DECEMBER 1954

RADIO - ELECTRONICS
TELEVISION • SERVICING • HIGH FIDELITY

HUGO GERNSBACK, Editor

In this Issue:

Balanced Modulators For Low-Level Audio

Transistorized Voltmeter

Cathode Followers

Build a Radio for Junior's Car (See page 4)
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DECEMBER, 1954
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VETERANS: Give date of discharge.
TV PIONEER AND INVENTOR Dr. V. K. Zworykin has been appointed the first honorary vice president in the history of RCA. The appointment was made at a scientific seminar and dinner at which he was the guest of honor, and which marked his retirement as an active vice president of the RCA Laboratories.

David Sarnoff, chairman of the board of RCA, and many of the nation’s leading scientists, educators and industrialists paid tribute to Dr. Zworykin for his epoch-making achievements in the fields of television and electronics.

Dr. Zworykin is best known for his contributions to all-electronic television, including the invention of the iconoscope and development of the kinescope. The electron microscope, developed under his direction, has been acclaimed as one of the most useful scientific tools of this century. He will remain as technical consultant at the RCA Laboratories.

ONE MEGAWATT effective power output will be achieved by a u.h.f. TV station when equipment ordered by WILK-TV (channel 34, Wilkes-Barre, Pa.) is put into operation. The equipment is a 45-kilowatt u.h.f. transmitter, which when used with a new antenna and transmission lines will result in one million watts ERP.

ULTRASOUND may be of considerable value in attacking brain diseases, according to reports by University of Illinois scientists at the recent Madison meeting of the American Physiological Society. The waves of supersound can be used to eradicate small areas of diseased brain tissue, tumors, etc., with less damage to surrounding areas than either the surgeon’s knife or other earlier methods. Brain tissue can be destroyed in an area as small as .05 inch across, without affecting areas immediately adjacent.

FREE SERVICE in addition to free replacement of defective parts for a period of 90 days after date of purchase, is being offered by both Westinghouse and General Electric to buyers of their radio receivers.

G-E representatives, meeting with delegates of the Federation of Radio Service Associations of Pennsylvania at Harrisburg, agreed to channel G-E service work to independent shops throughout Pennsylvania. Delegates from the National Electronic Technicians and Service Dealers Association, the Empire State Federation, and from the Eastern Service Conference were present at the meeting. It is expected that General Electric might extend its plan to cover independent service shops in other states.

Westinghouse radio owners should return a new set that needs servicing to the dealer from which it was purchased. The dealer will forward the radio to the nearest authorized Westinghouse radio service station where it will be serviced at no charge to the owner for either parts or labor, and then shipped back, according to the official statement.

NINE NEW TV STATIONS have gone on the air since our last report. These are:

- KTIV Sioux City, Iowa
- KALV-TV Alexandria, La.
- KPLC-TV Lake Charles, La.
- KFVS Cape Girardeau, Mo.
- KETC St. Louis, Mo.
- WLOS-TV Asheville, N. C.
- KTVX Muskogee, Okla.
- WBTW Florence, S. C.
- KUW Salt Lake City, Utah

Two stations, KOPR-TV, Butte, Mont., channel 4 (suspending until March 15), and WTOV-TV, Norfolk, Va., channel 23, have gone off the air.

Four new Canadian stations have gone on the air: CFPA-TV, Port Arthur, Ont., channel 2; CFRN-TV, Edmonton, Alta., channel 3; CJCB-TV, Sidney, N.S., channel 4; CHCT-TV, Calgary, Alta., channel 2.

The Editors wish you a

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DECEMBER, 1954
Once you gain the confidence of customers, you're on your way to increased volume and profits. The Raytheon Bonded Electronic Technician program is designed to help you do just that. The Raytheon Registered Bond Certificate, the Raytheon Creed Display Decal and Identification Cards, featuring your bonded status and the fine Raytheon "Code of Ethics", inspire customer confidence. And a recent survey proved that wherever Raytheon Bonded Technicians took full advantage of their Bonded status — used it to inspire customer confidence — business improved by at least ten per cent.

Ask your Raytheon Tube Distributor if you can qualify for this important sales advantage. If you can, the bond is yours at no cost to you — it is Raytheon's investment in your future.
DECEMBER, 1954

Carl E. Smith, E.E.
President

**How To Pass FCC COMMERCIAL LICENSE RADIO OPERATOR EXAMINATIONS**

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If you have had any practical experience—Amateur, Army, Navy, radio repair, or experimenting.

**TELLS HOW...**

Employers Make JOB OFFERS Like These to Our Graduates Every Month!

Letter from nationally-known Airlines, "We would also appreciate it if you would place the following additional advertisement in your bulletin—Wanted—Superintendent of Communications Salary $4644.54 per month." Letter from nationally-known airplane manufacturer, "We need men with electronic training or experience in radar maintenance to perform operational check-out of radar and other electronics systems—Starting salary amounting to $375.33 per month." Letter from nationally-known Airlines, "We are contemplating placing an Airline Ground Radio Engineer." Starting salary $385.00 per month.

These are just a few examples of the job offers that come to our office periodically. Some licensed radio men filled each of these jobs...it might have been you!

**HERE'S PROOF FCC LICENSES ARE OFTEN SECURED IN A FEW HOURS OF STUDY WITH OUR COACHING AT HOME IN SPARE TIME**

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2210 Wilshire Blvd., Bakersfield, Calif.
Drum E. Vogt  1st Phone  20
9418 Davie, Fla.
Frank X. Forth  1st Phone  38
39 Bleeker Pl., Bergenfield, N. J.
6 West Ben H. Davis  1st Phone  29
517 North Roosevelt, Lebanon, Ill.
Alfred Schnell  2nd Phone  33
110 West 11th St., Escondido, Calif.
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Carl E. Smith, E.E., Consulting Engineer, President
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Helps CIRE Students Get Better Jobs

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**GETS STATE POLICE JOB**
"I have obtained my 1st class ticket (thanks to your school) and since receiving same I have held good jobs at all times. I am now Chief Radio Operator with the Rhode Island Police."—Kevin P. Flynn, 381 E. 2nd St., London, Ky.

**GETS BROADCAST JOB**
"I wish to thank your Jan-Finding Service for the help in securing me the position of transmitter operator here at WODA in Pittsburgh."—Wallace Roberts, 142 Ridge Ave., M. Brockport, Pa.

**GETS AIRLINES JOB**
"Due to your Jan-Finding Service, I have been getting more offers from all over the country, and I have leased a job with Capital Airlines in Chicago, at a radio station."—Harry Clark, 1071 W. Broadway, Champaign, Ill.

Your FCC Ticket is recognized by most employers in the electronic field as a proof of your technical ability.

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Outstanding wide-range speaker system at moderate cost. Uses the famous "Royal Eight" speaker and Super Tweeter. Unique new-type back-loading horn enclosure is matched, octave by octave, to speakers... assuring undistorted reproduction from 35-16,000 cps. In selected 3/4' Mahogany or Korina veneers. Exclusive: Special connection for headset extension cord. Size: 24" W, 23" H, 14" D. Impedance, 8 ohms.

THE DIMINUETTE
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Suggested Audiophile Net... $49.50
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—Audio Forum Dept., Oct., 1954 issue

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THE RADIO MONTH
(continued)

RESOLUTION OF APPRECIATION
was adopted unanimously by the Federal Communications Commission to accompany a scroll paying tribute to Commissioner George E. Sterling who retired September 30. The scroll commemorated his retirement and said in part:

"George Edward Sterling, during his 31 years of Federal service with this commission and its predecessors, advanced from radio inspector to organizer and director of the FCC wartime radio intelligence activities, chief engineer and, in 1948, to the office of Commissioner. He brought to that office an unprecedented experience in radio which had its beginning in 1908, and as operator, engineer, author and policy maker has played a prominent role in the development of the radio art."

JOHN SCOTT AWARD, presented to "ingenious men and women who make useful inventions," has been made to Marvin Camras, inventor of modern magnetic wire recording. Camras, senior physicist at the Armour Research Foundation of Illinois Institute of Technology, Chicago, will receive the 198-year-old award—consisting of $1,000, a medal, and a scroll—from the City of Philadelphia, trustee.

KENNETH A. NORTON, chief of the radio propagation engineering division of the Bureau of Standards, has been presented the Stuart Ballantine medal by the Franklin Institute (Philadelphia, Pa.). The award was given for his work in the study and description of propagation phenomena. He is either author or co-author of a long series of papers, all of which contribute to our knowledge of radio propagation or outline methods for making easy engineering calculations in that field.

ELECTRONIC BRAIN designed to simulate aerial fights on a three-dimensional screen and help to solve guided missile problems has been developed in Great Britain. The machine is called the "Traidec" (three-dimensional analogue computer) and uses 8,000 tubes. When the brain calculates what would happen when a fighter chases a bomber, diagrams are automatically drawn showing the sequence of events. END

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WHAT IS CORRECTNESS?

Dear Editor:

This letter is in reference to High-Quality Audio by Richard H. Dorf in the September issue.

Mr. Dorf does not distinguish between impedance and resistance. When speaking of impedance, power is not merely the product of current and voltage; another factor, the phase angle between I and E, must be considered. This is expressed as cosine θ and must be included in the formula for power. . . .

Mr. Dorf calculates that to drive a 4-ohm speaker with 8 watts through a 40-1 turns ratio transformer requires what he calls a plate current of 35 ma. . . . I would like to point out that this 35 ma is the r.m.s. value of an a.c. signal with peak values of approximately 50 ma in each direction from the steady-state plate current. Thus the steady-state current must be in excess of 50 ma. . . . After listing some pentode and beam-power tubes (6F6, 6L6, etc.) Mr. Dorf says, "Triodes can pass more current than the others and thus require smaller turns-ratio transformers." This statement is in error and is misleading. A 6L6 can pass more plate current for a given plate voltage because its plate dissipation is higher. Yet it should look into roughly twice the load resistance of the 2A3. . . .

I mention these points because I feel RADIO-ELECTRONICS has an obligation to provide accurate information to its reading public.

William Holm

East Lansing, Mich.

(Mr. Holm's attitude is typical of a group whose manuscripts make up nearly 75% of those rejected by this magazine. What these scholars do not understand is that material must be tailored to its application. Most of them would consider it insane to design a 10-watt amplifier for use in a hearing aid; yet they do not hesitate at the equal absurdity of firing "cosine theta" at a reader whose Greek has been neglected and whose trigonometry is nil. Mr. Dorf's articles indicate by their very content—as well as the statement in the opening installment—that they are aimed at the reader who is not already an expert, in facts for "everyone who would like to know through look into the purposes, ingredients and techniques of high-quality home music systems." To reach this group it is necessary to simplify, round off corners,
Don't let any Antenna Manufacturer pitch curves to you!

new WINEGARD INTERCEPTOR

with Exclusive Electro-Lens*

Focusing means...

Don't be misled by any manufacturer's "homemade" charts on antenna performance. We can show you how to plot curves that will "prove" that a busted steel guitar on an 8-ft. flagpole outperforms any antenna you can name—even our own superb Winegard INTERCEPTOR.

No—all that counts is PERFORMANCE IN ACTUAL USE. Therefore, DO THIS: Use a single INTERCEPTOR on your next installation where you'd ordinarily stack two bays. See if one INTERCEPTOR doesn't give better results than any stacked installation you have made nearby. Once in a blue moon, a second bay is necessary even with our INTERCEPTOR. But rarely. So we say to you, "Why use two when ONE will do?"

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A FAR BETTER PICTURE OR YOUR MONEY BACK!

Antenna No. Winegard Trade Name List Price
L-4 ............... INTERCEPTOR ........ $24.95 a boy
L-5 ............... PIXIE ............... $14.95 a boy

Shipped, one L-4 to a carton (stacking bars available)
Shipped, two L-5's to a carton with stacking bars.

THE INTERCEPTOR gives the clearest, sharpest pictures obtainable... not just on one or two channels... but on all 12 VHF channels.

WINEGARD PIXIE
A quality all VHF channel antenna for top performance at a low price.

See your jobber now or write us for complete information.

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DECEMBER, 1954
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model AR-1... same as AR-2 without thrust bearing

model TR-12
A special combination value consisting of complete rotor including thrust bearing. Handsome modern cabinet with meter control dial, uses 4 wire cable.

model TR-11
The same as the TR-12 without thrust bearing, complete with meter control dial cabinet, uses 4 wire cable.

model TR-2
The heavy-duty rotor with plastic cabinet featuring "Compass Control", illuminated "perfect pattern" dial, uses 8 wire cable.

model TR-4
The heavy-duty rotor complete with handsome, modern design cabinet with meter control dial, uses 4 wire cable.

Cornell-Dubilier
South Plainfield, N.J.

The Radiant Corp.
Cleveland 13, Ohio

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It took more than engineering promises before leading set manufacturers invested in production of color TV sets. It took a practical big-screen color picture tube . . . the CBS-Colortron "205."

It took creative engineering to conceive this advanced design in color picture tubes. And it took advanced-engineering knowledge to produce it.

Today, these set manufacturers are demonstrating their recognition of the leadership of CBS-Hytron. For they are shipping color TV sets and these sets are equipped with the CBS-Colortron "205.

There can be no better evidence of the advanced-engineering knowledge of CBS-Hytron . . . Advanced-engineering knowledge you can depend upon to produce the finest in television tubes . . . for color or for black and white.

The One Big Reason
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"205"

205 SQUARE INCHES
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Four pages of engineering application data
on the "205."

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DECEMBER, 1954

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CORRESPONDENCE (continued)

ignore insignificant but very real factors, and otherwise avoid any confusing points which do not aid the reader to grasp the essentials of the matter, but whose inclusion would very effectively prevent him from grasping any of it at all.

The material is correct as far as it goes. All Mr. Dorf's articles were read critically by three persons of different viewpoints and different degrees of audio knowledge, and any doubtful points were discussed with the author and agreed upon before printing. As an example, take Mr. Holm's stand on impedance. No one will deny that X as well as R must be taken into account. But, from a practical point of view, if Mr. Holm's own audio system's output impedance doesn't look pretty much like R over the whole frequency range, he had better look to the quality of his audio! Dorf's equivalent circuit follows the electrical facts of life within all practical tolerances, and has one decisive advantage over conventional equivalent speaker circuits: it's comprehensible!

Again, his reference to plate current, taken out of context, seems a clumsy and "incorrect" way to describe the difference between a triode and pentode. But Dorf has referred to a current ratio four paragraphs before (those 35 ma) and is explaining triode action in terms of the earlier paragraph. Only one thing is incorrect in the paragraph, and the editors apologize for it, since it could have been removed with the stroke of a pencil. That is the reference to "better speaker damping" which is not explained anywhere in the article. We should not glibly have used a term too advanced for the discussion.

On one point we agree heartily with Mr. Holm, that RADIO-ELECTRONICS has an obligation to provide accurate information to its readers. If material is not clear and simple, it is not information. And if it is not information, it cannot be accurate information. If Mr. Holm has overlooked that point, we are sure that he will agree—after looking at the articles over again—that they were a correct approach to the job attempted. If, on the other hand, he belongs to the small but widespread group who feel that the would-be learner has no right to knowledge unless he has sufficient mathematical training to take it in "correct" form, we have little sympathy for him.

There remains another possibility—that a series of articles could be written which would suit the purist and yet be useful to the beginner without engineering training. We have not seen such articles, but would love to stand corrected! If any reader can supply articles which would meet all the objections of Mr. Holm, yet be interesting and understandable to the average intelligent radio-TV technician, we can guarantee for them the highest prices we have ever paid, plus a book contract!

(—The Editors)

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A Reliable, General-Purpose 5" Oscilloscope

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PRECISION MODEL

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PRECISION-engineered in response to industrial requirements for a reliable, low-cost, FACTORY-wired and FACTORY-calibrated 'scope for basic production and testing applications.

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Plus additional engineering and performance features never before incorporated in an oscilloscope designed for such general application and at such economical price!

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ADVANTAGES:
1. Lowest losses at UHF and VHF frequencies.
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3. No time-consuming end seal required; easy to install.
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9. Can be clamped tightly in stand-off insulators without crushing. No special fittings required.
10. Conductor spacing is constant even when the lead-in is transposed.
11. No stripping problem for attaching the conductor.

Thousands of separately sealed tiny cells, filled with inert gas, make his waterproof cable stable and efficient electrically.

This heavy wall of brown virgin polyethylene protects the cable against mechanical abuse and damage from ultraviolet sun rays.

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This completely new 300-chm line results from the development of a new cellular plastic core where each separate cell is filled with an inert gas to make an efficient cable with the lowest possible losses at both UHF and VHF frequencies. With this absolutely waterproof cable, no sealing of the ends is necessary. Celluline cable can be fixed in stand-off insulators without crushing. The thick outer wall of polyethylene serves to protect the cable from abrasion and sun damage.

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The copper-covered steel strands, which make up the conductors, assure 49% greater resistance to breaking from flexing or stretching than any all-copper conductor.

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Four hundred and six stations are on the air! Many more are building or in the plans stage; 32,859,915 sets are in use. Color TV is just starting to really come alive! This same phenomenal growth picture is repeated in every phase of business employing electronics — crime prevention, aeronautics, fire-fighting, communications, to name but a few. CREI has the plan to keep you moving upward, to help you assume your rightful place!

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CHECK: □ Residence School □ Veteran
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HEAVIEST BOOMS!

Thick-wall, extra-sturdy 11/4" diameter Booms. Nothing approaching them for strength! Now used on ALL low-band yagis.

New "VARI-CON" HEAD

Four Hi-strength aluminum adjusting arms, interlocking Butterfly sections, Heavier snap-action spring assembly. The "Vari-Con" is the only antenna with spring dampeners to lessen vibration and breakage. The "Vari-Con" head also used on the popular TRIO 88 Series.

INSTA-LOK CLAMP (Good-Bye Nuts)

This revolutionary clamp permits instant flip-out assembly, permanent alignment with ultra strength. Nothing stronger - nothing faster! Insta-Lok employed on ALL TRIO Antennas that have parasitic elements.

New INSTA-LOK CONICAL HEAD

Swing out element mounting plates, fan out elements into snap-fastenings and it's set! Used throughout conical line.

New MYCASTYRENE INSULATORS USED THROUGHOUT TRIO LINE

New TRIO ARISTOCRAT ROTATOR
NOW AVAILABLE IN FOUR GLORIOUS COLORS!

TRIO 88
Far superior construction. Rugged, foolproof - easily installed. Parasitic elements supported by TRIO's revolutionary new "Insta-Lok" clamps. Low-channel dipoles supported by the strongest conical head made. No vibration - No element shedding. Completely pre-assembled. Available in single or two bay models.

TRIO 66
Three dipoles provide exceptionally high gain on all VHF channels. Exclusive TRIO grid reflector gives improved performance. Extremely rugged yet lightweight. Pre-assembled - simply unfold and tighten reflector and dipole assemblies. Three vertical braces on reflector screen for increased strength. Available in single or two bay models.

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The New TRIO "ARISTOCRAT"

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RADIO-ELECTRONICS
Only Rotator Available
in Four Glorious Colors:
Mahogany
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THE TRIO
"ARISTOCRAT"

NOW...
America's most beautiful!

The sleek, modern, low silhouette of the new TRIO rotator control case marks a new high in styling.

Beauty, here, is more than skin deep since its low center of gravity makes it tip-proof! Note, too, that there are NO unsightly control knobs or switches to spoil its beauty. These are located at top rear of case — where your hand naturally rests in operation of rotator!

There is no obscuring the easily-read lighted dial.

Available in either blonde or mahogany, the unit with its graceful flowing lines blends perfectly with any decor.

Yes, America's most dependable rotator is now America's most beautiful as well!

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Here are the two famous 980 Line instruments that form the basis of the new Weston simplified method of TV receiver alignment... eliminating the troublesome, time-wasting procedures heretofore involved, and enabling servicemen to cut alignment time almost in half. This new method is possible when these two instruments are used with the Weston scope, or scopes with provisions for Z-axis intensity modulation. They also can be used with available test equipment in the conventional method of alignment. For the complete story, write... WESTON Electrical Instrument Corporation, 614 Frelinghuysen Avenue, Newark 5, N. J.

WESTON MODEL 985 CALIBRATOR
FEATURES

SCALE CALIBRATION: Crystal calibrating points are available at 1.5 and 4.5 megacycles throughout the entire scale. A scale shift knob is provided to align the scale with the crystal calibrating dots.

SCALE PRESENTATION: Slide rule type in which one scale is visible at a time. Ten scale range bands available... total scale length of 8 1/4 ft.

DUAL MARKERS: 4.5 mc side band markers permit simultaneous observation of video and sound carrier.

INTERNAL MARKERS: Special circuitry provides an internal marker of either a positive or negative pulse suitable for Z-axis intensity modulation of the scope pattern. Marker is visible even at the sound trap frequencies.

HETEROODYNE DETECTION: With an input sensitivity of 500 microvolts, the local TV receiver-tuner channel oscillator frequency can be determined without tuner disassembly.

BAR PATTERN GENERATOR: Amplitude modulated signals of the band oscillator at 400 cycles and 300 KC are available for linearity checks.

SPECIFICATIONS

Frequency Range [with Variable Frequency Oscillator]: 4-110 megacycles in 7 bands. 170-260 megacycles in 3 bands.

Output Attenuator Range: 100% to 1%

Crystal Marker Accuracy: 1.5 mc position ± 0.01%; 4.5 mc position ± 0.01%

Internal Modulation Frequencies: 400 cps, 300 KC, 4.5 mc

Heterodyne Input Sensitivity: 500 microvolts (VFO)

Linearity Adjustment: Horizontal—400 cycles, Vertical—300 KC

Dual Markers: video and sound... available for either Z-axis intensity modulation of scope or conventional marker pip display.

WESTON MODEL 984 SWEEP GENERATOR
FEATURES

BLANKING: Special circuitry produces a zero output reference base which is essential for relative gain measurements.

RF OUTPUT: Frequency modulated signal, TV channels 2 to 13 inclusive, complete FM coverage available by means of two preset selector positions. Frequencies are fundamentals of the oscillator frequency.

IF/VIDEO OUTPUT: Frequency modulated signals ranging to 50 megacycles, continuous tuning, signals free from harmonics.

SWEEP WIDTH: Full 10 megacycles on all channels.

Z-AXIS TERMINAL: For use with the Model 985 Calibrator.

SPECIFICATIONS

Sweep Width: 0-10 Megacycles (continuously variable for both IF and RF)

Output Voltage (RMS): 0.1 Volt... sweep is linear

RF Output: TV channels 2 to 13 preset. Complete FM coverage available by means of two additional preset selector positions.

IF/VIDEO Output: 50 Megacycles (continuous tuning)

Horizontal Sweep for Oscilloscope: Phase adjustment range... 165° Frequency... Power Line 60 cycles per second.
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You can be a TELEVISION TECHNICIAN

I'LL TRAIN YOU AT HOME IN YOUR SPARE TIME FOR BETTER-PAYING JOBS IN THE DYNAMIC RADIO-TELEVISION-ELECTRONICS FIELD.

Not since the beginning of the Industrial Revolution in the 18th Century have there been such amazing technological advances in the wide open Radio-Television-Electronics field. And there's no end in sight! . . . FM — UHF — HIGH FIDELITY — COLOR TELEVISION — RADAR — LORAN — all magic words holding out the promise of good-paying jobs with a secure future . . . for qualified men.

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You get two weeks, 50 hours, of intensive laboratory work on modern electronic equipment at our associate school in New York City — Pierce School of Radio & Television. I give you this AT NO EXTRA COST whatsoever, after you finish your home study training in the Radio-FM-TV Technician Course or the FM-TV Technician Course. Your home study course is complete without this two-week laboratory session, but it is one of the many Extras available to you from RTA.

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23
this great all-channel antenna discovery is smashing sales and performance records in every TV area!

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these 3 revolutionary, power-packed design features — found in no other antenna today!

1. New spacing formula: Radical new spacing arrangements between the directors and reflectors has, for the first time, extended the full efficiency and high gain of the basic narrow band Yagi over the full width of an entire VHF band.

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PLUS Channel Master's patented, super-gain TRI-POLE . . . the unique triple-power dipole that made the Champion America's most wanted antenna.

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- features 3 telescoping sections.
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the ALL-VU®
- all VHF . . . all UHF
- the only indoor antenna with this "2-Way" feature.

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Aluminum Masting.
The new idea in antenna masting —
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Lightweight ALUMast is
⅓ the weight of steel, making it so easy to install — it swings right up!
Stronger than steel, ALUMast is easier to stock and actually more economical.

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5

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for . . . more effective installations
. . . greater customer satisfaction
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TV ROTATOR
with features found in no other rotator today:
- flexible worm gear, built-in thrust bearing.
- removable motor, electrical and mechanical stops.
- weatherproof, lightweight, strong.
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- extremely high torque.

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SELECTENNA COUPLING SYSTEM
permits unlimited antenna combinations with only one transmission line!
- for the first time, you can tie together an unlimited combination of antennas, including separate antennas operating on the same band.
- ideal for areas currently using rotators, manually-operated selector switches, and "omnidirectional" antennas.

list price: $542 each
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This interlocked stack consists of 4 antenna couplers and 1 HiLo coupler joins 4 antennas.

Beautifully-styled cabinet has great consumer appeal — is smallest on market (2¾" x 4").
Finger-tip control bar.

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FULL RANGE WEATHERPROOF COAXIAL SPEAKER

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Response 70-15,000 cps
Power 25 watts
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Mounting 180° adjustable "U" bracket
Dimensions 22½" diameter, 6" depth

Better Low's: BALANCED"COMPRESSION"TYPE FOLDED HORN, starting with eight inch throat and energized by top quality low frequency "woofer" driver provides more lows than other bulky designs.

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Television is Growing Fast
Making New Jobs, Prosperity

More than 25 million homes now have Television sets and thousands more are being sold every week. Well trained men are needed to make, install, service TV sets. About 200 television stations on the air with hundreds more being built. Think of the good job opportunities here for qualified technicians, operators, etc. If you're looking for opportunity get started now learning Radio-Television at home in spare time. Cut out and mail postage free card, J. E. Smith, President, National Radio Institute, Washington, D.C. OUR 40TH YEAR.

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Sample Lesson & 64-Page Book
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This card entitles you to Actual Lesson on Servicing, shows how you learn Radio-Television at home. You'll also receive my 64-Page Book, "How to Be a Success in Radio-Television." Mail card now!

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Training PLUS opportunity is the PERFECT COMBINATION for job security, good pay, advancement. When times are good, the trained man makes the BETTER PAY, GETS PROMOTED. When jobs are scarce, the trained man enjoys GREATER SECURITY. NRI training can help assure you and your family more of the better things of life.

Radio-Television is today's opportunity field. Even without Television, Radio is bigger than ever before. Over 3,000 Radio Broadcasting Stations on the air, more than 115 million home and Automobile Radios are in use. Then add Television. Television Broadcast Stations extend from coast to coast now with over 25 million Television sets already in use. There are channels for 1,800 more Television Stations. Use of Aviation and Police Radio, Micro-Wave Relay, Two-way Radio communication for buses, taxis, trucks, etc. is expanding. New uses for Radio-Television principles coming in Industry, Government, Communications and Homes.

My Training is Up-to-Date You Learn by Practicing

Get the benefit of my 40 years experience training men. My well-illustrated lessons give you the basic principles you must have to assure continued success. Skillfully developed kits of parts I furnish "bring to life" the principles you learn from my lessons. Read more about equipment you get on other side of this page.

Naturally, my training includes Television. I have, over the years, added more and more Television information to my courses. The equipment I furnish students gives experience on circuits common to BOTH Radio and Television.

Find Out About the Tested Way to Better Pay

Read at the right how just a few of my students came out who acted to get the better things of life. Read how NRI students earn $10, $15 a week extra fixing Radios in spare time starting soon after enrolling. Read how my students want their own businesses. Then take the next step—mail card below.

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OUR FABULOUS INDUSTRY

... It is difficult to comprehend the vast extent of radioelectronics...

Back in our November, 1951, issue in the editorial "Radioelectronic Giant," based on an appraisal of the industry, we made the following forecast:

"The war effort has also blurred the picture somewhat, because it is difficult to disentangle war orders from others throughout the industry. One thing is quite certain: even with no war orders it seems likely that annual nonmilitary sales of the entire radioelectronic industry by 1960 should reach no less than $10 billions."

That figure of $10 billions will probably prove to be too conservative. It would appear that by 1957 the total sales of the radioelectronics industry will reach $11.8 billions.

We take pleasure in reprinting here a condensed version of an address by Frank M. Folsom, president of the Radio Corporation of America, and Investment Analysts Society of Chicago, gave a highly illuminating talk on "The Business of Electronics." Excerpts from Mr. Folsom's address follow:

Broad Scope of Electronics. The domain of electronics is so vast and so rapid is the rate of development that it seems hard to believe that eight years ago the industry was only one-fifth its present size. Yet that was the case.

In those eight years, electronics has assumed a stature that commands high respect in all branches of finance, trade and industry. It is the vital factor in all forms of modern communications; all modern means of mass entertainment depend upon it—radio, television, talking movies, tape sound systems, and phonograph recording and reproduction; modern transportation must have its controls and communications; the military uses it in myriad ways, and even atomic devices depend on it. Growth of the electronics industry has, of course, accompanied the increased use of many new products and services.

We have recently conducted a survey covering the 12-year period from 1945 through 1957 and analyzing the principal industry components. We believe this survey is the first attempt at so comprehensive a review of the industry.

Facts About Industry Growth. Total annual sales of the electronics industry grew from $0.5 billions in 1946 to $24.4 billions in 1958. Further growth is projected, as follows:

- 1954: $8.8 billions
- 1955: $9.5 billions
- 1956: $11.8 billions
- 1957: $14.1 billions
- 1958: $17.6 billions
- 1959: $21.9 billions
- 1960: $26.4 billions

Our study took into consideration principal components of the electronic industry, as follows:

Home and Portable Radios. This field once represented the chief source of revenue in our business. Today, because of television and changing habits of the people, sales are declining gradually—from a postwar peak of $000 millions in 1947 to an estimated $101 millions in 1956.

Auto Radios. Relatively stable sales somewhat in excess of $100 millions annually.

Black-and-White Television. Postwar growth was spectacular, with sales increasing from $1 million in 1946 to $14.4 billions in 1957, in fact totaling $1.2 billions, and nearly $1 billion is expected in 1958. A drop to $888 millions is projected by 1957, due to the shifting of the mass market from black-and-white TV to color.

Color Television. Following commercial introduction in 1954, increased volume is expected to more than offset reduced sales of black-and-white television, reaching $264 millions in 1955, $767 millions in 1956, and $952 millions in 1957. This would mean a total of nearly $2 billions at factory prices during color television's first three years.

Repair Parts (chiefly renewal tubes). Steady growth is expected to continue in support of increased receivers in service. Sales increased to $250 millions in 1955, and is estimated at $454 millions by 1957.

Servicing and Installation. This important element has grown from $145 millions in 1946 to $1.4 billions in 1953. Continued growth to $2.7 billions by 1957 is indicated.

Industrial and Commercial Equipment. Steady growth is also indicated here, from $50 millions in 1946 to $267 millions in 1953, and a projection of $320 millions by 1957.

Government Electronics Expenditures. Volume totaled $2.5 billions in 1955 (30% of total electronics industry sales). During the four years 1954-1957, Government electronics expenditures are estimated at $2.7 to $2.8 billions annually.

Electricity. Total electricity costs to operate TV and radio instruments have increased at $49 millions in 1946, $524 millions in 1955 and $634 millions in 1957.

Broadcasting and Communications. Total revenues in 1946 were $4.8 billions, or 25% of the total sales by the electronics industry. Revenues increased to $1.1 billions in 1953. By 1957, total revenues of $1.5 billions are estimated. Nearly all of this growth is due to TV broadcasting revenue, with radio and communications revenues showing only marginal gains.

Television: Black-and-White and Color. Never before in history has a service to the public expanded with such vigor. When the first postwar television sets were introduced in 1946, there were less than 10,000 sets in the entire country and only six TV stations were on the air. Today, there are more than 31 million sets in American homes and 415 stations in operation.

Color television is expected both as a technical marvel and one of the most significant advances in the history of electronics. Its prospects are brighter than the prospects of black-and-white television were eight years ago. We will see a day when virtually every American home will have a color TV set!

During the rest of this year and next year, it is estimated that more than 350,000 color sets will be produced and sold by the industry. During 1956, unit sales should reach 1,780,000; during 1957, 3,000,000; in 1958, about 5,000,000. These annual sales add up to the very satisfactory estimate of more than 10,000,000 color sets in American homes by 1959.

From a standpoint of economic gain, the estimated output would require a mountain of raw materials. Our exports in such matters have informed us that 10,000,000 color sets would use more than 350,000 tons of wood, 15,000 tons of steel, 5,000 tons of brass, 3,500 tons of solder, 3,000 tons of copper, 2,500 tons of zinc, and enough glass, plastics and miscellaneous materials to bring the total well over 1,000,000 tons—or two billion pounds!

Making This Potential Materialize. This great economic potential can and shall be transformed into a reality. It will take a lot of doing, but the foundation is sound.

Our (RCA and NBC's) expenditures in establishing black-and-white television exceeded $50,000,000 before a single dollar profit returned. We have already spent another $50,000,000 in color television research and development, manufacturing and broadcasting.

Because of the compatibility feature, pioneered and developed by RCA, color offers no threat of obsolescence to the millions of black-and-white sets now in American homes. Color augments the black-and-white service and—I would like to re-emphasize—color starts the industry on a new growth curve. From the standpoint of manufacturing alone, it is estimated that in the next five years the American public will invest nearly three billion more dollars in color set purchases than did in buying black-and-white sets during the past five years.

Beyond this period, the outlook is equally bright for the electronics business. We are far from realizing the full potentialities of this industry. We have complete confidence in the future of electronics as a science, art and industry.
Part I—The electrical and mechanical construction of a magnificent musical instrument

By RICHARD H. DORF*

AND then came a crashing chord from the mighty Wurlitzer!" In the days of our fathers and grandfathers the "mighty Wurlitzer" was as much a byword in the realm of organs as the Hammond is today.

The mighty Wurlitzer is today a streamlined reed organ actuated and heard with the help of electrons, and sounding like a small pipe organ, with nary a reed wheeze to be heard.

The model 44 organ, shown in Fig. 1, is possibly the most interesting for several reasons. As the photo shows, it is a spinet, but has two short manuals in what is coming to be something of a style among modern spinet electronics, each with 44 notes, from F to C. It also has a toe-pedal clavier of 13 notes, C to C. The power amplifier and speaker are built into the console, which is a complete organ with nothing coming out but the a.c. cable and the music.

What particularly distinguishes the model 44, however, is the fact that all reeds are blown all the time at low air pressure. The keying is done entirely electronically, resulting in the elimination of the characteristic slow speaking of reeds which has bothered some people. The system tends to reduce mechanical ciphers, too, because dirt has little chance to accumulate on a reed that is waving in the breeze all the time. Electronic keying also makes the keying much simpler, doing away with both direct action and its Rube Goldberg lever system and electrical keying with its solenoids by the ton. Another advantage is that attacks may be made as sharp or as delayed as the designer wishes.

I have often been asked, "What ever happened to the Everett Organtron?" The answer is that Wurlitzer took it over, refined it, and built it into a line of large electronic organ models. The one pictured in Fig. 2, for instance, is a Wurlitzer series 50, a complete two-
Principles of operation

The Wurlitzer uses the same basic principle as most other electrostatic transducer devices (Fig. 3). The basic component is the brass reed. A stream of air strikes the reed through a slot in the reed cell. If the stream comes upward, the first breath pushes the outer end of the reed upward. Since the reed is a high-Q mechanical resonator, the burst tends to make it overshoot the point to which the wind pressure would push a nonresonant body. Then the energy stored in the reed because of its elastic properties pushes it down and again it overshoots the initial resting position, after which it springs back up again. The angular velocity of the movement is a function of its resonant frequency and the mechanism always finds some air pressure at the proper point in its cycle to give it back the energy it expended in the last cycle, thus keeping it moving at its resonant frequency.

The electrostatic transducer system is necessary to convert the reed motion into audio voltage since the reeds themselves are enclosed in a soundproof compartment. The reed and the pickup screw are the two "plates" of a capacitor whose capacitance varies with the position of the reed. With each variation in capacitance there is a rush of electrons from one plate to the other. The electrons pass through the load resistor and create an audio-frequency voltage drop through it that is transferred to the grid of a tube. The model 44 has a bank of 73 reeds that provide all the tones and tone colorings for the organ.
The keys are mounted on a strip of spring metal. The rear of the spring metal is screwed down to a transverse rail, a part of the formed metal key support. When the key is pressed, all parts back to the end of the spring holding the assembly to the rail are depressed. The usual bearing and leveling provisions are toward the front of the key to keep it in line and equalize its action with the others.

When the key is pressed it forces down the activator, a vertical strip of insulating material with a hole for each of the 10 key contacts (9 for stops and the tenth for gating the amplifier). Each key contact is a spring wire fastened at the rear to the key contact mounting block and free at the other end. Since each contact passes through a small hole in the activator, when the key is pressed and the activator lowered, the contacts are pushed downward.

Beneath each contact is a rod of phenolic material with a heavy nichrome wire running its length. The manner in which the nichrome wire is mounted on the rod can be seen in Fig. 9. Each rod is pivoted axially (pivots not shown). Each is controlled by a stop tablet, through the mechanical linkage system. With the tab in the off position (face most nearly horizontal) the corresponding phenolic rod is so rotated that the key contact above it strikes the phenolic instead of the nichrome. In the on position, the rod is so rotated that the nichrome is uppermost and is contacted by the key contact.

