves step-by-step progressions initiated by the operator.

An example, the sequence of solving a parallel-resistance problem might go like this:

\[
\begin{align*}
R_1 \times R_2 &= R_s \\
R_1 + R_2 &= R_s
\end{align*}
\]

1. Enter value for \( R_1 \).
2. Add value for \( R_2 \).
4. Enter value for \( R_1 \).
5. Multiply by value for \( R_2 \).
6. Divide by sum stored in register.

The operator takes the place of the program that is normally stored in a big computer. The 10-digit keyboard is used by most manufacturers for data entry, and additional keys select the function to be performed.

There seems little doubt that desk-top computers will increase in number by the tens of thousands during the next decade. Indeed, some expert observers compare the effect of the "computer revolution" to that of the machine-based technology that sparked and nourished the industrial revolution of the last century. If that is true, and it seems to be, the predictions of the AFIPS may prove to be extremely conservative.

In any case, opportunities for technically trained electronics specialists are certain to grow along with the expanding computer technology. Take a good, hard look at a fascinating field. You well might find yourself a part of it.  

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With the Wang 360 you can find the sum of a set of digits and also the sum of their square roots at the same time.

Mathatronics Mathatron supplies a paper-tape answer showing the complete problem solution with all operational signs.

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Wyle Scientific calculator uses a cathode-ray-tube display.

This microelectronics calculator by Victor employs 30 MOS (metal-oxide-semiconductor) devices for circuit reliability.
"Scientists in Florida have discovered a new form of electromagnetic radiation which propagates under water as well as radio does in air."

"I must protest. Has a serious trade journal resorted to sensationalism? It appears that either a hoax is being perpetrated or that . . . ."

"A new mode of communications—via underwater electromagnetic radiation—is being explored by Hydronics Corp. Wallace Minto, inventor of the system, says signals have been transmitted over a distance as great as 30 miles by this method."

Is "hydronic radiation" a fact or a fraud? It has raised a storm of controversy as illustrated by the quotations above, all of which were written by responsible engineers or technical editors. The fact is that hydronic radiation does work. And the equipment is easy to build once a basic understanding of the new concept is attained.

But first, let's see how the system's inventor describes hydronic radiation. Wallace Minto of Sarasota, Fla., describes it as "a new vector field related to the electromagnetic and magnetohydrodynamic forces, characteristiclly propagated through a water medium and associated with electronic oscillations."

Translation: Hydronic radiation is the same as radio, except that it works through water instead of air.

Only the antennas distinguish the hydronic-radiation system from a conventional radio system. Receiving and transmitting antennas both have large plates, at each end of an insulating separator, to make contact with the water. A halfwave radio antenna has insulators on each end and radiates at right angles to the wire, as shown in Fig. 1-a. The hydronic-radiation antenna appears to radiate off the ends of its plates (Fig. 1-b).

Early hydronic-radiation experiments were made in the salt water of the Atlantic Ocean, and it's in the oceans that the most exciting possible applications for hydronic-radiation systems exist. Static-free communication between ships, dependable communication with submarines, and trans-Atlantic communication without cables have all been suggested by the concept's proponents.

But hydronic radiation also works through the fresh water found in lakes and rivers. As a matter of fact, the transmitter described here can be used in ordinary tap water to perform experiments in your bathtub.

When working in water, it's dangerous to use ac-powered equipment; the transmitter, therefore, is designed to operate from a 9-volt transistor radio battery. Power drain is low, and the battery will last for many hours. The transmitter schematic is shown in Fig. 2. Q1 is the rf oscillator, tuned by L1 to transmit in the standard broadcast band, allowing use of an ordinary transistor-type AM radio as a receiver. Q2 is an audio oscillator, operating at about 1 kHz, which modulates Q1 through transformer T. The tone-modulated signal stands out clearly among the regular stations and thus makes testing easier.

The transmitter is built in a 5 x 7 x 3-inch aluminum box, with all components mounted on a perforated board except for L1 and the terminal strip. These are mounted on one end of the box. Standard construction techniques are used, and parts placement and lead lengths are not critical.

L1 is modified by winding a 25-turn coupling coil at its lower end. The two terminals marked TRANSMIT are used as the on-off switch. When the terminals are connected by a shorting wire, the transmitter will operate.

To test the transmitter, leave the box open and place a transistor radio next to the circuit board. Tune across the band and listen for the modulated tone signal. It should appear between 550 and 800 kHz. If you can't find it, set the receiver dial at 550 kHz and tune L1 until the tone is received.

The transistor radio becomes our hydronic-radiation receiver with a simple modification. Open the case and wind 25 turns of No. 24 enameled wire around the transformer radiator.
Build Hydronic-Radiation Transmitter

around its ferrite-loop antenna. Leave about 6 inches of wire at each end of the winding and twist the ends together to hold the winding in place. Bring the wires out of the cave and snap the cover shut to hold them in place.

Two identical antennas are needed, one for the transmitter and one for the receiver. Fig. 3 illustrates the simple construction. End plates are 2 inches square, of 18-gauge brass cleaned to the bare metal to make good contact with the water. The 6-inch spacer can be of Bakelite, Lucite or other insulating material.

Two 6-foot lengths of plastic-insulated hookup wire are used for the feeder. One wire is connected to each plate by a solder lug passed over the end of the plate, then fed through holes in the spacer which allow the wires to be stretched taut. The rest of the wire is twisted together to make a balanced feed line.

Strip the enamel insulation from the ends of the antenna-feeder wires and from the two receiver leads, then solder them together. Tape the connections to prevent shorting.

Connect the second antenna to the transmitter. Place the antennas close together and check to see that the tone-modulated signal can still be heard at 550 kHz. Adjust I 1 if necessary.

Final equipment checks should be made with the antennas in water. A pool of water at least the size of a bathtub is needed, and the water should be a foot or more deep. Place the antennas facing each other about 3 feet apart. Return the receiver to find the tone. It will appear at about 700 kHz (the water loads the transmitter heavily and shifts its frequency).

To make sure that the hydronic-radiation signals are being received through the water, lift the receiving antenna out of the water. The signal should disappear or at least drop considerably in volume.

With the gear ready, we can perform a few experiments, to see how the system operates. A few questions may be answered, and a few more may be raised.

Hydronic-radiation experiments

One of the interesting characteristics of a hydronic-radiation communications system is the apparent directional pattern of its antennas.

Place the antennas underwater with the end plates parallel to each other. Rotate one antenna 90° horizontally or vertically so the edges of its end plates are perpendicular to the other antenna's end plates. The signal will fade as you rotate, disappearing completely at the 90° position. As you turn toward 180°, the signal will come back and become strong again as the end plates once again become parallel. This experiment appears to show that the antennas radiate off the ends of the plates.

Antenna engineers, however, say, "No. The signal does not radiate from the surface of the plates." Instead, they explain, the signal radiates from the wires that connect the plates. These wires actually form a dipole antenna; the plates, they say, are just "ground rods." Furthermore, the signals do not go straight through the water at all. They travel from the transmitting antenna upward to the surface of the water, along the surface, then back down to the receiving antenna.

This description of hydronic radiation suggests that the behavior of signal radiation is opposed to our general experience. It is true that a horizontal-wire antenna will radiate its signal upward. The up-over-down theory implies that when hydronic waves reach the surface of the water, they must bend at right angles to travel along the surface. Then, when they are above the receiving antenna, they must bend downward so the antenna can pick them up.

Long-range antennas

The maximum distance for effective communication by hydronic radiation apparently depends on the spacing between the antenna plates. The greater the spacing, the longer the range. Plate spacings of 1,000 feet have been used to communicate over several miles.

For our experiment, a plate spacing of 6 feet is convenient and will provide a range of 100 feet or more. The antenna will use a 6-foot 2 x 2 wood spacer and 1-foot-square brass plates. The plates don't have to be that big, but they must be heavy enough so the antenna will sink in the water (the wood spacer tends to float, of course). A good electrical connection is made to each plate, the connecting wires are brought directly to the center of the antenna and twisted to form the transmission line, which should be about 15 feet long. As before, two identical antennas are required.

Up-over-down experiment

The antennas should be placed about 50 feet apart in water at least 6 feet deep. With the antennas just below the surface and pointed at each other, the signal should be received loud and clear. Now, if both antennas are slowly lowered deeper into the water, the distance between them being kept the same, we observe an interesting result. The deeper the antennas go, the weaker the signal becomes. If we put the antennas deep enough, the signal disappears completely. Since the distance between the end plates hasn't changed, we would expect to the signal strength to remain the same.

If the antenna engineers are correct, however, the up-over-down path is 50 feet long when the antennas are 10 feet below the surface. Thus, if the up-over-down theory is true, we would expect the signal to become weaker as the antennas go deeper into the water. Since the signal is, in fact, weaker with the antennas deeply submerged, our experiment apparently shows that the antenna engineers are right. The signal probably does go up-over-down.

Future of hydronic radiation

One experiment seems to "prove" that hydronic radiation travels off the ends of the antenna plates. Another seems to "prove" the signal somehow goes up-over-down. Is there an experiment that could prove that neither explanation is correct? If so, engineers haven't yet discovered it. But there is no reason you can't experiment with your radiation transmitter to see what conclusions might be drawn. After all, the last word on the concept isn't in yet.

So far, experimenters are strongly divided on whether hydronic radiation actually is a different form of electromagnetic radiation that will prove useful in underwater communication systems. One camp holds that rf energy generated by the "hydronic" transmitter is radiated through water in much the same manner as it would be through the air, though with some differences. Obviously the circuits employed are identical in equipment used for both propagation media: air and water. The second group feels there is some basically different phenomenon at work, one that promises efficient underwater communication.

Only extensive experimentation under carefully controlled conditions will provide the complete answer, of course. But, you can explore a phenomenon that's in the news today, and do it with very little cash outlay. The "hydronic" transmitter does work: why it does isn't apparent, at the moment.
WORLD'S TOUGHEST R/C JOB

Guidance in Outer Space

By ALLEN B. SMITH

NASA's spectacular success in soft-landing Surveyor I on the moon, and in obtaining high-resolution photographs of the lunar surface from Lunar Orbiter, has generated a tremendous surge of interest in America's plans to explore outer space. Project Apollo, though struck by the tragic fire that took the lives of astronauts Grissom, White and Chafee, continues to captivate the interest of the world public. By the end of 1968 or early in 1969, three American astronauts will embark on an historic 8-day trip to the moon and return to earth. Two of the three-man Apollo crew will explore the moon's surface during a projected 35-hour encumbrance.

 Virtually every system in NASA's diversified collection of spacecraft depends heavily on electronic instrumentation. Of all the electronics, the guidance and navigation package in all missiles and spacecraft is particularly interesting.

In NASA's space-exploration program there is a bewildering array of spacecraft. They fall roughly into three major categories: missiles, unmanned probes, and manned spaceships and capsules. Each is designed for a specific purpose, and each has a unique physical configuration. Yet in the center of all these vehicles nestles a group of gyroscopes and accelerometers, a computer and other precision components. These comprise the nucleus of the remarkable inertial guidance and navigation unit.

Ballistic missiles like Redstone, Jupiter, Atlas and Titan, originally conceived as force-deterrent weapons, have been developed into highly dependable vehicles for thrusting American satellites and spacecraft into space. This family of rocket-engineered craft uses the most direct form of inertial guidance. The system's computer is preprogramed to control the vehicle during its flight without reference to external signals.

Unmanned, instrumented spacecraft are far more varied than missiles. Explorer, Pioneer, Nimbus, Tiros, Early Bird and other orbital craft employ an inertial-navigation system, but only to maintain or correct the attitude of the satellite relative to the earth. Lunar probes of the Ranger and Lunar Orbiter series are equipped with restartable rocket engines and may be required to make midcourse corrections. They

must, therefore, be able to receive commands from ground-based stations so that their computers can develop course-change commands to the attitude-control system and the main propulsion engine. Surveyor-series craft can do that also.

In simplest terms, an inertial-navigation system is a self-contained instrument package requiring no external reference signals or commands once it has been programed to guide a vehicle to its destination. It must have three basic elements: (1) a directional reference provided by gyroscopes, (2) a memory device—usually a compact computer, and (3) velocity-sensitive accelerometers to detect external forces that alter the vehicle's course in any direction. The computer integrates information obtained from the gyro and accelerometers to produce electrical error signals that command the attitude-control system and main rocket engine.

The heart of the inertial system consists of the stable platform, also known as the inertial reference unit (IRU), or the inertial measuring unit (IMU). The platform is a three-axis gyro system. It establishes a stable element to which the accelerometers are mounted. Fig. 1 shows how the three gyro (G) and three accelerometers (A) are suspended by the innermost member of a gimbaled support framework. Gimbals permit the gyro assembly to tilt freely in any direction. Axes of the gyro and accelerometers coincide to make a kind of artificial horizontal/vertical frame of reference against which spacecraft motion is measured and computed.

The gyro platform "wants" to remain fixed as the vehicle tips and pitches around it. When the gyroes are disturbed in any axis or combination of axes, output signals are obtained. These signals are applied to small torque motors which reposition the stable platform. This action compensates for the axial disturbance and restores the correct frame of reference. Gyro output information also is fed into the computer to record the history of the flight for reference in future computations. As long as the computer knows where it is going and where it has been (in the form of integrated data from the stable platform), it can compute flight-path corrections to arrive exactly at the destination.

A gyro used in inertial systems is so sensitive that the weight of an oily fingerprint on its balance wheel would cause the entire gyro to drift enough to cause a serious aiming error and throw the spacecraft completely off course. The accelerometers, too, have startling capabilities. A typical unit is capable of

Containing over 1,500 components, this IRU device guided the Lunar Orbiter.

Fig. 1—Heart of inertial-guidance system is gimbaled platform with three gyro's.
WORLD'S TOUGHEST R/C JOB

measuring velocity changes as great as 10 G’s (equivalent to accelerating a body from 0 to 60 mph in 1/4 sec) or as slight as .0008 G (from 0 to 6 mph in 3 hours).

These elements are necessary to any inertial system. While such basic packages have been used in orbiting spacecraft and in missile and booster rockets, most of the state-of-the-art inertial systems (Lunar Orbiter and Apollo particularly) have complementary subsystems to increase the effectiveness of the guidance and navigation equipment. The Apollo, for example, uses a scheme like the one diagramed in Fig. 2.

The most spectacular performances of inertial guidance and navigation equipment to date have occurred in the Mariner, Lunar Orbiter and Surveyor spacecraft.

Launched on November 28, 1964, Mariner IV was lofted into an initial parking orbit around the earth by an Atlas booster. Reaching the optimum point on this orbital path from which it would begin its trajectory to Mars, the restartable Agena second stage thrust the craft into its interplanetary path. Separated from the booster, Mariner then was free to follow the commands programed into its computer sequencer or to receive commands from the control center on earth.

In flight for more than 7,375 hours before passing beyond the range of the Deep Space Network. Mariner performed literally thousands of attitude corrections and major orientation changes. Data, both from the scientific instruments carried on board and from the spacecraft support telemetry system, were fed back to earth continuously. The final signals, received far beyond the orbit of Mars, were at a level of 10^-16 (a billion-billionth) watt!

The first operation required of Mariner after separating itself from the Agena booster was to position itself to receive light from the sun. The craft’s sun sensor locked on the sun, positioning the solar-cell power panels that generated electrical power for all systems. This maneuver assured a continuing supply of power for the duration of the flight.

With Mariner locked onto the sun along its roll axis, the CC&S (central computer and sequencer) activated a second optical unit, this time to identify the bright star Canopus. Lying about 15° from the spacecraft’s south rotational pole, perpendicular to the sun-lock roll axis, Canopus was used to position the craft’s antenna systems. This maneuver enabled Mariner to receive ground-control commands required for midcourse correction. It also aided the two 10-watt transmitters used to relay scientific data back to the earth-based stations.

Midcourse commands were the only signals Mariner had to receive from ground control; the CC&S was capable of handling all other en-route computations and commands. Still, Mariner’s computer could respond to 28 additional ground-based commands, if an unplanned series of operations had to be performed during the voyage.

Midcourse trajectory was corrected December 5, 1964. Commands received by Mariner resulted in a change in attitude of 39° in pitch, 156° in roll.

A burn of 20 seconds by the rocket propulsion system changed the craft’s course 0.25° and increased its speed 37 mph. Mariner IV was now on a trajectory that intercepted Mars on the 228th day of its voyage. At the time of the Mars encounter, Mariner was 134 million miles from earth. The CC&S then triggered the television tape-recording system which transmitted pictures back to the ground-control stations.

Twenty-two complete electronic photographs were recorded by Mariner’s camera system, and data on tape were transmitted to earth at 8 1/2 bits per sec, requiring 8 1/2 hr for each of the 22 pictures. The tape was played back twice to make sure a complete set of video data had been sent down. Signals required slightly more than 12 minutes to make the trip from Mars.

The inertial guidance system had to do two things during this epic solar-system exploration: It had first to position the craft by the sun and Canopus sensors. Second, it had to function in harmony with the ground-control commands to execute the midcourse maneuver and assure an accurate trajectory. Because gyro's do drift and because magnetic, gravitational and light-pressure forces tend to change the attitude of spacecraft, the inertial system had constantly to detect minute variations in attitude and correct them to maintain the craft’s position in space. No other kind of system could do the job.

Lunar orbiter

A more recent unmanned flight, the highly productive photoreconnaissance mission of Lunar Orbiter, also had an eventful space history. As you’ll note in Fig. 3, several complex maneuvers were executed successfully during the 90-hr trip to the moon. Launched from Cape Kennedy at 19 hr 26 min GMT on August 10, 1966, the 850-lb photographic spacecraft was boosted into its initial parking orbit by the reliable Atlas/Agena rocket launch system. After 38 min and 3 sec in this preliminary orbit, during which the inertial system positioned the craft correctly for injection into the translunar trajectory, the Agena was restarted. The vehicle reached its escape velocity of approximately 25,000 mph. and the Agena parted from the spacecraft. Within 3 minutes, the high-gain and omnidirectional antennas and the four solar-power panels were deployed. The Woomera, Australia, station of the Deep Space Net picked up the craft at 20 hr 13 min 38 sec GMT. From that moment on, the spacecraft was under the control of NASA and Boeing Co. scientists and engineers from the Langley Research Center.

