NOW... see how you score on the
RCA BATTERY
"Get the Facts" Quiz

The battery for the Radio Trade!

Try yourself out on these leading questions about RCA Radio Batteries. You’ll profit by the answers whether you get a high score or not...because these are the facts about RCA Radio Batteries everybody should know.

Official Questions and Answers from the $10,000 RCA Battery "Get the Facts" Contest (Ended July 10, 1950)

1. Of all the radio battery brands sold—which brand is recognized as "The Battery for the Radio Trade"?
   (Answer) RCA.

2. Why is RCA recognized as "The Battery for the Radio Trade"?
   (Answer) Because RCA sells its batteries to Radio Dealers and Servicemen through Radio-Electronics Distributors.

3. What sales advantage does RCA's Battery distribution give retailers?
   (Answer) Virtually no RCA Battery competition from non-radio outlets.

4. How complete is RCA's line of Radio Batteries in terms of consumer needs?
   (Answer) RCA's line covers 99% of today's radio battery demand.

5. What is the most outstanding Battery in RCA's line?

6. Why is the RCA-V5036 an outstanding battery?
   (Answer) Because it contains a Special Radio Mix—top, bottom, and sides.

7. What important advantages does the RCA-V5036 offer retailers?
   (Answer) Sealed-in-Steel construction eliminates shelf-life problems—the battery stays fresh until used.

8. Why will customers prefer the RCA-V5036?
   (Answer) It's virtually leakproof and, in radio operation, gives more operating hours than competitive flashlight-type batteries.

9. Is the RCA-V5036 good for Flashlight use?
   (Answer) Yes. It exceeds the U.S. Bureau of Standards Household Flashlight Cell Capacity requirements by nearly 50%.

10. What RCA "Exclusive" helps you sell the RCA-V5036?
    (Answer) The "Carry Kit" merchandiser—a "carry-away" container that sells batteries like soda pop—eight at a time!

11. What is your assurance of top quality in RCA Batteries?
    (Answer) Only the finest materials and skilled workmanship go into RCA Batteries. Every cell is aged and individually tested.

12. How has it been proved that RCA Batteries give extra listening hours?
    (Answer) Exacting laboratory performance tests show that RCA Batteries exceed the average of competitive brands.

13. What steps are taken to safeguard the quality standard of RCA Batteries?
    (Answer) Constant research, product development, and the continued application of latest production techniques.

14. Are radio set manufacturers recommending RCA Batteries for their Portables?
    (Answer) Yes—in ever-increasing numbers!

15. What specialized knowledge does RCA bring to Radio Battery retailers?
    (Answer) RCA—World Leader in Radio—is closer to radio-battery market requirements than any other battery supplier. Such knowledge assures dealers availability of the right types—at the right time—backed by proved radio trade sales programs.

16. What selling aids does RCA provide?
    (Answer) All types of displays and merchandisers, together with literature—the most complete line in the industry—all geared for radio trade use.

17. What are some RCA "Firsts" in battery sales promotion?
    (Answer) The RCA "Carry Kit," Toy Shipping Container, Basic Sales Aid Kit, and, Radio-gear Sales Aids!

18. What is your guarantee of high consumer acceptance of RCA Batteries?
    (Answer) Today, as always, the RCA Trade Mark is your assurance of immediate customer acceptance. Consumers everywhere recognize RCA as "the Greatest Name in Radio."

19. What can radio dealers and servicemen do to prove to themselves that they can sell more RCA Batteries than any other brand?
    (Answer) Starting now—sell RCA Batteries. Watch your business grow!

See your RCA Battery Distributor for fast, reliable service.

RADIO CORPORATION of AMERICA
RADIO BATTERIES
HARRISON, N. J.
NEW

LEARN COMMUNICATIONS
by PRACTICING at Home in Spare Time

YOU BUILD
this Transmitter Power Supply
used in the basic experiments in
RF and AF amplifiers, frequency
multipliers, buffers, etc.

YOU MEASURE
current, voltage (AC, DC and RF),
resistance and impedance in cir-
cuits with Electronic Multi-
tester you build. Shows how
basic transmitter circuits be-
have; needed to maintain sta-
tion operation.

YOU PRACTICE
setting up code,
amplitude and frequency modula-
tion circuits (put voice, music,
etc., on "carrier signals") you
produce. You learn how to
get best performance.

YOU BUILD
this Wave-meter and use it to determine frequency
of operation, make other tests
on transmitter currents.

I TRAINED THESE MEN
"N.R.I. has been my step-
ing stone from a few hun-
dred to over $1,000 a year
as a Radio Engineer."—
ALTON B. MICHAELS,
Trenton, Georgia.

"Am Broadcast Engineer at
WLPY. Your NEW
Communications course shows
the kind of equipment we
use."—J. BANGLEY, JR.,
Suffolk, Virginia.

"I am employed by WEDO
as transmitter operator.
Have more than doubled
salary since starting in Ra-
dio full time."—A. HERR,
New Canmerland, Penna.

"4 years ago, I was a book-
keeper with hand-to-mouth
salary. Am now Radio
Engineer with ABC net-
work."—N. R. WARD,
Ridgefield Park, N. J.

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with parts I send. With this Transmitter you practice how
to put a station "on the air." You perform procedures de-
manded of Broadcast Station Operators, conduct
many experiments, make many prac-
tical tests.

YOU TRAIN AT HOME
while you get practical
training with your NEW
Transmitter, which you
build. There are no
exams. You receive
training when you
are ready for it.

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Commercial Operator's License puts you
in line for a good job in Radio or Tele-
vision Broadcasting, Police, Marine, Avia-
tion, Two-way, Mobile or Micro-wave Re-
lay Radio. Mail coupon below for 64-page
book FREE. It will give you complete facts
about my NEW Communications course.

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with MANY KITS of
RADIO EQUIPMENT I SEND

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is like. Find out how I get you ready for a
brighter future, better earnings, more
security in Radio-Television. Send coupon
now in envelope or paste on a postal.
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WILL CALL! My book, sent to you
FREE, tells the full story. J. E. SMITH,
President, Dept. OHX, National Radio In-
tstitute, Washington 9, D.C.

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MR. J. E. SMITH, President, Dept. OHX.
National Radio Institute, Washington 9, D. C.

Mail me your 64-page Book about Radio and Television
Communications opportunities and training. (No sales-
man will call. Please write plainly.)

Name_________________________Age________

City_________________________Zone_________State________

☑ Check if Veteran
☑ Approved Under G. I. Bill
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NBC's experimental ultra-high-frequency television station K2XAK at Bridgeport, Conn. Engineer Vic Barry is at the controls and technician John Piorek logging the dial.

Kodachrome by Avery Stock.

GOOD PAY and Unlimited Opportunities in JOBS LIKE THESE:

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- Manufacturing, Sales, Service
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- Sound Systems and Telephone Companies
- Oil Well and Drilling Companies
- Engineering Firms
- Theatre Sound Systems, Police Radio

And scores of other good jobs in many related fields.

YOU CONDUCT MANY EXPERIMENTS LIKE THESE!

- Checking action of condensers
- Experiments with AP and RP amplifiers
- Experiments with resonance
- Producing beat frequencies
- Calibrating oscillators
- Experiments with diode, grid-bias, grid-leak and infinite impedance detectors
- Practical experience in receiver trouble shooting
- Application of visual tester in checking parts and circuits
- Experiments with audio oscillators
- Advanced trouble-shooting
- ... and many, many others

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The same highly trained faculty, instruction materials and methods used here in our large, modern resident school, are adapted to your training in your own home. Shop Method Home Training has been proved by hundreds of successful graduates.

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YOU LEARN BY DOING

You receive special laboratory experiment lessons to show you how to build with your own hands various experimental units such as those shown at left, and how to conduct many tests.

You will find all lessons easy to understand because they are illustrated throughout with clear diagrams and step-by-step examples that you work out yourself. Every piece of the equipment and complete lesson material we send you is yours to keep and enjoy, including the multimeter, experimental equipment, all parts of the Superhetrodynne, tube manual, radio dictionary, and complete, modern Television texts. All parts are standard equipment.

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**THE COMPLETE TV-TUBE DATA CHART**

All your TV-tube information at a glance... up-to-date listings of ALL standard types... all-glass and metal-glass, round and rectangular...dimensions and drawings... electrical characteristics...ion traps...bulb contacts...base diagrams...etc.

**ASK YOUR JOBBER FOR THE DU MONT TV TUBE DATA CHART. OR WRITE US.**

ALLEN B. DU MONT LABORATORIES, INC. • TUBE DIVISION • CLIFTON, N. J.

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### PICTION TUBE DATA CHART

**MAGNETIC DEFLECTION AND FOCUS TELEVISION TYPE CATHODE-RAY TUBES**

<table>
<thead>
<tr>
<th>Type</th>
<th>Power Supply Voltage</th>
<th>Construction</th>
<th>Type</th>
<th>Plate Stabilisation</th>
<th>Plate Current</th>
<th>Plate Voltage</th>
<th>Plate Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>3452</td>
<td>6.3 VDC</td>
<td>Glass Tube</td>
<td>Type A</td>
<td>60 Hz</td>
<td>100 mA</td>
<td>150 V</td>
<td>100 mV</td>
</tr>
</tbody>
</table>

---

**USING THE DATA CHART**

Several important factors must be considered when using this tube data chart as an index to tube interchangeability:

**Bulb Diameter (approximate, overall length)**

Tube sizes present a problem of tube interchangeability, unless the designer is aware that two types in question would not interchange in this manner.

**Conduction Coating**

If a tube without conduction coating is replaced for a tube with conduction coating, a 500-V to 1100-V supply connected between the high-voltage output lead and ground will cause proper set operation.

**Deflection Angle**

In general, there are three types of deflection in the same of television tubes, the cavity and bell circuitry in all glass types, and the coil circuitry for types with a metal case induction. When making tube changes, the deflection coil must be used.

**Deflection Angle**

In practice, the series deflection tube usually may be employed with all tube types having deflection angles of less than 60°. Types with deflection angles of 60° or over will require a wide-angle deflection tube.

---

**FREE!**

**THE COMPLETE TV-TUBE DATA CHART**

All your TV-tube information at a glance... up-to-date listings of ALL standard types... all-glass and metal-glass, round and rectangular...dimensions and drawings... electrical characteristics...ion traps...bulb contacts...base diagrams...etc.

**ASK YOUR JOBBER FOR THE DU MONT TV TUBE DATA CHART. OR WRITE US.**

ALLEN B. DU MONT LABORATORIES, INC. • TUBE DIVISION • CLIFTON, N. J.
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TODAY THE CATHODE RAY TUBE can be the crystal ball that forecasts your future. Is the picture clear and bright—or is it fuzzy and out of focus?

Are you going to learn how to install and service all types of TV and FM receivers? There can be no doubt that TV is the important field for greater earnings: 83 stations on the air (many more authorized); two million new sets in 1949; twelve million predicted by 1953; practically every area in the nation soon to be in range of a TV station. Technicians with specialized TV-FM training will inevitably have the inside track installing and maintaining all these sets.

CREI offers just the specialized training you need. It’s a streamlined course for the top third of the men in the field—thorough and complete. It gives practical answers to the technical problems you run into while servicing today’s intricate TV and FM equipment. It is up to date, constantly revised to cover new developments as they are adopted by the industry.

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SAMPLE LESSON FREE! "Television & FM Trouble Shooting" devoted to live, "dollar-and-cents", practical practice based on day-to-day servicing problems. Read this interesting lesson! See for yourself how CREI training can help you. Mail coupon for sample lesson, free booklet and details.

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* PRACTICAL RADIO ENGINEERING  
Fundamental course in all phases of radio-electronics

* PRACTICAL TELEVISION ENGINEERING  
Specialized training for professional radio-men

* TELEVISION AND FM SERVICING  
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PRACTICAL RADIO ENGINEERING  
AERONAUTICAL RADIO ENGINEERING  
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RADIO-ELECTRONICS IN INDUSTRY

MAIL COUPON TODAY

NAME:  
ADDRESS:  
CITY:  
ZONE:  
STATE:  
AGE:  
IF RESIDENCE SCHOOL PREFERRED, CHECK HERE.
"We doubled our store space... increased our staff"

Sylvania Electric Products Inc.

Advertising Department

Emporium, Pa.

Gentlemen:

During January, February and March, I sent 12,000 postal cards to each occupant of three suburban towns, Elmhurst, Villa Park, Lombard and countryside. I had a local letter service handle addressing and mailing at a cost of only $7.00 per card.

Business began to pour in soon after the first mailing. In January, it was up 30%, February 35% and March up 40% compared to service business in similar months in 1963. Yes, service business increased an average of 38% the first three months of this year, over last.

Due to the increased business we've had as a result of your campaign mailing, we have doubled our store space, increased our service staff from two to six bench positions, and added an outside staff for antenna installations.

Just thought you would like to know! Thanks for your help in the rapid growth of "Better Radio".

Cordially,

Howard J. Neutrogen
President
BETTER RADIO INC.

Now let this SYLVANIA DEALER CAMPAIGN boost your business!

The above letter is actual proof of how Sylvania's Service Dealer Campaigns step up sales.

Now the new fall campaign is ready for you. It's tied in with the advertising your customers will be seeing in the Saturday Evening Post, Life, Look, Collier's and Radio and Television Best. It's sure-fire, powerful and complete... from colorful window and counter displays to bright, business-pulling postal cards... even radio spot announcements and ad mats.

All yours ALL FREE... you pay only the postage (1¢ for each card). So don't delay, mail the coupon TODAY!

Sylvania Electric Products Inc.
Dept. R-1008, Emporium, Pa.

Send full details about Sylvania's Fall Advertising Campaign for Radio-TV Service Dealers.

Name__________________

Company__________________

Street__________________

City__________________

Zone State__________________

RADIO-ELECTRONICS for
HERE'S YOUR BIG CHANCE!

D.T.I. Can prepare you for a Profitable Future in TELEVISION
RADIO
and
ELECTRONICS

NO EXPERIENCE NECESSARY

NOW... BUILD AND KEEP A 16” RECTANGULAR “BLACK” PICTURE TUBE TELEVISION RECEIVER (This is an optional feature...described below)

See how DeForest’s Training, Inc. can start preparing you now for the opportunities ahead in TELEVISION... RADIO... ELECTRONICS. We provide practical training in your own home to help you gain the confidence and “know-how” to fit you for a responsible, well-paying job or your own business. D.T.I. trains you rapidly, thoroughly, by using modern instruction methods and equipment. Write today for free facts on how you, too, may get started toward a profitable, exciting career.

16 Big Shipments of Parts—Plus Lessons

Here is everything you need to prepare you at home for FASCINATING WORK, GOOD MONEY and a THRILLING FUTURE in one of America’s most promising fields. Work over 300 electronic experiments and projects from 16 big shipments of parts. This includes building and keeping all test equipment and radio set shown at left side of page. Modern easy-to-read lessons with handy fold-out diagrams simplifies your entire training.

You Also Use Home Movies

D.T.I., alone, includes the modern, visual training aid... MOVIES to help you learn faster, easier at home. See electronics on the march and other fascinating “hidden action” — a remarkable home training advantage that speeds your progress.

EMPLOYMENT SERVICE

When you complete your training, our effective Employment Service helps you get started toward a real future in Television-Radio-Electronics.

Modern Laboratories

If you prefer, you can get ALL your preparation in our new, Chicago training laboratories... one of the finest of its kind. Ample instructors... modern equipment. Write for details!

DeForest’s Training, Inc.
CHICAGO 14, ILLINOIS
Associated with the De Vry Corporation
Builders of Movie & Electronic Equipment

MAIL THIS OPPORTUNITY COUPON NOW!

Mr. E. B. De Vry, President
DeForest’s Training, Inc.
2533 N. Ashland Avenue, Dept. RE-6-8
Chicago 14, Illinois

Please show me how I may get started toward a good job or a business of my own in Television-Radio-Electronics.

Name ____________________________ Age ____________________________
Address ____________________________ Apt. ____________________________
City ____________________________ Zone. ____________________________ State. ____________________________

10 reasons why!

1. You profit from our 19 years of “Know-how” in preparing men for real opportunities in the Radio-Electronics field—which now includes Television.
2. A faculty of more than 60 skilled residential and 15 extension instructors.
3. You may learn at home or in our new, modern laboratories in Chicago.
5. You receive many shipments of commercial type equipment which give you practical “on-the-job” experience in your own home.
6. You build modern-type test equipment which you can regularly use for analyzing, checking and testing purposes.
7. You also get lessons with many illustrations and schematic drawings that make electronic circuits easier to grasp.
8. EFFECTIVE EMPLOYMENT SERVICE is available when you complete your training and help you get started.
9. Consultation Service. After you complete your training, you are privileged to write for additional information to help you with “on-the-job” problems.
10. Build and keep a quality, 16-inch rectangular tube Television Receiver. This is an optional feature—available at slight additional cost after completing training described above.
ANNOUNCING

2 NEW MODELS

Of Sun Radio's Famous All-Triode Amplifier

The renowned Sun Amplifier Model CR-10 is now offered in three models, the two new ones featuring the famous Peerless transformers. New models are approved by Consumers' Research, original designers. Here's more good news -- for the first time in many months this much wanted amplifier is available for immediate delivery on all models. And prices are scheduled to go up September 1st, so better buy yours now...

CR-10, standard model, as engineered to original design by CONSUMERS' RESEARCH OF WASHINGTON, N. J.

- Lab wired, tested, ready to use... $42.50
- Kit...... $31.50

CR-10-P uses Peerless transformers throughout, including output transformer designed especially for this amplifier. Improved low frequency response adds "presence" to reproduction.

- Lab wired, tested, ready to use... $69.50
- Kit...... $45.95

CR-10-Q using Peerless transformers throughout, features famous Peerless 3-240-Q Output Transformer for real presence effect. Note these specs:

- Frequency Response ±1db, 20-20,000 cps.
- Less than 2% Harmonic Distortion at 10W output.
- Source impedance at 4 ohm tap is 1.3 ohms -- this provides excellent damping of loudspeakers.
- Delivers full power within 1db from 40 to 10,000 cycles.
- Output transformer vacuum impregnated, moisture resistant.
- Output impedances available for any load 2 ohms to 100 ohms. (500 ohms available on special order at no extra cost).
- Transformer has split windings interleaved with secondary, making for extremely high efficiency and low losses.

- Kit...... $54.95
- Lab wired, tested, ready to use... $84.00

Sold Exclusively by Sun Radio, N. Y.

MAIL ORDERS FILLED PROMPTLY

The Radio Month

BANKING BY TELEVISION is carried on at the Glyn Mills Bank in London. A customer may go into the London office of the bank, and at a moment's notice, see his balance sheet projected on a television screen. In addition, an officer of the bank may view at his desk any document kept in the vaults. All the bank's records are kept in a secure place about twelve miles out of town where they were moved during the war for safekeeping.

The remote location of the bank's files makes this television system necessary. Besides permitting safe storage of documents in the present vaults, this system will actually cut down expenses by reducing costly filing space in London.

STEREO TELEVISION with three-dimensional pictures has been developed both by RCA and by the Argonne National Laboratory in cooperation with Du Mont. The two systems are very similar, both using a twin lens arrangement with the lenses separated about the same distance as between the human eyes.

The RCA system uses two vidicons (Radio-Electronics, May, 1950) side-by-side in the camera and two kinescopes at the receiver. The Du Mont camera has only one camera tube and the images are placed side-by-side in the space normally occupied by one image. The Du Mont receiver requires only one ordinary picture tube.

In both cases, complementary polaroid filters are placed in front of the two adjacent images and the observer wears a pair of polarized spectacles so that he sees the right-eye image only with his right eye and the left-eye image with his left eye.

The RCA receiver has the two kinescopes mounted at right angles and the two images are superposed by a semireflecting mirror; while in the Du Mont set the observer has glass prisms in front of his eyes with which he can adjust the two pictures into a single image.

The stereoscopic television is intended for use in watching and remote control of industrial processes which might be dangerous to an observer and in other applications where a two-dimensional picture is inadequate.

ELECTRONIC BANKING has been introduced at the County Trust Company's eighteen branches in Westchester County, N. Y., to speed the processing of checks. Individual customers receive pre-punched checks with their account numbers and these are used in the ordinary way. When the check is presented for payment it is handled entirely by mechanical and electronic machines that can "read" the checks.

SAFETY MEASURES for electronic methods and equipment are increasing - ly necessary according to a safety panel of the American Institute of Electrical Engineers at Pasadena, California.

In the last ten years nearly 5,000 workers were seriously injured and 380 were killed by electrical accidents in California alone. Most of these resulted from the inexperience of novice electricians and attempts to put equipment to the wrong use.

All electronic equipment designed for use by non-technical personnel such as radios, sound gear, television, X-ray and other medical and dental equipment must have safety built in by manufacturers.

Carl A. Hermanson of the Argonne National Laboratory performs remote control operations which he observes with Du Mont three-dimensional stereotelevision.

RADIO-ELECTRONICS for
FASTEST COMPUTER, recently put into operation, was developed for the Air Force by the National Bureau of Standards. The new electronic brain will be used to handle the astronomical calculations of the military budget. It is called the SEAC (Bureau of Standards Eastern Automatic Computer).

While the Bureau of Standards made no direct comparison with other electronic computers, it called the SEAC "the first automatically sequenced, super-speed computer to be put into useful operation." It can multiply or divide 11-digit numbers in 250 microseconds and add or subtract 11-digit numbers 1,100 times a second.

An input-output unit, a memory unit, a control unit, and an arithmetic unit comprise the four main sections of the SEAC. It can make comparisons of numbers and make limited decisions based on these comparisons and also store intermediate results.

Coded pulses can be fed to the calculator at the rate of 1,000,000 per second to instruct it. It then takes over the calculations, checks its own mistakes, and divides the answer when it has computed the problem to a predetermined accuracy. A typical operation of the SEAC is to set up mathematical tables for each aircraft in the Air Force in relation to the material needed to keep it operating. Estimated budget requirements for each of these aircraft on any of 200 possible budget levels are supplied within minutes.

SHIPBOARD RADAR RULES governing the license requirements for marine radar operators have been issued by the FCC. The new regulations which will go into effect in January 1961 specify that only licensed radio officers may make repairs, adjustments, tests, or perform maintenance and servicing duties on radar units in normal use as navigational instruments.

Deck officers may operate the radar unit and replace fuses and receiving-type tubes, but only holders of first or second class radiotelephone or radiotelegraph licenses may perform the technical functions until the new rules go in effect. Special FCC examinations, in "Element Eight", are being given covering the theory and practice of installation, servicing, and maintenance of marine navigational radar.

Radio officers must get a special endorsement on their licenses by passing this examination before they can qualify as radar officers.

WINDBLOW SHOPPING has become a reality at New York’s Lewis and Conger department store. A coin-operated microphone installed on the outside of a display window is connected to a tape recording machine inside the window. The window shopper can order any items in the window simply by depositing 25¢ in the microphone and giving his order verbally. The following morning the orders are transcribed and delivered to the customer with a refund of the 25¢ deposit. The device operates 24 hours a day. Similar devices have been installed in several other cities.

Window shopping is now a literal term.

The "Tell-it-to", as the window shopper is called, is a development of the L. A. Darling Company of Bronson, Michigan. The recording apparatus is manufactured by the Wilcox-Gay Radio Corporation of Charlotte, Michigan.

The Radio Month

BIG MONEY IN RADIO and TELEVISION NOW!

See COYNE'S Brand New 7 Volume Set
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Merchandising News
Ward Products Corp. of Cleveland offers as a new sales aid a display rack for store counters which illustrates Ward's national advertising. The easel stands 18 inches high and 15 inches wide. Ad replacements will be sent to dealers several times a year.

General Electric Co. has recently issued a series of "Tele-Clues" designed to aid service technicians in localizing circuit troubles in TV sets. A series of photos shows pictures produced on the TV screen when various components are defective.

The Tele-Clues are available without charge through G-E and Ken-Rad distributors. They are inserted in the monthly publication, "Techni-talk" published by the tube division of G-E.

Jensen Industries, Inc., has developed a new replacement needle package for phonograph dealers and service technicians. Called the "Jensen No. 5 Dealer Pack," the combination consists of a balanced assortment of 26 different needles individually packaged in a plastic box. The assortment will take care of the majority of requests for new needles.

The Television Contractors Association, Philadelphia, has begun distribution of 20,000 copies of its new Television Service Guide Book. Ten chapters describe how television works, how to select a receiver, how to get the best reception, and how to reduce service calls. It is intended for prospective or present TV set owners. The book will be distributed after an installation or upon request.

The Chicago Parts Distributors Show management reported more than 2,000 distributor personnel of over 800 companies packed the 168 booths and 134 display rooms. An estimated 8,000 persons attended the show at the Hotel Stevens from May 22 to 25.

The sixth annual Pacific Electronic Exhibit, a joint meeting of the IRE and the West Coast Electronic Manufacturers Association, will be held September 15-15 at Long Beach, Cal.

The National Electronics Distributors Association (NEDA) will hold their first National Convention and Show in the Cleveland Public Auditorium from August 29 to 31. RADIO-ELECTRONICS will have a booth at both the Pacific and Cleveland shows.