A supply of plus 310 volts is fed from the amplifier power supply to the nichrome wire on each of the rods through a 47,000-ohm resistor and a bronze spring that facilitates rod rotation. When the key is played, each contact picks up this positive voltage from its stop rod (assuming the particular stop is pulled) and feeds it through a time-delay filter to the appropriate reed pickup. If the d.c. were suddenly applied to the pickup, there would be a distinct click. The filters allow the d.c. voltage to be built up gradually on the pickup, resulting in a smooth keying characteristic. The filters are printed circuits. There are 165 filters, each going to a pickup screw or strip—73 are for flute-tone pickup screws; 36 for accompaniment pickup screws, similar to the flute pickups but not so close to the reeds and therefore softer; 43 for trumpet pickup strips; and 13 for pedal pickups, similar to the flute-tone pickups for the manuals.

All the reeds for the lowest octave of notes are connected together and through 9.1-megohm and 3,900-ohm resistors to ground. The series combination is the load across which audio is developed. From the top of the load resistor the signal goes through a 0.1-μf capacitor to the bass input of the amplifier. The remainder of the reeds are connected in common to another 9.1-megohm resistor that goes through the same 3,900-ohm resistor to ground to make up the treble load. The signal passes through another 0.1-μf capacitor to the amplifier treble input. A 50,000-ohm potentiometer with its arm connected to the top of the 3,900-ohm resistor provides 0 to 5.7 volts that polarize the reeds slightly to prevent any signal when there are no keys being pressed.

Fig. 9—Section of the keying system.

Fig. 8—Diagram shows the basic keying system for the Wurlitzer model 44.
BALANCED MODULATORS for LOW-LEVEL AUDIO

By ALBERT H. TAYLOR

MODULATED vacuum-tube oscillators have long been used to play phonograph records through radio receivers. The oscillators give adequate volume from pickups delivering a volt or more but not as a rule from low-level pickups which deliver 100 millivolts or less. A recent author reports the same experience with his transistor phonograph oscillator.

This is not because the receiver lacks gain. It is probably designed for 100 microvolts or less input, but you can't expect much sound from the loudspeaker unless the signal is sufficiently modulated. A low-output pickup produces very weak sideband energy. What happens? The carrier creates a strong a.c. voltage, leaving very little gain for the weak sidebands. There are a few ways to overcome this:

1. Preamplify the audio signal before modulating. But this presents the usual problems of low-level audio and defeats the purpose of modulating—use the high r.f. and i.f. gain of the receiver, though it might well be the best approach from the quality point of view.

2. Reduce the carrier amplitude. This is more difficult with tubes than with transistors.

3. Balance out most of the carrier by a bridge or balanced modulator, as is regularly done in carrier telephony. This (double-sideband reduced-carrier) method has the advantage that while the percentage of modulation in the output can be made high, the individual elements which perform the mixing operate at low percentage modulation, that is, with the carrier very strong compared to the audio. This is accepted as the best condition for low distortion.

Of course, if the balance is too perfect, the receiver output will be highly distorted by overmodulation. Hence, with ordinary receivers without carrier re-injection, the modulator must be operated a little off balance. Distortion by the receiver detector will be least with the lowest percentage modulation (i.e., the most unbalanced carrier) that gives enough volume.

Suppressed-carrier transmission has a number of important applications (as in the carrier telephone systems already mentioned). The circuits are interesting, and the technician or experimenter will find the subject one well worth his study.

There are many possible circuits for suppressed-carrier modulation. All depend on some sort of bridge action and, like other a.c. bridges, most require at least one transformer with an electrostatic shield between its windings. I have built two balanced modulators that work very nicely with low-level pickups and microphones. One is a tube circuit, the other a transistor and varistor circuit. The tube circuit appears in this article.

A practical circuit

This modulator (Fig. 1) operates at 6200 kc because it is also a band converter for my BC-348-P receiver and I have a 6200-ke crystal. To change frequency I just plug in a different crystal and tuned circuit. The modulator being directly coupled to it through a capacitor to avoid misalignment, just as when using a signal generator.

The triode portions of the tubes form a push-pull oscillator controlled by the crystal but will also oscillate without it. It will not oscillate without the neutralizing capacitors. For 6200-ke, tank coil L has 24 turns center-tapped of No. 22 d.c.c. wire, a single layer 1 inch long on a ¾-inch wooden dowel set in a tube base. Needless to say, a much

Fig. 1—Electron-coupled modulator—push-pull oscillator, parallel output.

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better and smaller coil could be made. Tank capacitor C is a 100-µuf ceramic mounted on the coil. The neutralizing capacitors are 20-µuf ceramics mounted on the extra pins of the coil base. An 18-µuf trimmer for C is located beneath the chassis and can be screwdriver-adjusted for best results.

To adapt the crystal holder from a 274-N transmitter for push-pull operation, I disconnected the shield cover from pin 7 and connected it to pin 1. Pins 3 and 7 are the crystal electrodes and pin 1 is grounded.

**Modulator**

The oscillator feeds the hexodes as in ordinary converter circuits, but 180° out of phase in the two tubes. Since the plates are connected in parallel, the oscillator fundamental frequency is cancelled in the output if there is perfect symmetry. Of the many ways of adjusting for symmetry I have chosen the differential screen voltage control because it is bypassed and can be located anywhere. Perfect balance of an a.c. bridge requires reactive as well as resistive balance, but this one balances well enough for the Pickering D-140-S with the screen control alone. In the broadcast band the reactive error should be even less than at 6 mc.

If you need a finer balance for some weaker audio source, one way to get it is to insert an additional impedance (a 1,500-ohm resistor will do) between each plate and the output and hang a trimmer from each plate to ground, as indicated in Fig. 1. The bridge then shows the same minima typical of a.c. bridges, approaching the true balance by successive adjustments of the screen voltage control and trimmer, and the fundamental output can be reduced to a few microvolts. Even harmonics are not balanced but are excluded by the receiver tuning.

The pickup or microphone is connected push-pull to the two signal grids or single-sided to one of them. If it feeds both in parallel, the desired sidebands are cancelled. The r.f. bypass capacitors across it are a precaution only and may not be necessary.

The strong carrier developed by the oscillator must be kept from getting out by any route except the bridge, otherwise it will mask the sidebands. Thus, all leads entering the chassis and especially the oscillator compartment must be filtered and bypassed, and the shielding must be good. My modulator is in a BC-1028-A cabinet and uses some BC-1028-A parts. The under-chassis space is divided into a bridge compartment next to the panel and an oscillator compartment at the rear. The copper-shield partition passes across the middle of each tube socket between pins 1 and 8 and pins 4 and 5. Cellulose tape keeps it from grounding the terminals.

**Band conversion**

With inputs as in Fig. 2, this unit does double duty as a phonograph modulator and band converter. Using the 6200-ky crystal, the BC-348-P covers the missing broadcast band and below when tuned to either 4700–6185 or 6215–7000 kc. The balanced modulator offers no advantage in this application except for receiving such low frequencies—approaching bottom at 10 kc—that the receiver must be tuned very close to the local oscillator frequency and an unbalanced converter might greatly overload it.

With no variable tuning ahead of it, this converter requires no tuning control except that of the receiver, but it is subject to spurious responses. The low-pass filter removes the worst of these, namely the receiver tuning range itself, the image frequencies which lie still higher, and any sums of two transmitter frequencies which might fall into the receiver tuning range. Its elements are computed for $f_c = 1730$ kc, $f_{ac} = 2070$ kc, $Z_o = 400$ ohms. They lie within manufacturing tolerances of the nearest stock values. A tuned circuit covering the broadcast band costs less and works better but of course means another tuning control. It can be switched in after searching has been done with the receiver tuning alone. A later article will deal with a similar unit using transistors.

**Fig. 2—Circuit for band conversion.**

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**References**

DECIBEL MEASUREMENTS

with SIMPLE EQUIPMENT

By HECTOR E. FRENCH, W1JKZ

PROBABLY every service technician, radio amateur and audio enthusiast is familiar with decibels. Even the television viewer who watches talent shows sometimes decides that the best performer of the evening is the sound-level meter, which is calibrated in decibels.

Unlike the television viewer, the technician, radio amateur and audio enthusiast should know that a decibel is not primarily a fixed value, like a volt or an amper, but is a very useful way to compare different sound intensities.

For example, two different receivers may be compared in terms of a noise figure stated in decibels, or the response curve of an amplifier at 15,000 cycles may be stated as being a certain number of decibels below the response at 1,000 cycles, or an amateur phone station may be 20 decibels above S9. (I'm usually the same distance below S2.) It's always a comparison, no matter what the particular measurement may be.

The values to be compared can usually be measured with very simple equipment; the familiar volt-ohm-milli-ammeter will often be adequate. The response curve of an audio amplifier, for example, can be found merely by applying a signal of constant strength to the input and measuring the output voltage across the load as the frequency is varied through the audio range.

But when these values of output voltage are converted to decibels, there is usually trouble—complicated mathematics or a confusing decibel chart. Because of this process of measuring in linear values and then converting to decibels, the usefulness of decibel measurement is often controlled by the labor involved, to say nothing of the possibility of error. Later in this article a simplified conversion method will be shown. But first—how are the measurements made?

No matter how easy or how difficult the conversion to decibels may be, some sort of measurement must be made first. Fortunately, if the readings are to be expressed in decibels an extremely accurate meter is not necessary. A meter accurate to only 10% will be accurate to within 1 db when the readings are converted from linear values.

A meter of only moderate accuracy may be used, with one important restriction. This has to do with the meter resistance. An a.c. meter with an iron-vane magnetic-repulsion movement, for example, can take so much power from the circuit that the readings are completely worthless.

To illustrate, assume that the response of an audio amplifier is to be found by measuring its output voltage as shown in Fig. 1. With a 5-volt meter connected as shown, it is obvious that the meter will show the voltage across the voice coil of the speaker. It also seems obvious that with an 8-ohm speaker, a full-scale reading on the meter should correspond to a little over 3 watts output from the amplifier.

But if this meter is of the iron-vane magnetic-repulsion type, the readings may be completely worthless because such a movement can have a resistance as low as 5 ohms. This would mean that the meter would be drawing more power than the speaker! In this particular case, full-scale reading indicates 7½ watts, not 3.

Obviously, another type meter must be used. For our purposes either a rectifier type d.c. meter or a v.t.v.m. is indicated. The rectifier type meter can be used because its resistance is usually at least 1,000 ohms per volt. In this circuit, the 5-volt range would have a resistance of at least 5,000 ohms, so much greater than the voice-coil impedance that the effect of adding the meter to the circuit can be ignored.

There is one possible source of error in a rectifier type d.c. meter which cannot be ignored, especially when measuring at the upper audio frequencies. It is the effect of capacitance in the rectifier elements, which reduces the rectification efficiency as the frequency rises and causes the meter to read progressively lower as the frequency is increased. With many instruments an error of 1% per 1,000 cycles is not uncommon. This error will vary from model to model, and the only certain way to find the amount is either to calibrate the meter at different frequencies or to ask the manufacturer for the frequency characteristics.

Another common location for measuring audio voltages is not at the output terminals but at some particular spot inside the amplifier, such as at the plate of one stage. This is usually done as shown in Fig. 2. The signal is being measured at the plate of the output stage, using a rectifier type d.c. meter to avoid loading down the stage and a blocking capacitor to protect the meter from the d.c. component of the plate voltage.

In this case, if a 0-10-volt meter is used, with an effective resistance of 1,000 ohms per volt, the meter resistance will be 10,000 ohms. But if the tube is operating into a plate load of
AUDIO—HIGH FIDELITY

5,000 ohms, adding this meter will drop the loading to 3,300 ohms. Since this change is too great to get, it will be necessary either to use a more sensitive meter (one with more ohms-per-volt), or the same meter at a higher full-scale reading. If the meter to be used has a resistance of 50,000 ohms, for example, its effect on the loading is merely to drop it to 4,550 ohms, probably too small a change to cause difficulty.

However, even this particular circuit has a source of error which can somehow be corrected. When the resistance of the blocking capacitor in Fig. 2 starts to approach the resistance of the meter as the frequency is lowered, much of the output voltage is dropped across it. A value of 0.5 at or more will be satisfactory down to the lowest frequencies of the audio range when using a 50,000-ohm meter resistance.

This same measurement applied to a voltage amplifier would have the appearance of Fig. 3. In this circuit, the most rectifier type a.c. meters cannot be used at all, because of their low resistance compared to the plate load resistance. This problem is made even more troublesome by the low levels of signal voltage usually found in the voltage amplifier stages. Thus, these measurements will almost always require a v.t.v.m. with an effective input resistance of at least a megohm.

Picking the right meter for a particular measurement is only half the job. Even when the measurements are completed, the results are still just a series of readings in terms of voltage or current or power. The question now becomes, how are these measurements converted into decibels?

As previously described, the corresponding values of decibels can be calculated (using a table of logarithms) or the conversion can be made from a table of decibels. And even when a table is right at hand, the ratio of the currents, voltages or powers must be found before the table can be used, since all the values are tabulated with respect to 1,000.

Conversion Scales

To simplify this roundabout process, the set of decibel conversion scales shown in Fig. 4 were designed. These scales could also be called a decibel-conversion nomogram, using the (imaginary) pointer of the meter to indicate the values.

The lower part of each scale is a linear measurement, with one scale having a range of 0–3, the next 0–12, and the last 0–60. The actual measurements in terms of voltage, current or power are represented along this linear scale, using whichever range is appropriate. The three ranges are chosen to correspond to the most popular scales used in multimeter type instruments.

The upper part of each scale has a series of numbers which are squeezed together at the low end of the range and spread out at the high end. These are the numbers which represent the decibel equivalent of the voltage, current or power being measured.

By using the conversion scales not only can linear measurement be transformed at a glance to decibel measurement, but with a little practice the readings on the meter being used can be converted directly to decibels during the process of making the measurements. The result is that values are found directly in the required terms, with no intermediate steps being required.

To avoid confusion, there are two sets of decibel values for each range of linear values: the large-sized figures used for current and voltage measurements, the smaller figures (in parentheses) for power measurements.

The reasons for using two scales lies in the nature of the values being measured. With a given resistance, the power is proportional to either the square of the voltage across the resistance or the square of the current through the resistance. Since the conversion to decibels is being made for some values that are squared (voltage and current) and for some values that are not squared (power or watts), two separate sets of decibel values are required.

For convenience, these scales are all drawn with respect to 1.5 as the reference value, since this permits the number of decibels to be a whole number at full scale. In addition, 1.5 volts of audio applied to the 5-ohm voice coil of a good loudspeaker represents close to 1⁄4 watt of electrical power, which corresponds to an average sound level of music reproduction with neither bass nor treble boost in a family-size living room—a convenient basis of comparison. A 1.5-volt signal across a 500-ohm load also represents 4.5 milliwatts of power, which is reasonably close (1.3 db.) to the older standard power reference level of 6 milliwatts.

To use these scales, find the value of current, voltage or power along the bottom of the scale (linear measurement) and read up to the upper scale which is calibrated in decibels. If the heavy figures directly above the (curved) scale are used for voltage and the small figures in parentheses directly above are used for power measurements, the proper decibel values will be found quickly, with no calculations and no values to extract from a decibel conversion table.

Transforming these measurements, which are given with respect to 1.5, into measurements referred to some other value is very simple. If, for example, the measurements are to be obtained with respect to 10 volts, first perform the conversion in the usual way, with respect to 1.5.

Now operate on the 10-volt value. Referring to the 0–12 range, 10 volts in linear measurement corresponds to 16.5 db, still with respect to 1.5.

To find the final set of values, take each of the decibel values of the measurements (with respect to 1.5) and find the difference between them (subtract 16.5 decibels from each value). This will be in decibels, referred to 10 volts. Or, for example, if the power readings are to be given with respect to 1 watt, the readings are first converted to decibels by using the values in parentheses. Then, since the 0–3 range shows that 1 watt corresponds to −1.75 db, the difference is found between −1.75 db and the other power readings.

Each value in the measurement has 1.5 decibels added to it. In this way, the final power measurements are found in decibels, with respect to 1 watt.

There are no doubt other simple ways to arrive at decibels, and some technicians may find the curved meter type scales strange. They are more like a nomogram than a straight graph and easier to read. This may not be the best and simplest approach—merely the nearest thing to it this author has been able to find.

END
SERVICING

Part X—Turntable hum, rumble and wow; record care and storage

By JOSEPH MARSHALL

HE phonograph turntable or changer affects performance principally by masking portions of music with hum and rumble, and by increasing IM distortion because of the hum, rumble and wow. In one or two turntable models, hum, rumble and wow have been reduced to a point where they have no significant effect on the playback characteristics, but all others have a certain amount of hum and rumble which must be tolerated. When these have increased over their original or design value, the service technician may be able to correct the trouble.

Hum is due to the motor type, its location in relation to the pickup, and the sensitivity of the pickup to hum. In general, the lower the output of a pickup, the more susceptible it is to hum. So, when hum is very annoying, it may be worth while to replace the cartridge with one of higher output and better signal-to-noise ratio. In many cases, the new high-fidelity ceramic cartridges might well be used. Their frequency range and distortion characteristics are scarcely distinguishable from those of the lower-priced magnets and they are much less susceptible to hum pickup.

Sometimes, a low-impedance magnetic pickup will improve the hum ratio because the hum picked up by the leads is lower in proportion to the desired signal. Such a pickup will require a matching transformer which must be positioned carefully. Incidentally, when hum is bothersome with low-impedance pickups using a transformer, check the orientation of the transformer not only to the phono motor but to any power transformers on the same or adjoining chassis. Such transformers are often triple-shielded and sometimes have two grounding points; it may help to ground the outermost shield to the turntable or motor and the inside shield to the preamp chassis or common ground bus.

More often than not, any increase in hum is due to an increase in the hum picked up by the cable from pickup to preamp. Check to see that this cable is well grounded on both ends and does not pass close to the motor or a power transformer. Sometimes a considerable improvement can be produced by using a two-conductor shielded cable with the shield grounded only at the preamp (see diagram). Often the hum can be decreased by simply transposing the power-line plug of changer or amplifier; in other cases, it may be worth while to ground the motor frame.

When assembling a single-play turntable and arm, check to see that the arm is positioned so the cartridge is always the maximum distance from the motor.

Short of replacing a turntable or changer with a low-noise model the above are the only steps a service technician can take to improve hum level.

Rumble is produced by motor vibrations reaching the needle by way of the turntable. Ordinarily, the rumble is set by the design and there is little a service technician can do about it. Occasionally one of the rubber mounts of the motor ruptures. This either increases the vibration or decreases the damping—check for it.

Most tables and changers are mounted on springs which not only absorb the external shocks but also damp the motor vibrations. Check to see that the table is actually floating on these springs; often the bolts have been tightened down to such an extent that the cushioning effect has been nullified.

The turntable runs on some sort of bearing requiring grease or oil. The lubricant not only reduces friction but often also serves as a viscous damping element. Check the lubrication as per the manufacturer's recommendations. Be careful not to over lubricate—excess oil may spray the drive wheel, causing slippage; excess grease may increase friction or catch dust and dirt.

The padding of the turntable, whether felt, flocking, rubber or what not, also serves as a damping element. If its resiliency is gone, the damping is decreased. The new foam rubber turntable cushions may prove a help in reducing rumble.

This thing called wow

Variations in turntable speed, due either to variations in the line voltage or to play and wear in the drive mechanism, produce wow. Check with a stroscopic disc; wow will be indicated by a more or less rhythmic tendency of the pattern to slip backward or forward one or more times each revolution. A simpler way is to listen to the tones on a test record. If the pitch of the tone is constant, there is very little wow; but if the pitch rises and fails, there is considerable wow.

Most hi-fi changers and turntables are driven by a rubber idler, and wow is almost invariably the result of wear or distortion in this drive wheel. With the table revolving, touch the edge of the turntable to increase the friction

Shield is grounded only at the preamp.

and drag. The turntable will slow down but the pull should remain constant and smooth. If the drag the finger feels is reduced in some portions of the turntable's revolution, the idler is slipping at these points.

If idler and turntable remain in contact in the "off" position, flats may form on the idler, especially if the changer or turntable is used infrequently. Such flats are easily felt when the above test is made. The drive wheel or idler should be replaced whenever the wow increases. In fact, it is probably a good idea to replace the idler every 6 months.

Always examine both the idler and the turntable rim for any evidence of oil or grease. Overoiling sometimes results in the turntable spraying the
rim with oil, causing slippage. The idler can be cleaned with a cloth saturated in either alcohol or cleaning fluid. Unfortunately, there is no telling which should be used except by reading the manufacturer's recommendation. The wrong cleaner may damage the idler. The rim, of course, should also be cleaned thoroughly.

In most cases the idler is held in position against both the turntable rim and the motor shaft by a spring and a slotted mounting. Check to see that this spring is still "alive." Some service technicians like to adjust this spring by cutting off a portion of it to increase the tension. This is not a safe procedure for hi-fi. Though it may temporarily reduce the wow of a worn idler, it is very likely to increase the rumble by reducing the damping the spring provides.

Changers, of course, present many mechanical problems. This magazine published a series on this subject some months ago. The technician who does a lot of audio servicing should obtain service manuals for all the commonly used changers—servicing procedures differ widely from make to make.

Care of records

There is no full agreement on the proper care of records. That may well be an indication that records are more tolerant of mishandling than we have thought. Record problems fall into three categories: reducing wear; reducing noise, and maintaining physical stability or preventing warping, shrinkage, etc.

Wear is a function of use and the condition of the pickup. The best guarantee of long life is to use a pickup with low pressure and to maintain good needle condition, alignment and tracking as detailed previously.

With today's LP plastic records, noise is the result of static electricity or dirt. Since static charges attract dirt, the two are interrelated. There are methods of reducing the static charge by spraying or wiping the record with special liquids. Both the spray and the special antistatic cloths do work. Some say that scratches due to wiping will cause more damage than do the dust particles and static it is supposed to remove; others say that any liquid film sprayed or wiped on the record simply serves as an adhesive and trap for dirt. Some advocate wiping the record occasionally with a damp cloth; others recommend a periodic washing of the record in soap or detergent and water. You pay your money and take your choice.

One procedure is recommended by almost everybody: The needle should be cleaned of dust and lint after every record with a fairly stiff but fine brush.

Good storage conditions are probably the best and least used insurance against deterioration and damage of records. The plastics of which records are made are sensitive to changes in temperature and humidity; they are also very subject to distortion and deformation. A record cabinet with doors is undoubtedly the best. The doors tend to maintain a more uniform temperature and to exclude dust.

When the records are few enough in number so that they cannot fill a shelf snugly, stacking them flat is probably the best insurance against warping. They can be filed vertically provided they are enclosed in fairly stiff cases which will take up most of the weight and that they stand straight up and down, instead of leaning sideways. If you do not have enough records to fill a shelf snugly, use bookends or some other means to keep them standing straight up and to prevent them from slipping into a leaning position.

Choose a place for record storage that is not overheated or overcooled and maintains a fairly uniform temperature. Usually that means putting the case against an inside partition or wall and well away from radiators, convectors, or air conditioners. Be careful, too, that the case is not in a place where the sun can hit the records. On a warm day the heat produced by this direct radiation may do very great damage. Humidity is also important; a range of from 40 to 60% is best. Since this is also the best for human beings, care in maintaining it will do both you and your records the most good.

Occasionally, despite all precautions, records may acquire a marked wave, warping or dishing. In most cases they can be restored to flatness by placing them between two panes of glass, a little larger than the records, and weighing them down with 10 or 15 pounds of books or whatever. Leave them weighted down for at least a week or, if badly warped, longer.
MISSING LINK IN SPEAKER OPERATION

By D. J. TOMCIK*

Part I — The amplifier damping factor and its application to speaker performance

In audio reproduction, a subject of considerable importance to the high-fidelity enthusiast is amplifier damping factor and its effects on speaker operation. Misconceptions have arisen concerning this subject, and vague and incomplete answers have too often been given to the many questions involved.

Are the high damping factors found in present high-fidelity amplifiers byproducts of high-feedback circuits and, as such, unimportant in the operation of the system? Or is the ultimate, as some loudly proclaim, to have the highest possible damping factor built into the amplifier? Why does a particular speaker sound better with amplifier A than with amplifier B, although both show identical frequency response and power capabilities under bench checks? Why does that $2 speaker with the 6-ounce magnet (inefficiency and distortion included) seem in some cases to have more bass than the high-fidelity unit with the 5-pound magnet? Why is it that one enthusiast found reproduction more pleasing when he used a little current feedback from the output circuit and another didn't when using the same circuitry?

Some simple laboratory experiments and straightforward analysis can clear the air and answer these questions.

Speaker mechanics

The speaker can be considered, for the moment, a purely mechanical device. As such, the cone with its inherent mass and the cone suspension with its compliance or stiffness make up a resonant system. Some mechanical damping is present but such a slight amount that the system can be considered highly underdamped — if the cone is displaced and then released, it oscillates about its normal resting position for several cycles at its natural resonant frequency. This oscillation decreases in amplitude and finally reaches a state of rest due to the small amount of damping.

If this underdamped speaker is driven by a voltage source having a very high internal impedance so as to maintain the underdamped condition, the cone will vibrate at a greater amplitude at frequencies close to its natural resonance. (This action is similar to pushing a swing or pendulum "in time" with its natural period so as to obtain large amplitudes.) The frequency-response curve of the speaker under these conditions will show a peaked output near cone resonance, usually between 30 and 80 cycles per second. Operation in this manner produces high transient distortion and is undesirable in high-fidelity systems. This can be shown by depressing and then releasing the cone. The oscillation which results is all distortion, since the cone does not follow the applied square waveform of depressing and releasing it.

The speaker as a transducer

To reduce the transient distortion as well as the peaked bass response, it is necessary only to damp the cone. If the speaker were purely a mechanical device this would be difficult. But since it is an electromechanical transducer, damping is obtained easily. In analyzing the electrical portion of the speaker we find a coil of wire wound around a form and attached to the cone. The coil is placed in a magnetic field and in this way constitutes a simple motor or generator.

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If a voltage is applied to the coil, it moves in the magnetic field which in turn moves the cone. If the cone is made of a material that cools the motion of the coil in the magnetic field generates a voltage in the coil. From this it can be seen that cone damping can be obtained by using the magnetic braking action present when the coil terminals are externally closed through a resistance. The cone in trying to oscillate generates a voltage in the coil. This voltage produces a current flow through the coil and external resistance. The current flow tries to move the cone by motor action, but opposite in direction to the motion producing the current. Therefore the cone is damped in its free motion.

For a given speaker, the amount of damping can be varied by changing the value of the external resistance and consequently the value of the braking current. There is one point of value of damping at which the cone returns to rest in the quickest possible time without going past the rest position. This condition is called the critically damped state. Transient distortion is greatly reduced and the frequency response is more nearly uniform.

Excessive damping returns the cone slowly to its rest position. If the speaker is driven by a voltage source with very low internal resistance, the low-frequency response lacks intensity. The action now is similar to pushing a pendulum while it is submerged in grease or heavy oil.)

Fig. 1 shows the extent to which a speaker's output is affected by various damping values. Actual speaker performance is shown for overdamped, critically damped and underdamped conditions. To obtain these curves, the amplifier driving the speaker had means for varying its internal resistance. The method of doing this will be discussed later.

### Speaker-damping variables

The easiest method for determining the factors as well as their effect on the damping action is to resort again to the magnetic brake. First, we know that the induced voltage in the coil is directly related to the amount of flux cut by the coil. Therefore a larger magnet or smaller gap volume will induce more voltage. As a result, a larger external resistance is needed to limit the current so that the desired braking action is retained.

Second, the amount of induced voltage is directly related to the length of conductor cutting the magnetic field. Since both the coil diameter and the number of turns are directly proportional to the length of wire, we may conclude that these factors also enter into the determination of the critical damping resistance (CDR).

Third, the CDR is made up of two components. The d.c. coil resistance as well as the external resistance limits the amount of braking current. So the CDR is the sum of the coil resistance and the external, or amplifier, resistance. Conditions might exist, and they surely do, where the d.c. coil resistance of a given speaker itself is greater than the resistance of the coil mass and suspension stiffness on the value of the CDR. It is logical that to stop a heavier cone from moving in a given time, a greater force in opposition to the motion is required. To increase the opposing force, it is necessary to have a larger braking current, obtainable by lowering the circuit resistance. Also, the stiffer suspension, when displaced, possesses a greater restoring force. The tendency for the cone to overshoot and undershoot is increased. Therefore the damping force necessary to overcome this restoring force must be increased proportionately and can be obtained by again decreasing circuit resistance. It must be remembered that the effective mass and stiffness of a speaker are dependent to some extent on the type of enclosure in which the speaker is housed.

To summarize, the factors that determine the CDR can be mathematically expressed as follows:

$$\text{CDR} = C(BJ)^2 \sqrt{\frac{V}{M}}$$

where CDR is the critical damping resistance; B, flux density; I, length of conductor in the magnetic field; k, effective stiffness; M, effective mass and C is a constant.

This formula is not given to encourage experimental calculations—the test equipment necessary is far beyond the means of most audio enthusiasts. Rather, it is used to indicate the relationship of the various factors in determining the critical damping resistance.

### Experimental results

Figs. 2, 3 and 4 show actual response curves of three speakers for various values of damping resistance. Infinite baffles were used in obtaining all curves. The speaker whose curves are shown in Fig. 2 is an inexpensive unit with a 0.8-ohm magnet and 1-inch coil diameter. But referring to the equation we can expect the CDR to be low because of the low values of B and I. The curves bear this out since both show the speaker in an underdamped condition. The d.c. coil resistance of this speaker is greater than the CDR. Even though the amplifier resistance is 0.5 ohm for the lower curve, the speaker is still underdamped. Fig. 3 shows curves for a high-fidelity 12-inch speaker with a 3-pound magnet and a 2½-inch voice-coil diameter. As expected, the CDR is much higher than in the first case. The overdamped, critically damped and underdamped curves show plainly that the smoothest response occurs when the speaker is critically damped. Fig. 4 indicates that this 15-inch speaker with a 5-pound magnet and large voice coil has still a higher CDR. The overdamped curve, No. 5, shows what can be expected from the speaker when the amplifier damping factor is 10 or greater, commonly considered to be the criterion of a good high-fidelity amplifier. However, to obtain critical damping with this particular speaker, the amplifier damping factor had to be adjusted to a value of 0.4—an internal resistance of 40 ohms on the 16-ohm tap! The increase in efficiency by operating this speaker critically damped instead of overdamped is 5 db or 8 times the acoustic power output.

### Conclusion

To return to our questions in the early part of this article, let us now see how many can be answered. High damping factors should not be considered byproducts of the amplifier design but should be controlled. They play a very important part in the reproducing chain. Neither should an amplifier be designed with very high damping factors only. A good high-fidelity amplifier demands that the damping factor be variable within wide limits. It is important not only to present the correct load impedance to the amplifier, but also to present the correct load impedance to the speaker. These two load values are seldom the same. The means of true amplifier-to-speaker matching is obtained with the aid of correct amplifier damping factor selection. The answer to the question of why amplifier A works better than amplifier B with a given speaker is that the damping factor of amplifier A more nearly critically damped the speaker than did amplifier B. During the bench test with resistive loads, the two performed in an identical manner. With the variable load of the speaker, the operation was entirely different. The inexpensive speaker seemed to have more bass than the high-fidelity unit because it was working in an underdamped condition, even with a high amplifier damping factor, whereas the hi-fi speaker was heavily overdamped. And finally, the man who improved his speaker performance with current feedback was merely altering his amplifier damping factor to suit his speaker combination. Your speaker, being entirely different, demands that that particular value of damping as it would with critical damping. It's all as simple as that.

From the foregoing we see that speakers vary greatly in their requirements of source impedances to critically damp the cone and achieve optimum speaker performance. It also has been conclusively shown with laboratory curves that best speaker performance occurs with critical damping. No one value of the damping factor can satisfactorily match all speakers and enclosures. The missing link has been found in critically damping the speaker.
HIGH-QUALITY AUDIO

Part XVI—Series concluded—What price "high fidelity"?

By RICHARD H. DORF*

The Bogen model DO10 audio amplifier.

This is the final article in this series in which we have tried to show the basic requirements of home music systems and something about their components. The series has been far from encyclopedic and has steered clear of uncompromising engineering discussions and mathematical analyses. This was done in the hope that the hobbyists and service technicians for whom the series was written would gain from it a familiarity with the problems of high-quality sound reproduction.

The title of these articles has been "High-Quality Audio"—not "High Fidelity." An author does not normally decide the title which will appear in print, but in this case did make a definite point of avoiding "high fidelity" in both title and text for good reasons.

"High fidelity" is a catchword and a slogan today. Like so many words (democracy, for instance) it has come to mean different things to different people. If you make McIntosh amplifiers or sell Altec speakers "high fidelity" means sound as much like the original as possible. If you sell CBS-Columbia model 560 phonographs, you mean sound that is about as good as can be had from a table machine (considerably better than previous table models ever could produce, but not comparable to concert-hall timbre). And if you are a salesman for Acme Phonola† ($29.95—incorporating-a-spare-permanent-needle) you mean that you have included an extra 95-cent speaker in an exact replica of yesterday's squawkbox. One manufacturer has come out with a "high-fidelity" piano.

Obviously, the sense has been lost from what started out as a meaningful term, and we are reduced to finding a substitute in "high quality." But whatever the term we use, its meaning will sooner or later be confused. Industry people here and there have made motions to clean up this nomenclature mess, for example, by placing various equipments into classes—High-fidelity Class I to, say, High-Fidelity Class VII (which would correspond to no fidelity or pure unintelligibility). Inclusion of the words "high fidelity" in each case would, some innocents assume, satisfy advertising managers whose products fall into Class V or thereof, but few companies have yet displayed any inclination to adopt classifications that would tend to indicate its offspring are not as good as somebody else's. Very little can be done about the situation, other than for the individual to exercise some informed initiative when he buys. This can be done in a few relatively simple steps, especially by readers of this magazine.

The first step is to decide whether a high-quality (or high-fidelity or high-something-else) music system is really a good idea for the individual's enjoyment of music. To imply that a question exists here is heresy in certain circles, but history's heretics (Galileo, for instance) are largely responsible for human progress. There are many, including a great number of musicians, who honestly feel that the time, thought, space and money involved aren't worth while. If you have heard the Apassionata played 73 times by as many artists, it is very likely that your mind's ear will supply the missing overtones if you listen to the 74th rendition on a cheap phonograph—which requires no connecting, adjusting or space, is easily portable, and relieves you of a minimum of dollars. But there is no denying that most people—if they will put forth the rather modest effort involved and the (believe it or not) rather modest financial expenditure—will genuinely enjoy music which sounds as much as possible like the original. If you are in this majority, read on.

Step two is to realize that there are different grades of sound reproduction. The aforesaid Acme phonograph may be the lowest quality but the music is generally recognizable, and for the price the instrument may well be a bargain if we disregard the "hi-fi" claims in its advertising. The two-speaker trend in table models has produced considerably better sound than the same size machine gave in former years. A product like the 360 produces better sound than you have any right to expect in such a small, compact "system," though it still leaves much to be desired. Most department store type radio-phonograph combinations, even the "console" models, don't do as well, the money you pay being poured into wood rather than sound.

But from here the next step is an abrupt transition upward in sound quality—when you begin to look in "hi-fi" catalogs for components—separate amplifiers, tuners, speakers, turntables, and so on. Even a $55 amplifier, with other components amounting to perhaps another $150 or so, gives you a real system, with sound comparable to almost nothing obtainable otherwise. And when you get some of the really first-line components and put them together, you have the kind of sound for which we can no longer find any unabused name.

But even here, let us soft-pedal the extravagant touch with a nice, controversial, flat statement: If you spend more than about $500 to $600 for components (with speaker cabinet but no other woodwork), you will have to be extraordinary indeed to hear your money's worth—and most of those who can are so busy picking apart the sound quality they have no time to hear the music. The quality of components rises much more slowly than their cost. To an Ivory-soap percentage of ears, a $200 amplifier does not sound anything like twice as good as a $100 amplifier. There is generally an improvement with higher price, but be very sure you can hear it before you pay for it. The Bogen DO10 amplifier (see photo), for instance, costs only around $40, but its sound will amaze and delight people who have been used to ordinary instruments. The McIntosh 50-W2 (whose photo also appears here) is a far better performer (at about six times the price)

*Audio Consultant. N. Y. C. The name is fictitious.

†Audio Consultant. N. Y. C. The name is fictitious.
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on several counts—but the newcomer to good sound (or even some of the self-styled "golden ear") won't appreciate it six times as much.

The same is true of other components: You can buy a tuner for $50 or for $200, a speaker for $20 or a speaker system for $500. If you have the money, by all means buy the better one. But if you can't ascend to the heights, don't think the modest hills aren't worth while.

For a really satisfactory system at any price, certain rules do apply. For instance, a dual speaker system (woofer and tweeter) is almost always more satisfying than a single speaker and is usually worth the extra cost. A tuner with a.f.c. for FM is a very fine thing. A preamplifier or control box with step type equalizers tailored to the various phonograph curves is almost a necessity, while the presence of continuous "tone controls" is less important. A loudness control rather than a simple volume-control potentiometer is a big contributor to good sound.

Magnetic phonograph cartridges are no longer the only choice. There are good crystals and ceramics, though it is true most are not the equals of a good magnetic. Equals do exist, however. Good crystals, and ceramics we predict, will come into strong competition when people learn enough about them to avoid instinctively grouping them with the old-fashioned crystal groove-grinders.

Diamond styli on pickups, at least for LP use, are cheaper in the long run than "precious metal" or sapphire points. True, there is a big difference between their initial prices, but there is a much bigger difference in the length of time they last, and they keep records in good condition right up to the end of their lives. Record changers are not ideal for either record or stylus life because of the constantly changing vertical arm angle as the turntable fills. This is minor and the worst objection to even good changers is that the pickup arm is rarely if ever correctly made with respect to lack of audible resonance. There is little reason to buy a changer if only LPs are to be played, since they play as long as 20 minutes a side and must be turned over.

