# WirelessWorld 

February 1969 Three Shillings

## Operational amplifiers Multimeter construction




## Cut the operational and maintenance costs of your HF radio station right now -with STANFAST

## Here's how

STANFAST Systems-the STC concept of automated h.f. radio stations-permit transmitting and receiving installation to be controlled completely by one man from a central location.
STANFAST Systems provide high speed frequency changing, automatic performance monitoring and rapid fault location affordingoptimum traffic handling capability and maximum revenue.

STANFAST Systems use the latest techniques in radio design, demand smaller sites and require less maintenance than hitherto. Initial capital cost is lower and return on investment is greater.

Standard Telephones and Cables Limited, Communications Division, New Southgate, London N.11. Telephone: 01-368 1200. Telex: 261912.

## When is an Avo meter not an Avometer?



## When it tests nuvistors, compactrons \& 13 -pin valves

The new Avo VCM163 Valve Characteristic Meter is one of the most versatile valve testers ever developed. With facilities for testing valves with as many as 13 pin connections (and 2 top caps), plus recently introduced types such as nuvistors and compactrons, the VCM163 provides both rapid fault diagnosis and comprehensive static/dynamic characteristics data. Nevertheless, it is even simpler to use than previous models - no backing-off is required. A separate meter displays mutual conductance values continuously during testing, and there is pushbutton monitoring of screen parameters. The full range of $h . t$. voltage -12.6 V to 400 V - can be applied to anode and screen, heater voltage is adjustable in 0.1 V steps from 0 to 119.9 and grid voltage may be varied continuously from 0 to 100 V (calibrated). Get complete information about the VCM163 from your local dealer or Avo Ltd, Avocet House, Dover,
 Kent. Telephone Dover 2626. Telex 96283.

# Ferrograph Series 7a lifetime of recording 

Ferrograph Tape Recorders have been famous ever since 1949. A lifetime's experience of making fine recorders goes into every one of Ferrograph's brilliant new Series 7.
And there is a lifetime's recording in every Ferrograph instrument. Many of the earliest Ferrographs are giving perfect service today, nearly twenty years later. You can be sure your Ferrograph will do the same for you. It will give dependable service for many, many years to come. It will keep its value. It will need the minimum of service. Spare parts will remain available for a lifetime's recording. That's how Ferrograph got its name.

Available in Mono, and in Stereo with and without end amplifiers: combining a unique range of 30 recording facilities, including:

- All silicon solid-state electronics with FET input stages and wide input overload margins. - Vertical or horizontal operation.

Unit construction: The 3 individual units i.e. tape deck, power unit and amplifier complex are mounted on a single frame casily removable from cabinet for service or installation in other cabinets or racks.

- 3 motors (no leelts). 3 tape speeds.
- Variable speed spooling control for easy indexing and editing.
- Electrical deck operation allowing pre-setting for time-switch starting without need for machine to be previously powered.
- Provision for instantaneous stop/start by electrical remote control.
- Single lever-knob deck operation with pause position.
- Independent press-to-record button for safety and to permit click-free recording and insertions.
- $84^{\prime \prime}$ reel capacity.
- Endless loop cassette facility.
- Internal loud speakers (2)--1 each channel on stereo, 2 phased on mono.
- 4 digit, one-press re-set, gear-driven index counter.
© 2 inputs per channel with independent mixing (ability to mix 4 inputs into one channel on stereo machine).
- Signal level meter for each channel operative on playback as well as record.
- Tape/original switching through to output stages.
Re-record facility on stereo models for multiplay, echo effects etc, without external connections.
- Meters switchable to read 100 kHz bias and erase supply with accessible preset adjustment.
- Three outputs per channel i.e. (1) line outlevel response. (2) line out-after tone controls.
(3) power outpul-8-15 ohms.
- Power output low per channel.
- Independent tone controls giving full lift and cut to both bass and treble each channel.
- Retractable carrying handle permitting carrying by one or two persons.

U.K. Retail prices from $\mathfrak{f} 150$ incl. P.T.

See and hear Ferrograph Series 7 recorders at your local Ferrograph stockist, or post coupon for details and address of nearest Ferrograph specialist (or ring 01-589 4485)


Don't take our word for it-test EEV flash tubes against the equivalents you're now using and learn why other users think so highly of those made by EEV. Incorporating extra heavy duty electrodes, EEV flash tubes are renowned for their reliability, long life (up to $10^{6}$ flashes) and high conversion efficiency. EEV liquid-cooled and air-cooled xenon flash tubes for pumping laser rods offer a wide range of input energy levels and they are capable of operation at high repetition rates.
Full details of the range are available on request-

## Outstanding in quality, reliability in quality, reliability and performance

 but if your application calls for a flash tube that is not in the present range, tell us your requirement because we can probably make it for you.
## REPEAT PERFDRMANCE FROM GARDNERS

## Exceptionally wide band microphone and audio line matching transformers



FREQUENCY RANGE
$100 \mathrm{~K} . \mathrm{ohm}$ models $\pm 1 \mathrm{~dB} 30 \mathrm{c} / \mathrm{s}$ to $20 \mathrm{kc} / \mathrm{s}$ All other models $\pm 0.5 \mathrm{~dB} 30 \mathrm{c} / \mathrm{s}$ to $20 \mathrm{kc} / \mathrm{s}$ MAXIMUM AUDIO LEVEL $+12 \mathrm{dBm}(16 \mathrm{~mW})$
INPUT IMPEDANCE maintained to within $\pm 10 \%( \pm 20 \% \mathrm{j})$ at all frequencies within the range $50 \mathrm{c} / \mathrm{s}$ to $8 \mathrm{kc} / \mathrm{s}$ (to $5 \mathrm{kc} / \mathrm{s}$ only for 100 K .0 hm models)
MAGNETICALLY SCREENED
-50 dB reduction in hum pick up.

For professional recording and broadcast transmission equipment, these Octal-based plug-in transformers have a frequency response extending well beyond the audio range. The design achieves dynamic performance with minimum distortionat all levels

Type MU. 7525 may be used in "Hybrid" circuits, as shown, to establish 2 to 4 wire operation in telephony. Accurate balancing of the windings enable guaranteed rejection of better than - 55 dB from $50 \mathrm{c} / \mathrm{s}$ to $10 \mathrm{kc} / \mathrm{s}$. Up to - 75 dB may be expected for normal rejection levels.

WRITE FOR A.F. BROCHURE GT. 4
giving full details of these and other types of A.F. transformers



EEV glass and ceramic hydrogen thyratrons are extensively used to provide more precise and efficient high speed switching. Here are some of the reasons why:
1 Their short anode delay time of between 20 and 120 nanoseconds depending on triggering method.
2 Low jitter generally of 1 to 2 nanoseconds but down to less than $\frac{1}{2}$ nanosecond depending on heater supply.
3 The negligible change in anode delay timetypically only 10 nanoseconds over a long period of use.
4 A high peak inverse voltage capability of 20 kV immediately following pulse.
5 The low trigger power required.
6 The wide operating voltage range of $1 \mathrm{kV}-120 \mathrm{kV}$ with four tubes.
7 The ability to control anode delay time and rise time of current, using reservoir.
8 The wide reservoir range for maintenance of gas pressure typically 4.5 V to 5.7 V .
The standard range plus EEV's ability to meet special requirements means that virtually any high speed switching application can be met. Here are a few:
Radar modulators with a system output power of $10 \mathrm{~kW}-10 \mathrm{MW}$.
Medical linear accelerators with RF accelerating powers up to 15 MW .
Particle linear accelerators with RF accelerating powers up to 50 MW . They may also be used in first-stage particle beam choppers Particle beam benders where a network of stored energy needs to be discharged into a deflection coil or other device somewhere on the accelerating ring.
Spark chambers
For pulsing light shutters such as Kerr or Pockel cells.
Electronic crowbars and energy diverters

## EEV thyratronsfor better high speed switching

|  | Peak <br> power <br> output <br> max <br> $(M W)$ | Heating <br> Factor <br> (V.A.p.p.s.) | Peak <br> forward <br> voltage <br> max <br> $(\mathrm{kV})$ | Peak <br> anode <br> current <br> max | Mean <br> anode <br> current <br> max |
| :--- | :---: | :--- | :--- | :--- | :--- |
| CX1154 | 50.0 | $30 \times 10^{9}$ | 40 | $(\mathrm{~A})$ | $(\mathrm{A})$ |
| CX1157 | 3.5 | $7 \times 10^{9}$ | 20 | 2500 | 3.0 |
| CX1168 | 100.0 | $70 \times 10^{9}$ | 80 | 350 | 0.35 |
| CX1171 | 150 | $70 \times 10^{9}$ | 120 | 2500 | 2.5 |
| CX1174 | 120 | $60 \times 10^{9}$ | 40 | 2500 | 2.5 |
| CX1175 | 200 | $140 \times 10^{9}$ | 80 | 6000 | 6.0 |
| CX1180 | 12.5 | $9 \times 10^{9}$ | 25 | 5000 | 6.0 |

Send for full details of the complete range of EEV thyratrons.


English Electric Valve Co Ltd
Chelmsford Essex England Telephone : 61777 Telex: 99103 Grams: Enelectico Chelmsford


I am particularly interested in using a thyratron with the following parameters

## Application

Peak power output
Peak forward voltage

Please send me full data on your complete range of glass and ceramic hydrogen thyratrons
$\qquad$
NAME

COMPANY


## Comprehensive range for civil and military authorities as well as domestic users in more than 50 countries.

Teonex now supplies a full range of British made valves and semi-conductors (or their Continental equivalents) to authorities operating stringent quality control, and to private individuals right across the world. Current price list and further particulars available on request from:

## TEONEK LIMITED

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The secret is in the fixing qf the Brewster window
-the angled glass plate at each end of the tube
In many tubes the seal is made with an epoxy resin which eventually cracks and ruins efficiency by letting in air. EEV, on the other hand, use fu'sion sealed windows where the seal is as strong as any other part of the tube. Fusion sealing allows the tube to be heated to a very high temperature during manufacture, driving out all the gases in the tube surface which would otherwise contaminate the helium-neon filling. EEV tubes have been life tested up to 6000 hours which is two or three times the life generally expected from tubes employing epoxy sealing techniques. There is a standard range of EEV laser tubes available, full details of which can be obtained by filling in the coupon. If your laser design calls for a special tube give us brief details of what you need as we can probably meet your requirements.

## Why EEV gas laser tubes



## last longer



Send for full details of the complete range of EEV gas laser tubes.


English Electric Valve Co Ltd
Chelmsford Essex England Telephone : 61777 Telex : 99103 Grams: Enelectico Chelmsford


Please send me full data on your range of gas laser tubes.
I am particularly interested in using a tube with the following parameters.
Wavelength ( nm ) Power Output (mW) POSITION
NAME
COMPANY
ADDRESS

ADDRESS


## Now Plessey cartridge recorders



## offer 48 Volt DC operation

The CT85 is the latest model to be added to the wide range of Plessey endless loop cartridge recorders. This unit provides emergency interception services in' Telephone Exchanges and other special services where 48 Volt D.C. operation is required.
The CT85 is suitable for continuous or intermittent service. Audio output is 2 watts into 50 ohms for multiple telephone line distribution. A second track is used for stop cues together with an auxiliary cue for other functions such as redirecting telephone traffic from non-operative numbers. Start can be local or by an external earth signal.
In common with all CT80 cartridge recorders the CT85 operates from an exclusive integral direct drive capstan motor, solenoid and puck wheel assembly. For
long term reliability the motor is an AC type driven through an electronic switching module. Solid state silicon devices are utilized throughout.
The CT85 is supplied in a portable cabinet that can be locked to prevent access to the cartridge, circuit boards and operating controls by unauthorised personnel. The unit can also be supplied for desk top, built-in and rack mounting.
For full details of Plessey broadcast standard CT80 recorders contact your local Plessey office now.

## PLESSEY <br> Electronics

## Choose your duplexer devices from EEV's extensive range


BS390

BS800

BS452
k

| Product | Type No | Band | $\begin{aligned} & \text { range } \\ & (\mathrm{MHz}) \end{aligned}$ | $\begin{aligned} & \text { power } \\ & \text { (kW) } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: |
| Pre TR cells | BS834 | - | 2000-12000 | 2500 |
|  | BS870 | - | 1240-1365 | 2500 |
| TR cells | BS390 | S | 2925-3075 | 1250 |
|  | BS800 | S | 2840-3100 | 1250 |
|  | BS824* | S | 2700-3100 | 250 |
| - | BS156 | X | 9000-9600 | 200 |
|  | BS452 | $x$ | 9310-9510 | 100 |
|  | BS810 | X | 9250-9550 | 75 |
|  | BS850 | X | 9300-9500 | 50 |
| TB cells | BS310 | X | 9375 | 5-200 |
| TR limiter cells | BS814 | X | 9000-9700 | 200 |
|  | BS828 | X | 9325-9425 | 50 |
| Solid state microwave switches | BS392 | S | 2925-3075 | 0.5 |
|  | BS460 | X | 8500-12000 | 0.5 |

Send for this booklet giving full details of the complete range of EEV duplexer devices and waveguide switches.


## English Electric Valve Co Ltd

Chelmsford Essex England Telephone : 61777 Telex: 99103 Grams : Enelectico Chelmsford
*For protection of travelling waveguide amplifiers


Please send me a copy of "Duplexer Devices". I am interested in a tube with the following parameters:
Frequency range
Power
Type of cell
$\qquad$
COMPANY
ADDRESS


Cameras a plenty... but how quickly can you find the right low cost tube?

There is a growing range of closed-circuit equipment available, ranging from the simple black and white camera to sophisticated full-colour facilities. The time inevitably arrives when a replacement vidicon tube is needed quickly. This is the service EMI sets out to provide. Our vidicon range provides a type for virtually every camera, where reliability. good resolution and high sensitivity are required Send for the EMI Vidicon replacement chart. Then, when
 contact your distributor or EMI.

Systems are helping industry's key personnel to stay put while controlling distant out-stations.

The equipment covers transmission of information between out-stations and control rooms by various forms of telemetry, including: Teledata; Teleshift; Telecode and Teleducer, and provides visual presentation of conditions of controlled equipment on mimic diagrams, using the well-known GECAEI Modular Mimic Systems.
 Write today for telemetry literature/modular mimic systems literature.

$$
\begin{aligned}
& \text { Communications } \\
& \text { Division } \\
& S \in C
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## BREAK THE SOUND BARRIER



MODEL 488 SONO-BAR


## WithProvid sulpe setine the woalios standard in sturnd

## NOISE CANCELLING MICROPHONES

When the chips are down, and noise levels are high, Shure Noise Cancelling microphones with their exclusive Controlled Magnetic cartridges, distancediscrimination design, and specially tailored response get the message through ... even when noise level is so high the operator cannot hear himself! They have been field-tested and proved in such ear-shattering environments as: drop forges, helicopters, police power boats, "hard surface" gyms among cheering crowds, motorcycles, jets revving up, fire engines, etc.

SHURE MODEL 488 SONO-BAR
Rugged, impact resistant "Armo-Dur" case. Four types: High or low impedance transistorized for direct replacement of carbon microphone; and FAA Certified Transistorized Aircraft version.

## SHURE MODEL 419 RANGER II

New small size. Only about half the size and weight of conventional mobile communications microphones. Unsurpassed for use with portable or miniaturized equipment.

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Erie Blue Chips are resistors. Thick film, metal glaze resistors built to satisfy the circuit designer's need for reduced physical size and increased reliablity. Designed specifically for printed circuit boards, Blue Chips, with radial terminations at 0.2 in centres, make possible a packaging density up to double that of conventional cylindrical resistors.
Resistance range:
1 ohm to 1 megohm with tolerances of $1 \%, 2 \%, 5 \%$ and $10 \%$.

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$\frac{1}{8} w, \frac{1}{4} w, \frac{1}{2} w$, with maximum
voltage 350 V d.c.

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the only UK custom-built thickfilm hybrid
microcircuit manufacturer also producing the active devices for hybrids in house.

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12 actual weeks from first enquiry to delivery of custombuilt batches for full production (only 6 weeks to prototype production).

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typical annual requirement is 500-5,000 off:
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Whether your products are individually assembled or on a flow line, missing components spell loss of time, delayed deliveries - and maybe tied-up capital. When it comes to meters, there's no excuse. Anders carry the largest stocks of meters in the U.K. Standard meters are off-the-shelf and on their way to you within 24 hours of your order. Nonstandard instruments take very little longer. Anders have a fast moving production team of well-equipped specialists in assembly, calibration, and even hand-lettering of dials. In fact the only things missing from the Anders' service are excuses: we take care to see that we don't have to make them. So when it comes to meters, come to Anders.
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## LEVELL measure $\mu$ V's from 1 Hz to 450 MHz VOLTMETERS

TYPE TM3B £63
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TYPE TM3A
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Complete with battery and input lead.
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EXTRAS
Leather case
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A.C. Power Unit


Complete with battery and in. put lead.
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VOLTMETER RANGES
$15 \mu \mathrm{~V}, 50 \mu \mathrm{~V}, 150 \mu \mathrm{~V}$.... 500 V f.s.d. Accuracy $\pm \% \pm \%$ ois.d. $\pm$ M VatikHz

## dB RANGES

100 dB to +50 dB in 10 dB steps. Scale

FREQUENCY RESPONSE
Above $500 \mu \mathrm{~V}$ : $\pm 3 \mathrm{~dB}$ from $1 \mathrm{~Hz}_{2}$ to 3 MHz . $\pm 0.3 \mathrm{~dB}$ from 4 Hz to 1 MHz On 50031 V: $\pm 3 \mathrm{~dB}$ from 2 Hz to 2 MHz on $50 \mu \mathrm{~V}$ : $\pm 3 \mathrm{dr}$ from 7 Hz 2 10 Hz On 15 $\mu \mathrm{V}$ : $\pm 3 \mathrm{~dB}$ from 20 Hz to 200 kHz .

AMPLIFIER OUTPUT
150 mV at f.s.d. on all ranzes. Will drive a load of $200 \mathrm{k} \Omega$ and $50 \mathrm{p} F$ withoue loss.

## POWER SUPPLY

One type PP9 battery, life 1000 hours; or
A.C. mains when Power Unit is fitted
$\star \star \star \star \star \star \star \star \star \star$

BROADBAND VOLTMETERS
As A.C. Microvaltmeters plus H.F. probe to extend response to 450 MHz . Two versions differ only in meter size and L.F. bandwidth switch on type TM6B.

TYPE TM6A
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battery and in. put lead. OPTIONAL EXTRAS

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OPTIONAL
EXTRAS
Leather Case A.C. Power Unil 67/10/-.
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ImV, $3 \mathrm{mV}, 10 \mathrm{mV}$. . . . 3 V f.s.d. Square law scaies. Accuracy ${ }^{ \pm}$. $\%$ of reading
$\pm 1 \%$ of $\mathrm{I}, \mathrm{s}$. at 30 MHz . of.s.d. at jomHz.
H.F. dB RANGES
$-50 \mathrm{~dB},-40 \mathrm{~dB},-30 \mathrm{~dB}$
Scate -10 dB to +3 dB . OdB $=1 \mathrm{~mW}$ into 508.

## H.F. RESPONSE

$\pm 0.7 \mathrm{~dB}$ from 1 MHz to 50 MHz
$\pm 3 \mathrm{~dB}$ from 300 kHz to 400 MHz $\pm 6 \mathrm{~dB}$ from 400 MHz to 450 MHz
L.F. RANGES

As TM3A and TM3B except for the omls sion of $15 \mu \mathrm{~V}$ and $150 \mu \mathrm{~V}$.

## POWER SUPPLY

One type PP9 battery. life 1000 hours on L.F. ranges and 400 hours on H.F. ranges; is fisced.

# Sounds exactly what <br> you want 

Here's a professional tape recorder that you can use in the studio and in outside broadcast vans. Philips Pro' 12 meets a long standing requirement of studio sound engineers. This portable two-channel recorder is designed to meet the high standards of sound quality and versatility expected of professional equipment, yet it is small and competitively priced. Recording and playback quality of the Pro' 12 is of a very high standard. Tapes prepared on a Pro' 12 are sui, table for immediate broadcasting. Even at the lowest tape speed of $33 / 4$ $\mathrm{in} / \mathrm{s}$, the sound quality is at least equal to the DIN 45511 studio equipment specification.
It features: - Twin-track stereo, twintrack mono and dual-track mono operation on $6.25 \mathrm{~mm}(1 / 4 \mathrm{in})$ wide tape (standard version). . Extra quartertrack stereo (special version). - Tape speeds of 9.5 and $19 \mathrm{~cm} / \mathrm{s}(33 / 4$ and $71 / 2 \mathrm{in} / \mathrm{s}$ ). Unique "constant load" tape transport. - Microphone, diode and line inputs for each channel.

- Facilities for mixing input signals of both channels. - Multiplay, sound on sound and echo effect. - Fade in and out and dubbing facilities. - Cueing and pause keys. - Line and monitoring outputs for each channel. - Monitoring with stereo headset or built-in loudspeaker, before or after tape. - VU-control of either channel. - End-of-tape switch. . Remote control connection. - Horizontal or vertical operation.


## Technical data

Tape speeds
$33 / 4$ and $71 / 2 \mathrm{in} / \mathrm{s}$ ( 9.5 and $19 \mathrm{~cm} / \mathrm{s}$ )
Tape
longplay ( $1800 \mathrm{ft}-540 \mathrm{~m}$ ) or
doubleplay ( $2400 \mathrm{ft}-720 \mathrm{~m}$ )
Reels
Ciné type, max. 7 in ( 180 mm ) Playing time
for longplay tape on 7 -inch reel: at $7 \frac{1}{2} \mathrm{in} / \mathrm{s}: 45 \mathrm{~min}$
for doubleplay tape on 7 -inch reel:
at $71 / 2 \mathrm{in} / \mathrm{s}: 60 \mathrm{~min}$
Deviation on absolute tape speed
less than $0.8 \%$
Wow and flutter
measured acc. to DIN 45507 with EMT 420, at $71 / 2 \mathrm{in} / \mathrm{s}: 0.08 \%$
at $3 \mathrm{3} / 4 \mathrm{in} / \mathrm{s}: 0.1 \%$
Frequency response
acc. to DIN 4551 , playback
at $71 / 2 \mathrm{in} / \mathrm{s}: 60 \ldots 12000 \mathrm{~Hz}, 0-1.5 \mathrm{~dB}$

at $71 / 2 \mathrm{in} / \mathrm{s}: 40 \ldots 18000 \mathrm{~Hz}, 0-2.5 \mathrm{~dB}$ at $33 / 4 \mathrm{in} / \mathrm{s}: 60 \ldots 10000 \mathrm{~Hz}, 0-1.5 \mathrm{~dB}$ at $3 \mathrm{3} / 4 \mathrm{in} / \mathrm{s}: 40 \ldots 15000 \mathrm{~Hz}, 0-2.5 \mathrm{~dB}$ overall at $7 \mathrm{1} / 2 \mathrm{in} / \mathrm{s}$ :
$60 \ldots 12000 \mathrm{~Hz}, 0-3 \mathrm{~dB}$
overall at $71 / 2 \mathrm{in} / \mathrm{s}$ :
$40 \ldots 18000 \mathrm{~Hz}, 0-5 \mathrm{~dB}$.
overall at $3 \mathrm{3} / 4 \mathrm{in} / \mathrm{s}$ :
$60 \ldots 10000 \mathrm{~Hz}, 0-3 \mathrm{~dB}$
overall at $3 \mathrm{3} / 4 \mathrm{in} / \mathrm{s}$ :
$40 \ldots 15000 \mathrm{~Hz}, 0-5 \mathrm{~dB}$
Signal-to-noise ratio
acc. to DIN 45405, weighted,
at $7 \frac{1}{2} \mathrm{in} / \mathrm{s}:-56 \mathrm{~dB}$
at $33 / 4 \mathrm{in} / \mathrm{s}:-52 \mathrm{~dB}$

## Inputs

a. line: $100 \mathrm{mV}, 100 \mathrm{k} \Omega$
b. microphone: $\leq 1 \mathrm{mV}$ (unbalanced), suitable for microphones from 50 to $2000 \Omega$
c. diode: $2-40 \mathrm{mV}, 20 \mathrm{k} \Omega$

Other inputs are av-ilable of conally

## Outputs

a. line:
nom. 0.775 V , max. $4 \mathrm{~V}, 10000 \Omega$
b. monitor (stereo):
nom. $0.775 \mathrm{~V}, \max .4 \mathrm{~V}, 10000 \Omega$
c. diode: $0.5-2 \mathrm{~V}, 100 \mathrm{k} \Omega$

Other outputs available optionally.

## Power supply

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Power consumption: 80 W
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# Wireless World 

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The front cover design introduces a short series of articles on operational amplifiers, beginning in this issue. Combined with the familiar triangular graphical symbol for an operational amplifier is a magnified photograph of the semiconductor chip of an integrated-circuit type of op.amp.-actually the ZLD 709 made by Ferranti. This d.c. linear amplifier has differential inputs and a Class B output stage.

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[^2]

# Why we decided to make eyery part in this PAL delay line 

The PAL delay line is a precision item. But it also has to be inexpensive, and therefore mass-produced. The problems involved in getting the delay time of $63.94 \mu \mathrm{~s}$-an adjustment to a few thousandths of a microsecond-for just one, are quite formidable. To achieve it on an assembly line is practically impossible, unless you have everything under your own control.
When the PAL system was being developed, we found ourselves in an excellent position to develop the special glass delay line needed for the chrominance decoder. Delay lines weren't new to us. For the previous five years we'd been producing them for the computer industry. We therefore had considerable experience. Experience which few others in the television industry had and which enabled us to develop our delay line in parallel with the development of colour television itself.

Critical factors. The set designer's demands pose problems in design and in production (remember we're concerned with price too!). Our considerable experience gained in the computing industry made the design problems
relatively easy to overcome. But marrying them to mass-production was something quite new. Again we were fortunate in having vast experience in mass producing complex items for other areas of the electronics industry.
Any old glass? The Mullard delay line is made of glass and works on an electromechanical principle.

The glass is specially compounded to ensure consistent behaviour propagation velocities and good stability with changes in temperature. The blocks are cast to ensure complete uniformity and an absence of any internal stressing. One end is ground with two optically flat faces which are at a slight angle to each other and to which two transducers are connected. The electrical television colour signal enters one transducer and is converted into vibrations. These vibrations travel through the glass until they are reflected back from the end face to the second transducer. This converts them back into an electrical signal. In this way we halve the size of the delay line and help save space within the set.