The success of these early operations was due to the 13.2-lb inertial ref-

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Fig. 2—The man-and-systems-augmented inertial-guidance package designed for the Apollo spacecraft is a brilliantly versatile and fantastically accurate vehicle control system.
ene unit (IRU) shown in the photo. Designed and built by Sperry Gyroscope, the IRU supplied continuous attitude- and rate-error signals for the computer sequencer to use in executing commands from its programmer and from ground-based stations.

Like Mariner IV, Lunar Orbiter carried six Canopus sensors for positional references during the lunar voyage. Sun-lock was achieved without trouble at 20 hr 32 min; but when the Canopus sensor was activated, it saw a visual light level of nearly 3 volts rather than the expected dark level of 0.56 volt, as shown by an examination of the telemetric data tapes. The 3-volt level was more than enough to blind the Canopus sensor, preventing it from locking on the star to position the craft before midcourse maneuver. Scientists eventually determined that the sensor was seeing light reflected from some part of the spacecraft itself, probably from the omnidirectional antenna.

For several hours it seemed that the mission might be doomed by this unforeseen development. By analyzing the telemetric data received from Orbiter, however, and by rolling the craft around its sun-locked axis, scientists in the control center discovered that the sensor could be locked onto the moon itself. Appearing to the sensor as a thin crescent, the angular dimension of the moon was small enough to provide an accurate position reference to the IRU. This was possible only because a 1-day delay in launching put Orbiter in the only trans-lunar flight path on which it would be able to see the moon within the narrow angle of recognition of the Canopus sensor! So coincidence, telemetric data analysis and remote-control capability all contributed together to the mission's success.

Once locked to the moon and the sun, the spacecraft was commanded to realign itself at midnight GMT on August 11. The needed thrust, computed from data received from the IRU and from the ground-control center, was supplied by the craft's 100-lb-thrust reaction engine, which burned for 32 sec. This maneuver changed the craft's course from a lunar-miss distance of 5,200 miles to an aiming point 3,950 miles behind the moon's trailing edge. Attaining the planned trajectory got the mission over one major obstacle, but for the critical retro-firing that would place the craft into an orbit around the moon, the Canopus sensor had somehow to be locked onto its star. Locked onto the moon's positional accuracy was rather crude—about 2°: an error signal accurate to 0.1° was required for the deboost engine to start.

By rotating the spacecraft so the Canopus sensor's viewing angle began to include the sun, a protective bright-object sensor was triggered into closing a shutter over the Canopus sensor's eye, and a normal background level of 0.56 volt was obtained. This confirmed that the electronics in the star tracker was operating as designed.

A specialist on the system was flown to the control center from Boeing's Seattle plant, and he suggested that the sensor might lock on Canopus despite the high background that reap- peared when the craft was once again stabilized in its sun-lock axis, if the spacecraft could be pointed directly at the star before the lock circuits were energized. Orbiter was then stepped in small rotational increments around its sun-lock axis across the area in which Canopus was expected to be found. After several attempts, the star was located and the system locked on at 13 hr 49 min GMT on August 13.

Rather than place the sensor in its automatic-track mode, however, the loop was kept open to include corrections fed from ground-control equipment. This removed the possibility that the sensor might lose the star momentarily and begin a searching-mode roll that could not be stopped by the attitude-control system.

The retro-fire maneuver was executed at 15 hr 34 min GMT August 14, and the spacecraft was injected into its intended lunar orbit. The craft's rocket engine fired for 9 min 49 sec. The IRU's 90-sec timer cut off the engine when the velocity reached 790 meters/sec. The planned orbit was inclined at an angle of 12.04° to the lunar equator and had an apolune (high point) of 1,150 miles and a perilune (low point) of 124 miles. So great was the accuracy of the inertial system that the actual orbit was inclined at 12.18° with an apolune of 1,158.7 miles and a perilune of 117.4 miles. The orbital period was 3 hr 37 min 45 sec.

On August 18, Orbiter began taking the first pictures of the moon from a lunar orbit, and the IRU once again correctly positioned the spacecraft to point the cameras to the desired sectors of the lunar surface. On August 21, a second retro-fire burn of 24 sec was calculated by the inertial system, lowering the craft to within 30 miles of the surface. Velocity data supplied by the PIP accelerometers provided the cutoff signal at the calculated de-boost speed. Picture-taking continued through August 29, at which time the IRU locked onto earth to permit transmission of the video information. Everyone is aware by now of the spectacular quality of Orbiter's photography.

Thanks to the remarkable capability of the Deep Space Network, and to the analytical ability of the scientists at the control center, Lunar Orbiter was an outstanding success. It furnished us new scientific data about the moon.

**We're on our way**

Inertial systems like those in Mariner and Lunar Orbiter have enabled NASA's space scientists to expand tremendously the range of man's exploration in our solar system and of what lies far beyond. Within 2 or 3 years, the mighty Apollo moonship—364 feet high on its launching pad—will soar to the moon carrying three astronauts, two of whom will step out of their lunar module onto the moon's surface. Guided by inertial systems and a full range of complementary systems, men will take their first steps toward the stars.
Something New in Color Generators

This one has two unique features

By LARRY ALLEN

"FIND A NEED AND FILL IT."

Good philosophy for any business anxious to prosper. I ran across it in Addison, Ill., recently while I was investigating a lead about a new color generator. This one was supposed to be a new design that would eliminate the problems service technicians have been having with color generators since color-TV servicing began.

When someone tells me a new test instrument is unique, I can't suppress a twinge of doubt; unique means "single in kind"—no other like it." I'll tell you what I found out about this one. You can decide for yourself if it solves the problems you've been having with color generators.

The company is Sencore, Inc. The new generator is their Color King model CG-141.

The first thing I did was talk with Sencore president Herb Bowden, engineering vp Bob Baum, and sales manager Dick Reed. How do they decide what goes into an instrument; indeed, why does the industry need another color generator anyway?

Herb Bowden answered first: "We evaluate what we hear from our distributors, reps and field men. For the Color King, we asked them and ourselves what are the chief problems service technicians have when they use color generators."

"And they were . . . ?"

Bob Baum supplied the answer: "One thing is stability. It seems impossible to make a solid-state timer circuit that isn't sensitive to temperature. You bring the generator in out of a cold truck and it's at least a 30-minute waste of time trying to get a pattern to stop jumping and jittering. The Color King licks that problem."

"Anything else?"

Dick Reed fielded this one. "In the servicing seminars we hold all year long, we've found that another trouble is trying to figure out which dot is the center one while looking in a mirror. Our answer is a single dot."

"Is that anything new?"

"Ours is movable," Dick pointed out with a trace of triumph. "You can move it to other places on the screen, too, for careful analysis of exactly how

Like the lid of a soapdish, this panel flips up to expose calibration adjustments. The 5 screwdriver controls at upper left are used to zero-bias the 5 TV channel outputs.

The Color King, like other color generators, makes a screenful of dots in usual manner. Conventional crosshatch pattern is useful for checking linearity on both h-v and color. The CG-141 uses a keyed rainbow display for chroma-circuit setup and troubleshooting.
the convergence is off. No lines around it to hide the edges of the dot."

Maybe I didn't look convinced, because Dick went on: "But that isn't all. We've also heard many technicians complain about the difficulty of dynamic convergence. It seems the crosshatch gets confusing after you've watched it for a while. We decided what was needed was a pattern that could be moved to whatever section of the screen was called for on the convergence chart. Then adjustments could be made for each part of the screen without the distraction of seeing other parts of the pattern shift.

"We all know that convergence adjustments are easy if they're done in the proper order. Our movable single-line crosshatch takes away the temptation to correct other parts of the pattern before the first steps are done. We think this will simplify dynamic convergence for technicians."

Dick speaks with some authority. He and five other Sencore men travel the entire country 26 weeks a year, holding a training meeting each night. He talks to from 30 to 50 technicians 130 nights every year. Since the meetings include training in servicing color TV, the field men hear a lot about the troubles the technicians encounter.

Then Bowden picked up the conversation again. "In conferences here at Sencore we decide what problems we're going to solve with our next piece of test equipment—what need we're going to fill. Then it's Bob Baum's job to figure out how to do it. He does, and we build trial models for the engineering staff and the field men to try out. Once all the bugs are worked out, we start planning production."

Dick Reed anticipated my next question by saying, "But we don't go into production right away. We build a few pilot models for the field men and for more lab testing. The Color King you're going to examine is one of this pilot run. We'll be field-testing others just like it before we actually release the model for production."

And so on. After our talk, I was walked out to the lab, given my choice of CG-141's from the pilot run, and left free to make whatever lab tests I could dream up.

I was curious about the temperature problem. What could they do in the CG-141 to cure that all-time headache—constant readjusting of timer or "hold" controls to keep the patterns steady? (This isn't a problem with most generators if the unit is in the shop and left on constantly, but it is when you take the generator along on calls. I know.) A temperature-compensating diode has helped in some generators, and I could see by the schematic that the

Color King had this little helper, too.

But that isn't enough.

What Sencore did in the CG-141 is simple, really. A heating element and a thermostat bring temperature inside the case, especially near the timer and counter transistors, quickly to 90° and then keep it there.

How well did it work? I put my Color King into an environmental chamber and ran it down to 30°F. After it had time to chill through and through, I took it out and connected it to a color set. Only 7 minutes after I plugged in the generator, the temp control light clicked off and I set the timers with the little knobs under the soapdish along the top of the front panel. Not once after that did the pattern even jitter. Repeated tests at Sencore have shown that the unit will warm up to complete stability from 20° below zero in 15 minutes or less.

Not satisfied, I ran the power cord out the access hole of the chamber. With the Color King on, I ran the chamber down to 30° again. Not a jitter. The thermostat light kept clicking on and off, but the counters stayed right on their timing.

For one last test, I unplugged the generator and let it get very cold again. Then I took it out of the chamber and turned it on. After 6 minutes, the temp control light went out. I didn't have to touch the timer controls; the pattern had locked in solid after the first 5 minutes of operation.

I thought this was great, but the feature Sencore seems most proud of is the CG-141's movable single-line crosshatch and movable dot. The photo series shows what the pattern looks like with the lines and dot in several positions. In their lab, I was interested mostly in how they developed this unusual pattern. The diagram will help you follow

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**Secret of Color King's stability is the space heater and thermostat which controls it.**

**Single crosshatch and single dot are each produced by two cross-coupled multivibrators.**

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May 1967

43
Something New in Color Generators

the explanation that follows.

Pulses from the 15,750-Hz counter trigger a multivibrator that develops a single output pulse for each horizontal line of sweep. As each line scans the picture tube of the receiver, this video pulse forms a single vertical line. A time delay is built into the multivibrator, controlled by the VERT LINE position pot, which determines how long after the start of the horizontal scan the pulse appears. Thus the pot determines the position of the vertical line.

The horizontal line for the movable single-line crosshatch is generated the same way. Taking its trigger from the 60-Hz counter, the horizontal-line delay multivibrator puts out one pulse for each frame. The timing of that pulse is set by the HORIZ LINE position control, thus selecting the location of the line on the screen.

The two are mixed in a resistive network and fed to the video section through the FUNCTION switch just like any other pattern. To form the dot, the two lines are fed through a clipping diode, and the dot is formed from their point of intersection. Thus the dot is movable by the same controls as the two crosshatch lines.

While I was still in the lab, Bob Baum pointed out other features of the Color King. The ones I particularly noted were the Zener-regulated line to the critical stages and the individually tuned rf channels. The CG-141 can be fed into channel 2, 3, 4, 5 or 6, and each is separately tunable by a screwdriver adjustment in the soopdish.

Lab work is okay, but I wanted to use the Color King. So I took my unit along with me when I left Sencore. Lugging it around with me might have some effect on it.

First thing I did was leave it outside in the cold overnight in below-zero weather. Next morning, connected to a color set, it warmed up in 10 minutes and never once shifted pattern again through more than 30 hours of continuous operation. I didn't even have to reset the timers after the Color King warmed up; they had held their stability just the way they were set when I left Sencore.

It wasn't long after that when I had a chance to try out that convergence gimmick. Would a movable crosshatch really make much difference?

Maybe a full-screen crosshatch doesn't make your eyes cross after staring at it steadily a while. Maybe you have the self-control to concentrate on only the part of the pattern you're adjusting in each step. I don't. Consequently, yes, the single-line crosshatch helped me. Step by step, here's how I used it.

Static convergence. The rf cable was connected to the antenna input and the three gun-killer leads to the solid-color CRT-grid leads coming from the demodulator board in the receiver. (Ground for the gun killer is via the rf cable and the balun transformer in the tuner input circuit.) I set the pattern for single dot and put the dot in the center of the CRT screen.

I killed the blue gun and ran the red and green together with the static magnets on the CRT neck. I reactivated the blue gun and superimposed the blue dot on the yellow one with the other static magnet and the blue lateral (that ion-trap-looking thing just behind the static convergence magnets). At this point, I also set up the gray-scale tracking adjustments.

Dynamic convergence. How you use the Color King depends on the procedures recommended by the manufacturer of the color receiver. Nevertheless, the versatility of the Color King's patterns permits uncluttered examination of convergence in any portion of the CRT screen. You can inspect how the dots go together over all the screen with the main dot pattern, or inspect one spot with the single dot.

In the set I was converging, I killed the blue gun and used the single vertical line. moved to the middle of the screen, to check convergence at top and bottom center. With the blue gun back on, I adjusted convergence controls for a single vertical white line from top to bottom.

The horizontal single line served the same purpose, first with red and green only, then with blue added.

Once, in the dynamic convergence, I reached a point where the lines (red, green and blue) simply wouldn't merge at top and bottom and still remain converged in the center. I lined them all up as parallel as possible, then used the single dot and the static magnets to re-converge the center. When I tried again to superimpose all three lines, it was no longer a problem.

One other feature I noticed and liked was the interlace control. By choosing exactly where on the raster each field begins, the operator can make the horizontal line(s) separate into two thin lines or merge as a thick line(s). For dots this is even more helpful; you'll like it best if you merge the two thin pips into one square or oblong dot. You do this with the interlace control on the front panel, right next to the single-line positioning controls.

There you have my impressions of the Color King. It is stable, it warms up fast and stays there, and it has the movable single-line crosshatch and movable single dot. It is fully solid-state. Those are the facts. You decide for yourself if it is unique and if Sencore has "found a need and filled it."
Why Servicers Like Servicing

By DICK GLASS and JEFF TRACY


What does a group of men so varied in accomplishments have in common? Each one is a service technician. The pursuits mentioned above are all part-time; these men work full time as technicians servicing home electronic equipment.

Why are these fellows in a business that carries little glamour and has been tagged as one of the toughest and most unrewarding jobs around? With the skill needed to do a good job of servicing the countless varieties of home entertainment, any of them could have a job in industry with an impressive title and a good salary.

We decided to go to men in shops and try to get the picture. We picked three that we felt were representative: a one-man dealer-technician shop where the owner both services and sells; a service-only establishment with several technicians, and a three-man partnership.

Woody’s TV (not the real name) was first. Essentially a one-man operator, Woody does use a part-time college student in his 1,600-square-foot shop. The front is a neat TV and stereo display area. Woody’s wife, Flo, helps in the office, but Woody does most of the paperwork himself. The day we walked in, Woody was in the back, tied up on a portable transistor TV.

Our first question was. “Have you had much experience on transistor TV, Woody?”

“No, this is about the tenth one I’ve ever touched. But I’m making out okay by following the manufacturer’s service notes. I’m getting more familiar every time I get one in. Not too many shops will take them. I hear, so it makes me feel sort of special.”

“Do you charge extra?”

“A little. These things are harder to get into than most portables. And I don’t have as much chance to substitute; when a transistor is bad, it isn’t like just popping in another tube.”

“What do you have planned for this afternoon?”

“Well,” he looked thoughtful. “If

Flo gets in to mind the store, I’ve got to deliver and set up a new 19-inch color set I sold last night. Then there’s a call-back on a 1960 Sylvania I looked at last week. It’s started flipping again, and I’ll bet I’m going to find a bad lead-in since it didn’t start until today. I guess you noticed the wind.”

We nodded and he went on. “That means I’ll get to play Tarzan all over their roof. Then I have another call on a stereo right across the street, and, if I have time, the church wants me to check their PA.”

“How do you feel about the service business, Woody?”

“I like it or I wouldn’t be in it. We do pretty well, but we really have to stay on top of things. I get my biggest kicks out of fixing sets like this one. Sorted of proves to me that I’m pretty good. It took enough time and study to get good at it. So I charge enough to make it worth my while, and the customers must not mind too much—they keep coming back.”

We turned to go. “Then we’d better let you get back to work. Thanks, Woody. We’ll see you at the association meeting Tuesday night.”

Next stop was Video Service Corp. (again disguised). With 13 full-time men, three part-time helpers, and a pair of office girls, VSC is the biggest independent service shop in town. VSC covers the entire metropolitan area and handles service for several department stores and appliance dealers.

Bruce Herald, president and manager, is a good promoter. He’s been in electronics servicing since the KCS 47 was a new chassis. He wasn’t in, so we talked to one of the employees, Jim Jones. Jim has been with VSC for 12 years.

“Jim, where did you get your training?” we asked.

“In the Air Force, mainly. After I got out, I went to a local school for a year and studied TV servicing. Bruce also sends me to all the service clinics the distributors hold, and I’m going evenings right now to a 12-week transistor course the association is sponsoring.”

“What do you like most about your job here?”

Jim grinned a little. “Sometimes it’s hard to be sure I like anything about it except paydays.” We waited, and he went on. “To be fair about it, I have it pretty good. I only do bench work. I don’t care much for stopping around in the snow and rain, and Mrs. Customer can be downright unreasonable at times. I let the outside men worry with that: they bring them in and I troubleshoot them.”