Actual choice of components is something that people are too concerned over. Today's audio-components market is a large one, with many manufacturers making equipment in a variety of price ranges, styles and ratings. It is not difficult, therefore, to fill any kind of requirement, and it is rare indeed that any one brand and model of component is very much superior to others in the same price range.

The sound departments of the dealers are there for your use and you should use them. Before the visit, decide on a budget. For about $200 you can get a system (complete except for cabinetry but including a speaker enclosure). This is about the minimum and will satisfy a very large number of people—and we are not speaking in a contemptuous tone about people without discrimination. For $400, under the same conditions, you may not get a great deal of improvement in sound (except perhaps in the loudspeaker department) but you will have a wider choice of "luxuries," such as tuners with a.f.c., more flexible control systems, and the like. For $600 you can buy a system with very good components indeed, and for $800 you can have the best of amplifiers and an excellent speaker system.

If you spend $1,000 or more, you are, acoustically speaking, in the yacht- owner's class, with the best of everything. This may satisfy some inner craving but it is hard—at least for this writer—to see how it can be justified in terms of sound—unless, of course, you are doing something fancy like wiring a whole house or installing a binaural system. If you spend more than about $1,200, you are paying some one an annuity and should require an itemized list with prices and compare it with readily available catalogs.

All of this, of course, leaves out cabinetry for containing amplifier, tuner, turntable and controls, but one of the joys of this game is the originality and artistry you can summon forth in making, having made or utilizing existing or antique-shop pieces for this purpose. Costs of cabinetry can range from almost zero for a home-made job to fortunes for converted period pieces.

If you are a technician or skilled hobbyist, you will quickly see how to connect the components together, for manufacturers today usually provide everything—jacks, plugs and full instructions. If not, ask the dealer to furnish cables, with verbal (and drawn) instructions.

One last word, a bit of special pleading occasioned by too many hours of having our ears assaulted by drums, triangles and trumpets, coupled with admonitions to "Just listen to those highs!" and "You really feel that bass, don't you!" There is little point in wasting several hundred dollars and a lot of time and space, followed by nothing but "high-fidelity" demonstrations to yourself and your friends when the local music store can sell you a set of real drums and a couple of triangles for less money. The way to enjoy and demonstrate your system is to play music. If you like Eddie Fisher, put on Eddie Fisher records, adjust the equalizer and volume, and let him send you. If you like chamber music, put on a good quartet and allow everyone present to revel in Ravel. But, just as a Parker 51 is useless unless you use it to write intelligibly, the best music system is so much expensive junk if you don't use it to reproduce music.

As Shakespeare used to say, the theatre is beautiful, the air-conditioning perfect, and the costumes magnificent—but, the play's the thing!  

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Above—The new Ronette crystal cartridge.

Left—McIntosh model 50-W2 audio amplifier.
DEMONSTRATION TRANSISTOR CIRCUITS

By EDWIN BOHR

Kit for the transistor experiments.

TRY all of these circuits and really get the feel of transistorized equipment. Experiment for personal pleasure or for an audience—the layouts are ideal for club demonstrations.

A single transistor is used in each circuit. Other parts are standard and easily obtained—if not available from the junkbox. Every circuit has been checked out with several CK722 production transistors and found to be foolproof. (A demonstration circuit must work without embarrassing tinkering and fumbling.) Components are mounted breadboard style, allowing the individual circuits to be connected or torn down quickly. Since a single transistor is involved in all the circuits, it should be especially protected from continuous handling and soldering.

Two types of breadboard mountings for the transistor are used. In one layout, the leads are run to Fahnestock clips held to a heavy base of clothboard or masonite by 6-32 screws. This base should be heavy and large enough to prevent accidental upsets. The other breadboard model sports a small five-prong hearing-aid tube socket into which the transistor can be plugged. Only the first, third and fifth pins of this socket are used. The other two pins are given a slight twist and pushed out. It is a good idea to mark clearly which lead is the emitter, base and collector. Another suggestion: Slip spaghetti over all wires that might touch or short.

Microwatt oscillator

The most remarkable feature of the transistor is its ability to perform useful electronic tasks with a minimum of power consumption. The first project is an audio oscillator requiring a total power input of 4 microwatts—a perfect example of low power consumption! A simple electrolytic cell, constructed from a dime and a piece of absorbent paper, supplies the energy.

The circuit (Fig. 1-a) must use transformer feedback to operate at this low power level. For this purpose, and in keeping with the size of the transistor, a tiny U.T.C. SO-3 or SSO-3 subounce transformer is used. Larger transformers will do, but the data given in this article is for the SO-3.

Fig. 1-a—The microwatt oscillator.
Fig. 1-b—Wiring diagram for the SO-3. Fig. 1-c—Experimental "power supply."

The exact connections for the SO-3, with the correct feedback polarity, are shown in Fig. 1-b. One wire from the "plate" side of the transformer goes to the collector and the other wire is connected to one side of the headphones. The two low-impedance wires from the transformer are connected to the base and emitter.

Either grounded-emitter (point 2) or grounded-base (point 1) operation is available by switching the wet-cell connection. Operation of the two circuits is essentially the same, but the grounded-base circuit gives a tone of slightly higher pitch. The frequency of the oscillator changes with the voltage of the wet cell. Increasing the voltage lowers the tone.

Make the wet cell (Fig. 1-c) by placing a piece of absorbent paper moistened with saliva against a dime. The dime forms the positive electrode and a wire lead held against the other side of the moist paper provides the second electrode. A few drops of soft drink also make an excellent electrolyte.

Several of these dime-saliva cells were checked on a potentiometer and the voltage was always around 0.7. When the cell is connected to the transistor oscillator, the voltage drops to about 0.5. With this voltage, the oscillator draws 8 microamps of collector current—a total input power of .000004 watt!

By making a megaphone out of paper and placing it to one of the earphones, the tone can be heard several feet away. For larger groups, an audio amplifier input can be connected directly across the earphones for room volume.

The microwatt oscillator operates from other fee-power sources. Replace the saliva cell with a self-generating photocell and the oscillator will "sing" from the power produced by ordinary room illumination. Automobile headlights at night as far as 30 feet away will give enough light for oscillation. The tone changes with light intensity. This could possibly be used as a blind man's light meter or in a steering device to bring an electronic animal to its nest.

A peak a.c. output voltage greater than the applied d.c. voltage can be obtained from transistor oscillators. This offers some novel transistor uses. For example, it is often necessary to send small d.c. signals over long-wire distances from remote locations. The microwatt oscillator can operate from these small voltages and, at the point of origin, change them to a.c. for more
convenient transmission and amplification.

High-gain radio

Astounding! is the only way to describe the feelings of most people when they first hear this set (Fig. 2) in operation. Only a single tuned circuit, transistor, SO-3 transformer and a battery are used.

The tuned circuit is coupled to the emitter by a tap on the coil, a 6SA7 oscillator type—only here it is used as an antenna coil. A broadcast tuning capacitor (410 μF) covers the band with perhaps only a few stations missing on the low end of the dial. The tuning is broad enough so these stations will be picked up anyway.

The high-gain radio demonstration.

A small 0-100-microamp meter connected in series with the collector makes a very good tuning indicator. With a single dry cell for current, and no signal, the collector draws about 10 microamps. As a signal is tuned in, the current increases. A signal producing 20 microamps of collector current is loud enough to be heard faintly 3 or 4 inches from the earphones, while 40-microamp signals are exceedingly loud. Generally, a ground is not needed, but it does eliminate hand-capacitance tuning effects. For demonstrations, just touch the tip of a transformer type instantaneous-soldering iron (plugged in but not turned on) to the antenna terminal of the receiver. This makes a very effective antenna with a whopping big signal of over 50 microamps. For even more volume, simply increase the collector supply to 3 volts.

Audio amplifier

The secret of the receiver's performance is the SO-3 matching transformer. Ordinary medium-impedance earphones connected directly in the collector circuit do not realize the full gain possible from the transistor. But, with the circuit of Fig. 2, much higher gain is possible. Even Army type moving-coil (low-impedance) earphones connected across the output winding give very good performance.

A small 0-100-microamp meter connected in series with the collector makes a very good tuning indicator. With a single dry cell for current, and no signal, the collector draws about 10 microamps. As a signal is tuned in, the current increases. A signal producing 20 microamps of collector current is loud enough to be heard faintly 3 or 4 inches from the earphones, while 40-microamp signals are exceedingly loud. Generally, a ground is not needed, but it does eliminate hand-capacitance tuning effects. For demonstrations, just touch the tip of a transformer type instantaneous-soldering iron (plugged in but not turned on) to the antenna terminal of the receiver. This makes a very effective antenna with a whopping big signal of over 50 microamps. For even more volume, simply increase the collector supply to 3 volts.

Audio amplifier

The grounded-emitter circuit has become the established transistor audio circuit. There are two reasons for this: First, the gain is highest and, second, the proper bias voltage can be obtained most easily and with the least wasted power by grounding the emitter.

The transistor emitter is basically a rectifier circuit and a bias voltage must be placed on it to cause forward current to flow. For the CK722 this voltage must be positive. Without this bias, any negative swing of the emitter signal would drive the emitter into its "backward" region and produce collector cutoff distortion. In the grounded-emitter circuit, the base is made slightly negative by a series-dropping resistor from the negative collector supply, which is equivalent to making the emitter positive.

Fig. 3 is a typical amplifier circuit of this type. The input impedance being roughly 1,000 ohms, it can be fed from a single magnetic headphone used as a microphone, the transistor radio or a variable-reluctance cartridge.

To try out the circuit, connect a single earphone to the input and another to the output. Place the input earphone (microphone) in one room and the output one in another room. Sounds near the microphone can be heard clearly. The transistor amplifier can be built complete with a mercury-cell power supply in a volume as small as that of the earphone. You can't do this with a vacuum tube! On the debit side, the transistor amplifier has a noise level higher than an equivalent vacuum-tube circuit. This noise, heard as a soft hissing sound in the earphones, is not obtrusive.

Utility oscillators

The utility oscillator is a demonstration circuit operating from a single-
cell. It is useful as a self-contained signal source for Wheatstone bridge circuits, toy musical instruments or an audio test generator. When operating at very low frequencies (60 cycles or less), the output is sufficiently rich in harmonics to radiate signals into the lower end of the broadcast band. The circuit of Fig. 4 is the same as Fig. 1, except for the power source and a variable resistor inserted in the emitter circuit to vary the frequency. Figs. 5-a, b, c and d give the waveshapes and frequencies obtained as the resistance is varied from zero to 7,000 ohms. Notice that the waveshape improves and the harmonic content decreases as the emitter resistance is increased.

Output is sufficient to operate headphones or a speaker at very low volume. Either earphones or the speaker output transformer may be connected to the output terminals of the oscillator shown in Fig. 4.

A short antenna can be connected to the collector for signal radiation. The audible tone generated by the oscillator can be heard all across the low end of the band where it is strongest. Just turn on any radio within a few feet of the antenna.

Fig. 4—Variation of Fig. 1 oscillator. Frequency controlled by potentiometer.

Another utility oscillator circuit is shown in Fig. 6. This is an "easier starting" circuit than that of Fig. 4. Some circuit disturbance is always necessary to start an oscillatory circuit in operation. A pendulum or tuning fork, for example, must be given that first push or impact to start it oscillating at its natural frequency. Vacuum tubes receive this push the instant the plate voltage is applied. When the plate voltage is switched on, the grid has zero bias and heavy plate current flows. This heavy surge of current excites the tuned circuit which oscillates and in turn builds up a grid-leak bias that limits the plate current to a lesser value.

With the transistor, this same situation is not obtained. Like the vacuum-tube grid, the emitter also has zero bias at the instant the collector voltage is switched on. But, conversely, zero emitter bias allows very little collector current to flow, since a positive emitter bias is needed to increase the collector current to a point higher than the normal back current.

The Fig. 4 oscillator develops a starting bias across the variable base resistor. As the collector starting current flows through this resistor, it places a positive bias on the emitter.

Fig. 6—An "easy start" oscillator.

Loading the circuit with an antenna or hand capacitance changes the waveshape considerably but causes little if any frequency variation.

These circuits will provide plenty of new entertainment and experimental material. You will hate to disassemble some of them—especially the radio.

Parts for transistor demonstration
1-CX272 transistor; 1—50-3 (U.T.C.) transformer; 1—5A4G oscillator coil; 1—section tuning capacitor, 410 µf. 1—5,000-ohm potentiometer; 1—resistor, any value from 100,000 to 270,000 ohms; 1—electrolytic capacitor, 25 µf or greater; 1—mercury or dry cell; several Fahnestock clips.

However, even one transistor is too expensive to limit to any single circuit. Furthermore, the clip-breadboard technique makes it possible to set up any of the other circuits in short order.

DECEMBER, 1954
A receiver that can be installed conveniently in small-fry vehicle—realism beyond the imagination

By JAMES S. MICHAEL

ONE of the most prized possessions a child can have is his or her own automobile. Unfortunately, manufacturers of these little autos do not provide a line of accessories such as is available for dad's car. To make my little boy's fire engine more realistic, I installed a radio. The result of my trials and tribulations is shown on the front cover.

The first step was to see how much room was available for the installation. In this and just about all other children's vehicles on the market, there is ample room for a decent-sized radio. I decided a 3- or 4-tube circuit would give the desired results.

After some experimentation I settled on the simple 4-tube t.r.f. circuit shown in the schematic. The sensitivity and selectivity of the receiver are about the same as in most t.r.f. circuits without an extra r.f. stage. For my purposes it was very satisfactory.

The circuit is very flexible and can be adapted to almost any installation. Only the usual construction requirements need be followed, such as keeping the plate and grid leads separated.

I used unshielded antenna and r.f. coils because they were on hand. But an improvement could be made by using shielded or iron-core coils.

The unit is powered by a 67.5-volt B battery and two flashlight cells, installed in a separate container. The radio draws very little current, resulting in long battery life.

The r.f. and detector stages (The 1T4 detector circuit is operating without a grid-leak resistor and capacitor in the grid circuit. The author reports satisfactory detection with the circuit as shown.—Editor) use 1T4's. A 1S5 is used as the first audio amplifier. It drives a 3V4 power amplifier. The speaker is a 2-inch job small enough to install in the dashboard and helps along the realism.

In constructing the unit, one precaution must be observed. Do not use a s.p.s.t. switch in the common negative lead to turn the unit on and off. It will be much cheaper to buy a d.p.s.t. and use it as shown. Or if you must buy a.

Diagram of Junior's car radio. Unit is not critical—will stand rough treatment.
NOW—THE TRANSISTOR POCKET RADIO

The transistor portable, predicted by many as a certain 1955 development, is now in production. It is a super-heterodyne and uses four transistors of new design plus a diode detector. The set is only 3 x 5 x 1½ inches in dimensions and weighs less than 12 ounces. Design of such a set is made possible by new high-performance transistors developed by Texas Instruments Inc., Dallas, Texas. Power is supplied by a single 92.5-volt battery, whose life is naturally expected to be very long.

The little set is made by Regency (a division of Industrial Development Engineering Associates) of Indianapolis, Ind., and is priced at $49.95. It is expected to be on the market in time for the Christmas trade.
A practical approach to the understanding of this popular circuit

Getting the Cathode Follower Straight

By NORMAN H. CROWHURST

THE well-known cathode follower is a very useful but frequently misunderstood circuit. The commonest misconception concerns its use to produce a low output impedance. It is true the follower circuit does produce a low output impedance, but the kind of misunderstanding referred to can be seen in a recent case where the circuit was used as a buffer stage to isolate an oscillator tube. The essential requirements were high stability and good waveform. The circuit designer assured me the follower would feed satisfactorily into a 600-ohm line. I had my doubts, which proved well founded.

Let's not get involved in formulas at this stage. Other articles on cathode followers have done a good job of explanation from this approach. Instead, let's take a diagram of a cathode follower and put some figures in. A cathode follower is essentially a tube in which the plate load is connected between cathode and ground instead of its usual place between plate and B plus (Fig. 1).

Start by assuming that the signal voltage from grid to cathode is, say, 1 volt. This is the tube's viewpoint—it can have no way of knowing what the voltage is from grid to ground, although we may prefer that starting point—and it will control the plate-cathode voltage and current. Suppose that, with the plate load chosen (connected in the cathode), the tube has a gain of 20. There will be a signal voltage of 20 across the plate load resistor, in this case from cathode to ground.

We find the grid to ground signal voltage by adding these two—21 volts. Not sure of the phase? Well, a positive swing of grid-cathode voltage increases plate current, increasing the drop across the load to produce a positive swing at the cathode. So the two are additive.

Thus, 21 volts input, from grid to ground, will give an output, from cathode to ground, of 20 volts.

What happens when we connect a 600-ohm load (Fig. 2)? Let's take the figuring further: Suppose the tube has an amplification factor of 30, a plate resistance of 15,000 ohms, and the plate load resistor is 30,000 ohms. This works out to give a gain of 20 without feedback. To complete the tube values, its transconductance would be 2,000 microhmhos. From the tube's viewpoint, for signal voltages, its plate load is effectively 30,000 ohms in parallel with 600 ohms—not much less than 600 ohms.

The gain of this tube with a 600-ohm load will be $20 \times \frac{15,000}{600} = 1.2$. If the input is still 21 volts, this voltage will divide so that the voltage across the cathode load (11.5) is 1.2 times the voltage from grid to cathode (9.5).

Connection of the 600-ohm load has dropped the output voltage from 20 to 11.5.

This drop can be regarded as due to the effective source impedance of the cathode follower. As the drop is $20 - 11.5 = 8.5$ volts, the effective source impedance, calculated by ratio with the load impedance, can be called $600 \times \frac{8.5}{11.5} = 440$ ohms. The accurate value of the cathode follower impedance could be calculated directly from the formula $Z_{eff} = \frac{R_P + R_L}{R_P + R_L + R_F}$.

Take a look at what has happened to the grid voltage. Before the 600-ohm resistor was connected, the grid to cathode signal voltage was only 1 volt; when the 600-ohm resistor is connected, this signal voltage rises to 9.5 volt. Think a moment and you will realize what the arithmetic failed to tell us. Do you know of a tube with 15,000-ohm plate resistance, 2,000 microhmhos transconductance, that will handle 9.5 volts signal on its grid—requiring a grid bias of at least 23 volts? But one to handle 1 volt would be feasible, wouldn't it? So adding the 600-ohm resistor would obviously produce distortion.

At this stage, many people come up with the argument: "But what about the feedback—won't that take care of the distortion?" To which we can imagine our poor little overworked tube squeaking up: "What feedback?" Look again. When the tube works into the changed plate load of 600 ohms, its gain is knocked down to 1.2. Before the load was connected, the tube gain was 20, reduced by feedback to 0.95—a reduction of a little over 20 db; with the load connected the gain is 1.2, reduced by feedback to 0.575—a reduction of a little over 6 db. And a mere 6 db will not do much toward cleaning up a waveform produced by giving an input of 9.5 volts to a tube intended to accept not much more than 1 volt.

All right, someone says, but all you're pointing out is that using a low load on a cathode follower restricts the signal it can handle. But keep within a smaller signal and the cathode fol-
lower will still produce lower distortion than other circuits, won’t it? To satisfy you, let’s go into this a bit further.

Suppose we do turn the input down from 21 volts grid-to-ground to about 2 or 3 volts, so as to keep within the grid-handling capability of the tube. Take a look at the tube characteristics shown at Fig. 3: The load line AB represents a plate load resistance of 30,000 ohms; assuming the circuit in Fig. 4 provides a bias of 2 volts, the load line CD represents the result of connecting 600 ohms in parallel by a.c. coupling through the capacitor. Notice that the line AB crosses the curves representing successive grid voltages at uniform intervals, while CD cuts at an angle so the spacings taper off noticeably at the bottom end, while the top end extends into high plate current. This means serious distortion, even before full grid swing is reached, and with only 6-dB feedback the distortion left will be much more than is expected of a respectable amplifier.

Otherwise expressed, the feedback has been “used up” to provide impedance matching and not enough remains to take care of distortion or for other purposes.

Isn’t it surprising how resentful people can get when they think an old favorite is being attacked? So some, to whom the cathode follower has seemed to be a valiant knight slaying all the electronic bugs, take a fighting stand. When triodes aren’t “good enough” in other circuits we turn to better tubes, such as pentodes; pentodes have been used successfully as cathode followers, the circuit of Fig. 5 being typical. Doesn’t this arrangement enable low distortion to be obtained into a load value, since pentodes “like” working into low impedances better than triodes?

It’s not as easy as that. Continuing with the example figures used: With a triode, reduction of the plate load from 30,000 to 600 ohms—a ratio of 50:1—only reduced the gain from 20 to 1.2 because of the triode’s relatively low plate resistance. But because of the high plate resistance of a pentode, the corresponding reduction would be almost in the same ratio as the change in plate load resistance. Even if the pentode had a high gain to start with, loading it down with 600 ohms would reduce the gain as with the triode, so again the effect of feedback in cleaning up waveform would be negligible. In the end, the pentode does no better under these conditions than the same tube used as a triode.

Now don’t start throwing out cathode followers everywhere. Let’s get things really straight by recounting what they can do: It is true that they reduce output source impedance. Their limitation is that this feature must not be imposed upon by assuming that we can go ahead and use a load impedance much lower than the regular load impedance for the tube used. If we try that, we run into trouble.

Another good use for cathode followers is to obtain a very high input impedance. This they also do very well, provided the plate load (connected in the cathode) is kept high enough to preserve the normal gain of the tube.

Now let’s look around to see how all this affects the circuits commonly used. Probably the most frequent everyday application of the cathode follower is in the output of preamplifiers to provide a low source impedance suitable for line-impedance connection between the preamplifier and the main amplifier. But before throwing that poorly operating preamplifier out, check what the main amplifier input impedance does to it. If it has a line-to-grid transformer, the cathode follower is undoubtedly being loaded down too much, and distortion occurs. If the main amplifier is arranged for input direct to grid, the cathode follower will be quite happy, and its low impedance will give the advantage of line impedance coupling between amplifiers, i.e. freedom from loss of “higs” and from “static” pickup.

Summarizing the foregoing further, the right answer works out this way: if the main amplifier uses transformer input, either for phase splitting or to save gain or both, then the preamplifier output should be properly matched down through a transformer, whether plate or cathode coupled. On the other hand, if the input end of the amplifier is all resistance-loaded, with nice high-value resistors, the cathode follower direct is O.K. in the preamplifier output. So take another look at those circuits: you may rid yourself of an unsuspected source of distortion.

PHILCO'S NEW EASILY INSTALLED ADAPTER

U.h.f. Tuner-Adapter has been announced by Philco (see photo). The unit consists of a u.h.f. tuner, adapter cables and plugs, a planetary tuner driving assembly and mounting hardware. A feature of the adapter is that it can be installed without removing the chassis or cabinet back.

DECEMBER, 1954
NOVEL HIGH-VOLTAGE SUPPLY

By HAROLD PALLATZ*

The novel high-voltage supply in Fig. 1.

The design and construction of high-voltage, low-current power supplies for portable photoflash units, G-M counters, breakdown and insulation testers, meggers, and similar devices poses a number of problems to the engineer and constructor. The usual solution is to use high-voltage B batteries, r.f. power supplies, hand-cranked generators, or vibrator-type supplies operated from a storage battery. In each of these cases, the supply is bulky, heavy, and inefficient.

This newly developed vibrator converter is designed to be used as the basic component of compact, lightweight, and efficient high-voltage, low-current power supplies. Basically, the unit consists of an efficient step-up transformer with a built-in vibrator. Current flowing in the transformer primary provides the magnetism to operate the vibrator. The secondary output is a series of high-voltage pulses which may be rectified and filtered to provide a constant d.c. voltage. The output can be varied between 500 and 6,000 volts by connecting suitable shunt resistors across the load.

Input voltage is 1.5 to 6 volts d.c. or 60-cycle a.c. at about 300 ma. Standard size-D flashlight cells operate the converter nicely. Cells designed for photoflash service are recommended because of their reserve capacity.

For reliable vibrator operation spring brass is used on the vibrating reeds to provide low-resistance paths. A double leaf-spring arrangement is used to make the contacts self-cleaning. A silver-tungsten alloy is used for the contacts. This produces low-resistance contacts with high resistance to welding and tarnishing.

The circuit and constants

A typical experimental supply delivering up to 6,000 volts d.c. is shown in the photo and in Fig. 1. The rectifier is a subminiature 5642. Other recommended types are the 1X2-A, 1B3-GT, 1B54 and 2X2A. The latter should be used when power rather than physical dimensions is the determining factor.

High-voltage selenium rectifiers may be used when filament connections become complex as in voltage doublers and bridge circuits. Selenium rectifiers for this service should have an inverse rating of at least 10 kv. The CK5517 cold-cathode gas-filled rectifier tube is usable but requires careful circuit design for reliable results.

For most applications, a .005-µF 6,000-volt filter capacitor will do. Electronic photoflash units require much larger capacitors. Suggested values are 2 to 40 µF with voltage ratings up to 4 kv. Charging time depends on the size of the capacitor, the final voltage, the primary voltage and current, and the type of rectifier. An 8-µF capacitor can be charged to 1,000 volts in about 15 seconds using a 2X2-A rectifier and 2.5 volts a.c. on the primary. Charging time with d.c. input is about the same if the battery resistance is low. Fig. 2 shows how the output of the supply in Fig. 1 varies with load resistance and current.

Vibrator hash can be reduced by connecting a buffer capacitor directly across the high-voltage (a.c.) output of the vibrator-converter transformer. The capacitor value depends on the circuit used. About .005 µF is a good starting point. A buffer that is too large reduces the output voltage.

Voltage-regulator circuits

Various circuits have been used with the supply to stabilize its output voltage. Fig. 3-a shows the circuit for 500 volts output. The regulator consists of a fairly heavy filter capacitor and nine series-connected NE-2 neon lamps shunted across the output. No-load stability is better than 2%. When using a 2X2-A rectifier and 2.5 volts, 60-cycle a.c. input, output regulation was better than 4% with a load ranging from 0 to 250 ma. We haven't tried the neon-lamp regulator above 500 volts but we feel that it will work up to several thousand volts if enough lamps are used in series. Intermediate voltages can be tapped off.

Fig. 3-b is the circuit that we use for high voltages. A capacitor and a CK5517 rectifier are paralleled across the output of the supply. At 2,500 volts output, the regulation is 6% with current drains up to 40 µA. Note that the tube must be connected with its cathode going to the positive side of the supply and its plate to the negative. If the tube is reversed it operates as a relaxation oscillator and the voltage drops to a low value.

The vibrator unit was one specially developed for the application (Precise Measurements model 5-MVT). Readers who wish to construct their own supply along the lines of the 5-MVT vibrator converter may also try a model aircraft ignition coil as the transformer, with parts from a doorbell or buzzer as the vibrator. Experimenting and careful adjustment may be necessary because it is unlikely that the vibrator will match the coil closely enough for high efficiency. It may also be necessary to raise the battery voltage to 4.5 or 6.

*Precise Measurements Co., Brooklyn, N.Y.

RADIO-ELECTRONICS
WHILE self-educating myself in TV, I accumulated a rather large collection of “dgos” — receivers bought for training purposes only. As time passed and it became more and more evident that these 10-inch sets would never be worth repairing, I salvaged what I could, and threw the rest away—except a certain RCA which I converted into a portable and serviceable TV set, but which eventually became quite a valuable piece of equipment.

In the course of time this old RCA 10-incher has been pretty well breadboarded (see photo). It’s mounted on a low (24-inch) table. The chassis is on top of that and bolted through into the wooden shelf. All vital parts not easily accessible have been brought out and draped around the chassis edges in various panels, sets of tie-points, etc. Together with a few patch cords made up from 300-ohm line, some of them broken with decoupling capacitors, the old chassis has become the dandiest little dynamic substitution box a man could want.

With a piece of gear like this it’s easy to quickly and accurately double-check the occasional picture tube which is apparently kaput, but where some shadow of doubt still exists. Just roll the box up to the back of the customer’s set, plug the C-R tube socket of the former into the C-R tube base of the latter, or vice versa if the situation calls for it. It takes practice. If you are worried about blowing anything up—which is possible, especially where a-c-d-c sets are involved, insert 1/4-amp SAG instrument fuses in the potch cords.

To cite another use, a sure method of checking a suspected video amplifier is by cutting it loose from its detector load resistor and patching in the know-to-be-good video amplifier in your dynamic box. Or vice versa if your angle of attack calls for it.

Need 10 kv to supply another set you’re checking, or to flash out a short in a C-R tube? Fire up the substitution box. Running down a toughie in the horizontal oscillator circuit? Just drive the suspected output stage with the 15,750-cycle sawtooth from your substitution box and narrow things down in a hurry. Or vice versa, again. You take it from here.

A chassis like mine is an absolute necessity when the TV student getting his experience at home arrives at the “scope stage.” A scope is a fearful piece of machinery for a student working all by himself at the outset; it’s only human to dodge the issue as long as possible. That’s exactly what I did, until I ran into a really rough case of sync separator trouble—and no amount of static checking did me the slightest good. It was a job for a scope and nothing else. However when I finally got mine into working condition, I couldn’t tell whether those patterns were right, wrong, upside-down, or hind-side-to. Sure, I’d done some practicing with that scope, observing patterns of sets in good working order, but there are quite a few patterns, and I don’t have a photographic memory. That’s when I “invented” my dynamic substitution box.

It gave me a standard of comparison. The instant I got lost in checking a defective set with the scope, or the instant I wasn’t sure whether this pattern or that one was good or bad, I could pretty well get an answer by checking a similar part or stage in my box. It’s lot like having a TV shark at your elbow all the time, to straighten things out when required.

Similarly, for the student sweating it out all alone, a chassis like this provides the easy way to get a working knowledge of sweep and signal generators in general. Learning this angle is a rough business; merely setting things up properly is a headachy assignment for the solitary novice. The lessons must be rigged so that he learns to walk before he tries to run—and a breadboarded TV set is just what he needs.

You work out on simple things first: an i-f stage screwed out of alignment and restored; ditto for an adjacent-channel trap, then sound pickup, sound trap, and so on. Try to take notes on symptoms; it might save you hours later on in busting some job in the “stinker” category.

The usefulness of this kind of dynamic substitution box extends in a great many directions for both student and service technician. The moment a meter is suspected of having gone haywire, or whenever the scope seems to be acting abnormally, if what I get through a scope doesn’t seem to make sense, I can settle all suspicions within 60 seconds. If for instance, my grid-dip oscillator persists in acting half-witted when I try it out on the box, the answer is obvious. Equally obvious is the answer when a crystal probe behaves normally in the box but plays dead when applied to the customer’s set.

Then there are the instances when resort to the substitution box quickly shows I’m the one who’s acting half-witted. I recall in particular when—after a particularly long day—the substitution box informed me the reason there was no plate voltage where I knew there was plate voltage lay in the fact that the range switch on the v.t.v.m. was in the wrong position! And had been there for about fifteen minutes!

As a final, parting shot of sage advice, I will tell you what I do in a case like this. I forthwith close up shop, go home, and spend the rest of the evening watching the rasslers and the Westerns. If anyone has a better prescription I will be pleased to hear from him.
HIGH values of current amplification are obtained in the grounded-emitter junction transistor circuit by using the base as the input electrode. A small change in base current produces a rather large change in collector current.

Such base-to-collector current amplification is designated by the Greek $\beta$ (beta), and in commercial transistors has a value many times higher than the familiar emitter-to-collector amplification $a$ (alpha). $\beta$ is equal approximately to $a/(1-a)$. From this relationship, we see how a junction transistor having an alpha of only 0.909 can give current amplification of 10 in the grounded-emitter circuit. It is clear also that junction transistors with the highest alphas, approaching 1, also show the highest betas.

High-beta performance permits operating the grounded-emitter circuit as a d.c. amplifier with low power drain. The amplifier has a current gain of 10 when alpha is 0.909 and can have a gain of approximately 100 when alpha is very slightly greater than 0.99. Taking advantage of this gain, a d.c. milliammeter in the collector circuit will respond to microamperes applied to the base-input circuit.

Fig. 1-a shows such a simple amplifier type microammeter. Input current of 100 microamperes will deflect the 0–1 d.c. milliammeter to full scale. For simplicity, zero-setting circuits for bucking the static collector current out of the meter have been omitted from each circuit in Fig. 1. In Fig. 1-b, a second common-emitter amplifier stage has been added in cascade for 10-microampere operation of the meter. Note that transistor V2 has been "turned over" so that current flows through its base in the correct direction to increase collector current. The collector current of transistor V1 flows directly through the base-emitter input circuit of V2, so no load resistors are required and interstage impedance-matching problems disappear. Transistor V2 is operated in a class-C fashion, with the plate of V1, plate of V2, and the a.c. lead to supply both transistor stages. This has been done in the final circuit shown in the Fig. 1-c diagram.

It is entirely possible to use low-priced junction transistors, such as the CK722, in this d.c. amplifier circuit. However, for a total current gain of 100, the transistors would have to be selected for a minimum alpha of 0.909 each. There is gain to spare when using a combination of one CK722 and one higher-alpha unit, such as the CK721, in cascade. This combination requires no special picking of transistors.

An interesting and useful application of the transistorized microammeter is as the basis of an electronic d.c. voltmeter—the transistor equivalent of the v.t.v.m.—by adding suitable multiplier resistors to the input of the d.c. amplifier circuit. Since the two junction transistors require only about 2 ma from a 1.5-volt cell for complete circuit operation, the result is a completely portable electronic voltmeter having operating economy not obtainable with conventional battery-operated v.t.v.m.'s.

A 0–10 d.c. microammeter has been shown before as the basis of a d.c. voltmeter having 100,000 ohms-per-volt sensitivity.1 A 10-microampere d.c. meter is expensive, however, and not obtainable except on special order. The transistorized d.c. amplifier permits use of the more rugged, inexpensive, and readily obtainable 0–1 d.c. milliammeter.

To readers who compare a sensitivity of 100,000 ohms per volt unfavorably with the input resistance of the conventional v.t.v.m., we would like to point out that the 100 volt range has an input resistance equal to that of many v.t.v.m.'s, and on all higher ranges the transistor voltmeter has a higher input resistance than the v.t.v.m.!

Instrument circuit

The complete circuit of the transistorized electronic voltmeter is shown in

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Fig. 2. To the simple two-stage transistor microammeter circuit of Fig. 1c have been added the input multiplier resistors (R1 to R5) and the zero-setting meter circuit (R7 to R10). The zero-setting circuit is of the four-arm-bridge type common in v.t.v.m.'s.

Resistors R1 to R6 must be selected to the exact specified values. The 50-megohm value required for the 500-volt range is obtained by series-connecting one 10- and two 20-megohm resistors. For highest accuracy on the 1-volt range, the input resistance of the CK722 (approximately 2,000 ohms) should be subtracted from the normal 100,000-ohm value of R1, making it 98,000 ohms. If this is not done, the 1-volt range will read 2% low. To check the input resistance of the first transistor in the complete circuit, feed in an input current of 10 microamperes and measure the voltage drop between base and emitter. Determine the resistance by dividing the voltage by .00001.

One all but the last range, the 0–1 scale of the milliammeter can be used by merely adding zeros mentally where necessary. The author found the 1-, 10-, 100-, and 500-volt ranges suitable for his purposes. Other ranges may be included if those shown are undesirable. The table shows multiplier resistor values for common voltage ranges other than those shown in Fig. 2.

### ADDITIONAL VOLTAGE RANGES

<table>
<thead>
<tr>
<th>Range (volts)</th>
<th>Multiplier Resistance (megohms)</th>
</tr>
</thead>
<tbody>
<tr>
<td>2.5</td>
<td>0.25</td>
</tr>
<tr>
<td>3</td>
<td>0.3</td>
</tr>
<tr>
<td>5</td>
<td>0.5</td>
</tr>
<tr>
<td>7.5</td>
<td>0.75</td>
</tr>
<tr>
<td>15</td>
<td>1.5</td>
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<tr>
<td>25</td>
<td>2.5</td>
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<td>50</td>
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<tr>
<td>250</td>
<td>25</td>
</tr>
<tr>
<td>300</td>
<td>30</td>
</tr>
<tr>
<td>750</td>
<td>75</td>
</tr>
<tr>
<td>1,000</td>
<td>100</td>
</tr>
</tbody>
</table>

One pole of the range selector switch disconnects the battery. For protection of the instrument, the OFF position is placed after the highest voltage range. Phone jack J1 is provided for the “high” d.c. input lead which in this circuit is connected to the negative pole of the voltage source under test. A conventional shielded input lead and probe are advantageous when working around strong fields. The “low” (positive) lead is connected to pin jack J2.

### Construction

Being a straight d.c. instrument, the problems of stray coupling and frequency dependence are absent. The model shown in the photos is built in an aluminum utility box 6 inches high, 4 inches wide, and 5 inches deep.

The two transistors and components R8, R9, and R10 are mounted on a 1 1/4 x 4 1/2-inch bakelite board attached to an inner wall of the case with long 6-32 screws. Transistor and resistor leads are pulled through small holes in the board and connections are made underneath.

Resistors R1 to R6 are soldered directly to the range switch. The flashlight cell is held to an inner wall by a curved bracket which does not appear in the photo.

Leads from the meter, battery, zero-set rheostat and battery section of the range switch are cable together and run to the component board, underneath which connections are made.

The instrument can be built much smaller than shown here. Smaller meter, components, battery and case are entirely feasible.