New Plants and Expansions
RCA Victor expansion plans include the purchase of a large new building at Harrison, N. J., to provide 126,000 square feet of additional space for electron tube manufacturing facilities. Increased production facilities for radio and television manufacture are also planned. A new radio set factory will be erected in Canonsburg, Pa., and the Bloomington, Ind., plant will step up TV production by several hundred percent. Radio and phonograph manufacture will be transferred from Bloomington to Canonsburg and added TV facilities will be installed at Bloomington. . . . Sylvania Electric Products, Inc., has begun construction on a new $1,500,000 radio tube plant in Shawnee, Okla. When completed, about the first of the year, the plant will produce more than a million radio tubes per month. Sylvania has also temporarily reopened its Altoona, Pa., radio tube plant. Crest Transformer Corp. has taken over an entire new building in Chicago. This new plant now more than doubles the company's previous factory facilities. . . . The Square Root Manufacturing Corp. has purchased a 30,000-square-foot building on its 3-acre site in Yonkers, N.Y. The company will now produce a new projection TV set, a complete line of capacitors and flyback transformers in addition to its indoor and outdoor TV antennas. . . . Hy-Lite Antennae has moved its plant and general offices to larger quarters in New York City. . . . Aerovox Corp. has rented 200,000 square feet in the Nonquitt Mill, New Bedford, Mass. The company plans to add 800 additional employees by fall as part of its expanded production program. . . . RMA reported that TV production by members in April equalled the peak levels established in March with 420,026 sets manufactured. Home radio production hit 648,352 sets and auto radio production 234,554—a slight increase over the March rate. Of the 648,352 new home radios, 78,088 included FM reception facilities. In April, 498,624 cathode-ray tubes were sold to TV manufacturers. Of these
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Radio Business

51% were 14 inches or larger and 99% were 12 inches or larger. The RMA stated that total receiver manufacture reported by members for the first four months of 1950 was 1,447,782 TV receivers, 2,375,012 home radios, and 900,646 auto radios. Hoffman Radio Corp. is now occupying its new Plant No. 5 in Los Angeles.

Financial Notes

Sylvania Electric Products declared a 35¢ dividend on common stock and $1 on preferred.

Sprague Electric Co. announced a 20% boost in its dividend rate. Latest quarterly dividend is 30¢ on common stock.

Cornell-Dubilier Electric Corp. declared dividends of 20¢ per share on common stock and 13.31¢ on preferred.

General Electric Co. announced a 60¢ dividend payable July 29.

Hazeltime Corp. issued a 25¢ dividend on common stock.

Noblitt-Sparks Industries Inc. paid a 50¢ stock dividend—one share for each two held—and a 50¢ cash dividend on common stock.

Stewart-Warner Corp. announced a 25¢ quarterly dividend on common stock.

Philoic Corp.'s regular quarterly dividends amounted to 50¢ per share on common stock and 934¢ on preferred.

Stromberg-Carlson Co. paid a quarterly dividend of 50¢ on preferred stock.

Emerson Radio & Phonograph Corp. stockholders voted an increase in the authorized number of capital shares from 1,000,000 to 3,000,000. This increases outstanding shares from 879,805 to 1,759,610 and leaves 1,240,390 authorized but unissued shares.

Business Briefs

Astron Corp., East Newark, N. J., which makes metallized paper capacitors, will introduce a complete line of electrolytics and molded paper capacitors.

Remington Rand, Inc., has sold the television picture tube division of its advanced research laboratory to the Reeves Soundcraft Corp.

Motorola, Inc., has transferred a record $1,654,120.03 to its Employees’ Savings and Profit Sharing Fund.

Television Engineering Corp., a Westfield, N. J., service contractor, announced a new pay-as-you-go television service contract plan by which the set owner pays $5 for a one-year registration and then for service only as needed—not to exceed a fixed sum.

JFD Manufacturing Co., Inc., has established new terms of payment for its distributors.

R. C. Sprague, chairman of the RMA Town Meetings Committee, announced the appointment of committees to supervise the production of four sound slide films to step up radio and TV dealers’ management and sales methods. Films will be shown in 60 cities from coast to coast during August and September under the sponsorship of at least 16 radio and TV manufacturers and their local distributors. An estimated 40,000 to 50,000 TV receivers will be produced in Canada in 1950, according to RCA Victor, Montreal.

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  - Laboratory Technician
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opening for Engineer-Announcer, basic salary $62.50—real future for right man.”
Letter, January 30, 1950 from Chief Engineer, Broadcast Station, Tenn. “Rate openings
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RADIO-ELECTRONICS for
Television Problems

The rapid growth of television constantly brings new problems...

By HUGO GERNSBACK

AUGUST, 1930

The rapidly growing television art is and will continue in constant flux for a number of years to come. In the meanwhile, new problems constantly occur; these must be solved to keep abreast with television and the constant pressure and demand by the public.

One of the bad problems in our larger cities and suburbs is the inter-receiver interference between neighboring sets. Each television receiver is a miniature transmitter in itself and if aerials or sets are close together and if several receivers are on simultaneously, interference can take place. For example, the frequency of the local oscillator in some receivers tuned to Channel 2 is located squarely on Channel 5. All neighboring receivers tuned to Channel 5 will thus suffer interference. This interference manifests itself in various annoying patterns on the screens of all the receivers thus affected. The condition often becomes so annoying that it is almost impossible to follow the action on the screen. At times such interference is so strong that everything is completely blurred. Because the non-technical public does not know where the interference originates many people jump to the conclusion that there is something wrong with the receiver itself. The upshot is a call to the service technician. Thousands of such calls are made routinely nowadays. This makes inter-receiver interference a serious problem to the servicing industry, which finds it difficult—often impossible—to eliminate the trouble.

We have noted as many as 20 television antennas on a single apartment house roof. Some of these antennas are so close together that they almost touch. It also follows that the closer the aerials approach to each other, the greater the interference. Furthermore some receivers will cause interference over several hundred yards.

Inter-receiver interference also exists even if there is no roof aerial, when only indoor antennas are used. There are also many instances where television sets are close together as in apartment houses (as well as in other houses), where there may be a television set on one floor with another set on the floor above or below, or one on the opposite side of a wall on the same floor. Very often the separation of the sets is only six to ten feet or even less—in other words close enough to cause inter-receiver interference.

Some set manufacturers are belatedly waking up to this situation and improving their product. Several new front-end units which will be incorporated in next fall's sets have greatly reduced oscillator radiation. And it is only fair to say that not all present-day sets cause interference of this type. The problem has also been under serious consideration by the FCC.

In the meanwhile, what is needed is some reasonable-priced interference eliminator that will do the work and do it satisfactorily.

If such a device can be manufactured at a sufficiently low price it might become standard practice to install one when a new set is installed, at least until such time that non-radiating TV receivers are on the market.

Such a device would certainly greatly help the servicing industry. It would save untold thousands of dollars on useless calls, which service technicians have to make today.

How close are we to overseas and international television? The normal range of our present-day television transmitters is about 50 miles. The only long distance transmission means of today is by television networks, as is well-known.

But other means are in the offering. Early last June Brig. General David Sarnoff, Chairman of the Board of the Radio Corporation of America, when he was awarded the Grand Medal of the Association des Ingénieurs-Docteurs de France for his war and peace services, made the following significant announcement:

"International television is not far off. The scientific principles for linking the hemispheres by television already are known. The continued growth of this new art will some day make it possible to see as well as hear around the world.

"When this time comes, New York will look in on Paris and Paris will look in on New York. In fact, the peoples of all nations—large and small—will be able to see and hear each other directly and to understand each other better."

Coming from such a high informed source, we must listen respectfully and take it for granted that long distance television indeed is not far off.

It is reasonable to suppose that the radio-controlled television robot airplane, which the writer first projected in the November 1924 issue of the EXPERIMENTER magazine will then also become a reality, particularly for military use. While practical today, its range is limited to the horizon, but with long distance television such a plane could be conducted safely over thousands of miles because its dispatchers could watch the plane as it flies over strange territory. For war purposes, without any human being on board, such a plane would have tremendous military advantages.

Another pressing problem today is that our television sets are "one sided." Unlike radio (where you can sit in any part of the room and listen to your set) a television receiver can be watched from one direction only. Four- or six-way television which would permit the receiver to be placed in the center of the room where onlookers could see it from every point of the compass so far has not materialized, although it is technically feasible today. The writer projected such a receiver in the August 1935 issue of RADIO-CRAFT.

Today—with a projection television tube mounted vertically instead of horizontally and using a simple optical system of mirrors and lenses—it should be possible to construct a television receiver with four or six screens mounted vertically on top of the receiver. The single television cathode-ray tube would serve the multiple screens in such a way that all screens would carry the same image. Then, instead of grouping the viewers in one direction only, the television audience could enjoy the program seated anywhere in the room.

The cost of doing this should not come too high. The public will welcome such a TV receiver once it is introduced.
**Low-Voltage Supplies With No Transformers**

Cost and weight reduction make the transformerless low-voltage supplies increasingly popular in modern television receivers

By H. G. CISIN

**Elimination** of the low-voltage power supply transformer permits considerable reduction in the cost of a television receiver. Hence designers have been investigating the possibilities of transformerless low-voltage supplies. Several manufacturers have used power supplies of this type, not only to reduce costs, but also to obtain lighter and more compact receivers.

The working principle of the transformerless power supply is illustrated in Fig. 1. Invented by the writer in 1932 (U.S. Patent 2,056,256), this circuit since that time has been used in millions of a.c.-d.c. midget radio sets. It consists of a half-wave rectifier with its anode connected to one side of the source and its cathode connected to the input of a filter. This unit provides rectified and filtered d.c. suitable for operating vacuum tubes. Tube heaters in this circuit are connected in series across the unfiltered a.c. source.

This basic circuit has been applied with only slight modifications to a number of low-voltage power supply circuits in television receivers. Some such circuits use cascade voltage doubling and, in some cases, voltage tripling, obtained by additional rectifiers and electrolytic capacitors.

One of the problems of the transformerless power supply circuit is obtaining a suitable negative bias supply. If a power supply transformer is used, it is a simple matter to connect a tapped resistance between the center tap of the transformer and ground to provide bias potentials.

A method of providing negative bias in a television transformerless circuit was patented by the writer several years ago (U.S. Patent 2,917,178). Fig. 2 shows how this is done. Two supplies like that of Fig. 1 are used back-to-back with a common B-minus and C-plus lead, but the positions of the anode and cathode of the bias supply are reversed. Since electrons can flow only from cathode to anode, the current in the bias supply is opposite in direction to that in the B-supply and the bias voltage is negative rather than positive.

A common ground, one a.c. plug, and one off-on switch is needed. This is then a transformerless power supply which provides d.c. plate and screen voltages as well as negative grid bias voltages.

The commercial application of Fig. 2 is shown in Fig. 3, the low-voltage power supply of the Teletone model TV 149 television receiver. In this receiver, one of the diodes of a 2326 is used as the bias rectifier. One of the B-plus potentials is obtained from the portion of the circuit which uses a selenium rectifier. A higher B-plus voltage is supplied by using the other half of the 2326 as a doubler. Still higher potentials are obtained by using a 6X5GT/G as a tripler.

The heaters are connected in series in two parallel branches, except that total heater current flows through the heaters of the 6X5GT, 7FP4, and 6J6. Since the 6J6 draws less heater current than the other two, a 43-ohm shunt must be connected in parallel with it. One of the series branches has a 60-ohm voltage dropping resistor. Low-pass filters in the heater circuits of television receivers are used to keep high frequencies from entering the receiver via the power supply. The r.f. chokes generally have an inductance of a few microhenries.

Sets which use this type of supply range all the way from the Pilot TV-37 which has electrostatic deflection and a 3-inch picture tube to the General Electric 805-821 series which has electromagnetic deflection and a 12-inch tube.

Fig. 4 shows a typical low-voltage
The tube heaters are arranged in series with suitable limiting resistors and r.f. chokes. The series strings are connected in two parallel branches combined at one end to permit total current to pass through the 12KP4 tube.

The Motorola models VT71, VT73, and VT73A use a low-voltage power supply similar in most respects to the General Electric design. The Silver-tone model 8130 also has selenium rectifiers in a voltage doubler arrangement, but this circuit differs in the connections of the vacuum-tube heaters. Only three tubes are connected in the conventional transformerless series arrangement, the other heaters being energized by a special filament transformer having two separate 6.3-volt windings.

The Stewart-Warner AVC1, AVC2, and AVT1 television receivers are of the a.c.-d.c. transformerless type. They use selenium rectifiers for a.c. operation and a polarizing relay for removing the rectifiers from the circuit when the set is plugged into a d.c. source. Truetone model D2985 television set is also an a.c.-d.c. receiver with a ballast tube and selenium rectifiers. Truetone's D1990 and D2087 and Reventon's A-100X24, etc. are similar in design to the Stewart-Warner models, using a polarized relay for a.c.-d.c. operation. The entire circuit diagrams of these Stewart-Warner, Truetone, and Raytheon sets are shown in this article.

Emerson models 571, 606, 611, 612, 619, 620, 624, and 627 are of the low-voltage transformerless power supply type. These are available either with selenium rectifiers or with 25Z6 vacuum-tube rectifiers. Fig. 5 shows the low-voltage power supply of the models with selenium rectifiers. Voltage doubling is used both for the B-plus supply and the negative grid bias supply. This requires four selenium rectifiers. A filament transformer is used for supplying the heater current to the tubes in this receiver.

This circuit makes novel use of a ballast tube for shunting the 10-ohm protective resistors and the 75-ohm filter resistors with resistances of equal value. Unusual current surges will cause the ballast resistors to open in place of damaging more expensive components such as the selenium rectifiers or electrolytic capacitors. The ballast tube is relatively inexpensive and readily replaced. When one of these Emerson units stops working completely, the ballast tube is the first place to look for trouble.

Fig. 6 is a diagram of the same Emerson models' low-voltage power supply, except that 25Z6 rectifier tubes have been substituted for the selenium rectifiers. Voltage doubling is used as in the previously described circuit. In the branch which supplies the B-plus potentials, three diodes are connected in parallel to supply sufficient rectified current, which in this case is about 0.215 amp. In the negative grid bias branch, two diodes are parallel connected since this branch supplies only 0.136 amp. The ballast tube and the filament transformer are used in the same way as in the sets using selenium rectifiers and have not been redrawn in Fig. 6.

One of the most recent trends in television is to supply large-screen picture tubes which require accelerating potentials up to 25 or 30 kilovolts. In the early days of television such high voltages meant bigger and more costly power transformers. Since the development of the r.f. oscillator-rectifier high-voltage supplies, horizontal flyback supplies, and pulse-type high-voltage systems, combined with paddlers and triplers, voltages in excess of 30 kilovolts may be obtained at lower cost than was formerly required to get 3 or 4 kilovolts.
NOW that literally millions of television antennas are being installed on chimneys and roofs, the practical problems of erecting these antennas properly have become important. Fig. 1 shows three common methods of installing rooftop antennas. Two major troubles of the average TV set owner—whether he has installed his own antenna or had it erected by the local service technician—are severe swaying of antenna and mast in high winds, and an annoying hum caused by the wind vibrating the tubular members of the antenna. Many TV antennas are fastened to chimneys, with or without guy wires to prevent their swaying in the wind. If the mast is over 10 feet high, it is advisable to install guy cables. The guy cables (1/8-inch diameter stranded steel or other cable may be used for low masts, although 1/4-inch cable is preferable) are best anchored by clamps about two-thirds of the way up the mast.

If the mast is 20 feet or more in height, two sets of guy cables should be installed, anchoring the first set about 1/3 and the second set 2/3 of the way up the mast. Turnbuckles in each cable are useful to help tighten them.

Much of the humming noise heard when these TV antennas vibrate in a high wind will be reduced by proper guying, not to mention the reduced strain on the mounting clamps and the chimney or other support on which the antenna mast is mounted. We have seen many relatively high TV masts anchored on a small brick chimney with no guys at all! How long it will take to rip the mast loose, tear bricks out of the chimney, or even bring the whole chimney down during a strong blow is anybody’s guess. This is not only a menace to the chimney or roof, but there is also the danger that a falling brick, or the antenna itself, may injure someone.

Many TV installations work fine for the first six months or so and then suddenly develop trouble. A recent case of this kind where the image was jumping badly was traced to a sloppy antenna installation. Strong winds had loosened the antenna elements and the lead connections. When the elements were tightened, reception cleared up and the image was steady again. Part of the trouble here was due to the lack of guy cables to hold the mast steady.

Another cause of trouble is the use of iron or plated iron screws and nuts on the antenna to fasten the ribbon lead terminals in place. Brass screws and nuts are best, and it is a good idea to cover the connections with tape to protect them from the weather.

If the mast is to rest on the roof, install it on a substantial metal or wooden base support to protect the

![Diagram of TV antenna installation](image)

Fig. 1—Three common ways of mounting TV antennas on a rooftop.

![Diagram of guy wires](image)

Fig. 2—Fasten guys firmly to the roof.

![Diagram of guy cable anchor](image)

Fig. 3—How to secure the guys to mast.
pieces of joist near the outer edge of the roof to prevent leaks later. Sometimes it is possible to drill holes through the overhanging rafters, through which the guy cables can be anchored. The guys must be securely attached to the mast. Ways of doing this are suggested in Fig. 3.

One of the most annoying troubles with TV antennas is a loud humming when the wind is high. The first step to reduce this humming is to guy the mast properly. Next the ends of the tubular members of the antenna should be plugged or capped so that the wind blowing through (or across) the tubes will not cause them to act as "whistles" or resonant tubes, which vibrate at a startling rate. The tubular mast also must be plugged or capped at top and bottom to eliminate this whistle or singing effect. Filling the tubes with sawdust or even sand helps in stubborn cases.

Further, the base of the mast may have to be mounted in thick rubber to prevent the vibrations from being transmitted down the chimney. In one case the humming was so strong that even the steam and water pipes vibrated every time the wind was strong. The noise even interfered with the family's sleep. The trouble was finally solved by mounting the base of the mast in thick rubber liners placed inside the supporting clamps. (See Fig. 4.) The rubber may be placed inside the clamping straps fastened around the chimney or other support if preferred.

Fig. 4—Annoying hum is eliminated by mounting the mast in rubber supports.

Twin-leads are often left free to flap about in the wind. This is another source of trouble. The twin-lead should be fastened in insulators made for the purpose and anchored down along the house wall and the antenna mast as well. Where separate twin-leads are used for the high-low-frequency antennas, separate the leads at least 5 to 6 inches. Fig. 5 shows a few suggestions for installing lead-in wires.

Where the twin-lead passes through a wall an insulating tube should be used as indicated in Fig. 6. It is preferable to mount lightning arresters outside the building and run a ground wire (no smaller than No. 10 copper wire—it may be bare wire) to a piece of pipe driven into moist earth or to the nearest water pipe. If lightning arresters are mounted inside the building, they should be placed just inside the wall where the lead wires enter and the shortest possible ground wire used to connect the arresters to the nearest water pipe.

In some installations long runs of twin lead have been placed parallel to steam or water pipes, with twin lead close to the pipe. This is undesirable as it may change the impedance of the lead wire. In no case should the ribbon lead be closer than 6 to 8 inches to a metal pipe (including rain gutters and conductor pipes).

Most of our observations have been made about 30 miles from New York City. At this distance, a high antenna is usually considered essential but we found that an indoor antenna (even one located on the first floor of the house) gave excellent results. Another compromise which has worked very well in several cases is to place the TV antenna in the attic. This eliminates wind troubles and makes it more accessible for experiments such as changing its direction with respect to the transmitting station. Some TV owners have "fished" the twin-lead down through the wall as shown in Fig. 7. Make sure you are not running between or beside metal lath. Once the best position of the antenna has been determined, the tubular elements may be fastened to the under side of the roof rafters. With suitable insulators no mast is necessary.

While a high, outdoor TV antenna is still best in most cases, it will pay to do a little experimenting with an indoor or attic antenna. An indoor antenna in an attic may not be very effective if the roof covering is metal; the writer's experience has been with roofs covered with wooden or slate-surfaced shingles.

If your roof is of these materials, you may be agreeably surprised to find that for your location the indoor antenna will give comparable results to an outdoor antenna. An indoor antenna with telescopic arms can be purchased at a nominal price. A little experimenting with one or two of these is better than a theoretical study of the pros and cons of indoor antennas. "The proof of the pudding is in the eating."
D. C. Restorer Circuit

By ROBERT L. DONALDSON

A LARGE number of present and would-be television set owners have declared that “it tires my eyes to watch television for more than a short time,” and “the pictures seem to jump.” Their usual solution to both problems has been: “Keep the lights on in the room and it won’t tire the eyes.”

In running down these complaints, it became evident that what was really complained of was the usual intermittent change in average brightness of the screen from second to second and scene to scene. A few checks made with simple light-measuring equipment disclosed that, during an average telecast, the total light output from the screen varied from second to second so much so that the ratio of total light output between extremely light scenes and dark scenes was greater than 20 to 1. To make matters worse, the light output dropped to zero when no scene was being presented and during tuning.

Fig. 1—Combination d.c. restorer and sync clipper as used in many TV sets.

This extreme variation in light intensity forces the iris of the viewer’s eye to dilate and contract continually to counteract these variations. That is the source of the eye strain. Any doctor will verify the fact that flickering or intermittent light tires the eye very rapidly.

The common practice of leaving other lights on in the room is an attempt to wash out such extreme variations by having a constant, high light level entering the eye from other sources.

It results in less eye strain, but excessive ambient light makes the picture look weak and washed out. The blacks in the picture lose all detail, because the black presented cannot be any blacker than an unlighted screen.

Motion-picture practice has always been to darken the theater to allow a good dynamic light range on the screen, but to control the average brightness of the scenes carefully so that a fairly constant total amount of light is presented to the eyes of the audience. Movie makers and exhibitors have found out through the years that it pays to be kind to their viewers’ eyes. Television receiver practice should take a leaf out of the book of movie experience.

The entire blame for this unsatisfactory performance can be laid to the d.c.-restorer circuit used in almost all television sets. The author proposes certain circuit changes to correct the condition.

Usual circuit

Many sets use the combination d.c. restorer and sync clipper diagrammed in Fig. 1. Assuming no video input to the picture-tube grid, the grid is biased negative by the brightness control to such a point that its screen current is just cut off. This bias value may be regarded as black. Since the video output from the screen varies from second to second, the average light intensity of the scene will change. The sync pulses are negative in relation to the steady-state voltage, and thus are blacker than black. This relation of white, black, and blacker-than-black can be seen from the waveform of one line.

Now consider the action of the diode in the circuit. This is placed in shunt with the video load, and in series with the picture-tube grid return. It rectifies the video output and develops on its cathode a positive d.c. voltage derived from the a.c. average value of the video signal. The d.c. voltage is averaged out by the storage action of the capacitor and is applied in series with the fixed bias to the grid of the picture tube. This raises the picture-tube grid to such a positive point that the variations in voltage of the video signal are depicted on the screen as various shades of gray and the scanning retrace lines become visible.

To counteract this, the fixed bias is reset more negative by the brightness control. Thus, unless a strong video signal is present, the screen is completely black. A strong video signal containing much white is pushed even more positive by the average d.c. voltage, making the whites whiter, the blacks grayer, and the sync pulses visible. A weak video signal containing mostly dark grays and blacks cannot develop much d.c. voltage, so the whole scene is reproduced much darker than average. The sync pulses are not seen in the latter condition, but neither is there any detail visible in any darker portions of the picture, because both of these values are now more negative than the threshold black point. To present a more normal-looking picture, the operator must readjust the brightness control. That means more attention and knob twiddling during the telecast.

Time constants of the circuit are important. Time constant is the time that the value of the capacitor and voltage drop to 1/e or 37% of their initial value. In the case of the 2-microseconds capacitor, the voltage drop will be 37% of the original voltage to 1/e or 37% of its original value. It is for this reason that the memory constant of the capacitor is important. This value of time is used to set the brightness control of the set. The time constant of the circuit is the time that the voltage drop to 1/e or 37% of its original value.

The modification

The d.c.-restorer circuit proposed by the author, diagrammed in Fig. 2, has only one basic difference from the circuit described above. The diode is reversed in polarity so that a negative
Television

AUGUST, 1950

Novel High-Voltage Test Probe

By WALTER H. BUCHSBAUM

A NODE voltages of 8 to 15 kilovolts on modern television picture tubes present somewhat of a problem to the service technician when he has to measure them. The usual method is to use a high-voltage probe with a standard voltmeter or estimate the voltage by drawing an arc.

Fig. 1 shows a novel device which has the features of both these methods. When the two round balls receive opposite charges and are close enough together, a discharge takes place through the air separating them. The critical distance for discharge depends on the potential. For given atmospheric conditions, the relation between voltage and distance is linear and the gap indicates the voltage.

To measure high voltage, the two balls are first separated so that the micrometer-type scale reads higher than the rated voltage to be measured. The tip of the device is applied to the voltage under test and the tip is screwed in. When the separation between the two balls reaches the critical discharge distance, a bluish purple glow becomes visible and a hissing noise is heard. The correct voltage can then be read directly on the scale.

The vertical scale on the glass tube is calibrated in steps of 5,000 volts, and one complete revolution of the cap is equivalent to 6,000 volts. The cap is further divided into steps of 250 volts.

The device is sold under the name Seg-Hy-Volter. It is small, inexpensive, and easy to use. The technician handy with tools can make one himself.

Fig. 2 shows the disassembled unit. The two balls are machined brass and are each ½ inch in diameter. The disk which holds the upper ball bolt is also brass. The metal parts are chrome finished, and the balls have an extra smooth surface.

If one full turn of the top cap is to be equivalent to 5,000 volts, the screw pitch must be 16.7 turns per inch.

Fig. 1—Closeup of the novel Hy-Volter.

Fig. 2—A few simple parts are all that is needed for the high-voltage probe.
Satellite Television Station

KC2XAK, the world’s first ultra-high-frequency television station operating on regular program schedules, is now on the air in the 480-920-mc u.h.f. television band. The station is a satellite, operated by the National Broadcasting Company, and rebroadcasts the programs of WNBT, New York City, from a hill between Bridgeport and Stratford, Connecticut. The new station started operation at full power (1 kilowatt) early in the year, and at the beginning of June was operating on a regular five-day weekly schedule from 9 a.m. to WNBT’s signoff.

The tower shown on the cover is 250 feet high to the top of the “stovepipe” mast. Both the antennas for receiving the signals from New York, 55 miles distant, and for retransmitting them at u.h.f., are on it. The transmitter is installed in the building at the foot of the tower.

Originally the satellite station picked up the signals from WNBT on its regular frequency (channel 4) and rebroadcast them. This was not entirely satisfactory, and a relay on 2,000 mc was arranged.

