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Free Information Card—up-to-date information on the latest products
Who is out there?

Every editor flatters himself by pretending to know everything there is to know about his field of interest. Therefore, if electronics is used to make yogurt, I should know about it! So, to get to the point of this editorial; I received a very interesting article in the mail on a computerized model-railroad track power-supply. Now I know that some project builders will use 20 ICs to dim a light, and thus, I thought, using a microprocessor chip to control a model-railroad engine was too much!

But, I liked the idea and the article was prepared in excellent style. I asked the staff to read the manuscript and rave notices came from them. In fact, one staff member was into model railroading and thought that the project in question was the best to come down the track in a long time (to coin a phrase). I purchased it with some reservations.

The very next day I was in the hands of my chiropractor and he was chatting as I was grunting. I discovered he was into model railroading. In fact, he made very complex layouts that he bragged about. I let him test the power supply and he gave it rave notices. To make the story short, a section of his track layout appears on the cover.

At the photo studio I barely got past the front desk when a model-railroad fan stopped me as I was carrying the props to the studio area. And I could go on with a few other stories about instances where model-railroad buffs identified themselves once they knew that I was involved with their hobby.

My unofficial census tells me that there must be 50-million people in North America who are into model railroading. You see, now I know everything about model railroading and that the Comp-Throttle project in this issue is the best power supply a model railroader can build.

That's all for this month. For the next 30 days I may be looking into yogurt and electronics.

Julian S. Martin, KA2GUN
Editor
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MARCH 1988
Communications Break Down
I never miss your magazine, I have found your articles very informative, but unfortunately, I have not seen much in the area of communications. Your last issue touched frequency modulation as well as AM. Would you help us readers understand Pulse Modulation Techniques (PPM, PWM)?
Thank you for your help and keep up the good work.
—B.D., West Haven, CT

If you look back a little ways, you’ll find an article on modems in the September 1987 issue. It is some information on phase modulation, but unfortunately nothing on pulse-width modulation techniques. (You experts get your keyboards out.)

Are You Board
This letter is in reference to your June 1987 issue to the person who wanted to build a musical doorbell, and others like him. We are a company that designs PC boards. We are small, but do have several boards available for the hobby-minded. Our list includes musical-doorbell boards and speech-synthesiser boards.
Each kit has a PC board; double sided, silk screened, solder masked, of excellent quality. It also comes with manual parts placement and schematic diagrams (full size). All parts are easily obtained from local electronics stores.
We take orders by mail or phone and accept MC or Visa. Write to RFJ Engineering, PO Box 4166, Sanford, FL 32771; Tel. 305/323-9039.
—RFJ, Sanford, FL

Our readers are happy to know about you. One thing comes to mind though, assuming you don’t want to remain a small company, why don’t you advertise with us? Your competition does. (That goes for the rest of you in the hobby market place reading this, face it—you bought the mag.)

Good Reception
In reply to D.K., Waldorf, MD in the January 1988 issue on building a short-wave receiver. I would refer him to a book distributed by Electronics Technology Today. Their ads are carried by Hands-On Electronics.

From the book “How to Build Advanced Short Wave Receivers,” I have built two receivers and they perform very well. You can use many junk box parts if you have them. The book is easily understood and construction could be carried out by any average experimenter.
—G.W.T., Peoria, IL

That’s a good one for the Bookshelf column. We’ll review it in an upcoming issue. Thanks for your friendly advice.

Hooked, Not Hooked Up
I am writing this letter in the hopes you can help me with a problem I haven’t been able to resolve up to this time.
My hobby is music, and I’m interested in home-brewing an amplifier for an electric guitar. Unfortunately, I have seen nothing which would give instructions for constructing one. I’m not interested in a toy but a reasonably-powered guitar amp which I could use with a small group. Can you advise where I might get the plans? Is there a particular publication which might deal with this, or do you have any construction plans? I’d be glad to pay you if you have!

The other question is: in the event I can’t locate plans to build an amp, can I use a regular stereo amp to play the guitar through, using a “Y-adaptor” to play through both channels simultaneously? If so, do I need a preamp or any other equipment to adapt the guitar for playing through the stereo system?
Thanks very much for your help! This is the first time I’ve read Hands-On Electronics, and it looks like I very well might be hooked.
—S.D.C., Walton, KY

Checking the classifieds in the back pages is a good way to dig up suppliers for almost anything. I saw an advertisement for a company dealing only in amplifier kits with the address Amplifiers Unlimited, Rt. 2 Box 139, Highmore, SD 57345. Give them a try.
If that doesn’t work out, you can hook your guitar though a stereo as you say, but there are two things worthy of note: The phono input has a preamp, but adjusts the frequency equalization distorting your guitar, an auxiliary input may not have preamplification. Using a tape deck input on your amp is therefore the best way to go.
Be sure that you have speakers that can handle the power, especially if you use fuzz boxes or other distortion-creating techniques.

Tech Trek
I have bought the last several issues of Hands-On Electronics and have found it more informative than most comparable magazines. The educational quality of your special columns are great. Also, you publish some very interesting circuit projects. (I’ve built a few and of course, they work fine).
There is one thing I would like help with. I have been trying to locate information on where I can find a schematic or other circuit diagram (photofact, etc.) for either a dual—or single—trace oscilloscope.
If you should have any information or suggestions on such a schematic, please respond.
—M.H.R., Minneapolis, MN

The best you can do is go straight to the horses mouth: check with the manufacturer. Look for a mailing address on a part they manufacture. If you are uncooperative try asking a repairman what text he would use.

No Energy
The other day I picked up a copy of the December 1987 issue of Hands-On Electronics. What caught my eye was your article on replacing rechargeable batteries. I am an avid photo buff and recently I was given two Honeywell 770 strobes, two AC charger/adapters and four NiCd packs. What a bargain! But the NiCds in one of the packs was starting to leak a little. I immediately took them out and checked the other NiCds out. No damage to the flash units or the NiCd clusters whatsoever. Whew!!
Many circuits Mostly jar
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But the problem is finding the NiCds. I checked around and found a wholesale electronics place in town that has the sub-C NiCds with the clips. They are $6.75 apiece plus tax. Right now I need four cells. As I stated before, I checked the other NiCds out and they were not leaking, but they had dates on them of the late 70's, 1977, 1978, and 1979. I checked with the source that I got them from and found out that the dates were the replacement dates of the cells. So they are old but they still seem to charge pretty well.

I read your article and was somewhat confused. The pictures showed Radio Shack NiCds. At the end of your article you stated "Also, I discovered a few surplus mail order houses that sold odd size NiCd cells" but you never did mention names or addresses. Would you give them to me? I would appreciate it greatly. I am shopping for the best prices as I would like to replace all 16 cells. Kind of an expensive proposition if you know what I mean.

—S.M.L., Ames, IA

There are some C and sub-C cells for $4.25 available from All Electronics, PO Box 567, Van Nuys, CA 91408; Tel. 800/826-5432. Tell them we mentioned them in the letters column.

Colorless letter
My article in the December issue turned out great especially with the addition of color photos of the targets. However, there are errors in the captions of Figs. 4 and 5.

In Fig. 4, the signal Y is not “yellow” as printed, but denotes the luminance signal “Y” as detailed in the text. In Fig. 5, the word “yellow” should of course, be changed to “Y.”

I am certain that you will get a few letters from the readership, so I though I'd give them the answer to their letters ahead of time.

—Adolph A. Mangieri

Thanks for the correction, and you're right about the letters. One thing's for sure our readers are really on the ball. A Pat on the back to those of you who wrote in.

Giving 150 Percent
I have been an electronics experimenter since I was about 12 (1939). In the Hands-on Electronics, December 1987, about the article on modulation on page 80—I have had a different theory of what causes modulation.

What it amounts to is, on negative audio modulation the plate or collector RF final voltage is reduced to zero, therefore turning off the source of the carrier. On positive audio modulation positive voltage is added to the plate or collector, therefore turning off the source of the carrier.

Further, on positive audio modulation, positive voltage is added to the plate or collector, therefore increasing RF power, the only limit here is the amount of power the final can handle without burning out. Thanks for your fine magazine.

—F.J.S., Eiermont, CA

You're absolutely right. That view of modulation was used to create Taylor super modulators. The idea being that you can actually get over 100% modulation out of a transmitter.

However, there is a legal limit on a super modulator's modulation capacity. That is because their sidebands can have large amplitudes causing them to interfere with the side bands of other stations close to them on the dial.

Rights and Lefts
Concerning The Pirate Radio article in the December 1987 issue: It seems that if the operators of RNI were arrested in international waters, they have a good cause to sue the government and that their rights have been violated. The FCC nor the U.S. government should have jurisdiction in international waters.

It also seems that the FCC should allow the usage of some of "their" unused air space for small local operators with a music format, and low wattage.

Pirate radio changed the whole radio system of Europe, especially France who finally gave in and deregulated and allowed the pirate operators some leeway. In our country of controlled commercial format, I find it appalling, as well as pathetic, that the FCC is so afraid of private broadcast. With a little regulation to keep the psycho-factions from starting a propaganda wave, we, the intellectuals with good taste, could really have something nice. Besides that, if something doesn't give, those hardcore punk broadcasts to Russia might start again. Long live the 1st Amendment. Good luck RAI.

—A.A., Cincinatti, OH

It is interesting to note that the FCC exists for much the same reason as the maritime services do: traffic control. I think the FCC's job is to set up and police broadcasting standards, not programming ones. If I'm not mistaken they have done just that.

Cars are great, but would you want to drive without stop lights, or the qualifications the Motor Vehicle Department demands of drivers? I wouldn't like to have to listen to a radio station whose sidebands could possibly interfere with all other programming.

If you want to get into a business (and Radio is just that) then why not do it by the book? Surely all intellectuals appreciate order; to wit, is it permissible to yell fire in a crowded movie theater?

Picture Imperfect
We were pleased to find coverage of our color Video Printer in the November issue of Hands-On Electronics. However, the Freeze Frame article in the Gadget section mentioned several points which need to be clarified for your readers.

The article incorrectly mentioned a cost per print of $5, based on the accessory Paper/Ink set (20 print) which is included at no charge with the printer. The standard Paper/Ink set (VY-S100A, Retail $99.95) makes 100 prints at a cost of less than $1 per print. The Transparency Paper/Ink set (VY-T50A, Retail $99.95) makes 50 prints at a cost of less than $2 per print.

The picture memory noted was incorrect, but that was probably due to an error in a press release. Actual picture memory in the VY-100A is equal to 1,008,000 bytes consisting of 36 RAM memories of 224,000 bits each. In the VY-50A, 15 RAM ICs are used (about 420,000 bytes), since it memorizes and prints a single field. Composite video input from the Video Output of a camera, VCR, or other video source, is separated into red, green and blue, and converted to a 6 bit (5 bit, VY-50A) digital signal for red, green and blue. The six bits represent 64 levels of each color, so that 256,144 dif-
lerent colors can be represented in memory for each pixel (picture element). The digital signals representing each pixel are then stored in RAM. During printing, the RAM is ready to make the print. When the memory is pressed to store a new picture or the power is turned off, prior pictures are cleared.

—Tom Rocco, Technical Support Manager, Hitachi.

Thanks for the corrections and extra information. It would appear as though even photography will soon be all electronic (which is upsetting to some of our staff). While we’re on the subject of Gadget, we hope the readers will like the new Gizmo column replacing it. Please write in and let us know what you think.

A Bit More

I got in on the computer craze when it was in its infancy, upgading to new and better systems as they came along. Now I’ve got this great big IBM compatible computer, with multiple hard disks and floppy’s, controlling nearly every repetitive operation that you could imagine. I’ve even installed the Voice Synthesizer project that was presented in your December 1986 issue. However, I now wish to add a voice recognition system to my hookup. I’ve searched high and low, but thus far have been unable to locate a schematic diagram for such a circuit. Have you ever printed or plan to print such a project, or do you know of any other publication that has.

—Louis A. Sullivan, National City, MI

You got a good question there Louis, and you’re not alone. I too am interested in a voice recognition system for my computer, but, much to my dismay, simply can’t afford such a high-priced toy. Anyway, Louis, the answer to your question is no! But stay tuned, something may turn up in the future. Until then, why not try some of the other computer related projects that so often appear in these pages.

Hats Off to Hands-On

I never trusted product reports in magazines, because advertisers pay a part of your expenses. It was with a bit of apprehension that I decided to try the ByteSize Calculator software that you raved about in the February, 1987 issue. Gee, it was good. Thanks for an honest report.

—D.C., Cherry Hill, NJ

The Editor re-read the article and the "raving" part was actually removed when the two-page story was assembled. We ran out of space for the "raving" paragraph at the end. Nevertheless, we talked about what the program can do, and the features we found in ByteSize Calculator did all the selling. When a product is good, we take our hats off to it!

Help!

Having read the Electronic Dazer article, which appeared in your November 1987 issue, I decided to put one together for a lady friend of mine, to waylay her fears about traveling the streets of New York late at night. However, as I began collecting the necessary parts, I found it impossible to get a hold of the MJE731 transistor (Q2 in the schematic). Would you please tell where I can get one of the little monsters, or at least an ECG replacement.

—Alton Ripley, Fleetwood, NC

Well Alton, first off let me say that the problem you’ve been having in locating that transistor was brought about by a typographical error. The part number should have read MJE371. The MJE371 is available from Jameco Electronics, 1355 Shoreway Road, Belmont, CA 94002; 415/592-8097. And there is also an Sylvania replacement for it, the ECG185. Sorry for any inconvenience that the error may have caused you.
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**24-Bit Digital I/O Interface**

MetraByte's uCPIO-12 24-bit parallel digital I/O board provides a simple and inexpensive means of interfacing an IBM PC/PS/2 Model 50, 60 or 80 in a wide variety of digital-I/O applications. In addition to the 24 TTL/NMOS/CMOS compatible data lines, the uCPIO-12 offers full access to the PS/2's interrupt lines, and allows external connection to the computer's power supplies. For applications with other than logic-level signals, Metra-Byte offers the following signal conditioning boards: SRA-01 8-channel Solid-State I/O module board; ERA-01 8-channel electro-mechanical relay board; ERB-24 24-channel electro-mechanical relay board. The uCPIO-12 can also be easily interfaced to industry standard PB-16A and MS-16 solid-state I/O module racks.

The 24 Digital I/O lines are provided by an Intel 8255 programmable peripheral interface. The 8255 divides the 24 bits into three separate 8-bit ports (PA, PE, and PC), which may be set independently as inputs or outputs. In addition to operating as a standard 8-bit data port, the third port (PC) can be subdivided into two 4-bit ports, or can be set to provide handshaking signals for the other two 8-bit ports.

In keeping with the design specifications for the PS/2 adapter cards, there are no dip switches or user adjustments on the uCPIO-12. Base address and interrupt level selection are performed by the PS/2 set-up program.

Along with the uCPIO-12, MetraByte supplies a complete users manual, and a 3-1/2-in. disk with all required installation and operation software. The board has been designed to plug into any available Micro-Channel expansion slot within the computer.

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The uCPIO-12 can easily be configured to interface to BCD equipment, Monitor switches and contact closures, control relays, etc. Applications for the board include equipment control, key-pad scanning, printer interface, motor control, intruder alarms, energy management, interface to custom circuitry plus many, many more.

Pricing is as follows: uCPIO-12 $299.00; SRA-01 $109.00; ERA-01 $135.00; ERB-24 $395.00; STA-U, Screw terminal board $115.00.

For more information contact Metra Byte, 440 Myles Standish Blvd., Taunton, MA 02780.

**ECONOMICAL MULTI-OUTLET PROTECTORS**

Now you can have economical surge protection for your computers and other valuable electronics equipment by simply retrofitting any of your existing duplex wall outlet.

Model SP-300 upgrades your standard
The members of the Electronic Industries Association Consumer Electronics Group (EIA/CEG) through the Product Services Committee, has marketed the illustrated parts kit for vocational schools, educators and technicians. This is the same material used in the Digital and Microprocessor Course during EIA's summer workshop programs. These workshops are organized by the Consumer Electronics Group and co-sponsored by national service organizations and state departments of vocational education.

Parts and components are contained in a lightweight tool box with individual compartments. It includes a breadboard, power supply, pre-dressed jumpers, resistors, capacitors, and integrated circuits to perform all digital exercises 1 through 25 of the Digital/Microprocessor course book listed in the table of contents. Some parts have been included for the microprocessor section but other components will have to be acquired (as listed in the Introduction to Exercises 26–31).

Individual and classroom size quantities are available at the following cost: quantities 1–9, $69.95 each, quantities 10–19, $67.95 each, and for quantities 20 or more, $64.95 each (cost includes shipping and handling). The kits will also include the Digital and Microprocessor Course book. Additional books are available at the cost of $2.00 per copy.

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Chief among them is a dual speed capability that allows you to run both the new higher speed 3.5-inch disks and existing 5.25-inch software that may require slower speeds. You also have the option of built-in drives for both 5.25-inch disks and 3.5-inch disks in any configuration you want.

Additionally, Model 35 has six full-size expansion slots, giving you more than twice the capacity for add-on features than you’ll find in competing and even higher priced computers. And it offers you more speed than other comparable computers and more than twice the data storage capacity.

Model 35 includes enhanced state-of-the-art graphic capabilities. Your color displays will be clean and brilliant. Two hundred and fifty-six colors can be seen on the screen at any one time, from a palette of more than 256,000 shades. Monochrome displays are equally sharp with a range of up to 64 shades or gray. And Pro/One Model 35 comes fully equipped with built-in surge protection and EMI filters, so your protection is already provided.

Model 35 also offers you what no other computer can. As a special feature, it comes with Profit Technology’s exclusive BreakThrough Mentor software, a real-time multi-tasking package that enables you to display five or more programs on the screen simultaneously, and to perform interrelated tasks.

Mentor also serves as a unique information delivery system that can flash up key points of information from its self-contained business library and display them in sequential order in up to five windows on the screen.

Profit Technology backs its commitment with nationwide service and a two-year Warranty that exceeds industry standards. Extended Maintenance Agreements are available from authorized representatives in more than 500 locations nationwide. You’re assured of having your service requirements met on-site and at recognized centers.

For more information contact Profit Technology Inc., 17 Battery Plaza, 14th Floor, New York, NY 10004.

Fast ATE Multimeter

Speed plus Mate capability make the Solartron 7061 a strong contender for military ATE applications, such as the ADINTS (Air Defense Initial Navigation Test System) for which it has just been selected. The new version of the 7061 fully implements the CITL language specified by the Mate standard and does not require the use of any external hardware adapters.

The 7061 can capture 1500 readings per-second into an internal memory of 8000 readings, output 300 readings per second over the IEEE bus. respond within 3 ms to an external trigger, and signal an alarm condition in 1 ms. Measurements may be tracked until a defined input level is detected and tracking continues until a specified number of results have been collected. Memory then contains a record of measurements, before and after the desired event. An extensive math and statistics program can then process the output results in any desired format.
Tracking Generator

This unit allows swept RF measurements to be made directly, thereby eliminating the need for an external RF detector. The Tracking Generator's span width is selectable from 20 kHz to 10 mHz in a 1-2-5 sequence at output levels of -50 dBm or -15 dBm. That allows the test set to make measurements with a dynamic range up to 90 dB.

Voice Recognition System

Having a talking and listening computer is now an affordable dream. A low-cost version of the original Voice Master, the Covox Voice Master Junior, is a self-contained micro-sized speech digitizer and voice recognition hardware device that plugs into the joystick port of popular 8-bit computers.

The custom style case with built-in microphone allows for either hands-free desk-top use, or hand held, whichever is convenient. Speech or other sounds are recorded into memory which can be saved to disk. Digitized speech can be incorporated in Basic programs and played back without the Voice Master Junior attached.

For word recognition, recognition templates are made. Up to 32 words or phrases can be utilized at a time with others accessible from disk. Wedged into Basic are new command statements which make the Voice Master Junior easy to use. Bonus software includes the Covox Composer program, which lets the user write and compose music by merely whistling a tune. The score can be edited, saved, or printed out.

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If Voice Master Junior itself isn’t amazing enough, then its suggested retail price of $39.95 should be, which includes software and manual. Available for Apple II +, IIc, IIe IIgs (DOS 3.3 only), Commodore 64, 128 (in 64 mode), and Atari 800, 800XL, 65XE, 130XE. Contact Covox Inc., 675 Conger Street, Eugene, OR 97402; Tel. 503/342-1271.

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The standard features for the FM/AM-1200S include: FM/AM Signal Generator; 2-µV Receiver for FM/AM/SSB; Duplex Generator; RF and Audio Frequency Error Meter; Modulation Meter; RF Wattmeter; SINAD and Distortion Meter; 1 kHz and Variable Frequency Audio Sources; Oscilloscope; Spectrum Analyzer; 0.2 PPM TCXO; and Internal Rechargeable Battery.

For more information, contact IFR Systems, Inc., 10200 West York Street, Wichita, KS 67215-8935; Tel. 316/522-4981.

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Grantham College of Engineering is a specialized institution catering to adults who are employed in electronics and allied fields such as computers. These fields are so enormous that opportunity to move up is always present. Promotions and natural turn-over make desirable positions available.

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rust spots, cleaning electrical contacts, roughing surfaces before using epoxy or glue, cleaning lead in stained glass, de-burring small metal parts, cleaning stains off pots and pans, and a number of other household and office uses.

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Excell AA rechargeable batteries provide up to 50% more flashes, more music, more play time, yet are priced approximately the same as lower capacity batteries currently on the market. Other AA Rechargeables are only 450 to 500MAH. Excell AA's are 750MAH. Excell Batteries are backed by a 5 year/1000 charge warranty.
Compact-Disc Troubleshooting and Repair Guide
By Neil R. Helen and Thomas Bentz

This book discusses much more than how to repair a compact disc, it is a concise and complete introduction to the field of digital signal processing and compact disc players. The book describes in detail the hardware and technology involved in this expanding field.

The book clearly explains the audio signal and its problems, discusses basic digital theory, and presents hardware, lasers, circuits, adjustments, tracking and more. Its 208 pages include charts, tables and illustrations useful for beginning and experienced audio/video technicians and hobbyists.

Topics covered include: the audio signal and signal processing; basic digital theory, the CD modulation system, construction of the compact disc, the laser pickup, detailed CD circuit descriptions, adjustments and troubleshooting, the CD-ROM, glossary of terms.

Compact Disc Troubleshooting & Repair Guide, No. 22521, retails for $19.95 and is available at bookstores, electronics distributors or direct from the publisher by calling 800/428-SAMS or by writing to Sams, 4300 West 62nd Street, Indianapolis, IN 298-5400.

Using Microsoft Works
By Ron Mansfield

Microsoft Works, the popular integrated program for the Apple Macintosh, combines word processing, database, spreadsheet, and communications into one easy-to-use package. Using Microsoft Works, makes this program easy to learn and

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Combining Quick Start tutorials with more advanced tips and techniques, Using Microsoft Works explains both basic and advanced features of Works' four applications. This book also contains a review of basic Macintosh techniques, a hardware and software troubleshooting guide, and information on exchanging data between applications. Readers will learn to use all four Works modules: Word Processing—open word-processing windows, use

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different formats and fonts, work with multiple documents, and integrate graphics with text; Spreadsheet—create spreadsheets, select printing setups, manipulate cells, and produce effective graphs; Database—set up a database, enter and save data, search for data, and create and print reports; Communications—send and receive text, store and print complex files, and automate log-on procedures.

This book is available for $18.95 in most bookstores and computer stores. To order directly from Que, call 800/428-5331 and ask for a sales representative.

44 Power Supplies for Your Electronic Projects
By Robert J. Traister and Jonathan L. Mayo

This is a sourcebook that will make a valuable addition to any electronics library. For the novice looking for a practical introduction to power-supply technology, an advanced hobbyist in search of power supply designs for specific applications, or a technician in need of quick reference to power supply circuitry, 44 Power Supplies

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Francis J. Keenan and Christopher M. H."Advanced Power Supplies

By Robert J. Traister and Jonathan L. Mayo

This book contains 256 pages, costing $15.60 in paperback, available from Tab Books Inc., PO Box 40, Blue Ridge Summit, PA 17214; Tel. 717/794-2191.

WordPerfect Advanced Techniques
By Daniel Traister

WordPerfect is one of the most versatile, powerful word-processing programs available. Many of its advanced features were essentially inaccessible to most users until the release of Que's latest book, WordPerfect Advanced Techniques. This easy-to-follow text shows WordPerfect users how to do all the things they couldn't do before. WordPerfect Advanced Techniques reviews WordPerfect basics, provides troubleshooting tips, and explores the program's advanced features. The timesaving techniques and special applications presented in this book, including comprehensive coverage of the WordPerfect Library, turn WordPerfect into one of the most versatile word-processing programs available.

WordPerfect Advanced Techniques will show readers how to alter the program's default settings and choose from a number of start-up options, manipulate individual columns of text, create macros to automate complicated procedures, efficiently select, sort, and merge data, exchange data with 1-2-3 dBASE III Plus, and other popular

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programs, configure and customize WordPerfect for various printers, including PostScript supported printers and the Hewlett-Packard LaserJet line, exploit WordPerfect's desktop publishing capabilities, use WordPerfect with many of the more popular memory-resident programs.

The book contains 400 pages for $19.95 and is available from Que Corporation, 11711 N. College Ave., Carmel, IN 46032; Tel. 800/428-5331.

Hard Disk Management Techniques
By Joseph David Carrabis

This 250-page tutorial highlights installation of utilities, hardware, software and its applications for the experienced business professional working with a hard disk drive.

The book is software-oriented for the serious user as opposed to sophisticated programmer or developer. Each fundamental technique, based on the author's own consulting experience with Fortune 500 companies, is emphasized and explained in detail.

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Learn the Basics the NRI Way—and Earn Good Money Troubleshooting Any Brand of Computer

The biggest growth in jobs between now and 1995, according to Department of Labor estimates, will occur in the computer service and repair business, where demand for trained technicians will actually double.

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You start with the step-by-step assembly of the new, highly rated fully IBM compatible Sanyo 880 Series computer. You install and troubleshoot the "intelligent" keyboard. Then you assemble the power supply, install the disk drive, and add extra memory to give you a powerful 256K RAM system. The new 880 computer has two operating speeds: standard IBM speed of 4.77 MHz and a remarkable turbo speed of 8 MHz, making it almost twice as fast as the IBM PC.

Next, you'll interface the high-resolution monitor and begin to use the valuable software also included with your complete computer system.

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Topics covered include introduction to hard disks and DOS, servicing and maintenance, setting up and organizing a hard disk, hard disk managers, maintenance, file, and security utilities.

*Hard Disk Management Techniques,* No. 22580, retails for $19.95 and is available at bookstores, electronics distributors or direct from the publisher by calling 800/428-SAMS or by writing Sams, 4300 West 62nd Street, Indianapolis, IN 46268.

**Microprocessors in Industrial Measurement and Control By Marvin D. Weiss**

The microprocessor revolution has made tremendous impact on process measurement and control. It has made control systems more compatible with the plant environment and more sympathetic with the operator's needs. At the same time, it has meant a reduction in control equipment costs, start-up costs, and installation time. Now, Marvin D. Weiss provides engineers and managers with practical insight into control requirements for specific processes and shows how the microprocessor uniquely fills those needs.

Starting with a look at the roots of digital systems—including the relay, the tube, the transistor and diode—Weiss traces the evolution of control and automation in the process industries. He then shows how to determine where microprocessors can yield maximum dollar dividends for the process industry by improving quality measurement and control at any level of automation—from pneumatic and analog systems to microcomputer-based distributed systems and expert systems. Also covered is the development of control hardware and control software with a preview of what's in hardware and software technology.

*Microprocessors in Industrial Measurement and Control* provides insight into how the newest microtechnologies and trends are affecting the automation of processes in industry. Other topics highlighted include classification of levels of automation with examples of how microprocessors can aid automation at each level; measurement techniques and standards; distributed control systems; descriptions and examples of unit control; control methods for coal gasification and energy management; the use of micro systems for selecting or designing safety systems; micro-based pollution control systems...and much more.

*Microprocessors in Industrial Measurement and Control* is order No. 2818, has 622 pages, available from Tab Professional and Reference Books, PO Box 40, Blue Ridge Summit, PA 17214; Tel. 717/794-2191.

**Understanding Expert Systems By Louis E. Frenzel**

Louis E. Frenzel, Jr. uses a self-instructional format to lead the reader to an understanding of expert systems, and how they work, where they are used, and how to develop them. Readers will learn how an expert system can act as an intelligent consultant to answer questions, solve problems, and help make decisions.

Ideally suited for the technical professional or manager interested in artificial intelligence, the material is completely illustrated, clear, concise, and application-oriented with marginal notes and review questions and answers.

Topics covered include types of expert systems and their applications; knowledge representation and search; how expert systems work; expert system development tools; expert system tool vendors; and AI language vendors.

**The Benchtop Electronics Reference Manual By Victor F.C. Veley**

The real strength of this huge reference lies in its unique format, which divides each of 160 topics into a three-stage exploration: Basic Principles—a simple discussion of the basic concepts that gives newcomers to electronics a solid foundation in theory; Mathematical Derivations—in modern SI metric units)—relevant equations that help more advanced electronics practitioners to gain a deeper understanding of the subject; and Practical examples—giving interested readers concrete illustrations of each topic. The result of this format is an easy-to-use reference for electronics and communications technicians now working in industry and a valuable learning aid to high school and college-level students and hobbyists.

Topics are drawn from the most common subjects in electronics, including direct-current principles, alternating current principles, solid-state devices, tube circuits, and principles of radio communications;
The book begins with general information on how small gas engines work, describing the distinguishing features of small engines and presenting practical techniques for servicing, starting, operating, and storing these engines. Proper operation, maintenance, and attention to minor repairs are the keys to prolonging the life of a small engine, allowing it to run safely at its top productivity level. All of the routine maintenance procedures so necessary to smooth operation—such as lubrication and refueling—are described in complete detail. Readers then learn why certain accessories such as carburetor air cleaners, crankcase breathers, spark plugs, and storage batteries require frequent servicing, and they learn how to care for these items.

As an added bonus, this handbook is also a complete engine systems reference, explaining in step-by-step detail the operating principles, maintenance and repair procedures for starters; cylinders; fuel, piston-and-rod, and camshaft assemblies; ignition and lubrication systems; and valves. With the expert advice and detailed instructions in this useful guide, readers can solve almost any small engine problem—from minor adjustments and maintenance to complete overhauls.

Small Engines: Operation, Maintenance and Repair

By the American Association of Vocational Instructional Materials

With an estimated seventy million small engines in use in the United States today, small gas engines have definitely been established as the power source for everything from lawn mowers and chain saws to pumps and generators. And repair of these engines is a very profitable and booming business. Now, with this complete guide, anyone can learn how to service, operate, maintain, and repair these air-cooled, spark-ignition, 1/2-15 hp engines, to obtain maximum power and efficiency—and to save money.

The book contains 450 pages costing $19.95. To order directly from Que, call 800/428-5331 and ask for a sales representative.
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IBM Peripheral Troubleshooting & Repair Guide
By Charles J. Brooks
With this book in the Troubleshooting & Repair Series, personal computer owners, service technicians, students, and hobbyists can learn to service and repair computer peripherals with this step-by-step procedures manual.

Using WordPerfect Workbook and Disk
Learning to use WordPerfect is now easier! Que Corporation's latest release, the Using WordPerfect Workbook and Disk, will help novice users quickly learn the features and functions of the WordPerfect word-processing program.