### Initial adjustment

After the wiring has been inspected, make the initial adjustment in the following manner: 1. Set the range switch to the 10-volt position, and set R8 about halfway between its minimum and maximum rotations. 2. Zero the meter by adjusting R7. 3. Apply an accurately known 10-volt d.c. potential to the input terminals. 4. Adjust R5 for exact full-scale deflection of the meter. 5. Remove the voltage. If the meter does not read exactly zero, reset by adjusting R7. 6. Reapply the voltage and readjust R8, if necessary, for full-scale reading. 7. Repeat steps 4, 5 and 6 until the meter reads full scale when the voltage is applied and falls back to zero when the voltage is removed.

Even when substituting transistors, the author found the circuit response surprisingly close to true linearity. This had made it possible to obtain good accuracy with the smaller meter scale. Such operation was not anticipated, since we expected alpha and beta would vary greatly with input current. However, where highest possible accuracy must be insured, the builder should calibrate as many scale points as possible during the initial adjustment. Suggested points would be 1-volt apart from 1 to 10 volts. They should be checked after completing the full-scale adjustment. If the calibration then did not follow the milliammeter scale, a special meter card might be drawn or a calibration chart prepared.

### Performance

The transistorized electronic voltmeter will give a portable rating of the instrument as a completely portable instrument having high input impedance and excellent economy of operation.

It can be used in place of the v.t.v.m., which it usually will supplement, especially in tests involving voltages which can be read on the 100-volt and higher ranges. The input resistance on these ranges equals or better than of the tube type d.c. instrument.

### Parts for electronic voltmeter

1. 2–500 ohms, 1–100,000 ohms (see text). 2–1 megohm, 2–10 megohms, 2–20 megohms, 1/2 volt, resistors; 1–200 ohms, 1–100 ohms, potential meters, wiredown or Ohmite A9 composition; 1–0.1 d.c. milliammeter; 1–phone jack; 1–pin jack; 1–1.5-volt flashlight cell; 1–25–125 ohm wirewound or Ohmite potentiometers, 1–molded rotary selector switch; 1–CK721; 1–CK722; 1–cabinet; 1–mounting board.

All operating power is furnished by a 1.5-volt cell. Total current drain, at full deflection of the milliammeter, is approximately 2 ma. A jumbo-size flashlight cell will give long life even when the instrument is accidentally left running. A penlight cell will give somewhat shorter service, while a mercury cell will very nearly give its shelf life of several years continuous operation. Operation is instantaneous, without warm-up periods, as result of battery and transistor operation.

With good transistors, zero-setting drift is negligible except during wide changes of temperature—the static collector current increases rather severely with temperature. However, the author finds that resetting zero compensates for this drift. The instrument was kept in continuous operation for 5 hours at a controlled temperature of 30° C with no zero drift.

Other operational features, such as current, resistance, and a.c. voltage ranges, may be added in the conventional manner, borrowing from v.t.v.m. techniques.
HAVING a continuously variable power supply, this tester permits application of the proper d.c. voltages for forming electrolytic capacitors—the leakage current being read on a meter—and testing paper and mica capacitors for opens, shorts, leakage by neon tube indication. The power supply consists of a grid-controlled rectifier using a 6V6 (6L6, 6F6, 42) with 500–600 volts a.c. on the plate, supplied by a standard power transformer (center tap not used). A 1,000-ohm resistor and 0.5-µf paper capacitor provide ample filtering. The resistor also serves as a protective device if the grid bias is removed or the test terminals shorted. An additional 0.5-µf input filter capacitor will boost the output d.c. voltage by approximately 50, if necessary.

The high-resistance bleeder is included as a protection against shock. A lower value of bleeder resistance might be used if it does seriously reduce the d.c. output voltage.

The dual-feature capacitor tester. Capacitor in dotted lines is optional.

The potentialmeter controlling the grid bias on the tube is calibrated in d.c. output volts with a high-resistance voltmeter at the output filter capacitor.

Forming electrolytics

It is good practice to form electrolytics before connecting them into a circuit. I use the tester for forming them directly in sets that have been brought in for repairs and have been idle for any length of time. This is done before plugging in the set after preliminary tests for shorts. (The current indicated on the meter might not, in this case, be the true leakage current, depending on the presence of bleeders or voltage dividers.)

This practice has undoubtedly preserved or lengthened the life of rectifiers, especially directly heated ones, by not exposing them to heavy initial surge currents.

For proper forming, gradually increase the d.c. voltage across the electrolytic up to the specified working voltage. For best results the leakage (forming) current should not exceed 10–20 ma during any period in the forming process.

The meter indicating the leakage current has two ranges, 0–10 and 0–100 ma. Always use the 100-ma range first. After the leakage current has dropped to a value lower than 10 ma, switch the meter to the lower range for a more accurate reading.

A maximum value of 0.25 to 0.5 ma per microfarad can be considered normal for a good, fully formed electrolytic capacitor.

Should the forming current increase instead of decrease at a forming voltage much less than the working voltage, the film is breaking down and cannot be properly formed. Such electrolytics may—after a new forming attempt at a lower voltage—still be used in low-voltage circuits and should be plainly marked as such. It will generally be found that due to the thinner film formed at a lower voltage, the capacitance has increased.

Since a very low leakage current during the forming process might be due to a dried-out electrolytic that has lost most of its capacitance, the formed electrolytic should be measured for capacitance or its action observed in an actual circuit before grading it as good.

Due to rather poor power-supply regulation the tester has an interesting semiautomatic feature, which makes it unnecessary to increase the voltage manually during the forming process.

The initial heavy leakage current causes most of the voltage to be dropped in the tube and filter resistor. The voltage increases automatically—due to voltage division action—to the preset value as the leakage resistance increases while the electrolytic is being formed. The forming d.c. voltage can thus usually be preset to the full value.

As the supply will not deliver more than about 20 ma—or may be adjusted by using higher values for the filter (protecting) resistor—no damage will be caused to the supply or to the electrolytic to be formed.

The following procedure should be used for forming electrolytics:

1. Set function switch to discharge position.
2. Set the forming voltage on the voltage selector to the desired value (relying on the semiautomatic feature

This leakage tester also forms your electrolytics

By DANIEL NOF
of the supply) or start with a low voltage and gradually increase manually.
3. Make sure meter is in the 0–100-ma position.
4. Connect electrolytic to test terminals, observing polarity.
5. Turn power on.
6. After leakage current has dropped below 10 ma, switch meter to the 0–100-ma range for more accurate reading.
7. Increase forming voltage if necessary, always watching meter. Should current begin to increase after application of higher voltage, the film is breaking down and the electrolytic cannot be properly formed to normal working voltage. Try to re-form with lower voltage and use in low-voltage circuit.
8. After forming is complete (leakage current steady at 0.25 to 0.5 ma per microfarad as maximum), discharge electrolytic through resistor by switching to discharge.
Do not turn power off before disconnecting or discharging the electrolytic, as the discharge current through the bleeder deflates meter backward and may damage it.

Capacitor tests
Leakage in paper and mica capacitors is a constant source of trouble in coupling and high-resistance decoupling circuits. It is important to determine the permissible amount of leakage before connecting a capacitor into a circuit.

The capacitor to be tested is connected by turning the function switch to the paper-mica position. This activates the neon-tube relaxation oscillator circuit and puts the leakage resistance of the capacitor under test in place of the series resistor.

Flashes in the neon tube indicate the leakage resistance, the approximate value being found by connecting—for comparison—resistors of 30–200 megohms into the test terminals. A continuous glow indicates a shorted capacitor; rapid flashes, heavy leakage.

Glow on both electrodes of the neon tube indicates that the capacitor is not open. With no glow on the electrodes or with a very weak glow even with comparatively high voltage, the capacitor is open.

### Parts for capacitor tester
- 100–1000 ohms, 0.5 watt, resistor; 10–22,000 ohms, 1 watt, resistor; 10–100,000 ohms, 5 watts, resistor
- Shunt resistor (for 100-ma meter range); 2–megohm resistor for 0–10-ma meter
- 1–megohm capacitor
- 0.05–µf paper, capacitor
- 12–µf, 600 volts, capacitor
- 0.05–µf, 630 volts
- 0.1–µf, 500 volts
- 0.2–µf, 250 volts
- 0.3–µf, 100 volts
- 0.5–µf, 100 volts
- 1–µf, 100 volts
- 2–µf, 100 volts
- 4–µf, 100 volts
- 6–µf, 100 volts
- 10–µf, 100 volts
- 12–µf, 100 volts
- 20–µf, 100 volts
- 47–µf, 100 volts
- 100–µf, 100 volts
- 470–µf, 100 volts
- 1000–µf, 100 volts
- 10,000–µf, 100 volts

**Fig. 1—Schematic of the compact a.f. - r.f. tracer probe.**

**Fig. 2—Physical layout of components.**
Simple junkbox instrument makes fine resonance indicator

GRID-DIP ADAPTER WITHOUT TUBES

By ROBERT H. MITCHELL

Adapter shape makes for easy handling.

I HAVE wanted a grid-dip oscillator for several years. Like a lot of other hams, I didn’t feel like spending the cash to buy one or expending the effort necessary to build one. Besides, I had a friend who had a grid-dipper that he was willing to part with for a few hours occasionally. Unfortunately one Sunday morning I had to measure the resonant frequency of an inductance-loaded vertical antenna, and neither the friend nor his grid-dipper was available. Then and there I decided to build my own.

Perusing magazines of the past several years, trying to determine just which instrument could be built most easily from the junkbox, I came across an article on a tubeless grid-dip adapter which had been designed and sold by a company which had since gone out of business (McMurdo Silver). This interesting gadget, when inserted between a signal generator and a tuned circuit, indicated the resonant frequency of the circuit by the drop in output when the signal generator was tuned to that frequency. It looked simple and efficient, and I had a signal generator and enough parts in the junkbox for the adapter.

Having no spare 0–1 milliammeter, I used the 0–1-ma range of my multi-meter as a resonance indicator. The original instrument had two separate components, a coupling head and an indicator. I combined everything but the meter into the coupling head. The resulting circuit is shown in the diagram. A piece of 1 x 2 x 7-inch wood formed a handle and a mounting platform for the components, as shown in the photo. The coil socket was mounted on a piece of 2 x 3-inch aluminum which was bent into an L 2 x 2 inches, with a 1-inch upright. The jack was mounted on a smaller aluminum L. A 4-foot length of RG-59/U coaxial cable was used to connect the adapter to the signal generator. The coil consists of 10 turns of No. 22 enamelled wire spaced to cover 1 inch on a 1/4-inch diameter plug-in coil form. This permits operation in the range from approximately 5 to 40 mc. Total construction time was less than two hours, and most of that was spent in carving out the handle.

I plugged the multimeter leads into the jack, hooked the RG-59/U to the signal generator and started measuring resonant frequencies. Results were extremely disappointing. The meter indicated a moderate amount of output on the lower frequency ranges of the signal generator, but scarcely moved off the zero position when it was set above the broadcast band. This resulted in a frantic check of the wiring, substitution of a new 1N34 and several coil rewindings. I found the optimum value of R, for maximum meter reading, lay between 600 and 1,000 ohms. That appeared to be the only change that could be made to improve the performance of the gadget. After considerable thought, the obvious became apparent. The signal generator simply lacked enough output on the higher ranges to drive the adapter.

My signal generator was constructed from an Eico kit, output being taken through a capacitor and isolating resistor. When a jumper was placed across the isolating resistor, the output increased enough to drive the grid-dip adapter on all ranges, and I was in business. Shorting the isolating resistor resulted in some “pulling” of the signal generator when approaching resonance with a tuned circuit. However, the signal generator frequency is checked with a receiver and this is not a serious problem.

The adapter has several uses other than as a resonance indicator. One of the most interesting is checking receiver oscillator frequencies. To do this, plug headphones into the jack, couple the coil loosely to the receiver oscillator and tune the signal generator until a squeal is heard in the phones. This has helped me in the alignment of several receivers in which the oscillators were considerably off frequency.

The ham doesn’t need a commercial signal generator to drive this adapter. The station v.f.o. is an excellent signal generator.

After using the circuit for a while, it became apparent that a similar circuit could be used as a tubeless grid-dip adapter with a vacuum-tube voltmeter and a diode probe. All the components in the circuit other than the coil can be omitted. Connect the diode probe across the coil and resonance will be indicated by the drop in signal generator output at resonance. I have tried this circuit and found it effective.
TRANSISTOR OVERTONE GENERATOR

Point-contact unit makes excellent high-frequency generator

By I. QUEEN

THE April, 1954, issue of Radio-Electronics carried an article on a subharmonic generator which used a CK722 junction transistor to produce low-frequency signals from high-frequency crystals. For instance, a 400-kc crystal acted just like a 100-kc crystal—signals could be heard at 100, 200 and 300 kc, etc. By properly retuning the circuit the 400-kc crystal can be made to behave like a crystal ground for 80 kc, 133 kc, or any other desired subharmonic. Here is a new circuit that does just the opposite: it generates high frequencies from relatively low crystal frequencies. A point-contact transistor of the 2N33 type is used.

The circuit of the new overtone generator is shown. Coil L is tuned to the desired overtone (or multiple) of the crystal frequency. It is not tuned to the fundamental frequency. For high-frequency operation near 16 mc, L can be wound on a small slug-tuned ceramic coil form. The instrument shown here uses a CTC coil form, type L65, ¾ inch in diameter. It was wound with 23 turns of No. 22 enameled wire. The winding length is about ½ inch.

A 3575-kc crystal generated an overtone of about 18 mc. A 3552-kc crystal was heard near 17.5 mc. Both of these are fifth overtones, that is, frequencies five times that of the fundamental. A 4900-kc crystal was heard near 15 mc, the third overtone. These are all ordinary crystals, not special units treated for overtone circuits. All crystals (with third or fifth overtones near 15 mc) that were tried worked perfectly. In each case the slug may be tuned for maximum output and most stable signal. Tuning is not at all critical.

Note the difference between a harmonic and an overtone. In the overtone circuit, the 3552-kc crystal actually oscillates at 18 mc. There are no intermediate signals near 3.5, 7 and 10.5 mc, etc. This is a big advantage for the overtone circuit. There are no intermediate harmonics that may confuse identification. Furthermore, there is no progressive weakening from one harmonic to the next. For example, the 18-mc output mentioned above sounds like a fundamental should—very stable and strong. In an ordinary harmonic generator, because this signal would be the fifth harmonic, it would be considerably weakened and perhaps not as stable as a fundamental. Of course, harmonics of the 18-me signal exist at 36, 54 mc, and so on, well into the u.h.f. and v.h.f. regions. Therefore even 80-meter crystals can generate clear signals, well separated in frequency, deep into the high-frequency spectrum. (Overtone generators do not have even multiples of the fundamental frequency. Only odd multiples exist. Also, overtones may not be exact integral multiples. They may not be exactly three or five or seven times the original frequency, but they are always nearly so.)

Next, a 1-mc crystal was tried in the overtone generator. Tank coil L was a jumble-wound broadcast oscillator coil commonly used for replacement purposes. Its center tap was not used. A strange thing happened when the battery was turned on. The overtone was heard at 2.5 mc! This, of course, is not the expected odd multiple of 1 mc. On the possibility that the crystal was oscillating at 500 kc, the receiver was tuned to that frequency but nothing could be picked up. The crystal was, in fact, actually oscillating at 2.5 mc and this was the lowest frequency emitted from the transistor circuit. Strong harmonics were present, of course, at 5, 7.5, 10 mc, and so on. This greatly increases the usefulness of a crystal otherwise limited to integral multiples of 1 mc.

The generator is easily assembled and put to work. Components were not critical. The photo shows a special plug and socket used in connection with the 2N33. This was done because the transistor was frequently plugged in and removed during experimental work on the circuit. The pins of a transistor cannot stand too much handling since they are only very thin wires.

The signal generator has a jack in the B plus return for a key or modulation input. This safeguards the transistor by shutting off power when it is not needed. When properly tuned, the circuit could be keyed rapidly. Therefore the overtones are suitable as a signal source for communications purposes.

The power supply may be anything from 3 to 7.5 volts. On most crystals tried, good results were obtained with the lower voltage. The battery drain is approximately 3 ma.

With some crystals, the instrument may seem slow starting and may require touching the base terminal with a finger. This indicates that L is not correctly tuned. Adjust the core while monitoring the signal on a receiver. As you tune the core the output will become louder, but if you go too far the signal will disappear. Back off the slug adjustment slightly from this point for maximum output and optimum stability.

The short piece of wire shown extending from the generator is an antenna. With it, the signal can be heard a short distance from the radio receiver.

There are many applications for this overtone generator. The signals may be used for calibrating and aligning high-frequency receivers and experimental apparatus. Ordinary ham-band and other relatively low-frequency crystals become the source of signals deep into the v.h.f. and u.h.f. regions. The well-tuned frequencies are much easier to identify than if we attempted to use harmonics of low-frequency crystals. The high-frequency signals can serve as markers for TV and high-frequency receivers or to define the upper ham bands. They can even be used for high-frequency communications over short hauls.

Diagram of overtone generator circuit.

TEST INSTRUMENTS

<table>
<thead>
<tr>
<th>Parts for overtone generator</th>
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<tbody>
<tr>
<td>1—1,000 ohms, 2—4,700 ohms, resistors, ½ watt; 3—300 µuf. mica capacitor; 1—0.05 µf. disc ceramic; 4—2N33 point-contact transistor; 5—2N33 (see test); 6—socket for transistor; 7—3-pin socket and plug; 8—battery, 1 to 6 volts; 9—headphone jack; 10—crystal socket; 11—chassis; 12—length of wire.</td>
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www.americanradiohistory.com
New antennas often evolve from standard types. This all-channel unit is derived from stacked u.h.f. V’s and a u.h.f. dipole and reflector.

By R. F. KOLAR*

A u.h.f. antenna providing good gain and directivity is the stacked V’s in Fig. 2. Since the dipole and reflector and the stacked V’s are adequate in most areas for good signal reception on v.h.f. and u.h.f. respectively, their combination provides desirable performance characteristics over all TV channels. The two antennas are usually combined and connected to a common lead-in by using a crossover network consisting of bandpass filters. However, such networks are relatively expensive and unnecessarily introduce about 1 db loss. Furthermore, when two such antennas are placed near each other, the mutual coupling results in a degradation of the performance of each, particularly the one for v.h.f. Therefore, an array which has gain within one or two db of the individual antennas and eliminates the need for a crossover network would be a valuable addition to currently available antenna types. This article describes a new type of antenna which incorporates the features of the v.h.f. dipole and reflector and the stacked V u.h.f. type in a unique manner.

The impedance between any two points symmetrically placed about the center of a resonant antenna will depend on the distance between them. The greater the separation between the feed points, the higher the impedance. By using this principle, together with a transmission line of continuously increasing spacing, it is possible to match a resonant dipole to a high-impedance transmission line. Such a system, shown in Fig. 3, is known as a delta-matched antenna.

Design consideration

If the wings on the v.h.f. dipole are modified and the horizontal sections of the wings are extended and connected together as shown in Fig. 4, the configuration of the all-channel antenna is complete. The antenna may be considered to be made up of two vertically stacked u.h.f. long-wire V antennas, with a v.h.f. dipole located between the two V’s. The two systems are connected electrically as shown.

In the design of the array, primary consideration was given to the u.h.f. section—that is, the stacked V’s. Fig. 5 shows that the gain of the V antenna...
TELEVISION

Increases rapidly with length until the elements are about 3 wavelengths long, thereafter the rate of increase becomes less. At 450 megacycles, 3 wavelengths is approximately 78 inches. This length would be a good choice for high gain and minimum material requirements. However, 55 inches is about the largest practical length which can be supported with sufficient rigidity to prevent wind modulation.

Having established the length of the V elements at 55 inches, which is 4.2 wavelengths at 900 megacycles, Fig. 5 shows that the included angle should be 50 degrees to maintain maximum gain in the forward direction at 900 megacycles. Polar pattern measurements indicate that when a 50-degree angle is selected for 55-inch rod lengths, the field pattern is essentially unidirectional throughout the u.h.f. television spectrum.

With this length and included angle the vertical separation must be between 9 and 13 inches. Less than 9 inches substantially reduces the u.h.f. gain, whereas anything greater than 13 inches causes the vertical polar pattern at 900 megacycles to become too sharp. Twelve inches was selected as a good compromise between high gain and acceptable beam width.

Fig. 6 shows the all-channel v.h.f.-u.h.f. antenna developed by combining stacked u.h.f. V's and a v.h.f. dipole and reflector. Performance on the v.h.f. channels depends on the distance between the dipole and the open ends of the V, measured along the V. The performance of the new array with an 8-foot v.h.f. dipole is shown in Fig. 7. The optimum location for the dipole is 26 inches from the ends of the V. If it is greater than 27 inches from the ends of the V, the horizontal pattern on channel 6 breaks into multiple lobes and the gain decreases. Also if it is located nearer than 24 inches the gain is lowered.

Similar measurements were taken for dipole lengths other than 8 feet. If the dipole is made appreciably longer than 8 feet, the horizontal polar pattern on

Fig. 4—All-channel v.h.f.-u.h.f. array.

Fig. 5—The gain of a V antenna varies with leg length and the included angle.

Fig. 6—Delta Vee all-channel antenna. Reflector increases low-channel gain.

Fig. 7—The v.h.f. power gain varies with position of the dipole radiator.

DECEMBER, 1954
The dipole diameter is not critical, but if it is to be used as a support for the array it should be large enough to provide ample strength for mounting. A diameter of 3/4 inch is adequate for the mounting shown in Fig. 6. The dipole supporting elements must be less than 1/2 inch in diameter in order to provide sufficient inductance to prevent their presence from impairing the u.h.f. performance of the antenna.

A 106-inch reflector, 46 inches behind the dipole element in Fig. 6, substantially increases the low-frequency gain of the array. Improvement on certain channels may be realized by slightly veering the v.h.f. dipole element. More than one dipole element can be used, one above and one below the u.h.f. stacked V, with or without one between the V's. However, such improvements at one frequency are invariably accompanied by serious degradation of the gain at other frequencies in the television bands.

Operation and performance

From 51 to 88 megacycles the array operates as a dipole with a delta-matched feed, using the rear portion of the two stacked long-wire V antennas for the delta section. From 174 to 216 megacycles the rear portion of the V again acts as a delta feed, while the forward extensions of the V's produce a current distribution to provide a maximum of gain in the forward direction. On the u.h.f. channels, from 470 to 890 megacycles, the antenna acts as two stacked long-wire V antennas. Since the antenna is a combination of a delta-fed dipole and a u.h.f. V antenna, it is called the Delta Vee.

The characteristics of the antenna are given in Figs. 8, 9, and 10. The v.h.f. power gain, shown by Fig. 8, is within 3 db of that of the v.h.f. dipole and reflector of Fig. 1. On channel 2, the gain of the new antenna is substantially greater than that of the dipole and reflector, whereas on channels 5 and 6 the dipole and reflector is 3 db better than the Delta Vee. The gain on channels 5 and 6 is no worse than would be expected from mounting the two antennas near each other and combining their outputs with a crossover network.

The u.h.f. power gain, shown in Fig. 9, compares very favorably with that of the u.h.f. stacked-V antenna.

The horizontal patterns in Fig. 10, display desirable characteristics. The v.h.f. curves show considerable forward gain. Although there is an indication of some v.h.f. pickup on the u.h.f. elements, the undesired response is small. The u.h.f. patterns have a narrow forward lobe with multiple secondary lobes which are characteristic of long-wire antennas such as the stacked V. It would be necessary to increase the length of the V elements many times in order to reduce the amplitude of the secondary lobes.

Fig. 10—Horizontal response patterns of the Delta Vee on various channels.
**Fourteenth conversation, first half—**

**The d.c. component and image brightness**

**WILL**—The other day, when we were talking about video frequency, you put in a lot of time on high-frequency problems. But don't you run into difficulties with the other end of the spectrum—the very low frequencies?

**KEN**—Maybe—but just how does all this come up?

**WILL**—Well, I began by wondering if—in certain cases—the video signal didn't come down to a simple direct current. For example, suppose the picture has a large horizontal band of one shade. But you can't transmit d.c. through the interstage-coupling capacitors...

**KEN**—You might have a real difficulty there if you didn't have sync pulses at the end of each line. They make a very sudden change in the voltage and prevent it from becoming the simple d.c. you're thinking about. But—as you say—a capacitor can't transmit an unvarying direct current. And that poses some problems—only they're not the ones you're thinking about!

**WILL**—Suppose you lay off the mystery and give me a line on some of those problems! Who knows—I might understand you!

**KEN**—O.K. Suppose we start by thinking of the way an alternating voltage is transmitted through a coupling capacitor C with its leak resistor R.

**WILL**—We did that a long time ago, and nothing could be simpler. Let's say that an alternating voltage is applied to the left-hand plate of the capacitor. During the positive alternation it draws electrons away from that plate. Because the atoms on the left-hand plate are short of electrons, they try to attract a few from the other plate. They can't get them across the dielectric, but a lot of electrons are dragged onto the right-hand plate. Where from? The only place they could come from is the ground, through resistor R. That current flow through R causes a voltage drop across it, and its top end becomes more positive than the bottom. So it's just as if the positive alternation had passed through the capacitor.

**KEN**—With this one important difference: If an alternating current is superimposed on a direct one (as in a plate circuit where you usually find your high alternating voltage), the d.c. won't be passed through the capacitor. But tell me, what goes on during the negative alternation?

**WILL**—Much the same thing. The electrons flow into the plate at the left, and of course chase others off the right-hand one, because there's nothing more repulsive to an electron than another electron. The displaced electrons have nowhere to go except to ground through R. That makes its top end negative. So again, it's just as if the negative voltage had gone through the capacitor.

**KEN**—You're right all the way. The way these electrons rock back and forth balancing themselves so neatly and symmetrically reminds me of children on a seesaw.

**Symmetry and equilibrium**

**WILL**—I knew all this a long time ago. Is there some special reason behind this refresher course?

**KEN**—There is. And the reason is that things are just a bit different for the video signal.

**WILL**—How come?

**KEN**—Because—unlike modulated r.f. or sound a.f.—the video-frequency signal is not symmetrical. It isn't made up of equal positive and negative alternations that are exactly like each other—like an object and its reflection in the mirror.

**WILL**—I see! Once it's detected, our video signal is either all positive or all negative. Only the extreme whites even approach zero. We haven't got that axis of symmetry you can always draw on an a.c. waveform.

**KEN**—Let's take a look at the way those signals are transmitted through the coupling circuits.

**WILL**—Do you want positive or negative polarity?

**KEN**—Let's take the commonest case—a set with one video stage and the control voltage to the picture-tube grid.
Then the detected signal would, of course, be...

**Will.**—...positive-going. Then we'd always have more or less of a shortage of electrons in the left plate of the capacitor, except during the odd periods when we have a "whiter than white" signal. So there'd be a greater or lesser number of electrons crowding into our right-hand capacitor plate.

**Ken.**—That's it. As the positive voltage reaches a peak (during the sync pulses) the largest number of electrons are being dragged from ground through R and to the right-hand plate. So the top of the resistor becomes more positive.

**Will.**—I think I see what you're driving at. As the signal becomes less positive—for instance during a white patch on the image—some of the electrons in the right-hand plate leave it and try to get back to ground through R. That would make the top of R negative...

**Ken.**—You see, the signal we get on the right-hand side of the capacitor is a voltage of the same form as the one that came from the detector. But it's no longer entirely positive (or entirely negative if the detection was in that direction). The voltage on the grid of the video amplifier, while not entirely symmetrical, is composed of alternations which are negative and positive as compared to the ground potential, and equitably disposed around it.

**Will.**—What do you mean "equitably"?

**Ken.**—The number of electrons that leave the plate in a negative alternation is equal to the number that come back into it in a positive one. They're good accountants and always try to even up the receipts and expenses. I think you can realize—without going into integral calculus—that the quantities of electricity are proportional to surfaces inside these curves.

**Will.**—In other words, to find the point where the curve passes through the zero axis, you have only to cut through it in such a way that you have the same areas on each side?

**Ken.**—At least that would be a way of seeing if the curve is well drawn... You see, in sending your video signal through the coupling capacitor, you have made it alternating—have taken away its polarity. And above all, you have separated the sync pulses from the constant level as compared with zero volt.

**More defects of capacitance**

**Will.**—Is that all?

**Ken.**—It's not as catastrophic as was once thought, but it's not good! For instance, your sync pulses are going to vary in height according to the strength and modulation of the signal. That makes it harder to set the controls for correct synchronization. On top of that, the shade of the picture itself is likely to be altered.

**Will.**—I don't get that!

**Ken.**—Let's try a simple example to show what the situation really is. Suppose the image is a white equilateral triangle on an even black background. Now try to trace the form of a detected video signal (with positive polarity) for three of the scanning lines: one at the top, the next near the middle, and the third at the base of the triangle.

**Will.**—Easy! We have the sync pulse at 100% of maximum amplitude, then a black line at about 70%, with the exception of a short point where the amplitude drops down close to zero. The black lines occupy only about half the middle scan, and the white one lengthens out accordingly. And it keeps on lengthening out, till it occupies practically all the bottom line.

**Ken.**—Perfect. Now can you draw me—in dotted lines—the zero axis for each of these signals as it would be situated after passing through a coupling capacitor?

**Will.**—Nothing simpler. Here are your axes!

**Ken.**—You see now that the sync pulses are all at different voltages. In fringe areas that would mess up synchronization. But that isn't all! If you were to apply signals exactly like these to a picture-tube grid and regulate the brilliancy so as to get a good white for the top line, it wouldn't stay the same for the other two. The middle one would be a decided gray and the bottom one even darker. So our triangle, instead of being a uniform white, would get darker and darker toward the base. *To be continued*
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TELEVISION

Progress—the new RCA 21-inch color tube shown alongside old 15-inch tube.

THE previous article of this series explained the operation and control adjustment procedure for the three-gun, 15 1/2-inch color picture tube. This article deals with the larger 19-inch three-gun tube and the associated purity, convergence and high-voltage circuitry.

The 19-inch tube is very similar to the 15 1/2-inch color tube in that it contains three electron guns along with a viewing screen composed of red, blue and green phosphor dots. The only major differences between the 15 1/2-inch and 19-inch color tubes are that the convergence electrode is absent in the larger tube and that the three electron guns are mechanically tilted during manufacture so that the beams will converge or come together at the center of the tube. Tilting the three guns toward a common axis permits improved d.c. convergence control and makes convergence adjustment easier. Since the 19-inch color tube does not include a convergence electrode, external means are necessary to provide convergence so that each electron beam strikes only its own proper phosphor dot at the correct time. Three convergence coils are placed around the neck of the 19-inch color tube between the deflection yoke and purity coil. The coils, one for each electron gun, deflect the electron beams so that convergence is correct at all times.

Three convergence or beam-positioning magnets were required for the 15 1/2-inch tube. Due to the tilted electron guns in the 19-inch tube only one such magnet (blue, in this case) is required. This single magnet is located on the neck of the tube (Fig. 1). To get equal spacing of the three electron beams (to form a triangle) at the center of the phosphor screen, the d.c. through each of the three convergence coils is varied. The effect of the blue-positioning magnet is to move the blue dots at right angles to the direction in which the blue-convergence coil d.c. current would move them. This combined action aids in forming the triangles.

Fig. 1—Components on 19-inch tube.

Fig. 2—Horizontal and vertical waveforms are applied to convergence coils.

COLOR TV CIRCUITS

Part VII—The 19-inch color tube and associated circuits

By KEN KLEIDON* and PHIL STEINBERG*

Fig. 3—Schematic diagram of a convergence channel for the 19-inch color tube.

Fig. 4—The high-voltage doubler circuit for second-anode and focus voltages.

*Raytheon Manufacturing Company Television and Radio Operations.

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NEW 1955

Heathkit Engineering Features

New PRINTED CIRCUITS
One of the many tremendous improvements in the new 955 Heathkit is the use of anetched metal process printed circuit board. Printed circuits will be used in Heathkit whenever they will affect construction simplification, performance, and lend themselves to instrument design. Now for the first time it is possible to combine the advantages of modern printed circuit instrument construction technique. For the first time a complete panel has been given towards a new kit assembly line. As of this is the first time that printed circuit boards have been hand soldered on a volume basis. Offered only by Heathkit, the pioneer and leader in kit instrument design.

New HIGH READABILITY PANELS
New 1955 Heathkit feature complete panel redesign. Sharp white lettering is applied to the device to create a beautiful contrast against the grey panels. A new high in readability. Lettering is clear to read over style and panel utilization. Vividly clear against the panel's white background. New kinds of exclusive Heathkit designs.

New 3" UTILITY SCOPE
The new 3" Scope is a "natural" for the well rounded line of Heathkit instruments. Small in size, 11.5" deep, 6.1" wide, 3.7" high, yet big in performance. Just think of this value, an Oscilloscope for $26.50. Brilliant intensity, sharpness, using wide positioning range. An ideal portable Scope for the TV technician—second sweep scope—modulation monitor for you same (deflection plane) terminals in rear of cabinet. Performance to spare for all general scope applications. See specifications on following page.

New SCOPE SWEEP CIRCUIT
10 CYCLES - 500 KC
New 1955 Heathkit Model 0-10 Scope features a new wide frequency range sweep generator covering 10 cycles to 500,000 cycles. This coverage is available in five virtually decimating sweep ranges and is five times greater than the sweep frequency range usually available. Excellent frequency range sensitivity, actually less than 5% at 500 KC. Use of the new freedom in Heath circuit provides a larger margin of stability and a new high in Heathkit Scope performance.

New PEAK-TO-PEAK VTVM CIRCUIT
New SAL 5 half wave rectifier in AC input circuit permits full scale peak-to-peak measurements. Seven ranges—upper limits 4000 volts peak-to-peak. Just the thing for TV service men have needed in making TV circuit voltage checks. Precision rectifier voltage divider limits AC RMS level to 50 volts. Precision overload protection the recifier — upper limits upper lines AC RMS ranges to 1000 volts. Another definite example of continuing Heathkit design leadership in the kit instrument field.

New STYLING New COLOR
New styling and coloring is responsible for tremendous improvement in Heathkit appearance. The new instrument panel color combination in high definition white lettering in a soft charcoal gray panel. Caisson once silver, now feather grey. The satin gold is now enamel rainbow for the WA-72 Preamp. Another example of the modern progressive trend in Heathkit styling.

Continuing PROGRESS FUTURE LINE EXPANSION
The outstanding improvements featured in the 1955 Heathkit line are representative of the progress characteristic of Heathkit operation. Long range planning will provide a continuing succession of new kits. The new kits feature larger panel color combinations, higher performance, and more comprehensive features. Another indication of the dedication to the Heathkit line which already represents the world's greatest selection of electronic kits. The innovations in the Heathkit line are representative of the continuing improvements and development of Heathkit's line of electronic instrumentation.
The basic function of the Heathkit Electronic Switch Kit is to permit simultaneous oscilloscope observation of two separate traces which can be either separated or superimposed for individual study. This is accomplished through the use of two individually controlled inputs working through amplifier, multivibrator, and blocking stages. The output of the Electronic Switch is connected directly to the vertical input of the Oscilloscope. A typical example of usefulness would be simultaneous observation of a signal or waveform as it appears at both the input and output stages of an amplifier.

APPLICATIIONS

An Electronic Switch has many applications to increase the over-all operating versatility of your oscilloscope. It can be used to check amplifier distortion—audio crossover networks—phase inverter circuits—to measure phase shift—special waveform study, etc. The instrument can also be conveniently used as a square wave generator over the range of switching frequencies, often providing the necessary wave form response information without incurring the expense of an additional instrument. Ownership of this instrument will reveal many entirely new fields of oscilloscope application and will quickly justify the modest cost of the Electronic Switch Kit.

Another useful oscilloscope accessory particularly in circuit development work in TV and radio service work. The Voltage Calibrator provides a convenient method for making peak-to-peak voltage measurements with an oscilloscope, by establishing a relationship on a comparison basis between the amplitude of an unknown wave shape and a known output of the voltage calibrator. Peak-to-peak voltage values are read directly from a calibrated panel scale without recourse to involved calculations.

FEATURES:

To offset line voltage supply irregularities, the instrument features a voltage regulator tube. A convenient "signal" position on the panel switch by-passes the calibrator completely and the signal is applied through the oscilloscope vertical input, thereby eliminating the necessity for constantly transferring test leads.

RANGES:

With the Heathkit Voltage Calibrator it is possible to measure all types of complex waveforms within a voltage range of .01 to 100 volts peak-to-peak. Build this instrument in a few hours and enjoy the added benefits offered only through combination use of test equipment.

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TELEVISION

The signal currents applied to the three convergence coils come from the vertical and horizontal output circuits and are parabolic in shape, similar to the signal applied to the convergence electrode of the 15½-inch tube. The vertical convergence parabolic signals are derived from the vertical output tube. The same vertical sawtooth waveform that is coupled to the deflection yoke is also coupled to the vertical convergence circuits that change the sawtooth waveform to a parabolic waveform at the vertical frequency. The parabolic signal is then separated into three different channels, the red, blue and green convergence coils mounted on the neck of the tube. Each of the three channels has three controls: one varies the amplitude of the parabolic signal, another changes the shape by tilting the parabola either one way or the other to secure better dynamic convergence, and the third is the d.c. convergence control.

The derivation of the horizontal and vertical convergence signals is similar. A special winding on the horizontal output transformer couples the horizontal flyback pulse to the horizontal convergence circuit. A separate channel is used for each convergence coil and two controls are provided in each channel to vary the tilt and amplitude of the parabolic waveforms. This is similar to the vertical convergence circuit. The horizontal and vertical parabolic waveforms from each channel are added together before being applied to the three separate convergence coils (Fig. 2).
NEW Heathkit
5" PUSH-PULL
OSCILLOSCOPE KIT
FOR COLOR TV

BRAND NEW DESIGN: The new Heathkit Model O-10 Oscilloscope would be something special at any price, but is almost unbelievable at $69.50. Completely re-designed scope has broadband amplifiers for color TV work and offers brilliant overall performance. Vertical frequency response within 3 db from 5 cps to 3 mc. Even more astounding, the response is down less than 1 mc at 33.8 mc. The color TV sync burst frequency. It is essential that scopes for color work have true broadband characteristics.