Signals beamed on 2,000 mc from the Empire State Building are received on the small parabola on KC2XAK’s 210-foot tower. (The larger parabola was used to receive the channel 4 transmissions.) The signals are fed into the u.h.f. transmitter and radiated on channel 24 (529-535 mc) from the 40-foot slot antenna at the top of the tower.

The radiating elements of the antenna are slots cut in the 10-inch diameter tube. They are a half wavelength long and are separated a half wavelength from each other. A special coaxial feed system provides power and phase balance between the upper and lower halves of the antenna array. The resulting pattern is perfectly circular with practically all the energy radiated horizontally.

The transmitter itself—shown in the insert in the cover—is believed to be the first commercial-type u.h.f. transmitter. TV transmitters have been operated in the u.h.f. band before, but they were special rigs which seldom long outlived the test or demonstration for which they were thrown together. In contrast, this new job is a prototype for a commercial model which can be duplicated on order for any u.h.f. station when such stations can go on the air.

Designated the TTU-1A, the transmitter contains a number of units identical with those of RCA’s v.h.f. transmitter TT-500B, plus a tripler stage to increase the frequency to the required 530 mc, a 1-kilowatt power amplifier stage, and a video amplifier stage which grid-modulates the power amplifier. Identical tripplers and final amplifiers are used for the sound and picture carriers. Each uses eight 4X150A u.h.f. power tetrodes in special ring-type resonant cavity circuits.

The audience is small but enthusiastic. There are about 100 receivers scattered in carefully selected locations throughout the area, from 1 to 23 miles away from the station, with a few at much greater distances. Approximately half these receivers are specially constructed, the other half are u.h.f. converters attached to standard television receivers. No notable difference has been recorded in the performance of the two types, though of course the self-contained special receiver is more convenient from the set-owner’s point of view. One of the results of the KC2XAK project has been to show that converters are practical.

While the transmission range is not yet fully determined (determination of u.h.f. transmission ranges was one of the chief reasons for the project, and is still under study) the 500-microwatt contour will lie at a radius of approximately 8 to 8 miles from the station. Satisfactory pictures have been received at considerably greater distances than that, of course. Within a radius of 10 miles, 84% of the viewers report good or excellent results and, within a radius of 23 miles, the same results are obtained at 65% of the receivers.

Satellite station operation has been considered by many the answer to television’s economic problems. By greatly increasing the television audience it may solve the problems of expensive television programs for the big networks while at the same time making television possible for the small community which could not support an independent station. The country’s first regular u.h.f. station will answer many questions about satellite station operation, as well as a number of questions about the operation of independent u.h.f. television transmitters.

The results of KC2XAK’s transmission tests may even help to hasten the end of the present freeze on TV station allocations. Two important problems now facing the FCC are providing TV outlets to as many communities as possible and eliminating co-channel and adjacent-channel interference. If this station proves successful, the FCC may find the satellite-type station one of the answers to its problems.
Fundamentals of Radio Servicing

Part XVIII—Demodulating the R.F.

By JOHN T. FRYE

AUGUST, 1950

For a station in the middle of the broadcast band, there would be about 1,000,000 per second. Fig. 1-b shows a complex audio-frequency wave consisting of various audio frequencies of different intensities. At 1-c we show the carrier wave “envelope” that results when the audio wave modulates that carrier. Note that when the audio signal is zero, our carrier is of normal amplitude. As the audio signal increases in a positive direction, the carrier amplitude increases just as does the amount of water issuing from the hose when you relax your pressure on the kink. As the modulating voltage increases in a negative direction, the amplitude of the carrier is greatly reduced, just as the water flow is reduced to a trickle when you clamp down on the hose.

This modulated carrier is the signal our radio set receives. You will notice that the frequency of the actual carrier remains unchanged. Only its amplitude changes. Under modulation the carrier develops a series of bulges and notches like the neck of an ostrich swallowing different-sized oranges. If the modulating voltage swings are strong, the bulges and accompanying dips are large; if this voltage is weak, the variations are barely perceptible. Also, if the modulating frequency is high, many of these bulges in amplitude appear in a short space of time; if low, only a few appear.

There is one marked difference between the audio voltage and the modulated carrier envelope produced by this voltage: the audio voltage appears first on one side of the zero line and then on the other so that its force is exerted first in one direction and then in the opposite; but our modulated carrier expands and contracts simultaneously and similarly both above and below the center line (because a half-cycle of a.f. modulates many cycles of r.f.). That means that the equal and opposite voltages tend to cancel each other. The effect is about the same as if you held a strip of rubber in the center while someone pulled on the two ends with equal pressure. As far as you
could tell by the feel at the point you were holding, nothing would be happening.

Demodulating the signal

If we are to recover our audio voltage from this modulated carrier, we
must find a way of changing this “two-way stretch” back into a one-way stretch. The method of doing this is simple: split the carrier in two right down the center and use half of it. Since the top and bottom halves are identical, either will contain all of the information needed about the original modulating voltage.

This “splitting” is not hard if we remember that the top-and-bottom graph is just an engineer’s way of saying that the current in the circuit flows only half the time and the other way the other half. So what must be done is to rectify the r.f. current.

That leaves us with a pulsating direct current; actually one with two sets of pulsations—those of the r.f. and those of the a.f. (See Part XIII, March, 1950.)

We filter this pulsating current just enough to smooth it out but r.f. peaks but not enough to bother the a.f. changes caused by the modulating voltage. (Look at the March article again.) Since the lowest radio frequency used in broadcasting is over 500,000 cycles per second and the highest audio frequency that can be heard somewhere around 15,000 cycles, that is easy.

Then we are left with a current whose pulsations reproduce those of the current which originally modulated the r.f. signal.

Fig. 2 is a diagram of one of the earliest practical methods of detection. The crystal detector (indicated by XTAL) consists of a piece of galena (a common ore of lead) contacted by the sharp point of a phosphor-bronze wire called a catwhisker. It is a peculiarity of this tiny contact that it will allow current to flow easily from the catwhisker to the galena but prevents a very high resistance to the flow of current from the galena to the catwhisker.

In our diagram the radio signal is picked up by the antenna and inductively transferred to the tuned circuit L1-C1. The alternating radio-frequency voltage across this circuit is applied to the headphones through the crystal detector. Remembering that the crystal will allow current to flow in only one direction, we can easily see how half of our radio signal will be cut off as shown in Fig. 3.

At Fig. 3-a is the modulated carrier. This envelope is altered as shown in Fig. 3-b by the crystal’s refusal to pass pulses of current going in one direction. Only the pulses going up from the zero line are allowed to pass through. These pulses of current work together to charge capacitor C2 with a varying voltage as indicated by the dotted line of Fig. 3-b. This capacitor discharges this varying voltage through the headphones.

Since the charge of C2 is made up of the average of the voltage pulses contributed by the expanding and contracting carrier, the voltage on this capacitor looks like Fig. 3-c. This voltage wave is a faithful reproduction of the original audio voltage that modulated the carrier. C2 and the coils in the headphone make up our r.f. filter.

Catwhisker crystal detectors are seldom used these days. They are too hard to adjust and to keep in adjustment. The efficiency of the detector action is critically dependent upon the pressure of the catwhisker on the crystal and also varies greatly from one spot on the crystal surface to another. Furthermore, the crystal detector does not handle either extremely weak or extremely strong signals very well. It has been superseded by the vacuum-tube diode type of detector diagrammed in Fig. 4.

The diode detector

Although this circuit looks much more complicated, it functions just as did the crystal detector. The signal voltage appears across the tuned circuit L-C just as before, and the alternating r.f. voltage appears between the plate and cathode of the vacuum tube. While the plate is positive, as it is every half-cycle of r.f. voltage, electrons flow from the cathode to the plate. The more positive the plate swings, the more electrons it attracts. This means that when the carrier is expanding and the positive pulses of r.f. energy are increasing in amplitude, an increasing number of electrons go to the plate when the carrier is decreasing and each pulse of voltage is weaker than the one just before it, the total electron flow decreases. The electrons actually flow in spurts as dictated by the evenly spaced positive pulses of r.f. voltage, but these spurts are blended together by the filter C1-R1-C2 (see next paragraph) to produce a continuous flow that follows faithfully the variations in the amplitude of the modulated carrier.

The electrons return to the cathode through L and resistors R1 and R2. Resistor R1, together with capacitors C1 and C2, act as a filter to smooth out the “spurtly” nature of the current delivered from the diode plate. The capacitors are large enough to offer a very low resistance bypassing action on the high-frequency variations resulting from the rectification of the individual r.f. pulses, but not large enough to have any material effect on the comparatively low-frequency variations resulting from demodulating the carrier envelope.

When this current, varying in accordance with the original audio voltage, flows through R2, it causes a voltage to appear across the resistor that is an exact reproduction of the original audio signal. Because of the resistor C3, this voltage also appears across R3, the volume control. The slider on this control permits any fraction of the total audio voltage to be selected and passed along to the audio amplifier.

The grid leak detector

While the diode detector is much like the crystal detector in its “check-valve” action, it was not the immediate successor of the catwhisker to the grid leak detector actually supplanted the galena and catwhisker combination. Fig. 5 is a diagram of a triode grid leak detector.

If you think of the grid as taking the place of the diode plate, you can see that the input portion of this circuit is really quite similar to that of our diode detector. The resistor across which the audio voltage develops has been moved to the grid leg of the input coil, and a capacitor is placed across this resistor so that the r.f. voltage can still be applied between the grid and the cathode; but we still have a condition in which the electrons will flow to the grid, just as they did to the diode plate, when that element is made positive by the half-cycle swings of r.f. voltage. The electron current to the grid, and consequently the voltage across the resistor Rg, the “grid leak”—varies in accordance with the amplitude of the modulated carrier.

The resistance value of the grid leak is sufficiently high (usually 2 to 10 megohms) so that the electrons attracted to the grid during the positive half-cycles of the r.f. carrier do not...

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**Note:** The text continues on the next page. The diagram for Fig. 4 and Fig. 5 are not transcribed here. The full text can be found in the source material provided by www.americanradiohistory.com.
have time to leak off during the negative half-cycles; but it is at the same time small enough so that the negative voltage on the grid can go up and down in unison with the changes in amplitude of the modulated carrier. This leaves only the audio signal on the grid.

Now stop thinking of our grid as a diode plate and look upon the tube as an ordinary amplifier. On the grid we have a varying voltage that exactly reproduces the audio voltage originally impressed on the carrier. The triode amplifies this audio voltage in the normal manner, and we have an amplified audio signal of transformer T that leads into the audio amplifier. C2 and C3, together with the radio-frequency choke RFC, remove any trace of the radio-frequency ripple that might get through to the plate of the triode.

The grid leak detector is very sensitive because it combines the functions of detection and amplification. A pentode can be used in place of the triode to still further increase sensitivity. One big drawback of the grid leak detector is that, if it is made very sensitive to weak signals by the proper selection of values for the grid leak and the grid capacitor, it will distort strong signals badly.

The power detector

The “power detector” of Fig. 6 will handle very strong signals without distortion. It looks like and is an ordinary amplifier circuit except that the bias is set so the tube is working very near the plate current cutoff portion of the curve, as is shown in Fig. 7. Bias batteries are generally not used in practical circuits. For description of types of bias see Vacuum-tube Grid Bias by H. B. Davis in the May issue of Radio-Electronics.

Since the plate current is practically zero with no signal on the grid, the negative swings of the r.f. voltage impressed on the grid have no effect on the plate current (it can’t go below zero) but the positive half-cycles cause pulses of r.f. current to flow in the plate circuit as shown. These positive pulses of current increase or decrease in step with the amplitude of the voltage pulses on the grid. The charging and discharging action of the plate circuit results in a combination of rectification (only the positive half-cycles of voltage affect the plate current) plus amplification. Since the grid bias is very negative, extremely large signals can be handled without overloading the detector.

This completes the types of detectors that have been used to receive amplitude-modulated radio signals. Each rightfully has enjoyed its day in the sun. The crystal detector had one decided advantage: it was entirely self-sufficient. No filament or plate batteries were needed to help it do its job.

The grid leak detector was much more sensitive—in fact, it is still the most sensitive of all detectors—but it could reproduce faithfully the audio portion only of weak signals. When the signal grew strong enough, the sensitive grid leak detector was over-loaded and caused distortion in the signal. The power detector can handle much larger signals without introducing distortion, and it is between the diode and the grid leak detector as far as sensitivity is concerned. What is more, since it causes no rectified current to flow through the tuned input circuit, it does not “load” this circuit and so reduce its sharpness of tuning. All other detectors mentioned exhibit this effect.

With the introduction of the pentode amplifier tube sensitivity in a detector became of little importance. R.f. amplifiers could easily build up a weak signal to almost any value before it was presented to the detector. That is why the diode, in spite of its low sensitivity, became the most popular detector. It is unbeatable in its ability to handle any strength of signal without introducing distortion. What is more, it can do its job of detecting and, at the same time, produce other useful byproducts, such as automatic volume control voltage which will be discussed in the next chapter.

Quite recently a new form of crystal detector has become popular. This is the tiny germanium crystal diode seen in the photograph. The catwhisker in this unit is sealed in permanent optimum adjustment, and it will handle much more current than would the old galena job. That frees it of the two main drawbacks of its ancestor.

Thus, once more we see demonstrated that amazing and oft-repeated cycle wherein one generation of radio engineers banishes a particular gadget to the electronic attic, and then the next generation hauls the item out of the garret, dusts it off, makes a few changes, and puts it back into use until something better comes along!

CHECK THAT ADVERTISING!

T’S time to start making out that advertising budget on a scientific, businesslike basis, using actual results as a guidepost.

With business in the radio repair field due to become more competitive in the next months, advertising will play an increasingly important part. Naturally, any radio repairman wants to spend his hard-earned dollar for advertising that will do the most good.

Sad to say, however, the majority of businessmen in the radio maintenance field shoot in the dark when it comes to advertising. They dote out those precious dollars in a hit or miss fashion, and more often than not, they miss! In other words, they have no plan. A good advertising program needs a workable, business-producing plan, and it’s simple to work out an effective program.

The thing to do is check your present customers. Conduct a survey of your own right in your shop for the next week or two to help you find just where your advertising dollars bring the most returns. Your customers will be glad to help you—and they’ll be able to tell just how to budget to do the most good.

Ask your customers when they come in just how it happened that they visited your radio repair shop in the first place.

Better yet, prepare some small mimeographed sheets explaining that you are trying to test the pullline power of your present advertising program, and hand one to each customer over a period of a week or 10 days.

The card could read something like this:

“I brought my radio for repair to the............store because I noticed:

Newspaper advertising ( ).
Radio program ( ).
Window display ( ).
Friend recommended ( ).
Billboard sign ( ).
Ad in phone book ( ).
Handbill ( ).
Business card ( ).
Roadside sign ( ).
Neon sign in front ( ).

Just happened to be passing ( ).
Other reasons (leave a space big enough for them).

You can probably think of some more categories, but the above is the general idea.

National companies, who spend millions of dollars annually for advertising, send crews to all parts of the nation to check their advertising program. You can do your own checking right in your own shop—and it won’t cost a cent. Just a little added effort is all that is needed.

As you spot check your customers for a few days and compile the results, you’ll have a better than fair idea where your advertising dollar can do the most good.—Don Valentine

AUGUST, 1950

Fig. 6—This detector looks like an amplifier but it is biased at cutoff.

Fig. 7—These curves show how the incoming r.f. signal is both detected and amplified by the action of the power detector circuit.
Radio Set and Service Review

A new 3-inch oscilloscope designed for TV servicing

By ROBERT F. SCOTT

The type WO-57A 3-inch television service oscilloscope is the latest addition to RCA's line of specialized TV test equipment. Although its design makes it particularly suitable for TV servicing, it is versatile enough to permit its use in all phases of radio servicing. Its size, 13½ inches wide, 10 inches high, and 7½ inches deep; and weight, 18 pounds, make it a truly portable instrument. Its blue hammetrex case and chrome panel match the WR-39A television calibrator, WR-59A sweep generator, and similar equipment in the RCA test equipment line for TV servicing.

Features which are particularly useful to the TV service technician are:

A two-stage direct-coupled, vertical amplifier having response flat within plus or minus 1.2 db from d.c. to 500 ke, 3.5 db at 1 mc, and useful beyond 2 mc.

Linear sweep generator having fixed 30- and 7,875-cycle positions for viewing sync pulses and sweep oscillator waveforms. These positions make it unnecessary for the operator to adjust coarse and fine sweep controls each time he goes from a horizontal to vertical deflection circuits while servicing a TV receiver. The oscillator will sync equally well on positive or negative pulses being fed into the vertical amplifier and on externally applied sync signals. Internal 60-cycle phase-controlled sweep.

Separate vertical input terminals, one for a.c. and the other for d.c. Combined a.c. and d.c. voltages can be applied to the input terminals.

A calibrated screen or grid marked in tenths of an inch for making voltage measurements.

A 1-volt peak-to-peak calibrating voltage source.

A voltage-calibrated and frequency-compensated, step-type vertical attenuator.

The circuit

The complete circuit of the WR-57A is shown in the large schematic. The signals to be observed or measured are fed into the vertical amplifier through a.c. or d.c. input jacks and the frequency-compensated attenuator, V range. Fig. 1 is a diagram of a conventional potentiometer-type gain or volume control. Capacitors C1 and C2 represent the stray capacitances between the arm and the grounded and hot ends of the control. The values of these capacitances are constant and fairly low so that they act as a capacitive voltage divider in parallel with the variable-resistance control. Therefore, the high-frequency components of the input signal will be handled by the fixed capacitive divider and the low-frequency by the variable resistive voltage divider. Consequently, the attenuation of the high- and low-frequency components of the input signal will not be equal and serious frequency distortion will develop in the gain control.

A simplified version of a compensated attenuator is shown in Fig. 2. Here the resistive elements are fixed. Stray capacitances C1 and C2 are shunted by fixed capacitors C3 and C4 which are much larger. The values of the resistors and fixed capacitors are selected to give the desired attenuation with a linear voltage distribution over a wide frequency range. C3 is usually made variable so the network can be adjusted for different values of stray capacitance in otherwise identical setups.

The attenuator feeds into a push-pull, direct-coupled amplifier which is rather unusual in its design. The input signal is fed into the control grid of V1. The control grid of V2 is grounded. At low frequencies, the 1800-µf capacitor C12, between the cathodes of V1 and V2, is effectively an open circuit and no signal appears in any part of V2. This tube serves only to balance the current and voltage distribution in the circuit. The signal is amplified by V1 and direct-coupled to the control grid of V3. The grid of V4 is effectively grounded for a.c. so this tube operates as a grounded-grid amplifier being fed from the cathode resistor R25 which it shares with V3. V3 appears as a cathode-follower when viewed from V4, and V4 looks like a grounded-grid amplifier to V3.

At high frequencies, C12 has a low impedance and effectively couples the cathodes of V1 and V2 together. Its control grid being grounded, and having a high-frequency signal applied to its cathode, V2 operates as a grounded-grid amplifier and sees V1 as a cathode-follower. Thus, phase inversion takes place in the first stage at high frequencies and in the second at low frequencies. Because V2 operates only at high frequencies, the high-frequency gain is twice that of low frequencies. C15 compensates for high-frequency losses caused by stray capacitance. The vertical deflection plates of the 3MP1 C-R tube are direct-coupled to the plates of V3 and V4.

---

Fig. 1. left—Conventional gain control.
Fig. 2—Compensated attenuator circuit.
The deflection sensitivity of the vertical amplifier in r.m.s. volts per inch is .03 with the direct cable and .3 with the low-capacitance WG-214 probe. Its input resistance and capacitance are 1 megohm shunted by 90 µf with the direct cable, and 1 megohm shunted by 14 µf with the low-capacitance probe.

The horizontal deflection circuit is simpler than the vertical. It has an uncompensated potentiometer - type gain control (H GAIN) and a single 12AX7 amplifier and phase inverter. A part of the signal from the amplifier is fed to the grid of the phase inverter section through an R-C network. The deflection sensitivity of the horizontal amplifier is 0.6 volt r.m.s. per inch. Its input resistance and capacitance is 1 megohm shunted by 33 µf.

The sweep generator is a Potter-type cathode-coupled multivibrator using a 6AU8. Its frequency is adjustable from 15 to 30,000 cycles in five ranges. A SWEEP VER (vernier) R38B varies the frequency of the oscillator within the ranges set by the SWEEP RANGE switch. Two additional positions on the switch provide fixed 30- and 7,875-cycle sweeps for observing sync and blanking pulses and oscillator waveforms in the vertical and horizontal circuits of TV receivers. Screwdriver-adjust controls are provided on the back panel to permit setting the fixed sweeps to their proper frequencies.

The power supply is similar to those found in most scopes of this size. A full-wave rectifier supplies B-voltage for the amplifiers, oscillator, and second anode of the 3MP1. Approximately 500 volts negative is developed by the half-wave rectifier and applied directly to the control grid and to the cathode and anode No. 1 through a voltage divider.

Controls

Intensity, centering, gain, sweep, phase, focus, deflection, and calibrating controls are on the front panel. Controls for setting the TV horizontal and vertical sweep frequencies, vertical balance, second-anode voltage, astigmatism are on the rear of the chassis. The first three can be adjusted with screwdrivers through holes in the rear cover. The cover must be removed to reach the others.

The H and V centering controls are variable resistors in the cathodes of their respective amplifiers. Moving these controls causes the bias on the push-pull tubes to vary in opposite directions, thus causing the plate voltages and the position of the beam to move along its horizontal or vertical axis.

The DEFLECTION switch reverses the direction of the trace so it is possible to make the trace being observed have the same time-versus-frequency direction as the set manufacturers' service note illustrations. This switch reverses the connection between the horizontal deflection plates and the horizontal amplifier. The sweep direction is conven-
Ayrton Shunt Protects Meter Movement

Many technicians prefer to build their own multimeters. Milliammeters are often designed around a 200- or 500-μA movement with individual shunts calculated from the formula

\[ S = \frac{M}{N-I} \]

where \( M \) is the meter resistance, \( S \) the shunt resistor, and \( N \) the meter multiplier. When several shunts are used, the circuit looks like Fig. 1.

Fig. 1, left—A shunt for each range. Fig. 2, right—The Ayrton or ring shunt.

The Ayrton, or ring shunt, is not as widely used by home constructors, although it is generally found in lab and industrial equipment. It consists of a resistor across the meter (Fig. 2) with a switch to tap off a portion through which the total current flows. Sometimes the taps are brought out as terminals or binding posts, and the switch is eliminated.

![Ring shunt with typical values](image)

One of the advantages of an Ayrton shunt is greater meter safety. If the switch in Fig. 1 should become defective, the full current is apt to flow through the meter with probable damage to it. This cannot happen in Fig. 2.

If a contact should develop an open or a high-resistance, no current flows through the meter. Either a shorting or a nonshorting switch may be used here without danger.

Another advantage of the Ayrton shunt is the constant damping across the meter. Most technicians know that deflection becomes sluggish when the meter is shunted by a low resistance. This happens in Fig. 1 when \( N \) is large. The total resistance across the meter in Fig. 2 is constant, regardless of the range chosen.

The tapped sections of Fig. 2 are easily calculated from the formula

\[ S = \frac{M + T}{N} \]

Here \( T \) is the total shunt and \( S \) is the resistance between the common terminal and the desired tap. Another advantage is now evident. Usually \( N \) is 10, 50, 100, etc. Therefore in Fig. 1 we must divide the meter resistance by 9, 49, 99, respectively. The formula just given generally gives more convenient values. Assume for example that shunts are to be designed for a 200-μA meter with internal resistance of 500 ohms. The table below gives the correct values for both types of shunt.

<table>
<thead>
<tr>
<th>( N )</th>
<th>( I_{\text{max}} ) (μA)</th>
<th>( S ) (Fig. 1)</th>
<th>( S ) (Fig. 2)</th>
</tr>
</thead>
<tbody>
<tr>
<td>2</td>
<td>400</td>
<td>500</td>
<td>500</td>
</tr>
<tr>
<td>10</td>
<td>2 ma</td>
<td>500/9</td>
<td>100</td>
</tr>
<tr>
<td>25</td>
<td>5 ma</td>
<td>500/24</td>
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<tr>
<td>100</td>
<td>20 ma</td>
<td>500/99</td>
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<tr>
<td>500</td>
<td>100 ma</td>
<td>500/499</td>
<td>2</td>
</tr>
<tr>
<td>1000</td>
<td>200 ma</td>
<td>500/999</td>
<td>1</td>
</tr>
</tbody>
</table>

Note that the Ayrton shunt values are inversely proportional to \( N \) and need not be recalculated each time from the formula.

The Ayrton shunt circuit appears in Fig. 3—1. Queen

Sweeping across the tube. The Astigmatism Adjust control varies the bias on the 12AX7, thereby changing its plate voltage and the voltage on the deflection plates.

The oscillator can be synchronized with negative or positive pulses on the incoming signal or on external sync pulses. R38B is the sweep vernier. The C-R tube can be driven directly from an internal 60-cycle source or from a signal being applied to the horizontal amplifier. When the beam is driven from the 60-cycle source, R38A is the PHASE Adj control which permits superimposing the forward and backward traces during circuit alignment. Characteristics of the horizontal deflection circuit with the exception of frequency and width are controlled by the SYNCH SEL control. Negative or positive sync pulses are taken from the outputs of V4 or V3, respectively.

The vert. CAL (vertical calibration) control is a vernier gain control for the vertical amplifier. Its purpose is to vary the load impedance of V1 and V2 without disturbing the d.c. balance of the circuit. If it is adjusted for exactly half of peak-to-peak deflection and the direct probe is touched to the AC CAL terminal and the vertical range switch is on 1.0, input voltages can be read directly. In this case, 1 volt input will produce a 1-inch deflection.

The vertical attenuator and the calibrated scale can be used to measure other voltages. The vertical attenuator (V RANGE) is calibrated in steps of .01, 0.3, 1.0, 3.0, 10, and 30. A.c. or d.c., and combined a.c. and d.c. voltages can be read on the scale. The input terminal has a 0.22-μF capacitor between it and the attenuator, the d.c. input terminal has not.