Electrostatic Loudspeaker Design and Construction
By Ronald Wagner
For many true audiophiles, electrostatic loudspeakers are the Porsches of the speaker world. The units produce the ultimate in sound reproduction, but the cost of a unit can run anywhere from $3,000 and up. Now, with publication of Ronald Wagner's Electrostatic Loudspeaker Design and Construction, even budget-minded music enthusiasts can afford to own and enjoy their own custom-designed electrostatic sound.

I'm so depressed. My artificial intelligence is the only thing about me that's real.
Oscillators Simplified, With 61 Projects
By Delton T. Horn

An excellent learn-by-doing book that includes just about everything there is to know about oscillators in one compact volume. Although oscillator circuits are covered to a degree in virtually every text on basic electronics, materials beyond the basic level are often difficult to find. This book is a totally comprehensive guide on the subject, covering both theory and practical applications, at an intermediate level. Horn discusses not only sine, rectangle, and square waves, but also sawtooth, spike, triangle, and staircase waves. And he explains how to generate those waves with many active devices.

More than 60 projects provide readers with the designs for many different signal generator circuits, such as the Hartley oscillator, the Colpitts oscillator, a complex signal generator, a pulsed signal generator, a basic 555 astable multivibrator, a tone decoder/demodulator, a musical tone generator, and many others. Tips for adapting these basic circuits are also given to encourage readers to experiment with new designs on their own.

The second section is designed to function as a C language reference manual giving quick access to listings of the most frequently used C statements and functions as well as language operators, storage classes and data types, and compiler directives.

The final section with its C-toolbox library, however, will probably prove most useful to all levels of C programmers from novice to experienced. Here is a total of 196 ready-to-run functions and subroutines for such diverse uses as printer control, string manipulation, data input, array manipulation, measurement conversion, data functions, screen output functions, and mathematical functions.

The C Trilogy: A Complete Library For C Programmers
By Eric Bloom

Far more than just another programming tutorial or collection of subroutines and functions, this is a complete, single-volume C reference library. It gives instant access to the full range of information needed by both novice and more experienced programmers. Just some of the highlights include C language rules, syntax, and standard functions, a discussion of C within the MS-DOS video, cursor placement, and more; PC printer control—with details on the double strike, compressed print, and other special printing requirements, structured programming techniques, data manipulation techniques, data manipulation techniques, accounting and financial calculating techniques, plus array manipulation techniques.

Secrets of 123 Classic Science Tricks and Experiments
By Edi Lanners

This fact-and-fun-filled book reveals the secrets and scientific principles of mind-teasing tricks and experiments sure to please any age group. A dancing egg, an electric walnut, a coin suspended in midair, drawing with fire, these are but just a few of the amazing "impossibilities" explained in Secrets of 123 Classic Science Tricks and Experiments.

Microprocessor Theory and Operation
A Self-Study Check With Experiments
By A. J. A. St. John

A respect for and interest in science can be cultivated through these entertaining projects. Step-by-step directions and illustrations make mastering these feats as simple as pulling a rabbit from a hat! No special skill or experience is required to perform the tricks in this collection. Only ordinary household items are needed for the experiments and games that demonstrate such phenomena as card tricks, centrifugal force, gravity, a hovering ladder, turning water into wine, coin tricks, inertia, mirror
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images, optical illusions, and a tower of matches.
Parents and teachers will find this a handy source of school and science-fair projects—principles of electricity, physics, chemistry and physiology are the basics for all of these fascinating—and educational tricks, experiments and games.
From first page to last, this book of easy-to-perform tricks and illusions will provide hours, even years of delightful fun!
Secrets of 123 classic Science Tricks and Experiments, order No. 2821, has 192 pages for $7.70 in paperback. Available from TAB Books Inc., PO Box 40, Blue Ridge Summit, PA 17214; Tel. 717/794-2191.

Introduction To Telecommunications System
By P.H. Smale
This guide will make the process of transmitting data electronically over long distances amazingly easy to understand—from the history, theory, and basic principles to performance characteristics, uses, current research, and the most recent technological developments.
Detailed information that is technical enough for electronics hobbyists, students, and those working in telecommunications, yet clear, practical, and useful enough for business and management students and professionals, Smale covers the whole range of telecommunication principles, beginning with a basic discussion of wave theory and continuing with the principles behind radio, television, telephone, and telegraph networks. Other highlighted and digital networks for national and international telecommunications systems.

CIRCLE 99 ON FREE INFORMATION CARD
Along with necessary background information on each subject, the author explains how each emerging technology has changed human communications and how telecommunications systems have evolved to accommodate the rapidly changing business, commercial, and industrial requirements of users.
Introduction to Telecommunications Systems, order No. 2924, has 160 pages, for $14.60 in paperback, is available from TAB Books Inc., PO Box 40, Blue Ridge Summit, PA 17214; Tel. 717/794-2191.

Microprocessor Theory and Operation--A Self-Study Guide with Experiments
By J.A. Sam Wilson and Ron Walls
Here is a long-needed sourcebook that effectively bridges the gap between digital electronics theory and microprocessor technology. Beginning with the heart of the computer, this guide explains bit-by-bit exactly how microprocessors and their circuitry work, and then examines the power supply and peripherals.
Wilson and Walls provide a detailed picture of the whole microprocessor system and how it functions, with specific information on busses, clocks, memories, and interfaces. Dedicated microprocessors, 6802 microprocessors, microprocessor instruction sets, and some special features of microprocessors are also fully examined—as are specific microprocessor systems: the 74181 ALU (arithmetic logic unit), the PLA (programmed logic array), and more. Finally, the authors look at the many different uses for microprocessors, how information is obtained, and what the microprocessor does with that information.
To provide readers with hands-on familiarity with microprocessors and their circuitry, 15 easy-to-build, inexpensive projects are included. These innovative designs range from a bounceless switch and a counter to more advanced projects that explore ROM, RAM, and interfacing elements. Full instructions are included for all 15 projects, and for the power supplies they require.
A review section is provided at the end of each chapter, ensuring that readers will master all of the material in the book, and an appendix and index make reference easy. Since understanding microprocessors involves more than just learning how the hardware is put together, a simple program is provided which demonstrates how each part of the microprocessor is used. Finally, Wilson and Walls include some useful programming tips and provide pointers on how to troubleshoot microprocessor systems.

80386—A Programming and Design Handbook
By Penn Brumm and Don Brumm
Exciting new standards in microcomputer power, speed, and versatility have been established by the revolutionary 80386—Intel Corporation's new 32-bit processor. In 80386—A Programming and Design Handbook, computer experts Penn Brumm and Don Brumm examine the impact of this advanced hardware on desktop computing, and detail information about the microprocessor's architecture, capabilities, and operation.
This handbook to the 80386 provides the facts prospective users will want to know before making a purchasing decision. The authors worked side by side with 80386 designers in producing this sourcebook.
to make certain the information and explanations included are complete and accurate.

The book covers 80386 design concepts and development history; 80386 architecture—block diagram, functional description, hardware/software compatibility, operational modes, pipelining, clock speeds, advanced features, and execution rates; memory organization and management overview, multitasking, multiple execution and concurrent operating system; performance tests/ratings overview; hardware protection mechanism, plus many more features are reviewed.

80386—A Programming and Design Handbook, scrutinizes every design and operating parameter of the microprocessor. To the fullest extent possible, programmers and designers' questions have been anticipated and answered.


Transmitter Hunting: Radio Direction Finding Simplified By Joseph D. Moell, and Thomas N. Curlee

Radio direction finding has been an important technology since its inception during World War I. But until recently most practitioners were scientists and engineers, who were primarily interested in generic studies of antenna theory and signal processing—little of which is readily understandable to the average hobbyist. Now that amateur radio operators have discovered this fascinating science, a great need has emerged for one comprehensive beginner's guide. This book is designed to fill that gap.

In the first chapters, Moell and Curlee review the important role radio direction finding has played during this century, and describe the many opportunities for fun and satisfaction that T-hunting provides. Then readers are guided through the beginning steps of setting up for the hunt, including an overview of mobile hunting techniques and the kinds of hardware now available. The authors even include complete plans for constructing inexpensive, easy-to-build antennas and Doppler units.

CIRCLE 99 ON FREE INFORMATION CARD

Readers receive practical guidance for involving other area hams in T-hunting, with ideas for rules, scoring, and some unusual types of hunts. Special topics include T-hunting in Europe and Asia; search and rescue hunting; commercial and military direction finding systems; and other uses for radio direction finding skills, such as wildlife preservation, cable TV interference searching, and finding power line interference. The authors also describe how a programmable calculator can be used to do accurate triangulation calculations in the field—and they include complete program listings, with explanations. Finally, the book takes a look ahead at what the future holds for transmitter hunting. A bibliography for further information and a complete list of manufacturers round out this all-inclusive reference.

There is a whole lot more to T-hunting than meets the eye—it is a public service, a technical challenge, a club builder, and a lot of fun. With this comprehensive and easy-to-understand guide and just one afternoon's work, anyone with an amateur radio license can begin locating downward aircraft, searching out radio frequency jammers, and enjoying the fascinating hobby of radio direction finding.

Transmitter Hunting: Radio Direction Finding Simplified, order No. 2701, 336 pages $17.60 paperback. Available from TAB Books Inc., 40 Blue Ridge Summit, PA 17214; Tel. 717/794-2191. (Continued on page 105)
ADVANCED DIGITAL COMMUNICATIONS: Systems and Signal Processing. Edited by K. Feher. 768 pp., 436 illus. Emphasizes the newest advances and developments in telecommunications systems and networks. Chapters on subjects such as ISDN, speech coding algorithms, digital speech interpretation systems and interference are all written by international authorities to give you on-the-job expertise.

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**CALL TOLL-FREE 1-800-2-MCGRAW**
EVEN WISH YOUR CHEAP PLASTIC-MODEL LOCOMOTIVE would run as smoothly as those expensive brass engines? Do your brass engines run poorly, because you have a budget power pack? Does your model railroad run like an inexpensive toy? Do your trains go from standstill to 60 miles-per-hour instantly when the throttle is opened? Do they skid to a halt when you brake or power down from 50 mph?

Well, Bucky, hold your head up high for here is a project that will help coax better running out of any locomotive. Now you will get Ooohs and Aaahs from visitors the instant that you throttle up!

That performance is made possible by using a model train speed-control unit ("throttle" or "power pack") based on a Motorola MC68705P3 Micro Computer Unit (MCU). You build it from the plans in this article. It is called the Compu-Throttle.

Having a microcomputer in the track circuit allows the operator to enjoy locomotive response to the throttle that duplicates the real thing. Now, the power output to the tracks using Compu-Throttle will be controlled precisely for optimum performance.

The author has been experimenting with the power control system for several years. The Compu-Throttle includes the Handheld Control Box with two pushbutton switches and a three position slide switch, plus the Power Unit with the remaining circuitry. The two units are connected by a four-wire cable. Hook-up to the track is via two wires, the same as done with a regular power pack. The electronic circuit of Compu-Throttle has three major circuit divisions: power supply, control circuitry, and power output. Each is discussed separately.

Power Supply Circuit

The primary purpose of the power-supply circuit for the Compu-Throttle is to provide a source of 16- to 24-volts DC for the output section to shape into controlled power that is supplied to the locomotive. The power-supply circuit also supplies +5-volt DC to run the computer and associated circuitry, and a +1.4-volt DC reference level. Refer to Fig. 1A. (Note: Figure 1 comes in three parts: IA, IB, and IC. This was done for ease in preparing the text for circuit discussion.) Bridge rectifier BR1 converts AC from transformer T1 (18 to 24-volt AC) to full-wave rectified DC, which is smoothed by capacitor C8. The electrolytic capacitor C8, is divided into two 1000-µF units for physical-size reason only. A small amount of ripple on the 18-volt DC output signal does not matter, but the +5-volt DC regulated supply to the computer circuit must be clean. Resistor R13 and capacitor C5 provide additional filtering and isolation from any possible motor noise that may be coupled back into the regulated supply. Resistor R13 also helps reduce the amount of heating on chip U5. Voltage regulator U5 provides regulated +5-volt DC. Capacitor C6 helps reduce noise on the +5-volt DC supply bus. Resistors R11 and R12 form a voltage divider to provide a

PLAYING ON THE RAILROAD WITH COMPU-THROTTLE

A single-chip microcomputer assists in procuring maximum performance from your model railroad locomotives.

By Eric W. Brom
+1.4-volt DC reference for U2, and capacitor C7 helps stabilize it. C1, C2, and C3, located near U1, U2, and U3, respectively, provide supply decoupling from noise spikes generated in each IC chip—U1 through U4.

**Control Circuity**

The control circuit for the Compu-Throttle (refer to Fig. 1B) consists primarily of the Motorola MC68705P3 microcomputer unit, U1 (MCU). It reads the inputs from the control switches through U3, and provides output signals which are amplified and shaped by U2 and U4. To keep the number of wires between the Power Unit and the Handheld Control Box switches to a minimum, they are scanned.

Wires 1 and 2 are outputs, while 3 and 4 are inputs. The wires are part of a telephone cable that normally connects between the telephone main housing and the handset. The cable has four wires that are color coded. You can select a coiled-cord version that can stretch up to 20 feet or more. This cable can be purchased almost anywhere—even food supermarkets carry them.

There are two parts to the scan cycle. First, wire 1 is set high and wire 2 is set low (U1: pin 22-low, pin 23-high). The inputs (wires 3 and 4) are held low by resistors R5 and R6. The Momentum switch, S3, is a two position switch with a center-off return position. If the Momentum switch, S3, is pushed, D2 will be forward biased, making U1 pin 20 low. Similarly, if the Up switch is pushed, D3 conducts, making U1 pin 21 low. Since wire 2 is low, D4 cannot be forward biased even if the Down switch is pushed. When the computer reads the inputs, it makes wire 1 low and 2 high. Now the Starting Speed and Down switches can forward bias D2 and D4, so the computer can read them. Resistors R3 and R4 ensure that the input levels to U3 are high enough.

**Fig. 1A**—The Compu-Throttle's schematic diagram is served up in three functional parts. The circuit shown above is for the power-supply circuit that powers the other two sections.

**Fig. 1B**—The control circuitry for the Compu-Throttle consists primarily of the Motorola MC68705P3 microcomputer chip.
circuit complexity and cost. Semiconductors U6, U5, BR1, and Q1 through Q4 dissipate enough heat to need heatsinking. If you are using the printed-circuit board (refer to the Parts List and Figs. 9 and 10) specified in this article, these components are all along one edge so they can be mounted on a common heatsink. They must be electrically isolated from each other, so insulation must be used. The leads may have to be bent to provide an offset if the bodies do not extend far enough to reach the heatsink. There are several ways to handle BR1: it could be mounted on the printed-circuit board with its own heatsink or an angle bracket to the main heatsink; it could be mounted flat against the heatsink and the leads bent and soldered into the board; or it could be mounted off-board on another heatsink and the rectified output connected to the + and − pads on the board. If bridge rectifier BR1 is mounted flat on the board, a hole may have to be drilled in the printed-circuit board for a mounting screw to pass through.

If you are using the printed-circuit board and make it either from scratch or the template supplied in this article, inspect it carefully for shorted and broken traces before starting on the assembly.

Some components cover others, so the order of assembly is important: 1. Jumper wires (four). 2. IC's or IC sockets. 3. Resistors. 4. diodes and bridge rectifier. 5. Capacitors. and 6. transistors and voltage-regulator chips.

If you decide to use modular telephone cables as suggested, be careful: Modular telephone cord sets contain two pairs of wires, the first is the red and green wires, the second is the black and yellow wires. For a reason unknown to the author, most cords reverse the two wires of each pair. This can lead to some confusion when using them in non-telephone applications. Although the actual choice of wire colors is not important, wiring it as suggested will maintain compatibility between all units built. One sure technique is to ring out the telephone cable and number each of the corresponding ends 1 through 4.

Installing the program in an MC68705P3 requires either a Motorola EVM board (about $500) or a small board which copies the program from a 2764 EPROM. Since most readers probably do not have this equipment, the microcomputer chip U1 is offered pre-programmed (See Parts List). If you do have the equipment and wish to program the chip yourself, please contact the author.

**Handheld Control Box**

The Handheld Control Box is the user interface for the Compu-Throttle. Each builder will probably prefer something a little different, so only general notes are presented here. Several factors must be considered:

1. Size and shape factor of box. Because of the small number of components, almost any size box can be used.
2. Button feel! Since the buttons are pushed frequently, light touch, reliable buttons should be used. The author prefers the computer keyboard type which actually have several sets of contacts that come together when the key is pushed (bifurcated).
3. Since pushing both buttons (S1 and S2) activates the

**Parts List for the Compu-Throttle**

<table>
<thead>
<tr>
<th>Semiconductors</th>
<th>Capacitors</th>
<th>Resistors</th>
</tr>
</thead>
<tbody>
<tr>
<td>BR1—10-μ, 100-PIV, silicon, bridge rectifier</td>
<td>C1, C2, C3—0.1-μF, 100-WVDC, miniature ceramic disc</td>
<td>R1—12,000-ohm (See note in Initial Checkout and Testing text section)</td>
</tr>
<tr>
<td>D1—5-μ, 200-PIV, silicon diode</td>
<td>C4—1-μF, 50-WVDC, Tantalum</td>
<td>R2—2000-ohm</td>
</tr>
<tr>
<td>D2—5-μ, IN914 silicon, small-signal diode (or equivalent)</td>
<td>C5—1000-μF, 35-WVDC, electrolytic</td>
<td>R3, R4, R14, R17, R18, R21—5100-ohm</td>
</tr>
<tr>
<td>LED1—Light-emitting diode, tri-color</td>
<td>C6—10-μF, 50-WVDC, electrolytic</td>
<td>R5, R6—10,000-ohm</td>
</tr>
<tr>
<td>Q1—Q3—TIP126 5-μ, 80-volt, PNP, Darlington power transistor (or equivalent)</td>
<td>C7—2-2-μF, 35-WVDC, electrolytic</td>
<td>R7—R10—56,000-ohm</td>
</tr>
<tr>
<td>Q2—Q4—TIP121 5-μ, 80-volt, NPN, Darlington power transistor (or equivalent)</td>
<td>C8A, C8B—1000-μF, 50-WVDC, electrolytic</td>
<td>R11—4700-ohm</td>
</tr>
<tr>
<td>U1—Motorola MC68705P3 MCU integrated circuit with program installed. See text and ordering information below.</td>
<td>C8: The published circuit board layout is designed for Panasonic SU Series with axial leads.</td>
<td>R12, R15, R16, R19, R20—1800-ohm</td>
</tr>
<tr>
<td>U2—LM324 quad op-amp integrated circuit</td>
<td>Parts List.</td>
<td>R13—75-ohm, 2-watt, 10%</td>
</tr>
<tr>
<td>U3—74HC14 or 74C14 CMOS hex Schmidt trigger</td>
<td></td>
<td></td>
</tr>
<tr>
<td>U4—16-pin DIP header with wave-shaping circuit components (see text)</td>
<td></td>
<td></td>
</tr>
<tr>
<td>U5—7805 or 78M05 +5-volt, fixed, voltage regulator integrated circuit</td>
<td></td>
<td></td>
</tr>
<tr>
<td>U6—LM350T 3-ampere, adjustable, voltage regulator integrated circuit</td>
<td></td>
<td></td>
</tr>
</tbody>
</table>

**Additional Parts and Materials**

- F1—Fuse, 1-μ, 250-volt, slow blow and holder
- J1, J2—RJ11 4-conductor, modular, telephone jack (line cord size)
- S1, S2—Keyboard-type pushbutton switch
- S3—Single-pole, double-throw, center-off toggle or slide switch
- T1—Power transformer, miniature: 117-volt AC primary winding, 18- to 24-volt, 3-A secondary winding
- AC line cord, modular phone cord, printed-circuit board materials, IC sockets (if desired), suitable enclosures for Power Unit and Handheld Control Box, hardware, decals, wire, solder, etc.

The following parts are available from Brom Micro-systems Engineering, Box 616, Winona, MN 55987. VISA/MC accepted.

High-quality etched and drilled printed-circuit board with silk-screened parts placement diagrams—$7.00; MC68705P3 microcomputer chip U1 with program installed in EPROM—$23.00; kit of above plus all parts on PC board, IC sockets, LED1, and R24—$75.00. Please add $1.00 per order for shipping.
**panic stop** function, some provision should be made to keep **one finger** from accidentally pushing both. Wide spacing, or a piece of plastic between the buttons can prevent this.

4. Since the Handheld Control Box may be dropped occasionally, durable switches should be used.

5. Durability is also a factor in selecting the slide switch. A “Bat handle” toggle switch takes quite a beating. A slide switch with well defined detents works well.

6. A piece of self adhesive “Velcro” on the back makes a convenient way of hanging the Handheld Control Unit on the edge of the layout. The Handheld Control Unit does not have to be handheld, the same buttons could be mounted in a panel attached to the layout. When the Handheld Control Box is hanging on the side of the layout, it can be used without detaching it.

**Initial Checkout and Testing**

Since the most common problems will probably be short circuits between traces of the printed-circuit board wiring; and improperly installed components, check everything over carefully. This will save much time later. The author prefers to perform initial checkout in stages. The entire circuit should be built up, but the three IC’s left out. Power can now be applied for the **smoke test**. With power applied, check to see that the +5-volt (4.75- to 5.25-volt) and +1.4-volt (1.2- to 1.6-volt) power supplies are operating correctly.

The output to the tracks may be slightly turned on due to the lack of drive. Resistor R2 helps keep U6 off when the header or U2 are out.

Now remove power and install U2 and the header, then apply power again. The output should now be off. The output circuit can be tested, if desired, by pulling pins of U2 high. See Table 1. Chip U3 and the Handheld Control Box can be tested by applying various levels to U2. See Table 2.

If all is satisfactory thus far, then remove power, install U1, and reapply power. The first test of the CPU is to put a voltmeter or logic probe on U1 pin 27, which is a squarewave with a period of 10 seconds (high for 5 seconds, then low for 5 seconds). This signal can be probed with a voltmeter. If it is present, the CPU is running. The period will be exactly 10 seconds.

**TABLE 1—OUTPUT LOGIC**

<table>
<thead>
<tr>
<th>Conditions</th>
<th>Test Point</th>
</tr>
</thead>
<tbody>
<tr>
<td>U5 Pin 12</td>
<td>U2 Pin 10</td>
</tr>
<tr>
<td>+ Volts DC</td>
<td>Signal Level</td>
</tr>
<tr>
<td>0</td>
<td>Low</td>
</tr>
<tr>
<td>X</td>
<td>Low</td>
</tr>
<tr>
<td>5.5</td>
<td>Low</td>
</tr>
<tr>
<td>5</td>
<td>High</td>
</tr>
<tr>
<td>5</td>
<td>High</td>
</tr>
</tbody>
</table>

Connect a 1000-ohm, ¥Watt resistor across the output terminals. A Low level will be 3 volts or less, a High level will be approximately the Zener (on the header) voltage. X indicates either a Low or a High.

**TABLE 2—KEYBOARD SCAN LOGIC**

**COMPU-THROTTLE FEATURES**

*Easy to use:* The locomotive engineer does not have to be an electrical engineer, too.

*Easy to set up:* Small enough so that it is easy to carry to a show, hooks up just like a regular power pack.

*Economical to build:* The cost does not outweigh the features.

*Walk-around hand control:* The Handheld Control Box that the operator uses is a small box with 2 pushbuttons and a slide switch, connected to the main unit with a 4-wire modular telephone cable.

*Memory:* If the Handheld Control Box is unplugged from the Power Unit for any reason, the train continues at the same speed. *Momentum:* An adjustable amount of momentum is provided to simulate different power trains.

*Starting speed adjust:* Since most engines will not start until approximately 5-volts DC is applied to the track and each locomotive is different, the system allows the starting voltage to be adjusted.

*Choice of pulse power, pure DC, or any combination:* The output waveform can be changed from pure pulse power to pure DC, or any combination in between, just by changing a few parts, which are mounted on a DIP header. When pulse is used, it is low frequency at low speeds, providing very smooth starts and stops, but a higher frequency at high speeds reducing unnecessary heating.

*Reliable operation:* The fewer components in the circuit, the better. The single-chip computer design helps this. The voltage regulator that controls the output contains internal protection that makes it almost impossible to destroy. LM350 regulators have been used from the beginning and one has yet to burn out in operation.

*Usable with all model scales:* Almost any electrically powered model train can be operated, from Z gauge to O to G gauge (LGB), as long as the Compu-Throttle’s ratings are not exceeded. Even older “toy” trains that normally run on AC can be run, except that reversing them may be different than with DC powered trains. (Minor adjustments to the output wave-shaping circuit may be needed for optimum operation).
seconds if the CPU clock frequency is 4 MHz, so even a voltmeter can be used to determine the exact operating frequency of the CPU. The value of R1 needed to produce 4 MHz varies from resistor to resistor. If the value is too low, the oscillator may not run. If this is the case, the value of R1 may need to be changed. Start by substituting about a 22,000-ohm resistor in place of the 12,000-ohm unit. This will probably result in too low a clock frequency. The resistance value can then be lowered until the correct frequency is obtained. The actual operating frequency is not critical, so anything close to 10 seconds will do. If the frequency is too high (squarewave period less than 10 second) with the 12,000-ohm resistor, the CPU might run erratically, so it would be wise to increase the resistance value.

Pin 19 of the CPU presents the results of the checksum test of the EPROM. If no error is found, this pin will be low. If an error is found, it will be high. If an error is indicated, the output pin could be bad, or one or more bytes of EPROM may have the wrong values. Do not be afraid to try using the chip anyway, as the error may not cause a problem.

Since the keyboard is scanned, it is difficult to test that area of the circuitry with the MCU installed. Instead, 4 output bits contain the results of the scanning. Each bit will be high if the corresponding switch is pushed; low, if not pushed:

(Continued on page 98)

**OPERATOR INSTRUCTIONS**

**SPEED UP**
1. Make sure the slide switch (S3) is in the center position.
2. Push the Up button (S1).
3. As long as the Up button is held, speed will increase until maximum speed is reached.

**SLOW DOWN**
1. Make sure the slide switch (S3) is in the center position.
2. Push the Down button (S2).
3. As long as the Down button is held, speed will decrease until the train has stopped.

**PANIC STOP**
1. Make sure the slide switch (S3) is in the center position.
2. Push the Up and Down buttons (S1 and S2, respectively) simultaneously.
3. The train will stop immediately.
4. Release both buttons.

**CHANGING DIRECTION**
1. Make sure the slide switch (S3) is in the center position.
2. Bring the train to a stop.
3. Release all buttons.
4. Push, then release, Down button (S2).

**NOTE:** The train may stop before the output is completely off, which will prevent reversing. In this case, hit PANIC STOP (both Up and Down buttons) once the train has stopped, then reverse.

**ADJUSTING STARTING SPEED**
1. Set the slide switch (S3) to the STARTING SPEED position.
2. Using the Up and Down buttons (S1 and S2, respectively), adjust the speed so the train is just about to start.
3. Set the slide switch back to the center position.
4. This speed is now the lowest speed above stop.

**ADJUSTING MOMENTUM**
1. Set the slide switch (S3) to the MOMENTUM position.
2. To increase the momentum (slower rate of acceleration/deceleration) hold the Down button (S2) for a few seconds.
3. To decrease the momentum (faster rate of acceleration/deceleration) hold the Up button (S1) for a few seconds.
4. Set the slide switch to the center position.
5. Operate the train.
6. Repeat the above steps until the desired amount of momentum is reached.
CHEMICAL POLARITY EXPERIMENT

Lets face it, electronics is a science and science is becoming all electronic. To prove it, try this simple multimeter test to determine chemical properties.

Phenolphthalein is a member of a large class of substances known as indicators. Indicators signal the presence, absence, or concentration of other materials by means of some detectable change, often a change in color. Phenolphthalein is colorless in the presence of neutral or acid solutions but turns a bright pinkish red in the presence of an alkali (basic) solution.

With some phenolphthalein, some alcohol, some salt, a dish, and a few odds and ends from the junkbox you can build a simple polarity meter and learn something about electrochemistry besides. Where will the phenolphthalein come from? That's easy. It's the active ingredient in Ex-Lax.

Obtain a small glass dish or bowl with steep sides. The piece of hardware in the photograph is a laboratory culture dish, but any sort of dish will do fine. Place a circle of filter paper or paper towel in the center of the plate. Run two lengths of stiff wire or narrow metal rod from the surface of the towel to a couple of binding posts. A couple more terminals may be added to facilitate wiring (Photo 2). The ends of the contact rods should be approximately one inch apart. It is not necessary to attach the terminals to the dish; they can be located elsewhere, on a block of wood, for example.

Now crush a single Ex-Lax tablet with the round end of a large screwdriver or something similar. Dissolve the powder in one ounce of solvent alcohol. Be careful, solvent alcohol is very flammable. In another container, dissolve one-half tea-

(Continued on page 108)

PARTS LIST
FOR THE CHEMICAL POLARITY EXPERIMENT

Glass dish
Ex-Lax tablets (Phenolphthalein)
Filter paper or paper towel
Solvent alcohol
Table salt (Sodium Chloride)

Two short lengths of stiff wire or narrow metal rod, binding posts, hook-up wire, and low voltage DC power source.

For those of you who may want to use genuine laboratory chemicals and glassware, the author recommends: Hagenow Laboratories, 1302 Washington, Manitowoc, Wisconsin (catalog $1.00) and Chem-Lab, 13814 Inglewood, Hawthorne, California 90250 (chemical catalog $3.00, equipment catalog $4.00, both for $5.00).
There is nothing quite as frustrating as to awaken on a cold rainy morning, open the front door, go outside and open the newspaper delivery box, only to find that the newspaper has not yet been delivered. That's a scenario that is repeated several times daily in many households. But, fortunately, there is a solution—Delivery Indicator.

The Delivery Indicator is a circuit that can be used to report that your newspaper or mail has been delivered by means of a flashing LED installed within the home. Once the delivery is retrieved, you only have to push a reset button to prepare the circuit for the next event.

The Indicator requires only one integrated circuit, a diode, LED, and a few passive support components, all of which are inexpensive and readily available. So the entire project can be built for less than ten dollars.

Circuit Description

The schematic diagram of the Delivery Indicator is shown in Fig. 1. A single 9-volt battery (B1) provides all of the power required by the circuit. The circuit was designed to be as efficient as possible, thereby maximizing battery life.

The heart of the circuit is U1, a 4001 CMOS quad 2-input NOR gate. A CMOS logic device was selected specifically because of its low-power requirement. Two of the NOR gates, U1a and U1b, are configured as a positive edge-triggered flip-flop. The edge-triggered configuration is used to avoid disallowed state conditions. The resistor-capacitor networks (R2/C1, and R4/C2) provide a 20-microsecond pulse to accept the input from the switches and prevent a disallowed state from occurring.

Resistors R1 and R3 are pull-down resistors, which provide a low-state input in the absence of a switch closure. Switch S1 is a momentary pushbutton switch that is used to reset the flip-flop. A Mercury tilt switch (S2) is used to indicate that a delivery has occurred, by setting the output of the flip-flop to a low state.

