

There's no fun in being an mi product. Long before it begins its working life - way back as a design prototype, in fact - it's being vibrated, bumped, sent hot and cold, and subjected to other horrid experiences. And very much the same sort of things have happened to its components long before they got anywhere near it at all.

That's only the start. For instance during
production, an instrument may undergo as many as 60 separate electrical and mechanical inspections adding up to 120 hours on inspection alone - after having endured 500 hours of those shock tactics at design and trial batch stages.

That's typical mi thoroughness for you. In fact, when it comes to reliability you can be quite sure of one thing: at $\mathbf{m i}$ we're not playing at it.

# In: THE PERFECTIONISTS 

# LOW COST TESTERS LEVELL 

PORTABLE INSTRUMENTS

## INSULATION TESTER



A logarithmic scale covering 6 decades is used to display either insulation resistance or leakage current at a fixed stabilised test voltage. The current available is limited to a maximum value of 3 mA for safety and capacitors are automatically discharged when the instrument is switched off or to the CAL condition. The instrument operates from a 9 V internal battery.

## RESISTANCE RANGES

$10 \mathrm{M} \Omega$ to $10 \mathrm{~T} \Omega\left(10^{13} \Omega\right)$ at $250 \mathrm{~V}, 500 \mathrm{~V}, 750 \mathrm{~V}$ and 1 kV
$1 \mathrm{M} \Omega$ to $1 \mathrm{~T} \Omega$ at $25 \mathrm{~V}, 50 \mathrm{~V}$ and 100 V .
$100 \mathrm{k} \Omega$ to $100 \mathrm{G} \Omega$ at $2.5 \mathrm{~V}, 5 \mathrm{~V}$ and 10 V .
$10 \mathrm{k} \Omega$ to $10 \mathrm{G} \Omega$ at 1 V .
Accuracy $\pm 15 \%+800 \Omega$ on 6 decade logarithmic scale.
Accuracy of test voltages $\pm 3 \% \pm 50 \mathrm{mV}$ at scale centre.
Fall of test voltages $<2 \%$ at $10 \mu \mathrm{~A}$ and $<20 \%$ at $100 \mu \mathrm{~A}$.
Short circuit current between $500 \mu \mathrm{~A}$ and 3 mA .

## CURRENT RANGE

100pA to $100 \mu$ A on 6 decade logarithmic scale.
Accuracy of current measurement $\pm 15 \%$ of indicated value. Input voltage drop is approximately 20 mV at $100 \mathrm{pA}, 200 \mathrm{mV}$ at 100 nA and 400 mV at $100 \mu \mathrm{~A}$.
Maximum safe continuous overload is 50 mA .

## MEASUREMENT TIME

$<3$ s for resistance on all ranges relative to CAL position. $<10$ s for resistance of $10 \mathrm{G} \Omega$ across $1 \mu \mathrm{~F}$ on 50 V to 500 V . Discharge time to $1 \%$ is 0.1 s per $\mu \mathrm{F}$ on CAL position.
RECORDER OUTPUT
1 V per decade $\pm 2 \%$ with zero output at scale centre.
Maximum output $\pm 3 \mathrm{~V}$. Output resistance $1 \mathrm{k} \Omega$.

TRANSISTOR TESTER


Tests bipolar transistors, diodes and zener diodes. Measures leakage down to 0.5 nA at 2 V to 150 V . Current gains are checked from $1 \mu \mathrm{~A}$ to 100 mA . Breakdown voltages up to 100 V are measured at $10 \mu \mathrm{~A}, 100 \mu \mathrm{~A}$ and 1 mA . Collector to emitter saturation voltage is measured at $1 \mathrm{~mA}, 10 \mathrm{~mA}, 30 \mathrm{~mA}$ and 100 mA for $I_{\mathrm{C}} / \mathrm{I}_{\mathrm{B}}$ ratios of $10,20,30$. The instrument is powered by a 9 V battery.
TRANSISTOR RANGES (PNP OR NPN)
${ }^{\prime}$ с в ${ }^{\&} I_{\text {E в }}: 10 \mathrm{nA}, 100 \mathrm{nA}, 1 \mu \mathrm{~A}, 10 \mu \mathrm{~A}$ and $100 \mu \mathrm{~A}$ f.s.d. acc. $\pm 2 \%$ f.s.d. $\pm 1 \%$ at voltages of $2 \mathrm{~V}, 5 \mathrm{~V}$ $10 \mathrm{~V}, 20 \mathrm{~V}, 30 \mathrm{~V}, 40 \mathrm{~V}, 50 \mathrm{~V}, 60 \mathrm{~V}, 80 \mathrm{~V}, 100 \mathrm{~V}$, 120 V , and 150 V acc. $\pm 3 \% \pm 100 \mathrm{mV}$ up to $10 \mu \mathrm{~A}$ with fall at $100 \mu \mathrm{~A}<5 \%+250 \mathrm{mV}$.
BV CBO: $\quad 10$ or 100 V f.s.d.acc $\pm 2 \%$ f.s.d. $\pm 1 \%$ at currents of $10 \mu \mathrm{~A}, 100 \mu \mathrm{~A}$ and $1 \mathrm{~mA} \pm 20 \%$
$\mathrm{I}_{\mathrm{B}}: \quad 10 \mathrm{nA}, 100 \mathrm{nA}, 1 \mu \mathrm{~A} \ldots 10 \mathrm{~mA}$ f.s.d. acc. $\pm 2 \%$ f.s.d. $\pm 1 \%$ at fixed $I_{E}$ of $1 \mu \mathrm{~A}, 10 \mu \mathrm{~A}, 100 \mu \mathrm{~A}$, $1 \mathrm{~mA}, 10 \mathrm{~mA}, 30 \mathrm{~mA}$, and $100 \mathrm{~mA} \mathrm{acc} . \pm 1 \%$.
$h_{\text {FE }}: \quad 3$ inverse scales of 2000 to 100, 400 to 30 and 100 to 10 convert $I_{B}$ into $h_{\text {FE }}$ readings.
$V_{B E} \quad 1 \mathrm{Vf.s.d.acc} . \pm 20 \mathrm{mV}$ measured at conditions on $h_{\text {FE }}$ test.
$V_{\text {CE (sat) }} \quad 1 \mathrm{~V} . \mathrm{s.d}$. acc. $\pm 20 \mathrm{mV}$ at collector currents of $1 \mathrm{~mA}, 10 \mathrm{~mA}, 30 \mathrm{~mA}$ and 100 mA with $\mathrm{I}_{\mathrm{C}} / \mathrm{I}_{\mathrm{B}}$ selected at 10,20 or 30 acc. $\pm 20 \%$.

## DIODE \& ZENER DIODE RANGES

${ }^{1} \mathrm{DR}$ :
As $\left.\right|_{\text {E b }}$ transistor ranges
$V_{Z}: \quad B r e a k d o w n$ ranges as $B V_{C B O}$ for transistors.
$V_{D F}: \quad 1 \mathrm{~V}$ f.s.d. acc. $\pm 20 \mathrm{mV}$ at $\mathrm{I}_{\mathrm{DF}}$ of $1 \mu \mathrm{~A}, 10 \mu \mathrm{~A}$, $100 \mu \mathrm{~A}, 1 \mathrm{~mA}, 10 \mathrm{~mA}, 30 \mathrm{~mA}$ and 100 mA .

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# wireless world 

# Electronics, Television, Radio, Audio 

## APRIL 1975 Vol 81 No 1472

## Contents

151 Outlook for cable television
152 Using ferrite pot-cores by D. E. O'N. Waddington
155 Sixty years ago
155 April meetings
156 Artificial vision progresses by T. E. Ivall
158 Announcements
159 News of the month
TV standards converter
IEE on engineering profession
First production c.c.d.
16175 years of magnetic recording-2 by Basil Lane
165 Letters to the editor
169 Noise-confusion in more ways than one-2 by K. L. Smith
173 HF predictions
173 Wireless World noise reducer announcement
174 Space news
175 Circuit ideas
Deflection amplifier for 'scopes
Low battery voltage indicator
Sinewave oscillator uses c.d.a
177 An aerial rotator servo by D. J. Telfer
182 Loudspeaker developments
183 Circards 21: voltage to frequency converters by J. Carruthers, J. H. Evans, J. Kinsler and P. Williams

185 Vision cassette recorders
190 Literature received
191 Transistor-aided ignition by G. F. Nudd
192 Project: National Electronics Council Link scheme
193 World of amateur radio
194 New products
198 Real and imaginary by "Vector"
a80 APPOINTMENTS VACANT
al04 INDEX TO ADVERTISERS

[^1]

This month's cover picture, showing part of an uncompleted implant for the human head and a thick film implant receiver, made by Eric Sayer, introduces the article on artificial vision on p. 156 of this issue. (Photographer Paul Brierley)

## IN OUR NEXT ISSUE

## Wireless World noise reducer

Constructional project based on the Dolby principle for which we are supplying a kit of parts (see page 173)

## Build an oscilloscope

Professional standard design for home construction with $50 \mathrm{MHz} \quad Y$-amplifier bandwidth and extensive facilities

## Display devices survey

Review of techniques used in alpha-numeric display devices and characteristics of types now on the market

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## Outlook for cable television

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It's a pity that the cable television companies' experiments with local origination of programmes, at Bristol, Greenwich, Sheffield, Swindon and Wellingborough, have not proved very successful. Greenwich is virtually closed down (except for three hours at weekends when the station is run by outside volunteers), other stations are facing criticism about uninteresting programmes and some are doubtful whether they will be able to carry on after 1976. This may prove that the companies are not very good at producing programmes, that the type of material they are providing is not wanted or that they have insufficient finance from their private sources to produce the programmes they would like. It would be a pity, though, if this experience threw doubt on the whole principle of originating programmes locally and distributing them on cable, for this is what the cable television companies are well fitted technically to do, especially if in the future a large number of programmes and/or interactive information services is required.

Whereas in broadcasting the number of programmes that can be transmitted is limited by the amount of electromagnetic spectrum available, there is theoretically no limit, from an engineering point of view, to the number of programmes/information services that can be distributed by cable. In practice, judging from recent developments such as the "dial-a-programme" system and experimental work on using bundles of optical fibres for local distribution, it should be possible to bring $30-40$ interactive channels into a household. The fact that programme material is originated locally does not mean that it has to be about local affairs, in the manner of a local newspaper. By analogy with the education service it can be material of general or national interest but with a "mix" adjusted to local circumstances and demand.

But now into this scene steps the Post Office with a claim that it should take over the whole of cable television. In its evidence to the Annan Committee on the future of broadcasting it says "The transmission of information is Post Office business . . ." and "If and when there is an increase in television broadcasting, leading to a wide-scale requirement for cable-TV networks, the Post Office is, we believe, the organization to provide such networks on a national basis to meet the demands both for enhanced television and for the broadening range of telecommunication services-voice, vision and data-that we foresee".

Of course the Post Office is the right organization to handle the large-scale transmission of information-probably by integrated digital systems in the future-on trunk routes between cities. And this rightly includes the long distance transmission of television signals. But this doesn't mean that the Post Office is necessarily the best organization to handle local distribution. It certainly has extensive plans but lacks the experience of the cable television companies (at present it provides networks in six new towns) and as a public corporation it does not have the spur of competition that gives a keen assessment of the market and often leads to valuable technical developments.

The cable companies have made a considerable investment in their networks. This is not to say they should necessarily be guaranteed a good return-after all it was a risk they took. But this is also a national investment and as such should be taken into account in any plans for the future.

# Using ferrite pot-cores 

# Basic inductor design for the development engineer 

by D. E. O'N. Waddington, M.I.E.R.E

When, as a schoolboy, I became interested in radio, I blamed all my failures to persuade crystal sets to work on the coils. As I lived about 50 miles from the nearest transmitter which radiated a meagre 2 kW , I now feel that this was a bit unjust to the coils. Nevertheless, coil design remained a bogey for many years. One of the main reasons was that so many variables are involved that the design is always complicated. To design a single layer coil with a specified inductance value, one has to assume diameter, winding length, wire diameter and winding pitch before starting on the calculation of the number of turns. The odds are that the first try will produce a ridiculous answer and it will probably be necessary to try several times before a practical result is achieved. Even at the end of all this, there will be a nagging doubt as to whether the result is correct or not! Multi-layer coils are even worse, if possible, as the dimensions predicted by theory are seldom realizable in practice. In fact the only method appears to be to take an "educated" guess at the coil design and to check by calculation. Thus it was with a great sense of relief that I learned to use ferrite pot cores. At last here was an inductor which could be designed.(most of the time!).

I will start with a short description of the core material and manufacture. It is well known that placing a magnetic core inside a coil increases its self-inductance. However, the alternating magnetic field causes eddy currents to flow within the core absorbing energy from it and reducing the effective $Q$ of the coil. This loss occurs mainly because of the low resistivity of the core material. It also increases with frequency. In transformers it is usual to reduce this loss by laminating the core material and insulating each lamination from its neighbour. The thinner the laminations, the lower the eddy current loss, and the higher the frequency to which the core may be operated. However, a practical limit is reached very quickly so that this technique, while giving a substantial improvement, does not provide the answer for radio-frequency coils

One method of overcoming the limitations of laminations is to use a powdered iron dust core in which finely divided
particles of iron, or other ferro-magnetic material, are suspended in an insulating medium and moulded into a core. This effectively insulates the particles from each other and reduces eddy current flow but, at the same time it reduces the effective permeability of the core to ten or less. Nevertheless, these iron dust cores are very useful at radio frequencies as not only do they increase the effective inductance of coils, but, when used in cup form, they tend to confine the magnetic fields within the coil, providing a measure of screening. For high frequency work iron dust cores are superior to ferrites both in performance and cost. The design methods which I will be describing can also be applied to iron dust cores.


Fig. 1. Cross section of a typical pot cure.


Fig. 2. Variation of permeability with temperature for a low-frequency ferrite material.

Unlike iron dust cores, ferrite cores are primarily made of non-conducting materials, which belong to the family of ferrites. The ferrites are non-metallic refractory materials composed of the oxides of iron and other metals, usually cobalt, copper, manganese, magnesium, nickel or zinc. The most important ferrites for pot cores are manganese zinc and nickel zinc-ferrite ${ }^{1}$. In manufacture the correct proportions of the relevant oxides are milled together so that they are thoroughly mixed. They are then moulded into the desired shape in a press and fired at a temperature in the range from $1000^{\circ} \mathrm{C}$ to $1300^{\circ} \mathrm{C}$. During this process chemical reactions occur and when the resultant cores are cooled to room temperature, they are hard and brittle. This firing or sintering process is a very critical one as the properties of the finished core depend largely upon the precise firing temperature and the time for which it is "cooked". The cores shrink appreciably (between 20 and $25 \%$ ) during the firing process and, as the ferrite is very hard to machine, it is also essential that the density of the moulded core must be correct before firing as subsequent adjustment would be very costly. The cores used for inductors are said to be "soft". In this context soft means that the core does not remain magnetized to any appreciable extent after a magnetizing field has been applied. This is analogous to "soft" iron cores recommended in text books for electric bells, etcetera.

For use in inductors, the cores are usually made in the form of cups as shown in Fig. 1. The mating surfaces are ground smooth and polished so that the air gap is reduced to a minimum. The effective permeability of the basic core material will be of the order of 2000 for low frequency ferrites, reducing to 100 for high frequencies. This basic permeability is very sensitive to temperature variations, the degree of sensitivity depending upon the composition of the ferrite. Normally the permeability increases fairly steadily with temperature until it suddenly falls off very rapidly to the Curie point (see Fig. 2). Curie point is generally defined as the temperature at which the permeability has fallen to $10 \%$ of its maximum value and lies in the range from 150 to $200^{\circ} \mathrm{C}$ for
most ferrites although some ferrites have Curie points as high as $500^{\circ} \mathrm{C}$. For inductors, the cores are usually modified by grinding the centre spigot so that there is an air gap in the magnetic path. The working permeability of the finished core depends upon the length of this gap which also confers two very desirable properties. Firstly, the temperature coefficient is greatly reduced and now depends to a greater degree upon the physical dimensions of the core. Thus it is possible to specify the temperature coefficients of various cores with fair accuracy. Secondly, by adjusting the position of a ferrite slug so that it "bridges" the air gap, it is possible to adjust the working permeability of the core and hence the inductance of a coil wound on it. As would be expected, cores with small gaps (high permeability) have less adjustment range than those with large gaps although neither has a very large range ( $5 \%$ to $25 \%$ ). In early cores. the adjuster was not a built-in feature and it was necessary for the user to grind the core himself to adjust the inductance. This was done by rubbing the core on fine emery paper taking great care to keep the surfaces flat. I mention this method as it still has its uses when an inductor is just out of the adjuster range. However I would not recommend its use as cores are easily cracked by the overheating which can be produced by too vigorous rubbing. For repeatable and stable performance. it is essential that the two halves of the core are adequately clamped together. Most manufacturers supply excellent clamping systems although gluing, with Araldite for example. is a very effective assembly method. Cores are usually made in matched pairs so it is best to keep them in pairs. Sorting is both tedious and frustrating.

## Core losses

The losses which occur in ferrite cores are of three main types; hysteresis, eddycurrent and residual.

Hysteresis loss. This is usually very small compared with the other losses and, at low drive levels, it may be ignored. At high signal levels, however, it can contribute an undesirable effect in the form of nonlinear distortion, mainly third order. The degree of distortion depends upon the flux density and can be predicted by calculation ${ }^{1}$. Normally this effect is of little significance but, in some audio applications. it may become important. The cure is

Fig. 3. To work out the number of turns required to give a specified inductance value, lay a ruler across the abac connecting the required inductance (on scale A) with the $A_{L}$ or $\alpha$ (on scale $C$ ) of the core used. The number of turns is read from scale B.
Note. For micro-henries use the right-hand calibration of scales $A$ and B. For millihenries use the left-hand calibration of scales $A$ and $B$. For henries use the same scales as for micro-henries but multiply the number of turns by 1000 .

## Abac to determine number of turns from core data


either to run the core at a lower level or to use a larger core (which amounts to the same thing!).

Eddy-current loss. This depends mainly on the resistivity of the core material. Thus, in most ferrites this loss is small so that it is normally lumped in with the residual losses. There are exceptions where the eddy current losses "resonate" with the dimensions of the core at high frequencies ${ }^{2}$. The discussion of them, however, is beyond the scope of this article.

Residual losses. These depend upon the composition of the ferrite and will vary with the different grades. These losses are frequency dependent, usually increasing relatively slowly up to a "critical" frequency after which they increase drastically. Thus the grade of ferrite determines the high frequency operating limit.

## Coil losses

These are far more severe than in air-cored coils because, in addition to skin effect, there are eddy-current losses in the conductors caused by proximity effects. This means that the $Q$ of the inductor will also depend upon the type of wire used as well as the core losses. In general, solid conductors give a maximum $Q$ at a very much lower frequency than that for maximum $Q$ with stranded wire and the $Q$ will also be lower. One manufacturer quotes the following:-

$$
\begin{array}{ll}
\text { Solid wire } & Q_{\max }=200 \mathrm{at} \\
& 20 \mathrm{kHz}(10-100 \mathrm{mH}) \\
\text { Stranded } .06 \mathrm{~mm} & Q_{\max }=600 \mathrm{at} \\
& 150 \mathrm{kHz}(.2-1 \mathrm{mH}) \\
\text { Stranded } .04 \mathrm{~mm} & Q_{\max }=700 \mathrm{at} \\
& 200 \mathrm{kHz}(.2-1 \mathrm{mH}
\end{array}
$$

This information is usually included in the manufacturers' data books in the form of typical ISO- $Q$ curves although it is sometimes in tabular form. The word "typical" seems to have the meaning ascribed to it by a cynical engineer; namely "It has actually been achieved once!" In all fairness, however, the quoted $Q$ can be attained under ideal conditions with all details fully under control. However, even if the final $Q$ is less than that predicted, it should be far higher than could have been obtained using an air-cored coil and, of course, the dimensions of the coil will be considerably smaller.

## Inductor design

The calculation of the number of turns necessary to achieve a particular inductance value is very easy as manufacturers quote either $A_{L}$ (induction factor) or $\alpha$ (turns factor). These can be defined as follows:-
$A_{L}$ (induction factor)-The self-inductance, in nano-henries, that a coil wound on the core should have if it consisted of a single turn.
$A_{L}=\frac{L}{N^{2}}$ or $N=\sqrt{\frac{L}{A_{L}}}$
$L$ is in nano-henries
$N$ is the number of turns.
The term $\alpha$ (sometimes $C$ or $K$ ) is the turns factor or the number of turns required for a coil wound on the core to give an inductance of 1 milli-henry.


Fig. 4. This family of curves shows how the induction factor varies with the "fullness" of the available winding space for an 18 mm pot core. Other core sizes will exhibit similar variations.


Fig. 5(a). These curves apply to round cores conforming to B.S. 4061 range 2 or I.E.C. Pub. 133.


Fig. 5(b). These curves apply to R.M. (rectangular module) cores.
$\alpha=\frac{N}{\sqrt{L}}$ or $N=\alpha \sqrt{L}$

## $L$ is in milli-henries

$N$ is the number of turns
e.g., required-a 9 mH inductor. The core selected has an $A_{L}$ of 400 or $\alpha$ of 50 . $N=\sqrt{9 \times 10^{6} / 400}=3 \times 10^{3} / 20=150$ turns or $N=50 \times \sqrt{9}=50 \times 3=150$ turns The abac shown in Fig. 3 provides a simple alternative method of determining the number of turns. Lay a ruler across the abac connecting the $A_{L}$ or $\alpha$ on the righthand scale with the required inductance value on the left-hand scale and read the number of turns from the centre scale.

Normally the winding factors given in the manufacturer's data will refer to a coil wound so that it fills a predetermined percentage of the winding space and it may be necessary to adjust the number of turns slightly depending upon whether the bobbin is fuller or emptier. Fig. 3 shows the sort of variations which can be expected with a typical core. In general it will be seen that, with high permeability (i.e., "small gap"), the degree of "fullness" of the bobbin has very little effect upon the turns factor. On the other hand, lower permeability cores (i.e., "large gap"), are more affected by the "fullness". This effect is caused by fringing of the magnetic field in the gap. It is good practice however to choose a wire gauge which fills the winding space as completely as possible. This gives the lowest d.c. resistance together with the highest $Q$ value. Most core manufacturers give tables or charts showing the numbers of turns which will fill the various bobbins. Now that there has been a degree of standardization of core sizes (British Standard B.S. 4061 range 2 and International I.E.C. Pub.133) it has been possible to prepare some winding charts which have fairly universal application. Fig. 5(a) shows winding data for the round cores and Fig.5(b) gives data for r.m. (rectangular module) cores. The numbers of turns which should fit the cores are nominal so that it is generally safer to use a slightly thinner gauge than suggested by the chart.

I feel that a word of warning is necessary here. As George Orwell says, "All animals are equal but some are more equal than others." This comment could well be applied to ferrite pot cores. So far the standardization only goes as far as specifying the dimensions of the cores and formers and $A_{L}$. Nothing is said of clamping systems, termination methods or adjusters so far as I know. At least, if it is specified, it is frequently ignored. In general British manufacturers produce reasonably compatible systems but the same cannot be said for all the imported products. This means that it is necessary to study alternative core types very carefully before accepting them as equivalents.

Earlier in this article I referred to the temperature coefficient of the permeability. Obviously this will affect the stability of the finished inductor. In practice there are one or two more points to be watched if the best stability is to be obtained. Movement of the coil in the core will change the
inductance slightly so the coil should be locked in position. Similarly movement of the individual turns of the coil can also introduce instability. This makes it desirable to impregnate the coil. Actually, if moisture penetrates the coil it can degrade the $Q$ so there is a second reason for impregnation.

If the impregnation is carried out with the coil fitted to the core care must be taken that the adjuster system is kept clear. While moisture does not affect the permeability of the basic core to any measurable extent, it can affect the adjuster system so that it is wise to check this point. Personally I have found that the adjusters which consist of a ferrite tube fitted on a plastic sleeve with a hole up their centres to screw onto a brass screw are the best. A further point to watch is a phenomenon known as "disaccommodation". This is a temporary change in permeability which occurs if the core is subjected to a thermal or mechanical shock. However, provided that final adjustment of the inductance is not carried out until 24 hours after the shock, this effect should not prove troublesome.

In conclusion I would like to thank Mullard Ltd for permission to reproduce illustrations of their cores and graphs.

## References

1. Snelling, E. C. ${ }^{-}$Soft ferrites, properties and applications", Butterworths, London 1969.
2. Mullard Ferroxcube. Mullard Components Division, May 1955.

## Sixty Years Ago

In 1915, spy. scares were getting well into their stride and the still-new invention of "wireless" was fuel to the fire. Suitcase transmitters were still in the future, however, and it seems that people's imaginations tended to become a little over-heated. A note in our April 1915 issue commented: "Mr Charles R. Gibson has been contributing long articles recently to the Glasgow Herald on the present use of wireless by the belligerents, and in the course of one of them tells an amusing story which, according to the writer, was repeated to him with portentous seriousness as an incident of the greatest gravity which had recently come under the narrator's personal observation:
'Two German workmen had been arrested as spies, and there had been discovered, hidden beneath the hearthstone of the kitchen in their two-roomed tenement house, a complete wireless installation capable of transmitting messages to Berlin.'

Mr Gibson comments that it is possible to send wireless messages as far as from here to Berlin, but not with apparatus that can be stowed away beneath a kitchen hearthstone, or even contained in a large room."

## Meetings

## LONDON

7th. IEE-"Mechanical shock protection in the design of electrical equipment" by L. A. Ward at 17.30 at Savoy Pl., WC2.

8th. IEE-Discussion on "Microprocessors versus programmable logic arrays" at 17.30 at Savoy Pl., WC2.

9th. IERE-Colloquium on "Radar and associated systems for vehicle guidance" at 14.00 at 9 Bedford Sq., WC 1 .

9th. IEE-Discussion on "Is there a future for pointer instruments?" opened by G. D. H. Keen, Dr R. B. D. Knight and A. H. Silcocks at 17.30 at Savoy Pl., WC2.

10th. IERE/IEE-Colloquium on "Computers in transport" at 10.00 at 9 Bedford Sq., WC I .

10th. IEE - "The work of the House of Commons Select Committee on Science and Technology" by Airey Neave followed by discussion at 17.30 at Savoy Pl., WC2.

10th. RST-The Fleming Memorial lecture "Television: parliament and the people" at 19.00 at the Royal Institution, Albemarle St., W I.

1 th. IEE--Colloquium on "Spectrum allocation management and engineering in radio communication" at 10.00 at Savoy PI., WC2.

11th. IEE-Colloquium on "Innovatory ideas in energy generation and conversion" at 10.30 at Savoy Pl., WC2.

14th. IEE-Colloquium on "Electronic counter measures--components and systems" at 10.30 at Savoy PI., WC2.

15th. IEE-"Replaceable control systems" by B. Welch at 17.30 at Savoy Pl., WC2.

15th. AES-"Speech perception and speech synthesis" by D. B. Fry at 19.15 at the IEE, Savoy Pl.. WC2.

16th. IEE-Colloquium on "Hardware and software aspects of parallel processors" at 11.00 at Savoy P1., WC2.

16th. BKSTS-"Portable power systems for cinematography" by V. F. Saunders, R. W. Scarr and F. R. Cloke at 19.30 at Thames Television Theatre, 308-316 Euston Rd., NW 1.

17th. IEE-Colloquium on "Techniques for designing for reliability" at 10.30 at Savoy Pl., WC2.

17th. IEE-"Engineering management and the professional unions" by A. Grosschalk at 17.30 at Savoy Pl., WC2.

18th. IEE-Colloquium on "Alpha numeric display devices" at 14.30 at Savoy PI., WC2.
23rd. IERE-Colloquium on "Recent developments in turntable design" at 10.00 at 9 Bedford Sq., WC1.
23rd. BKSTS - "The factors affecting the image quality of 16 mm film for television" by Arthur Branson at 19.30 at Thames Television Theatre, 308-316 Euston Rd., NW 1 .
24th. RTS-A.G.M. followed by "Television special effects using electronics and photography" by A. B. Palmer at 18.30 at London Weekend Television South Bank TV Centre, Upper Ground, SE1.
25th. IEE-Colloquium on "Digital microwave relay systems above 10 GHz " at 14.30 at Savoy Pl., WC2.

28th. IEE-Colloquium on "Message switching" at 10.30 at Savoy PI., WC2.

30th. IERE-Colloquium on "Trends in testing telecommunications materials" at 10.00 at 9 Bedford Sq., WC 1 .

30th. IEE-"Artificial vision" by P. E. K. Donaldson at 17.30 at Savoy PI., WC2.

## BIRMINGHAM

9th. IEETE-EASCON 75 on "The Partnership? training-education" at 10.30 at the City of Birmingham Polytechnic.

## EDINBURGH

28th. IEETE-"Electronics versus the criminal" by J. S. T. Charters at 19.00 at Carlton Hotel, North Bridge.

## GLASGOW

29th. IEETE-"Electronics versus the criminnal" at 19.00 at Royal Stuart Hotel,

GRAVESEND
3rd. IERE-A.G.M. followed by "The application of digital computers to radar and navigation at sea" by Bruce Williams at 19.00 at the Tollgate Motel, Watling Street.

## LIVERPOOL

8th. IEETE-"Automatic flying controls" by D. I. Jackson at 19.30 at MANWEB Social Club, Thingwall Road.

## MANCHESTER

10th. IEETE-"Intruder alarms" by E. Tanham at 19.30 at UMIST, Reynold Building, Sackville St.

## READING

8th. IERE-"Project management" by Dr. I. Maddock at 19.45 at the J. J. Thomson Physical Laboratory, University of Reading, Whiteknights Park.

## SWINDON

29th. IEETE-"Technician engineers and tech-nicians-their role, status and qualifications" (speaker from IEETE secretariat) at 19.30 at The College, Regent Circus.

Tickets are required for some meetings: readers are advised therefore to communicate with the society concerned.

## Literafure Received

## ACTIVE DEVICES

We have received a 24 -page catalogue giving specification of the Siemens range of charge storage varactors, varactor diodes for frequency conversion. tuning varactor, PIN, Schottky, tunnel, backward. IMPATT diodes and Gunn elements. Siemens Ltd Great West House, Great West Road, Brentford. Middx TW8 9DG

WW401

Also available from Siemens is an applications booklet on the subject of surge voltage protection . . . . . . . . . . . . . . . . . . . . . . . . . . . WW402

A price list and complete set of data sheets describing the Monolthic Memories Inc. range of semiconductor memories is availabe from Memory Devices Ltd, Central Avenue, East Molesey, Surrey KT8 0SN . . . . . . . . . . . . . . . . . . . . . . . . . . . . WW403

Books on c.m.o.s. logic devices and applications by Motorola are available from Jermyn. The McMOS Handbook (applications) is available free with each order for the McMOS Data Book, which is priced at $£ 2.50$. McMOS is the Motorola name for c.m.o.s. Jermyn Distribution, Sevenoaks, Kent.

The seven-volume set of RCA Data Books for 1975 is now available. The complete range of RCA semiconductors is described, including diodes, transistors, integrated circuits of all kinds, thyristors and microwave devices. Each volume costs ${ }^{*} £ 1.80$, the price for the complete set being $£ 8$. RCA Ltd, Solid-State-Europe. Sunbury-on-Thames, Middx.

We have received from OVUM a bibliography of charge-coupled devices, containing abstracts on general information, theory, technology, bucket brigades and several other subjects. The book is well-indexed and is entitled "International abstracts on charge-coupled devices 1970-74". It is available from Ovum Ltd, 22 Grays Inn Road, London WC 1 at $£ 30$.

Many applications of a variety of semiconductor devices are described in a new book by Siemens Both discrete semiconductors and integrated circuits, both digital and analogue, are dealt with in applications from industrial control to audio. The book is available free of charge from Marketing Services Department, Siemens Ltd, Great West House, Great West Road, Brentford, Middx. . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . WW404

## Artificial vision progresses

# Improved design of microelectronic implant giving more stimulation points on the brain and greater reliability 

by $T$. E. Ivall<br>Editor, Wireless World

Work by the Medical Research Council's Neurological Prostheses Unit in providing some degree of vision for blind people by means of microelectronic implants in the head was described in our May, 1971 issue*. Two implants have been made and fitted, one in a female patient in 1968 and one in a male patient in 1972, and both have given encouraging results. Since 1972 research and development led by P. E. K. Donaldson has been continued with the object of improving the design of the implant, notably to increase the number of electrical stimulation points on the visual cortex of the brain and to make the implanted electronic devices neater and more reliable.

The principle of the MRC's visual prosthesis is directly to stimulate a large number of points on the visual cortex of the brain of a person who has become blind, for example, through damage to the optic nerve ${ }^{1,2,3}$. As a result the patient "sees" spots of light, called phosphenes, which are fixed in the visual field. By suitably organizing the electrical stimula-

[^3]tion these phosphenes can be arranged into meaningful patterns for the patient, such as letters of the alphabet or Braille characters. The stimulation is applied by $500 \mu \mathrm{~s}$ pulses of current fed to electrodes mounted in two flexible silicone rubber cups which fit round the two occipital lobes of the brain. The stimulating pulses come from microelectronic inductive-loop receivers and logic units which are implanted, as packages in a silicone rubber "cap", between the skull and the scalp (Fig. 1). No implanted battery is needed as all the required power comes from the external bank of inductive-loop transmitters (mounted in a hat-shaped shell similar to a hair-drying hood) which activate the implanted receivers.

The d.c. outputs of the implanted receivers are electrically arranged to form a matrix of rows and columns, so that when a particular "row receiver" and a particular "column receiver" are energized simultaneously by their transmitters their d.c. outputs activate a particular AND gate (at the "intersection" of that row and that column). The d.c. output of the AND gate then provides the stimulating pulse for a particular electrode on the visual cortex.


Thus if there are $x$ row receivers and $y$ column receivers in the implant it is possible to identify $x y$ unique pairs of receivers and therefore to have $x y$ stimulating electrodes. The external transmitters are arranged in a corresponding matrix of rows and columns. Row-transmitters generate $500 \mu \mathrm{~s}$ pulses of r.f. at 10 MHz while alternate column-transmitters give $500 \mu$ s pulses at 8 MHz and 6 MHz (this arrangement of different frequencies for adjacent column-transmitters being a means of avoiding crosstalk).

In the implant described in our May, 1971 issue there were nine row receivers and 20 column receivers, giving 180 stimulation points. It was designed for a 64 -year-old male patient who had been blind for 30 years with retinitis pigmentosa. When a dummy of the device was handed over to the neurosurgeon who was to perform the implantation operation he said it was too big and he would have difficulty in closing the scalp over it. Would the engineers please think again? It was therefore decided to reduce the number of row receivers from nine to five and the number of column receivers from 20 to 15 . In addition it was decided to eliminate the $1.0 \mu \mathrm{~F}$ tantalum capacitors in series with all but three of the AND gate outputs (see May, 1971 issue, p.216). Capacitance is needed in these outputs to keep the mean stimulating current zero and so avoid electrolysis at the electrodes and consequent tissue damage, but it is possible to rely on the capacitance-like properties of the electrode-tissue interface. (More about this later.)

A dummy $5 \times 15$ device was accepted as satisfactory by the surgeon in July, 1971 and the actual $5 \times 15$ implant, giving 75 stimulation points instead of the 180 originally intended, was implanted into the patient on February 4, 1972. As a result of testing ${ }^{4}$ it was found that the

Fig. I. Completed second implant, showing the stimulating electrode assembly at the end of its cable (bottom left), before surgical implantation in the head of a 64-year-old male patient. See also front cover.
patient could in fact "see" 55 phosphenes of the theoretically possible 75. These, however, were disappointing for pattern organization purposes because the phosphenes were larger than those experienced by the first patient and when pairs were elicited simultaneously they tended to fuse together into a single, bigger phosphene. Finally six good phosphenes-bright and clearly defined-were chosen lying in two vertical columns of three, the format for Braille characters, and the subsequent tests on this patient were mainly confined to the reading of Braille text fed to the transmitters character by character by a punched tape apparatus.

Experience gained from this second implant clearly showed that it was desirable to provide many more stimulation points on the visual cortex to make possible more detailed patterns of phosphenes and to allow for possible failures of stimulation points after implantation. Moreover it was believed that the patient had himself inadvertently put some electrodes out of action simply by scratching and bumping his head, and this suggested some mechanical fragility in the wiring between the microelectronic packages. It was therefore decided to produce a third design of implant which would overcome these problems.

To provide more stimulation points it is of course necessary to put more receivers and AND gates into the implant, but in order to keep the implant size down to that required by the neurosurgeon the packages obviously have to be made smaller. The most bulky packages in the second implant were the hermetically sealed logic units containing transistor and diode AND gates. These logic units are also the most environmentally sensitive of the packages-the environment being a warm saline "mist" produced by the body fluid-hence the need for particular care in sealing them. It was decided, however, to do away with the hermetic sealing, which required rigid ceramic packages measuring $29 \mathrm{~mm} \times 20 \mathrm{~mm}$ with projecting connection tags. Instead, after experiments with various materials, straightforward encapsulation with silicone rubber adhesive was chosen. At first sight this seems a very unsuitable process, for silicone rubber has a high permeability and a low water absorption, and it must therefore transmit water vapour rapidly. But in fact hybrid electronic components are not very susceptible to water vapour. Thick film resistors, chip capacitors, cross-over glazes, conductors and passivated semiconductor devices can be operated successfully in the presence of such vapour. On the other hand, water as

Fig. 2. Thick-film microelectronic circuit of the logic package, carrying, on a $26 \mathrm{~mm} \times 6 \mathrm{~mm}$ substrate, transistors, diodes and resistors for 19 AND gates (see Fig. 3).