Ground away. The end of the glass block opposite the transducers is then ground away under automatic control until the response is exactly right. We have found that this constructionapart from saving space-greatly simplifies the problem of delay time. adjustment to $63.943 \mu$ s at 4.433619 MHz .

Insertion loss. While the glass has some effect on the insertion loss, the major loss is in the transducer and the coupling to the glass. The transducers themselves have been developed from
ceramics selected for their long term stability as well as good mechanical properties. We have further reduced insertion loss by developing a new metal deposition technique and adhesives which create an intimate bond. As a result the overall insertion loss is only about 13 dB over the bandwidth 3.43 to $5 \cdot 23 \mathrm{MHz}$.

The final step is the assembly of the delay line on its mounting plate with the associated input and output coils before final testing and inspection.

Worth it? Right from the beginning we've had everything under our control. So we can be sure that the product will give consistent service. And that we're producing it at the best possible price.

Consistently achieving these two aims with all our products has helped us build our reputation. A reputation which stretches across the electronics industry. Before we embark on any new project we can draw on the insight and experience we have gained-sometimes from unusual areas. We can employ our resources to provide the technically excellent products our customers demand.

## Mullard components for consumer electronics

[^3]
# Wireless World 

## $R \& D$ and $£ \mathrm{~s} \mathbf{d}$

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Closer collaboration between the three "components" trade associations (R.E.C.M.F., V.A.S.C.A. and B.V.A.) is foreseen by the announcement elsewhere in this issue that they are now together under one roof (even if not the same ceiling!). With the Electronics Components Board overlord, Sir Alan Dudley, in the role of matchmaker it augers well for an eventual marriage. With the increasing use of integrated circuits and therefore a gradual decline in the number of discreet components used in equipment it would seem a logical step. Otherwise we can foresee a situation arising calling for a judicial pronouncement on "what is a component?"

It is not without significance that Dr. F. E. Jones is the current chairman of three of the four bodies mentioned, i.e. E.C.B., R.E.C.M.F. and V.A.S.C.A. He certainly has some strong views on many of the problems which beset our industry (and indeed the whole country) and, since the publication of the report on the manpower committee, of which he was chairman, his knowledge and advice is much sought after.

He recently had some trenchant things to say about the reasons for the "technological gap" which exists between this country (and indeed the whole of Europe) and the U.S.A. This gap is not, in fact, due to our inability to create new and worthwhile technical advances in the various fields of technology, but because of our apparent inability to make full use of them industrially. This, says F.E.J., is why the average output of, or wealth created by, each employee in this country in 1967 was only $£ 1,300$ whereas in the United States it was four times this figure.

An instance of our tardiness to make capital out of new ideas is mentioned by one of our contributors in this issue (p. 85) who comments on the fact that although Piccolo-the refined teleprinter transmitting system-was developed by the Diplomatic Wireless Service ten years ago it is only just being exploited commercially.

To get back to the title we have given to this month's leader. One of the major factors affecting our industry today, and not ours only, is the vast expenditure on R \& D which in many instances appears to bear little relationship to the eventual profitability of the end product. It is interesting to note that the cost of R \& D in both the U.K. and the U.S.A. over the past ten years or so has increased threefold, but, the overall figure in America is in fact some six times the U.K. total of $£ 1,000 \mathrm{M}$. However, the wealth created per employee in America makes the return on this expenditure considerably higher than in this country. What then is the answer? Dr. Jones has stressed that the only way of obtaining effective correlation between the cost of $R \& D$ and the creation of wealth, is to ensure that research is "conducted under pressure from the market". This problem has already been explored by the Central Advisory Council for Science and Technology of which Dr. Jones is a member. In the Council's report "Technological Innovation in Britain" it recommends "the direct linkage of $R \& D$, production and marketing into a single interacting operation, and, planned programmes of innovation related to market opportunities".

No longer is it practicable, as in the past, to have one or two backroom boys "on the strength" in the hope that one day they will come up with a bright idea. R \& D must be geared to production. Over the past few years the curve for R \& D has risen sharply in this country but that for the gross national product is fairly flat.

# Device characteristics and what they mean: Methods of testing commercial units 

by G. B. Clayton,* B.Sc., A.Inst.P.

An operational amplifier is basically a very high gain d.c. coupled amplifier which makes use of feedback to control its response characteristics. The term 'operational amplifier' was in fact originally introduced by workers in the analogue computer field to denote an amplifier circuit which performed various mathematical operations such as integration, differentiation, summation and subtraction. Operational amplifiers are still widely used for analogue computation but their range of applications has now been vastly extended to include a great number of other fields, for example in the many branches of instrumentation and control. Non-linear applications of operational amplifiers are also now quite common; voltage comparators, non-linear function generators, and ultra linear rectifiers are but a few examples of their use in this last category.
Early operational amplifiers used thermionic valve circuitry, but these have now largely been replaced by solid state circuits. A wide range of solid state amplifier modules is commercially available, including f.e.t., chopper stabilized, and parametric amplifiers in discrete-component, thinfilm hybrid integrated circuit, thick-film hybrid integrated circuit and monolithic integrated circuit forms.
Discrete-component amplifiers are assembled on printed circuit boards and are often supplied in epoxy encapsulated modules which may be either soldered directly to a printed circuit board or plugged into a suitable mating socket. They can be economically produced in small quantities. In general thin-film hybrid integrated circuits are used for highly complex, close tolerance analogue applications requiring extreme resistor stability. Thick-film hybrid integrated circuits are widely used for high production requirements where a large number of circuit types is needed but where the quantity of circuits produced per type may be small.

Semiconductor integrated circuits find most applications where highly repetitive requirements exist and where relatively few circuit types are involved.

The manufacture of electronic systems in integrated form dates from about 1959 and attention was initially focused on digital

[^4]systems; it is only comparatively recently that a variety of linear integrated circuits have become available at moderate prices. An important characteristic of the i.c. manufacturing process is that the tooling cost for each individual circuit is high and circuits can only be produced economically if large numbers of the same circuit are made. The electronic designer accustomed to selecting individual components that exactly meet his requirements must now adopt rather a different approach if he is to make use of i.cs. He must accept the available massproduced circuits and modify them as necessary to fulfil his requirements by connecting external discrete components to them.

One may well ask the advantages of making this change over to i.cs and the reasons that may be put forward are those of reliability, cost, performance, size and weight-all quite formidable reasons. The improved reliability results from factors such as fewer connections of dissimilar materials, therefore less connection failure, and less handling of individual component parts and therefore more uniformity of product. The cost advantage is not always immediately apparent. The actual cost of an i.c. may be more than the cost of the discrete components that would be required to make a similar circuit; however, if one takes into account the reduction in design and construction time the i.c. usually turns out to be the more economical. The superior performance characteristics of i.cs result from the close matching of components and the very short interconnections between components that are possible in these circuits. Size and weight reductions that are obtained are obvious although not always necessary.

The actual silicon chip on which an integrated-circuit operational amplifier is formed (see front cover) is very little bigger than a single transistor chip, one millimetre square being fairly typical. There are several different methods of packaging the chip. Mounting in a transistor-type metal can is one method (Fig. I(a)), a nother is the ten-lead flat pack (Fig. $\mathbf{I}(\mathrm{b})$ ), and more recently the dual-in-line plastics package has been introduced (Fig. I(c)). Many manufacturers produce economy versions of their amplifiers in the last style of package. The dual-in-line plastics pack has fourteen pin-like connections arranged to mate up
conveniently with a printed circuit board although once these pins are soldered into position it is somewhat difficult to remove the amplifier. To overcome this difficulty it is often convenient to use one of the dual-in-line sockets that are available, and these eliminate the need for soldering the i.c. leads. The sockets themselves may be printed-circuit or chassis mounted; they are made with solder or wire wrap terminations.

It is not essential that the user of i.c. op. amps be familiar with the intricacies of the internal circuit details of the amplifier (he can't get at them!), but he must understand the function of the external connections provided by the manufacturer and he must understand the terms used to specify the amplifier's performance if he is to be able to select the best device for a particular application. The desired response of an op. amp. is normally obtained by connecting feedback components externally to the input terminals of the amplifier. An amplifier with a differential input allows a greater flexibility in the choice of feedback configuration and most i.c. op. amps are made with differential input and single-ended output. In fact most direct coupled amplifiers invariably use the differential configuration because of the superior drift performance that can be obtained by the matching of characteristics. Op. amps are generally designed to operate from symmetrical positive and negative power supplies to permit an output voltage which may be positive or negative with respect to earth and a frequency response extending down to zero. The functions mentioned require five external connections to the amplifier. Most amplifiers are provided with several other external connections, the purpose of which will be discussed later. It is possible to operate the amplifiers with a single power supply if the particular application does not require a d.c. response ${ }^{1}$.

## I.C. op. amp. characteristics

The graphical symbol commonly used for an op a amp. is a triangle indicating the direction of signal flow (Fig. 2(a)). Ignoring certain important errors which will be discussed later the output of the amplifier is related to its inputs by the transfer curve shown in

[^5]Fig. 2(b), and with open loop gains greater than $10^{3}$ being quite typical only a very small voltage between the two input terminals is needed to cause saturation of the amplifier output.

The open loop voltage gain, $A_{V o L}$, is defined as the ratio of the change in output voltage $\Delta e_{o}$ to the change in voltage $\Delta e_{\epsilon}$ between the two input terminals; it is nor.nally specified for d.c. and may be determined from the slope of the nonsaturated portion of the transfer curve.
Op. amps are in fact seldom used open loop, but are used with negative feedback to improve accuracy. The significance of the open loop gain is that it determines the accuracy limits in such applications. The two basic feedback configurations are the inverting (Fig. 3) and non inverting (Fig. 4) circuits.

The closed-loop voltage gain, $A_{V C L}$, of the inverting amplifier is defined as the ratio of the change in output voltage to the change in the input voltage applied to the input resistor $R_{1}$. If the amplifier is considered to be ideal, i.e. infinite input impedance, zero output impedance, infinite open-loop gain and bandwidth; feedback maintains the error voltage $e_{\epsilon}$ between the two input terminals zero at all times and a simple analysis shows that the closed-loop gain $A_{V C L}=-\frac{R_{2}}{R_{1}}$

If the amplifier is ideal except for finite open-loop gain, the closed-loop gain,
$A_{V C L}=-\frac{R_{2}}{R_{1}}\left\{\frac{1}{1+\frac{1}{\beta A_{\text {VOL }}}}\right\}$
(see Appendix 1)
where

$$
\beta=\frac{R_{1}}{R_{1}+R_{2}}
$$

The quantity $\beta A_{V O L}$ is called the loop gain, and it is a most important factor in determining closed-loop performance. The error in closed-loop gain due to finite open-loop gain may be expressed by the error factor

$$
\frac{1}{1+\frac{1}{\beta A_{V O L}}}
$$

which is approximately

$$
1-\frac{1}{\beta A V O L}
$$

for $\beta$ Avol much greater than 1. The percentage error due to finite open-loop gain thus $\frac{100}{\beta A_{\text {VOL }}} \%$, which is a direct function of loop gain. The error in closed-loop gain is not in itself very significant since the ratio $R_{2} / R_{1}$ can always be adjusted to compensate for this error, but the closed-loop gain stability, the closed-loop output impedance and closed-loop distortion are all directly related to loop gain.

Closed-loop gain stability:

$$
\frac{\Delta A_{V C L}}{A_{C L L}}=\frac{\Delta A_{V O L}}{A_{V O L}} \cdot \frac{1}{\beta A_{V O L}}
$$

Closed-loop output impedance :

$$
Z_{O C L}=\frac{Z_{0 O L}}{\beta A_{V O L}}
$$

Closed-loop distortion;

$$
D_{C L}=\frac{D_{O L}}{\beta A_{V O L}}
$$

The closed-loop gain of the non-inverting amplifier assuming ideal amplifier performance, is

$$
A_{\text {VCL }}=1+\frac{R_{2}}{R_{1}}
$$

If the amplifier is ideal except for finite open-loop gain

AVC $_{L_{0}}=\left\{1+\frac{R_{2}}{R_{1}}\right\}\left\{\frac{1}{1+\frac{1}{\beta A_{V O L}}}\right\}$.
(see Appendix 2)
where again

$$
\beta=\frac{R_{1}}{R_{1}+R_{2}}
$$

and as in the case of the inverting amplifier the loop gain $\beta$ Avol plays an important part in determining closed-loop characteristics.
In the case of both amplifier configurations if the closed-loop gain is greater than


Fig. I. Types of encapsulation for integrated-circuit operational amplifiers:
(a) "transistor outline", (b) "fat-pack", (c) "dual-in-line".

$$
\beta A V O L \approx \frac{A_{V O L}}{A_{V C L}}
$$

If as is usual gains are measured in dB we have the relationship: loop gain $(\mathrm{dB})=$ open-loop gain (dB)-closed-loop gain (dB).

The maximum output voltage swing, Vomax. This is the maximum output voltage swing (positive and negative) measured with respect to earth that can be achieved without clipping of the signal waveform.

An ideal differential amplifier with equal voltages applied to its input terminals would give zero output voltage, but under these circumstances real amplifiers are found to give a non-zero output voltage called an 'offset voltage'. In many amplifiers provision is made for zeroing the amplifier output voltage with an external trim potentiometer.

Input offset voltage, $V_{i o}$. This is the difference in the d.c. voltages which must be applied to the input terminals to obtain a zero quiescent output. It is indicative of the degree of matching in the differential amplifier stages of thȩ integrated circuit, and in general represents the main source of offset error when the amplifier is used with low source impedances. Integrated-circuit op. amps with the smallest input offset voltages in general exhibit the smallest output drift with temperature variations.

All i.c. op. amps require some small and relatively constant current at each input.
(a)

$$
1-e_{0} \text { positive }
$$



Fig. 2. (a) Graphical symbol for an operational amplifier; (b) ideal transfer curve for an op. amp.

Fig. 3. Inverting amplifier feedback
configuration.

Fig. 4. Non-inverting amplifier feedback configuration.
one, i.e. if $R_{2} / R_{1}>1$ then $\beta$ is approximately equal to $R_{1} / R_{2}$ and the loop gain


Input bias current, $I_{b}$, is defined as the average value (half the sum) of the currents at the two input terminals with the quiescent output voltage zero. It constitutes the bias currents drawn by the transistors in the differential input stage of the i.c. and if these were perfectly matched the two currents would be equal. It is normal practice to use balanced impedances at each input so that offset due to bias current is cancelled at the output. In practice some degree of mismatch always exists.

Input offset current, $I_{i 0}$, is defined as the difference in the input bias currents into the two input terminals. With equal source impedances connected to the two inputs it is only this mismatch or difference current which causes an offset error. The effects of $I_{i n}$ tend to overshadow the effects of input offset voltage when the input source impedances are high.

Temperature drift. The output voltage of all d.c. coupled solid-state amplifiers changes or 'drifts' from its initial value if the temperature changes. The temperature drift of i.c. op. amps is specified by the temperature coefficients of input offset voltage, input bias current and input offset current.

Supply voltage sensitivity. The output voltage of an i.c. op. amp changes if the supply voltages are changed. The effect is usually specified by the effect of supply voltages on input offset voltage, input bias current, and input offset current. With well regulated power supplies offset errors due to this effect are usually negligible compared to temperature drift.

It is very instructive for the new user of i.c. op. amps to gain an initial familiarity with the devices by wherever possible setting up test circuits to measure their characteristics. A simple test circuit for measuring the transfer curve of an op. amp. is shown in Fig. 5.

An oscilloscope with d.c. coupled $X$ and $Y$ channels is used to obtain a visual display of the transfer curve. The same signal is used to drive both the input of the amplifier and the horizontal sweep. Op. amps will normally be found to have gains in excess of 1000 and a resistive divider is placed at the input of the amplifier in order that the amplitudes of the $X$ and $Y$ inputs presented to the oscilloscope shall be of the same order. A low frequency sinusoid ( $f<20 \mathrm{~Hz}$ ) may be used as the drive signal. If a low frequency signal generator is not available or if the amplifier under test shows appreciable hysteresis effects it may be found convenient to use a low frequency ramp as the drive signal. A circuit that has been found quite suitable for producing a ramp drive is shown in Fig. 6. If this ramp drive is used it will be found that the retrace sweep rate is so much faster than the trace rate that the retrace is effectively blanked off from visual presentation.

In order to obtain the display the oscilloscope inputs are initially earthed and the spot centred in order to establish the vertical and horizontal references. The oscilloscope is then connected into circuit and the


Fig. 5. Test circuit for obtaining a transfer curve; significant parameters of the transfer curve.


Fig. 6. Circuit for generating a ramp function for test purposes.
amplitude of the input drive is turned up until the amplifier is in saturation. The maximum positive and negative output voltage swing $\left(V_{o \text { max }}^{--}\right)$is read directly from the trace; Arol is calculated from the slope of the transfer curve. The horizontal displacement from the horizontal zero reference to the point where the curve crosses the vertical zero reference gives the input offset voltage $V_{i o}$. Any departure from linearity shown by the device is also readily apparent from the display. If supply voltages are changed the dependence of these parameters on supply voltages can be measured. The oscillograms (Fig. 7) were obtained using the test circuit and show the transfer curve of R.C.A. type CA3029 amplifier for different values of supply voltage.

A circuit suitable for the measurement of input bias current $I_{b}$, and input offset current $I_{\text {in }}$ is shown in Fig. 8. The effects of input offset voltage are swamped by the use of the large input source resistors ( $100 \mathrm{k} \Omega$ ). The measurement procedure consists of adjusting the voltage $V_{b}$ to bring the output voltage of the amplifier to zero and then measuring the voltages $V_{i n_{1}}$ and $V_{b}$ with a high input resistance d.c. voltmeter. The bias current is calculated from $I_{b}=\frac{V_{i n_{1}}}{\operatorname{Lok} \Omega}$ amperes; the input offset current is calculated from $I_{i o}=\frac{V_{b}}{100 \mathrm{k} \Omega}$ amperes.

Maximum voltage between inputs. The voltage between the input terminals of an op. amp. is maintained at a very small value under most operating conditions by the feedback circuit in which the amplifier is used. If the application is such that the voltage between the input terminals might be appreciable care must be taken to ensure that it does not exceed the maximum allowable value for the particular amplifier, otherwise permanent damage may be caused. The connection of parallel back-to-back diodes across the input terminals is one way of protecting the circuit.

Maximum common mode voltage. The voltage at both inputs of a differential amplifier can be raised above earth potential. The input common mode voltage ( $e_{c m}$ ) is defined as the voltage above earth at each input when both inputs are at the same voltage. The maximum common mode voltage $E_{c m}$ is the maximum value of this voltage which can be applied without producing clipping or excessive non-linearity at the output.

If an amplifier is to be used under conditions in which excessive common mode voltage might cause permanent damage, protection can be provided by the use of a suitable pair of zener diodes. The circuit of Fig. 9 illustrates protection both against excessive voltage between inputs and excessive common mode voltage.

Common mode rejection. An ideal differential amplifier responds only to the difference in the voltages applied to its input terminals and produces no output for a common mode voltage. In practical amplifiers, because of slightly different gains between the inverting and non-inverting inputs, common mode input voltages are not entirely subtracted at the output. The gain of an amplifier for common mode voltages is known as the common mode response and the ratio of the gain with the signal applied differentially to the common mode response is called the common mode rejection ratio (c.m.r.r.). It is often expressed in dB by taking 20 times $\log$ (base 10) of the ratio. An alternative way of defining c.m.r.r. is as the ratio of input common mode voltage $e_{c m}$ to the output common mode error voltage referred to the input $e_{\text {cem }}$ (divided by the differential gain). A dittle consideration will show that the two definitions are of course identical.

Common mode rejection presents no problems in the case of amplifiers used in the inverting configuration (Fig. 3), for with one input earthed the input common mode voltage $e_{c m}$ must be zero. In the case of the non-inverting circuit (Fig. 4) feedback causes the voltage at the inverting input to follow that at the non-inverting input. The input common mode voltage thus varies directly with the input signal. This introduces a basic error which affects the overall circuit accuracy. For example, consider an amplifier with c.m.r.r. of 1000, used in the non-inverting configuration. With an input signal of say I V , the input common mode voltage $e_{c m}$ would also be


Fig. 7. Illustrating supply voltage sensitivity: transfer curves of CA3029 i.c. amplifier with different supply volrages. Left: positive supply $+4 V$, negative supply $-4 V$. Middle: $+4 V,-5 V$. Right: $+3 V,-4 V$. (Vertical scales: $I V / d i v i s i o n$. Horizontal scales: ImV/division).

I $V$ and the common mode error voltage (referred to the input) would be $e_{\text {ecm }}=$ $e_{c m} / \mathrm{c} . \mathrm{m} . \mathrm{r} . \mathrm{r}$. equals ImV and representing a $0.1 \%$ measuring error. If common mode error voltage efem varies linearly with common mode voltage this error is not very important, since it can be compensated for by adjustment of closed-loop gain (adjustment of $R_{2} / R_{1}$ ). Linearity of common mode error voltage with common mode voltage is thus in many applications more important than the value of c.m.r.r., and a graph of the type shown in Fig. Io is particularly useful in specifying the common mode behaviour of an amplifier.

A suitable test circuit for practically obtaining such a graph is shown in Fig. I I. An oscilloscope with d.c. coupled $X$ and $Y$ channels is used for the visual display; the


Fig. 8. Test circuit for determining input bias current and input offset current.


Fig. 9. Method for protection against excessive common mode voltage and excessive input voltage


Fig. 1o. Graph of input common mode input voltage vs. input common mode voltage.
oscilloscope may be single-ended provided, as shown in the diagram, the power supply is floated. A low frequency signal generator is used to provide both the input common mode voltage and the oscilloscope horizontal sweep. The amplifier output, which is equal to the closed loop gain multiplied by the equivalent input common mode error voltage

$$
\left[1+\frac{R_{2}}{R_{1}}\right] \cdot e_{\varepsilon c m}
$$

provides the oscilloscope vertical sweep. A variable d.c. bias is included in addition to the signal generator to provide the common mode input voltage $e_{c m}$ for amplifiers in which the positive and negative values of maximum common mode voltage differ appreciably. To allow a measurement of this parameter to be made directly from the trace the horizontal zero reference is established by earthing the oscilloscope horizontal input and aligning the trace with the central vertical graticule line. The vertical position of the trace is not of significance in the measurements. The oscillogram shows a result obtained with the test circuit; the polarity of the horizontal scale has the opposite sense to that of the graph in Fig. 10. For the particular amplifier tested, positive and negative values of maximum common mode voltage are seen to be +1 V and -2.5 V respectively. The trace is fairly linear between these limits; its average slope is used to give the c.m.r.r. from the relationship

$$
\text { c.m.r.r. }=\frac{X}{Y} \cdot\left(1+\frac{R_{2}}{R_{1}}\right)
$$

The effects of power supply voltage on c.m.r.r. and maximum common mode voltage can easily be measured by changing the power supply voltages and observing any changes in the trace.

Open loop bandwidth and frequency response. The importance of open-loop gain $A_{V O L}$, and loop gain $\beta A_{V O L}$ in determining the closed-loop performance of an op. amp. has already been discussed, but the assumption was made that the amplifier had an infinite bandwidth. Practical amplifiers have, of course, a finite bandwidth, and the effect of this on closed-loop performance must be taken into account. The open loop bandwidth is defined as the frequency at which the open-loop gain


Fig. 11. (a) Test circuit for determining common mode rejection ratio. (b) Oscillogram showing result obtained with test circuit. Vertical scale(0.IV/division) represents rooesem. Horizontal scale (IV/division) represents $\boldsymbol{e}_{\text {cin }}$.

is 3 dB down on its value at low frequencies. Many applications of op. amps require closed-loop gains over bandwidths of only a few hundred hertz, nevertheless the openloop gain characteristics at much higher frequencies are of great importance. Amplifier gain attenuation with frequency is always accompanied by phase shift and because of this phase shift a negative feedback circuit may in fact provide positive feedback at high frequencies resulting in peaking of closed-loop gain or in high frequency oscillations.

A typical operational amplifier open-loop frequency response rogether with a closed loop response for a gain of 40 dB is shown in Fig. 12. A response of this type can be represented mathematically by the equation:

$$
\begin{equation*}
A_{V O L(\omega)}=\frac{A_{V O L}}{1+j \frac{\omega}{\omega_{0}}} \tag{3}
\end{equation*}
$$

This equation describes what is known as a first order high frequency response; $\omega_{0}$
is called the break frequency. The function is conveniently approximated by its asymptotes (see Appendix 3) and this approximation has been made in Fig. 12. An amplifier having an open-loop response of this form with a 6 dB per octave ( 20 dB per decade) roll off will be stable (will not oscillate) for all values of resistive feedback, for the amplifier phase shift never exceeds 90 and the feedback is negative at all frequencies. An expression showing the effect of the frequency dependence of open-loop gain on closed-loop gain is obtained by inserting equation (3) into the equation for closed loop gain (eq. 1), thus
$A_{\operatorname{VCl}}^{\Delta(\omega)}=-\frac{R_{2}}{R_{1}}\left\{\frac{1}{1+\frac{1}{\beta \operatorname{AvoL}(\omega)}}\right\}$
$\operatorname{AvC}^{\prime} L(\omega)=-\frac{R_{2}}{R_{1}}\left(\frac{1}{1+\begin{array}{c}1+i\left(\omega / \omega_{0}\right) \\ \beta \text { AVOL }\end{array}}\right) \ldots$
The closed loop 3 dB frequency is obtained by equating real and imaginary parts of the denominator, giving

$$
\omega_{o C L}=\left(1+\beta A_{\text {VOL }}\right) \omega_{0} \ldots(
$$

The closed-loop bandwidth is greater than the open-loop bandwidth by the amount $\beta$. Avol. wo At frequencies higher than $\omega_{0} \subset L$, closed-loop and open-loop gains become equal. This may be seen from inspection of eq. (4) remembering that $\beta こ R_{1} / R_{2}$. These points are illustrated in Fig. 12 and also the fact that loop gain in dB is the difference between open-loop gain and closed-loop gain. Loop gain decreases with increase in frequency because of the attenuation of open-loop gain, and consequently closed-loop gain stability, linearity and other parameters that depend on loop gain are degraded at higher frequencies.