The output of the flip-flop is fed to the input of a gated astable multivibrator that is comprised of U1c and U1d. A high flip-flop output prevents the multivibrator from oscillating, while a low output permits oscillation to occur. The output waveform of the astable multivibrator is dependent upon the following:

\[
T_{HI} = (1.1)(R6)(C3)
\]

\[
T_{LO} = [(1.1)(R5)(R6)(C30)]/(R5 + R6)
\]
The LED is illuminated when the output of the astable is high and goes dark when the output of the astable is low. The on and off times of the LED were selected to provide a minimum "on" time and a maximum "off" time to conserve battery power. The on/off times based on the component values shown in Fig. 1 are given by substituting the actual component values for the designations in the above equations, for example:

\[
T_{\text{OFF}} = 1.1(10^7)(1.6) = 1.1 \text{ seconds}
\]

\[
T_{\text{ON}} = 1.1(10^6)(10^7)(1.6)/(10^7 + 10^6) = 0.1 \text{ seconds}
\]

Resistor R7 is included to provide a cleaner squarewave output than is possible without it. The value of R7 is determined by:

\[
R_7 = (10)(R_5)(R_6)/(R_5 + R_6)
\]

The value of R7 based on the values of R5 and R6 given in Fig. 1 is:

\[
R_7 = (10)(10^6)(10^7)/(10^7 + 10^6) = 9.1 \text{ megohms}
\]

The nearest common value is 10 megohms, which is the value selected for R7 (as revealed in Fig. 1).

The LED selected was specified primarily because of its low-power requirement. The Hewlett Packard HLMP-1700 LED requires approximately 10 mA. The current-limiting resistor, R8, is selected to provide the maximum current required by the LED, and its value can be calculated according to the following expression:

\[
R_8 = (V_{\text{source}})/(V_{\text{LED}} - I_{\text{LED}})
\]

or in more practical terms,

\[
R_8 = (9 - 2)/10^3 = 700 \text{ ohms}
\]

Although there several standard resistor values between 700 and 1000, R8 is selected as 1000 ohms, which serves to reduce current drain. If the circuit is built using the HLMP-1700 LED, the standby current is 0.005 mA. A standard carbon 9-volt battery can provide close to 80 mA hours.

The author's prototype of the Delivery Indicator was built on a piece of perfboard, measuring about 3 x 1.5 inches. A home-made battery holder secures the battery to the board. Short lengths of wire connect the circuit to the LED and S2, which are mounted off-board in some convenient location.

The life of the battery in standby mode is:

\[
80 \text{ mAH}/0.00053 = 1.6 \times 10^4 \text{ hrs. or 1.83 years}
\]

Note: You may substitute with a standard red LED, but R8 must be changed to 1000 ohms. Using a standard LED will only reduce the battery life significantly if the LED is flashing for long periods. A long-life battery can be substituted for the standard type specified if battery life is a significant problem.

Construction

Construction of the Indicator is non-critical; therefore, any construction technique may be used. Radio Shack parts (which are also available through other suppliers) were selected wherever possible for convenience. The unit was built on standard perfboard (having a 0.100 x 0.100-inch grid pattern) using point-to-point wiring.
To minimize the size of the completed project, 6-inch wire leads were provided to splice into the LED and switch wiring, rather than to provide a separate connector. The circuit should be built as small as possible to facilitate installation.

The perfboard was mounted on standoffs from a second board of the same size to provide a flat mounting surface. Care should be taken when handling U1 because it is a CMOS device, which is inherently sensitive to static discharge. Do not handle U1 by its leads and keep it in its container prior to installation. Do not solder directly to U1; instead use the socket recommended in the Parts List.

Installation

The home in which the Delivery Indicator was installed has a mail chute as shown in the photos, which makes an excellent location for installation of the unit. Double-faced tape was used to mount the circuit to the side wall of the chute, with board located near the inside mail-chute door to facilitate battery replacement.

The knob on the inside mail-chute door was replaced with a larger knob which could accommodate the LED and reset switch more easily than the original knob, as shown in the photos. The reset switch was mounted into a hole countersunk on the front and rear of the knob to make the switch less obtrusive. The switch was brought in from the rear of the knob and secured in place. The LED was mounted into a countersunk hole from the rear of the knob. It was retained in the knob by RTV adhesive silicon sealant.

The knob was mounted on the mail-chute door by a screw located at its center. The leads for the reset switch and the LED were brought through two holes drilled through the mail-chute door, one on each side of the center mounting screw.

Lengths of number 26 AWG wire were soldered to the reset switch and LED, and brought back to the circuit board. Only three wires are required because the +9 volt source is common to both the switch and the LED. Duct tape was used to retain the wires in place once they were properly routed. Care was taken to provide enough slack in the wires to permit the mail-chute door to swing freely.

The set switch is installed on either the outdoor newspaper-box door, or outdoor mail-chute door, or both as desired. The set switch, a Mercury tilt switch, must be installed properly to permit correct operation. Observe the motion of the door carefully. Follow the same motion with the switch in hand. From that exercise determine the proper mounting angle to ensure that the mercury makes contact with the door open and breaks contact with the door closed. A shim may be required to provide the proper mounting orientation. The switch is held in place by duct tape.

Two number 26 AWG wires are routed from the mercury switch to the mail chute and to the circuit board. If more than one mercury switch is used, each switch is wired in parallel with the others.

Once a delivery has been made, the LED will begin flashing. The flashing will continue regardless if the delivery door is re-closed or left open.

The mail or newspaper is received and the reset switch is depressed to shut off the LED and prepare for the next delivery.
Get use out of both the IBM and Okidata character sets with a switch that adds greater capacity to your printer setup.

By Jeff Holtzman

Okidata printers have long been of the highest quality, inexpensive, and well-supported by most major applications programs, including both word processors and graphics programs. Advantages include rock-solid construction, user-defined character sets, and high-quality graphics.

In fact, the only real disadvantage of early Oki printers was incompatibility with the IBM graphics printer. Okidata soon realized that fact and introduced an upgrade option that allowed owners of the original Microline 92 and 93 models to attain IBM compatibility.

Later models provided similar options. Installation was a snap: remove an EPROM or two, depending on the model, and install the new EPROM (or two), known as the Plug "n' Play kit.

The only problem was that by gaining IBM compatibility, you lost Okidata compatibility—and many software packages were written for the Okidata printer standard before IBM introduced its own printer. So you had to choose: IBM or Okidata. Otherwise, if you wanted to print something in the non-installed mode, you ended up physically switching EPROM sets. And obviously that's a real hassle.

Articles in computer magazines suggested a solution: burn the contents of both EPROM sets into one large EPROM, and use a switch to select IBM or Okidata mode. The problem was that many people simply didn't have the necessary equipment or expertise to do the job.

So the UKI Switch Corporation (326 Linden Place, West Hempstead, NY 11552) came to the rescue. The company offers a small plug-in circuit board with an EPROM containing both the Okidata command set and the IBM Plug "n' Play options. This means that if you have an Okidata printer with just the Oki command set, you don't have to buy the Plug "n' Play option from Okidata separately. UKI sells different configurations for various models, including the Microline 83, 92/93, and the 192/193.

Installation

Installation is simple, but requires care. The UKI EPROM is mounted on a small circuit board that bridges both EPROM sockets on the Okidata's circuit board. 28 pins on the UKI board plug into the Q4 socket; another pin plugs into pin 20 of the Q5 socket. You must get the alignment right, or your printer won't work. After installation, the circuit board rides about 1/8-in. above the main circuit board, providing plenty of clearance between it and the rear chassis panel.

Trailing from the UKI board is a sub-miniature toggle switch that you can mount in one of the unused holes on the rear panel of the printer. Then, to change modes, you just turn off the power, flip the toggle switch, and turn the power back on.

To make a quick check of your work, turn the power on (in either mode) while pressing the LINEFEED switch. The printer then enters a self-test mode in which it repetitively prints all characters it is capable of printing. Repeat the test in the other mode. If everything is working correctly, you'll get different sign-on messages and different character sets in the two modes.

If you have a problem, make sure the circuit-board pins

*(Continued on page 100)*
If you live in an area in which a nuclear accident is likely to occur (perhaps in close proximity to Three Mile Island), and are concerned that radiation could contaminate life’s essentials (water, food, and air supply), this little circuit can restore some degree of security to your existence by alerting you to the contamination.

Nuclear contamination can stalk the land, leaving little indication of its presence until it is too late—it is a silent, phantom-like killer, in that it’s usually not visible and is never audible. But the Rad Hound (with its dog-like dedication) will “sniff” out the source of contamination and “bark” its warning to indicate the presence of radiation. Needless to say, such a device (under some circumstances) can mean the difference between life and death for your family.

If the area you live in has been definitely identified as a contaminated area, you may just up and leave—taking your family and all your worldly possessions with you. But in doing so, you must also know whether you’re carrying away a contamination “doggy bag” that might possibly harm you and your household over a period of time. It could be in your clothing, in your hair, on your body, even inside your body. You’ve got to seek it out and then work on the problem of separating it from your body before it’s had a chance to do serious damage. That’s just the type of problem that the Rad Hound is design to help remedy.

If you live in an area where nuclear contamination is a definite possibility, then you may have considered purchasing one of the store-bought Geiger counters, retailing for at least $400, which can communicate exact amounts of radiation (in calibrated units) present within a given area either by means of an analog or digital readout. But if, like the average citizen, you are more concerned with knowing where the contamination is than you are with knowing how much, then the Rad Hound can certainly handle that task.

The sensing device used in radiation detectors is something called a Geiger-muller tube. Recently, the manufacturers of the Geiger-muller tube developed a version that is small and economically priced ($58). Plus, the money you spend for that one item alone is well worth the price—after all, who would try to save a “buck” by gambling on the survival of their family?

That's Rad

Figure 1 shows the schematic diagram of the Rad Hound, which consists of (in addition to the Geiger-muller tube) a six volt DC power supply; a solid state oscillator with an output of 2 kilohertz; a high voltage power supply; and an audio amplifier. The power supply can be as simple as four batteries (connected in series), for portable and stationary use. Or you might consider building a power supply, or using a wall-mounted power pack.

The operation of the circuit is lot simpler than one might imagine. Transistor Q1, which is configured as a variation of the Hartley oscillator, generates a low voltage AC signal. That alternating signal is fed across the primary of step-up transformer T1, creating a rising and collapsing field which is, in turn, inductively transferred to its secondary as high-voltage AC. The voltage at the secondary of T1 is fed through D1 and D2. Those diodes form a voltage doubler/full-wave rectifier, to provide an extra boost to the output voltage. The magnified/rectified voltage is then filtered by capacitor C4 into 500-volts DC, which is then fed to a Geiger-muller (G-M) tube.

Inside the G-M tube is a gas that ionizes (conducts) whenever a pulse of radiation enters the tube window. Each pulse causes a brief voltage drop across a 4,700,000-ohm resistor, R5. Capacitor C5 passes that pulse into the audio amplifier (U1) which, in turn, sends its amplified output to the loudspeaker (SPKR1), where it’s converted into a pulse of sound.

Be aware that the 500-volts DC on the G-M tube is not regulated. It will drop slightly whenever the tube is busy, but that won’t have a noticeable effect on the sound you’ll hear. Transistor Q1 is a little “tough guy” that is capable of dissipating up to 115 watts of power. But it just hums along at ten percent of that maximum and doesn’t need a heat sink. Trimmer potentiometer R1 in Q1’s base circuit, varies the oscillator’s frequency just enough so that you can get 500-
Fig. 1—The Rad Hound, whose operation is almost totally dependant upon a Geiger-muller (G-M) tube, consists (in addition to the tube) of a six volt DC power supply; a solid state oscillator with an output of 2 kilohertz; a high voltage power supply; and an audio amplifier.

volts DC for the G-M tube at the point where R4 and C5 join. (Make absolutely sure your DC voltmeter has a very-high impedance so that you don’t load it down.)

It is very important that you use the exact same type transformer for T1 that is specified in the Parts List: a center-tapped 24-volt unit capable of handling at least 450 milliamps of current. A different one may not work at all; or worse, it could put out too much voltage and just burn up your little G-M tube.

Audio Amplifier

For the audio output section, an LM386 audio power amplifier (U1) was pressed into service. That 8-pin chip saves lots of assembly time, and is capable of driving 400 milliwatts of audio power into a 500 milliwatt 8-ohm loudspeaker. Included in the circuit is a, bass-boost, which is used to emphasize the low-frequency pulses and to de-emphasize the high-frequency components of the 2-kHz oscillator that might otherwise leak through to the amplifier. The loudspeaker can be any 500-mW unit that you might have on hand. And, when installed in the finished project, it should be protected from damage by covering it with a wire screen or perhaps a grille.

If you want to get fancy, you can use a pair of low-impedance earphones from a portable FM-radio headset as a substitute for the speaker. However, no one else will be able to hear the radiation pulses except the wearer of those headphones. Or you may want to use a normal-through jack (also called a closed-circuit jack by hobbyists). Then when an earphone plug is inserted into the jack, the audio signal is re-routed to the earphones, and when removed, the signal is once again fed to the speaker.

Figure 2 illustrates how to connect such a jack. As shown, the speaker is connected in the circuit. But when a plug is inserted into the jack, the speaker is disconnected and the audio signal is re-routed through the earphones.

Construction

The Rad Hound can be built using the construction method of preference—printed-circuit, perfboard (with point-to-point wiring), or pre-etched experimenter board—and components laid out in an arrangement most convenient to you. The authors prototype of the circuit was built on perfboard using point-to-point wiring (see photos). If you opt to go the perfboard route, using the authors layout as a guide might make things go a bit smoother.

Start by placing the IC socket and other components on the board and securing them in place by temporarily bending the leads. Use short lengths of insulated wire to make the interconnections between components, beginning with the passive components (connected to the IC socket). Note that LED1, SPKR1, the power switch (S1), and the power jack (J1) are mounted on the front panel of the housing. So put those parts aside until the board components have been installed and your work checked for correctness.

When you are satisfied with your work, put the board to the side and go to work on the panel-mounted components. First drill and cut appropriate size holes in the panel to accommodate the parts. For the speaker hole, depending on what the cabinet is made of (metal plastic, etc.) you may be able to use a hole cutter (like those used by locksmiths).

After cutting the holes mount the components. Be careful when soldering to the LED. While it can withstand quite a bit of heat (for a semiconductor), it is nonetheless, thermal.
PARTS LIST
FOR THE RAD HOUND

RESISTORS
(All resistors are ½-watt, 5% units, unless otherwise noted.)
R1—1000-ohm, 15-turn, trimmer potentiometer
R2—8200-ohm
R3—200-ohm
R4—470,000-ohm
R5—47,000,000-ohm
R6—10,000-ohm PC-mount trimmer potentiometer (RS 271-218)
R7—10,000-ohm

CAPACITORS
C1—500-µF, 12-VWDC electrolytic
C2—C4—0.005-µF, 1000-VWDC ceramic disc
C5—50-pF, 1000-VWDC ceramic disc
C6—0.033-µF, 100-VWDC ceramic disc
C7—100-µF, 12-VWDC electrolytic
C8—300-µF, 12-VWDC electrolytic

ADDITIONAL PARTS AND MATERIALS
D1, D2—1N4007, 1-A, 1000-PIV diodes
J1—to match plug on T2 (RS 274-1567)
LED1—Jumbo red light-emitting diode
Q1—2N3055 power transistor (RS 276-2041)
U1—LM386 audio amplifier (RS 276-1731), integrated circuit
S1—Single-pole, single-throw subminiature toggle switch (RS 275-662)
SPKR1—8-ohm, 500-mW loudspeaker (RS 40-245)
T1—117-volt AC primary to 24-volt AC, 450-mA, center-tapped secondary, power transformer (RS-273-1366)
T2—6-volt DC, 150-mA, wall mount power supply (RS 273-1454)
Printed circuit or perfboard materials, microphone housing for G-M tube (RS 33-1034), metal cabinet (RS 270-253), wire screen or grill for speaker, carrying handle for project box, alarm clock dial face with luminous paint, 4 C-cell battery holder (RS 270-390), spring clip, wire, solder, hardware, etc.

Note: The Geiger-Muller Tube part # 712 is available, priced at $58, plus $2 for postage and handling (postal money order or cashiers check only) from A.P. Campione, 464A Hudson Street, Box H16, NYC, NY 10014. Please allow 6 to 8 weeks for delivery.

sensitive. Once that's complete, the next thing is the housing for the G-M tube.

G-M Tube Housing
The biggest problem in designing the Rad Hound, was in finding a suitable ready-made housing for the G-M tube. Nobody makes them cheaply. But I discovered an old microphone case that had plenty of room inside. Once the on/off switch was removed. You can do the same thing if you buy that microphone from Radio Shack for $7. Just remove the dynamic mike element and the on/off switch, and very carefully place the G-M tube behind the front grill.

The tube itself is small, measuring only .59 inches in diameter and 2.12 inches in length. Pack some plastic foam into the space behind it so that nothing rattles around, after you’ve soldered the cable’s center conductor and shield to the G/M tube. For added protection, you might want to surround the tube in silicon within the microphone housing.
Video on the Run


Nobody asked us, but it's clear that the consumer-video marketplace faces a major shake-out sooner or later. There's the unresolved question of competing cassette formats, at last count VHS, Beta, 8-mm, and VHS-C. Coming up from behind are newer developments, the now near-standard "high-quality" (HQ) circuitry and, beginning last year Super-VHS. All of that has transpired in the little more than a decade since video became a consumer product.

Technical superiority is no guarantee of eventual market conquest. In the home-video beginning, Beta was clearly the superior format. But the buying public's attention was arrested by a brouhaha regarding recording time (the one category in which VHS excelled) and soon Beta (at least in this country) was on its way to competitive status.

Amid the mounting marketplace uncertainty, the video industry has noticed a somewhat surprising development. Even as VCR sales level off and consumers display uncertain, its about jumping into newer formats, camcorder sales continued steady and even increase. That despite price tags that put these portable combo recorder/player at the upper end of the video market.

If you have an opportunity to get to...
know a state-of-the-art camcorder, there's little mystery to this sales pattern. Camcorders are the kind of electronic toy worth saving up for.

Consumers have become understandably blase about VCRs. Nowadays, they're just another home-entertainment appliance. But for a first-time user, there's still a trace of magic to the camcorder. We suspect that for a generation of consumers who grew up during the home-movie era, these electronic instant motion-picture machines are nothing short of miraculous. Especially if the camcorder is an elegant unit like the Minolta Master Series V-1400. Some experience with the V-1400's predecessor model made this rig even more impressive.

Minolta has made this camcorder impressively lightweight (4.4 lbs.), cutting down on body and battery bulk while retaining the previous model's features and capabilities.

This camcorder in its fully automatic mode knows more about capturing good video images than we do. In several weeks of use, we never got around to testing the V-1400's manual capabilities (focus, iris control, and shutter speed can all be automated), if only because automatically the camcorder recorded such technically splendid pictures.

Even when the images in the electronic viewfinder were seen as little more than gray shadows, the resulting taped image was surprisingly sharp and detailed. Minolta partially credits the unit's fine HQ circuitry. More important is what the firm calls "the electronic equivalent of an adjustable shutter." The camcorder's "MOS image sensor" has a "variable scanning rate." Constructured with 300,000 picture elements (pixels), its ability to select "effective shutter speeds" makes possible rates ranging from 1/60 of a second to 1/2000 of a second.

Minolta says the V-1400 can provide "good images even in light levels as low as seven lux." In real life that translated into worthwhile images captured by the light of a single, heavily shaded table lamp and, outside, in near-dusk. There's a vivdness and intensity to the colors captured under good lighting far beyond the capacities of home movie systems of the recent past. It's seemingly modest audio pickup is as detailed and vivid in its recording as the imagery it accompanies.

The V-1400's controls and grips have a naturalness which is a tribute to the Minolta design department. The right hand fits through a fabric strap and the thumb rests on the record control. Once engaged, it needn't be held down.

The abbreviation REC appears in the viewfinder to remind the operator that tape is running. Actually, the earlier model's red light signal is to be preferred. In that the superimposed letters are in the way of the image, that isn't necessarily a change for the better.

The right-hand forefingers control the camcorder's automatic zoom. While always a stylish capability, in its temptation to overuse, the zoom is particularly reminiscent of pre-electronic home movie making. Operational and secondary controls (like playback features and camera/player selection) are arrayed topside. Further functions are controlled from the left side of the camcorder body.

Information is conveyed to the operator via graphics which appear in the viewfinder. Besides the data (which appears on the recorded tape if engaged) the REC, a battery-power indicator and tape count are superimposed. In the playback mode, rewind and fast-forward, as well as the tape count, appear.

The viewfinder can also function as a quick review monitor. Engaging the function causes the last few recorded seconds to run backward and forward.

As a playback device, the V-1400 was entirely satisfactory. Recordings taken from a television broadcast were as clear as dubs from a regular VHS recorder.

Eventually, we predict, somebody is going to market a multi-event timer add-on for camcorders, furnishing the capability which separates portable units from home VCRs. Whether camcorders are built to do the work of a home VCR is another question. One video cameraman we know says it's a bad idea to use the 160 minutes VHS cassettes with camcorders, as in rewind they strain the mechanism.

One aspect we liked about the V-1400 kit was the variety of adapters supplied. That is in contrast to camcorders of a couple of seasons ago whose makes assumed that the equipment would be used only with the newest generation of televisions. At camcorder prices, we don't blame an owner for not rushing to replace his or her old, pre-home-video tube.

Using the V-1400 cleared up one video mystery, namely the continuing market popularity of VHS-Compact formats. In comparison with this Minolta camcorder, any weight or bulk advantages to the VHS-C units we've examined are virtually nonexistent. What we discovered with this standard VHS camcorder is that the task of filling up a cassette's hour-plus running time in an interesting way can be a definite strain. Particularly since operating the camcorder tends to overwhelm whatever activity is being recorded.

But with the much abbreviated running time of compact VHS, the user is back on familiar, home-movie terrain. Faced with the possibility of convincing friends and relatives to sit through full-length videos of stupefying tedium, the VHS-C format may be the more humane and effective choice.—G.A.

Typographical Era


The first primitive version of what we might recognize as a typewriter was patented in England in 1714. The manual typewriter of the recent past was a product of the 19th Century, with the first electric machines appearing in 1935 and electronic units coming along a couple of decades later. But it was the development of computers which brought about the current stage of typewriter development. Today's electronic machines are a hybrid, using computer technology (like electronic memory), but functioning like any typewriter of the last century—a machine for putting words onto paper.

The Canon S-58 Personal Typewriter, on the market for a little over a year, is an inexpensive electronic machine which integrates computer-related capabilities. Plastic components help keep the price low, although as any typewriter repair technician will tell you, these price-cutting parts also shorten the unit's life.

For anyone who's spent their typing life using manual or pre-electronic machines, the Canon S-58 takes some getting used to. Its excellent instruction manual, a model of organization and clear language, helps make this process just about as painless as possible.

Each of the S-58's features and capabilities are carefully described, logically presented in order of importance and accompanied by clear illustrations and diagrams. The booklet begins with how to insert the paper and continues through a final troubleshooting section which attempts to anticipate problems a novice might run into.

Right at the beginning we discovered a quirk which can be maddening, or at least
disconcerting, to an experienced typist. Common to all but a few expensive, electronic typewriters, there's a lag between hitting the keyboard and the letter going onto the page. This brief delay puts the ear out of sync with the fingers.

Comfortable typing, at least before this current generation of electronic machines, depends on rhythm and the Canon S-58's key-to-print sequence can be surprisingly distracting.

In the machine's automatic return mode, as the return is being made and the user continues to type, there's a point when the next line begins and the keys start clicking out the word typed while the machine was returning.

The keyboard itself is certainly no IBM Selectric. It has that fragile, plastic feel that seems to be a feature of computer-style keyboards. It takes some time getting used to the required light touch to the keys. Eventually, even the worst hunt-and-peek keyboardist can adjust.

For a veteran typist, all of this may just be the price of admission to the advantages of contemporary electronic typewriters. In general, we found the Canon S-58 to be a well-designed and very adaptable personal typewriter.

Tabs, margins, line spacing and automatic bold face and underlining are all easy to master and use. And the S-58's correction system demonstrates the unique advantages of memory-equipped electronic typing.

A touch to the keyboard's "correct" button and the previous letter (or up to 30 letters) is wiped out clearly and fully. The user can type the corrected word into the S-58 even as the correction is being made. Corrections beyond the 30-character memory are a little more complicated. But again the instruction manual is very clear on how to move the carrier back to the proper position in order to make the required correction. The carrier can be moved back in increments of 1/20 of an inch. Useful in making a correction when the page has been removed from the machine and reinserted.

An element of electronic typing which struck us as a disadvantage is the relatively short life of cartridge-style ribbons. The one supplied with this machine had to be replaced after about 35 double spaced pages. Convenience, clearly, has its price.

An added feature, as the advertisements might put it, to the Canon S-58's capabilities is its use as a simple calculator. "Your electronic typewriter," the manual explains, "can do arithmetic (addition, subtraction, multiplication and division) with numbers of up to 8 digits. " Code and the "1-2-3-4-5-6-7-8" key put the S-58 into the calculator mode, with the machine automatically typing out the results of computations. This is at least a clever extra for budget-priced electronic typewriter.

As for the disconcerting sonic delay between key and striking action, that crude 18th Century version of a typewriter was designed as a tool for the blind. Perhaps on the verge of the 21st Century, the electronic typewriter is best used by the deaf.—B.R.O.

Just One Touch

PHONE-MATE PERFORMANCE SERIES 7000 ANSWERING MACHINE.
Manufactured by: Phone-Mate, 325 Maple Ave., Torrance, CA 90503. Price $119.95.

A four-foot drop to the floor taken by an expensive combination telephone-answering machine taught us something about these popular home and office personal communication devices. When we took the unit in for repairs, we were told that the cost of fixing it would just about equal the cost of replacing it. With the answering half of the combo out of commission, we were left with a functioning but rather bulky telephone.

The experience made us wary of combination units and also raised doubts about paying top dollar for an answering machine, given that repair charges equal replacement cost. The Phone-Mate Performance Series 7000 is an answering machine which addresses both of these lessons.
Strictly an answering unit, it's often sold for considerably less than the manufacturer's suggested retail price. We've seen it retail for as little as $69.99 by discounters. Despite its budget price, *Phone-Mate* is a long-established brand in the answering machine field, suggesting that its products are both durable and dependable.

Compact and simple to activate and operate, the *Performance Series 7000* has a number of sophisticated features. Its remote capabilities allow the user to hear, replay, erase or save incoming messages from another phone, without carrying a beeper unit to activate the machine. Instead, the 7000 responds to tones generated by a touch-tone telephone. Of course, as the owner's guide explains, "not all push button telephones are touch-tone. In which case, Phone-Mate offers a portable "Touchtone Dialer" (model 10118) so "you can get your messages from a pulse or dial telephone."

One feature which may be appealing to users who spend a lot of time talking to answering machines, is this Phone-Mate's instantaneous interface between the pre-recorded greeting message and the beep. The seemingly intangible dead air between message and beep can be one of telephonic major life's irritations. Voice activation cuts out dead time on the incoming message tape as well, especially appreciated when you're listening through a day's worth of accumulated calls.

The machine's one-touch operation is another user-friendly feature. The single "Play Messages" button rewinds the incoming message tape, plays to the end of the final message on the cassette, stops automatically and then rewinds the tape and resets the machine to take messages, all with a single tap of a finger.

The features and capabilities described here are all fairly common to answering machines in a number of price categories. But Phone-Mate has come up with some further, useful gimmicks.

If the user is out of town and wants to collect his or her messages, but doesn't want to pay long-distance rates just to hear again incoming messages already received, the *Performance Series 7000* has a nifty tollsaver device. Setting the machine to "TS" activates the unit to pick up on the second or third ring only if there are new messages. If the machine allows four or five rings, that tells the user there are no new messages. Which gives him or her a chance to hang up before the toll charge.

Forgetting to turn on the 7000 before leaving home or workplace no longer means a trip back to flip the switch. Instead, as long as the machine is plugged in, the user can call and let the phone ring ten times, which activates the unit to receive incoming messages.

Some more-expensive answering machines feature a user-selected, two-digit message retrieval code. With this unit, the single-digit retrieval code is permanently set (and printed on a sticker on the underside of the lid covering the cassettes). Unless you've got some reason to be seriously concerned with message security, this is a minor limitation. Similarly, the 7000 indicates the number of received messages with a light. The user has to count the flashes to know how many are stored and waiting. A numerical display would have been a nice refinement, although for many users probably not one worth paying that much more for.

Finally, the machine—in common with most answering machines we've examined—is outfitted with a Lilliputian phone connection cord. Extensions are easily obtained, but why do answering machine manufacturers assume that the user is always going to place the unit next to the phone?

At a reasonable price, the Phone-Mate Performance Series 7000 offers convenience and a number of genuinely useful features. Including at least one the company may not be aware of.

After our experience with a previous machine taking a disastrous fall, we decided to keep this unit safely on the floor. Thanks to its design and one-touch operation, all of its controls can be manipulated with your toes, even though a shoe. For our money, that's real user-friendliness.—E.L.

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**Slipped Discs**


The record industry has had a difficult time keeping up with the music public's demand for compact discs. In 1983, its first year on the market, the compact disc format sold some 800,000 copies. By 1986, some 53 millions CDs were sold, a dramatic 134 percent increase over the previous year. Clearly, this format is a good bet to eclipse the LP in the near future.

The advantages offered by the CD over the LP format are nearly as great as the LP's unquestioned superiority over the 78. With the compact disc, an hour's worth of programming is stored on a disc smaller than a vinyl single. CDs play back at higher volumes and, depending on the care taken in the remixing and mastering stages, with much higher definition.

Musical reproduction is one area in which the highly-touted digital revolution has paid big dividends. Subtle details drowned in the background (and surface) noise of LPs reveal themselves on CD. The audio sales pitch most time honored—"it's as if the musicians were actually in the room"—has taken on some validity in the CD era.

It's no accident that the group which ushered in the LP era in rock, the Beatles, spearheaded last year's CD retail gold rush. Producer George Martin did a spectacular job preparing all the Beatles CD's for release in '87. Every one of the titles reached at least the number four spot in the CD sales charts. Both Sgt. Pepper and *A Hard Day's Night* hit number one and last June, five Beatles CDs were in the top 30. Other artists have also garnered good sales with CD releases of their originally LP-recorded music.

Just in time for these multi-CD retrospective re-releases, disc changers became an available option for players, at first only at the high end of the retail mar-

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CIRCLE 52 ON FREE INFORMATION CARD
ket. But now units in the market middle come equipped with multi-disc capabilities.

Sharp's DX-C6000 includes many features popularized by more expense makes. Such as programmable play, random play, remote control, high-speed search, two-speed cue and review, reserve memory and tape dubbing. But its crowning feature is its six-disc CD changer. Looking like a stack of jewel box CD cases, the unit slides into the front of the DX-C6000 and allows the user to program up to seven hours of uninterrupted music. The CDs can be programmed in custom sequence or at random, tracks from one CD playing after tracks from another and so on.

Given the newness of the CD format, it's slightly ironic that changers are basically mechanical devices, variations on a theme which dates back to the cumbersome record changers of the Pre-LP era. The magazine swings each CD out for play and then swings it back into the stack for storage. The audible clicks and whirring heard during changes sound incongruous coming from a device which represents the very latest in digital audio reproduction.

After living with the unit for some time, we began to wonder just how "random" its random play capability is. Although not a scientific test, everyone in our household agreed that in this mode, the same tracks turned up with disconcerting regularity.

Finally, we had an experience with the Sharp Multi-Play CD Changer which made us wonder about the state of this particular art. After several hours of trouble-free listening, we attempted to change some discs in the magazine and reprogram the player's sequence of play.

For reasons still unknown, the unit went daffy on us—clicking through its six discs, as vividly recorded by the LCD display, but refusing to play any tracks at all. Regardless of the control we engaged, and even after shutting down power and turning it back on, the player and changer continued to misbehave. It was one of those awful moments, known to most electronic consumers, in which we decided we must have done something to break the component, some combination of contradictory commands or an improper sequence of button pushing. But, for whatever reason, the only sounds we could get out of the DX-C6000 were clicks and whirring. Vigorous manipulation of the remote control offered no relief. So we finally admitted defeat and turned it off. As of this writing, we haven't had the heart to try again.