Jim led us over to his workbench. It and the rest of the benches were neat, even though all of them had work in progress. “You see, I’m all set up to check alignment if a set looks suspicious. The scope is always hot, of course, so I use it every chance I get—sure saves me a lot of time. I have just about every piece of troubleshooting equipment I could want to make servicing fast and easy.”

We asked Jim, “Do you ever get tired with just bench work?”

“No. We work on all makes, and there always seems to be something new. I guess that’s what keeps us in this hectic business. We like to solve problems.”

VSC’s top outside service technician walked in, so we decided to quiz him a little. Ed has worked for VSC about 10 years.

“Ed, you don’t seem to like inside work. Why?”

“Well, I don’t really dislike bench work. In fact, I enjoy getting my hands in occasionally. You get rusty if you stay away from a schematic too long. But mainly, I like to meet people, and dealing with some of the people you meet in TV servicing is a real challenge. I like to think that I’m a public relations man as well as a good technician.”

We asked Ed what equipment he uses in the home.

“We don’t try to go too far in the home. There are too many other things that may be wrong when some under-chassis part goes out. But we carry a soldering gun and tube checker, and we stock on-off controls, thermistors, horizontal phase diodes, and stuff like that. I carry a volt, a color generator and a degaussing coil.”

“Is there anything you dislike about in-home servicing, Ed?”

“I don’t like having to deal with uncorrected children and dangerous pets. And once in a while I have a customer who has been watching TV shows that
Why Servicers Like Servicing

say we're all crooks and so she won't let me take the set out of the house."

"What do you do about that?"

"I don't argue. I do the best I can and the regular customers know I'm trustworthy. If someone wants to think I'm a gyp, let them find someone else. I explain and then the customer has to decide for herself."

"Do you usually convince her?"

"Yes, as a matter of fact. I usually do. But I can't help resenting that I have to do it."

Ed picked up his clipboard. "I've got six more calls to make, so I hope you'll excuse me." And with that, he left.

We decided to leave, too, and on the way out met Bruce on his way in. We had a few questions for him, too.

"As manager of such a large shop, what do you spend most of your time doing these days?"

"I keep busy, that's for sure," he answered. "I never get to service anything anymore, though. Sometimes I think I need another manager to order parts, fill out warranty forms, check the trucks and caddies out and in, do the payroll, dream up the advertising, and a hundred other things." Bruce took a deep breath. "But the bookkeeper says we're making money, so I'll keep at it."

"Wouldn't you rather be back at the factory, making 10 G's as an engineering technician?"

"No. I guess not. I make that here, and although the work is much more demanding I get a lot more money out of it."

We thanked Bruce and moved on to our last visit. Before we got there, we noticed a truck from a local factory-service branch parked at a drive-in restaurant. A good chance to get something on that picture! We stopped.

Bill Grainger formerly worked for one of the independents in town. He's been with factory service now for a little over a year. We asked him why he quit his old job for this one.

"It was the pay, I guess. Jake's wasn't really a bad place to work. Here, though, we have all the fringe benefits factory workers get, days off, and I get to do what I like best: repair TV and stereo."

"Do you repair other brands of sets besides your own?"

"Only rarely," Bill replied. "That's another thing I like about it. You really get to be an expert, servicing the same sets over and over. We carry the parts we need and can fix them in the home usually."

"Then you're glad you made the move," we suggested.

"Yes. I have fewer worries and problems here. At Jake's, there was always some crisis coming up and I had to be responsible for too much sometimes."

We left Bill and drove on over to Triple TV, our final stop. Three fellows who once owned one-man service shops had banded together 5 years ago and formed Triple TV. Now, besides service, they have a display area for TV, stereo, auto radios and tape recorders. Jerry, Harold and Neal were all in the shop when we arrived. Neal was checking a car radio for an elderly man who was waiting in his car.

"Hello, Harold," we said, grinning. "Looks like Neal is doing all the work. You catching up on your reading?"

"I guess it looks that way," Harold replied. "I was reading up on these new auto tape machines. We've been thinking about adding them to our auto radio department."

"Is auto radio servicing profitable? Some techs seem to detest them."

"It is for us. It's like so many other things in electronics nowadays. There aren't many places a man can get decent service. It's especially true of anything that has a transistor in it, and car radios do. So . . . we do a good job and we get all we can handle of them."

We grinned at Jerry and said to Harold, "Why is Jerry dressed fit to kill today?"

"Oh, he has a lunch meeting. He's president of the Northside Businessmen's Association now."

"Well, congratulations, Jerry. All you guys seem active in some business or civic group. Neal is secretary of the technicians' association, isn't he?"

"Yes, we stay pretty active," Harold answered.

"Do you find it helps business?"

"It does, but that's a side benefit. We just like to work with people. In our technical association, the business community, church—everywhere we go we meet people we know and like. We all participate in many extra activities. With all three of us putting our heads together on the business problems, we find we handle them quickly and have time left over for civic duties."

It sounded good to us. But we wondered about the technical work of Triple. Who does it?

Harold answered, "We all do, and like it. No matter how much money a guy makes, he usually likes to work with his hands as well as his head. Servicing does both for us. I can't think of any other profession that is so completely fulfilling."

As we drove back to the office, we were thinking that the story was very much the same everywhere we'd been. From the one-man shop to the big incorporated service operation, servicing was the one way of life most appealing to employee and owner alike. Each has his own preferences, but all are united in wanting to do the very best job they could. In their shops, in their customers' homes, and in their communities. Electronic servicing gives them the chance.

END

WHAT'S YOUR EQ?

Crystal Mike Input

A high-impedance crystal microphone is connected, as shown, to an amplifier with low input impedance. This looks plausible, as the transistor gets more current drive than if it had a higher input impedance (say with the emitter bypassed). Therefore the hookup should produce more output voltage at the collector. Yet this method is undesirable. Why?—C. S. S. Sheno

Conducted by E. D. Clark

Another Black Box

Close the switch, and the lamp lights. Then open the switch, and the lamps remains lighted. The lamp can be turned off only by disconnecting the battery. No diodes or relays are inside the box. What is?—Dale Terdan

www.americanradiohistory.com
Remote Control for PA Systems

Several clever techniques for turning the amplifier gain control up or down when you’re not close to it

By JACK DARR

A REMOTE-GAIN CONTROL CAN BE A VALUABLE addition to any public address system. It comes in especially handy when there are several orators on the program. No two people have the same mike technique; this is especially true of untrained speakers (oops)—sorry. At any rate, if the sound man has a remote control, he can sit with the audience and adjust the PA gain as needed.

There are a couple of particularly neat and direct systems, both of which can be added to any PA circuit very handily. Both handle any number of microphones at the same time. They also can be used on home music systems, either mono or stereo. Commercial versions can even control as many as four channels.

A workable remote-control system must be able to adjust the gain of any channel, yet have no effect on sound quality. Wiring must be simple, and no dangerous voltages should be used. Both circuits meet these requirements.

The circuit shown in Fig. 1 is used in an amplifier popular among professionals. Only one channel is shown since all are identical. The theory is simple—the gain of each preamplifier stage is regulated by controlling its plate voltage, B+ is applied to the plate of V through a 220K resistor and a 47K resistor with a 100K plate load. At the junction of the 220K and 47K resistors, the remote control itself is a 2-meg pot connected to ground as a rheostat.

With the remote rheostat turned to maximum resistance, the applied plate voltage is unaffected. Turned to zero resistance, the applied voltage is shunted to ground; the gain of the stage then drops to zero. This might seem poor practice, but the series resistors in the circuit are so high in value they can drop the whole plate-supply voltage without getting warm. Voltage actually is present on the remote-control-box units, but—again because of the high-value resistors—current is kept so low there is no shock hazard.

To adjust the system, the remote control is turned full on. The amplifier-chassis gain control provided for each channel should be set for a little more gain than actually needed. The remote control can reduce the level but can’t raise it. If adjustments are made before the audience enters the auditorium, controls should be set so that words spoken into the mikes in a normal voice have a very perceptible “tail” or ring. This gain setting will be very near the point at which the system sets up the familiar feedback howl.

During the actual performance with the audience seated, sound-absorption will be much higher. Background noise, too, will be higher, so the increased gain will be needed.

This particular remote is a plug-in type, with the remote unplugged, the amplifier works normally. Wiring can be the standard intercom type. There is very little current flow, so very small conductors will do. And, since the circuit handles no signal voltages, the wires need not be shielded. Only one more wire than the total number of controlled channels is needed. Four channels, five wires: four hot wires and a common ground.

The second system also uses a variable resistor, but you’d never guess by just looking at it. The “resistor” is the “dark” resistance of a photoelectric cell (see Fig. 2). As in the first unit, this is an auxiliary circuit—one that can be unplugged without affecting the amplifier. A local gain control is used to set maxi-
Convert Your Heathkit WA-P2 To Feed Transistor Amplifiers

By HANK OLSON

The Heathkit WA-P2 preamp has been sold together by literally tens of thousands of Hi-Fi enthusiasts. A good number have been put together by Hi-Fi fans' wives, impatient for hobby to make music. The old preamp is still doing yeoman duty, but as the newer all-transistor power amplifiers have come upon us, many of these old "tubers" have been retired. Why? First, WA-P2 draws its B-plus and heater supply from the power amplifier. Second, its output impedance is too high to be suitable for many transistor power amplifiers.

The WA-P2 can be converted easily to a fine preamp for an otherwise all-transistor Hi-Fi system. No one will even guess you're guilty of having a tube in your system unless he actually opens the preamp.

First build a small power supply. It need not put out the 300 volts from which the WA-P2 was originally fed, since we will not need to waste voltage in RC decoupling to isolate it from the amplifier. The circuitry needs only 180 volts; the rest of the 300 went up as BTUs in the decoupling. Therefore, short together three sections of the 40–40–40–40 µF can-type electrolytic capacitor. This will be the output filter capacitor of our little power supply. See Fig. 1 for details.

The power supply is a full-wave bridge with four silicon rectifiers. Any rectifier with a piv of 400 volts or greater (M150, MS50, 1N1695, etc.) can be used. My unit was constructed in an LMB No. 141 box chassis, fitted with four rubber feet. See Fig. 2 for the complete circuit.

The last change is to lower the output impedance of the cathode-follower stage. Replace the 2,200- and 47,000-ohm resistors in the cathode of V1-b with a 680-ohm 1/2 watt and a 15,000-ohm 1-watt resistor, respectively. Replace the output coupling capacitor (0.1 µF) with a 20-µF 150-volt tubular electrolytic as shown in Fig. 1. Then put a 120,000-ohm 1/2-watt resistor across the output jack to dissipate any charge that might be caused by leakage through the filter electrolytic.

Plug the new preamp into your transistor amplifier, turn it on and zero the hum control.

Of course, the modified WA-P2 will work just as well as before, if not better, with tube-type power amplifiers.

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**Fig. 1—Output stage and decoupling network of Heathkit WA-P2 preamp/control unit. All necessary changes are shown. Used units are available in many large audio or surplus stores from $10 to $25.**

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**Fig. 2—Simple power supply for converted WA-P2 frees it from tube amplifier forever. Any well-filtered supply that delivers about 180 volts at 20 mA or so will work okay.**
COMSAT: Communication in the Space Age

Not experimental, but commercial, instant worldwide information transmission by satellite

By RAY D. THROWER

In the 17th century, it took about 4 months for news of the New World to reach Europe. Now, with satellite communication, news whips around the globe in seconds. In less than 3 years, instant global communication will be a reality. Advanced communications equipment and the space-age vehicle, the Communications Satellite Corp. and its international partner, Intelsat, are all together responsible for that.

"Just what is COMSAT?" is a question one frequently hears. Many have the idea that COMSAT is a government agency, staffed by Federal civil-service personnel. This mistaken idea probably comes from the fact that COMSAT was authorized by the Communications Satellite Act passed by Congress in 1962. The basic Communications Act of 1934 made no specific provisions for satellite communication. In fact, in 1934, satellite communication was placed in the category of Buck Rogers space adventure stories, popular in the late 1930’s. COMSAT’s relationship to the Federal Government is about the same as the relationship of other communication companies such as General Telephone & Electronics, American Telephone & Telegraph, and International Telephone & Telegraph. They are all Government-regulated, profit-making stockholder-owned organizations.

RADIO-ELECTRONICS visited the new earth-station facilities at Brewster Flat, Wash., and Paumalu, Oahu, Hawaii, and obtained an interview with Wallace M. Lauterbach, Western area manager for the Communications Satellite Corp. Lauterbach has been in communications for about 25 years.

He was graduated in 1941 from the US Military Academy with a BS in electrical engineering. He obtained his MS from the University of Illinois. During World War II, he commanded signal troops in the Pacific Theater. Since then, he has been executive officer to the Chief Signal Officer, Department of the Army; a member of the US delegation to the International Telecommunications Union in Geneva; military assistant to the telecommunications adviser to the President; and first Commanding Officer, US Army Strategic Communications Command.

When Colonel Lauterbach retired from active duty in June, 1965, he was an obvious choice for Western area manager, Communications Satellite Corp.

After we toured the COMSAT site at Brewster Flat, Wash., Lauterbach invited us into his office for some discussion about COMSAT and the future of space-age communications.

MAY 1967

RADIO-ELECTRONICS: What is COMSAT’s purpose?
COLONEL LAUTERBACH: It’s to be a world-wide commercial communications satellite network to provide communications services to business, government, and individuals. Understand one thing: When we speak in terms of “communications” here at COMSAT, we mean not just telephone conversations, though they will be an important part of COMSAT’s activity. But I think the important contributions will be data transmission and, to a lesser degree, video communication.

R-E: Well, will the communications satellites be flexible enough to handle the different kinds of communications circuits you’re talking about? For example, can one single satellite take care of voice, data and video traffic, too?
LAUTERBACH: I’ll give you a qualified yes to that question. Qualified only because of the way it was worded. Yes, the present satellites can handle voice, data and video. But not all at the same time. They can handle a mixture of voice and data. The exact number of circuits depends on the speed. and therefore the bandwidth, of the data circuit. The real limiting factor is the terminal equipment used at the earth stations. The receivers and transmitters are the same for all modes, but the demodulating and modulating equipment is different for voice, data and video.

R-E: How many of each type circuit can satellites handle?
LAUTERBACH: Early Bird, which was our first program, can handle 240 two-way telephone conversations, or 6,200 full duplex, simultaneous teletype circuits, or one television video circuit. It can handle a few computer circuits or hundreds. As I mentioned before, the exact number of computer circuits will depend on the speed of transmission of the data.

R-E: I see. I’d guess that the communications satellites launched early this year can handle more than the 240 voice circuits of Early Bird, true? Do you have a name for the current program?
LAUTERBACH: Let’s take those in reverse. Early Bird was one name for what we call Intelsat I. That’s a single satellite located over the Atlantic off the east coast of South America. The current program, the one that affects us here at Brewster Flat and at Paumalu as well is called the Intelsat II series. We have several satellites for this second phase. One did not achieve a usable orbit and is idle. The second is stationed above the Pacific Ocean about halfway between here and Australia. A third will be put on the opposite side of the globe over the Atlantic off the west coast of Africa.

As to channel capacity, Intelsat II spacecraft have the same capacity as Early Bird but more than twice the area of

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COMSAT: Communication in the Space Age

coverage. However, we're constructing what we call Intelsat III. That will be what we call a "multiple-access" type communications satellite. These so-called global satellites, for use starting in 1968, will have a capacity in excess of 1,200 voice circuits each.

R-E: You've mentioned Intelsat several times. What is that?
LAUTERBACH: Intelsat stands for International Telecommunications Satellite Consortium. It's an organization made up of a group of the member nations of the ITU, the International Telecommunications Union, which is an arm of the United Nations. Right now we have more than 55 member nations in Intelsat. Intelsat owns the satellites. COMSAT holds a majority interest, and acts as manager of Intelsat. Each member nation, or its commercial representative, will own its own earth station. We expect to have as many as 30 earth stations operational by 1968.

R-E: You also mentioned "multiple-access" satellites. What do you mean by that?
LAUTERBACH: Well, by using a single broadband input receiver, a large number of earth stations, say, 10 or more, can communicate through the same satellite simultaneously, even though each earth station transmits on a different frequency. In fact, for the system to work, each earth station must transmit on a different frequency. Each station is assigned a band in the satellite receiver's spectrum so one earth station's transmissions won't interfere with those of another.

Actually, you know, the communications satellite is a glorified translator, comparable to the vhf/uhf translators used to serve a lot of communities with TV. Our translation frequency is 2.225 GHz.

R-E: What bands do you operate in? I read that it was in the 6-GHz and 4-GHz bands, but there are already so many microwave systems operating in those bands, it would seem you'd have quite an interference problem.

LAUTERBACH: Exactly. Actually, you have no idea of the number of common-carrier microwave systems in operation.

R-E: What are common carriers?
LAUTERBACH: A common carrier is an organization, like a telephone company, that sells communications services. There are so many in operation in the bands we operate in that we've had to get sort of a special dispensation from the FCC that any future systems in our vicinity will be installed and operated on a noninterference basis. General Telephone Co. of the Northwest brings in the microwave relay channels that carry the COMSAT circuits out of Brewster Flat. They had to do some special engineering to get their microwave in here in the 11-GHz band, so as not to interfere with our 4- and 6-GHz operation.

R-E: What about the case where there was already a system in operation in your band? What do you do then? I'd think this might be pretty important when it comes to site selection.

LAUTERBACH: You've just hit on one of the most difficult things...
Future satellites will use "multiple-access" receivers. They will have extremely broadband front ends and will be able to receive from 10 separate transmitters on different frequencies simultaneously. Bandwidths of 250 MHz will be divided into 25-MHz channels and each channel assigned a discrete frequency. Each frequency will be assigned to one earth station. Any one of the inputs will be able to accommodate up to 960 voice channels or one video circuit or a variety of data circuits or even combinations of data and voice circuits. This change in configuration will be done at the earth station by switching in the appropriate terminal equipment. Each channel now carries one video or 240 voice channels.

about setting up an earth station: site selection. Yes, we have to have an "electronically quiet" environment. Our receivers, which are cryogenic systems by the way, have a sensitivity of -159 dBm, so, not just any place will do. We looked for quite a while before finding the Brewster Flat site. We're in the bottom of a saucer-shaped depression between several mountain ranges. The mountains shield us from other microwave systems. Of course, we have a certain maximum angular elevation limit on our surroundings. Anything above 4° might obstruct the path to the "bird."