The INTENSITY and FOCUS controls vary the brightness and sharpness of the trace by varying the potential between the cathode and control grid and between the first anode (pin 3) and the cathode, respectively. The vertical balance (V BAL) control is a variable resistor common to the cathodes of V1 and V2 and connected between the individual cathode resistors R14 and R16. It adjusts the linearity of the vertical input stage so the trace will not move up or down as the vertical calibrator is rotated.

The scope which we tested was just off the production line and had not been pre-aged, therefore it was necessary to let it run for approximately 32 hours before it would stabilize itself within a 15 or 20-minute warmup period. The initial drift does not greatly affect the performance of the scope for most servicing applications; however, there was some nonlinearity in the vertical amplifier. Voltage measurements cannot be made until the scope has reached operating temperature. Disregarding the inconvenience of the long warmup during the first operation, the unusual features of this will make it a useful and treasured addition to any radio-TV service bench.
Midget Signal Generator is Hum Modulated

An easy-to-build 1-tube generator that covers the needs of broadcast receivers

By LYMAN E. GREENLEE

This simple hum-modulated signal generator can be thrown together in a couple of hours and, if carefully calibrated, will take the place of an expensive piece of test equipment. It is particularly useful in locating stations when setting up push-button controls on new radios. Fig. 1 is the circuit diagram.

A 50L6 tube is used as a triode in a conventional Hartley circuit. The oscillator coil may be wound by hand on a 1-inch cardboard or Bakelite tube with No. 30 enameled wire. Wind 100 turns and bring out a tap for the cathode; then wind 40 more turns and finish with a coat of coil dope to hold the winding in place. The assembly is virtually self-explanatory from the photographs and wiring diagram. The two-gang t.r.f. capacitor (the two sections in parallel) can be salvaged from a junked midget radio. Frequency coverage is from about 400 to 1,800 or 2,000 kc which is adequate for servicing household radios covering the broadcast band and having a 455-ke i.f. This includes almost everything made in recent years except FM and short-wave sets.

The assembly shown in the photos was made in an old shield can, but the parts can be mounted on a board if no case is available. No attenuator is provided as there is sufficient radiation to couple to the radio under test and the input can be controlled by moving the oscillator away from the set. Since the signal generator is used mainly for station finding or testing, an accurate attenuator is not needed.

To avoid a hot chassis, the tuning capacitor is insulated from the shield can and the capacitor shaft is connected to the tuning knob with an insulating coupler. For best stability, the can itself should be grounded to earth.

The 50L6 tube is operated at a reduced filament voltage of approximately 30 and this is also used as the plate voltage. This makes the oscillator stable with adequate output for service work, but it may be a little slow in warming up. The signal is hum-modulated. An unmodulated signal could be obtained by using d.c. for the plate supply, either from a small B-battery or a selenium rectifier in series with the power lines.

The instrument may be calibrated by using a radio set having a dial reading in kilocycles. Tune in a station and listen for the announcement to identify its frequency (a list of broadcast stations and their operating frequencies will be a help) then tune the signal calibrator to the station and mark the frequency on a paper scale. Repeat the process until a sufficient number of points have been located. It is usually convenient to locate the i.f.'s of 455, 465, etc., and the r.f.'s of 600, 700, 800, to 1700 or 1800 ke. Points between the 100-ke calibrations can be located or estimated and a reasonably accurate dial prepared. The paper dial can be inked in and covered with a piece of celluloid to prevent the numbers from rubbing off. The i.f.'s can be located by listening for their harmonics at 900, 910, 950 kc. Divide the dial reading on the radio set by 2 to get the fundamental or i.f. frequency.

To set up push-buttons, tune the signal calibrator to the frequency of the station wanted on the pushbutton and it will be easy to identify the oscillator signal when the trimmers are adjusted. This method saves a lot of time and requires no direct connection to the radio. Plug both the radio and signal generator into the same outlet. Maximum range is about 30 feet under normal conditions, which is short enough to prevent interference with other radio sets and is within FCC regulations. Do not connect the oscillator to an antenna of any kind as radiation will then be excessive.

**MATERIALS FOR SIGNAL CALIBRATOR**

- Resistors: 1-100,000 ohm, 1/2 watt; 1-600 ohm, 10 watt
- Capacitors: 1-0.0005 µf mica; 1-0.005, 1-0.05, 1-0.1 µf, 600 volt, paper; 1-0.00365 µf, 2-gang variable; 1-50 pf trimmer.
- Miscellaneous: 1-1-inch diameter coil form; 1-40 mh r.f. choke; tube, socket, hookup wire.

The small metal case allows ample room for all of the parts with no crowding.

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*Fig. 1—Hookup of the 1-tube oscillator.*

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**AUGUST, 1950**
An Electronic Therapy Machine

A simple and dependable device that will produce soothing facial massages

By IRVING GOTTLIEB

A front view of the electronic therapy machine showing controls on the panel.

Rear view with the top cover removed.

ELECTRICAL impulses are commonly used to produce involuntary muscular contraction. Many beneficial physiological results have been ascribed to such treatments although not all of these will survive authoritative medical scrutiny. In such afflictions as polio and some forms of arthritis, it is often desirable to induce activity in the affected muscles. Many beauty parlors now feature a so-called “electric facial massage.” This treatment may not be medical therapy in the accepted sense, but the quality of the apparatus should be on a par with that intended for more serious medical work.

In the electric facial, two electrodes are applied to the various nerve motor points on the face. An alternating potential, usually a low audio-frequency pulse of several to perhaps a dozen cycles per minute, is delivered to the electrodes by the machine. This is claimed to provide relief from tension and to induce relaxation in muscles which have been unduly contracted by the tempo of modern living.

The circuit of Fig. 1 was designed to eliminate some of the deficiencies found in existing equipment. In much apparatus presently being marketed, mechanical or thermomechanical means are used to pulsate the voltage. Failures in the switching mechanisms and the corrosion and pitting of the contact points reduce the reliability of such equipment. Trouble from this source is more frequent than tube failure in the cheaper machines. The pulsing can, of course, be done electronically. The author, after servicing many machines, is in favor of eliminating all mechanical motion to make the machines more reliable.

Another shortcoming of many such machines is the difficulty of applying two electrodes to the patient’s face and simultaneously adjusting the amplitude control on the panel to the proper level. This is a very important procedure. The correct amplitude is usually determined experimentally by advancing the amplitude control a bit at a time and alternately asking the patient if there is any feeling of discomfort. The operator also watches for the first visual evidence of stimulation.

This procedure is not entirely satisfactory. If the voltage is too high, the patient will get an unpleasant shock. Second, it requires a little time for the electrolytic resistance between skin and electrode to become stabilized. Therefore, when the operator must repeatedly put down the electrodes to readjust the amplitude control on the machine, considerable experimentation is sometimes necessary—to the discomfort and annoyance of the patient. This situation is aggravated because the neuromuscular system does not immediately decide upon the most pleasant amount of excitation.

Third, the sensitivity to electrical stimulus varies with different nerve motor points. What may be a tolerable or pleasant stimulation at one position of electrode application may be painful half an inch away. This makes it highly desirable for the operator to have means of reducing and increasing the intensity of the stimulus quickly and smoothly.

Furthermore, for an extensive range...
of stimulus intensity, pain results more from too rapid a change in intensity than from the level of the intensity itself. Consequently, the electrodes must not be applied or removed abruptly when the amplitude is above the threshold of initial response.

Because of these considerations, an auxiliary amplitude control was installed directly in the handle of the electrodes. The series variable resistor used for this purpose is mounted so the operator can manipulate it with her fingers as she applies the electrode to the patient's face. This amplitude control provides sufficient variation of electrode potential so the main control on the panel of the machine can be set permanently at some nominal position. The electrodes make contact with the skin through small felt pads about 5/8 inch in diameter. These pads are made conductive by soaking them in an electrolyte such as a solution of bicarbonate of soda.

The machine delivers either 500-cycle alternating current or direct current. In the terminology used in electrotherapy work, a periodic alternating potential is called a "contractive mode." Similarly, a direct potential is known as the "galvanic mode." Pulsing of the electrical stimulus is called "surging." This machine enables either of the modes to be surged or to be applied at a constant level. There are, therefore, four combinations of electrical stimulus available. These galvanic preparations are usually used to drive the ions of certain pharmaceutical preparations into the tissues.

The circuit is basically a blocked-grid audio oscillator followed by a class-A amplifier and a rectifier for the galvanic mode. The R-C combination R4-C3 is a feedback link which sustains the duration of the pulses or surges generated by the 6F6 blocked-grid oscillator. Without this feedback network, the pulses are short and very short in duration compared to the time between successive pulses. The feedback makes the surge periods approximately equal to the quiescent periods. This holds for the range of surging frequency provided, which is from 5 to 40 surges per second.

The control which varies surge frequency and the switch which selects either surge or constant-level operation is in the grid circuit of the 6F6 oscillator. R1 and C1 are the conventional grid leak and grid capacitor, respectively. R2 and C2 in the grid circuit vary the R-C time constant of this part of the circuit. R2 varies the frequency of the surges of 500-cycle oscillations developed by the 6F6 tube. The switch S1 determines whether the 6F6 will operate as a blocked-grid or constant-oscillation oscillator. When S1 is in the SURGE (open) position, the oscillations are pulsed at a frequency determined by R1-C1 and R2-C2.

When S1 is in the CONSTANT position (S1 closed), resistor R3 is shunted across the grid-leak network. Although the basic circuit is not changed, the blocked-grid oscillator will not pulse when the grid-leak resistance is too low because the reduced time constant of the grid circuit allows the charge on C1 to leak off fast enough that it cannot become sufficiently negative to block the grid. The circuit then acts as a constant-output oscillator.

When S1 is in the surge position, the panel-mounted neon bulb blinks at the surge rate. When S1 is in the CONSTANT position, this bulb glows with a steady light.

The four series-connected, 6-watt bulbs give the leading edge of the surges a gradual rise because of the change in resistance of the lamp filaments. When the output of the 6L6 amplifier is impressed across these lamps immediately after the blocked-grid oscillator has burst into oscillation, the cold resistance of the filaments is comparatively low and a large amount of the audio energy is dissipated. As the filament temperatures rise, their resistance increases and more power is delivered to the output terminals. Thus, even though the oscillator bursts into oscillation abruptly, the output of the amplifier becomes maximum at a much slower rate.

The feedback network R4-C3, not only prolongs the duration of the oscillation pulses, but also delays the decay of the pulse. As a result, both the rise and fall of the audio-frequency energy pulses are gradual enough to avoid the sensation of electric shock caused by too high a rate of change of the stimulus.

When S2 is in the GALVANIC position, the output of the 6L6 is rectified by the 2050 thyatron which functions as a half-wave rectifier. The output of the rectifier is filtered by a conventional capacitor input.

No metering provision is made for the contractive mode. Electrical stimulus of a nerve motor point causes muscular contraction if alternating current is used and the frequency is not too low nor too high. Direct current causes no such contraction if the rate of change of the stimulus is not too high, and burning of the skin tissue, nerve damage, and pain may occur before the operator is aware of it. Because of this, the milliammeter is required with direct-current treatments, so the correct milliampere "dose" prescribed for the particular treatment is given. With alternating current, the operator merely watches for muscular contraction. If this is obtained gradually and carefully, the muscles will be exercised rhythmically with a soothing sensation. There will be no feeling of electric shock.

The patient is isolated from the high-voltage d.c. by capacitor C4 and

Photo of the two electrodes. One has a built-in potentiometer for easy control. The 10,000-ohm resistor R5. This capacitor must be of such quality and voltage rating that the probability of di-electric failure is negligible. In this case an 800-volt transmitting type is used. Any other safety precautions which might occur to the experimenter should be used.

The laws concerning the commercial use of this kind of apparatus vary in different states. In most cases, the operator must show evidence of formal training in therapy or nursing. However, many of the beauty operator schools provide the necessary training and meet the legal qualifications of most states.

**MATERIALS FOR THERAPY MACHINE**

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<thead>
<tr>
<th>Resistors</th>
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<th>Notes</th>
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<tbody>
<tr>
<td>1-10,000</td>
<td>1-33,000</td>
<td>1-220,000</td>
</tr>
<tr>
<td>100,000</td>
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<table>
<thead>
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<td>1-470</td>
<td>1-1,000</td>
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<thead>
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<tr>
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<td>1-10</td>
<td>1-15</td>
</tr>
</tbody>
</table>

www.americanradiohistory.com
The phon meter, complete. Most constructors will use standard microphone plug.

The noise of whirring machinery and pounding punch presses can annoy factory workers, decreasing their efficiency. Traffic noises may produce unfortunate effects, and noises in many other locations may affect people and processes more or less profoundly.

Measurements of these noises are becoming more important every day as methods for increasing production efficiency and comfort advance. It might appear that the matter is simple—just set up a microphone and amplifier with an output meter. The difficulty is that the microphone and amplifier would function more or less linearly, while the human ear does not. The problem is to make the meter readings indicate as closely as possible, not the absolute level of the noise, but its apparent loudness as the ear hears it.

The response of the ear to changes in volume level is neither linear nor logarithmic, nor does it follow any other straight-line function. The closest approximation, however, is logarithmic, which is the reason for the use of the decibel as the measure of sound-level change. To complicate matters further, the ear does not respond in the same way to all frequencies. The graph of Fig. 1 presents the well known Fletcher-Munson curves, which show the required relative levels of sound at various frequencies for a sensation of equal loudness.

The unit of apparent loudness is the phon. It takes into account the ear's frequency response and the ear's change in frequency response with changes in intensity. It tells how many decibels a sound is above the threshold of hearing.

The level in phons of any sound is the same as the number of db above .0002 dynes per square centimeter of a 1,000-cycle tone which sounds exactly as loud to the average listener as the sound being measured. When the curves of Fig. 1 are being used, the level in phons of a measured sound pressure is the same as the label on the curve in which that pressure at that frequency can be found.

These curves represent the loudness level of pure tones and cannot be applied to sounds containing a spectrum of frequencies such as the noise level in a room. They do indicate the ear's response to different sound levels.

<table>
<thead>
<tr>
<th>TABLE 1</th>
</tr>
</thead>
<tbody>
<tr>
<td>TABLE OF PHONS</td>
</tr>
<tr>
<td>Threshold of audibility</td>
</tr>
<tr>
<td>Cat purring</td>
</tr>
<tr>
<td>Quiet countryside</td>
</tr>
<tr>
<td>Rainfall on roof</td>
</tr>
<tr>
<td>Quiet conversation</td>
</tr>
<tr>
<td>Small chamber orchestra</td>
</tr>
<tr>
<td>Loud speech</td>
</tr>
<tr>
<td>Symphony orchestra</td>
</tr>
<tr>
<td>Rivet hammer</td>
</tr>
<tr>
<td>Unmuffled aircraft engine</td>
</tr>
<tr>
<td>Threshold of pain</td>
</tr>
</tbody>
</table>

Table I lists a number of typical noise sources with the usual noise level in phons for each. The table should enable the newcomer to the phon to orient himself and gain a little familiarity with the orders of the numbers.

**Meter characteristics**

A meter designed to measure sound level in phons must have two characteristics that make it comparable to the human ear. First, its frequency response must approximate the curves shown in Fig. 1. Second, the frequency response must change with the sound level, becoming flatter as the level rises, according to Fig. 1. For example, shows two curves that would be required at two different levels. While their shapes are the same in general, one attenuates from 200 cycles and the other from 400. The dotted lines show the crossings of the asymptotes, at which points the turnovers are considered to take place. Of course, a compromise is necessary to avoid great complications in the circuits.

The easiest compromise in practice is simply to reduce the highest and lowest frequencies. That can be done very readily by reducing the value of a coupling capacitor and using another capacitor to bypass highs. The circuit and its response are given in Fig. 3. Notice that this curve is within a decibel or two of the upper curve in Fig. 2 at every point. Fig. 3 corresponds roughly to the 60-phon curve of Fig. 1, which, as can be seen in Table I, is at about the middle of the sound range usually measured.

**A practical meter**

The instrument described here does not have high accuracy. It is intended for use primarily by the radio service technician, amplifier enthusiast, public-address specialist, advanced architect, designer of sound insulation systems, and automobile engineers.

The basic design is a three-stage voltage amplifier followed by a diode-type vacuum-tube voltmeter. The complete circuit is shown in Fig. 4.

Although the 6SF7 is a 6.3-volt tube, it is fed from a 5-volt filament winding to keep the cathode of the diode fairly cool. If the cathode were extremely hot, the excess emission of electrons would produce a false reading on the meter. It is not necessary to reduce the voltage below about 5½ volts in most cases—sometimes the full 6.3 volts can be used, but many power transformers have a 5-volt winding.

Sound levels are continually varying and the meter must be highly damped; otherwise each tap or scratch will cause a large flicker of the needle and read-

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*Fig. 2—These curves complement two in Fig. 1. They are ideal for noise meter.*

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*Fig. 3 corresponds roughly to the 60-phon curve of Fig. 1, which, as can be seen in Table I, is at about the middle of the sound range usually measured.*

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*Fig. 4—The basic design is a three-stage voltage amplifier followed by a diode-type vacuum-tube voltmeter. The complete circuit is shown in Fig. 4.*
ing the meter will be impossible. A large capacitor (1,000 µf) is connected directly across the meter and another of lower capacitance (24 µf) but higher voltage rating is shunted across part of the diode load. Both of these damping capacitors must have *very low leakage* or there will be a big drop in sensitivity.

![GAIN IN \( \text{sensitivity} \) of voltage](image)

Fig. 3—This simply realized response is produced by the circuit diagrammed above.

The amplifier should have reasonably good logarithmic characteristics to keep the "phons" scale from being too cramped at the low end. This is achieved partly by the use of a number of ranges and partly by the use of grid-leak bias for the first two tubes, particularly for the second one. The gain of a grid-leak-biased pentode varies in a surprising way with the signal level. As the signal level rises from zero to a low value (usually provided by hum, stray r.f., and tube noise), the gain rises very rapidly to a maximum. As the signal increases, the gain decreases so that the circuit is more sensitive to small signals than large.

The range control consists of a voltage divider, the resistors of which are rather critical, preferably less than 21% tolerance.

The markings on the range control carry, not the usual factors or multiplication signs, but those of addition. At the sensitive position, labeled as SCALE on the panel, the effective range of the meter is 40 to 60 phons, which occupies the central two-thirds of the scale. At the other positions the range control reduces the gain by factors of \( \sqrt{10} \), 10, 100, 1000, etc., and so the corresponding markings on the front are +10, +20, +30, +40 etc. Highest range being +50, the loudest sound that can be measured is 110 phons (actually 113 at extreme end of scale). A range of 40 to 110 phons covers whispers to boiler riveting.

Slightly greater accuracy on the higher ranges can be achieved by inserting between resistor R (Fig. 4) and the chassis a 5,000-ohm resistor shunted by a 0.25-µf capacitor. This helps to match the ear's increase in low-frequency response at high levels.

The diode load resistance is adjustable in value, providing a measure of sensitivity control. It is not intended to act as a compensating device for variation in values, but rather as an initial adjustment before calibration so that the lowest range (40 to 60 phons) fits nicely on the meter scale with a little clear space at each end. When adjusting the 0.1-megohm potentiometer, a position of maximum sensitivity will be found and the sensitivity should be reduced slightly to the required value by increasing the resistance rather than reducing it. If you are interested solely in a *relative* calibration, the potentiometer could be replaced by a fixed resistor of about 25,000 ohms.

High voltage for the three tubes comes from a conventional power pack using a 6X5-GT rectifier and a simple R-C filter. Anode voltages are not critical, and the secondaries of the transformer can deliver anything from 550 to 750 volts. In the particular instrument illustrated in the photographs the rectified voltage was 315. Further smoothing is provided for the first two tubes by the decoupling network.

![Diagram of the noise meter](image)

Fig. 4—Complete diagram of the noise meter. The output is registered in phons. This works a 10,000-ohm resistor and a 24-µf capacitor.

The layout is not critical and requires only a few standard precautions. All components normally grounded to the chassis are separated by fiber insulating washers and a grounding of bare copper wire connects them all together and to the chassis at one point only, as close as possible to the cathode lug of the first tube. This keeps down hum level and should be adopted in all amplifiers, test equipment, and the like.

Sensitivity is adjusted by means of a screwdriver fitting into a slot hack-sawn in the shaft of the 0.1-megohm potentiometer, which is mounted under the scale on one side of the meter. A tube socket is used for the microphone input.

### Calibration

Absolute calibration of the instrument can be performed only if you have access to a calibrated microphone or an acoustics laboratory, but a relative calibration (differing from a true calibration by a constant number of phons, probably only a few) can be performed quite readily by using a good-quality a.c. voltmeter. The circuit is shown in Fig. 5.

Set the first, 1,000-ohm potentiometer to give a reading of 1.5 volts. Then adjust the second potentiometer to make the noise meter read about three-quarters of full scale on any convenient range—preferably one of the middle ranges. Mark the position of the needle on the scale. Now, keeping the second potentiometer fixed, reduce the voltage with the first potentiometer to 0.47 volt and again mark the needle position. The first marking should be labeled 60 phons and the second 50 phons. Following are a series of voltages and corresponding phons: voltages, 1.5, 1.2, 0.94, 0.75, 0.6, 0.47, 0.27, and 0.15; phons, 60, 58, 56, 54, 52, 50, 45, and 40.

The most reliable part of the scale will be that between 50 and 60. The microphone must be non-directional—including upward and downward. It must be pressure-operated because the response of a velocity-operated microphone varies with distance from the source. Fortunately, the requirement is easily met, as many non-directional pressure microphones are available. A crystal of the sound-cell type is ideal except that its response varies with temperature. Probably the best bet is a moving-coil or dynamic microphone.

Electronics amateurs suffering from talkative wives can educate said wives to refrain from talking by connecting a sensitivity relay in place of the meter. Any time the noise level rises above a certain value, the relay closes a circuit setting the phonograph playing some suitable record.

*Teachers may find a similar use in class control, the phonograph being replaced by a 500-volt supply and contacts on the seat of every desk!*
Square Wave Analysis
For Audio Amplifiers

Part II—Circuits for square wave generators and more uses for them

By EUGENE J. THOMPSON

A

n ideal square wave is the algebraic sum of a fundamental frequency and all its harmonics. (In practice, harmonics up to about the thirty-first are enough for a good square wave.) Therefore, the action of an electronic circuit on all of these frequencies can be observed, simultaneously, by passing a square wave through the circuit. This makes it unnecessary to study the effect on each frequency individually, as was formerly the practice. The result is a great saving of time and work. In the preceding article, the square wave method for checking the frequency response of an audio amplifier from 500 to 15,000 cycles, in one operation, was described.

Approximations of square waves can be produced with clipping networks similar to those used for timing, control and trigger voltages and for wave-shaping in radar, television, and other pulse transmission systems.

Fig. 1—Simple parallel diode clipper.

Fig. 1 shows a simple parallel diode clipping circuit. The diode is biased so that it will not conduct until a predetermined positive voltage is applied to the plate. As the input voltage rises from its zero value, the diode remains open until the voltage is high enough to make it conduct. As soon as

Fig. 2—Series diode clipping circuit.

the diode conducts, it is effectively a short circuit compared to the other circuit components, and the further rise in input voltage does not show up at the output.

A series diode clipping circuit is shown in Fig. 2. In this case, the diode is biased so that it conducts and is a short circuit until the input voltage rises high enough to make it stop conducting. When this happens, there is no current in the load resistance, and again the top of the sine wave is clipped off. In either of these circuits, the bottom of the sine wave signal could be clipped off just as well by reversing the diode connections and the bias.

Fig. 3—Two diodes will clip both peaks.

By using two diodes, both the top and the bottom of the sine wave can be clipped. Fig. 3 shows the diodes in a parallel arrangement; in Fig. 4 the diodes are in series. In each of these circuits, the clipping action is the same as with the single diode.

Fig. 5 shows a triode used as a clipper. The sine wave input must be large enough to drive the triode to saturation on the positive half cycle and to drive it to cutoff on the negative half cycle so that both peaks of the sine wave will be clipped. The load resistance of the triode must be large enough so the saturation current will not exceed the tube's current rating.

Fig. 4—Two diodes in a series clipper.

With any of these circuits, an ideal square wave can be approximated better by amplifying the clipped wave and clipping it again so that the sides of the square wave will be only the steepest part of the sine wave.

Better square waves can be obtained with the generator in Fig. 6. This is a type of relaxation oscillator known as a free-running multivibrator and is used for generating control and trigger voltages in radar and television equipment. It is a two-stage, resistance-capacitance-coupled amplifier with regenerative feedback. The output of each of the 6J5's is capacitance-coupled to the grid of the opposite tube. As the

Fig. 5—Clipper circuit using a triode.
signal applied to the input of an R-C-coupled amplifier is reversed in phase at the output, it is of the right polarity to reinforce the signal applied to the grid of the opposite tube. The oscillations are produced by the regenerative switching action resulting from one tube conducting while the opposite tube is nonconducting. The output signal is then passed through a cathode follower which isolates the multivibrator from the following stage, a class-A amplifier. This circuit is the one used in the Du Mont type 185-A electronic switch and square wave generator.

The frequency response of any circuit depends upon its constants. A given combination of R and C will pass a certain range of frequencies and no others. The problem is to determine if a certain circuit will pass the required frequencies. Because square waves are a composite of a wide range of frequencies, this often can be done in one operation.

A square wave is square because it consists of a fundamental and all its harmonics in definite phase and amplitude relation to each other. Any circuit which changes the phase relation or the amplitude of any of the components will distort the square wave. Only when the circuit passes all the frequencies of the square wave without attenuation or relative phase shift can the output be undistorted. This is the principle of square wave analysis.