Pioneering has its price. Our first CD player suffered from serious defects, including shrill, brittle high-end response and an annoying tendency to leap suddenly to disconcerting volume. So perhaps this first generation of CD changers can't be held to different standards.

Apart from this, we think the DX-C6000 is an effective compromise between big-ticket CD players with changers and the many bargain basement models with no special features. The multi-disc format is perfect for CD players, because it enhances the format's outstanding characteristic, more music in less space. We just wish we weren't recommending an item which experienced an electronic nervous breakdown while we were testing it. —J.S.

Penny Savior

J.C. PENNEY VIDEO CASSETTE RECORDER (MODEL NO. 686-6184).


Price: $450.

Had you asked us six months ago what we thought about buying video gear from J.C. Penney, we'd have said that it was like buying lobsters from F.W. Woolworth. Sure, they look like lobsters, they walk like lobsters, their spiny antennae twitch like lobster-antennae are supposed to twitch—but are they lobsters on the test of taste?

Can a department store known for its underwear compete in the world of retail electronics? Until six months ago, we had never sampled electronic products from Penney's. Now we have. And we're impressed.
The J.C. Penney tabletop Video Cassette Recorder (model 686-6184) is a four-head, HQ, eight-program, three-week marvel of the complicated made simple enough even for pre-electronic grandmother. Here's a VCR so user-friendly that you might consider inviting it into the kitchen after supper for a few hands of pinochle. And of course, it would probably win.

The 6184 comes with everything you'd expect from a current-generation VCR. It offers three-speed playback, on-screen display, fast-forward and fast-reverse plus those standard accessories, cables and adapters, for whatever hook-ups you might require.

And even grandma can follow the easy instructions in the Penney operating guide. Beyond basic features, this unit offers a double-speed playback plus variable-speed slow-in and single-frame advance that can be used against the pause/still function.

There's a nifty channel-memory setup and a one-touch record feature, this in addition to the standard recording mode. We weren't quite sure how or why we'd ever need this one-touch capability until the phone rang during the premier of one of our favorite new shows. Bingo—we pressed the button and didn't miss a minute of "Jake and the Fatman."

An Help function guides first-time users or the electronically fearful. But the machine is so easy to program that not even Uncle Jim is going to need to engage this more than once.

If we've got a complaint, it's that the wireless remote's buttons are a little too small for adult fingers trying to piano-up from one channel to another. The 6184 hasn't got the most stylish box on the block, but then the price doesn't come close to what you might pay for swank name-brands available from other retailers. Ours was purchased at a sale price considerably below Penney's suggested retail tag.

The only feature lacking is high-fidelity sound. Given a sound capability beyond one-track, the unit would be darn near perfect, especially as a first VCR in just about anyone's home. And yes, there are still some people without one.

For those who want to use the 6184 in conjunction with a video camera (or with a second VCR) for recording or re-recording, the operations manual gives clear instructions on how-to. And beyond all that, this Penney's jewel runs about as quiet as a kitten lost in yarn.

Penney was unwilling to name the manufacturer of this store-brand VCR, but there may be a hint in the manual's section on service which directs the owner to the nearest RCA factory service center. Information Penney did share was that after July, the 6184 will no longer be supplied by the company to its stores. Except for consumers who insist on the latest model, the Penney change-over most likely means that exceptional prices will be offered on what we found to be an exceptional, home video recorder.—L. B.

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### Tale of the Tapes

**TEAC STEREO DOUBLE CASSETTE DECK (D-525C).** Manufactured by: TEAC Corp. of America, 7733 Telegraph Rd., Montebello, CA 90640. Price: $259.

The long-play record, a staple of the recording industry's advance into the modern era of audio technology, is fast becoming a thing of the past. After decades of dominating music software sales (predating the term software itself), the LP began a precipitous decline early in this decade.

Since 1983, more cassette tapes than LPs have been sold. In 1986, 125.2 million albums were purchased, a 25 percent drop from a year before. Meanwhile, 344.5 million tapes were sold, only a slight percentage increase over the previous year's sales.

Clearly, despite the excitement (and hype) surrounding the compact disc and yet-to-be-marketed digital audio tape formats, the cassette is the dominant audio software of the 80s. This has as much to do with the proliferation of affordable playback systems as it does with any inherent advantages of the tape cassette format.

The record industry freaks out around the question of the public's ability to record and copy its own music programming. This is why the cassette format met plenty of music and record industry opposition in the 1970s. The industry put its early money on eight-track tapes, especially because they were inefficient as a copying or duplication medium. Eight track tapes eventually became the recording industry's Edsel.

Now practically every kid on the street, or in the mall, has a portable cassette player. Tape cassettes are much more durable than LPs. The tapes stretch and break and all but the highest quality tapes eventually fade as particles are scraped off of the surface each time it passes over the heads of a playback system. Likewise, portable cassette players are seemingly designed to fall apart relatively rapidly. Flimsy cases, break-off buttons and the inevitable fall-to-the-pavement contribute to a neat bit of planned obsolescence that has most street players being replaced as often as sneakers.

That's why the reliable but inexpensive TEAC Dual Cassette Deck, the D-525C, is such a superior machine. No portable unit priced at even twice its tag (often under $150 on sale) can offer the advantages of this sturdy component.

Of course, it won't be used for on-the-spot mixing or dubbing by urban street rappers. But anyone interested in making a high-quality copy of a favorite record before the inevitable vinyl wear-and-tear manifests itself, or making a duplicate of a favorite music tape will find this an invaluable home machine.

Although the D-525C lacks such electronic amenities as auto-reverse (a favorite feature for portable machines) or remote control, its high speed and synchronized dubbing capabilities provide the discriminating do-it-yourself tape engineer with the kind of quality only found in a recording studio a decade ago.

This is the kind of unit the record business would prefer the public didn't have. The music industry opposed the cassette format when it was first marketed, fearing a loss of revenue from home copying.

Few of those same naysayers would any longer question a format that is now the industry's bread and butter. But it's ironic that their fears regarding the new format, digital audio tape (DAT), has distracted industry watchdogs enough to allow a dual cassette machine like the D-525C to become available to virtually any consumer.

Although it comes arrayed with the sort of visual display with which manufacturers proclaim their audio sophistication, the D-525C is a simple-to-operate unit. It may not be the most sophisticated dual cassette deck on the market, but it's far from the most expensive, either.

Finally, if you're inclined to feel a little guilty about home taping of music, consider these figures: In 1973, 614 million records and tapes shipped and were estimated to be worth a little more than two billion dollars. In 1986, the numbers of records shipped (618.3 million) was roughly the same. But the estimated list price dollar value of those records and tapes had more than doubled to 4.651 billion dollars. For consumers, this might make the music business's fears about home taping difficult to take seriously.—J.S. (Next follows items that would normally go in the Bits & Pieces section of the old Gadget. These now go under the head Gizmo/Bytes.)
Blaupunkt Dallas AM/FM-Stereo Radio SQM 88

Car stereos get installed, they get stolen but one thing which doesn't happen much to them is getting reconfigured. Blaupunkt (2800 S. 25th Ave., Broadview, IL 60153), the long-established, prestige audio manufacturer has changed that with its new Dallas FM/AM Stereo Radio (SQM 88). Instead of being mounted in the dashboard, the Dallas sits on a flexible stalk, readily maneuvered into any desired position. It's easily removed for theft deterrence and is small enough to carry in a coat pocket or handbag. As Blaupunkt points out, "this configuration also frees the normal radio mounting hole for a tape player, CD player or future DAT player." A separate module, which can be installed anywhere in the vehicle, carries the circuitry for the Dallas radio. Among features of the radio itself are dynamic noise reduction, scan, preset scan, 15 AM and FM presets, base and treble controls and switching and inputs for two auxiliary audio sources. Price: $369.95.

CIRCLE 57 ON FREE INFORMATION CARD

Panasonic VHS camcorder

Camcorders, in all their many splendored formats seem out to individualize themselves with non-duplicated special capabilities. A new VHS Camcorder (PV-330) from Panasonic Co. (1 Panasonic Way, Secaucus, NJ 07094) offers what are described as "digital effects," namely a single-touch freeze frame and, fancier still, "combination effects," with a still image appearing in one of the corners of the viewing screen. The freeze-frame image can then be made to disappear with a "wipe" from right to left replacing "the freeze frame with a live image." In a second mode, "the live picture expands outward from the center of the screen." Finally, an "overlap, undoubtedly better seen than described, in which "the moving image automatically overlaps the still frame." Other advanced features include a CCD image pick up light sensitivity "approaching one lux," a "three-way high speed shutter mode" and a tape indexing system. What's next. 3-D? Price: $2,000.

CIRCLE 58 ON FREE INFORMATION CARD

Shinton VCRs

Henry Ford, it will be recalled, became world famous for low-price, no-frills transportation. Shinton West Corp., (20435 S. Western Ave., Torrance, CA 90501) is apparently out to do for home video what Henry did with his basic unit. The firm is marketing a VHS VCR (4040N), with one-touch recordings, HQ circuitry and 14-day/six-program timer) at surprisingly no-frills suggested retail. The unit utilizes an automatic playback system for the ultimate in passive video viewing. You front-load the cassette and the VCR does the rest, right through rewind. Besides the 4040N, Shinton has rolled out the 4030N, virtually identical except for its more subdued box, which brings a slightly higher price. Shinton says the silver-finished 4040N is the mass merchandiser unit. Price $289.95.

CIRCLE 59 ON FREE INFORMATION CARD

Plus U.S.A. Halftime-Scissors

If there was such a category as "designer office accessories," Plus U.S.A. Corp. (10 Reuten Dr., Closter, NJ 07624) would be recognized as a fashion leader. The firm's newest designer gadget is the Halftime-Scissors, three-inch stainless steel blades are contained in a two-piece matte plastic case not much bulkier than a fountain pen. Removing the protective cover opens the blades for business. The item is available in red, white, blue and black. Price: $9.95.

CIRCLE 76 ON FREE INFORMATION CARD

Healthcheck blood pressure monitor

It used to be nobody was much interested in blood pressure except doctors. Now, everybody seems to be monitoring their own, with the help of coin-operated testing units, street-corner pressure testing stations and home health care devices. Including the compact Healthcheck Digital Blood Pressure Monitor. Sold by Markline Co., Inc. (14 Jewel Dr., Wilmington, MA 01887), this unit has the advantage of taking a blood pressure and pulse reading via an adjustable band that fits onto the user's finger. Featuring something called "Pulsonic technology," the Healthcheck features digital display and is small enough to fit into a briefcase or purse, complete with protective carry case. Price: $79.

CIRCLE 64 ON FREE INFORMATION CARD
Polaroid PhotoMagic System

Given all the changes in the delivery and reproduction of images that are afoot in the consumer market, it doesn’t pay to rest on your previous achievements. Even when the achievement is something epoch-marketing like instant photography. Polaroid Corp. (575 Technology Square, Cambridge, MA 02139) has a new product which it’s presenting as “an exciting new profit-making opportunity” for enterprising individuals as well as “an ideal fund-raiser for charitable and community groups.” Dubbed the PhotoMagic System, this instant camera automatically superimposes graphics on the film, framing the subject in the instant picture. The system includes a way to quickly die-cut the finished picture and mount it in specially designed buttons or picture frames. In addition to the camera, the PhotoMagic system includes a portable picture button cutter, acetate art-forms, plastic buttons and specialized frames. All packaged in a black vinyl case. Price: $525.

CIRCLE 60 ON FREE INFORMATION CARD

Panasonic Bread Bakery

The staff of life goes space age with a newly introduced kitchen appliance from Panasonic (1 Panasonic Way, Secaucus, NJ 07094). Developed for the Japanese consumer market, the Bread Bakery (SD-BT2P) is a counter-top unit that manufactures one-pound loaves of bread. First a revolving mechanism mixes and kneads the ingredients. After the loaf rises and assumes the shape of the Bread Bakery’s bread canister, it’s baked by a single heating element. The process takes some four hours, but a built-in timer can set the unit to remain dormant for up to nine hours before bread-making commences, assuring a piping hot loaf when the user wants it. Panasonic says its Bread Bakery can also knead dough for such foods as croissants, doughnuts and even pita dough (using the “leavening mode”). But the Bread Bakery obviously will rise and fall on its merits as a hot bread source, and consumer interest in that baked delicacy. Price: $329

CIRCLE 63 ON FREE INFORMATION CARD

Hewlett Packard Business Consultant II

Not only do today’s sophisticated calculators do laborious computations for us, but they’re beginning to do them in ordinary, non-technical language. For example, the recently introduced Business Consultant II Handheld Calculator can use words and symbols as variable names in equations entered by the user and permits entry of equations without programming. From Hewlett Packard (3000 Hanover St., Palo Alto, CA 94304), the Business Consultant II offers messages and prompts in English, German, Spanish, French, Italian and Portuguese and, according to HP, may be “the first business calculator to offer statistical and cash-flow graphics capabilities.” It can produce scatter diagrams, curve fits and histograms, as well as cash-flow graphs. Finally, the unit also has special TEXT application for name-and number lists, bonds and depreciation, currency and unit measurement conversions and an appointment menu with an alarm clock. Price: only $175.

CIRCLE 69 ON FREE INFORMATION CARD

Onkyo CD Player with changer DX-C600

You don’t have to be clairvoyant to see what the next stage in the developing CD player wars is likely to be. Onkyo U.S.A. Corp. (200 Williams Dr., Ramsey, NJ 07446), like a number of its competitors, is offering a CD Player/Changer (DX-C600) with six-disc capacity. Onkyo hopes that features like its “acclaimed opto-coupling” technique which electronically separates digital and analog circuit blocks in order to free the unit’s playback sound from the harshness sometimes associated with the CD medium will put the DX-C600 over the top with consumers. The unit’s programming is so sophisticated that you don’t even have to have a disc or disc magazine loaded when you’re entering memory play. In other words, you can program discs that aren’t even there, kind of spooky when you think about it. Anyway, the next CD player/changer king of the hill, for at least a couple of weeks, will be the brand that markets a seven- or eight-disc changer. In the meantime, the Onkyo DX-C600 joins the battle of the sextets. Price: $600.

CIRCLE 81 ON FREE INFORMATION CARD
Canon VM-E2N 8-mm camcorder

Camcorders in the 8-mm format are keeping pace with their more widespread VHS competitors. Canon U.S.A., Inc., (1 Canon Plaza, Lake Success, NY 11042) has introduced a new 8-mm Video Camcorder (VM-E2N) which boasts a 250,000 pixel charge-coupled device image sensor, and recording capability at illuminations as low as five lux. Something Canon dubs a “flying erase head” aims to eliminate blanks and video noise between scenes, ensuring “clean in-camera editing.” An intriguing extra is a Sepia tone button. The function will record in brown-toned Sepia, giving the image an antique quality. The VM-E2N also boasts, among other features, high fidelity audio, automatic focus, zoom lens, two-hour recording capability and five-second review. Price: $1,699.

CIRCLE 61 ON FREE INFORMATION CARD

Adjustable Exercise Bike Reading Stand

A healthy mind and a healthy body are supposed to have something to do with each other. Now from a leading manufacturer of fitness equipment, Bodyguard (40 Radio Circle, Mt. Cisco, NY 10954), comes a product to allow you to improve one while working on the other. It’s an adjustable Exercise Bike Reading Stand, to engage the mind while exercising the body. The unit clamps onto the neck or handlebars of any brand exercise bike and features an adjustable page-holder. Made of smoked Plexiglas, we just hope these don’t start appearing on non-stationary bicycles. Price: $21.99.

CIRCLE 71 ON FREE INFORMATION CARD

Olympia USA, Inc. paper shredders

Perhaps inspired by last year’s Iran-Contra Congressional hearings, Olympia USA, Inc. (Box 22, Somerville, NJ 08876-0022) has entered a new area of business equipment, paper shredders. The firm is offering a full-service line of 13 of these security tools, including the Olympia Desk-Side Shredder (2206). “Straight-cut models,” the firm says, “turn documents into illegible strips,” while “cross-cut” units reduce “a single sheet of letter-size paper to create 2,000 confetti-size particles.” In common with other models, the 2206 can take staples and clips, features electronically controlled automatic reverse which eliminates paper jams and messy overflow and dust-free shred collection in fully enclosed plastic bags. All this and styles ranging from pearl gray to walnut wood finish. Price: $1,349.

CIRCLE 72 ON FREE INFORMATION CARD

Zenith home-satellite receiving systems

Scrambling of signals was supposed to bring big changes to home satellite TV. But if a new product from Zenith Electronics Corp. (1000 Milwaukee Ave., Glenview, Ill 60025) is any indication, the change may not be so big. Zenith, one of the few electronic giants to dive head-first into home satellite reception, is now offering Home Satellite Receiving Systems with built-in decoders. The ZS-6000 features an integrated combination receiver/decoder with something called Vid-eocipher II, along with on-screen displays, stereo sound, parental security, automatic antenna positioner, wireless remote control, automatic program recall, front panel displays and a ten-foot antenna. Price: $2,799.

CIRCLE 74 ON FREE INFORMATION CARD

Canon Cat work processor

We’re not sure what the implications of this are, but if the product succeeds, it probably represents a new stage to the computerization of daily life, or something. From Canon U.S.A., Inc., (1 Canon Plaza, Lake Success, NY 11042), the company is calling it the Cat Work Processor. While the press release avoids the word computer, the Cat is described as combining “typing, word processing, information storage and retrieval, calculating and communicating with other Cats” or even computers. It’s storage capacity is approximately 80 pages or 180 KB. Designed to do away with menus, windows, files, booting, formatting or computer terminology or jargon, it only requires the desk space of a regular typewriter. Its most distinctive feature is a jump function, “the first improvement in cursor control since the mouse.” Two jump keys control text editing, rearrangement, spelling checking, revision, underlining and bold facing. Aimed at office users, the Canon Cat sounds ripe for consumer sales. Price: $1,495.

CIRCLE 75 ON FREE INFORMATION CARD
Nakamichi CA-5A II Control Amplifier

Even as capabilities become more complex, audio equipment is being simplified. This according to Nakamichi America Corp. (19701 S. Vermont Ave., Torrance, CA 90502) in announcing its new Control Amplifier (CA-512). This new pre-amp features defeatable tone controls, two-way tape dubbing capability with five high-level inputs plus phono input. Especially noteworthy, Nakamichi proclaims, is the unit's "elegantly simple topology... the simpler the circuit, the better the sound." Price: $995.

CIRCLE 70 ON FREE INFORMATION CARD

Sonin ultrasonic measuring device

It looks as if the measuring tape and the yardstick might be on their way to the museum of obsolete tools. Sonin, Inc., (672 White Plains Rd., Scarsdale, NY 10583) markets a device which measures the electronic way. Called the Sonin 60, the device uses ultrasonic waves, processed via microchips, to measure distances. Pocket-size, the unit is placed against a wall and pointed in the direction to be measured. Readings appear on an LCD and the Sonin 60 includes dual memories and a calculator to assist in computing areas and volumes of measured space. According to the company, the device incorporates three electronic safeguards. One prevents false readings due to variations in air temperature and humidity, a second eliminates stray signals from objects near the measuring path and the third adjusts the Sonin's response to temperature changes. Price: $99.95.

CIRCLE 65 ON FREE INFORMATION CARD

UPS Shutdown/Electronic Specialists, Inc.

So what happens if a micro-computer's battery back-up system discharges before interrupted regular power resumes? Most PC users probably don't want to find out which is where the Power Loss Shutdown from Electronic Specialists (171 S. Main St., P.O. Box 389, Natick, MA 01760) comes in. The device will "shut down an entire computer system if commercial AC power has not yet been restored when UPS or SPS batteries have discharged," thereby preventing "undesired computer and power system start up after a power outage." In case you're wondering, Electronic Specialists say "models with adjustable AC line drop-out voltage level" are available. Price: $150.

CIRCLE 66 ON FREE INFORMATION CARD

Career Navigator

Here's an unfortunately timely new software package, the kind PC users may not want, but one which they eventually could need. The Career Navigator is dubbed "the computer-powered job search system" and it's been developed by Drake Beam Morin, Inc. (100 Park Ave., New York, NY 10017) which calls itself, "the world's largest career counseling and out-placement firm." Career Navigator includes a job search handbook and four disks which facilitate self-assessment, creation and storage of up to five different resumes and ten different cover letters, a file of up to a hundred names and addresses, a progress monitor and a permanent personal database for lifetime updates and reference. Drake Beam Morin calls the program "highly inactive and individualized." Career Navigator is available for the IBM PC, XT, AT and compatibles with 256k, color or monochrome monitor and dot matrix or letter quality printer. Price: $95.

CIRCLE 73 ON FREE INFORMATION CARD

Oster four-cup coffeemaker

Although this product is pitched for households with one or two members, we think it might be just the ticket for people trying to cut down on their coffee intake under any circumstances. It's a new Four-Cup Coffeemaker (3400) from Oster (5055 N. Lydell Ave., Milwaukee, WI 53217) with a 20-ounce capacity which heats water to brewing temperature in just 90 seconds. The model comes equipped with separate on/off signal lights for both the coffeemaker and its warming tray and the water reservoir is stainless steel. Price: $45.95.

CIRCLE 67 ON FREE INFORMATION CARD
HAVE YOU EVER NEEDED TO TALK ON THE PHONE FOR long periods of time while having both hands free? I had that problem while working as an operator at a telephone solicitation company and found that I was getting many a neck ache from trying to pinch the phone between my shoulder and ear while writing. To solve the problem I invented a circuit to connect my stereo's headphones to the telephone. I hope that you will find it useful too.

What the Circuit Does

As well as being able to use both hands, we also want to be able to increase the volume of weak incoming calls and decrease the background noise. At the same time the circuit should still allow the use of the phone in the regular way without switching cables.

The Connection of Handset to Phone

This circuit will work only with a handset that has no electronics in it. If you have the right kind of phone it will have a cable between the phone and handset with four wires in it. Two of the wires go to the mike and two go to the speaker, see Fig. 1. The wires to the mike form a current loop with the microphone, which is a carbon granular mike, acting as a variable resistor. Its resistance changes with air pressure and therefore varies the current in the loop. The speaker connection is just like a speaker connection on a stereo. An AC signal comes from the phone and is applied across the speaker terminals.

Connecting the Headphones

In order to connect the headphones to the phone we have to do two things. First we must get the signal that would normally go to the speaker in the handset routed to the headphone speakers. Next we must replace the handset microphone current loop with a circuit which will keep the DC components the way they used to be while injecting the AC voice component from the electret microphone used on the headphones. Note that this microphone does not work the same way a carbon granular mike does and it is not a direct replacement for the carbon granular mike.

Figure 2 shows the replacement circuit diagram in block form. The signal from the phone that used to go to the handset speaker is broken. One wire is taken as ground and the other is connected to a volume control. Next the signal is AC...
coupled (capacitively coupled) to the incoming amplifier which amplifies the signal and drives both of the headphone speakers.

The replacement circuit for the microphone must have the same DC resistance as a carbon granular mike and allow the injection of AC signals into the loop. The circuit consists of an audio isolation transformer and a resistor in series to get the DC resistance correct. A capacitor bypasses the AC component of the injected signal around this current-setting resistor. The signal from the headphone's electret mike is connected to the outgoing volume control. The signal is then AC coupled to the outgoing amplifier and the output of the amp drives the primary of the isolation transformer. This injects the audio signal into the DC current loop.

The Talk-Back Problem
Talk-back is not something the phone does when it wants to hassle you, but rather it was designed into the phone to make you feel more comfortable while using it. To control the volume at which a user speaks, the phone purposely feeds a small percentage of the signal it receives from the mike back to the speaker so that the person will hear themselves talk and know that the phone is working properly. Without talk-back users tend to yell into the phone.

In two situations our device suffers from this talk-back system. Fig. 3 shows the difference between a regular and weak call. The phone always feeds the same portion of the mike signal back to the speaker no matter what the incoming signal strength is. When we receive a weak call our device can compensate and bring the call back up to normal level by amplifying the signal between the phone and the speaker. The problem is that the amplifier amplifies both the incoming signal and the talk-back signal that the phone is sending to the speaker. The result being that the talk-back signal ends up

Fig. 1—In a telephone, resistance varies with air pressure from the user, thereby changing the current in the loop. That shows up on the phone line as a small AC signal superimposed on the DC voltage.

Fig. 2—The headphone interface must pass the mike current while imposing the head set's mike signal on the line. That is done with the isolation transformer shown.

Fig. 3—A phone pipes some of the mike's signal back to the earpiece so you sense the phone working. Because of that, any external amplifier set to amplify a weak incoming call will amplify the signal being fed back as well.
way too strong and hurts the ears of the user. The user's natural reaction is to talk softer so that it doesn't hurt, but if they do this the other party now has a hard time hearing them because the outgoing signal is now weaker.

Another problem with talk-back is that the phone is feeding back the noise that is being picked up by the microphone when the user is not talking. If the incoming amplifier is set for a high gain to compensate for a weak connection, it will also be amplifying the background noise. The background noise can then become as loud as the other party. Further amplification will not help as it will still amplify both signals.

**Beating Talk-Back Problems**

One way to fix the trouble would be to reduce the amount of talk-back on weak calls and restore it to normal for strong calls. Unfortunately, the amount of talk-back is determined inside the phone. The only way to change the amount of talk-back is to alter the inside of the phone.

If we instead change the gain of the incoming and outgoing amps as the user speaks and listens we could correct the problem. Here is how it works: Assume the incoming call is weak, therefore it needs to be amplified to be brought up to a normal volume. When the user is not talking, the gain of the outgoing amp is set low. Since the gain of the incoming amp is set high, the call is being amplified to the proper level. Since the gain of the outgoing amp is set low, little background noise is going in the mike input of the phone, therefore the amount being fed back is also low, and there is no background noise heard while listening to the other party.

When the user speaks we change the gain settings so that the outgoing gain is at the normal level and incoming gain is at unity. Since both gains are at the normal levels the talk-back is just right and the user feels comfortable and speaks at the correct level. Our output signal is thus the correct strength.

Note that we could not hear the other party with poor settings because they would be very weak and we would hear a lot of background noise, but gain is necessary so that the outgoing signal and the talk-back level is right.

Figure 4 shows our solution to the problem. We provide two volume controls for each amplifier. One for whenever the user is talking and the other for when the user is listening. Those are referred to as the talk mode and the listen mode respectively. We provide a voice activated circuit that switches between the volume controls automatically when the user speaks.

**Circuit Operation**

Figure 5 shows the complete schematic. There are actually four sections to this circuit. The first is the incoming amp. The second is the outgoing amp. The third is the voice activation circuit, and last is the power supply, which simply consists of a switch, nickel-cadmium batteries, and a plug for connecting an AC adapter to charge the batteries. The Parts List shows all the catalog numbers to obtain the parts either from Radio Shack or through the mail from Mouser electronics with their address.

**Incoming Amplifier**

The signal from the phone enters the box and goes directly to the two 5K incoming volume controls, R1 and R2. They are mounted on the front panel for user access. A shielded cable goes from these controls up to the PC board and connects to the two coupling capacitors, C1 and C2.

Note that all shielded cable used has three conductors in it. There are two center conductors, colored red and white, and the shield surrounds both of those.

Capacitors C1 and C2 connect the two different volume incoming signals to U1, the 4066 switch. The switch selects the signal to be passed on to the incoming amplifier. Another shielded cable is used to take the signal to the base of Q1. Transistors Q1 and Q2 form the incoming amplifier, which is a dual-stage class-A amplifier. The amplified signal is taken from the collector of Q2 and sent to the headphone speaker jack. J1. Switch S1 allows the handset speaker to be connected either straight to the phone, to the output of the incoming amp, or shut off.

**Outgoing Amplifier**

The outgoing signal originates at the electret microphone mounted on the headphones. It is brought from here via shielded cable to plug J4. The shield is used for ground, the white is the audio signal from the mike, and red carries V+ to the mike to power it. Plug J4 is plugged into jack J2. Both of these have three connectors to carry the signals mentioned above.

From J2 a shielded cable runs to volume control R3, the background volume mounted on the front panel. The shield is used as ground, the red wire takes the audio signal to one end of the pot, and the white wire brings the audio back from the
As you can see, the circuit is composed of four main sections: incoming and outgoing amplifiers, voice activation, and headset circuits. The incoming and outgoing amps are identical except for their output coupling.

Another shielded cable originates at jack J2 and goes to the PC board. The shield of that cable is ground and the white wire carries the audio from the microphone. The red wire takes the signal coming back from volume control R3 up to the PC board.

At the PC board the red wire is connected to coupling capacitor C4 which leads to U1, the 4066 switch. The white wire goes to R4, which is the outgoing volume control and is mounted on the PC board so that the user cannot change the output level of the circuit.

The signal from R4 is fed to the switch via capacitor C3. The switch selects which signal is to be used and the signal is fed to the base of Q3.

Transistors Q3 and Q4 form the outgoing amplifier. Their arrangement is the same as that used for the incoming amplifier. The outgoing signal is taken from the collector of Q4 and is hooked to the primary of the isolation transformer, T1, by capacitor C8.

The secondary of transformer T1 and resistor R24 form the DC current loop that replaces the carbon granular mike from the original handset. R24 allows for the adjustment of the DC current value. Capacitors C13 and C14 bypass the AC audio around resistor R24. They are connected back to back to form a bipolar capacitor because the polarity of the connection with the phone can not always be counted on. Switch S2 allows either the circuit to be hooked to the phone or the original microphone in the handset. It's center position allows both microphones to be disconnected so the caller cannot hear what is being said.

**The Voice Activation Circuit**

The signal for the voice activation circuit is taken from the white wire of the cable that goes between jack J2 and the PC board. The signal is the audio as it comes from the mike without going through any volume controls. capacitor C9 couples the signal to IC2, an LM324 Quad op amp.

The first opamp boosts the signal 393 times. The amplified signal is brought back down to ground by capacitor C11 and resistor R18. Diode D1 gets rid of the bottom halves of the signal and capacitor C12 filters the signal so that a DC voltage of the value of the peak of the AC signal appears at the + terminal of the second opamp.

The DC value depends on whether, and how loudly the user is talking. Resistor R19 discharges capacitor C12 to provide a short delay in the switch of volumes after the user stops talking.

Trigger-level control R5 determines what voltage will be used as the reference for determining what is talking and what is not. The output of the second op amp is the talk signal and the third op amp inverts this to get the talk-bar signal. Those two signals are connected to the control pins on U1, the 4066
switch to determine which volume controls to use. LED2 displays the status of the talk signal, making it easy to adjust the trigger level.

**The Power Supply**

Six nickel-cadmium rechargeable batteries supply the power to operate the circuit. Switch S3 allows power to the circuit to be turned on or off. Jack J3 is hooked to the batteries to allow an external charger to be connected. Note that the size of jack J3 is different depending on the source of the charger used, check the parts list carefully. The circuit will run between 10 and 15 hours with the batteries listed. Charging time is about the same.

**Constructing the Circuit**

Be sure to use shielded cable in the places I have noted. I housed my circuit in a small plastic box from Radio Shack, catalog number #270-223, and hand wired it using a perforated board.

Put four of the potentiometers, the three switches, and the LED on the front panel. Both of the incoming volume controls, R1 and R2, and the listen-mode outgoing volume (background volume) control, R3, should be mounted on the front panel. The trigger level potentiometer, R5, should also be mounted on the front panel.