R:E: You mentioned your receivers are cryogenic devices. This means they're supercooled to reduce the natural electron noise, doesn't it?

LAUTERBACH: Yes. They're cooled to 4° Kelvin. And that's close to absolute zero.

R:E: That should keep anything quiet!

LAUTERBACH: It does a good job of it. Actually, we're not the first to use cryogenics. Radioastronomy systems have been using them for years and many of the telemetry systems for space work use cryogenics.

R:E: Besides the use of cryogenics, are there any other specific technical details in the COMSAT system that aren't used in the usual communications system?

LAUTERBACH: Oh, yes. One thing that seems to surprise quite a few technicians and even some of the younger engineers is the fact that we transmit and receive simultaneously on the same antenna.

R:E: Could you explain how that works?

LAUTERBACH: The technique has been used for years in microwave and vhf and uhf communications. We use what we call a duplexer. It's a resonant-cavity device, actually two cavities, one tuned to the transmit frequency and one to the receive frequency. At the resonant frequency, the cavity represents a low impedance to any energy it sees. At any other frequency it looks like an extremely high impedance, so the transmitter output is effectively isolated from the receiver input, but the receiver can still "see" any signal that's on its frequency.

R:E: Sounds like something very useful. It lets you get away from having to build two of these "monster" antennas for each direction of transmission. doesn't it?

LAUTERBACH: It sure does. And that cuts down on the overall. There are some microwave systems that connect as many as eight transmitters and eight receivers to the same antenna, all operating simultaneously.

R:E: Whew! Let's see. COMSAT was organized in 1962, and

* 6 dBm equals 1 mW.

MAY 1967
COMSAT: Communication in the Space Age

Communication needs, we'll need 12,000 circuits by 1980. We added 240 circuits with Early Bird. This amounted to an increase of about 30%, but the most impressive improvement is the instantaneous availability of these circuits over an area of tens of thousands of square miles.

R-E: What kinds of customers will COMSAT serve?

LAUTEBACH: The most often mentioned example is NASA. We're providing just about every conceivable type communications circuit to NASA for the Apollo program. Probably one of the most interesting services we propose is to provide voice and data communications to aircraft in flight on transoceanic runs.

R-E: Oh, I think I understand. On long over-water flights, vhf communication won't work, and the hf radio bands are pretty crowded—and not always reliable.

LAUTEBACH: Exactly. Direct communications will play an important role in air traffic control in the future, especially when the 2,000-mile-per-hour passenger liners go into service. Recent estimates show that at any given moment there are over 280 aircraft over the Atlantic alone. And don't forget the ships at sea. We can provide them with telephone and data service to the home office. That way, if there's a change in the price of, say, oil, in a certain port, the home office can direct the tanker to go to another port where the price is better.

R-E: What about the possibilities of satellite communications systems being used for worldwide educational television? Does COMSAT or anyone else have anything along these lines?

LAUTEBACH: Yes. AIB, CBS, and NBC have already expressed interest in this area. Certainly it would be technically feasible. Actually, when we consider the ETV aspect of satellite communications, the only thing that keeps us from doing it is "doing it." The technology exists. The only thing still necessary is the political and economic backing. COMSAT has already outlined a program for a domestic US satellite system that would serve the major TV networks as well as handle ETV.

R-E: How about computers? Couldn't they be tied together through communications satellites? This would help in making data available on a world-wide scale. Hugo Gernsback, editor-in-chief of Radio-Electronics, in editorials for December 1959 and May 1964, urged the establishment of a "national facts center." Using your facilities, a facts center could be international, couldn't it?

LAUTEBACH: Someone's been reading our mail! Seriously, though, the establishment of information grids, connected by relay satellites, has already been proposed. Some authorities think that in less than 10 years a student will be able to dial a local computer on his home telephone and program problems into it. This is already being done on a limited scale, but not with relay satellites for computer interconnect. But it could be done.

R-E: I'll bet engineering firms and other businesses would benefit from being able to tie into such a system.

LAUTEBACH: They certainly would. And they'd find the cost not much more than a monthly telephone bill and a lot less than owning and maintaining their own computer.

R-E: Seems like you're going to have a lot of people relying on your satellite. What happens if it goes had after just a few days of operation? Or what if it doesn't work to begin with? You can't send a man up to fix it—not yet, anyway. What do you do?

LAUTEBACH: To begin with, our systems are designed to minimize failure. Each component and each unit is designed and tested to meet extreme requirements. The chance of failure is pretty remote. If a failure should occur in a critical component after the bird is up, we still wouldn't have a failure because the equipment has built-in redundancy. That means there is a parallel unit that will take over the function of the defective unit. And, if, just if, the bird should be a total failure, we do have a couple of spares we can send up. But that's expensive.

R-E: I guess you're pleased with Early Bird's performance. It went up in, let's see, April of 1965, wasn't it? And it's still operating.

LAUTEBACH: Yes. Early Bird had a life expectancy of 18 months. It exceeded that by quite a margin. And looks like it will keep going for a while yet. The satellites orbited this year are designed to operate for 3 years and the ones planned for Intelsat III are being designed for a life of 5 years.

R-E: What is the power of the transmitter in the satellite?

LAUTEBACH: Six watts.

R-E: Six watts? But the one at the earth station is 12,000 watts!

LAUTEBACH: It does seem strange, but remember that right now our techniques don't permit a very high power-to-weight ratio. We're limited to low-powered transmitters on the satellites. We make up for this by using the large antennas and cryogenic receivers at the earth station. Going the other way, we can transmit from earth with high power and large antennas, with their high gain, and come up with a respectable signal level for the satellite receiver. This way, we can use fairly conventional circuits for the receivers in the birds and get away from having to put huge antennas and cryogenic receiver systems in orbit.

R-E: Then, actually, the complicated circuits are at the earth stations, more so than in the satellites?

LAUTEBACH: In a manner of speaking, that's true. But that isn't to say that the circuits in the satellites aren't up to the state of the art. Some of our equipment is far advanced from the equipment of the more conventional, earthbound systems. It has to be, because of size and weight limits.

There's a great future for satellite communications and its engineers and technicians—a future where not even the sky is the limit.

[In late January, Intelsat II's Pacific satellite Lani Bird began to serve in two major functions. AT&T started using the satellite for commercial telephone service—with 6 circuits to Hawaii and 30 to Japan. And ITT initiated commercial TV use of Intelsat II with transmission of an NBC newscast to Nippon Television Corp. Fulltime commercial service is now underway between North America, Hawaii and Japan. The Atlantic satellite Canary Bird was launched March 22. —Editor]
Selected Circuits from an Experimenter's File

Eight nifty circuits more people ought to know about

By ROBERT F. SCOTT

An avid experimenter, I collect and file hundreds of circuits that I hope to be able to use in one form or another. Some of the most interesting and useful of these appeared in foreign publications. Others were developed by component manufacturers and published in application notes, catalogs or ads in specialized electronic publications not generally available to experimenters. I hope you will be able to use some of these selections—or maybe add them to your files.

Remote-tuned superregenerator

Fig. 1 is one of many circuits developed by Vari-L. Co. (207 Greenwich Ave., Stamford, Conn.) to show typical applications of their electrically variable inductors. The detector tunes from 20 to 50 MHz with the Vari-L type M-1 inductor as the control current is varied from 0 to 60 mA. The antenna is connected across a 100-µH rf choke in the emitter circuit. The audio output can be fed to a simple transistor or tube amplifier.

The electrically variable inductor is a two-part assembly. The control winding is wound on the legs of a U-shaped ferrite or laminated core. (There may be more than one control winding.) The signal winding—used as the tunable coil in a Colpitts oscillator or similar two-terminal network—is wound on a ferrite core placed across the open end of the U.

When current, ac or dc, is passed through the control winding, it varies the magnetic flux of the assembly and decreases the permeability of the signal-winding core. This reduces the inductance of the signal winding and increases the frequency of the tuned circuit. Vari-L current-controlled inductors in the company's Mite series are available with inductances ranging from 100 mH to .05 µH at zero control current. They cover frequency bands from 2 to 30 kHz for the 100-mH unit to 250 to 300 MHz with the .05-µH inductor.

SCR time-delay relay

This circuit (Fig. 2), developed by G-E, controls loads up to 100 watts direct from the 117-volt ac line without intermediate relays. When the switch is pressed, D1 conducts, charging C to around 140 volts and discharging a negative voltage on the SCR gate. When the switch is released, D2 conducts and starts to charge C positive through R1 and R2. The SCR does not conduct until D2 has had sufficient time to reverse the polarity of the voltage across C and raise it to firing potential.

The timing range is from 0.1 to 30 seconds with a maximum error of approximately 1%. Reset time is less than 0.1 second. (Note well that C is a metalized paper capacitor such as Sprague type 118P or equivalent.)

Line-voltage indicator

Technicians and plant maintenance personnel need a device to identify quickly the voltage and frequency of the various power lines installed in some plants and labs. Fig. 3 is a simple indicator that can be mounted in a non-metallic instrument case and fitted with insulated test leads. The instrument was developed at Clevite Corp. and described in Signalite Application News.

The indicators are NE-2's but Signalite T2-27-1R100's are recommended because of their closer tolerance in breakdown and maintaining voltage. Breakdown voltage is 66 to 74 and maintaining voltage 52 to 59.

When the test leads are clipped across a de source, C2 and C4 prevent the 60- and 400-Hz indicators from firing. The de voltage appears across C1 and fires the dc indicator. Since only one electrode glows on de, the indicator can be calibrated to show voltage polarity.

On a 400-Hz ac source, C4 and R7 form a voltage-dividing impedance network that develops enough voltage to fire the 400-Hz indicator. Dividers R5-C1 and C2-R6-C3 prevent the dc and 60-Hz indicators from firing. On 60-Hz sources, enough voltage is developed across C3 to fire the 60-Hz indicator.

When the line voltage is 115, the 115-volt indicator fires with R3 and R4 serving as current limiters. There is not enough voltage drop across R2 and R4 to fire the 440- and 220-volt indicators.

When 220 volts is applied to the test leads, the 220- and 115-volt indicators fire. Raising the input voltage to 440 causes the 440-, 220- and 115-volt lamps to fire.

Remote-controlled attenuator

The internal resistance of a vacuum tube has been used as a variable resistance in tone controls, vibratos and similar circuits. To date, transistors have not been widely used in such applications because of their low impedance. Field-effect transistors are ideal for such audio control jobs. Fig. 4 shows a remote-controlled attenuator circuit designed for the 3N98. See RCA
Selected Circuits from an Experimenters's File

Application Note AN-201.

The circuit is an L-attenuator with the 1-megohm resistor as the series element and the drain resistance of the transistor as the shunt element. Drain resistance is a function of gate voltage so the attenuation can be controlled by varying the dc voltage applied to the gate. With the circuit constants shown, the maximum attenuation is 60 to 70 dB and the minimum is 1 or 2 dB. Values of C1, C2 and C3 depend on the desired low-frequency limit. For 30 Hz, .05 \( \mu F \) is ample for C1, and 25 \( \mu F \) for C2. A value of 0.5 \( \mu F \) for C3 will be sufficient for loads of 100,000 ohms or higher.

(Insulated-gate field-effect transistors have extremely high input resistances and are likely to be damaged if electrostatic charges are allowed to accumulate through improper handling—allowing the leads to brush against silk or nylon fabric, for instance. These transistors can also be damaged by electrostatic charges that the human body picks up under normal conditions. Read manufacturer's handling and operating instructions before removing FET's from their package.)

Audio-frequency AM detector

The detection of slow changes in the amplitude of audio-frequency carriers is often used to telemeter specific operating or environmental conditions in medical electronics, communications, oceanographic and space surveys, and industrial and scientific instrumentation projects.

Fig. 5 is the circuit of an AM detector for a 250-Hz carrier. The incoming modulated carrier is amplified, rectified and then passed through the pi-network low-pass filter to remove the carrier. (This and equivalent circuits can be replaced by the L55-PD postage-stamp size—0.5 cubic inch—low-pass active filter developed by Guilemin Networks, Inc., 170 Brookline Ave., Boston, Mass. 02215.)

Waveform generators

Various types of signals other than sine and square waves are often used in timing circuits, triggers, sweeps and other applications. The circuit in Fig. 6 develops a symmetrical triangular waveform whose frequency is determined by the capacitor's value. The slope of the waveform is \( 5.5 \times 10^5 \) volt per second.

The circuit in Fig. 7 is a linear-voltage ramp generator whose frequency is variable from \( 7.3 \times 10^4 \) to \( 6.95 \times 10^4 \) volt per second. The circuits in Fig. 6 and 7 were developed by Texas Instruments.

Polarity-reversal hint

Fig. 8 shows two circuits commonly used to reverse the direction of a dc motor or the polarity of the voltage applied to a similar inductive load. Both circuits do the job but Sigma Instruments points out that the circuit at b is the preferred arrangement.

Sequential tone control

Sequential-type remote controls that operate only after a second tone pulse (of different frequency from the first) has been received are generally complex circuits requiring filters or high-Q networks and stepping or ratchet relays. This circuit (Fig. 9) is a very simple one developed by Bramco Controls Div., Ledex Inc., Piqua, Ohio, to illustrate one of the uses of their two-channel resonant-reed relays.

When the first tone is received the F1 reed vibrates and charges C through its contact connected to one side of the power supply. When the second tone comes in, the F2 reed vibrates and transfers C's charge to the grid or base of a control tube or transistor.

END
An Engineer Talks About Transistors in Audio

Part 2—A discussion of how semiconductors are used in hi-fi FM tuners

By PETER E. SUTHEIM

Last month we discovered some reasons why transistors are making better audio amplifiers possible—better than the tube amplifiers of a few years ago. The improvements, we learned, are not only smaller size and greater efficiency, but better performance as well.

In this second and final part of our Transistor Appreciation Course, we continue our conversation with Daniel R. von Recklinghausen, chief research engineer of H. H. Scott, Inc. This article, about semiconductors in tuners, comes from the same taped interview.

Dan started with a little historical background.

"The very first circuit in an FM tuner to use solid-state components was the detector: germanium diodes. Well, why was that?"

"They were small and made wiring simpler, I suppose," I said.

"That's true, but that isn't the big reason, from the designer's viewpoint. The ratio detector in an FM set is at the highest signal level in the whole circuit. Now, in a ratio detector, one diode gets the signal at its plate, from one side of the transformer, and the other diode gets it at the cathode, from the other side. So in a tube you have all this 10.7-MHz stuff being coupled into the heater winding, through heater-to-cathode capacitance in the tube.

"To filter that out and keep it from causing regeneration in the whole i.f. strip was a very tough job. That's where the germanium diodes came in."

"What was the next step? I asked.

"The multiplexer: switching. Diodes or transistors may make very lovely switches for stereo multiplex."

"And then?"

"The i.f. amplifier. Originally, FM i.f. amplifiers were just an extension of AM transistor-radio technology, with germanium transistors. You had to neutralize each stage to prevent regeneration—right back to the old triode r.f. and i.f. amplifiers. Neutralization is a tricky thing; it’s quite critical, and you have to sort of tune it. Then came high-gain, high-frequency silicon planar transistors, and we could design i.f. strips without neutralization, by mismatching impedances and sacrificing a little gain.

"Now by mismatching, we gained something else. We could make transistor stages that did not load the tuned circuits and reduce their Q and selectivity. So the passive networks—transformers—alone determined the bandpass of the i.f. system."

"What followed that?"

"Well, actually, then we started putting field-effect transistors in the front end, but that's getting ahead a little. While we're on the subject of i.f. amplifiers, I should go on to the next step—the use of integrated circuits for i.f. amplifiers."

"In a transistor you always have some collector-to-base capacitance—a couple of picofarads. It limits the absolute maximum gain. An integrated circuit, such as the one we use, has a fraction of a pF more like pentode tubes. So the gain-handwidth product of an IC is higher than that of most transistors, and the input and output impedances are higher. Now you can either go to still higher gain per stage or get a lot less dependence on individual IC characteristics affecting the tuning or Q of transformers. A 2-to-1 change in the gain of an IC affects the tuned circuit maybe 5%; with old germanium transistors it's more like 50%. Silicon transistors are somewhere in between."

"Is that pretty much all? I mean, that's nice enough, but is it just that ICs do the same job somewhat better?"

"No. Let's look at something else. Suppose you get a big, fat ignition pulse riding in on top of your FM signal. Now the age is too slow to change the system gain that quickly. So something will overload. The base-emitter junction of a transistor is like a diode and so at least one of the transistors will rectify the pulse and store up the current from that in the filter capacitors of the bias network. That changes the operating point for a time, until the capacitors discharge down to normal. So you get a lump in your signal."

"Now the IC we use as an i.f. amplifier is an emitter-coupled circuit, which is a beautiful current limiter. It limits before it can change voltages and currents. Not only that, but the IC operating points are set by built-in diodes, not by resistor networks. The diodes have a very low impedance, so they don't need to be bypassed with capacitors. So there's a tremendous improvement in overload recovery."

"Let's look at another facet: capture ratio."

Dan digressed here for several minutes to explain some of the intricacies of capture effect.

Briefly, capture effect is the ability of one FM signal only slightly stronger than another on the same or a nearby frequency to "capture" the detector and dominate the other—almost to the complete exclusion of the weaker signal. The arbitrarily chosen figure for gauging this exclusion is 30 dB (audio ratio) at the output of the detector. Capture ratio is the ratio of the two incoming signals, expressed in decibels, that the tuner must "see" to achieve 30-dB suppression of the weaker one compared to the stronger. The smaller the number, the better the tuner. The best modern FM tuners..."
An Engineer Talks About Transistors in Audio

offer capture ratios between 1 and 2 dB—quite remarkable.

Dan continued.