In this relatively brief treatment, not all the possible uses of square wave technique can be discussed. The applications which follow have been selected to suggest other uses to the reader in his own particular field of interest and experience. The only equipment re-
Fig. 6—This multivibrator-controlled generator is a square wave generator and an oscilloscope, having a frequency response that will pass the fundamental and all the desired harmonics of the square wave without attenuation or relative phase shift. The square wave circuit will produce good square waves. will be distorted. The same is true if the frequency response is not good. Video amplifiers can be serviced in the same way. In addition to this, their low-frequency characteristics sometimes need adjustment. Fig. 8 is a typical video amplifier circuit.

If the R-C constants C1 and R3 change in value, they may introduce phase distortion in the lower frequencies. The phase distortion can be reduced by substituting a variable resistor for R2 and trying different values for C2 and adjusting them until the oscilloscope trace approaches the square wave shape as nearly as possible. This is illustrated in Fig. 9.

As in any procedure using cathode-ray oscilloscopes, some skill is necessary to interpret the patterns, but this is readily acquired with a little experience. In general, three major types of patterns are seen (see Fig. 7). These represent: (a) no attenuation or relative phase shift, (b) a loss of low-frequency response, and (c) a loss of high-frequency response.

Square wave analysis can be used for signal tracing public address systems and the audio amplifier sections of receivers. The usual stage-by-stage test procedure is used. If a stage is dead because of an open coupling capacitor or a shorted bypass capacitor, or for some other reason, there will be no output from the stage on the oscilloscope. If the grid bias is incorrect, the square wave output for that stage

Fig. 7—Three types of pattern: b and c show low- and high-frequency losses. generator is connected to the input of the circuit under test, and the oscilloscope to the output.

Fig. 8—Two ways to use square waves for adjusting video amplifier response.

AUGUST, 1950

Fig. 9—A scope pattern showing square wave pips superimposed on sine wave. A more satisfactory procedure (Fig. 8b) employs the combination of R4, R5, and C3. C3 is selected to have a small reactance compared to the sum of R4 and R5 at the lowest frequency to be passed. If the values of these components are adjusted until the output has a square shape, they will compensate for variations which exist in the other circuit constants and reduce the relative phase shift. When the circuit has been properly adjusted by either of the above methods, the values of the variable capacitors and resistors are measured and fixed components substituted for them.

An accurate square wave generator can be used to check the calibration of audio-frequency oscillators. The outputs of both instruments are applied to the vertical plates of the oscilloscope and the a.f. oscillator is tuned to a number of different frequency settings. At each setting, the horizontal sweep of the oscilloscope is adjusted to get several complete a.f. cycles on the screen. The frequency of the square wave generator is then tuned until the little square wave "pips" stand still on the sine waves (see Fig. 9). If the readings of the sine wave generator and the square wave generator agree, the calibration of the former is correct. The above procedures are also useful for checking receivers and the modulator stages of transmitters, frequency

Fig. 10—Setup for impedance matching.

Fig. 11, left—Reflection distort wave. Fig. 12, right—An attenuator circuit. calibration, etc. The correct terminal impedance of transmission lines also can be found by square wave analysis. Fig. 10 represents a typical transmission line. If the terminal resistance is incorrect, the reflections within the line will distort the waveshape as shown in Fig. 11. The stepping up of the voltage because of these reflections is quite pronounced. The correct terminal impedance can be found by varying the resistance of the decade box until a square wave is obtained. After this, the oscilloscope should be connected to the input of the line to check for reflections.

Fig. 12 is a schematic diagram of an attenuator network employed in some types of oscilloscopes to limit the input voltage. If the values of the components are set so that the ratio R1/R2 is equal to C2/C1, the attenuator will have a flat frequency response. Stray capacitance and slight variations in the values of C1 and C2 will cause amplitude distortion.

To remedy this by the square wave method, the fundamental frequency of the square wave is selected a little above the frequency at which XCl = R1, to obtain the full effect of the capacitive reactance for all harmonics. If the output waveshape indicates that the low frequencies are being attenuated, the ratio C2/C1 is too small; and if the high frequencies are being attenuated, the ratio is too large. Compensation can be made by substituting a number of capacitors for either C1 or C2 until the output is a square wave.

In addition to the above applications, the square wave generator and oscilloscope have numerous other uses. Square wave generators can be used as electronic switches to permit the simultaneous viewing of two or more different electrical phenomena which are to be compared.

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The designer and experimenter with electronic music has lots of room in which his imagination and ingenuity can run around. He need not be bound by what has been done before, but he will almost always get inspiration and assistance from a knowledge of the best of the many systems tried and used in the past. For that reason, this article and following ones present a number of designs and circuits in commercial musical instruments and in an exhaustive survey of U.S. patents, as well as some that are original or partly so with the writer.

The tempered scale

The most important basic interval between tones, the octave, is common to all musical scales—it is the starting point and initial building block. The octave of any note has a frequency equal to twice that of the note itself; it is, in other words, the note's second harmonic. It is important because the human ear senses a kind of identity between any note and its octave, a feeling that the note is being repeated at a higher pitch.

The most familiar system of music in the western world uses the so-called tempered scale consisting of 12 tones, with the interval between adjacent tones equal as far as the ear can detect. Since the ear's response to frequency (as well as volume) is logarithmic, the frequency difference between adjacent tones is progressively greater as pitch increases.

The piano keyboard drawing in the tempered-scale frequency chart in Fig. 1 can be used to illustrate the tone structure of music. The white keys are always identified by letters A through G. The keys in this chart are also numbered from 1 to 88. Middle C is key No. 40. The major scale consists of seven notes plus an eighth which is the octave of the first.

Beginning at middle C, we can play a C-major scale by pressing in turn each of the white keys—Nos. 40 through 52. Key No. 52 is the octave of key No. 40. Additional octaves of C are Nos. 64, 76, and 88. The notes in each octave are known by the same letters as those in other octaves.

Notice that the frequency intervals between adjacent notes of the major scale are not constant. To begin with, most are not adjacent to each other because black keys are interposed. The black keys are known as accidentals.

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Fig. 1—Drawing of a piano keyboard showing the tempered scale. The frequency of each note in cycles per second is given.
(appropriate because they do not occur in the scale itself but are used for a sort of embellishment). Each accidental may be identified in two ways: it is the sharp of the white key below it and the flat of the white key above it. Thus, note No. 41 may be called either C-sharp or D-flat. The interval between any key and its fourth, fifth, or adjacent one, black or white, is a half-tone. Since two halves make a whole, the interval between two notes with one key between them is a whole tone. Example: C to D is a whole tone; E to F is a half tone.

One of the terms that will be used in these articles also refer to intervals. We already have the half-tone and the whole tone. Next is the third. That is the interval between one note of the scale and, not the next, but the second higher one—between, for instance, C and E in the C-major scale, or between F and B. There are also fourths, fifths, sixths, and sevenths. In each case, the number is an ordinal, not a fraction, and refers to the third, fourth, etc., note from the starting point, counting the starting note as the first. These numbers refer only to notes in the scale, not to accidentals. Thus, a seventh, with C as the bottom note, has B as the top. By the key of C, but B-flat is the seventh of C in the key of F, since, as we illustrated, B-flat is part of the F-major scale, and B-natural is not.

Neon-lamp oscillators

Probably the most truly electronic method of tonal generation is the use of space-discharge tubes—hard and soft electron tubes—in which nothing moves except electrons and perhaps ions. Of these, the simplest types are gas-tube relaxation oscillators.

Fig. 2—The basic relaxation oscillator and sawtooth waveform it produces.

The basic circuit is shown in Fig. 2, using a neon lamp. When the switch is closed, the capacitor starts charging. The lamp cannot ionize or fire until the voltage across the capacitor is above its breakdown point. The rush of electrons from one capacitor plate to the other passes through the resistor, the voltage drop across which reduces the voltage available to charge the capacitor. As the capacitor continues to charge, however, the electron flow diminishes and the drop across the resistor becomes smaller until finally the capacitor is fully charged to the voltage of the battery.

In practice, the capacitor is never allowed to charge fully. When the charge reaches the breakdown voltage of the neon lamp, the neon ionizes suddenly and the lamp becomes a fairly low resistance. This near-short across the capacitor quickly discharges it.

There being now little or no voltage across the lamp and capacitor, the lamp deionizes and the voltage across the lamp begins to rise. More charge is taken across the lamp or across any part of all of the resistor. The output wave is a sawtooth like that shown in Fig. 2.

The frequency of a neon-lamp relaxation oscillator is not stable. Even if all components and voltages are held constant, the discharge may not always take place between the same points along the two electrodes in the lamp. The gas is somewhat temperature-sensitive, too, and the replacement problem is important, since no two "identical" lamps ever oscillate at the same frequency.

The Trautonium

A single neon lamp has been used in a solo instrument with continuously variable frequency by (among others) the German inventor, Friedrich Trautwein, in his Trautonium and similar instruments. The basic idea appears in Fig. 3. The circuit is standard except that R2 has been added as an output load. Its value is a fraction of that of the main tuning resistor R1. A sawtooth wave is obtained from across the lamp and a pulsed output, the result of the periodic neon-lamp discharges, from across R2.

R1 may be a long coil of resistance wire or a long composition element like those used in potentiometers. The contactor is a strip of flexible conducting material suspended slightly above R1 and topped with a nonconducting (insulating) layer. The player presses his finger on the contactor at any point to short out part of R1 and change the frequency.

Typical values for R1 and C with a 1/25-watt neon lamp experimented with by the writer are 1 megohm and .005 µf. R2 may be anywhere from 1,000 to about 100,000 ohms. The lower values are safer because the following circuits will affect the tuning less, but of course more amplification is necessary.

The circuit to which the sawtooth output goes must have a very high impedance and high resistance (and usually a blocking capacitor) must be inserted in series with the output.

Feeding the two outputs to different amplifiers and mixing the results gives various tone qualities. The actual Trautonium circuit is more complex and will appear in another article.

Synchronized oscillators

Neon-lamp oscillators can be used in a lamp organ if they are synchronized by a stable source of frequency.

For a full scale, this requires 12 such sources, one for each of the notes in the top octave, which may be vacuum-tube oscillators, tuning forks, or any other stable devices. Each source synchronizes one note in the organ.

A typical way of doing this appears in a patent issued to Nicholas Langer, who is probably the most prolific worker with gas-tube oscillators for music use. The idea is shown in Fig. 4.

Two neon oscillators, shown, one tuned roughly to the desired frequency and the other to about twice that frequency or an octave higher. There would be one oscillator in this string for each note and one for each of its octaves which appear in the organ; for example, six if the note were C, on a 61-note instrument. The neon lamps are fixed-tuned by resistors R1 and capacitors C. Across the common resistor R2 appears a small part of the variations in current caused by each oscillator. This is a synchronizing voltage, which locks all the oscillators together in octave relationship. R2, which should have a maximum value of about 10,000 ohms when 1-megohm resistors R1 are used to tune the oscillators, is varied so that locking is obtained with a minimum of resistance in the circuit.

R3 is across the secondary of a transformer which is fed by a primary sync source—a tuning fork or vacuum-tube oscillator of high stability with a frequency equal to or one octave above the highest note wanted. By adjusting R3, enough primary sync signal is brought in to lock all the oscillators to it.

Output is taken through the transformers associated with each oscillator. The primaries are in series with the lamps and the secondaries are in series with each other. Each secondary is normally shorted. When a key is pressed, the short is removed and the desired tone goes through to the output. The output transformers should have fairly low-impedance primaries to avoid effects on oscillator frequency and somewhat higher-impedance secondaries to minimize disturbances fed back from other circuits. Ordinary 3-to-1 audios do the job very well.

Another system for syncing neon lamps will be found in the article "Simple Electronic Organ" in the January, 1947, issue of this magazine. Additional gas-tube tone generators will be discussed in the next article.
It is relatively simple to add modulation to those BC-221 frequency meters which do not have this feature. A BC-221-F was modified as shown in the diagram. Almost all models can be converted by following the outlines of this article and the diagram. Some are easier to modify than others because of the parts layout.

Remove the a.f. amplifier socket and replace it with an octal socket which will be used for a 6SN7 modulator and amplifier. This is not difficult in the model F because the a.f. socket is out in the clear. Break the B-plus lead to the plate and screen grid of the variable oscillator and insert the primary of a 3-to-1 interstage transformer. Connect the secondary to cathode and grid of one half of the 6SN7 with a 47,000-ohm resistor and .01-mfd capacitor in the grid lead. Replace one of the phone jacks with a s.p.s.l. toggle switch connected between B-plus and the modulator plate. This switch turns the modulation on and off. These modifications are shown in the diagram.

The modulation transformer is not critical. The cheaper it is, the better it appears to work in this circuit. Remove laminations from its core until the oscillator tunes to 400 cycles. The transformer may be mounted on the back of the panel along one side. Connect the remaining half of the 6SN7 as an a.f. amplifier. If you plan to use batteries, ground the amplifier cathode to the hot heater lead as in the original circuit. If you use an a.c. supply, connect the cathode to ground through a 3,500-ohm resistor.

The power supply consists of a selenium rectifier in a half-wave circuit. Two 6-volt filament transformers are connected back-to-back to isolate the B-minus lead from the a.c. line. Heaters are supplied by the secondary of the transformer next to the line. Because the BC-221 was designed to hold calibration over a comparatively wide range of battery voltages, a regulated supply is not a necessity. Voltage regulator tubes such as the OB2 and OC3 may be included as refinements.

Some components in the diagram have values and others have reference numbers, reference numbers indicating components which appear in the original schematic. Add only those components which have values. Locate all components by comparing the diagram with the original schematic. BC-221 diagrams will be found in technical manual TM 11-300 which may be obtained from the Superintendent of Documents, Washington 25, D. C.

Some models have a crystal corrector—reference number 3-3 or 30—connected in parallel with the crystal to set it to exactly 1 mc. In some models, this variable capacitor is adjusted by passing an aligning tool through the hole behind the name plate on the panel. On others, the chassis must be removed from the case.

Turn on the meter and let it warm up for 15 to 30 minutes. Set the control to CRYSTAL or XTAL ONLY, then adjust the corrector until the crystal zero beats exactly with WWV. We find it easier to make this adjustment while WWV is modulated because it is easier for the ear to detect a change of a few cycles in an audio note than a difference of a few cycles at 5 or 10 mc. A 1N34 and milliammeter can be used as a zero-beat indicator.

Notes on the BC-221
Do not rotate the dial of the BC-221 through the black (recalibrated) area as this will upset the delicate bearings and throw the calibration off so the instrument must be recalibrated to restore its original accuracy.

Some BC-221 and LM-type frequency meters are available without calibration books and crystals. If these are in operating condition, they can be used. Recalibration is a simple but fairly lengthy process.

The original crystal is enclosed in a shell which resists a metal tube. Any reliable 1-mc crystal can be used to replace it. However, it may be necessary to modify the crystal socket.

The low band of the BC-221 covers from 125 to 250 kc on fundamentals and the high band from 2,000 to 4,000 kc. The 125-250 kc range can be calibrated roughly by tuning in its harmonics on the broadcast band. Beating the harmonics against broadcast carriers will provide several good points from which to start calibration.

Zero-beat the crystal against WWV, and use the beats between the crystal harmonics and the variable oscillator for closer calibration of the dial. Use the same method to locate beat points in the high band.

A 10-ke multivibrator is useful in calibrating between crystal check points. Adjust the multivibrator to exactly 10 kc, then loosely couple its output to the antenna of the BC-221. Select a point where the crystal and multivibrator signals are zero-beat. Turn off the crystal and leave the multivibrator running. Note the dial setting. Tune higher or lower until another 10-ke beat is heard. Record the exact number of divisions between the 10-ke beats. Divide the number of divisions by 10 to determine the number of kilocycles per dial division.

Check both bands and enter the frequencies and dial settings on a calibration chart or in a book. Check the settings and beat points several times.

TUNING AN SSSC SIGNAL
Have you ever heard an amateur phone signal which sounded one moment like a phonograph playing at half speed and like the same record at twice normal speed the next? Was the S meter waving like crazy? If so, you were listening to one of the ever-increasing number of SSSC (single-sideband, suppressed-carrier) transmitters. Because the signal seemed so distorted, you probably wondered how anyone could read it. It's really simple.

Your first step is to forget everything you have learned about tuning your receiver. Tune in the station for maximum deflection on the meter. Turn the r.f. gain all the way off and the a.f. gain full on. Turn off the a.v.c. Increase the r.f. gain until the signal is just audible. Turn on the b.f.o. and adjust its pitch so the signal sounds natural. Advance the r.f. gain for desired volume. It will probably take a transmission or so to get the hang of it; but when you do, you won't have any more trouble. Remember: A.v.c. off. Maximum a.f. Minimum r.f. B.f.o. on. Tune with pitch.

Operating the set with maximum a.f. and minimum r.f. gain prevents overloading.

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By W. S. KEMPER, W4KOF

RADIO-ELECTRONICS for
**A Compact 75-Meter Rig**

**A portable phone-c.w. station made from surplus gear**

By ERNEST J. SCHULTZ

WITH the supply of ARC-5 equipment still plentiful, it is a simple matter to assemble a complete portable phone-c.w. station with little effort. This article describes an assembly making a compact 75-meter phone-c.w. station and also a unit which vastly improves the utility of the command receiver as a communications receiver.

A glance at the surplus ads showed a scarcity of the 274-N transmitters covering the 3-4 megacycle range, indicated by the higher prices now being asked for these units. However, the supply of the BC-457-A's covering the 4-5.3 mc spectrum still seems plentiful and their cost is moderate. With no further incentive, a 457 unit was acquired together with a 3-4 mc matching receiver, a readily available item.

Conversion of the transmitter to cover the required range was just a matter of adding three turns of solid hookup wire close wound to both the oscillator and amplifier coils and realigning the slugs and trimmers in both circuits. The transmitter alignment is a simple process. The oscillator adjustment is merely a normality as the transmitter dial is discarded and the selection of dial position in regard to frequency is arbitrary. (In this case 3.5 mc was made to fall at 4.1 mc on the old dial.)

The amplifier is aligned by inserting a milliammeter in the plate lead and tuning the slug for minimum current at 3.5 mc and the trimmer for minimum at 4.0 mc. The 4.6 mc crystal check point is still retained and is found near the high frequency limit of the dial.

**Adding an i.f. stage**

The receiver was usable as obtained but the performance was unsatisfactory on the congested 75-meter band. It seemed to receive at least half the band at one setting of the dial. It was necessary to supplement the receiver to give a higher degree of selectivity and adequate loudspeaker volume.

An outboard chassis was made of a piece of thin aluminum sheet bent to a U to fit in the space formerly occupied by the dynamotor. The device contains a 12BE6 converter and a 12ATC diode triode (Fig. 1). A fixed 960-ke oscillator—using a broadcast oscillator coil in the converter reduces the 1415 kc i.f. frequency to 455 kc. The 455 kc signal is passed through two 455 kc i.f. transformers in tandem loosely coupled by a 3-muF capacitor. The 455 kc signal is detected by the 12AT6 diodes and the audio is fed through a volume control to the triode section which is R-C coupled back to the 12A6 output tube in the receiver. A volt. voltage is taken from the 12AT6 diode load and fed back to the grids of the r.f. and i.f. stages in the receiver, providing a further refinement.

Bandspread was added to the receiver by removing all but two rotor plates from each section of the tuning capacitor and winding an additional 10 to 15 turns on the coils. (The serrated rotor plate and its adjacent neighbor are left intact.)

Alignment of the receiver was facilitated by drilling out the rivets holding the r.f. coil shield cans to the mounting bracket to give access to the slugs. The receiver is aligned in the conventional manner, the slugs being adjusted at the low end of the band and the trimmers at the high end. With these modifications and additions the performance is comparable or superior to some of the other units found on the market. Two 12A6 tubes and a 274-N is all that is required to complete the assembly.

---

**Fig. 1—Circuit of the outboard i.f. stage showing hookup to the receiver.**

---

AUGUST, 1950

www.americanradiohistory.com
Front view of the complete unit. The frame is made of aluminum angle stock.

communications receivers in the moderate price class.

The original dials were discarded after using them as templates for the new ones. The replacement dials were made out of flat aluminum with white poster board glued to the surface. The frequencies were inked in with India ink and were colored with different shades of transparent oil paints to indicate band edges as well as c.w. and phone regions.

The oil colors were those normally used for coloring photographs. These colors are easy to apply with a small wad of cotton on a toothpick and if errors or changes are made, they can be removed with a solution supplied with the oil colors. To make the job permanent, clear lacquer or varnish is applied over the colors after allowing several days for drying. An ample supply of colors for a great number of dials is available at very low cost in a “sampler” kit available in all photography supply stores.

This method of dial construction is particularly suited to those ARC-5 owners who use their sets as multiband v.f.o.’s and have to stop to multiply mentally each time they change frequency. A dial plate with concentric scales indicating the frequency in use on each band can be made to replace the original.

Two power supplies are used (See Fig. 2). The lower voltage supply furnishes voltage to the oscillator plate when on c.w. and powers the modulator tube plus the amplifier plate and screen when transmitting c.w. The larger power supply furnishes voltage to the amplifier plate when on c.w. and powers the modulator tube plus the modulator plate and screen when transmitting on phone. A push-button applies voltage to the oscillator plate so that the signal can be spotted when receiving, and a monitor switch allows the receiver to operate independent of the T-R switch if the signal is to be monitored when transmitting.

Three relays are used in the unit. One (RY-1) switches the antenna from the transmitter to the receiver, another (RY-2) does the job of changing from transmit to receive and a third (RY-3) is used for keying. The relay coils may be any type available to the constructor. In our case a low-current 50-watt bias winding on the large power transformer was put to use with a selenium rectifier and an R-C filter to supply 24 volts for the antenna and T-R relays. The keying relay is a 6.3-volt a.c. relay and operates from one of the filament windings.

The heaters in both the transmitter and receiver are connected in parallel and are supplied from two 6.3-volt windings on the large transformer connected in series. The speech amplifier and modulator tubes. Filaments are supplied from the 6.3-volt winding on the small power transformer.

The modulator circuit

The speech amplifier and modulator are conventional in design. The coupling components and the microphone load were chosen to favor the middle of the audio range, thereby avoiding loss of modulator power in useless low frequencies. This pays dividends in another way, as low frequency motor-boat ing sometimes occurring in high gain speech amplifiers along with hum is eliminated before it starts. This does not impair the quality of speech, all reports being “excellent”, “crisp”, etc.

The single 6L6-G Heising modulator is run somewhat above its ratings with 400 volts on the plate and 300 volts on the screen, but no coloring of either element has shown up in service. The plate input to the transmitter varies somewhat with different types of antennas, but is about 30 to 40 watts. The 6L6 under the conditions in this setup furnishes adequate audio power to modulate the 1625’s sufficiently. While it doesn’t have enough power to modulate 120 percent and create a lot of unwanted splatter, no reports have been received of too little audio on the signal.

The antenna loading coil in the transmitter would not do a job on long antennas. A 350-mf fixed mica transmitting capacitor connected between the antenna terminal and ground makes good loading possible with end fed antennas from 66 to 400 feet long. An external ground makes no apparent difference in signal reports so its use is optional but desirable for safety.

The power supplies and modulator are mounted on a 14-inch chassis and the receiver and transmitter are supported on a “skeleton like” platform made of % × ¼ × ½-inch aluminum angle stock secured to the chassis. The height of the receiver and transmitter above the lower chassis is adjusted so that the power supply and modulator tubes can be replaced without removing the transmitter or receiver. Trunk handles screwed to the ends of the chassis provide easy handles for carrying.

It is not necessary to sell anyone on what can be done with low-powered transmitters, so a list of stations contacted and reports obtained seems superfluous. Many hours have been spent on enjoyable phone and c.w. QSO’s both far and near. The portable is a dandy to stow into the car trunk and bring along when you are visiting friends in that cliff dwelling in the Bronx or spending a couple of weeks at some vacation spot. The only accessories are a mike and a key, a piece of wire for an antenna, and last but not least, power.

Fig. 2—The schematic of the modulator power supply and control system unit.
**High-Accuracy Timer For Short Intervals**

This versatile all-electronic timer has many uses in both shop and home.

By R. L. PARMENTER

Accurate short-interval timing is necessary for spot welders, punch pressers, plating, heat-treating, and many other industrial uses. Around the house and workshop an electronic-type timer is often more useful than clockwork mechanisms. For very short intervals (less than a second) timers are preferable to mechanical devices because the general accuracy is better and they have greater reset accuracy. This last feature is important in photographic work—printing and color work especially. Such a timer has many uses around the kitchen. Why have burnt toast in the morning when an electronic timer is available?

Almost all electronic timers operate on the same principle. A vacuum tube is generally used to operate a relay. A negative charge on the grid of the tube cuts off the plate current under normal conditions. When a timing switch is closed, a circuit is completed so that this negative charge gradually leaks off until the tube conducts. The plate current is used to operate a relay which in turn controls an external circuit.

This circuit is frequently varied by adding components to provide for controls and sometimes an additional tube is used as a diode rectifier to provide the d.c. biasing potential for the control tube.

**How the timer works**

A simplified circuit of the timer described in this article is shown in Fig. 1. The single tube is a thyratron, type 2050, 2021, or 2051. The 2050 has been on surplus market lists at reasonable prices. The basic circuit is simple and, since it is practically bug-free, will operate dependably.

No power supply is used. A filament transformer (not shown) provides 6.3 volts for the heater. The a.c. line voltage is applied to the circuit so that, with the timing switch open, the plate of the tube is at one polarity at the same instant the grid is at the other. By grid rectification, the grid of the tube has a negative voltage high enough to cut off plate current flow. When the timing switch is closed, the grid return is completed, and the negative potential stored by C leaks off through R1 to the cathode, gradually allowing the grid to become more positive until the tube ionizes and conducts. A heavy plate current flows under these conditions (75 ma maximum for a 2051 and 100 ma for a 2050 or 2021), which is sufficient to operate relays having a wide range of resistances.