Jacks J1, J2, and J3 should be mounted on the sides of the cabinet. Remember to leave two holes for the handset cables going to the phone and handset. You make the two cables by cutting the one handset cable specified in the parts list in two. I made the piece to the phone shorter than the one to the handset so I could tell them apart.

I mounted the microphone on the headphones by drilling a hole for a number-six screw in the left side of the headphone's band and bent a piece of heavy aluminum wire to hold the mike. You will need a windscreen over the microphone so that air rushing in and out of your mouth doesn't disturb the microphone. I used a scrap one from one of those old Mr. Microphones. Any headphones will work. You will have to find a suitable way of mounting the boom for the mike.

**Preliminary Setup**

Before adjusting your circuit you will have to set it up in the proper way. First you must charge the batteries. Set switch S3 to the off position. Put the batteries in their holders, connect the AC adapter to jack J3 and plug the adapter into an outlet. If you observe the voltage across the batteries it should go up to between 7.5 and 9 volts when you plug in the adapter. Allow the batteries to charge for about 12 hours. An alternative would be to use regular batteries to test the circuit.

Monitor the current flowing out of the batteries when switch S3 is turned on. It should be no more than about 30ma. If it isn't, check the circuit for shorts or wrong-value resistors.

**Adjustments**

You will have to adjust a couple of things once you have your basic circuit working. First you should test your phone and find out what the normal value of current is in the microphone current loop. Once you know this value you change the value of R24 to make the value the same when using this circuit. To make this adjustment be sure that the cable to the phone is connected and that switch S2 is set to the headphone position.

The second thing you should do is to check the voltages at the collectors of the four transistors. They should all be within one or two volts of half the power-supply voltage. If they are not, you can adjust the voltages on Q1, Q2, Q3, or else Q4 by changing bias resistors R5, R8, R10, or R12...
The third thing you must adjust is the output volume of the unit. Before you can do that however, you must set it up correctly. Connect the headphones to jack J1 and the microphone to jack J2. Connect the correct cable to the phone and the other one to the handset. Leave the handset on the phone at this time. Set switch S1 to the center (off) position and S3 to the on position. Speak into the microphone and observe LED2. When you speak the LED should light and then go out when you stop speaking. Adjust the trigger level control R5 if the LED is not lighting at the correct time.

Final Connections

Now connect either an oscilloscope or another phone to the phone line. If you use an oscilloscope be sure to use a three- to-two wire adapter on the power cord because the phone company grounds one of the wires in the phone line.

Set the mike switch S2 to the handset position and lift the handset. Speak at a normal level and observe the strength of the signal on the oscilloscope or the other phone. Next change switch S2 to the headphone position and speak into the headphone’s mike. Continue to go back and forth between the two while adjusting volume control R4 to get the levels the same.

Using the Circuit

Before using the circuit connect it to the phone as in the previous section. Also connect the headphones and microphone to the circuit.

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**PARTS LIST**

FOR THE TELEPHONE-CONTROL BOX

<table>
<thead>
<tr>
<th>SEMICONDUCTORS</th>
<th>RESISTORS</th>
</tr>
</thead>
<tbody>
<tr>
<td>D1—1N4001 small-signal diode, Mouser #333-1N4001</td>
<td>R1, R2—5000-ohm, linear-taper potentiometer, Mouser #31VA305</td>
</tr>
<tr>
<td>D2—Red light-emitting diode, Mouser #35BL501</td>
<td>R3—10,000-ohm, linear-taper potentiometer, Mouser #31VA401</td>
</tr>
<tr>
<td>Q1, Q3—PNP small-signal transistor, Mouser #333-KN3904</td>
<td>R4—10,000-ohm, PC-mount potentiometer, Mouser #32NA401</td>
</tr>
<tr>
<td>Q2, Q4—PNP small-signal transistor, Mouser #333-KN3906</td>
<td>R5—100,000-ohm, linear-taper potentiometer, Mouser #31VA501</td>
</tr>
</tbody>
</table>

**CAPACITORS**

C1—C5, C7—22-mF, Mouser #23AK422
C6—470-mF electrolytic, Mouser #20WH470
C8—4.7-mF electrolytic, Mouser #20WK005
C9, C12—47-mF tantalum, Mouser #551-47M35
C10, C11—2.2-mF tantalum, Mouser #551-2.2M16
C13, C14—1000-mF electrolytic, Mouser #20WH901

**ADDITIONAL PARTS AND MATERIALS**

J1—¼-in. stereo phone jack, Mouser #16PJ081
J2—¼-in. phone jack, Mouser #16PJ108
J3—¼-in. phone jack (for Mouser AC adapt given below), Mouser #16PJ125
J4—¼-in. stereo phone plug, Mouser #17PP004
M1—electret microphone, Mouser #25LM044
S1,S2—SPDT, center-off toggle
S3—SPST toggle
T1—600-600-ohm audio isolation transformer, Mouser #42TM016

Handset cable, gridboard (6 x 2-in.), Mouser #ME153-1102; case (6 x 3-1/8 x 2-in.), Mouser #546-1591DS-BK; 2-“AA” battery holder, Mouser #128H321; 4-“AA” battery holder, Mouser #128H348; 9-Volt battery snaps (for R.S. holders only); 6-“AA” nickel-cadmium batteries, Mouser #547-PS-AA, AC adapter jack J3 depends on adapter, Mouser adapter #41AC109; knobs for volume controls, solder, wire, etc.

Set the three volume controls on the panel to about one quarter turn. Leave the trigger level set to where it was found to work best in the previous section. Set the mike switch, S2 to headphone and the handset speaker switch, S1, to off. Turn the power switch, S3, on and adjust the trigger level control R5, if the LED is not lighting at the correct time when you speak into the headphone microphone.

When you lift the handset you should hear a dial tone in the headphones. Adjust the volume of the dial tone with the caller volume control, R2. Set the volume you hear yourself with the talk-back volume control, R1. Adjust the amount of background noise you hear with the background volume control R3.

To have another person listen to your conversation just set the handset speaker switch to either normal or amplified and give them the handset. The mike in the handset is still off so the caller will not hear the person with the handset.

To use the handset normally set the mike switch to the handset position and the handset speaker switch to the normal position. While using the handset normally another person can listen on the headphones without being heard by the caller. If the headphones are not to be used while the handset is being used, the power can be turned off, thus saving the batteries.

Hopefully your circuit will serve you as well as mine has served me. I have used mine for about 500 hours now and it has never failed. With a device like this you can let your fingers do a lot of walking!
In the old days, variable power-control applications were accomplished through large variable resistors and thyratron tubes, which made the circuit bulky and extremely expensive. However, the development of the thyratron-transistor—more commonly known as a thyristor, and sometimes referred to as a reverse-blocking triode thyristor—changed all of that; making possible a more reliable and efficient replacement for its predecessor, the thyratron tube.

The thyristor is not a device unto itself, but rather a whole family of devices. That family includes the SCR (Silicon-Controlled Rectifier), most widely used as a power-control element: CSCR (Complementary, Silicon-Controlled Rectifier), defined as a negatively-triggered SCR; Triac (quite popular in low-current, under 40-ampere, AC circuits); SCS (Silicon-Controlled Switch); ASR (Asymmetrical Silicon Rectifier); SUS (Silicon Unilateral Switch), and SBS (Silicon Bilateral Switch), which are most commonly used as gate-trigger devices for power-control elements.

And that’s not all there is to this family: There are a few other devices that can be used as solid-state switches, in variable power-control circuits, and for low-voltage control of high-power circuits.

To successfully use this family of devices, an understanding of their characteristics, ratings, and limitations is imperative. Therefore, we’ll cover some of the more significant thyristor characteristics, the basis of their ratings, and their relationship to circuit design.

SCR’s

Figure 1A shows the schematic symbol for the SCR—the first, and probably the most important member of the thyristor family. Its structure is like that of a four-layer diode, with a gate connected to the P-section nearest its cathode.

Fig. 1—Shown here is the schematic symbol for the SCR; B illustrates the PNPN structure the schematic symbol represents. A two-transistor SCR construct (similar in structure to the SCR) is shown in C and the four-quadrant characteristic curve for the SCR is shown in D.
as shown in Fig. 1B. A two-transistor SCR construct (which is similar in structure to the SCR) is shown in Fig. 1C, and Fig. 1D shows the IV-quadrant characteristic curve for the SCR.

The interconnection of the transistors (see Fig. 1C) is such that if current is injected into any leg of the construct, the gain of the transistor (if sufficiently high) causes current to be amplified (regenerated) in the other leg. In order for regeneration to occur, the sum of the common-base current gains (a) of the transistors must exceed unity. Because junction leakage current is relatively small and current gain is designed to be low at the leakage-current level, the PNPN device remains off unless externally acted upon.

When sufficient trigger current is applied to the gate of the construct (raising loop gain to unity) regeneration occurs, and the on-state principle current is limited primarily by external circuit impedance. Removing the initiating trigger does not effect turn-off provided that the unity-gain criteria (or latching-current requirement) is met.

![Diagram showing interaction of gate current and voltage in an SCR.](image)

**Fig. 2**—Here is the interaction of gate current and voltage in an SCR. When gate current (I_g) is zero, the applied voltage must reach the breakover point before switching occurs. But, as gate current increases, the SCR's ability to block the applied voltage is reduced, and turn-on occurs.

![Graph showing gate sensitivity as a function of temperature.](image)

**Fig. 3**—Gate sensitivity is shown here as a function of temperature. Since junction-leakage currents and current gain increase with temperature, the magnitude of required gate-trigger current decreases. The voltage drop across the gate also decreases as temperature increases.

In order to induce turn-off, the SCR's loop gain must be brought below unity. From Fig. 1C, it appears that shorting the gate to the cathode would bring that about; however, in SCR operation very little current is diverted via such a short. Principle current must fall below a certain level before gain falls below unity and turn-off can commence.

In fabricating SCR's, a shorted-emitter design is generally used in which (schematically) a resistor is added between the SCR's gate and cathode. Because current is diverted from the N-base through the resistor, the gate-trigger (I_g) is increased (I_1), and holding (I_H) currents all increase. The prime reasons for the shunt resistor—without which leakage current could initiate turn-on—is to improve performance at elevated temperatures.

Sensitive-gate SCR's (and thyristors in general) use a high-resistance shunt or none at all. Consequently, their characteristics can be altered dramatically by way of an external resistance, which has only minor effect on most shorted-emitter designs. Junction temperature is the fundamental variable affecting SCR characteristics.

Increased temperatures make SCR's easier to turn on and keep on. Thus, circuit conditions that determine turn-on must be designed to operate at the lowest anticipated junction temperatures. At the same time, circuit conditions that are to turn off the SCR or prevent false triggering must be designed to operate at the maximum anticipated junction temperature.

**Pulling the Trigger**

The turn-on of the SCR has been described as requiring the injection of a current—in the form of gate current (I_g), anode-leakage current, or avalanche breakdown of a blocking junction—to raise the loop gain to unity. The breakover voltage of a SCR can be varied or controlled by the injection of gate current. Figure 2 shows the interaction of gate current and voltage in an SCR. Note that when I_g is zero, the applied voltage must reach the breakover point before switching occurs. (There is an I_g level at which the SCR's, and thyristors in general, behave more like rectifiers.)

Because SCR turn-on can induce high instantaneous power dissipation, die failure may result unless the magnitude and rate of rise in principle current (du/dt) is restricted to tolerable levels. Normally, SCR's are operated at voltages lower than their breakover rating, and turn-on is induced by gate signals of sufficient amplitude to assure complete turn-on independent of the applied voltage.

**Trigger Sensitivity**

Because junction-leakage currents and current gain (see Fig. 1C) increase with temperature, the magnitude of the required gate-trigger current decreases, and the gate—which can be regarded as a diode—displays a lessening voltage drop as temperature increases. (Figure 3 shows gate sensitivity as a function of temperature.) It's important that the gate trigger circuit be designed to deliver sufficient I_g at the lowest temperature anticipated.

Further, it's advisable that maximum gate ratings, and peak- and average-power dissipation ratings be observed. Although turn-on criteria has been discussed in terms of current, it's more basic to consider the SCR as being charge controlled. As the trigger-pulse duration is reduced, its amplitude must be correspondingly increased.

The gate-pulse width required to induce SCR conduction also depends on the time required for the anode current to reach the latching value. It may be necessary to maintain a gate signal throughout the conduction period in applications where the load is inductive, or where the anode current may
Fig. 4—SCR turn-on time has two stages: a delay time (t\(_d\)) and rise time (t\(_r\)). The total gate-controlled turn-on time (t\(_{on}\)) is measured from the 50% point of the gate-trigger to 90% of the principal current value. The rise time (t\(_r\)) is the time required for principal current to achieve 90% of its maximum value.

Fig. 5—Turn-off time is a property associated only with unidirectional thyristor devices. The recovery time of the SCR is comprised of two stages: a reverse recovery time and a gate-recovery (forward-blocking) time.

swing below the holding value within the conduction period.

When triggering an SCR with DC, excess reverse leakage normally occurs if the trigger signal is maintained during the reverse-blocking phase of the anode voltage. That happens because the SCR acts like a remote-base transistor. When high gate-drive currents are used, substantial dissipation may occur in the SCR. Therefore, some means must be provided to remove the gate signal during the reverse-blocking phase.

**Latching and Holding Characteristics**

As mentioned, it's necessary to have sufficient principal current flowing for the SCR to remain on when the trigger is removed. The required principal-current—called latching current, I\(_{L}\)—is primarily affected by temperature in shorted-emitter structures. To induce turn-off, principal current must fall below the latching level. The current at which turn-off occurs, called the holding current (IH), is similarly affected by temperature, and dependent on gate impedance.

Reverse current on the SCR gate markedly increases the latch current and hold levels, while forward biasing its gate may significantly lower those levels compared to those specified, which are normally given with the gate open. Sometimes failure to account for such instances causes latch or hold problems when SCR's are driven by transistors with saturation points of a few tenths of a volt. SCR's of shorted-emitter design are obviously not as sensitive as those that have no built-in shunt.

**Switching Characteristics**

When triggered by a gate signal, SCR turn-on time has two stages: a delay time (t\(_d\)) and rise time (t\(_r\))—the time required for the principal current to rise from 10 to 90% of its maximum value—as shown in Fig. 4. The total gate-controlled turn-on time (t\(_{on}\)) is defined as the time between the 50% point of the leading edge of the gate-trigger voltage and 90% point of the principal current.

The peak off-state voltage (which is related to the magnitude of the gate-trigger current, having a nearly inversely-proportional relationship) increases as the delay-time is decreased. Since high voltage causes increased regenerative gain, t\(_d\) is influenced primarily by off-state voltage.

It's important to note the relationship between the principal voltage and current flow through the SCR during the rise-time interval. During that time, the voltage drop is high and current density, due to rapid-charge, can produce localized hot spots on the die, which can permanently degrade the blocking characteristics. It's important that power dissipation at turn-on be restricted to safe levels.

**Recovery Time**

Turn-off time is a property associated only with SCR's and other unidirectional devices. The recovery time of an SCR is comprised of two stages: a reverse recovery time, t\(_{rr}\), and a gate-recovery time, t\(_{gr}\), (see Fig. 5). When the SCR's forward current is reduced to zero (at the end of forward conduction), the application of reverse voltage (V\(_R\)) between the anode and cathode of the SCR causes reverse current (I\(_R\)) to flow until I\(_R\) is decreased to the leakage level. Reverse recovery time (t\(_{rr}\)) is measured from the point where the principal current changes polarity, to a specified point on the I\(_R\) waveform, as shown in Fig. 5. During that period, the anode and cathode are being swept of their charge so that V\(_R\) may be supported.
Fig. 6—In general, SCR's exhibit longer \( t_q \) with increasing \( I_{TM} \). \( I_{TM} \) and \( \frac{dV}{dt} \) have the strongest effect on \( t_q \). If an SCR must have a short \( t_q \), a low \( I_{TM} \) is present, a large gate-trigger pulse (\( I_{GT} \)) is required to trigger it.

A second recovery period, called the gate-recovery time (\( t_r \)), must elapse for the charge stored in the forward-blocking junction to recombine so that forward voltage can be successfully blocked by the SCR. Gate-recovery time is usually much longer than \( t_q \). The total time from the instant reverse recovery current begins to flow, to the start of the forward blocking voltage, is referred to as circuit-commutated turn-off time or \( t_o \).

Turn-off time depends on a number of circuit conditions, including on-state current prior to turn-off, rate of current change during the forward to reverse-voltage transition, reverse-blocking voltage, rate of change of reapplied forward voltage, gate bias, and junction temperature. Increasing junction temperature and on-state current increases \( t_o \) and has a more significant effect than any of the other factors. Negative gate-bias decreases \( t_o \).

For applications where an SCR is used to control AC power, \( V_R \) is applied during the entire negative half of the sinewave. Turn-off is easily accomplished for most devices at frequencies up to a few kilohertz. In applications where an SCR is used to control the output of a fullwave rectifier, there's no \( V_R \) available for turn-off. Thus turn-off can only be accomplished if the rectifier output drops below the SCR's \( I_H \) value for a sufficiently long period. Turn-off problems may occur at 60-Hz, particularly when controlling an inductive load.

Jumping the Gun

Certain circuit conditions can cause SCR's to turn on without a gate-trigger signal having been applied (i.e., false triggering). For example, a rapid rise in anode voltage (symbolized by \( \frac{dV}{dt} \) and defined as the rate of change in voltage with respect to time), transient-induced anode breakover, and spurious gate signals can all cause false triggering.

When a voltage is suddenly applied to an SCR that's turned off, turn-on might be initiated even though no trigger signal is present at its gate. If, however, the SCR is controlling AC, false triggering due to a transient-imposed voltage is limited to no more than a half cycle of the applied voltage. Turn-off occurs at the zero-crossing point.

On the other hand, if the principal is DC, the transient may cause the device to switch on—due to a rapid rise in anode voltage, resulting from the internal capacitance of the SCR—in which case, turn-off can henceforth only be accomplished through circuit interruption.

A voltage applied across the anode/cathode terminals of an SCR causes capacitance-charging current—a function of the applied off-state voltage (\( I = C \frac{dV}{dt} \))—to flow through the device. If the rate of the voltage rise exceeds a critical value, capacitance-charging current exceeds gate trigger current, causing the SCR to turn on. Due to increased sensitivity, operating the device at high temperatures reduces the SCR's ability to support a steep rising voltage.

Solutions and Cautions

In sensitive-gate devices and shorted-emitter designs (to some extent), \( \frac{dV}{dt} \) ability can be markedly limited by inserting a resistance from cathode to gate. Reverse-bias voltage has even more influence in an SCR. More commonly, a snubber network is used to keep \( \frac{dV}{dt} \) within the limits of the SCR when the gate is open. Transient voltages, occurring as a result of disturbances in the AC line caused by various sources, such as energizing transformers, load switching, solenoid closure, and the like may generate voltages that exceed the SCR breakover rating.

Generally speaking, SCR's switch on (transferring energy to the load) whenever breakover is exceeded. Thus, care should be taken to ensure that breakover—which could result in blocking-characteristics degradation—does not occur. For applications where low energy, long-duration transients may occur, it's advisable that an SCR with a higher voltage rating than the highest anticipated transient be used.

It's a good practice to provide some form of transient suppression; an MOV or Zener diode, for example. An RC snubber is effective in reducing the effects of high-energy, short-duration transients frequently encountered. The snubber is usually used to prevent the static \( \frac{dV}{dt} \) limit from being exceeded. And may be effective in limiting the amplitude of the transient, as well.

Under no circumstance should the device \( \frac{dV}{dt} \) limit exceed the minimum rate of rise off-state voltage applied immediately to its anode/cathode terminals after the principal current of the opposing polarity has decreased to zero.

In noisy electrical environments, enough energy could cause a trigger current to be induced in the gate wiring via stray capacitance or electromagnetic induction. Thus, it is necessary to keep the gate lead short and provide a direct common return to the SCR's cathode. In extreme cases, shielded cable may be required. Another commonly used method is to place a capacitor (on the order of 0.01 to 0.1 \( \mu \)F) across the gate and the cathode, increasing the SCR's ability to withstand high voltage changes by forming a capacitive divider with the anode-to-gate capacitance. The gate capacitor also reduces the rate of application of gate-trigger current, which may cause \( \frac{dV}{dt} \) failures if a high inrush load is present.

To ensure long life and proper operation, it's important that the operating parameters be kept within the SCR's ratings—the most important being temperature and voltage, which are loosely related. The voltage rating is valid only up to the maximum operating temperature of that particular device type. The temperature rating may be designed to ensure suitable voltage ratings, switching speeds, and \( \frac{dV}{dt} \). (Various manufacturers may choose different criteria to establish their ratings.) Current ratings are chosen so that at a practical case temperature, the power dissipation won't cause the junction-temperature rating to be exceeded.
SCR Turn-off

SCR's, in addition to traditional power-control applications, are used in areas where turn-off characteristics are vital. For example, reliable high-frequency (greater than 20 kHz) inverters and converters require a precisely controlled $t_q$. But, it's difficult to tailor the $t_q$ of a particular SCR for a given set of circuit conditions.

We've already established that SCR's turn on and remain on even after the gate trigger is removed, as long as the loop-gain criteria is satisfied—i.e., as long as the sum of the common-base current gains (a) of the equivalent PNP and NPN transistors exceed unity. To turn off the SCR, the loop gain must be brought below unity; e.g., the on-state principal anode current ($I_a$) must be reduced below $I_H$. For AC-line applications, that occurrence is automatic during the negative transition of the waveform.

However, in DC applications (inverters, for example), the anode current must be interrupted or diverted. Once anode current in the SCR ceases, a period of time (which is dependent on temperature, forward current, and other parameters) must elapse before the SCR can again block forward voltage. The SCR $t_q$ can be understood by considering the three junctions that make up the SCR. When the SCR is conducting, each of the three junctions is forward biased and the N and P regions (base regions) on either side are heavily saturated with holes and electrons.

Turning off an SCR

To affect turn-off in a minimum amount of time, it's necessary to apply a negative (reverse) voltage to the anode, causing the holes and electrons near the two end junctions to diffuse, and $I_a$ to flow through the SCR. When the holes and electrons near junctions have been removed, $I_a$ ceases, causing the SCR to assume a blocking state.

However, that doesn't complete recovery, inasmuch as a high concentration of holes and electrons still exist near the center junction. That concentration is decreased by recombination and is largely independent of the external circuit. When the hole and electron concentration near the center junction reaches some low value, the junction assumes its blocking condition and a forward voltage can, thereafter, be applied without the SCR switching back to the original conduction state.

When measuring SCR $t_q$, it's necessary to establish a forward current for a period long enough to ensure carrier equilibrium, since $I_{TM}$ has a strong effect on $t_q$. The current is reversed at a specified $dv/dt$, usually by shunting the SCR anode to some negative voltage through an inductor, causing the SCR to show a reverse-recovery current. Further waiting time must elapse, as the charges recombine before forward voltage can be applied.

The forward voltage is ramped up at a specified $dv/dt$, with $dv/dt$ delay time being reduced until a point is reached when the SCR can no longer block the forward-voltage ramp, turning it on as the ramp-voltage collapses. The time between that point and the point where forward-current passes through zero and starts to go negative (reverse-recovery phase) is the SCR's $t_q$.

To see how the various circuit parameters—forward current magnitude ($I_{TM}$), forward current duration, rate of change of turn-off current ($dv/dt$), reverse-current magnitude ($I_{RM}$), reverse voltage magnitude ($V_{RM}$), rate of reapplied forward voltage ($dv/dt$), magnitude limit of reapplied voltage, gate-cathode resistance, and gate-drive magnitude ($I_{GT}$)—can affect $t_q$, one condition at a time is varied, while the others are held constant. (Some parameters affect $t_q$ more than others.)

Forward/Reverse-Current Magnitude

Forward-current magnitude and $dv/dt$ have the strongest effect on $t_q$. Varying $I_{TM}$ can change $t_q$ by about 30.0%, and the change is attributed to varying-current densities (stored charge) present in the SCR as $I_{TM}$ magnitude is changed. If an SCR must have a short $t_q$ when a low $I_{TM}$ is present, a large gate trigger pulse ($I_{GT}$) would be advantageous. That turns on a large portion of the SCR to minimize high-current densities that exist only if a small portion of the SCR were turned on (by a weak gate pulse) and the low $I_{TM}$ did not fully extend into the turn-on region. In general, SCR's exhibit longer $t_q$ with increasing $I_{TM}$, as shown in the curves of Fig. 6.

---

Fig. 7—Increasing temperature increases $t_q$ for SCRs.

Fig. 8—Varying the turn-off rate of change of anode current also has some effect on $t_q$. 
Increasing temperature also increases $t_q$ (see Fig. 7). Varying the turn-off rate of change of anode current ($di/dt$) does have some effect on the $t_q$ of SCR’s, as shown in Fig. 8. Reverse current ($I_q$) is actually due to the stored charge exiting the SCR junctions when a negative voltage is applied to its anode. Reverse-current magnitude ($I_{RM}$) is closely related to $di/dt$. Increasing $di/dt$ causes an increase of $I_{RM}$, while a decreasing $di/dt$ causes a lower $I_{RM}$. By using different series inductors and changing the negative anode turn-off voltage, it’s possible to keep $di/dt$ constant, while changing $I_{RM}$. $I_{RM}$ has little or no effect on $t_q$ when it’s the only variable changed.

Reverse anode voltage has a strong effect on the $I_{RM}$ magnitude and $di/dt$, but when $V_{RM}$ alone is varied, with $I_{RM}$ and $di/dt$ held constant, there is little or no change in $t_q$ time. $V_{RM}$ must always be within the $V_R$ rating of the device.

Reapplied $dv/dt$

Varying the reapplied $dv/dt$ can vary the $t_q$ of a given SCR by more than 10%. The effect of $dv/dt$ on $t_q$ is due to the anode-gate capacitance. $dv/dt$ applied at the SCR anode injects current into the gate through the anode-gate capacitance ($C_{GT} = C dv/dt$); therefore, as the $dv/dt$ increased, $I_G$ also increases and can induce turn-on.

To complicate matters, injected current adds to the leakage in the junctions just after turn-off. The charge remaining in the center junction is the main reason for long $t_q$ times and, for the most part, is removed by the recombination process. If the reapplied $dv/dt$ is high, more charge is injected into the center junction and prevents its returning to the blocking state, as if it had a slow $dv/dt$. The higher $dv/dt$, the longer the $t_q$ time.

Changing the magnitude of the reapplied $dv/dt$ voltage has little or no effect on an SCR’s $t_q$ time when the maximum applied voltage is well below the device’s breakdown voltage. The $t_q$ time increases if the SCR is being used near its breakdown voltage. Since the leakage current near breakdown is higher than at lower-voltage levels, leakage lengthens the time it takes for the charge to be swept out of the center junction, thus lengthening the usual recovery time of the SCR.

Changing the gate-drive magnitude has little effect on an SCR’s $t_q$ time unless it’s grossly over or underdriven. When overdriven, there’s an unnecessarily large charge in the SCR’s junction. When underdriven, it’s feasible that only a small portion of the gate region turns on. If anode current is not sufficiently large to spread the small turned-on region, there’s a high current and charge density in the region that consequently lengthens $t_q$ time.

Forward Current Duration

Forward-current duration has no measurable effect on $t_q$ time when varied from 100 to 300 microseconds. Longer $t_{TM}$ durations heat up the SCR, causing very short $t_{TM}$ durations, which affects $t_q$ times in view of the lack of time for the SCR to reach equilibrium.

As in transistor operation, reverse biasing the gate of the SCR decreases $t_q$, due to the rapid depletion of the stored charge. The reduction in $t_q$ for standard SCR’s is notable, nearing perhaps 50% in some cases. For fast SCR’s only nominal improvement might result. Due to the internal, monolithic resistor of most SCR’s, the actual reverse-bias voltage between the gate-cathode is less than the reverse bias supply.

MOS SCR’s

Developed from the vertical structured MOS, MOS-SCR operation is like that of its predecessor, while combining the benefits of the high input impedance and fast turn-on of a power MOSFET, and the regenerative, latching action of a thyristor into a single package. Figure 9 shows the proposed symbol for the MOS-SCR, as well as the evolution of MOSFET to MOS-SCR.

Referring to Fig. 1D. Turn-on commences when a positive gate-cathode (source) voltage $V_{GS}$ is applied to the FET. The resulting drain current is the base current of Q1. Like the SCR, the collector current of Q1 supplies base current to Q2, and if there is adequate loop gain, regeneration occurs, latching the device into conduction. The gate signal can then be removed, without affect, until the principal current falls below the hold-current level.

Due to its MOSFET input, MOS-SCR’s exhibit the high impedance and fast turn-on characteristics of a MOSFET. Its high input impedance allows the MOS-SCR to be readily driven by standard CMOS logic.

Problems with MOS-SCRs

Despite the MOSFET’s high static input impedance, providing fast switching (in the tens of nanosecond range) means that the gate input capacitance must be quickly charged (or discharged during turn-off). Thus the gate-source and gate-drain capacitance (and its Miller-capacitance effect) may require that peak charging currents of .5 to several amperes be fed through the gate.

Rather than have a larger charging current capability, a lower one is more desirable, and lower current (larger gate (Continued on page 102)
TRIANGLES AND VECTORS

By Louis E. Frenzel

Current and voltage in a circuit with reactance do not behave the same way as they do in a purely resistive circuit. Here's a simple method to figure out just what's going on in those mysterious circuits.

Trying to visualize what's going on inside an electronic circuit is usually difficult. We can't see electrons flowing, we can't see voltage, and the concepts of frequency and phase shift are vague at best. To help get a handle on circuit operation, we often resort to the use of graphic techniques which try to visualize circuit action. Triangles and vectors are examples of visual aids that add some dimension to circuit analysis. Anything that helps you picture what's going on inside a circuit will improve your understanding of circuit operation.

Of course, you know what a triangle is. It's a three-sided geometric figure you learned about in high school. Of particular interest in electronics is the right triangle. That's a triangle with one right angle (90°). A vector is just a straight line or arrow that represents some electronic quantity such as voltage, current, resistance, impedance or whatever.

In this installment, you will learn to use triangles and vectors to represent and compute quantities in electronic circuits.

Triangle Basics

A triangle is a geometric figure with three sides and three angles. An angle is formed where two lines meet at a point. Each angle is formed by the intersection of two sides of the triangle. Figure 1 shows a triangle. Note that the angles are designated by capital letters while the sides are indicated by lower case letters.

Angles, as you know, are measured in degrees. One degree is a unit of measure equal to the angle between two lines drawn from the center of a circle that define 1/360th of the circumference of the circle. See Fig. 2. If we divide the circle up into 360 “pie slices,” the angle between any two lines is one degree.

There are three basic categories of angles: acute, obtuse, and right angles. These are illustrated in Fig. 3. An obtuse angle is one between 90° and 180°. An acute angle is one that is less than 90°. A right angle is exactly 90°. One line is perpendicular to the other.

Several different types of triangles are shown in Fig. 4. A right angle is illustrated in Fig. 4A. The triangle in Fig. 4B is an isosceles triangle which has two equal sides and two equal angles. The equilateral triangle in Fig. 4C has all three sides and all three angles equal.

Knowing the Angles

The sum of all three angles in any triangle is 180°. For example, the sum of angles A, B and C in Fig. 5 is:

\[
A + B + C = 180° \\
105 + 30 + 45 = 180°
\]

![Fig. 1](image1) This simple triangle can help you solve some not-so-simple circuit equations. Note that the sides are given the same letter as their opposing angle.