Fig. 3. One of the 16 logic packages fed by (left) one of the 16 row-receiver coils. Outputs from the 19 column-receivers are fed into the diodes. All resistors are $10 k \Omega$.
liquid affects the operation of the circuit in two ways: it can provide spurious conduction paths which cause malfunction; and it may allow electrolysis to occur, filling up the package with electrolytic debris, which causes further shorts, and allowing the generation of gas under enormous pressure. Water as liquid will condense from water vapour in any voids which may be present in the encapsulating material, and it is the voids at the interface between the encapsulant and the electronics which cause the trouble.

The requirements of the encapsulant in the implant packages are therefore not only the usual one, that it shall penetrate the surface convolutions of the microelectronic circuitry everywhere so that no voids are left when the device is first made, but also that it shall discourage the formation of new voids subsequently. This means that the adhesion of the encapsulant must be good, and that the encapsulant
should be a rubber and not a resin, so that small strains set up at the encapsulantelectronics interface (as a result of, say, slight swelling of the encapsulant because of its water load) do not nevertheless set up large stresses which break the adhesive bonds. Fortunately silicon rubber adhesives are both rubbers and very adhesive in the presence of water, and this is why they work. The conclusions of the MRC workers are, therefore, that moistureprotecting encapsulants work not because they are in any sense a wall, but because they displace water as liquid from the surface of the microelectronics.

Using this encapsulation technique, logic packages containing 19 AND gates have been produced measuring only 26 mm (long) by 7 mm (wide) by 3 mm (deep) as shown in Fig. 2. Unlike the previous hermetically sealed packages they have flying leads. The hybrid microelectronics forming the circuit (Fig. 3) are laid on a

$25 \mathrm{~mm} \times 6 \mathrm{~mm}$ ceramic substrate and the thick film parts of the circuit are successive layers of resistor material, palladium silver conducting pads and cross-overs, glass for insulation, and gold for transistor connecting pads and the two bus-bars shown in Fig. 3. The transistors and diodes attached to this thick film circuit are beam lead devices, and have a silicon nitride impermeable skin put over them.

Although there are fewer AND gates in this new logic package (19 as against the previous 20) the smaller size of the package allows more logic units to be used and in fact the third implant will contain 16 of these units (instead of nine). These will be fed with d.c. pulses from 16 row receivers (one is shown in Fig. 3) and 19 column receivers, giving $16 \times 19$ unique pairs of receivers on the matrix and therefore allowing 304 stimulating electrodes. Thus, relative to the second implant with its 75 outputs, the number of stimulation points will be quadrupled in this new implant.

The row and column receivers will be encapsulated in the same silicone rubber adhesive as is used for the logic packages. Samples of units made in this way have been tested by operating them under normal electrical conditions while immersed in a warm saline bath ( $1 \%$ sodium chloride solution at $50^{\circ} \mathrm{C}$--a more literal meaning for "soak testing"!-and in many months of continuous testing no insulation failures have been detected.

The thick film receiver circuits themselves have been modified to make them smaller and safer. Most notably the inductive pick-up coils are now made in thick-film form (see illustration on front cover) instead of being coils of copper wire. The coiled conductor deposited on the ceramic substrate is of platinum, gold plated to bring the coil resistance to below 2 ohms. This has the advantage of allowing a thinner receiver package and avoiding the use of copper, which could be chemically harmful to the patient. Moreover it allows more stringent cleaning methods to be used on the circuit. The tuning capacitors are chip devices while
the detector rectifiers are passivated diodes.

The final space-saving expedient is to do away with, as mentioned above, the $1.0 \mu \mathrm{~F}$ wet tantalum electrolytic capacitors connected between the outputs of the AND gates and the stimulating electrodes. In the project described in our May, 1971 issue these were housed in packages each containing 15 capacitors. The alternative, as explained, is to combine the capacitor and stimulating electrode into one by coating the electrodes with a suitable dielectric layer. Thus the metal electrode forms one plate of the capacitor while the surrounding biological tissue forms the other plate. Experiments at the MRC Unit have in fact shown that tantalum electrodes coated with tantalum pentoxide can perform stably as capacitor anodes when implanted into biological tissue ${ }^{5}$. Capacitances and leakages (typically $1 \mu \mathrm{~A}$ at 5 V in a $1 \mu \mathrm{~F}$ device) do not differ much from those obtained in the electrolytes of conventional tantalum capacitors. It therefore seems likely that this technique will be successful when such capacitorelectrodes are used in future implants.

Resistance to mechanical shear forces on the implant, with consequent breaking of inter-package wiring, will be sought by virtually wiring the implant with springs.

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Fig. 4. Two of the new thick-film logic circuits compared in size with the hermetically sealed logic package (above) used in the second implant.

## Announcements

The basic methods and techniques used in Industrial Digital Control Systems and their applications in both computer and non-computer systems. is the subject to be studied at a vacation school on Industrial Digital Control Systems. It is being organized by the Control and Automation Division of the Institution of Electrical Engineers, Savoy Place, London WC2R OBL, in association with the Institute of Measurement and Control, to take place at the University of Oxford between April 7 and 11th.

An agreement has been signed between Keighley Instruments, 1 Boulton Road, Reading, Berks RG2 0NL. and Neff instruments of Duarte, California, USA for marketing Neff's range of data amplifiers for laboratories in the UK and Ireland.

The fourth Salon International Audiovisual et Communication (International Audiovisual and Communication Show) will be held in Paris, Porte de Versailles, from April 2-8th. On April 3rd, the presentation-discussion will be concerned with "Cable Television in France Today".

Arrow-Hart (Europe) Ltd have announced the appointment of Radio Resistor Co Ltd, 9-11 Palmerston Road, Wealdstone, Harrow, HA4.7RS, to their network of stockists and distributors for Arrow switches including the subminiature ranges

GDS Sales Ltd, Michaelmas House, Salt Hill, Bath Road, Slough, Bucks, has announced that its franchise for Hewlett Packard optoelectronic components has been extended. Under the new agreement GDS will be stocking HP Schottky and PIN v.h.f./u.h.f. diodes.

A new business ITT Instrument Services is being established by the Distribution Division of ITT Components Group, Edinburgh Way, Harlow, Essex, with effect from the beginning of February. ITT Instrument Services takes over from the Instrument Product Group of ITT Electronic Services. It is a much larger and independent marketing operation with its own field force, catalogue and internal sales engineers, but utilizes the computer system and stores operation of ITT's Distribution Division. The catalogue covers the following product areas: oscilloscopes, digital multimeters and voltmeters, analogue multimeters, analogue and digital panel meters, signal sources, counter timers, testers, modular power supplies, bench power supplies and variable transformers.

This year's AGM of the National Association of Hospital Broadcasting Organisations is to be held on April 12 and 13th. The host station will be Radio Whittington, Whittington Hospital, North London.

Calvert Engineers Ltd has moved to new premises at 44a Elmsdale Road, Walthamstow. London E17 6PW. CEL has been involved in the manufacture and installation of telecommunication equipment and with this new move production capacity is being increased to include cable television equipment.

Celdis Ltd. 37/39 Loverock Road, Reading, Berks RG3 IED have announced that they are the UK agents for the range of small electric motors manufactured by Papst Motoren KG in Germany.

The Electronic Component Show (RECMF) will this year be held at Olympia, London from May 13th to 16th, 09.30-17.30 daily. Organizers of the show are Industrial and Trade Fairs Holdings Ltd, Radcliffe House, Lennon Court, Lode Lane, Solihull, W Midlands.

## DICE throws a double

Latest version of DICE, the Digital Intercontinental Conversion Equipment designed by engineers of the Independent Broadcasting Authority, can convert 525line NTSC colour pictures as used in the USA and Japan into 625 -line PAL or SECAM pictures used in most other parts of the world, and will now also operate in the reverse direction, taking advantage of the availability of higher-speed integrated circuits. Improvements are mainly in the field of vertical resolution, particularly relevant to pictures coming in to the UK. This two-way DICE was first demonstrated outside the IBA Engineering Laboratories in December 1974 and an agreement has recently been signed giving Marconi Communication Systems exclusive world-wide manufacturing and marketing rights.

The standards conversion is essential not only for "live" relays via satellite, but also where programme material or videotape is exchanged between countries working to different television picture standards. A number of different types of standards converters have been developed over the years, but IBA engineers were the first to develop the unit based on digital techniques to eliminate the need for careful alignment and adjustment and to provide conversion without perceptive picture impairment.

The latest DICE occupies no more floor space than the original unit and uses about 8,000 integrated circuits, while the main storage devices alone represent the equivalent of more than 15 million transistors. Five-line interpolation is now used rather than the three-line integration of the experimental digital converter and the spatial filters have been improved. The converter is available for operational use within 30 s of switching on from cold.

## IEE recommends reconstruction of engineering profession

The following is a summary of the conclusions reached by a council of the Institution of Electrical Engineers concerned with the future organization of the engineering profession.*

The council agreed that the structure of the engineering profession was in need of change and endorsed the President's proposal that any change, whether in the form of an adjustment of the Council of Engineering Institutions or the setting up of a new central body to replace it, should be based on these principles:

- Authority and responsibility for learnedsociety and professional matters affecting special branches of engineering must remain in the hands of the individual specialized institutions.
- The central professional body should progressively become the single effective authority and instrument for qualifying chartered engineers, assisted wherever appropriate by experts nominated by the specialized institutions.
- The central body should not include technician engineers.
- The central body should not be federal in structure but should comprise individual engineers of all disciplines, the members of its council being elected in a suitable manner by the chartered engineers.
- Provision should be made to enable wellqualified members of certain non-chartered societies to become chartered engineers, provided that their education, training and experience were judged by the central body to be of sufficiently high standard.
- A person should not be eligible for registration as a chartered engineer unless he was a member of a specialized institution recognized for the purpose.
*"The importance of status", Wireless World,
Oct, 1974, p.363.


## First production

 c.c.d. memoryThe first c.c.d. memory to be put into largescale production has been introduced. The new device is a 1 -kilobyte serial storage element claimed to represent a significant advance in the density of solid-state memory storage. It is aimed at memory applications in terminal buffers, video display refresh; microprocessor-control data stores and electronic switching in data communication networks. The memory utilizes a buried channel, ion-implanted barrier structure in the storage registers combined with nchannel silicon-gate m.o.s. structures for timing, charge detection and level conversion circuitry. The nine two-way data lines are t.t.l. compatible and have three-state output buffers for wired-OR application.
The device is organized as 1,024 words by nine bits each. It contains nine 1,024-bit low power c.c.d. registers which are shifted in parallel to provide storage and retrieval of nine-bit words in a byte-serial mode. Each register is accessed by its own two-way data line and all nine registers are serviced by common two-phase data transfer clocks and read/write control functions. The device operates in four modes: read, write, read/modify/write and recirculate. Power dissipation in the read and write modes is said to be 250 mW maximum and only 30 mW in standby recirculate mode. Average random byte access time is $200 \mu \mathrm{~s}$. The device uses simple two-phase clocking and is packaged in a standard 18 -pin ceramic

Engineer uses EMI's portable Privateer telephone scrambler device to transmit confidential information back to head office.

d.i.1. Data rate is 50 kHz to 3 MHz . Evaluation quantities of the CCD450, manufactured by Fairchild are available on fourweek delivery, while production quantities will be available in the fourth quarter of 1975.

## High-speed waveform recorder

Since 1969 the National Research and Development Corporation has been supporting a work programme at the University of Manchester aimed at developing a novel type of storage cathode-ray tube to be used for signal averaging. A 16 -channel laboratory prototype has been built and NRDC would now like to hear from companies who would be interested in completing the development and assessment of the instrument and in its subsequent commercial exploitation.

The basic principle of the waveform recorder is as follows. The electron beam in a c.r.t. is focused so as to form a beam whose cross-section at the face of the tube is narrow (approximately 0.5 mm ) in the $x$-direction but broad in the $y$-direction (approximately 1 cm ). This rectangular beam falls upon a faceplate consisting of a series of parallel, electrically isolated strips of aluminium that are also narrow in the $x$-direction but broad in the $y$-direction. The electron beam can be scanned across the strips in the $x$-direction. The signal being investigated is fed to the c.r.t. electrodes controlling the $y$-deflection of the beam and the strips are located within the tube so that the amplitude of the signal determines how much of the beam's area falls upon any particular aluminium strip. With zero signal there is no overlap and when the signal is
maximum the entire beam falls on an aluminium strip. Each strip is connected to a storage capacitor which is charged by the impinging beam, the quantity of charge being determined by the degree of beam overlap. As the beam scans repeatedly in the $x$-direction, charge is accumulated and, by monitoring the potential of each capacitor, the average signal can be extracted.

## Study on teleconferencing

The Stanford Research Institute in California has recently undertaken a ninemonth study of "teleconference" systems that enable people to communicate to a mass audience across the span of a continent. Audio and visual systems that are substitutes for bringing together conference participants offer an attractive means of saving costs, but only if people use them. An engineer-economist of the institute states, "We know a lot about the technology of such systems, but we need to know a lot more about their psychological and sociological aspects". The findings will document effectiveness of the systems, usage patterns over a period of time and how the cost, quality and types of capabilities offered by a system affect its usage. In the UK the Post Office runs a service of this type called Confravision.

## TV deliveries still down

Deliveries to UK distributors of UK made and imported colour television receivers reached 165,000 in December, a $23 \%$ decrease compared with December 1973, according to the latest statistics compiled


## Charge coupled

 image sensor, the 'eye"'of RCA's new tubeless TV cameraheld below. The image sensor and camera will be available in Europe early in 1976.by the British Radio Equipment Manufacturers' Association. This brought the total for the year to $2,209,000$, a fall of $20 \%$ compared with 1973.

Total monochrome television deliveries for December of 51,000 brought the total for the year to 821,000 , a fall of $42 \%$ compared with January to December 1973. BREMA members delivered 59,000 audio stereo systems in the month, a fall of $27 \%$ compared with December 1973, bringing the year's total to 831,000 , a fall of $17 \%$ compared with 1973. Deliveries of radio receivers reached 259,000 for the month, a $44 \%$ drop on December 1973, bringing the 1974 total to $4,798,000$ compared with $6,681,000$ in 1973 , a fall of $28 \%$.

These figures are for deliveries of UK made and imported deliveries to home distributors including those to rental and relay companies.

## Bell Laboratories celebrate fifty years

The research and development unit of the Bell System marked its 50th anniversary in January. In its first 50 years Bell Labs scientists and engineers have been awarded more than 17,000 US patents, two Nobel physics prizes (in 1956 for the invention of the transistor), three National Medals of Science and hundreds of other prizes.

One of the largest industrial laboratories in the world, Bell Labs is now an organization of about 16,000 employees, with 18 locations in nine states of the USA. It was established in New York City in 1925 with the reassignment of 3,600 staff members of Western Electric Co's engineering department and some additional supporting personnel from the American Telephone \& Telegraph Co.

## Briefly

Radio City on v.h.f. The Independent Broadcasting Authority's new v.h.f. f.m. stereo transmitters at Allerton Park, Liverpool are now in operation on 96.7 MHz , carrying the programmes of Radio City, previously available only on 194 metres medium wave. The IBA's local 95.9 MHz relay station in Rotherham is also now in service, carrying the programmes of Radio Hallam.

Merseyside slant polarized. Since the start of programmes on January 24, the v.h.f. service of BBC Radio Merseyside $(95.8 \mathrm{MHz})$ has used slant polarization. This will provide improved reception for portable receivers and v.h.f. car radios, particularly towards the limit of the service area. Listeners using outdoor horizontal aerials should find reception unchanged.

New SERT president. The Council of the Society of Electronic and Radio Technicians has elected as its third President, Sir Cyril English, who took the chair on the occasion of SERT's 10th anniversary on January 30.

# 75 years of magnetic recording 

## 2-The dark years

by Basil Lane

Assistant Editor, Wireless World

Up to about 1915 the use of valves had been extremely limited and rarely applied to the telegraphone type of recorder. However, from that date on until the mid-1950s it was to play a massive part in turning a declining technology into a brilliant new era. The dark years of World War II were also approaching to produce a remarkable dichotomy in recording media. In this article the story advances to 1945.

The combination of World War I and mismanagement of the technical development of the Telegraphone, brought about the demise of the Poulsen companies by about 1918. From then on there are only passing references to magnetic recording in the literature, mostly connected with Poulsen models or slight variants of them. As mentioned in Part I of this series, it was Kurt Stille who revived interest in magnetic recording and this through the medium of the Dailygraph, later developed into the Textophone ${ }^{-6}$, and a steel tape machine originally conceived for synchronized film sound track.

In Britain, Stille's ideas were exploited by Ludwig Blattner, who, according to a contemporary account. ${ }^{27}$ was a small, lively man with a keen showman's mind. He , with his engineers, developed a machine called the Blattnerphone, an early model of which was used to provide synchronized sound for demonstration films. These films were used as part of a sort of "circus show" where a public audience would come to see the "talkies" and in the intervals Ludwig Blattner, also a keen dancer, would pull ladies from the audience to dance with him on stage to recorded music from the Blattnerphone!

More seriously, the BBC took an interest in these machines and by 1931 at least one had been bought and installed at Savoy Hill (Fig. 1). This was a machine that used steel tape 6 mm wide running at a speed of 1.5 metres per second with a playing time of 20 minutes. Since the drive was by d.c. motor, it suffered from wow and speed drift, which had to be corrected by operating a rheostat and observing a stroboscope attached to the capstan.

Pressure was increasing within the BBC to provide an Empire Service and since the government of the day had taken so long to produce a decision to allocate

Fig. I. An early 6 mm Blattnerphone machine installed in Savoy Hill in 1931. (Courtesy BBC).
funds for the capital investment, the BBC took an independent decision to finance the initial stage and open service just after Christmas 1931. Since the longdistance transmissions had to be timed to obtain reasonable hours of receptionusually early evening local time-broadcasts were beamed by using directional aerials, with the transmitters switched to
each aerial at two hour intervals. Thus, to enable a programme broadcast to Australia to be heard in Canada the material had to be available for repeat. Disc recording had not been used in the BBC up to that time, and in any case the playing time was rather limited. The Blattnerphone seemed to provide just the right answer.



Fig. 2. A Marconi-Stille recorder installed in BBC Maida Vale studios from 1934.


Fig. 3. The Stille erase head assembly showing the saturating fux fields.


Fig. 4. The record head fux field of a Marconi-Stille machine.

Having pointed out the deficiencies in the 6 mm Blattnerphone, the BBC then encouraged an engineer, von Heising of the British Blattnerphone Company, to develop a machine meeting the BBC requirements ${ }^{28}$. After only three months, two prototypes were produced and installed, first at Savoy Hill and then at Broadcasting House. Further technical details on these and later machines follow, but for the moment, suffice it to say that the speed stability was improved and tape width reduced to 3 mm . Apart from the somewhat dangerous operating conditions, the steel tapes were also difficult to edit.

Nevertheless, it was obvious that this represented somewhat of a challenge to the engineers of the day, since several magazine programmes were broadcast during late 1932 and in 1933. One of these included a composite programme of the 1932 Economic Conference in Ottawa which was compiled from seven miles of recorded steel tape ${ }^{29}$. However, the fact that steel tape was a new recording medium coupled with the prospect of being able to erase the tapes made it unreliable, in the minds of the BBC, as a source of archival recordings. What confirmed this thought was that part of the first Christmas Day, 1932 feature programme was accidentally erased. Godfrey ${ }^{29}$ goes on to say that subsequently arrangements were made, with the British Homophone Company, to record highlights onto disc from Blattnerphone tapes, the signal being fed from Maida Vale to Kilburn by telephone lines. He also remarks that this must have been the first time discs were produced from magnetic recordings.
Shortly after 1932, the Marconi Company bought rights in the Blattnerphone machine and produced a slightly lighter version which was mounted on a wooden table. By 1934 this, however, was
superseded by what surely must have been one of the largest audio magnetic recorders ever - the Marconi-Stille machine. This was mechanically very sophisticated and six were ordered and installed in Maida Vale from 1934 (Fig 2). Two more were added during the war and these machines were in constant use during this period and after, the last one being taken out sometime around 1950.

A fascinating tale is told of one of the early Blattnerphones. This machine was one of the original two 3 mm recorders installed at Savoy Hill and as part of the move to Broadcasting House they had to be shifted overnight. It had just been connected, though not tested, when a telephone call came through to the tape room to get a machine going, whatever it took. The switches were thrown without further ado and with, it would seem, a good deal of finger crossing, to record an historic interview with Amelia Earhart. The date was May 21,1932 , the very day she landed after an epic flight across the Atlantic.

History was to repeat itself since during 1939 it was resurrected from the embryo BBC museum to be the first tape machine installed at the dispersed BBC wartime location in Worcestershire. Once again, the same engineer, with other colleagues, had hardly completed the installation when they were told to get the machine going, this time to record the Prime Minister. The date was September 3, 1939 and the Prime Minister was Chamberlain broadcasting the declaration of a state of war between Britain and Germany ${ }^{30}$.

This self-same machine was again resurrected to record some items for the 50th Anniversary of the BBC and now rests in a well earned retirement at Bristol City Museum, awaiting location in a new gallery.

## Technical specification

The second generation Blattnerphones were driven with an a.c. synchronous motor which improved speed stability. Since this was an era before the adoption of a.c. bias, the tape was erased and biased with d.c. set from preset controls on the amplifier rack. The replay amplifier was a standard BBC type A amplifier ${ }^{31}$, modified to permit an equalization circuit to be connected to the grid of one of the valves. A power output stage, capable of giving up to 10 W , provided theloudspeaker monitoring facility. The microphone and head-driving amplifier were specially designed for the job. The Blattner machines were only fitted with three head block assemblies, the later Marconi types having five, the reason for which was not at first obvious to the author. Contact was therefore made with the engineer mentioned in the previous anecdotes, R. C. Patrick, for an explanation. It would seem that the idea originated with Patrick, who at that time was working in BBC Research. Marconi had just taken over the licence to produce the machines and had asked the BBC, as largest users of Blattnerphones, what improvements could
be made.
Editing of steel tapes was then quite common but unfortunately the actual edits, which consisted of a soldered joint, destroyed the knife-edge pole pieces of the record and replay heads. Patrick suggested that two standby heads, one record and one replay, were fitted which during operation of the machine were left out of contact with the tape. After the passage of an edit, the spare heads would then be quickly brought into contact and the damaged heads opened to permit replacement of the spring-loaded pole pieces and wait for another edit!

Of the three basic types of head assembly used, one was erase, one record and one replay. The design consisted of two simple pole pieces, solenoid wound, one on either side of the tape. The erasehead pole pieces had a flat contact surface with the tape and were made of Stalloy, also used for the record head. The assembly could be hinged open to facilitate threading.

Erasure was by saturation magnetization of the tape ${ }^{31}$, illustrated in Fig. 3. Briefly, a direct current of about 20 mA was passed through the coils connected in series. When the tape approached from the left, the field $h_{t}$ applied, the strength being above tape saturation as it passed under the first pole piece. There then followed a reversal of flux under the influence of field $h_{2}$ and finally another reversal caused by $h_{3}$. The tape was left in a saturated state in the direction of this field.

The record head was of similar construction, though the interchangeable pole pieces were this time shaped to a knife edge to improve short wavelength performance. Of the alternative arrangements possible, single pole piece or double narrow stagger, double wide stagger or double pole piece with one being idle, the BBC adopted the double pole narrow stagger arrangement (Fig. 4).

Again, the coils were connected in series and a 4 mA direct current bias applied with the signal. Here the tape saturation field $h_{1}$ was reduced by field $h_{2}$, restored to saturation by $h_{3}$ and finally subjected to the demagnetizing influence of $h_{4}$. Since $h_{4}$ was also modulated by the signal the remanent flux in the tape followed the current fluctuations in the head.

Finally, the replay head used by the BBC. had only one pole piece, made of Permalloy, since the setting of two pole pieces, which produced better highfrequency performance, was too critical for practical purposes.

The actual tape deck of the MarconiStille machine represented a considerable advance on early models with the tape drive being achieved through three motors. Tape was drawn off the feed spool by drive No. 2 and fed into a box reservoir

Fig. 5. Tape drive system for the MarconiStille machine.

Fig. 6. The first model Magnetophone shown at the Berlin Radio Exhibition, August 1935.
where a loop would build up. When the earthed loop contacted a metal surface at the bottom of the box the bias was removed from the grid of a thyratron and a relay in the anode operated, to switch a resistance into the motor circuit, slowing the motor down.

The tape was drawn from this reservoir by a capstan drive, which in turn fed a loop of tape into a second, larger reservoir. Again, when the loop of tape contacted the bottom of a reservoir a thyratron operated relay would remove resistance from the winding motor circuit speeding the motor up.

Despite sterling service and a surprisingly good performance for its day, disc recording gained the ascendancy during the World War II and after 1947, the impact of plastic based tape was to sound the death knell for this remarkable machine.

## Recording in Germany

Going back to the late 1920s the seeds were being sown, in Germany, of a new-
old idea which, in later years, was to revolutionize the art of magnetic recording. This was the revival of the idea of coating a flexible insulated base with a finely divided magnetizable substance. An independent engineer from Dresden, Fritz Pfleumer, was struggling to develop both a recording tape which had a flexible insulated base with a magnetizable surface and also a suitable machine. Presumably his funds and resources were too limited, since although he had secured a patent ${ }^{32}$ for such a tape (filed in February 1929), by 1930 he soon after sought the help of a German electrical company Allgemeine Elektrizitats Gesellschaft of Berlin (A.E.G.).

It is not too clear just how good a chemist Pfleumer was, since his early patent sounds rather more like a recipe for a pudding than a tape coating! In the introduction, he acknowledges that there prior inventions regarding the use of magnetizable substances on a flexible base but then goes on to describe the methods for his type of tape. I quote, ". . . a powder of soft iron is mixed with an organic binding medium such as

dissolved sugar, molasses or the like, which is then dried and finally caramelized or carbonized, that is, the carbon chemically combined in the iron by heating. The steel powder so produced is then, while in a heated state quenched in water or other liquid, dried and again powdered. The use of such a material has for its object that phonograms are thereby obtained which last many years without loss of strength of sound."

He went on to suggest that this powder could be then mixed with a water-insoluble binder and coated onto paper or cellulose type films. Also in the patent he suggests the coating of sound stripes on moving picture film. Several alternative magnetic materials were included in the specification, such as nickel-iron alloys, ferrosilicon or iron-hydrogen compounds. At least one reference ${ }^{33}$ indicates that Pfleumer did succeed in making paper tape, and also one coated on a cellulose hydrate film.

Fortunately for Pfleumer, A.E.G. were very interested in the proposition, but very soon realized that specialists would have to be used to manufacture a suitable tape. They chose I. G. Farbenindustrie Aktiengesellschaft of Ludwigshaven. This company specialized in the production of a wide variety of chemicals including fast opaque pigment dispersions and carbonyl iron used in the manufacture of loading (Pupin) coils for the telephone system.
Hermann Bücher of A.E.G. was soon in contact with a brilliant physical chemist at I. G. Farben, Wilhelm Gaus, who readily responded to the proposals and set to work on a suitable tape. The pace thus far seemed to have been a little slow from Pfleumer's first ideas, but now it increased-though not without quite a few problems, both technical and in company politics.

Some eighteen months after the initial approach Gaus reported back to Bücher that progress was good having received favourable reports on the quality of the first tapes delivered to A.E.G. In return, A.E.G. suggested that their machine was nearing completion and should be ready for launch in 1934 at the autumn Radio Exhibition in Berlin. With this air of optimism circulating, the two companies prepared for a grand launch. Designers at Ludwigshaven produced an exhibit which ran riot with ideas of the potential at domestic and broadcast level.

In July 1934, a decision to produce the first 10,000 metres of tape was taken, and by August this was in the hands of A.E.G. A further 40,000 metres was to be produced in time for the exhibition which was to be held from 17th to 27th August. With time getting short, internal politics started to show, since press releases and a prior announcement to a meeting of the TechnischLiterarische Gesellschaft du Berlin showed considerably greater restraint than the designers of the Ludwigshaven exhibit. Here an emphasis was laid on the speech recording aspect of the invention, rather than on music. Someone had suggested that any flaws in the performance would damage the prospects of the invention if exaggerated claims had been made initially. So, the plan was to underplay the potential, but as
events were to show, this sudden pessimism was the precursor to real problems. A joint meeting of management from both companies, was held one week before the exhibition and demonstration given. The result was that the recorder was withdrawn, delivery of tapes stopped and the press information suppressed as far as possible.

The trouble was two-fold, first that the prototype machine made in breadboard form, suffered from amplifier instability when condensed into a practical cabinet. Second, the performance did not come up to that of the competition. Remember, the Marconi-Stille and its predecessors had been in practical service in broadcasting for at least two years and similarly, in Germany C. Lorenz had introduced the StahltonBand Maschine ${ }^{34,35}$. This was a steel tape machine using Stille's principles, but considerably smaller than the British versions having a frequency response up to 5 kHz . The best achieved by the prototype A.E.G. machine was 3 kHz at a tape speed of $1 \mathrm{~m} / \mathrm{s}$. In addition the noise performance was hardly up to broadcast standards, so it was natural that there should be much soul-searching before taking any further commercial decisions.

Eight weeks later, the A.E.G. engineers announced that they had overcome the problems and a second demonstration was arranged. The resulting decision was favourable and so development went ahead to finally produce, in the summer of 1935 , a completely redesigned model meeting all requirements and available for the 1935 Radio Show in Berlin.

With a potential success on their hands, I. G. Farben suddenly ran into internal political problems with two of their factories -Ludwigshaven, who had developed and produced the first tape, and Wolfen entrenched in film coating, squabbling over who should mass-produce the tape. Wolfen, by the way, was later to be split, by an Occupying Forces Commission, away from I. G. Farben to become the Agfa tape and film concern-but that is a separate story to be told later. The final decision was delayed until 1938, due to vacillation by the Reichs-Rundfunk-Gesellschraft, (German Radio) on which recording system to adopt. By 1938. Ludwigshaven was so firmly in full production that no decision needed to be taken.

However, this takes us beyond August 1935 and the Radio Exhibition where the first eight A.E.G. machines, now called the Magnetophone, were shown and demonstrated with success, indeed with so much success, they were all immediately sold. The first Magnetophone tape was cellulose acetate, coated with carbonyl iron powder. Since at the time, the steel tape, wire, and direct-cut disc were firmly entrenched in broadcasting it was to be some years of hard selling before A.E.G. was successful in getting the Magnetophone accepted by the German broadcasting stations and during that time several stages of evolution were to occur. The first model (Fig. 6) was to be superseded by the FT2 an elegant console model, and the K3, a portable machine in three parts-deck, amplifier and loudspeaker. These appeared in $1937^{36}$
to be followed later by the K4, a broadcast machine made in portable or rack-mounted form. One interesting incident occurred in 1936 during the period of promotion; Sir Thomas Beecham was invited, with the London Philharmonic Orchestra, to go to Ludwigshaven to record the first public concert on magnetic tape. Beecham, being quite interested in recording, accepted and on November 19, 1936 made a tape recording parts of which survive to this day.

However, even he could not have been too impressed with the Magnetophone, since during that season he purchased two German optical sound recorders and had them installed in Covent Garden, where he later made private recordings of his seasons in 1937 and 1938!

Iron powder produced by the carbonyl process was not ideal as a magnetic material for tape since it had low coercivity and the individual particles were still too large to permit high-frequency recording. In addition the particles were spherical, a disadvantage not realized until much later when a study of small particle magnetics was to reveal the advantages of shape anisotropy.

However, there were other promising materials and one of these was magnetite $\left(\mathrm{Fe}_{3} \mathrm{O}_{4}\right)$ suggested in 1934 by Erwin Leher. Some tape was eventually produced using this oxide, but it had rather too high a coercivity which made erasure a problem, and so brown gamma-ferric oxide, still with spherical particles, was eventually adopted.

It was in January 1938 that seal of success was to be set upon the Magnetophone when the technical manager of Reichs-Rundfunk Gesellschaft, Dr. HansJoachim von Braunmuhl gave an announcement at a lecture that the Magnetophone had been adopted by R.R.G. for broadcast service.

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AMPLIFIER
CLAIMS
I was much amused by the letters from Mr Paravicini and Mr Radford in the January issue.

If one accepts that the best equipment from the two companies can look each other in the eye without too much neck stretching, then one must give Mr Rad ford the laurels for reaching this rarefied level with a much lower component count, and hence a better cost/price ratio.

Unfortunately, this does not guarantee success.

The lesson that British manufacturers, whether of amplifiers or motorcycles, have signally failed to learn is that the buying public is notoriously indifferent to specifications.

Lux will win the battle in the shops because, sadly, the most important parameter of all is the shiny knob area.
R. A. J. Glowacki,

London, N.W.3.

## RIBBON <br> MICROPHONES

John Dwyer's statements with regard to ribbon microphones in your "Microphone survey" in the October 1974 issue would seem to be drawn from references which relate to microphones produced in the 1930s and not of present day manufacture.

Beyer Dynamic have, for the last twelve years, been producing a hand held ribbon microphone. In fact within the range they have three different microphones serving the entertainment industry. All of these are supplied to broadcasting authorities and corporations throughout the world. They are also much in demand within the club circuits where microphones are not always treated very well and the Beyer microphones withstand the rough treatment in this area.

We would like to draw your attention to an extract from a letter we have received from the Revox Corporation of New York.
"I had thought that the English reviewers were somewhat more au fait with current ribbon microphone tech-
nology than their American counterparts, as here in America, I am constantly battling to overcome odious remarks and comparisons made against the ribbon transducer technique.
"It therefore came as some shock to note Mr Dwyer's same old hackneyed statements: to whit: 'The ribbon corrugations provide some control of the tension as well as increasing the mass of the ribbon and making it more rigid: it is still delicate, though, and susceptible to rumble and wind. The ribbon exhibits the worst susceptibility to handling noise.'
" 6
Ribbon microphones tended to be bulky in the past and their delicacy has tended to encourage them being abandoned in favour of the capacitor or moving coil types. They can be used for pressure operation by providing a cavity at the back of the ribbon to provide an acoustical resistance.'
"These damaging remarks, of course, cannot be applied to the Beyer ribbon. However, all ribbons seem to be 'tarred with the same brush' no matter whether they are described on your side of the Atlantic, or mine."

This we feel expresses the views of Beyer and, of course, of the Revox Corporation.
Douglas Ireland,
Eyeline Communications Ltd,
London WC2.

## Mr Dwyer replies:

Naturally the ribbon microphone can be constructed in such a way as to make it as good as other types.

The article was intended as a guide to the basic principles of operation of the various types of transducer now in wide use for good quality sound reproduction. All of the various types of transducer have disadvantages of one kind or another if only the basic construction is used. It is obviously true that a well designed unit of any type can overcome its inherent limitations. Nevertheless it is equally true that the cost of doing so may become an added limitation, as may the complexity of the unit so produced, and I think, if I may suggest so, that the simpler a unit is the more reliable it is. This may explain why, on the numerous occasions on which I have visited recording studios, the type of microphones predominantly in use were those either of the capacitor or the moving coil type. Every studio has at least one ribbon, but the occasions on which it is used tend to be rather specialised. I can only rely on the use to which the microphones are put as a guide to their value, though it may be that British recording engineers, like those elsewhere, have been subjected to a propaganda campaign of massive proportions conducted by the makers of capacitor and moving coil types in concert. If that is the case I can only say that I am sorry I have become an unwitting instrument of such propaganda. In addition. I am sure that Beyer microphones mentioned in the letter are every bit as good as Mr Ireland says Beyer say they are. My remarks were not intended to suggest that no ribbon microphone
could be as robust or as rumble-free as any other type, and it would be misleading to suggest that that was what I was saying.

## dB CONVERSION <br> ON A SLIDE-RULE

The article by Mr Nelson-Jones "Electronic engineers" slide rule" in the February issue prompts me to mention a technique for dB conversion using the LL2 and LL3 scales on a standard slide rule. If " 6 " on the C scale is set opposite " 1000 " on the LL3 (corresponding to $60 \mathrm{~dB}=1000$ ), other ratios may be converted to dB by reading from the LL3 scale to the $C$ scale; 6 on the $C$ scale is also opposite 2 on LL2 (corresponding to $6 \mathrm{~dB}=2$ ) so lower ratios may be read from the LL2 scale to the C scale.

Certainly the new rule should be a great deal more convenient, but the above technique may be of use to someone.
R. A. Scott,

Bury St. Edmunds, Suffolk.

## Mr Nelson-Jones replies:

I have tried the method suggested by Mr Scott and it is certainly ingenious, but I find it hard to remember which scale is which, and in addition the accuracy is not good. I am sure I would soon get used to the method, but I find it much easier to use the new scale with the $A$ and $B$ scales, and the accuracy is much higher. I had in fact heard of the method before, but had never tried it out until Mr Scott's letter arrived.

## EMERGENCY POWER GENERATOR

Congratulations to Mr J. M. Caunter for tackling the power disruption problem (February issue), but I feel that the car dynamo could have been more effectively converted by making use of the principles embodied in the most recent alternators fitted to cars. In these designs it is the rotating armature which is excited by the battery and the fixed stator windings which are used to generate the a.c. This has several advantages: the currents flowing into the armature via the brushes are smaller, and steady, and the armature heat dissipation is lower. The stator, by contrast, being heavily heat-sinked can develop quite large amounts of power, and, since plenty of winding space is available, can be more readily wound for 240 V . In modern car alternators, the regulating equipment is carried within the frame of the alternator, and consists of a power transistor controlling (on/off system) the armature current. The armature current is reduced whenever the output voltage causes a zener diode to conduct, so that the armature current is rapidly pulsed. This method of control, though suitable for battery charging, would not be suitable for a mains-output
alternator, and a voltage-controlled current regulator with a non-pulsed output would be needed.
I. R. Sinclair,

Braintree,
Essex.