"I said before that an emitter-coupled circuit is a fine symmetrical limiter. But it's very hard and very expensive to make one with separate transistors. In transistors with the same type number, there are often differences of 50 to 100 mV of bias for the same current. The diffusion is slightly different, and so on. Now this can give you a current variation between two transistors of as much as maybe 30 to 1. So the variations in transistors, even within the same batch, make it necessary to have separate adjustable bias networks for each one of the pair. Otherwise, it's not symmetrical.

"Which brings up this operating-point shift again."

"Right. Now in an IC, the advantage is that the transistors in it are formed close together, on the same silicon chip. They're bound to be nearly alike. A typical variation is perhaps 1 mV for the same current—5 mV is often a reject limit. So this gives you your first chance to make a good emitter-coupled limiter.

"And the current in an IC emitter-coupled amplifier is limited by a current-source transistor—the emitters of the emitter-coupled pair are tied together and fed from the collector of another transistor, connected as a constant-current source. So all the pair can do is divide that current between them. As we drive one to zero, we drive the other one toward maximum, or back the other way. When the transistors are alike, this has to be symmetrical."

"Now what about FM front ends? Why do you use field-effect transistors there?"

"I explained before that the input impedance of a transistor is that of a diode. If a diode, or a transistor, is subjected to one signal, its nonlinearity will make harmonics come out. Assuming we have a tuned amplifier, the harmonics of one signal won't bother you, because they're way outside the tuning range. If the signal is at 100 MHz, the second harmonic is at 200, which is way beyond the tuning range of the receiver. Now let's put in two signals. Because of the nonlinearity, you'll get intermodulation.

"First-order intermodulation, which corresponds to second-harmonic distortion, gets you sum-and-difference frequencies. If one signal is 100 MHz and the other 101, the sum is 201 and the difference is 1. Still no problem. So having a second-harmonic characteristic isn't bad. It may even be convenient—we can apply two signals and come up with the product, like the desired signal and dc. We can change the gain with that trick. This is what we call a square-law device.

"Let's look at second-order intermodulation, third harmonics."

Dan pointed out that you could get "harmonic babies" at 99 and 102 MHz, which is well within the FM tuning range.

"So we want, in an FM front end, a device that does not create any third- or higher-order harmonics. Or as few as possible. A diode is pretty bad. A tube is better—it works on a three-halves-power law instead of a square law. You have some higher-order products, but not so many. That's why some of the better tuners that used transistors everywhere else used nuviators or some other kind of tube in the front-end.

"So, the world was looking for that perfect square-law device. Turns out it was described back in 1952 by one of the inventors of the transistor—Shockley. A unipolar field-effect transistor. Strictly theory. It wasn't until a few years ago that the first ones came on the market—at $30 apiece in production quantities! With transconductances of only a few hundred micromhos and capacitances of only a few picofarads. Not too good for high-frequency work.

"Nobody even thought of using them for rf—everybody looked at that high input impedance and out came timing circuits, dc amplifiers and so forth. Manufacturers tried to make FET's with a constant transconductance—a nice, linear class-A amplifier. Not linearly changing transconductance, which is what was needed so we could apply age.

"Well, we cooperated with Texas Instruments to go back to the original Shockley thing—high low transconductance, low capacitance, and square-law."

"So that explains why you use field-effect transistors."

"Yes, that's one reason. But there's another. Are you familiar with the concept of equivalent noise resistance?"

"A bit vaguely, yes."

"Equivalent noise resistance is a fictitious resistor connected in series with the input of an amplifier. You can't ever short it out, and it makes noise. This equivalent noise resistance is what determines noise figure. Now in a triode tube, which is what is most commonly used now as a low-noise rf amplifier, the equivalent noise resistance is equal to 2.5 divided by the transconductance. Say a good rf tube has a transconductance of 100,000 micromhos. Its noise resistance would be 250 ohms.

"Now where does that 2.5 come from? The formula says noise resistance is inversely proportional to Gm—in other words, proportional to 1/Gm. A tube works about 2.5 times higher than room temperature, expressed in absolute temperature (°K). That's about 750°K, which is quite hot.

"In a field-effect transistor, which operates around room temperature, the noise resistance really is 1/Gm. You can get many thousands of micromhos transconductance now in FET's, but with much less noise than the same value in tubes."

"Better than transistors?"

"Oh yes—better. Far better than transistors. Also, high-Gm tubes lose Gm as they age. FET's don't seem to do that.

"So I think you see we've improved things at both ends. Lower noise gives a higher useful sensitivity, and less cross-modulation means you can handle bigger signals. You've increased the dynamic range. You know how everybody fights to get a fraction of a dB more sensitivity. Well, that fraction of a dB at the other end—high signal level—is perhaps more important."

"I see."

"Now let's look at figures. With tubes we had maybe 80 to 85 dB cross-modulation rejection. With conventional transistor circuitry, you run 60, 70—maybe 75 dB. If you're really careful. With FET's, we have measured in excess of 100 dB."

So that's the story. If you've noticed the repeated trend in the introduction of new semiconductors and circuits from esoteric military/industrial applications, to high-quality high-fidelity equipment, to low-cost radio-phones sets, you can predict that FET's will soon begin appearing in less and less expensive equipment as their cost drops and designers become more familiar with them. While there will always be manufacturers eager to produce junk, recent developments make it almost impossible to avoid producing radios and amplifiers considerably better than those of a few years ago.

For most of the home-entertainment equipment we're likely to run across in our daily work, it seems safe to say that tubes are pretty much extinct!
Imaginary Numbers are a Cinch

Part 1—Math tools that aren’t real, but which work in a real way

By NORMAN H. CROWHURST

IT ALL STARTED WHEN GEORGE NEEDED a filter for the new cross-multiplexing system he was developing. He called me up about his problem and I asked what he wanted the filter to do.

"Come on over, bring your design stuff, and I’ll show you," was his response.

I went over to his lab and he explained the problem to me. After a little figuring, which he watched, I sketched a schematic and put the values in.

"We’ll have that made up in a jiffy," he said, as he picked up the phone to call the storeroom for parts.

In less than an hour a messenger brought the parts. George’s technician wired them, and George had a working filter. Satisfied that George had what he wanted, I turned to leave, but he called me back.

"Just a cotton-pickin’ minute," he said. "Can’t you show a fellow how you figured those values in such short order, so I can do it myself next time?"

It took me twice as long to explain it as it had taken me to figure it. He realized that what made the calculation difficult for him was the use of imaginary numbers, although that was what made it easy for me. As I left, he made me promise to explain imaginary numbers to him in easy stages.

One slack afternoon about a week later, not long before closing time, George and I got together again. "Let’s see why the idea seems difficult," was how I started.

"Do you remember when you first encountered irrational numbers?"

George remembered learning about them, but couldn’t remember what they were.

To clarify them, I quickly went over rational numbers, particularly fractions. I showed, using a number line (Fig. 1.), that \( \frac{3}{4} \) and \( \frac{5}{8} \) are fractions of close to the same magnitude. "If we compare various fractions, it seems as if we can make up just about any size of part we want by taking a suitable denominator, or bottom, and then selecting an appropriate count, or numerator, of that denominator," I illustrated this by fitting \( \frac{19}{80} \) between \( \frac{3}{4} \) and \( \frac{5}{8} \).

"That seems logical," George said, "but you say it as if there’s a catch. What is it?"

"Well now. George, I’m sure you remember some numbers that could not be accurately written, either as decimals or as fractions. No matter what denominators and numerators you use, nor how many decimal places you use, you can reach only an approximation."

"You mean like \( \pi \)?" George asked, and then before I could answer he added, "or root two?"

"That’s it. You do remember. That’s the difference between rational and irrational numbers. Any number that can be written accurately with fractions using whole numbers for numerator and denominator, or that can be written with a terminating decimal, is called a rational number. It fits the known pattern of numbers before irrational numbers were admitted to exist. Then mathematicians found impossible- seeming numbers like root two, or \( \pi \) ...

"And they’re irrational numbers?"

George butted in.

"That’s right."

"Makes sense," he responded, "although I don’t remember learning it that way in school. But irrational numbers aren’t the same as imaginary numbers, are they?"

"No, they’re not, but notice this: At one time math scholars thought any number could be represented by a fraction. Later they realized that irrational numbers belong to a different class than rational numbers. And do you remember when you first learned about negative numbers?"

"I should say," George replied.

"They gave me a hard time. Especially that bit about minus times a minus making a plus. I never did understand that fully."

"Oddly enough, understanding imaginary numbers will make that easier too. I then asked him. “Do you know how negative numbers differ from positive numbers?”

"Well, as I see it," he said, "it’s like another world, an upside-down world. When you combine negative numbers with positive numbers, you subtract the negative ones whereas you would add them if they were all positive."

"You’re right. I don’t know if a math teacher would accept that answer, but it shows you know. Now, do you remember doing squares and square roots?"

"It’s a bit rusty," he replied, "but I remember it."

"Let’s take some simple cases," I suggested. "What is 2 times 2?"

He looked at me as if I must think him stupid, volunteered "4," and then looked quizzical, as if he thought that answer might somehow be wrong in higher mathematics.

"Right. Now 3 times 3?"

"Nine," he replied, still puzzled.

"So you know what squares are. Now what is \(-3\) times \(-3\)?"

"Nine," he replied, hesitantly.

"Yes, but is it +9 or -9?"

"Well, I remember having trouble with that," he admitted. "but if I remember right I got it through my head that it was +9."

"Right. Then I asked, "So what’s the square root of \(+9\)?"

"Three," he replied.

"Plus 3 or minus 3? I asked."

He thought for a moment and then said, “It could be either, couldn’t it?”

"That’s right." I assured him. "So, if the square root of \(+9\) can be \(+3\) or \(-3\), what is the square root of \(-9\)?"

"Didn’t we learn that you can’t actually have a square root of a minus number?"

"Maybe, but let’s recap a little. Before we knew about irrational numbers, no such number as \(\pi\) or root two seemed possible. Before we knew about negative numbers, we were told we couldn’t subtract 8 from 3. Later, we accepted the existence of irrational and negative numbers, and these impossible numbers became possible. So let’s imagine there
Imaginary Numbers are a Cinch

are roots of negative numbers."

"You mean we just accept them, and then learn to use them." I could see his interest growing; he wanted to find out where imaginary numbers would come from.

"What we do," I went on, "is to write the letter j and assign it the mean-
ing square root of -1. All along we have believed there isn't such a thing. Now we imagine there is, although all we know so far is that we just gave it a name: square root of -1. But from that very fact, we know that squaring it will make -1."

"That's right enough, but aren't we going round in circles?"

"We soon will be," I replied, "but not in the way you think."

I showed him that, just as the square root of +1 is either plus 1 or -1, the square root of -1 can have two signs, +j or -j. The square root of -9, for example, is either +3j or -3j.

"It doesn't make sense yet," said George, "but go on, 'cause it's getting interesting."

"Remember that minus times a minus makes a plus," I said, drawing out vectors to illustrate (Fig. 2-a). "A minus reverses direction and the second minus reverses it again, bringing us to the original direction. With j numbers, multiplying j by j brings us to negative. As a math teacher would say, 'By definition, j times j makes a minus.' So, if negative represents reversal, what does the j sign mean?"

"From the way you twiddled your pencil." George responded. "I'd say it could mean halfway to reversal, or 90°. Is that it?"

"You've spotted it, George." I said, sketching a vector representation (Fig. 2-b) for him. "The vector diagrams illustrate..."

"Hey, I begin to see daylight," he interrupted me. "It's a way of writing quadrature in math symbols, without spelling it out. But does it make the calculations easier too?"

"Yes, the j tells you what to do with the number that follows it, just like plus and minus signs have been doing."

"Just a minute, why? What does it stand for? Wouldn't i be better—for imaginary?"

"As a matter of fact, that's what mathematicians and physicists use. But in electrical and electronic work, i already stands for current—although I never knew why—so, to avoid confusion, electrical people started using the next letter of the alphabet—j—for root of -1."

"Is that what they call the operator j?" George again butted in.

"That's right."

"And operator j means the root of minus one?"

"Correct again."

"Well, I'll be a monkey's uncle!" George exclaimed. "I don't know whether that clarifies anything for me, but I've a suspicion it will."

"In modern math classes," I went on, "they use number lines, as I did to demonstrate the fractions in my first sketch. Lengths along the line represent numbers." And I showed him how the concept of negative numbers is developed on a number line, following the concept of reversal a step further.

"They didn't use number lines when I was in school," George commented. "Maybe things would have come easier if they had."

"The main thing to note," I continued, "is how you add numbers on a number line. Start the second number from where the first one finishes; the result, or sum, is at the end of the second line segment."

"Makes sense," said George. "In fact, it seems obvious."

"Do you think now you could add an imaginary number to a real number, either positive or negative?"

"Is it at right angles?"

"That's right. So we have a right-triangle vector addition." I went on, sketching Fig. 3 as I talked.

"Hey, that really is familiar, though the old right-triangle bit always gave me a headache!"

"Then how would you go about finding the total, or resultant vector?"

"Oh, that's old Pythag... what's his name? The sum of the squares on the other two sides—what did they call it, hypot... 2?" George wondered aloud.

"It's Pythagoras' theorem," I filled in, "and the side opposite the right angle is called the hypotenuse. I'll bet you've done quite a few exercises in school, squaring two sides and finding the square root of their sum for the result."

"Yes." George said, "we seemed to have endless exercises in that at technical school. Will imaginary numbers take the sweat out of all that?"

"Sure will. Right now, it may seem just another name for what you've done before. But as we move along, you'll find imaginary numbers lead to more and more shortcuts, making calculations easier."

"That's for me," declared George.

"Now to complete the picture. For this one vector in my sketch, do you know how to find the phase angle?"

"You have me there. It has something to do with the ratio of two sides— is that right?"

I nodded and he went on, "I had some trig. It's called the sine or cosine or something like that, isn't it?"

"The sine and cosine are two other ratios," I told him. "This one's the tangent. It's the side opposite the phase angle, divided by the side adjacent." And I sketched it (Fig. 4).

Fig. 2-a—Minus times minus makes plus. h—j denotes only 90° change in phase.

Fig. 3—A complex number is a combination of a real and an imaginary number. It is shown graphically as a vector quantity, the result of two other numbers.
"The one opposite is the imaginary part, while the one adjacent is the real part. Now, to every angle there is just one tangent ratio, and to every tangent ratio there is just one angle. So we can find the angle from the tangent ratio, or the tangent ratio from the angle—either way, so long as we have one of them to start with."

George said, "Let's get this straight: I know the ratio between imaginary and real parts, so I look up the tangent of this ratio to find the angle. Is that right?"

"Ordinarily, it's the other way round. If you knew the angle in degrees, you'd look up the tangent of the angle to find the ratio between the imaginary and real parts. Here, you know the ratio, so you look up the arctan to find the angle. As we don't happen to have arctan tables, we can use the tangent table backward, or a slide rule."

"Let's take something practical and relatively simple, like an inductance with some series resistance. The real part of its impedance, producing in-phase voltage and current, is the coil resistance. The imaginary part, producing quadrature voltage and current, is the inductive reactance of the coil. Do you remember inductive reactance?"

"Let's see," pondered George, "it depends on L—the inductance in henrys—and there's a 2-pi-j in it, isn't there?"

"When we're using it a lot, it's much easier to write a lower-case omega for 2-pi-j. I wrote this equation on our scrap paper (Fig. 5). "Now, does inductive reactance get bigger or smaller as we increase frequency?"

"Isn't inductive reactance directly proportional to frequency?"

"Right. If L is in henrys, and ω is in hertz and omega is 2-pi-j, then inductive reactance is simply omega times L..."

"That does make it look simple," commented George.

"A reactive voltage is in quadrature with current," I went on. "so that means we need a j to represent the value properly. If the coil resistance is R ohms and the inductance is L henrys, the complete expression for impedance is Z = R + jωL. By multiplying various values of ω by 2π to get omega, we can make this expression tell the whole story, at all frequencies. Now what's the numerical value, or magnitude, of this expression?"

"From what we were saying earlier," George replied, "you'd square R and square omega-L, add the two squares together and then take the square root."

"You're catching on fast. Now what's the phase angle?"

"That's not so easy, but I know it has to do with the ratio between R and omega-L," he replied.

"Let's put in some numbers, and see how this works," I suggested, turning again to Fig. 5. "Suppose a 10-mH coil has a resistance of 50 ohms. Let's figure its impedance and phase at a few frequencies. To start with, take 1,000 Hz. First, what is omega at 1,000 Hz?"

"Well, 2-pi is 6.283, so 1,000 Hz makes omega 6,283."

"Correct. So what is omega-L at 1,000 Hz?"

"10 mH is one-hundredth of a henry," George mused, and doodled a moment, "so omega-L comes out to 62.83, right?"

"I showed him how to use his slide rule to sum the squares and take the square root at that sum, getting 80.3 ohms as the impedance Z. I asked him about the phase angle. He had the idea, but needed to be shown how to do it."

He solved the ratio of imaginary to real, which was 1.257, and I showed him on the slide rule the angle—about 51°. Then he took the slide rule and calculated impedance and phase angle for several other frequencies, while I watched. His results are shown in the table. He was thrilled at how simple it seemed.

Then I threw in the notion of using admittance values instead of impedance values for parallel combinations whose phase is in quadrature. He wanted to know why. He'd always preferred thinking in impedances.

I showed him that, with a specific voltage, he'd have to divide by the complex number representing impedance to figure out current. With admittance, he could simply multiply.

"Is that so much easier?" he wanted to know.

Just then the evening whistle blew. George and I both had to go."

"If you don't have too much work ahead of you," George asked, "can you drop by another time soon and finish telling me that bit?"

"Sure thing," I replied.

TO BE CONTINUED
Troubleshooting Chroma with VTVM

It's not always what service instruments you have, but how well you know them and how skillfully and carefully you can use each of them.

By CARL H. BABCOKE

THE OSCILLOSCOPE HAS BEEN REPRESENTED as the perfect TV servicing instrument. Leaders in the electronics servicing industry have persuaded and cajoled technicians to increase their skill in using it. In many circumstances, the scope is certainly useful. In other troubleshooting situations, however, a scope can waste valuable time.