![Fig. 2](image2) This is how a degree is defined. It is 1/360th of a circle's circumference. The use of the 360° circles actually dates back to before the Summerian astrologers.
If you know two angles, you can always find the third. If angles A and B are 76° and 22° respectively, then their sum is $76 + 22 = 98°$. Angle C can be found by subtracting the sum of A and B from 180°.

$$C = 180 - (A + B)$$
$$C = 180 - 98 = 82°$$

The right triangle is a special case. Refer back to Fig. 4A. Since one angle, angle C, is 90°, the sum of the other two must be 90°.

$$A + B = 90°$$

You can find A or B by simply rearranging the formula:

$$A = 90° - B$$
$$B = 90° - A$$

In electronics, it is the right triangle that will be used to help visualize and explain circuits. That will become clear later once we get to working some sample problems.

**Side Wise**

The lengths of the sides of a triangle are interrelated. In a right triangle, if two sides are known, the third side can be computed. Refer to Fig. 6. Assume the sides of a right angle are labeled a, b, and c. Side c is opposite to the right angle and is called the hypotenuse. The expression below shows the relationship between the sides.

$$c^2 = a^2 + b^2$$

The actual length of the hypotenuse is:

$$c = \sqrt{a^2 + b^2}$$

Of course, you can use algebra to rearrange the formulas to compute either of the other two sides.

$$a^2 = c^2 - b^2$$
$$a = \sqrt{c^2 - b^2}$$

and:

$$b^2 = c^2 - a^2$$
$$b = \sqrt{c^2 - a^2}$$
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That relationship is known as the Pythagorean rule. It is named after the Greek philosopher, Pythagoras, who discovered it in 6 B.C. and used the famous Pythagorean Theorem to prove it.

As you will soon see, That simple formula is extremely powerful and you will use it frequently in electronics.

Let's take a simple example. If side a of a triangle is 40 inches and side b is 30 inches, the hypotenuse or side c is:

\[
\begin{align*}
c &= \sqrt{a^2 + b^2} \\
c &= \sqrt{40^2 + 30^2} \\
c &= \sqrt{1600 + 900} \\
c &= \sqrt{2500} = 50
\end{align*}
\]

**Practice Problems**

To see how much you remember, try these practice problems.

1. A right triangle has angle B that is 39°. What is angle A?
2. A right triangle has a hypotenuse that is 15 centimeters (cm) long. Side a is 9 cm long. How long is side b?

See answers at the end of this article.

**Vector Basics**

A vector is just a straight line or arrow that indicates the size or amplitude of some physical or electrical quantity. The length is proportional to the magnitude of the quantity while the direction in which it points indicates movement or its angular relationship to another vector. For example, a vector one inch long might represent a voltage of 2 volts. A vector 3-in. long represent 2 x 3 = 6 volts.

Many times you will need to add or subtract vectors. If the vectors are pointing in the same direction, all you do is add their lengths. Suppose a .1-in. vector represents 10 mA. See Fig. 7A. A 3-in. vector, therefore, represents 30 mA as Fig. 7B shows. If these two vectors are pointing in the same direction, you would add them as shown in Fig. 7C. The total is 1 + 3 = 4.0-in., representing 40 mA.

**Fig. 7**—Any vector (A) in any units can be added to any other vector (B) provided that the point in the same direction and have the same scale.

If the two vectors above are pointing in exactly the opposite directions as shown in Fig. 8, you subtract the two values to get the algebraic sum. The result is 3–1 = 2. The sum vector is pointing in the direction of the largest vector.

If the vectors are not pointing in the same or opposite direction, they cannot be added directly. Special techniques must be used. Take a look at Fig. 9. Vector A represents a pulling force of 30 pounds on the object in the direction shown. Vector B is a pulling force of 40 pounds in another direction. The big question is: what is the total net pulling force, and in what direction will the object move?

The solution to that problem is done graphically. Assume 0.5-in. is equivalent to 10 pounds. Vector A then is 1.5-in. long and vector B is 2-in. long. We first construct a line at the end of vector A equal in length and parallel to vector B. Then, we construct another line at the end of vector B equal in length and parallel to vector A. The result is a four sided figure called a parallelogram. The sum of the two vectors and the direction to the composite pulling force is the length and direction of the diagonal drawn from the object to the far corner. Measuring its length, we see that it is approximately 3.25-in. which is equivalent to 65 pounds of force.

**Fig. 8**—If vectors point in opposite directions they have to be subtracted to obtain the proper value of their result.

**Fig. 9**—We can figure out the sum of two vectors not pointing in the same direction by using the parallelogram method. Doing this gives you a new vector at a new angle.

**More Vector Addition**

Another example of vector addition is shown in Fig. 10. Here the two vectors a and b are perpendicular to one another thus forming a right (90°) angle. We can add these together the same way. Draw in the parallel lines as shown. That creates a parallelogram that, in this case, is a rectangle. The diagonal line is the sum of the two vectors.

**Fig. 10**—Vectors at right angles are easy to sum with the Pythagorean relation, that is because they can be laid out to look like a right triangle.

But wait. When you drew in the diagonal, you constructed a right triangle. Now, instead of finding the sum by measuring the length of the diagonal, you can use the Pythagorean Theorem to compute it. Using the values in Fig. 10, the diagonal is:
c = \sqrt{a^2 + b^2}

If \( a = 2 \) and \( b = 4 \), then \( c \) is:

\[
\begin{align*}
\text{c} &= \sqrt{2^2 + 4^2} \\
\text{c} &= \sqrt{4 + 16} \\
\text{c} &= \sqrt{20} = 4.472
\end{align*}
\]

In electronics, many of the vectors will be at right angles to one another making vector addition very easy. So now, let’s get started and apply all of this to some real electronic circuits.

**Practice Problems**

Verify what you have learned by solving these problems.

3. A 2 centimeter (cm) vector represents 10 volts. A 12 cm vector represents how many volts?
4. Two force vectors \( a \) and \( b \) are pulling perpendicular to one another. Vector \( a = 110 \) lb and \( b = 80 \) lb. What is the total pulling force?

Check your answers at the end of this article.

**Analyzing an RL Circuit**

There are many electronic circuits made up of passive components like resistors (R), capacitors (C), and inductors (L). Examples are filters, phase shifters, equalizers, impedance matching networks, tuned circuits etc. We can use the triangle and vector math to simplify the analysis and design of those circuits.

For example, take the simple series RL circuit shown in Fig. 11. An AC voltage causes current to flow. In a series circuit, the same current flows in both the resistor and inductor because there is only one path for the current. We know from our general knowledge of electronics that current in an inductor lags the applied voltage by 90°. That is caused by the inductor’s inherent characteristic that it opposes changes in current with a self induced voltage. Figure 12A shows the AC current-voltage relationship in an inductor. If there is resistance in series with the inductor, the phase shift will be some value between 0 and 90°.

The current-voltage relationship for a resistor is shown in Fig. 12B. The current and voltage are in phase with one another.

Now, when current flows through a component, it causes a voltage drop to occur across each component. See Fig. 11. We can compute these voltages with Ohm’s law if we know the current value and the values of resistance and inductive reactance. The voltage across the resistor is \( V_R \) and the voltage across the inductor is \( V_L \).

![Fig. 12](https://via.placeholder.com/150)

**Reactance**

Reactance is the opposition to alternating current flow offered by reactive components such as inductors (coils) or capacitors. Like resistance, reactance is expressed or measured in ohms.

Inductive reactance \( (X_L) \) depends on the value of inductance \( (L) \) and the frequency \( (f) \) of the applied AC in Hertz (cycles per second). It is computed with the simple formula:

\[
X_L = \frac{2\pi f L}{\text{Hertz}}
\]

Capacitive reactance \( (X_C) \) depends upon the frequency \( (f) \) and the capacitance in farads \( (C) \):

\[
X_C = \frac{1}{2\pi f C}
\]

For given values of \( L \) and \( C \), \( X_L \) is directly proportional to frequency while \( X_C \) is inversely proportional to frequency. If \( f \) increases, \( X_L \) increases but \( X_C \) decreases.

The total amount of reactance in a circuit determines the current flow.
According to Kirchhoff's voltage law, the sum of the voltage drops across series components is equal to the applied voltage. But there is a problem here. We can't just add the two voltages directly together because they are out of phase with one another. Since the current in the inductor lags the voltage across it by 90°, we can say that the inductor voltage leads the current by 90°. It means the same thing. The voltage across the resistor is in phase with the current. Therefore, the voltage across the inductor leads the voltage across the resistor by 90°. Let's draw a picture of that with vectors to see it better.

Refer to Fig. 13. First we draw a vector pointing to the right representing the current I. Current is the reference in a series circuit since it is common to all components. Of course, the length of the vector is proportional to the actual current value.

Next, we draw a vector representing the resistor voltage (V_R). Since it is in phase with the current, that vector overlaps or coincides with the current vector.

![Fig. 13—We can use vectors to show the relationship between the voltages of all components in an AC circuit.](image)

Now we construct the inductor voltage (V_L) vector. The voltage leads the current by 90°. When drawing vector diagrams in electronics, leading is counter-clockwise from the reference (current) vector while lagging is clockwise from the reference. So, the V_L vector leads the current vector by 90° as Fig. 13 illustrates.

At last we can add the voltage vectors. And we do that using the parallelogram method described earlier. The result is a rectangle where the diagonal or hypotenuse represents the applied or source voltage (V_S) (see Fig. 13). And since a right angle is involved, we can use Pythagorean Theorem to compute the exact value of the applied or source voltage (V_S).

\[
V_S = \sqrt{(V_R)^2 + (V_L)^2}
\]

\[
V_S = \sqrt{3^2 + 1.6^2}
\]

\[
V_S = \sqrt{9 + 2.56}
\]

\[
V_S = \sqrt{11.56} = 3.4 \text{ volt}
\]

The angle formed between the current and source voltage vectors is the actual phase shift in the circuit. As you can see, it is less than 90°. In the next installment, we will show you how to compute it.

**Resistance, Reactance, and Impedance**

The vector diagram in Fig. 13 shows voltages. But we can convert it to represent resistance, reactance and impedance. Here's how.

We said that the current was common to all components in the series RL circuit. If we divide V_L by the current I, we get X_L as Ohm's law tells us. Dividing V_R by I, we get R. Dividing V_S by I, we get Z, the total circuit impedance. The new vector diagram looks like that in Fig. 14. It looks just like our voltage vector diagram only we are dealing with different quantities.

The current in any circuit is determined by the amount of source voltage (V_S) applied and the total opposition to current flow. That's Ohm's law, right? The opposition comes from the resistance (R) and the inductive reactance (X_L). The total opposition is the sum of the resistance and reactance and is called impedance (Z). But as you can see by Fig. 14, the R and X_L vectors are 90° out of phase so we cannot add them directly. Again we use vector addition. The hypotenuse of the right triangle is the total circuit impedance. It can be calculated using Pythagorean Theorem.

\[
Z = \sqrt{R^2 + X_L^2}
\]

Going back to the values given earlier (R = 1500 ohms and X_L = 800 ohms), we can use the formula to calculate the total impedance.

\[
Z = \sqrt{1500^2 + 800^2}
\]

\[
Z = \sqrt{2.25 \times 10^6 + .64 \times 10^6}
\]

\[
Z = \sqrt{2.89 \times 10^6}
\]

\[
Z = 1.7 \times 10^3 = 1700 \text{ ohms}
\]

**Fig. 14—The total impedance of the circuit cannot be computed by adding the impedances of the components together. Since the impedances are 90° out of phase, the must be summed vectorally which is made easier by drawing the vectors.**

Knowing the source voltage which we computed earlier (3.4 volts) and the total impedance, we can use Ohm's law to calculate the circuit current.

\[
l = \frac{V_S}{Z}
\]

\[
l = \frac{3.4}{1700}
\]

\[
l = 2 \times 10^{-3} = 2 \text{ mA}
\]

Of course, that jibes with the original value given earlier. As you can see, the vectors and triangles, if accurately constructed, allow you to see the relationship of the various circuit parameters. It makes an otherwise nebulous thing more concrete.

**Analyzing an RC Circuit**

You can use exactly the same procedures to analyze a resistor-capacitor (RC) circuit of the type shown in Fig. 15. Here you have to remember because of the charging and
discharging action of the capacitor, the current flowing leads the voltage across it by 90°. That’s the same as saying that the voltage across the capacitor lags the current.

Assume that \( R = 375 \) ohms and \( X_C = 500 \) ohms. The current is 8mA. Using Ohm’s law, you can find the voltage drops.

\[
V_R = IR = (8 \times 10^{-3})(375) = 3 \text{ V} \\
V_C = IX_C = (8 \times 10^{-3})(500) = 4 \text{ V}
\]

Now let’s draw the voltage vector diagram. Refer to Fig. 16. The current, being common, is represented by the vector pointing to the right. The voltage across the resistor is in phase with the current so the \( V_R \) vector overlaps the current vector.

![Voltage Vector Diagram](image)

**Fig. 16**—Since the voltage of the capacitor lags behind the supply (and the resistor), it is drawn pointing downward. Just the opposite of the inductor’s vector.

The capacitor voltage \( V_C \) lags the current by 90°. So we move in a clockwise direction from the current vector 90° and draw \( V_C \) pointing down.

Kirchhoff’s voltage law says that the sum of the component voltage drops equals the source voltage (\( V_S \)). Ohm’s law tells us the component voltage drops are:

\[
V_R = IR \\
V_C = IX_C
\]

where \( X_C \) is the capacitive reactance.

These voltages have to be added vectorally. Therefore, we use the parallelogram method. The length of the diagonal or hypotenuse is the source voltage. Of course, since we constructed a triangle with the sides being \( V_R \) and \( V_C \), we can use the Pythagorean relation to compute \( V_S \).

\[
V_S = \sqrt{V_R^2 + V_C^2} \\
V_S = \sqrt{3^2 + 4^2} \\
V_S = \sqrt{9 + 16} = \sqrt{25} = 5 \text{ V.}
\]

Just keep in mind that you can always rearrange Pythagorean relation to calculate \( V_R \) or \( V_C \) if you ever need to:

\[
V_R = \sqrt{V_S^2 - V_C^2} \\
V_C = \sqrt{V_S^2 - V_R^2}
\]

Now, we can also construct an impedance triangle as we did with the RL circuit. We already know \( R \) and \( X_C \). To get the impedance, we divide the source voltage by the current.

\[
Z = V_S/I = 5/8 \times 10^{-3} = 625 \text{ ohms}
\]

The resulting triangle is illustrated in Fig. 17. And, of course, you can also use Pythagorean theorem to calculate the impedance.

**Practice Problems**

Now, check what you’ve learned with these exercises.

1. In a series vibration, the resistor voltage is 20 volts, the applied source voltage is 35 volts. What is the coil voltage?

6. What is the total current in a series RC circuit with a resistance of 20 ohms and a capacitive reactance of 30 ohms? The applied voltage is 7.2 volts.

Look for the answers at the end of the article.

**Fig. 17**—The impedance of a circuit containing a capacitive load must be computed using a vector diagram (or at least the math for one). Note that the magnitude is still a positive number. That is because the components are squared for computation removing negative signs.

**Fig. 18**—An LCR circuit does not really make the situation more difficult to analyze. The components for the inductor and capacitor are merely subtracted from one another.

**Analyzing LCR Circuits**

An LCR circuit contains inductance, capacitance and resistance. An example circuit is shown in Fig. 18. We can draw the vector diagram of the circuit voltages as before. See Fig. 19. The inductor voltage leads the current by 90° and the capacitor voltage lags the current by 90° as shown.

Note one key fact. The inductor and capacitor voltages are 180° out of phase with one another. That is illustrated in the vector diagram of Fig. 19 as vectors pointing in opposite directions. Vectors 180° out of phase with one another are combined by adding them algebraically. That is the same as
Fig. 19—The component resulting from the difference between the inductor and capacitor vectors is used to find the voltage source magnitude.

Fig. 20—Calculating the impedance for the LCR circuit is done more easily by subtracting the components of the inductor and capacitor from one another.

Fig. 21—Even though the resultant vector points downward its magnitude is a positive value.

saying we subtract them. One voltage cancels the other.
Assume \( V_L = 15 \) volts and \( V_C = 8 \) volts. The difference \( V_{LC} \) is:
\[
V_{LC} = V_L - V_C
\]
\( V_{LC} = 15 - 8 = 7 \) volts

\( V_C \) cancelled 8 volts or \( V_L \) leaving a total voltage of 7 volts. Since \( V_L \) is larger than \( V_C, V_{LC} \) points in the same direction as \( V_L \). See Fig. 19. Our LCR circuit acts like it is just an inductor and a resistor in series.

Since \( V_L \) is greater than \( V_C \), the effect of \( V_C \) is totally cancelled. The equivalent circuit is just an RL circuit like that shown back in Fig. 11.

If we convert the voltage vectors to impedances by dividing by the current, we get the result in Fig. 20. Assume the current is .4A.

\[
X_L = V_L/I = 15/0.4 = 37.5 \text{ ohms}
\]
\[
X_C = V_C/I = 8/0.4 = 20 \text{ ohms}
\]
\[
R = V_r/I = 12/0.4 = 30 \text{ ohms}
\]

Note that the inductive reactance \( (X_L) \) is greater than the capacitive reactance \( (X_C) \). Since they are pointing in opposite directions, they cancel one another. \( X_L \) is greater than \( X_C \) so \( X_C \) is totally cancelled. The total circuit reactance is:
\[
X_{LC} = X_L - X_C
\]
\[
X_{LC} = 37.5 - 20 = 17.5 \text{ ohms}
\]

Since \( X_L \) is greater than \( X_C \), it is as if the circuit only contains an inductive reactance equal to 17.5 ohms. That’s why the LCR circuit acts as if it is on an RL circuit. We say the circuit is inductive.

You can now calculate the total circuit impedance as you did before.
\[
Z = \sqrt{R^2 + X_{LC}^2}
\]
\[
Z = \sqrt{30^2 + 17.5^2}
\]
\[
Z = \sqrt{900 + 306.25}
\]
\[
Z = \sqrt{1206.25} = 34.73 \text{ ohms}
\]

A more general version of the impedance formula takes into consideration both \( X_L \) and \( X_C \).
\[
Z = \sqrt{R^2 + (X_L - X_C)^2}
\]

You can also write it by reversing \( X_L \) and \( X_C \), but the outcome is the same.
\[
Z = \sqrt{R^2 + (X_C - X_L)^2}
\]

We could also have a condition where \( V_C \) is greater than \( V_L \) or \( X_C \) is greater than \( X_L \). All of \( X_L \) is cancelled leaving a smaller net reactance which is capacitive. Therefore, the circuit acts like a resistor and capacitor in series just like the one shown previously in Fig. 15.

You use the same formulas as before to compute values in the circuit. For example, the source voltage is:
\[
V_S = \sqrt{V_r^2 + (V_L - V_C)^2}
\]

The total impedance is:
\[
Z = \sqrt{R^2 + (X_L - X_C)^2}
\]

Since the effect of inductance is exactly opposite that of capacitance, the two cancel one another, the large reactance dominating. If \( X_L \) is larger, we say the circuit is inductive. If \( X_C \) is greater, we say the circuit is capacitive. It’s that simple.

(Continued on page 102)
Whenever I visit a hamfest flea market, I almost always come away with some ideas for this column. There’s usually a fair amount of vintage gear and literature to look at. And sometimes—if the price is right and the piece sufficiently interesting—I’ve even been known to make a purchase or two.

A few weeks ago, I came home with an Instructograph code-practice machine (mid-1940’s model). For those of you who’ve never met one, an Instructograph generates Morse code characters by means of a perforated paper tape. The spoolled tape is pulled along between a pair of contact points by a motor drive. As they’re opened and closed by the perforations, the points key an audio oscillator, forming the Morse code characters.

The Instructograph was supplied with a collection of ten different tapes, so that the prospective ham or commercial radio operator could practice receiving a variety of messages at different speeds. For sending practice, a couple of binding posts were provided for a manual code key. And my Instructograph happened to come with a key that was as interesting to me as the machine itself—a World War II surplus J38.

In a B17 Over Berlin

As a teenager in the 1940’s, I could never hope to afford an Instructograph—but I certainly had a J38. Like most other military surplus gear dumped on the civilian market at the end of World War II, it was dirt cheap. A couple of bucks would get you one of those beautifully-crafted brass, nickel, and Bakelite specimens.

Connect your J38 to a 69-cent high-frequency buzzer (audio oscillators were also a little expensive for the typical 1940’s teen budget) and a dry cell, and you had put together a nifty code-practice set for less than five dollars. If you had a buddy with similar interests, you could trade off sending and receiving until you had both mastered the code. At the same time, with that army key in your hand, you could easily imagine that you were a radio operator on a B17 over Berlin or at a front-line command post.

The fact that a J38 was included with the hamfest Instructograph certainly clinched the deal for me. My own had disappeared years ago, and I didn’t think I’d ever own one again. J38’s are hard to find today.

In fact, I looked through about a dozen 1950’s vintage QST magazines in the vain hope of finding a nostalgic J38 ad for this column. The ads in the ham radio publications were certainly loaded with surplus material—including such classics as the T17B mike and the HS33 headset (both were companions of the J38 in the aircraft and command posts of World War II). But apparently, most of the loose J38’s had been snatched up by the early 1950’s.

The ARRL’s 25-Cent Masterpiece

The J38 acquisition really brought back memories about what it was like to prepare for the ham radio-code exams back in the 1940’s. One essential piece of literature for the serious prospective ham was the American Radio Relay League’s excellent pamphlet Learning the Radio-telegraph Code.

The 1951 edition now in my library carries a publisher’s price of 25 cents, and the one that I had in the mid-1940’s may have cost even less. That was certainly no strain on my allowance, and I remember wondering at the time if the League may have been partly subsidizing the price in order to encourage as many prospective hams as possible.

The well-illustrated, slick-paper 1951 edition is loaded with suggestions not only about how to learn the code, but about such things as how to put together an oscillator or buzzer system for code practice; how to adjust and hold the key; how to write when transcribing code; where and when to listen for suitable practice transmission on the shortwave bands; and even how to operate once the license was obtained. There’s a wasted word...
in this tightly-organized, 32-page pamphlet. I consider it a minor masterpiece of the editor’s art.

Dots and Dits

But there was one issue on which I wasn’t sure I agreed with the ARRL. The League felt that people should be strongly discouraged from thinking about Morse code as a pattern of dots and dashes. They felt that such a convention was, perhaps, better suited to represent the clicking noises given off by the old-time, land-line telegraph sounders. The long and short tones used in modern radiotelegraph communication were better represented by the words “da” and “dit,” respectively.

Ads for Candler System Morse training always featured a testimonial from code champion Ted McElroy.

Up to a point, I agreed. When speaking the code for the letter “D,” for example, it certainly made more sense to say “dash dot” than “dash dot dot.” However, when reading Morse characters off the printed page, I was always able to imagine the sounds better—and quicker—when they were more graphically represented as actual dots and dashes. Learning the Radiotelegraph Code’s chart of the Morse characters was made up of printed “dots” and “dahs.” And I remember deciding to make up my own “dot-dash” chart for easier study and memorization. It’s interesting to note that the instruction book that came with the Instructograph machine (which dates from about the same period) didn’t get cute with the printed Morse. It represented the characters with big, bold, graphic dots and dashes.

Quips for Code Practice

The code practice sending and receiving exercises in Learning the Radiotelegraph Code were set up something like typing lessons. They started out by using just a few basic characters and then, little by little, added new ones until the full set of numbers, letters and punctuation marks was involved. Looking over those exercises today, I can see that the writers used quite a bit of ingenuity and humor to come up with meaningful practice phrases stressing all of the newly-added letters.

The Candler system would help you build code speed “without tension or nervousness.” Candler offered both beginning and advanced classes, and every ad featured a testimonial by Candler’s star pupil. Ted McElroy. Ted was the “official champion radio operator, with a speed of 75.2 wpm.”

Like the Instructograph, Candler’s system wasn’t available to most of us kids. You had to be an adult to come up with the cash or sign up for the time payments. But, with our surplus keys and ARRL’s wonderful two-bit booklet, most of us did just fine.

Last Word on the EC-1

Without a doubt, the Echophone EC-1 restoration project has drawn more reader mail than any other subject I’ve covered in this column. (For those of you who intend to begin an antique-radio restoration project of your own, particularly those for whom this will be the first time you’ve pulled apart one of yesterday’s gems, its a good idea to go over the material covered in those columns, which began with the July 1987 issue.) Here are a few of the most recent comments.

Roland Ratcliffe of Lockport, NY does service work for the local gas company and found his in a customer’s basement. It was free “for the mere price of asking.”

John Hall (Newark, DE) purchased his in 1942. He got it on a stopover in Seattle while being shipped to Alaska with the Army Corps of Engineers. Private shortwave sets weren’t allowed on the troop ship, but Jim got around the rules by loaning the EC-1 to the ship’s radio operator for the rest of the voyage. John and the operator were able to enjoy the radio all the way to Kodiak.

Stephen Shaw (Randfontein, South Africa) inherited an EC-1B from his father. As the 1946 ad he sent along shows, Shaw’s set has a slightly different dial face than the EC-1, which we restored in the column. His model has a face that looks more like that of its famous descendant, the Hallicrafters S-38.

But perhaps the last word on the subject comes from reader Jacques Guertin (Quebec, Canada). He sent along an EC-1 ad from the February, 1942 issue of QST Magazine. I enjoyed it so much, that I couldn’t resist running it in this column even though we finished the EC-1 project several issues ago.

Coming Up

Stay tuned for the next issue, when we’ll take a close look at the Instructograph machine that indirectly provided my inspiration for this month’s column. I’ll be looking forward to hearing from you. Write to Marc Ellis, C/O Hands-on Electronics, 500-B Bi-County Blvd., Farmingdale, NY 11735.
Speaker "Z" Tester
The simple bridge circuit shown in Fig. 1 can make the job of determining the impedance of an unmarked speaker a snap. When a speaker of unknown impedance is connected to the Z Tester (at the test terminals), the impedance of the speaker in question is compared to that of the circuit's internal speaker (SPKR1).

An audio signal, fed through T1 (an audio impedance-matching transformer), is input to the two speakers. Then by adjusting R1, the bridge is brought into balance. When the bridge is brought into balance, the impedance of the unknown speaker can be read on the dial of the balance potentiometer, R1.

Almost any audio sinewave oscillator with an output of 3 to 10 volts will do to drive the bridge circuit. And a pair of headphones or an oscilloscope can be used as the balance indicator. The entire circuit can be built in a small plastic or metal cabinet, or merely breadboarded with clips for all of the necessary external connections.

Making an escutcheon for the circuit is as simple as cutting a three-inch circle from a piece of white cardboard, punching a ¼-inch hole in the center and mounting it over the shaft of R1. Turn R1 counter-clockwise until it stops, and place a pointer-type knob on the shaft facing the 7:00 position and lock it in place.

Connect an audio generator to T1, and set the frequency to 1 kHz with a sinewave output of 3- to 10-volts peak-to-peak. Connect a scope or headphones to the bridge’s output, and a resistance decade box to the test terminals. Set the decade box for 16 ohms and turn R1 for a null in the 1-kHz tone. Mark that point with a dot in front of the knob and write in the number 16. Do the same for resistor values of 3, 4, 8, 10, and 100. Balance each resistor value and mark each value on the white escutcheon.

Connect the speaker in question to the test terminals and balance R1 for a null output. Read out the speaker’s impedance on the dial. If you just happen to check out a speaker with a short between turns on the voice coil, the bridge will probably not balance with a null as sharp as with a good speaker.

Auto-Power Adapters
If you are fed up with replacing batteries in electronic gear that you use in your automobile, and would like to tap the power reservoir of the auto’s battery and save those hard-earned bucks, then try one of the Auto-Power Adapters circuits shown in Fig. 2 or Fig. 3.

The circuit in Fig. 2 can be adjusted to supply 0- to 10-volts at currents up to 1 ampere, while the circuit in Fig. 3 covers the same voltage range with a current output of over 3 amperes. The two Auto-Power Adapter circuits operate alike in the following manner.

Looking at Fig. 2, we see that transistor Q1 is connected in an emitter-follower circuit and its base connected to the wiper of R1. The voltage at the emitter of Q1 follows the voltage at its base from zero to slightly less (minus the junction voltage drop) than the 12-volt source. The output of Q1, at its emitter, is fed to the base of Q2, forcing it to follow the voltage at the emitter of Q1.

Since Q2 is to pass a pretty hefty current, Q2’s metal tab should be bolted to a suitable-size heatsink for the load used. The larger the heatsink the better. If the maximum load current requirements are less than 50 milliamperes, leave off Q2 and connect the load to the emitter of Q1.

To really juice up the output current
PARTS LIST FOR THE AUTO-POWER ADAPTERS
R1—1000-ohm potentiometer  
Q1—2N2222 general-purpose NPN silicon transistor  
Q2—TIP31 NPN, tab-mount, silicon power transistor  
Q3—2N3055 NPN silicon, power transistor  
Perfboard or breadboard materials, heatsink material, cabinet, wire, solder, etc.

capacity, go to the circuit in Fig. 3, where another more-powerful transistor is added to the output circuitry. The 2N3055 with a good size heatsink can dish out all the required current.

Neither of the circuits are voltage regulated and the output voltages will vary proportionately with the supplied battery voltage. But for most applications, that will cause no problem. Connect a voltmeter to the output of the circuit and adjust R1 to obtain the desired output voltage level, and then connect the load across the output terminals. Re-check the voltage and fine tune R1 if necessary.

Electronic Circuit Breaker

Anytime you work with a high-current, DC power source, such as an automobile's battery, the chances of sparks flying and metal burning is a real possibility. Fuses make a good, and generally safe, stopgap measure for protecting against disaster caused by overloads and shorts. But if the problem is an intermittent one, and fuses are blowing like leaves falling from autumn trees, then some other plan is going to be called for.

The Electronic Circuit Breaker in Fig. 4 will help keep the bulk fuse boxes intact and allow a broad range of current trip adjustments. With the two power transistors mounted on heatsinks, currents of over 10 amps can be handled. That makes it an ideal circuit to use when making a test run of a newly built project or any equipment that is to be powered from the auto's electrical system.

After the smoke clears, the debugging is completed, and all is functioning OK, determine the current requirement of the project and select a fuse to replace the Electronic Circuit Breaker. You'll see that the Electronic Circuit Breaker is to be used in a high-current application, for a long period of time, then keep an eye on the temperature of the two transistors. If the Electronic Circuit Breaker is to be used in a high-current application, for a long period of time, then keep an eye on the temperature of the two transistors.

Fig. 4—The Electronic Circuit Breaker provides a broad range of current trip points, making it an ideal when making a test run of a newly built project or any equipment that is to be powered from the auto's electrical system.
Ever since I received my first license in 1959, I have loved 40-meter CW, with some flirtation with 20-meters. In those days, almost anyone with a surplus command set receiver (ARC-5 anyone?) capable of homing in on 40-meters could work the CW portions of the band with impunity...and have a heckuva lot of fun doing it.

Besides the skill factor, CW operation is cheap and can be done with relatively low-cost equipment. But today, no one in his or her right mind wants to work 40-meters anytime after sundown...or even within an hour of sundown. If Radio Moscow doesn’t nail you with its multimegawatt signal, then the BBC or the Voice of America will. The 40-meter band is probably the worst example of an overcrowded—nearly useless—band.

So what to do? Of course, with modern HF transceivers, we can buy a 250-Hz or 500-Hz filter to cut through the crud. But that is expensive—at least two filters are needed in practice, and that in itself, pushes the price to over $100. Another disadvantage of the fixed-filter approach is that it is a fixed filter. That is to say, the bandwidth is the same all the time. We cannot custom tailor the bandwidth of the rig to fit the situation.