Not all op. amps are designed with a 6 dB per octave attenuation. Some are designed with a much faster roll-off, and these fast roll-off amplifiers allow an improved closed-loop performance at the higher frequencies, but without compensation they are not stable under all conditions of resistive feedoack. Consider an amplifier having an open-loop response of the form shown in Eig. 13, with a 6 dB per octave roll-off breaking at $\omega_{o_{1}}$, and a second break at $\omega_{o_{2}}$ followed by a 12 dB per octave roll-off. Mathematically this is represented by a combination of two first-order high frequency response functions with break frequencies $\omega_{o_{1}}$ and $\omega_{o_{2}}$; where the slope reaches 12 dB per octave the amplifier phase shift approaches $180^{\circ}$. With resistive feedback this amplifier would give stable closedloop operation for closed-loop gains in excess of 60 dB , and at frequencies above the closed-loop bandwidth open-loop and closed-loop gains would become equal as before. If feedback components were changed in order to obtain a closed-loop gain less than 60 dB instability would occur. The criterion for stable closed-loop operation is simply that the rate of closure between the open-loop and closed-loop response curves should be less than 12 dB per octave.


Fig. 12. Amplifier with 6 dBloctave of open-loop gain.


Fig. 13. Amplifier with attenuation of open-loop gain greater than 6dB/octave.

## Appendices

## I. Inverting amplifier

With ideal amplifier, i.e., infinite input impedance, infinite open loop gain, infinite bandwith, zero output impedance

$$
e_{ह}=-\frac{e_{o}}{A v O L}=0
$$

and

$$
i_{i}=i_{f}
$$

But $\quad i_{i}=\frac{e_{i}}{R_{1}} \quad$ and $\quad i_{f}=-\frac{e_{o}}{R_{2}}$,
which gives

$$
A_{V O L}=\frac{e_{0}}{e_{i}}=-\frac{R_{2}}{R_{1}}
$$

With ideal amplifier except for finite open loop gain

$$
\begin{aligned}
i_{i} & =\frac{e_{i}-e_{G}}{R_{1}}=\frac{e_{i}+\frac{e_{O}}{A_{V O L}}}{R_{1}} \\
i_{f} & =\frac{e_{\epsilon}-e_{O}}{R_{2}}=-\frac{\frac{e_{O}}{A_{V O L}}-e_{O}}{R_{2}} \\
& =-\frac{e_{O}\left\{\frac{1}{A_{\text {loL }}}+1\right\}}{R_{2}}
\end{aligned}
$$

As before $i_{i}=i_{f}$, and substitution and rearrangement gives,

$$
\begin{equation*}
A_{\text {VOL }}=\frac{e_{0}}{e_{i}}=-\frac{R_{2}}{R_{1}}\left\{\frac{1}{1+\frac{R_{1}+R_{2}}{A_{\text {VOL }} R_{1}}}\right\} \tag{A1}
\end{equation*}
$$

## 2. Non inverting amplifier

With ideal amplifier, $e_{\varepsilon}=0$ and

$$
e_{\ell}=e_{o} \frac{R_{1}}{R_{1}+R_{2}}
$$

which gives

$$
A_{V O L}=\frac{e_{0}}{e_{i}}=1+\frac{R_{2}}{R_{1}}
$$

With ideal amplifier except for finite open loop gain

$$
e_{i}-e_{e}=e_{o} \frac{R_{1}}{R_{1}+R_{2}}
$$

but

$$
e_{\epsilon}=\frac{e_{0}}{A V O L}
$$

Substitution and rearrangement gives

$$
\begin{gather*}
\text { AVCI }=\frac{e_{o}}{e_{i}} \\
=\left\{1+\frac{R_{2}}{R_{1}}\right\}\left\{\frac{1}{1+\frac{R_{1}+R_{2}}{A \text { vOI }_{1} R_{1}}}\right\} \tag{A2}
\end{gather*}
$$

## 3. First order systems

Consider the function

$$
A v(\omega)=A_{V}(O) \frac{1}{1+j \frac{\omega}{\omega_{c}}}
$$

At low frequencies the straight line given by $\left|A v^{\prime}(\omega)\right|=A v^{\prime}(0$; is an asymptote and at high frequencies the curve is asymptotic to

$$
\left|A_{V(\omega)}\right|=A_{V(0)} \frac{\omega_{c}}{\omega}
$$

The high frequency asymptote has a slope of 20 dB per decade, i.e. if the frequency is increased ten times $\left|A v_{(\omega)}\right|$ is reduced by 20 dB . (If the frequency is doubled $\left|A_{V(\omega)}\right|$ is reduced by 6 dB , i.e. 6 dB per octave.)
The asymptotes intersect at $\omega$ equals $\omega_{c}$ and here

$$
\left|A_{V^{\prime}(\omega c)}\right|=\frac{A_{V(0)}}{1+j 1}
$$

which gives

$$
\left|A_{V(\omega c)}\right|=A_{(v o)} \cdot \frac{1}{\sqrt{2}}
$$

$\left|A_{V(\omega)}\right|$ is 3 dB down on $A V(0)$ The angle (phase) as well as the magnitude of $A V(\omega)$ is of importance. It may be sketched as a function of $\omega$ by noting the following: I. As $\omega \rightarrow 0$, $A_{V^{\prime}(\omega)} \rightarrow A_{V^{\prime \prime}(O),}$ which is real, therefore the phase shift produced by the amplifier is zero at low frequencies 2. As $\omega \rightarrow \infty, A V^{\prime}(\omega)$ becomes imaginary corresponding to a $90^{\circ}$ phase lag.
3. At $\omega=\omega_{c}, A V(\omega c)$ has real and imaginary parts equal, and the phase shift is thus $45^{\circ}$.

The two plots of magnitude and angle of $A V(\omega)$ are referred to as the Bode plot. The magnitude of $A_{V^{r}(\omega)}$ is usually expressed in dB and is plotted against a $\log$ (base 10) scale of frequency.

# The Notion of "State" 

# A unifying concept in the diverse world of electronics 

by James Franklin

Some of the older readers of Wireless World tend to be worried by the fact that the contents of a typical 1960s issue do not seem to reflect a well-defined area of technology-say "radio" as it was in the '2os and 'zos. To them, and perhaps some younger people too, it is disturbing to find articles on computers and switching circuits cheek-by-jowl with articles on receivers and audio amplifiers. It is not enough to say "well, all these things are embraced by electronics" and explain that there is now one body of engineering theory which is equally relevant to television sets and industrial process-control systems. Within electronics the techniques, and the languages used by their practitioners, often seem worlds apart, and it may certainly be difficult for a particular reader to understand two adjacent articles in Wireless World.*

One of the fundamental divisions in electronics is, of course, between what we call continuous or analogue techniques and digital techniques, and this has arisen largely because analysis and design is done mainly in the frequency domain for the first and in the time domain for the second. In reality, of course, all events take place in the time domain, and the frequency approach, depending on the convention that all signals are made up of pure sine waves, is a specialized, blinkered view which avoids the direct handling of time relationships because it is convenient to do so. At the basis of the frequency approach, however, are differential equations which express the behaviour of circuits and systems with respect to time. It is these differential equations which enable us to predict what will happen after a specified interval of time in, say, an $L C R$ circuit with the same sort of certainty that we can predict what will happen after a specified interval of time in, say, a shift register of a digital computer (given the initial conditions and input signals in both cases, of course).

In Fig. I this common dimension of time is used to compare the action of an analogue circuit with that of a digital circuit. This diagram may be rather obvious and elementary, but it is intended to bring out the fact that both the behaviour of the analogue circuit and the behaviour of the digital circuit can be considered in terms of
their having a state which changes from instant to instant (indicated on the two graphs by the dots at $t_{1} t_{2} \ldots \ldots t_{n}$ ). As another example imagine the mechanical "system" of a ball being propelled through the air by the foot of a small boy (see Fig. 2). The state of the system can be defined as the position of the ball with respect to a threedimensional frame of reference, and the $x$, $y$ and $z$ distances giving this position are a set of variables-the state variables-which, of course, change from instant to instant during the flight of the ball.

In an electrical system the principal variables which define its state are, of course, the charges, voltages and currents at various points. Again, the values of these variables change from instant to instant during the operation of the system. In Fig. $\mathrm{I}(\mathrm{a})$ we have selected for examination one state variable which tells us practically everything we want to know about the behaviour of the circuit-the voltage across the capacitor, $V_{c}$. This variable has a sequence of values, each of which is different from the previous ones. In Fig. $1(b)$ a significant state variable is the voltage $V_{B}$ representing the binary condition at the output of the register. Here again the state variable has a sequence of values at $t_{0}$, $t_{1} \ldots \ldots t_{n}$ but at each instant it can be only one or the other of two possible values

In more complex dynamical systems the number of state variables necessary to describe behaviour will obviously be greater. For example, in an $L C R$ circuit one must know about the movement of charge through the inductor (e.g. current through
it) and the accumulation of charge in the capacitor (e.g. voltage across it). It is not necessary, however, to know anything about any variable associated with the resistance because $R$ is non-reactive and has no memory and does not itself produce any change with time in the flow of charge through it. What are significant as state variables in systems, then, are variables associated with elements producing operations in the time domain-stores or delays in digital systems, reactive or energy-storing elements in continuous systems.
This view of system behaviour as a state which changes from moment to moment is by no means new. For example, Newton talks about the state of rest or state of uniform motion of a body in the laws of motion he propounded in the 17 th century. A more conscious use of state, in analysing dynamical systems, was made in the 19th century by the French mathematician and philosopher Henri Poincaré who introduced the whole mathematical basis of what are now called state space techniques in his three-volume work "Les Méthodes Nouvelles de la Mécanique Céleste". Also in the 19th century the Russian mathematician Liapunov (well known to students of control theory) used the concept of state in analysing the conditions for stability in a dynamical system. During the past decade there have emerged various methods for applying these ideas to engineering systems. The remainder of this article is intended to provide no more than an introduction to the many learned works that have been written on the subject.


Fig. I. Illustrating the idea that any system-here an analogue one in (a) and a digital one in (b)can be considered as having a state, which changes with successive instants of time, $t_{0}, t_{1}$, $t_{2} \ldots$. etc.

Fig. 3 is a block diagram of an electrical or mechanical system. As already explained its state can be represented by a set of variables, which here we shall call $x_{1}, x_{2}$, $x_{3} \ldots \ldots x_{n}$. Similarly, its input will consist of another set of variables which we shall call $u_{1}, u_{2}, u_{3} \ldots \ldots u_{n}$ and its output yet another set of variables $y_{1}, y_{2}, y_{3} \ldots y_{n}$. (In many electronic systems, of course, there is only one input variable, $u_{1}$, and one output variable, $y_{1}$.) For convenience these sets of variables are represented by single symbols in heavy type, $\mathbf{x}, \mathbf{u}$ and $\mathbf{y}$, on the diagram.
Now if Fig. 3 is an analogue system we are interested in its state (and input and output) continuously, and in graphical representations we use a continuous time scale $t$ as in Fig. 4(a). But if Fig. 3 is a digital system, although time is still a continuum we are only interested in what happens at particular instants on the Fig. 4(a) scale, $t_{0}, t_{1}$, $t_{2} \ldots$. etc. We will call these discrete instants of time $t$, where $k$ ranges over the integers. In most digital systems the interval between the discrete instants $t_{k}$ is constant, and here we will call it $T$. Thus the actual time at which an event takes place in the digital system is given by a value $k T$ (with respect to $t_{0}$ ). But since $T$ is a constant it is $k$ which is the significant factor and we can then use a discrete-time scale as in Fig. 4(b). Of course, the scale in Fig. 4(b) could also be used for a continuous system if it is assumed that $T$ is infinitesimally small.

We can now use the symbology of Figs. 3 and + to write down generalized equations for the two types of system. In the case of the continuous or analogue system we can say first that the state $\mathbf{x}$ is a function of time, which is expressed formally as $\mathbf{x}(t)$. At any instant the state of a system depends on its initial condition when switched on or otherwise started-an instant we shall define as $t_{0}$-and on the input during the period between $t_{0}$ and the present instant, $t$. This can be written in algebraic form as

$$
\begin{equation*}
\mathbf{x}(t)=f\left[\mathbf{x}\left(t_{0}\right): \mathbf{u}\left(t_{0} . t\right)\right] \tag{1}
\end{equation*}
$$

where $f$ is some function.
The output of the system can be considered as one or more of the internal state variables which have been made available externally for observation, measurement or further use in some way. Thus $y(t)$, the present output, depends on $\mathbf{x}(t)$ the present state, and algebraically we can write

$$
\begin{equation*}
\mathbf{y}(t)=g[\mathbf{x}(t)] \tag{2}
\end{equation*}
$$

where $g$ is some function.
Equations (1) and (2) are what are called the "state equations" for a continuous system.

With a digital or discrete-time system we are concerned with the state $\mathbf{x}$ in Fig. 3 at particular instants of time, so using the symbology of Fig. 4(b) the present state is represented by $\mathbf{x}(k)$. Consider, for example, a simple digital system such as a bistable, with an input consisting of a triggering voltage and an output consisting of a voltage taken from one side of the circuit. It has two possible states, expressible in binary symbols as " $\circ 1$ " and


Fig. 2. The state of this mechanical "system" is the position of the ball within the frame of reference-given here by the distances $x_{t n}, y_{t n}$ and $z_{t n}$ which fix it at the instant $t_{n}$ "frozen" in the picture.

Fig. 3. Block schematic of any system. Each of the symbols $\mathbf{u}, \mathbf{x}$ and $\mathbf{y}$ represents a complete set of variables. In many electronic systems there is only one input variable and one output variable.


Fig. 4. Two scales for representing events in the time domain graphically: (a) for a continuous system; (b) for a "discrete-time" system, where we are only interested in what happens at particular instants.


"I o". Which state it occupies at any given instant depends on (a) its state at a previous time and (b) the presence or absence of a triggering voltage at that instant. Thus $\mathbf{x}(k)$ depends on $\mathbf{x}(k-1)$ and $\mathbf{u}(k)$. This can be expressed algebraically as

$$
\begin{equation*}
\mathbf{x}(k)=q[\mathbf{x}(k-1) ; \mathbf{u}(k)] \tag{3}
\end{equation*}
$$

where $q$ is some function. And since the present output $y(k)$ is an observable part of the present state

$$
\begin{equation*}
\mathbf{y}(k)=r[\mathbf{x}(k)] \tag{4}
\end{equation*}
$$

where $r$ is some function.
Equations (3) and (4) are the corresponding state equations for the discrete-time system.
At this point the reader may have become rather lost in the abstractness and generality of this form of system representation, so a look at some concrete examples may be helpful. Let us, for an example, consider the simple $L C R$ system in Fig. 5 in terms of the state concept. Here the input is a single variable, a voltage $e_{1}$. The output is also a single variable, the voltage $e_{2}$ developed across $R$. We must now decide how the state of the system is to be represented and what are the state variables. Earlier it was stated that the state variables are those variables associated with the elements which cause changes in the time domain-in particular energy-storage elements in continuous systems. Here, then, the state variables are clearly linked with $L$ and $C$. In an inductance the energy stored is given by $\frac{1}{2} L i^{2}$, so an obvious variable to choose as a state variable is the current $i$ flowing through the inductance. Similarly with the capacitance, the energy stored is given by $\frac{1}{2} C v^{2}$ and the most obvious state variable is the voltage $v$ across the capacitor. A further, practical, reason for choosing these two particular variables is that when the differential equations for such circuits are being solved the initial conditions necessary
for solution are usually most readily obtainable in terms of inductor current or capacitor voltage. But in practice other variables can be chosen, provided that they are linearly related to the basic state variables, the $i$ and $v$ in this case. For example, another possible pair of state variables in Fig. 5 would be $e_{2}$, which is obviously derivable from $i$, and $\mathrm{d} e_{2} / \mathrm{d} t$ which is related to $v$.

Since we wish to examine the state of the Fig. 5 system in the time domain, we must use differential equations to represent the rates of change of the variables concerned. In fact the general rule for expressing the behaviour of continuous dynamic systems in "state" form, developed from Poincaré's work, is to derive from the system equation a set of first-order differential equations which express the behaviour of the energystorage elements. In this case we have chosen $i$ and $v$ as the state variables, so the relevant rates of change to be used in the differential equations are

$$
\frac{\mathrm{d} i}{\mathrm{~d} t} \text { and } \frac{\mathrm{d} v}{\mathrm{~d} t}
$$

The system equation for Fig. 4, set up from Kirchhoff's voltage law, is


Fig. 5. A continuous system consisting of a simple LCR circuit. Considering it in terms of the "state" concept, one set of state variables can be the current through the inductance and the voltage across the capacitance.


Fig. 6. Diode "pump" circuit used as an example of a discrete-time system. The input is a sequence of pulses of amplitude $V_{i}$.

From which

$$
\begin{equation*}
\frac{\mathrm{d} i}{\mathrm{~d} t}=-\frac{R}{L} i-\frac{1}{L} t+\frac{1}{L} e_{1} \tag{5}
\end{equation*}
$$

The charge on the capacitor $q=C v$, so $v=q / C$

$$
\begin{equation*}
\therefore \frac{\mathrm{d} v}{\mathrm{~d} t}=\frac{\mathrm{d}(q / C)}{\mathrm{d} t}=\frac{1}{C} i \tag{6}
\end{equation*}
$$

(Since rate of change of charge, $q$, is current, $i$ ).

Equations (5) and (6), then, are the required first-order differential equations. It will be noted that these have been obtained from what can be considered as a second-order differential equation if the system relationships are expressed in terms of charge:

$$
e_{1}=L \frac{\mathrm{~d}^{2} q}{\mathrm{~d} t^{2}}+R \frac{\mathrm{~d} q}{\mathrm{~d} t}+\frac{1}{C} q
$$

The same principle could be applied to third, fourth or higher order differential equations-all can be reduced to sets of first-order equations.

It is now possible to bring the particular state equations (5) and (6) into a form corresponding with the generalized state equations (1) and (2). For this purpose we can present the set of first-order differential equations (only two in this case) in matrix algebra form, thus


Reverting to the original terminology of expressing state variables as $x_{1}, x_{2}$, $x_{3} \ldots$ etc. and input variables as $u_{1}, u_{2}$, $u_{3} \ldots$. etc., this becomes:

$$
\left[\begin{array}{c}
\frac{\mathrm{d} x_{1}}{\mathrm{~d} t} \\
\frac{\mathrm{~d} x_{2}}{\mathrm{~d} t}
\end{array}\right]=\left[\begin{array}{rr}
-\frac{R}{L} & \frac{1}{L} \\
\frac{1}{C} & 0
\end{array}\right]\left[\begin{array}{l}
x_{1} \\
x_{2}
\end{array}\right]+\left[\begin{array}{c}
\frac{1}{L} \\
0
\end{array}\right] \mathbf{u}_{1}
$$

And if we label the two matrices containing the $L C R$ constants as $A$ and $B$, we can express this in generalized matrix form as

$$
\frac{\mathrm{d} x}{\mathrm{~d} t}=A \mathbf{x}+B \mathbf{u}
$$

or

$$
\begin{equation*}
\mathbf{x}=\int[A \mathbf{x}+B \mathbf{u}] \mathrm{d} t \tag{7}
\end{equation*}
$$

which is the particular version of equation (1) for the particular system Fig. 5 .

As a concrete example of a digital or discrete-time system, consider a situation where a capacitor is not charged continuously, as in Fig. I(a), but only at particular instants of time. This occurs in the familiar diode pump circuit, an example of which is shown in Fig. 6. There is no need to describe the action of this circuit in detail here, $\ddagger$ suffice to say that when pulses are applied to the input the capacitor $C_{2}$ becomes charged in steps, the actual voltage across $C_{2}$ being determined by the number of pulses that have already occurred (assuming the pulse amplitude to be constant). To be more precise the $C_{2}$ voltage after' the $n$th pulse will depend partly on the amplitude of the $n$th pulse and partly on the number of previous pulses (of the same amplitude) that have occurred, $n-1$. The electrical formula expressing this condition is:**

$$
\begin{equation*}
V_{n}=V_{1} A\left(1+B+B^{2}+\ldots+B^{n-1}\right) \tag{8}
\end{equation*}
$$

where $V_{n}$ is the voltage across $C_{2}$ after $n$ input pulses, $V_{i}$ is the amplitude of the input pulses, and $A$ and $B$ are the following constants:

$$
A=C_{1} /\left(C_{1}+C_{2}\right) ; B=C_{2} /\left(C_{1}+C_{2}\right)
$$

Looking at this system in terms of its state, we can see that the state variables will be associated with the energy (and information) storage elements $C_{1}$ and $C_{2}$. The state variable which is of greatest interest to us (because it is used as the output) is $V_{n}$, so this will be called $x(k)$-remembering that $k$ is an integer. Similarly the exponent $n-1$ in equation (8) will be called $k-1$, The input pulse amplitude $V_{i}$, again in accordance with the earlier terminology, becomes $u$, and because the voltage is not a function of continuous time but of particular instants (see Fig. $4(\mathrm{~b})$ ) it becomes $u(k)$. Thus from this "state" point of view equation (8) would be expressed as
$x(k)=u(k) A\left[1+B+B^{2}+\ldots B^{k-1}\right]$
which is a particular version of the generalized expression for a discrete-time system given in equation (3).

This, then is the kind of symbolical language that is being used for describing systems in a precise manner in terms of the "state" concept. As such it does no more than sharpen and solidify the concept, just as verbalizing a vague idea in speech or writing sharpens and solidifies it. Are there any practical ends to served by analysing systems in this manner? It seems there may be, although it is too early yet to be certain. The process of breaking down the mathematical descriptions of complicated continuous systems into sets of first-order

[^6]differential equations is particularly useful when one realizes that computers, both analogue and digital, can be used to solve these sets of first-order equations quite easily, whereas the original equations would be relatively difficult to handle. So there may be some applications in computeraided design of electronic circuits. In addition state variable analysis avoids the problems which arise when one tries to apply classical linear theory to non-linear systems -the point being of course, that it is a form of analysis based on the time domain instead of the frequency domain. The method may also be particularly useful in dealing with complicated control systems with a multiplicity of inter-related inputs and outputs, as occur in process control and flight control. Some idea of these more practical aspects can be gathered from references 4 and 5. For the average Wireless World reader, however, the state concept must be at present little more than a way of looking at things. But, because it is a fresh view through another window, it may be helpful to the understanding.

Acknowledgement. I would like to thank Dr. A. G. J. MacFarlane, of the University of Manchester Institute of Science and Technology, for his kind help in the preparation of this article.

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## Corrections

"Test Your Knowledge-8"': As a result of a printing error, question No. 1 in the set on Electromagnetic Radiation (January, p.47) was presented incorrectly. The first two possible answers should read: (a) $\alpha$ radiation, (b) X radiation.
"Digital Exposure Timer": The following corrections should be made to three diagrams published in the January issue. The pulse outputs B and C in Fig. 6 should be relabelled A and B to key them with Fig. 3. (The related text on page 25 should also be amended.) In Fig. 5 the collector of $T r_{2}$ should be marked $V$ as this is the input for the output of NAND gate $\mathrm{G}_{3}$ shown in Fig. 3. Switch $\mathrm{S}_{8}$ drawn at the top left of Fig. 3 should be labelled stari, not stop. The diode between point $C$ and switch $\mathrm{S}_{4 b}$ in Fig. 3 should be marked $D_{23}$.

## Some other Measuring Rectifiers

by 'Cathode Ray'

Last month we considered the very simplelooking rectifier circuit shown here as Fig. I, as used for measuring alternating voltages. TQ recap., $C$ charges up through $D$ to very nearly the positive input peak voltage, and $R$ allows it to discharge slowly enough not to lose voltage significantly between one cycle and the next, but fast enough to make the voltmeter ready for another reading. The unidirectional output voltage, which is thus a close approximation to the peak input, can be indiçated either by a microammeter in series with $R$ or by a less delicate meter via a stable amplifier. There is a more often used variety of this rectifier in which $D$ and $C$ in Fig. I change places.
What we found was that the behaviour of these circuits is a good deal less simple than their appearance, or than it is represented to be in some of the specifications of electronic a.v. voltmeters in which such a rectifier comes at the front end. In particular, the quoted input resistance is valid only when applied to resonant circuits and is seriously misleading when applied to resistive circuits. What might look like an instrument suitable for measurements in quite high-impedance circuits could in fact be worse for that purpose than an ordinary metal-rectifier voltmeter, with the additional disadvantage of causing an unpleasant form of distortion.

Most ordinary a.v. voltmeters for power and audio frequencies use a bridge rectifier as shown in Fig. 2 in which $R$ is the rangesetting resistance. During positive halfcycles, current flows through the moving coil milliammeter $M$ via $D_{1}$ and $D_{4}$, and during negative half-cyles in the same direction through $M$ via $D_{3}$ and $D_{2}$. The forward resistances of the diodes are normally negligible compared with $R$, so the behaviour of the instrument conforms to the simple pattern in Fig. 3, which shows that the defiection will be proportional to the mean (average) value of the input voltage, regardless of polarity. Almost always such instruments are calibrated in I•II $\times$ mean value, which directly indicates r.m.s. values so long as the waveform is either sine or square. Unless the Fig. I type is specifically called a peak-reading voltmeter, it too is usually calibrated in what are correctly r.m.s. volts with the same proviso. Both types are misleading when used on other waveforms.
There is no doubt about input resistance with the Fig. 2 type-except perhaps on the


Fig. 1. One variety of the simple peak rectifier circuit considered last month.


Fig. 2. Bridge type of rectifier circuit used in many multirange a.v. voltmeters.


Fig. 3. Voltage and current waveforms for the Fig. a circuit.
lowest range-because the diode and meter resistance is, normally made small compared with $R$. On any ranges, and whatever kind of circuit one is measuring, one can be sure that the resistance of the instrument is at least $R$. And it remains practically constant throughout the cycle, so there is not the distortion one gets with the Fig. I type as a result of its drawing a sharp pulse of current at each positive (or negative) peak. But Fig. 2 is limited, as I said, more or less to power and audio frequencies and perhaps the lower radio frequencies, owing mainly to the shunting effect of unwanted capacitances.

Another circuit that hás been coming into use is usually drawn as in Fig. 4, which suggests that it is akin to Fig. 2. Sometimes it is even called a half-bridge rectifier circuit. But really it is a pair of Fig. I type, with diodes connected oppositely so as to give outputs of opposite polarity. It is therefore a voltage-doubling rectifier, which Fig. 2 is not. In fact, because its output is nearly peak-to-peak whereas Fig. 2 yields the mean value, with a sine wave it gives $\pi$ times as much.