## Mr Caunter replies:

While I agree with Mr Sinclair that most alternators work on the principles he describes, and there are several obvious advantages to be gained from using this method of construction, his suggestion is not applicable to the conversion of a dynamo for two important reasons.

Firstly, the dynamo has a solid steel yoke and cast-iron pole pieces and is therefore not designed for rotating field operation. If this were attempted, a large amount of power would be lost in circulating eddy currents within the solid stator. The armature, on the other hand is laminated to reduce this loss to a minimum when rotated within the stationary field supplied by the existing field winding. Secondly, since the stator is not of true annular form, the variation in reluctance of the magnetic circuit seen by the rotor as it rotated would produce serious distortion to the output waveform.

The best way to improve the performance of the alternator is to get as much copper as possible into the armature slots. This necessitates using a finer gauge wire to improve the filling factor, and either winding for 240 V in a single winding taking great care over the insulation, or by winding several parallel windings together and operating at a lower output voltage as in the present design. It is quite possible that the output could be increased to over 300 W in this way.

Incidentally, if anyone has been put off the idea of building this generator because of the machining needed to construct the slip rings, and has no scruples about passing a current through the dynamo bearings, the following suggestion passed to me by a colleague may be worth trying. Connect one end of the armature winding to the shaft and the other to all the commutator segments shorted together. With the earth brush removed, the output can now be taken from the alternator casing and the live (insulated) brush output.

## A NOVEL CLASS B OUTPUT?

As far as I know all class B output configurations are based on the same principle: two emitter followers are tied together and the circuit is improved, in a more or less elaborate way, by replacing a single emitter follower by a two- or three-transistor circuit in an attempt to approach an "ideal" emitter follower.

An example of this is the Quad 303 which has two triplets in the output stage. Although a very fine amplifier, it exhibits clearly the shortcomings of this type of output circuit, which are: (a) the quiescent current has to be adjusted; (b)

the quiescent current is dependent on the temperature; and (c) too much quiescent current results in a kind of "take over distortion". This kind of distortion is due to a signal current flowing through the resistors $R_{1}$ or $R_{2}$ (Fig. 1), cutting off the quiescent current of the other stage, which results in a voltage shift at the input necessary to keep the output following the signal.

It is obvious that crossover distortion decreases with increasing bias current. From the facts mentioned before it is also obvious that an increasing bias current causes an increasing "take over distortion". So, with this type of output there is an optional value for the quiescent current.
It is possible to overcome all these shortcomings by using the circuit shown in Fig. 2. This circuit has none of the limitations mentioned in (a), (b) and (c). The quiescent current is set at 15 mA by $\mathrm{Tr}_{7}$. (Later on 5 mA proved to be sufficient.) For d.c. this transistor forms a constant current source as long as diode $D$ is not forward biased. For small signals $T r_{1}$ and $T r_{4}$ can be regarded as a long-tailed pair without a tail, for positive signal the upper half ( $\operatorname{Tr}_{1}, T r_{2}, T r_{3}$ and $T r_{4}$ ) is active behaving as a super emitter follower. The same for negative signals,
but this time with $T r_{1}, T r_{4}, T r_{5}$ and $T r_{6}$.
Since $T r_{1}$ and $T r_{4}$ are used in both modes of operation and the output resistors are missing, no "take over distortion" is possible.

One advantage is a lower output impedance due to the missing output resistors.
Nico M. Visch,
Rotterdam,
Netherlands.

## DIGITAL <br> SPEEDOMETER

I read the articles on the digital speedometer by Bishop and Woodruff in the September and October issues with great interest, but I feel that "average speed" is not really the parameter of interest. What one really wants to know is the difference between the elapsed time and the time which should have been taken to travel that distance at a particular speed.
The above comment arises from the fact that one usually knows the distance to be travelled and a reasonable average speed which one can hope to maintain during the whole journey. What is required is an indication of how much time you have in hand or how far you are behind the clock at any time during the journey. This is the information provided mechanically by the Halda Speed Pilot used by many trials drivers.
I would thus be interested in a modification to the design of the average speed part of the project to substitute an electronic equivalent of the Speed Pilot. This only requires multiplying the actual distance travelled by the inverse of an


Fig. 2
average speed set in by hand and subtracing this from the actual time elapsed, to arrive at a plus or minus indication of the time in hand.
G. B. Weston,

Wooburn Moor,
Bucks.

## SOUND BROADCASTING DYNAMIC RANGE

There has recently been comment in the press ${ }^{1,2}$ on the undesirability of a large (but relatively natural) orchestral dynamic range, as broadcast by the BBC . The opinion expressed is that a lightly compressed programme is unsuitable for domestic loudspeaker reproduction. Thus a reduction in transmitted dynamic range is demanded. Such a step would be regressive and could not be easily compensated for by those who have the facilities to appreciate a natural dynamic range.

My suggestion is that domestic amplifiers should incorporate a switchable dynamic range compressor. Thus the transmitted dynamic range could remain high, and those people (including myself, at times) who require music at reduced dynamic range could then adjust the compression as necessary, while retaining the option to appreciate the full dynamic range.

It is well known that simple compressors are unsatisfactory on high-quality equipment-manufacturers would be expected to fit circuits and controls appropriate to the quality of the rest of the equipment. It is my belief that most people who demand an increase in compression would not notice the transient distortion which automatic control introduces. This innovation would also encourage the record companies to decrease their compression.
J. M. Hughes,

The University,

## Nottingham.

## References

1. Angus McKenzie, Hi Fi News and Record Review, January 1975, p. 107.
2. Tim Souster, "The Mike Oldfield concert", The Listener, January 23, 1975.

## TWIN VOLTAGE STABILIZED POWER SUPPLY

Mr Linsley Hood is to be congratulated on an excellent piece of writing and a very nicely conceived design ("Twin voltage stabilized power supply", January issue). Nevertheless there are one or two points about which I am not entirely happy, and on which he may care to comment:

1. An output smoothing capacitor has been used, of $32 \mu \mathrm{~F}$. This is far too big since it will make nonsense of the currentlimiting under conditions of initial connection (i.e., the current-limit won't work
until the capacitor has discharged its surplus coulombs into the luckless load). In theory there is no need for an output smoothing capacitor at all: in practice one will probably be found necessary to maintain stability, but it should not need to be greater than $1 \mu \mathrm{~F}$ or so.
2. I am not at all happy about the 12-volt reference supplies. As Mr Linsley Hood rightly points out, the overall performance of the whole circuit depends basically on the stability of the reference voltage; and the simple series-fed zener which he uses is not really good enough. A further defect is that he has chosen a 12 -volt zener, and this will have quite a large voltage/temperature coefficient. Three possible solutions to these defects present themselves: (a) change the zener voltage to 5.6 , which is a zener with practically zero temperature coefficient; (b) use two zeners in series $(8.2 \mathrm{~V}+3.9 \mathrm{~V}$, say) so that their temperature coefficients, which of course will be of opposite sign, cancel to near zero; (c) replace the zener with a suitable proprietary potted regulator.

Solutions (a) and (b) have, apart from the stated advantage regarding tempera-ture-coefficient, no other virtues. In fact they also have a number of fairly obvious drawbacks. Solution (c), on the other hand, is ideal-potted regulators are cheap (Signetics, for instance, do a very high quality one for 67 p ); their stability, both long and short term, is excellent; and the external circuitry with them is not only simple but allows for a precise adjustment of regulated voltage. In short, a suitable choice of potted regulator provides such an obviously ideal reference source for Mr Linsley Hood's excellent design that I cannot for the life of me see why he has failed to use it!
J. F. K. Nosworthy,

Cranleigh School,

## Surrey.

## Mr Linsley Hood replies:

I am grateful to Mr Nosworthy for his kind letter and his helpful comments. To take his second point first, the suggestion of replacing the zener stabilization of the regulators appears to be an excellent one. I only wish I had thought of that idea myself! However, the intention of the design in its published form was not to provide a very high degree of precision and the simple arrangement shown was adequate in practice.

On Mr Nosworthy's first point, concerning the size of the output capacitor, and the magnitude of the energy stored in this, the answer is not so simple. In practice, all engineering design is a matter of compromise between conflicting requirements; between performance and economy of means; between versatility and simplicity. Depending on the design specification or the order in which the designer places his priorities, so the nature of the design which will be evolved.

Because, in this instance. I was prepared to accept the use of a $32 \mu \mathrm{~F}$ output capacitor, it became practicable to use a relatively simple loop stabilization con-
figuration, having a straightforward 20 dB / decade roll-off in open-loop gain and a good gain and phase margin with a wide range of output load reactances coupled with a very high d.c. stabilization factor. The use of a smaller output capacitor would have demanded a lower open-loop gain and a flatter open-loop frequency response, and a different balance between the conflicting requirements of source and load ripple rejection.

## LOW-COST PRACTICE ELECTRONIC ORGAN

Electronic organs have continued to improve and prices are still competitive. In fact the "pop" enthusiast who is happy with a one-octave pedalboard is well catered for. However the "straight" organist who wishes to practise at home and needs two manuals and a 32 -note radiating and concave pedalboard to RCO dimensions has much less choice and faces a much higher outlay. A low-cost practice instrument is therefore proposed, on which one manual and the pedalboard are monophonic, i.e. capable of playing only one note each at a time. If the other manual is polyphonic (i.e. chords can be played on it) much of the classical repertory can then be practised on it, including Bach's trio sonatas. Much "pop" music can also be played on it.

Monophonic manuals already exist and the u.j.t. gives single-resistor tuning though not an ideal waveform; other circuits are available ${ }^{1}$ and tunable i.c. tone generators are now on sale. No monophonic pedalboard with 30 or 32 notes has yet been marketed, though a separate one-octave pedalboard is on sale. It would appear desirable to market a 32 -note monophonic pedalboard which could be used in conjunction with instruments lacking a pedalboard, and/or incorporated in the low-cost practice instrument proposed. In either event the pedalboard might be arranged to tip on end when not in use. 32 -note pedalboards are priced at $£ 40$ or more without circuits, and it might prove cheaper to mould the pedals etc. in plastic. A more drastic price reduction might perhaps be achieved by moulding the whole pedalboard in flexible plastic. The further alternative of moulding the whole pedalboard in rigid plastic and relying on proximity detectors to actuate the note played seems unlikely to find favour.

It is possible that a further reduction in cost might be achieved by limiting the polyphonic manual to a maximum of four notes at a time, as described by J. Asbery ${ }^{2}$; other methods might be developed for selecting from four tunable oscillators, e.g. by the interruption of light beams, but the devices used have of course to be shown to be cheaper than a conventional full range of oscillators. The practice organ might well have a headphone socket (with safe isolation), so that practice can be made inaudible to other people.

Opinions are invited from users as to
whether a low-cost practice instrument is worth developing, and if so what features should be included.
K. J. Young,

61 Madeley Street,
Derby DE3 8EZ.

## References

1. Oscillators and networks with singlepotentiometer frequency control, Young, K. J. Electron. Compon., Vol. 12, No. 19, Oct. 1971. 2. Multiphonic organ, Asbery, J. Wireless World, June 1973.

## IMPEDANCE OF A <br> TRANSMISSION LINE

I read with interest the articles on transmission lines: "Graphical analysis of pulses on lines", in the September 1972 issue and "Transmission lines for the birdwatcher" by P. I. Day in the September 1974 issue. They have been very useful to me, as I could take some hints from them and they led me to a successful method of analytical and graphical resolution of transmission line problems which is different from that of the Smith Chart. I have been able to achieve a thorough knowledge of the Smith Chart, and it seems to me that it cannot help to solve the problem of matching a transistor to a line without a stub, as suggested in the second article. The problem in fact is to find the impedance of the matching line, and its length, and it is impossible to properly enter into the Smith Chart if the impedance of the line is unknown.


I am sending the resolution of the first part of that problem which may also help in the use of the Smith Chart to solve many other problems.

$$
R_{o}=\sqrt{R_{i} R_{t}-\frac{\overline{R_{i} X_{i}^{2}}}{R_{i}-R_{t}}}
$$

In the transistor-matching circuit of the example in the article ( $R_{i}=50 ; R_{t}=$ $5+j 5$;) the impedance is 14.9 and the length $0.193 \lambda$. I think the formula is original, and hope it will be useful.
Romolo Aratari,
Gioia Dei Marsi,
Italy.

## Mr Day replies:

Sr. Aratari has obtained a result which certainly enables one to enter directly the Smith Chart, but he is incorrect in assuming that the Smith Chart cannot be used to determine the line impedance. There is a very simple construction by which we can find the impedance of a line needed to transform from one complex impedance to another. Obviously the situation he describes when we are transforming to a resistance is a special case of the general construction.


Basically we rely on the fact that a circle centred on the Smith Chart real axis can be transformed by a change of normalizing impedance to a circle anywhere on the chart axis. So to match $5+j 5$ to 50 ohms, as in the example, a possible procedure is as follows.

Choose any normalizing impedance, say 10 ohms, then the normalized impedances are $r_{t}+j x_{t}=0.5+j 0.5$

$$
\text { and } r_{i}=5
$$

Enter both points on the Smith Chart and construct a circle passing through both with its centre on the axis. The circle intercepts the axis at 5 and 0.45 so the required line impedance is

$$
10 \sqrt{5 \times 0.45}=15 \mathrm{ohms}
$$

To find the line length the simplest method is to re-enter the Smith Chart using 15 ohms as the normalizing impedance. The original choice of normalizing impedance is completely arbitrary, but greater accuracy is obtained the nearer the circle is to being central.

The constructed circle must not intersect the chart boundary. If it does then the simple matching is insufficient: this condition is identical to the requirement that $R_{o}$ be real, so $R_{i}$ must lie outside the range $R_{t} \rightarrow R_{i}+X_{t}^{2} / R_{t}$.

Unfortunately I am not aware of any references covering the use of the Smith Chart in this off-centred mode, but undoubtedly they must exist somewhere in the technical literature.

## ELECTROLYTIC CAPACITORS

I was most interested to read the survey on capacitors by Mr R. A. Fairs (December issue) and feel that the presentation was extremely useful. There is, however, one criticism which I would offer on his article, where he refers to the practice of etching aluminium foil in electrolytic capacitors (see p.512). The point is that etching does not increase the permittivity of aluminium oxide, which is generally between 7 and 10 . Eiching is applied to the base aluminium foil and this can increase the surface area by up to three times that of a plain foil. The oxide layer is then formed over the etched foil, resulting in the subminiature capacitors which we see today.

The etch factor and permittivity "con-
stants" can be better recognized when the formula for a capacitor is examined.

$$
\begin{aligned}
\text { i.e. } C_{p F} & =\frac{\Sigma \times A}{4 \pi t \times 9 \times 10^{11}} \\
\text { which becomes } & C_{p F}
\end{aligned}=\frac{0.0885 \Sigma A}{t}
$$

where $\Sigma=$ permittivity, i.e. $7-10$ for aluminium oxide, $A=$ the area of each plate in sq.cm, and $t=$ distance between plates in cm ; (in the case of electrolytic capacitors this is the thickness of the oxide layer).

The question of etched foil capacitors being unable to withstand high currents is not entirely correct as multi-tab internal connections ensure that the high peak and/or ripple currents can be applied. Certain limitations to ripple currents do exist with regard to low $C V$ products, due to the dissipation of heat (generated by the $I^{2} R$ loss inherent in the electrolyte and connections) from the surface area of the can, but, in the main, etching of the foil only marginally degrades the tangent of loss angle $(\tan \delta)$.

## P. D. Habermel,

Mullard Ltd,
London W.C. 1.

## Mr Fairs replies:

The statement concerning the increase in permittivity of etched aluminium foils was not entirely correct. The point here is that, although etching increases the effective area of the foil, it does not alter the thickness of aluminium oxide coating applied after etching has taken place.

In the early days of manufacture of etched aluminium electrolytic capacitors, the aluminium oxide coating may have been inconsistent due to imperfections in the etch; this explanation would support my statement on this matter, the material for this part of the article being drawn from several research papers on this topic.

I do take Mr Habermel's point on this "increase" in permittivity and support the arguments he gives showing that etching does not cause an increase in permittivity of the aluminium oxide in present-day manufacture of electrolytics.
As Mr Habermel points out, the current rating of etched foil electrolytic capacitors is only slightly different from plain foil types. My statements on this matter were not intended to deter any would-be purchasers of etched film capacitors (which are usually adequate for almost every application) but merely to point out the design limitation that exists between the two types. It was unfortunate that space in the article did not permit a more fuller discussion on the differences between the two types of capacitor.
There is not much I can add to Mr Habermel's informative letter except that one can argue a slightly greater dissipation factor in etched film electrolytics due to tortuous paths in the etched film followed by the oxide layer: this argument can be considered trivial in present-day technology.
I thank Mr Habermel for his kind comments and his interest in the article.


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# Noise-confusion in more ways <br> than one 

## 2-Noise temperature and noise generators

by K. L. Smith<br>University of Kent at Canterbury

## In part 1, temperature was shown to play a large part in discussions about noise. In this part the noise temperature concept is discussed more fully, together with methods of measurement at low frequencies using a noise generator.

If a resistor at room temperature is connected across the input terminals of an amplifier of bandwidth $B(\mathrm{~Hz})$, the available noise power $k T_{0} B$ is amplified by the gain $G_{A}$. This means that the output power from the amplifier is $G_{A} k T_{o} B$. The noise power added by the amplifier must also be taken into account. If this amplifier contribution is $P_{N a}$ at the output, it can be added to the above expression directly, because noise powers from different sources can simply be summed if they are unrelated. The total available output noise power $P_{N O}$ becomes $G_{A} k T_{0} B+P_{N a}$ as shown in Fig. 5(a).

This is the point at which we think up our first bit of convenient fiction. We imagine that the amplifier is completely noiseless and account for $P_{N a}$ by a (now fictitious) extra noise power available at the input terminals. So we write $P_{N a}=G_{A} k T_{e} B$. By this dodge we can replace a noisy amplifier by a noiseless equivalent, Fig. 5(b), whose output is

$$
\begin{gathered}
P_{N O}=G_{A} k T_{0} B+G_{A} k T_{e} B \\
P_{N O}=G_{A} k B\left(T_{o}+T_{e}\right) .
\end{gathered}
$$

The whole thing is equivalent to an input source resistor at a temperature of $T_{o}+T_{e}$ connected to a noise-free amplifier, where $T_{o}$ is the room temperature of the actual

Fig. 5. It is more convenient to replace a noisy real amplifier (a) with a noiseless one (b), and account for the noise by inventing a fictitious noise temperature $T_{e}$ at the input.
resistor at the input terminals ( $=290 \mathrm{~K}$ ) and $T_{e}$ is the effective input noise temperature of the amplifier. Like available gain, $T_{e}$ varies with input matching conditions, so there is not a unique $T_{e}$ for every system. It will depend on how the system is used. An amplifier with a low $T_{e}$ is better noisewise than one with higher temperatures. other things being equal. The idea of $T_{e}$ is a little abstract because it is not a physical temperature (the input of an amplifier with $T_{e}=4000 \mathrm{~K}$ would not be glowing white hot!).
One or two points arise at this stage. The first is that we are not limited to a source temperature of $T_{0}$ in every case. Thus the noise power output for a receiver whose effective input temperature is $T_{e}$ and connected to an aerial whose aerial temperature is $T_{a}$ is

$$
P_{N O}=G_{A} k B\left(T_{a}+T_{e}\right)
$$

Another point arising is to do with the bandwidth $B-\mathrm{I}$ have been assuming that we know all about it. $B$ is not the normal $3-\mathrm{dB}$ bandwidth used by radio engineers. but is the noise power equivalent bandwidth and involves notions about the available gain-bandwidth product. $\left(G_{A} B\right)$. I have relegated these ideas to a brief discussion in Appendix B.

There is another very easily overlooked complication and that is the possibility of more than one channel allowing signals and/or noise to pass through the system. An obvious example is the superhetrodyne receiver with a response at the image frequency. I often wonder how many experimenters measure the noise perform-
ance of their v.h.f. converters, oblivious of the fact that they have a wide open channel at the image frequency. Incidentally, this "improves" the (erroneously) measured single-channel noise performance figures. so one should beware of excellentlooking figures on some manufacturers equipment specifications.

A useful concept in connection with the above arguments is that of the operational noise temperature. $T_{o p}$. This is a measure of the overall system performance. A knowledge of $T_{o p}$ enables the all important output signal to noise ratio to be calculated. As an example of how this idea arises, consider a superhet with a gain $G_{s}$ at the signal frequency and $G_{i}$ at the image frequency, as outlined in Fig. 6. The noise bandwidth is usually $B_{I F}$ for all channels, because it is set by the i.f. amplifier. The signal may occupy a bandwidth $B_{s}\left(B_{s}<B_{I F}\right.$ because if the i.f. is narrow it will limit $B_{s}$ to $B_{I F}$ ). The total available output noise power from this receiver will be

$$
\begin{gather*}
P_{N O}= \\
k\left(T_{s}+T_{e}\right) B_{I F} G_{s}+k\left(T_{i}+T_{e}\right) B_{I F} G_{i} \tag{3}
\end{gather*}
$$

where $T_{s}$ is the temperature of the aerial. signal generator etc., at the frequency of the signal channel, and $T_{i}$ is the same quantity but at the image frequency. If the temperature is constant over the two channels, then $T_{s}=T_{i}$.
The question arises. how do we handle $P_{N O}$ for signals to noise ratio purposes? The answer is that if the available output signal power is $P_{\text {so }}$, the signal-to-noise ratio is given directly by $P_{s o} / P_{N o}$, a little thought shows this to be the important


final parameter in any data link or communications system. The effect of the noise power is as though all of it is concentrated into the signal bandwidth $B_{s}$. Therefore we define another temperature, the operating noise temperature, $T_{o p}$ as $P_{N o} / k B_{s} G_{s}$.
Notice the particular gain bandwidth product used. You will be pleased to know this is about the limit of abstract thinking we need, so we will soon be back to more concrete things!) Substituting for $P_{N o}$. by using equation (3), and assuming for simplicity that $T_{s}=T_{i}$ and relabelling them $T_{a}$, the aerial temperature, operating noise tempetature becomes

$$
\begin{aligned}
T_{o p} & =\frac{\left(T_{a}+T_{e}\right) B_{I F}\left(G_{s}+G_{i}\right)}{B_{s} G_{s}} \\
& =\frac{B_{I F}}{B_{s}}\left(T_{a}+T_{e}\right)\left(1+\frac{G_{i}}{G_{s}}\right) .
\end{aligned}
$$

Fig. 7. Overall noise temperature of a cascade of amplifier stages can be deduced as. shown here.

This equation offers considerable meat to get one's teeth into. First, it illustrates the rationale of using temperatures in noise discussions. Awkward Boltzmann's constants cancel out and one is left with the various temperatures and parameters of the amplifier only. Clearly, the output signal to noise ratio degrades as $T_{\text {op }}$ becomes larger. The lowest $T_{e}$ should be the aim when designing the equipment and is achieved by noise matching and low noise components in the front end.

Care should be taken to understand the significance of $T_{a}$. For instance, the signal from a satellite is not enhanced when it is originating from the direction of the sun! ( $T_{a}$ shoots up.) Significantly, simple but all too easily-forgotten pieces of work need to be attended to, such as making sure $B_{I F}$ is not greater than $B_{s}$. If the receiver bandwidth is twice as wide, say as that required to pass the signal, then $T_{o p}$ is doubled. The noise coming in via the image channel increases $T_{o p}$. If $G_{i}=G_{s}$ (as in microwave receivers) $T_{o p}$

Fig. 6. In a superhet receiver there are usually at least two channels through which noise can pass to the output. Unless signal information is also coming in via the image frequency $f_{i}$, it is always advantageous to reduce $G_{i}$ to the smallest possible value. The "shape factor" of the i.f. bandpass, $B_{I F}$ also has a significant effect on the noise performance.
is again doubled. The receiver designer should reduce the bandwidth to the minimum ( $B_{s}$ ) and filter out the image, (make $G_{i}=0$ ) to obtain the minimum operating noise temperature. Then $T_{o p}=T_{a}+T_{e}$.

There are certain wideband signals which are received with a sensitivity advantage if both channels are wide open. Radio astronomical signals are themselves wideband noise powers. This means that useful signal powers are received in both sidebands. In fact the wider the bandwidth of the radio astronomy receiver the more signal power will be received. There is a worsening of signal-to-noise ratio by a factor of two if a double-channel receiver is used to receive a single-channel signal.
If the gain of the first stage of an amplifier or receiver is high, intuition might suggest that noise power contributions by later stages are not significant. Although intuition is not very trustworthy sometimes, in this example it is all right, as the following argument shows.

If we consider the three stages with gains and effective temperatures as shown in Fig. 7 then the output noise power is

$$
\begin{equation*}
P_{N O}=G_{l} G_{2} G_{3} k B\left(T_{i}+T_{e}\right) \tag{4}
\end{equation*}
$$

The noise output of the first stage is the noise power from the resistor times $G_{t}$ plus the contribution represented by $T_{e t}$.

Therefore the available noise output from stage one is $G_{I} k B\left(T_{i}+T_{e l}\right)$. The output from the second stage is its own noise, represented by $T_{e 2}$, plus the input from stage one multiplied by $G_{2}$. The output from stage two is

$$
G_{2} k B\left[G_{i}\left(T_{i}+T_{e l}\right)+T_{e 2}\right] .
$$

Similarly the output from stage three, which is the final output noise power, is

$$
G_{3} k B\left\{G_{2}\left[G_{I}\left(T_{i}+T_{e I}\right)+T_{e 2}\right]+T_{e 3}\right\}(5) .
$$

Equations (4) and (5) are both expressions for $P_{N 0}$, therefore,

$$
\begin{gathered}
G_{I} G_{2} G_{3} k B\left(T_{i}+T_{e}\right)= \\
G_{3} k B\left\{G_{2}\left[G_{l}\left(T_{i}+T_{e l}\right)+T_{e 2}\right]+T_{e 3}\right\} .
\end{gathered}
$$




Fig. 8. Still an extremely useful noise source for measurement purposes, the saturated thermionic diode is an absolute noise generator. (a) shows a typical circuit using an A2087, (b) is the equivalent' circuit for calculation purposes.

This cancels down to the final simple equation:

$$
T_{e}=T_{e I}+\frac{T_{e 2}}{G_{I}}+\frac{T_{e 3}}{G_{I} G_{2}} .
$$

Notice that the term containing $T_{i}$ conveniently subtracts from both sides. This equation shows that if the first stage gain is, for example, 100 times and the effective noise temperature of the second stage is 300 K , then the contribution to the overall $T_{e}$ by stage two is only 3 K . Usually the third term can be neglected, unless $G_{2}$ is very small. The gain of stage three $\left(G_{3}\right)$ has not entered into the picture. The argument can be extended to any number of stages. The equation is conveniently termed the cascading formula and in effect describes how the various noise temperatures throughout a chain of stages can all be referred to the front-end terminals as a single $T_{e}$, the system of stages is regarded from then on as noiseless.

## Measuring $T_{e}$

The way in which I have discussed the role of the absolute temperature in noise problems, shows the convenience of dividing the output noise power from a signal handling system into two parts. One part is the noise that comes in with the signal represented by $T_{a}$ and the other is that introduced by the local equipment, which accounts for the $T_{e}$ term. This means that all the various noise powers produced in the local equipment are lumped together under the title $T_{e}$ -whether they originate as thermal noise in the resistors, shot effect in the transistors or valves, flicker noise and so on.

If you have just built a receiver or a
customer has ordered a system to be designed with a stated maximum $T_{o p}$, it is essential to be able to make fairly accurate measurements of $T_{e}$, so that you know what you are talking about. The basis for any noise measurements involves generating accurately known noise powers. The standard noise generators are based on physical mechanisms including the saturated thermionic diode, the gas discharge tube and the noise generated in a reverse biased semiconductor diode. Sinewave signal generators can be used as standard power sources, but because they produce narrow band signals, their use in noise measurements involves difficulties interpreting what bandwidths mean and errors are very likely.

Before going on to the construction of noise sources, I will discuss a technique for measuring $T_{e}$. The following way for determining $T_{e}$ might be termed the ratio method. A noise source with an effective temperature $T_{\text {hot }}$ when it is fired, is coupled into the amplifier or receiver and the output $P_{\text {No(hal) }}$ is noted on a power meter. The noise source is now switched off but still connected to the system. The temperature when the noise source is not fired can be labelled $T_{\text {cold }}$, with a corresponding output power from the system, $P_{\text {Nofcold })}$. It is not necessary to know accurately the actual values of the output powers, only their ratio, $A$.

As an example, consider the superhet receiver for which equation (3) applies. Putting in the appropriate values for the "hot" and "cold" conditions, gives

$$
\begin{gathered}
P_{N O(h o l)}=k B_{I F}\left(T_{\text {hot }}+T_{e}\right)\left(G_{s}+G_{i}\right) \\
\text { and } P_{N O(\text { cold })}=k B_{I F}\left(T_{\text {cold }}+T_{e}\right)\left(G_{s}+G_{i}\right) .
\end{gathered}
$$

## Dividing them gives $A$

$$
A=\frac{P_{\text {Norfolt }}}{P_{\text {NO(cold })}}=\frac{T_{\text {hot }}+T_{e}}{T_{\text {cold }}+T_{e}} .
$$

From which we get

$$
\begin{equation*}
T=\frac{T_{\text {hot }}-A T_{\text {cold }}}{A-1} \tag{6}
\end{equation*}
$$

All we require to know is $T_{\text {hot }}, T_{\text {cold }}$ and $A$. The bandwidths, gains and $k$ have cancelled. This straightforward result applies for any system whether there are more channels than two or any other complexities. For best results, the value of $A$ is often chosen to be two (the minimum error occurs near this value).
As usual, the assumptions made should be considered to avoid, or at least understand, errors that might creep in. $T_{\text {cold }}$ is usually taken to be $T_{0}$, but the temperature of the lab or workshop in which the measurements are made could very well differ by a few degrees from 290 K , and there will be a corresponding error introduced. $T_{\text {hot }}$ must be known accurately for the particular noise source. The matching conditions of the noise source to the receiver or amplifier should duplicate the conditions that will apply in the operational system. It is not certain that the source impedance of the noise generator when it is fired will be the same as when it is cold. Any difference that does exist will introduce an error in $T_{e}$, but it is difficult to establish any such impedance shift.
The output meter should be a true square-law device with voltage or current. In other words it should be accurately linear as a function of power. Any overloading or non-linearity in the amplifier will introduce errors. For instance, the common f.m. receiver is non-linear for amplitude changes, and cannot be investigated by the above method. (The front end could be checked separately, but we are discussing a.m. noise, which would normally be eliminated in this kind of receiver anyway. In f.m. systems the more difficult f.m. noise has to be considered). Errors also arise at the higher frequencies, mainly because of the usual effects of the stray reactances.

## Sources of wideband noise, diode noise generators

One of the most useful noise generators for frequencies up to a few hundred megahertz is based on the temperature limited
diode. The full shot-noise generated by a thermionic diode operated under these conditions can be calculated exactly, but involves fairly complex statistical ideas such as Campbell's theorem. A treatment can be found in reference 9 (see part 1). Pierce derived the shot noise equation very simply but his method lacks the rigour demanded by, purists. It is an interesting derivation and I have included an outline of it in Appendix C. The full shot noise produced on a direct current $I_{a}$ in a bandwidth $B$, is

$$
\overline{i_{\text {shot }}}=2 e I_{a} B
$$

where $e$ is the charge on an electron.
Because the diode is saturated, the effective source resistance of the shot noise generator is very high indeed. Fig. 8 shows a typical circuit for a diode noise generator with a source resistance, $R$. The equivalent circuit is also shown. The total available noise power from the generator is the sum of the noise power from the shot source and that from $R$, which is at the ambient temperature $T$. The two sources of noise power are not correlated, so that their outputs add directly as we have seen earlier. From Fig. 8(b) the available power from the two current generators is

$$
P_{N}=\frac{\overline{i ⿻}^{2}}{4 G}=\frac{\overline{i n}_{\text {shot }}+{\overline{i^{2}}}^{2}}{4 G}
$$

where the conductance $G$ is equal to $1 / R$.

$$
P_{N}=\frac{e I_{G} B R}{2}+k T B
$$

Excess noise temperature, $T_{D}$, for a saturated diode is obtained by equating the first term on the right hand side of this last equation to $k T_{D} B$, so that $T_{D}=e I_{a} R / 2 k$. The numerical values of the physical constants, $e$ and $k$ give the value 11,600 for the quotient $e / k$. Therefore $T_{D}=5800 I_{a} R$. The total noise temperature of the fired source is $T_{D}$ plus the contribution from $R$

$$
\begin{equation*}
T_{h o t}=5800 I_{a} R+T \tag{7}
\end{equation*}
$$

The cold temperature is simply $T$, because with the diode off, $I_{a}=0$ and no contribution is forthcoming from the shot noise term. From these considerations we know the values of $T_{\text {hor }}$ and $T_{\text {cold }}$ to use in equation (6). Putting in the quantities gives

$$
T_{e}=\frac{5800 I_{a} R+T-A T}{A-1}
$$

which conveniently simplifies to

$$
T_{e}=\frac{5800 I_{a} R}{A-1}-T
$$

A number of authors have used the ideas of the noise ratio and excess noise ratio. I think we have enough detail from the preceding discussions to illustrate at this point, how these ideas are used. You may recall the definition involves the ratio of the temperature to 290 K or the ratio of the excess temperature to 290 K respectively. The ratios obtained are really
noise power ratios, in which the bandwidth and Boltzmann's constant cancel. Being a power ratio, the results are often expressed in decibels. By dividing the equation above by 290 K we obtain the noise ratio $t_{e}$

$$
t_{e}=\frac{T_{e}}{290}=\frac{20 I_{a} R}{A-1}-\frac{T}{290}
$$

Often $T$ is taken equal to 290 K (but see my earlier cautionary note); in that instance this equation becomes

$$
t_{e}=\frac{T_{e}}{290}=\frac{20 I_{a} R}{A-1}-1
$$

The excess noise ratio for a diode generator can be obtained from equation (7) by subtracting 290 K from both sides, then dividing by 290 K

$$
\frac{T_{\text {hoi }}-1}{290}=20 I_{a} R+\frac{T}{290}-1
$$

and again if $T=290 \mathrm{~K}$

$$
\frac{T_{h o t}}{290}-1=20 I_{a} R
$$

The diode noise source is very convenient because the temperature and noise ratios are directly proportional to $I_{a}$, and by just winding up the filament temperature, $I_{a}$ can be set to any convenient values on an accurate anode current meter. (With due care not to burn out the filament of course!)

Ordinary lumped-component circuitry begins to fail as the frequency of operation rises toward the GHz region. The diode noise generator is no exception and errors begin to affect the result when measuring at the frequencies in question. Another effect becomes important at the same time: transit time of the electrons across the cathode to anode space is significant in the hundreds of megahertz range and the shot noise equation begins to break down. To be continued

## Appendix B

## Noise equivalent bandwidth

Perhaps you have noticed in the discussion so far, I have blandly assumed that $G_{A}$ is "the power gain", without any real attempt to discuss how this quantity varies with frequency. Most amplifiers, whether intended or not, are severely
limited in their frequency response. This means that $G_{A}$ is a maximum somewhere near the centre of the band and drops off towards zero at both ends of the response, except for d.c. amplifiers. If you think of a constant distribution of energy over the frequency spectrum (white noise) then the bandpass function "weights" the contribution in each very small band at points across the response. The total output power is a sum of all these weighted contributions. This is the kind of reasoning we do when finding averages. Fig. B shows an example to make the point clear.
We can imagine $G_{A}$ to stay at its maximum value for a bandwidth $B$, then drop off sharply to zero at each side. If the width B of this fictitious rectangular bandpass curve is such that the output power is the same as from the actual response, then $B$ is defined as the "equivalent noise power bandwidth". What we have really said is that the area of the rectangular curve is made the same as the area of the actual curve. This gives us a clue about the mathematical approach to writing down the definition. If the available noise power is constant over the band then the available noise power in any small band $d f$, is $K d f . K$ is the constant level. Therefore the available output power is $G_{A}(\rho) K d f$ and the total outpat power is
$P_{N o}=K \int_{\text {bandpass }} G_{A}(\rho) d f$.
By definition, the total output power is also
$P_{N o}=K B G_{A(\text { max })}$
Equating these gives
$B=\frac{l_{\text {bandpass }} G_{A}(f) d f}{G_{A(\text { max })}}$
This is alright if you can do the integration or look it up in tables, but if, as usual, no simple function exists for $G_{A}(f)$, then the integral would have to be solved numerically. Equation (B1) shows that the amount of noise power emanating from the output of a system is proportional to the gain-bandwidth product $B G_{A(\max )}$.
Note that $B$ is not the ordinary "halfpower" bandwidth; a simple exampleshows this to be true by relating the two bandwidths.
Consider the bandwidth to be limited by a series tuned circuit. The reactance at any frequency will be $X=\omega L-(1 / \omega C)$. Using the equation for $G_{A}$ (p.110) available gain is

Fig. B. Area of the shaded rectangle is equal to area under response curve. Height is $G_{A}(\max )$ and width is definition of equivalent noise
bandwidth, $B$.


$$
G_{A}=\frac{K^{\prime} R_{i n}^{2} R_{g} n^{2}}{\left(n^{2} R_{g}+R_{i n}\right)^{2}+\left(\omega L-\frac{1}{\omega C}\right)^{2}},
$$

which can be written

$$
G_{A}=\frac{\text { constant }}{R^{2}+\left(\omega L-\frac{1}{\omega C}\right)^{2}}
$$

where $R$ has been written for $n^{2} R_{g}+R_{\text {in }}$. From this, $G_{A(\max )}$ is constant $/ R^{2}$. At the 3-dB points $|X|= \pm R$ because $G_{A}$ is then equal to $\frac{1}{2} G_{A(\max )}$. This condition enables us to write down the frequencies of the 3-dB down points. From $R=\omega C-(1 / \omega L)$ and $-R=\omega C-(1 / \omega \quad L)$ we get two quadratic equations whose solutions are

$$
\begin{aligned}
\omega_{1} & =\frac{R}{2 C} \pm\left(\frac{R^{2}}{4 C^{2}}+\frac{1}{L C}\right)^{\frac{1}{2}} \\
\text { and } \quad \omega_{2} & =-\frac{R}{2 C} \pm\left(\frac{R^{2}}{4 C^{2}}+\frac{1}{L C}\right)^{\frac{1}{2}}
\end{aligned}
$$

Subtracting gives the frequency difference

$$
B_{3 d B}=f_{1}-f_{2}=\frac{\omega_{1}-\omega_{2}}{2 \pi}=\frac{R}{2 \pi C}
$$

Using equation (B2)

$$
B=\frac{1}{2 \pi} \int_{0}^{\infty} \frac{1}{1+\left(\frac{\omega C}{R}-\frac{1}{\omega L R}\right)^{2}} d \omega
$$

The integral is a "do-able" one, and involves $\tan ^{-1}$ type solutions. Carrying out this solution, $B$ is $R / 4 C$, which means that the relationship between $B$ and $B_{3 d B}$ for a single tuned circuit is $B=\pi B_{3 d B}^{\prime} / 2$. Thus $B$ is somewhat wider than $B_{3 d B}$. The Table shows a few relationships for other band-limiting filters.

| Circuit | Relationship |
| :--- | :--- |
| Two cascaded tuned | $B=1.22 B_{3 d B}$ |
| circuits |  |
| Three cascaded tuned | $B=1.16 B_{3 d B}$ |
| circuits | $B=1.11 B_{3 d B}$ |
| A staggered pair | $B=1.11 B_{3 d B}$ |
| A 4-pole Butterworth |  |
| filter | $B=1.05 B_{3 d B}$ |
| A 6-pole Butterworth |  |
| filter |  |

The noise bandwidth approaches the 3 dB bandwidth more and more closely as the "shape factor" improves. For ordinary i.f. amplifiers with a number of tuned stages, there is very little error if you assume $B \dot{=} B_{3 d B}$.