If we select too wide a bandwidth, then we will hear far too many signals, and weak signals are lost in the roar. Alternatively, if we select too narrow a filter, then we must tune too slowly and pay too close attention to what’s going on. Most operators that I have talked to claim that a 250-Hz CW filter is actually too narrow for practical use.

The solution is to use an adjustable, active bandpass filter, such as the Heathkit model HD-1418, shown with my Kenwood TS-430S in the photo. After using that product on the air, I have dubbed it a “crud buster” because it makes it possible to work CW and SSB on the 40-meter bands—damn the Russians, full speed (well, 15 WPM) ahead!

The HD-1418 uses operational amplifiers and other chips in an active filter arrangement that permits one to either notch out or peak up certain frequencies in the audio spectrum. The HD-1418 is placed in the audio-output line between the receiver AF output and either the loudspeaker or the headphones. There is also a low-level AF output designed to drive tape-recorder inputs and other devices (which allows you to record the received signals).

The front panel controls are as follows:

- **Power, phones, gain, input overload, low/center, high/width, notch/peak** and also a series of mode-selection pushbuttons.

- **Phones**—This connection is a ¼-inch phone jack that accepts an earphone plug.

- **Power**—Turns the filter on and off. The power supply is an external transformer pack.

- **Gain**—While the filter power button is on, the gain control determines the gain of the amplifier section, hence the drive to the loudspeaker or earphones.

- **Input overload**—A light-emitting diode warns when the input signal is too high for the filter circuitry. The audio output of a receiver can be at quite a high level, so some means of warning is needed.

The controls below are both interactive and dual use. Their function depends upon the settings of the mode pushbuttons on the right side of the front panel.

- **Low/Center**—When used as a low control, this knob moves the lower -3 dB end of the audio passband over the range 300 to 2500 Hz. The “low” function is typically used on AM and SSB signals. When selected as a center control, which is used on CW or RTTY signals, the center of the passband is moved up and down over the same 300-2500 Hz range. In this mode, the center control is interactive with the width control.

- **High/Width**—In AM/SSB modes this control sets the high-frequency end of the passband, and also moves the upper -3 dB end of the passband over the 300 to 2500 Hz range. Care must be taken not to make the high frequency lower than the low frequency. In the CW/RTTY modes, this control sets the width of the passband.

- **Notch/Peak**—This control is activated only by pressing the notch or peak buttons, and is disabled if any of the SSB
buttons are pressed. In the notch mode, the center frequency of the passband is severely attenuated; in the peak mode, the center frequency is amplified much more than the other frequencies located in the passband.

**Operation**

The HD-1418 can be tuned as a bandpass filter with bandwidths of 200 to 2500 Hz, mimicking the behavior of some very expensive filters. Figures 1 to 3 show the various modes of operation in graphic form.

Figure 1 shows a bandpass mode. There are two ways to achieve that mode. In the SSB mode, select high and low frequencies for the -3 dB points. Make the selections by listening to the signal on the air, while adjusting for minimal interference and noise.

The other way to achieve the bandpass mode is to use the CW or RTTY modes, and then set the center frequency and bandwidth using the center and width controls.

The other way to achieve the bandpass mode is to use the CW or RTTY modes, and then set the center frequency and bandwidth using the center and width controls. The usual way to use this mode is to position the desired signal in the passband with the center control, and then adjust the width control to attenuate unwanted interfering signals.

Figure 2 shows the "peak" mode. In this case, the center control is used to place the passband on the desired frequency. The desired signal is therefore enhanced, while interfering signals outside of the narrow passband are attenuated.

Figure 3 shows the "notch" mode. Although you might not think of this mode at first glance, it is possibly the most-useful for eliminating interference. The idea is to adjust the width and center controls such that the notch overlays the unwanted signal, and the desired signal falls into the non-attenuated passband. This mode of operation takes a little getting used to, but it is easily mastered and proves exceptionally valuable.

**Construction**

The construction of the Heathkit HD-1418 is easy. As an experienced constructor, I built mine in one afternoon, while visiting my in-laws ...and their kitchen table with my kids crawling all over the place is not the most ideal workshop. I suspect that even novice builders can successfully put the HD-1418 together in no more than two evenings, or a whole Saturday.

The entire project is built onto a single printed-circuit board. All IC devices are socketed, which makes both construction and servicing extremely easy. Two devices that Heath packages with this kit are imaginative...and fortunately come with most Heathkits.

The first is the little red plastic nut starter. It picks up a ¼-inch nut on one end and a ¼-inch on the other...and makes starting tiny nuts on machine screws duck soup. The other is the IC remover. Merely a bent piece of sheet metal, it is invaluable for removing IC’s from their sockets without, in any way, damaging the device.

**Conclusion**

No one should have to put up with the cacophony of 40-meters after sundown...it’s barbaric. But 40-meters is a very useful band that allows communications over both limited groundwave ranges and distant skip ranges. I still find it exciting from my Virginia QTH to simultaneously hear K9/WC80 signals along with W6 and DX.

As to what kind of DX is available, at 0500 EST one morning, I heard what was first taken to be a 52...only to find it was a VK2 in Sydney, Australia. After saying "73" to that one, I worked an EI8 (Irish) for awhile until sun-up shortened the skip. And no, my antenna is not a portable 40-meter beam, it’s a trap vertical.

Well that’s the end of the space set aside for us this month. So until next month, if you have any comments or suggestions, or perhaps just want to chit-chat about the hobby, write to Joe Carr, K4IPV, at PO Box 1099, Falls Church, VA 22041.
FRIEDMAN ON COMPUTERS

Get a universal monitor driver.

By Herb Friedman

The Hercules Graphics Card is full size. The two connectors on the backplane—on the right—are for a 9-pin TTL monitor and a conventional parallel printer.

When IBM introduced their first personal computer, there were actually two separate systems. One was intended primarily for text work such as word processing and calculations. It featured a special kind of monitor, called a TTL Monochrome Monitor, and a matching monitor adapter card that plugged into one of the computer's five "slots." The monitor card also provided an output for a printer. The TTL monitor system provided a very clean and legible character display and, because of its printer output connector, was particularly well-suited for word processing.

The other system was intended to compete with the "home computers" of the day. Its monitor was a special kind of color monitor having separate color and sync inputs, and was known as an RGB monitor—the letters R, G, and B representing the red, green, and blue primary-color signals from the computer. Its monitor adapter, known as a CGA (for Color Graphics Adapter), was intended primarily for displaying graphics; text characters were a secondary consideration.

They were larger than those of the TTL monochrome adapter, but not necessarily as clear. To compete with the "home computers" of its time, in addition to the RGB monitor output, the CGA card also had a composite output that could drive inexpensive composite input color monitors (such as those supplied with Commodore computers), and there was an output for an RF modulator that allowed the computer's signal to be viewed on a conventional TV.

(IBM discontinued the RF modulator very quickly when they realized that the future of personal computing was with the business user and not home and family.)

Cheap Monitoring

Discount computer dealers soon realized that since the CGA's composite output could also drive a composite monochrome monitor—which was less than half the cost of a TTL monitor—they could provide an "IBM system capable of both word processing and graphics" at substantially reduced prices.

The average consumer couldn't tell the

Fig. 1—The only major difference between the Hercules and the CGA adapters is that the Hercules tends to scrunch some graphic displays.

Compare the display shown on a true monochrome TTL system (A), the Hercules/TTL system (B), and a EP CGA composite monitor system (C). If you can see any difference other than screen brightness, you've got a good imagination. (The Hercules/TTL screen has more contrast.)
difference until his eyes got strained from the not-so-clear composite monitor display of the CGA card's character set. Also, the CGA card tended to distort the display of certain popular programs that were specifically intended for a TTL monochrome monitor. In fact, some of the very best "mono" software was not even readable, or legible, when using CGA/composite monitor equipment.

But create a void or a need and someone will rush to fill it. In this case, the solution was something called the Hercules Graphics Card, a device which has also been cloned, and what is usually provided in many rock-bottom-priced computer systems.

**One Card, Two functions**

The Hercules card was a really clever idea. In basic terms, it is a TTL monitor card intended specifically for TTL monitors, but it has the capability to display graphics on the TTL monitor. When running monochrome software, such as a word processor, the text characters are exactly the same as the high-quality characters you find on any of IBM's own monochrome adapter cards.

But, when running in the graphics mode, the entire display resembles that of IBM's CGA. The major difference is in the text characters, as you'll see soon. The Hercules characters are slightly more legible because they resemble the TTL characters. The only major difference between the Hercules and the CGA adapters is that the Hercules tends to **scratch** some graphic displays.

For example, Fig. 1A shows the screen display from an IBM TTL monochrome adapter. Figure 1B is the same TTL monitor, but with the monitor driven by the Hercules card. If you can see any difference other than screen brightness, you've got a good imagination. (The Hercules/TTL screen has more contrast because the Hercules card puts out a slightly stronger signal.)

Figure 1C is the same display using a CGA adapter and a composite monitor. Firstly, the vertical height is not as great. But more important, the characters are the CGA characters, which are not as easy to read as the TTL characters because they are separated by only one scanning line.

**Another Tough One**

Figure 3A and Fig. 3B show Printing Press, another popular "Can't run on TTL" program that actually looks better from the Hercules card (Fig. 3A) than it does from a CGA (Fig. 3B).

Figure 4A and Fig. 4B show a major problem with some graphics software. Figure 4A is Greeting Card running on a CGA system. The display is large and easy to read. Figure 4B is the same program running on the Hercules/TTL system. The vertical size is so greatly scrutinized that you need really fine eyesight to read some of the screen cues: the text size is smaller than many persons—particularly those wearing bifocals—can read.

But regardless of whether a graphics display is full-size or scrutinized, or whether or not the text characters of the Hercules system matches that of the CGA, the important point is that you can display graphics software on a TTL monochrome monitor, which is one heck of an advantage.

**Usually Supplied**

Because the Hercules card provides both monochrome and graphics capability, discount dealers have discovered that it saves their having to face disgruntled customers wanting to know why their computer can't run the same games that run on their friends' computers.

And so, many of the low-cost computer systems now use a Hercules-compatible graphics card even though the system is supplied with a TTL monitor. The cus-

(Continued on page 107)
To many, shortwave is a means of keeping up on the latest news as it happens around the world. It is a window to the cultures and music of other countries: a chance to learn or brush up on another language. But for many naturalized citizens and second-generation Americans, world-band radio is the way to keep contact with the land of their—or their parents—birth.

Often I get inquiries from listeners who want to know when and where to tune shortwave programs from the homeland—be it Germany, Greece, the Ukraine, or Turkey. So for those of Polish heritage, and other SWL's too, here's the information you've asked about.

Broadcasting began in Poland back in February 1925, with a feeble transmitter of less than 500 watts. Today's Radio Polonia—the foreign service of Polskie Radio Telewizja—has studios only a mile from the small building in Warsaw's Mokotow district where the original station was located. And the power of the Polish shortwave transmitters is 200 times greater than that first broadcaster.

The best time to hear Radio Polonia is during its daily English transmission to North America, from 0200 to 0400 UTC. Frequencies are 6.095, 6.135, 7.145, 7.270, 9.525, 11.815 and 15.120 kHz. Polish radio also airs 8 half-hour (actually 23 minutes is closer to the mark) English programs for Europe during the day, broadcasting its programming to the winds at 0630, 1200, 1600, 1730, 1830, 2030 and 2230 UTC, on various frequencies from the 6 to the 15 MHz bands.

Some of those transmissions can be heard in North America, too, even though they are not beammed in that direction. Worth trying, for instance, is the 2230 UTC program on 7.270 kHz.

How's the Programming?

Here what some British SWL's, commenting in Contact—the monthly publication of the World-DX Club—had to say about the programming:

J. Sheppard, commenting on the News, said that it "seemed reasonable enough for a Marxist country: not too obvious on the propaganda, and a fair mix of internal and international news items. The current affairs review...was notably more aggressive."

K. O'Reilly: "Anyone looking for magazine type programs will be disappointed in Polonia."

M. Leigh: "The Top 20 Show I found not bad—all the music was Polish and concentrated mostly on easy listening."

E. Southwell: "One of the better stations broadcasting from the Eastern Bloc."

M. Ward: "Manic and anarchic...three presenters compete to see who can talk the loudest, talk the most, and (oddly enough) who can sound spontaneous while reading their prepared script. The appalling state of Radio Polonia's audio quality is, possibly, without a rival on the shortwaves."

The program reviews, clearly, are mixed. But why not hear for yourself? If you want to write to the station for more schedule information, letters go to Radio Polonia, all Niepodleglosci 75/77, Warsaw, Poland.

A Smidgen of Pidgin

It began in the 17th Century, when western traders began dealing with local merchants along the China coast. The ship captains and brokers, arrogantly assuming that the merchants with whom they were dealing could never command a proper understanding of the English language, began talking to them in a sort of "baby talk."

In time, that form of communication between buyer and seller became a language in its own right, Pidgin English—pidgin being a Chinese corruption of the word "business."

From Canton and Hong Kong it spread south and east, with the white traders, to New Guinea and the smaller islands. Here, in today's Papua New Guinea, Pidgin is firmly entrenched (a living language), expanding rapidly with increased vitality. Listening to spoken Pidgin on the shortwave stations of the National Broadcasting Commission of Papua New Guinea is a fascinating experience.

In its development, Pidgin took the course of least resistance. As linguist Mario Pei puts it, the language adopted simple English words, strung together with local syntax, and disregarded things like gender and number, having verbs that were timeless and invariable.

The result is strange, to say the least. Missionaries, for example, translated the English word, bishop, into Pidgin thusly: topside-piece-Heaven-pidgin-man. And a butcher is manbelong-bullanae.

There are 19 Papua New Guinea shortwave stations operated by the government NBC. Several, such as the main NBC stations in Port Moresby, are not difficult to hear in North America. Probably the best frequencies to try are 4,890 and 9,520 kHz, during the early morning hours, from about 0000 UTC onward.

But you might want to look for some of
the other outlets, such as Radio North Solomon at Kietu in the Solomon Islands, on 3,325 kHz. In Pidgin, it's identification is Maus Bilong Sankamap. (Mouth Belong Sun-come-up, or, Voice of the Sunrise).

Other broadcasters include Radio Gulf at Kerema on 3,245 kHz; Radio East New Britain at Rabaul on 3,385 kHz; Radio Milne Bay at Alotau on 3,365 kHz, and Radio Western at Daru, 3,305 kHz.

In mid-winter, you may find some of the broadcasts with decent signals in the "around dawn" time slot, but a number of them can be very tough DX's to log, indeed.

What's Current?

Among the more interesting of the new shortwave voices to hit the airways is Radio for Peace International, set up in Costa Rica by the World Peace University, headquartered in Oregon, with United Nations support. Initial broadcasts have been aired on 7,380 kHz, but the station should also be on 15,405 kHz by the time you read this.

Programs on the 15-MHz frequency are to be aired between 2000 and 2300 UTC, and on the 7 MHz channel from 0000 to 0300 UTC. The programming, in both English and Spanish, is aimed at furthering world peace.

Scheduled to be on the air by now is the new international service of an old line Caribbean broadcaster, Radio TV Domi-
ticana (RTVD), the government station at Santo Domingo, Dominican Republic. RTVD, some years ago, did broadcast on shortwave. Now it is supposed to return to SW on 15,045 kHz, via the transmitter of Radio Discovery.

Jeff White, spokesman for the Florida-based World Radio Network that operates Radio Discovery, announced that RTVD has leased the air time from the station. RTVD's International Service will mostly feature Spanish programs from the domestic Dominican medium-wave (MW) network, relayed on shortwave. But twice a day, World Radio Network will produce English programs for RTVD's shortwave service. The programs, "This is Santo Domingo," is to air daily from 2000 to 2100 UTC, and again from 0000 to 0100 UTC.

Down the Dial

Here's where you get your say. In this corner of our monthly column, we look at what SWL's are hearing. Your reports could be here next month. Drop me a line--the address is Jensen on DXing.

Hands-on Electronics, 500-B Bi-County Blvd., Farmingdale, NY 11735—and let me know what you are logging on shortwave. Include where you heard it (frequency in kHz) and when in UTC (Universal Coordinated Time). Any questions about SWL'ing also are welcome, as are photos of you and your equipment, to appear in these pages in future columns.

3,905 kHz—one of the Papua New Guinea SW stations, Radio New Ireland, has been logged in California at 1045 UTC with Pidgin talk and native vocals.

7,145 kHz—Radio Algiers can be heard by North American SWL's during an afternoon transmission, until 2300 with music, political talk, and identification—all in Arabic—prior to close-down for the day.

7,240 kHz—Here's another ethnic special for those who might want to hear programming from the land of their heritage. You can hear English news from Radio Yugoslavia in Belgrade at 2100 UTC.

9,545 kHz—A nice piece of DX is the Solomon Islands Broadcasting Corporation's (SIBC) station at Honiara, which can be heard, in English, around 0730 UTC with news and weather, including information on the tides for shippers and fishermen.

13,650 kHz—Some broadcasters can be found in this new shortwave band, including Radio Baghdad, Iraq. It has been reported here about 1820 UTC with nonstop Arabic music.

15,150 kHz—Radio Canada Interna-
tional's "SWL Digest," a weekly program for shortwave enthusiasts, hosted by Ian McFarland, is one of the more popular SW shows on the air. Look for it Saturdays at 2130 UTC. And, if it's the third weekend of alternate months—March, July, Sept., etc.—you will hear yours truly with a regular shortwave feature. Don Jensen's Journal.

Credits: Ken Kashiwabara, CA; Richard D'Angelo, PA; Dustin Brann, CA; Peter Dillon, MD; Art Harris, NY; Raymond Slagt, Aruba, Netherlands Antilles

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MARCH 1986
The SR-15 is the fifth model in Cobra's recently introduced family of sophisticated scanners. User-programmable, the unit tunes from 29.0 through 54 MHz, 108 to 174 MHz, 406 through 512 MHz—covering public safety, ham, federal government, business, industrial, maritime, aero and land transportation bands.

The memory feature of the SR-15 (each of the five banks can store up to 20 frequencies) enables added monitoring convenience. Channels can be grouped within the banks in any order for faster and easier access: the memory banks can be monitored separately as well as together in any combination.

A complete array of automatic operating functions are featured: including normal scan, automatic search, channel lockouts, channel priority, selective scan delay, channel hold and manual scan, keyboard lock, and light button.

The Cobra SR-15 has a back-lit, 6-digit LCD display that indicates the frequency, channel position, plus the status of the priority, lockout, delay, and the 5-position memory bank.

The unit offers excellent program memory protection. It will retain your frequencies for up to an hour without batteries, saving you the time and effort of recalling which frequencies you had programmed into the SR-15, and then reprogramming the set.

Ultra compact (6-in. high, 2-3/4-in. wide) and weighing in at only 10 ounces, the SR-15 is housed in a weather-resistant anodized aluminum and high-impact ABS plastic case. Accessories include a rubberized antenna, rechargeable battery pack, AC adapter/charger, earphone, and carrying case. Its suggested retail is $299.95.

For more information, contact Cobra/Dynascan Corp., 6500 West Cortland St., Chicago, IL 60635, or circle No. 40 on the Free Information Card.

Question Box

Paul J. Stever of Placentia, CA writes to say that not long ago he purchased a Realistic PRO-32 handheld scanner. He'd like to know how he can get it to scan through frequency ranges that weren't designed into the unit—for instance above 800 MHz. Also, he wants to increase the range of the unit with a different antenna.

The Realistic PRO-32, with its 200 programmable channels, is a fine scanner, but it's not easy finding add-on accessories that will turn a handheld into much more than you started out with. Several 800-MHz band converters have appeared, but they're intended for base station installation. In other words, if you want 800 MHz in a handheld, you're going to have to check out a handheld like the new Uniden Bearcat BC-205XLT that includes that band in its original design.

Same thing with the antenna: unless you're willing to hook the PRO-32 to a base-station antenna, you're going to have to accept the fact that a couple of inches of rubber-duckie antenna just doesn't equal a roof job. No way, no how.

Walter Kibler of Cedar Rapids, IA would like to monitor some of the activities of the Army Corps of Engineers and asks if we can point him in the right direction.

Well, Walter, that agency operates on a great many VHF and UHF frequencies, although I've found that 163.4125, 163.4375, 163.5625 and 165.1875 MHz seem to be quite active in several areas. An excellent listing of the Corps' engineer stations/frequencies is contained in the 6th Edition of the Top Secret Registry of U.S. Government Radio Frequencies, available at $17.95 (plus $2 postage/handling) from CRB Research, PO Box 56-GP, Commack, NY 11725.

Reader Barry Hirsh says that in recent months, the huge New York City Transit Authority communications system was modified and modernized. That's just fine, except for the fact that the new system appears to have different frequencies than the old system. Basically, Barry wants to get a handle on the new frequencies used by the NYC Transit Police. We suggest that you try out 160.905 MHz for Manhattan, 160.50 MHz for the Bronx, 160.305 MHz for Brooklyn, and 160 MHz for Queens.

Kyle Merrweather, who hails from southern California, points out that the TV news always shows law enforcement officers negotiating with aircraft hijackers via two-way radio. Since that practice is so widespread, he wonders how the aircraft are able to operate on police VHF/ UHF FM frequencies when the VHF-aero band uses AM and is on other frequencies. That's a good question.

The police and FBI communicate with the hijackers over frequencies in the VHF-aero band, thus eliminating the need for the plane to have police-band communications capabilities. Mostly, frequencies in the 121.6 to 121.925 MHz portion of the band are used. Those frequencies are all set aside for airport ground-control use and won't cause interference to in-flight communications. At airports where there is an assigned ground-control frequency.

(Continued on page 105)
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To order your first module for a 15-day trial examination, simply complete the card and send today! If the card is missing, write to us for ordering information.
As most of you may be aware, I grew up with vacuum tubes and had just gotten a bite on transistors, when I did what most editors do—I went into public relations where the “real” money is! I learned to use (and abuse) adjectives, and sing loudly the praises of equipment I usually hadn’t actually seen, much less tested.

I was hired for that particular job because they had just picked up a new client (a Japanese electronics manufacturer) and needed someone with a sci-fi background. My first assignment was to write a new-product release about a portable tape recorder—which I quickly did, and passed along to the boss. He read it, understood none, and told me “It had damned well better be right...Two Japanese engineers would be looking it over.”

We gave them the release, which was quickly tucked into a jacket pocket; it first had to be translated into Japanese. Two days later, my boss got a call to come (with me) to the office, as “corrections had to be made.” The drive to their office was horrid. The boss wouldn’t talk to me, and kept slamming the steering wheel with one hand. When I tried to start up a conversation, he turned the radio to near maximum volume.

When we finally arrived, there were the two engineers, along with a translator. We all smiled, bowed, and exchanged a few good mornings. The interpreter said, “See? Where begin with name of town? Is followed by NJ. Must have decimal points after N and J.” That was it. The only thing wrong with my release was that they wanted “decimal points” after the N and J in New Jersey. Needless to say, We didn’t play the radio on the way home. Instead, we talked about my raise. But living under the sword of Damocles isn’t my way, and I got back into editing.

I remembered that vacuum-tube symbols had grid, cathode, and plate symbols inside the circle representing the tube envelope. Even transistors showed a base, collector, and emitter.

Now, there were rectangular boxes (integrated circuits or IC’s) with nothing but a part number inside, with numbers all about the perimeter. I was lost. And my old familiar Remington Upright finally gave way to a computer terminal.

I struggled...I sweated...I was even ready to go back to public relations writing, where knowing little could be an advantage! Then I walked into one of my fellow editors’ offices and asked “What’s inside these boxes?” “Who cares?” He said, “as long as it does the job intended.”

Maxwell, Faraday, and even Volta wouldn’t know what an IC did or was for, anymore than I did! But I could, at least, still learn about them. I did, and we’ll probably go further into this later on. Meanwhile, let’s check the mail and see how many copies of the Fips Book it’s going to cost the boss!

Welcome Ladies!

Apparently, not all of our readers are men. We’re happy to welcome this letter from Cathy Primerano, of Bloomfield, NJ, who points out that it seems to surprise everybody, but she’s also an electronics experimenter! Well, why not—why should the boys have all the fun?

Cathy’s offering, shown in Fig. 1, is a rather unusual adaptation of the many “locator” projects that have appeared in these pages from time to time. Taking a pair of old, portable (under 100-mW) CB transceivers the kids weren’t using any longer, she removed the loudspeaker from one of the units. In place of the loudspeaker, a small Piezo buzzer (BZ1) was wired in with a small relay and battery to operate the buzzer. That unit is then wired to her dog’s collar in such a way that the antenna will be upright when worn.

Then came a period of training the dog. With a long cord attached to the dog’s collar, she let the dog out to the end of the rope. Then by pressing the transmit switch on the other CB unit, and (at the same time) dragging the dog back in she got him to return home, where he’d be rewarded. By repeating that procedure a few times, the dog learned that when the buzzer went off, he’d get a reward if he rushed home.

Once trained to respond to the buzzer, Cathy says that the rope was eliminated. Now when she wants Fido, she just presses the transmit button, and the next thing you know, he’s scratching at the back door to get in!

Thanks Cathy, it’s nice to know we have some distaff readers out there too. And by the way, your copy of the Fips book should be there by now!

Wake Up Caller

Hey! John Macklin of New Florence, MO...Look what you went and did! You’ve gone and earned yourself a copy of the Fips book, which should be in your hands by the time you read this.

John explains that he and a friend of his were discussing problems, when his friend said that his worst problem was oversleeping because his alarm clock didn’t do the job. The design they came up with worked so well, that John decided to share it with us.

Adapt the circuit in Fig. 2 to any radio alarm clock and connect it directly to the

---

Fig. 1—The locator circuit is built around portable CB transceivers with the speaker from one of the units. Then a relay is wired in its place, and the relay contacts are wired to operate a piezoe buzzer (BZ1).
earphone jack on the radio. The audio frequency current fluctuations to which
your loudspeaker usually responds, trig-
gers a relay which, along with the correct
components, will control an AC outlet
which, in turn, will control any sort of
alarm device from a loud bell, to a light
bulb, or anything else that will do the job
of getting you from sleep to wakefulness.

Be careful of the RMS AC output of the
radio as it can make or break this circuit.
The linear bias region of the Q1 transistor
which has been set for a fixed gain is
basically a common base-biased tran-
sistor configuration which allows the mid-
range audio frequencies to be linearly am-
plified, transformed and then rectified by
the combination of transformer T1 and
diodes D1–D4. Capacitor C3 filters out
current fluctuations over Q2 by charging
to the peak value of the rectified current
allowing a positive-negative voltage drop
across C3. The current flow through R5 is
part of a biasing network which chooses
the transistor biasing region which, for
silicon is 0.7-volts base-to-emitter. When
C3 is charged to at least 0.8 volts. That
causes Q2 to become forward biased, al-
lowing the transistor to conduct produc-
ing a voltage drop across the relay which
must reach eight volts to trip.

Just be careful that the relay’s contact
ratings can handle the applied load!

Simple Alarm Circuit

Tim White of Elmore City, OK (I swear,
you guys are making up the names of
those towns!) has come up with a winner
in the form of a simple burglar-alarm
circuit.

Turn your attention to Fig. 3. Tim says,
and I quote, “It works like a million
bucks.” However, checking the sche-
matic, you’ll find that it doesn’t cost any-
where near that much!

Here’s Tim’s description: Gate bias for
the SCR is provided by the voltage divider
formed by resistors R1 and R2. However,

R2 is shorted out by whatever is connected
to points A and B, which could take the
form of switches, phototransistors, foil
tape, etc.

When the short is opened (presumably
by a would-be intruder), the SCR fires and
your alarm—be it a bell, buzzer, siren,
relay or light—will indicate an un-
authorized entry. The two potentiometers
can be replaced by fixed resistors once
you are satisfied with the operation of the
circuit. And in keeping with our infor-
mality of design, Tim adds that “R3 could
be removed.”

Tim finished his letter by asking if his
circuit was worth a Fips book. Watch the
mails Tim, watch the mails!

Tone Generator

Here’s some happy news for Roger In-
gersoll, of Irving, TX. He tells us that a
couple of years ago, he was installing a
massive sound system in a convention
center, and found himself in need of a tone
generator to test the cables. Well, there
seemed to have been some 20 other men
working on the same project, but there
were only two tone generators! Roger
checked out the price of a commercial unit
and when the salesman said “$30,” Roger
made a wise decision—he de-
signed and built his own.

Roger’s circuit is shown in Fig. 4. The
circuit is built around two 555 oscillator/ti-

Fig. 3—The Alarm Circuit uses SCR1
connected in a closed-loop) to detect
unauthorized-entry. When the loop opens,
the SCR turns on, sounding the buzzer.

Fig. 4—The Tone Generator is built
around two 555 oscillator/timers (although
a 556 dual oscillator/timer would have
minimized the parts count) each configured
as astable multivibrators, and operating at
different frequencies.
timers, with both set up for astable operation, each providing different output frequencies in order to produce an alternating tone. The output capacitor, C1, keeps the tone stable as the output drops in resistance. The LED acts as a pilot lamp, while flashing in sync with the operating tone.

The component values are not at all critical—use what you have. The output frequency can be computed for each half by the formula:

\[ F = \frac{1.44}{(R1 + 2R2)C1} \]

where R1 is the resistor between pin 7 and 8, and R2 is the resistor between pins 6 and 7.

The circuit can be powered from an ordinary nine-volt transistor-radio battery: for added convenience, do add a single-pole, single-throw switch to power down the circuit when it’s not in use. As an afterthought, a 556 timer would have done a better job, but I didn’t happen to have one in my junkbox.

By now Rog, you should have received your copy of the Fips book. Hope that you enjoy it.

Battery Substitute

David R. Bell, of Reisterstown, MD, sends us a brief letter, with a brief explanation, and a brief diagram. (See Fig. 5.) Well, they say that Sampson was a piker—he slew a thousand men with the jawbone of an ass, and that millions of sales are killed each day with the same weapon!

The problem Dave, is that we have to provide sufficient verbiage to at least wrap some copy around the diagram! One of the first things that most authors learn is that magazines pay by the amount of space a story takes. But that always results in the editor having to kill some copy because it’s too long.

You needn’t concern yourself however, because your schematic looks like a serviceable adjunct to any project that requires battery power. With so few parts, it’s certain that any active electronics hobbyist will quickly recoup the cost of putting that circuit together in batteries he no longer has to buy!

Dave explains that he originally designed his circuit as a battery replacement in a VTVM. It provides good voltage regulation, and the AC source can be an ordinary 6.3-volt filament transformer.

Hmmph! That about does it, and it should be sufficient to cover the Fips book we’re sending out to you.

Ripple Suppressor

Got an interesting and useful project from Stephen M. Hart, of Omaha, NE. Steve, let’s start with basics. You addressed me as “Mr. Wels.” Mr. Wels is my father. I’m “Byron.” The fact of the matter is we’re pretty informal around here, and I’d like to think that we can function best on a first-name basis.

With that said, let’s now turn our attention to Steve’s schematic diagram, which is shown in Fig. 6. Most AC adaptors are loaded with an unacceptable ripple content. For a ripple of 450 mV, this circuit suppresses that component to less than 5 mV. Resistor R1 provides the necessary bias current to keep the transistor in the active region. Capacitor C1 provides the necessary attenuation for the ripple factor.

Transistor Tester

Lou Supek, of Brunswick, OH, writes to say that he’s an old subscriber to Radio Craft Magazine, (or did he mean he was a subscriber to the old Radio Craft?) and his comment that he wound many plug-in coils on 20IA tube bases, brought back a flood of memories. I wonder if you (or any of our other readers can still name the grids in a pentagrid converter? Or maybe tell me what the unique characteristic of the old 6E5 tube was?