Instead of having two separate leak or load resistors, we can connect them in series between the output terminals to make one. The potential of their junction or centre point varies only slightly with respect to that of the bottom 'terminal ('earthy'), so the behaviour is practically the same as that of two separate Fig. I rectifiers, without the necessity for connecting the half-loads across $C_{1}$ and $C_{2}$, so one gets the advantage of the modified Fig. I ( $R$ across D) without its disadvantage (a.v. in the output as well as d.v.). Because $C_{1}$ and $C_{2}$ can be large, the output device is all nearly at earth potential so far as the a.v. is concerned. All the same, if this circuit is used for high radio frequencies (and at least one commercial example is rated up to 1.5 GHz ) it would be asking for trouble to connect a microammeter quite so near the input terminal, and for this and other reasons a high-input-impedance amplifier is used. The fact that it has to be a balanced one is no real disadvantage, because that would be preferred anyway for stability and interference rejection.
But there is still the same complaint that we had about its single-phase prototype, Fig. I-unless you think the fact that both peaks equally have to provide capacitor-


Fig. 5. If voltage-operated, the Fig. 4 circuit behaves as a pair of Fig. I type rectifiers, giving voltage and current waveforms like these.
charging current pulses is any mitigation. Some might regard it as an aggravation!

However, although we cannot be particularly enthusiastic about this kind of rectifier at the front end, it is interestingly different in the class of electronic volimeter in which the amplification comes first. Here the amplifier must not only be very constant -that goes without saying in a measuring instrument-it must also be so over the full frequency range of the voltages to be measured. That rules out the hertz to gigahertz specification that is possible with the front-end rectifier, and concentrates design skill on the amplifier. Negative feedback is essential, of course. Usually one doesn't think of including a rectifier in a feedback system, because rectification is a drastic kind of distortion, which negative feedback would do its best to iron out. However, this is not necessarily so with a full-wave rectifier such as Fig. 2 or Fig. 4. The current through $R$ in Fig. 2, and therefore the voltage across it, has the same waveform as the input, except to the extent that the non-linear diode forward resistance is appreciable compared with $R$. So if this rectifier circuit were fed from an amplifier, the voltage across $R$ could be used for negative feedback. It would be current feedback, because that voltage is proportional to the amplifier output current. One effect of negative current feedback is to increase the apparent output resistance of the amplifier, making it more like a constantcurrent source. So the non-linearity of the diodes is made even less significant than it was, and the rectifier more nearly ideal.

The same technique works with Fig. 4. However, making the diodes more like perfect on-off switches is not the most significant or interesting effect of current
feedback in this case. Without feedback, the voltage and current diagram is something like Fig. 5, which is the full-wave voltagedoubling counterpart of the one we studied last month in connection with Fig. I. Here the current waveform is obviously drastically distorted. If however Fig. 4 is worked from a source impedance which, owing to the use of current feedback or otherwise, is very large compared with the load resistance connected to its + and - terminals, and also linear, the current waveform is forced to be that of the source. The impedance of the rectifier as a whole is too relatively small to distort it. Clearly the operation of the rectifier must be quite different from that represented in Fig. 5, which applies to a more or less constant-voltage source.

Let us continue to assume that the applied waveform is sinusoidal, but we must begin again on the basis of current instead of voltage. In Fig. 6, then, we draw a sine wave for current; 1 . The corresponding equivalent circuit diagram is shown as Fig. 7, where the diodes are represented by a twoway switch, $S$. Let us suppose that the capacitances of $C_{1}$ and $C_{2}$ are equal to one another and large enough to maintain their charges almost constant between switch movements (half cycles). Then point $a$ will be at an almost constant positive voltage which we can represent approximately in Fig. 6 by the horizontal line marked $V_{a}$, and $b$ is at an equal but negative voltage represented by the line marked $V_{b}$. The nearly constant voltage between $a$ and $b$ drives a nearly constant current, $i_{L}$, through the load resistance $R$, and this current is approximately represented by the horizontal line $i_{L}$. During the first half-cycle, when the switch is to the left, $i_{L}$ and $i_{b}$ are necessarily identical. During the same period $i_{a}=I-i_{L}$. So we can draw $i_{a}$ as a positive half-cycle, lower than $I$ by the amount $i_{L}$. During the other half-cycle everything is reversed: $i_{a}$ is constant and negative, and $i_{b}$ has the same form as $I$ but is more positive by $i_{L}$. Because the current through (more correctly, into and out of) a good capacitor can have no d.c. component, the $i_{a}$ and $i_{b}$ waveforms must enclose equal areas above and below the zero-current line. That fact fixes $i_{L}$. And if you use any of the various methods for equalizing the areas, you should arrive at the result that $i_{L}=I_{\text {(peaki) }} / \pi$. (Students of duality will note with pleasure that this corresponds to the voltage relationship, already mentioned, between the voltage output in Fig. 5 and that in Fig. 3.)

We can now get a better approximation to the voltages at $a$ and $b$. At the start of the cycle $I$ is zero, so the current $i_{L}$ is kept going by the discharge of $C_{1}, i_{a}$ being negative. This phase lasts only a short time and soon $i_{a}$ goes positive and remains so for most of the remainder of the half-cycle, positive current being a charging current. These charges and discharges cause the voltage at $a$ to vary in the manner shown rather exaggeratedly by the dotted line marked $v_{a} ; V_{a}$ is now the mean voltage at $a$. The $v_{b}$ waveform is arrived at in the same way, and we now see that the voltage across the load, $v_{L}$, varies even less than $v_{a}$ and vo separately. Except near the low-frequency end of the range, $v_{a}$ and $v_{b}$ should
be hardly distinguishable from $V_{a}$ and $V_{b}$. And the voltage, $V$, of the top terminal, being $v_{"}$ during the first half-cycle and $v_{0}$ during the second, is a near-square wave.

As the Fig. 4 circuit used in this way as a current instrument produces readings proportional to mean values, that fact must be kept in mind if it has an r.m.s. calibration and is used to measure non-sinusoidal and non-square waveforms. An example of this type of instrument was described by D.E.O'N. Waddington in the March 1966 issue.

There is another variety in which resistors are used in place of the capacitors $C_{1}$ and $C_{2}$, but it seems to lack any advantages.


Fig. 6. If Fig. 4 is current-operated, its voltage and current waveforms are like this. The overall voltage $V$, not shown, is a near-square wave, the first half-cycle being like $v_{a}$ and the second like $v_{b}$.


Fig. 7. Equivalent circuit diagram corresponding to Fig, 4, the diodes being simulated by a switch. The symbols correspond to those in Fig. 6.

# News of the Month 

## Giant electronics merger

A company with an electronics turnover in excess of $f, 100 \mathrm{M}$ has resulted from the move to bring together the electronics interests of G.E.C.-A.E.I., English Electric and Elliott Automation. The new company, called G.E.C. Marconi Electronics Ltd will have Robert Telford as managing director. Robert Telford also continues as managing director of the-Marconi Company.

The new company has interests in defence and broadcasting equipment and communications and navigational equipment for land, sea, air and space.
G.E.C.-Marconi Electronics Limited will be responsible for the management of the following units:-
from Marconi Company:
The Marconi Company Limited (Chelmsford and district, Basildon, Billericay, Gateshead, Hackbridge and Wembley); including Marconi Instruments Lid. (St. Albans and Stevenage). Marconi-Elliott Microelectronics Ltd. (Witham and Glenrothes).
Eddystone Radio Ltd. (Birmingham). and including all Marconi subsidiary companies overseas;
from G.E.C.-A.E.I.:
G.E.C.-A.E.I. (Electronics) Ltd. Radar, Aerospace and Defence Division (Stanmore, Portsmouth, Leicester and Watford).
G.E.C.-A.E.I. (Electronics) Limited, Communications Division, (Coventry);
from Elliott-Automation:
Elliott Flight Automation Limited (Rochester).
Elliott-Automation Radar Systems Lid. (Borehamwood, Hillend).
Elliott Space and Weapon Automation Limited (Frimley, Borehamwood, and Hillend).
E.-A. Space and Advanced Military Systems Limited (Camberley).

## Domestic monochrome-tocolour converter

A converter that enables colour pictures to be viewed on a standard monochrome receiver has been built by A. Becker of Scottish Television. The operating principles of the converter are not new, in fact they can be traced back to Baird's experiments.

The electronics of the converter consist of about a dozen transistors mounted on a small board that in most cases would fit into the cabinet of the monochrome receiver. The unit converts a chroma signal from the receiver's video stage into a sequential $R-Y$, $\mathrm{B}-\mathrm{Y}, \mathrm{G}-\mathrm{Y}$ signal which is fed to the grid of the receiver c.r.t. The picture on the screen is viewed through a rotating optical filter which, in its simplest form consists of two tri-colour filter groups rotating at 500 r.p.m. in synchronism in phase and frequency with
frame drive. A block diagram of the system is shown in the accompanying drawing.
Mr. Becker tells us that in the kit form the system would cost less than $£ 30$ and he is seeking commercial exploitation of the idea.

## Watt output convention?

It would appear that manufacturers of amplifiers are going to have to watch the watts they claim that their products will produce. The Association of Public Address Engineers point out that the figures used to express the output power of amplifiers would seem to be covered by the new Trade Description Act.

The correct method of expressing the power output of audio amplifiers is given in BS 3860:65. This states that two power ratings should be given, these are rated and maximum. The rated power is the output an amplifier will provide continuously at some value of harmonic distortion lower than is claimed for the amplifier. The maximum power output is the power developed across the load when the level of harmonic distortion equals that specified for the amplifier. In both cases the input is a 1 kHz sine-wave.

It is difficult to see why some amplifiers are rated using the American I.H.F.M. convention. Is it because this figure is about $50 \%$ higher than the continuous rating and that the less informed members of the public may be deceived?

## U.K. Audio Engineering Society?

As a result of a meeting of some 70 former members of the defunct British Sound Recording Association an approach is being made to the American Audio Engineering Society with a view to forming an affiliated society in the U.K. The meeting, held at the Hotel Russell on December 10th, set up a small working committee, comprising R. E. Cooke, J. C. G. Gilbert, J. Maunder, R. Baldock and N. Leevers, all members of the A.E.S. John Gilbert, who convened the meeting, has been assured of the financial support of some 20 audio manufacturers. Readers interested in this proposal may obtain further information from 10 Museum Street, London W.C.1. Letters should be marked "Audio Engineering Society".

## Hospital computer system

The Department of Health \& Social Security, the Birmingham Regional Hospital Board and International Computers Limited are co-operating in the development of a large hospital information system and communication network. The system will employ remote access terminals connected to a central system 4-50 computer and will be installed at the North Staffordshire Hospital Centre. Initially the equipment will handle out-patients' bookings and clinical records and will be followed by the development of similar facilities for in-patients. In addition to the computer, 12 video terminals and 4 teleprinters, will be installed in a new building in Stoke-on-Trent in April 1970. Further
remote terminals will be added to the system as the project develops.

Video terminals located in the outpatient clinic areas will be used for making appointments, requesting tests, displaying test results, entering patients on hospital bed waiting lists, and also to book patients for examination in other clinics.

In addition summary clinical information on patients, such as diseases diagnosed, test results, operations performed and drugs prescribed, will be printed out by the computer for the doctor.
In the first instance video terminals will be installed in the admission offices of the Hospital Centre, the North Staffordshire Royal Infirmary and the City General Hospital. These terminals will be used for admission and bed allocation procedures.
The next stage will involve the extension of the system into selected wards of the two main hospitals by the installation of teletypewriter terminals at strategic points to serve groups of wards.

The terminals in wards will primarily be used to make requests and receive reports on investigations, to aid the prescription and administration of drugs and for putting patients on operating lists. They will also help to establish the existence and location of all records relating to a patient. In certain cases, for instance, the reporting of biochemical tests, the computer will detect and draw to the attention of the medical staff any abnormal results. A further use of ward teleprinters will be the recording of diagnoses.

## Detecting loose articles

Foreign particles left in equipment after manufacture can be detected with the aid of a device developed by the General Electric Company of U.S.A. The presence of particles that may be too small to be seen with the naked eye can be detected.

The equipment to be tested is placed on a standard vibration machine and a very sensitive listening device enables one to hear foreign particles bouncing.

Accelerometers mounted to the vibrating


Showing the loose article detector in operation.
table feed signals into the electronic listening device so that unwanted sounds are cancelled.

The signals detected by the listening system can be monitored over a loudspeaker system or on an oscilloscope. It is asserted that different articles can be identified by the nature of the bounce or by their "sound signature" to use the jargon coined by G.E.

## Racal expands

The 88.9 M bid made by Racal Electronics for Controls and Communications has been favourably received. At the time of going to press a spokesman for Racal said that his company had succeeded in obtaining more than $50 \%$ of C and C's shares. Racal's offer follows the breakdown of talks between Plessey and $C$ and $C$. Subsidiaries of $C$ and $C$ are Airmec, British Communications Corporation; Modern Aerials, Thermionic Products and Vectron Electronics.

## British sensing equipment in HEOS-A

ESRO's highly eccentric orbit satellite (HEOS-A) which was successfully launched on December Sth, from the Kennedy Space Centre has an apogee of about $225,000 \mathrm{~km}$ (two-thirds of the distance to the moon) and an orbital period of 4.5 days. Because of the vast distance from the earth during orbit it was necessary to employ two methods of attitude measurement to ensure that the satellite is aligned correctly to the earth. The first, for close-range alignment, depends on infrared sensors which use the earth's horizon as a reference. The second, for longrange alignment, uses a unique sensor to measure the reflection of sunlight from the face of the earth. A solar aspect sensor measures the attitude of sunlight striking the satellite enabling it to be aligned correctly with respect to the sun. The scientific mission of HEOS-A is to study interplanetary physics, particularly magnetic fields, cosmic radiation and solar winds outside of the earth's atmosphere. The attitude sensing and control system was designed by British Aircraft Corporation.

## U.S.A. to have Pay-TV

America's Federal Communications Commission has recently authorized subscription television operations on a national scale, the rules governing this type of operation to become effective on June 12th, 1969. This announcement is in direct contrast to the state of affairs in this country. After the Government's decision to refuse the British Pay-TV experiment to expand the company was run-down as rapidly as possible, as reported in Wireless World, December 1968, page 444.

## Canada's own satellite

Canada could have its own communications satellite by 1971 if proposals resulting from a study are accepted by the Department of Industry. The study was carried out under a six-month contract by the Northern Electric Company Lid, with Canadair and the


A mock-up of the proposed Canadian communications satellite.

Hughes Aircraft Company as sub-contractors.

The proposed six-channel satellite would be placed in a high synchronous orbit and would have a total capacity of 6,000 one-way voice circuits or six colour television channels. Primary power source would be formed by 18,780 solar cells. During times when the satellite is in eclipse power would be provided by a battery of nickel-cadmium cells.

## Component Associations, more co-operation

Closer integration between the various trade associations concerned with active and passive components has been achieved since Sir Alan Dudley was appointed director of the Electric Components Board last year. In January, 1969, the Radio and Electronic Manufacturer's Association moved into premises at Mappin House, 4 Winsley Street, London, WIN ODT., where the British Radio Valve Manufacturers Association, the Electronic Valve and Semiconductor Manufacturers Association and the Electronic Components Board are already accommodated.
Although each Association will continue to deal separately with matters of specialist interest, matters concerning the electronic components industry as a whole will be handled by the Board.

## Radio and space research report

In the recently published first triennial report of the Radio and Space Research Station (H.M.S.O. 7s 6 d ) some details are given of studies being carried out on investigating the mechanism of solar flares. It is thought that it may be possible to predict their occurrence. This would be of great value to operators of supersonic aircraft, such as the Concorde, who would be able to predetermine whether or not dangerous levels of radiation are likely to exist at high altitudes.

Studies of the physics of the ionosphere have continued using data obtained from the

Canadian satellites Alouette 1 and 2 received by telemerry at the Radio and Space Research Stations situated at Winkfield, Berks, and Singapore. Work done in this connection, derived from satellite drag data, has indicated that it is atmospheric winds at heights between 100 and 700 km that are responsible for the many ionospheric anomalies.

More work has been done on relating meteorological data to the characteristics of u.h.f. signals propagated in the troposphere. This work will be extended with the aid of the Chilbolton Tracking station.

In anticipation of the commercial use of millimetre waves for communications investigations of their transmission characteristics have begun. The scintillation of signals arising from variations in atmospheric refractive index has been studied and measurements made of attenuation caused by rain on 2.9 mm waves propagated over a path of 300 m .

Research at 0.793 mm has been concerned with generation and detection and absorption by water vapour.

The Station's experiment to measure high-frequency radio noise distribution over the world, started in May 1967 in Ariel 3, has progressed well. Noise intensities above major storm areas are in broad agreement with expectations and much new information has been obtained from areas not covered by ground stations.

## Libyan television

The television service of Libya was inaugurated on December 24th 1968 by the Crown Prince. The B.B.C., which was represented at the opening ceremony by a twelve-man outside broadcast team, has been associated with the project since 1964 and will have seventeen engineers seconded to the service for a year.

The service is operating on 625 -line standard $G^{*}$ which differs very slightly from the U.K. standard. For colour the PAL system is employed and we understand that experimental colour broadcasts have already begun using a 16 mm telecine machine as the source of programme material.

* Standard G: 625 lines; channel width 8 MHz ; vision bandwidth 5 MHz ; sound separation +5.5 MHz ; vision modulation negative going; sound f.m.


## Powerful ion laser

A powerful gas laser system which emits beams of the three primary colours has been
developed by the Japanese Hitachi Company. Coherence and power of the beams are of sufficiently good quality to enable the system to be used to produce full-colour holograms or to project coloured images on to a screen. The system comprises a krypton laser which emits red light and an argon laser which emits blue and green light, a single blue or green line being selected by a mirror in the optical resonator. Outputs are of the order of 5-6W.

## "WIRELESS WORLD" INDEX

The Index to Volume 74 (March-December 1968 ) is now available price 2 s . 6 d . (postage 4 d .) Cloth binding cases with index cost 11 s .6 d ., including postage and packing. Our publishers will undertake the binding of readers' issues, the cost being 40 s . per volume including binding case, index and return postage. Copies should be sent to Associated lliffe Press Lid., Binding Department, c/o 4 lliffe Yard, London, S.E.17, with a note of the sender's name and address. A separate note confirming despatch, and enclosing the remittance, should be sent to the Publishing Department, Dorset House, , Stamford Street, London, S.E. 1 .

## Colour TV in Sweden

When Sweden's colour television service starts in 1970 a major role will be played by a British equipment manufacturer. Twentythree of the existing black and white v.h.f. television stations in Sweden incorporate Marconi transmitters and these will all be modified to enable them to transmit colour pictures as part of that country's first colour TV service.

## DX listeners' award

The B.B.C. is offering a certificate award (called World Radio Club Award) to listeners who correctly report on a number of B.B.C. transmissions received from different locations. It applies to one frequency schedule period only, from March 2nd to May 3rd. To qualify for the award, listeners must be members of the World Radio Club run by the B.B.C. and must give evidence of reception of three B.B.C. transmissions from each of the following areas: Great Britain and the Atlantic, east Mediterranean and Far Eastern relay stations. Membership of World Radio Club can be obtained by writing to B.B.C., Bush House, London W.C.2.


Gas laser developed by Hitachi of Japan emits beams of the three primary colours.

## U.H.F. transmitters

The following table gives, alphabetically, B.B.C. and I.T.A. u.h.f. television stations with the channels allocated for the three proposed programmes. Those transmitters already in operation (to date BBC-2 only) are shown in heavy type. Relay stations are shown after the main station of the group. For extended coverage, further relay stations are, of course, planned. It is intended that individual BBC-1 and I.T.A. stations will "come on the air" simultaneously, using the 625 -line standard, and they will transmit programmes in colour from the start. This list is provisional and the information it contains is subject to alteration.


Angus ( H )
BELMONT (H)
Bilsdale West Moor (H)
BLACK HILL (H)
Caradon Hill (H)
Craigkelly (H)
CRYSTAL PALACE (H)
Guildford (V)
Hemel Hempstead (V)
Hertford (V)
High Wycombe (V)
Reigate (V)
Tunbridge Wells (V)
DIVIS (H)
Larne (V)
DOVER (H)
DURRIS (H)
EMLEY MOOR (H)
Chesterfield (V)
Halifax (V)
Keighley (V)
Sheffield (V)
Hannington (H)
Heathfield ( H )
Newhaven (V)
Limavady (H)
Londonderry (V)
LLANDDONA (H)
Betws-y-Coed (V)
Mendip (H)
Bath (V)
Bristol (V)
West Sussex ( H )
Moel-y-Parc (H)
OXFORD (H)
PONTOP PIKE (H)
Fenham (V)
Newton (V)
Weardale (V)
ROWRIDGE (H)
Brighton (V)
Salisbury (V)
Ventnor (V)
Sandy Heath (H)
SUDBURY (H)
SUTTON COLDFIELD (H)
Brierley Hill (V)
Bromsgrove (V)
Fenton (V)
Kidderminster (V)
Lark Stoke (V)
Malvern (V)
TACOLNESTON (H)
Aldeburgh (V)
West Runton (V)
WALTHAM (H)
WENVOE (H)
Aberdare (V)
Kilvey Hill (V)
Merthyr Tydfil (V)
Mynydd Machen (V)
Pontypridd (V)
Rhondda (V)
WINTER HILL (H)
Darwen (V)
Haslingden (V)
Pendle Forest (V)
Saddleworth (V)
Skipton (V)
Todmorden (V)
$\mathrm{H}=$ Horizontal polarization
$\mathrm{V}=$ Vertical polarization

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Model ES- 25 watts. Fitted with $\frac{1}{4}$ " bit. Interchangeable spare bits $\frac{3}{32}$. $\frac{3}{7^{\prime \prime}}{ }^{\circ}$ and $\frac{1^{\circ}}{4}$. For 240, 220, 110, 24 and 12 volts. FROM 35/-


Model F- 40 watts. Fitted with ${ }^{5}{ }^{56}{ }^{" 1}$ bit. interchangeable spare bits $\frac{1^{* *}}{4}$. $\frac{3}{188^{\circ}}, \frac{1}{8}{ }^{\prime \prime}$, $\frac{3}{32}$. For $240,220,110,24$ and 20 volts. FROM 42/6.



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Giro No. 2581000
transient conditions. These outputs operate the printer or other mechanism, and at the same time set stage $C$, which then is set again on every cycle, thus inhibiting the striking of, say, B. As soon as the glide finishes this inhibition is automatically removed and normal operation resumed. The other actions can be followed easily.

## Recognition of words

The main grouper (Fig. 4) discussed above can, by itself, and without the VC-L circuits act as a recogniser for words carefully chosen $t 0$ match its capabilities. For instance consider the words of command shown in Table 5 , where they have been broken down into their basic constituents. At first sight it might appear a reasonably satisfactory selection, but closer inspection shows some defects. For instance, apart from the beginnings, 'not' and 'right' are identical and both ' $r$ ' and ' $n$ ' tend to be somewhat illdefined phonemes; moreover if an attempt were made to add 'left' as another control action, the difficulties would be increased to an impossible extent. The solution here would be to replace 'right' and 'left' by, for instance, 'starboard' and 'port'. Alter-
natively manufactured words of well defined photenic content could be used. Table 6 lists suitable phonemes for constructing strong words from.

However, with the equipment in the form described, values for $A, B, C$, and $G$ are available, and the next step is to use them. Thus instead of saying that a phoneme was a vowel, it could be specified as a vowelA1; or in further detail as VC-V: $\mathbf{A}_{1}: \mathbf{B}_{1}$ : $\mathrm{C}_{3}$-thus identifying itself as ' $a$ ' from Table 1 in Part 1. This approach will be considered in more detail below, but meanwhile the digits $0-9$ may be examined using the main groupings and $A$ values only (Table 7). It will be seen that separation appears to be quite good, except for ' 2 ' and ' 3 ', which can however be very easily separated on the value of $B$. However the position is not as simple as this, because the weak phonemes, show in brackets, may or may not be detected. Thus '5' can easily be confused with '9', and perhaps also with ' 1 '. However, in spite of troubles such as these, a recogniser for the spoken digits can be built along these lines and will give some $90 \%$ correct responses for well enunciated speech.



Fig. 7 Delay, burst frequency, and amplitude for unvoiced stops

It is perhaps unfortunate that it seems to be a feature of languages generally that opposites are very often phonetically or actually similar. Thus is Latin 'altus' means both high and low. In English we have 'left' and 'right' which are phonetically close, or 'starboard' and 'larboard' which are even closer. The tendency is however for separation to occur: thus 'larboard' has become 'port' and 'yea' and 'nay' have become 'yes' and 'no', and later 'affirmative' or 'negative'.

## Recognition of phrases

In a machine for identifying words, the gaps between the words are marked and are unlikely to be confused with other gaps, as those in stopped utterances. Moreover, if a single word is missed, the machine can demand an immediate repetition. In a machine designed to recognise phrases, however, doubtful gaps and missed phonemes can very quickly ruin the most perfect identification scheme. Under these conditions the machine is best arranged to operate as the brain is believed to-that is to compare the incoming message with the most likely of the repetoire of possible messages, on the basis of what has gone before, and to mark off points of resemblance as they occur. The problem is made more difficult by the inclusion in any worthwhile practical situation (such as air traffic control) of variable data, the position of which in the message is known, but not the content, and which must be detected accurately for the message to be effective. There are clearly many more problems here than those of phonetics only.

## Recognition of phonemes

The recognition of phonemes is a useful study in itself, for it could lead to a machine producing a phonetic script from a spoken input, and possibly ultimately to an orthographically correct script. It also has other uses, such as determining which phonemes can be most reliably detected. Using the coder of Fig. 1 and the main grouper of Fig. 4 to supply inputs as before, about half the phone-
mes can be identified with accuracies of $\mathbf{8 0 \%}$ or above in carefully enunciated connected speech, and of the remaining phonemes many of the errors are near misses. So far the resulting script has proved barely intelligible if the subject matter is unknown, but moderately intelligible if the subject is familiar. Much further work is needed before a commercial machine can be made, but the remainder of the article will be devoted to a description of the method used in one system. The same nomenclature as that used above is employed.