When you test a chroma circuit, does your scope light up in beautiful green letters that say, "Secondary of the burst transformer is open?" Mine does not. Instead, the corner of the pattern tilts up in a sardonic leer as if to say, "This is the voltage and waveform at the end of my low-capacitance probe. Interpret it if you can." As can any skilled scope jockey, I eventually arrive at a correct diagnosis, but only after using a vtvm to pinpoint the defective part or parts.

Many chroma defects exhibit visual symptoms on the picture tube that suggest the general trouble area, thus eliminating that part of the troubleshooting usually reserved for the scope. These symptoms are half of the chroma troubleshooting technique to be described here. [The same is true for black-and-white.—Editor] The other half is multiplying the usefulness of the lowly vtvm by minor probe modifications.

A basic problem in obtaining accurate readings with any test instrument is that the device itself may radically change the circuit being tested. Vtvm's are recommended over von's because of higher input impedance (especially on the low scales, where it is the most important). But a dc vtvm can still upset a tuned circuit and make it oscillate, or detune it, because of capacitance in the probe.

There is a simple way to minimize the stray capacitance of any dc probe (see photo). The exposed resistor lead should be kept as short as possible. The 1-meg resistor decreases readings by about 10%, unless you use a probe without an internal resistor or slide the switch to the ac-ohms position to restore the original calibration.

In spite of the fact that the ac response of many vtvm's is flat to 3 MHz or higher, any attempt to measure the signal at the chroma-bandpass plate or at the grid of a 3.58-MHz oscillator is doomed to failure. The combined capacitance of the probe, cable and meter will kill the signal. This occurs because the total meter capacitance is approximately 60 pF, equal to about 1,000 ohms of capacitive reactance at 3.58 MHz.

The situation is not completely discouraging, however. The low-C probe used with most scopes has a capacitance of about 11 pF and can be used with the vtvm (Fig. 1). Of course, it no longer will give an exact 10-to-1 reduction in reading, so you'll have to check the loss ratio of your own meter and probe, for accuracy.

To do this, measure the 6.3-volt heater supply with the probe in the direct position and record the peak-to-peak reading. Change the probe switch to low cap X10, measure again and record the second reading. Divide the second reading into the first, and you have the ratio at 60 Hz. Now find a low-impedance point in a 3.58-MHz oscillator (for example, secondary terminal C of the oscillator plate transformer, as
indicated in Table 1) and measure this in the direct and low-cap positions. Calculate the ratio for the higher frequency. The two ratios differed in my own calibration, so I used the one for 3.58 MHz—approximately 12.

Even with this relatively high level of attenuation, the low-capacitance probe will permit us to read virtually any signal in the entire chroma circuit, even at the bandpass grid. Where the signal typically would be about 12 volts peak to peak.

Fig. 2 is a block diagram of chroma circuits in an RCA CTC-16X color chassis. The circles contain numbers corresponding to measurement conditions given in Table I. Let's use these two illustrations to find out how much of the circuit can be investigated with a vtvm compared to a scope. (A color-bar generator was used for a signal source.) Table I gives the circuit test point, the readings with the vtvm (using two kinds of probes) and a reading with a calibrated scope (using low-C probe). Any change in color-bar display when a probe was touched to the circuit is also listed—as a guide to the loading effect of the testing equipment.

Several conclusions can be drawn from the chart. The first is that the vtvm plus the low-capacitance probe will give fast and fairly accurate readings on any circuit where a good scope could be used. Furthermore, the vtvm will not change receiver performance any more than the scope would. An unexpected bonus is the news that the existing probe can be used with fair accuracy on 14 out of the 20 tests. Only one chroma voltage was too small to read on the low-capacitance probe—the grid of the G-Y amplifier—and it has only some crosstalk on it and is a point seldom measured.

One more hint if you prefer to use the normal vtvm probe: Almost any circuit that's detuned badly by a probe has another branch that's not so affected. The bandpass-amplifier plate-transformer primary is greatly changed by probe capacitance, while the secondary (also the top of the color control) is virtually unchanged. The oscillator plate transformer is a similar example. While primary measurement with a normal probe may kill all color, the secondary (the suppressor grid of the X demodulator) can easily be measured. This will tell you if the oscillator is running and has enough output.

To avoid the bother of changing probes, just keep a scope probe on the vtvm at all times and use the switch for normal or low-cap function. For dc readings attach the external 1-meg resistor to the tip, as described earlier.

Regardless of the probe used, a 1-meg resistor on the probe tip is a must for fast, efficient measurements.

---

**Table 1**

<table>
<thead>
<tr>
<th>Circuit Position</th>
<th>Vtvm normal probe</th>
<th>Vtvm low-C probe actual</th>
<th>Vtvm low-C probe corrected</th>
<th>Scope low-C probe</th>
</tr>
</thead>
<tbody>
<tr>
<td>1. Chroma input coil</td>
<td>4.5 changes phase much less color</td>
<td>2.4 more color, changes phase</td>
<td>2.6 more color, changes phase</td>
<td></td>
</tr>
<tr>
<td>2. Bandpass grid</td>
<td>6.0 less color</td>
<td>1.03 slight change</td>
<td>13 slight change</td>
<td></td>
</tr>
<tr>
<td>3. Bandpass plate</td>
<td>11 much less color, much phase change</td>
<td>1.6 no change</td>
<td>18 no change</td>
<td></td>
</tr>
<tr>
<td>4. Top of color control</td>
<td>18.5 slight phase change</td>
<td>0.8 no change</td>
<td>12 no change</td>
<td></td>
</tr>
<tr>
<td>5. Arm of color control</td>
<td>10 slight phase change</td>
<td>0.6 no change</td>
<td>10 no change</td>
<td></td>
</tr>
<tr>
<td>6. Plate pin 6 of X demod</td>
<td>18 phase change</td>
<td>22 no change</td>
<td>26 no change</td>
<td></td>
</tr>
<tr>
<td>7. Supp. pin 7 of X demod</td>
<td>21 no change</td>
<td>26 no change</td>
<td>26 no change</td>
<td></td>
</tr>
<tr>
<td>8. Plate pin 6 of Z demod</td>
<td>26 phase change</td>
<td>30 no change</td>
<td>38 no change</td>
<td></td>
</tr>
<tr>
<td>9. Supp. pin 7 of Z demod</td>
<td>20.5 blue to right</td>
<td>2.6 slight phase change</td>
<td>25 slight phase change</td>
<td></td>
</tr>
<tr>
<td>10. Grid pin 2 R-Y ampl</td>
<td>17 slight phase change</td>
<td>18 no change</td>
<td>21 no change</td>
<td></td>
</tr>
<tr>
<td>11. Grid pin 7 B-Y ampl</td>
<td>24.5 slight phase change</td>
<td>25 no change</td>
<td>30 no change</td>
<td></td>
</tr>
<tr>
<td>12. Grid pin 7 G-Y ampl</td>
<td>3.6 no phase change</td>
<td>(can't read) no change</td>
<td>5 no change</td>
<td></td>
</tr>
<tr>
<td>13. Plate pin 1 R-Y ampl</td>
<td>200 blanking &amp; chroma 115 no color</td>
<td>22 both 264 no color</td>
<td>20 color 110 blanking no change</td>
<td></td>
</tr>
<tr>
<td>14. Plate pin 6 B-Y ampl</td>
<td>220 both 110 no color</td>
<td>24 both 288 no color</td>
<td>260 color 210 blanking no change</td>
<td></td>
</tr>
<tr>
<td>15. Plate pin 6 G-Y ampl</td>
<td>145 both 120 no color</td>
<td>14.5 both 174 no color</td>
<td>75 color 120 blanking no change</td>
<td></td>
</tr>
<tr>
<td>16. -Y amplifiers common cathode</td>
<td>10.8 both 9.8 no color</td>
<td>0.9 both 10.8 no color</td>
<td>6 color 9.5 blanking no change</td>
<td></td>
</tr>
<tr>
<td>17. Oscillator grid pin 2</td>
<td>12 far off freq.</td>
<td>1.2 slightly off freq.</td>
<td>14 slightly off freq.</td>
<td></td>
</tr>
<tr>
<td>18. Oscillator plate pin 6</td>
<td>8 no color</td>
<td>5.6 little color</td>
<td>50 little color</td>
<td></td>
</tr>
<tr>
<td>19. Oscillator screen pin 3</td>
<td>zero no color</td>
<td>0.42 no change</td>
<td>5 no change</td>
<td></td>
</tr>
<tr>
<td>20. Osc. plate transformer terminal C</td>
<td>21 no change</td>
<td>22 no change</td>
<td>21 no change</td>
<td></td>
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</table>

**Table 2**

<table>
<thead>
<tr>
<th>Test point</th>
<th>Black-and-white program</th>
<th>No antenna, no signal</th>
<th>100 Percent color bars</th>
</tr>
</thead>
<tbody>
<tr>
<td>6JUB pin 1 (phase det plate)</td>
<td>15</td>
<td>26.5</td>
<td>-56</td>
</tr>
<tr>
<td>6JUB pin 3 (phase det cath)</td>
<td>15</td>
<td>29</td>
<td>-56</td>
</tr>
<tr>
<td>Junction of two matched resistors</td>
<td>0.05</td>
<td>-0.3</td>
<td>+0.05</td>
</tr>
<tr>
<td>6JUB pin 7 (killed det plate)</td>
<td>15</td>
<td>-27.5</td>
<td>-55</td>
</tr>
<tr>
<td>6JUB pin 9 (killed cath)</td>
<td>13</td>
<td>-14</td>
<td>+47</td>
</tr>
<tr>
<td>Junction of two matched resistors</td>
<td>-0.83</td>
<td>-1.0</td>
<td>-3.3</td>
</tr>
</tbody>
</table>

*With color-killer control correctly set

**Note:** Pin 1 had -8.5 volts with burst tube removed.
Troubleshooting Chroma with VTVM

where grids and diodes in the circuit are actually built-in ac rectifiers. The 6JU8 phase and killer detectors shown in Fig.
3 are prime examples. Let's measure the dc voltages on pins 1, 3, 7 and 9 under several signal conditions and record them in the chart of Table II. Where do these voltages come from? On color bars from burst rectification. on black-and-white from video harmonics above 3 MHz and on snow from random-noise harmonics.

But that's not all. With the burst tube removed from its socket, there is still -8.5 volts on pin 1. Where did this voltage come from? From a source we may have neglected. the sample of 3.58 MHz from the oscillator applied to the plate and cathode of pin 2. This 3.58 signal is compared against the burst and, if the phase or frequency of the two are different, error dc voltage is formed at the junction of the two matched 1-meg resistors (point A). This error voltage controls the reactance tube and thus the 3.58-MHz oscillator.

At the moment we are only concerned with the diode voltage, and there are two important things we can learn by measuring the voltage on plate pin 1. If the voltage is -40 or more we can conclude there is burst present. However, this does not prove the oscillator signal is there. Disable the burst stage, and the voltage should be -8 to -9 if the oscillator is running.

Now let's see how we can use several of these techniques to make a fast diagnosis on a colorless color set. Tune the receiver to a color-bar pattern from a generator, with the sound carrier switched on at the generator. Tune into the heat pattern on the faint vertical bars that mark the proper color-bar position. Then back off the smallest amount necessary to eliminate the heat. This will assure that the fine tuning is correctly set to receive color. Preset the color control almost full up and the tint control to the center of its range. Now measure the voltage (with external resistor on the probe) at pin 1 of the 6JU8. Less than -30 volts indicates there is no burst coming from the burst keyer, but -40 or more is normal for good burst.

You will notice in Fig. 2 that the bandpass amplifier (color) and burst keyer stages are nearly in parallel—they share a common input from the takeoff coil. Therefore, if we have good burst to the 6JU8, it is nearly 100% certain we have chrominance sideband information at the grid circuit of the bandpass amplifier. Since there are two demodulators and three -Y amplifiers, a complete loss of color is most likely to be in the bandpass amplifier.

To check this the easy way, we use the normal ac probe at the top end of the color control. Any voltage above about 10 p-p should be sufficient for good color, and we would go next to the demodulator stages. But just suppose we measured no ac voltage on the color control. This positively isolates the trouble to the bandpass amplifier. A bad tube or wrong plate, screen or cathode voltage can stop the color, but the best bet is the grid voltage because of the color-killer connection. Fig. 4 is a simplified diagram of this grid circuit including a very important test point. The photo shows the physical location. Negative 8 or 9 volts is enough to cut off the 6CH8. As a rule, -3 volts or less will permit normal amplification while -8 or more indicates the tube is biased for no gain.

A shortcut here is to ground resistor R703 or capacitor C702 with a clip lead and see if color of any kind appears on the screen. The color may be normal and stable if the sole defect is in the killer stage or its adjustment. More likely is the possibility of seeing out-of-lock color stripes because the killer is noise-sensitive. It biases the bandpass amplifier to cutoff unless the burst and local 3.58 signal are phase-locked. In this event, correct the poor locking and check tint range. Finally, remove the ground and set the color killer.

Should the burst-keyer and bandpass-amplifier stages check normal in this series of tests, we would check the 3.58 oscillator by one of two methods. The low-capacitance probe could be applied to pin 7—the suppressor of the X demodulator—where we would normally expect 2 volts p-p (34 volts p-p corrected). Or we could remove the burst-keyer tube and check the dc voltage at pin 1 of the 6JU8, expecting to find -8 to -9 if the oscillator is normal or zero if it's dead.

The demodulators and -Y amplifiers can be checked similarly, using ac and dc tests alternately, to pinpoint any possible loss of color there.

All kidding aside, a scope is a fine instrument to have and use effectively, but the familiar vtvm with its three kinds of probes has the advantage in two important characteristics: speed and accuracy of measurement. A calibrated voltage-reading scope is frustrating to use with calibration to check, centering controls to adjust, scales to select and finally a reading to be made by squinting and wondering if the trace is lined up right. By comparison, a vtvm reading is quick and easy. A scale is selected, the pointer rises to a definite point on the scale, where you find large, easy-to-read numbers.

Try it yourself, and you will probably use the vtvm for 90% of your chroma testing.

END
A Mechanical Look at Tape Recorder Servicing

Get to know the nonelectronic side of this branch of service work and you'll enlarge your troubleshooting ability

By JOHN MOHAN

THERE'S MORE TO HOME-ENTERTAINMENT SERVICING THESE DAYS THAN SIMPLE ELECTRONIC REPAIR WORK. THE MODERN TECHNICIAN IS CONFRONTED DAILY WITH MECHANICAL DEVICES OF ALL DESCRIPTIONS—motorized TV tuners, pushbutton auto radios, record changers, and the ever-popular tape recorder.

This latter product is attaining huge numbers, and the troubles in them are as often mechanical as they are electronic. But your customer expects you to handle all the problems in his home-entertainment equipment. It matters not if your training was electronic, not mechanical.

How can you meet this challenge? . . . Here, I feel are the best steps: (1) Study the mechanism. (2) Observe its actions. (3) Try to visualize correct operations; then (4) figure out why's of incorrect operation. In brief, you must understand the mechanism if you are to solve its mechanical problems. Good facilities such as lighting, tools, lubricants and cleaners, and reliable service literature (many hints are given in service data) are important aids.

And please note—use care . . . don't start by bending mechanisms or forcing moving parts until something breaks! Something we might call "courage over caution" causes parts to be bent or distorted to achieve what appears to be a repair but invariably results in more problems. Careless use of lubricants can also compound mechanical problems and result in annoying callbacks.

Each mechanism—whether it's the gear train of a television tuner, the change slide of a record player, or the mixed-up inside of a tape recorder—must be treated separately, because each accomplishes its mechanical results in a different way.

Tape-recorder drive systems are of four types. Although the basic principles of operation are similar, the manner in which the driving force is delivered to the moving parts differs.

A BELT DRIVE

Fig. 1 illustrates the essential elements of a typical belt-drive system. In spite of its simplicity, this system works well. The chassis plate and control knobs or pushbuttons and associated linkages are omitted from the sketch for clarity. The mechanism is shown in Fig. 1-a in the play position. (The only difference between this and the record position is in how the amplifier circuits are switched.)

One drive belt couples the motor to the flywheel, and another couples the flywheel assembly to the takeup reel. The pressure roller holds the tape against the capstan shaft, which pulls the tape past play/record heads.

The takeup reel, also driven by a belt, winds up the tape. Excess slack is removed from the drive belt by an idler pulley, but enough is left to allow slight slippage—so the tape won't be stretched too taut. The supply reel feeds the tape and turns counterclockwise solely from the pull of the tape. Uniform tape speed past the heads is maintained because the capstan shaft actually controls the travel speed of the tape. The capstan shaft is maintained at constant speed by the inertia of the flywheel and a governor in the motor. In this machine, a choice of tape speeds is provided by an accessory sleeve which can be placed on the capstan shaft to change its diameter.

When this recorder is in the rewind position, the mechanical drive conditions change. Fig. 1-b shows how rewind is accomplished. The pressure roller is pulled away from the capstan shaft. The idler pulley is pushed back, leaving the takeup belt slack. The pulley coupled to the supply reel is pushed into contact with the flywheel assembly. Under these conditions, the capstan and pressure roller no longer provide the driving force for the tape; instead, the supply reel is now the primary drive for the tape, being driven in a clockwise direction by the flywheel assembly. The takeup reel also is pulled clockwise (opposing the very slight force of the slack drive belt on the takeup reel).

Now consider mechanical problems which could arise. Suppose that in

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**TAPE-TRANSPORT SERVICING HINTS**

1. Equip your service bench with the necessary tools and facilities.
2. Figure out how the mechanism operates.
3. Be sure all driving surfaces are clean (use alcohol) and free of oil or grease.
4. Avoid excessive lubrication. Use only recommended lubricants.
5. Don't bend or distort mechanical parts.
6. Refer to service literature for guidance in adjustment and servicing.
7. Be sure the electronic performance is good before suspecting the mechanism.
8. Operate this machine for a reasonable period of time to check your repair.
9. Test all modes of operation (at all speeds).