PRINTED CIRCUITS: Two printed circuit boards used in this fine instrument to produce stable, consistent performance. Problems solved by pre-engineering of boards, and their use guarantees completed unit that will have same characteristics as lab development model. Printed circuits simplify construction and save labor.

NEW SWEEP CIRCUIT: Sweep circuit operates with exceptionally good linearity from 20 cps to over 500,000 cps, 5 times the usual range for scopes in this price range. An entirely new circuit introduced for the first time in any Heathkit. Simplicity, standardization, construction technique of vertical and horizontal amplifier construction made possible through the use of one single circuit board.

Clean, crisp, under -chassis construction and wiring. Possible only through use of new, and simplified printed circuit boards.

EXCEPTIONAL VALUE: The brand new Model OL-1 Utility Oscilloscope is designed especially for portable applications so that outside servicing or personal performing field tests can have the advantages of a scope available. Then too, it is ideal for home workshops, the ham shack, or as an "extra" scope for the service shop. It is compact, light in weight, and surprisingly versatile in operation. An outstanding instrument for the price.

Front panel controls are "built out" for ease of operation and convenience. Printed circuit board used for consistent circuit performance. Assembly time cut in half.

FEATURES: Other outstanding characteristics of this professional oscilloscope are: Built-in 1 V pk-to-pk reference for calibration of plastic or C1T take-off. 3 SUPC CRT; push-pull hrf and vrf deflection amplifiers, hrf trace expandable to 3 line; constant current of CR tube to allow inspection of any small portion of the signal. Deflection sensitivity, .055 volts per inch; wiring harness pre-formed and matted to save assembly time and insure professional quality control. Introduces efficient retract blanking. Frequency compensated sweep operation. Additional versatility provided by 10 line sweep, 200 kc sweep, and a sweep of any frequency adjustable from 20 to 500,000 cps. Also available: New Heathkit color stylus with charcoal gray panel and high readability white lettering. Printed circuit boards used in all coupling and by-pass applications. The "new -look" in Heathkit styling produces professional appearance in keeping with the professional performance of this instrument.

MODEL O-10
$69.50

MODEL OL-1
$29.50

VERSATILE INSTRUMENT: The new Model OM-1 general purpose Oscilloscope represents an outstanding dollar value in reliable test equipment. Full 5 inch CRT; Printed circuit boards ease of assembly; constant circuit characteristics, and rugged component mounting. Includes all the design features necessary for servicing, students, experimenters, radio amateurs, etc. Frequency response of amplifiers flat within 3 db from 10 cps to 10 kc, and down only 7 db from 10 cps to 500 kc. Sweep generator range from 20 cps to 100,000 cps. Also features New Heathkit color styling with charcoal gray panel and high definition white lettering for readability under non -ideal lighting conditions.

DESIGN FEATURES: A full-size, versatile oscilloscope at a price you can afford. Other features are: adjustable spot shape control; RF connections to detection plates; direct coupled centering controls; external and internal sweep and sync; 60 cycle line sweep; New Heathkit color styling and unusual performance make this instrument an outstanding value.

HEATH company
BENTON HARBOR 20,
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DECEMBER, 1954

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The Heathkit Multimeter is a "smart" to complete the instrument lineup of any well equipped service shop. Here is an instrument packed with every desirable service feature, many of which are not found in other Multimeters. All of the measurement ranges you need or want. High sensitivity 20,000 ohms per volt DC, 5,000 ohms per volt AC.

**ADVANTAGES**

Complete portability through freedom from AC line power operation—provides service ranges of direct current measurements from 0 to 1500 microamperes in 50 microamperes steps. Can be easily operated in RF fields without impairing accuracy of measurement.

**RANGES**

Full scale AC and DC voltage ranges are 0-15, 50, 150, 500, 1500 and 5000 volts. Direct current ranges are 150 microamperes, 150, 500 and 2000 milliamperes and 10 ampere. Resistances are measured from 2 ohms to 20 megohms in 3 ranges and db range from 10 to +65 db.

**CONSTRUCTION**

The Heathkit Multimeter features a unique resistor ring switch mounting assembly procedure. With this method of assembly the precision resistors are wired to the rings and range switch before actual mounting of the switch to the instrument panel. This procedure affords the advantage of simpler construction yet complete accessibility of precision resistors in event replacement is ever required. Ohmmeter batteries were selected for convenience of replacement and only standard commercially available types are used. Batteries consist of type C flash light cell and 4 Penlite cells. All batteries and necessary test leads are furnished with the kit.

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**TELEVISION**

showed. The parabolic signal is then coupled through the vertical parabolic amplitude control and isolation choke L1 to one winding (L4) of the convergence coil. The other convergence-coil winding (L3) in conjunction with the vertical tilt control—causes the vertical parabola to lean either left or right. This is done by coupling the vertical sawtooth through the tilt control and isolation choke L2 to the convergence coil. The polarity, or direction, and amplitude of tilt are governed by setting of the tilt control with respect to the tap (point of signal injection).

The horizontal sawtooth, obtained either from a special winding of the horizontal output transformer or as a sawtooth generator circuit or any other convenient point in the horizontal circuit, is coupled through the horizontal sawtooth amplitude control to the grid of the amplifier V1. This sawtooth waveform develops a parabolic current through the convergence coil L4 in the plate circuit due to the reactance of the coil at the horizontal frequency. The horizontal tilt control varies the polarity and amplitude of the pulse that is also coupled to the grid of the amplifier. Since a pulse voltage is changed to a sawtooth current by the reactance of L4 to the horizontal frequency, the same tilt effect is obtained as for the vertical parabola.

The d.c. convergence control in the plate circuit of V1 varies the d.c. through the convergence coil and therefore permits convergence at the center of the color tube. For each of the three coils, 5 convergence controls are provided, making a total of 15 for a 19-inch color tube. Only one other convergence control is required—the blue beam-positioning magnet mentioned previously. This magnet is a static control and mainly affects convergence at the center of the picture tube. The purity coil and control for the 19-inch color tube is identical to that of the 15½-inch tube. The purity coil around the neck of the picture tube (Fig. 1) is supplied approximately 255 ma of d.c. that can be varied by the color purity control. The magnetic field of this coil electrically aligns all three electron guns so that each gun strikes only its respective phosphor dot. Three other major components affect color purity: the deflection yoke, blue-beam-positioning magnet and and convergence coils. The deflection yoke axial position and rotation affect purity but should not present trouble after the yoke is correctly positioned. As with a 15½-inch color tube the interaction between convergence components and the purity coil may be troublesome. However, due to the smaller d.c. magnetic field of the 19-inch receiver's convergence magnet and coils, less interaction should be experienced.

The high-voltage section of a color receiver is similar to that of a black-and-white set. The only difference is the addition of three tubes to obtain...
NEW Heathkit
VACUUM TUBE
VOLTOMETER KIT
PRINTED CIRCUIT DESIGN

Another outstanding example of continuing Heath Company pioneering and leadership in the kit instrument field. A new printed circuit VTVM. New peak-to-peak circuit—new styling and new panel design. A precision, precalibrated printed circuit board eliminates chassis wiring, cuts assembly time in half, assures duplication of Engineering pilot model specifications, and virtually eliminates possibility of construction error.

CIRCUIT:
A 6AK5 tube operated as a full wave AC input rectifier permits seven peak-to-peak voltage ranges with upper limits of 4000 volts P-P. Just the ticket for you TV serviceman. Voltages input in the 6AK5 input circuit limits applied AC input to a safe level. The circuitry and the deduction of the meter in the outside of the 12AU7 bridge circuit affords a high degree of protection to the sensitive 20 microammeter meter.

RANGES:
Seven voltage ranges, 1.5, 5, 15, 50, 150, 1500, and 15,000 volts DC and AC RMS Peak-to-peak ranges 5, 14, 40, 140, 400, 1400, 4400. Ohmmeter ranges X1, X10, X100, X1000, X10K, X100K. X1 meg. Additional features are a dialscale, a center scale zero position, and a polarity reversal switch. IMPORTANT FEATURES:
High impedance 11 megohm input—transformer operated—4% precision resistors, 6AK5 and 12AU7 tube—solid power rectifier—individual AC and DC calibrations—smoother improved zero adjust control action—new panel styling and color—new placement of pilot lights—new positive contact battery mounting—new knob—spot bulb included. The new V7 also sets the pace as a kit instrument style leader. Sears, good-looking channel gray panel and soft-feather gray cabinet. High reliability panel with sharply contrasting white, black, and red calibrations. The plastic, eye catching, modern styling is in harmonious balance with the outstanding circuit design improvements. Easily the best buy in kit instruments.

Heathkit AC VACUUM TUBE
VOLTOMETER KIT
MODEL AV-2
$2950

Extreme sensitivity has been emphasized in the design of the Heathkit AC VTVM. Ten full scale RMS ranges are .01, .03, 1, 3, 10, 30, 100, and 300 volts. Frequency response is substantially flat from 10 cycles per second to 50 KC with input impedance of 1 megohm at 1 KC. Will accurately measure as low as 1 millivolt at high impedance. Total db range is —52 db to +152 db. An excellent kit for measuring the output of phono cartridges and the gain of amplifier stages. Use it also to check power supply ripple, as a sensitive null detector, and for compiling frequency response data. Features one knob operation, 200 microammeter Simpson meter and precision resistors.

Heathkit AUDIO WATTMETER KIT
Read audio power output directly without using external load resistors with the new Heathkit Audio Wattmeter. Built-in non-inductive load resistors provide impedances of 4, 8, 16, and 600 ohms. Flat response from 10 CPS to 250 KC. Full scale power ranges are 0-5 MW, 0-50 MW, 0-500 MW, 0-5-W and 0-50 W. Model AW-1 will operate continuously at 25 watts and has a duty cycle of 3 minutes at 50 watts. Total db range in five positions is —50 db to +48 db, using the standard 1 milliwatt 600 ohms.

MODEL AV-1
$2950

Heathkit 30,000 VOLTS DC
PROBE KIT
MODEL V-7
$24.50

Heathkit PEAK-TO-PEAK
PROBE KIT
No. 338-C
$550

Heathkit RF PROBE KIT
No. 309-C
$350

DECEMBER, 1954
a regulated 27-kv output and supply a high potential to the focus anode of the picture tube. To obtain 27 kilovolts, a voltage-doubler circuit is used (Fig. 4).

The high potential from the output transformer is applied to the plate of V1 and charges C1. The potential across C1 is coupled through the focus control to the plate of V2 (high-voltage coupling diode). Since the plate of V2 is now positive with respect to its cathode, it conducts and charges C2. The charge in C2 adds to the pulse voltage from the television high-voltage transformer winding and is sufficient to develop 27 kv at the cathode of V3.

The action of the high-voltage regulator is similar to having two variable resistances in parallel (Fig. 5). For proper operation, the total parallel resistance will always be constant. Thus the total supply current will remain constant and as a result the voltage will also remain constant. One of the variable resistors may be considered the high-voltage regulator, and the other the picture-tube load.

Assume that the picture-tube current increases—effectively, a decrease in picture-tube load resistance. This would cause a decrease in high voltage that would be coupled back by diodes V2 and V3 to the bipolar network and a decrease the positive control-grid bias on the high-voltage regulator. The high-voltage regulator would then conduct less current and have the effect of increasing the regulator plate resistance. The plate of the regulator is connected to 27-kv point and the cathode is tied to B plus, which is, effectively, ground for the high voltage. The high-voltage regulator is therefore in parallel with the picture-tube load resistance.

Since the picture-tube load resistance originally decreased and caused the regulator’s resistance to increase, the original requirement for constant parallel resistance is maintained. The opposite effect may also occur where the picture-tube load resistance increases and causes the regulator to conduct more current due to its grid going more positive and thereby lowering the regulator’s resistance to compensate for the increased picture-tube load resistance.

This article has been particularly related to the 19-inch three-gun color tube and circuitry involving purity, convergence and high voltage. The next installment will discuss circuitry as it applies to a complete color television receiver.

TO BE CONTINUED
NEW Heathkit
TV ALIGNMENT GENERATOR KIT

Here is the most radically improved Sweep Generator in the history of the TV service industry. The basic design follows latest high frequency techniques which result in a combination of performance features not found in any other sweep generator.

Sweep action is obtained electronically through the use of a newly developed controllable inductor, thereby eliminating all moving parts with their resultant hum, vibration, fatigue, etc. Frequency coverage entirely on fundamentals, is continuous from 4 MC to 220 MC at an output level well over a measurable 1 volt.

MARKER:
The same instrument incorporates a triple marker system with a crystal controlled reference. A variable marker provides accurate coverage from 30 to 220 MC on fundamentals, and 57 to 180 MC on calibrated harmonics. A separate fixed crystal controlled 4.5 MC marker can be used for checking IF, bandpass, calibration, reference, etc. Provisions are also found for external marker use. A 4.5 MC crystal is supplied with the kit.

POWER SUPPLY:
The transformer operated Power Supply features voltage regulation for stable oscillator operation. Three sets of shielded cables are furnished with the kit. Sweep range is completely and smoothly controllable from zero up to a maximum of 20 MC, depending upon base frequency. Here is a Sweep Generator that truly no serviceman can afford to be without for rapid, accurate, TV alignment work.

NEW Heathkit
SIGNAL GENERATOR KIT

The new Heathkit service type Signal Generator, Model SG-8 incorporates many design features not usually found in this instrument price range. Frequency coverage is from 160 KC to 110 MC in five ranges, all on fundamentals, with useful calibrated harmonics up to 220 MC. The RF output level is well in excess of 100,000 microvolts throughout the frequency range. The oscillator circuit consists of a twin triode tube, one-half used as a Cathode oscillator, and the other half as a cathode follower output which acts as a buffer between the oscillator and external load, thereby eliminating oscillator frequency shift usually caused by external loading.

All coils are factory wound and adjusted, thereby completely eliminating the need for individual calibration and the use of additional calibrating equipment. The stable, low impedance output, features step and variable attenuation for complete control of RF level. A separate 6C4 triode acts as a 400 cycle sine oscillator, and a panel mounted switching system permits choice of either external or internal modulation.

NEW Heathkit
BAR GENERATOR KIT

The Heathkit BG-1 produces a series of horizontal or vertical bars on a TV screen. Since these bars are equally spaced, they will quickly indicate picture linearity of the receiver under test without waiting for transmitted test patterns. Panel switch provides "steady—horizontal and vertical position. " The oscillator unit uses a 12AX7 twin triode for the RF oscillator and video carrier frequency. A new relaxation oscillator provides low frequency for vertical linearity tests. The instrument will also provide an indication of horizontal and vertical sync pulse stability as well as overall picture size. Operation is simple and merely requires connection to the TV receiver at a terminal. Transformer operated for safety.

HEATH COMPANY
BENTON HARBOR 20, MICHIGAN

DECEMBER, 1954

www.americanradiohistory.com
The new Heathkit Visual-Aural Signal Tracer features a special built-in RF input channel used in conjunction with a newly designed wide frequency range demodulator probe. High RF sensitivity permits signal tracing from the receiver antenna input. Separate low gain channel and probe available for audio circuit exploration. Both input channels are consistently monitored by an electron-ray beam indicator so that visual as well as aural indications may be obtained.

NOISE LOCATOR:
A dually tuned feature is a noise locator circuit used in conjunction with the audio probe. With this system, a DC potential is applied to a suspected circuit component, and the motion of the image in the component can be seen as well as heard. Invincible for ferreting out noisy or intermittent components, noisy rectifiers, potentiometers, IF and power transformers, etc.

WATTMETER:
Built-in calibrated wattmeter circuit will prove useful for quick preliminary check of total voltage consumption of a complete system under test. Separate panel terminals provide external use of the speaker or output transformer for substitution purposes. Same useful service tone by eliminating the necessity for speaker tone-masking service job. The same panel terminals also provide easy access to a well filtered B+ supply for external use. Don’t overlook the many interesting preening service possibilities provided through the use of this instrument, and let the Signal Tracer work for you by saving time and money.

Heathkit CONDENSER CHECKER KIT
Here is a handy test instrument for any Service Shop. Unknown values of capacitance and resistance are quickly determined on the direct reading condenser-checker dial. Capacity is measured in four ranges from 1,000 mf to 100 mf. Resistance in the range from 10 ohms to 5 megohms. DC polarization voltages of 25, 150, 250, and 450 volts are available for leakage tests on all types of condensers. For electrolytics, a power factor control is provided to balance out inherent leakage and to indicate directly the power factor of a condenser under test. Proper balancing of the AC bridge is reflected in the degree of closure of an electronic beam indicator tube. Model C-3 uses a transformer-operated power supply, spring return leakage test switch, and a convenient combination of panel scales for all readings. Test leads are furnished in addition to precision components for calibrating purposes. Check and easy to operate, the Heathkit Condenser Checker will have valuable time and increase your shop efficiency.

Heathkit "Q" METER KIT
The Heathkit QM-1 represents the first practical shorted Q meter available within the price range of schools, laboratories, TV service men, and experimenters. This instrument will enable the operator to simulate conditions encountered in practical circuits and to measure the performance of coils or condensers at the operating frequencies actually encountered. All indications of value are read directly on the 15/2 microammeter Simpson calibrated meter scale. Measure of transformers, RF reactors, and the distributed capacity of coils. Oscillator section supplies RF frequencies 150 KC to 15 MC or four ranges. Calibrate capacity with range of 10 MF to 45 MF with pointer of 1.5 MF. Invincible for the many services this instrument can perform for you.

Heathkit AUDIO OSCILLATOR KIT
The Heathkit Audio Oscillator will produce both sine and square waves within the frequency range from 20 CY to 20 KC in three ranges. Thermostat controlled linearity results in a variation of not more than ±1 ib in a 40 volt (full scale) output level. There will be less than 0.5% distortion from 100 CY through the audio range. Low impedance 600 ohm output. Precision 15 emitters, used in the range multiplier circuits to provide accurate calibration.

Heathkit VISUAL-AURAL SIGNAL TRACER KIT
The Heathkit Visual-Aural Signal Tracer features a special built-in RF input channel used in conjunction with a newly designed wide frequency range demodulator probe. High RF sensitivity permits signal tracing from the receiver antenna input. Separate low gain channel and probe available for audio circuit exploration. Both input channels are consistently monitored by an electron-ray beam indicator so that visual as well as aural indications may be obtained.

NOISE LOCATOR:
A dually tuned feature is a noise locator circuit used in conjunction with the audio probe. With this system, a DC potential is applied to a suspected circuit component, and the motion of the image in the component can be seen as well as heard. Invincible for ferreting out noisy or intermittent components, noisy rectifiers, potentiometers, IF and power transformers, etc.

WATTMETER:
Built-in calibrated wattmeter circuit will prove useful for quick preliminary check of total voltage consumption of a complete system under test. Separate panel terminals provide external use of the speaker or output transformer for substitution purposes. Same useful service tone by eliminating the necessity for speaker tone-masking service job. The same panel terminals also provide easy access to a well filtered B+ supply for external use. Don’t overlook the many interesting preening service possibilities provided through the use of this instrument, and let the Signal Tracer work for you by saving time and money.

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TELEVISION
TV SERVICE NOTES
By DAVID GNESIN
THESE items from the author's extensive TV trouble-shooting experience should prove helpful to other service technicians. In a few instances the notes apply to specific receivers (the author was servicing Philco and Transvision at the time), but most of them are fundamental suggestions that may help you out of a difficulty with any make set.

Transvision 12-inch kit
If the right side of the picture folds over, check to see if the horizontal hold control is erratic. There may also be insufficient horizontal sweep. Look for a leaky or shorted 150-muf capacitor in the grid circuit of the 6B6G-G output tube.

This would put a positive d.c. voltage on the grid of the horizontal output tube and upset the horizontal sweep current pattern.

Vertical linearity
Vertical linearity may sometimes be off just enough to be annoying, but not bad enough to warrant tearing the set apart to improve it. In some cases a quick check is simply to reverse the a.c.-line plug.

If this is done while a test pattern is on, a dramatic change in vertical wedge sometimes results. (Of course this may also indicate video-circuit or horizontal sweep troubles.) In any case, the vertical wedges sometimes do show improvement with this simple technique.

AM interference
Complaints turn up in the New York area on channel 4, in the form of parallel vertical lines traveling across the screen. These are produced by beats between the channel 4 picture carrier and an AM station in the immediate vicinity. It generally happens when the 300-ohm lead in—which acts as vertical antenna at low frequencies—is not accurately balanced to ground.

To correct the condition take a 1,000-ohm, 1/4-watt resistor, wind 15 turns of No. 32 enamelled wire over it, and solder this across the antenna-input terminals of the receiver.

Switching 6SN7's
The sync separator and amplifier have to handle both 60-cycle vertical sync and 15,750-cycle horizontal sync. The 6SN7's sometimes used in these circuits may perform better at one frequency that at the other. These tubes may test O.K. on the tube checker, but because of unequal or excessive internal capacitances, or leakage between triodes, they may not act efficiently in a particular circuit.

In receivers that have more than one 6SN7, try switching them around. You may find your whole service call consists of swapping the vertical oscillator tube with the horizontal oscillator or the sync separator tube.
THE HEATHKIT TC-2 TUBE CHECKER KIT

The Heathkit TC-2 Tube Checker was primarily designed for the convenience of radio and TV servicemen and will check the operating quality of tubes commonly encountered in this type of work. Test set-up procedure is simplified, rapid, and flexible. Panel sockets accommodate 12 standard pin tubes, 5 pin Hytron, and a built-in neon short indicator. Individual 3-position lever switch for each tube element, spring return test switch, 14 filament voltage ranges, and line-set control to compensate for supply voltage variations, all represent features of the TC-2.

Examples of tube tests are read directly from the large 4½" Simpson 3-color meter. Checks emission, shorts, elements, open elements, and continuity. Wiring procedure has been simplified through the use of multi-wired color coded cable providing a harness type installation between tube sockets and lever switches. This procedure insures standard assembly and imports a "factory built" appearance to the instrument. New Construction Manual furnishes detailed information regarding tube set-up procedure for testing of new or unlisted tube types. No delay necessary for release of factory data.

DECADE RESISTANCE KIT

MODEL DC-1
Shpg. Wt. 3 lbs.
$16.50

The Heathkit Decade Condenser KIT provides a study source of capacity values from 100 mmf to 111 mmf inclusive in capacity steps of 100 mmf. Silver plated contacts on husky ceramic switches, assure positive contact for each switch position. Precision silver wire condensers ±1% accuracy for close tolerance work.

REGULATED POWER SUPPLY KIT

MODEL PS-2
Shpg. Wt. 15 lbs.
$33.50

Here is a source of regulated D.C. voltage for circuit development work. Power supply voltage and current drain to the circuit under test are constantly monitored by the 4½" panel mounted meter. Separate 6.3 volt and 4 ampere A.C. filament source available. The regulated and variable output voltage will be constant over wide load variations, and hum ripple will not exceed .012% at 250 volts under a 50 MA load. Completely isolated circuit, standby switch, and other desirable features make the Model PS-2 extremely useful in a wide variety of applications.

HoTHKIT TV PICTURE TUBE TEST ADAPTER

No. 355
Shpg. Wt. 4 lbs.
$4.50

The Heathkit TV Picture Tube Test Adapter used with the Heathkit Tube Checker Kit will quickly check picture tubes for emission, shorts, etc. and determine tube quality. Consists of standard 12-pin TV tube socket, four feet of cable, octal socket connector, and data sheet.

AUDIO GENERATOR KIT

MODEL AG-8
Shpg. Wt. 11 lbs.
$29.50

Here is an Audio Generator with features generally found only in the most expensive instruments. Sine wave coverage from 20 cycles to 1 Megacycle—response flat ±1 db from 20 cycles to 400 Kc—continuously variable and step attenuated output. Because the output voltage is relatively constant over wide frequency ranges, the AG-8 is ideal for running frequency response curves in audio circuits. Once set by means of the attenuator, this voltage may be relied upon for accuracy within ±1 db. Instrument features low impedance 600 ohm output circuit and distortion less than .4 of 1% from 100 CPS through audible range.

DECEMBER, 1954
NEW Heathkit HIGH FIDELITY PREAMPLIFIER KIT

Here is the exciting new Heathkit Preamplifier with all of the features you Audiofiles have asked for and at a down-to-earth price level. Beautiful satin-gold baked enamel finish, striking control knobs and arrangement, attractive custom appearance and entirely functional design.

DESIGN: Uses three twin triode tubes in a shock mounted chassis, 2-12AX7 and 1-12AU7. Features tube shielding, plastic sealed color coded capacitors, smooth actung controls, good filtering, excellent decoupling, low hum and noise level, and all aluminum cabinet. Special balancing control for absolute minimum hum level, Cathode follower, low impedance output circuit for complete installation flexibility.

SPECIFICATIONS: Provides five switch selected inputs, 3 high level, and two low level, each with individual level controls. 4 position L.P., RIAA, A.E.S. and early 78 equalization switch. 4 position roll-off switch, 8, 12, 16 with one flat position. Separate tone controls, base 18 db boost and 15 db cut at 80 CPS, treble 15 db boost and 30 db cut at 15,000 CPS. Power requirements from Heathkit Williamson Type Amplifier power supply 6.3 volts AC at 1 ampere, and 300 volts DC at 10 ma. Over-all dimensions 12½" wide x 5 1/4" deep x 8 3/4" high.

APPLICATION: The new Heathkit WA-P2 Preamplifier has been designed to operate with any of the Heathkit Williamson Type Amplifiers and is directly interchangeable with the previous Model WA-P1 Preamplifier unit. Order your kit today and enjoy completely smooth control over the operation of your Hi-Fi system. Obtain the exact tonal balance of base and treble with the precise degree of equalization you want. Note that the design of the WA-P2 accommodates the newly established RIAA curve.

HEATHKIT TRANSMITTER KIT

The Heathkit AT-1 Transmitter has established a high reputation and has been enthusiastically accepted by hundreds of experienced operators as well as beginners. Power input up to 35 watts for the novice and suitable for a standby exciter for your higher powered rig later on.

Model AT-1 can be crystal or VFO excited and operates on 80, 40, 20, 15, 11 and 10 meters. The pre-wound coils with the oscillator and amplifier are switched simultaneously by the rugged hand switch. Meter switch allows a reading of the final grid and plate current on the panel mounted meter. Modulator input and VFO power sockets are provided as well as a key jack for CW operation. Other features include a crystal socket, standby switch, key click filter, AC line filter, good shielding and a 52 oz coaxial output. The 425 volt. 100 ma rectifier power supply of the AT-1 transmitter are more than adequate for the 6AG7 oscillator multiplier and 6L6 amplifier doubler.

HEATHKIT VFO KIT

The new Heathkit VFO is the perfect companion to the Heathkit Model AT-1 Transmitter and it has sufficient output to drive any multi-stage transmitter of modern design. Good mechanical and electrical design ensures operating stability. Coils are wound on stable, heavy duty, ceramic forms using Litz or double cellulose wire coated with Poly-urethane cement and baked for humidity protection. Variable capacitor of differential type construction, especially designed for maximum bandspread. Kit is furnished with a carefully pre-calibrated scale which provides well over two feet of scale length. Smooth acting vernier reduction drive and illuminated dial provides easy tuning and zero beating. Power requirements 6.3 volts AC at 45 amperes, and 250 volts DC at 15 mls. Just plug it into the power receptacle provided on the rear of the AT-1 transmitter. Seven band coverage 160 through 10 meters with 10 volt average RF output. Uses 6AU6 electron coupled Chapp oscillator and OAS voltage regulator.

HEATHKIT IMPEDANCE METER KIT

Determine antenna resonance and resistance, transmission line surge impedance, and receiver input impedance. Works with coaxial and one-quarter wave lines. Safe-wave and folded dipole, harmonic mobile and home antenna. Resistance type SWR bridge -100 megohm meter-frequency range 0-150 MC-impedance range 0-600 ohms.

HEATH company
BENTON HARBOR 20, MICHIGAN

LOW FREQUENCY COILS:

Low frequency range extended to 355 KC by the use of two additional coils. Complete with dial correlation curves. Set 341-A for C11-18 and 341 for GD-1A. Shpg. wt. 1 lb. Price $3.00

MODEL AM-1

$14.50

Shpg. Wt. 2 lbs.

ANTENNA COUPLER KIT

For the Heathkit AT-1 Transmitter or any comparable Amateur Transmitter. Will handle power up to 25 watts at its 32 ohm coaxial input. Maintains a wide range of antenna impedance with its L type tuning net-work and never indicated. A tapped inductor provides coarse adjustment and a transmitting type variable capacitor sets it "right on the nose." Will operate on the 16 through 80 meter bands.

HEATH company
BENTON HARBOR 20, MICHIGAN
Here is the newest Heathkit Hi-Fi Amplifier at the lowest price ever quoted for a complete Williamson Type Amplifier circuit. The W-1 Model has been designed for single channel construction, and only for the new Chicago "super-range" output transformer. This transformer, a new development in the Hi-Fi field, is being offered at substantial saving over transformers of comparable quality. It is outstanding in performance and on the basis of our tests, we find it equal in every respect to transformers used in the W-2 and W-3 Heathkit series.

**LOW PRICES:**
Through utilization of a single chassis with resultant economy obtained through elimination of duplicate sheet metal fabrication, connecting cables, power supply, and a new Chicago "super range" output transformer, a 20% price reduction has been made possible without sacrificing kit quality.

**COMPONENTS:**
The new Heathkit W-1 uses the same heavy duty power transformer and choke. It has all of the features of previous models including individual jacks and a built-in circuit board to balance the output tubes—plastic high quality capacitors and the correct components, size, and quality used in Williamson Type Amplifiers. Intermodulation distortion and harmonic distortion are both at the same low level as in the W-2 and W-3 models.

**CONSTRUCTION:**
Here is the opportunity for even the economy minded Hi-Fi enthusiast to enjoy all of the advantages offered through Hi-Fi reproduction at its lowest costs of Hi-Fi music lovers. Virtually distortion, clean musical reproduction, full range frequency response, and more that only Heathkit offers are reserved.

**FEATURES:**
Outstanding features of the Heathkit 20 watt Amplifier include frequency response of ±1 db from 20 CPS to 20 KC, Separate (boost and cut) bass and treble tone controls. Four switch selected input jacks and a special hum balancing control. Flexibility is emphasized in the input circuits and proper equalization for all input devices is incorporated.

**TUBE LINEUP:**
12AX7 magnetic preamplifier and first audio amplifier. 12AU7 two stage voltage amplifier and phase splitter. 2BU6 push-pull beam power output and 5U4G rectifier.

**MODEL A-9B:**
$35.50
Shpg. Wt. 24 lbs.

**MODEL A-7B:**
$15.50
Shpg. Wt. 10 lbs.

**WILLIAMSON TYPE AMPLIFIER KIT**

**NEW Heathkit 20 WATT High Fidelity AMPLIFIER KIT**

In keeping with the progressive policy of the Heath Company, further improvements has been made in the already famous Heathkit High Fidelity 20 Watt Amplifier. Additional reserve power has been obtained by using a heavier power transformer. A new output transformer designed and manufactured especially for the Heath Company, now provides output impedences of 4, 8, 16 and 500 ohms. The harmonic distortion level will not exceed 1% at the rated output.

**SPECIFICATIONS:**
Frequency response within ±1 db from 10 cycles to 100,000 cycles. 1M distortion at 5 watts equivalent output. 0.5% between 20 cycles and 30,000 cycles. 1M distortion at 5 watts equivalent output. 0.5% using 60 and 3000 cycles. Output impedences of 4, 8, or 16 ohms. Overall dimensions for each 7 1/2" h x 13/4" w x 13/4" d. Heigh, weight 29 lbs. Express only...

**COMBINATIONS AVAILABLE:**
W-3 Amplifier Kit (Includes Main Amplifier with Associated Output Transformer, Power Supply and W-4P2 Preamplifier). Shipping weight 35 lbs. Shipped express only...

**HEATH company**
BENTON HARBOR 20, MICHIGAN

**COMBINATION W-4 with Chicago "super-range" transformer only. Single channel main amplifier and power supply. Shipping weight 25 lbs. Express only...

**COMBINATION W-4M**
W-4M with Chicago "super-range" transformer only includes single channel main amplifier and power supply with W-4P2 preamplifier kit. Shipping weight 35 lbs. Express only...

**SIX WATT AMPLIFIER KIT**

An outstanding value, this economically priced 6 watt Amplifier is capable of performance expected only in much more expensive units. Only 2 or 3 watts output will ever be used in normal home applications and Model A-7B will be more than adequate for this purpose.

**SPECIFICATIONS:**
Two switch selected inputs are available for crystal and ceramic phonograph pickups, transistor, TV audio, tape recorder, and carbon type microphone. Model A-7B features separate bass and treble tone controls. Push-pull balanced output stages, output impedences of 4, 8, 16 and 15 ohms, and extremely wide frequency range ±1 1/2 db from 20 CPS to 20 KC. Not just a souped up AC-DC job. Full wave rectification, transformer operated power supply and good filtering, results in exceptionally low hum level.

**MODEL A-7C:**
Provides a preamplifier stage and proper compensation for the variable reluctance cartridge and low level microphone. $17.50

**W-3 Amplifier Kit (Includes Main Amplifier with Associated Output Transformer, Power Supply and W-4P2 Preamplifier). Shipping weight 35 lbs. Shipped express only... $69.50

W-3P3 Amplifier Kit (Includes Main Amplifier with Associated Output Transformer and Power Supply) Shipping weight 29 lbs. Express only...

$49.75

**HEATH company**
BENTON HARBOR 20, MICHIGAN
Heathkit COMMUNICATIONS RECEIVER KIT

An excellent example of typical Heath Company ability to produce top quality kit merchandise at ridiculously low prices is the AR-2 Communications Receiver. Here is a transformer operated all-wave receiver with all of the desired features and none of the disadvanges commonly considered in so-called "economy sets.

Receiver employs high gain miniature tubes and 1P transformers, chassis mounted 555 PM speaker, headphon jack, slide-rule dial with Ham Bands plainly identified, and easy tuning with direct planetary drive. Continuous frequency coverage from 550 KC to 35 MC on 5 bands, with electrical high-speed tuning and logging curves. Other features are IF gain control with AGC on-off switch-phonostandby-CW panel switch - prewound coils in shielded turret assembly and copper plated chassis and shielding.

Uses 12B6 mixer-oscillator, 12B6A IF amplifier, 12AV6 detector-first audio, 12A6 beam power output, 12BA6 RFQ oscillator, 5Y3 IF transformers, and 5V3 rectifier. A lettered control plate is provided for the labeling of your chasis or you can order the optional Heathkit cabinet featuring the full size aluminum panel.

RECEIVER CABINETS

Pre-assembled cabinets generally plywood chassis available for BR-2 and AR-2 receivers. Espe practically excellent panel speakers. Use for BR-2 Receiver, Cabinet 91-9

Shipping weight 5 lbs ....................................... $4.50

AR-2 Receiver, Cabinet 91-10

Shipping weight 5 lbs ....................................... $4.50

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RADIO-ELECTRONICS
In trouble-shooting the picture-tube power supply the service technician runs into an ironic situation. Voltages in this circuit are so high the technician has no difficulty determining whether they are present. But just because these voltages are so high conventional trouble-shooting techniques and equipment cannot be used.

The flyback or "kickback" power supply has proved to be the most efficient system and is almost universally used. Technicians will seldom come across an r.f. power supply these days. We will consider only the flyback system.

High voltage is a byproduct of the horizontal deflection circuit. Though many modifications have been made in the design of the horizontal system to improve high-voltage circuit operation, it nevertheless remains true that a properly operating second-anode supply can result only from a properly operating horizontal deflection circuit.

The dependence of the high-voltage circuit on other circuits can be seen in Fig. 1. The a.c. voltage necessary to drive the high-voltage rectifier is generated in the horizontal deflection coils. During each horizontal sweep the current flowing through the deflection coil increases at a linear rate, causing the electron beam to sweep horizontally across the face of the screen. When the plate current of the horizontal output tube is cut off at the end of each horizontal trace, the magnetic field around the horizontal deflection coil collapses rapidly, generating a large inductive kick. This kick consists of an approximately 1,200-volt negative pulse.

With the horizontal output tube cut off and with the deflection coil connected directly across the secondary (L2 in Fig. 1) of the horizontal output transformer, the deflection coil acts as a generator of large pulses. During this time the output transformer acts in reverse. The usual secondary winding acts as the primary, and the usual stepdown function of the transformer becomes one of voltage stepup.

The stepdown ratio in these output transformers varies—let us assume ours is 5 to 1. Thus we can expect a 1-5 enormous pulse voltage that a note appears on almost all TV schematics warning the service technician not to attempt direct voltage measurements at the plate of the horizontal output tube. If measurement is necessary, a high-voltage probe should be used.

There is an additional winding (L2) connected in series with the primary; the two together constitute an auto-transformer. The number of turns in L2 is usually several times that in L1. The exact ratio determines the positive pulse voltage that—increased by autotransformer action—is fed to the plate of the high-voltage rectifier tube.

The flyback transformer used to create these high potentials has undergone a great deal of modification. Since the development of large-screen tubes, the older powdered-iron units have given way to the present ceramic-core high-efficiency types, making possible the necessary higher voltages.

Rectifier

The high-voltage rectifier does the same job as any conventional rectifier. The high pulse voltage is applied to the plate of the rectifier tube and converted into a d.c. voltage (Fig. 2-c) suitable for second-anode use on a TV picture tube.

Because of the relatively high frequency of the pulse voltages applied to the rectifier tube (15,750 cycles) and the small current drain (approximately 200 μa), the filtering circuit is simple.