## Appendix C

## Shot noise equation

A simple but not very rigorous derivation of the shot noise current equation was ingeniously put forward in J. R. Pierce's paper, "Noise in Resistances and Electron Streams" published in the Bell System Technical Journal, volume 27 (1948). It goes something like this:


Fig. C. Artificial double cathode "diode" used by J. R. Pierce to derive the shot noise equation.

If a diode consisting of two emitting cathodes (Fig. C) has a potential $V$ between them, a current $I$ will pass equal to $I_{V}=I_{0} \exp (e V / k T)$, where $I_{0}$ is the current that passes when $V=O$; that is, by the thermally energetic electrons "bridging the gap". Differentiating gives

$$
\frac{d I_{V}}{d V}=\frac{1}{r_{a}}=\frac{I_{o} e}{k T} e^{e V / k T}
$$

As the mean square noise current expected from a resistance $r_{a}$ is $i^{2}=4 k T B / r_{a}$ and the diode "resistance" at zero volts is $r_{a}=k T / I_{o} e$, it follows that $\overline{i^{2}}=4 e I B$ after substituting for $r_{a}$. This is the total noise current produced by the special case of two cathodes exchanging current. Because noise powers add, then the mean square current of one cathode is half the value and therefore $\overline{i^{2}}=2 e I B$, which is the shot noise equation.

## (To be continued)

## Wireless World noise reducer

Next month's issue will contain the start of an article describing the Wireless World noise reducer, an add-on Dolby processor mainly for use with magnetic tape cassette machines. This constructional design, the only one of its kind, has been planned in close collaboration with Dolby Laboratories and will be available from Wireless World in kit form.

The unit includes a stereo Dolby B processor that is switchable for both encoding and decoding. This means that as well as decoding commercial Dolby B cassettes, encoded tapes can be prepared. For recording stereo broadcasts, a switched $\cdot 19 \mathrm{kHz}$ pilot-tone filter is included. And should B-type encoding be adopted for f.m. transmissions, as in the USA, the unit will also decode those. There is another use of the processor. Because of the improved signal-to-noise ratio obtained with the unit, recordings can be made at a lower level that would otherwise be possible. Consequently some of the noise reduction can be traded for a lower distortion level at peak recorded levels.

The Wireless World Dolby processor can be aligned without using additional instrumentation. The circuit board has been designed to include the required alignment facilities- 400 Hz and 5 kHz oscillators are constructed from components in the WW kit, together with a $1-\mathrm{kHz}$ meter calibration oscillator. Full alignment and calibration instructions are included in the article, which starts in the May issue with a description of the Dolby system and its functioning.

## HF predictions

Predicted disturbed periods are March 23-28, April 4-10 and 19-25.

Seasonal trend and low solar activity combine to produce FOTs and LUFs which give a restricted choice of time and frequency for reliable day-to-day communication. The charts show that the restriction is severe when both ends of a circuit are in the northern hemisphere.



## Seventh Intelsat IV Launch

The seventh in the series of Intelsat IV commercial communications satellites was launched on February 20 after a delay of two weeks from Cape Canaveral.

Final position of the 1400 kg satellite is on the equator over the Indian Ocean. When in position there will be three Intelsat IVs over the Atlantic Ocean, two over the Pacific and two over the Indian Ocean, completing the world-wide network originally planned. Each of these satellites is able to carry approximately 3,500 twoway telephone conversations and 12 television channels. Despite the growth already experienced, the pressure of rapidly growing demands for international telephone, television and data transmission has led to the development of even larger communications satellites. The first of a new series designated Intelsat IV-A, with twice the capacity of Intelsat IV is scheduled for launch in the summer of 1975. The satellites are owned by the International Telecommunications Satellite Organization. Frequency re-use by means of a modified communications subsystem using 20 transponders and a novel antenna configuration with separate antenna beams will aid the capacity doubling of the A series. The opening of the frequency spectrum above 10 GHz to satellite communications approved by the World Administrative Radio Conference in 1971 will eventually provide communications capacity at least five times higher than that available at the presently used frequencies of 4 and 6 GHz .
The delay in launching the most recent Intelsat IV satellite which was scheduled for launch on February 6, was due to the failure of a single electronic component in the spacecraft.

## Weather Satellite for Western States

The second in a series of weather satellites, Synchronous Meteorological Satellite-B, was scheduled for launch by NASA from Cape Canaveral aboard a Delta rocket at the end of January.

SMS-B (called SMS-2 in orbit) is to be placed in geosynchronous orbit over the equator at $36,357 \mathrm{~km}$ altitude at 135
degrees west longitude, which is directly south of Sitka, Alaska, and about 15 degrees southeast of Hawaii. From this position it can view the western half of the United States and Hawaii while its sister spacecraft, SMS-1, can view the eastern US from its perch at 75 degrees west longitude on a line with New York City and just south of Bogota, Columbia.

The two spacecraft will be able to keep a 24 -hour watch on the western hemisphere and provide cloud-cover pictures every 30 minutes to weathermen of the National Oceanic and Atmospheric Administration. Each carries a visible and infrared spin-scan radiometer that returns visible light daytime pictures of 0.9 km resolution day and night. This continuous coverage is of special importance for short term phenomena such as the severe storm conditions that precede tornadoes. In addition, the west coast of Africa, breeding ground for hurricanes that strike the Caribbean, Florida, Gulf of Mexico and US east coast areas, will be kept under the surveillance of SMS-1. The primary types of data to be obtained consist of meteorological, seismic and

Copy of Indonesia's national communications satellite, scheduled to be in orbit after mid 1976, is pictured at Hughes Aircraft Company in California.

tsunami information. Both SMS spacecraft also carry a space environment monitoring system that monitors solar particle flux, X-ray emission and magnetic field direction and strength.

The US synchronous orbit spacecraft are expected to be joined, beginning in 1977, by similar spacecraft placed in orbit by the European Space Research Organization, Japan and Russia to form a global network of synchronous orbit satellites The two SMS spacecraft, including all on-board instrumentation cost about $\$ 60 \mathrm{~m}$, the Delta launch vehicles about $\$ 4.5 \mathrm{~m}$ each.

## Self-repairing Memories

A technique for the self-checking of a faulty memory on board space-craft* is under development by Intertechnique, a firm best known for its nuclear instruments and minicomputers. The concept was reported at the Large Scale Integrated Circuits Conference in Paris early in December. The self-check of a memory removes the data stored in it so the feature would be of. little value in missiles after they have been fired but could be valuable for satellites. The check can be made after the memory is dumped and its contents transmitted to a ground station. but before it starts to store information again. Intertechnique's concept. which has been patented, is for the memory to check itself at two levels. The lower level is in the basic memory elements, made up of one or several shift registers. These basic elements are grouped together on a printed circuit board along with complementarym.o.s. test and control logic integrated on a custom chip. Each element has a set of control logic and when a test sequence reveals a faulty one, the associated logic. in effect, shunts it out of the shift-register chain. These so-called elementary cards in turn work under control of a "system card" which contains circuits that interface the memory with the rest of the telemetry system plus c.m.o.s. logic to start tests of the elementary cards. If the test logic on one of them is faulty, meaning that reorganization at the lower level can no longer be made, the system logic shunts around the card and reconfigures the memory accordingly. The European Space Research Organization has so far funded the work.
*Electronics International, December 1974. pp.14E, 16E, 18E.

## Briefly

The Mariner-Jupiter mission scheduled by NASA for launch in 1977 will be the first deep-space probe to use X-band for telemetry and video transmission.

Skynet II, Europe's first communications satellite has recently been accepted as an operational system by the RAF acting on behalf of the Ministry of Defence.

## Deflection amplifier for oscilloscopes

The circuit combines the advantages of a differential output stage and a high-impedance j.f.e.t. input stage. The silicon input diodes form a crude overload protection for the input of the f.e.t. amplifier. Transistors $\operatorname{Tr}_{1}$ and $\operatorname{Tr}_{2}$ act together as both an amplifier and a level shifter, the quiescent output voltage of $\mathrm{Tr}_{2}$ being set by $R$, to approximately 15 V . This also sets the gain of the amplifier unfortunately. A multi-turn preset was used for this purpose as the setting can be quite critical.
Transistors $\operatorname{Tr}_{3}$ and $\operatorname{Tr}_{4}$ form a differential output stage enabling an output saving of about 400 V pk -pk. Feedback
is introduced through the 220 ohm emitter resistors and high-frequency compensation is brought about by $R_{2}$ and $C_{1}$. Resistor $R_{3}$ forms the Y -shift control.

To set up for operation, set $R_{2}$ and $C_{1}$ to their maximum values. Set +15 V at the collector of $\operatorname{Tr}_{2}$ using $\dot{R}_{1}$. Inject a 10 kHz square wave into the amplifier and increase $C_{I}$ to give the sharpest possible corner to the display without overshoot. Then increase $R_{2}$ as far as possible without losing too much of the squareness of the display.
G. A. Johnston,

Stechford,
Birmingham.


## Oscillator uses passive voltage-gain network

It is frequently necessary to make a simple oscillator when a limited range of components are available. Most phase sensitive networks used to define the frequency of oscillation have attenuation at zero phase shift. A Wien bridge attenuates three times, a three-stage RC iterative filter 29 times. It is therefore necessary to use an amplifier, but the bandwidth of the convenient 741 is limited and it is a significant item of expense in this context.

Consider the circuit of Fig. 1. When $a=2+2 \sqrt{2}$,

$$
\frac{V_{o}}{V_{i}}=\frac{2}{2 \sqrt{2}-1}=1.094
$$

for zero phase shift and $\omega C R=1$.
It is easy to obtain an output from an emitter follower greater than $1 / 1.094$.

The circuit of Fig. 2 was tried using a super-alpha pair.

An output of $20 \mathrm{~V} \mathrm{pk}-\mathrm{pk}$ was obtained. Factor $a$ was fixed at $4.7 \approx 2+2 \sqrt{2}$ as a preferred value.
A $4-\mathrm{V}, 0.2 \mathrm{~W}$ capless pilot bulb was used to stabilize the loop gain to unity, rather than the rarer R 53 thermistor.
W. R. Jackson,

University of Bristol.

## Low battery voltage indication

May I add something more to the interesting idea of P. C. J. Parsonage (Circuit Ideas, January 1973).

- The low battery voltage indicator circuit can be modified to work as a high battery voltage indicator, or simply a high voltage indicator, just by interchanging gate and cathode connections of the thyristor. In particular, say a battery voltage is 8.3 V and needs to be charged to 9 V , then the circuit of Fig. 1 can be used. The l.e.d. lights when battery charges to 9 V .


Fig. 2


Fig. 3


- The l.e.d. in the circuit of Fig. 2 lights up when the input voltage is $>$ or $<(V+\Delta V)$, where $V$ is the normal voltage at which circuit is designed and $\Delta V$ is the change in input voltage at which I.e.d. lights up.
- The cost of the equipment can be cut slightly by replacing the thyristor with a less costly silicon switching transistor, Fig. 3. This circuit can return to its original state (l.e.d. off) when the voltage returns to its design value.
P. R. K. Chetty,

ISRO,
Bangalore.

## Circuit Ideas

## Sine oscillator uses c.d.a.

The circuit, new in realization but not in principle, produces moderately low-distortion sinewaves (typically $0.5 \%$ t.h.d.) which have negligible amplitude bounce on changing frequency. Further advantages are the ability to alter frequency with a single component and the low cost of the quad differential amplifier (LM3900N).

When the supply is switched on the comparator output initially goes to $+V_{c c} \dot{ }$ after about a second $C_{4}$ has charged and the output rapidly slews to 0 V . This shocks the bandpass filter, formed by the two integrators $I C_{I a}, I C_{l c}$, and the inverting amplifier $I C_{l b}$, and causes it to ring. The resultant sinewave causes the comparator to produce a square wave which

is fed back into the loop to sustain oscillation. Sinewave amplitude is stabilized by virtue of the constant square wave input and is typically $0.25 \mathrm{~V}_{c c} \mathrm{pk}-\mathrm{pk}$, its purity being proportional to filter $Q$.

Frequency of oscillation $(2.34 \mathrm{kHz})$ and $Q(62)$ are:
$2 \pi f=\sqrt{\frac{R_{4}}{C_{1} R_{1} C_{2} R_{5} R_{3}}} \quad Q=\omega C_{1} R_{2}$
Note that owing to the internal compensation of the amplifiers significant $Q$ -
enhancement occurs at frequencies greater than a few kHz and this may lead to oscillation of the filter itself.

To vary the frequency the inset network can be used in place of $R_{3}$, the effective impedance being

$$
R_{13}=R_{17}+R_{18}+\frac{R_{I 7} R_{I 8}}{R_{19}}
$$

T. J. M. Rossiter, Corpus Christi College, Cambridge.

## Pulse height modulator

This circuit reduces the spike feedthrough in series f.e.t. gates by always limiting the gate voltage swing to between the source voltage and the pinch-off voltage. Referring to Fig. 1, if the input voltage ( $V_{i}$ ) is

varied between 0 and 13 V , say from an op-amp, then the f.e.t. gate would have to be swung from +13 V to $-V_{p}$ volts ( $V_{p}$ is pinch-off voltage). Fig. 2 shows one version of a circuit used to limit the voltage swing on the f.e.t. gate to approximately $V_{i}-V_{p}$. Input voltage is monitored by the emitter followers $\operatorname{Tr}_{1}$ and $T r_{2}$ and $T r_{1}$ emitter is maintained at $V_{i}-V_{b e} \approx V_{i}-0.7 \mathrm{~V}$. Zener diode $D_{2}$ is matched as nearly as possible to the measured $V_{p}$ of the particular f.e.t. in use. If $V_{p}<1 V$ a forward-biased diode (e.g. 1N916) may be used. The emitter of $\operatorname{Tr}_{4}$ is therefore established at $V_{i}-V_{p}-1.4 \mathrm{~V}$. $\operatorname{Tr}_{2}, \operatorname{Tr}_{3}$ and $D_{1}$ establish the upper limit of the voltage swing to $\approx V_{i}$. The switching waveform, a $\pm 15 \mathrm{~V}$ squarewave with
fast rise and fall times, drives the base of $T_{5}$. Clearly from Fig. 2 the output waveform cannot go below $V_{i}-V_{p}-1.4 \mathrm{~V}$ or above $V_{i}$.

Capacitors $C_{3}$ and $C_{4}$ are optional. Capacitor $C_{4}$ increases the rise time of the output signal and $C_{3}$ increases the fall time. Very slow turn off times can be obtained by suitable adjustment of $C_{3}$ thereby giving further spike reduction. Resistor $R$, should be kept high because for low values of $V_{p}$ and high values of $V_{i}$ the emitter-base junction of $\operatorname{Tr}_{5}$ will become reverse biased. Alternatively a diode can be placed between the emitter and $R_{5}$.

The modulator was tried with a number of different types of f.e.ts and always reduced the spike amplitude when compared to the spike produced by a full $\pm 15 \mathrm{~V}$ swing on the gate. By using a slow fall time the spike amplitude for this edge could easily be reduced by an order of magnitude. The circuit may need slight modification to suit individual requirements but works well with a slowly changing analogue signal and with switching rise/fall time of the order of $1 \mu \mathrm{~s}$.
M. D. G. Dabbs,

Home Office Central Research Establishment, Aldermaston, Berks.



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# An aerial rotator servo 

by D. J. Telfer, A.R.I.C


#### Abstract

This article describes a circuit for remotely adjusting angular displacements in a drive shaft, for use with $12-24 \mathrm{~V}$ d.c. motors at continuous currents of up to 250 mA . The system is well suited to a wide range of applications and has been very successfully employed as an automatic aerial rotator. The advantages of proportional control are available while preserving low cost and simplicity of design.


Sometimes there are applications in which the full potential of elaborate control equipment may not be fully exploited. In such instances, a less complex and more economical system could adequately perform the required operations. The control system to be described in this article is simple and yet has been found to be reliable in operation and particularly well suited for use in automatic aerial rotators. Although it was initially designed, while the author was with the Department of Physics, UMIST, for remote positioning of furnace charges, the circuit lends itself to many other possible applications, not least in the teaching laboratory as a technique for demonstrating the use of feedback systems and the principles of proportional control.

## Proportional control system

A block schematic diagram is shown in Fig. 1. Use of a Wheatstone bridge to provide positive or negative error signals follows conventional practice. The spindle of one potentiometer $R V_{1}$ is mechanically coupled to the final drive (signified by the dotted line) and the other potentiometer $R V_{2}$ is the final drive position selector. A difference in the relative positions of the wiper arms of $R V_{I}$ and $R V_{2}$ produces an error signal which is amplified, firstly by the differential amplifier $A_{1}$ and then by an output stage $A_{2}$ connected to the motor, whose direction and speed depend on polarity and magnitude, respectively, of the voltage applied across its terminals. The final-drive shaft keeps turning until the wiper arm position at the motor-driven potentiometer catches up with the selected setting of the control potentiometer. The error signal is thereby continually reduced until the motor stops with the final-drive shaft in the desired position. In the author's design, operational amplifiers are used for $A_{1}$, and $A_{2}$ consists of two pairs of complementary emitter followers. An additional feature is the


Fig. 1. Block schematic diagram of the proportional control system.


Fig. 2. The final drive shaft of motor $M$ is coupled to the wiper arm of $R V_{1}$.
electronic bridge shunt $S$, which is activated at the final stage of operation to ensure that the motor is switched off.

Amplifier. In Fig. 2 the d.c. error voltage is taken to a pair of differential amplifiers $I C_{1}$ and $I C_{2}$, whose gain is adjusted with preset potentiometers $R V_{3}$ and $R V_{4}$ respectively. When the wiper arm of $R V_{2}$ is more positive than that of $R V_{1}$, the output of $I C_{1}$ goes negative and that of $I C_{2}$ goes positive. Under these conditions, $T r_{2}$ and $T r_{3}$ are turned off, while $T r_{1}$ and $T r_{4}$ are turned on, affording a low resistance path through which the motor is connected across the supply. If the wiper arm of $R V_{1}$ is more positive than that of $R V_{2}, \cdot T r_{1}$ and $T r_{4}$ are turned off and conduction is through $T r_{2}$ and $T r_{3}$, whereupon polarity of the voltage applied to the motor is reversed.

Proportional control. The mode of operation is conveniently described by assigning three states to the system. Fig. 3(a) shows how the output voltage of $I C_{1}$ ( $V_{1}$ ) and of $I C_{2}\left(V_{2}\right)$ varies with angular displacement, $\theta$, of the driven potentiometer spindle with respect to the setting chosen for $R V_{1}$, which is represented by $\theta=0$ at A.

In region D-C, the input signal is large enough to saturate both amplifiers $I C_{1}$ and $I C_{2}$. Motor voltage, which depends upon the difference between $V_{1}$ and $V_{2}$, is held at a maximum value. The final-drive shaft rotates at a constant angular velocity, and the spindle of $R V_{2}$ is driven towards the selected rest position that it will eventually take up at A.

At an angle $\theta_{2}$ from A, which is predetermined by the setting of $R V_{3}$, the error voltage falls below that level required to saturate $I C_{1}$, and $V_{1}$ steadily decreases. Passage through $C$ represents the onset of proportional control.

In the region $\mathrm{B}-\mathrm{A}$, amplifier $I C_{2}$ is no longer held at saturation. However, the setting of $R V_{4}$ is such that it has greater gain than $I C_{l}$. Its proportional control bandwidth, given by $2 \theta_{i}$, is correspondingly narrower than that of $I C_{1}$. The value of $V_{1}-V_{2}$ continues to fall, ideally reaching zero at A . If these conditions are faithfully transmitted to the motor, there is no residual current in the windings and the final-drive shaft comes to rest with the spindle of $R V_{2}$ exactly in the position determined by $R V_{1}$. In practice, the motor may stop when an appreciable voltage is still being applied to its terminals. Since at B the value of $V_{1}-V_{2}$ is just over half its maximum value at $C$, this event will be captured within the narrow region BA, provided that mechanical loading is not excessive and that the motor is not severely under-run. Although the author has experienced no difficulty on occasions when 24 V motors were run using a 12 V supply, it is recommended that the h.t. voltage should be at least $60 \%$ of the voltage rating for the motor.

The output voltages of $I C_{1}$ and $I C_{2}$ are not transmitted faithfully to the motor because of the emitter-base voltage drop incurred at the power transistors. In Fig.

3(b), the emitter voltages of transistor pairs $\operatorname{Tr}_{1}, \operatorname{Tr}_{3}\left(V_{13}\right)$ and of $\operatorname{Tr}_{2}, \operatorname{Tr}_{4}\left(V_{24}\right)$ converge to plateaux centred at A . The difference between $V_{13}$ and $V_{24}$ is therefore the voltage applied to the motor. However, the range of $\theta$ values over which the motor is stationary, SAS', may be compressed by increasing the gain of $I C_{2}$. This will not affect the overall proportional control bandwidth of the system, which is given by $2 \theta_{2}$, and is dependent on the gain of $I C_{1}$.

Protection of transistors. Quite low values of residual voltage across the motor can give rise to standing currents high enough to justify an automatic switching arrangement for protection of the conducting pair of output transistors, which will dissipate maximum power just before they become


Fig. 3(a). Voltage at IC outputs plotted against $\theta$; at (b) is shown the voltage at emitters of power transistors plotted against $\theta$. Limiting voltage of the motor is reached at $S$ and $S^{\prime}$.


Fig. 4. Circuit diagram of bridge shunt.
biased to cut-off, when the emittercollector voltages approach their highest values. The motor may be made to cut out below a certain applied voltage, within the region BA of Fig. 3(b), by connecting a suitable relay across the motor. For example, a motor rated at 24 V maximum was run with 20 V on the h.t. rail of the circuit. Satisfactory action was obtained from a reed switch having a solenoid resistance of $800 \Omega$, operating at 7 V .

Any such cut-out device must come into operation before the motor has actually stopped, resulting in a dead zone about A in Fig. 3(b) which is greater than SAS'. This state of affairs may be avoided by introducing a time delay so that the motor can stop at its limiting voltage before being switched off.

An alternative switching method, incorporating a delay, is shown in Fig. 4. This solid-state approach, which the author has found to be very effective, uses a complementary pair of transistors shunted across the bridge potentiometers. Conductance of the transistor pair $T r_{s}$ and $T r_{6}$ is appreciable only when both base voltages are within a limited range centred on half h.t. potential. The state of this circuit may first be considered with the input diodes $D_{3}$ and $D_{4}$ disconnected from the output of $I C_{2}$.

The bases of $T r_{s}$ and $T r_{6}$ are connected by a resistor through which most of the mutual base current will flow, since $D_{3}$ and $D_{4}$ are reverse-biased by the small potential difference reflected across this resistor. Base bias is forward at both transistors, which conduct and act as emitter followers. Their mutual load is the bridge, across which the voltage falls to a value approaching the sum of the voltages across the interbase resistor and the emitter-base junctions. In practice this total amounts to about 2 V .
Next the connexion of $D_{3}$ and $D_{4}$ to the output of $I C_{2}$ is restored, via a limiting resistor $R$. No significant change will occur at the bridge shunt until the small reverse bias voltage at either $D_{3}$ or $D_{4}$ is cancelled by a voltage swing at $I C_{2}$, transmitted through $R$. When the output of $I C_{2}$ goes sufficiently positive, conduction through $T r_{6}$ is maintained but $T r_{5}$ is cut off. Conversely, $\operatorname{Tr}_{5}$ conducts and $\operatorname{Tr}_{6}$ is turned off during negative excursions. The diodes $D_{l}$ and $D_{2}$ protect $T r_{5}$ and $T r_{6}$ from Zener breakdown of their baseemitter junctions under reverse biasing conditions.

Finally, the onset of shunting action is delayed by introducing a capacitor $C$ between the input of the shunt circuit and ground. A suitable choice of time constant for $R C$ is about one-fifth of the duration of the proportional control régime.

## Practical circuit

The circuit diagram of a practical design for use with 24 V d.c. motors appears in Fig. 5. An electronic bridge shunt is employed and the unit may be run from a 15 to 28 V supply. Feedback capacitors are included to lower the a.c. gain of the operational amplifiers in order to reduce transient response and provide a safe-

guard against instability at settings of high d.c. gain. An interference suppression capacitor is also connected across the motor terminals. Inclusion of offset null controls (the $10 \mathrm{k} \Omega$ potentiometers) is recommended. Adjustments are carried out with the wiper arms of the bridge potentiometers brought to the centre of their tracks and then short circuited together. The offset null potentiometers are then set to give an output of exactly half h.t. potential at each operational amplifier.

A panel meter for monitoring the behaviour of the motor is a useful asset. Totat current may be measured, as shown in Fig. 5, or, alternatively, motor voltage or current may be displayed, using a centre-zero instrument to follow directional changes.

The power supply should be capable of delivering 1 A at the operating voltage and be well smoothed. Otherwise, requirements are not critical.

Performance. Operation with the bridge shunt is not critically dependent upon supply voltage, so long as the input capacitor value fulfils the time constant requirements mentioned above. Fine adjustments may be made with the $220 \mathrm{k} \Omega$ preset potentiometer, which is normally set near mid-range. Efficacy of the shunt is improved if bridge resistance is high compared to the value of resistance presented by the shunt during its turn-on period. However, the values of bridge circuit resistors shown in Fig. 7 were found to be more than adequate and may be considered to represent an upper practical limit above which the performance of the differential amplifiers becomes adversely affected. This arises
because of the differences in d.c. input resistance of the inverting and non-inverting inputs, and variation in amplifier gain with wiper arm position at the bridge potentiometers. The operational amplifiers see highest source resistance, and experience concomitant reduction in gain, when the wiper arms are near track centre. In this region, therefore, the proportional control bandwidth becomes relatively expanded.

Measurements of amplifier output voltage were made with the bridge wiper arms positioned at similar track intervals and then shorted together. Experimental conditions and data are summarized in Table


1 for an h.t. of 15 V and feedback resistors of $680 \mathrm{k} \Omega\left(I C_{1}\right)$ and $4.7 \mathrm{M} \Omega\left(I C_{2}\right)$. The behaviour pattern shown in Fig. 6 represents the situation with the wiper arms at the positive end of the bridge. The crossover point did not deviate markedly from the $\theta=0$ axis, but was displaced in voltage, being more positive than the halfh.t. potential axis, which is taken as the zero of voltage reference. At the negative end of the bridge, an approximately equal negative displacement relative to a centre offset potential of 0.25 V was observed. The bridge shunt was removed during these measurements, which confirmed that the effective common mode gain of the amplifiers was near to unity. This tends to produce a degradation in symmetry of the proportional control characteristics, which change progressively from one end of the bridge to the other. Therefore, the ratio of voltage across the bridge to peak swing at the amplifier outputs should not exceed 0.15 if good symmetry is to be preserved.

Voltage reflected across the bridge is directly proportional to the supply voltage whether or not the bridge shunt is used, so that proportional control bandwidth at
given gain settings remains practically constant above 20 V h.t. At lower h.t. voltages the discrepancy between peak output swing of the operational amplifiers and the supply voltage must be taken into account. During conduction, approximately 1.5 mA base current flows at the power transistors. This loads the amplifiers sufficiently to produce a total discrepancy of about 1.5 V . As the supply voltage is reduced, there is little change in this value, but its effect in decreasing the bandwidth becomes more noticeable.

In addition, the difference between h.t. and peak motor voltage amounts to approximately four volts, and this becomes an important consideration when using the circuit to drive motors at lower peak voltages.

Circuit properties are considered further in the light of other practical experiments. A small 24 V d.c. motor (see Fig. 4) was connected to the circuit of Fig. 7, which was operated at 24 V h.t. and with fixed
feedback resistors; $330 \mathrm{k} \Omega$ for $I C_{I}$ and $4.7 \mathrm{M} \Omega$ for $I C_{2}$. Maximum potential across the bridge was 3 V , falling to 0.7 V at cut-off, when the motor current was reduced to less than one microamp.

An $x y$ plot of motor voltage against amplifier input voltage $V_{i}$ measured at the bridge wiper arms is presented in Fig. 7. Total proportional control bandwidth CC' was 82 degrees, centred at mid-scale, for a driven potentiometer electrical rotation of 280 degrees. The bandwidth of $I C_{2}$ was seven degrees, giving a practical dead zone of $\pm 2.5$ degrees for a limiting motor voltage of 3 V .

Potentiometer drive. There are various possible mechanical arrangements at the bridge potentiometers, and only the rotary type is considered. To cover rotation through a complete circle, a 360 -degree potentiometer with $1: 1$ coupling is required at the final-drive shaft and also at the control box. Alternatively, the more usual

pattern with electrical rotation in the region of 280 degrees may be used in conjunction with pulley, chain or gear coupling of the correct ratio. If the absence of a 90 -degree sector from the rotation range can be tolerated, direct $1: 1$ coupling may be retained, as in the rotator.

## Aerial rotator

In point-to-point v.h.f. and u.h.f. communication, well-sited portable equipment incorporating a low-power transmitter can be capable of very encouraging results, particularly if a high-gain directional aerial is used, in conjunction with a reliable and accurate means of turning the mast. In aerial rotator applications the servo system may be used in conjunction with a variety of mechanical arrangements, depending on the requirements of the operator.

Basic construction of a portable aerial rotator for mounting on the roof-rack of a stationary vehicle is shown schematically in Fig. 8. The drive unit is readily demountable and an alternative type may be fitted if desired; Fig. 9 shows how the gearbox adapted by the author was installed. This was part of an ex-government switching unit having rubber mounting bushes and a 24 V d.c. motor coupled to the final-drive shaft through 625:1 reduction gearing.

Removal of the lower cover plate exposed the switch wafers, which were then discarded to allow a feedback potentiometer to be coupled to the finaldrive shaft through the $1: 1$ gearing as illustrated. Drive was transmitted to the mast through a simple dog clutch. A similar arrangement was employed at the potentiometer spindle, into which a slot was cut to accommodate a blade filed on the end of the coupling shaft. Although the potentiometer could have been mounted in a carefully positioned hole drilled in the lower cover plate, compactness was preserved by fixing the potentiometer case to the inside of the cover plate with soft solder, in the position shown. In order to mount the component in this way, the threaded part of the spindle collar was shortened. The spindle was of nylon to minimize damage in the event of accidental servo overruns at track limits.

Above the deck, short lengths of mild steel slotted angle were bolted together to act as a support for the vaned tube containing the aerial mast socket bush (Fig. 8). Grease was applied liberally to the mast socket bearing before fitting it to the bush. Positioning of the lower retention bolt allowed the mast socket assembly to be lifted just clear of the gearbox dog to permit easy and rapid alteration of the aerial reference direction by 180 degrees. An upper retention bolt was also fitted to secure the mast. Steel J-clamps were used to firmly fasten the rotator to a secure vehicle roof-rack (Fig. 10) and dimensions of the grooved mounting blocks were adjusted to suit the type of rack. Protection from the weather was afforded by fitting an aluminium cover over the gearbox and applying paint to exter nal surfaces.

Upper torsional limits for the above

Fig. 9. Underside view of rotator showing adapted gearbox unit with lower cover plate detached to expose the feedback potentiometer (bottom right). Final drive shaft and dog clutch are in the centre, with the motor at left centre.

Fig. 10. Rotator secured to horizontal roof-rack bars.
gearbox were approached in normal weather conditions with an eight-element conventional Yagi array cut for the twometre amateur band, which was supported at its centre of gravity on a five-foot mast. Aerials of greater physical size were not considered practicable on a free-standing mast fixed to this type of rotator.

Mechanical backlash in the blade and slot feedback potentiometer coupling has the effect of allowing the aerial to overrun its selected heading, but by judicious use of the relative sizes of blade and slot, can be made to correct any slight lag which may otherwise be present.

Feedback and control potentiometers should preferably have a linearity better than $2 \%$, and the system be calibrated before operation.

In practice, the portable rotator has performed with consistent reliability in conjunction with the control unit described. Aerials have also included a 16 element aerial for the 70 cm amateur band. using a five-foot mast.

When the portable rotator was used with the above proportional control unit, time taken for complete rotation of a 2 m eight-element Yagi array was about 20 s at 15 V h.t.

Circuit assembly. Components in the prototype were mounted on a $2 \frac{1}{2}$ in square piece of 0.1 in matrix Veroboard, in a $4 \frac{1}{2} \times 3 \frac{1}{2} \times 2$ in diecast box, with the control potentiometer and dial on the largest face. A five-cored cable from the motor and driven potentiometer was plugged into a DIN socket on the control box, allowing different motor units to be activated.

If the motor connections are reversed, an aerial rotator will become an automatic beam heading avoider. Care must be taken to connect the control and driven potentiometers in the correct sense, and to prevent mechanical damage to the latter component, operational checking should be carried out with both wiper arms near track centre.

## Other applications

In common with other proportional control systems, the above design commends itself to a wide variety of possible functions. Simple modifications may greatly extend its range of capabilities.

By connecting a suitable amplifier (such as another 741) in place of the driven potentiometer, the system may be coupled to external probes or sensors. For instance, the e.m.f. across a thermocouple junction may be used for remote automatic position-

ing of a furnace charge. Position is manually pre-set with the control potentiometer.

If the driven potentiometer is mechanically disengaged from the motor, the unit becomes a manually adjustable reversible motor speed controller.

## Law and insurance

It is of the utmost importance to ensure that, as a load attached to a vehicle roofrack, the rotator and aerial conform to legal requirements.

There must be no danger to people inside or outside the vehicle. On a public highway the aerial and rotator also become illegal if any part extends beyond the front, rear or sides of the vehicle by more than 12 in .
Any effect that the presence of the aerial and rotator may have on the vehicle's insurance should be ascertained.

The author has found that the authorities are very willing to help in these matters, and if the operator has any doubts about his position, he should not hesitate
to seek advice from the Traffic Department of the local police.

## Suppliers

Transistors and integrated circuits were obtained from Texas instruments. Minimum size of heat sinks for the power transistors will depend on circuit applications, and manufacturer's literature should be consulted. For the rotator, the TO-92 plastic encapsulation may be bonded to the diecast box with epoxy adhesive.

The surplus gearbox unit, and also separate d.c. motors, were obtained from North West Electrics, 769 Stockport Road, Manchester.
An extensive range of small gearboxes is manufactured and supplied by S. H. Muffet Limited, Mount Ephraim Works, Tunbridge Wells, Kent. For driving the rotator, the author recommends that a unit is chosen with an output ratio of $500-1,000$ which is capable of delivering at least 30 lb . in. continuous torque at the output shaft.

# Recent loudspeaker developments 

Consider the performance of a practical loudspeaker system in which the sealed volume of the enclosure, cone area and mass of the moving parts are kept constant. The results of changing the motor strength are plotted in Fig. 1. In the 70 to 500 Hz range, 'if $B l$ (product of magnet strength $B$ and coil winding length $l$ ) is increased output will increase, if it is decreased output will decrease. However, around resonance the reverse happens. Increasing $B l$ decreases output and decreasing $B l$ increases output. In other words efficiency or cone velocity for a given input at frequencies above resonance are directly proportional to $B l$, while at frequencies around resonance these two factors are inversely proportional to $B l$.

It can be seen from Fig. 1 that for a given loudspeaker system, and where a flat amplitude response is desired, the motor must be of the correct strength. If the motor is too small, efficiency is low and there is a bump in the bass. If it is too large, efficiency is high, but the bass response is down. This also shows that purchasing the speaker with the larger magnet could result in the use of a speaker with less than optimum bass response. By juggling motor parameters, there is apparently an inevitable trade-off between bass response and efficiency in the flat band. Restating the requirements, then, we need a large motor for high efficiency above resonance and a smaller motor for similar efficiency at bass frequencies.