---

**Fig. 1—In a simple belt-drive tape recorder the basic mechanism is not so complicated.**
A Mechanical Look at Tape Recorder Servicing

the PLAY or RECORD position the tape piles up at the heads, failing to accumulate properly on the takeup reel. Analyze Fig. 1. The takeup reel is obviously not receiving drive enough to wind the tape as it leaves the capstan. Dirt or grease on the takeup-reel drive belt might cause this. Or the idler may not be keeping the belt tight enough—check the control linkage that actuates the idler. Another check would be for excessive friction which may prevent the takeup reel from turning freely.

Failure to rewind properly suggests inspection of the supply reel and its drive belt, which might be broken, slipped off, oily, or too loose. Check the pulley that drives the supply reel for freedom of motion. Make sure it is in good contact with the flywheel assembly.

Uneven tape speed when recording or playing back requires a check of the motor, the power source (batteries in some cases), the drive belt from the motor to the flywheel, and the capstan shaft and pressure roller. Dirt or oil on the drive belt or pressure roller, or too little pressure between roller and shaft, could cause the trouble. If a sleeve is used on the capstan, the sleeve may be loose or improperly installed.

By understanding how the mechanism operates normally, you will find it is easy to analyze where a difficulty might originate.

A rim drive

Another tape-recorder mechanism variation is the rim drive. Power is transferred through rim-to-rim contact of the various rotating parts.

Fig. 2a illustrates the basic components of a rim-drive mechanism in the PLAY or RECORD position. The horizontal motor is free to pivot in a vertical plane. By spring action, the motor shaft contacts the underside of the flywheel, driving it at the rim. An idler wheel contacts the outer rim of the flywheel and also the rim of the takeup reel. The supply reel is free to turn since it is disengaged from the mechanism. The main drive for the tape is again provided by the "pinching" action of the pressure roller and capstan shaft. The takeup reel in two parts: the upper, or turntable, is coupled to the lower, or drive, by a slip clutch. The takeup reel winds up the tape as it leaves the capstan. The slip clutch prevents the takeup reel from upsetting the established constant tape speed by preventing the reel from pulling too hard on the tape as it comes through the capstan shaft and pressure roller. Inertia of the flywheel and controlled motor speed insure constant-speed tape travel. The supply reel simply gives up the tape, rotated counterclockwise by the pull of the tape.

In the rewind position, the drive conditions change as shown in Fig. 2b. The idler that furnished drive to the takeup reel is disengaged, and the pressure roller is separated from the capstan. The supply reel becomes the driving force, because it is pressed into rim-to-rim contact with the flywheel. The tape travels from right to left, accumulating at a fast rate on the supply reel.

The service approach for this mechanism is also based on understanding its normal operation, then determining where trouble could arise. Take special care to clean the driving-rim surfaces with alcohol. Make sure good contact is being made, and that rotating parts are free.

A reel-to-reel transport

A study of the seemingly more complex mechanism of Figs. 3-a and 3-b will show that basic operation and the servicing approach are similar to other mechanisms.

This recorder provides a FAST FORWARD in addition to the usual PLAY/RECORD and REWIND. The PLAY and RECORD positions are shown in Fig. 3-a. The pressure roller is against the capstan shaft. Pressure pads also hold the tape against the play/record head and the tape guide. The motor drives the flywheel (and capstan shaft) through a belt. Another drive belt runs from the motor to the takeup reel. Light drive pressure is applied to the takeup pulley by its clutch assembly. As the tape is propelled by the capstan shaft and pressure roller, the right-hand (takeup) turntable winds up the tape but does not exert excessive pull on the tape coming through the capstan shaft and pressure roller because of the slippage inherent in the takeup clutch.

Both the right and left turntables have brakes, but both brakes are disengaged in the PLAY and RECORD modes. By a special system actuated by the counterclockwise motion of the left-hand turntable only the left (supply-reel) brake is applied when the mechanism is stopped, thus avoiding tape slippage due to coasting.

A PAUSE function, when actuated, simply moves the pressure roller slightly away from the capstan shaft and applies the left-hand brake. This stops the movement of tape. Takeup clutch slippage is continuous during the pause.

In the FAST FORWARD position (Fig. 3-b), the pressure roller is moved away from the capstan shaft, and the pressure pads are released. Strong pressure is applied to the takeup pulley clutch, and the takeup reel becomes the primary driv-
ing force for the tape, winding it up at high speed. The left-hand reel simply feeds out the tape as it is pulled by the takeup reel. When the mechanism is stopped, only the left brake is applied to prevent tape spillage.

Conditions in the rewind position change very much as in other systems. A rewind pulley which is coupled to the motor by a small belt comes into contact with the rim of the left-hand supply turntable. This turntable becomes the primary driving force, rotating clockwise. The tape travels rapidly from right to left, winding up on the left-hand reel. When the mechanism is stopped at the end of rewind, the brake system applies only the right-hand brake, and spillage is again prevented.

The servicing procedure, once normal operation is understood, is to associate failure with function. Then check to see why the mechanism does not perform. Give extra attention to clutch pressures, brake action and brake pressures, and the usual cleaning and lubrication aspects.

A tape-cartridge transport

This mechanism handles preloaded two-hub tape cartridges. The same basic elements of any transport are found in the tape-cartridge machine. The top deck of this type mechanism is dimensioned and shaped so that the tape cartridge drops into place, engaging the hubs of the cartridge with the turntables of the mechanism and aligning the tape so it fits between the capstan and pressure roller.

Fig. 4-a shows a partial view of a tape-cartridge mechanism in play position. Driving force is transmitted from the motor to the flywheel (and capstan shaft) by a drive belt. The same belt rotates the takeup turntable. The tape is pulled across the play/record head in the usual way by the pressure roller and capstan. The takeup turntable winds the tape as it leaves the capstan. Here again, a clutch permits the tape to wind up without upsetting the tape speed established by the capstan. Pressure pads hold the tape against the play/record head. (NOTE: In this system, the tape heads face toward the turntables, so the oxide side of the tape faces outward in the tape cartridge.)

When the tape-cartridge mechanism is in the rewind position (Fig. 4-b) the idler pulley pushes the drive belt away from the takeup pulley and against the rewind pulley. The flywheel and capstan turn, but the pressure roller is away from the capstan, so the tape can travel freely from right to left. The rewind pulley turns in the direction shown in Fig. 4-b. Tape accumulates on the supply hub of the cartridge.

A fast forward is also provided by releasing the pinch action of the pressure roller and capstan shaft, releasing the brake, and tightening the clutch pressure of the takeup pulley. The right-hand turntable then becomes the driving force, winding the tape at high speed onto the takeup hub of the cartridge.

As in the other mechanisms described, the service procedure is to study normal operation, then consider the various causes of trouble which could prevent proper operation. Through cleaning and proper lubrication are vital. Also check the condition of the drive belt, pressure roller, and driving surfaces. Binding within the tape cartridge itself should be considered as a possible trouble.

Again let me stress that the few minutes required to study a tape mechanism and learn how it operates is time well spent. After gaining some experience in mechanical servicing, you can find it as interesting and rewarding as electronics.
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MAY 1967
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Everything electrical has a watt (power) rating. This goes for hi-fi components, too, whether stereo or mono. How many or how much you need depends to a large extent on your listening area and its acoustical conditions. A room with thick carpeting, heavy drapes and overstuffed furniture absorbs a great deal of sound. For adequate listening levels, such a room will require more amplifier power (watts) to the loudspeakers than would a room with hard surfaces, little drapery and modern furniture. The same is true of big, open rooms vs. small, compact rooms.

At maximum volume (watts) some amplifiers may tend to develop distortion. Loudspeakers will simply reproduce any distortion along with the high fidelity music. So, if your components are used in a big or "overstuffed" room, make certain the amplifier has sufficient wattage.

To be sure of your requirements, ask the expert—your Jensen dealer. He'll be glad to help plan your hi-fi system. He will also demonstrate Jensen loudspeakers—how they preserve amplifier watts and fidelity.

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EQUIPMENT REPORT

Heathkit GR-180 Color TV Receiver

Circle 26 on reader's service card

You sit down and thumb through the booklet. Lists of parts, pictorial sketches of printed boards, columns and columns of instructions, pages with colored pictures to show how your GR-180 color TV set should look when you get it built. Even hints and charts that show what to do if it doesn't work the first time you turn it on. You get the impression, too, that here's a lot of work.

Well, you're anxious to watch it and sitting there admiring the manual won't get anything built. Out comes more packing, a bunch of smaller boxes, metal plates, a plastic facia panel.

This latest color TV creation from Heath is a 19-inch set (called 180 square inches now to please the FTC). Smaller and less expensive than the GR-53 series or the recent 25-inch (oops, excuse me—295-square-inch) model GRA-295, the new GR-180 is nevertheless filled with virtually the same circuits used in the larger models, including the automatic degaussing system of the 25-inch model.

You tape the big folded schematic to the wall, not because it will help you build the set but so you can study it while you rest between sessions of building. And it's worth getting acquainted with. Not that it's much different from the earlier models, but you don't recall much about them, either.

Let's see . . . X and Y demodulation, with separate red, green and blue amplifiers feeding the picture-tube grids. A video cathode follower precedes the high-gain pentode video (Y-channel) amplifier. Three high-gain i.f. stages, with the first two stacked—a well-designed i.f. strip with four traps to shape the response precisely. Sync and sound have their own separate detector diode (not the video detector) and are amplified together in one sound-i.f. stage before the sync is taken off for the 6HS8 sync circuit. Shunt HV regulator, with the usual 6BK4. Nothing very unusual, but really designed for performance.

You like to build kits, so the job is fascinatingly simple though time-consuming. It would seem monotonous at times, were you not so anxious to get it finished and see the result. You mount and solder all the small parts onto the printed boards. Bolt together the larger parts and subassemblies. Install the wiring harness.

The whole job of putting together
the chassis is simplified by an already-wired i.f. board and high-voltage cage. The chassis mounts vertically around the picture-tube neck and swings out on hinges from a cabinetlike metal "shield." Once all the hardware is in place, you set the chassis aside, open the other box and set about putting the picture tube and the plastic faceplate together. Then the deflection-yoke and convergence-yoke assemblies. Bolt the faceplate, CRT, chassis "shield" and chassis all together and now you’re ready to turn the set on.

This is one of those rare times when you’ve done everything exactly by the instructions and the GR-180 works the first time you try it. Well... it shows three images, but the book says that’s normal until you converge the beams.

Convergence turns out to be surprisingly simple. The instructions are easy to understand and are illustrated. Using the built-in dot generator (see pattern in photo), you get the first part done in short order—5 minutes for static convergence of the center dot. Next you do purity, then gray-scale adjustments.

Finally, what is usually the toughest part of all: dynamic convergence. This time it seems so simple. Everything just slides into place. You follow the book step by step, and suddenly the GR-180 is converged. You tune in a station and the black-and-white picture looks great.

Well, it’s a color set, so you try a color program. Odd colors? You twist the phase coil just a little with the alignment stick furnished (the same one you used for convergence) and now the TINT control will get those flesh tones.

Gee, a kit without any problems? You wonder. Aha!... the picture looks funny. It seems grainy. Takes a while, but you finally notice the graininess can’t be seen from a few feet away. Closer examination solves the riddle. The raster lines are forming an optical moire with the triad dot pattern of the small picture tube. Natural, you conclude.

Better let it cook out. For hours it runs, and runs, and runs. Colors are bright, black and white is pure.

You’ve run it now for a few weeks and played with it, learning all you can about how it operates—getting the "feel" of it. So far it’s still in solid convergence and showing no signs of any kind of trouble. You like it, don’t you?—Jeff Tracy

Price: $379.95

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MAY 1967

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CAPACITANCE SWITCHES, the TSM-114B and the TSM-94B. Dynaquad touch control switches operate either as contact or proximity switches, depending on sensitivity setting. Stray rf signals will not cause false firing. Stable operation over temperature range of -20° to +170°F. Basic mechanism is relaxation oscillator with matched pair of transistors. Control 100-W lamp load. Handle larger external loads with external switching pro-

vion. Operate on 105-125 Vac and 45-100 Hz. Both trigger on 1pF capacitance change. Sensitivity level of TSM-114B is adjustable from 6-25 pF. It operates as a momentary switch. The TSM-94B has a sensitivity level adjustable from 25-175 pF and operates as a momentary or a latching switch. Plug-in connections. 3 x 2 x 1½ in.—Tung-Sol Div., Wagner Electric Corp.

Circle 46 on reader’s service card

SELF-CONTAINED BATTERY PACK, the Port-A-Pak, model CAP-1. When these two units are used together, CB rig can be used as a base station. Collapsible antenna, Teflon case, battery meter, charging connector, mounting hardware, microphone bracket and shoulder strap. Port-A-Pak, $80.95; Charge-A-Pak, $12.95.—Courier Communications, Inc.

Circle 47 on reader’s service card

TRANSISTOR TESTER AND SET ANALYZER, model 1000. Self-powered, portable. Checks all transistors for gain, leakage and oscillating ability, in or out of circuit. Determines transistor pnp or nnp. Can be used as signal tracer. Checks battery voltage and current drain. Tests

The only Theatre Organ available in kit form — for only $1350.

You’ve asked for it and here it is — the NEW Schober THEATRE ORGAN that you assemble yourself. For the first time in kit form, a real Theatre Organ with that rich, full, old time theatre pipe organ sound. You create the organ, then you create the music!

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[Enclosed please find $3.00 for 12-inch qual-

ity LP record of Schober Organ music.

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NEW PRODUCTS continued

diods and rectifiers. Sockets mounted on face of instrument. Comes with Interchangeability Guide, which gives replacement information on approximately 5,000 semiconductors, and operation booklet. For limited time, also free transistor and pair of rubber-covered, color-coded needlepoint test prods. $34.95—Semtronics Corp.

Circle 48 on reader's service card


Circle 49 on reader's service card

23-CHANNEL SIDEBAND AND AM TRANSCIEVER, the Imperial. Receives on three modes in each of 23 CB channels, provides 69-signal reception capability. Transmits on 2 modes within each CB channel for 46-signal capability. Operator can conduct 2-way transmissions with receivers tuned to either upper or lower sideband. Receiver R/VFO control permits tuning to exact frequency of incoming signal. Double conversion is used with 23-channel synthesized circuit with 0.5 µV or less sensitivity. Transmitter is capable of 30-watt peak-to-peak modulation. "Pi-L" network built in for matching 30–75-ohm antenna loads. Illuminated channel-selector switch, 2-scale, 4-function S-meter. Final tune control. Adjustable squelch, volume control, and microphone plug. Operates on 120 Vac or 12 Vdc. Optional mounting bracket. $299.—Regency Electronics, Inc.

Circle 50 on reader's service card

MUTUAL-CONDUCTANCE TUBE TESTER, Mercury model 2000, for testing latest tube and transistor types. Tests new magnoval, 7-pin inductors, 10-pin decals, transistors, diodes and power rectifiers. Tests for dynamic mutual conductance, gas and grid emission (sensitivity over 100 megahms) and for shorts and leakage between any tube elements. Tests color and black-and-white picture tubes with use of optional multthead adapter. Leatherette case.—Singer Products Co., Inc.

Circle 51 on reader's service card

50-W HI-Fi STEREO AMPLIFIER, Knight model KN-960. All silicon transis-tors. Separate on-off switch. Power: 50-W IHF; 100-W peak; 17-W rms per channel at 8 ohms, ± 1 dB, 20–20,000 Hz. Under 1% distortion. Outputs: 4 to 16 ohms, stereo headphones. 2 ac convenience out-
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lets. 22 semiconductors. For 110-130 V 60-Hz. 3.3 KΩ x 13 x 10 in. Price $99.95. Oiled walnut case, $14.95 if purchased separately: $1 if purchased with the amplifier.—Allied Radio Corp.

Circle 52 on reader’s service card

SOLID-STATE COLOR BAR GENERATOR, model GC660. Gun-killer controls for fast purity checks. 4.5-MHz crystal-controlled sound carrier for setting fine tuning. Stability over temperature range from -5°F to +120°F. Color bars: gated 10-bar pattern covers all color alignment signals. Level variable from 0% to 20% for checking color sync, ACC, etc. Chroma signal: crystal-controlled, provided at rf or video, controllable from 0% to 200% and referenced to zero. Crosshatch: available as crosshatch or 18 separate vertical or horizontal "thin-line" bars presented at 60-Hz rate. Dot pattern provides 324 dots of 0.1-angle duration. Video output: provided at 2V p-p with choice of positive or negative polarity. Special separate -2V p-p sync output for sets requiring this feature. RF output is tunable from front panel to channel 3, 4, or 6, matched to 300-ohm load. Operates from 105-125 V, 60 Hz, 3W, 10½ x 10⅛ x 5 in., 6 lb. Leatherette-covered portable case. $159.50.—Hickok Electrical Instrument Co.

Circle 53 on reader’s service card

BOOKSHELF-SIZE ALL-SILICON RECEIVER, model S-8600. Amplifier powers 2 pairs of speakers. Front-panel meter-action switches. Front-panel controls for tape monitoring, bass, treble, balance, loudness, tuning, FM/phone and mono-stereo. Stereo headphone jack. Power output: music power, 80 W at 4 ohms, 30 W at 8 ohms. Continuous power each channel: 30 W at 4 ohms, 20 W at 8 ohms for 1% distortion. FM (HIF) sensitivity: 0.1 μV for -30 dB noise and distortion below 100% mod. FM distortion: less than 0.25%. FM at 100% mod. FM hum and noise level: 70 dB below 100% mod. Frequency response: FM mono, 20-20,000 Hz ±3 db; FM stereo, 20-15,000 Hz ±5 dB; ampli-

plifier, 20-20,000 Hz ±5 dB at rated output. Outputs: 4 to 16 ohms, main and remote, left and right speakers, stereo headphone and record output. Power consumption: 115-125 V, 60 Hz, 30 to 120 W; fused. 35 silicon transistors, 4 silicon rectifiers, 9 silicon diodes, 1 Zenep diode. 16½ x 12 x 4½ in., 26 lb (with case). Chassis $299.50, with walnut-grained leatherette case, $298.50.—Sherwood Electronic Laboratories, Inc.