The rectifier is usually a half-wave unit such as the 1B3-GT or 1X2. A high-voltage rectifier tube is unique only in its ability to handle large pulse-shaped a.c. waves, sometimes as high as 20 kv. Since the rectified high voltage is taken from the filament of the tube, heating the filament with a conventional transformer would be impractical. Extremely heavy insulation would be required. So filaments are heated by connecting them to one or two loops of heavily insulated wire wound around a section of the horizontal output transformer. Both the 1B3-GT and the 1X2 draw about ¾ watt of filament power (1.25 volts, 0.2 amp).

In series with many filaments is a
WHEN YOU THINK

Each Collins Tuner Kit is complete with punched chassis, tubes, power transformer, power supply components, hardware, assembly, tuning eyes, knobs, wire, etc., as well as the completed sub-assemblies: FM tuning units, AM tuning units, IF amplifiers, etc., where applicable. All sub-assemblies wired, tested and aligned at the factory make Collins Pre-Fab Kits easy to assemble even without technical knowledge. The end result is a fine, high quality, high fidelity instrument at often less than half the cost — because you helped make it and bought it direct from the factory.

FM/AM Tuner Kit

The original 15 tube deluxe FM/AM pre-fab kit redesigned on a smaller chassis. The tuner now measures 14" wide by 12" deep by 7½" high. This attractive new front end and dial assembly opens up new applications where space is at a premium. Kit includes everything necessary to put it into operation — punched chassis, tubes, wired and aligned components, power supply, hardware, etc. Kit comprises FM/3 tuning unit, IF-6 amplifier, AM-4 AM tuning unit, magic eye assembly and complete instructions. All tubes included. Shipping weight 19 lbs.

AM-4 Tuning Unit

If you think of tuners, think of Collins Audio Products.

When you think of tuners, think of Collins Audio Products.

FM Tuner Kit

The FM-11 tuner is available in kit form with the IF Amplifier mounted in the chassis, wired and tested by you. You mount the completed RF Tuning Unit and power supply, then after some simple wiring, it's all set to operate. 13 tubes: 6J6 RF amp, 6AG5 converter, 6C4 oscillator, 6BA6 1st IF, 6AU6 2nd and 3rd IF, 6AL5 discriminator, 6AL5-GT double tuning eye, 5Y3-GT rectifier. Sensitivity 4 to 10 microvolts, less than ½ of 1% distortion, 20 to 20,000 cycle response with 2DB variation. Chassis dimensions: 12½" wide, 8" deep, 7½" high. Illustrated manual supplied. Shipping weight 14 lbs.

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AM-4 Tuning Unit

This high fidelity tuning condenser gives 3 tuned stages with high sensitivity and selectivity. Assembly is completely wired, tested and aligned ready for immediate use. Frequency coverage 550 KC to 1650 KC at a sensitivity of 5 microvolts. Tubes 6BA6 RF amplifier, 6B6A converter, 6BA6 IF amplifier and 6AL6 detector. 30 ma @ 200 volts. Mounts on a chassis plate measuring 4½"x7½". Shipping weight 2½ lbs. Dial available at $3.85.

FM/3 Tuning Unit

The best for FM. The most sensitive and most selective type of "front end" on the market. 6 to 10 microvolts sensitivity. Image ratio 500 to 1. 666 tuned RF stage, 6AG5 converter, 6C4 oscillator. Permeability tuned, stable and drift-free. Chassis plate measures 6½"x4½". In combination with the IF-6 amplifier, the highest order of sensitivity on FM can be attained. Tubes included as well as schematic and instructions. Draws 20 ma. Shipping weight FM-3: 2½ lbs. Dial available @ $2.85.

IF-6 Amplifier

$19.75

6 Tubes. Shipping Wt. 3 lbs.

AM-4 Tuning Unit

$24.50

7 tops in AM superhet performance! A 3-gang tuning condenser gives 3 tuned stages with high sensitivity and selectivity. Assembly is completely wired, tested and aligned ready for immediate use. Frequency coverage 550 KC to 1650 KC at a sensitivity of 5 microvolts. Tubes 6BA6 RF amplifier, 6B6A converter, 6BA6 IF amplifier and 6AL6 discriminator. 50 ma @ 200 volts. Mounts on a chassis plate measuring 4½"x7½". Shipping weight 2½ lbs. Dial available at $3.85.

FM Tuner Kit

$55

with AFC $58.50

The FM-11 tuner is available in kit form with the IF Amplifier mounted in the chassis, wired and tested by you. You mount the completed RF Tuning Unit and power supply, then after some simple wiring, it's all set to operate. 13 tubes: 6J6 RF amp, 6AG5 converter, 6C4 oscillator, 6BA6 1st IF, (2) 6AU6 2nd and 3rd IF, (2) 6AL5 limiters, 6AL5 discriminator, 6AL5-GT double tuning eye, 5Y3-GT rectifier. Sensitivity 4 to 10 microvolts, less than ½ of 1% distortion, 20 to 20,000 cycle response with 2DB variation. Chassis dimensions: 12½" wide, 8" deep, 7½" high. Illustrated manual supplied. Shipping weight 14 lbs.

FM/AM Tuner Kit

$77.50

with AFC $81.00

The original 15 tube deluxe FM/AM pre-fab kit redesigned on a smaller chassis. The tuner now measures 14" wide by 12" deep by 7½" high. This attractive new front end and dial assembly opens up new applications where space is at a premium. Kit includes everything necessary to put it into operation — punched chassis, tubes, wired and aligned components, power supply, hardware, etc. Kit comprises FM/3 tuning unit, IF-6 amplifier, AM-4 AM tuning unit, magic eye assembly and complete instructions. All tubes included. Shipping weight 19 lbs.

TELEVISION

current limiting resistor. It is frequently imbedded in high-voltage wax and helps to prevent filament burnout. This resistor requires special attention. Because of the pulsed nature of the filament current, it tends to char.

The filter circuit is usually a simple R-C network. The resistor is generally about 1½ megohm and the capacitor 500 µf. In some cases even the filter resistor is omitted.

To simplify the circuit further, the 500-µf filter capacitor frequently consists of the capacitance between the Aquadag coating in the picture tube and a second conductive coating on the outside surface of the tube. The glass of the tube acts as the dielectric and has a very high voltage breakdown rating. The outer coating is usually grounded by spring wires connected to the deflection yoke bracket.

Since the capacitor is so small, and the current drain low, a dangerous charge cannot be stored, making this type of power supply relatively safe. Regulation is also very poor, the output voltage decreasing sharply with overload. Conventional high-voltage capacitors used in this type of circuit are constructed as to present a high-resistance leakage path to the high voltage.

This type of power supply is referred to as the flyback type because the pulse voltage is developed during the horizontal retrace period. During this time, an ac signal of extreme interest to the service technician takes place. Fig. 2-b shows the falling edge of the inductive kick overshooting its starting position. This makes the plate of the output tube extremely sensitive with respect to the screen grid. As a result, electrons oscillate around the screen grid, resulting in a light vertical bar at the left edge of the picture tube (see photo). This is the usual symptom of Barkhausen oscillations.

These are radiated from the output tube and frequently picked up by the antenna lead and fed back through the receiver. A strong signal will usually eliminate this interference, because the large a.c. voltage developed will make the receiver insensitive to this usually weak signal.

The negative swing is a resonant effect caused by leakage inductance in the flyback transformer. Elimination of this and other defects involving the high-voltage rectifier will be discussed in our next issue.

Any imperfection in the horizontal deflection circuit is bound to reduce the amplitude of the inductive kick and thus the second-anode voltage. This will cause defocusing and a "soft" beam, permitting the deflection voltages to have greater influence and creating an effect known as "blooming."

Many receivers contain an R-C network between the windings of the auto-transformer (Fig. 2-a) included as a safety factor to prevent damage in case the plate cap of the high-voltage rectifier tube accidentally touches the chassis. The network will

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TELEVISION
also provide protection against a burn in case the technician contacts the high-voltage line. It also limits the amount of current flow in short circuits and minimizes shock hazard in transformerless sets.

These components have no effect on circuit operation unless they become defective. Should the capacitor open, the high voltage drops considerably and there will be no brightness. Its value is approximately 0.001 μF, and it provides a path for the 15,750-cycle pulses across the resistor.

Spooks
I have a Sparton receiver, model 24U174, and get a very thin vertical line at the extreme left edge of the raster. At first I took it for Barkhausen oscillations, but I notice it only on the lower channels and it seems thinner and differs in appearance from Barkhausen oscillations. It appears to be the effect of some kind of radiation. I have changed the horizontal oscillator and output tubes, the damper and the high-voltage rectifier. I have also checked the entire output circuit for arcing but there doesn't seem to be any. I would appreciate your help.—A. R., Webster, N. Y.

The probability is that the effect you describe is "spooks." It is seldom troublesome because it generally appears beyond the visible portion of the raster and can be ignored. It is only comparatively recently that particular attention has been paid to this trouble, and it is the effect of radiation.

Many harmonics are generated in the damper circuit as a result of the rapid rise in deflection current. These are radiated, picked up by the antenna, r.f. or i.f. circuits, and appear on the screen. Aside from adjusting the deflection circuit for maximum linearity, there are very few remedies. The most effective check on this radiation consists of inserting r.f. chokes of 1-15 μH in the cathode, heater and plate circuits of the damper tube. Small bypass ca-

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TELEVISION

Capacitors may also be used. See Fig. 3 for a typical circuit.

This effect is most noticeable on the low-band channels because the strength of harmonics decreases with increase in frequency.

Since the vertical line appears on the extreme left edge of the picture, should you have difficulty eliminating this interference, simply increase the width slightly to move the line behind the mask. Most receivers are capable of a little extra width.

Burned plate resistor

I have an Admiral model 21D1 on the bench for the second time. In each case the complaint was the same. No pictures on the weak-signal channels and weak pictures with lots of snow on the stronger ones. The set uses a high-pass cascode tuner. The first time in the shop I replaced a few components in the tuner and that corrected it. Now one of the components I replaced the last time, the cascode plate resistor, is defective. I don't want just to replace it and have another callback. I'd appreciate some help.—C. E., Richmond, Va.

The plate resistor was probably burned out by excessive plate current in the r.f. amplifier. This is especially true in the type 6BK7 that is used in this receiver. The r.f. amplifier circuit is far too critical to attempt reducing the plate current. A more practical approach would consist of replacing the plate resistor with a 1-watt unit. It will easily handle the current flow.

In many cases this failure is the result of a shorted bypass capacitor between ground and the plate end of this resistor. There have been several instances where this capacitor shorted to the cover plate of the tuner. After replacing the resistor, carefully check the bypass capacitor before reassembling the tuner.

Critical focus

I would like some information on a Capehart receiver, chassis CX-33. I have had this set for over two years and cannot find the cause of its critical focus. There is no instability involved. When the focus is adjusted for optimum sharpness, it will stay put for a few weeks. Then it will go out of focus again and I have to re-adjust the focus control. I have checked every component in the focus circuit and all voltages. Every voltage, as far as I can determine, is exactly as the manufacturer's schematic.—S. E., Biloxi, Miss.

Check the direction of taper of the focus control. If it is reversed, the focusing point, while obtainable, will be extremely critical—so much so that even the slight aging of any of several tubes will cause defocusing. Set the control at the center of its rotation and measure the resistance between the center terminal and the unused outer terminal—you should get about 3,500 ohms. If the resistance measures about 1,000 ohms, reverse the outer terminal connection.

END

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Signal Outputs
1. Composite video of either polarity, adjustable amplitude to 2 volts across 90 ohms.
2. Modulated R.F., channels 3 through 5, 0.1 volt across 300 ohms.
3. Horizontal sync., positive polarity, 8 volt across 200 ohms.
4. Color subcarrier, 4 volts across 200 ohms at burst phase.

Synchronizing Signals
1. Horizontal sync. and blanking (F.C.C. standards).
2. Vertical sync., serrated and locked to horizontal.
3. Vertical blanking.

Video Signals
1. Dots (nominal 108 dots)
2. Crosshatch (nominal 12 by 9)
3. Color bars (5 switch positions)
   Position 1—Multi.
   8 Bars in the following sequence:
   1. White, relative luminance 1.0, chrominance zero
   2. Yellow, relative luminance 0.89, chrominance 0.44
   3. Cyan, relative luminance 0.70, chrominance 0.63
   4. Green, relative luminance 0.59, chrominance 0.59
   5. Magenta, relative luminance 0.41, chrominance 0.59
   6. Red, relative luminance 0.30, chrominance 0.63
   7. Blue, relative luminance 0.11, chrominance 0.44
   8. Black, relative luminance zero, chrominance zero
   Luminance and chrominance held to 10 percent, phase angles to ±3 degrees.
   Position 2—Color Difference.
   7 Bars of zero luminance in the following sequence:
   1. Black, relative chrominance zero.
   2. I, relative chrominance 0.25
   3. Q, relative chrominance 0.25
   4. Black, relative chrominance zero
   5. R-Y, relative chrominance 0.25
   6. B-Y, relative chrominance 0.25
   7. Black, relative chrominance zero
   Phase angles held to ±2 degrees.
   Positions 3, 4, and 5—Single bars, luminance 0.3, chrominance 0.5, occupying approximately 60% of screen width.
   1. Red (position 3)
   2. Green (position 4)
   3. Blue (position 5)

Sound Carrier, approximately 25% of peak picture carrier, placed 4.5 megacycles from picture carrier.

Panel Controls
1. R.F. carrier tuning, channels 3 through 5.
2. Video output amplitude.
3. Horizontal lock.
4. Standby switch (sound on, sound off).
5. Video output polarity switch.
8. Color bar switch (Multi, Color Diff., Red, Green, Blue).

Internal Adjustments
1. Burst amplitude.
2. 3.58 frequency vernier.
3. Sync. lock controls.

Tube Complement
12A17 — 8
12AX7 — 4
6U4-G — 1
6AL5 — 1

Circuit Operation
1. Color subcarrier and sound frequencies are determined by crystal oscillators.
2. Color phase angles are determined by an accurate, low impedance delay line.
3. Direct gating of proper chrominance phase is employed for each color bar to attain maximum stability and reliability rather than usual methods utilizing quadrature encoders.
4. Serrated vertical sync. is maintained on integral divider of horizontal rate.
5. Luminance and chrominance levels are reliable and stable. No multivibrators are employed in generating color bars.
6. No internal or external adjustments are necessary for proper bar widths, luminance or chrominance levels. For use on 105-125 volts 60 cycle AC.

Physical Characteristics
Dimensions — 16½" wide x 9½" high x 10½" deep. Designed to match other Jackson TV Equipment, both in styling and exterior dimensions.
Weights — 27 net pounds. Shipping 32 pounds.

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DECEMBER, 1954
MERELEY because the technician succeeds in localizing the trouble in a “dog” chassis, it should not be assumed that the receiver can be easily restored to normal operation. For example, it may be found that a horizontal output transformer has some shorted turns. But unless a replacement can be obtained, it may be a big job to re-store the receiver to normal operation. This situation becomes aggravated when the transformer operates in a complex arrangement, feeding back pulse voltages to the a.c., horizontal oscillator, and keyed a.g.c. circuit, as well as to the booster and high-voltage circuits. In some arrangements, pulses are also fed to blanking circuits to eliminate retrace lines and Barkhausen lines, for operation of noise gates, and to peak the horizontal drive voltage.

A big job can be avoided by using a replacement transformer close to the original specifications. But sometimes the remaining differences are sufficient to prevent proper operation. If the operating waveforms and peak-to-peak voltages are available, the technician can sometimes design correcting circuits to compensate for the replacement transformer differences. This procedure comes under the heading of redesign, which most service shops (wisely) do not care to tackle.

In some instances, technicians suggest converting the entire circuit to a standard arrangement. Since this is costly, the customer often refuses to permit it. In such a case it becomes necessary to advise the customer that the receiver is obsolete and no longer worth fixing. Any reasonable customer will accept the logic of the situation.

Sometimes waveforms and peak-to-peak voltage data cannot be obtained for a particular chassis. In a situation of this kind, there are two possible procedures. In some cases, the necessary data can be found for another make of receiver which uses the same circuit arrangement. This requires considerable searching. A great deal of time can be saved if cross-reference data are available to the technician. Since cross-reference data are not published commercially, the technician must build up his own file of pertinent data. These are best filed in card-catalog form for ready reference.

The other procedure involves obtaining another receiver of the same type for comparative measurements of wave shapes and peak-to-peak voltages. If this method is used, the technician should record the peak-to-peak voltages and sketch the waveforms, so that he will have them available if a similar situation arises in the future.

When another receiver of the same type can be obtained, whole sections of the faulty receiver can be checked out. As shown in Fig. 1, the entire sync section of an auxiliary receiver (of the same type) can be substituted for the sync section of a faulty receiver. The video signal from the cathode of the video amplifier is applied to the grid of the sync separator.

To check the entire sync section by substituting the sync section of an auxiliary receiver, the chassis are grounded together and the signal paths broken at suitable points. Test leads are then used to patch the section; they should be kept reasonably short. If there is trouble in the sync section of the faulty receiver, normal operation will be resumed when the test leads are connect- ed. Various stages in the sync circuit can be eliminated by similar procedure, thus “closing in” on the faulty stage or component.

If normal operation is not resumed, it is apparent that the fault does not lie in the sync section, but in another section of the receiver. The sweep section or the video section of the receiver can be patched in the same manner.

*Field Engineer, Simpson Electric Co.

Fig. 1—Substituting the sync section.
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Here's Mighty Mo's proof...
...the results of ACTUAL FIELD TESTS.

- On channels 2 to 13, Mighty Mo outperforms every other antenna manufactured today.
- Higher uniform gain over all channels. Does not vary more than $1/2$ D.B. on any channel across band. Perfect on color TV.
- Clearer, sharper, deeper pictures on all channels.
- Higher average gain than 6 of the most advertised antennas.

MIGHTY MO'S FEATURES

- DDP (Double Diamond Phasing) precision-timed phasing regulator enables the weakest of signals to be trapped and then boosted to a clear, magnificently sharp, photo-like picture.
- Flat response... a must for color reception.
- Largest screen area... over 70 sq. ft. Screen elements spaced less than 1/10 wave length apart for maximum reflector efficiency.
- Highest front to back ratio ever achieved.
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TELEVISION

Caution must be used in the case of transformerless receivers in which the power plugs may be “crossed,” resulting in a short circuit. Transformerless receivers should always be powered by a line isolation transformer. The auto-transformer type of unit is unsuitable and will not isolate the receivers—separate primary and secondary windings must be provided.

Waveforms in faulty receivers

Interpretation of waveforms is not always an easy matter. When an auxiliary receiver is available, comparison checks can be made back and forth between the faulty receiver and the normal receiver. When an auxiliary receiver is not available, the technician may have trouble because of the incompleteness of the published data. In many cases the waveform and peak-to-peak voltage may check out correctly, but the circuit action will still be faulty because of minor spurious components in the waveform. Whether or not such minor spurious components are tolerable cannot be determined when the published waveforms are sketched rather than photographed; the technician making the sketch almost always deletes minor components which may be present in normal operation.

Trouble-shooting a dog

A dog receiver was brought in with the complaint that the picture pulled when the contrast control was set to a normal level. New tubes did not improve the situation and all d.c. voltages and resistances measured correctly. As the contrast control was advanced, the test-pattern circle became distorted. Inspection showed that the picture pulling followed black lettering in the pattern or followed black areas such as the horizontal wedge. This in itself was a clue since it was apparent that the distortion was due to the entry of video signal into the sync circuits.

In cases of this kind it is usually advisable to check the d.c. supply lines for spurious a.c. voltages. Spurious voltages can couple from one receiver section to another in the event of failure or partial failure of decoupling or bypass capacitors. Applying the scope across the B plus supply line to the sync section showed that various spurious a.c. voltages were present, at a level of 0.3 volt peak-to-peak. This level is high enough to cause trouble in some chassis, but not in all. Since the service data for the receiver did not state the normal level of a.c. voltage in this line, it was necessary to regard the spurious voltage with suspicion, and to eliminate the possibility of trouble from this source. Additional filter capacitance was shunted from the B plus line to chassis until the spurious voltage dropped to 0.1 volt peak-to-peak—with no apparent change in the picture pulling. Hence, the B plus line was given a clean bill of health.

Next the waveform at the horizontal phase detector was checked on the
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TELEVISION

scope with the aid of a low-capacitance probe. The waveform appeared to have the same shape as the published waveform and the peak-to-peak voltage checked within 20%, usually considered satisfactory. However there was evidently some small amount of fuzzy spurious voltage also present, which did not appear in the data. It was questionable whether this spurious voltage was video information because, if the spurious voltage actually was residual video signal, it could easily account for picture pulling at this point in the circuit.

To make the check, the technician touched his finger very lightly to the tip of the low-capacitance probe while also touching the probe tip to the plate of the phase-detector tube (the body always picks up 60-cycle hum around the test bench). This mixed a small amount of 60-cycle hum with the waveform voltage at the phase detector. This type mixture causes a velocity modulation of the waveform, which develops a picture on the screen of the scope if the waveform contains video information. As a result of the test, the technician observed half the test pattern appeared in bas-relief on the scope screen and the suspicion that the spurious voltage consisted of video signal was verified. (See Fig. 2.)

Fig. 2—Testing of spurious voltages.

Only half the test pattern appears because the 60-cycle hum voltage produces a folded-over vertical sweep in the scope, just as a 60-cycle sine-wave voltage produces a folded-over picture when applied to the vertical deflection coils of the picture tube. The picture on the scope screen appears to be in bas-relief because the picture is being developed by velocity modulation, with the result that the high-frequency portions of the video signal are much more apparent in the reproduced signal than are the low-frequency portions.

Since it was now known that the fuzzy portion of the display consisted of spurious video signal, it was apparent that the trouble was in the sync separator where the camera signal is normally removed from the sync pulses. From this point on, the job was simple enough.
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DECEMBER, 1954
TELEVISION

As an illustration of a similar situation, there is a large amount of vertical sync pulse present along with horizontal sync pulse voltage at the grid of various pulse-width a.f.c. arrangements. In case of trouble with the circuit, the presence of this vertical sync pulse, might be a matter for concern, if idealized waveform sketches are referred to. However, the presence of the vertical component is normal and does not disturb satisfactory operation.

Waveform photographs are usually more satisfactory from this point of view if taken in pairs; one photograph should be taken using a 60-cycle scope sweep, another using a 7,875-cycle scope sweep. The reason is that low-frequency spurious voltages may be all but invisible on fast sweep, and vice versa. Some service data are available in which the waveforms are illustrated for half the line and field sweep rates. These will be valuable in servicing "dog" receivers.

The mixing of 60-cycle hum voltage with the input signal to the scope is also useful to determine whether distortion of low-frequency video voltages is being caused by 60-cycle hum or by reactive distortion. For example, when aligning an i.f. amplifier, the technician sometimes observes that the trace and retrace are displaced as shown in Fig. 3. This distortion may be caused by 60-cycle hum or by capacitive reactance. The distortion opens up the base line (when the scope is operating on 60-cycle sine-wave horizontal sweep) and may be caused by heater-cathode leakage in a tube, by a by-pass capacitor in circuits carrying the video signal, or by a coupling capacitor having too small an value.

In any case, the technician should determine whether the distortion is hum distortion or reactive distortion. As shown in Fig. 3, this determination is made by injecting a small amount of 60-cycle hum voltage into the scope input circuit by touching the finger lightly to the vertical input post of the scope.

If the end of the display closes up into a straight base line, but the other end of the display opens up into a large loop, reactive distortion is present. But if the display opens or closes equally at both ends, hum distortion is present.

Thus it is seen that hum voltage can be a very useful tool in practical service work, when the operator knows how to use it.

Fig. 3—Trace and retrace scope curves.

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Thus it is seen that hum voltage can be a very useful tool in practical service work, when the operator knows how to use it.
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A recent exhibition I was taken gently by the hand by a competitor friend and led like a sheep to the slaughter to be shown a most beautifully constructed control panel several yards long that was covered with dials, gauges, switches, knobs and buttons, from real big grandfathers to lovely little babies that simply screamed to be twisted. By inserting a Yale key into a slot somewhere in the side, my friend opened this wonder of wonders. It was a picture to behold. There were rows of tubes from big inverters and rectifiers to tiny little subminiatures that should have been in hearing aids but had been mislaid. The complete chassis layout was a mass of hinged breadboards.

I gasped and held my breath. Here was a masterpiece. Stifling my natural jealousy, I applauded and crawled away, back to my little stand of ordinary simple meters and controls. My heart was full. This was competition with a capital C.

Three months later I was called north to a large textile plant to fit a control to a calender in a hurry. The job was to fit a continuous thickness gauging unit to indicate automatically the thickness of the linoleum plastic as it left the cylinder and, when a change in thickness occurred, to readjust back to standard within a matter of seconds.

The calender was a massive machine some 60 feet long with two heads raised about 20 feet in the air, each head consisting of three rollers and bearings with the necessary standards to carry them. The width of the machine was about 12 to 14 feet and the calender cylinders measured about 100 inches long, one being 60 inches in diameter and weighing 26 tons, the next or middle 18 inches in diameter and weighing 4 tons, the last or polishing roller 30 inches in diameter and weighing 6 tons.

The polishing roller was fed with freezing brine through the bore to keep it cold, the middle roller was fed through the bore with boiling water to keep it hot. Plastic material of the thermoplastic type was fed between these two rollers and the thickness of the linoleum plastic was controlled by the gap allowed between the two peripheries. Heat and cold from the two cylinders also affected the thickness, as did the speed of each roller, each of which was driven independently by a variable-speed 100-h.p. d.c. motor.

There—covered with a dust sheet in one corner of the shop—was a familiar sight. A control panel several yards long, covered with knobs and meters and a nice little slot where one could insert a key to open the side and see inside. For three months it had stood there coupled to the big calender, and for three months every time one little thing went wrong a specialist had to be called from 200 miles away in London.

---

Fig. 1—Schematic of a simple control which gauges the thickness of linoleum strip continuously and automatically.

---

To be useful, they must be simple enough to be operated and adjusted by the average plant electrician.

By J. R. CORNELIUS*
to open that lock and lift up one or more of those bread baskets until the panel had cost more in spoiled loaves than it had cost in cash. Why? It was a beautiful panel for the laboratory where what in this country is called a boffin is always available to put it right, but it was too, too complicated for the plant electrician who was supposed to be its day and night nurse.

We want something simple, said the foreman, something that John there can wrestle with in between his motors and lights. Give us something quick, said the manager, and it doesn't have to look pretty. Give us something that works, said the president, that's all we want.

This is what I gave them. It fitted into an oblong wood box 20 inches long and 12 inches square. There was not a knob to be seen except one on-off switch. The only adornment was my nameplate and some dirty finger marks, but the unit worked as soon as it was put in, and is working still. It's not pretty, but against the bulk of the whole machine that is not noticed, in fact it's not noticed at all.

That's how electronic instruments should be. Once they are installed, they should be unobtrusively efficient. They can't be continually going wrong and needing a wet nurse night and day. Let us look at this simple job. It combines practically everything in old-fashioned electronics, simple straightforward engineering. The device is covered by British patents and by U.S. patent No. 2,387,946.

The circuit

The general scheme is shown in Fig. 1. There is an oscillator, frequency multiplier, frequency discriminator, double-diode detector, signal discriminator, and two power tubes. The d.c. is obtained from metal rectifiers. There are two voltage regulator tubes, not shown.

The oscillator uses a 100-kc X-cut high-quality crystal and is coupled to the amplifier by a frequency-doubling tuned circuit. The second tube amplifies the signal and feeds the discriminator section via a tuned circuit also. The primary is fixed-tuned; the secondary is detuned some 15 kc and includes in its circuit the control capacitor Cl. It acts as a frequency discriminator in the manner of the "slope detector."

The signal from the frequency discriminator is fed to the plate of the first half of the double-diode detector and a voltage from the 60-cycle line is fed to the cathode of the other half. The signal from the first half of this tube is positive; that from the second half negative. By adding these two voltages together in the grid resistor of the signal discriminator we have a result that varies with the value of the signal fed to the first half of the detector, and can be set to zero with the variable resistors in the cathode circuit of this tube and the grid circuit of the signal discriminator. Now any change in the signal voltage from the frequency discriminator will drive the signal discriminator grid more positive or more negative.

The operation of capacitor Cl is shown in Fig. 2, where A is the middle calender cylinder; B, a shoe pressing onto this cylinder and moving up or down in block C with any movement of the cylinder, and F, springs holding B against the cylinder face. D is one plate of the capacitor Cl, and E the other grounded plate, which is integral with a micrometer screw J operated by the control motor G, which acts through the step-by-step transmitter switch H to change the reading on the indicator dial. Fig. 3 shows the unit (minus the hot plate) complete and also a part view of the reduction gear driving the micrometer and transmitter.

If capacitor C1 is adjusted to a definite value to give, say, zero volts at X, then an increase in the gap between the plates will cause a positive voltage at X and a decrease in the gap will cause a negative voltage.

Since an increase in capacitance causes a negative voltage at X (Fig. 1), and a decrease in capacitance causes a positive voltage, the output of this double triode, if fed correctly to the two power tubes, will cause an unbalance to appear across the plates in one direction or the other. Thus if a gap of 0.1 mm between the control plates gives a balanced pair of power tubes, an increase in this gap will cause a voltage to appear at X which will unbalance the power tubes and cause an unbalance of current in the split fields of the control motor coupled to the micrometer drive. This voltage will persist until the motor has closed the gap to the correct value or until the cylinder control motor closes the cylinder gap, whichever is included in the circuit. A decrease in the gap will unbalance the fields in the opposite direction and cause the coupling to open the gap.

Assume that the equipment is set up for 6-mm thickness and the dial is adjusted to indicate this value. Temperature and other factors affect the characteristics of the raw material and the

Fig. 2—Drawing shows physical adjustment of the control capacitor.

Fig. 3—Mechanical control unit, also a closeup showing the reduction gear.

Fig. 4—the dial indicates thickness of plastic in mm and hundredths.
ELECTRONICS

Color's bigger than 96 in color's bigger than 96.75 mm, the control motor immediately starts and restores C1 to its original value and simultaneously operates the rotary switch to stop the pointer around to indicate the thickness of the material at that instant. The roller adjustment screws can then be operated manually to restore the thickness to the desired value.

This system also can be connected (at a little greater cost) so the spacing between the rollers is automatically adjusted to maintain the thickness that is indicated on the manually preset thickness indicator. In this case the dial remains fixed and the correction is applied to the screws controlling roller spacing. There are no tricky circuits. Seven simple tubes and three little potentiometers. It was a masterpiece when I put it in, but—like a good servant—it has already been forgotten. That's how it should be.

MULTIPURPOSE TOOL

Solderless connections is one of the many functions of a new tool called the Super Champ. An extremely versatile tool, it shears stove bolts and machine screws; cuts wire, sizes 22 to 10; strips wire, sizes 22 to 10; serves as a wire size gauge, and crimps insulated solderless terminals and connectors. This tool should prove especially useful in antenna work where soldering usually requires drawing long power lines. The tool is manufactured by Aircraft-Marine Products, Inc., Cincinnati, Ohio.

PORTABLE—115 or 230 V, @ 50 to 60 cycles—KEYER TG-34A is an automatic unit for reproducing audible code practice signals previously recorded on paper tape. By use of the self-contained speaker, the unit will provide code practice signals to one or more people on provide a keying chart with a hand key. The unit is compact, in portable carrying case, and complete with tubes, photo cell, and operating manual. Size: 10½" x 10½" x 15[½]". Shipping weight: 45 lbs.

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MUSIC INTO LIGHT

By C. M. BRAINARD*

WHILE touring one of our local aircraft plants, I noticed a group of men who did a lot with their time besides replacing spent fluorescent lamps. It seemed to me that a lot of time and money was being wasted. So I set about to develop a method of lighting seemingly dead lamps. I found many different types of devices which would do the trick. Among these were r.f. generators and high-voltage (neon) type transformers. These worked, but it cost approximately 50 times as much as it should to produce enough light to make it worthwhile.

During the process, I discovered that a standard fluorescent lamp, when connected in various ways across the output of an audio amplifier, would fire when certain frequencies were produced in sufficient strength. Further experimentation I found that all fluorescent lamps did not necessarily light on the same frequencies. Therefore, by connecting a large number of them in parallel, I found that a very unusual and almost hypnotizing transformation of sound into light could be achieved. The effect was further improved by the use of various colored lamps. They are available commercially in pink, green, blue, white, and "daylight" colors. They all appear white until lighted. The use of lamps with colored glass did not prove too practical because of the reduction of light intensity caused by the coloring of the glass. The length or number of tubes doesn't seem to make a great deal of difference.

We have made small models for the home with as few as 12 18-inch lamps or for giant commercial jobs using many as 260 4-foot lamps. The arrangement of the lights is up to the builder. We found that mounting them parallel to each other and about one inch apart gives the best results. The tubes are covered with several panels of curved frosted glass mounted about 6 inches in front of the lamps.

Our largest model is located in the auditorium of the North Island Naval Air Station in San Diego, California. It is mounted across the top of the stage, and is 4 feet high and 52 feet long, containing 260 48-inch fluorescent lamps. It has been in use for 7 years, and none of the lamps have ever needed replacement. Since the gas inside these lamps is all that is used and since there are no filaments or ballast transformers to worry about, their life, for all practical purposes, is infinite.

Since the efficiency of many things can be improved by a power source with a frequency considerably higher than 50 or 60 cycles, we feel that power for lighting homes in the future will be in the range of 1 to 10 kc. Of course this would completely eliminate the stroboscopic effect, so annoying with standard fluorescent installations. The intensity of the light will be controlled by the strength of the audio signal. However, for the present, we'll content ourselves with further details on what has come to be known as the "Light Organ".

A few construction hints follow:

Wire all lamps in parallel as shown in Fig. 1. Wrap a fine wire (approximately No. 24) around each of them about ¼ the distance from either end. Ground this wire to any good ground or the amplifier chassis. Connect one end of the parallel group through a 1-megohm potentiometer, to B minus. Connect the other end of the group through a 0.1-µf 600-volt capacitor to the plate of the output tube. If a conventional amplifier is used, load the secondary of the output transformer with a suitable resistance. Increased variation of light patterns may be obtained by reversing the lamps' terminal connections. The intensity of the light will be controlled by the potentiometer. The input to the output tube should be adjusted for the greatest variation in patterns. If too little volume is used the effect will be weak; if too much volume is used, the lights will fire simultaneously and little or no variation will be obtained.

Do not attempt to run the Light Organ and a loudspeaker from the same output transformer because considerable distortion results from the discharge of the lamps. Fig. 2 shows the diagram of a typical amplifier that can be used to drive the organ. Transformer T1 may be an intercom input transformer or a high-impedance plate-to-voice coil output transformer connected in reverse. The low-impedance winding connects across the loudspeaker voice coil. This input connection loads the output transformer, so it should not be used with a high-quality system. Instead, remove T1 and substitute a 470,000-ohm grid resistor and a 0.02-µf coupling capacitor. Input can be tapped off at a suitable high-impedance point in the main amplifier.

The effect of the Light Organ, when used in the home for entertainment, is unusual. The startling potentials of creating light from sound are unlimited and provide a vast field for experimentation.

Lamps are mounted behind curved frosted glass.

Substitute a 470,000-ohm grid resistor and a 0.02-µf coupling capacitor. Input can be tapped off at a suitable high-impedance point in the main amplifier.

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[www.americanradiohistory.com](http://www.americanradiohistory.com)
STUDY COLOR TV

The Associated Radio and Television Servicemen (Chicago) began a series of lectures on color TV in October. Further winter meetings are: Color TV Tubes, Nov. 30; More on Color TV Tubes, Dec. 7; Colorimetry, Jan. 11; and a further series to start at the end of January. Speakers from leading manufacturers have promised their services.

SERVICE TECHNICIAN WEDS MISS EUROPE

A 23-year-old radio-electrician of Amiens, France, Michael Waremberg, has married Miss Europe of 1953, Sylviane Carpentier, 20 years old, according to news reaching us simultaneously from two sources.

The marriage was due not so much to the glamour surrounding the radio-TV repair business as to the fact that they had been childhood friends, according to E. Aisberg, editor of Tout le Radio, the leading radio magazine of France. Instead of movie contracts, Miss Europe is settling down to life in a four-room apartment on a radio-technician's salary of 40,000 francs (about $114) per month.

The event attracted great attention in France, and the magazine Paris-Match (called the Life of Europe) printed the young bride's picture on the cover, as well as giving the story two pages of photos and text.

HEAR KCOR PLANS

The San Antonio Radio and Television Association held its first October meeting at Radio Station KCOR to listen to a presentation of KCOR's plans for a u.h.f. TV station and to discuss the problems of u.h.f. conversion and antenna installation that will be faced in the near future.

The second October meeting was scheduled as a Color School, with a buffet supper served at the meeting place. Conductor of the meeting was
WE ARE SHARING the success of our products WITH YOU as a CUSTOMER'S DIVIDEND in the most SENSATIONAL PRICE SLASHER EVER!

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**#630 SUPER DE LUXE TV CHASSIS**

LICENSED UNDER RCA PATENTS COMPLETE READY TO PLUG IN AND PLAY—Similar in components and features to the TV Kit at half • Manufactured especially for us by nationally known manufacturers • No extra or repair have been learned in construction or material, to make this #630 SUPER DE LUXE TV CHASSIS the best suitable for fringe areas chiefly and all-round performance. Customers report experience better than 600. Each set is factory aligned and pre-wired • All kits carry the latest features • That most volume of business on this CHASSIS (millions of satisfied customers) new tubes are possible for us to reduce the price.