## Dual motors

The usual practice for adjusting the power output of the motor is to vary the magnet strength, $B$. To construct a speaker with two different magnetic field densities to drive the same cone would be both expensive and difficult to manufacture. Suppose instead it was possible in effect to make $l$ vary with frequency in such a manner that a lower value of $B l$ in one frequency range would not affect a higher $B l$ product in another range and vice versa. Fig. 2 shows a simple method.

A second voice coil is wound over or under the conventional voice coil and is driven via a series $L C$ resonant circuit adjusted to resonate at the same frequency as the fundamental mechanical/acoustical resonant frequency of the woofer. The $L C$ circuit presents almost zero impedance at resonance and a sufficiently high impedance one octave either side of resonance to effectively remove voice coil 2 from the circuit. Thus $l$ of voice coil 2 can be adjusted to eliminate the high value of motional induced back e.m.f. at the fundamental resonance, $f_{o}$. A lower impedance path is provided at $f_{o}$ to maintain current


#### Abstract

According to a recent article in the American journal Audio ${ }^{1}$ it is possible to adjust separately the amplitude response of the upper range and bass frequencies of a single loudspeaker drive unit without one affecting the other, thus reducing the necessity for the motor (coil and magnet system) to be of an optimum strength for a system.


flow and the bass response at $f_{o}$ can be adjusted at will and independently of midrange response above $f_{o}$. In effect a second motor is added that generates less back e.m.f. and offers a lower impedance to the amplifier at the tuned frequency.

A voice coil gap of twice the width is not required to accommodate the extra voice coil. The output and inner clearance spaces are the same as usual and since a single layer coil with a notch in the top plate for the return end of the coil has apparently proved satisfactory, the gap width need only be increased by $25 \%$.

Summarizing, the design (it is claimed) "does not involve trade-offs in areas of performance, requires no additional amplifier power or equalizer, has the


FREQUENCY ( Hz )
Fig. 1. Loudspeaker amplitude response for different values of motor strength (a) motor too large, (b) motor optimum, (c) motor too small.


Fig. 2. Circuit for a dual motor woofer.
advantage of simplicity of construction and offers an improvement readily discernible on listening".

## Soft speaker

Further to the item "Flexible speaker cone" (News of the Month, March issue), this system, under development in W. Germany by JWM Systems, has caused somewhat of a stir in the technical press, be it on an academic level of interest only. The flat diaphragm structure of this new loudspeaker is a flexible, highly viscous, elastic material. The diaphragm is driven at its centre by a conventional voice coil and magnet system. The area of radiation is inversely proportional to the radiated frequency.
To ensure that the radiation area is symmetrical around the centre point of the diaphragm, the voice coil is split into two, each section being fed with an offset current of opposite polarity. In similar manner to a differential input, like signals cancel so that the voice coil remains centred in the gap between magnet pole pieces. The system requires equalization to account for a 6 dB per octave drop in response above 6 kHz .
The new diaphragm ${ }^{2}$ consists of a flexible bearing structure and a filling compound. A mesh of polyamide threads is used as a bearer, which is capable of stretching. In its manufacture, the diaphragmis radially pre-loaded (stretched) to a small degree and the visco-elastic filling compound is applied as a lacteal dispersion which dries like varnish and becomes interlaced in the polyamide.
The diaphragm is held at its circumference by the speaker basket frame and in addition there is a firm star-shaped support in the centre. In operation, amplitude is limited to a maximum of 3 mm at the coupling point between voice coil and diaphragm. With the amplitude of vibration decreasing out from the centre, a smaller and smaller part of the diaphragm vibrates as frequency rises so that the large area required for moving a large air volume at bass levels and the low mass favourable for rapid movement at high levels is achieved with the single diaphragm. The prototype drive requires a continuous input of 3.2 W for 96 dB s.p.l. at 1 m at 1 kHz .

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# Voltage-to-frequency converters 

# This article complements set 21 of Circards 

by J. Carruthers, J. H. Evans, J. Kinsler and P. Williams

Paisley College of Technology

Voltage-controlled oscillators-astable multivibrators-waveform generatorsfrequency modulators: under each of these headings one finds circuits that have an important common property, that the output frequency is a function of some reference or control signal. Such circuits are multi-variable systems in which several parameters of the output waveform are controlled singly or in various combinations by other parameters at the input. Thus the same circuit can appear under different headings depending on which input/output relationship is of priority concern.

As an example, some recent integrated circuits have been designed as waveform generators with square/triangle/sine wave outputs. If the output waveform is of no particular concern, the fact that the frequency of each output is proportional to a direct control voltage assumes a greater importance. The circuit can then be called a voltage-controlled oscillator. Now assume that the control voltage is set to a particular quiescent value with a smaller alternating voltage superimposed. Then the output frequency is modulated by the a.c. input, with the carrier frequency corresponding to the quiescent value of control voltage. The label for this circuit is frequency modulator.

In set 21 of Circards the primary property of interest is the relationship between an input voltage or current and the frequency of the output, with much less importance being attached to the wave shape or amplitude. A particularly desirable property is that the voltage-to-frequency relationship be linear, and in extreme cases departures from linearity of as little as $0.01 \%$ may be desired. In the process of achieving this, the output pulse height and width may have to be equally well controlled but these are a means to the end and not an end in themselves. There are other cases where the frequency needs to be varied only over a limited range, demanding only a small linear region to the $V / f$ characteristic. A good example is found in the design of v.c.os for high-frequency phase-locked loops. Restriction of the frequency range and of linearity is a compromise accepted more or less willingly in exchange for a speed capability that matches that of the associated digital circuits.

In nearly all of these examples, the basic timing mechanism is that of charging a capacitor from a control voltage or current. The voltage change across the capacitor is sensed by some level-detecting circuit which activates an electronic switch


Fig. 1. Constant charging current allows repetition frequency to be made proportional to current.


Fig. 2. To cause charging cycle to recommence, a low-value resistor is switched across the capacitor to discharge it quickly.


Fig. 3. If discharge time is made small enough the charging current can remain connected. Level of capacitor voltage is used to operate discharge switch.


Fig. 4. Triangular waves with repetition frequency proportional to current are produced by reversing capacitor charging current.
to discharge the capacitor and restart the cycle. Two categories of circuit can be clearly distinguished:

- where the discharge time of the capacitor is made short compared with the shortest charging time and need not be under the control of the input voltage, and
- where both charge and discharge times are controlled in common by the input. The first-mentioned circuits produce sawtooth waveforms across the capacitor and short duration output pulses, while the lastmentioned commonly develops a triangular wave across the capacitor, in association with a square wave at a separate output.

These ideas are illustrated in Figs 1 to 4. In Fig. 1, constant current results in a constant rate-of-change of voltage across the capacitor, i.e. the time taken to charge to a given p.d, will be inverse to the charging current. If that level can be sensed and caused to end the cycle or half-cycle, then the repetition frequency (being inverse to the period of the waveform) will be proportional to the current and a linear $I / f$ converter results. The simplest way of causing the cycle to recommence is to place a lowvalue resistor across the capacitor to discharge it in the shortest possible time. If the discharge current is large compared to the charging current, then it is immaterial whether the charging current is disabled or not and Fig. 3 represents the basic principle of many $V / f$ converters, with the switch periodically closing at the instant when the p.d. across the capacitor reaches a defined value.

An alternative principle is shown in Fig. 4. The current generator is applied to the capacitor in the reverse direction giving an opposing slope to the ramp but of equal magnitude. The resulting waveform is triangular with the repetition frequency linearly related to the current if the points at which switching is initiated are defined. The provision of a purely electronic twopole change-over switch is difficult, and the reversal of current direction is more often achieved by using a single-pole switch or its equivalent to control the carrent generator directly.

A second problem that often arises is that the changing p.d. across the capacitor affects the nominally constant current. This is obvious in terms of the non-linearity of the ramp, but may not affect the linearity of the


Fig. 5. Using the charging capacitor in an op-amp integrator ensures current is independent of capacitor p.d.


Fig. 6. Simple form of triangular. wave generator uses principle of Fig. 5.


Fig. 7. Circuit provides equal $+V$ and -V inputs for Fig. 6 with an op-amp of -1 gain.
$V / f$ function provided the waveshape is well controlled, e.g. accurate $V / f$ conversion is possible with simple $R-C$ charge and discharge circuits though the wave shape is highly non-linear. Where waveshape is also of importance, the capacitor forms part of an operational amplifier integrator circuit, with the virtual earth action ensuring that the charging current is independent of the p.d. The discharge element now has no point connected to ground which can raise problems in activating it. (Fig. 5.)

This technique leads to a simple form of triangular-wave generator shown in Fig. 6 where both the $+V$ and $-V$ inputs have to vary together if the slopes are to remain of equal magnitude. By using both the input and the output of an amplifier with a voltage gain of -1 this is readily achieved (Fig. 7). Alternative methods include the design of amplifiers whose voltage gain is switched from +1 to -1 , and of integrators in which the direction of capacitor current is reversed by a switch while the magnitude is controlled by a single input voltage.

In all of these circuits there remains the problem of the level sensing circuitry that is to determine the instant of switching; both switching speed and accuracy of level are important making the design of a fast, accurate $V / f$ converter a difficult one.

The term charge-dispensing is a big one in the literature on precision $V / f$ converters. A feedback system is set up in which the output pulses from a generator (basically
monostable in form) are arranged to feed back a constant amount of charge for each output pulse. If these units of charge are combined at the input of the system with the control signal, and the overall feedback is negative, then the pulse rate will be proportional to the control signal.
In block-diagram form in Fig. 8, the principle is illustrated by a combination of $V / f$ and an $f / V$ converter. Assuming that the amplifier gain is high, and that the $f / V$ convertor is very linear then the feedback overcomes any non-linearities in the $V / f$ converter, i.e. $V_{o}=V$ to a high accuracy because of the feedback while $V_{o} \propto f$ ensuring that $f \propto V$ without reference to the linearity of the $V / f$ converter. The $f / V$ converter might be of the diode-pump variety which with suitable design can transfer a fixed charge into a load for each output pulse rate.

A level-sensing monostable gives an output pulse when the input level rises above a critical value. If the input then falls a second pulse is generated on the next excursion through the set level in the same sense. An important restriction is that the capacitor shall have been completely discharged prior to the second pulse-otherwise the time taken for recharging will be shortened and the output pulse-width reduced. The output of such a monostable would ideally be a train of constant-amplitude constantwidth pulses, which could be smoothed and fed back to the input amplifier as in Fig. 9.


Fig. 8. In this "charge-dispersing" system, a constant amount of charge for each output pulse is fed back so that pulse rate can be proportional to the control signal.


Fig. 9. Monostable circuil produces output pulse when input exceeds a certain level, in either sense.


Fig. 10. An alternative arrangement is to dispense charge into a summing integrator. Output pulse rate is a linear function of control voltage or current.


Fig. 11. Technique of using two current sources, but switching only the one having twice the value of the other, is used in some i.cs.


Fig. 12. Triangular wave generator using technique of Fig. 11 . Comparator reference inputs are set to $2 \mathrm{~V} / 3$ and $\mathrm{V} / 3$, the capacitor voltage ranging between these limits.

A better arrangement dispenses these units of charge into a summing integratorFig. 10. For positive pulses a negative control voltage is required, the integrator output ramping up until a pulse is produced from the monostable. The charge dispensed into the summing junction causes the output of the integrator to fall, again rising slowly under the action of the control current. On average, the net charge inflow has to be zero, the charge dispensed per pulse is constant and hence the pulse-rate is a linear function of the control voltage/ current.

Other recent i.cs revert to the separate constant current circuit for timing circuits and waveform generators, and the resulting I/f linearity can be accurate enough for many applications. One technique is to have two current sources one set by the external control voltage, the other of opposite polarity but of twice the magnitude-Fig. 11. Keeping the former permanently on and switching the latter on and off makes the net current in the capacitor change from $+I$ to $-I$. A circuit configuration to use this technique to produce a triangular-wave generator is shown in Fig. 12.

Two comparators sense the capacitor voltage, their reference inputs being set to $+V / 3$ and $+2 V / 3$ by an internal potential divider. Assume the current at $I$; the capacitor charges until its p.d. reaches $+2 V / 3$. Comparator 1 changes its output and resets the flip-flop. This reverses the direction of current flow until the capacitor discharges to $+V / 3$. The comparator 2 operates setting the flip-flop into its original state and restarting the cycle.

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# Vision cassette and cartridge recorders 

## Facilities and performance of models on the UK market

In attempting to assess the current state of domestic, industrial and educational video activity, one is reminded of the sub-title of a recent article on the computer industry. It read: "Where âre we now, and how did we get into this mess?". For it seems that commercial and political considerations dictate that each new development is attended by a flurry of alternative ap-proaches-some only slightly different to each other-and that the eventual emergence of one or two practical solutions to the problem can take many years. It is all very wasteful, expensive and uncivilized, but nonetheless entertaining.

It seems likely that John Logie Baird, having worked out his system of seeing at a distance which, with fine impartiality he named "tele-vision", was the first to record a picture. There was, in 1927, nothing new needed to do this, as he simply used a 10 -in, 78 r.p.m. record and called it"Phonovision". Magnetic recording was not well developed and it was not until the 1950s that an acceptable, recorded image was possible. In 1951, Crosby Enterprises were using a longitudinally-recorded tape at $100 \mathrm{in} / \mathrm{s}$ for black-and-white pictures, in which the spectrum was separated in 10 bands, each being recorded separately, with two more tracks for control and sound. This was followed rapidly by RCA in 1953 with a longitudinal system capable of recording colour at $240 \mathrm{in} / \mathrm{s}$ on three tracks for RGB and two more for sync and sound. The longitudinal method, wherein the video tracks were recorded along the length of the tape as in audio tape recorders, was wasteful of tape (high speeds of up to 360 in/s for adéquate bandwidth) and caused problems of speed control, particularly in colour. Head-to-tape contact was difficult to ensure and even at high speeds, the theoretical maximum bandwidth of the tape recording process ( 10 octaves) is insufficient for the 20 or so octaves of a vision signal.

Ginsburg and Anderson, together with a man named Dolby, of Ampex, originated the modern approach to vision recording in 1956 with the transverse-track recorder, using frequency-modulated vision signals to avoid the bandwidth problem. In this principle, the tape is slowed to a canter and the tape-to-head speed is maintained by moving several heads. across the tape, giving transverse tracks. Four heads were
used by Ampex, and the term "quadruplex" was applied.

From then on, the transverse-scan recording method was to become standard throughout television broadcasting, using tape up to two inches wide and eventually producing a picture indistinguishable from the original. Much programme material is now transmitted from tape. Only the BBC continued the longitudinal method in


Fig. 1. The IVC cartridge of 1-in tape is shown at (a) and the tape path in the machine is at (b).
"VERA" (1956), but soon acknowledged that this was not the way to do it.

All this time, the idea of the domestic and educational use of television recording was being pursued, albeit rather spasmodically. RCA had a $\frac{1}{4}$-in tape system for home use in 1956, and throughout the 50s and 60s one saw optimistic announcements from time to time that the ideal had been achieved, but they all sank without trace. Many systems have been tried, but the "electronic" kinds have now narrowed to several types of tape recorder and a few systems using discs, optical and electro-mechanical. Several manufacturers produce tape systems with open-reel tape handling, but our impression is that, for domestic and educational use, the open-reel machine has had its day and that the enclosed tape storage machine will reign supreme within two or three years. Seven thousand are said to be in use in the UK now.

The tape enclosures take three forms. A cassette, familiar in one form as the audio cassette, possesses two spools mounted either side-by-side as in the Sony machines

or one on top of the other, as in machines made to the Philips pattern. A further type of tape enclosure is the cartridge, which possesses only one spool and is analogous to the 35 mm film cassette (we should, perhaps, have mentioned that it is all very confusing!) in which the tape is pulled out of the enclosure, past the heads and on to an external take-up spool contained in the machine. When the tape is used up, it is rewound into the cartridge and ejected, a process which points to one disadvantage of the cartridge-it cannot, unlike the cassette, be removed until the tape is rewound.
There is, as yet, very little standardization in the use of enclosed-tape machines. They differ in enclosure type, tape width, signal-processing, tape type and many smaller parameters. Sony and Philips are the leaders in their own fields, and there is standardization in machines using these two different systems as there is in another group, the EIAJ v.c.r. standard used by Matsushita and several others. Both types use helical scanning, which is a half-way stage between longitudinal and transverse scanning. The tape is pulled out of its enclosure and wrapped part of the way round a drum, rising or falling on the way round. The drum is provided with two, three or four heads, which revolve inside the drum about a vertical axis, "looking" at the tape through a circumferential slit in the drum.

As the tape is slightly inclined, moving on a helical path round its part of the drum, and the head axis is precisely vertical, the tracks recorded on the tape are inclined at about $3^{\circ}$ to the horizontal. As one head finishes its track, the next one starts the next track and the effect is as though there were one continuous track, recorded at high speed instead of the five or seven inches per second of the actual tape speed. In this way, a low tape speed provides band width of up to 3.5 MHz and a horizontal picture resolution of up to 360 lines. Either two audio tracks or one audio and a control track are recorded along the edges of the tape in the normal way.

Signal processing is rather more complicated than in the ordinary audio tape recorder, particularly when a colour signal is being handled. As transmitted, the broadcast colour signal consists of a vestigial-sideband luminance carrier with the upper sideband extending to 5.5 MHz , and a chroma signal with a suppressed sub-carrier extending from 3 MHz to 5.5 MHz . Neither of these signals can be handled directly by the tape machine and must be turned into a recordable form.

The chroma signal is simply transposed to a centre frequency of 562.5 kHz (in the Philips system) with a bandwidth of 650 kHz and recorded in the normal way as a.m. The reduced bandwidth, and hence

resolution, does not have as serious an effect on colour as it would on the luminance information which determines the sharpness of the picture. Luminance is not recorded directly but is remodulated as f.m. with a deviation of $3-4.4 \mathrm{MHz}$, thereby avoiding the effects of imperfect head-totape contact and completely eliminating the need for tape bias, as the waveform is no longer important. Bias for the chroma signal is automatically provided by the luminance f.m. signal, the two being combined in the recording amplifier.

Problems are introduced by the transposition of the chroma signal to a different frequency and also by inevitable phase jitter in the tape transport. This would, of course, be disastrous for the chroma decoder and would also result in an increased amount of sub-carrier patterning on the screen due to the loss of interleaving of chroma sideband energy peaks between those due to time-base repetition rates. Circuitry is therefore needed to overcome this defect, and a description appeared in Wireless World, December 1972.

On playback, the luminance information is passed to an f.m. detector and the chroma is reinstated in its proper position at 4.43 MHz , prior to being impressed as modulation on a u.h.f. carrier and passed to the aerial socket of the television receiver. Not every recorder possesses an r.f. output and if the output is at video frequencies, modifications to the receiver are needed. Many receivers will need modification for other reasons in order to be compatible with video recorders. For instance, the fly-wheel time-constant will need to be shorter to accommodate the "drop-out" time-the time between one head finishing a track and the next one starting. During the time when no signal is being played back, the flywheel will try to compensate unnecessarily, only to be caught on one leg by the arrival, on time, of the next set of information. The result will be "hooking" or bending of verticals at the top of the picture as the time-base slowly comes back into sync and this effect

Fig. 2. Loading a Philips cassette. Lowering the cassette engages the pins behind the tape and takes it round the drum.
Fig. 3. A Sony cassette is loaded in roughly the same way as the Philips type.
Fig. 4. An IVC cartridge being loaded.

can, without modification, reach half-way down the screen. The time-constant must therefore be shortened so that the hooking occurs invisibly during the blanking time or during a small amount of over-scan. If, however, it is shortened too much, the object of having a flywheel is lost and noise again becomes a problem.

It seems possible that future television receivers will make some provision for the connexion of recorders, preferably in video form, thereby eliminating the cost of a u.h.f. modulator. There will then, of course, be the old question of live chassis, as manufacturers still have not found it necessary to use mains transformers. A. C. Smaal of Philips set out his views on this in Wireless World March 1975. On this question of compatibility, it should be pointed out that the different systems are mutually incompatible. The two cassette systems-Sony and Philips-are possibly the closest in conception, but are still incompatible because of the different tape width and cassette type. Compatibility between two machines of the same model is better than it initially was; control circuitry is improved and there does not appear to be an insuperable problem. Dealers have told us that they can choose any machine in stock and play any tape on it with every chance of success.

Some recorders possess their own u.h.f. tuner, which means that the recorder can receive and storeinformation on one channel while the television receiver is displaying another. Others take a video feed from the receiver, necessitating yet another modification.

The feeling expressed fairly freely is that off-air recording is not going to be enough to make a success of these machines. A supply of programme material is essential if they are to enjoy the success of an audio system, but the number of competing formats must be drastically reduced before any programme supplier is likely to commit himself. There is also the question of copyright. It is, after all, an infringement of the Copyright Act of 1956 to record a broadcast programme. The Whitford Committee are unlikely to report for some time and this unenforceable law will continue to be broken daily, but it is an unsatisfactory situation.

The other source of "programmes" is to buy a monochrome television camera (colour is far too expensive a proposition) and to use the camera and recorder as a kind of up-market home movie system, but one would have to be very single-minded about immortalizing Dad and the kids on tape to go to such lengths.

The facility of stop motion or still frame is obtained by stopping the tape feed, while the heads continue to turn. As each track contains one field of information, the same field is scanned continuously. There is a slight problem in that the heads do not now cross the tape at precisely the same angle as when it is moving, so that they may start on one track, cross the guard band between tracks and finish on another. A drop-out then exists and it is necessary to ensure that this drop-out occurs in the blanking interval between frames. Most
machines incorporate a drop-out compensator, the Philips type consisting of a dropout detector which, when a defect is noted, substitutes for the line of information containing the drop-out a previous line, delayed by $64 \mu s$. The difference between the two is usually negligible and preferable to a total loss of information.

Our impression of the two "standard" machines is that the Philips, being smaller, cheaper and possessing a tuner and timer is better suited to the domestic scene than the Sony, but that Sony's performance is a little better and should be more at home with a camera input for education and training.

## Other systems

Although this article is intended to cover methods of video recording using tape in "convenience" form, it is well to note that several other contenders exist which use discs-an area of activity on which we intend to report in detail in the near future. Most of these (Thomson-CSF, Zenith, Philips/MCA) use optical methods in either the transmission or reflection modes and have playing times of between 20 and 30 minutes. The records are thin plastic or glass discs and the information is encoded in the form of pits or holes, which tends to render them somewhat vulnerable to dust and grease. The recent Philips/MCA link would appear to give the VLP (video long player) a distinct advantage over othersMCA have a vast library of material and are to manufacture the discs.

The Telefunken-Decca system (TeD) uses what is effectively an up-rated audio record of 8 -in diameter playing for ten minutes. Hill-and-dale recording is used. The disc systems were described in detail in Wireless World, November 1973.

A recently-announced development from BASF is the LVR (Longitudinal Video Recorder) which again employs a singlespool enclosure. As its name implies, longitudinal recording is used, but the extremely high rate of tape usage common to this method is avoided in the LVR by the 28-track format employed. Quarter-inch tape is used and playing time can be as long as 120 .minutes using $6 \mu \mathrm{~m}$ thick tape. Little is known of this machine at the time of writing, except than an unusual tape handling system is used. The cartridge opens to reveal the spool of tape, the leader being extracted by a large-diameter capstan, passed through the recording/playback station and again past the capstan onto the take-up spool. Feed spool, capstan and take-up spool are in continuous contact, leaving very little free tape. The extremely thin tape is therefore protected.

A speed of $3 \mathrm{~m} / \mathrm{s}$ is adopted, the reversing at the end of each of 28 passes taking 80 ms . Colour recording is offered in conjunction with several audio tracks. Bandwidth is 3 MHz . BASF claim that the area of tape used is less than a quarter of that used in the Sony system and even less than in the Philips method. The unit is not expected to make its appearance for at least two years.

A recent announcement is the MDR, developed by Erich Rabe. MDR is Mag-
netic Disc Recording and offers the facility of recording to the user-unlike the optical or stylus-pickup discs. An ordinary record turntable modified to run at 200 r.p.m. carries a disc whose inner section has a helical guidance groove which guides a stylus and, by a link, steers a magnetic head over the outer, magnetic, section of the record. All colour systems can be recorded and played for 15 minutes. Alternatively, the turntable can be slowed to 33 r.p.m. and used to record up to 16 hours of audio.

The RCA Selectavision Magtape, not yet available here, uses a new type of tape handling and head format. Four heads are used, a layout which, amongst other benefits, allows all the tape to remain in the two (side-by-side) reel cassette, as only $90^{\circ}$ of the drum must be wrapped. The drum protrudes into the cassette to achieve this amount of wrap. Cassette size is $9 \times 6 \frac{1}{4}$ $\times 1 \frac{1}{2} \mathrm{in}$.

In the following section, the machines mentioned are the ones we have found to be available in the UK. There are many more, but they are not obtainable here and so have been omitted.

## Philips <br> N1500 VCR

Cassette-loading
(vertically-stacked
spools)
Record/playback of colour and monochrome
Tuner for off-air recording
U.h.f. modulator output

PAL standard 625/50
Two heads
$\frac{1}{2}$-in tape
Automatic tape threading
Cassette size: $12.7 \times 14.6 \times 4.1 \mathrm{~cm}$
Recording time: 30,45 or 60 min
Bandwidth: 2.7 MHz
Tape speed: $14.29 \mathrm{~cm} / \mathrm{s}$
Dimensions: $56 \times 33 \times 16 \mathrm{~cm}$ (with cassette lid up)
Sound is on two tracks on tape edges
Mains supply: $110-245 \mathrm{~V} \pm 10 \%$ at 50 Hz
$\pm 1 \%$. Any frequency drift must be slow to remain tolerable
Price: $£ 462.84$ (plus v.a.t.) (1500)
cassettes: $£ 11, £ 14.50$ and $£ 17.00$
( $1500 / 15 \mathrm{M}$ ): $£ 537.04$ (plus v.a.t.)
This is one of the group leaders in these machines. It provides an acceptably sharp picture but not, perhaps, as finely resolved as in some others. It must be said that most dealers tend to demonstrate Philips machines on large-screen receivers, whereas other systems seem to be shown on small Trinitron sets-a procedure which does emphasize the difference. Controls are provided for tracking, audio record level auto/ manual, the usual function controls and a timer for use when recording a programme in one's absence. Cassettes are available for 30,45 or 60 minute playing times; the cassette holder is raised, the cassette inserted and the whole lowered, thereby engaging pins which pull out a loop of tape and wrap it round the scanning drum. The $1500 / 15 \mathrm{M}$ is similar, but the input and output are at video frequencies for direct connexion of colour or monochrome camera
and monitor. The 1520 , at $£ 820$, is a semiprofessional machine with no timer or tuner, having assemble and insert editing provision, and facility for sound dubbing on two sound channels. It has an extended bandwidth $(3.2 \mathrm{MHz})$ for monochrome, and is intended for a video input from a camera. Stop motion is possible, and the output can be either video or u.h.f. Head life on this range of machines is up to 500 hours with chrome tape, and replacement during the first year is free; after this the cost of new heads is $£ 35-£ 40$. The N1500 and $1500 / 15 \mathrm{M}$ are handled by Philips Electrical Ltd, Century House, Shaftesbury Avenue, London WC2H 8AS, and the N1520 by Pye Business Communications Ltd, Cromwell Road, Cambridge CB1 3HE.

## Radio Rentals

## Model 8200

This is based on the Philips N1500. The performance and appearance are the same essentially, but the programme timer and u.h.f. tuner are not included. The 8200 can be bought or rented (not privately) from Radio Rentals Contracts Ltd, 1/15 Clyde Road, Tottenham, London N15.

## Sony

## Vo-1810UK

Cassette: (Sony U-matic, spools side-byside)
Record/playback colour and mono-chrome U.h.f. output modulator/video input

PAL standard on 625/50
Two heads
$\frac{3}{4}$-in tape
Auto tape threading
Cassette size: $3.3 \times 22.1 \times 14 \mathrm{~cm}$
Recording time: 20, 30 or 60 min
Resolution: 300 lines monochrome, 240 lines colqur
Tape speed: $9.53 \mathrm{~cm} / \mathrm{s}$
Dimensions: $67.6 \times 23.8 \times 46.6 \mathrm{~cm}$
Two sound channels
Mains supply: $110-240 \mathrm{~V} \pm 5 \%$ at 50 Hz $\pm 0.5 \%$
Price: $£ 765$ (plus v.a.t.)
Cassettes: $£ 8.20, £ 9.80$ or $£ 14.50$
The leader of the U-matic group of recorders, that being a Sony trade mark. The operation of the two types of machine is broadly similar in all external essentials. The VO-1810UK uses chrome tape and possesces a tape winding memory feature, which enables continuous repeat of a full tape or part of a tape, starting and finishing points being pre-set by the operator. Sound dubbing on one channel is provided for. A u.h.f. tuner, type TU1000B is available for off-air recording at $£ 99$.

## VO-1210

A similar machine to the VO-1810UK but intended solely for play-back of recorded cassettes.

VO-2850 U-matic is a semi-professional machine in the cassette format with assemble and insert editing facilities and sound dubbing. Stop-motion is offered. Price is $£ 2,500$.
Sony (UK) Ltd, Pyrene House, Sunbury-on-Thames, Middx.


Philips N1500


Radio Rentals 8200


Sony V0-1810UK


## Action Video

## Rentals



If you can't justify purchasing all the video equipment you occasionally need why not rent it from us. We have a vast stock of all types of video recorders, cameras etc.
(including complete monochrome and colour portable studio units).

Action Video
45 Great Marlborough St London W1V 1DB
Phone 01-734 7465/7
Midlands Representatives:
Foxall \& Chapman,
51 North Street,
Cheetham, Manchester.
Phone 061-834 5786



## PHILIPS N1500 VCR

Low cost video cassette recorder for up to 1 hou recording time. With built-in tuner and fime clock Available from stock. SPECIAL OFFER $\mathbf{~} 425$ plus VAT N1500/M VIDEO INPUT AND OUTPUT VERSION. Available from stock. $\quad 539$ plus VAT N1520 ELECTRONIC EDIT VCR
Complete with electronic editing facility with insert and assemble modes • 2 audio tracks • extended band width • stop/motion facility. Available from stock. £850 pius VA

## AKAI GREAT SAVING!

For one month only we are able to offer the Aka VTS-110 DX complete $\frac{1}{4}{ }^{\prime \prime}$ mono kit for an amazing $\mathbf{£ 2 5}+\mathrm{V}$ +.T. This is a saving of $£ 150$ on the normai price.
Incredibly versatile, the kit comprises portable video tape recorder, video camera, portable monitor and adaptor.
Power source is no problem. You operate from built in battery, mains, or a car battery. High quality pictures can be relayed from camera, "off air" or other VTR equipment, and can be played back through clip-on monitor, VTR monitor, or any UHF receiver.

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## SONY V01810 UK

PAL NTSC U-MATIC RECORDER PLAYER High resolution colour, auto repeat and memory facilities, 2 audiotracks and R/F output. Available from stock. $\mathbf{E} 765$ plus VAT

SONY VP1210 UK PAL NTSC PLAYER ONLY Available from stock. $\mathbf{£ 6 4 5}$ plus VAT

## Video Tapes

We offer highly competitive prices on all makes of $\frac{1}{2}$ in., 1 in., VCR and U-Matic Video Tapes including Sony, Scotch, Ampex, BASF, Philips and Memorex. Next time you are ordering tapes, ring our Sales office at Colliers Wood and let us quote you. You'll be surprised.

REW have been in the Video Industry for over 10 years and their accumulated wealih of experience offers you the finest Video service in the country. All the equipment you need is always available from stock and at their London Video Cassette Centre at Centrepoint in London you can view and compare all the latest equipment REW also offer first class studio and production facilities. Why not contact us when you want to talk video - REW know better than most. REW are Main Agents for:
AMPEX, AKAI, NATIONAL. PANASONIC, SONY, HITACHI-SHIBADEN. J.V.C NIVICO, MALHAM LIGHTING, ELECTROCRAFT, QUICKSET TRIPODS. ASTON, FUJINON, CANON, DECCA, RANK IANIRO, LIGHTING
In our Centrepoint showroom we have a permanent demonstration of the full range of Shibaden colour cameras, all available from stock.

DES COLOUR 18" RECEIVER/MONITOR


The DES monitor is based on the SONY KV1810 UB $18^{\prime \prime}$ Colour Receiver and provides full video and audio input and output facilities plus an EIA-JVTR connector
This high quality monitor is compatible with most video tape recorders and video Cassette recorders currently available on the market.


## $\left.\begin{array}{c}\text { available } \\ \text { NOW } \\ \text { at ONLY }\end{array}\right] 36$

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London Showroom - Centrepoint, 21 St. Giles High Street, London WC2.
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National NV-5120A-B


## JVC CR-6000E



Hitachi Shibaden SV-630E(K)

## National (Matsushita)

## NV-5120A-B

Cartridge loading (one reel-one in machine) Record/playback of colour and monochrome U.h.f. output converter as optional extra

PAL standard 625/50
Two heads
$\frac{1}{2}$-in tape
Automatic tape threading
Cartridge size: $12.8 \times 13 \times 2.9 \mathrm{~cm}$
Recording time: 36 min
Bandwidth: $4 \mathrm{MHz}(-20 \mathrm{~dB})$ monochrome resolution 240 lines, $3 \mathrm{MHz}(-20 \mathrm{db}$ ) colour (resolution 260 lines)
Tape speed: $16.322 \mathrm{~cm} / \mathrm{s}$
Dimensions: $48.5 \times 38.8 \times 20.8 \mathrm{~cm}$
Mains supply: 240 V 50 Hz
Price: $£ 595$ (plus v.a.t.)
Tape: $£ 8.50$ for 36 mins (National tape)
A cartridge recorder which conforms to the only "standard" in existence, if one defines a standard as a specification arrived at by agreement rather than by force majeure. The A-matic cartridge uses a single reel, the tape having a stiffened leader which is automatically pulled past the heads onto the tape-up reel. A programmer for repeated playing of selected parts of the tape (search) is provided and there is provision for stop-motion. Controls are solenoid-operated. A timer is obtainable as an extra, as is a remote-control unit. Sound can be dubbed.

A point to note about the EIAJ $\frac{1}{2}$-inch cartridge is that it can be loaded by the user with $\frac{1}{2}$-inch tape, so that tape recorded on an open-reel machine can be used in cartridge form.
Collett Dickinson, Pearce \& Partners Ltd, Howhand House, 18 Howhand Street, London W1P 6AT.

## JVC

## CR-6000 E

Uses the U-matic $\frac{3}{4}$-inch cassette and is compatible with other U-matic equipment. Records and plays back in colour or monochrome with an r.f. or video input and video output (u.h.f. converter as an extra). It possesses the "search" facility, solenoidoperated controls, audio dubbing and two sound channels. A remote-control unit is an accessory. Playing time is up to 60 minutes. Price $£ 749$ plus v.a.t.
Bell and Howell, Alperton House, Bridgwater Road, Wembley, Middlesex.

## Hitachi Shibaden

## SV-630E(K)

A cartridge machine conforming to the A-matic (EIAJ) $\frac{1}{2}$ inch standard of the National NV-5120A-B, with a similar specification and range of facilities. A sound-dubbing facility is provided, as is automatic re-wind. The price is $£ 580$ plus v.a.t. The price of tape is $£ 12$ for 36 minutes (Shibaden tape).
Shibaden (UK) Ltd, Lodge House, Lodge Road, Hendon, London NW4.

## Loewe-Opta

## Optocord 700

Basically similar to the Philips $1500 / 15 \mathrm{M}$ but with r.f. and video in and out. Auto tracking is provided, as is drop-out compensation and a colour killer which operates on playback only, thereby
avoiding the possibility of recording colour in black and white. A sevenselector r.f. tuner is included, and a timer, and the machine offers stop-motion operation. Price $£ 744.17$ plus v.a.t. and the Optocord 700 uses the same type of cassette as the Philips machines. Hokushin Audio Visual Ltd, 2 Ambleside Avenue, London SWI 6 6AD.

## IVC

## VCR-101C

Cartridge loading (one spool)
Record/playback in colour
Video input and output
PAL on 625/50
One head (tape completely encircles drum)
1 -in tape
Automatic threading
NAB 8-in reel mounted in cartridge
Recording time: 60 min
Bandwidth: 3.2 MHz luminance, 1.4 MHz colour
Tape speed: $17.1 \mathrm{~cm} / \mathrm{s}$
Dimensions: $18 \times 13.5 \times 8.5$ in
Two sound channels
Mains supply
Price $£ 1812$ plus v.a.t.
Tape around $£ 21.00$ per hour
Yet another cartridge machine, this time using a standard 8 -in NAB reel of 1 -in tape,


## IVC VCR-101C

with obvious compatibility with open-reel machines. Stop-motion is provided and the instruments (this is one of a range) are fitted with audioamplifiers and speakers. The machines are available in monochrome or PAL versions, and are controllable
electrically by t.t.l.-compatible voltage levels. The scanning mechanism is directdriven, having its own printed-circuit motor. Head life is claimed to be 2000 hours.
Bell and Howell, Alperton House, Bridgwater Road, Wembley, Middlesex.

## Literature Received

A booklet entitled "Photocouplers" is now avail able from Mullard, describing the characteristics, operation and application of these devices. Requests for copies, on headed notepaper, should be sent to Computer Electronics Division, Mullard Ltd, Mullard House, Torrington Place, London WCIE 7HD.