While the exact resistance of the relay is not critical, if the resistance of the particular unit used is too far from 2,000 to 2,500 ohms, it is well to insert a series limiting resistor if the relay multiplier switch is thrown to x1, the original intervals will be divided by 10. Since the basic range capacitor is .05 µf, by switching we arrive at .5 µf for the x10 range and 2.5 µf for the x50 range.

The main time set control is an 11-point rotary switch with ten 500,000-ohm, ½ watt resistors soldered to the terminals. (The 510,000-ohm preferred values may be used with only a slight change in the time interval.) This method of determining the interval gives a better degree of reset accuracy than the alternative of using a high-resistance potentiometer which would be difficult to reset at a specific value. A ½-megohm potentiometer in series with the string of resistors is a vernier adjustment. Although this potentiometer is on the front panel, it would be preferable in most cases to make this a screwdriver adjustment behind the front panel.

A terminal board was installed to simplify connections for different uses. The time switch is mounted on the panel but leads are brought out to terminals 3 and 4 to permit connection of a remote, momentary type switch in parallel or in series with it. Likewise, one set of relay contacts are brought out to terminals 5, 6, and 7 with the moving contact connected to terminal 6. This is for the same purpose. The time switch can be connected in series with one pair of relay contacts for such functions as repeat cycle, etc.

The other set of relay contacts is connected to the a.c. line through a single-pole double-throw reverse switch and an a.c. outlet. This switch connects the outlet to the line either at the start of a cycle or at the end of a cycle. Other arrangements will suggest themselves to the builder as the need arises.

**Construction details**

The construction of this timer is not difficult. No exact layout of parts need be followed, no shielding is necessary, and, if it is for one specific use only, some of the parts may be omitted. The
photos may serve as a guide for the builder.

The size of the cabinet is approximately 6¼ x 7 x 5 inches, allowing ample room for the components. Masonite is used for the panel and sub-base, and light wood and plywood for the cabinet. A fairly tight cover on the back of the box keeps dust and dirt from the relay contacts. Little or no ventilation is required as there is negligible heat even during continuous use. A couple of small vent-holes may be left, one at the top and one at the bottom, for circulation.

Considerable space may be saved if the unit is for limited use. For instance, if the timer is only for photographic work, the terminal board, the multiplier switch, and the reverse switch may be eliminated. The relay could be a more common s.p.s.t. The author has included all the extras so that other arrangements can be tried.

The range of time intervals covered with the components shown in the circuit diagram is from instantaneous to a maximum of about 1 minute and 25 seconds. The table indicates the different settings used to obtain a fairly wide range of time intervals. These were obtained with the vernier adjustment set at zero (maximum resistance). The intervals can be made shorter with the vernier if desired. Four representative intervals are given for the ×50 to illustrate the multiplication factor.

Fig. 2—Complete diagram of the timer. It does not need its own power supply.

(There is a slight error in the ×50 intervals as compared to the ×10 intervals because the 2-µf capacitor was somewhat under its capacitance rating.)

**Adjusting the intervals**

If other time intervals are desired, the values of C1, C2, and C3 and R1 can be changed. For smaller intervals, the total resistance of R1 and the capacitances could be made smaller; they should be made larger for longer time intervals. It would be convenient to make the time intervals as close to one second as possible for some applications. A little cut and try may be necessary to adjust the interval to the desired value.

In selecting capacitors C1, C2, and C3 use a capacitance bridge if possible to get capacitances as nearly as possible exact multiples of one another. This makes calibration easier. The same applies in selecting resistors for the set time unit. The response of the circuit is not linear, but it will be nearly so if you make the resistor values equal. If less vernier action is required, the potentiometer may be as little as 75,000 ohms and still retain some control.

This electronic timer is an interesting project for the beginner as well as a functional device for industrial uses.

**INFINITE IMPEDANCE TEST PROBE**

Voltage measurements of high-impedance high-voltage sources may be in error when ordinary measuring equipment is used because an additional current drain of even a few microamperes may reduce the output voltage appreciably. The ideal voltage measurement would take no current from the source.

A voltmeter that does this must have an infinite input impedance. One such device, shown in Fig. 1, consists of a high-voltage probe containing a high-voltage capacitor and a low-voltage capacitor and switch which are applied across the terminals of any ordinary vacuum-tube voltmeter. When the probe is applied to the high-voltage source, the high-voltage capacitor is charged.

The charge is then transferred to the low-voltage capacitor, and this charge is measured directly by the v.t.v.m. on a low-voltage scale. The switch is used to discharge the low-voltage capacitor.

The capacitance of the high-voltage capacitor is 1/1,000 that of the low-voltage capacitor. Thus the voltage transferred to the low-voltage capacitor is 1/1,000 that picked up by the high-voltage capacitor. A voltage of 10,000 volts would give a full-scale reading on the 10-volt scale of the v.t.v.m.

Millivac Instruments is producing this device. Their model, the PR6-NL, measures up to 30,000 volts. In addition, the probe can be used with other Millivac equipment to measure extremely small values of direct current.

![Photo of the high-voltage test unit.](image)

Fig. 1—A high-voltage charge is transferred from the high- to the low-voltage capacitor and measured with the v.t.v.m.

For best possible accuracy the two capacitors must have extremely low leakage. The accuracy of the instrument then depends only on the tolerance of the capacitors and the accuracy of the v.t.v.m.

**INTERVAL TABLE**

<table>
<thead>
<tr>
<th>Set Time</th>
<th>X10 (Secs.)</th>
<th>X50 (Secs.)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>3½</td>
<td>15</td>
</tr>
<tr>
<td>2</td>
<td>5</td>
<td>24</td>
</tr>
<tr>
<td>3</td>
<td>7</td>
<td></td>
</tr>
<tr>
<td>4</td>
<td>9½</td>
<td>45</td>
</tr>
<tr>
<td>5</td>
<td>10</td>
<td></td>
</tr>
<tr>
<td>6</td>
<td>12</td>
<td></td>
</tr>
<tr>
<td>7</td>
<td>14</td>
<td></td>
</tr>
<tr>
<td>8</td>
<td>19½</td>
<td></td>
</tr>
<tr>
<td>9</td>
<td>17</td>
<td></td>
</tr>
<tr>
<td>10 * *</td>
<td>18</td>
<td>85</td>
</tr>
</tbody>
</table>

**MATERIALS FOR UTILITY TIMER**

- Resistors: 10–500,000 ohm, ½ watt; 1–100,000 ohm, 1 watt; 1–15,000 ohm, ½ watt; 1–20,000 ohm, 10 watt; 1–500,000 ohm potentiometer.
- Capacitors: 1–0.5, 1–2 µf; 400 volt paper; 1–10, 450-volt electrolytic.
- Miscellaneous: 1–2,500-ohm, d.p.d.t. relay; 1–4.3-volt, 1-amp transformer; terminal board; tubes; socket; chassis; o.c. receptacle; switches; hookup wire.
He seals out trouble...

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To make cable joints tight and strong, splicers formerly used lots of solder. Then, Bell Telephone Laboratories developed a new technique for making better joints with much less solder. This saves one million pounds of solder a year — helps keep the price of your telephone service low.

Two kinds of solder are now used. One makes the splice strong; the other seals it. First, the splicer builds up a joint with a solder of lead and tin, which flows easily under his wiping cloth. To seal the joint, he applies a light coating of low-melting-point solder, composed of lead, tin and bismuth. On contact with the still hot joint, it flows into and seals every pore.

Cable-sealing solder is only one of 30 low-melting-point alloys which Bell metallurgists have developed for special uses — in fuse wires, for example, and in the solder connecting hair-like wires to piezoelectric crystals for electric wave filters.

Continuing research with a substance seemingly as commonplace as solder demonstrates again how Bell scientists help keep your telephone service the world’s best.
Capacitance Relay Operates Display

By W. G. ESLICK

This circuit will operate a device in a store window when a hand is held near a metal plate on the window, but it can be used as well as an alarm for many other purposes.

L1-L2 tunes to 456 kc or to 227 kc with no difference in operation. I used an oscillator coil from a midget aircraft beacon radio that tuned from 200 to 430 kc with 135-ke i.f.'s. The coil can be made from an r.f. choke. Tune the signal in on a broadcast receiver and notice if there is a harmonic of 456 kc (or thereabouts, just so the i.f. transformer tunes to the oscillator).

The neon bulb across the control R1 is a midget taken from a fuse. The voltage across R1, measured with a 20,000-ohm-per-volt meter, is 165 volts d.c. The neon bulb indicates if the oscillator is tuning to the correct frequency. A voltmeter could be used but the built-in neon bulb is simpler. The relay was taken from a two-tube marker receiver and is very sensitive. C1 must be made a smaller value if the circuit is hooked to a large object which pulls the oscillator out of oscillation. The relay can be made to close or open, depending on the setting of R1.

Set R1 so the relay is open, but on the grade of closing. When a hand is held near the metal plate hooked to C1, the relay should close. If R2 is set so the relay just closes, then the added hand capacitance will open the relay. When the oscillator is tuned to the frequency (one-half the frequency) of the i.f. transformer, the r.f. voltage is applied to the plate of the 12H6. This voltage is rectified and applied as cutoff or near cutoff bias on the 12A6. When the oscillator plate is tuned to 456 kc, the bias is removed and the 12A6 plate current operates the relay. The i.f. transformer must be peaked sharply to have the slightest change in oscillator frequency operate the tube.

Do not use a tuned circuit that requires a large value of padder capacitance to reach 456 kc because a small change in capacitance will not detune the oscillator sufficiently to operate the relay. Sensitivity depends on the setting of R1 and the sharpness of the i.f. transformer. Use a voltmeter across R1 to adjust the i.f. transformer.

The circuit can be hooked up with an isolating resistor and capacitor between the common ground and chassis, as shown, to avoid a hot chassis. If this is not satisfactory, a little experimentation may be necessary to get the circuit operating properly.

Materials for Capacitance Relay

- Resistors: 1-420, 1-150,000, 2-700,000 ohm, 1/2 volt; 1-100 ohm, 10 volt; 1-250,000-ohm potentiometer.
- Capacitors: 1-0.001, 1-0.02, 1-0.002 mf mica; 1-0.05 mf, 400 volt; 1-0.2 mf, 25-, 50-uf trimmers; 1-100-uf variola.
- Miscellaneous: neon bulb (midget), 1-105, ma relay, 456-ke i.f. transformer, oscillator coil, tube socket, chassis, hookup wire, switch.

\[ \text{Continuing} \]

$1,200.00 Prize Contest

Radi-o-Electronics in the Home

Entries for the August contest hit a new summer low. Of the few ideas submitted, practically all were electrical rather than electronic, and therefore did not fall within the scope of this contest.

The only entry which appears to fulfill the conditions is that of Simon Wyrrn, 79-18 Elks Road, Elmhurst, N. Y., who sends us a description of a radio nurse. Since his entry meets the requirements of the rules, and since no better idea has been submitted, Mr. Wyrrn was awarded the $50 first prize.

Unfortunately, the radio nurse idea he submits resembles too closely those embodied in a number of articles and radio-electronic circuits previously printed in this magazine to justify present publication, and will therefore not be printed.

There is still money to be made by working out really good ideas for radio-electronics in the home. If your idea amounts to a full-dress article, regular space rates will be paid in addition to the prize money.

**FIRST PRIZE** $50

**SECOND PRIZE** $25

**THIRD PRIZE** $15

**FOURTH PRIZE** $10

Please Note the Following Rules

1. This is a monthly cash Prize Contest for the best idea submitted during the month for a practical new radio-electronic application in the home.

2. The highest prizes will go to those contestants who have actually built the devices they describe and who submit photographs to prove it. Lesser prizes may be given for "ideas" not reduced to practice and for entries unaccompanied by photographs.

3. Entries of constructed devices must be accompanied by photographs, full description, and complete circuit diagrams.

4. Ideas for devices which have not actually been built must be stated in complete detail and accompanied by complete diagrams, drawings, and all other possible descriptive material.

5. All the descriptions and photographs of the prize-winning devices or ideas will become the property of Radi-o-Electronics, which will publish a descriptive article on each device or application. The prize winners will be paid regular rates for their articles, in addition to the prize money. Entries not winning prizes will be returned.

6. If two or more entries submitted during the same month are judged to be of equal worth, identical prize awards will be made for both entries. Devices which have been awarded prizes in previous contests will not be considered unless they show marked improvement on earlier entries.

7. All entries will be judged by the Board of Editors of Radio-Electronics. Prizes will be awarded in accordance with novelty, general importance of the application or device, smallness of cost involved in building it, and practicability. The decisions of the Board of Editors of Radio-Electronics will be final.

8. Excluded from this contest are Radio-Electronics employees and their relatives.

9. The fifth monthly contest closes August 31 at midnight, Eastern Standard Time. All entries postmarked not later than August 31 will be judged in the fifth month's contest. Address all entries to "Contest Editor, Radio-Electronics, 25 West Broadway, New York 7, N. Y.

10. Announcement of the third monthly prize award will be made in the September issue of Radio-Electronics. The third month's prizes will be paid on the publication date of the September issue.

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This antenna is for high signal areas only, but in these areas it will perform as well as any antenna on the market today. It eliminates ghost images, gives you strong, sharp pictures on all channels, and receives FM.

You don't have to worry about weather with the Econ-a-Ray Butterfly. It is constructed from Dural, with Polystyrene and stainless steel fittings, to withstand winds up to 75 mph, and is unaffected by snow, rain or any other weather conditions. It cannot corrode.

The Econ-a-Ray Butterfly will give you good television at the lowest cost possible. Ask your dealer for it today, or write for information. Remember, the Butterfly is a primary antenna only.

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- Dural elements solidly mounted in weatherproof polished Polystyrene.
- Perfect for low cost, high quality television in strong signal areas.

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ACCURATE ALIGNMENT of any TV Set with only one instrument

The New Model TV-30
TELEVISION SIGNAL GENERATOR only
$29.95 NET
Aligns Television I.F. and Front Ends Without the Use of an Oscilloscope or TV Sweep Generator

Save $150

TV SWEEP GENERATOR $80
OSCILLOSCOPE $100
Approx.

An investment of $29.95 for the TV-30 saves you $150. And besides the huge sum of money saved, you also save time. The TV-30 does a three-instrument test in a fraction of the time.

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GENERAL ELECTRONIC DISTRIBUTING CO.

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- Double shielding of oscilatory circuit assures stability and reduces radiation to absolute minimum.
- Provision made for external modulation by A.F. or R.F. source to provide frequency modulation.
- All I.F. frequencies and 2 to 13 channel frequencies are calibrated direct in Megacycles on the Vernier dial. Markers for the Video and Audio carriers within their respective channels are also calibrated on the dial.
- Linear calibrations throughout are achieved by the use of a Straight Line Frequency Variable Condenser together with a permeability trimmed coil.
- Stability assured by cathode follower buffer tube and double shielding of component parts.

SPECIFICATIONS
Frequency Range: 4 Bands—No switching
18—32 Mc, 35—45 Mc, 54—98 Mc, 150—250 Mc.
Audio Modulating Frequency: 400 cycles (Sine Wave)
Attenuator: 4 position, ladder type with constant impedance control for fine adjustment.
Tubes Used:
6C4 as Cathode follower and modulated buffer.
6C4 as R.F. Oscillator.
6SN7 as Audio Oscillator and power rectifier.
Model TV-30 comes complete with shielded co-axial lead and all operating instructions.
Measures 6" x 7" x 9". Shipping Weight 10 lbs.
Manufactured by Superior Instruments Co. under license agreement with Western Electric.

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### The New Model 200

**AM and FM SIGNAL GENERATOR**

**SPECIFICATIONS:**
- **R.F. FREQUENCY RANGES:** 100 Kilocycles to 150 Megacycles.
- **MODULATING FREQUENCY:** 400 Cycles. May be used for modulating the R.F. signal. Also available separately.
- **ATTENUATION:** The constant impedance attenuator is isolated from the oscillating circuit by the buffer tube. Output impedance of this model is only 100 ohms. This low impedance reduces losses in the output cable.
- **OSCILLATORY CIRCUIT:** Hartley oscillator with cathode follower buffer tube. Frequency stability is assured by modulating the buffer tube.
- **ACCURACY:** Use of High-Q permeability tuned coils adjusted against 1/10th of 1% standards assures an accuracy of 1% on all ranges from 100 Kilocycles to 10 Megacycles and an accuracy of 2% on the higher frequencies.
- **TUBES USED:** 12AU7—One section is used as oscillator and the second is modulated cathode follower. T-2 is used as modulator. 6C4 is used as rectifier.

*The Model 200 operates on 110 Volts A.C.*

Comes complete with output cable and operating instructions.

**$18.85** NET

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### Superior's new model 770

**AN ACCURATE POCKET-SIZE VOLT-OHM MILLIAMMETER**

(Sensitivity: 1000 Ohms per Volt)

**FEATURES:**
- Compact—measures 3/4" x 5/8" x 2 1/4".
- Uses latest design 2% accurate Millivoltmeter type meter. Same zero adjustment holds for both resistance ranges. It is not necessary to adjust when switching from one resistance range to another. This is an important time-saving feature never before included in a V.O.M. in this price range. Housed in round-cornered, molded case. Beautiful black etched panel. Depressed letters filled with permanent white. Insures long-life even with constant use.
- **SPECIFICATIONS:**
  - **A.C. VOLTAGE RANGES:** 0—15/30/150/300/1500 VOLTS.
  - **D.C. VOLTAGE RANGES:** 0—7.5/15/30/75/150/300 VOLTS.
  - **4 D.C. CURRENT RANGES:** 0—1/5/15 MA. 0—1.5 AMP.
  - **2 RESISTANCE RANGES:** 0—500 OHMS, 0—1 MEYCOHMM.

*The Model 770 comes complete with self-contained batteries, test leads and all operating instructions.*

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**SUPERIOR'S NEW MODEL TV-10**

**SPECIFICATIONS:**
- Tests all tubes including 4, 5, 6, 7, Octal, Lock-in, Peanut, Bantam, Hearing Aid, Thyratron, Minatures, Sub-Miniatures, Novaks, etc. Will also test Pilot Lights.
- Tests for "shorts" and "leakages" up to 5 Megohms.
- Uses the new self-cleaning Lever Action Switches for individual element testing. Because all elements are numbered according to pin-number in the B.N.A. base numbering system, the user can instantly identify which element is under test. Tubes having tapped filaments and tubes with filaments terminating in more than one pin are only tested with the Model TV-10 as any of the pins may be placed in the neutral position when necessary.
- The Model TV-10 does not use any combination type sockets. Instead individual sockets are used for each type of tube. Thus it is impossible to damage a tube by inserting it in the wrong socket.
- Free-moving built-in roll chart provides complete data for all tubes.
- Newly designed Line Voltage Control compensates for variation of any line voltage between 96 Volts and 100 Volts.

The Model TV-10 operates on 105-130 Volt 60 Cycles A.C. Comes housed in a beautiful hand-rubbed oak cabinet complete with portable carry case.

**$39.50** NET

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**GENERAL ELECTRONIC DISTRIBUTING CO. DEPT. RC-8, 98 PARK PLACE, NEW YORK, N. Y.**

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*www.americanradiohistory.com*
FUSE HOLDER
Litetfuse Corp.
Chicago, Ill.
The Snap-On TV fuse holder is a fiber attachment which slips on the leads of blow fuses. It eliminates the usual difficulties of replacing such fuses. The pigtail of the old fuse and the fuse holder become permanent attachments in the set. Each time a new fuse is needed, it can be slipped into the holder.

TV OSCILLOSCOPE
Sylvania Electric Products Inc.
New York, N. Y.
The new type 400 oscilloscope, with 7-inch type 7DF green-screen cathode-ray tube, has a vertical sensitivity of 10 millivolts per inch and a vertical response useful up to 4 mc.

Other features include: four-position frequency-compensated attenuator for uniform frequency response at any gain setting; vernier gain control; low internal hum level; internal 60-cycle sine-wave sweep which eliminates one set of leads during TV alignment operations; laboratory shield; integral control box; and low-parallel, cross-tined screen.

OSCILLOSCOPE
Triplet Electrical Instrument Co.
Bluffton, Ohio.
The Triplet model 3410 5-inch oscilloscope for TV and general use has a pattern-reversing switch, a calibrated extra rod mounts to permit slab addition and rearrangement of elements to provide combinations for any signal strength area. Preassembled of aluminum and cadmium-plated steel, the antenna matches 71, 150, and 300-om transmission lines.

TUBE TESTER KIT
Electronic Instrument Co.
Brooklyn, N. Y.
Conventional receiving and TV tubes including 4, 6, 7, large and small octal, metal, and electron-ray tubes as well as pilot light lamps can be tested with Eico model 425X tube tester kit. A blank space socket is provided for future new tubes, and grid caps are provided for convenience, and a protective bulb individual and transformer overload and acts as a fuse. An illumination and driven speed control chart is provided for setting up any type tube.

NEW V.T.V.M.
Radio City Products Co.
New York, N. Y.
An electronic balanced bridge-type push pull circuit with an input impedance of 25 megohms is found in the model 654 V.T.V.M. It makes both a.c. and d.c. measurements and has a discriminator alignment scale with a zero center.

It can accommodate antenna masts from 4.4 to 57 inches in diameter.

LOW-LOSS LEAD-IN
Gosset Company
Burbank, Cal.
Using polystyrene spacers to minimize dielectric losses and 1-inch spaced to minimize line pick-up and radiation losses, Gosset Line has only 5.6 db loss per 100 feet at 200 mc and the losses do not increase appreciably under unfavorable weather conditions and aging. Its impedance is 100 ohms. Gosset Line is particularly well adapted for fringe area installations and where long runs of lead are required. The low attenuation permits runs in some cases as long as 200 feet, to bring TV reception to locations previously unable to get satisfactory reception.

ANTENNA ROTATOR
U. S. Devices
Plainfield, N. J.
Available in two models, the 501 with an end-of-rotation light and the 502 with a directional indicator which rotates 360° both clockwise and counterclockwise, this combination antenna rotator and control box can handle any TV antenna or stacked guy.

The rotator has an aluminum main casting, a heavy-duty tandem motor, a magnetic brake release, bell thrust bearings, and insulated strain relief.
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See your Stancor distributor today for your free copies of these books. If he is out of stock, we'll be glad to send you copies. Write Standard Transformer Corporation, 3592 Eison Avenue, Chicago 18, Illinois.

PROTEST LICENSE BILL

More than 1,000 New York City radio service technicians crowded the auditorium of the Central Commercial High School to protest the proposed licensing bill recently introduced into the City Council. The bill, introduced by Councilman Chas. Keegan, would require a $500 license fee from each service shop, and provide that no shop with a staff of less than 1 person or less than 500 square feet of floor space would be permitted to service television receivers. Every organization (no matter how large or small) must employ one person who has been certified by the Commissioner of Licenses as a qualified television technician.

Among those speaking against the proposed bill were L. B. Calamaras, representing the National Electronic Distributors Association, Tim Alexander, Motorola service chief, who spoke for the servicing committee of the Radio Manufacturers' Association, and John F. Rider, who besides speaking for himself brought messages of opposition from the New York district head of the American Legion and from Bruno-New York, distributor of RCA radios and televisions.

Numerous speakers from the floor joined in denouncing the bill, and the forum continued till well after eleven o'clock. Chief point made was that it would not prevent or reduce the two evils it complained of: poor and unskilled workmanship, and the loss of sums of money paid in advance for yearly service contracts. On the contrary, it would tend to drive out of business the smaller independent service shop which as a rule does not do contract work.

A number of speakers agreed that some measure of protection should be provided for purchasers of service contracts, but that licensing was not the correct approach; in fact the bill made the direct provision for protecting the customer, merely providing that financial statements be filed with the license commissioner annually (at a fee of $25) with interim statements to be demanded at his discretion as often as he would require. (These interim statements would be accompanied with a fee of only $10.)

The meeting was held under the auspices of the Associated Radio-Television Technicians Association of New York City, John F. Rider Publisher, the local section of the National Electronic Distributors Association, and the district organization of the American Legion.

TEXAS HAMS TO MEET

The twentieth Annual West Gulf Division Convention (ARRL) will be held in San Antonio, Texas, at the Gunter Hotel, August 18, 19 and 20 under the sponsorship of the San Antonio Radio Club. Features include choice of preregistration prizes, a hidden transmitter hunt, and special prizes and entertainment for the ladies, including a visit to the historic Alamo.

RADIO-ELECTRONICS for
PRSMA TO HOLD  

Second annual radio, television, and electronics exhibit sponsored by the Philadelphia Radio Service Men’s Association will be held at the Broadwood Hotel, Philadelphia, from September 25 to 27. Sessions will start at 7 pm on the 25th and will run from 10 am to 10 pm the next two days.

Purpose of the exhibit, according to Dave Krantz, show committee chairman, is to acquaint service technicians, dealers, and other interested groups with the latest developments in electronics. Leading manufacturers and others have engaged booths, and a program of educational lectures is being worked out. The first evening of the show will feature addresses by key industrial figures and civic officials, after which the trade displays in the main ballroom of the Broadwood Hotel will open.

The educational program, which was arranged by a panel of trade per-

TISA ADOPTS CODE OF ETHICS  

Attempts by the Television Installation Service Association of America to enlist the aid of television manufacturers and distributors in a campaign to clean up service abuses have so far been unsuccessful. At a meeting of service representatives from Chicago, Toledo, Detroit, Philadelphia, and other cities a code of ethics designed to raise service standards and guarantee good workmanship was adopted unanimous-

WCAU-TV PLANS  

Station WCAU-TV of Philadelphia has invited service organizations and others to present the case of the service technician and the television viewer in open forum discussions. One such program was scheduled (at the time of writing) for late June, and others may follow if response shows that the audience is interested.