The first piece of additional equipment is a circuit designed to select the appropriate A value at the start of a vowel or diphthong. In Fig. 5, as soon as bistable B (Fig. 4) is set, the four bistables 10-13 (Fig. 5) are primed, and whichever output from the averaging $A$ stages is effective fires the appropriate bistable which is reset only on the resetting of bistable B. Once one of the group is set, all are inhibited. The output from this group are labelled $A_{1} I, A_{2} I$, etc.
As an example of the subsequent processing the coding for Group $A_{1}$ vowels and diphthongs (Fig. 6) may be considered. All the bistables shown are reset when the printer or other display apparatus receives a command to print. All the outlets are scanned in the order given in square brackets after the symbol, and the first one found energized is printed, the remainder being reset. This enables subsequent characters to be selected whilst a relatively slow display is acting. The command to print is given whenever the main grouper (Fig. 4) changes state, or on the receipt of a valid ending to a diphthong, indicated by an asterisk(*) on the diagram. In the latter case the print command on the next change of state of the main grouper is inhibited. In Fig. 6, the setting of $A_{1} I$ (Fig. 5) primes bistables $46,42,36$, and 47 , which can thereafter be set by $\mathrm{B}_{1}, \mathrm{~B}_{3}, \mathrm{~B}_{4}$, and $A_{3}$ or $A_{5}$ (uninhibited). Assume that $B_{1}$ is firing: then 46 will set-this represents ' $a$ ' or the first stage of the diphthong 'ie'. If the former condition, then ' $a$ ' will be printed when the main grouper changes state. If the input is the diphthong, though, as soon as the first formant period increases to energize $\mathrm{A}_{3}$, then 44 and 45 will be primed, and in the case quoted, 45 will set, thus energizing the 'ie' print line, and immediately executing the print command, as described above.
Similar groups of bistable elements deal with the other vowel groups. Fricatives are handled similarly. Stops require a more complex arrangement as several features need to be taken into consideration. As a fully satisfactory coding arrangement has not yet been found, a full description will not be given. The factors involved are:
(a) The amplitude of the fricative burst formed on the release of pressure. This is greatest for ' $k$ '.
(b) The high frequency content of the burst.
(c) The duration of the silent period.

Fig. 7 shows the relationship between the length of silent period and the peak high frequency ( $C$ channel) response for the voiced and unvoiced stops. Using the additional parameter of high energy, ' $k$ ' can be separated from ' $t$ ' and ' $p$ ', which can be separated between themselves by high frequency level. With the voiced stops
(' $b$ ', ' $d$ ', and ' $g$ ') however, separation is much more difficult, and is perhaps impossible to achieve except on a syllabic basis which would take account of the context.

In conclusion it may be said that the field of speech recognition is only just beginning to be explored. It is of the greatest interest and is not so complicated that an individual worker cannot make useful progress. What are needed at present are more good ideas to make for simpler and more direct precoding. The remaining operations are then perhaps better done on a general purpose computer.

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## Announcements

"Advances in microwave applications of semiconductors" is the title of a course of ten lectures to be given on Wednesday afternoons commencing January 15 th at the Borough Polytechnic, Borough Road, London S.E.L. There is also a course of nine lectures entitled "Transistors in communication circuits" to be held on Tuesday evenings commencing January 21 st Fees are 3 gns and $\ell, 210$ s respectively.

A course of eight laboratory sessions for students who have attended a course on transistor theory and wish to gain practical experience will be held at Hendon College of Technology, The Burroughs, Hendon, London N.W.4, on Thursday evenings commencing February 27 th . There is also a nine-lecture course on hi-fi sound reproduction on Wednesday evenings from January 291h.

The Electronics Division of the I.E.E. is organizing a vacation school on aerials to be held at the University of Birmingham from July 7ih to 19 th .

A lecture course entitled "Semiconductor devices and circuit techniques" will be held at Bournemouth College of Technology commencing February 4th for eight weeks.

Show cancelled. The Electronic I'roduction Equipment Exhibition arranged for March 10th-14th at Earls Court (see list of 1969 Conferences and Exhibitions in our January issue) will not now be held.

Welwyn-Berco agreement. British Electric Resistance Company and Welwyn Electric have agreed to exchange know-how and commercial interests in the areas of fixed and variable power vitreous enamelled urits. By 1 st April this year Welwyn will stop making theostats (Berco taking responsibility for any of Welwyn's outstanding commitments) and Berco will stop manufacturing fixed vitreous enamelled resistors (Welwyn taking over their outstanding commitments)

The Westinghouse Electric International Company of New York, and Ferranti Lid., jointly announced a technical exchange agreement involving Westinghouse electro optical systems and Ferranti air-to-surface radar systems. For the past three years both companies have been working on avionics and navigation systems for the Phantom F-4 aircraft purchased by the United Kingdom.

Audio equipment hire. Carston Electronics have formed a new division to provide long- or short-term hire for a range of tape recorders, microphones, and other studio sound equipment as well as audio test equipment. Hire periods may be from a week up to three years or longer and equipment will be maintained throughout the hire period at no extra charge. Carston Electronics Lid, 71 Oakley Road, Chinnor, Oxon (Tel. Kingston Blount 8561)

SGS-Fairchild have changed the name of the company to "SGS (United Kingdom) Lid".

Honeywell Controls Lid of Brentford, Middlesex, have changed the name of the company to Honeywell Lid.
R.E.C.M.F. move. The Radio \& Electronic Component Manufacturers' Federation have moved from 6 Hanover Street, W.I, to Mappin House, 4 Winsley Streel, Oxford Street, London, WIN ODT. (Tel: 01-580 8562.)

The address of Muliard's Order Department (Disributor Sales), is now New Road, Mitcham, Surrey, CR4 4SR. (Tel: 01-648 3471.)

The southern sales and export offices of F. C. Robinson \& Partners Lid are now at Rilton House, Uxbridge Road, London W.S. (Tel: 01-579 2041.)
E.M.I. Lid, have completed an agreement with B. \& F. Instruments Inc., an electronic instrument company of Philadelphia, U.S.A, in which they are taking an equity investment, with the option to acquire a majority interest, to direct the marketing of their own instrument range, particularly the products of their subsidiary S.I). Laboratories

Guest Electronics L.td., Nicholas House, Brigstock Road, Thornton Heath, Surrey CR4 7JA, have acquired the exclusive marketing rights within the U.K. of Luft Instruments Inc., U.S.A

Rastra Electronics I.td., 275 King Street, Hammersmith, L.ondon W. 6 , have been appointed official agents for part of the Redpoint range of small heat sinks.

A contract valued at $£ 125,349$ has been placed with Marconi Instruments, for the supply of 127 signal generators, Type TF 2002AS, by the Canadian Department of Transport. This instrument covers the carrier frequency range 10 kHz to 72 MHz with provision for a.m. up to $100 \%$ at modulating frequencies from 20 Hz to 20 kHz .

Orders worth over $f, \frac{1}{2}$ million have been placed by the G.P.O. with S.T.C's Microwave and Line Division, Basildon, for 6 GHz equipment to provide up to seven broadband radio links between Carlisle and Belfast, Carlisle and Manchester, and L.eeds and Newcastle.

The Indonesian Government have placed an order valued at $£ 55,000$ with Racal for a large quantity of the "Squadeal" Manpack h.f. transmitter-receiver together with other mobile radio and ancillary equipment.

The Industrial Products Group of British Aircraft Corporation's Guided Weapons Division have been awarded a product on contract valued at almost $\mathcal{L} 200,000$ by the G.P.O. The contract is for semiautomatic test set equipment, known as TRT 116.

Labgear L.td, a member of the Pye of Cambridge group, have received an order valued at $\{60,000$ from the Iraq mobile police force for s.s.b. packset transceivers.


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| :--- | :---: | :---: | :---: |
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## Circuit Ideas

## Readers' tricks and trifles



Fig. 1. Circuit for rumble suppression.


Fig.2. Output voltage with respect to frequency for 1 signal level to each channel.
circuit which effectively gives mono at low frequencies and allows for signal separation at frequencies above say 200 Hz .

The circuit given (Fig.1) allows for such action and also boosts the bass frequencies. The $2 \mathrm{M} \Omega$ potentiometer across the two outputs allows balance to be controlled. The main filtering action of the circuit is slightly disturbed when the balance control is offcentre. The potentiometer can be replaced by fixed resistors if the two inputs are equal. The effect of the filter is shown in Fig. 2.
David Ralph,
Northfleet,
Kent.

## Switching decades of capacitance

In order to switch decades of capacitance with a single control it might seem that the alternatives are either a single wafer specially made, or a rather buiky 4 -wafer switch (Radio Lab. Handbook 7th edition, pp.188-190). However, the following solution presented itself using two dissimilar Oak wafers. The 2 -pole 5 -way wafer repeats its pattern twice in a revolution. Similar results can be obtained using a 3 -pole 3 -way switch to repeat the $0,1,2,3$ pattern (switching 1 and 2 units) and a 1-pole 11-way switch to bring in 4 and 8 units as required. R. Massey,

Planet Instrument Co. Ltd., Leeds.


Wiring diagram of wafer-swich assembly.

# Wireless World Colour Television Receiver 

## 9. Miscellaneous

Except for a few matters, which will be dealt with in this article, details of the receiver have now been given with but one important omission, the colour decoding circuits. To most people these circuits are probably the most interesting of all for there is nothing like them in an ordinary monochrome receiver. The rest of the apparatus bears a superficial likeness to a black-and-white set, the differences being chiefly those dictated by the use of a colour tube.

It was pointed out at the beginning of these articles that the first essential to the attainment of a good colour picture is a good monochrome one. It is quite hopeless to attempt to get the colour circuits working until the receiver, as so far described, is working well and reliably. The picture obtained should compare well with that given by an ordinary black-and-white set. In a comparison between the two, the monochrome picture of the colour set will normally have a somewhat lower maximum brightness and the horizontal definition may not be quite as good as that of the best black-and-white sets, but may be better than that of the poorer ones. In addition, there may be some colour fringing towards the corners of the picture, due to the impossibility of obtaining perfect convergence.

## Delay line

So far no details have been given of the $0.6 \mu \mathrm{sec}$ delay line in the luminance channel which is shown in Fig. 1, Part 7. The purpose of this is to equalize the transit times of the monochrome and colour components of the complete signal. What may be variously termed the luminance signal, the Y signal, or the monochrome signal passes through equipment with an overall bandwidth of some 4.5 MHz . The chrominance components are picked out of the first video stage and pass to the tube through a separate chrominance channel which has a bandwidth of about 1.2 MHz only. Because of this narrower bandwidth, the chrominance components of the signal take longer to pass through the circuits and reach the tube than do the luminance components. Therefore, a delay line must be inserted in the luminance amplifier to make up for the difference in transit times. Without it, the colour would appear on the tube displaced to the right relative to the picture detail.

Since it is included in the i.f. unit, it would be logical to deal with the construction of the line at this point. However, it is not necessary for a good monochrome picture, and we are regarding it as a special colour part, and shall deal with it in the colour section of these articles.

So far little or nothing has been said about grey-scale adjustments. These are important for a good colour picture; they are also important for a good monochrome picture, but the effect of poor adjustment is usually less noticeable.

We normally think of grey as a kind of colour in its own right. In reality, however, it is not; it is merely white of a low brightness. This is easily seen if we consider a normal mono-


This photograph shows the video and i.f. amplifier boards mounted side by side. Both are hinged for ready access to their other sides and also to permit the convergence magnets to be reached from the front.
chrome tube and gradually increase the grid bias so that a blank raster passes from peak white to black through all depths of grey. There is clearly no change of colour involved (except possibly and undesirably for small changes in the spectral efficiency of the phosphor with excitation). This same result is required with a colour tube, but here white is obtained by combining the effect of three guns exciting red, green, and blue phosphor dots on the screen.

The characteristics of the three guns are unlikely to be identical nor are the relative efficiencies of the three phosphors. As a result, if adjustments are made to produce white at maximum brightness then as the tube bias is increased to give grey the balance between the three colour components may alter and give a colour cast to the grey. Typically, the grey may become brownish, but it might equally well become greenish or bluish. The result on a colour picture, of course, is to make the hue dependent on brightness.

There are five controls which affect grey-scale tracking.

These are the three potentiometers $R_{17}, R_{18}$ and $R_{19}$ (Fig. 1, Part 5) mounted in the convergence unit which control the first-anode voltages of the three guns, and the two potentiometers $R_{1}$ and $R_{2}$ (Fig. 2, Part 2) which control the signal drive to the green and blue cathodes. The first three are adjusted to give a pure grey at a very low brightness level; the second two are then adjusted for a pure white at maximum brightness.

A signal is needed for proper adjustment. If possible this should be from a colour signal generator which provides a set of standard colour bars. In monochrome these reproduce as a set of vertical bars varying in steps from black to white. These bars are broadcast by BBC-2 during the trade test transmissions only and have the drawback that they are not always there when one wants them or for as long as one wants. There is also a grey scale in the Test Card D which is also broadcast and more frequently than the colour bars.

First of all set the brightness and contrast controls so that all steps of the grey scale are clearly visible, but without defocusing on peak white. The red anode voltage control $R_{17}$ should normally be set at little below maximum. Then adjust the green and blue controls $R_{18}$ and $R_{98}$ for a pure grey in the darkest tone value for which changes of colour can be observed. If it happens that either of these comes to its maximum setting, turn down the red control a bit and start again.

Having done this leave these controls alone and adjust $\boldsymbol{R}_{1}$ and $R_{2}$ on the tube base for a pure white on the brightest tone step. The proper white is one which is much less blue than that of the ordinary monochrome tube, one which may appear brownish by comparison but which should not really be brownish at all.

Provided that the anode voltages are adjusted first, adjustment of the drive to the cathodes does not affect the first set of adjustments.

These adjustments provide two-point tracking and some errors may still exist away from these points. These are quite unimportant at very low brightness levels, for the eye there loses its sensitivity to colour, and levels above peak white should never occur. If examination of the grey scale between the levels used for adjustment of the grey scale shows any appreciable colouration, it probably means that the characteristics of the guns of that particular tube differ from each other more than usual. There is not much that can be done about it, but it may prove better to adopt one grey tone brighter for adjustments of the anode voltages. This will reduce the error between the tracking points at the expense of increasing it on the darker tones. As the eye is there less sensitive the net result may be better.

The drive potentiometers $R_{1}$ and $R_{2}$ (Fig. 2, Part 2) have

| Line-Scan Transformer |  |  |  |
| :---: | :---: | :---: | :---: |
| Terminals | $\begin{aligned} & \mathrm{L} \\ & (\mathrm{mH}) \end{aligned}$ | Q | $\begin{aligned} & \mathrm{R} \\ & (s) \end{aligned}$ |
| 3-4 | 5.6 | 50 | 0.3 |
| 4-5 | 1.42 | 41.5 | 0.25 |
| 6-7 | 12.5 | 64 | 0.6 |
| 8-9 | 3.4 | 44 | 0.8 |
| 9-10 | 265 | 49 | 6.5 |
| 10-13 | 45.1 | 68 | 3.2 |
| 13-12 | 69 | 64 | 4 |
| 3-5 | 12.5 | 68 | 0.8 |
| 3-7 | 50 | 74 | 1.4 |
| ( 5 \& 6 joined) |  |  |  |
| 6-12 | 1350 | 60 | 14.9 |
| (7 \& 8 joined) |  |  |  |

on $\div 100$ range.

Fig. 1. Circuit diagram of simple a.f. amplifier. Screened leads for the triode grid and input connections are essential. The transformer ratio must be chosen to give a load of $5.6 k \Omega$ on the pentode
capacitors $C_{1}$ and $C_{2}$ associated with them. These are for the purpose of equalizing the frequency responses of the green and blue drives with the red. If they are not properly adjusted some slight colouration of a vertical edge may appear because the step responses of the three cathode inputs will not be alike. For this to be observable it is, of course, necessary that the vertical edge be in a part of the picture at which the convergence is perfect.

In practice, the effect of these capacitors is quite small and on an ordinary moving picture may not be detectable.

## A.F. amplifier

Turning now to the sound channel, as so far described this terminates in an audio output from a ratio detector in the i.f. unit. An a.f. amplifier and loudspeaker are naturally required additionally. Compared with colour television this is very much bread and butter stuff? It seems unnecessary to give much detail but a suitable circuit is shown in Fig. 1. This can be tucked away in any convenient place, but the loudspeaker magnet must, of course, be kept away from the tube.

In view of past correspondence in Wireless World many readers will doubtless deplore the fact that so little attention is being paid to the sound side and would expect that, at the least, provision would be made for feeding an external highquality sound system. This is, however, a very difficult thing to do and, because we feel that there is this expectation, we shall go into the matter in some detail.

The difficulty arises because, in common with almost all modern television receivers, colour or not, no isolating mains transformer is used. Series-connected heaters must be used for the valves because there are no suitable types available for parallel connection. This in itself does not preclude the use of a mains transformer, of course, but the heater supply is 72 watts. The h.t. supply is of the order of 200 watts, so that the transformer would have a rating approaching 300 watts. This would not be cheap and it would be quite heavy. Much more important, however, is the fact that it would not be at all easy to dispose without its stray field affecting the picture tube. This can be very troublesome with a black-and-white tube; it will obviously be more so with a colour tube, where screening against the effect of the earth's magnetic field is needed!

When there is no isolating transformer the circuits of the receiver are live to the mains. This at once raises difficulties in connecting an external audio amplifier. A high-quality audio amplifier will normally have a mains transformer and so be itself isolated from the mains. It is, therefore, really only practicable to connect an audio output from a set which is not

so isolated via a double-wound audio transformer which is insulated to withstand the full mains voltage.

We have not investigated this, but it is quite likely that there is no suitable component commercially available. In any case, the purist will object to the use of an a.f. transformer in high-quality equipment; we shall not argue whether this is a valid objection.

The use of isolating capacitors will doubtless be suggested. This is not so easy as it sounds because they must pass 50 Hz from the quality point of view and if they do they will hardly be isolating at mains frequency! In addition, there would almost certainly be serious hum problems.

We can, in fact, think of only one way in which an external quality amplifier can be used safely and successfully with a television receiver which is live to the mains. This is to take the output, not at audio, but at the sound i.f. To do this, the sound i.f. amplifier would have to be split at its middle, only the first stage remaining in the receiver, with the second stage and the ratio detector forming an external unit which is mounted close to the a.f. amplifier and which is fed either from this amplifier's h.t. supply or from its own separate supply through a mains transformer.

The problem of insulation is now transferred to an r.f. transformer at 6 MHz and is relatively easy. In fact, the transformer installation could be supplemented by capacitors since these need now pass only radio-frequency.

These are our ideas on the subject. Quite a lot of development would be needed, no doubt, to produce a practicable scheme and we do not intend to do this unless there is a considerable demand for it. This is because we do not think that the use of a separate sound system is usually right for television. We agree entirely that it is desirable to have high-quality sound, but there is a factor which most critics of television sound quality overlook. This is the fact that for natural results the sound must appear to originate at the picture. If the sound comes from a loudspeaker which is separated some distance from the picture the effect is quite unpleasant. In our view high-quality sound from a loudspeaker which is appreciably distant from the picture is far less pleasing than ordinary television sound which seems to come straight from the picture.

It is normally impracticable to build high-quality audio equipment into a television set; the size of the loudspeaker alone prevents that. The nearest that one can get to it is to have the loudspeaker mounted behind the television set to minimize the effect of its displacement from the set. In a normal living room this may well be impracticable.

## Some measurements

Before concluding this article, we give some miscellaneous notes on the complete equipment as so far described. First, we give some measurements made on a Marconi Instruments Universal Bridge type TF868B at 10 kHz for inductance and $Q$ between various terminals of the line output transformer. These refer to a particular sample but they at least give an indication of the order of magnitude which may be useful if one suspects a defective component. Short-circuited turns, for instance, would reduce inductance and particularly $Q$, but because of the tight coupling between the sections, probably all the figures would be considerably affected. Such a fault could happen but is not likely. More probable is a high-voltage flash over between parts of the windings which does not show up at all on such a test but which makes itself very evident in operation.

It can happen that the anode of the PL509 gets red hot. This is a sign of greatly excessive anode dissipation and the valve will be damaged if it is operated for any length of time in this condition. It means that the anode current and/or voltage is
much too high. One cause may be the wrong grid drive waveform. It should be a negative pulse of some 200 V amplitude and some $20 \mu \mathrm{sec}$ duration with a fairly slow exponential recovery.

The dissipation will be too great if for any reason the scan amplitude is much too small; in this case the back e.m.f. on the anode will be too small, so that the anode voltage during the scan will be too high.

Measurements on the line output stage are difficult because there are so few earthy points. Referring to Fig. 1, Part 3, it is useful to disconnect $C_{19}$ and $R_{26}$ from the chassis, together with which ever of $L_{1}$ or terminal 6 of the transformer is normally connected to the chassis. Connect all three to chassis through a $1 \Omega$ resistor. This should be wire-wound and adjusted to $1 \Omega$ within $1 \%$. If desired it can be included permanently. The total mean cathode current can be checked by measuring the voltage across this resistor and is normally 390 mA . The waveform can be checked with an oscilloscope and the peak cathode current measured; this should be 880 mA .

Some indication of the deflector-coil current can be obtained by connecting a $1 \Omega$ resistor between terminal 5 and chassis, the line convergence circuit being disconnected. An oscilloscope connected across it indicates the current in the transformer, which is not necessarily exactly that in the deflector coils, but the two are not very different.

In the field timebase it has been found desirable to connect a capacitor of $0.01+\mathrm{F}$ between chassis and the junction of $R_{13}$ with the output transformer primary. This reduces the linefrequency voltages fed into this circuit and improves interlacing.

## Point-to-Point Review (see opposite)




Monthly figures and smoothed values of the ionosphere index ( $\mathrm{IF}_{2}$ ) and of the number of sunspots for the past 12 years.

# Point-to-Point Review, 1968 

by David Wilkinson,* B.Sc., M.I.E.E.

On the whole, 1968 was a good year for h.f. radio communications. Solar activity was sufficient to ensure maximum usable frequencies (MUFs) high enough for all practical purposes and yet conditions remained fairly stable. The efficiency of most h.f. circuits remained good, a monthly average commercial time of $99.1 \%$ being maintained on three representative circuits received in the U.K.

For a year of peak solar activity, however, 1968 was remarkably uneventful, due to comparatively low values of sunspot number and ionospheric index (IF2). There seems to be some justification in relating the present solar cycle to that of $1878 / 1882$ (minimum /maximum), both of which have been very quiet and followed by a maximum of enhanced activity.

In order to align their figures with those of Zurich Observatory, the Royal Observatory, Greenwich, twice adjusted their sunspot ' $k$ ' factor; from 0.8 to 0.7 in September 1967 and from 0.7 to 0.75 in November 1968. (The Zurich sunspot number is given by $k(f+10 g)$, where $g$ is the number of sunspot groups and $f$ is the number of spots within those groups. $k$ is a factor to allow for variations between observers and telescopes.) These alterations have not been taken into consideration when calculating smoothed values, which has resulted in the peak of solar activity being slightly masked and offset. They do account, however, for the apparent drop in the Greenwich provisional monthly mean sunspot number for the first eleven months of the year from 103.2 in 1967 to 94.8 in 1968. The peak of solar activity is represented more accurately by the smoothed IF2 curve ${ }^{\dagger}$ which indicates that the maximum probably occurred early in 1968. It is interesting to note from the latest figures available at the time of writing, however, that a second upsurge took place during more recent months, resulting in a plateau effect from the end of 1967 to June 1968.

Forty-five sunspot groups of area greater than 500 millionths of the visible solar hemisphere were reported. Seven were over 1000 millionths, the largest of which grew to 3100 millionths (approximately 3600 million square miles) before splitting in two. It was the largest group since 1947, but caused only partial Dellinger fade-outs and little magnetic disturbance during its passage from June 24th to July 6th. One of the most active sunspots was only 400 millionths in area, flares from which resulted in several Dellinger fade-outs and a magnetic disturbance from October 29th to November 4th, which was more severe than any since May 1967.

Dellinger fade-outs totalled 27 and were evenly distributed throughout the year although it is worth noting that none occurred during March, April or May.

Magnetic activity remained at a comparatively low level during the year, the monthly mean " C " value of readings recorded every three hours at Hartland, Devon, being 0.75, which represents a small increase on the 1967 figure of 0.62 .

[^7]The activity was fairly evenly spread and, apart from the above-mentioned storm, those magnetic disturbances which did occur were not severe and had little effect on h.f. circuit performance.

## Satellite communications

There was considerable activity in the field of satellite communications during the year. The first of the Intelsat III satellites was scheduled to be put into orbit on September 18th but, due to a launcher failure, the spacecraft had to be destroyed a few minutes after launch. The subsequent enquiry disclosed that one of the rate gyros on the launch vehicle had failed; there was every reason to expect that the Intelsat III spacecraft was itself perfectly satisfactory. It was fortunately possible to provide coverage of the Mexico Olympic games by employing the NASA experimental satellite ATS III. (The replacement for Intelsat III was successfully launched over the Atlantic Ocean on December 18th). It is intended that additional satellites be launched in 1969, to cover the Pacific and Indian Oceans and also a second over the Atlantic, to cater for the rapidly increasing traffic.

In September, the contract for the Intelsat IV satellites was placed with the Hughes Aircraft Corporation. The satellites should be ready for launch in 1970. A significant part of the later satellites will be built by the British Aircraft Corporation, under a contract worth $£ 2.8$ million. This new generation of satellites will have some twelve times the capacity of Intelsat III.

More earth stations were completed during the year, including those in Moree (Australia), Chile, Mexico, Panama, Spain, Etam and Jamesburg (U.S.A.), Puerto Rico, Thailand and the Philippines. The second paraboloid at Goonhilly was completed; this is now the first earth station to have two large aerials, though second aerials are under construction in Germany and France. The stations in Hong Kong and Bahrain are expected to be complete in mid-1969.

The Piccolo system seems, at last, to have achieved some measure of success, having been commercially produced and installed in the QE2. This system is designed to carry teleprinter channels over noisy h.f. circuits and employs a different tone frequency for each character ( 32 in all). The first patents of this system were taken out, by members of the Diplomatic Wireless Service, in 1957 and, other than a mention in an I.E.E. Convention in 1963, little interest seems to have been shown until the 1968 R.S.G.B. Exhibition $\ddagger$. This seems to be another example of the much-publicised British inability to see the commercial possibilities of an invention.

The year seems to have been relatively quiet in the fields of trunk waveguides and optical communications though those working in these fields might not agree.

The author wishes to thank Mr. B. Priestley for collecting all the data on h.f. performance.