## EQUIPMENT

A leaflet describing a deglitched d.a.c. system, the DMC Digisweep, which takes in digital data and drives c.r.t. deflection amplifiers to draw vectors and write alpha-numerics, is available from Amplicon Electronics Ltd, Lion Mews, Hove BN3 5RA ...................................WW41I

A publication is available illustrating and giving technical data on radiotelephones types M202 (v.h.f.) and M212 (u.h.f.). Pye Telecommunications Ltd, Cambridge CB5 8PD ...................WW412

A basic guide to data communications is the subject of a new brochure relating to computer installations where remote control terminals are connected by telephone lines to a control computer. SE Labs (EMI) Ltd, Spur Road, Feltham, Middlesex WW413

A catalogue from Burns describes a range of equipment intended for the amateur radio market, including a frequency standard, wavemeter, test oscillator and many modules for building into other equipment. Burns Electronics, 43a Chipstead Valley Road, Coulsdon, Surrey CR3 2RB ........ WW4 14

The Spellman range of high-voltage power supplies is shown in a catalogue 7400 from Hartley. Solid-state, regulated and unregulated, miniature, rack-mounted and modular units are described. Hartley Measurements Ltd, HML House, London Road, Hartley Wintney, Basingstoke, Hants. WW415

We have received a leaflet on the lightweight v.h.f./ f.m. marine radiotelephone, Model RF-440, made by Harris. Complete performance details and facilities provided are described, as are several accessories. Harris Corporation, RF Communications Division, 1680 University Avenue, Rochester, New York 14610, U.S.A. ................. WW416

Tally have sent us a leaflet on their latest range of paper tape peripheral equipment, which is designed for use with the PDPI1, Nova and Digico 16 V minicomputers on a plug-in basis. The.leaflet describes a reader and two punches for low and high speeds. Tally Ltd, 7 Cremyll Road, Reading RG1 8NQ ...................................WW417

Farnell have produced a leaflet on the PG5000 series of five pulse generators. Types 5111 to 5222 provide between them, dual channel output with delay or positive or negative-going pulses. Repetition frequency is up to 5 MHz . Farnell Instruments Ltd, Sandbeck Way, Wetherby, Yorkshire LS22 4DH ......................................WW4 18

We have received a leaflet describing a loudspeaking intercommunication using mains-borne f.m. or a.m. and made by the NOA Corporation. The unit is the Model FN-113S and the leaflet is obtainable from Hadley Sales Services, 112 Gilbert Road, Smethwick, Birmíngham
. WW419
The latest Heathkit catalogue is now available. New equipment this time includes a digital clock/ radio, a scientific calculator, a car clock, a 15 MHz oscilloscope and a function generator. Heath (Gloucester) Ltd, Gloucester GL2 6EE $\therefore .$. WW420

A leaflet describing the Digipet electronic weighing machine for top-loading is available from Transducers (CEL). The weighers are by Shinko-Denshi,
provide a digital indication and automatically select ranges of $0-19.99 \mathrm{~g}$ or $0-199.9 \mathrm{~g}$. Transducers (CEL) Ltd, Trafford Road, Reading RGI 8JH ......................................WW42I

The recent informative advertisements for Wayne Kerr have been reprinted in booklet form entitled "Some Notes on Bridge Measurement". The publication is obtainable free from The Wayne Kerr Company Ltd, Durban Road, South Bersted, Bognor Regis, Sussex PO22 9RL

Peerless loudspeaker kits and drive units are described and pictured in a leaflet from Ross Electronics, 32 Rathbone Place, London W1P IAD

WW423
A description and specification of the Philips time division multiplexer type 3 TR 1500 is given in a brochure from Philips' Telecommunicatie Industrie BV, PO Box 32, Hilversum, The Netherlands WW424

## MATERIALS

Data sheets describing the applications for and properties of four new silicone resins specially developed for use in the electrical and electronics industries are available. The resins M15 and P22 can be used for binding high-temperature-resistant impregnating varnishes, while PO5 and P15 are additives for use in the manufacture of base cements for electric bulbs. TH Goldschmidt Ltd, York House, Station Road, Harrow, Middlesex .. WW425

A booklet from DuPont describes the company's range of products for the manufacture of microcircuits, optoelectronics, and potentiometers, together with basic information on thick-film compositions. R. G. Paterson, DuPont Information Service, DuPont (UK) Ltd, 18 Bream's Buildings, Fetter Lane, London EC4A 1HT .......... WW426

In a new leaflet, EGM Solders give full details of their ranges of solders, fluxes and chemicals. (EGM is the amalgamation of Enthoven, Grey and Morton and McKechnie.) EGM Solders Ltd, Wolseley Road, Mitcham, Surrey CR4 4JQ WW427

# Transistor-aided ignition 

## A simple solid-state switch for ignition coils

by G. F. Nudd

The contact breaker is, in the author's opinion, the bugbear of a modern car. Many vehicles require the contact breaker to be adjusted, if not replaced, every three months or so. In a recent survey by the Automobile Association one in 15 breakdowns was found to be caused by points failure. Various types of electronic ignition have been designed to overcome the drawbacks of standard systems, notably the capacitor discharge method. However, in the case of mass-produced cars, these systems could be considered overdesigned as they are generally costly, usually requiring a special transformer. Also, in some cases, electronic revolution counters cease to operate correctly.

As a car works perfectly well when the points are in good condition and correctly adjusted all that is needed is an electronic switch to isolate the points from the heavy current and high-voltage backswing of the ignition coil. Until recently transistors capable of the 300 V or so needed were not readily available. Now one can obtain the so-called "triple diffused device" that not only offers high-voltage operation but a much better second breakdown region because of its higher switching speed. The author has used the Texas BUY23/23A which, when operated with ten ohms between base and emitter, is capable of withstanding 600 V . Some designs have used a high-voltage, high-power zener diode across the transistor for protection. This, however, has been found unnecessary with the author's circuit.

Concerning the driver circuitry, normal amplifying stages have been used in some designs. This, however, gives rise to excessive power dissipations in components when a worst-case circuit is designed for operation between 7 and 15 V limits. To overcome this problem, a constant-current driver is used, which results in quite reasonable dissipations, and the design is suitable for all cars using a 12 V ignition system. If the car does not have a ballast resistor system, $R_{2}$ can be increased from 1.2 ohms to 2.2 ohms, giving less dissipation in the driver transistor. When using the positive-ground version, the ignition coil is connected to ground instead of battery voltage. The capacitor $C$ can be a 600 V , electronic type or alternatively a "points" capacitor as normally used in the car, the normal capacitor being left in situ to facilitate disconnecting the unit. The capacitor
should be soldered into the i.a. unit because if, for example, it became disconnected through a faulty slide connector, the ensuing high voltage might damage the transistor. Diodes $D_{4}$ and $D_{5}$ are protection measures for the transistors against voltage transients.
The i.a. unit may be built on a piece of aluminium and attached to the car body under the bonnet for heat dissipation. In the case of glassfibre cars the chassis must be used. Also modern aluminium oxide insulating washers for the power transistors should be used.
The points should be replaced and the engine timed accurately when the unit is fitted. The sparking plugs should be replaced or regapped as normal. It has been remarked that when electronic ignition is fitted there is no need to check the
ignition system. This may, in fact, be true with an older type of car but with a more modern one the engine, timing must be within a couple of degrees accuracy to obtain optimum power output.

The unit has been functioning in two cars for many months with no troubles. The points themselves wear slowly, both parts receiving slight indentations which causes the unused outer surfaces to gradually be used leaving the engine timing unaltered. The fibre surface of the points which rubs on the cam also wears to the extent of one or two thou at the points-gap in a year.

## Components list

Tr $\quad$ BUY23A/BUY23
Texas Instruments
or BDY96/97/98 Mullard
$\mathrm{Tr}_{2} \quad$ 2N3789/90/91/92
$\mathrm{Tr}_{3} \quad 2 \mathrm{~N} 3055$
$D_{1,2,3,4} 1 \mathrm{~N} 4001$
$D_{5} \quad 18 \mathrm{~V}$ zener diode 400 mW
$R_{1} \quad 56 \Omega 2 \mathrm{~W}$
$R_{2} \quad 1.2 \Omega 2 \mathrm{~W}$ or $2.2 \Omega 2 \mathrm{~W}$, see text
$R_{3} \quad 1000.5 \mathrm{~W}$
$C 600 \mathrm{~V}$ d.c. working, same capacitance as the points capacitor, see text.
Aluminium oxide TO3 thermal insulating washers:
2 off for negative earth A26-2004
1 off for positive earth
Jermyn
Industries


Complete circuit diagrams for positive- and negative-ground systems.

# National Electronics Council Link Scheme 

The NEC Link scheme has just entered its second year of successful operation. It is an organization devoted to linking schools wanting to start electronics projects with advisers based in industry and commerce. A good example of a successful link is described in their newsletter and which is reproduced below with their kind permission. Those wishing to contact Link should write to The Organiser, Peter Noakes, Link Scheme, Department of Electrical Engineering Science, University of Essex, Colchester CO4 3SQ.

## A link in operation

In October 1973, having received an offer of help from Mr Short, an engineer at Recording Designs (E.M.I.) Limited, Link Scheme put him in touch with Mr Ellerker, a teacher at the Robert Haining School, Surrey. Both were obvious electronics enthusiasts and after the initial introduction we retired to await the outcome. Following initial discussions concerning what each side expected to get from the link, it was decided to develop an introductory electronics course for 12 year olds. After considerable thought, careful design and preparation the course has now been introduced, and I was pleased to receive from the individuals involved in this link the following report. If you are interested in receiving more information, please contact directly any of the individuals mentioned at the end of the report.
Electronics at the Robert Haining School. The lives of most of us today are increasingly influenced by technological development; because of this we have organised a series of courses for our 12 year olds which expose them to a variety of technologies. The basic courses are intended to act as a stimulus, creating interest and enthusiasm.
Introduction to electronic work units. In the case of electronics a set of six work units offers the pupils the opportunity to gain familiarity with and confidence in handling components, plus intrigue and excitement through seeing and using their completed projects. They very soon show their newly gained knowledge through their ability to select resistors, capacitors, diodes and transistors with confidence.

In designing the units we had to look for efficient ways of producing attractive software which would involve the young
pupils at all stages. A short introduction describing the project and its possible uses is followed by an "items sheet" which involves the selection of components and the placing of them alongside their respective symbols on the sheet. On the next sheet is drawn a 1 cm square grid depicting the component positions as they appear on the actual circuit board and numbered and lettered to correspond with the items sheet. The pupils transfer the components from the items sheet to the grid. It is now a simple matter to transfer the components from the grid to identical positions on the circuit board.

The circuit board is made from white faced hardboard marked with a 1 cm square grid and numbered to assist in the transfer of the components. The components are held to the board by tension springs, mounted vertically, which may be stretched upwards to allow the component leads to be slipped between the coils of the spring. This technique is shown in diagrammatic form and is studied before the transfer takes place. A sheet of step-by-step instructions ensures that each component is placed correctly on the circuit board.

When the project is satisfactorily completed the pupils are required to fill in a questionnaire which is designed to test their understanding of the project.
Selection of projects. Selecting suitable projects for the six units of work was not a simple task. The choice was constrained by a number of factors, some dictated by the objectives of the course and others by practical considerations.

The most important objective is that the child should enjoy the work unit and this implies that each project should have a degree of novelty, such that when complete it is fun to use. A further implication of the "fun factor" is that the completed project must be guaranteed to work, provided the components are not faulty and are inserted in the correct positions. Many youngsters have been turned away from electronics as a pastime due to the repeated experience of building projects described in some of the many electronic magazines and finding they cannot make them work. To avoid this pitfall the circuits must be designed to tolerate wide variations in transistor gain, poor tolerance components and a variation in supply voltage consistent with battery operation.

Also, because battery supplies are used, current economy must be considered at the design stage.

A further objective of the course is to demonstrate a range of tasks to which electronics can be applied. However, certain categories of projects were not considered. For example, the obvious applications of electronics in radios and audio amplifiers were deliberately avoided. As 12 year olds do not own cars, electronic gadgets for cars were not included. Also, electronic test instruments were excluded because they have no appeal unless their purpose is understood. In all cases the theory of operation was not considered.

The projects finally selected were as follows:

1. Moisture detector
2. Simple electronic organ
3. Light beam burglar alarm
4. Sound operated switch
5. Two-way intercom
6. Reaction time tester

Future work. In order to provide continuity of work as the present group of 12 year olds moves up through the school, further courses will be developed. At present, consideration is being given to a set of work units based on circuit blocks such as multivibrators, amplifiers, level detectors etc. The object will be to demonstrate how a wide variety of tasks may be tackled by various arrangements of a small number of basic circuit blocks. At some stage it will be necessary to change from the "spring terminal" method of construction to the more conventional technique of soldering. To this end a work unit entitled "An Introduction to Soldering" is being produced, including a video tape presentation demonstrating the technique.

Anyone who would like further details of this work is welcome to contact either:
Ted Ellerker or Brian Burtsell,
Technical Studies Department, Robert Haining School,
Mytchett Place Road,
Mytchett, Surrey
or
Lawrence Short,
Recording Designs (EMI) Ltd., Victoria Avenue, Camberley, Surrey.


## Proposed changes to American licences

The long-awaited FCC proposals for the major "restructuring" of amateur licence conditions in the United States have now been outlined in a 29 -page document, Docket 20282. Among the many changes suggested is a 2000 -watt p.e.p. output power limit for those holding an "Advanced Class" permit, thus effectively doubling the already very high powers permitted in the USA. Amateurs with h.f. licences would be restricted to operation below 29.0 MHz until they obtain an "Experimenter" licence. "Novice" licensees would be able to use up to 250 watts d.c. input (for c.w.-only operation) instead of 75 watts, and these licences would be renewable in five-year terms. A new "Communicator" class of licence would not require a Morse code test and would permit use of all amateur frequencies above 144 MHz but restricted to frequency-modulation (F3). "Extra" class licences for h.f. and v.h.f. would require a 20 w.p.m. Morse test but no further theoretical examination. Extra facilities on 50 and 144 MHz would be given to "Technician" class licensees.

Generally it seems that the FCC wants to make entry into the hobby easier and to give newcomers more facilities, including new Morse-free licences, but would retain the existing "incentive" structure by providing progressively more operating privileges. The FCC has invited comment by June 16,1975 , so it will be some time before these proposals become effectiveand of course they may yet be modified.

## The r.t.t.y. facilities at ZS3B

Interest in radio-teleprinting continues to grow and many well-equipped stations are using this mode. But surely one of the most elaborate installations must be that of Gerhard Schlorf, ZS3B, in what used to be known as South-west Africa. The following description of his station appeared recently in Radio $Z S$ : "The station operates auto start on 14075 kHz and offers a number of facilities. In response to a code contained in the incoming 45 baud, 170 Hz shift signal a message generator responds: 'ZS3B attended' or unattended, whichever is the case, or
'ZS3B printing'. In response to a different incoming coded signal, a stored message can be activated. Another form of coded input signal records the incoming signal which, if ended appropriately, would by using a memory, switch on the transmitter, switch off the receiver and retransmit the incoming signal to another address.
"Another feature permits an incoming 7 MHz signal to be retransmitted at the same time on 14 MHz , and vice versa, to allow retransmission to another area.
"The installation includes two teleprinters and the whole station is operative 24 hours of the day with any incoming signal printed, with those signals addressed specifically to ZS3B printed on one teleprinter, so that the operator need not wade through reams of paper to see if anything has come in for him.
"The 14075 kHz frequency is crystal controlled and maintained to within $\pm 30 \mathrm{~Hz}$. The station forms part of a world wide amateur network."

## Good winter for "Top Band"

The low sunspot levels of activity which have restricted operation on 14 MHz (and above) fairly strictly to the hours of daylight this winter have brought compensating benefits to the considerable number of "Top Band" ( 1.8 MHz ) dx enthusiasts, to judge by the latest Bulletin from Stewart Perry, W1BB. He reports that many amateurs have this season completed the by-no-means-easy feat of achieving "worked all continents" on this band (KV4FZ even completed a WAC in a space of eight hours!). Much sought after have been VS6DO in Hong Kong and a growing string of stations in South and Central America. Helena de Kertesz, YV5CKR, after a visit to Europe and the United States returned to Caracas, Venezuela, to become possibly the only "young lady" operator currently active on 1.8 MHz dx , and has made many longdistance c.w. contacts. One of the new countries to appear on the band this winter was ST2AY in the Sudan, operated by Roger Crofts, G3UPK. The "first-timers" tests were handicapped by rather poor conditions, but the ARRL 1.8 MHz tests in December provided many excellent contacts particularly on the second night. Stew Perry, W1BB, has this season worked 150 dx stations in 46 countries compared with 116 stations in 37 countries in the equivalent season of 1973-74.

## 50 years of REF

This month, French amateurs are marking the 50th anniversary of the formation of the Reseau des Emetteurs Français in April 1925. This society-long-established as the French national society for radio amateurs-was by no means the first radio society to be formed in France; for example, in 1914 there was the "Groupe Français des amateurs de TSF" and others in the early 1920s included the rather sinisterly named "Club des 8". But in 1925, Jack Lefebvre, F8GL, invited licensed amateurs to join an association that would
be concerned exclusively with amateur radio activities and promised to eschew the intrigues that were plaguing some of the other groups that were attempting to embrace also broadcast listeners. Some 50 amateurs responded and Jack Lefebvre became founder-president.

Although amateur activity has always been on a fairly modest scale in France (currently there are about 5000 French amateurs) at least two of Europe's most successful pioneers of h.f. were located there: Leon Deloy, F8AB of Nice and Pierre Louis, F8BF.

The South African Radio League similarly reaches it 50th anniversary in May.

## From all quarters

A suggested "facsimile standard" for British amateurs is put forward in $C Q-T V$ by J. J. Wilcox, G8GGU: drum speed 3 Hz ; drum size 70 mm diameter by 70 mm long for $1: 1$ aspect ratio; scan rate 64 or 96 lines/inch; co-operation index 264 or 176; sync/phasing 15 second period, $4 \%$ white pulse in black level at start of line; scan direction left to right; modulation a.f.s.k. to A4, F4 or A4J; tones carrier 1700 Hz , white 1300 Hz , black 2100 Hz , stop 1100 Hz with pictuie inversion available.

The Radio Amateurs Old Timers Association (open to amateurs who have held a licence for 25 years) is holding its 1975 annual reunion on Friday, May 16 at the Bonnington Hotel, London WC1 (details Miss M. Gadsden, 79 New River Crescent, London N13 5RQ). Its official "net" is at - 1100 hours on the first Thursday of each month on 3740 kHz .
Following representations from the RSGB, the Home Office has agreed to a simplication of log-keeping for mobile operation. Logs will now have to show only time of the start and finish of the journey; starting and finishing points of the journey; and frequency bands used during the journey.

The Sunday-morning GB2RS news stations on v.h.f. are now all using the same frequency of 144.5 MHz .

## In Brief

Letters reaching me from the RSGB are usually franked with the slogan "Radio Society of Great Britain guards the interest of the radio amateur"-but recently the Post Office substituted the rather perverse message: "Collect stamps a great hobby". . . To counter overcharging of the Oscar 6 battery amateurs can now make use of the morning "descending" orbits on Mondays, Wednesdays and Saturdays ... Allan Mears, G8SM, has been elected as President of the Thames Valley Amateur Radio Transmitters Society, now in its 42nd year . . . The Amateur Radio Mobile Society's 1975 rally will be on Sunday, May 18 at The Clinical Research Centre, Northwick Park Hospital, Watford Road, Harrow, Middx (near Northwick Park underground station).

PAT HAWKER, G3VA

## New Products

## Column indicator

This indicator consists of two columns of light, the lengths of which represent an analogue quantity. Two separate analogue values can be displayed on the columns which are formed by 100 elements, each being illuminated in turn to form a continuous column of light up to 126 mm long and 2.54 mm wide. The indicator is manufactured by Burroughs and available from Walmore Electronics Ltd, 11-15 Betterton St, London WC2H 9BS.
WW300 for further details

## Temperature detector

The "thermafilm" temperature detector is a thick-film unit which matches the BS1904 and DIN43760 specifications and can therefore replace conventional wirewound platinum resistance detectors. Response time of the device is claimed to be half that of platinum detectors. Thermafilm can be used over a temperature range of -50 to $+600^{\circ} \mathrm{C}$. Matthey Printed Products Ltd, William Clowes Street, Burslem, Stoke-on-Trent, Staffs.
WW303 for further details


WW300

## Microwave filters

Models TYG-100 and TYG-400 are con-tinuously-tunable bandpass filters having bandwidths from 1 to 20 GHz and 4 to 18 GHz respectively. These filters are YIG types offering an error of less than $1 \%$ and a resolution on the frequency dial of 10 MHz . Maximum average r.f. power from the instrument, which measures $4 \frac{1}{2} \times$ $4 \frac{1}{2} \times 3 \mathrm{in}$, is 100 mW . Telonic Industries UK, 2 Castle Hill Terrace, Maidenhead, Berks.
WW313 for further details

## Heat-sinks

A range of black-anodized heat-sinks for TO-5 and TO- 100 packages have thermal resistances from $30^{\circ} \mathrm{C} / \mathrm{W}$ and are manufactured from copper-based alloys. Dau UK Ltd, 42A Main Road, Barnham, Sussex PO22 0ES.
WW327 for further details

## Vacuum relays

Latest additions to the Kilovac Corporation range of vacuum relays are the $\mathrm{KC}-3$ rated at 8 kV , the $\mathrm{KC}-10$ and $\mathrm{H}-26$ both rated at 15 kV , and the $\mathrm{KC}-20$ rated at 28 kV . The relays offer a dielectric strength of around $1000 \mathrm{~V} / \mathrm{mil}$ when operating, which permits closer contact spacing and low-bounce mechanisms. Walmore Electronics Ltd, 11 Betterton Street, Drury Lane, London WC2H 9BS.
WW320 for further details

## Frequency synthesizer

The Rockland model 5100 programmable frequency synthesizer uses digital techniques to provide outputs in 0.001 Hz steps from d.c. to 2 MHz . Programming is accomplished through t.t.l.-compatible circuits or contact-closures to ground.


WW303


WW313

Either a binary or 8.4 .2 . 1 b.c.d. format can be used, with up to 46 parallel bits or four 12 -bit bytes. Output amplitude of the instrument is variable continuously and in 1 dB steps to 85 dB from a maximum of 10 V pk-pk with $50 \Omega$ source impedance. Wessex Electronics Ltd, Stover Trading Estate, Yate, Bristol BS 17 5QP.
WW315 for further details

## Contactless keyboard

Plessey Keyboards have announced a contactless electronic keyboard-the PCK 2000. The unit, which has been produced primarily for the professional computer market, features capacitance coupled keyswitches. These switches operate into encoding logic based on an l.s.i. r.o.m. which provides various design options. The options can be selected on the basis of specification or cost requirements. Plessey Keyboards, Wood Burcote Way, Towcester, Northants NN12 7JN.
WW316 for further details

## Opto-isolated switches

Two new solid-state switches consist of a low-level voltage switching control, suitable for direct drive from logic pulses, and - optical isolation between input and output circuits. The input voltage range is from 3 to 32 V d.c., which will switch an alternating current of 10 A r.m.s. at a voltage of either 120 or 240 V . Hamlin Electronics Ltd, 14 New Road, Southampton.
WW317 for further details

## Thick-film amplifiers

A 12 W class A power amplifier, type TF008, requires an input of 0.5 V for full rated output and a claimed distortion figure of $0.05 \%$. The supply voltage range is from $\pm 12$ to $\pm 20 \mathrm{~V}$, and the frequency response is 10 Hz to 30 kHz . Type TF 009 is a 25 W class B design requiring a supply voltage of between $\pm 17$ and $\pm 25 \mathrm{~V}$ at 2 A maximum. Frequency response is


WW327


WW320

20 Hz to 60 kHz with a typical harmonic distortion figure of $0.2 \%$. Both units require external power transistors, and measure $1.35 \times 1 \times 0.25$ in. Guest Distribution Ltd, Redlands, Coulsdon, Surrey CR3 2HT.
WW318 for further details

## Phasor meter

The model STD 10,000 phase-sensitive multimeter will give direct readings; of in-phase and quadrature components of voltage or current, on two separate meters. Five voltage/current ranges from 500 mV to 500 V and 1 mA to 10 A f.s.d. are provided on the instrument, which operates at $50 / 60 \mathrm{~Hz}$ or 12 to 2400 Hz with the aid of an adaptor. J. J. Lloyd Instruments Ltd, 1 Brook Lane, Warsash, Southampton, Hants.
WW305 for further details

## Digital multimeter

A multimeter offering a voltage range from $1 \mu \mathrm{~V}$ to 1000 V , a resistance range from $\operatorname{lm} \Omega$ to $2000 \mathrm{M} \Omega$, and a current range from 10 pA to 2 A has been introduced by Keithley Instruments. Other features of the model 160 B are a 1200 V floating capability, a $0.2 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ stability, and several options/accessories including a b.c.d. output, a r.f. probe, and a 50A shunt. Keithley Instruments Ltd, 1 Boulton Road, Reading, Berks.
WW325 for further details

## Load simulator

The model EL750 is a portable d.c. power tester suitable for checking power supplies. The unit will dissipate up to 750 W d.c. and will operate in a constant-resistance or constant-current mode, selected manually in steps by push buttons. Load-current programming can be accomplished by applying an external direct voltage through a connector on the rear panel. Data Technology Corporation, Sherwood House, High Street, Crowthorne, Berks.
WW308 for further details

## Silk-screen service

Circuitape Ltd have introduced a made-toorder service for silk-screen printed aluminium panels. The panels can be produced in any shape and size with punched holes to specific requirements. Silk-screening can be in any colour with legends in any language. Delivery is normally around five weeks, but a special rapid service is also available. Circuitape Ltd, New Street, Aylesbury, Bucks.
WW306 for further details

## Elapse timers

A custom range of elapse timers from Longmore Systems enable time periods between 1 ms and 99990s to be measured. Five-decade selection is provided but different ranges may be specified. Control is by voltage-trigger and push-button startstop with separate reset. Instrument read-


WW305


WW325


WW308
out is on a four-digit display which is accurate to within 10 p.p.m. Longmore Systems Ltd, Environment House, 875 Sidcup Road, London SE9 3PP.
WW307 for further details

## YIG counter

The model 331 microwave counter will automatically measure frequencies from 825 MHz to 18 GHz . The centre frequency of signals with up to 200 MHz f.m. devi ation can be measured directly and an optional plug-in circuit permits the measurement of signals as low as -25 dBm . Remote programming, b.c.d. output and rear input options are also available for systems application, where up to 80 readings a second can be made. Dana Electronics Ltd, Collingdon Street, Luton, Beds.
WW329 for further details

## Function generator

The Hewlett-Packard model 3312A function generator contains two independent generators in one case. The main generator has a frequency range from 0.1 Hz to 13 MHz in eight ranges while the modulator generator delivers signals from 0.01 Hz to 10 kHz . Both generators provide sine, triangle, square, pulse and positive/negative ramps. By combining the generators, sweep, a.m., f.m. and tone
bursts can be created with an output, from the main generator, of 10 V pk-pk into 50』. A four-position attenuator with variable control adjusts the output over a 60dB range. Hewlett Packard Ltd, King Street Lane, Winnersh, Wokingham, Berks RG115AR.
WW326 for further details

## 36 position switch

A single-pole, 35 -way switch rated at 2 A continuous with a breaking figure of 50 mA at 300 V a.c./d.c. has been added to the N.S.F. range of rotary wafer switches. Both $10^{\circ}$ and $20^{\circ}$ indexing versions are available from N.S.F. Ltd, Keighley, Yorkshire DD2 15 EF .
WW302 for further details

## Plastic pots?

Two ranges of conductive plastic potentiometers designated P4100/4200 and P4400, the latter being a low cast version of the former, are now available in the UK. The precision range is rated at 1.8 W and offers a resolution of $0.003 \%$ with $352^{\circ}$ angle of rotation, a linearity within $0.2 \%$ and an operating torque of $0.2 \mathrm{~cm} . \mathrm{g}$. Both models are manufactured in servo size 13 and can be supplied with up to ten ganged tracks. Variohm Components, The Barn, Wood Burcote, Towcester, Northants NN 12 7JR. WW311 for further details


WW329


WW326

## Active filter

The UAF31 is a two-pole active filter in which, with the addition of three or four external resistors, the Q -factor, resonant frequency and gain can be controlled. Three separate outputs provide low, high and band-pass transfer function-by summing the high and low pass outputs a band-reject transfer function can be obtained. Frequency accuracy is within $1 \%$ and the Q range is from 0.5 to 500. Burr-Brown International Ltd, 25A King Street, Watford, Herts WD1 8BT.
WW310 for further details

## Pyrometer

The Litesold pyrometer has been designed for measuring soldering-iron bit temperatures. A fine thermocouple tip, which causes negligible cooling, is placed on the bit and temperature is read off a meter calibrated to $500^{\circ} \mathrm{C}$ f.s.d. Light Soldering Developments Ltd, 97 Gloucester Road, Croydon, Surrey.
WW332 for further details

## Conductive plastics

3M have announced a conductive plastic called Velostat. This product is available as a material or as a variety of manufactured items. For the benefit of any organic-chemists that may be reading the material is a carbon-loaded polyolefin


WW302


WW332
plastic which is conductive throughout its volume. 3M UK Ltd, 3M House, Wigmore Street, London W 1A 1ET.
WW330 for further details

## Logic panel meter

This panel meter has a six-digit display and can be used for frequency counting, time, and period measurements. The unit, which occupies $3.3 \times 1.4$ in of panel space, consists of three modules-a six digit decimal counting and display section, a pre-scaler and timing generator module, a clock and offset module which consists of an internal 500 kHz crystal oscillator and a programmable divider. Power requirements are $\pm 5 \mathrm{~V}$ at 800 mA . Tony Chapman Electronics Ltd, 80A High Street, Epping, Essex CM16 4AE.
WW331 for further details

## Circuit tester

A pocket-sized tester that will check voltage, polarity and continuity is now available in the UK. The instrument has a l.e.d. indicator which glows when either a voltage between 3 and 600 V a.c. $/$ d.c. is present, continuity in a circuit exists, or the polarity of a circuit/component is correct with respect to the coloured probes. British Central Electrical Co Ltd, Ringwood, Hants.
WW321 for further details

## Variable transformers

Cherishaw Ltd have introduced a new range of single- and three-phase variable transformers with current ratings from 2 to 28A. Each model is manufactured in either an open form for panel mounting or enclosed for bench use and all versions are designed for a 240 V supply. Cherishaw Ltd, 103 Mount Pleasant, Tunbridge Wells, Kent.
WW322 for further details

## Inductors

The 1537 series of moulded r.f. inductors will operate in the temperature range $-55^{\circ}$ to $+125^{\circ} \mathrm{C}$ and are available in inductances from 0.15 to $240 \mu \mathrm{H}$. Maximum current ratings range from 115 mA , for the $240 \mu \mathrm{H}$ device, to 2.74 A for a $0.15 \mu \mathrm{H}$ type. The components measure 0.155 in dia $\times 0.375$ in and are manufactured by Amphenol Ltd, Thanet Way, Whitstable, Kent CT5 3TF.
WW323 for further details

## P.r.o.m. eraser

An ultra-violet power source designed for erasing p.r.o.ms has a built-in timer, variable from $0-30$ minutes, and can erase up to six memories in a single run. The unit is manufactured by Stolz A.G. of Switzerland and is available in the UK from Memec Ltd, The Firs, Whitchurch, Aylesbury, Bucks.
WW324 for further details

## Sinewave oscillators

A series of low-distortion, amplitude-stable signal sources manufactured by Frequency Devices Inc., provide a single, specified
frequency in the range 100 Hz to 10 kHz . Features include a stability of $0.02 \%$ per ${ }^{\circ} \mathrm{C}$, amplitude stability of 0.1 dB , adjustable output from 1 to 20 V p-p. Distortion of the device is $0.1 \%$ and the impedance is less than $10 \Omega$. The oscillators are shortcircuit protected, measure $1.5 \times 2.0 \times$ 0.4 in and are available from Lyons Instruments Ltd, Hoddesdon, Herts.
WW304 for further details

## Solid Stafe Devices

Names of suppliers of devices in this section are given in abbreviation after each entry and in full at the end of the section.

## Photodiodes

The TIXL471 gallium-arsenide l.e.d. and the TIXL451 silicon avalanche photodiode are both high-speed diodes for use in fibre optic application. The devices will connect directly and self-align with Corning T-19H optical waveguide terminations. WW350 for further details

Texas

## Regulators

Fixed-voltage regulators for both positive and negative supplies are available with outputs from 5 to 24 V and current ratings up to 1.5 A . The regulators are supplied in either a plastic package or TO-3 encapsulation.
WW351 for further details
GDS

## Switching transistor

A triple diffused n-p-n- power transistor, type SCA100-120, appears as only a $0.002 \Omega$ resistance with a 100 A collector current. Saturation voltage at the maxi-mum-rated $I_{c}$ is 1.7 V and the maximum voltage is 120 V .
WW352 for further details Impectron

## Microprocessor

The Mostek eight-bit parallel microprocessor, type MK5065 is a 40 -pin single chip-device. It offers 51 basic instructions or 81 with modifications, and has t.t.l. compatible inputs and outputs.
WW353 for further details
Lock

## Miniature bridges

A. new range of 1.5 A silicon full-wave rectifiers comprises seven devices-the MDA 100 to 110 designed for voltages between 50 and 1000 V . These bridges will operate over a junction-temperature range from -55 to $150^{\circ} \mathrm{C}$ and will withstand a 45 A surge for one cycle of operation.
WW354 for further details Motorola

## Alarm i.cs

A range of i.cs designed for alarm application are now available in the U.K. Devices in the range include the 3010 tone alarm which compares an input signal to an adjustable reference voltage if the reference voltage is exceeded a pulsating
or constant tone for driving an external loudspeaker is generated. The 3020 tristage alert/alarm has three l.e.d. drivers. Each of the three drivers has two t.t.l. compatible inputs. The 3030 temperature alarm activates both a steady t.t.l. compatible output and a tone output if the temperature of the i.c. package exceeds a preset level.
WW355 for further details
Adrian Electronics

## A/d system

A low-cost a/d system can be realized by using the MC14435 d.v.m. i.c. and the MC1505L dual-ramp generator and comparator i.c. One external capacitor and two potentiometers are required to complete the circuit.
WW356 for further details Semicomps

## TV-sound i.c.

The TDA1 190 is capable of carrying out all the functions of a television sound channel. These functions include an i.f. amplifier/limiter, an active low-pass filter, f.m. detector, a d.c. volume control and a power amplifier.
WW357 for further details
SGS-Ates

## Zener diodes

A new range of zener diodes are plastic package types with a power dissipation capability of 1.32 W and a zener voltage range from 3.3 to 200 V .
WW358 for further details
Siemens

## C.m.o.s. a/d converter

Analog Devices Ltd have announced what is claimed to be the world's first micro-processor-compatible i.c. analogue-todigital converter to provide up to 10 -bit accuracy. The device, designated AD7570, uses c.m.o.s. construction and is designed specifically to interface with microprocessors, and is fully t.t.l./d.t.l./c.m.o.s. compatible. The AD 7570 features a conversion time of $20 \mu \mathrm{~s}$ and a throughput rate of 50 kHz .
WW359 for further details
Analog Devices
Texas Instruments Ltd, Manton Lane, Bedford.
GDS Sales Ltd, Michaelmas House, Salt Hill, Bath Road, Slough, Bucks.
Impectron Ltd, 23 King Street, London W3 9LH.
Lock Distribution, Neville Street, Middleton Road, Oldham, Lancs OL9 6LF.
Motorola Ltd, Semiconductor Products Division, York House, Empire Way, Wembley, Middx.
Adrian Electronics Ltd, 28 High Street, Winslow, Buckingham MK 18 3HF.
Semicomps Ltd, Northfield Industrial Estate, Beresford Avenue. Wembley, Middx HA0 1 SD.
SGS-Ates Componenti Elettronici SpA , Via C. Olivetti 2, 20041 Agrate Br., Milan. Italy.
Siemens Ltd, Great West House, Great West Road, Brentford, Middx.
Analog Devices Ltd, Central Avenue, East Molesey, Surrey.


There dwelt in the land of Brit certain high priests who served in the temples of Elektron, which is an invisible god who darteth around in ever-decreasing circles but never into his own nucleus. And the priests of Elektron were devout men, serving no other god but he. And Elektron looked with favour upon them and rewarded them each according to his worth with divers strange gifts. To some he gave power to converse with those from afar off and to others he brought visions of strange happenings in distant lands; yea, even of the United States cavalry in glorious Technicolor.

And to certain other of his high priests Elektron gave powers of levitation, so that they walked with their feet ever-soslightly off the ground; these dwelt in glass temples called, in the native tongue, Researchlabs or Funnifarms, which were set apart from the common people and to which entrance was denied to all, saving only those having scrips of authority from the chief priest. And these priests were called by the common people Egbonces which meaneth he who knoweth the square root of minus one. And the Egbonces were cunning at fashioning curious devices from boot-latchets and wax so that the populace were astonied and continually cried out, saying, Behold, these are great wonders but of what use be they?

Yet other high priests of Elektron were followers of the prophet Babbage and these were set in authority over divers machines that brought much benefit to the common people; some computed the numbers of the tribes and the taxes that each man should pay; others controlled the paycheks of those that laboured, so that each man received less than his hire, while others suggested that the inter-city chariots were tardy in arrival. And Elektron taught the high priests to feed the engines with curious symbols engraven upon tablets that they might print out likenesses of the sex-goddess Bardot devoid of her apparel, which gave satisfaction to many. And these priests likewise withdrew the hems of their garments from the common populace and, by conversing in the alien tongues of Fortran and Algol, preserved their mysteries jealously.

At this time the skies were filled with
heavier-than-air machines of many nations which flew with the noise of emasculated hornets and carried the peoples to and fro, even unto the ends of the earth. These machines were under the auspices of the god Hijak. And certain of the nations had air machines which could drop unpleasantness on the land beneath to discomfort the people; but certain other nations who were poor and backward and, as the saying goeth, not with it, did not possess these amenities. Thus it came to pass that the acquisition of such machines was regarded by all as an outward and visible sign that the possessor nation was emerging from savage practices and an ensample to others.