I remember reading about a great electronics inventor who was being presented with an award, and responded that “If I have been able to see farther than my fellows, it was only because I stood on the shoulders of giants.” In our electronics business, that was never quite as true as it is today—Past is prologue, and the future will be the result of the past. See Fig. 7.

Today’s electronic gadgetry, such as VCR’s, stereo components, TV’s and the like, are loaded with transistors. Normal testing procedures usually require that the component be unsoldered and re-soldered—a task that not only becomes a time-consuming pain in the neck, but can also result in mistakes, making the possibility of leaving the unit being serviced in worse condition than it was when we started on it.

They say a doctor buries his mistakes, a chef eats his mistakes, and we in electronics have shelves full of projects that we haven’t finished working on yet! Lou’s circuit, although he didn’t provide step-by-step, component-by-component description, is almost self-explanatory. The bottom line is that the three clip leads get connected to the transistor terminals, and you don’t have to unsolder them!

Lou, your copy of the Fips book is on the way. Enjoy.

5-volts On Tap!

Dave Martin, of Knoxville, TN, built this super-simple circuit and says that (Continued on page 100)
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MAH-1008
COMPU-THROTTLE

(Continued from page 38)

U1, Pin 12—Momentum switch
U1, Pin 13—Up switch
U1, Pin 14—Starting Speed switch
U1, Pin 15—Down switch

Because of the way modular telephone extension cords are built, it is easy to get them wired up wrong (Refer back to the construction section of this article).

Further Experimentation

Several of the components in the output circuit are on a DIP header that is referred to in Fig. 1B as U4. The DIP header allows for experimentation by providing easy access to the circuit. There are many possible component configurations and many possible output signals that can be generated.

Several signals are available:

- Pins 1 through 6, and 8—Output to regulator
- Pins 11 and 12—Main pulse signal
- Pin 9—Average pulse signal
- Pin 16—Filtered DC signal
- Pins 13, 14, 15—Ground
- Pins 7 and 10—Spare

Fig. 6—To obtain the two pure pulse outputs shown in Figs. 2 and 3, connect the following circuit elements into the DIP header (U4) using the terminals indicated.

Fig. 7—Either pulse output shown in Figs. 2 and 3 may be filtered with a resistor and capacitor into a pure DC voltage like the one shown in Fig. 4 by using the above hookup.

The Main Pulse Signal has a duty cycle which is 80% on and 20% off at full speed. The Average Pulse Signal is the same below half speed. Above half speed, it increases slightly faster so that at full speed it is fully on.

A word or two about motors and power packs is in order here: Many commercially available power packs have a provision for “Pulse Power.” The normal output is full-wave rectified DC. When pulse power is selected, the output is changed to half wave. While this gives better low-speed performance, it also causes the motor in the locomotive to heat up at higher speeds. This is due to the “back emf” of the motor, caused by its inductance. Since an inductor resists a change of current, when the power is turned off, it acts like a flywheel. The higher the frequency, the greater the flywheel effect. Low frequencies cause large pulses of current, which cause heating. Since the Compu-Throttle’s pulse output starts at a low frequency (30 Hz), but increases at higher speeds (up to 400 Hz), this problem is minimized.

Many waveforms can be generated by reconfiguring the components on the header. Some that the author has experimented with are:

1. A pure pulse output using just the main pulse output (See Figs. 2 and 6).
2. A pure pulse output using just the average pulse output (See Figs. 3 and 6).
3. Either pulse output can be filtered with a resistor and capacitor into a pure DC voltage. (See Figs. 4 and 7). This is similar to many “transistor throttles” on the market.

4. The circuit shown in Fig. 8 provides a combination of pulse and DC power (See Fig. 5). The main-pulse output is used to provide a full-voltage pulse while the average-pulse output is used to provide the DC component. The two signal diodes in this circuit cause the output to be the higher of the two signals, so the main-pulse output always provides the full “on” part of the pulse while the average output provides the low (or “off”) part of the signal. This signal may reduce the heating of some motors.

The difference between configurations 1 and 2 is that configuration 2 will cause the train to run slightly faster. Conversely, for a given maximum speed, the first one allows a higher voltage starting pulse. This may be needed for starting some especially sluggish locomotives.

In all configurations, the Zener diode determines the maximum “on” voltage. If it is omitted, it will vary with the unregulated, filtered, power supply. This will probably not be noticed unless the load changes while a train is running (for example, another train enters the same circuit), at which time both will change speed. An advantage of eliminating the Zener diode is that U6 will have a minimum voltage drop and will not generate as much heat.

The author has used all but configuration 3, and all seem to work well.

5- and 10-ampere adjustable voltage regulators are available if the 3-ampere output rating is not enough for your application. Keep in mind that T1, BR1, D1, Q1, and Q4 also carry the entire output current, so they must also be changed.

Although the author has not tried it, with some additional circuitry it should be possible to detect the presence of a train on the track circuit. See discussion of the tri-color LED below.

Operation

The Compu-Throttle is easy to operate once you get the
feel of it. Full instructions are given in the Operator Instructions.

The tri-color LED1 in conjunction with the LM350 provides some useful information on the condition of the track circuit. The track circuit can be open, shorted, or have a normal load (the motor in the locomotive). The throttle can either be off, on forward, or on reverse. The color of LED1 indicates the currently selected direction. The LM350 and the output circuitry have an interesting side effect: Even if the throttle is in the stopped position, if there is no load (an open circuit), there will be enough voltage to light LED1 dimly. If you watch LED1 while you put a train on the track, you will notice that it glows while the circuit is open, and goes out when the circuit through the motor is complete. While the train is running, the brightness of LED1 increases as the speed is increased, unless there is a short circuit, at which time LED1 will go out.

These indications are summarized in Table 3.

**TABLE 3—LED1 COLOR**

<table>
<thead>
<tr>
<th>Compu-Throttle:</th>
<th>Track Condition</th>
<th>Open</th>
<th>Normal</th>
<th>Shorted</th>
</tr>
</thead>
<tbody>
<tr>
<td>Stopped forward</td>
<td>Dim Green</td>
<td>Off</td>
<td>Off</td>
<td>Off</td>
</tr>
<tr>
<td>Stopped reverse</td>
<td>Dim Red</td>
<td>Off</td>
<td>Off</td>
<td>Off</td>
</tr>
<tr>
<td>On forward</td>
<td>Bright Green</td>
<td>Bright Green</td>
<td>Off</td>
<td></td>
</tr>
<tr>
<td>On reverse</td>
<td>Bright Red</td>
<td>Off</td>
<td>Off</td>
<td>Off</td>
</tr>
</tbody>
</table>

**Conclusion**

Having a microcomputer as the heart of the Compu-Throttle gives much flexibility to the final design and a reduced parts count. It also makes it easier to change the performance of the whole circuit. Self-test software makes it possible for the builder to check out the circuit and the microcomputer itself with just a voltmeter.

Even though this project will improve the performance of your model train locomotives, the better conditions the locomotives are in, the better they will run.

The author has enjoyed designing, building, and using this project. Even more enjoyable is seeing other model railroaders use and enjoy it. The other member of the model railroad club that the author belongs to enjoy using it so much that they have been heard to say that they would rather not run trains if the Compu-Throttle were not available.

---

Try to think of it as a "digital" transmission. One bang on the pipe means 'five minutes to dinner'—two means 'get up here right away!'
THE UKI OKI PRINTER SWITCH
(Continued from page 45)

are inserted all the way into the Oki's EPROM sockets. Those pins are larger than regular IC pins and require extra force to seat properly.

Documentation and Extras
A single IBM-format disk is included with the UKI switch. It contains short documentation files for installing the circuit board in the supported printers, and a utility that allows you to set up your printer for pitch (10, 12, or 17 characters per inch), double-width characters, print quality (draft or correspondence), enlarged or emphasized print, line spacing (six or eight lines per inch), page length, tab columns, and left margin (in Oki mode) or paper-out alert (in IBM mode). The set-up codes can be sent directly to the printer or saved in a disk file. You can then create several set-up files for different kinds of print jobs.

For example, you might want to run off BASIC program listings in compressed type at 8 lines per inch. You'd select the appropriate items from the menu, and then save them as a file called LISTINGS.OKI. Then, to put your printer in the desired mode, you'd type the following (assuming you're working from a hard disk):

C>TYPE LISTINGS.OKI>PRN

Other files might contain settings for letters, extra-wide

WELS' THINK TANK
(Continued from page 96)

what makes this his favorite, is that it is also his first effort at independent circuit design. (See Fig. 8.) He's using a 7805 voltage regulator and Dave admits to one flaw, in that the problem is that 9-volt transistor radio batteries usually don't last very long. However, this unit allows one to transport a TTL circuit away from the workbench and it also can serve as a temporary 5-volt power supply until you get around to building something a little more suitable.

Those of you who are considering a circuit of this sort would do well to go back and look at some of the other circuits in this very issue! We've even got regulators and filters and all sorts of goodies that could work together to do all sorts of outstanding things.

Fig. 8—This circuit, built around the 7805 5-volt regulator, can power low-current (1 amp or less) 5-volt circuits, such as TTL or CMOS based projects.

Headlight Reminder
David Herman, of Laurel, Montana, tells us that when it comes to his car's headlights, he's totally absentminded. Don't sweat it Dave: we've all of us done

Q2. Q1 is de-energized, and current flows through R3, the buzzer, Q2 and D1, so guess what? Right! The buzzer sounds.

There really isn't a critical part in the entire circuit, and your junkbox should yield everything that you'll need, maybe even the buzzer. The hardest part of the job is finding the proper wires under the hood or the dashboard.

Power Supply
In the mail, Randy! Randall Mason, of Raleigh, NC, not only sent in a beautiful power supply idea, he provided full and excellent descriptive text, and a schematic diagram (see Fig. 10) that made our art director totally flip out! "I wish they could all come in like that," he said. "But if they did, I'd lose my job."

Randy explains that this power supply circuit is capable of converting a positive 12- or 15-volt source to provide a negative 5.2-volt output, while delivering 0-50 mA of current. In operation, the 555 timer's frequency varies with the load current. The resulting positive output pulses (at pin 3) vary in repetition rate, but have a relatively constant duration. Each pulse turns on Q1 and rapidly charges C4. When pin 3 pulls the positive side of C4 low, the charge on C4 is transferred to C3. Line and load regulation are approximately 1%. C1 charges through R1 while pin 3 is high. Pin 3 goes low when pin 6 reaches 2/3 of the Vcc.

Capacitor C1 then discharges through R3, Q2, and the IC's open collector discharges the transistor at pin 7; transistor

(Continued on page 103)
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Practise Problem

Here’s a problem that will help you review all the key facts you have learned here. 7. A series LCR circuit like that in Fig. 18 has the following specifications:

\[
R = 1000 \text{ ohms} \\
X_L = 2268 \\
X_C = 4000 \text{ ohms} \\
\text{source voltage (V_S)} = 40 \text{ volts.}
\]

Answers to Practise Problems

1. 51°
   If A + B = 90° in a right triangle, then:
   \[
   A = 90° - B \\
   A = 90 - 39 = 51°
   \]

2. 12 cm
   \[
   b = \sqrt{c^2 - a^2} \\
   b = \sqrt{15^2 - 9^2} \\
   b = \sqrt{225 - 81} \\
   b = 144 = \sqrt{12}
   \]

3. 60 volts
   If 2 cm is equivalent to 10 volts, then:
   \[
   1 \text{ cm} = 10/2 = 5 \text{ volts}
   \]
   A 12 cm vector then is
   \[
   12 \times 5 = 60 \text{ volts.}
   \]

4. 136 lb
   If the pulling forces are perpendicular, then a right angle is formed so the Pythagorean relation can be used.
   \[
   c = \sqrt{a^2 + b^2} \\
   c = \sqrt{110^2 + 80^2} \\
   c = \sqrt{12100 + 6400} \\
   c = \sqrt{18500} = 136 \text{ lb}
   \]

5. 9 volts
   \[
   V_L = \sqrt{V_S^2 - V_R^2} \\
   V_L = \sqrt{40^2 - 20^2} \\
   V_L = \sqrt{15^2 - 12^2} \\
   V_L = \sqrt{225 - 144} \\
   V_L = \sqrt{81} = 9 \text{ volts}
   \]

6. .2 A
   First find the total circuit impedance:
   \[
   Z = \sqrt{R^2 + X_C^2} \\
   Z = \sqrt{20^2 + 30^2} \\
   Z = \sqrt{400 + 900} = \sqrt{1300} = 36 \text{ ohms}
   \]
   Next, use Ohm’s law to find the current.
   \[
   I = \frac{V_S}{Z} \\
   I = \frac{7.2}{36} = .2 \text{ A}
   \]

7. a. 2000 ohms
   \[
   Z = \sqrt{R^2 + (X_C - X_L)^2} \\
   Z = \sqrt{1000^2 + (4000 - 3368)^2} \\
   Z = \sqrt{1000^2 + 1732^2} \\
   Z = \sqrt{1000000 + 3000000} \\
   Z = \sqrt{4000000} = 2000 \text{ ohms}
   \]
   b. 20 mA
   \[
   I = \frac{V_S}{Z} = \frac{40}{2000} = .02A = 20 \text{ mA}
   \]
   c. 80 volts
   \[
   V_S = IR = (.02)(1000) = 20 \text{ volts}
   \]
   \[
   V_L = IX_L = (.02)(2268) = 45.36 \text{ volts}
   \]
   \[
   V_C = IX_C = (.02)(4000) = 80 \text{ volts}
   \]
   d. 40 volts
   \[
   V_S = \sqrt{V_R^2 + (V_C - V_L)^2} \\
   V_S = \sqrt{20^2 + (80 - 45.36)^2} \\
   V_S = \sqrt{20^2 + 34.64^2} \\
   V_S = \sqrt{400 + 1200} \\
   V_S = \sqrt{1600} = 40 \text{ volts}
   \]

**Thyristors (Continued from page 70)**

Series resistance is more compatible with the output-sourcing capability of standard CMOS logic gates; the less current that is sourced, the higher the CMOS high-level output voltage. That allows the high logic-level output to be used for driving other CMOS logic, if so required.

However, the high gate impedance can create dv/dt problems. Junction capacitance can couple a capacitive current to the device input when a fast positive-going step voltage is large enough, conduction might be inadvertently induced. To minimize the dv/dt effect, the gate of the MOS-SCR should be terminated in an impedance similar to the gate-cathode resistance of a sensitive gate SCR or gate-source resistance of a power MOSFET.

The resistance by-passes the capacitive current from the input of the device; the lower the resistance, the greater the degree of shunting. With low resistance, the MOS SCR has greater dv/dt immunity and higher blocking voltage capability, as illustrated by the curve in Fig. 10.

For optimum CMOS to MOS SCR operation, two impedance conditions should be satisfied: The gate circuit should limit the degree of the MOSFET turn-on, and the gate-cathode impedance should be low to increase device blocking-voltage capability.

**LASCR's**

The Light-Activated Silicon-Controlled Rectifier (LASCR) is another off-shoot of the SCR, which can be triggered into conduction by either a beam of light (a form of electromagnetic radiation) or an electrical pulse applied to its light/current-sensitive gate area.

Its basic four-layer PNPN construction is similar to that of ordinary SCR's with one exception: The PN junctions are formed on a silicon pellet in an elongated manner to permit irradiation by a light source. The spectrum of visible light
for activation ranges in color from red to violet, corresponding to wavelengths between 0.7 microns (red) down to 0.4 microns (violet). (One micron is one-millionth of a meter.) Photosensitive devices such as the LASCR respond more efficiently to the longer wavelengths of the infrared region.

The LASCR’s uses include power switching and control applications such as optical sensing, phase control, computers, and related digital-electronic control systems.

**WELS’ THINK TANK**
(Continued from page 100)

Q2 stays off as long as the negative output voltage is large enough to turn on the 6.0-volt Zener, D5. When load current discharges C3 enough to turn off D5, Q2 turns on and begins another discharge of C1. The resulting output ripple is about 100 mV. Unless the load current is light, ripple frequency is above the audio range. Either way, even if load current varies, ripple is inaudible because no chokes or transformers are present.

The output does increase slightly with temperature, and you can (if you wish) generate a more-negative output by selecting a higher-voltage Zener (D5); but be aware that diode losses limit the output to ~8 volts at low-current levels. U1 should never be driven by more than 15 volts, especially at full load.

**Sump-Pump Controller**

Remember, you saw it here first! I’m telling you, we bring you the latest, the utmost, the ipissima verba, the be all and end all… Al Disciullo, of Ashland, MA, offers a circuit (See Fig. 11) to completely control your sump pump! Al, what are you guys going to come up with next?

Al tells us that this circuit does away with such things as float switches that can get stuck, and he speaks from personal experience. He says the circuit is extremely reliable and easy to build. And he put his together from junk box parts. He put his into service over two years ago, and hasn’t had to touch it since then.

**Fig. 10**—This circuit, built around a 555 oscillator/timer, digests a positive 12- to 15-volt input, to provide a negative 5.2-volt output.

**Fig. 11**—The Sump-Pump Controller is designed to do away with such things as float switches that can get stuck. Its operation begins when water in the sump places a short across the probe input to the circuit, placing the input to U1-a at ground potential.
The circuit's operation begins when water in the sump causes a short across the (probe) input, placing pin 1 of U1a (a hex Schmitt trigger) at ground potential. That causes the output of U1a at pin 2 to go high. The high output of U1a is fed to pin 2 of U2 (a dual non-retriggerable monostable multivibrator), forcing its output to go high which, in turn, causes the output of U1b to go low.

With the output of U1b now low, LED1 lights and the sump pump is triggered into operation via solid-state switch S1. With the RC values given in the schematic diagram, the circuit sets the pump's run time at about 22 seconds, reducing the water level within the pump. The cycle repeats whenever the water level rises sufficiently to contact the probe. Also included, but not required, is a power-on reset, formed by R2 and C2. Reset switch S2 and test switch S1.

You can make a simple probe out of a piece of coaxial cable by separately tinning both inner conductor and shield. The tinned ends serve as the probe for the circuit. Using coax also improves the circuit's noise immunity, thereby, preventing false triggering due to external sources. And because the probe is used to sense the presence of water, a line-isolated power transformer should be used for the 12-volt supply.

What's it good for? Well, we like it as-is, but imagine its use as a flashing pilot lamp. No matter what application you find for it, it's certain to attract a good-deal more attention!

LED Flasher

Here's a quickie from T.R. Harris.

Fig. 12—The LED Flasher is just about the simplest do-nothing circuit that you can build. So if you're into blinking lights, here's your chance to get your hands dirty.

Okay, T.R. Because you sent us so little written material with this one, we're only sending you half of a FIPS book.

Take a look at Fig. 12 and you'll see just about the simplest do nothing circuit we've ever done. Using an LM3909 flasher chip and a few extra components, you can make the LED flash on and off. If you aren't happy with the flash rate and want to speed it up or slow it down, all you have to do is vary the capacitance of capacitor C1. The circuit can easily be powered by an ordinary "AA" cell, which should be good for about four days of continuous duty.

If you think the LED isn't lighting up brightly enough, you can add additional voltage, but be careful not to exceed four volts. If you do go over four volts, you're either going to fry the chip or blow the LED.

LED Flasher

Here's a quickie from T.R. Harris.
SAXON ON SCANNERS

(Continued from page 90)

that's the channel used.

As an interesting sidelight, pilots and air controllers have a secret codeword that permits the pilot to clue in the ground

CIRCLE 99 ON FREE INFORMATION CARD

BOOKSHELF
(Continued from page 27)

The Complete Computer Career Guide
By Judith Norback, Ph.D.

The rapidly expanding computer job market can be confusing at best to those who would like to get into the field, but don't know where to begin. Now, computer expert and educator, Judith Norback provides a realistic assessment of 25 major computer career options ranging from technician, data entry operator, systems programmer, and systems analyst to software engineer, hardware designer, computer salesman, and computer consultant.

Based on interviews with those currently working in the computer field, The Complete Computer Career Guide delves into real-life job requirements, benefits, and drawbacks. In addition to job description, educational and experience

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<td>SP10 Special Projects #10 (Spring 1984)</td>
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<td>111 Hands-On Electronics  #1</td>
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<td>116A Hands-On Electronics  #6</td>
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<td>116B Hands-On Electronics (Jan-Feb 86)</td>
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<td>116C Hands-On Electronics (Mar-Apr 86)</td>
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<td>116D Hands-On Electronics (May-Jun 86)</td>
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<tr>
<td>117 Hands-On Electronics back issues (1987)</td>
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<td>120 TV Descrambler</td>
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<tr>
<td>121 Home Experiments Handbook</td>
<td>$3.95</td>
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<td>122 8-Ball Satellite TV Antenna</td>
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<tr>
<td>123 Radio-Electronics back issues (1987)</td>
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<tr>
<td>124 Radio-Electronics back issues (1988)</td>
<td>$3.75</td>
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To order any of the items indicated above, check off the ones you want. Complete the order form below. Include your payment, check or money order (DO NOT SEND CASH), and mail to Hands-on-Electronics, Reprint Department, P.O. Box 4079, Farmingdale, NY 11735. Please allow 4-6 weeks for delivery.

**ARTICLE**

**PAGES MONTH YEAR**

**TOTAL PAGES @50¢ each TOTAL PRICE**

| MAIL TO: Hands-on-Electronics | Reprint Bookstore, P.O. Box 4079, Farmingdale, NY 11735 |
| TOTAL PRICE | All payments must be in U.S. funds |
| SHIPPING CHARGES IN USA & CANADA | |
| $0.01 to $5.00 | $1.00 | $30.01 to 40.00 | $4.75 |
| $5.01 to $10.00 | $1.75 | $40.01 to 50.00 | $5.75 |
| $10.01 to $20.00 | $2.75 | $50.01 and above | $7.00 |
| $20.01 to $30.00 | $3.75 |
| Total price of merchandise | $ |
| Sales Tax (New York State Residents only) | $ |
| Shipping Charge (see chart) | $ |
| Name | Total Enclosed | H-388 |
| Address | |
| City State Zip | |

MARCH 1988
CLASSIFIED AD ORDER FORM

To run your own classified ad, put one word on each of the lines below and send this form along with your check to:
Hands-on-Electronics Classified Ads, 500-B Bi-County Boulevard, Farmingdale, N.Y. 11735

PLEASE INDICATE in which category of classified advertising you wish your ad to appear. Special headings, there is a surcharge of $10.00.

- Plans/Kits
- Business Opportunities
- For Sale
- Education/instruction
- Wanted
- Satellite Telephone

Special Category: $10.00

PLEASE PRINT EACH WORD SEPARATELY, IN BLOCK LETTERS.

(No refunds or credits for typesetting errors can be made unless you clearly print or type your copy.) Rates indicated are for standard style classified ads only. See below for additional charges for special ads. Minimum: 15 words.

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<td>34 ($47.60)</td>
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We accept MasterCard and Visa for payment of orders. If you wish to use your credit card to pay for your ad in the following additional information (Sorry, no telephone orders can be accepted.):

Card Number

PRINT NAME

SIGNATURE

IF YOU USE A BOX NUMBER YOU MUST INCLUDE YOUR PERMANENT ADDRESS AND PHONE NUMBER FOR OUR FILES. ADS SUBMITTED WITHOUT THIS INFORMATION WILL NOT BE ACCEPTED.

CLASSIFIED COMMERCIAL RATE: (for firms or individuals offering commercial products or services)

$1.40 per word prepaid (no charge for ZIP code). MINIMUM 15 WORDS. 5% discount for same ad in 6 issues within one year; 10% discount for 12 issues within one year if prepaid. NON-COMMERCIAL RATE: (for individuals who want to buy or sell a personal item) $1.15 per word, prepaid...no minimum. ONLY FIRST WORD AND NAME set in bold caps at no extra charge. Additional bold face (not available as all caps) 25c per word additional. Entire ad in boldface. 17c per word. TINT SCREEN BEHIND ENTIRE AD: $1.75 per word. TINT SCREEN BEHIND ENTIRE AD PLUS ALL BOLD FACE AD: $2.05 per word. EXPANDED TYPE AD: $1.85 per word prepaid. Entire ad in boldface, $2.20 per word. TINT SCREEN BEHIND ENTIRE EXPANDED TYPE AD: $3.20 per word. TINT SCREEN BEHIND ENTIRE EXPANDED TYPE AD PLUS ALL BOLD FACE AD: $2.70 per word. DISPLAY ADS: 1" x 2½"—$160.00; 2" x 3½"—$320.00; 3" x 4½"—$480.00. General Information: Frequency rates and prepayment discounts are available. ALL COPY SUBJECT TO PUBLISHERS APPROVAL. ADVERTISEMENTS USING P.O. BOX ADDRESS WILL NOT BE ACCEPTED UNTIL ADVERTISER SUPPLIES PUBLISHER WITH PERMANENT ADDRESS AND PHONE NUMBER. Copy to be in our hands on the 18th of the third month preceding the date of issue (i.e.: Aug. issue copy must be received by May 18th). When normal closing date falls on Saturday, Sunday or Holiday, issue closes on preceding work day. Send for the classified brochure. Circle Number 49 on the Free Information Card.

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TRADESTORS-TUBES:

MF421 $24.00,
MF454 $15.00,
MF454-MF477 $12.00,
MF492 $16.95,
SFR2072 $13.50,
RFPC $18.95,
2SC2879 $25.00,
6LF6-6L06-6J56 $10.95,
959 $15.95,
New Ranger AR3500 all mode 10 meter transceiver $319.
Quantity discounts! Best prices on hard-to-find parts, antennas, mics, power supplies, & equipment! Catalog $1.00 prepaid...or free with orders. RFPC, Box 700, San Marcos, CA 92069.

CRISTAL radio sets, plans, parts, kits. Catalog $1.00. MIDCO, 660 North Dixie Highway, Hollywood, FL 33020.

PROJECT-TV...Convert your TV to project 7 foot picture. Results comparable to $2,500 pro-jectors. Total cost less than $30.00... Plans and 8" lens $21.95. Illustrated information free. MACROCOMA-HG, Washington Crossing, PA 18977.

HANDS-ON MARKETPLACE
Pilot if the use of Code 7500 was intentional. If the pilot answers in the affirmative, or doesn’t reply at all, then public safety agencies will be made aware of the situation. However, communications with the aircraft will remain routine so as not to make the hijacker(s) aware that the alarm has gone out.

Oscar Cooper of Dayton, OH says that he saw a mobile unit used by the USAF’s Office of Special Investigations. There was a VHF-low band whip antenna on the vehicle, and Oscar asks how he can scan in on these people. Nationally, the USAF’s OSI operates in this band on 40.17 and 40.19 MHz. Punch those frequencies into your scanner and see what happens.

Keep those letters coming, gang. We are always on the lookout for your photos, questions, suggestions, frequency information, and just about anything related to scanners and VHF/UHF two-way communications. Write to: Marc Saxon, Six-on On Scanners, Hands-on Electronics, 500-B Bi-County Blvd., Farmingdale, NY 11735.

FRIEDMAN ON COMPUTERS
(Continued from page 87)

Consumer knows nothing from a few scrunched graphics displays, and is perfectly happy because he or she can run “all software.”

Well...Not Really All

Unfortunately, there is no truly “universal” monitor card, just as there is no such thing as a perpetual-motion machine. (Although Hands-on receives at least four projects a year that claim to be perpetual-motion devices. You should see the latest one: It generates two gallons of gasoline for each gallon used to run the machine.)

Some software, usually games, will run only on a true CGA card; they won’t run at all on the Hercules card. Hercules provides a booklet that lists compatibility if you suspect that certain software won’t run. But all the biggies, such as Lotus, will run on the Hercules card.

Best Buy

I have virtually all the various monitor adapters, even a TTL piggybacked with a CGA card, so that they both take up only one slot; and I believe that the Hercules type card is the best buy if you’re trying to keep computer costs at rock-bottom. Naturally, nothing equals having separate adapters and two monitors—one TTL and the other RGB—but when you’ve got the budget stretched tighter than a Soprano’s vocal chord’s as she hits high-C, a Hercules-type card and a TTL monitor is the best dollar/performance value.

So if you’re out buying your first PC-compatible and you’re trying to keep overall costs at rock-bottom, go for a Her-
cules-compatible card and the TTL monitor. Oh, yes! And forget that nonsense about having to pay extra for a printer card if you don’t get a “true” monochrome adapter. The Hercules Graphics Card and all its real clones provide the conventional parallel-printer output connector. In fact, the back end of the Hercules card looks just like a mono card; it has one 9-pin D-connector for the TTL monitor and a 25-pin D-connector for a parallel printer.

RAD HOUND
(Continued from page 48)

within the safe operating range of the tube. Set R6 (the audio volume control) for 90% of maximum and close up the cabinet for Rad Hound’s final checkout.

Operating Rad Hound

Turn on the Hound and give it 10 seconds to charge up. Place the clock’s dial face about one inch away from the G-M tube’s window. You should get a good volume of sound from the loudspeaker if your cabinet is closed up.

To get a working idea about the effective range of radiation, collect a few insulators—sheet of paper, a piece of lead plate, a piece of aluminum; all three should be just big enough to cover that dial face.—and make a laboratory type test. Keep the dial face always separated from the G-M tube face by one inch space.

You should hear a radiation count of about 5000 counts per minute. Now place the paper insulator between the two and notice the slightly reduced radiation count. Replace the paper insulator with the piece of lead plate. Notice how much more reduction is made by the lead plate than was made by the paper.

Then replace the lead plate with the aluminum plate. Notice that the aluminum is not as good as insulator as the lead, but it’s much better than the paper as an insulator. Naturally, the thicker the metal plate the greater its insulating quality.

While you are still testing it, notice that the Inverse Square Law is at work. That means, doubling the distance between the G-M tube and the source of radiation reduces the radiation count by 75% not 50%. So the best insulator that you can have between you and the contamination is plenty of distance.

Go to your local library and get whatever information is available on radiation detection. Learn how to differentiate between alpha rays, beta rays, and gamma rays. Each has its own characteristics and can be separated once you know the proper procedure.

If a nuclear emergency does occur in your area, you can feel secure in knowing that you are prepared to deal with it in your own small way with the Rad Hound radiation detector. Finally, if your set works fine, show it to all the people you live and work with. They should be told what it’s for, how to use it, and to treat it with respect.

Once completed, Rad Hound’s operation can be checked using a source of radiation, such as a luminous clock face.

CHEMICAL POLARITY EXPERIMENT
(Continued from page 41)

spoon of ordinary table salt (sodium chloride) in one ounce of water. Wait a moment for the undissolved materials to settle. Pour the two solutions into a small bottle and shake gently. That gives you two ounces of indicator solution.

The Test

Pour some of the indicator solution onto the paper towel. Use just enough to soak the paper thoroughly. This is a good time to make sure the two contact rods are actually touching the paper; otherwise, the meter will not work. Now connect the terminals to any source of low-voltage direct current, a battery, a power supply, whatever. Keep your eyes on the paper towel. As you watch, the area beneath one of the electrodes will turn a beautiful shade of red. This indicates the negative pole of the power source (Photo 3).

It works like this. When the sodium chloride goes into solution the sodium and chloride ions take up an independent existence. The chloride ions, which carry a negative charge, are attracted to the positive electrode, or anode, where a tiny amount of neutral chloride gas is released. The sodium ions, which carry a positive charge, are attracted to the negative electrode, or cathode. The sodium reacts with the water to form sodium hydroxide. Sodium hydroxide is an alkali and so the phenolphthalein changes color.

The red ring around the contact is proof that the ions have separated and migrated. Notice no change at the other electrode.