[^8]
## Personalities

Alan L. Gray, B.Sc., A.Inst.I', for the past 20 years with the Plessey organization, has joined Allied Research Laboratories Ltd., of Luton, as chiefengineer. After graduating in physics at Kings College, London, in 1941, Mr. Gray, who is 45, gained a commission in the R.A.F.V.R. and specialized in radar, serving in 60 Group. He joined Plessey's Braxted Research Laboratory in 1948 and later transferred to the Caswell Research Laboratory where he was concerned with the development of ceramic piezoelectric materials and semiconductors. From 1956 until joining Allied Rescarch Laboratories he was with Plessey Nucleonics, latterly as technical manager.

Roderick Mclnnes, Grad.Inst.P., aged 30 , is appointed applications manager of Philbrick-Nexus Research, the Chichester, Sussex, subsidiary of Teledyne Inc., of Los Angeles. He spent twelve years at the Royal Aircraft Establishment, Farnborough, working on the design and construction of instrumentation and control systems for aircraft and missiles. He was latterly, for one year, with Elliott Automation's Space and Guided Weapons Division working on guidance systems.
The City and Guilds of London Institute has recently presented its Insignia Award (C.G.I.A.) to P. Beckley, B.Sc., A.M.I.E.E., A.M.I.E.R.E., aged 32 , a senior physicist with the Steel Company of Wales Lid., Newport, for his thesis "Sensors for automation"; 10 J . Hel szajn, M.I.E.R.E., aged 34, who is reading for a PhD. degree at the University of Leeds, for his thesis " $A$ general characterisation of the three port ferrite junction circulator"; and to E. G. Jarvis, aged 48, an executive engineer with the Post Office Research Laboratory, Backwell, Bristol, for his thesis "Multiaccess satellite repeaters".
P. J. Smith has joined the staff of Anglia Transformers Lid., of Farnham, Surrey, as technical manager. Mr. Smith served his apprenticeship at Foster Transformers with whom he remained for thirteen years before joining Gardners Transformers, where be became chief designer and latterly technical sales engineer.

John Matchett, B.Sc., who is 24 and joined Honeywell Controls Lid. in 1960 as a craft apprentice, has been selected as the company's "apprentice of the year" and his prize is a three-month trip to the United States to Honeywell's industrial division at Fort Washington. He studied for his degree which he gained with first class honours from Glasgow University, under the company's training scheme. He is now an applications engineer in the industrial products group.

The appointment of two new assistant superintendent engineers in the Transmitter Group was recently announced by the B.B.C. G.I.F. Tupper, M.I.E.E., A.M.I.E.R.E., is appointed to the post in Transmitter I department, dealing with v.h.f. and u.h.f. stations and M. Clough, M.I.E.E., in Transmitter II Department, which is concerned with 1.f., m.f. and h.f. stations in the United Kingdom and overseas. Mr. Tupper joined the B.B.C. in 1943 at Droitwich. In 1960 he transferred to the headquarters staff of Transmitter Department where he has latterly been concerned with the general development of the v.h.f. and u.h.f. transmitter network and the extension of colour television. Mr. Clough also joined the Corporation in 1943 and after service at several transmitting stations joined the headquarters staff of Transmitter Department in 1959. Since 1961 Mr. Clough has been head of the site acquisition section.

Kenneth L. Smith, B.Sc., who is well known in British amateur transmitting circles (his call is G3JIX), has been appointed to a senior research associateship in the department of electronics in the University of Kent at Canterbury. He graduated in physics at London University in 1962 and then did three years' research in infrared spectroscopy at the Northern Polytechnic, London. Since 1965 until his recent appointment he was head of the physics department at Holloway School, North London.
E. Trevor Thomas, formerly managing director of Sound Coverage Lid., has joined Leevers-Rich Equipment Lid., of Wandsworth, London, as general manager.

William Logan, a director of Baird and Tatlock (London) Lid., is the new chairman of the Instruments, Electronics and Automation Exhibition, held biennially in London. Mr. Logan retired, with the rank of major, from R.E.M.E. after fifteen years' service (during which he work ed for some time at the Radar Research Establishment at Malvern) and spent 16 years with Avo Lid., latterly as sales director, before joining Baird and Tatlock two years ago. He was president of the Scientific Instrument Manufacturers' Association from 1964 to 1965 and is a member of the Court of the Worshipful Company of Scientific Instrument Makers.

Maurice Cufflin, B.Sc., has inined English Electric Valve Co. Lid., at Chelmsford, as sales engineer. He graduated from Queen Mary College, University of London, with an honours degree in physics in 1935 and joined Marconi, working on communications and navigational equipment for aircraft. In 1949 Mr . Cufflin went to the Marconi Research Laboratories at Great Baddow where he was appointed chief of the measurements research group in 1957, in which work ranged from d.c. to millimetric waves. In 1965 he became engineering manager of Marconi's Automation Division.

John Locke, who has been with Marconi Instruments Lid. since I950, is appointed service representative and will be particularly concerned with customer liaison with the company's repair and calibration service. Before joining M. I. he was for eight years with E. K. Cole Lid., working on the test and calibration of airborne telecommunication and direction-finding equipment.
J. H. W. Costin, B.Sc., A.M.I.E.E., recently ioined Emihus Microcomponents L.id., of Glenrothes, Fifeshire, Scotland, as product sales manager (welding and special assemblies). He graduated in electrical engineering from Kings College, London University, in 1959. He then spent two years with Associated Electrical Industries as a graduate apprentice. From 1962 to 1965 he was with Standard Telephones and Cables at Footscray as a senior process engineer concerned with the manufacture of semiconductors. Mr. Costin was latterly with Mullard as a specialist semiconductor sales engineer.
V. G. Oastler, London area manager of the Marconi International Marine Co., has retired after nearly 40 years' service with the company. He began his career as a sea-going radio officer and then served on the technical staff of several of the company's depots. In 1966 he was appointed manager, London area, with responsibility for co-ordinating the company's installation and maintenance activities, not only in the vast Port of London, but in all the other ports of south east England from Clacton 10 Shoreham.

## New Year Honours

Among the recipients of life peerages conferred by the Queen in the New Year Honours were Professor P. M. S. Blackett, O.M., C.H., president of the Royal Society.

New Knights Bachelor include Morien Bedford Morgan, C.B., controller of guided weapons and electronics, Ministry of Technology, and F. Neil Sutherland, C.B.E., M.A., F.I.E.E., chairman of the Marconi Company and of the Conference of the Electronics Industry.

## C.B.

J. V. Dunworth, C.B.E., M.A., Ph.D., F.I.E.E., director, National Physical Laboratory.
W. Millward, C.B.E., superintending director, Government Communications Headquarters.
C.B.E.
J. Howlett, director, Atlas Computer Laboratory, Science Res. Council.
Col. R. Knowles, F.I.E.R.E., late R.E.M.E.
D. B. Weigall, deputy director of engineering, B.B.C.
O.B.E.

Lt. Col. A. C. Bate, M.B.E., Royal Corps of Signals.
Lt. Col. P. H. Flear, Royal Corps of Signals.
R. M. Billington, T.D., M.Sc.(Eng.), F.I.E.E., inspector of wireless telegraphy, G.P.O.
R. H. W. Burkett, B.Sc., F.I.E.E., managing director, Welwyn Electric Lid.
B. R. Greenhead, director of studio and engineering, Thames Television.
A. C. Heathcote, director, Posts and Telecommunications, Lesotho.
M. D. Mason, M.B.E., superintending electronic/communications engineer, Government Communications Headquarters.
Lt. Col. F. P. Nurdin, sales director, British Communications Corporation.
F. D. Outridge, director, Scientific Instrument Manufacturers Association.
Lt. Col. J. L. Purdon, F.I.E.R.E., Royal Corps of Signals.
Lt. Col. J. J. H. Swallow, B.Sc.(Eng.), F.I.E.E., Royal Corps of Signals.
J. M. J. Whellens, B.Sc., M.I.E.E., principal engineer, G.E.C.-A.E.I. (Electronics).
R. C. G. Williams, Ph.D., B.Sc.(Eng.), F.I.E.E., chief engineer, Philips' Electronic and Associated Industries I.td.

## M.B.E.

G. F. Budden, M.I.E.R.E., assistant engineer-in-charge, operations, North Region B.B.C.

Royal Victorian Medal
Chief Radio Supervisor L. L. Fuller, Royal Navy.

## World of Amateur Radio

## Big Increase in "B" Licences

During recent months the rate of increase in the number of current U.K. Amateur (Sound) "B" licences-which do not require the passing of any Morse test but which restrict operation to frequencies above 144 MHz -has risen very rapidly when compared with the issue of new " $A$ " licences. The " $A$ " licences permit operation on all amateur bands after passing a 12 -word-per-minute Morse test. Both " $A$ " and " $B$ " licences require that applicants should have passed the written Radio Amateurs' Examination. Callsigns in the sequence G8 plus three letters are issued to " B " licensees. In the three months, August 31 to November 30, 1968, "B" licences increased from 1058 to 1327 , a rise of 269 or about $25 \%$. This compared with an increase of 155 in the "A" licences, from 12903 to 13058 or roughly $1.2 \%$. Sound mobile " B " licences are similarly increasing at a rapid rate, the total having risen by almost six times in a year, from 22 to 128. U.K. licences at November 30, 1968, were: Sound "A" 13058; Sound "B" 1327; Sound "A" mobile 2578; Sound "B" mobile 128; Amateur television 186. The position twelve months earlier was: 12597; 693; 2396; 22; and 176. During this period model radio control licences have risen from 12016 to 14978. The marked rise in popularity of the " $B$ " licence follows the concession in spring 1968 when the licence was extended to include operation in the 144 MHz band; it is also similar to the experience of a number of other European countries where it has been shown that non-Morse v.h.f. licences tend to attract a substantial proportion of total applicants. " $B$ " licences thus seem certain to have important repercussions on the future pattern of amateur activity.

## Security Risk ?

British amateurs have been disturbed at what they believe to have been unduly adverse publicity directed at the hobby following the publication of the official Security Commission report (HMSO, Cmnd 3856) on the circumstances surrounding the case of D. R. Britten, the R.A.F. chief technician who pleaded guilty to offences under the Official Secrets Acts. Britten had held the amateur licence G3KFL since the early 1950 s , and admitted to supplying secret
information over a long period to Russian intelligence during personal meetings. In view of the opportunities which amateur radio affords for a hostile intelligence service to talent-spot potential agents and to communicate with them, the Commission suggested that the case justifies a reassessment of security risks attaching to amateur radio activities by members of the armed forces and public service. At the same time, although this received far less press and broadcast comment, the report made it clear that the R.A.F. "see great merit in amateur radio clubs, which have a high interest and training value for many on signals work". The report also shows that the widely reported story of Britten being initially hailed by his amateur callsign by a Russian short-wave listener at the Science Museum is not accepted as true by the Commission. Amateur radio circles believe that most enthusiasts have long appreciated the need for reasonable prudence in communicating with overseas stations, and are alert to any possible attempt to use their privileges for purposes of covert communication. Furthermore they can point to the role of amateur operators in assisting military and special communications during World War II.

## Open Weekend

An "Open Weekend" at the new R.S.G.B. headquarters at 35 Doughty Street, recently attracted some hundreds of amateurs and friends. Three stations, specially installed for the occasion, made nearly 400 contacts. The R.S.G.B. state that consideration is being given to setting up a permanent headquarters station.

## ARRL DX Contest

The 35th ARRL International DX Com-petition-the doyen of the big amateur longdistance contests-is being held this year, as usual, over four weekends, two for telephony operation and two for c.w. Dates and times are: Telephony, February 1, 00.01 g.m.t. to February 2, 23.59 g.m.t. March 1, 00.01 g.m.t. to March 2, 23.59 g.m.t. C.W. February 15, 00.01 g.m.t. to February $16,23.59$ g.m.t.; March $15,00.01$ g.m.t. to March 16, 23.59 g.m.t.

## New Moonbounce Record

A new earth-moon-earth (EME) record for the $1296 \mathrm{MHz}(23 \mathrm{~cm})$ band has been estab-
lished by Peter Blair, G3LTF of Chelmsford who successfully contacted on this mode the Californian amateur station WB6IOM. Blair was using a $15-\mathrm{ft}$ dish aerial and 150 -watt transmitter. WB6IOM had 500 watts of power and a $10-\mathrm{ft}$ dish aerial. Both stations used parametric amplifiers and bandwidth was about 75 Hz with signals just audible above the noise level.

## "Radio News of 1968"

A new R.S.G.B. $16-\mathrm{mm}$ sound film "Radio News of $1968^{\prime \prime}$ was well received at its first showing at the society's a.g.m. in December. This magazine-style film runs for 29 minutes and includes items on amateur radio astronomy, reception of cloud-cover maps from weather satellites on homeconstructed equipment, the GB2LO station at the 1968 City of London Festival, glimpses of amateur-operating during National Field Day, moonbounce equipment used by station G3LTF, and the 1968 R.S.G.B. exhibition. A similar film venture, including shots of British radio amateurs and stations, is currently being undertaken by the American Radio Relay League.

## Australian Youth Radio Clubs' Scheme

The first national conference of the Youth Radio Clubs' scheme of Australia took place recently at the headquarters of the Victorian Division of the Wireless Institute of Australia, Melbourne. The purpose of the conference was to organise the state groups on a national basis, and to ensure uniformity of syllabuses, certificates and examinations throughout Australia. This was successfully achieved and the scheme has been adopted by the Wireless Institute of Australia (national amateur radio society for that country), as part of its educational activities.

## Jamboree-on-the-air

L. F. Jarrell (VE3EWE/G3UXZ / HB9AMS), director of administration at the Boy Scout World Bureau now established at 72 Bd. St. Georges, Geneva, is anxious that amateur radio societies throughout the world shall avoid arranging international contests to clash with future Jamborees-on-the-air. The 1969 event-the 12th-has been fixed for the weekend October 18-19, and it is planned that future J.O.T.A. shall be held during the third full weekend in October each year.

## Amateur Radio in Hungary

The Hungarian Amateur Radio Society (M.R.S.) has made application to join the International Amateur Radio Union. Amateur radio in Hungary is organized on lines similar to those in force elsewhere in Europe, licences are issued by the Ministry of Posts and Transport who also arrange Morse code tests and technical examinations. There are 750 licensed amateurs in the country and the address of the society is Magyar Radio-amator Szovetseg, Budapest VI, Gorkij Fasor 6, Hungary.

## New Products

## Signal Recovery Equipment

The third generation of Brookdeal Instruments signal recovery systems, based on the principle of correlation detection, are devised around their phase-sensitive detector type 411. This instrument has a specified zero stability, linearity and frequency response which are claimed to lead this particular field. Complementary instruments are the type 451 nanovolt pre-amplifiet and type 450 low-noise amplifier. A different type of signal recovery system has recently been introduced by this company for plotting the waveform of signals buried in noise such as are found in experiments on fluorescent decay, radar signals and spin-lattice relaxation time. The system is designated Boxcar Detector type $415 / 425$. It is able to plot waveforms with a resolution of 10 ns . The range of repetition rate is 0.2 Hz to 1 MHz and sampling pulse width 0.5 s to 10 ns . Scan time is $1-2000 \mathrm{~s}$ and zero drift less than $0.01 \%$ deg $\mathrm{C}^{-1}$. Financial backing for expansion has recently been secured by Brookdeal by an agreement with Fulcra Finance Ltd. which has acquired a $33 \%$ interest in the company. Brookdeal Electronics Lid., 2 Myron Place, Lewisham, London, S.E.13.
ww $\mathbf{3 0 5}$ for further details

## I.C. Logics Circuits Check

The presence of " O " or " 1 " logic states and the occurrence of single pulses as short as 30 ns are indicated when checking the performance of logic circuits with a new hand-held probe by Hewlett Packard. Designed for use with i.c. circuits, the probe, type 10525 A , is compatible with t.t.l. and d.t.l. 5 V logic systems. Threshold level is +1.4 V . Above this level an indicator light near the tip of the probe is on; below +1.4 V the light is off. No triggering or threshold adjustments are required and protection against accidental overlead up to 200 V is provided. Input impedance is $10 \mathrm{k} \Omega$ Power for the probe can be obtained from an independent supply or from the power line of the circuit under test via a cable

with adaptors. Consumption is 75 mA at 5 V and operating temperature range $0-55^{\circ} \mathrm{C}$. Price: $£ 43$ 12s. Hewlett Packard Lid., 224 Bath Road, Slough, Bucks.
Ww 312 for further details

## Die-cast Boxes

Four new die-cast boxes by Electronic Services extend their range to nine sizes from $100 \times 75 \times$ 25 mm to $280 \times 180 \times 150 \mathrm{~mm}$. The new larger sizes incorporate the slot guide system used on

the original range. The die-cast construction offers ruggedness with lightness and the screening properties may be valuable in some applications. Electronic Services, Edinburgh Way, Harlow, Essex.
WW 313 for further details

## Thermally Stable Soldering Irons and Long-life Bits

Two temperature-controlled soldering irons, Litestat 50 and Litestat 70 have been introduced to the range of irons manufactured by Light Soldering Developments Lid. Of 50 and 70W loading respectively, they have handles moulded of translucent plastics through which an internal indicator lamp shows when the elements are energized, Temperature stability is within $\pm 2 \frac{1}{2}^{\circ} \mathrm{C}$ during idling and is achieved by a mechanical system operating a micro-switch inside the handle, via a push rod, in response to thermal expansion of the copper element core unit. Close thermal coupling ensures rapid response will.,ut temperature overshoot. Operating temper ir. e is continuously variable, without dismantlins, between approximately 200 and $420 \mathrm{deg} \mathrm{C}. \mathrm{Screw-on}$ copper bits are available in four sizes for each model. Basic prices are: $£ 416 \mathrm{~s}(50 \mathrm{~W})$ and $£ 5$ (70W). A range of Philips iron-coated bits are now

available for Litesold soldering irons giving an estimated life up to 75 times that of copper bits. These new long-life bits are coated with iron to a radial thickness of up to $250 \mu \mathrm{~m}$, covered with a protective coating of nickel, then chromium plated. The coating extends for the full length of the bit but it is not so thick as to impair the heat flow. The bits can be supplied in three standard shapes: chisel (single face), screwdriver and conical. They are not cheap; the price varies from 14 s 6 d for a 10 W version through to $£ 112 \mathrm{~s} 3 \mathrm{~d}$ for a 60W version but this cost is likely to represent a saving in the long term. Light Soldering Developments L.td., 28 Sydenham Road, Croydon, Surrey, CR9 2LL.
WW 306 for further details

## Step Recovery Diode

Reverse recovery transients of diodes, occurring during the brief time following forward conduction, can be used to generate significant amounts of power at microwave frequencies. This is because the very abrupt transient is a strong source of high order harmonics and because high conversion efficiencies can be achieved. The augmented reverse conductivity results from the presence of minority carriers, injected and stored during forward conduction. Diodes which are specially designed to enhance storage and to achieve an abrupt transition from reverse stor-age-conduction to cut-off are called step recovery diodes. A new step recovery microwave diode announced by Mullard has a cut-off frequency of 150 GHz and has been developed for use in high-order frequency multipliers with outputs up to 13 GHz . It will provide output powers in the region of 20 mW at 13 GHz when multiplying by a factor of 8 . When tested, with 500 mW 1 GHz input in a "muluiply by ten circuit", the output power at 10 GHz is 15 mW . The diode, type BXY32, is encapsulated in the standard varactor pill package. Other details are: junction capacitance $\left(V_{R}=6 \mathrm{~V}\right) 0.75 \mathrm{pF}$; transition time (max) 150 ps , life time 50 ns and $V_{R} \max .20 \mathrm{~V}$. Mullard Lid., Torrington Place, London, W.C. 1.
ww 323 for further details

## "Boxcar" Integrator

Recovery of repetitive waveforms with pulses as short as 10 ns from random noise is the function of a signal averaging instrument announced by Nuclear Measurements of Luton. Known as the PAR model 160 Boxcar Integrator, the instrument's applications could be in all areas where repetitive complex signals must be recovered from noise including pulsed laser studies, evoked response in physical and biological research, and absorption and emission spectrophotometry. The Boxcar Integrator recovers repetitive waveforms or any incremental waveform portion by time averaging a small portion of a coherent waveform over a large number of repetitions. This results in an average value of noise which approaches zero,


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WW-112 FOR FURTHER DETAILS

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yielding only the output from the coherent content of the sampled portion. The entire waveform is recovered by scanning the interval of interest. The range of aperture times available is 10 ns to 0.55 s , and variable averaging time constants are from 3 ns to 100 s . A variable bandwidth prefilter and $\mathbf{a} . \mathrm{m}$. limiting circuit are provided. Nuclear Measurements, Dalroad Industrial Estate, Dallow Road, Luton, Bedfordshire.
WW 310 for further details

## Miniature Capacitors

A tubular capacitor for mounting on printed boards has been introduced by Oxley Developments. This capacitor employs polytetra-

fluoroethylene (p.t.f.e.) as the dielectric medium. Uniformly smooth adjustment and linear reversalfree tuning is claimed. Capacitance value is 2 pF minimum, swing is 5 pF and power factor is better than 0.0005 at 10 kHz . Temperature coefficient is 50 p.p.m. per deg $C$ and insulation resistance $>16^{6} \mathrm{M} \Omega$. Oxley Developments Co. Lad., Priory Park, Ulverston, North Lancashire.
WW 307 for further details

## Mains Fuseholders

The irritation of Fuseholders which turn in the panel, twisting the connecting leads, or fuses which defy all attempts to dislodge them from the holder for replacement, has been remedied by Bulgin in three versions of a new panel mounting design to accept $\frac{1}{4} \mathrm{in}$. diameter fuses. A lug prevents the unit turning in the panel, and the fuse is withdrawn with the fuse cap; which is screw fitting. When the fuse is withdrawn the rear (live) contact cannot be reached by the B.S. test finger. Maximum ratings are 15 A at 250 V or 20 A at 32 V and the panel hole required is 15.9 mm plus a keyway, 2.4 mm wide and 1.2 mm deep.


Maximum panel thickness is 5.5 mm . A. F. Bulgin \& Co. Ltd., Bye-pass Road, Barking, Essex.
WW 311 for further details

## Radiation Thermometer

Temperature measurements of sea surface, clouds, terrain or any large-scale subjects are made possible by an American radiation thermometer which employs non-contact infrared techniques. The instrument takes the form of a light-weight battery-powered infrared radiometer, type PRT-5, which can be used from aircraft, ships or other platforms. Temperature measurements can be made in any selected range between $-50^{\circ} \mathrm{C}$ and $+150^{\circ} \mathrm{C}$. The standard model comprises an optical head and a separate solid-state control unit. The optical head may be hand-held or tripod-mounted. It has a $2^{\circ}$ field of view
restricted to the $8-14 \mu \mathrm{~m}$ spectral region. Measurements from $-20^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$ are made on three overlapping meter scales. Wavelength region, meter scale and field of view can be modified to suit special measurement requirements. Measurements are independent of distance, provided the target fills the instrument's field of view. Temperature sensitivity is $0.05^{\circ} \mathrm{C}$ above zero and $0.1^{\circ} \mathrm{C}$ below zero, and accuracy is $\pm 0.5^{\circ} \mathrm{C}$. A recorder output is available, of 1 V d.c. at $10 \mathrm{k} \Omega$ or 50 mV d.c. at $500 \Omega$ impedance. The internal batteries can be re-charged from 115 V or 230 V $50-400 \mathrm{~Hz}$ mains supplies. Barnes Engineering Co., Stamford, Connecticut, U.S.A. U.K. enquiries to: B \& K Publicity Division, 59 Union Street, London, S.E. 1 .
WW 316 for further details

## Voltage Regulators

Three silicon monolithic voltage regulators designed for a regulated output at -15 V contain their own voltage reference element for voltage regulation to $0.1 \%$. They are types 2103, 2104 and 2105 by Philbrick-Nexus. Internal shortcircuit protection is provided and a thermal feedback circuit prevents excessive operating temperature. Type 2103 features an electrical zero adjustment for output voltage in the range -13 to
-17 V and remote sensing is optional for regulation with respect to voltage at the load instead of at the regulator output pin. Type 2104 provides voltage sensing at the output pin with a preset regulator output. The 2103 and 2104 are housed in a TO-99 case and have a power dissipation of 300 mW in free air; one watt with heat sink. Type 2105 also dissipates 300 mW and, with only three terminal leads, provides easy installation. It is housed in a TO-105 non-conductive case. Vol-tage-divider resistors are built into all three regulators and need not be provided by the user. Prices for small quantities are: $2103 £ 6$ 2s 6d, $2104 £ 4$ 15s and 2105 \& 10 s . Philbrick-Nexus Research, 81 a North Street, Chichester, Sussex.
WW 315 for further details

## Portable R/T

Compact and light-weight is the claim made by Interplanetric for a new $450-490 \mathrm{MHz}$ band transmitter/receiver designed to be worn by the user. The complete unit weighs 2 kg and comprises a transmitter, receiver, switchblock, microphone, earphone, batteries and either a quarter-wave or half-wave aerial. It is suspended on the person by an adjustable harness. The system is crystal-controlled and provides good line-of-sight short-range communication. Transmitter power output is 100 mW and receiver audio output power 5 mW . Modulation is $\mathrm{f} . \mathrm{m}$., single channel working, with spurious radiation 25 dB down from the power output level. A carrier-operated squelch circuit mutes the audio output during "no signal" periods. The equìpment is powered by mercury cell batteries located in the transmitter and receiver casings. Interplanetric, 39-49 Cowleaze Road, Kingston on Thames, Surrey.
WW 343 for further details

## Motor-potentiometer

A motor-potentiometer with principal applications in the automatic adjustment of potentiometer circuits and the production of very long time constants for feedback in control loops has been announced by U.C.E. The unit comprises a motor driving a low torque potentiometer through a compact gear train. Motors are of three types, a measuring motor, a stepping motor and a reversible synchronous motor. Gear ratios available are up to $1: 500,000$ and units with six-speed gearboxes with ratios of up to $1: 500$ for use with low

torque potentiometers can be supplied. Special features include wiper reset, limit switches, wired self-balancing potentiometer systems and wired servo systems for position transmission or computing circuits. Universal Control Equipment Lid., 38 London Road, Stroud, Gloucestershire.
WW 308 for further details

## X-band 2W Varactor

The upper frequency limit of Bimode diodes by the Bomac division of Varian has been extended from 8 GHz to 10 GHz by the introduction of a new type, the VAB-824A. Output power is quoted as being typically 2.5 W in a $50 \%$ efficient tripler circuit with measurements being taken in the mid-point of the $7-10 \mathrm{GHz}$ range. Varian Associates Lid., Russell House, Molesey Road, Walton-on-Thames, Surrey.
WW 314 for further details

## 70W Amplifier

A new solid-state power amplifier from Bradmatic provides 70 W output into $4 \Omega$ with total harmonic distortion of $0.25 \%$ at full output. Input sensitivity is 850 mV into $10 \mathrm{k} \Omega$ for 70 W output and frequency response is $20 \mathrm{~Hz}-20 \mathrm{kHz}$. The amplifier, type SSP2, will operate from $220-250 \mathrm{~V} 50 \mathrm{~Hz}$ or $105-115 \mathrm{~V} 6 \mathrm{~Hz}$ mains input and has a power consumption of 130 VA at full output. A 28 V 100 mA unsmoothed output is available to power auxiliary equipment. The output stage is openand short-circuit proof. Bradmatic Lid., 338 Aldridge Road, Streetly, Sutton Coldfield, W'arwickshire.
WW 326 for further details


# A Folded Exponential Horn Loudspeaker 

## A design for a bass speaker, and details for building a full-range system

Abstract of an article by J. Jecklin*



Fig. 1. Side view of half the horn.