The organizations taking part are the Philadelphia Radio Service Men’s Association, the Federation of Radio

RADIO-TV SHOW  

sonalities including Lewis Winner and Fred Shunaman, editors, Sanford Cowan and John Rider, radio publishers, and Stanley Myers, John Zagury, Dave Krantz and Harry Bortnick representing PRSMA, will differ considerably in presentation and subject matter from earlier forums and lecture series. Separate sessions will be held for dealers, service technicians, and students, and the subjects will deal with matters of immediate present and future interest to the groups addressed. Among the subjects tentatively chosen are: “Problems of U.H.F.,” “Radio vs. TV,” “Television,” and for the student, “Where Do You Go from School.”

It is expected that displays will be exhibited by the Signal Corps, the Armed Forces and the Bell Telephone System and that color television will be demonstrated by different manufacturers. An attendance of more than 8,000 is expected.

SERVICE FORUMS  

Servicemen’s Associations of Pennsylvania, the Philadelphia Television Contractors Association and the Philadelphia Better Business Bureau. They will discuss such questions as “How much can the television owner expect from his receiver? When should the service technician be called? What can the service technician do and what can he not do? How much truth is there in the current ‘television racketeering’ rumors?”

AUGUST 1950
You'll save on tools and time with the new Weller Soldering Gun WD-250. Whether the job is rugged or delicate, your Weller Gun does it with the same ease and efficiency. Chisel-shaped RIGID-TIP provides more soldering area for faster heat transfer. New “over-and-under” terminal design gives bracing action to tip. Your Weller Gun is lightweight and compact, gets into the tightest spots.

Weller Guns actually pay for themselves in a few months. Fast 5-second heating saves time on every job. Trigger-switch control saves power—no need to unplug gun between jobs. Prefocused spotlight and longer length mean easy soldering, even when the job’s buried deep. No other soldering tool gives you so much time-and-money-saving features. Order your new 250-watt Weller Gun from your distributor today, or write for bulletin direct.

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 Do BOTH WITH
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5-SECOND HEATING
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DUAL HEAT
single heat 200 watts; dual heat 200/250 watts; 115 volts 60 cycles.

WELLER MANUFACTURING COMPANY
828 Packer Street, Easton, Pa.

REVIEW OF NEW TUBES

As in recent months, most of the new tubes for this month are designed for television applications. Hytron has announced the type 1X2A half-wave high-voltage rectifier for picture tube anode supply. This miniature filamentary type tube is similar to the 1X2 but has somewhat higher ratings. In typical operation as flyback pulse rectifier the filament potential is 1.25 volts; d.c. output potential is 14 kv; and output current is 175 microamperes.

Another new Hytron tube is the type 12R5 tube, which has double period having semi-high permeance. One section of the tube can be used as a vertical deflection amplifier for picture tubes having high deflection angle. In typical class-A operation its amplification factor is 21; and transconductance is 6,200 umhos. It can supply a sweep height of 10% inches on a 16-inch picture tube.

Sylvania's type 6BF7 is identical to the type 6BG7 described last month except for its mounting. The 6BF7 has tinned leads for wiring directly into the circuit, while the 6BG7 has short leads for use with an 8-pin subminiature socket.

A twin triode, the 6SN7-GTA, is being produced by General Electric. The principal difference between this tube and the 6SN7-GTA is its increased maximum ratings to make it suitable for a combined vertical oscillator and vertical deflection amplifier for television receivers. Maximum plate dissipation is 5 watts, maximum plate voltage is 500 volts, and maximum heater-cathode voltage is 200 volts.

A 17-inch rectangular picture tube, the 1AP4, is announced by Du Mont. The tube has a 160-square-inch picture size and features a single-magnet beam binder in its bent-gun construction. Its anode voltage is 12,000 volts; grid No. 2 voltage, 300 volts; focus coil current, 115 ma, and grid No. 1 circuit resistance, 1.6 megohms max.

A grid-controlled inert-gas rectifier by Westinghouse, the type WL-5796 thyatron, is designed for industrial control and ignitor firing service. For general control service maximum cathode current is 20 amps peak and 1.6 amps average. For ignitor firing service, maximum cathode current peak is 30 amps and average is .5 amps. For both uses the minimum negative control-grid voltage is 250 volts before conduction and 10 volts after.
NEW EICO Instruments and KITS give you Laboratory Precision at Lowest Cost!

VACUUM TUBE VOLTMETER

Versatile top-quality laboratory-precision VTVM for trigger-fast operation and lifetime dependable service. 15 different ranges, Large rugged 4½" meter, auto-burnout circuit. New zero center for TV & FM discriminator alignment. Electronic AC & DC ranges: 0-10, 100, 500, 1000 v, 150,000 volts and 200 MC with HVP-1 and P75 probe. Ohmmeter ranges, 2 ohms to 1000 meg, 6½ digit. New stable double-needle balanced bridge circuit—extreme accuracy. 26-meg DC input impedance. Protective overload bulb, Electronic resetter, HandiDrive Scouter-etched, rubproof panel, rugged steel case. 115 v., 60 cycle AC, 0.71/2 x 9.8 x 8.0".
Model 221-K, KIT, only $23.95
Model 221, factory wired, $49.95

NEW TUBE-TESTER

The brand new professional tube tester and merchandise EICO Service-Engineered for unbeatable values!

Large 4½" full-scale meter. Tests conventional and TV tubes including, tube microwave. New information switches—tests every tube element. Illuminated "Speedrail Chart": 2 grid cavities. Short and open-element tests. Spare sockets for new tubes. Protective overload bulb, Electronic resetter, HandiDrive Scouter-etched, rubproof panel, rugged steel case. 115 v., 60 cycle AC, 12½ x 9½ x 4".
Model 625-K, KIT, only $29.95
Model 625, factory wired, $44.95

HIGH VOLTAGE PROBE

A new professional EICO-engineered 200-kv probe carefully designed and built for extra safety and stability. Extends range of VTVMs and voltmeters up to 250 kV. Handle, trigger-action, interlocking, shockproof.
Model MVP-1, only $6.95

NEW BATTERY ELIMINATOR & CHARGER

Model 1040-K, KIT, only $22.50
Model 1040, factory wired, $39.95

Electronic Instrument Company, Inc.
276 Newport St., Brooklyn 12, N.Y.
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NEW DELUXE KIT

AM-FM-TV SIGNAL GENERATOR
A laboratory-precision generator EICO Service-Engineered with 1% accuracy. Extremely stable, frequency 75 KC—150 MC in 7 calibrated ranges. Illuminated harmonic exciter tuning. SR stabilized line supply. 400-cycle pure sine wave with less than 3% distortion. Tube complement: 645, 777, 6C4, 6F6-150. HandiDrive 3-color etched rubberproof panel, rugged steel case. 115 v., 60 cycle AC, 12 x 13 x 7".
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Model 315, factory wired, $59.95

See these other EICO Instrument KITS:
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Model 360-K Sweep Gen KIT, $29.95
Model 145-K Sig Tester KIT, $18.95
Model 511-X VOM KIT, $14.95

5" PUSH-PULL OSCILLOSCOPE
All new laboratory-precision scope gives you all the extra sensitivity and response for precise servicing of TV, FM & AM receivers. Push-pull undistorted vertical and horizontal amplifiers. Balanced sensitivity, 0.1 to 1 ms ranges. XHI to 2.5 MC. TV-ray multivibrator sweep circuits, 15 cps—75 KC. Zero instability modulation frequency. Dual positioning controls move trace anywhere on screen. Complete with 2415, 50047, 2515, 501 CRT. HandiDrive 3-color etched rubberproof panel, rugged steel case. 115 v., 60 cycle AC, 9½ x 17 x 11½.
Model 425-K, KIT, only $39.95
Model 425, factory wired, $69.95

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Be sure to look at the EICO line before you buy any advertised equipment. From tube replacement parts to modern, engineering design, each EICO Kit and instrument is constructed with unequaled reliability. Specify the top-quality, professional equipment. You'll be satisfied—guaranteed.

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Deluxe 5th higher on West Coast.
This instrument may be used aboard ship, submarine, airplane, etc. to indicate the exact vertical direction, regardless of tilt or motion of the vehicle. The apparatus is simpler than a gyroscope which is often used for this purpose.

A pendulum P is used in the zimbal arrangement shown fixed to a base B. The two perpendicular axes AA and BB are free to move, therefore P (which is fixed to BB) is free to swing in any direction whatever. As P swings it causes motion of either AA, BB or both.

Both AA and BB control a linear potentiometer at one end and each has a magnetic lock at the other ends. The lock prevents excessive shaft displacement due to sudden shock or acceleration. The potentiometers are shown in the insert. As the shaft (either AA or BB) moves, it carries the center arm over a corresponding arc. Each potentiometer has its own d.c. input. At any particular instant the output from each potentiometer depends upon the position of the corresponding shaft. For example, if the shaft oscillates about a position of equilibrium, there will be an a.c. component superimposed over the d.c.

The output of each potentiometer is averaged out by an integrating network and then amplified. Since a pendulum swings in an arc, the amplified voltages are the same whether P is stationary or whether it oscillates. These voltages will depend only upon the average angle which exists between the pendulum rod R and the base rod R. Each output may be measured by a separate d.c. meter calibrated in terms of angular displacement of each shaft.

A more complicated arrangement is to use a second zimbal as shown within the dotted lines. This one is fixed to an extension of the base B. Each of its axes is controlled by the amplified voltages so that each undergoes the same angular displacement as the corresponding axes of the first zimbal. Thus if a pendulum, however, an indicator IND is used. The angle between R and C will be the same as that between D and E, therefore IND will always indicate true vertical.

BRIEFE COMPENSATOR
Patent No. 2,494,449
Henry T. Wilhelm, Long Island City, N. Y.
(Assigned to Bell Tel. Lobts., Inc.)

One difficulty in designing a precision bridge is overcoming stray capacitance. This is especially troublesome across large capacitors. Fig. 1 shows a typical Maxwell bridge for measuring inductance and series resistance. R2 is usually large compared with R1. G1 and C1 are the standard conductance and capacitance, respectively. The unknown inductance L1, and its series resistance, R1, are calculated from the bridge constants by the equations:

\[ R_1 = R_1 \times R_2 \times G_1 \]
\[ L_1 = R_1 \times R_2 \times C_1 \]

The stray capacitance, C, across R2 often tends to appreciable error. A novel and effective method of compensation is shown in Fig. 2a. A neutralizing capacitor C2 is connected between point B and a tap point E on resistor R2. It is well known that the Y network, Fig. 2a may be transformed into an equivalent as in Fig. 2b. Of the 3 impedances shown in Fig. 2b, BD is across the a.c. source and cannot affect the bridge. All comes out to be a small capacitance so it may be neglected in comparison with the relatively large C1 shunting it. The remaining arm AD becomes an almost pure capacitance when C2 and point E are properly chosen.

The inventor shows that this occurs when C2 is made approximately equal to K\(C_1\).

\[ \frac{K}{1} \]

where K = G3

For example, if E is the midpoint of G2, K = 2. Therefore C2 should be 4 (C1) to balance out the stray capacitance C. For exact balance C2 should be slightly variable. Alternatively, a small variable capacitance may be connected between tap E and ground for exact compensation.

CABLE IMPEDANCE MEASUREMENT
Patent No. 2,490,827
Fred A. Muller, Newark, N. J.
(Assigned to Federal Tel. & Radio Corp.)

The impedance of a parallel line or coaxial cable is quickly determined with this instrument. The principle is illustrated in Fig. 1. An r.f. voltage is connected across a bridge which has 2 fixed arms R1 and R2. The third arm is P, a length of transmitter line terminated in variable resistor R3. The fourth arm is another variable resistor R4. These two resistors are equal and are gauged. When R4 equals the impedance of TL, the resistance across P is equal to R4. Therefore the bridge is balanced and the detector D indicates

RADIO- ELECTRONICS for
null. This circuit is simple enough but the problem is to design a satisfactory R3 and R4. Each of these must remain purely resistive at high frequencies, say up to 100 mc and they must be adaptable to gradual variation and gearing.

![Diagram](image)

**Fig. 1**—The circuit of the r.f. bridge is conventional.

R3 and R4 are each made up as in Fig. 2. The primary and secondary circuits are resonated at the desired frequency by adjusting the capacitors.

**Fig. 2**—R3 and R4 are tuned circuits.

The impedance becomes a pure resistance, determined by the value of R4 plus whatever resistance is reflected into the primary by the coupled circuit. The resistance increases as the coupling is increased.

Actual construction of the resistance may follow that shown in Fig. 3. L2 (and its capacitor C2) is made to slide along a track as the knob is rotated. This controls coupling between the coils and therefore the resistance. By means of gears and a chain drive the same knob controls a similar arrangement in an adjacent shield box. To use the equipment, the calibrated dial is adjusted until the bridge is balanced. Then the cable impedance is read off directly from the dial.

**AUTOMATIC FIRE ALARM**

Patent No. 2,507,357
Paul B. Weiss, Swarthmore, Pa. (assigned one-half to Herbert Friedman, Arlington, Va.)

An open flame is detected in this device by a special, quick-acting phototube that does not depend upon direct heat radiation. The phototube P is a form of Geiger tube. It has an anode wire near a flat quartz window and a curved metal cathode. An atmosphere of xenon with 3% butane mixture has been found sensitive and self-quenching.

The cathode emits electrons when deep ultra-violet light falls upon it. Sensitivity is restricted to radiation below 3,000 Angstrom units, which is much lower than radiation from most artificial lights or from sunlight which has been filtered through glass. Electrons are attracted to the wire anode which is connected to a B-supply through a large resistor. About 10⁻⁶ amperes flow through this resistor, producing a drop of 0.1 volt.

The voltage to V1 is amplified to drive thyratrons V2. This closes the relay contacts to an external circuit containing an alarm. R controls the thyatron bias.

**Fig. 3**—Mechanical details of R3-R4.

Recent structural refinements give the WORKSHOP Double-Vee a sinewy ruggedness unmatched by other TV antennas. Coupled with the inherent strength of heavy gauge aluminum elements, the strong, new supports and fittings recently added provide a corrosion-resistant assembly equal to the toughest wind and weather conditions.

**MODEL VV**

Write for Bulletin D

**THE WORKSHOP ASSOCIATES, Inc.**

135 CRESCENT ROAD, NEEDHAM 94, MASS.

![Image of DUBL-VEE TV ANTENNA](image)

**DUBL-VEE TV ANTENNA**

*UNMATCHED STRENGTH with minimum Wind Resistance*

- **Clearer Pictures**—higher gain brings in stronger signal—especially on higher channels
- **Clearer Pictures**—narrow beam cuts down multi-path ghosts
- **Clearer Pictures**—better impedance match on all channels maintains high signal strength
- **Clearer Pictures**—true horizontal polarization—no out-of-phase ghosts
- **Clearer Pictures**—no parasitic elements—all driven
- **Clearer Pictures**—designed by the pioneers in the antenna industry

**$10.95 LIST**

Model 2VV Double-Stack $21.90 List

Specialists in High Frequency Antennas
Radio-Electronic Circuits

CALIBRATED BALANCE ATTENUATOR

Balanced attenuators are used in telephone lines, between stages of audio and broadcast equipment, and between signal generators and bridges and other equipment. Though such units are highly useful and have many applications, they are often too expensive for the average experimenter or service technician. A balanced attenuator providing up to 55 decibels attenuation in steps of 5 db in a 200-ohm line was described in Wireless World (London).

The circuit shown has an insertion loss of 5 db. Although preferred-value resistors are used, the actual values may vary above and below the indicated values by factors determined by the tolerances. Therefore, it is recommended that 5% resistors be used whenever possible. Check the components on a bridge or ohmmeter and match them in pairs. Matched pairs should be used in corresponding sections of the series and shunt elements. If one of the pairs is low, the pairs immediately before and following it should be high so the calibration error is not cumulative.

HIGH-SENSITIVITY VOLTOMETER

An a.c. voltmeter having sensitivity of 100,000 ohms per volt is described here through courtesy of Cornell-Dubilier Electric Corp.

The circuit is shown in Fig. 1 and the probe in Fig. 2. The meter, a Marion model 53RN 100-10 d.c. microammeter, has a resistance of 2,000 ohms which with the 50,000-ohm isolating resistor is subtracted from the calculated multiplier values for all except the 1,000-volt range. The multipliers are mounted on two 3 1/4 x 1 1/2-inch strips of 1/16-inch Bakelite. Resistors for the low ranges are on one strip and those for the 1-kv range are on the other. They should be selected to have the exact values shown on the diagram. An accurate ohmmeter or a bridge will prove useful in selecting the resistors.

Fig. 2—The test probe is well shielded.

(Surplus precision resistors are available from some dealers. Use them if they are available.)

The meter reads up-scale when positive voltages are applied to the probe. It can be read to 25 mv.

Other voltage ranges may be added or substituted for those on the diagram. Two or more resistors must be connected in series to make multipliers for each range. For 2.5 volts use 180,000 and 18,000 ohms; 5 volts, 470,000, 22,000, 2,700, and 100 ohms; 25 volts—24 meg, 91,000, 2,700, and 1,100 ohms; 50 volts—4.7 meg, 220,000, 27,000, and 1,000 ohms; 250 volts—12 meg, 12 meg, 91,000, 27,000, 10,000, and 1,000 ohms; 500 volts—22 meg, 22 meg, 4.7 meg, 91,000, 150,000, 27,000, 10,000, and 1,000 ohms. All the resistors may be 1/2-watt carbons.

BASS-TREBLE TONE CONTROL

This circuit has a single control which attenuates highs or lows depending on its setting. A typical application of this circuit is shown.

Maximum high-frequency attenuation takes place when the arm of the control is close to C1. The control has no effect when the arm is at midpoint.

Turning the arm to the end nearer to C2 removes the highs from the feedback network which consists of R1 and R2 thus permitting the feedback voltage to attenuate the lows.

If this tone control is used between two voltage amplifier stages, the feedback network must be adjusted to ac-

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count for the higher gain. Values of R1 and R2 may be varied to suit the circuit as long as R1 and R2 are equal.—V. Woychowski

HEADPHONE CONNECTION
I find this circuit to be an effective method of connecting magnetic phones to receivers, recorders, audio amplifiers, etc. Phones can be connected in parallel with a high-impedance circuit but they are likely to reduce its gain and alter its frequency response. Secondly, phones cannot be added to or removed from the circuit without affecting it.

The system shown uses a cathode-follower as the connecting link between the amplifier and phones. It does not load the amplifier or affect its response, nor does it require complex switching arrangements for adding or removing phones from the circuit. Furthermore, connecting extra phones does not affect the volume of those already in the circuit. Certainly the addition of a single tube and socket is worthwhile in high-quality equipment.

A separate volume control for the phones can be made by replacing grid resistor R with a volume control and connecting the grid of the cathode-follower to its arm.—Charles Erwin Cohn

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100 pages, invaluable information that will help you re-double the value of your basic test equipment.
CONVERTER TUBES AS AMPLIFIERS

Converter tubes which are no longer useful or serviceable as such, can often be used as voltage amplifiers in experimental circuits. The control grids of the oscillator and mixer sections are paralleled as are the mixer plate and the oscillator anode. The circuit shows how a typical converter can be connected as an amplifier. The screen and cathode dropping resistors and bypass capacitors may or may not be needed.

—Charles Erwin Cohn

SPEAKER INSTALLATION KINK

A number of home, portable, and auto radios have the speakers mounted some distance from the chassis and are connected to it through flexible leads soldered at both ends. The speaker leads must be unsoldered or the speaker removed from its mounting whenever the chassis is pulled.

I have made a practice of soldering a phono jack firmly to one of the voice coil terminals on the speaker frame and replacing the speaker leads with shielded wire and a phone plug whenever a set of this type comes in for repairs or a replacement speaker.

The cost of this modification is absorbed in a saving of labor during future jobs on the set.—W. H. Gibson

UNUSUAL SCOPE APPLICATIONS

If your scope has a ruled grid in front of the C-R tube, you can use it for comparing the strength of permanent magnets, or as an electroscope.

If a negatively charged body is brought close to the ruled plate while the horizontal sweep is on, the trace will open in the center. The separation will depend on the settings of the intensity and focus controls and on the potential of the charged body. For best results, keep the intensity control turned down and bring the charged body to the exact center of the screen. If the plate becomes charged, run up the intensity control for a moment and then return it to its original position.

To compare the strength of permanent magnets, turn on the horizontal sweep and adjust the focus and intensity controls to the desired level. Place the horseshoe magnet against the plate with its sides parallel to the face of the tube. The pole faces should be just under the horizontal center line and on the center line. The beam should be deflected upward; if not, turn the magnet over. The distance through which the beam is deflected is a measure of the magnet's strength.

—Walter J. Woltowetch
SAVING SOLDERING IRONS
I have found a way to prevent soldering iron tips from corroding and pitting when the iron is left on for long periods. Dissolve one ounce of potassium silver cyanide in a pint of water in a small plating tank or crock. Using the copper tip as the cathode and any silver object such as an old knife or fork as the anode, deposit a fairly heavy coating of silver on the tip. I use the circuit shown and allow the current to flow over night. This will not discharge the average battery because the current quickly tapers from a 300-ma rate to approximately 5 ma steady state.

![Soldering Iron Diagram]

One tip has been in service for at least 25 hours per week for several months without pitting or corroding. A quick wipe with a clean cloth removes any residue which may form on the iron. It has not been necessary to re-tin the iron since it's initial tinning immediately after the plating.—Arnold B. Margolis

(TAKE CARE! Cyanides are DEADLY POISONS and must be handled carefully. All operations should be carried out in a well-ventilated room or out of doors. Use rubber gloves in handling the solution and plating equipment. WASH the tank, anode, and work in running water.—Editor)

PILOT LIGHT TESTER
The tester shown in the diagram does away with the tedious process of juggling test leads when testing pilot lamps with an ohmmeter. I used a 1.5-ma meter, 1,000-ohm resistor, and a 1.5-volt flashlight battery. Other meters can be used by selecting a battery and resistor which will cause the meter to read full scale when the lamp socket is shorted. A good lamp will cause the meter to deflect to nearly full scale.

Separate screw and bayonet sockets can be wired in parallel or you can use a candelabra socket of the type used for 117-volt pilot and Christmas tree lamps. This type of socket will take screw- and bayonet-type pilot lamps.—W. H. Leitch

MICROPHONE COVER
Unused microphones are easily damaged by dust or dampness. Protect them with a double layer of Reynolds Wrap or similar metal foil.—Fred Scott

AUGUST, 1950

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In addition to the Model TR, General Industries will continue to offer the ever-popular Model TS, 3-speed neoprene belt-driven model for both manual and record-changer use.

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**Question Box**

**D.C.-OPERATED BOOSTER AMPLIFIER**

Please print a diagram of a booster amplifier delivering approximately 10 watts when operated from a 115-volt a.c. source. The input will be 1 watt into 500 ohms, and the output 1,000-ohm load to the plates of the tubes when connected across the 50-ohm tap. If you have several 16-ohm speakers, then two of them can be connected in parallel across a 24-ohm tap.

![Diagram](https://example.com/diagram.png)

Impedance should be 16 ohms. Can I use six 50L6's in push-pull parallel?—R.G.F., Hemlock, Texas

**A.** This amplifier should meet your specifications. The plate-to-plate load impedance is in the order of 1,000 ohms. It being hard to find a suitable output transformer with a primary impedance this low, a standard 15-watt output transformer with a 3,000-ohm center tapped primary and multistap secondary is specified.

A 16-ohm speaker or two series-connected 8-ohm speakers will reflect a similarly, three can be connected in parallel across a 16-ohm tap or in series across a 150-ohm tap.

The input transformer should be designed to couple a 50-ohm line to push-pull grids and be rated at 1 watt or more.

Heaters in each pair of tubes are connected in series with a dropping resistor. The resistor should be adjusted for a 50-volt drop across the heater in each tube. All grounds should be insulated from the chassis to avoid shock hazard. Allow air to circulate to keep the tubes and resistors cool.

**SIMPLE SUPERHET USES FEW TUBES**

**Please prepare a diagram of an a.c. superhet receiver using a 6SA7, 6SQ7, 6K6-4T, and a power supply. This set is to be used for local reception, and I do not believe that I'll need a separate i.f. stage.**—J. E. G., Winter Garden, Fla.

**A.** The circuit has been prepared to your specifications. We recommend that you use a high-gain i.f. transformer. The best bet is to purchase one having powdered-iron cores and ceramic or air trimmers. If space permits, select a large size transformer rather than a midget. Having only one i.f. and one r.f. transformer, the selectivity of the set will suffer in strong-signal areas.

The tuning capacitor should have 365 µuf per section. The oscillator coil should be designed for use with a 6SA7 or similar tube. The oscillator padder capacitor will probably be 250 or 400 µuf; however, follow the coil manufacturer's instructions for best results.

![Diagram](https://example.com/diagram.png)

A small set like this is a nice gift for a child or it can be used as an extra radio for any room in the house or easily carried on a trip.

---

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NEW 1950 SENCOR RECEIVING TUBE BASISING DIAGRAM CHART

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DIODE DETECTORS

Why must the diode load resistor of a diode detector have a high value?

—C. W. B., Washington, D.C.

A. The diagram shows the basic diode detector circuit. Each positive half of the input cycles produces a positive voltage pulse across the diode load resistor R and filter capacitor C in parallel. C charges to a value equal to the peak signal voltage minus the drop across the diode. R must be much larger than the reactance of C at the signal frequency, yet must be considerably smaller than the reactance of C at audio frequencies. Therefore, if C is too large, or R too low, high frequency response will suffer.

The efficiency of the diode detector depends directly on the ratio of the load resistance to the diode plate resistance, so the load resistance should be made as large as the a.f. and r.f. frequencies will allow for best efficiency.

The tuned circuit feeding the diode may be considered as being paralleled by a resistance equal to approximately 1/8 R. Therefore, R should be as high as practical to keep the power (E/R) absorbed by the detector low.

SHORTWAVE CONVERTER

Please design a converter to cover 18 to 31 mc when used with an audio receiver tuned to 550 kc. If you can design the converter to tune over a wider range while using standard components, please do so.—A. E., Forest Hills, N. Y.