Circle 54 on reader’s service card

REPEATING MAGNETIC TAPE CARTRIDGE, the Audio Vendor, model 1510, for operation in small, battery-operated portable tape recorders. Works on standard "reel to reel" tape recorders which normally use 3- or 4-in. reels. Playing-time cycles (before repeating) of 6 or 10 min. at 15-ips tape speeds. Fits over feed spindle of almost any recorder.—Orrtronics, Inc.

Circle 55 on reader’s service card

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HEATH 1967
NEW PRODUCTS continued

usual "fall-off" of energy at extreme ends of polar pattern. New design gives extremely flat, wide-angle vertical dispersion. Mounts vertically or horizontally. Position easily changeable with Lever-Lock bracket. Bell made of Impex A with molded-in color impervious to weather conditions. Frequency response, 250-13,000 Hz; power rating, 30 W; dispersion angles, 120° x 60°; microphone sensitivity, -23 dBm.—Oxford Transducer Co.

Circle 56 on reader’s service card

STEREO HEADPHONE CONTROL CENTER, model PH-127. Permits use of 2 sets of stereo headphonest with separate volume control for each. 5-ft cable with 3-circuit plug connects to amplifiers. Plug is removable for connection to amplifiers without a phony micro. 5 1/2 x 3 1/2 in. $7.98.—Olson Electronics, Inc.

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SOLID-STATE CB SIGNAL GENERATOR, model 5404. For testing, servicing CB radio equipment. 24 crystal-controlled channels, 25 discrete CB frequencies. 1 spare. Modulation, from accurate internal 400-Hz oscillator, adjustable 0-100%. External modulation 0-100% with audio signals from 60 Hz to beyond 5,000 Hz. —Pace Communications Corp.

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Circle 113 on reader’s service card
splicing include No. 515 barrier strip terminal and No. 516 U-Grip terminal which presses into place, automatically strips away insulation and snakes the connection. 2- or 3-conductor widths in 100-ft rolls. 2-conductor roll starts at $6.10.—

3M Co.

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HEAVY-DUTY 10-IN. SPEAKER FOR GUITARS, ORGANs, model MI-10PC. Uses a 30-oz magnet and a 1½-in. voice coil. Each speaker handles up to 32 watts peak audio. Impedance 8 ohms. Rubberized cone edges, cotton-covered 66-strand voice coil. 3½ in. deep.—Utah Electronics

Circle 60 on reader’s service card

HIGH-CURRENT RECTIFIERS. Hi-Pac insulation process is incorporated into standard bridge rectifier assemblies

MAY 1967 87
NEW PRODUCTS continued

from 30-1,200 PIV and from 4-20 amp de output. Also high-voltage rectifiers from 1,600-80,000 PIV and 2-5 amp de half-wave output. Allows size reduction on units which previously required open construction and finned assemblies. Example of typical packaging: single-phase full-wave bridge delivering 12 amps de with 1,200 PIV per leg comes in package 2 x 2 x 1 in.; delivering 750 mA de with 5,000 PIV per leg, in a 1-in. cube.—Solitron Devices, Inc.

Circle 61 on reader's service card

AUTOMATIC HEADLIGHT CONTROL UNIT, the Turville Sentry. Turns auto headlights on at twilight, off at dawn. Controls lights during inclement daytime weather. Prevents driving with lights off at night. $19.95.—Dynamic Development

Circle 62 on reader's service card

BREADBOARDING KITS. 3/32 in.-thick phenolic deck mounts on aluminum base with slanting front. 1-, 2- and 3-in. holes on all 4 sides to mount switches, potentiometers, etc. Solder-type feed-through terminals, terminal-insertion tool, universal deck-mounting Z brackets, switch and potentiometer brackets, L brackets, small pieces of phenolic punched board. Multi-compartment plastic box. Assembly nuts and bolts. Various sizes, from $6.60.—Aladine Kitting Co.

Circle 63 on reader's service card

tips for cutting plastic and for sealing plastic bags, roll of rosin-core solder. Carrying case. Model 450K4, $15.95; model 222K5, $11.95, and model 75C, $4.95.—Wen Products, Inc.

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HANDS-FREE 2-WAY RADIO, the Marco Shoulder Talk. Antenna—speaker—microphone unit. Shoulder harness incorporates speaker, precision microphone is on gooseneck. Provision for variety of antennas, permanent or removable. Model illustrated uses safety hard hat as antenna. One model permits 2-way communication when wearing a gas mask. Units are compatible with all existing 2-way radio receivers. 5 oz.—American Teletronics Corp.

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CHURE. Stylus Force Requirements for Current Cartridges, 4-page brochure. Based on laboratory tests. Includes hints on record care.—Acoustic Research, Inc.
Circle 68 on reader’s service card

1967 CLOSED-CIRCUIT TV CATALOG, 20 pages. Includes TV camera kits (both tube and transistor), focus/deflection coils, monitors, lenses, vidicons, tripods and other items. Illustrations, descriptions, specifications, prices, order blank.—ATV Research
Circle 69 on reader’s service card

BUSINESS RADIO BROCHURE, No. D, Form 52A, 6 pages. Hole-punched. Describes portable models C-75G, C-75GK and 625C. Includes drawings, photos, system features, accessories, engineering specifications and prices.—Amphenol Distributor Div.
Circle 70 on reader’s service card

1967 AUDIOTAPE CATALOG, 12 pages. Describes 5 Audiotape formulations for specific recording applications. List of bases, lengths and reel sizes. Reference chart shows recording times for various tape lengths and speeds. Type and code numbers included.—Audio Devices Inc.
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HEP PROJECT BOOK, Field Effect Transi-
tor Projects, 93 pages. List of projects utilizing the field-effect transistor (FET). Oriented toward the hobbyist-experimenter. Available at HEP distributors or for $1.10 postpaid from Motorola Semiconductor Products, Inc., HEP Program, P.O. Box 955, Phoenix, Ariz. 85008.

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TECHNOTES

VERTICAL OUTPUT TRANSFORMER

A fairly common trouble in four-lead vertical output transformers (those having separate primary and secondary, not autotransformers) is a short from the plate end of the primary to the secondary. In sets using a circuit similar to Fig. 1 the shorted transformer can often still be used if the primary leads are reversed to place the short at the B+ end of the primary. This will make the picture "upside down," but it can be righted by reversing the secondary leads, as shown in Fig. 2. (Color codes will vary on different transformers. Typical color codes are shown on the diagrams to better illustrate the manner of changing connections.) The connection between primary and secondary should be removed; the short between windings replaces it. If the set takes a vertical blanking pulse from the secondary, the blanking may not work properly, and may have to be disconnected until a new transformer is installed.

Although this is recommended only as a temporary repair until a new transformer can be obtained, my own set

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NO MORE TV FRAUDS? Not if consumer “guardians” can help it. Editor Forest Belt interviews the publishers of a consumer magazine whose February 1967 issue carries a service-fraud story. See what can be done about some of the problems of the servicing industry.

CONSUMER ELECTRONICS SHOW The first show of its kind, will be held in New York June 25-28 this year. What the companies plan to exhibit, reasons for the Show, what you can learn there—all told in this article.

TECHNOTES continued

has operated with a shorted transformer. reconnected as in Fig. 2, for over a year. I have a new transformer, but have been waiting to see how long the shorted one would last.

—W. J. Stiles

EICO 435 OSCILLOSCOPE

The peculiar waveshape shown in the sketch was observed on a recently completed EICO 435 oscilloscope with no signal applied. It was very pronounced on the ×1 (50 mV/cm) scale and interfered with proper use of the scope. Shorting the vertical input circuit to ground, either before or after the attenuator switch, removed the pattern. When a 10X probe was used, however, the pips were present even with the probe tip shorted to ground.

Checkout with a laboratory-type scope eliminated the heater and B+ lines and the square-wave generator as possible causes. The problem was finally traced to neon indicator II, which apparently radiates large amounts of energy when it conducts on every half-cycle. The pips were eliminated by running the lamp on dc instead of ac.

To make the change, disconnect resistor R79 from pin 1 of tube socket XV8 and connect it to pin 3 instead. The neon lamp will now monitor the B+ voltage. There will be a slight delay before the lamp comes on, but the annoying pips will be gone forever.—Donald R. Hickie

END

HOMES OF THE FUTURE They’re loaded with electronics, and chances to make money are everywhere. Don’t be misled— these ideas are from the future, but how soon they will be put into use may depend on you.

PLUS . . .

• Housewife Builds a Stereo Receiver
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• Women in an Electronic World

The JUNE 1967 Issue of Radio-Electronics will go on sale May 25. Look for it at your newsstand or parts distributor (or order your subscription with the card between pages 46 and 47).
R-E PUZZLER

No vertical sweep is required to work this puzzle, only horizontal. Can you fill in all the blanks correctly? Note position of the tie-in letters.

1. Insufficient modification of carrier wave at transmitter.
2. Type of color picture tube using one electron gun.
4. Seven-electrode vacuum tube.
5. Device to reduce signal level.
6. To make uniform.
7. Compounds of this metal are sometimes used for cathodes.
8. Receiver cabinet that stands on floor.
10. Semiconductor device for switching and storage circuits.
11. One-millimicro unit of electrical current.
12. The reciprocal of reluctance.
15. Plating metal components with tin improves surface conductivity.
16. Alloy wire with qualities ideal for precision wirewound resistors.
17. A tone that slides smoothly from one pitch to another.
18. Charge remaining on capacitor after initial discharge.
19. Degree to which receiver responds to incoming signals.
22. An electronic musical instrument.
23. Pattern showing variations in amplitude with time.
24. Color of three bands when resistor value is 33 K.
25. Effect which causes wrong hue on face of color CRT.

Solution next month

MAY 1967 93
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Radio-Electronics
WHAT'S YOUR EQ?
These are the answers. Puzzles are on page 46.

Crystal Mike Input
A crystal mike is a capacitive transducer at audio frequencies, and can be represented as shown below.
The voltage across the input resistance $R_{in}$ of the amplifier is almost equal to $e_i$ at very high frequencies, when the reactance of the equivalent capacitor $\frac{1}{2} 2\pi f C$ is negligible in comparison to $R_{in}$. But at a low frequency when $\frac{1}{2} 2\pi f C = R_{in}, e_i$ falls to $\frac{e_i}{\sqrt{2}}$ (the 3-dB point). If $R_{in}$ is high (say 100K) then the 3-dB point is a low frequency (say 30 Hz). But if $R_{in}$ is low (say 1K) then the 3-dB point will be given by $\frac{1}{2} 2\pi f C = 1K$ at $f = 3$ kHz, and so the low-frequency response is badly affected. Thus to get a good low-frequency response $R_{in}$ should be as high as possible.

Another Black Box
The value of $R$ equals the value of LDR when the latter is illuminated by the lamp. With the switch open, LDR has a high resistance which prevents the lamp from lighting. When the switch is closed, the lamp lights through $R$. The illumination decreases the resistance of LDR, locking it "on."

30 Years Ago
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May 1967
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The Ovonic Threshold Switch (OTS) is a new development in solid-state switching. It is basically a symmetrical semiconductor that can be changed instantaneously from a high-impedance blocking state to a low-impedance conducting state simply by varying the voltage or current above or below a given threshold level. It responds to both negative and positive polarities.

The OTS is ideally suited for a wide variety of power-control applications where high switching speed is essential to circuit performance and long component life. It can also be used with a series resistor across an inductive load as a transient voltage and arc suppressor and as a staircase and square-wave generator (Fig. 1).

Capacitor voltage \( V_c \) follows the input but varies in increments equal to the OTS threshold voltage. The number of steps depends on the ratio of input to threshold voltage. Resistor voltage \( E_r \) consists of pulse groups whose number and spacing are indicative of the magnitude and slope of the input voltage. Pulse polarity indicates the direction of the slope.

The OTS symbol used by the manufacturer was developed from a plot of the E-I characteristics as seen in Fig. 2.

Electrical characteristics of the three OTSs in the line are:

- **Threshold volts \( V_T \)**: 10, 20, 30
- **Conducting volts \( V_C \)**: 1.4, 1.4, 1.6
- **Breakback volts \( V_T - V_C \)**: 8.6, 18.6, 28.4
- **Maximum current, rms**: 25 mA, 25 mA, 25 mA
- **20-µsec pulse time**: 1 amp, 1 amp, 1 amp
- **100 µsec pulse time**: 50 mA, 50 mA, 50 mA

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The MFE2133 is a Motorola JFET featuring low drain-source resistance (60 ohms maximum when turned on) and transfer capacitance of only 5 pF. These two characteristics combine to make this transistor an ideal choice for large-gate-voltage swings and eliminate the need for a transformer. The diagram shows a chopper circuit. Maximum chopping frequency is 200 kHz, maximum input is ±10 V.

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CRystal Switching With Diodes

In a crystal-controlled oscillator the "hot" lead between the crystal and tube or transistor should be kept as short as possible for best circuit performance. This complicates the design of multichannel oscillators because lead length is often controlled by the type of channel-selector switch and its location on the chassis or panel.

The diagram shows one method that minimizes the effects of switch and lead inductance. In this circuit, taken from Radio and Fernsehen (Berlin, Germany), the crystals are mounted close to the oscillator and connected to it by diode switches.

The diodes are connected through 10-mH rf chokes and RC filters to contacts on the selector switch. crystals for the channels not being used are isolated by nonconducting diodes. The crystal

For the desired channel is switched into the circuit by biasing its diode in the forward direction. This bias is a negative voltage applied to the cathode through the switch. Resistors R1 and R2 form the voltage divider that sets the bias level.

TV Color-Indicator Circuit

A number of TV set manufacturers have introduced indicators to show when a color broadcast is tuned in. The diagram shows the circuit used in the Olympic CTC-18 chassis. It can be adapted to any set with a similar bandpass amplifier circuit.

When a color program is tuned in, the 3.58-MHz burst causes the bandpass and blanker cathode voltage to rise. This voltage biases the transistor to conduction to light the indicator. Collector supply voltage is developed by rectifying the 6.3-volt heater supply.

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Locating a burned-out heater or filament in a series-string set can be time-consuming with a tube tester or an ohmmeter. Binary logic can locate the dead tube in just three quick measurements on an 8-filament string, and only four measurements for up to 16 filaments.

Plug in the line cord and turn the set on. This applies line voltage to the string, but since the string is not complete no current flows. Full line voltage will be present from the hot side of the line up to the point at which the string is broken. There will be no voltage from that point on.

Measure first at the midpoint of the string, point A in the sketch. (The other side of the meter is always at point D.) If there is line voltage, the string is complete from there back to the hot side and the break must be between point A and the cold side of the line. Make your second measurement halfway down this half of the string, point B. Line voltage here means the break is in either V7 or V8. The third measurement, at point C, pinpoints it. With the break in V7, line voltage is absent. Were voltage present at point C, the break would be in V8.—Jim Kyle

CONSUMER ELECTRONICS SHOW

Read about the first show of its kind—to be held in New York in June this year. What the companies plan to exhibit, reasons for the new Show, what you can learn there.

In June RADIO-ELECTRONICS

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solution to RE Puzzler for April 1967

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Top Hat A Epoxy 750 MA

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3. ADVERTISING INDEX

RADIO-ELECTRONICS does not assume responsibility for any errors which may appear in the index below.

Allied Radio Corp. ..... 77
Amperes Electronic Corp. . . . . . . Second Cover
Amplifier Corp. of America ..... 86
Arrow Fastener Co., Inc. ..... 86
Ariston Organs ..... 88
B & K Manufacturing Co., Inc. ..... 23
Brooks Radio & TV Corp. ..... 96-97
Capitol Radio Engineering Institute, The ..... 13
Castle TV Tuner Service, Inc. ..... 77
Centralab (Div. of Globe-Union Inc.) ..... 26
CLASS III ..... 100-103
Cleveland Institute of Electronics ..... 7, 18-21
Conor (Div. of National Radio Institute) ..... 88
Cornell Electronics Co. ..... 102
Datask Corporation, The ..... 95
Defa Products, Inc. ..... 87
DeVry Institute of Technology ..... 5
Dynaco Inc. ..... 82
Editors & Engineers, Ltd. ..... 95
EICO Electronic Instrument Co., Inc. Third Cover
Electronics Corp. ..... 6
Electronic Chemical Corp. ..... 97
Finney Co. ..... 86, 98
Genmarka Broadcasting Inc. ..... 74-75, 90, 94
Heald Engineering College ..... 95
Heath Company ..... 83, 85
Injecteall Electronics Corp. ..... 24
International Crystal Mfg. Co., Inc. ..... 104
Jensen Manufacturing Division (The Muter Company) ..... 76
Mallory Distributor Products Company (Div. of P. R. Mallory & Co., Inc.) ..... 17
Metrimex ..... 93
Music Associated ..... 46
National Radio Institute ..... 8-11
Olsen Electronics, Inc. ..... 84
Pols Paks ..... 103
Quam-Nichols Co. ..... 16
Quietco Co. ..... 84
Radar Devices ..... 1
RCA Electronic Components and Devices
Test Equipment ..... 25
Fourth Cover
RCA Institutes, Inc. ..... 28-31
Rye Industries, Inc. ..... 92
Sams & Co., Inc., Howard W. ..... 12
Schober Organ Corp., Inc. ..... 81
Scott, Inc., H. H. ..... 31
Sencore ..... 14-15
Shure Bros. ..... 73
Solid State Sales ..... 100
Sony Corp. of America ..... 24
Sprague Products Company ..... 89
Squires-Sanders, Inc. ..... 27
Surplus Center ..... 98
Sylvania (Subsidiary of General Telephone & Electronics) ..... 63-70
Turner Microphone Company, The ..... 91
United Radio Co. ..... 101
Warren Electronic Components
MARKET CENTER ..... 100-103
Chemtronics
Edmund Scientific Corp.
Fair Radio Sales
TAB
SCHOOL DIRECTORY
American Institute of Engineering & Technology
Graham School of Electronics
Northrop College of Science & Engineering
Tri-State College
Valparaso Technical Institute

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