Fig. 2. Perspective view of half the horn.


If a horn with a low frequency cut off of $40-50 \mathrm{~Hz}$ is to be used in a normal room it must, for aesthetic and space-requirement reasons, be folded. Unfortunately, when a horn is folded, its high frequency response deteriorates, but this can be overcome by a properly chosen combination of bass and medium/high frequency loudspeakers. Moreover, the radiation resistance of a folded horn is not independent of frequency, but rises as the latter is increased. This again is obviated in the system described here* by inserting a flared matching section between the hom proper and an acoustic low-pass chamber situated immediately in front of the loudspeaker unit.

The construction of the horn is based on an ingenious suggestion made by Klipsch in 1941. It is divided throughout its entire length into two symmetrical halves. The outer sides of the final section are constituted by the walls forming the rectangular corner of the room in which the horn is installed. The detailed construction is shown, with dimensions in mm, in Figs. 1-4, it being assumed that $\frac{1}{2}$ in. thick plywood or blockboard is used throughout. The dimensions are determined as explained by the author in Funk-technik No. 16 (1967) pp. 591-2.

An even low frequency response is obtained if the stiffness of the air mass in the horn is compensated by the stiffness of the air cushion in the closed housing and a loudspeaker with a very freely suspended diaphragm and low self-resonant frequency is employed.

Among various makes of loudspeakers available, the author has found as a result of listening tests that the Wharfedale W15RS, which proved highly reliable by virtue of its rigid construction, was particularly suitable for the bass section of the system. It is mounted within the front end of the horn assembly behind the rear wall of the closed housing as shown in Figs. 1-4.

For the medium and high frequency section, electrostatic, pressure chamber and cone-type loudspeakers were considered in turn. The first type was rejected owing to its low conversion efficiency and its relatively high bass cut off frequency, the second on the score of its high cost and not detectably superior performance to the cone type. Of

[^9]

Fig. 3. Section E-H of Fig. 1.


Fig. 4. Section I-J of Fig. 1.

Fig. 5. Crossover circuit for full-range loudspeaker system.



Fig. 6. Frequency response of the full-range system.
the various makes available the author found the Goodman's Axiom 80, which has a double cone and a frequency range of 20 $\mathrm{Hz}-20 \mathrm{kHz} \pm 4 \mathrm{~dB}$ the most suitable, mainly on account of its unusual cone suspension system. Owing to the narrowing of the radiated beam at frequencies above about 8 kHz , two such loudspeakers must be used in order to fill the $90^{\circ}$ angle formed by the walls in the corner of the room. Many tests were made to ascertain the optimal position for these loudspeakers. It was found that for stereo reproduction in normal rooms they are best mounted in the saddle of the horn, radiating at an angle of $20^{\circ}$ upwards and $30^{\circ}$ outwards, whereas for mono in small rooms angles of $20^{\circ}$ upwards and $15^{\circ}$ outwards are preferable.

For the combined system, the calculated cross-over frequency comes out at 350 Hz , the network and its method of connection being shown, with numerical values of the components, in Fig. 5. The overall frequency response of the complete system was measured in a low-reverberation resonancefree room using a narrow-band noise source and a previously calibrated condenser microphone placed in various positions. The mean of eight measurements is shown in Fig. 6.

Although the response curve is not absolutely level at very low frequencies, the unevenness is due to the horn being folded, and not to any artificial resonances which are virtually unavoidable with cabinets and bass reflex housings, and moreover, as compared with the latter, variations in the furnishings of the room have very litule effect on the effective frequency response of the horn system.

The efficiency of conversion of electrical into acoustic energy of the horn described is about $20 \%$, about ten times better than that of a compact box system. The maximum permissable loading is 12 W .

Although exponential horns are now out of fashion, both theoretical considerations and the practical results obtained with the system described here clearly show that they can offer unsurpassable performance, particularly in the bass range. The combination described is eminently suitable for the most exacting professional requirements.

## 'High Performance A.G.C.'

It has been pointed out that the collector of $T r_{1}$ in Fig. 6 (page 18, Jan. 1969) should go to the base of $\mathrm{Tr}_{2}$ as shown in Fig. 2.

## February Meetings

Tickets are required for some meetings: readers are advised, therefore, to communicate with the society concerned

## LONDON

4th. Inst. Electronics-"Practical aspects of acoustic insulation" by R. G. Monk at 18.45 at London School of Hygiene, Keppel St., W.C. 1.
Sth. Inst. Railway Sig. Eng.-"Radar for the railways" by Prof. H. M. Barlow at 18.00 at the I.E.E., Savoy Pl., W.C. 2.

Sth. B.K.S.T.S.-"The importance of colour separation accuracy for films and TV" by B. J. Rogers at 19.30 at Royal Overseas League, St. James's St., S.W. 1.

6th. I.E.R.E.-"Radar echo signatures from birds, insects and bats" by Dr. Glen W. Schaefer at 19.00 at 9 Bedford Sq., W.C. 1 .

7th. R. Instn.-"The hologram" by Dr. D. Gabor at 21.00 at 21 Albemarle St., W. 1 .

13th. I.P.P.S.-Symposium on "The preparation of thin films by the method of sputtering" at 10.00 at the I.E.E., Savoy PI., W.C. 2.

13th. I.E.R.E.-"M.O.S.T. arrays" by Dr. R. C. Foss at 18.00 at 9 Bedford Sq., W.C. 1 .

18th. I.E.R.E.-"Instrumentation and control in float glass manufacture" by G. P. Rigby and M. Hilton at 18.00 at 9 Bedford Sq., W.C. 1 .

18th. Radar \& Electronics Assoc.-"The use of radar in meteorology" at 19.00 at B.I.C.C. Lid., 21 Bloomsbury St., W.C. 1 .

19th. I.E.R.E.-"New approach to h.f. receiver design" by E. T. Wilson at 18.00 at 9 Bedford Sq., w.C. 1 .

20th. R.T.S.-"The process of learning-the future role of educational technology in highet education" by D. E. P. Jenkins at 19.00 at the I.T.A., 70 Brompton Rd., S.W. 3.
26th. I.E.R.E. \& I.E.E.-Colloquium on "Fail-safe techniques for high reliability computer equipment" at 10.30 at the Middx. Hospital Medical School, Cleveland St., W.1.
27th. R.T.S.-"The Yorkshire Television Centre at Leeds" by P. G. Parker and D. J. Whittle at 19.00 at the I.T.A., 70 Brompton Rd., S.W'3.
28th. R. Instn.-"Physics and music" by Prof. C. A. Taylor at 21.00 at 21 Albemarle St., W: 1

## BASINGSTOKE

13th. I.E.R.E.-"Management for engineers" by T. G. Clark at 19.30 at the Technical College.

## BELFAST

25th. I.E.R.E.-"Engineers must learn to manage" by Harley R. Sykes at 18.30 at the Ashby Inst., Queen's University, Stranmillis Rd.

## BIRMINGHAM

11th. Soc. Environmental Engrs.-"Application of random signal analysis to engineering problems" by Dr. C. Ashley at 19.30 at the University.

## BOURNEMOUTH

Sth. I.E.E.-"Post Office Tower" by D. C. Jones at 18.30 at the College of Technology.

## BRIGHTON

18th. I.E.R.E.-"Electronic organs and associated equipment" by L. F. Hawkes at 18.30 at the College of Technology.

## BRISTOL

19th. I.E.R.E. \& I.E.E.-"Active filters" by F. E. J. Girling and E. F. Good at 19.00 at the University.

## CAMBRIDGE

4th. I.E.R.E. \& I.E.E.-"Ergonomics in electronic equipment and system design" by B. Shackel at 20.00 at the University Eng'g Labs., Trumpington St.

## CARDIFF

12th. I.E.R.E.-"Long-range radio communication" by Dr. G. L. Grisdale at 18.30 at the University of W'ales Inst. of Science \& Technology
14th. R.T.S.-"Stcreophonic broadcasting" by R. S. C. Gundry at 19.00 at Broadcasting House, Llandaff.

## CHELTENHAM

18th. I.E.R.E.- "Computer aided circuit design" by E. Wolfendale at 19.00 at the Government Communications Headquarters, Oakley.

## EDINBURGH

11th. I.E.R.E. \& I.E.E.-"Electronic signal processing in hospitals" by Dr. J. M. M. Neilson at 18.00 at the Carlton Hotel, North Bridge.

## GLASGOW

10th. I.E.R.E. \& I.E.E.-"Electronic signal processing in hospitals" by Dr. J. M. M. Neilson at 18.00 at the University of Strathclyde.

## LIVERPOOL

19th. I.E.R.E.-"Electrical and electronic devices associated with railway signalling" by F. Bowyer at 19.00 at the University.

## LOUGHBOROUGH

11th I.E.R.E. \& I.E.E.-"A, flexible modular data handling system for satellite use, with particular refer ence to the Black Arrow programme" by E. K. Crampton at 18.30 at the University of Technology.

## NEWCASTLE-UPON-TYNE

12th. I.E.R.E.- "Moire fringe digitization of linear and circular movement" by C. N. 'W'. Reece at 18.00 at the Inst. of Mining \& Mech. Engrs., Westgate Rd.

## NEWPORT, ISLE OF WIGHT

7th. I.E.R.E_- "Yacht electronics" by Major R. N. B. Gatehouse at 19.00 at the Technical College.

## PLYMOUTH

18th. I.E.R.E. \& I.E.E.-"Application of microelectronics" by Dr. S. S. Forte at 19.00 at the College of Technology.

## READING

18th. I.E.R.E.-"Positional transducers and precision electronic measurement" by P. Wolfendale at 19.30 at the J. J. Thomson Physical Lab., the University.

## RUGELEY

6th. I.E.R.E., I.E.E. \& I.P.O.E.E.-"The design of high-quality transistor power amplifiers" by Dr. A. R. Bailey at 19.00 at Shrewsbury Arms, Market St.

## SHRTVENHAM

4th. I.E.R.E. \& I.E.E.-"Modern methods of traffic control" by D. G. Hornby at 18.15 at the Royal Military College of Science.

## SWANSEA

12th. I.E.E.T.E.-"Medical electronics" by R. G. Wood at 19.30 at the Applied Sciences Bldg., University College, Singleton Park.

## Literature Received

"Catalogue of Electronic Components and Equipment" (2nd edition) from G. W. Smith is now available at 7s 6 d . About half of the 190 pages are devoted to audio and "Ham" equipment. The rest of the catalogue lists a wide range of electronic components including some that are not to be found easily elsewhere. G. W. Smith \& Co. (Radio) Lid, $3 / 34$ Lisle St, L.ondon W.C. 2 .

Brief details of a range of relays are given in a pocket guide we have received from B \& R Relays Lid, Temple Fields, Harlow, Essex. Small conventional relays, heavy contactors and reed relays are included.

## WW 401 for further details

Industrial sound equipment, amplifiers, record players, tape recorders, microphones and loudspeakers manufactured by Magneta (B.V.C.). I.td, Ackmar Works, Parsons Green Lane, Fulham, London S.W.6, are described in a series of leaflets.
WW 402 for further details
A bulletin describing a high-speed transistor tester (type T217) for classifying semiconductors on the production line or for incoming component testing has just been released. The instrument is an improved version of the T207 with increased range. Teradyne Lid, 12 Swallow St, London W. 1.
WW403 for further details
"Concise catalogue of Industrial Instrumentation" contains details of a wide variety of tachometers, counters, electronic relays, frequency sensitive switches, position indicators, temperature measuring equipment and hours-gone indicators. It is produced by Smiths Industries Lid, Industrial Instrument Division, Kelvin House, Wembley Park, Middlesex.
WW 404 for further details
"Colour Television Training Courses by E.C.T.T." contains 12-pages which are devoted to details of courses covering various aspects of colour television training from servicing to sales. E.C.T.T. Lid., 45 Walton Rd, East Molesey, Surrey.
WW 405 for further details
The range of lasers and associated equipment available from Laser Associates Lid, 172 Bradford Kd, Slough, Bucks, is briefly outlined in a leaflet. A description of the company is also given.
WW 406 for further details
Switching circuits using a combination of Triacs and gas discharge tubes are presented in a leaflet from Cerberus Lid, CH-8708 Mannedof, Switzerland.
WW 407 for further details
The latest catalogue from Ariel Pressings pictures a range of pressed components, switches and connectors. Ariel Pressings Lid, Wollaton Rd, Beeston, Nollingham.
WW 408 for further details
A folding chart, for pocket or wall use, listing a range of Motorola semiconductors held in stock by Celdis Lid, $43 / 45$ Mifford Rd, Reading, Berks, is now available.
WW 409 for further details
Two new pamphlets from Mullard are (1) "Introducing Silicon Planar Transistors" and (2) "Simple Mptor Speed Control using a Thyristor". Both pamphlets orginate from the Mullard Educational Service, Mullard House, Torrington Place, London W.C. 1
(1) WW 417 for further details
(2) WW 418 for further details

A good deal of information on the types of microphone available and advise on their use is contained in a seven-and-half page quarto typescript we have received from Allbeury Coombs \& Partners. The report, which is called "Microphone Techniques and Applications", is "written round" Beyer microphones. Allbeury Coombs \& Partners, 29 Adam \& Eve Mews, London W. 8.

## WW $\mathbf{4 1 1}$ for further details

"Aerosol Aids to Industry" is the title of a small booklet listing such items as anti-oxidant, anti-corrosion, de-greasants, insulants, coolants, anti-static and colloidal graphite compounds in aerosol cans. These items are distributed by Special Products Distributors Lid, 81 Piccadilly, London W.1.

## WW $\mathbf{4 1 2}$ for further details

Modular power supplies for digital integrated circuits are described in a brochure received from Oltronix. A typical unit, type MB5-10, provides outputs at up to 10 A from 4.5 to 5.5 V . A $10 \%$ change in mains voltage results in about a $0.01 \%$ change in output. Oltronix, 99 Bancroft, Hitchin, Herts.
WW 413 for further details
A picoammeter that carries out measurements in the range of $10-2$ to 3 x $10^{-14} \mathrm{~A}$ is described in a leaflet from Keithley Instruments Inc., 28775 Aurora Rd, Cleveland Road, Ohio, U.S.A.
WW 414 for further details

## H. F. Predictions-February

MUFs are forecast by the two control point method. Thus, as regards reception in the U.K. from stations beyond about $60^{\circ}$ longitude and in the same general direction, MUFs will be the same after sunset for westerly, before sunrise for easterly, routes, duration of the effect increasing with distance. This means that the Montreal and Buenos Aires charts are a good guide to evening reception from the whole of North and South America. An ionospheric index IF2 of 104 has been used to calculate MUFs and a sunspot number of 97 for the LUFs. There is a similarity between recent solar cycles and those of 169 years previously which suggests that the smoothed sunspot number will not exceed 100 for the next 30 years and will be below 50 for two thirds of this period. There is no theoretical support to this observation, so its empirical nature must be borne in mind when considering such long-term predictions.



```
MMedian standard MUF
----- Optimum traffic frequency
-.--- Lowest usable HF
```


# Test Your Knowledge 

Series devised by L. Ibbotson*, B.Sc., A.Inst.P., M.I.E.E., M.I.E.R.E.

## 9. Aerials

1. An aerial which will radiate a coherent electromagnetic wave with the same power flux in all directions (an isotropic radiator) is
(a) theoretically impossible
(b) theoretically possible but has never been made
(c) available but never used because of its low gain
(d) used in space communications.
2. A transmitting aerial has a gain of 3 dB . This means that:
(a) it radiates twice as much power as is fed into it from the transmitter
(b) it produces in all directions twice the power flux that would be produced by a reference aerial fed with the same transmitter power at the same distance
(c) in some direction it produces twice the power flux that would be produced by a reference aerial fed with the same transmitter power at the same distance
(d) it radiates all its energy towards the receiver thus causing the received power to have twice the value it would have if energy were radiated equally in all directions.
3. The equivalent absorbing area (effective aperture) of a receiving aerial when compared to the physical area presented by the aerial perpendicular to the direction of propagation of the received wave:
(a) is the same
(b) is always greater
(c) is always less
(d) is in some cases greater, in others less.
4. An aerial when used to transmit has a highly directive polar diagram. When used to receive its polar diagram is:
(a) the same
(b) more directive
(c) less directive
(d) shows no directivity.
5. The radiation resistance of a centre-fed halfwave dipole is of the order of:
(a) 130 hms
(b) 70 ohms
(c) 600 ohms
(d) 1 megohm.
6. From four properties of a half-wave dipole given below select one which does not depend on its thickness:
(a) the radiation resistance
(b) the resonant length
(c) the bandwidth
(d) the gain.
7. A centre-fed dipole which is very much shorter than half a wavelength at the frequency to be radiated has an input impedance which is:
(a) purely resistive with a value much less than that at half-wave resonance
(b) purely resistive with a value much greater than that at half-wave resonance
(c) highly capacitive
(d) highly inductive.
8. A narrow rectangular slot in an infinite conducting sheet when fed between the centres of opposite sides of the slot forms an aerial with similar radiating properties to those of a centre-fed dipole which would just fill the slot. The two differ, however, in certain respects. Select from the four properties below one in which they differ:
(a) the half-wave resonant frequency
(b) the gain
(c) the polar diagram
(d) the plane of polarization.
9. A broadside array of dipoles, fed in phase, is to be built to produce a narrow beam of radiation in a direction perpendicular to the array, the aperture being given. The optimum separation between dipoles is:
(a) as close together as possible
(b) a quarter wavelength
(c) a half wavelength
(d) a wavelength.
10. The driven element of a Yagi array is frequently a folded half-wave dipole. The advantage of using a folded dipole rather than a simple dipole for this purpose is that
(a) it has an input impedance which is four times as great, so that it is much easier to match to the feeder
(b) it is much shorter and requires much shorter parasitic elements thus producing a smaller aerial
(c) it has better directional properties which improve the gain of the array
(d) the amplitude of side-lobes is reduced.
11. A vertical unipole, used to transmit a low frequency signal, usually has a network of buried wires connected to the earth terminal
beneath it so as to improve the conductivity of the earth. This is necessary
(a) to reduce the reactive component of the input impedance
(b) to increase the effective height
(c) to reduce the dissipation of transmitter energy in heat
(d) to increase the proportion of the radiated energy radiated in the ground wave.
12. A common aerial for receiving long-wave and medium-wave (vertically polarized) signals consists of a vertical wire a few metres high with a horizontal wire connected to the upper end. The purpose of the horizontal wire is:
(a) to give capacitance to earth thus increasing the effective height of the vertical section
(b) to receive any incoming signal which may have become horizontally polarized
(c) to make the aerial terminal impedance resistive
(d) to increase the gain of the aerial by making it directive.
13. At high frequencies directional aerial arrays, consisting of a number of coplanar half-wave dipoles mounted in front of a reflecting screen, are used. The elements are generally end-fed. They are fed in this way because:
(a) the input impedance is high so that many elements can be connected in parallel with the same feeder
(b) connecting the feeders to the ends of the dipoles causes least disturbance to the polar diagram
(c) this form of feeding reduces the effect of earth reflections on the radiation pattern (d) end-feeding reduces the effect of mutual impedance between the elements.
14. A rhombic aerial used for transmission is mounted with the plane containing its conductors horizontal. The radiation which it produces is:
(a) entirely horizontally polarized
(b) entirely vertically polarized
(c) horizontally polarized in the main lobe but not in all the side lobes (d) vertically polarized in the main lobe but not in all the side lobes.
15. Which of the following is normally used to transmit a circularly polarized wave?
(a) a dielectric rod aerial
(b) a helical aerial
(c) a lens-corrected horn aerial
(d) a Franklin aerial.
16. A uniformly illuminated aperture (an ideal radiator to which a number of high frequency aerial systems approximate) can produce a narrow beam of radiation. Under conditions in which it does so the beam width of the main lobe in a given plane containing the axis of the main lobe is:
(a) directly proportional to the width of the aperture in that plane
(b) inversely proportional to the width of the aperture in that plane
(c) directly proportional to the area of the aperture
(d) inversely proportional to the area of the aperture.
Answers and Comments page 95

## Letter from America

A recent press conference in a small office at Troy, Michigan, not only stirred up electronic circles but created a flurry of interest in Wall Street. The office in question belonged to one Stamford Ovshinsky of Energy Conversion Devices. The meeting was called to announce the publication of an article by the American Physical Society entitled "A description of an a.c. switch made of homogenous film containing no rectifying elements". Reporters were told about a new glassy compound which could supersede transistors in many areas and there was much talk of flat TV sets that could be. hung on the wall and so on. All this was very exciting to the non-technical press and the New York Times had large headlines saying "Glassy Electronic Device May Surpass Transistors" and the solemn Wall Street Journal devoted several columns to the Ovshinsky devices saying "Electron microcircuits using semiconductors are the basis of modern computers and the miniatured circuits used in transistorised radios, TV sets and other electronic machines. Mr Ovshinsky has developed an electronic switch with new materials. A microscopically thin layer of the material separates two electrodes which otherwise carry a current. The material blocks off the current until the force of the current hits a specific voltage. Then, in less than 50 trillionths of a second the material becomes a conductor . . . the material continues to conduct the current until the voltage drops. . . . What is exciting the physicists is that Mr. Ovshinsky's materials aren't crystalline in the atomic structure as all other semiconductors are. In a crystal the atoms are in an extremely evenly spaced geometric orderly arrangement. The Ovshinsky materials, however, consist of a variety of different atoms linked together in a disorderly fashion, a structure known to scientists as an 'amorphous' material. They are actually glasses composed of a mixture of tellurium, arsenic, silicon, germanium and other elements. By contrast, conventional semiconductors are crystalline forms of elements such as silicon and germanium. Ovshinsky materials are made fairly cheaply and easily. Common chemicals are weighed on a simple scale, mixed and then placed in a small furnace. There they are heated to 1000 deg for 24 hours until they fuse into a chunk of gray opaque glass . . . a small chip of this glass can be evaporated in a vacuum and laid down as a microscopic film.

Or the material can be simply painted on. One advantage of the Ovshinsky devices is that they can operate on alternating current. Conventional semiconductors can pass an electric current in only one direction but the new semiconductors work regardless of which direction the current is flowing. Reporters were shown a memory switch made of the new materials. This memory switch, as with the threshold switch, is a non-conductor until the current flowing through hits a certain voltage; it then becomes a conductor. However, if the current is suddenly turned off the device doesn't revert to a nonconductor. Instead, it will stay in its conducting state indefinitely until it is hit with an electric pulse. An array of these devices, they explained, could permanently store the on-off pattern as long as needed. . . . Thus, they suggested, computer memories could be mailed across the country for example.
Mr. Ovshinsky envisioned a picture frame sized TV set. Essentially, it would consist of an array of tiny dots of the new semiconductor behind a coating of phosphors that coat conventional TV screens. These would be sandwiched between a grid of strips of conducting materials in the rear and transparent conductors in the front. As the electronic TV signals swept through the grid of the conductors in the rear it would cause the tiny semiconductors either to block or pass the current. Where it would pass the current the phosphors would glow creating the TV image. . . . Non-exclusive licences for the new materials have already been granted to International Telephone \& Telegraph, L. M. Ericsson of Sweden and Danfoss of Denmark."

Publicity of this kind caused a tremendous rise in E.C.D. shares, but, the reaction from the industry was, to say the least, a little cautious. It was pointed out that these devices were not really new and that the first article appeared in Control as far back as April 1964. This was entitled "The threshold switch, a new component for a.c. control". Another article was published by Electronics in September 1966 and at that time Mr. Ovshinsky said his invention would shake the industry. At the moment the industry is still unshaken. RCA say they have tested the devices and Dr. Webster of the Princeton laboratories says "we have found no significant use for them", Raytheon are said to have found them erratic and Texas Instruments reported some instability problems and expressed doubt that they could be
produced in quantities. Dr. Sparks of Bell Telephone said "There are relatively few applications for two-terminal devices, also the materials are more complex than crystalline structures and are less understood". He went on to say "they are not easily controlled and in our experience have exhibited considerable jitter in switching characteristics". Bell have patents for amorphous devices dating back to 1961, but, they stated recently that these devices do not warrant a major development programme. On the other hand, Sir Neville Mott, director of the Cavendish Laboratories, said "the discovery of the Ov shinsky effect is the newest, biggest and most exciting discovery in solid state physics at the moment". Mr. Ovshinsky himself said he is delighted by all the publicity and, undeterred by all the controversy, his company is busily making some 150,000 devices a day.