**#630 PARTS IN COMPLETE SETS**

TV WIRE SOLDER KIT, for any SET...$1.25 VIDEO and F.I. KIT, 19 items...4.99 VARIABLE CONTROL KIT, 11 controls...3.97 CABLE RESISTOR KIT, 107 resistors...4.99 WIRE WOUND RESISTOR KIT, 4 resistors...1.76 BRACKET AND SHIELD KIT, 18 items...6.44 ELECTROLYTIC CONDENSER KIT, 6 cond...4.96 TUBULAR CONDENSER KIT, 38 condensers...3.63 CERAMIC CONDENSER KIT, 28 condensers...1.98 MICA CONDENSER KIT, 14 condensers...1.72 TERMINAL STRIP KIT, set of 20...0.69 COMPLETE SET OF TUBES, 29 tubes...24.64

**NATIONALLY KNOWN BRAND PICTURE TUBES**

BRAND NEW in Factory Sealed Cartons—With a Full Year Guarantee

17" CHASSIS...$26.62 21"...$39.21 24"...$59.99 27"...$74.31

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**CHRISTMAS SPECIALS**

All complete with tubes ready to play

- STANDARD RADIOS at almost 1/2 price $13.58
- 5-TUBE PORTABLE, superb in every detail $19.96
- 5 TUBE CLOCK RADIO, latest sensation 16.56
- ROUND THE WORLD PORTABLE, 3 band 43.87

Call showing these to many others on request

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**SELENIUM RECTIFIER**

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- 250ma $0.12
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- 104 ohms...$1.38
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**MICA CONDENSER**

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**SPEAKER**

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**CLOCK RADIO**

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**NEW**

- 24" $26.62
- 30"...$39.21
- 36"...$59.99
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**TELEPHONE**

- Cortland 7-2359

**DECEMBER, 1954**
VERSATILE for Replacement Use --
78-45-33 1/2 RPM

MODEL AX

WEBSTER ELECTRIC
Featheride CRYSTAL CARTRIDGE
--with twist mechanism

This exceptionally lightweight, two-needle crystal cartridge—MODEL AX, one of the "Famous Featheride Five"—is ideal for fast, easy replacement in literally scores of applications. MODEL AX comes furnished with two needles, removable twist mechanism, terminal lugs and complete installation instructions. It fits any standard RCA 1/2-inch mounting in tone arms with or without twist mechanism.

Write today for full information on the MODEL AX. It will give complete customer satisfaction—and will build sales and profits for you.

WEBSTER ELECTRIC
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"Where Quality is a Responsibility and Fair Dealing an Obligation"
WEBSTER ELECTRIC COMPANY, RACINE, WISCONSIN • EST. 1909

TECHNICIANS' NEWS (continued)
Dave Doss, national sales promotion manager of Hoffman.

NATESA CONvenes

The fifth annual convention of the National Alliance of Television and Electronic Service Technicians took place in Chicago September 24, 25 and 26. Approximately 300 delegates and visitors registered during the convention. More than 60 delegates from 41 localities attended the 24 hours of business sessions that took up a large part of the three days. Exhibitors, 27 in number, included the leading manufacturers and publishers of TV receivers, service materials and servicing books and magazines.

Some excellent information was dispensed during the technical sessions. Unfortunately the competition of business sessions and exhibits kept away some listeners. However, more than 100 technicians attended the final lecture, by Bill Ashby of Magnecord, on tape recorders and the service technician. A number of sessions had nearly as great an attendance.

New affiliates officially voted into membership were: Television Bureau of Elkhart, Ind.; King County Radio and TV Service Association, Seattle, Wash.; TV Technicians Association, Joplin, Mo.; Radio and TV Technicians of Boise, Idaho; Tulsa TV Service Association, Okla.; TV Service Guild, Day-

ton, Ohio; and the Radio and TV Technicians Guild of Alabama. Groups in Canada and Venezuela were also welcomed into the organization.

New officers follow: President, Frank J. Moch, Chicago. Vice presidents: (Eastern) Ferdinand Lynn, Buffalo, N. Y.; (East Central) Fred Colton, Columbus, Ohio; (West Central) Vincent Lutz, St. Louis, Mo.; (Western) Horace Collins, Boise, Idaho. Secretaries: (Eastern) Milton Klarfeld, Albany, N. Y.; (East Central) Chas. N. Burns, Memphis, Tenn.; (West Central) William Bixia, Omaha, Neb.; (Western) Jim Failing, Greeley, Colo. Secretary general: Walt Niswonger, Kansas City, Mo. Treasurer, Bertram Lewis, Rochester, N. Y.

Friends of Service Awards were made to Philco for its great efforts on behalf of independent service through the PFFS plan and the Philco stand on transshipping; Sprague Products for its forthright defense of independent
The ONLY Way You Can Get an FP Capacitor is to Specify MALLORY

Why?
Because Only Mallory Makes FP Capacitors

They were introduced by Mallory. For over 15 years they have set the standard of quality and performance in the industry. And they are available only from Mallory.

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When you ask for FP Capacitors... be sure you get Mallory, not an imitation.
TECHNICIANS NEWS (Continued)

service people against unfair charges of profiteering: Albert C. W. Saunders for his continuous tireless and effective work on behalf of the service technician. A continuing award given to Sylvania carried a special commendation for Mr. Walter Biele, field service engineer. Howard W. Sams Co. was cited for its continued support of the independent industry. Special citations were also voted to Mr. Dan Creato, vice president of RCA Service Co., and to Carroll Hoshour and Bill Ashby of Magnecord.

MOVIES AND TV SERVICE

A TV service technician is cast as the hero who saves the life of a U. S. president in the recent United Artists film "Suddenly."

Hoodlums conspiring to assassinate the President have mounted a powerful gun on a mica kitchen table in a house on a hill the President is to pass. The members of the household—including a TV service technician who was repairing the set—are held prisoners by the gangsters. While working on the set, the repairman manages to connect the high-voltage lead to the metal table. Then a glass of water is "accidentally" spilled on the floor near the table, making the floor conductive.

Just as the chief gangster sights his gun—moments before the Presidential train arrives—the TV technician makes his fateful contact. The hoodlum "freezes" onto the gun, setting it off. He is killed in the resulting melee.

The back of the set is prominently labeled "$5,000 volts." This time the movies underestimated!

TV IN DRAGNET

The television show "Dragnet" will present an expose of a television service racketeer, according to a report from the National Appliance and Radio-TV Dealers Association. A "gimmicked" TV receiver will apparently be the program's main character.

CETA PLANS MEETINGS

A series of meetings planned by the Certified Electronic Technicians Association of New York City has been technical lectures. Instead of asking manufacturers to send guest speakers, CETA members go to the manufacturers. A recent meeting was held at Radio Merchandise Sales, Bronx, N. Y., and future meetings and technical forums will be held at Precise Development, Precision Apparatus, CBS-Columbia and Motorola.

CETA, in composed of technicians who have finished the Advanced TV Training Course held under the auspices of the Radio-Electronic-Television Manufacturers Association. Since the pilot courses are being taught in New York City, it is at present a local association, but plans have been made to establish chapters in other localities where the RETMA Advanced TV Training Course is being given.

END
Where better-than-standard components are desirable...

**CLAROSTAT C Line Controls**

A new line of composition-element and wire-wound controls designed specifically for custom-built, industrial, laboratory or other semi-critical applications. These controls are *de luxe versions* — mechanically and electrically — of the popular Clarostat standard types widely used in radio-TV sets and for the servicing thereof.

With factory-assembled fixed 1¼" round shaft (C1) or ½" screwdriver-slot shaft (C2) with split locking bushing.

All exposed metal parts have corrosion-resistant finish. Terminals have suitable form for ease of soldering.

Close-fitting covers — no holes or cut-outs. Maximum protection against dust and dirt.

Ratings: 37C1 and 37C2 composition-element, 1 watt; 43C1 and 43C2 wire-wound, 2 watts; 58C1 and 58C2, 3 watts; 10C1 and 10C2, 4 watts.

Electrical tolerances plus/minus 10% for composition-element controls up to 100,000 ohms; plus/minus 20% from 100,000 ohms to 10 megohms.

Wire-wound controls within plus/minus 5% in all ohmages. Independent linearity to plus/minus 1% for 58C1 and 58C2, and 10C1 and 10C2 controls; for 43C1 and 43C2, plus/minus 2%.

Available with switches on special order.

Units sealed in dustproof plastic bags within standard Clarostat cartons. Factory-fresh appearance and condition, regardless of shelf life.

If you want still better-than-standard quality, performance and life, insist on Clarostat C LINE controls.

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In Canada: Canadian Marconi Co., Ltd., Toronto, Ontario

DECEMBER, 1954
NEW IMPROVED JSC STRIPS WIRE EASILY!

Secret Process Plus Pure Polyethylene Guarantees Quality of 300-Ohm Cable

No more complaints when you use newly improved J.S.C. wire. J.S.C.'s secret process plus the use of only pure Polyethylene make this wire easiest to strip. Compare with any other on the market; you will find none better.

Compare JSC Quality With Any Higher Priced Make


Installers Wild About New JSC 300-Ohm Wire

Installers everywhere, now using newly improved J.S.C. wire, are stripping cable on the market. Their enthusiasm is based on the fact that J.S.C. wire is of only pure Polyethylene. Send for your free sample today.

JSC invites you.

Send For Your Free Sample

Saying it differently, J.S.C. has so much faith in their newly improved 300-ohm cable that they want you to see it with your own eyes. Send a free sample today...then compare.

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CLEAR TV PICTURE PERMANENT INSTALLATION—WITH A WIN POWER TV TOWER

Permanent, one-man tower erection—fewer guys wire—no special tools. Offer better reception at lowest cost—eliminate mounting troubles and service calls.

“All-Angle” feet fit any roof.

5’ or 10’ height.

Adaptor mast or serves as rotor bracket.

Always look for the Orange Blue Disc of Quality.

THE FUND REACHES $11,492.61

HELP-FREDIE-WALK FUND

IT IS heartening to see that little Freddie, armless and legless, son of radio technician Herschel Thomason, Magnolia, Ark., is still remembered by readers of RADIO-ELECTRONICS. From all over the world they continue to send in contributions and messages of encouragement.

It is perhaps even more heartening to know that the $11,492.61 his friends have donated to Freddie's fund is responsible for the great strides he has been able to take. Not literally, of course, and not yet, but soon. For Freddie has been fitted with artificial legs and is learning to walk with the best of them. Writes his father:

"He is still walking fine. He can go over the house, and at a fair pace, too. He usually walks, and then takes a nap and exercises in the afternoon."

Recently we had the pleasure of meeting Freddie when his mother brought him East to the Kessler Institute for Rehabilitation, and we can happily report that he is so cheerful, so full of fun and the devil, that you almost forget that he is not completely like other children. It is a tribute to his parents and friends that Freddie, young as he is, shows every sign of becoming the contributing, well-adjusted adult we all hope he will be.

However, it is obvious that the special treatments, mechanical appliances, etc., that he will need throughout his life will cost thousands and thousands of dollars. Although RADIO-ELECTRONICS' readers have already contributed almost $11,500, we can't stop there. We urge each and every reader to send in his contribution whenever and as often as he can. No amount is too small to merit our sincere thanks and appreciation, and every donation is acknowledged. Make all money orders, checks, etc., payable to Herschel Thomason.

Address all letters to:
Help-Freddie-Walk Fund
C/o RADIO-ELECTRONICS Magazine
25 West Broadway
New York 7, N.Y.

FAMILY CIRCLE Contributions $602.50
RADIO-ELECTRONICS Contributions as of July 2, 1954 $10,942.61
Anonymous, San Francisco, Calif. 6.00
Anonymous, Hartford, Conn. 1.00
Anonymous, New York, N. Y. 1.50
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Roy Sunday, Camden, N. J. 1.00
Class B-2-14 United Radio Tel.Inst. 7.00
Newark N. J. 5.00

TOTAL CONTRIBUTIONS as of October 15, 1954 $11,492.61

MISSCELLANY
HANDY AMPLIFIER KINK

On a recent PA job I used an amplifier designed for high-output crystal phonographs and crystal microphones. In an emergency I had to use a low-output magnetic phono pickup. The amplifier did not have enough volume, even with the gain control set at maximum. No spare parts or tools were available.

The amplifier uses a 6SN7-GT amplifier and split-load phase inverter, a 6N7 push-pull driver, and a 6AS7-G push-pull output stage as shown in the diagram. Using a pair of borrowed scissors, I clipped the 10-µf capacitor lead running to the cathode of V2, stripped the insulation off one end, and then twisted the lead around the cathode pin of V1-b.

This converted the phase inverter (V1-b) into a straight amplifier driving the grid of V2-a. The grid of V2-b is now effectively grounded by the 10-µf capacitor at the cathode of V1-b while the cathode of V2 is now un-bypassed. Thus, V2 is converted into a cathode-coupled phase inverter with only half the gain of the original push-pull amplifier circuit. On the other hand, the voltage gain of V1-b as an amplifier is about 20 times that of the original phase-inverter connection. Thus, the over-all gain from the plate of V1-a to the grids of V3 is now about 10 times (20 db) greater than previously. This added 20-db gain was enough to do the job. Naturally, I do not recommend this modification except in emergencies where everything else looks hopeless.—Nathan Sokal

B.F.O. MODIFICATION

Many amateur operators fail to tune their receivers properly for CW signals. They usually tune in a signal and adjust its pitch with the tuning control, using the pitch control only to vary the pitch when fighting QRM. This method of tuning in a CW signal often results in the r.f. and i.f. stages being several kc off resonance.

In Short Wave Magazine, G3DXI describes how he modified the b.f.o. in an AR 77 to assure proper CW tuning. His modification, shown in the diagram, can be applied to almost any communications receiver regardless of b.f.o. circuitry.

The original b.f.o. tuning capacitor was removed and replaced with a 3-position rotary switch and a pair of 3-25-µf trimmers. For CW reception he feeds the output of the receiver through a tuned a.f. amplifier peaked at 1,000 cycles so his switch tunes the b.f.o. to frequencies 1 kc higher and lower than the receiver's i.f.

To align the b.f.o., set the switch to HIGH and feed into the receiver's mixer grid the unmodulated output of a signal generator tuned exactly 1,000 cycles

DECEMBER, 1954
**RADIO-ELECTRONICS**

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**RADIO-ELECTRONIC CIRCUITS (continued)**

(1 kc) higher than the i.f. Adjust the h.f.o. tuning slug (or padder capacitor) for zero beat.

Tune the signal generator to the i.f., set the switch to ZERO, and adjust C2 for zero beat. Throw the switch to LOW, tune the signal generator 1 kc before the i.f., and adjust C1 for zero beat.

To operate, set the b.f.o. switch to ZERO and adjust the tuning controls to zero-beat the desired signal. This assures that the r.f. and i.f. circuits are peaked on the incoming signal. Now, flip the switch to LOW or HIGH, and you are ready to copy. In QRM, you can simply flip the switch to the other side rather than tune the variable control and take the chance of losing the station. Of course, if you want to keep the variable pitch control for optional use, you can use a 4-position switch and connect the original variable pitch control as shown.

**SHORT-PULSE GENERATOR**

Multivibrators, sine-wave clippers, and differentiators are generally used to produce sharp unidirectional pulses for timing, triggering, and gating applications. The circuits are usually complex and component values are critical. In Electronic Engineering (London, England) F. A. Benson and G. V. G. Lawrence review the use of a simple shock-excited LCR circuit as a source of sharp unidirectional pulses.

The oscillator circuit consists of inductor L tuned by stray and distributed capacitance C. When a positive-going pulse is applied to the input, the L-C circuit is shocked into a simple chain of damped oscillations. Resistor R damps out the oscillations after two or three cycles. The diode clips off most of the unwanted negative portion.

With a 6,500-cycle, 20-volt peak positive pulse applied to the input and a damping resistor of 3,000 ohms, a 400-µh coil produces an output signal with a rise time of 0.8 µsec and amplitude of 1.6 volts. With 180- and 33-µh coils, rise times are 0.6 and 0.5 µsec and output pulse amplitudes 1.35 and 0.3 volts, respectively. With a 250-µh coil and 2,000-ohm damping resistor, a 6,500-cycle, 60-volt input pulse produces a 10-volt output pulse with a rise of 0.7 µsec and a width of 2 µsec.

For negative output pulses, reverse the diode polarity and use negative-going input pulses. When a negative pulse of greater amplitude is needed, replace the diode with the grid-cathode circuit of a triode and take the output from the plate.
Giving coils a high IQ...

Delco Radio carefully controls the production of original and replacement RF and IF coils.

*Inbuilt quality. The "Q" of a coil affects its performance in the circuit, and is of vital importance to the radio engineer. The "IQ" of all RF and IF coils used for servicing is of equal importance to you, and can mean real customer satisfaction. To assure high IQ, Delco Radio carefully controls the production of all RF and IF coils—both original equipment and replacement coils.

If you could visit the Delco Radio plant, you would see skilled operators sitting at expensive specially designed machines winding high-quality coils. On the production line, these coils are assembled with their powdered iron cores and shield cans. By controlling all steps of the production, Delco Radio gives its coils not only the required Q, but also the highest possible IQ—Inbuilt Quality. Delco Radio parts are available through your UMS Delco Electronics Distributor.
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FIX ANY RADIO or TV SET EVER MADE...easier...better...faster

Racked by the how-to-do-it methods so clearly explained in this one big 822-page book, you'll be prepared for fast, accurate service on any radio or television receiver ever made.

Best of all, the cost is only $8.76 (or see money-saving combination mail-in coupon).

Radio & Television TROUBLE-SHOOTING AND REPAIR by Ghirardi & Johnson is fair and away the world's most modern, easily understood guide. Step by step, it takes you through each service procedure from locating troubles with less testing to repairing them by professional methods...the kind that let you handle tough jobs as slick as you do the easy ones.

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For beginners, this giant book is your complete service training course. For experienced service men, it is a quick, easy way to "brush up" on specific jobs; develop better methods and short-cuts; and to find fast answers to tough jobs.

Here are just a few of the subjects covered: Components and Their Troubles; Basic Troubleshooting Methods..."Static" and "Dynamic" Testing; Practical Troubleshooting Tips and Ideas; AC/DC, 3-way Portable and Battery Set Troubleshooting Problems; Servicing Communications Receivers; Complete Trouble-Shooting and Repair Equipment;...and dozens more! Illustrated Read TROUBLE-SHOOTING AND REPAIR for 10 days AT OUR ISK!

It's amazing how much easier and faster you can repair radio, television sets and any industrial electronic equipment when you know all about circuits. You learn how to tackle each one "tick!" You locate troubles in a jiffy because you know what to look for and where to look.

You make repairs lots faster, better, and more profitably. Actually, there are only a comparatively few BASIC circuits in modern equipment. Radio & Television RECEIVER CIRCUITY AND OPERA-

TION by Ghirardi and Johnson gives you a complete understanding of these as well as their variations. It teaches you to recognize them...to understand their peculiarities and likely "troubleshooters"...and shows how to eliminate useless testing and guesswork in making repairs.

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Throughout, this 669-page book with its 417 clear illustrations gives you the all-important "above-average" kind of training that fits you for the better, higher paying jobs in electronic, radio, and television service...or general electronics. Covers all circuits in modern TV and radio receivers, amplifiers, phon-o-pickups, record players, etc. Price only $8.50...or see money-saving offer in coupon. Examine it 10 days at our risk!

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$7.25 for TROUBLE-SHOOTING & REPAIR: $7.00 for CIRCUITRY & OPERATION: $13.00 for both books; Order with order, but money refunded if return looks to 10 days.

INFRARED GENERATOR

Patent No. 2,683,794

Howard B. Briggs and James R. Haynes, Chatham, and William Shockley, Madison, N. J. (assigned to Bell Telephone Labs., Inc.)

The operation of a transistor is known to be affected by heat and light. Now it has been determined that semiconductors can radiate energy when large currents flow through them. This invention shows how a p-n junction transistor generates infrared energy.

In Fig. 1, C charges through R when the key is left up. When the key is depressed, C discharges through the resistive charges are injected into the n region of the semiconductor, preferably at a rate exceeding 10

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amps per sq. cm. These holes recombine with excess electrons in the n region, giving a pulse of infra-red rays.

Fig. 1 also shows how the radiation is detected. A lead sulphide photocell (PC) is energized by a battery. When radiation strikes the cell, electricity is generated. After amplification, it is fed to an oscilloscope with a synchronized sweep. The visible pattern indicates energy amplitude vs. wavelength.

The wavelength of the energy varies with junction temperature. (See Fig. 2.) Curve A shows energy distribution at room temperature (295° K). Curves B and C are drawn at the temperatures of liquid nitrogen (77.4° K) and liquid hydrogen (21.5° K), respectively. Note that wavelength drops as temperature decreases. Bandwidth also becomes narrower. Curve D is for room temperature but with a silicon junction instead of germanium.

Because infra-red rays easily pierce fog and haze, these pulses of energy are useful for short-distance signaling and communication.

The primary object of this invention is the creation of a readily controllable source of infra-red energy. A more specific object is to enable infra-red energy to be generated in narrow frequency bandwidths. This makes it particularly suitable as a radio-frequency carrier.

And today he's experimenting with high-powered audio.
Superior's new Model 670-A

SUPER METER

A COMBINATION Volt-Ohm Milliammeter PLUS CAPACITY REACTANCE INDUCTANCE AND DECIBEL MEASUREMENTS

SPECIFICATIONS:

P.C. Volts: 0 to 7.5/15/75/150/750/1,500/7,500 Volts
A.C. Volts: 0 to 15/30/150/300/1,500/3,000 Volts
Output Volts: 0 to 15/30/150/300/1,500/3,000 Volts
D.C. Current: 0 to 1.5/150 Ma. 0 to 1.5/15 Amperes
Resistance: 0 to 1,000/100,000 Ohms 0 to 10 Megohms
Capacity: 0 to 1 Mfd. 1 to 50 Mfd. (Good-Bad scale for checking quality of electrolytic condensers.)
Reactance: 50 to 1,500 Ohms 2,500 Ohms to 2.5 Megohms
Inductance: .15 to 7 Henries 7 to 70,000 Henries
Decibels: -5 to +18 +14 to +38 +34 to +58

ADDED FEATURE:

Built-in ISOLATION TRANSFORMER reduces possibility of burning out meter through misuse.

The Model 670-A comes housed, in a rugged, crackle-finished steel cabinet complete with test leads and operating instructions.

$2840 NET

Superior's new Model TV-11

TUBE TESTER

SPECIFICATIONS:

- Tests all tubes including 4, 6, 6, 7, Octal, Lock-in, Peanut, Bantam, Hearing Aid, Thyratron Miniatures, Sub-miniatures, Enter, Sub-miniatures, Proximity fuse types, etc.
- Uses the new self-clearing Lever Action Switches for individual element testing. Because all elements are numbered according to pin-number in the R.A. base numbering system, the user can instantly identify which element is under test.
- Tubes having lapped-filled filaments and tubes with filaments terminating in more than one pin are truly tested with the Model TV-11 as any of the pins may be placed in the neutral position when necessary.
- The Model TV-11 does not use any combination type sockets. Instead individual sockets are used for each type of tube. Thus it is impossible to damage a tube by inserting it in the wrong socket.
- Free-moving built-in roll chart provides complete data for all tubes.
- Newly designed Line Voltage Control compensates for variation of any Line Voltage between 105 Volts and 130 Volts.
- NOISE TEST: Phone-jack on front panel for plugging in either phones or external amplifiers will detect microphonic tubes or noise due to loose and internal connections.

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The Model TV-40 is absolutely complete and beneficial to one or more person who wishes to use their tubes.

Tests all magnetically deflected tubes . . . in the set . . . out of the set . . . in the carton!!

SPECIFICATIONS:

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The Model TV-50 GENOMETER

A versatile all-inclusive GENERATOR which provides ALL the outputs for servicing:

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<th>Amplifiers</th>
<th>Black and White TV</th>
<th>Color TV</th>
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7 Signal Generators in One!

- R. F. Signal Generator for A.M.
- R. F. Signal Generator for F.M.
- Audio Frequency Generator
- Bar Generator
- Cross Hatch Generator
- Color Dot Pattern Generator
- Marker Generator

**SPECIFICATIONS:**

**R. F. SIGNAL GENERATOR:**
The Model TV-50 Genometer provides complete coverage for A.M. and F.M. alignment. Generates Radio Frequencies from 100 Kilocycles to 60 Megacycles on fundamentals and from 60 Megacycles to 180 Megacycles on powerful harmonics. Accuracy and stability are assured by use of permeability trimmed Hi-Q coils. R.F. is available separately, modulated by the fixed 400 cycle sine-wave audio or modulated by the variable 300 cycle to 20,000 cycle variable audio. Provision has also been made for injection of any external modulating source.

**VARIABLE AUDIO FREQUENCY GENERATOR:**
In addition to a fixed 400 cycle sine-wave audio, the Model TV-50 Genometer provides a variable 300 cycle to 20,000 cycle peaked wave audio signal. This service is used for checking distortion in amplifiers, measuring amplifier gain, trouble shooting hearing aids, etc.

**BAR GENERATOR:**
This feature of the Model TV-50 Genometer will permit you to throw an actual Bar Pattern on any TV Receiver Screen. Pattern will consist of 4 to 16 horizontal bars or 7 to 40 vertical bars. A Bar Generator is acknowledged to provide the quickest and most efficient way of adjusting TV linearity controls. The Model TV-50 employs a recently improved Bar Generator circuit which assures stable non-shifting vertical and horizontal bars.

**CROSS HATCH GENERATOR:**
The Model TV-50 Genometer will project a cross-hatch pattern on any TV picture tube. The pattern will consist of non-shifting, horizontal and vertical lines interlaced to provide a stable cross-hatch effect. This service is used primarily for correct ion trap positioning and for adjustment of linearity.

**DOT PATTERN GENERATOR** (For Color TV)
Although you will be able to use most of your regular standard equipment for servicing Color TV, the one addition which is a "must" is a Dot Pattern Generator. The Dot Pattern projected on any color TV Receiver tube by the Model TV-50 will enable you to adjust for proper color convergence. When all controls and circuits are in proper alignment, the resulting pattern will consist of a sharp white dot pattern on a black background. One or more circuit or control deviations will result in a dot pattern out of convergence, with the blue, red and green dots in overlapping dot patterns.

**MARKER GENERATOR:**
The Model TV-50 includes all the most frequently needed marker points. Because of the ever-changing and ever-increasing number of such points required, we decided against using crystal holders. We instead adjust each marker point against precise laboratory standards. The following markers are provided: 180 Kc. 252.5 Kc. 450 Kc. 600 Kc. 1000 Kc. 1400 Kc. 1600 Kc. 2000 Kc. 2500 Kc. 3579 Kc. 4.5 Mc. 5 Mc. 10.7 Mc. (3579 Kc. is the color burst frequency.)

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CROSLEY 115 CHASSIS

To improve stability at the upper end of the shortwave band, a 15,000-ohm, 1/2-watt resistor has been added across the shortwave r.f. coil of late production sets, as shown in the diagram. Chassis containing this change are marked with code letter B (or higher). This resistor is not found in code A (early production) chassis.—Crosley Service Department

GASSY TUBES

When confronted with the complaint that the set "uses too much battery" or "doesn't pick up stations on the low-frequency side of the dial," check the oscillator tube carefully. Failure of the oscillator in battery sets and portable is often caused by a gassy oscillator-mixer tube. The most effective check is to replace with a new one—a gassy tube is not readily detected on the average tube tester.

Occasionally this defect can be detected by visual examination of the tube filament while under test. A flaring up of the filament glow together with the appearance of a bluish fluorescence around it is a good indication of gas. This is common in 1A7-GT and 1R5 types.—V. F. Woychowski

MUNTZ TV-16A1

A customer brought in a Muntz TV-16A1 TV chassis with a burned-out a.c. line choke used with the built-in antenna. (See a in the illustration.) A replacement choke not being avail-

RECORD PLAYERS

Many children's record players use 117-volt rectifier-pentode tubes such as the 117L7-GT in the amplifier circuit. Failure of the input filter capacitor usually burns off the cathode connection and makes it necessary to replace the tube. The high list price of the tube plus the cost of replacing the capacitor discourage some parents from having the player repaired.

When the pentode section of the tube checks O.K., you can greatly reduce the cost of the repair job by using the original tube and substituting a selenium rectifier for the defective diode section. Replace the capacitor, then connect a 22-47-ohm, 1-watt resistor and a 75-100-ma selenium rectifier between rectifier plate and cathode pins as shown. Be sure to paste a circuit-change label on the chassis.—Albert L. Sohl

JEWEL MODEL 920

Greater sensitivity can be had in this model by taking the resistor out of the cathode circuit of the i.f. amplifier and grounding the cathode directly to the chassis.—Ross Harris
FM AND TV AUDIO HISS

A high hiss level may be heard in the output of some TV and FM sets with ratio detectors. This is usually caused by insufficient AM rejection. When the usual corrective measures do not work, try placing a variable resistor of about 150 ohms across resistor R1 in series with the tertiary winding on the ratio-detector transformer as in the illustration. Mount the control on the chassis or cabinet with short leads and set it for minimum hiss and maximum AM rejection.

The usual value of R1 is between 200 and 400 ohms. Smoother balance-control action can be obtained by changing R1 to between 100 and 150 ohms.—Lyle Briggs

(This method of balancing ratio detectors reduces the circuit sensitivity, so take care when the set is installed in a fringe area or is used for dx'ing. —Editor)

G-E 16C113 DIAL CORD

I ran into a difficult repair situation recently when I had to replace the dial cord (fine tuning) on this receiver. I simplified the job by installing a General Cement phono-drive rubber belt. It made a friction drive tight enough to prevent slipping, and there is no danger of ever breaking the dial cord again.

There are numerous other applications where a phone-drive rubber belt can be used to replace a broken dial cord, and this procedure is becoming more and more popular among service technicians. It also eliminates duplicating complicated restringing jobs.—Hyman Herman

ZENITH MODEL 9S263

An open speaker voice coil is a common complaint on this model. Before replacing the speaker, remove the two ⅝-inch paper tabs near the apex of the cone and you will probably find the open circuit. You can save the speaker by resoldering the connection and then recementing the tabs.—Harold J. Weber

CAPEHART CX-33

Insufficient width in these chassis gave us lots of trouble because tubes and all components checked O.K. While studying the circuit of a later model I noticed an additional 30-µuf, 6,000-volt capacitor connected between ground and the plate of the 6W1-GT damper tube. Adding this capacitor to earlier models clears up complaints of insufficient width.

There are other ways of increasing width such as varying output-tube screen voltage, but they usually vary the high-voltage output.—Carl Hennig

RINGING COILS

Ringing coils, found in the multivibrator circuits of horizontal sweep systems, may be tested without pulling the chassis. Merely tune in a station and rotate the horizontal hold control toward the low-frequency (greater-resistance) side of synchronization. If the ringing circuit is working properly, the picture will slip into a number of stationary patterns as the control is rotated.

This pattern will slip abruptly into the following pattern, remaining for a short time in each position. However, if the ringing circuit is not working properly, there will be no stable patterns. Each pattern will appear briefly and intermittently. Of course, this technique will not work if the hold control is part of the ringing circuit itself, as is the case in some sets.—Charles Erwin Cohn

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NOVEL TV ANTENNA
This simple TV antenna costs not more than 50 cents and can be constructed in about 20 minutes. It is strictly an indoor affair particularly suited for attic installation, but, it will often outperform roof-top antennas costing many times more. I call it the Supermarket antenna because its bill of material consists of a roll of aluminum foil and a length of stout twine.

Refer to the diagram and proceed as follows: Select two points—suitable for attaching strong twine—of equal height on a line at right angles to the direction of the TV station. Tie a 14-foot piece of twine between the two points. Tie a second piece of twine to the center of the first piece, pull it taut in the direction opposite the TV station and tie it to a third point at the same height as the other two. The results should be an arrow-shaped framework pointing away from the station. Next, cut two 6-foot lengths from a 12-inch roll of aluminum foil. Fold one piece lengthwise over each leg of the V and slide them to within 1 inch of each other at the point. You now have a nonresonant 90° V antenna with each leg 6 inches high and 6 feet long.

Use 300-ohm ribbon for the lead-in. Connect a conductor to each sheet at the vertex. Because the foil tears easily and is not readily soldered, make the connection with paper clips.

The antenna has relatively high gain and sharp directivity so some orientation may be necessary in suburban areas. Handling and orientation will be easier if you use two lightweight sticks for the supporting framework. Performance can be improved for weak-signal areas by lengthening the legs

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**CLEANING TUBE SOCKETS**
Tub... (continues)

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**TV ANTENNA HINT**
Some of the better TV antennas are made of anodized aluminum to resist corrosion. Anodizing coats the aluminum with a “skin” that has a high r.f. resistance. Use emery cloth to remove it at all points of electrical contact to get maximum transfer of energy from the antenna to the lead-in. — Ger-ald Machak

**HANDY CHASSIS SUPPORT**
When test instruments such as v.t.v.m.’s and oscillators are removed from their metal cases, it is often difficult to balance the delicate exposed chassis on the bench for convenient servicing. Those instruments with one or more screws securing the back of the case to the chassis rear skirt can

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**TRY THIS ONE**
(continued)

of the V and slightly reducing the angle at the vertex of the antenna. — Harry W. Lawson

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TRY THIS ONE (Continued)

MEASURING METER RESISTANCE

Many technicians and experimenters have surplus microammeters and milliammeters that they would build into multimeters if the internal resistances of the meters were known so they could calculate shunt resistances.

Internal resistance cannot be measured with an ohmmeter without damaging the meter movement.

Here is a quick and simple method of finding the internal resistance of a meter. Suppose that the instrument to be measured has a full-scale deflection of 50 µA. Connect the meter in series with a 2-megohm potentiometer across a 45-volt battery. Make sure

that the resistance is fully in the circuit before making the connection to the battery. Carefully reduce the resistance until the meter deflects to full scale (50 µA). Next shunt a 1,000-ohm variable resistor across the meter as shown in the diagram and adjust its value until the meter reads exactly half scale. Remove the 1,000-ohm resistor from the circuit and measure its resistance with an accurate ohmmeter or a resistance bridge. Its measured resistance will be equal to the internal resistance of the meter. For a 200-µA movement, a 200,000-ohm resistor may be used in place of the 2-megohm unit.

Other values of series resistance and voltage may be used if the resistor is large enough to limit the current to a value less than the full-scale deflection with the applied battery voltage.—P. E. Clement

INSTRUMENT COVERS

Quite a lot of dust accumulates on test instruments that are not used often. While sitting in the kitchen a few weeks ago I noticed the neat plastic covers on the toaster, mixer and other kitchen accessories. After a talk with the Little Woman I went out and purchased a few yards of transparent plastic and some colored piping. My wife used this material to make instrument covers that she stitched together on her sewing machine. These covers now protect my equipment from dust and moisture.—Alfred Zanelli

BENDING LIGHT METALS

Wide-jawed canvas or webbing pliers, available from dealers in artists' and upholsterers' supplies, are handy tools for making neat bends in sheet metal used for shields, partitions, brackets, and chassis.—Van H. Ferguson END
S-METER FOR S-40B

I want to add an S-meter to a Halli- 
crafters S-40B receiver. Can I use a
1-ma meter in a bridge circuit to mea-
sure plate current of the i.f. tubes on
the a.v.c. line—F. H. B., Brooklyn,
N. Y.

Several circuits in which a meter
indicates signal strength by changes
in i.f. plate current will be found in
the 1853 and earlier editions of The
Radio Amateur's Handbook. These
circuits are sometimes difficult to adjust
and stabilize. You are less likely to
have trouble with a v.t.v.m. type of
indicator like that shown in the diagram.
The grid of the 6C4 S-meter (v.t.v.m.)
tube connects to the a.v.c. line in
the receiver. Probably the most convenient
spot for this connection is at the ter-
minal of the movable arm of the a.v.c.
switch. However, other points on the
a.v.c. line can be used.

To adjust the meter, short-circuit the
antenna posts and set the variable
resistor for full-scale deflection. The
needle should fall to zero on a strong
local signal. If it does not, try different
values in place of the 82,000-ohm
resistor.

TRAPS FOR BCI

I live about a half mile from the
broadcasting towers of KMPC (50,000
watts, 710 kc). This station is the only
one that I can receive on my table model
radios. It comes in all over the dial.

On two large console models I can tune
in other stations, but KMPC and a loud
whine can always be heard in the back-
ground. Is there any way that I can com-
eliminate this trouble—T. P., Van
Nyas, Calif.

BCI (broadcast interference) of this
type is usually fairly easy to eliminate
with tuned traps. The diagram at a shows
the connections for a series-tuned
trap in the grid circuit of the r.f.
amplifier or mixer in a receiver using
a loop antenna. When the set uses an
outside antenna, you can use either a
series-tuned trap across the antenna
and ground terminals or a parallel-
tuned trap in series with the lead-in
close to the set’s antenna post. In some
cases, you may have to use both types
of traps for complete elimination of
interference (b). Shielded traps are
more effective than unshielded types.

On a.c.-d.c. sets ground the shield to
the chassis. On other types, connect
the shield and chassis to a good outside
ground.

Tuned circuits L-C in the traps should
resonate at the frequency of the inter-
fering station. You can use the sec-
condary of a small broadcast antenna
or r.f. coil and a mica trimmer with a
maximum capacitance around 380 µf
and a minimum of not more than about
50 µf. J. W. Miller Co. has a series
of series-tuned traps designed especially
for receivers using tuned loops. The
type 816-BC1 tunes from 500 to 900 kc
and the 816-BC1 from 900 to 1,800 kc.
Others in the series are available for
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Note well that you will not be able
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Sensitivity as long as the traps are in the circuit. Series-tuned traps can be disabled by a switch that disconnects one end, and parallel-tuned traps by a switch that shorts the tuning capacitor C.

Traps are not effective when the interference is due to direct pickup in the audio circuit of the set. This usually occurs in the first audio-amplifier stage. Most sets subject to this type of interference have high-mu triodes with grid-leak bias. In this case, the trouble can be eliminated by shielding the tube or replacing it with its metal equivalent. If this does not help, try reducing the grid resistor—often as large as 10 megohms—to not more than about 2 megohms and bypass it with a mica capacitor of about 200 µµf as in the drawing at c.