And certain rich merchants searched diligently and redeemed many heavier-than-air machines; some from the knacker's yard; some which fell from the back of an hangar and yet others which were dislodged privily from the Science Museum. And they purposed to sell these to the heathen for many shekels of gold and at great profit. So it came to pass that the merchants sent envoys to a far country, even to the kingdom of Tsetse-Tsetse.

And the envoys said unto the king of Tsetse-Tsetse, $\mathbf{O}$ king live for ever but put not thy money upon it. And the king answered saying, What meanest thou? Then did the envoys reply saying, Surely thou knowest that thy neighbour the king of Beri-Beri hath cast covetous eyes upon thy lands and thy maidens? If only thou hadst an Air Force it would cause thine adversary to wind his neck in. Then did the king beat his breast crying, Woe is me! And the envoys made reply saying Not so, O king, for it so happeneth that we can supply thee with a squadron of Bleriot Mk.Is. And thus it came to pass that the king bought from the envoys for much fine gold and slept peacefully with his wives that night.

Then did the envoys depart and journeyed to the neighbouring land that is called Beri-Beri. And they said to the king of Beri-Beri, o king live for ever but begin not the reading of any long novels. And the king said What meanest thou? Whereupon the envoys replied saying, Knowest thou not that thy neighbour the king of Tsetse-Tsetse hath secretly purchased war-birds and purposeth to ravage thy country? At this the king went as pale as was possible and the end of the matter was that he became Commodore of a squadron of Cabbage White Mk. VIIs.

And it came to pass that in Brit the god Elektron gave unto his high priests the power to fashion magick bowls which could divine the presence and movements of heavier-than-air machines even at great distances. Yea, and not only this, for, by gazing into the bowl, yessels having their business in great waters could be made to broach each other with greater certainty. And on land its magick powers enabled the Fuzz to put the finger upon all charioteers who, like their forebear Jehu, drove furiously. And the name of this new wonder was radar, which, being translated, meaneth That which worketh by suction and mirrors.

And the rich merchants came unto the high priests of radar and said unto them, Lo, we have heard much of the wonders that thy god Elektron hath taught thee and it seemeth that we can do a deal with profit to all. Make for us great numbers of these magick bowls, we pray thee, that we may sell them to the nations for their greater safety. Do this and we will pay thee many shekels of gold; moreover, we will pull down thy temples which are but potting sheds and in their stead we will raise mighty glass temples to the greater glory of Elektron, wherein thou shalt find all the instruments that thy heart desirest. And we will clothe thee in white raiment and give thee charge over many. What sayest thou?

And the high priests conferred privily and agreed among themselves that they were on to a good thing. So it came to pass that the merchants caused mighty temples to be built wherein the god Elektron might be served, both by day and night; and the high priests, for their part, devised magick bowls with ever greater cunning and these the merchants sold to whoever was in the market place. Thus it came about that both the king of TsetseTsetse and the king of Beri-Beri were persuaded to buy the magick bowls with which to keep vigil each upon the other. Yea, both primary and secondary radar had they in plenty and certain inhabitants of the two countries were trained to interpret the signs and portents which appeared upon these bowls whenever an heavier-than-air machine was drawing nigh.

And behold, it came to pass that upon a certain night there was a watchman in the kingdom of Tsetse-Tsetse who was an exceeding dim lamp; moreover, when interpreting the symbols on the magick bowl, he was, as the saying is, unable to tell Squawk from Clutter. And this watchman, fearful of what he supposed he saw upon the face of the bowl, said unto himself The enemy is upon us, and thereupon smote the Panick Button. Whereupon the Bleriot Mk.Is rose (all excepting one which had broken its elastick band) and brought destruction to the sleeping land of Beri-Beri. But the Cabbage Whites, being forewarned by their magick bowls, were already riding the heavens and bringing affliction upon their neighbours. And, by morning, both countries were bathed in blood.

And in the temples of Elektron there was great commotion, for the hot lines were glowing red and the artificial moons which the high priests had raised were overburdened with coloured images of the slaughter, for the delectation of the common people. And when all was accomplished, overseers from the United Nations came and wagged their heads and voted Tsetse-Tsetse and Beri-Beri into their assemblies in recognition of their emergence.

# Sanyo Video Tape Recorder Systems 

In a changing world audio visual innovations and methods are developing with incredible speed. Keeping pace with this development is the range of uses to which this equipment can be applied Practical applications are virtually limitless and indeed appear to be bounded only by the employers imagination. Sanyo, acknowledged leaders in slow motion and 'stop frame' techniques, whose VTR products have been used world wide for many years in industry, commerce, education and sport have, with the aid of extensive research, produced a range of high quality competitively priced audio visual equipment. Cameras, recorders, monitors

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VCA 200 E Video camera kit with built-in viewfinder, microphone, zoom lens and tripod. For use with the VTR 1100 SLR and VTR 2000 3 " electronic viewfinder can also be used as a playback monitor.

VM 4120 (K) ponitor with R/F. Off-air portable video monitor/receiver with high resolution $12^{\prime \prime}$ CRT. Ideal for educational purposes. Can be used as conventional TV.

VTR 2000
A compact lightweight $\frac{1}{2}$ " Video Tape Recorder with a host of advanced reatures including an automatic level control system that eliminates video and audio signal adjustment, independent audio erasure for
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and tracking control system. automatic shut off switch, tape counter and a perfect ferrite crystal head assembly. Weight less than 29lbs.


VC 1150
ra with swirchable internalexternal sync and ALC. For use with the VTR 1100 SLR and VTR 2000


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Low light level video camera Operates at very low light levels for security purposes. Also for use with infra-red lighting.

## IP) IL.P.t.testronerevere



Mono electrical circuit diagram with interconnections for steren shown


The HY5 is a complete mono hybrid preamplifier, ideally suited for both mono and stereo applications. Internally the device consists of two high quality amplifiers- the first contains frequency equalisation and gain correction, white the second caters for tone control and
TECHN
NICAL SPECIFICATION
Inputs
Magnetic Pick-up $3 m$ V.RIAA
Ceramic Pick-up
Microphone
Tuner
Auxillary
Outputs
Tape
Tape $\quad 100 \mathrm{mV}$
Main output Oab ( 0.775 volts RMS)
Active Tone Controls
Treble $\pm 12 \mathrm{db}$ at 10 kHz
12 db at 100 Hz
Distortion $\quad 0.05 \%$ at 1 kHz
Signal/Noise Ratio $\quad 68 \mathrm{db}$
Overload Capability 40 db on most
$\begin{array}{ll}\text { Supply Voltage } & \begin{array}{l}\text { sensitive input } \\ \pm 16-25 \\ \end{array} \quad \begin{array}{l}\text { voits }\end{array}\end{array}$
PRICE E4. $50+0.36$ V.A.T. $P$ \& $P$ free
3 mV .RIAA
10 mv
100 mV
3.100 mV
$47 \mathrm{k} \Omega$ at 1 kHz
sitive input

- 25 volts


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Hi-Fi amplifier incorporating its own high conductivity heatsink hermetically sealed in black epoxy resin. Only five connec. tions are provided: input, output, power tines and earth.

TECHNICAL SPECIFICATION
Output Power 25 watts RMS into $8 \Omega$ Load Impedance $4-16 \Omega$
Input Sensitivity Odb ( 0.775 volts RMS) Input Impedance $47 \mathrm{k} \Omega$
Distortion Less than $0.1 \%$ at 25 watts typically $0.05 \%$
Signal/Noise Ratio Better than 75 db
Frequency Response $10 \mathrm{~Hz}-50 \mathrm{kHz}+3 \mathrm{db}$ Supply Voltage $\pm 25$ volts
Size $105 \times 50 \times 25 \mathrm{~mm}$.
PRICE $£ 5.98+0.48$ V.A.T. P \& $P$ free


The PSU50 incorporated a specially designed transformer and can be used for either mono or stereo systems.

TECHNICAL SPECIFICATIONS Output voltage 50 volts ( $25-0-25$ )
input voltage $210-240$ volts
Size L.70. D.90. H. 60 mm .
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## AMPLIFIER KITS OF $\mathscr{D}$ istinction

## DESIGNER－APPROVED KIT

In Hi－Fi News there was published by Mr Linsley－Hood a series of four articles（November 1972－February 1973 and a subsequent follow－up article（April 1974）on a design for an amplifier of exceptional performance which has as its principal feature an ability to supply from a direct coupled fully protected output stage，power in excess of 75 watts whilst maintaining distortion at less than $0.01 \%$ even at very low power levels．The power amplifier is complemented by a pre－amplifier based on a discrete component operational amplifier referred to as the Liniac which is employed in the two most critical points of the system，namely the equalization stage and tone control stage，positions where most conventional designs run out of gain at the extremes of the frequency spectrum Unusual features of the design are the variable transition frequencies of the tone controls and the variable slope of the scratch filter．There is a choice of four inputs，two equalized and two linear，each having independently adjustable signal level．The attractive slimline unit pictured has been made practical by highly compact PCBs and a specially designed Toroidal transformer

Hi－Fi News Linsley－Hood 75 W Amplifier
Mk III Version（modifications as per Hi－Fi News April 1974）


Full circuit description
in handbook
（pack 15－－price 30p）
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tor power amp．
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of 2 drilled，finned heat sinks
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－Set fre－amp
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Set of low noise，high gain
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| AC188K | 0.28 | BDY60 | 0.61 | OC44 | 0.08 | 2N3054 | 0.38 |
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| 150 | 100 |
| 154 | 200 |
| 152 | 250 |
| 153 | 350 |
| 154 | 500 |
| 155 | 750 |
| 156 | 1000 |
| 157 | 1590 |
| 158 | 2000 |
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| CD4024 | 1.24 | MC1310P | 2.59 | SN7430 | 0.16 | SN74121 | 0.37 | TAA621 | 2.03 |
| CD4025 | 0.32 | MC1330 | 0.90 | SN7432 | 0.28 | SN74122 | 0.50 | TAA6618 | 1.32 |
| CD4027 | 0.43 | MC1351P | 0.80 | SN7437 | 0.35 | SN74123 | 0.60 | TBA641B | 2.25 |
| CD4028 | 1.50 | MC1352P | 0.80 | SN7438 | 0.35 | SN74141 | 0.85 | TBA651 | 1.69 |
| CD4029 | 3.50 | MC1466L | 3.50 | SN7440 | 0.16 | SN74145 | 0.90 | tBa800 | 1.40 |
| CD4030 | 0.87 | MC1469R | 2.75 | SN7441AN | 0.85 | SN74150 | 1.50 | tbab 10 | 1.40 |
| CD4031 | 5.19 | NE555V | 0.70 | SN7442 | 0.65 | SN74151 | 0.85 | tBa820 | 1.15 |
| CD4037 | 1.93 | NE556 | 1.30 | SN7445 | 0.90 | SN74153 | 0.85 | tBA920 | 4.00 |
| CD404 1 | 1.86 | NE560 | 4.48 | SN7446 | 0.95 | SN74154 | 1.50 | DIL socket | 0.17 |
| 4042 | 1.38 | 61 |  | 44 | 0.95 |  |  |  |  |

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| 2N697 | 0.16 | 2N4037 | 0.42 | AF239 | 0.65 | BD140 |
| 2N69B | 0.82 | 2N4036 | 0.67 | AF240 | 0.90 | BF115 |
| 2N699 | 0.59 | 2N4058 | 0.18 | AF279 | 0.70 | BF117 |
| 2N706 | 0.14 | 2N4062 | 0.15 | AF2B0 | 0.79 | BF154 |
| 2N708 | 0.17 | 2N42B9 | 0.34 | AL102 | 1.00 | BF159 |
| 2N916 | 0.28 | 2N4920 | 1.10 | BC107 | 0.14 | BF 180 |
| 2N918 | 0.32 | 2N4921 | 0.83 | BC108 | 0.14 | 8F181 |
| 2N1302 | 0.185 | 2N4923 | 1.00 | BC109 | 0.14 | 8F184 |
| 2N1304 | 0.26 | 2N5245 | 0.47 | 8C1478 | 0.14 | BF194 |
| 2N1306 | 0.31 | 2N5294 | 0.48 | BC1488 | 0.15 | 8F195 |
| 2N1308 | 0.47 | 2N5296 | 0.48 | 8C1498 | 0.15 | BF196 |
| 2N1711 | 0.45 | 2N5457 | 0.49 | BC157A | 0.16 | 8F197 |
| 2N2102 | 0.60 | 2N5458 | 0.46 | BC15BA | 0.16 | BF198 |
| 2N2147 | 0.78 | 2N5459 | 0.49 | 8C1678 | 0.15 | 8F244 |
| 2N214B | 0.94 | 2N6027 | 0.45 | BC168B | 0.15 | BF257 |
| 2N2218A | 0.22 | 3N128 | 0.73 | 8C1698 | 0.15 | BF25B |
| 2N2219A | 0.26 | 3N140 | 1.00 | 8C182 | 0.12 | BF259 |
| 2N2220 | 0.25 | 3N414 | 0.81 | BC182L | 0.12 | BFS61 |
| 2N2221 | 0.18 | 3N200 | 2.49 | BC183 | 0.12 | BFS98 |
| 2N2222 | 0.20 | 40361 | 0.40 | 8C183L | 0.12 | BFR39 |
| 2N2369 | 0.20 | 40362 | 0.45 | 8C184 | 0.13 | BFR79 |
| 2N2646 | 0.55 | 40406 | 0.44 | BC184L | 0.13 | BFX29 |
| 2N2904 | 0.22 | 40407 | 0.35 | BC212A | 0.16 | 8 FX 30 |
| 2N 2905 | 0.25 | 40408 | 0.50 | 8C212LA | 0.16 | BFX84 |
| 2N2906 | 0.19 | 40409 | 0.52 | BC213LA | 0.15 | BFX85 |
| 2N2907 | 0.22 | 40410 | 0.52 | BC214LB | 0.18 | BFX88 |
| 2N2924 | 0.20 | 40411 | 2.00 | BC237B | 0.16 | BFY50 |
| 2N2926G | 0.12 | 40594 | 0.74 | BC238C | 0.15 | BFY51 |
| 2N3053 | 0.25 | 40595 | 0.84 | 8C239C | 0.15 | BFY52 |
| 2N3054 | 0.60 | 40636 | 1.10 | BC257A | 0.16 | BRY39 |
| 2N3055 | 0.75 | 40673 | 0.73 | BC258B | 0.16 | ME0402 |
| 2N3391 | 0.28 | AC126 | 0.20 | BC259B | 0.17 | ME0412 |
| 2N3392 | 0.15 | AC127 | 0.20 | BC301 | 0.34 | ME4102 |
| 2N3393 | 0.15 | AC128 | 0.20 | ВС307B | 0.17 | MJ480 |
| 2N3440 | 0.59 | AC151 | 0.27 | BC308A | 0.15 | MJ481 |
| 2N3442 | 1.40 | AC152 | 0.49 | BC309C | 0.20 | MJ490 |
| 2N3638 | 0.15 | AC153 | 0.35 | BC327 | 0.23 | MJ491 |
| 2N3702 | 0.12 | AC176 | 0.30 | BC328 | 0.22 | MJ2955 |
| 2N3703 | 0.13 | AC187K | 0.35 | BCY70 | 0.17 | MJE340 |
| 2N3704 | 0.15 | AC188K | 0.40 | BCY71 | 0.22 | MJE370 |
| 2N3706 | 0.15 | AD143 | 0.68 | BCY72 | 0.15 | MJE371 |
| 2N3708 | 0.14 | AD161 | 0.50 | BD121 | 1.00 | MJE520 |
| 2N3714 | 1.38 | AD162 | 0.50 | BD123 | 0.82 | MJE512 |
| 2N3716 | 1.80 | AF106 | 0.40 | 8D124 | 0.67 | MJE2955 |
| 2N3771 | 2.20 | AF109 | 0.40 | BD13 | 0.40 | MJE3055 |
| 2N3773 | 2.65 | AF115 | 0.35 | BD 132 | 0.50 | MP8113 |
| 2N3789 | 2.06 | AF 116 | 0.35 | 8D135 | 0.43 | MPF 102 |
| 2N3819 | 0.37 | AF117 | 0.35 | BD136 | 0.47 | MPSA05 |
| 2N3820 | 0.64 | AF118 | 0.35 | BD137 | 0.55 | MPSA06 |
| 2N3904 | 0.27 | AF124 | 0.30 | BD138 | 0.63 | MPSA55 |

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| :---: | :---: |
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| AM/FM MODULES |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |
| LP1157 AM/Module |  |  |  | ¢2.50 |
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|  |  |  |  | $f 4.50$ |
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| Mullard Modutes |  |  |  |  |
| LP $1157 \mathrm{AM} / \mathrm{T}_{\text {ype }}$ |  |  |  | 0.50 |
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| $470 \mathrm{C} 6 / 7 \frac{1}{2} / 9 \mathrm{~V} 300 \mathrm{~mA}$ |  |  | *P108145 volt 0.9A | £7.80 |
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4317 20K V with case tester steel case 0432320 KaV plus 1 KHz 465 KHz OSC with case T1-2 20KRV slim type THL33D (L33DX) 2 KDV
TP5SN 20K r V
(Case $\mathbf{f 2 . 0 0 |}$
TPS $10 \mathrm{~S} 2 \mathrm{~K} \Omega \mathrm{~V}$
W20S $20 \mathrm{~K} \cap \mathrm{~V}$
W50k 50K $\Omega$ V
ester
New Aevolutionerv
680 M Mulitestar
$\dagger$ IE40 AC mutivolitite
TE1 15 Grid dip moreq
$440 \mathrm{KHz}-28 \mathrm{MHz}$
$\dagger+$ TEES 28 range valy valtmeter
TE200 RF yenerator

$20 \mathrm{~Hz}-200 \mathrm{KHz}$

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TT1 145 Compact flansister tester

+ G3.36 R/C Osc
$20 \mathrm{~Hz}-200 \mathrm{KHz}$
C3042 SWF Meter
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C1-5 Scope 500.000 KHz t carir fle 1.001


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Marriot XRPS/ $17 \frac{1}{1}$-track high $£ 2.50$ Marriot XRPS $18 \frac{1}{4}$-track med $£ 3.50$ Marriot XRPS/63 $\frac{1}{2}$ track high $\mathbf{1 1 . 7 5}$ Marriot $\frac{1}{2}-\mathrm{fr}$. Erase $13 \times 12 \mathrm{E}$ 343
Marriot erase heads for XRPS 17/18/36 (XESII) R/RPI record/play $\frac{1}{2}$-rrack Bogen type UL290 erase Miniature stereo-cassette
reciplay

STC \& ITT
miniature relays
$\begin{array}{lll}150 \mathrm{~A} & 6 \mathrm{v} & 2 \text { p.c.o. } \\ 180 \mathrm{R} & 6 / 12 v & 2 \text { p.c.o. } \\ 185 \mathrm{R} & 12 \mathrm{v} & 2 \text { BRANB }\end{array}$

$\left.\begin{array}{l}1250 \mathrm{R} 12 / 55 \\ 1700 \mathrm{R} 1 \mathrm{~B} / 24 \mathrm{v} 2 \text { p.c.o. }\end{array}\right\}{ }_{60}$
$\left.\begin{array}{l}1700 \mathrm{R} 1 \mathrm{~B} / 24 \mathrm{v} 2 \text { p.c.o. } \\ 1800 \mathrm{R} 24 \mathrm{v} \quad 4 \text { p.c.o. }\end{array}\right\} \begin{aligned} & \text { p.p. } \\ & 15 p\end{aligned}$ 2500R 18
10.7 MHZ MINIATURE CERAMIC
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| $\mathbf{£ 2 4 . 5 0}$ |
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MODEL 43 RC BRIDGE
Null indicating bridge for resistors and capacitors. Resistance range 10 R to 10 M $\pm 2 \%$ at Centre Scale. Capacity range 10 pF to $10 \mathrm{pF} \pm 2 \%$ Centre Scale except 1 pF to 10 pF Range $\pm 5 \%$. Power Factor Measurement 0-70\%
23.50

## MODEL 44 INDUCTANCE BRIDGE

Measures $1 \mu \mathrm{H}$ to 100 H in four ranges $\pm 5 \%$ accuracy. $Q$ measurement from $0.1-1.000+10 \%$. $\mathbf{£ 3 4 . 0 0}$
MODEL 45 DIRECT READING FREQUENCY METER
10 Hz to 100 KHz in four ranges. Input from
10 mV to $5 \mathrm{~V} \quad \mathbf{£ 3 6 . 0 0}$

All models except Model 35 are internally powered from 9v battery (extra). Carriage and packing all models 37p.
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STABILISED POWER
£3.25 MODULE SPM80

SPm80 is especially designed to power 2 of the A' $\mathbf{L 6 0}$ Amplifiers, up to
15 watt (r.m.a.) per channet simultaneously. This module embodies the latest components and circuit techniques incorpos ating complete short cirevit protection. With the addition of the Mains "Transformer BaT80,
the unit will provide outputs of up to 1.5 amps ut 35 volts. Size: $63 \mathrm{~mm} \times 105 \mathrm{~mm} \times 20 \mathrm{~mm}$. These units ensble you to baild Audio Systems of the higheat quality at a hitherto uno btainable price. Also
ideal for many other applicationa ineludiag: Disco Bystems, Public

TRANSFORMER BMT80 $£ 2.75$ p. \& p. 40p
STEREO PRE-AMPLIFIER TYPE PA100

## INTEGRATED CIRCUIT PAKS

 Manufacturers "Fall Oute" which include Functional and Part-Functional Units. These are classed as 'out-ot spec' from the maker's very rigid specifications, Lut are fdeal for learning about I.C's and experimental workPak No. Contents
Price
Pas No. Contents UIC $00=12 \pm 7400$ $\mathrm{UIC01}=12 \times 7401$
$\mathrm{UIC} 02=12 \times 7402$ UCCO2 $=12 \times 7402$
UTC03 $=12 \times 7403$ UC03 $=12 \times 7403$
UIC04 $=12 \times 7404$
UIC $05=12 \times 740$ UICO5 $=12 \times 7400$
UC06 $=8 \times 740 \mathrm{i}$
UIC UTC10 $=8 \times 7407$
UTC20 $=12 \times 7410$ UIC20 $=12 \times 7420$
UIC $30=12 \times 740$
UIC $40=12 \times 7440$ $\mathrm{UIC40}=12 \times 7440$
UIC41 $=6 \times 7441$ UIC4 $2=5 \times 7442$
UIC $4=5 \times 7443$
UIC $44=5 \times 744$ LINEAR I.C.'S—FULL SPEC


Built to a specification and NOT a price, and yet still the greatert value on the market,
the PA100 stereo preamplifler has been concelvei from the latest circuit techniques Designed for use with the ALfio power amplifier system, this quality made unit
incorporates no less than elght milicon pianar transigtors, two of these are specially selected low noise NPN devices for use in the input stager.
 Paloo, which aiso treble controls.
variable basg and tren

| SPECIFICATION: |  |  |  |
| :---: | :---: | :---: | :---: |
| Frequency reaponse | $20 \mathrm{~Hz}-20 \mathrm{kHz} \pm 1 \mathrm{~dB}$ | Bass control | ${ }^{15 \mathrm{~dB}}$ at 20 Hz |
| Harmonic distortion | better than $0.1 \%$ | Treble control | $\pm 15 \mathrm{~dB}$ at 20 kHz |
| Inputa: 1. Tape head | 3.25 mV into $50 \mathrm{~K} \Omega$ | Filters: Rumble (high pass) | 100 Hz |
| 2. Radio, Tuner | 75 mV into $50 \mathrm{~K} \Omega$ | ${ }_{\text {Signal/nolse ratio }}^{\text {Scrate (low }}$ | ${ }_{\text {better }} 8 \mathrm{kHz}$ than +65 dB |
| 3. Magnethc P.U. | 3 myV into $50 \mathrm{~K} \Omega$ | Input overload | $+26 \mathrm{~dB}$ |
| Tape and P.U. inputs equa | ised to RLAA curve | Supply | +35 voits at 20 mA |
|  | 20 kHz . | Dimensions | $292 \times 82 \times 35 \mathrm{~mm}$ | with $1 n \pm 1 \mathrm{~dB}$ from 20 Hz to 20 kHz .

MK 60 AUDIO KIT TEAK 60 AUDIO KIT

## ALIO/AL20/AL30 AUDIO AMPLIFIER MODULES

 resulted in a range of output powers from
3 to 10 watts R.M.S. The versatility of their design makes them ideal for use in recoril players, tape reconders,
stereo amplifiers and cassette and cartridge
tipe plapers in the car and at honle.

| Parameter | Conditions | Performance |
| :---: | :---: | :---: |
| harmonic digtortion | Po $=3$ WATT $\mathrm{f}=1 \mathrm{KHz}$ | 0.25\% |
| LOAD IMPEDANCE | - | 8-1/is |
| INPUT TMPEDANCE | $\mathrm{f}=1 \mathrm{KHz}$ | $100 \mathrm{k} \Omega$ |
| FREQUENCY RESPONSE $\pm 3$ 3 | $\mathrm{PO}=2$ WATTS | $50 \mathrm{Hz-25KHz}$ |
| gENHITIVITY for kated o/P | $\mathrm{Vs}_{5}=25 \mathrm{~V} . \mathrm{Kl}=8 \mathrm{~S}_{2} \mathrm{f}=1 \mathrm{KHz}$ | 75 mV . RMS |
| DIMENSIONS | - |  |
| The above tatble relates to the AL10. AL20 and AL30 modules. The following table outlines the differences in their working conditions. |  |  |
| Parameter | All0 AL20 | AL30 |
| Maximurn Suphly Voltage | 25 - 30 | 30 |
| rower outwut fur $2 \%$ T.H.D. $(\mathrm{RL}=\kappa \Omega \mathrm{f}=1 \mathrm{KHz}$ ) PRTCE | 3 watts <br> RMS Min. 5 watts <br> RMS Min. <br> £2.50 22.85 | 10 watts RMS Min. <br> £3. 20 |

TRANSFORMERS
T538(Use with AL20 \& AL30) \&2.30 P. \& P P. 22 p BMT80(Use with AL60) \&2.75P. \&P.40p.
POWER SUPPLIES
P8 12. . (Uee with ALIN, A
SPM 80 . (Use with AL60)
95 p

PA 12. PRE-AMPLIFIER SPECIFICATION
 Treble control-
$\pm 14 \mathrm{~dB}$
at 14 KHIz - Input 1 . Impedance 1 Mck. Ohm
Sensitivity 300 mV nyut 2. Impedance

3 TERMINAL POSITIVE
OLTAGE REGULATORS



## G. F. MILWARD

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We are glad to say that it is now possible to supply from stock the following integrated circuits. ALL ARE BRANDED, FULL SPECIFICATION devices offered at unbeatable prices! This is YOUR chance to cut manufacturing costs and greatly increase profit margins!

\begin{tabular}{|c|c|c|c|c|c|c|c|c|c|c|c|}
\hline \& $$
\begin{array}{r}
1 / 99 \\
60.15
\end{array}
$$ \& $100 / 499$
60.125 \& $500 / 1000$
60.10 \& \& $1 / 99$
60.645 \& $100 / 499$
60.537 \& 500/1000 \& \& $1 / 99$

60.495 \& 100/499 \& 500/1000 <br>

\hline 7400 \& $$
60 \cdot 15
$$ \& \[

60.125
\] \& 60.10 \& 7442 \& 60.645 \& 60.537 \& 60.43 \& 7494 \& 60.495 \& 60.412 \& ¢0.33 <br>

\hline 7401 \& 60.15 \& 60. 125 \& ¢0.10 \& 7443 \& 61.275 \& 61.062 \& ¢0.85 \& 7495 \& 60.63 \& 60.525 \& ¢0.42 <br>
\hline 7402 \& C0. 15 \& ¢0.125 \& 60.10 \& 7445 \& ¢0.855 \& C0.712 \& E0.57 \& 7496 \& 60.72 \& 60.60 \& 60.48 <br>
\hline 7403 \& E0. 15 \& 60. 125 \& c0. 10 \& 7446 \& ¢1.05 \& $\underline{60.875}$ \& 10.70 \& 74104 \& ¢0.315 \& 60.262 \& 60.21 <br>
\hline 7404 \& E0. 18 \& c0. 15 \& 60.12 \& 7446A \& ¢1.05 \& C0.875 \& E0.70 \& 74105 \& 60.315 \& C0. 262 \& ¢0.21 <br>
\hline 7405 \& 60.18 \& ¢0.15 \& ¢0.12 \& 7447 \& ¢1. 05 \& ¢0.875 \& E0.70 \& 74107 \& C0.315 \& ¢0.262 \& ¢0.21 <br>
\hline 7406 \& 60.375 \& <0.312 \& 60.25 \& 7447A \& ¢1. 05 \& C0.875 \& E0.70 \& 74121 \& 60.315 \& C0. 262 \& 60.21 <br>
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\hline 7408 \& £0. 15 \& 60.125 \& E0.10 \& 7450 \& 60.15 \& 60. 125 \& 60.10 \& 74123 \& ¢0.63 \& E0. 525 \& 60.42 <br>
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\hline 7440 \& ¢0.15 \& 60.125 \& 60.10 \& 7492 \& 69.465 \& 60.387 \& ¢0.31 \& 74199 \& ¢2. 10 \& ¢1.75 \& E1.40 <br>
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This beautitul diffuser mada for Phitips sokd origanally at over $\mathrm{f6}$ These are in packs of 6 and we offer thase at the silly pnce of $\mathbf{5 4 . 5 0}$ per pack or $\mathbf{5 4}$ per pack in lots of 10 pacis of more
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B. SATCHWFLI DUOTRONIC CONTROLIER for the contral of ducting (through ZPM modulation motor which we can
suppply. These panals probably con f 50 - EBO ouch. fil each supplyl. These panels probably coant £50-EB0 esch. fit each
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to light it uses 22 Mullard OC 20 powel transistora
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to clear our stock of apprax 50
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500 ohms and over. all with multiple spring sets compnang
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bank, secondhand but usually only dity and will work pertectivy bank, secondhand but ussually only dirty and wilw work perrecty
once cleaned and adjusted 50 p each assorted tots of 50 ar 40p sach to clear our slock of approx 800
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11. SMITHS CLOCK SWITCHES without knobs or glass
fronts. as fitted to cookers. f 1.50 each in lots of 100 or f 1.35 froniz as fitted to cooker. f 1.50 each in
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each in lots of 500 . 51.25 each pel 1.000
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f 12 per drum in lots of 10 drums or f 10 per drum to clear our stock of apprax. 100 druma colours on 500 matra. drums. EA 4 per drum in late of 25 or f3.40 per drum to clear our stock of approx. 200 drums
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 Decca $12^{-}$high mith nubber ferrules. 25 p each in 100 lots ar
20 p each to clear our stock of approx. 2.000 $20 p$ each to cleer our stock of approx. 2.000
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12 V
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0.5
1
2
2
4
6
8
10
10
16
20
30
40
60


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PHASE LOCKED Stereo decoder with Stereo mute, see below
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PUSH BUTTON tuning (with AFC disable) over the FM band (88-104)
CERAMIC funing diodes in back to back
IC STABILISED and S/C protected power supply.
INTEGRATED circuit IF amplifiers for reliability and excellent limiting/AM rejection.
The Nelson-Jones Tuner has all of these features and many more, and more importantly the design is fully proven not just with a few prototypes but with many thousands of working tuners spread across the world.

Typ. Specn: 20 dB quieting 0.75 uV . Image rejection - 70 dB .I.F. Rejection -85 dB
Basic tuner module prices start as low as $£ 12.96$. with complete kits starting at $\mathbf{£ 2 6 . 9 5}$ (mono) + P.P. 65p. and of course all components are available separately
Our low cost alignment service is available to customers without access to a signal generator. Please send large SAE for our latest price lists which details all of the many options and special low prices for complete kits. All our other products remain available.
PORTUS AND HAYWOOD PHASE LOCKED DECODER (W.W. Sept. '70). Still the lowest distortion P.L. decoder available. THD typically $0.05 \%$ (at Nelson-Jones Tuner O/P level)! Supplied complete with Red LED.
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PLEASE NOTE. Existing tuners are readily convertible and kits/parts are available for this purpose.
TEXAN AMPLIFIER. We have designed the tuner case and metalwork to match the Texan amplifier (see photograph). Complete designer approved Texan kits are available at $\mathbf{£ 3 3 . 4 8}$ plus P.P. 65p including Teak Sleeve


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Basic stereo tuner $\mathbf{f 1 5}$ post free. Basic mono tuner f12 post free. 6 position push button units with integral pots $£ \mathbf{3 . 2 4}$.
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No alignment required. Mullard LP1186 front end module used with Ceramic IF and IC amplifier. Push button tuning ( 6 vosition) with Interstation Mute, restricted range AFC, single LED tuning indicator, phase locked IC decoder, and complete metalwork and veneered cabinet. Complete with IC regulated PSU and full assembly instructions. (Mechanically identical to $\mathrm{N}-\mathrm{J}$ Tuner.)

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Made in USSR U4312-low sensitivity
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| OC24 | $\mathbf{0 . 6 0}$ | ASZ17 | $\mathbf{0 . 8 0}$ |
| OC25 | $\mathbf{0 . 5 0}$ | ASZ18 | $\mathbf{0 . 8 0}$ |
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| OC28 | $\mathbf{0 . 7 0}$ | BD116 | $\mathbf{0 . 6 5}$ |
| OC29 | $\mathbf{0 . 6 0}$ | BD121 | $\mathbf{0 . 6 5}$ |
| OC35 | $\mathbf{0 . 5 0}$ | BD123 | $\mathbf{0 . 8 0}$ |
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AC CLAMP VOLT-
AMMETER


TYPE 491 Made in USSR

Measurement ranges: 10-25-100-250500A 300-600V Accuracy: 4\%

## Price $£ 14.00$

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Delivery to any destination can be arranged.

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|  |  |
| Marconi |  |
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| \% |  |
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$\begin{array}{ll}\text { able at extra charge) } & \mathbf{6 0 . 0 0} \\ 6-127 \text {. As above with } 12 \text { Channels } & \end{array}$ SIGNAL SOURCES

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SG70: Audio Oscillator.

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| :--- | :--- | ---: |
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$$
25 \text { Asmorted switches, rotary, lever, micro }
$$

$$
\begin{aligned}
& 25 \text { Assorted } \\
& \text { toggled etc. }
\end{aligned}
$$

$$
\begin{aligned}
& \text { toggled, etc. } \\
& 1050 \text { Preset.Potentiometers. } \\
& 11 \text { Trlal mixed component pack }
\end{aligned}
$$

$$
\begin{aligned}
& 1050 \text { Preset.Potentiometers. } \\
& 11 \text { Trlal mixed component } \\
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POTENTIOMETERS LInear or Log
Rotary Pots Rotary Pots
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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $5 / 64^{\prime \prime}$ or $3 / 32^{\prime \prime}$ or $1 / 88^{\prime \prime}-1$ oz |  |  |  | 1/32"-1 02 |  |  |  | 1/16"-1 oz |  |  |  |  |  |
|  | Single Sided |  | Double Sided |  | Single Sided |  | Double Sided |  | Single Sided |  | Double Sided |  | Single Sided |  |
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| $75 \mathrm{~mm} \times 100 \mathrm{~mm}$ | 14p | 12p | 15p | 13p | 8p | 8p | 8p | 8p | 16p | 15p | 14p | 13p | 8p | 8p |
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If you are interested please write to: Anne Dare, IBM United Kingdom Limited, 389 Chiswick High Road, London W4 4AL. Quoting ref: WW/92418.


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The Plessey Marine Research Unit is engaged in a broad range of research and development in the field of electronics and underwater acoustics. New projects are now creating a number of openings for Draughtsmen/Illustrators and Technical Authors. They will be assisting in the preparation of Admiralty Handbooks describing advanced sonar systems. The equipments contain the most modern digital circuitry employing TTL Logic, ROM's, RAM's and thin film techniques. The research and design laboratories are situated on a country estate at Templecombe, Somerset. Good educational and housing facilities are available in nearby towns like Yeovil, Sherborne and Wincanton, while the Dorset coast is less than an hour away.

## Principal Draughtsman/Illustrator

An experienced Draughtsman/Illustrator is required to take charge of the Illustration Section in a new Technical Publications Department.
The successful candidate must be capable of liaison with customers on technical matters relating to drawings and illustrations. He should be familiar with circuit diagram presentation to BS3939.
It will be an advantage if the candidate has been concerned with technical publications produced to Admiralty Specification NWS $1 / 70$

REF.DI. 50

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This is an ideal opportunity for illustrators to join an enthusiastic team in a new department engaged in the preparation of $M O D(N)$ Handbooks.
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REF.DI. 100

## Draughtsmen/Illustrutors

They should have had a minimum of three years' experience as illustrators in the Electronic industry, but
draughtsmen with a leaning to illustrative work and the ability to prepare good quality diagrams for photographic reproduction should apply. An understanding of reprographics is desirable.

REF.DI. 200

## Technical Authors

To prepare original material for publication, originate draft text illustrations; prepare final copy after approval; read and correct camera copy and printers' proofs. Would work largely on own initiative under limited supervision. Should be educated to O.N.C. standard in electrical engineering or equivalent standard in appropriate subjects coupled with an engineering apprenticeship or service in HM Forces, or other practical experience.

REF.TA. 516

[^5]4541

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# FEDERAL STATUTORY CORPORATIONS SERVICE COMMISSION, 

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(ii) Communication Engineers Grade I
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(b) Outfit and other approved allowances
(c) Terminal Gratuity of $25 \%$ of total Contract Salary earned.
5. Furnished accommodation will be provided where possible at the rate of $8 \frac{1}{3} \%$ of salary, up to maximum of N300 per annum. Alternatively, rent subsidy may be granted in accordance with the current Conditions of Employment (Officers).
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## 6. METHOD OF APPLICATION

Application forms are obtainable from:
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Nigeria House, 9, Northumberland Avenue, London, W.C.2.