A. If you use Miller type E-727-A and E-727-C coils or their equivalents in the antenna and oscillator circuits, respectively, the tuning range of the converter will be approximately 13 to 36 mc. The oscillator coil is designed for use with a 455-kc i.f. and no oscillator pad is used. However, if the converter must work into the equivalent of a 550-kc i.f. system a pad will probably be necessary. A .05-mfd capacitor is shown. It may be necessary to experiment with the size of this capacitor to insure tracking. A 15-muf capacitor is across the oscillator tuning capacitor for bandspread.

AUGUST, 1950

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GYROPHONIC PROJECTOR

Electronic organ tone deficiencies are generally attributed to the inability of the loudspeaker to provide the "mass" effect of a pipe organ or orchestra. To make the electronic organ sound less "electronic," the Allen Or- gan Company has developed a "Gyro- phonnic Projector," shown in the photo, which has in its speaker housing a ro- tating baffle on which are mounted two 25-watt speakers.

The rotating speakers provide tone dispersion similar to that of a pipe organ when several pipes are speaking at once if a single key is depressed. The Doppler effect of the rotation also gives the tone a certain indeterminateness of pitch similar to that of an organ or a large orchestra or choir.

The baffle can be rotated at three basic speeds. The lowest speed adds a singing quality as that of an organ or orchestra. The second speed produces a desirable celeste quality, and the third speed gives an acoustic tremulant that is superior to an electronically-pro- duced vibrato.

Electronic organs sometimes have more than one tone generator for a given tone to produce the ensemble effect of an orchestra. The Gyrophonnic Projector gives the effect of doubling or tripling the number of tone sources.

STUDENTS DEMONSTRATE TV

WOI-TV, at Iowa State College became the nation's first educational television station when it began opera- tion in February. The station marks a step forward in the 27-year-old develop- ment of educational broadcasting which has made Iowa State College a leader in the field with the most mod- ern and best-equipped broadcast facil- ities of any college in the country.

WOI-TV broadcasts programs for schools in the afternoons and general programs in the evenings each week- day. The area served is more than 55 miles in radius with a population of about 600,000 people in six major Iowa cities. The transmitter operates on channel 4 with a radiated output of 15.6 kw.

The television project has been under study since 1945, when Dr. Charles E. Friley, president of the institution, ap- pointed a committee to appraise the future roles of television and FM broadcasting for the school. The col- lege has long been mindful of the major role which WOI, the AM station at the college has played in off-campus activities.

As a result of the study, FCC per- mission to construct and operate a new FM station and a new TV station were applied for. When the permission was granted, the college became the first educational institution with permission to operate its own TV facilities. At present, Iowa State is the only college or university equipped to operate on all three of the modern broadcast media.

Video programs will be designed to serve the special and general interests of Iowa farmers, homemakers, schools, and adult education groups. Operation plans stipulate programming in three phases to provide maximum initial service at least cost.

In its first phase, WOI-TV will transmit all programs on film and slide projectors, making use of a 16-mm film crew and rapid entwickler devices for local programming. The second phase will begin with the purchase of a mobile television unit and a field camera which can be used for live broad- casts or remote sports events, meetings, classroom demonstrations, and labora- tory activities. The last phase will be the construction of television studios for live studio broadcasts.

In addition to providing programs to entertain and educate the public, WOI- TV will use its facilities to train young television engineers and program pro- ducers. Another area of activity will be research studies to determine the effect of television broadcasts and to find new techniques and methods in broadcasting.

Radio Thirty-Five Years Ago

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Electrical Experimenter

Radio News

Stereos & Television

Television

Radio-Printed Circuit

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RADIO NEWS

Wireless Association of America

August, 1916 Electrical Experimenter

Radio in N.Y. Police Preparedness

The Vacuum Detector and How It Serves, by H. W. Soster

New Radio Telephone Transmitter

Modern Radio Receiving Apparatus,

by Samuel Cohen

A 100 K.W. Radio Frequency Alter-

ator for Radio Telegraph

Long Distance Receiving With the

Audion Heterodyne, by A. S. Blatter-

man, B.Sc.

Sending on a Short Wave, by R. H.

Soster

Novel Detector Stand, by Harold S.

Dues

A Small Panel Type Radio Trans- 

mitter, by M. B. Sleeper

Making a Hughes Induction Balance.
SERVICE FOR FREDDIE

Funds are still pouring in for the Service to Freddie fund. Freddie, as readers will remember from the June issue (page 36) is the 2-year-old son of Herschel Thomason, Arkansas radio service technician, who was born without either arms or legs.

As reported in our July issue, Freddie has already been fitted with preliminary legs, which will have to be changed from year to year as he grows up. He is now beginning to use his artificial legs, but it will be many agonizing years before he can be rehabilitated to use first his legs, second his arms which are to be fitted later. This is a long uphill fight and a very expensive one.

Readers of RADIO-ELECTRONICS have joined together to help Freddie overcome this great handicap. To date June 22, $1836.00 has been collected to support him with artificial limbs.

No contribution is too large or too small in this campaign to help Freddie. Make all checks, money orders, etc., payable to Herschel Thomason. Address all letters to:

Help-Freddie-Walk-Fund
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Balance as of May 29.......$1057.50
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TESTER-
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- Tests all tubes including Nodal & sub-miniatures!
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BLOWER 115 Volt 60 cycle Blower, approx. 100 Cable 3/4" x 2 1/2" x 3/4" int, 2" outlet. Motor size: 3/4" x 3/4" at 1/2 HP, Copper coated, Brand New and Boxed. Order $7.95.

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AC ELECTRIC MOTOR—110 Volt. 60 cycle ball bearing, 1/150 HP. Shaft size: 3/16" x 1/8" Motor size: 7" x 4" Ideal for mobiles, fans, etc. Converted from Larr. surplus. Order $9.95.


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NEW TRANSFORMERS—CASED—115 Y.A.C. 60 CYCLE INPUT:

OUTPUT: 600-0-600 V.A.C., 3 amps. 12 V.A.C. 250 ma. 250 ma. Case.


NOW that summer is with us, the ionosphere is really bouncing those TV signals around and we've received a number of interesting reports from dxers. Most of the reports include KLEE TV, that cute little station in Houston, Texas, which seems to be operating on a nation-wide basis.

Mr. A. W. Crowl, of Welch, Louisiana, has sent in a list of 21 dx receptions in the past year and a half on his Air King TV, that cute little set with a 630-TS-type chassis. He uses one booster at the set and a remotely controlled booster at the antennas. The antennas are a Vee-DX RD-18A and a JFD double-stacked array cut near channel 3. A relay changes from one antenna to the other and a pair of 60-cycle, 110-volt alesa controls the booster as well as a rotating mechanism for the Vee-DX antenna. The antenna is 76 feet above the ground.

Besides KLEE-TV, some of the dx receptions which Mr. Crowl had with this setup were WBKB, channel 4, Chicago; WMAR-TV, channel 2, Baltimore; WJKB-TV, channel 2, Detroit; WWJ-TV, channel 9, Chicago; WTV-1, channel 4, Minneapolis; WBEN-TV, channel 4, Buffalo; WCBS-TV, channel 2, New York; and WFLF-TV, channel 6, and WPTZ, channel 3, Philadelphia.

One television dx viewer, Maurice Dubreuil of Lavaltrie, Quebec, found himself on the front page of the Sunday Section recently. La Patrie of Montreal devoted several pages to the story of a man who had never seen a television set on the air. The story is really quite interesting, and we hope that the following report will encourage others to get interested in television.

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ARMY AIRCRAFT RECEIVER—BC-496-B
Covers 500 Kc to 1500 Kc. Broadcast Band. Tubes: 8—12SK7, 1—
12FU7, 1—12K8. Designed for dynamotor operation; can be easily con-
verted to 110 volt or 22 volt use. Two IF Stages. Three-stage audio con-
version. BRAND NEW, in sealed carton, with tubes and instruction manual, less dynamotor...

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<tbody>
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<td>CVR.</td>
<td>$21.95</td>
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<tr>
<td>BC-454</td>
<td>CVR.</td>
<td>$9.95</td>
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<td>BC-455</td>
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<td>BC-456</td>
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<td>CVR.</td>
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<td>BC-458</td>
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The electronic equipment that saved many lives in the war. Set can be modi-

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2 AMP. Hour Rating...

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Made by Willard, for above storage batteries, 1 qt. sufficient for 2 two-volt cells. Hermetic...

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100-156 MC, 4 channels, XTAL-controlled, Ampli-
tude Modulated Voice, Electro-

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Suitable for 400 Mc citizens' band. Wire...

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New...

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Operates over a frac. range of 300-1600 MHz with nominal output of 10-30 watts. Comes equipped with 119 V. 56 CPS Blumen-

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transformer; blower; heavier wire test freq.

1-201A, 2-44AT, 2-44AT...

1-4L6G, 2-20P1 and 1-32SS (GL526 Oscil-

lato). Brand new in original carton, wading instructions, Manual...

www.americanradiohistory.com
If the volume control does not affect the audio output of the set, check to see that the grid caps are on the correct tubes. The lead from the first i.f. transformer goes to the 6K7 and the shielded lead to the 6Q7. If the leads are reversed, you cannot regulate the volume.

—William Gambone

A.C.-D.C. RECEIVERS

Always check the heater voltage on each tube in an a.c.-d.c. receiver when tubes are burning out too frequently. If one tube has a low-resistance heater, the voltage drop across the other tubes will be too high, reducing their life. Carefully check the oscillator tube. A low-resistance heater may keep it from oscillating or cause intermittent operation.—Peter J. Foradas

PHILCO 46-1201 AND 48-1201

If these models hum on radio and not on phonograph check the 15-µf capacitors in the voltage-doubling circuit before looking elsewhere for the trouble.—C. R. Luts

BELMONT CHASSIS 18DX21A AND 7DX21

These chassis include a dynamic limiter which is designed to reduce noise and external interference. It was later discovered that interference and noise were not as objectionable as expected, and therefore the dynamic limiter can be eliminated with an increase in a.f. sensitivity of approximately three times.

Remove the 1,000-µf capacitor C116 from between terminal 2 of transformer T9 and pin 6 on the 19T8 tube. If it is not convenient to realign T9, connect a 10-µf capacitor from terminal 1 to terminal 2 on the transformer. This small capacitor compensates for misalignment which resulted with the removal of C116. The diagram shows the location of the pertinent components.—Service Bulletin, Belmont Radio Corp.

TRUNETTE TX D1990 AND D2987

Complaint: Horizontal sweep is not stable or cannot be properly adjusted.

This trouble has been traced to two causes. The 1,500-ohm resistor in the plate circuit of the 12AU7 horizontal multivibrator may change its value enough to shift the natural frequency of the oscillator outside the range of the hold control.

Secondly, a 47-µf capacitor C-89 in the grid circuit of the horizontal multivibrator may be open or may be 470 µf. Replace this capacitor if its capacitance does not measure close to 47 µf.

—Sidney S. Goodkin

WIDENING TV PICTURES

This scheme can be used to increase the width of a picture on TV sets using kickback power supplies. Disconnect the lower end of the width coil from the transformer and connect it to position 1 of a four-position rotary switch. Connect the capacitors between the switch points and the other end of the width control as shown in the diagram. The width of the picture increases in steps as the switch is advanced.—Olaf W. Bailey

ADMIRAL 20A1, 20B1, and 21A1

In areas where two adjacent channels may be received, the sound from the lower may interfere with the picture on the upper channel. This type of interference may be reduced or eliminated by adding a sound trap to the second video i.f. amplifier.

Obtain a sound trap (part number T2A 88-1) and remove two turns from the coil at the end farthest from the slug screw, then resolder the coil to the lug. Do not remove the capacitor. Clip the white lead and bare tinned lead from the coil.

Remove the cover from the video i.f. strip and locate T302, the second video i.f. transformer. Wind 1½ turns (approximately 3 inches) of No. 24 or 26 insulated wire in a clockwise direction around the small-diameter portion of T302 at the end farthest from the slug screw. Loop one end of the wire under itself to hold the coil in place in a manner similar to the coupling coil on T301. Connect one end of the 1½-turn loop to the ground connection on T302.

On some chassies, there is a ¼-inch hole between the second and third video i.f. tubes. If the hole is not there, drill one and insert the new trap. Connect the black lead from the new trap to ground and connect the loose end of the 1½-turn coupling coil to the other lug on the trap.

Realign the video i.f. stages. Because of the shape of the video i.f. curve, it is difficult to set the new trap to 27.25
mc with a signal generator. Adjust the slug in the trap for minimum interference in the picture. Use a scope and sweep generator to make sure the video i.f. bandpass has not been affected by the trap adjustment. The sketch shows the connections for the trap.—Admiral Radio & Television Service Bulletin.

LEAKY 12JP4 C-R TUBES

Some 12JP4's develop leakage between grid and cathode after several months of operation. This condition causes horizontal tearing and loss of horizontal sync, which is not correctable by normal servicing methods. Substitute a new tube for a conclusive test. The internal short can sometimes be removed or burned out by connecting an induction-type spark coil between grid and grid pin on the tube.—Service Dept., Transcription, Inc.

PHILCO A.C.-D.C. SETS

Distortion after the set has been on for approximately 5 minutes can be caused by a leaky capacitor or a shorted 60BS output tube. Check the tube and capacitor because both may be contributing to the trouble.—A. O. Burden

TROUBLESHOOTING AUTO RADIOS

When an auto radio has a bad vibrator and a replacement is not immediately available, checking the rest of the circuits must wait until a vibrator is obtained. To save time when repairing the set or preparing for an estimate, remove the vibrator and connect the set to the 6-volt supply or storage battery. Using a standard 500-volt, center-tapped power transformer, apply 400 volts to each plate of the rectifier tube as shown in the diagram. This method can be used for a dead vibrator or open power transformer.—H. A. Blake

SILVERTONE 8117 and 8118

If these models start to fade after a few months of service, look for a defective antenna coupling capacitor. This is the .06-mfd capacitor (C1 on the manufacturer's service notes) mounted directly above the TCB power amplifier tube. The heat from the tube causes the wax to melt and run out of the capacitor, thus causing it to open. Replace this capacitor and insert a sheet of asbestos between the wax and the tube. This prevents a return of the trouble.—Floyd A. Roberts, Kearney, Neb.

It is advisable to use a plastic molded paper capacitor as a replacement. This type is designed for high-television service and is less likely to cause trouble in this position.—Editor

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With the latest 1950 improvements the 630 TV will out-perform all other sets in every way. The 30 plus tube circuit should not be compared to the cheaply designed 24 tube sets now being sold.

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  Assured by the new standard Tuner, which has a pentode RF amplifier and acts like a built-in High Gain Television Booster on all channels! The advanced 630 chassis will operate where most other sets fail, giving good performance in fringe Areas, and in noisy or weak locations.

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  Assured by use of the finest materials such as molded condensers, overrated resistors, RCA designed coils and transformers, etc.

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<tr>
<th>Price per set</th>
<th>品名</th>
<th>品牌</th>
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<tbody>
<tr>
<td>$39.50</td>
<td>Glass 16&quot; Black Face (round)</td>
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</tr>
<tr>
<td>$39.50</td>
<td>Glass 16&quot; Regular (round)</td>
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<tr>
<td>$49.95</td>
<td>Glass 16&quot; Rectangular</td>
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WRITE FOR COMPLETE CATALOG C-4

RADIO DEALERS SUPPLY CO.

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John G. Wilson, executive vice president in charge of the RCA Victor division, died June 1 in his Wynnewood, Pa., home after a brief illness.

A graduate of Northwestern University, Mr. Wilson served as a captain of artillery in World War I. He joined RCA in June, 1944, as administrator of accounts and finance of the Victor Division. In 1945 he was elected operating vice president. He rose to vice president and general manager in 1947 and executive vice president in 1948.

Vinton K. Ulrich has been appointed manager of the renewal tubes sales division of NATIONAL UNION RADIO CORP. A veteran of the radio trade since 1930, Mr. Ulrich was previously manager of Hytron's commercial engineering department.

James Calvin Affleck was appointed to the newly created post of sales promotion manager, receiver sales division of ALLEN B. ADAMS LABORATORIES, INC. For the past five years Mr. Affleck has been advertising and sales promotion manager for Radiomarine Corporation of America.

Robert B. Tomer has been appointed chief commercial engineer of the HYTRON RADIO & ELECTRONICS CORP. In this capacity he will supervise relations between the engineering divisions and customers.

Mr. Tomer has had 15 years experience in the radio and electronic industry. He was also instrumental in developing the proximity fuse and the Land camera. He has been associated with Hytron for the past three years.

Robert J. Cannon, eldest son of the founder of the CANNON ELECTRIC DEVELOPMENT CO., Los Angeles, has been elected president and treasurer of the company. He has served as vice president and general manager since 1942.

Burton Browne was awarded the degree of Doctor of Science in Business Administration by the University of Hollywood. Mr. Browne is head of the Chicago Advertising Agency bearing his name. The agency specializes in industrial advertising for radio and television products. The university also presented the degree of Doctor of Science in Audio Engineering to John K. Hilliard, chief engineer of ALTEC LANSING CORP.

Other officers elected were Richard L. Rowe, vice president and production manager, and John B. Milliken, secretary.

Caxton Brown resigned as chairman of the executive committee of the WESTON ELECTRICAL INSTRUMENT CORP. He will, however, remain on the board of directors and will be available to the company as consultant. Mr. Brown joined Weston in 1901.

Paul McKnight Deelely was elected a director of the CORNELL-DUBILIER ELECTRIC CORP. A recognized authority in radio engineering, Mr. Deelely was responsible for the design and installation of the first radio broadcasting station in Mexico. He has served as Cornell-Dubilier's vice president since 1932. During World War II, Mr. Deelely headed the capacitor division of the War Production Board.

Edgar K. Wimpy has been appointed director of quality control for HYTRON RADIO & ELECTRONICS CORP. He will direct the development and application of methods for controlling the quality of incoming materials, parts assemblies, and finished products. Mr. Wimpy joined Hytron in 1946.

Dr. Richard K. Cook, chief of the sound section of the NATIONAL BUREAU OF STANDARDS, has received the Washington Academy of Sciences Award for achievement in the engineering sciences by researchers under 40 years of age. He received this honor for his work in acoustics, particularly the development of an absolute method of calibrating microphones.

All incumbent officers of the STUART-WARNER CORP., including James S. Knowlson, president and chairman of the board; Frank A. Hiter, senior vice president; George L. Meyer, Jr., and Arden W. Le Fevre, vice-presidents, were re-elected to office. All incumbent directors were also re-elected.

Sam Norris has been elected president of AMPERE ELECTRONIC CORP. He had previously been executive vice president and has long been associated with the company.

Personnel Notes

Benjamin Abrams, president of EMERSON RADIO & ELECTRONICS CORP., was elected a director of the Better Business Bureau of New York City.

John S. Meck, Leslie Evan Roberts, Russell G. Eggo, and W. Adams were elected directors of SCOTT RADIO LABORATORIES, INC. Glenn E. Webster, former NBC operations supervisor, has been named distribution manager at Scott. C. E. Mead named vice president of LEAR INC. Daniel S. Rohrer assumed the post of advertising manager of CHANNEL MASTER CORP.

W. S. Hartford, general sales manager, and C. B. Dale, director of research of the WEBSTER-CHICAGO CORP., were elected vice-presidents in charge of sales and research, respectively.

RADIO-ELECTRONICS
IDEA FOR TV STATIONS
Dear Editor:
When television stations broadcasting test patterns are making adjustments to transmitters, they should make that fact known in some way. Too many stations have started to work on what we thought to be a bad receiver, only to find that the test pattern was distorted by the station itself during transmitter tuning, changing cameras or power, and so on. The vertical wedges play the mock-a-boo with you, the bandwidth changes every few minutes, horizontal interface goes crazy, and along about that time they decide to cut power in half.

My suggestion is that the letter T (or some other letter) be placed in the test pattern whenever it is not normal to notify the service technician that the fault may not be in the receiver. I have written this suggestion to all the local television stations and hope you will publish it to see what other service technicians think.

Wes Jayne
Woodhaven, N. Y.

GREEK VETERAN
Crippled by the war and now living on a meager pension, George Christodoulou, a Greek veteran of the Royal Hellenic Air Force, has expressed to the editors an earnest desire to study radio to become a good radio technician. He is now studying some books and past issues of Radio-Electronics we have sent him, but books and radio equipment are still difficult to get in war-torn Greece. If your Junk box is too full and your bookshelf crowded, why not give him a hand and send him the surplus. Any spare piece of equipment or parts that you can send him will be a big help in getting him back on his feet. The address is: George J. Christodoulou, Neon Eleftherochorion, Katerini, Greece.

500 FORMULAS TO SUCCESS
Many million dollar firms started with a single formula, for which they paid a crushing fee. Here are 500 ready-to-use formulas, recipes and processes for making things you use all the time. For example—there is your opportunity to start a business by making a product, which sells for a few dollars, but which will sell at 50 cents or a dollar and capital with little or no investment. The book you are also using in your business. You can also use these household expenses to the tune. If you can't save them to save on profit of your business, you might as well save them. You are saving money, buying not in your own business, but in the products you use. You can buy these savings in any book you like. You can also use these household expenses to the tune. You can buy these savings in any book you like. You can buy these savings in any book you like. You can buy these savings in any book you like.
BRITON PREFERS R-E
Dear Editor:

I have been a constant reader of your magazine for many years and wish to express my appreciation of your educational, constructional, and other interesting articles. I am not an American, but find that I personally prefer your publication to any other (including British) which may be on sale. I am of the opinion that, in general, British radio magazines (with possibly one exception) seem to have fallen into the old groove of repetition. Each month I can pick up a certain British magazine and be sure that 95% of the articles inside are rehashes with only minor changes of material previously published.

That particular magazine claims that it caters to the new experimenter in radio, which of course is essential to any radio magazine. But this is no excuse for only 5% of progressive articles being published. For this reason I prefer RADIO-ELECTRONICS, because I know that when I open its pages I will find an equal distribution of articles of past, present, and future, both theoretical and practical, and information on almost any piece of equipment connected with the forward trend of radio and TV.

To sum it up in as few words as possible, RADIO-ELECTRONICS publishes material for the guidance of the person who looks ahead, while most other magazines cater, on the whole, to the man who looks who looks already into his box of tools. When I open RADIO-ELECTRONICS forces the technical education of the individual to an ever-higher standard while still maintaining the interest of the newcomer to radio and TV.

E. W. MERCER
Liverpool, England

CRAZY LIKE A FOX
Dear Editor:

Your editorials are forward-looking and your predictions are usually borne out by facts—much to the chagrin of the people who scoff at Gernsback and his "crazy ideas."

The magazine hardly ever fails, too, to have something new, novel, unusual, to make the radioman think. You gave the transistor and the electret a big play; you have articles on the latest microwave developments, and a "New Patents" department, all of which keep the reader up to date and give him ideas for experiment. In fact, the magazine is highly imaginative, a much-to-be-desired quality. It is not a "how-to-do-it" magazine, but one that teaches basic theories so that the radioman has concrete ideas of what it's all about instead of blindly following instructions. It is a thinker's magazine.

PETER N. SAVESKI
Baton Rouge, La.

WANTS BETTER TV
Dear Editor:

Your directory of television receiver characteristics is of great value in estimating new sets.

P. B. WARD
Many of the new receivers are now using some form of intercarrier sound. Unfortunately, many of the manufacturers who use intercarrier are also reducing the overall bandwidth of their receivers.

Let's not try to kid ourselves. Video frequency components up to 4.6 mc are transmitted by TV stations. Although frequencies above 4 mc are attenuated at the transmitter, the receiver must be able to pass up to 4 mc to reproduce all the picture detail available in the signal.

Very few of the current intercarrier receivers pass 3.5 mc, and some don't recognize anything over 3 mc. Some manufacturers are using peaking coils in their video amplifiers to compensate for lack of bandwidth. Peaking coils are important in any TV set, but they can't replace high-frequency components which cannot get through preceding stages.

Frank M. Field

INDEX TO CONVERSION DATA
A 13-page booklet, Bibliography of Radio Conversion Articles, compiled by Elizabeth and J. Henry Lugg, is available upon request from U. S. Department of Commerce, Office of Technical Services, Washington 25, D.C. The booklet is free of charge.

Equipment is listed with the service nomenclature in alphabetical and numerical order. References give a description of the article and the issue and page of the publication carrying it.

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AUGUST, 1950

WASHINGTON, D. C.

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These new Gernsback Library Books belong on your bookshelf! Public-Address Guide tells how P. A. work can add dollars to your income every week. High-Fidelity Techniques by James R. Langham, RADIO-ELECTRONICS' popular audio writer, answers all the tough questions about audio—in your own language! Order these two new value-packed books from your distributor, today.

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DIAL CORD STRINGING GUIDE, DC2, compiled and published by Howard W. Sams & Co., Inc., Indianapolis, Ind. 5¼ x 8½ inches, 48 (unnumbered) pages. Price $1.00

The second volume of a series, this book begins with dial-stringing arrangement 558, where Volume 1 left off, and continues with 510 new diagrams representing dial cord setups in receivers manufactured since Volume 1 appeared some two years ago. The book uses the straight graphic presentation of the earlier book, with the diagrams numbered and an index by manufacturers.


Prepared for industrial plant managers, directors, and other nontechnical persons who wish to keep abreast of the latest developments in the field of industrial electronics, this book describes equipment which is used for heating, measuring, controlling, inspecting, and timing operations. Being non-technical in scope, it is devoted entirely to basic theory and possible applications of the equipment. No technical information, specifications, or diagrams are given.—R.F.S.


Similar to question-and-answer manuals designed for use in preparation for radio operators examinations, this book is a new and interesting approach to the home-study method of training for television. It is divided into twelve sections which are devoted to different phases of television. The sections are grouped in logical order which begins with antennas and related circuits and progresses through the receiver to the picture tube. Each subject covers regulations, transmitter theory, and the like.

The questions and problems are clearly presented and their answers are prepared in essay form with diagrams and formulas where necessary. Problems are solved by beginning with the formula and following it through step-by-step to the solution.

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