SENSITIVE R.F. INDICATOR

I am often called on to construct small transmitters, boosters, converters, and similar devices. I frequently have trouble that I believe to be due to spurious oscillations and incomplete neutralization but I have no way of checking it. How can I detect oscillations in an r.f. stage? Is there any indicator that I can use?—F. R., Memphis, Tenn.

The simplest way to detect oscillation is to measure the grid current or check for the presence of r.f. voltage on the grid with a v.t.v.m. Grid current can be checked with a microammeter shunted by a capacitor of 50–100 µµf. The bypass capacitor should be connected directly across the meter terminals. If the meter leads are long, shunt another capacitor directly across the points where they connect into the circuit. In some cases it may be necessary to use a meter with a sensitivity of 100 µa or less. Sometimes inserting the meter changes the circuit configuration so it won't oscillate or the strength of oscillation varies.

A simple r.f. detector probe like that shown in the diagram is useful in detecting weak r.f. oscillations in some circuits. The probe circuit is untuned so it will be easier to use but less sensitive than a wavemeter. The pickup coil may be three or four turns of self-supporting wire with an inside diameter of about 1 inch. Smaller diameter coils may be wound to meet the needs of the application. A wavemeter can be used to identify the frequency of the spurious oscillations.

Wavemeters and grid-dip oscillators working in the u.h.f. range are useful in detecting spurious oscillations well above the working range of the equipment under test. Such oscillations are often caused by inefficient bypassing, excessive lead lengths and poor choice in capacitor types.

TELEPHONE RECORDING

Can you tell me how to connect a tape recorder to a telephone line so I can record conversations. I would also like to have construction details on an automatic answering device using either tape or disc recorders. This will enable me to take incoming calls while I am out of the shop.—E. O. C., Nanarongie, N. Y.

Any device, circuit, apparatus or equipment not supplied by a telephone company but connected electrically, mechanically or by induction to its lines, equipment or facilities is a "foreign attachment." The "foreign attachment" clauses of FCC regulations and the Communications Act of 1934 make it illegal for any person to use or permit the use of any attachment, accessory or other device in connection with the services or facilities of a telephone company without the company's written permission. The FCC regulations also state that no direct recording of a telephone conversation shall be made unless a periodic "beep" is transmitted to inform the party at the other end that his remarks are being recorded. The "beep" is an audio tone of about 1,400 cycles transmitted for about 1/5 second at intervals of 12 to 18 seconds.

Since telephone companies are bound by regulations to provide certain services in a satisfactory manner, they retain the right to discontinue service or to remove any foreign attachment used without their approval.

Although a phone company is not likely to permit direct connection of a foreign attachment to its facilities, it will supply a "recorder connector"—a device for coupling your recording device to the line and for generating the beep of the required frequency, duration and interval. The recorder connector also equalizes the levels of speech fed into the recorder and feeds the beep into the phone circuit so it is about level with an average phone conversation. In many areas the recorder connector can be rented for about $2 per month, after an installation charge of around $5.

The regulations against the use of foreign attachments also apply to telephone answering devices not supplied by the company. You can probably obtain a type 1A telephone answering set from your phone company. The 1A is about the size of a portable typewriter. It connects to the line so the subscriber can have normal phone operation when he desires. When he is away, the instrument—automatically started by an incoming call—gives the caller a pre-recorded message, records a 30-second message and then disconnects to await the next call. Twenty-30-second incoming calls can be recorded. The device gives a "don't answer" signal if it is improperly adjusted or its recording time (10 minutes) is used up.

The outgoing announcement and the incoming messages are recorded on magnetic drums and can be erased at will.

END
Merchandising and Promotion
Astron Corp., East Newark, N.J., is promoting its Blue Point capacitors by offering $64.70 worth of the capacitors plus a free "Swing Bin" plastic and metal container for $28.96. The company is making the offer for a limited time only.

CBS-Hytron, Danvers, Mass., introduced a series of sales aids for service technicians, including two illuminated signs—a weatherproof outdoor sign and a jumbo illuminated clock sign. Other items include two-color stationery, which may be imprinted with the technician's or distributor's name, and a Certified Quality Service Repair sticker which may be affixed to sets serviced. The company also brought out indoor and outdoor illuminated signs for distributors.

JFD Manufacturing Co., Brooklyn, N.Y., is packaging its Roto-King rotator in a new "Valet" pack which serves both as a handy suitcase type package for carrying the rotator and as an attractive window or counter display. JFD also announced that it is

December 1954
BUSINESS

planning the biggest advertising campaign in its history for its antennas and accessories. The company is also mailing "ad paks"—including a window streamer featuring a picture of Imogene Coca—to dealers and service technicians across the country.

Electronic Measurement Corp., New York, is now packaging its test equipment in a display box with picture of the actual instrument on the cover. Copy on the package highlights the instrument's features.

Erie Resistor Corp., Distributor Division, Erie, Pa., is now packaging its disc ceramicics in a cardboard "Pallet-Pak."

National Company, Malden, Mass., developed a new hi-fi sales promotion kit for distributors. It includes current advertising, catalogs, advertising mats, window streamers, etc.

Jensen Industries, Forest Park, Ill., introduced a new needle-threader gimmick to help service technicians promote the sale of its phonograph needles to their housewife customers. The company has also developed a new white leatherette display kit which holds 64 different types of needles.

Alliance Manufacturing Company, Alliance, Ohio, has budgeted better than $500,000 for promotion of its Tenna-Rotor and other TV accessories, according to John Bentia, executive vice president.

Channel Master Corp., Ellenville, N. Y., was featured in a story in the October issue of Coronet Magazine. The company has built a promotional campaign around the story.

Simpson Electric Company, Chicago, announced that its chief field service engineer, Bob Middleton, had resumed his color and black-and-white TV service demonstrations. The most recent series was held throughout the Pacific Northwest.

Allen B. Du Mont Laboratories' Public Relations Department, Clifton, N. J., has produced a documentary motion picture, "A Story of Television," available to Du Mont distributors, dealers and service organizations.

Walsco, Los Angeles, Calif., introduced a new merchandising display for its "99" line of hardware and chemicals.

American Screen Products Company, Miami, Fla., is promoting its Unicorn TV antenna by field demonstrations in a ranch wagon on which the antenna is mounted.

The Winegard Co., and its advertising agency, Henry M. Hempstead Co., advise us they did not intentionally mean to illustrate in their advertisement in the November issue of RADIOTELE-ELECTRONICS any competitive antenna, and in particular an antenna that looks like the one manufactured by the JFD Mfg. Co., Inc. The advertisement, they state, merely attempted to illustrate any stacked antenna versus any...
unstacked antenna. Steps have been taken to correct the illustration in all future advertising.

H. G. Cisin, Amagansett, N. Y., publisher, announced the 50 winners in the recent contest for selecting the title of his forthcoming television service book. Ralph L. White, Springfield, Mass., won $100 in cash as first prize.

Finney Co., Cleveland, held a series of TV antenna clinics throughout Texas and the Midwest, co-sponsored by local distributors. The company also demonstrated its products at the Ohio State Fair in Columbus.

Production and Sales
RETMA reported the production of 3,785,619 TV receivers and 6,116,119 radios during the first eight months of 1954. This compares with 4,754,285 TV sets and 8,932,638 radios during the same period in 1953.

RETMA reported the record retail sale of 3,174,394 TV sets during the first seven months of 1953 as against 3,116,306 in the 1953 period. Retail sales of radios excluding auto sets reached 2,222,000 for the first seven months of 1954 as against 3,383,862 in 1953.

Calendar of Events
Symposium on Printed Circuits sponsored by RETMA, Engineering Department, January 20-21, University of Pennsylvania, Philadelphia.

New Plants and Expansions
Rohn Manufacturing Co., Peoria, Ill., has considerably enlarged its facilities for the manufacture of TV towers, masts and accessories, through the acquisition of a new plant which will provide an additional 20,000 square feet of factory and office space to the two plants it already owns. The company's general offices have been relocated in the new plant at 116 Farmington Road, Peoria, Ill.

CBS-Columbia opened a new headquarters building at 48-50 34th St., Long Island City, N.Y., which will contain a complete color TV service training school.

Permoflux, Chicago, officially opened a new plant in Bradford, Tenn.

Simpson Electric Co., Chicago, purchased the complete plant facilities of O. D. Jennings & Co., at 4307 W. Lake St., Chicago. The new plant will be used for production of new color TV test equipment and the company's new suspension type meter.

Clear Beam Antenna Corp. has combined operations with Tempo TV Sales Corp. and is now located in Canoga Park, Calif. Clear Beam also opened a new plant in Chicago.
Brig. Gen. James S. Willis, U. S. Army (ret.), joined Hallicrafters, Chicago, as coordinator of research and development. He was most recently commanding general of the Signal Corps Supply Agency in Philadelphia.

Nathaniel B. Nichols, manager of the Raytheon Manufacturing Co.'s research division, Waltham, Mass., was named assistant vice president. He is an international authority on servomechanisms and electronics automation. Raytheon also announced that Dr. K. C. Black, formerly business manager of Polytechnic Research and Development Co., had joined the company as head of the Communications Engineering Department.

Eugene E. Broker was named manager of Sylvania Electric Products, Shawnee, Okla., Tube Plant, succeeding Charles W. Hosterman, who was recently appointed assistant general manager of the Electronics Division. Broker was formerly manufacturing superintendent of the Sylvania Radio Tube Plant in Burlington, Iowa.

Jack Moore joined Skyline Manufacturing Co., Cleveland, Ohio, as sales manager. He has long been active in West Coast electronics sales.

Bruce E. Vinke, mulder joined Carter Parts Co., Chicago, as sales manager of the Electronics Division. He was formerly a sales executive with the Capacitor Division of Sangamo Electric. The Carter Co., organized by Nick Carter in 1922, was formerly a division of Utah Radio Products. The company is presently expanding its nonautomotive line of potentiometers, plugs, resistors and other electronic components.

Jack E. Wilson was appointed audio products manager of National Co., Malden, Mass. He was formerly with the Collaro Division of Rockbar Corp.

John J. Traveyas was promoted to general service manager of United Motors Service Division of General Motors, Detroit, Mich. He was formerly assistant general service manager.

Obituary
Bertram J. Grigsby, co-founder of the old Grigsby-Grunow Co., a director of Grigsby Allison Co., Arlington Heights, Ill., and a well known pioneer in the radio industry, died early this fall at his home in Barrington, Ill.

Personnel notes
...Elmer W. Engstrom was elected a member of the Board of Directors of RCA, filling the vacancy caused by the retirement of Walter A. Buck. Dr. Engstrom is executive vice president, research and engineering; head of RCA Laboratories and a director of RCA Victor, Ltd., Canada.

...O. O. Schreiber, assistant to the president of Philco Corp., Philadelphia, was named vice president of the corporation. He will continue to handle special assignments for the president and chairman of the board of Philco and act as secretary of the Policy and Management Committees and chairman of the Coordinating Committee of Philco, Inc.

...H. G. Cheney was appointed Midwestern regional sales manager for the Westinghouse Electronic Tube Division with offices in Chicago. He joined the Electronic Tube Division in 1951 as sales manager.

...A. Raymond Bermond, previously radio advertising manager of Hallicrafters, Chicago, was promoted to the post of advertising manager of the company, succeeding John S. Mahoney who left the company to join Sheriff-LeValley, Chicago advertising agency, as account executive on Hallicrafters' account. END
**CIRCOLotron** high-fidelity amplifier (Electro-Voice) includes new design principles. No d.c. in output transformer, less than quarter the plate circuit.

Impedance of conventional output circuits. Damping factor control permits perfect match of amplifier output impedance to loudspeaker's critical damping resistance. Output 20 watts rated, 40 watts on peaks. Frequency response ± 0.1 db, 20-20,000 cycles at 20 watts.

Circolotron has necessary controls for handling complete high-fidelity system. Inputs are Ultra-Linear phone, magnetic phone, tape or television, tuner or television, and high-impedance microphone. Controls consist of function selector, record compensation, level, loudness, bass, treble, damping factor, and two hum adjustments on chassis.


**10-Watt Amplifier**, Freed-Eisemann model 510, Williamson type. Frequency response 20-30,000 cycles at ± 1 db. Total harmonic distortion less than 0.5% at full rated output. Hum and noise level 80 db below rated output.

Freed-Eisemann, 200 Hudson St., New York 13, N. Y.

**Hi-Fi Equipment.** Espey model 710 (not illustrated) single chassis mounted receiver incorporating 12-watt Williamson type amplifier with integral control panel for ease of installation. Models 700 (tuner) and 501 (amplifier) designed for use together or with any standard audio system. The 700 and 710 use limiter-discriminator circuits. 501 is 24-watt Williamson type audio amplifier.

Espey Manufacturing Co., Inc., 529 E. 72nd St., New York 31, N. Y.

**12-Watt Amplifier**, Rauland model 1811 12-watt output at 0.6% harmonic and 2% IM distortion (100 and 7,000 cycles, 4-1). Response ± 0.5 db, 20 to 20,000 cycles. Hum and noise 75 db below 12 watts on tape and tuner inputs, 55 db below on mike and phone inputs. Controls: Separate bass and treble; volume, input (provides RIAA, frr, and "quiet" record compensation). Rauland-Borg Corp., 3515 W. Addison St., Chicago 18, Ill.

**Antenna**, CBS-Columbia Deline V-Reflector, designed for color TV reception, provides minimum of dips and peaks, maintaining flat response within 2 db across entire v.h.f. and u.h.f. spectrum. Average gain about 7 db relative to resonant dipole at u.h.f. and approximately 3 db over the v.h.f. range.

CBS-Columbia, Inc., 34-00 47th Ave., Long Island City 1, N. Y.

**NEW DEVICES**

Comparing... prove EMC superiority

**Model 206**

**Mutual Conductance Tube Tester**
- Checks mutual conductance on calibrated microhmm scale
- Checks tubes for open and short elements
- Tests all tubes from .75, to 117 filament volts
- Tests all tacho, actial, and miniature tubes
- Checks individual sections of multi-purpose tubes
- Built-in roll chart
- Uses lever type switches

MODEL 206P—With hand-rubbed oak carrying case (illustrated) $83.50
MODEL 206C—Stapling counter case $79.50

**Model 205**

**Tube Tester**
- Tests all tubes including Naval and sub-miniatures
- Completely flexible lever type switching arrangement
- Tests all tubes from .75 volts to 117 filament volts by standard emission test
- Tests all cold cathode, magic eye, voltage regulated, neon, tubes
- Has pilot light indicator
- Line voltage control compensates for line variations between 105 and 135 volts
- Checks for shorts and leakages
- Three-color hemantane panel

MODEL 205P—with hand-rubbed oak carrying case (illustrated) $47.50
MODEL 205C—Stapling counter case $46.50

**Model 204**

**Tube-Battery-Ohm Capacity Tube Tester**
- Tests all tubes including Naval and sub-miniatures
- Tests all batteries under rated load
- Emission testing method gives easy, direct readings
- Tests resistance to 4 megohms
- Tests condensers from 50 to 1 meg
- Uses four-position lever type switches
- Checks condenser leakages

MODEL 204P—Portable oak case, removable cover (illustrated) $55.90
MODEL 204C—Stapling counter case $54.90

**E.M.C.**

Write Dept. RE-12 for Free Complete Catalogue of tube and other Instruments.

Electronic Measurements Corporation
280 Lafayette Street
New York 12, N. Y.

Export Department 136 Liberty Street, N. Y. C. 6, N. Y.

DECEMBER, 1954
ANTENNA, Tezo Trapper Jr., available as single, two- and four-stacked array. Good gain on v.h.f. TV and FM. Covers u.h.f. band in medium signal areas. Driven element with wave traps backed by high- and low-frequency reflectors providing six working elements on high band. Fiberglass insulators on antenna sections. Technical Appliance Corp., Sherburne, N. Y.


COAXIAL SPEAKER, University model BLC, 22½ inches in diameter, 9 inches deep, 8-inch throat. Separate high-frequency tweeter, coaxially mounted, with its own driver unit coupled to wide-angle horn. Built-in frequency dividing network, crossover at 2,000 cycles. Rated frequency response 70-15,000 cycles. 25-watt rating. University Loudspeakers, Inc., 80 So. Kensico Ave., White Plains, N. Y.

TOWER, Bohn No. 6, for home TV installation and communication requirements. Self-supporting to 50 feet. Features 12½-inch triangular design with heavy-duty corrugated cross bracing. Bohn Manufacturing Co., 116 Limestone, Bellevue, Peoria, Ill.


Invisible hands make light work... of faraway TV tuning with Regency model RT-700... based on an entirely new principle in remote control devices. Now it's possible to change channels, sharpen contrast, brighten the picture, control the volume, and most important—adjust the picture from where it is seen. So simple to install, that a service man can do it in minutes. Handsome cabinet design complements any decor... blond or mahogany finish.

Regency, world's leading manufacturer of TV accessories including VHF boosters, FM boosters, voltage boosters, UHF converters, chair side TV control, high pass filters and a complete line of high fidelity equipment.

REGENCY DIVISION, I.D.E.A., INC. INDIANAPOLIS 28, INDIANA

List Price $69.95
NEW DEVICES


MIXER-SEPARATOR. Blender-Tongue model MJS, requires no a.c. power. Capable of mixing and equalizing signals from any number of v.h.f. antennas. Can separate individual channels being transmitted in single cable. U.h.f. signals handled by converting to unused v.h.f. channel at antenna. Includes complete series of plug-in pads to supply up to 24 db attenuation on any one channel.

THE TELE-FAL. Tele-Matic's external speaker for remote-controlled TV listening. Has switch for shifting from remote unit to TV speaker, and remote volume control. Easily connected to speaker in TV set.

DECEMBER, 1954

BUILD 15 RADIOS ONLY
AT HOME $19.95

WHAT THE PROGRESSIVE RADIO "EDU-KIT" OFFERS YOU

The Progressive Radio "EDU-KIT" offers you a home study course at a rock bottom price. Our Marketing Research Department, radio hobbyists, and construction experience in Radio Theory and Construction Practice expressed simply and cleanly. You will learn the basic principles of radio receiver operation. You will be taught step by step, in plain English, to build and operate an actual radio receiver. You will learn the radio code alphabet, how to test parts, how to operate equipment. You will learn the radio code alphabet, how to test parts, how to operate equipment.

THE KIT FOR EVERYONE

The Progressive Radio "EDU-KIT" comes complete with instructions. These instructions are written in English and make it possible for anyone to learn radio. The "EDU-KIT" is designed for do-it-yourselfers of all ages and all levels of technical ability. The "EDU-KIT" is designed to be built by the average person.

PROGRESSIVE TEACHING METHOD

The Progressive Radio "EDU-KIT" comes complete with instructions. These instructions are written in English and make it possible for anyone to learn radio. The "EDU-KIT" is designed for do-it-yourselfers of all ages and all levels of technical ability. The "EDU-KIT" is designed to be built by the average person.

OSCILLOSCOPES. Electronice large-screen, available with 17- and 21-inch rectangular tubes. Designed for data plotting, production test, waveform analysis, education and display uses. Features are: high resolution, 1% linearity, stable d.c. amplifiers, calibrated time base (from 10 microseconds to 1 second per inch), gain controls calibrated in two peak-to-peak voltage ranges (10 millivolts or 1 millivolt per inch peak-to-peak sensitivity), low drift rate and long-term stability. Performance not affected by line-voltage changes from 105 to 125 volts.

FREE EXTRAS

- ELECTRICAL & RADIO TESTER
- ELECTRIC SOLDERING IRON
- TV BOOK
- RADIO TROUBLE-SHOOTING GUIDE
- CONSULTATION SERVICE
- QUIZES
- F.C.C. TRAINING

MAIL TODAY: LEARN TO BUILD AT HOME "EDU-KIT"...

Progressive Radio "EDU-KIT" Inc., 497 Union Ave., Dept. RE-99, Brooklyn 11, N.Y.

MAIL TODAY—Order shipped same day received.

Mail Order—Paid in Advance. Include FREE EXTRAS:

Send "EDU-KIT" Postpaid. I enclose full payment of $19.95 (U.S. only).
Send "EDU-KIT" Postpaid. I enclose full payment of $20.95 (Outside U.S.A. only).
Send "EDU-KIT" 250 V. Adapter for "EDU-KIT"—$2.50
Send "EDU-KIT" C.O.D. I will pay $19.95 plus postage (U.S.A. only).

I wish additional information describing "EDU-KIT" No Obligation.

Name:

Address:

PROGRESSIVE "EDU-KITS" INC.
497 UNION AVE., DEPT. RE-99, BROOKLYN 11, N.Y.

129
NEW DEVICES

TV TESTER. Seco FB-4 checks flyback interval and inductance at a glance. No disassembling of parts necessary. Reveals condition of coil components as connected group, from the standpoint of self-resonant frequency established with distributed capacitance in coils and circuit. Verifies transformer and yoke matching, also tests flyback transformer and yokes individually. Checks linearity and ringing coils by comparison.

Seco Manufacturing Co., 5015 Penn Ave. So., Minneapolis, Minn.

V.V.M. Sylvania 300 Poly- meter, has subminiature vacuum tube r.f. probe, peak-to-peak scale, 5-inch meter movement, lighted scale, linearity circuit shielded a.e. and r.f. leads, and screw-on connectors. Input impedance: 17 megohms.

Voltage Ranges: Peak-to-peak, 200 millivolts to 2,800 volts; d.c., 200 millivolts to 2,800 volts; r.f., or 50 millivolts to 1,000 volts; a.e., 50 millivolts to 1,000 volts; r.f., 100 millivolts to 200 volts (10 ke to 300 mc).

Resistance: 0.5 ohm to 1,000 megohms.

Decibels: -20 to +61.4.

D.C. voltage range may be extended to 30,000 with type 238 (30 kv) d.c. voltage multiplier probe.

Sylvania Electric Products, Inc., 221 3rd St., Williamsport, Pa.

4-WAY TOOL (CBS-Hytron) has three parts: slotted steel barrel with hex sockets (4 inch at one end, 5 1/16 inch at other); double-ended screwdriver blade (one end with Phillips head; other with hollow-ground standard head which slides inside barrel); and sliding blade into any one of three recessed positions within slotted barrel. When sliding blade is centered with the barrel, either hex socket can be used or either screwdriver head. Phillips or standard, may be drawn out of barrel and locked into position for use.

CBS-Hytron, Danvers, Mass.

CAPACITORS, Cornell-Dubilier tubular subminiature. Require no voltage derating even at +130 C. Maximum moisture resistance (exceeding MIL-C-21 requirements). Noninductively wound extended foils soldered to wire leads for low r.f. impedance. Mylar polyester dielectric, tubular stylesheet cases. Ends sealed with Polviane fill.

Cornell-Dubilier Electric Corp., South Plainfield, N. J.

RESISTOR. Sprague subminiature 3-watt wirewound. 13/64-inch diameter by 17/32-inch long (same size as conventional 4-watt carbon). Maximum resistance, 30,000 ohms. Blue vitreous enamel coating.

He's using the telephone that lends an extra hand

"Sure ... I'm looking up the figures right now!" From busy executives to clerks, people in business can work more easily and efficiently with Bell's new Distant Talking Telephone. Small white rectangle is the loudspeaker.

For people who want to keep both hands free when they telephone, Bell Telephone Laboratories engineers have devised a new telephone with a sensitive microphone in its base.

To use it, simply press a button. The microphone picks up your voice and sends it on its way. Your party's voice comes to you through a small loudspeaker. Both hands are left free.

The volume can be adjusted to suit yourself. If privacy is needed, you simply lift the handset; this shuts off the microphone and loudspeaker and you talk just as you would on a regular telephone.

This new development of Bell Laboratories increases the number of ways your local Bell telephone company can serve in businesses and homes.

Bell Telephone Laboratories

Improving telephone service for America provides careers for creative men in scientific and technical fields.

DECEMBER, 1954
DU MONT has joined the contest for bigger and better color TV picture tubes with what is claimed to be the first 21-inch triode shadow-mask tube. The unit is of metal cone construction and provides approximately 225 square inches of picture area. Du Mont announced it expects to make quantity deliveries in 1955.

To improve picture performance in TV receivers using a single series-connected heater string, including the heater of the picture tube, RCA has announced the 6CG7, a 9-pin miniature version of the popular 6SN7-GT. The tube is intended for use particularly as a vertical and horizontal deflection oscillator, and is designed with a 600-ma heater having a controlled warm-up time.

The medium-mu triode contains an internal shield to prevent coupling between the triode units. It may also be used as a phase inverter, sync separator, and amplifier, and resistance-coupled amplifier.

RCA has announced an addition to its line of short, wide-angle picture tubes-the 21AMP-4A, a 21-inch, 90° deflection type using magnetic focus and deflection. Measuring 20¾ inches in length, the 21AMP-4A is a rectangular glass tube with an aluminized screen measuring 19¾ by 15 inches. It has a screen area of 263 square inches, an external conductive bulb coating, and uses a single-field ion-trap magnet.

Two 70 TV picture-tube types have been placed on the market by RCA: the 14HP1 and 21YP-4A. Both have been used as initial equipment in numerous brands of TV receivers and are now available for general replacement service.

The 14HP1 is a rectangular glass tube with a screen size of 11¾ by 8¾ inches. It uses low-voltage electrostatic focus and magnetic deflection, and has a maximum high-voltage rating of 14,000.

The 21YP-4A, an aluminized glass picture tube with a screen size of 19¾ by 14¾ inches, has a spherical filter-glass faceplate. It uses low-voltage electrostatic focus and magnetic deflection, and has a maximum high-voltage rating of 18,000. The tube has an external conductive bulb coating, and uses a single-field ion-trap magnet.

RCA has announced 20 new series-string tubes, not including the 6CG7 mentioned. They have 600-ma heaters designed to operate in a single series-connected circuit. These are in addition

**CUSTOM-DESIGNED**

for your individual needs

- Lowest Installation and Maintenance Cost of All Systems.

In the difficult Metropolitan New York area, more TRANVISION Master Antenna Systems were selected and proven out than any other system.

Features:
1. Highest power output allows longest cable runs...
2. Central system with only 1 point of service ...
3. Operates up to 500 TV sets from 1 amplifier...
4. Lowest cost per outlet...
5. Maximum signal to each set with minimum interference...
6. Standard units are provided...
7. They may be modified to meet your special reception problems and layout. We also offer you:

**FREE:**
Engineering advice. Send us your problem for quick solution.

**FIELD STRENGTH METER**
For 110V AC, and Battery Operation

- Saves 50% of installation cost...
- Measures all signal strengths directly from antenna... ...
- Operates without light source...
- Money Back Guarantee

Model FSM-5, for 110V AC only, complete with tubes, Weight 10 lbs., not $79.

Model FSM-5B, for Battery Operation and 110V AC, Weight 12 lbs., not $89.

TRANVISION, Inc., New Rochelle, N. Y.

JOBSBER INQUIRIES INVITED
These new books can help you get ahead in electronics

The new Gernsback Library books are the only technical books in the field dedicated to supplying technicians with a constant source of up-to-the-minute technical information at low cost. New titles are continually being added. These are the latest. Pick out the books you want and order them from your distributor today.

THE OSCILLOSCOPE—No. 52
A practical book that tells you how to use the scope to best advantage in all types of servicing. Gives you tips on use and warns about pitfalls to avoid. 192 Pages. Over 100 illustrations. $2.25

TRANSISTORS—THEORY AND PRACTICE—No. 51
Rufus P. Turner explains transistors for the practical man. Gives applications in well-known circuits. Contains guide to characteristics of commercial transistors. 144 Pages. $2.00

TV REPAIR TECHNIQUES—No. 50
Top technician-writers tell you how to recognize and correct quickly the tough servicing problems which stump even the experts. Will help you do a better job faster. 128 Pages. $1.50

RADIO & TV TEST INSTRUMENTS—No. 49
How to build just about every modern TV-radio audio servicing instrument, plus unique chapters on servicing broadcast and carrying case construction. 128 Pages. Over 100 illustrations. $1.50

See coupon on page 140
Gernsback Publications, Inc.
Publishers of RADIO-ELECTRONICS
25 West Broadway New York, N.Y.

NEW TUBES
(Continued) to the seven series-string tubes announced in the October issue. Except for changes in heater design, these new receiving tubes correspond electrically and mechanically to prototypes widely used in home TV receivers.

The tubes, followed by their prototypes, are: 2AF4-A (6AF4), 3AL5 (6AL5), 3AU6 (6AU6), 3AV6 (6AV6), 3BY6 (6BY6), 3CF6 (6CF6), 4BQ7-A (6BQ7-A), 4BZ7 (6BZ7) 5AQ5 (6AQ5), 5AS8 (6AS8), 5X8 (6X8), 6AU7 (12AU7), 6A4 (6A4), 6SN7-GTB (6SN7-GT), 12AX4-GTA (12AX4-GT), 12BH7-A (12BH7), 12BQ6-GTB (6BQ6-GT), 12BY7-A (12BY7), 12W6-GT (6W6-GT), 25CD6-GA (25CD6-G).

A new line of subminiature germanium diodes, using a miniature gold junction, has been announced by Hughes Aircraft, Culver City, Calif. The diodes have a high forward and high back resistance, making them particularly suitable for use in magnetic amplifier circuits, clamps and d.c. rectifiers.

These gold-junction diodes are made with a fusion-sealed, one-piece glass body (see photo). They are 0.265 inch long and 0.130 inch wide.

Radio Thirty-Five Years Ago In Gernsback Publications

HUGO GERNBACK
Founder
Modern Electrics 1908
Wireless Association of America 1908
Electrical Experimenter 1913
Radio News 1919
Science & Invention 1920
TVEXTRON 1922
Radio-Craft 1928
Science & Invention 1931
TV ENGINEER 1931

In December, 1920 Science and Invention (formerly Electrical Experimenter)

The Nauen Radio Station
Elieve Radio Station
Transmitting Photographs by Radio, by Pierre H. Boucheron
Light Ray Phonograph, by Eric A. Dine
Automatic Indicator for Storage Cells
New Precision Condenser and Wave-meter
Magnitude of Tesla High Frequency Currents

FLYBACK & YOKE TESTER KIT

The new line of subminiature germanium diodes, using a miniature gold junction, has been announced by Hughes Aircraft, Culver City, Calif. The diodes have a high forward and high back resistance, making them particularly suitable for use in magnetic amplifier circuits, clamps and d.c. rectifiers.

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Radio Thirty-Five Years Ago

In Gernsback Publications

HUGO GERNBACK
Founder
Modern Electrics 1908
Wireless Association of America 1908
Electrical Experimenter 1913
Radio News 1919
Science & Invention 1920
TVEXTRON 1922
Radio-Craft 1928
Science & Invention 1931
TV ENGINEER 1931

In December, 1920 Science and Invention (formerly Electrical Experimenter)

The Nauen Radio Station
Elieve Radio Station
Transmitting Photographs by Radio, by Pierre H. Boucheron
Light Ray Phonograph, by Eric A. Dine
Automatic Indicator for Storage Cells
New Precision Condenser and Wave-meter
Magnitude of Tesla High Frequency Currents

FLYBACK & YOKE TESTER KIT

Electronic Instrument Co., Inc.
84 Witherly Street, Brooklyn, N. Y.

BRAND NEW TUBES

FAR BELOW WHOLESALE

EICO's 38

FLYBACK & YOKE TESTER KIT

Checks all flybacks & yokes instantly—in or out of set.

Depends even on shorted turn.

Exclusive separate calibration for air & iron-cored flybacks for accurate testing of all types.

Tests continuity of coils, speakers, switches, etc.

Large 4½" meter, 3 colored scales.

Complete with easy instructions.

Compact, rugged, stylishly styled.

See it at your jobber today. Write for FREE Catalog CF-12 describing EICO's 38 Kits and 42 Wired Instruments. Prices 25¢ higher on West Coast.

EICOnight...the answer in electronics

STEVE-EL Electronics Corp.
61 Reade St.
New York 7, N. Y.

FREE 1955 Catalog!
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KEY TO SYMBOLS & ABBREVIATIONS

*Construction Articles

Section of a full-length article
Clinic ____________________ Television Clinic
Q&A ____________________ Question Box
REC ____________________ Radio-Electroni Circuits
RM ____________________ Radio Month
(For these items only—Use the terminology listed above except indicated)

Correction

There is an error in the diagram of the portable receiver in the article “A More Compact Battery Portable” on page 90 of the September issue. The 15,000- and 120,000-ohm resistors in the last two audio stages should return to the negative side of the 4.5-volt battery —not the positive side as is shown in the diagram.
REPLACEMENT GUIDE
An 8-page bulletin “TV Replacement Transformer Popularity Tables,” No. 409, lists 100 major set manufacturers and replacements for their components. These manufacturers produce over 90% of the sets listed in the Stancor TV Guide.
Chicago Standard Transformer Corp., Standard Division, Addison and Elston, Chicago 18, Ill.; or any Stancor distributor.

MINIATURE CONNECTORS
Bulletin SR-RM1 on miniature connectors gives in 4 pages test data on thermal shock, humidity, corrosion, contact resistance, insulation resistance, vibration, physical shock, durability, air leakage, moisture resistance, immersion, high temperature, arcing resistance, dielectric strength, and other pertinent information.
Cannon Electric Co., 3809 Humboldt St., Los Angeles 31, Calif.

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An 8-page catalog, No. 84A gives descriptions, specifications, performance, and replacement data on pickup cartridges. Also described are converters, boosters, recording heads, phonograph pickups, microphones, stands and microphone cartridges.

The Astatic Corp., Commentary, Ohio.

1955 CATALOG

Allied's 1955 catalog No. 140 contains 308 pages and lists over 25,000 items. Among the products described are the latest high-fidelity components, television chassis, professional and home recording equipment, amateur receivers, transmitters and other gear, and industrial v.h.f. radio and radiotelephone equipment.

Allied Radio Corp., 200 No. Western Ave., Chicago 80, Ill.

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Booklet P-84 (12 pages) contains illustrations, dimensional drawings, technical charts, data and general information on 12 new steppers, including midget, high-speed, vibration-resistant and interlock types for add-subtract, continuous rotation, electrical reset and other applications.

Guardian Electric Manufacturing Corp., 1021 W. Walnut St., Chicago 12, Ill.

DATA SHEETS

Five data sheets, 1801 to 1805, give detailed descriptions on industrial relays, inspection, servicing and maintenance. Dimensional drawings, wiring diagrams and specifications are provided for plug-mounted classes A, B, F, S and Z relays.

Automatic Electric Sales Corp., 1033 W. Van Buren St., Chicago 7, Ill.

DELAY LINES

No. 54, a 12-page brochure, gives illustrations, descriptions and technical data covering the lumped constant, distributed constant and ultrasonic delay lines. Also of special interest is a description of the test procedure used by the company as part of their strict quality control. Design, engineering and manufacturing facilities are described.

Richard D. Brew and Co., Inc., Airport Road, Concord, N. H.

TEST EQUIPMENT

Hickok's 4-page brochure describes a TV color-bar generator, white-dot generator, 5-inch oscilloscope, sweep-alignment generator, oscilloscopes and tube testers.

Hickok Electrical Instrument Co., 10519 Dupont Ave., Cleveland 8, Ohio.

RADIO SHACK CATALOG

Radio Shack's 1955 224-page catalog lists and illustrates 30,000 radio, TV and electronic products. It includes a manufacturer's index on the front cover, an 18-category thumb index and a product index.


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This is a radio communications course designed for those preparing to take FCC amateur license examinations. It discusses the written-examination requirements for the novice, technician, general, conditional, and advanced classes of licenses.

The book is divided into three sections: The first covers introductory theory and d.c. and a.c. circuits; the second, vacuum tubes, audio amplifiers, microphones, and loudspeakers; the third, transmitters, receivers, antennas, and FCC rules and regulations. Each section is divided into several lessons, with questions at the end of each. After each section there is an examination. Answers are provided to all questions. At the end of the book are two final examinations. One is for those preparing for any license, except the novice class; the other is strictly for the novice class. All questions are of the multiple-choice type.

This book by no means provides a comprehensive coverage of radio, but it does a good job for its purpose-license preparation.—JK
DECEMBER, 1954

BOOKS (Continued)


This new standard is a complete revision and coordination of five American Standards for graphical symbols. The document contains single-line diagrams and simplified symbols for radio electronics, telegraphy, telephone and power. It is prepared for engineers, designers, manufacturers and installers of equipment to show interconnection and functions of electrical circuits.


This book, the publisher states, represents the first of a series reporting the results of trouble-shooting a large number of television receivers in the John F. Rider Labs. No attempt is made to explain circuit operation in detail or to analyze the results from a theoretical standpoint. The book covers the four most popular horizontal a.f.c.-oscillator circuits.

They include the pulse-width a.f.c. oscillator (Synchroguide); phase detector-stabilized multivibrator a.f.c. oscillator; phase discriminator-sine wave oscillator (Synchrolock); and phase detector-sine wave oscillator.

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In addition, at the rear of the book is a long fold-out page containing the schematic diagrams and waveforms of each of the four a.f.c.-oscillator circuits discussed. This permits the reader to have the schematic before him while leafing through a chapter.

The book does a very good job for the service technician interested in a study of cause and effect.—JK END

CORRECTION

The parts list for the transistor oscillator on page 54 of the September issue specifies three .01-µf capacitors. The correct value is, as you have no doubt guessed, .01 µf.

The formula for distributed capacitance, in the middle column of page 57 of the October issue, is incorrect. Its correct form is

\[ C = \frac{C_1 - 4C_2}{3} \]

We thank William Pickering of Vancover, B. C., and Walter Planker of Camden, N. J., for bringing these corrections to our attention.

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