## 7. CLOSING DATE

Completed application forms with photostat copies of certificates and two recent passport photographs of the applicant duly signed at the back by the applicant must be submitted to reach the Nigerian Ports Authority Representative at the above address not later than 15th March 1975.

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[4510

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Candidates for post ( $h$ ) should be experienced scientists/engineers who have specialised later in one of the required fields. An ability to deal with nontechnical people is essential.
Appointments will be made within the grades of Higher Scientific Officer except for (e), (f) and (h) where appointments may also be made within the Senior Scientific Officer grade. In addition to the salary scales quoted, all posts attract the Threshold Agreement Payment ( $£ 229$ p.a.) and a non-contributory pension.

## HIGHER SCIENTIFIC OFFICER

Applicants should be under 30 years of age but this requirement may be waived if special qualification or experience can be offered. They should have one of the following qualifications:
(a) A degree in a scientific or engineering subject
(b) Degree-standard membership of a Professional Institution
(c) A Higher National Certificate or Higher National Diploma in a scientific or engineering subject
(d) A qualification equivalent to (c) above

In addition the following relevant experience is required:
(a) Applicants with Ist or 2nd class honours degrees-at least 2 years post-graduate experience.
(b) Applicants with other qualifications-at least 5 years post qualification experience.
Salary Scale: $£ 2,46 \mid-£ 3,371$ with entry point dependent upon experience beyond the minimum required.
SENIOR SCIENTIFIC OFFICER
Applicants should be at least 25 and under 32 years of age, although the upper age limit may be waived if experience of special value can be offered.
Applicants should have obtained a |st or 2nd class honours degree and have had a minimum of four years appropriate post-graduate experience. Salary Scale: $\{3,157-\{4,441$. Entry will normally be at the minimum of the scale but applicants with experience of special value may be entered above the minimum.

Applications, stating the field of work and grade required, should be made to
Administration Officer
HM Government Communications Centre
Hanslope Park
Hanslope
MILTON KEYNES MK19 7BH
[4478

## PROJECT DEVELOPMENT ENGINEER

To consolidate and further develop an established product and also look after the engineering and test requirements of the product in production. H.N.C. in Electrical/Electronic Engineering, with some years electro-mechanical experience would be required.
Applications giving full career details should be sent in confidence to:

> Mrs. J. I. Standfield,
> Personnel Officer,
> GEC Medical Equipment Limited,
> East Lane,
> North Wembley,
> Middlesex.


## OMAN

## DHOFAR REGION TELEVISION SERVICE

We are recruiting on renewable one-year contracts

## Staff-Engineers-Management

for the complete maintenance and operation of the television service.
If you work in television please apply for further information:
MANAGEMENT
PROGRAMME STAFF
ENGINEERS (STUDIO, TX \& M/W \& O.B.) OPERATIONS STAFF/ENGINEERS
ACCOUNTS \& ADMINISTRATION
OFFICE STAFF
LIGHTING
ELECTRICIANS
NEWSCASTING etc.
Let us discuss with you your abilities for these interesting and important positions.
Phone: Tony Owers 01-573 7352 for more information.

$\star$ VERY GOOD SALARY

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* HARD WORK IS NECESSARY

PERSONNEL \& ELECTRONICS LTD.

## GOVERNMENT OF BOTSWANA <br> EXECUTIVE ENGINEER

Required by the Department of Posts and Telecommunications to be responsible to the Assistant Director of Telecommunications for (a) co-ordination of planning, installátion and maintenance of all telecommunications equipment, (b) supervision of Senior Assistant Engineers, (c) expenditure control, (d) preparation of annual estimates and (e) short term planning for network extension.
Candidates, between 40-55 years of age, must possess a recognised degree in Telecommunications Engineering and have at least 5 years' professional experience.
Starting salary up to maximum of $£ 4,610$ in scale $£ 2,800$ to $£ 5, \mathbf{3 5 0}$ according to qualifications and experience, which includes an allowance normally tax-free in scale $\mathbf{£ 6 6 0}$ to $\mathbf{£ 1 , 7 5 2}$.
Engagement is for one tour of 24-36 months in the first instance. Gratuity $25 \%$ of total basic salary. Generous leave. Subsidised accommodation. Family passages. Children's education allowances and holiday visit passages. Interest-free car loan and tax-free Appointment Grant payable in certain circumstances.
The post described is partly financed by Britain's programme of aid to the developing countries administered by the Ministry of Overseas Development.
For further particulars you should apply, giving brief details of experience to: CROWN AGENTS, M Division, 4 Millbank, London SW1P 3JD, quoting reference number M2K/740818/WF.

# B-DNI <br> agents 

## Technical Officer (Components)

British Airways Group Management Services has a vacancy for a Technical Officer to organise and supervise the Central Technical Stores of the Telecommunications Engineering Department.

He will be expected to maintain a close liaison with manufacturers and distributors and keep abreast of rapidly changing technology in the compartment field. He will also work closely with the Accounts and Purchasing sections of Group Management Services.

The responsibilities also include arranging for the shipping and transport of equipment and components to UK and overseas stations, clearing equipment through Customs as required, meeting the requirements of maintenance terms for the supply of components, advising of suitable alternatives where appropriate and providing a technical advisory service on components and accessories.
Applicants should have at least five years' experience in an electronics design or maintenance environment, and preferably an HNC or equivalent certificate in elec tronics or communications. Experience in purchasing and components supply would be an advantage.

The job is based at Heathrow Airport and carries a starting salary of $£ 3,341$ which includes a London Weighting Allowance of $£ 200$.
Additional benefits include an excellent contributory pension scheme, a first-class sports and social club and opportunities for concessional holiday air travel worldwide.
Please write, giving details of age and experience, quoting reference 458/WW/MA, to:
Manager Selection Services, British Airways, PO Box 10, Heathrow Airport-London. Hounslow TW6 2JA.


## Applications are invited from Marine Electronic/ Communications Engineers

with a minimum of five years experience. The positions are available with a rapidly expanding Middle East Company and call for a wide background in the maintenance and servicing of Marime Electronics, Radar and Communications equipment Applicants should show evidence of having recognised and qualified experience in any three of the following fields:

1. Low and medium Power MF/HF

Transmitters and Receivers
2. Low Power Solid State SSB Transceivers.
3. Marine VHF Radio Telephone Equipments.
4. Marine navigating Radar equipments and' other navigational aids.
5. Gyro Compass and ship steering Servo Systems.
Successful applicants, following interviews in London will be engaged on a contract basis in the Arabian Gulf. Initial salary will be in the region of $£ 3,500-£ 4,000$ pa. plus fringe benefits and bachelor status accommodation, one month's overseas leave, return air passage paid annually, plus earned gratuity. Possibilities would be available for married status after proving ability in this area. Replies only will be sent to engineers actively servicing equipments at this time in a similar senior position. Reply in confidence to:

The Managing Director,
P.O. Box 1788.

Dubai,
United Arab Emirates.

## UNIVERSITY OF

 NEWCASTLE UPON TYNE Department of 'Photography and Teaching Aids Laboratory
## Colour <br> Television Engineer

To be responsible to the Head of the Film and Television Section for the operation of an off-air colour recording. playback and transeription service. He will begin to assist in the immediate planning of a new colour system to be commissioned in 1976 for the new Dental School and Medical School and or the subsequent phased development of University.

Applicants should have several years' experi ence of colour programme origination and video tape recording, and preferably some planning. He must be familiar with colour and monochrome studio equipment of all types, and capable of establishing and maintaining professional standards.
Salary at a suitable point on the scale £1,683-£2,931 plus a threshold payment of E 229.68 per annum, according to age, qualifications and experience. For an exceptionally well qualified and experienced candidate the appointment may be made on the higher scale £2,757. £4,341 (plus threshold) (scales are under review). Membership of an appropriate University superannuation scheme will be required.
Further particulars may be obtained from the Registrar, The University. Newcastle upon Tyne, NEI 7RU, with whom applications (three copies) stating age, education, job experience, availabitity for interview and later than 30th April 1975. Please auote reference W.W. 4568

## (1) | 1 LIMITED,

Manufacturers of modern FM radio communication systems for all branches of industry, transport and Public Authorities require additional

## TEST TECHNICIANS

based in Camberley to assist in the final testing of personal and mobile radio equipment and sophisticated control systems.
Knowledge of RF, digital and thick film techniques desirable with academic levels to ONC or C. \& G. Final, but for an applicant with exceptional experience and knowledge these qualifications may be waived.
Pleasant working conditions, good salary and overtime. Opportunities for further study and training.
Hours: Monday-Thursday:
$8.15 \mathrm{am}-1.00 \mathrm{pm} .1 .30 \mathrm{pm}-4.45 \mathrm{pm}$. Friday:
$8.15 \mathrm{am}-1.00 \mathrm{pm} .1 .30 \mathrm{pm}-3.30 \mathrm{pm}$.
Apply: The Personnel Officer,

## Storno LIMITED,

Frimley Road,
Camberley. Telephone: 027629131

-1 1 1 SERVES THE NATION.

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## $\begin{array}{lllllll}M & E & R & C & \mathbf{U} & \mathbf{R} & \mathbf{Y}\end{array}$

## PROJECT ENGINEERS BROADCAST TELEVISION

To cover and extend our increasing international commitments, we are seeking to further expand our team of engineers working on broadcast television systems design and installation.
This work involves both static studio installations and Outside Broadcast vehicle construction, and may be located at Uxbridge or Westbury, Wiltshire.
The potential ability and confidence to assume total responsibility for the planning and execution of complete broadcast systems is an essential requirement, together with the personality to deal with a wide variety of people in the course of this work.

The engineers we are looking for will have formal qualifications to at least HNC level or eguivalent, with detailed knowledge of one or more aspects of broadcast television. Experience of operational work within this sphere will be particularly useful. Overseas travel, occasionally for extended periods will be involved.
In return, we can offer you a varied, demanding and rewarding career with a young, vigorous company which is rapidly expanding and establishing a considerable reputation for itself in a highly competitive field.
Please write giving FULL details of your qualifications and experience or phone for an application form to:-

> UXBRIDGE 39876/39613 MERCURY ELECTRONICS, G ROCKINGHAM WHARF, ROCKINGHAM ROAD, UXBRIDGE, MIDDLESEX

## ENGINEER (WithTVService Experience)

R.S. Components Limited, Britain's biggest distributor of electronic components, requires an experienced Engineer or Service Manager who is currently employed on TV service work. Ideally, he should be between 25-45 and possess a good academic background.

This is a new post and an exciting opportunity for an engineer who is eager to further his experience by becoming our technical adviser on component requirements in the field of TV and Audio Equipment servicing.

Duties will include component evaluation, specification and assisting in answering customers telephoned enquiries. Additionally, the candidate will be expected to maintain a close liaison with the service industry and manufacturers.

We can guarantee an interesting career which may occasionally involve travel in the U.K.

This new important post commands a good salary commensurate with ability and there is every opportunity for advancement. Excellent working conditions, generous holiday entitlement and pension scheme.

Write giving brief résumé of your career to date or ring for an application form to

## Chief Engineer

## R.S. Components Limited

13-17 Epworth Street, London
EC2P 2HA. Tel: 01-253 1222

- An Electrocomponents Group Company


## CENTRAL BIRMINGHAM HEALTH DISTRICT

# ELECTRONICS TECHNICIAN 

(M.P.T. II)

A vacancy exists in the electronics section of the Medical Physics and Biomedical Engineering Department for an experienced Technician with H.N.C. or equivalent, competent to take responsibility for the servicing and development of biomedical electronic equipment throughout the Teaching District and to act as Deputy Head of the section. Experience of medical electronics advantageous, but good general electronic experience essential.

Salary: $£ 2,601-£ 3,390$ per annum, plus Threshold.

Further particulars and application form from the
PERSONNEL OFFICER,
QUEEN ELIZABETH MEDICAL CENTRE, EDGBASTON,
BIRMINGHAM B15 2TH.

## TONGA

## SUPERVISING BROADCASTING TECHNICIAN

required by the Tonga Broadcasting Commission to be responsible for the operation and maintenance of the Commission's two 10 Kilowatt sound transmitters, to install and maintain studio equipment, to run a radio retail store involving technical supervision in purchasing, selling and repairing of receivers and other equipment.
Candidates, under 55 years of age, MUST have a City and Guilds Telecommunications Technician Final Certificate Course 271 or equivalent with 10 years' experience in the operation of studio and transmitter equipment as well as in all aspects of a small broadcasting station with particular emphasis on sound transmitters.
Salary in scale $£ 2,125$ to $£ 3,400$ pa which includes an allowance normally tax free in scale $£ 504$ to $£ 1,404$ pa and $\mathbf{2 0 \%}$ cost of living allowance. Gratuity $20 \%$ of Local salary. Tour of 2 years.
Benefits include free passages, Government housing at moderate rental. Holiday visit passages and generous paid leave. An appointment grant of $£ 300$ and car loan of $£ 600$ may be payable.
The post described is partly financed by Britain's programme of aid to the developing countries administered by the Ministry of Overseas Development.
For further particulars you should apply, giving brief details of experience, to CROWN AGENTS, M Division, 4 Millbank, London SW1P 3JD, quoting reference number M2K/740928/WF.

CITY OF LONDON POLYTECHNIC

## SENIOR ELECTRONICS TECHNICIAN

(GRADE 5)
required immediately in the Department of Biological Sciences for the maintenance, design and operation of electronic and other instruments, especially those used in Neurophysiology. The successful candidate must possess the relevant qualifications at HND/HNC or CGLI level, together with at least seven years relevant experience (including training period). Salary $£ 2,439-£ 2,895$ plus $£ 411$ London Weighting (starting point dependent on qualifications and experience). Apply, in writing, giving full details of qualification, experience, etc. and including the names and addresses of two referees, to the Laboratory Superintendent, Biological Sciences, Calcutta House Precinct, Old Castle Street, London EI 7NT.

## University of Surrey

Audio Visual Aids Unit

## TECHNICIANS

## (T4 £2247-£2628)

(T2B £1860-£2187)
The AVA Unit is responsible for projection and allied services in 26 Central Lecture Theatres, and also provides services of photography, film and television for teaching and research throughout the University.

These new posts are for skilled technicians who will be responsible for installation, maintenance and repair of a wide range of Audio Visual equipment ranging from slide projectors to television. The Unit is well equipped for electronic and mechanical servicing, and instrument making and repair work.
For the Grade 2B post experience in one or more of these fields is essential, although training in specific techniques will be given where necessary:
For the Grade 4 post experience in electronic servicing is essential and candidates should hold an ONC or equivalent qualification.

Applications immediately on forms available from: Assistant Secretary (Personnel), University of Surrey, Guildford, or Tel: Guildford 71281 Ext. 452

## SERVICE ENGINEER

We are the distributors of World renowned Tandberg Products and are looking for a Service Engineer who has had experience in the Service and Repair of domestic $\mathrm{Hi}-\mathrm{Fi}$ Equipment. Up to date Service Facilities and good working conditions, 5 day week with 3 weeks Annual Holiday. Wages up to $£ 2,500$ per annum depending on experience.

Please apply in writing with details of Career to date to Mr. D. D. Hamilton, London Manager, FarnellTandberg Ltd, 167, Hermitage Road, London N4 ILZ. 14578

## Electronics Engineer

Our Research Function carries out innovative research through a number of project groups supported by certain essential specialised services. We are seeking an Electronics Engineer to join the Laboratory Services group in trouble shooting, maintenance and some development work.

Responsible to the Laboratory Manager, he will provide a service to all of the departments in our new research laboratories where the electronic equipment includes infra-red, ultra-violet, NMR and mass spectrometers as well as chromatographic equipment, calculators and recorders. There is also a Fourier transform NMR instrument incorporating a small computer.

The man we are looking for will be in his late twenties or thirties, qualified to HNC or possible degree level and he will have had some experience of service and development work preferably in a multi-disciplinary academic or industrial research laboratory. Specific experience in the field of NMR electronics would be an advantage. The person we appoint will be working largely without direct supervision and he should therefore be capable of accepting this degree of responsibility.

Roche Products Limited is part of one of the world's largest and most successful pharmaceutical companies and is itself one of the leading companies in the U.K. Working conditions are excellent and the Conditions of Service include some valuable fringe benefits.

Please apply in writing, quoting reference R50 to the Personnel Manager.

## ROCHE

Roche Products Limited, PO Box 8, Welwyn Garden City, Herts AL7 3AY

## Service Area

 PlanningEngineers£2488-£3019

The Independent Broadcasting Authority requires two Junior Engineers to assist Engineers in charge of field teams with the planning and execution of the UHF television service area surveys, RBL tests and other field work. The people appointed to the posts will also assist with the general UHF television and independent local radio planning work of the section.
Candidates should preferably be qualified to HNC or equivalent level and should have some basic knowledge of radio wave propagation and television principles, plus experience of radio frequency measurement.
The posts are based at Crawley Court, near Winchester, Hampshire, however a considerable amount of travelling throughout the UK will be involved for which appropriate allowances will be payable. Candidates should have a current driving licence and should preferably have the ability to climb acrial support structures up to about 150 feet.


INDEPENDENT BROADCASIING AUTHORITY

Please write or telephone for an application form quoting Ref. DT/2670 to:- Miss Vanessa Aldred, Independent Broadcasting Authority, Crawley Court, Winchester, Hants. SO2I 2QA. Tel: Winchester 822327.


4537

## ELECTRONIC CRAFTSMEN

Is your present job routine and uninteresting?
We are a research establishment and our craftsmen are engaged on a wide variety of work in the fields of prototype and small batch wiring and assembly, test and inspection, maintenance fault finding and repair. Why not join us and enjoy working in first class conditions in the country.

You can expect gross earnings including overtime of $£ 45$ per week, and we can offer good housing at low rental (for applicants who reside outside the radius of our Assisted Travel Area) together with 3 weeks paid holiday with holiday bonus, free pension and excellent sick benefit scheme.

Applicants who should have served a recognised apprenticeship or have had equivalent training together with experience in one of the fields detailed should 'phone Tadley 4111 (STD 07356 4111) Ext. 5230, or write to:

## INDUSTRIAL RECRUITMENT OFFICER <br> (PA/79/WW) PROCUREMENT EXECUTIVE <br> MINISTRY OF DEFENCE <br> AWRE ALDERMASTON <br> READING, BERKS. <br> RG7 4PR.

## ELECTRONICS TECHNICIAN JUNIOR ENGINEER

Systems Company requires 20/30 years old Engineer for development, Commissioning and Maintenance of minicomputer based remote batch terminals.

Good opportunity for either an experienced Man to establish himself in a fast growing and friendly Company or for a Young Man to acquire experience of the latest technology in mini computers and peripherals.
Full training will be given. Some travel U.K. Salary range $£ 1,800-£ 2,600$ p.a. Write or phone: Peter Rogers or Steve Clifford.

TASK TERMPNALS LTD.<br>117, Cleveland Street, London, W.I.<br>01-637 4516

## SIEMENS <br> MEDICAL ENGINEERING <br> Service and Sales Engineering

Service and Sales Engineers required for Electro-Medical Department, to work in the London area. Previous experience in this field an advantage, but knowledge of electronics essential.
Applications to:

## SIEREX LTD.,

Heron House, 109 Wembley Hill Road,
Wembley, Middlesex, HA9 8BZ.
[4570

## HARINGEY

## Education Services

## Full-time

## Luborafory Technician

required at Stationers' Company's School, Mayfield Road, N.8, to work 35 hours per week x 52 weeks per annum.
Salary rising to $£ 2,677$ per annum including threshold payments. Commencing salary according to qualifications and experience.
Minimum Qualifications: Ordinary National Certificate or Ordinary National Diploma; City and Guilds Laboratory Technicians Certificate; 4 G.C.E. passes with 2 at ' $A$ ' Level in appropriate subjects; Membership of Institute of Science Technology OR an equivalent suitable qualification OR 5 years suitable experience. Qualifications in Electronies would be an advantage.
Candidates will be responsible for the maintenance of the Language Laboratory and will be required to of the Language Laboratory and will be required to
assist in the upkeep of Audio Visual aids throughout assist in the upkeep of Audio Visual aids throughou
the school and help monitor a computer link-line.
The post is ideal for a candidate who wishes to gain experience in the maintenance of a fairly wide range of equipment.
Application forms obtainable from Chief Education Officer, Somerset Road, N.I7, to be returnable 7th Mareh, 1975.

## APPOINTMENTS

## BRUNEI

TELEVISION ENGINEER

* Posting Bandar Seri Begawan
* Engagement for three years initially
* Gratuity 25\% of total salary drawn
* Free family passages
* Furnished quarters at reasonable rental
* Children's education allowances and holiday visit passages
* Interest free car loan
* There is no income tax payable in Brunei at present

The Brunei Television Service require a Supervisory Engineer (Transmitters) to be responsible to the Superintending Engineer for the efficient operation and maintenance of all transmitting equipment; also routine inspection and maintenance of aerials and feeders on towers $400 / 450 \mathrm{ft}$. high and to undertake the training of local staff. Candidates, preferably under 55 years of age, must hold a recognised qualification in colour television engineering, and have spent at least 5 years in a supervisory position in a PAL colour television transmitting station. Experience should include parallel operation of Band III transmitters of 5 KW and higher output towers and the installation. operation and maintenance of microwave link equipment. Salary. according to qualifications and experience in the scale $£ 3,166$ to £5.750 approximately.

For further particulars you should apply, giving brief details of experience, to CROWN AGENTS, M Division, 4 Mitlbank, London SW1P 3JD, quoting reference number M2K/740804/WF.


## Maintenance Engineer

## £3238-£.3928

We require an experienced Aerial Engineer in the Station Operations and Maintenance Department to be responsible for the maintenance of UHF, VHF and SHF Transmitting and Receiving Aerial Systems.
The post is based at the Authority's North Regional Office in Leeds and the duties will mainly be confined to stations within the North of England, although duties throughout the UK may from time to time be required.
The work will require the successful applicant to travel extensively in the fulfillment of his duties (a car will be provided). In addition, because of the nature of the broadcasting service, duties outside 'normal office hours' will be required.

A minimum of three years' experience in the microwave transmission field on work involving aerial arrays and coaxial line assemblies and filters used in broadcasting bands is essential.
Applicants should preferably be qualified to HNC level or equivalent and/or should be able to demonstrate a sound theoretical understanding of aerial and transmission line systems.
Applicants should possess a current driving licence and should be prepared to climb masts up to 1250 feet in height.
The commencing salary will be within the above range, depending upon qualifications and experience.

Please write or telephone for an application form quoting reference number 2596 to:
Vanessa Aldred, Independent Broadcasting Authority, Crawley Court, Winchester, Hants. Telephone: Winchester 822599.

## GUY'S HOSpITAL <br> <br> MEDICAL PHYSICS <br> <br> MEDICAL PHYSICS TECHNICIAN GRADE II TECHNICIAN GRADE II AND AND <br> <br> ELECTRONICS TECHNICIAN/ <br> <br> ELECTRONICS TECHNICIAN/ ENGINEER GRADE III ENGINEER GRADE III <br> Department of Clinical Physics and Bioengineering

The Grade II Technician is a member of a team of physicists and technicians engaged in a variety of clinical instrumentation projects. ONC, HNC or higher qualification required together with 2 years electronics experience in Technician III Grade or other relevant technical experience. Basic salary from $£ 2,601-£ 3,390$, starting point according to experience.

The Grade III Technician post is for an Electronics Technician/Engineer engaged upon maintenance, repair and calibration of a wide maintenance, repair and calibration of a wide in electronics required plus at least 3 years elec. tronic instrument maintenance experience. Basic Salaryy from $£ 2,190-£ 2,817$.

Apply to Personnel, Guy's Hospital, London SE1 9RT. Telephone 01-407 3662 Ext. 68.
[4514

## RADIO TECHNICIAN FOR <br> CENTRAL AMERICA

Needed to work in Guatemala with the Radio Schools Movement, training a team of Guatemalans in the maintenance and repair of station equipment. A British Volunteer Pragramme post.
Information:
Paddy Coulter, Overseas Vołunteers/CIIR 41 Holland Park, London W.11. [\$577

## KRLINGBECK HOSPITAL, YORK ROAD, LEEDS 14

AN

## ELECTRONIC TECHNICIAN <br> (MEDICAL PHYSICS TECHNICIAN III)

is required for the Cardiovascular Unit. The hospital is the Regional Cardiothoracic Centre.
The work involves the servicing of patient monitoring and biochemical analysis equipment.
Basic qualifications required: ONC, HNC or HND. Experience in repair of audioamplifiers or TV servicing would be an advantage.
 plus current cost of living allowance.
Application form and job description from Personnel Officer, Seacroft Hospital, York Road., Leeds LS14 6UH. Telephone 648164 Ext. 253.

## BEACON BROADCASTING

## the Local Radio Station for Wolverhampton and the Black Country

 invites applications for the post of
## CHIEF <br> enginetr

The applicant must have a sound technical knowledge of local sound broadcasting and should ideally have had experience in setting up a local station and all the I.B.A. technical requirements.
Write giving details of past experience to:
BEACON BROADCASTING LIMITED
56/57 QUEEN STREET, WOLVERHAMPTON

## TEST ENGINEERS

Thorn Automation Limited, a recognised leader in the field of Electronic Industrial Control Equipment wish to appoint several Test Engineers to test a wide range of electronic industrial control equipment.
Applicants should have had experience in the testing of electronic control equipment, together with some experience in digital logic techniques and S.C.R. regulations.
The company is situated in pleasant rural surroundings within easy reach of new housing developments and several large towns.

If you would like to know more about these interesting and rewarding positions telephone or write to:

Peter Williams,


Personnel Officer.
THORN AUTOMATION LIMITED,
P.O. Box 4, Rugeley. Staffs WS 15 1DR

Telephone Rugeley 5151

## APPOINTMENTS

R EDIFON TELECOMMUNICATIONS LTD., Lon don, S.W.18, have a vacancy for an enthusiastic, practical man with some experience of Volume Production Testing in the electronics industry. Phone 01-874 7281 and ask for Len Porter.
"HE MIDDLESEX HOSPITAL, London, W.1. tions are invited for the post of Medical ApplicaTechnician Grade III in the Department of Clinical Measurement. Qualifications will be based on Whitley Councils Professional and Technical ${ }^{\circ}$ Scales. Duties will include a wide variety of work with apparatus used for physiological measurement and candidates should possess suitable qualifications i.e. some electronics experience and an ONC; HNC, HND or some other appropriate science degree. Applications should be made to the Establishment Officer, The Middlesex Hospital, London WIN 8AA
as soon as possible. Psychological Medicine at St, Pancras Hospital. The post is funded by a Grant from the Leverhulme Trust and will be available for two years, Salary according to age and qualifications within the range of $£ 1,860$ to $£ 2,187$ plus London Weighting- $£ 410$ and Threshold Payments. Applications to and further particulars from Dr. D. A. Sturgeon, Department of Psychological Medicine, St. Pancras Hospital, 4, St.
Pancras Way, London, N.W.I.

## SITVATIONS WANTED

CXPERIENCED Radio/Technician 23-years, seas/U.K. C \& G TT4, Fully conversant went overern radio telephones/UHF/VHF/control with modpractice. Available at reasonable short notice Box No. WW 4505.

## SITUATIONS VACANT

ELECTRONICS ENGINEER required for Central London recording studio. Experience in audio electronic work. Must be keen and prepared to work long hours. Box No. WW 4548.
ELECTRONIC Wireman and Tester, A vacancy ing company for small, West Country manufacturon printed circuits, to lay out circuits, to work prototypes, to evaluate, test and fault-find on standard and prototype units, and to generally carry out quality control inspection. Applicants must have previous industrial experience of this work. Qualifications are not important, provided that the individual has experience and enthusiasm. Applica tions, giving full details of age, experience, etc., to Eox No. WW 4519.
CLECTRONIC engineer to design equipment for use of the physically handicapped. Knowledge/ munications, computers preferred. Interesting, comof work with small Company in pleasant location Grange Electronics Ltd., Stone Lane, Wimborne Dorset.
[4551

## (1世 ARTICLES FOR SALE

A UTOMATIC TEST SET, teleprinter, tape punch, double 19 in . datum case on R. S. J. wheeled dolly. £60. Medway 55888 and 33168 . $[4546$
A ARVAK ELECTRONICS, 3 Channel SoundLight Converters from £17; Strobes, £21; Rainbow Strobes, £133. Free catalogue. 98A (W), West Green Road, (Side Door), London NI5 5NS. 01-800
8656. 23
BRENNEL M.K. 6 deck, new, unused, bought for WW Stuart Tape-Recorder project, but latter never built. Cost $£ 85$. Offers to Box No. WW 4503.
CLEARING distributor stocks, transistors, diodes, components, etc. Sample pack 65p incl., postage Lodge Close, Rickmansworth Herts. Mail Maple Only. [4499
CONSTRUCTION AIDS-Screws, nuts, spacers, etc., in small quantities. Aluminium panels punched to spec. or plain sheet supplied. Fascia printed circuit boards-masters, hoard, one-off or small numbers. Send 9p for list Ramar Constructor Services, 29 Shelbourne Road, Stratford on Avon, Warwks. Tel. Stratford on Avon (std 0789) 4879. [28
DIGITAL CLOCK CHIP, AY-5-I224, with data and circuit diagram, $£ 3.66$ plus VAT. 'Jumbo' LED digits ( 16 mm high) type DL-747, only $£ 2.04$ 94 New Chester Road, Wirral, Merseyside L62 94 N .

| $[83$ |
| :---: |

HEATH $10-102 \mathrm{DC}-5 \mathrm{MHz}$ Scope. Solid-State. 1 Like new. Less than 25 operating hours. £60. 26 Oberon Close, Hartford, Huntingdon, Cambs.

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## Appointments Vacant Advertisements appear on pages 80-100

|  | Page | Page | Page |
| :---: | :---: | :---: | :---: |
| Action Video |  | Harris Electronics (London) Ltd. ............... 13, 18 | Quality Electronics Lid. ................................ 17 |
| Aero Electronics Ltd. | 17, 20 | Harris, P. .................................................. 101 | Quartz Crystal Co. Ltd. .................................. 10.1 |
| Ambientaccoustics | ... 31 | Hart Electronics ....................................... 48 |  |
| Ancom Lid. | .. 10 | Hayden Laboratories Ltd. ............................. 22 |  |
| Anders Electronics Lid. | . 4 | Heath (Gloucester) Ltd. ................................. 2 |  |
| Arrow-Hart (Europe) | . 31 | Hengstler G.B. Ltd. ..................... Readers' Card | Radford Audio Ltd. ................................... 14 |
| Aspen Electronics Lid. | . 28 | Henry's Radio Ltd. ...................................... 59 | Radio Shop, The ........................................ 52 |
| A.S.P. Lid. ............. | . 70 | H.H. Electronics ....................................... 19 | Radio T.V. Components ............................ 62 |
| A. \& S. T.V. Components | . 57 | Hi-Fidelity '75' ........................................... 77 | Rank Audio Visual ..................................... 40 |
|  |  | Hi-Fi Designs .......................................... 45 | R.E.W. Audio Visual Co. ............................. 42 |
|  |  | Hitachi-Shibaden (U.K.) Lid. ......................... 40 | Rola Celestion Ltd. ........................................... 32 |
| Barrie Electronics Ltd. ................................ 53 |  |  |  |
| Bell \& Howell Ltd. ....... | .... 39 | Icon Design ............................................. 17 |  |
| Bentley Acoustic Corp. Ltd. | .... 69 | I.L.P. (Electronics) Ltd. ................................... 68 |  |
| Bi-Pak Semiconductors | 60,61 | Industrial Sub Assembly Lid. ........................ 27 | Sanyo Marubeni (U.K.) Ltd. ....................... 43 |
| Bi-Pre Pak Ltd. ... | ... 68 | Industrial Tape Applications Lid. . ...................... 19 | Samsons (Electronics) Ltd. ............................. 70 |
| Bias Electronics Ltd. | .. 27 | Integrex Ltd. ................................................ 72 | Scientronics ................................................. 57 |
| B.H. Components Factors Ltd. | .... 78 | I.T.F.-L.E.C.S. ${ }^{\text {a }}$. | Semicron Indexes Lid. ................................ 13 |
| Broadfields \& Mayco Disposals | $\ldots . .101$ | I.T.T. Instrument Services .......................... 15 | Scott. James Electronic Eng. Ltd. .................. 12 |
| Bul!, J. Electrical Ltd. ................................. 69 |  |  |  |
|  |  | Jackson Bros. (London) Ltd. ....................... 32 |  |
|  |  | Jasmin Electronics Ltd. ........................................ 103 |  |
| Cambridge Learning | .. 3 | J.E.F. Electronics $\ldots$................................... $10 . .$. | Shure Electronics Ltd. ................................ 23,35 |
| Carston Electronics | .. 73 | J.H. Assockates Ltd. ................................... 79 | Sinclair Radionics Ltd. ................................. 38 |
| Catronics | . 53 |  | Sintel S.................................................. 101 |
| Cerutti, S.A. \& Cie. | .... 77 |  | S.J. System Designs ........................................... 52 |
| Chiltmead Ltd. ...................... | 1,104 | K.E.F. Electronics Ltd. ............................... ${ }^{7}{ }^{7}$ | Sowater, E. A., Ltd. .......................................... 101 |
| Chromasonic Electronics Ltd. |  | Keytronics Ltd. ................................................ 101 | Sugden, J. E., \& Co. Ltd. .............................. 20.102 Surrey Electronics |
| Colomor (Electronics) Ltd. |  |  | Surrey Electronics ..................................... 102 |
| Computer Sales \& Services | . 78 |  |  |
| Concorde Instrument Co. | ... 52 |  |  |
| Crichton, J. ......... | 48 | Lynx Electronics ........................................................ 48 |  |
| C.T. Electronics Ltd. ............ | ... 76 | Lynx Electronics ........................................ 48 | Telequipment Products (Tektronix U.K.) Ltd. 34 |
|  |  |  | Teleradio Special Products .......................... 101 |
|  |  | Macfarlane, W. \& B. ................................... 58 | Teletape Video, The London Cassette Centre ... 41 |
|  |  | Macinnes Labs Ltd. ........................................ 30. | Teonex Ltd. ............................................... 6 |
| Dixons Technical CCTV Ltd. Drake Transformers Ltd. | .... 17 | Maplin Electronic Supplies .......................... 65 | Thurlby Electronics ....................................... 103 |
|  | ... 10 | Marco Trading Co. ..................................... 102 | Trampus Electronics ................................. 46 |
|  |  | Marconi Instruments Ltd. ......................... cover ii | Trec Consultants Lid. .................................... 40 |
|  |  | Marshall, A., \& Sons (London) Ltd. ................ 58 <br> McKnight Crystal Co. .............................................. 101 |  |
| East Cornwall Components | 50 | McLlennan Eng. Lid. ....................................... 16 |  |
| Eddystone Radio Ltd. ....... | . 21 | Mills, W. ...................................................... 57 | United-Carr Supplies .................................. 37 |
| Electronic Brokers Ltd. ..... | 74, 75 | Milward, G. F. ............................................. 64 |  |
| Electronic Mech. Sub Assembly | .... 101 | Modern Book Co. ......................................... 10. |  |
| Electrovalue ....... |  | Multicore Solders Ltd. ........................... cover iv |  |
| Elektor Publications Ltd. | ... 22 |  | Valradio Ltd. $20$ |
| E.M.I. Telecommunications ... | .... 36 |  | Vero Electronics ...................................... 18 |
| Eurotype .................................................. 27 |  | Naim Audio Ltd. ............................................ 26 <br> Newnes Buterworths .............................................. 50 | Vortexion Ltd. .................................. cover iii |
| Farnell Instruments | .. 28 | Ocli Optical Coatings Ltd. .......................... 12 | Wayne, Kerr, The, Co. Ltd. ....................... 8 |
| Ferrosraph, The, Co. Ltd. | ... 14 |  | West Hyde Developments Lid. ............................ 101 |
| Fi-Comp Electronics | .. 56 |  | Whiteley Electrical Radio Co. Ltd. ............... 20 |
| Future Film Development Lt |  |  | Wilkinson, L. (Croydon) Ltd. ........................ 57 |
| F.W. Electronics | .... 101 | P. \& B. Electronics Lid. .............................. 76 |  |
| Fylde Electronic Labs. Ltd. ... | .... 79 | Pattrick \& Kinnie ...................................... 66 | Wireless World (3 in 1) ................................ 49 |
|  |  | Phoenix Electronics (Portsmouth) Lid. ........... 26 | Wireless World Annual ................................ 30 |
|  |  | Physical \& Elec. Labs. Ltd. .......................... 56 |  |
| Gardners Transformers Ltd. |  | Precision Petite Ltd. ............................................. 77 |  |
| Grampian Reproducers Ltd. | . . 102 | Prosser Scientific Insts. ................................. 28 | Z. \& I. Aero Services Ltd. .................... 16,72 |
| Greenwood Electronics Ltd. | .... 11 | Pye Unicam ............................................. 26 | Zettler GmbH ............................................. 19 |

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| Sn62 | 179 | 179 | QQ-S-571E |
| Sn63 | 183 | 183 | QQ-S-57 1E |
| K | 183 | 188 | B.S. 219 |
| Sn60 | 183 | 188 | QQ-S-571E |
| F | 183 | 212 | B.S. 219 |
| Sn50 | 183 | 212 | QQ-S-571E, |
| Savbit 1 | 183 | 215 | DT1) 900/4535 <br> IUIN 1707 |
| R | 183 | 224 | B.S. 219 |
| G | 183 | 234 | B.S. 219 |
| Sn40 | 183 | 234 | QQ-S-571E |
| J | 183 | 255 | B.S. 219 |
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