At the recent Audio Engineering Society Convention (held in the Park-Sheraton Hotel, Washington, instead of the Barbizon Plaza because of the heavy demand for space in the exhibition section) Ampex were demonstrating a new high-speed cassette stereo tape duplicating unit which featured remote control operation, plug-in heads, separate recording of each instrument and up to 24 channels. Tape-Athon Corporation were showing their new tape recorder intended for station monitoring. At a tape speed of $11 / 32 \mathrm{in}$. per second it will operate over 400 hours using a $10 \frac{1}{2} \mathrm{in}$. reel of tripleplay tape! Signal to noise is given as 38 dB and a response of $\pm 3 \mathrm{~dB}$ from 200 to 3000 Hz . Not hi-fi, of course, but adequate for its purpose. Ray Dolby was demonstrating his system (this time he was using Lansing speakers); and again I was impressed with the remarkable improvement in signal-noise. One of the most interesting items was the Model 1925 Multifilter by General Radio. This new unit contains 30 channels of parallel octave band or one-third octave band filters in the frequency range from 3.15 Hz to 80 kHz . There are several options of frequency range and bandwidth; it can also be supplied without attenuators. It was demonstrated as a spectrum shaper showing an indication of the curve on the scale which is calibrated at 5 in per decade. Vertical range is 10 dB per inch and each attenuator has a 1 dB per step resolution with a control of 50 dB .
America is a country of contrasts, a land of extremes where you can see great wealth next to appalling poverty and the most selfish money grabbing side by side with incredible generosity. Huge corporations can be soul destroying, yet many have a higher regard for individual freedom and civic responsibitity than similar concerns almost anywhere ir the world. I am used to these contradictions so I was not surprised when I read about Lockheed Project LEND. The letters stanc for Lockheed Engineers for National Deploy ment and it is a programme which estab lishes a reserve of experienced engineers who are lent for specific periods to other com panies who have need of their particula talents. So far the company has lent ou more than 100 engineers to companies sucl as Philco, Chrysler, LTV and Genera Dynamics.
G. W. TILLETT

## Answers to "Test Your Knowledge" -9

## Questions on page 93

## 1. (a)

2. (c). The reference aerial is usually an isotropic radiator for which the power flux can be calculated even though it cannot physically exist. A Hertzian dipole, or a half-- wave dipole, radiating in the direction giving maximum power flux, sometimes used as reference.
3. (d). The effective aperture of a half-wave dipole is much greater than its physical area; that of a microwave horn aerial is less.
4. (a). This follows from the Reciprocity Theorem.
5. (b).
6. (d). If the thickness is increased the radiation resistance falls, the resonant length becomes shorter, the bandwidth increases, but the polar diagram, and hence the gain, remains the same.
7. (c). The resistive component, which accounts for radiated energy (so long as Joulean losses are small enough to be neglected) is much smaller than the radiation resistance at half-wave resonance.
8. (d). The dipole produces waves with electric vectors in planes containing the dipole, the slot produces waves with electric vectors in planes perpendicular to the length of the slot
9. (c). If the elements are spaced wider than half a wavelength "end-fire" lobes occur, causing energy to be radiated along the axis of the array. For spacing less than half a wavelength mutual impedance between the elements causes undesirable effects.
10. (a). Mutual impedance between the driven element and the parasitic elements reduces the input impedance of the driven element-typically to such an extent that the input impedance of a folded dipole in this application is a good match to 70 ohm coax.
11. (c). The functioning of a unipole aerial involves the flow of radial currents in the earth around it. Since the unipole must be short compared to a quarter wavelength at such a frequency the radiation resistance is only two or three ohms or less. Hence dissipative resistance has a marked effect on the aerial efficiency.
12. (a).
13. (a). Two adjacent dipoles fed from adjacent ends in antiphase so that they radiate in phase are equivalent to a centre-fed full-wave dipole. This typically has an impedance of several thousand ohms so that a number can be conveniently fed in parallel from a twin-wire feeder of characteristic resistance in the region of 600 ohms.
14. (c).
15. (b).
16. (b). Provided the aperture is many wavelengths across, the angle of the main lobe between half-power points is approximately $\lambda / a$ radians, where $\lambda$ is the wavelength and $a$ is the aperture width in the plane concerned.

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# COMPONENT BRIDGE SURVEY 

# D.C. and A.C. Bridges 

by T. D. Towers,* M.B.E., M.A., M.I.E.E.

How the electrical characteristics of components can be measured accurately by balancing them exactly against a known-value standard

Nearly one hundred and fifty years ago S. H. Christie first suggesped the bridge principle of measuring component characteristics in "Experimental Determination of the Laws of Magneto-electric Induction", Phil. Trans. Roy. Soc., 1833, Vol. 123, pp. 95-142. In essence the method was to insert the component being measured in a network of known components, two points in the network being connected to a current source and two others "bridged" by an instrument capable of detecting a potential difference. The values of the known network components were then adjusted until a zero reading on the detector showed the bridge points to be at the same potential. The characteristics of the one unknown component could then be calculated from the known components' values. Christie's paper is fascinating. Couched in elegant early Victorian style and with delightful illustrations, it discusses in detail the measurement of resistance for wires of different diameters, lengths and metals for nearly fifty pages without once using the word "resistance" (which had not yet been invented!)

Christie's ideas were neglected for ten years until Sir Charles Wheatstone gave them practical point in "An Account of Several New Instruments and Processes for Determining the Constants of a Voltaic Circuit', Phil. Trans. Roy. Soc. 1843, Vol. 133, pp. 303-327. This paper makes more comfortable reading as it uses the new term "resistance". The result was the original equal-ratio-arms "Wheatstone" bridge of Fig. I (a). In this, when $R_{v}$ is adjusted to give zero deflection in the detector, $I$, (a d.c. galvanometer or what we now call a centre-reading microammeter), then the unknown resistance $R_{X}$ is equal to the known $R_{V}$.

In this original Wheatstone arrangement, the calibrated variable resistance, $R_{V}$, had to be such that it could be set equal to the unknown, $R_{X}$, the ratio arms, $R-R$, being equal. In 1848 Siemens introduced the unequal ratio arms, $P-Q$, of Fig. I (b). The balance condition becomes $R_{x}=P R V Q$.

[^10]By selecting different ratios of $P$ to $Q$, it was possible to use one standard variable, $R_{V}$, to cover a much wider range of unknown resistances, $R_{X}$. This is the form in which the Wheatstone bridge has been known ever since.

The next basic development took place about 1865, when Clerk Maxwell used the Wheatstone bridge principle to measure inductance values in his "Inductance Ballistic Bridge", shown in basic form in Fig. I (c). In this, $L_{X}, R_{X}$ represents an unknown inductance $L_{X}$ of resistance $R_{X}$. The bridge is first adjusted for d.c. balance by means of $R_{V}$. Then the switch $S W$ is opened and the galvanometer, $I$, allowed to re-settle to zero. When the switch was closed again, the galvanometer needle kicked over before returning to zero. Knowing the ballistic properties of the galvanometer, you could measure the inductance value from the peak needle displacement. (Engineers who use an Avo in its resistance range for a rough estimate of a capacitanct from the kick of the meter needle are using the same sort of process.)

Maxwell then followed up with the "Inductance Null Ballistic Balance" shown in Fig. I (d). In this the switch was first closed, and, after the detector had settled, $R_{V}$ was adjusted for zero deflection. This gave $R_{X}$ in terms of $R_{V}, P$ and $Q$, all known. Next the switch was opened and closed repetitively, and the inductance $L_{V}$ varied in value until no deflection could be observed on the detector. This gave $L_{X}$ in terms of $L_{V}, P$ and $Q$.
Between 1870 and 1890 many different versions of the ballistic bridge were dreamedup both for inductance and capacitance measurements. The simple manually-operated on-off switch was soon replaced by a mechanical interruptor or commutator. The telephone (invented by Graham Bell in 1875) often replaced the galvanometer. But in 1891, Max Wien in Ann. der Phys., Vol. 44, pp. 681-712 (1891) introduced the modern a.c. bridge. He abandoned the interruptor method of energising the bridge, and fed the network, as in Fig. I (e), with alternating current of a definite frequency produced by an induction coil in which the primary current was made and broken at a fixed rate by a vibrating wire. He also designed a more sensitive detector in his "optical telephone", This was a magnetic telephone with its diaphragm tunable to
resonance with the a.c. in the bridge, anc with its output displayed by reflecting : beam of light from a mirror attached to th diaphragm. Wien adapted most of the ols ballistic methods to his new a.c. bridg Since Wien, much work has gone in developing more and more refined version of a.c. bridges.
The next big step forward was th "Wagner" earthing arrangement described in IgII by K. W. Wagner, "Zur Messunt dielektrischer Verluste mit der Wechselstrombrucke", Elekt. Zeits., 1911, Vol. 32, pp. 1001-2. This was designed primarily to get rid of what was known as the "head effect" by which the parasitic capacitances from the headphone detector could vitiate at a.c. bridge balance. Fig. I (f) illustrates the basic arrangement of the Wagner earth. Firs the bridge is balanced conventionally with th switch $S W$ in position one by adjusting $Z v$ Then the switch is changed to position tws and $R_{E}$ across the signal source is adjustet again for a null. By continually changin! between position one and two and re adjusting $R_{E}$ and $Z_{V}$ for a null, a fina position is reached where both points $b$ anc $d$ are at earth potential, so that the "heas effect" is eliminated.

Over the years many different a.c bridges have been produced as variants or the Wheatstone arrangement, all essen tially working with arm elements that do no have any cross coupling between them. Then in 1928, Alan Blumlein proposed in Britis! Patent No. 323037 a scheme for replacin the usual fixed isolated ratio arms with pair of tightly coupled inductors as shown i basic form in Fig. I $(g)$. When the bridg is balanced $Z_{X}=Z_{V}$ and equal current flow from $c$ to $d$ and from $c$ to $b$. Hence ther is no magnetisation of the inductor cort Thus, except for a small effect from th resistance of the windings, there is $n$ potential difference between $c$ and $d$, between $c$ and $a$, and thus between $b$ and This produces greater sensitivity in th bridge because the whole of the applie voltage appears across $Z_{V}$ and $Z_{X}$. Als if the point $c$ is earthed as shown in th diagram, strays from $b$ and $d$ to earth at innocuous, because these points are vi. tually at earth potential. Finally, stra! from $a$ to earth merely load the signal sours and do not affect the bridge balance.
The idea of the inductively-coupled rat arms bridge was not generally exploite
(a)

(C)
e)

(f)

(b)

(d)

(h)


Fig. I: The development of component bridges (a) 1843 Wheatstone: Resistance balance, equal ratio arms, $R-R$; adjusiable standard $R_{V}$; unknown, $R_{x}$; (b) 1848 Siemens; resistance balance; unequal ratio arms, $P-Q$. (c) 1865 Maxwell: inductance ballistic bridge.
(d) 1873 Maxwell: inductance null ballistic bridge. (e) 189I W'ien: a.c. bridge. (f) 1911 Wagner: "Wagner earth" bridge. (g) I928 Blumlein: inductively-coupled ratio arms bridge. (h) 1960 Short: two-signal bridge.
until just before and during World War II. Then in 1946 commercial development began and in 1949 a practical design was fully described by H. A. M. Clark and P. B. Vanderlyn in "Double-ratio A.C. Bridges with Inductively-coupled Ratio Arms", Proc. I.E.E., Part III, Vol. 96, 1949, pp. 189-202. Part I of that paper was taken almost verbatim from notes made by Blumlein before his death. C. G. Mayo (B.B.C.), quite independently, designed a similar bridge to Blumlein and was granted patents. Following these events, a whole family of inductively-coupled ratio arms bridges became commercially available. J. F. Golding in Wireless World, June 1961 gave a useful review of the practical constructional problems as well as of the principles of this kind of bridge.

All the Wheatstone-derived bridges described so far have the inconvenience that the bridge oscillator and the detector cannot have a common terminal. The connections of one diagonal of the bridge must be
floating. However, an interesting development was announced in 1960 by G. W. Short in "Two-signal Bridge", Electronic Technology, 1960, p. 452. This used two oscillators of different frequencies in a simple modification of the usual bridge arrangement as in Fig. $I(h)$, and made it possible for the input signal source and the output detector to have a common terminal without shorting one arm. The two input frequencies $f_{1}$ and $f_{2} \mathrm{mix}$ in the diode across the bridge and the bridge unbalance across the diode is sensed by the one-side-earthed detector across the $L C$ circuit tuned to the sum or difference frequency. The tuned circuit offers an effective short circuit to the two separate input frequencies. I am not aware that the system has been exploited commercially.

We have seen how component impedance characteristics can be measured by balancing them against known passive components in a bridge. Voltages can also be measured with great accuracy in a special type of "bridge".


Fig. 2. Voliage measurement by potentio-meter-type balanced bridge.

This is the so-called "potentiometer" circuit given in Fig. 2. Here a potentialdivider network across the known reference supply voltage, $V_{K}$ is adjusted until the indicator, $l$, shows zero current flow. Then points $a$ and $b$ are at the same potential. Since in balance no current passes through the indicator, there is no load on the potential divider and the unknown voltage at $a$ is given accurately by:

$$
R_{1} V_{K} /\left(R_{1}+R_{2}\right)
$$

With $R_{V}$ accurately calibrated, $R_{1}$ and $R_{2}$ are known and the unknown voltage can be computed. From Fig. 2 it can be seen that the potentiometer is really a three-arm

(b)

(c)


Fig. 3. Some basic arrangements used for d.c. resistance bridges (a) Wheatstone bridge for medium and high resistances as exemplified in internal connections of the P.O. box type of bridge. (b) Carey-Foster slide-wire bridge: for high accuracy midresistance measurements. (c) Kelvin Double-bridge : for low resistances, typically IS downwards.
bridge circuit, and like the full four-arm bridge works on null balance against a known standard.

## Classification of component bridges

Component bridges can be arranged to measure resistance (or conductance), capacitance, inductance, incremental inductance (in the presence of a standing current), mutual inductance, incremental mutual inductance, quality factor, phase angle, complex impedance (or admittance). In addition, bridges have been developed to measure active device characteristics such as transistor current gain, and valve or f.e.t. transconductance.
One way of classifying these diverse bridges is into "balanced" and "unbalanced". So far we have considered only balanced bridges, but it can be shown that near balance the reading of the null meter can be correlated closely with the deviation of the component characteristic from the centre value.
Nowadays the balancing of the bridge can be manual, semi-automatic or fully automatic. In semi-automatic bridges, near-balance is achieved manually by switches or push buttons bringing in fixed standards and the final small unbalance is read on a meter. Recently there has been a trend, however, to fully automatic self-balancing bridges. Analogue versions of these incorporate a feedback servo-mechanism and display the measurement as a displacement on a scale. Digital versions, using monolithic integrated circuits, switch themseives to the correct range, display the measured values in some form of digital readout (usually a row of figures) and normally also provide digital outputs for data logging, etc.
Bridges can also be classified by the degree of accuracy to which they can be read. Instruments fall into three main categories: general purpose, semi-precision and precision. Measurements in the first category ordinarily can be made from 0.3 to $3.0 \%$ accuracy, in the second 0.03 to $0.3 \%$ and in the last better than $0.03 \%$.
The tendency in commercial bridges is to make them capable of measuring as many different characteristics as possible, and the
so-called "universal" bridge has become a lab. commonplace, able to measure at least $R, C$ and $L$. For very refined measurement of a particular single characteristic, however you will usually find a specialist bridge devoted solely to that characteristic.'

Nowadays the bridge instrument often includes the signal source and detector internally in the equipment, but there may still be facilities for attaching a separate signal source or separate high-precision standards when required. Suitable standards are readily available. These range from an inexpensive $\frac{1}{2}$ to $1 \%$ decade capacitane or resistance box to expensive high-precision lab. standards supplied by specialist firms.

The student who is tempted to delve deeper into the vast literature on component bridges should find useful guidelines into the subject in standard references such as Radio and Electronic Laboratory Handbook by M. G. Scroggie (Iliffe Books Ltd.), The Principles of Electrical Measurements by H. Buckingham and E. M. Price (E.U.P.), Alternating Current Bridge Methods by B. Hague (Pitman), Electrical Measurements by F. K. Harris (Chapman and Hall), Electrical Measurements and Measuring Instruments by E. W. Golding and F. C. Widdis (Pitman), and Electronic Measurements by F. E. Terman and J. M. Pettit McGrawHill). At a more popular level you can consult such references as Bridges and Other Null Devices by R. P. Turner (Foulsham) or Electronic Lab. Instrument Practice by T. D. Towers (Iliffe Books).

If you do any reading into bridges, you will immediately be shaken by the vast number of different variants produced over the years. The first classified collection of a.c. bridge networks was published by Max Wien in 1891, following a collection of old ballistic methods by W. E. Sumpner in 1888. Subsequent collections were produced by Rowland (1898), Campbell (1908), Hay (1912), Cone (1920) and others. But the definitive classification was that of J. C. Ferguson in "Classification of Bridge Methods of Measuring Impedances", Trans. Am. I.E.E., 1933, Vol. 52, pp. 861-868. My own private "collection" of bridges now runs close on three hundred.
In the remainder of this article we will look only at the more common bridge


Fig. 4. Well-known bridges that used to be common around laboratories: (on left) P.O. Box d.c. Wheatstone, (on right) Mullard l.f. C-R bridge.


Fig. 5. Basic bridge types other than resistive-Wheatstone commonly used in l.f. bridges. (a) De Sauty: used for measuring C (in combination with a.c. resistive Wheatstone for $R$ ) in commercial $C-R$ tester.
(b) Maxwell : for low- $Q$ inductance.
(c) Hay: for high-Q inductance.
circuits widely used commercially, dividing them into three main groups: d.c., l.f., and h.f.

## D.C. bridges

Except for very special use, you will not nowadays find in the ordinary lab. singlepurpose d.c. bridges of the Wheatstone type, because so many universal bridges have adequate facilities for resistance measurements. However, where you do come across d.c. bridges, you will usually find that for medium and high resistance values the "straight" Wheatstone circuit of Fig. I(b) given earlier is used. A venerable example of this is the old P.O. box type for which the internal connections are shown in Fig. 3(a). The resistances are non-inductively wound coils connected between brass strips on top of the instrument. Tapered brass plugs, indicated as solid dots in the diagram, are inserted between the strips to short-circuit the resistances as desired. All the $\mathrm{s} / \mathrm{c}$ plugs are shown inserted in the diagram. If any one is withdrawn, it introduces into circuit a resistance of the value noted beside it in ohms. The arms $P$ and $Q$ are known as "ratio arms" and the plugs can be arrangedto produce resistance ratios of $100: 1,10: 1$ $I: I, I: 10$ and $I: 100$. The variable resistance arm $R$ can be adjusted by $1 \Omega$ steps from I to $11,110 \Omega$. Thus by using the ratic arms in conjunction with $R$ it is possible tc balance against an unknown resistance from 0 ol to 1, III, $000 \Omega$. A P.O. box Wheat-
 Measuring in will Measuring inductance with any certainty is difficult; and hit-or-miss methods not only waste time, but lead to expensive complications; which is why M.I. have designed the most comprehensive yet straightforward instrument available for complete inductance analysis.
The TF 2702 Inductor Analyser measures inductance at all frequencies from 20 Hz to 20 kHz , with currents up to $1,000 \mathrm{amps}-\mathrm{DC}, \mathrm{AC}$ or mixed! And it is sensitive enough for measurements at low current levels and for low- $Q$ inductances at high accuracy.

There is no interaction between balance controls, and operation is even further simplified by a
c.r.t. detector, which rapidly indicates the direction of unbalance. A tunable voltmeter gives final balance. A full range of accessories is available.

- Measurement range: $0.3 \mu \mathrm{H}$ to $20,000 \mathrm{H}$
- Internal or external excitation frequencies
- Variable a.c. current and d.c. bias facilities

MARCONI INSTRUMENTS LTD
Longacres, St. Albans, Hertfordshire, England
Telephone: St. Albans 59292 Telex: 23350

# it's clear to see. . . 


$\begin{array}{ll}81 r^{8} \\ 0 & D \\ 0 & 0 \\ 0 & 0 \\ 0\end{array}$

## DERRITRON ELECTRONICS LIMITED Instrument division

Sedlescombe Road North, Hastings, Sussex
Telephone Hastings 51372 Telex 95111
stone still stometimes in use in the author's lab. to select meter shunts can be seen on the left in Fig. 4.

Where higher accuracy of resistance measurement is required, the simple Wheatstone is replaced by the Carey-Foster slide-wire modification of Fig. 3(b). This is balariced in three steps. First arms $P$ and $Q$ are selected for an approximate null. Then the null is sharpened by moving the slider over the length $l_{1}$ of the slide-wire, $L$, which has a resistance of $r$ ohms per unit length. Next $R_{Y}$ and $S$ are interchanged and the slide-wire readjusted to $l_{2}$ to give null again. Then $R_{x}$ is given by:

$$
R_{X}=S+\left(I_{1}-l_{2}\right) r
$$

It should be noted that $P$ and $Q$ do not appear in the final equation. This bridge is specially suited to accurate comparison of an unknown resistance with a standard close to it in value.

When we come to measure low resistances, where contact resistances begin to make up a significant part of the total, we usually find the Kelvin Double-bridge of Fig. 3 (c) used. Here $R_{X}$ is the low resistance to be measured and $S$ is a standard of the same order of magnitude. $P, p, Q, q$ are four known ratio resistances, one pair of which ( $Q, q$ ) is variable. $Q$ and $q$ are variable in step together so that $P / p$ always equals $Q / q$. $Q$ and $q$ are varied until the galvanometer deflection becomes zero, when:

$$
R_{X}=S P / Q
$$

In wide-range precision commercial d.c. bridges, accuracies around $0.03 \%$ are typical, and the internal bridge arrangement used is switched from a Wheatstone in, say, the $10 \Omega$ to $10 \mathrm{M} \Omega$ range to a Kelvin for $10 \mu \Omega$ to $10 \Omega$ range. Usually neither a source nor a detector is included internally.

If you want to build your own precision d.c. resistance bridge, you will find a convenient design at p. 124 of R.P. Turner's Basic Electronic Test Instruments (Holt, Rinehart and Winstone).

## L.F. bridges

L.F. bridges fall into three main categories: fairly simple $C-R$ bridges, universal $L-C-R$ bridges, and special single-characteristic bridges for inductance for example.

Many relatively cheap commercial bridges are available for measuring $C$ and $R$ over a wide range at medium accuracy. One "old-timer" known to many is pictured on the right side of Fig. 4. Virtually all of these use the De-Sauty bridge circuit shown in Fig. 5(a) for the capacitance measurement, switching to a simple Wheatstone for resistance. M. G. Scroggie in his Radio and Electronic Laboratory Handbook (Iliffe) gives a useful practical design for a simple mainsfrequency $1 \%$ bridge for 10 pF to $10 \mu \mathrm{~F}$ and $10 \Omega$ to $10 M \Omega$, typical of these.
"Universal" l.f. bridges capable of measuring $L, C$ and $R$ with accuracies of $0.3 \%$ or better are widely available. Typically for $R$ and $C$ they use Wheatstone and

(b)

(c)

(d)


Fig. 6. Basic circuits commonly used in commercial h.f. impedance measuring bridges (a) Transformer inductive ratio arms bridge. (b) Adjusiable inducrivelycoupled ratio arms mutual admittance bridge. (c) Schering series-substitution capacitance ratio bridge. (d) Schering parallel-substitution capacitance rutio bridge.

De-Sauty respectively. For $L$ they generally use a "Maxwell" bridge-Fig. 5(b)-for low values and a "Hay" bridge-Fig. 5(c)-for high values.

Inductively-coupled ratio arms techniques described earlier are used in some commercial "in situ" l.f. bridges to measure components in conditions beyond the scope of conventional universal bridges, particularly in the presence of heavy shunting or with components wired in circuit.

Over the years many designs of universal $L C R$ l.f. bridges have been published. A useful complete construction was described recently by L. Nelson-Jones in "Universal Component Bridge", Wireless World, Dec. 1968. p. 434, covering with $1 \%$ accuracy the ranges $10 \mu \mathrm{H}$ to $100 \mathrm{H}, 10 \mathrm{pF}$ to $100 \mu \mathrm{~F}$, and

I \& to M』. This included a miniature c.r.t. output display.

Of single-function l.f. bridges quite a number have been on the market in the past, but they tend to be less common nowadays because freely-available widerange multi-function universal bridges make them unnecessary. Even so, such specialised instruments as an "inductor analyser" capable of measuring inductances up to $21,000 \mathrm{H}$ with standing current up to 10 A are on the market. Single-function bridges normally use one of the standard bridge circuits described earlier.

## H.F. bridges

Bridges to measure component characteristics above audio frequency become progressively more difficult to design, as the measurement frequency rises. Strays ultimately become of the same order of magnitude as some characteristics being measured. Bridge layout tends to become the main design problem. The basic type of h.f. bridge used tends to be either transformer ratio arms or the "Schering" capacitanceratio Wheatstone-derivative described below.

In practical instruments, the basic induc-tively-coupled ratio arms bridge illustrated earlier in Fig. $I(g)$ is usually the tappedtransformer driven as in Fig. 6(a), where the tappings $x, y$ and $z$, on the input transformer enable either an inductive or capacitive unknown to be measured. For inductance, tapping $x$ is moved to the right of $z$. To widen the bridge coverage with given variable standards $C_{8}$ and $R_{s}$, selective tapping of the inductive ratio arms can be used as shown in Fig. 6(b), thus introducing scale factors of $10: 1,1: 1,1: 10$, and 1 : 100 into the results. The unknown admittance is usually expressed in terms of parallel conductance and capacitance, with negative capacitance corresponding to an inductive unknown.

The other basic bridge common in h.f. measurements is the Schering-derived type of Fig. 6(c) and (d), whose balance equations do not involve the measurement frequency. In the series-substitution version of Fig. 6(c), the bridge is balanced with the test terminals, $T-T$, short circuited. The short circuit is removed and the unknown impedance, $R_{U}+$ $j X_{U}$, connected across the terminals and the bridge rebalanced. $R_{U}$ is then proportional to the change in $C_{R}$ and $X_{V}$ to the change in $C_{X}$ (capacitive if positive and inductive if negative). In the parallel substitution version, Fig. 6(d), balance is first set with nothing connected to the test terminals, $T-T$. The unknown admittance $Y_{U}=$ $G_{U}+\mathrm{j} B_{U}$ is then connected in and the bridge re-balanced. Then $G_{U}$ is inversely proportional to the change in $C_{2}$, and $B_{U}$ to the change in $C_{4}$ (positive for capacitance and negative for inductance).

Transformer ratio arms and Schering bridges are available commercially for measurements up to 250 MHz . Above this we move firmly into the realm of nonlumped characteristics, and generally find only very specialised admittance bridges using coaxial-line techniques measuring up to around 1500 MHz .

## Recent Products

Sullivan decade inductance bridge is a trans-former-ratio arm instrument using a modified Maxwell circuit and it measures inductance and resistance in terms of internal capacitance and conductance. It enables one side of the unknown and one side of the standard to be earthed. Accessories include a signal generator and detector. The bridge will measure up to 11.11 H with a discrimination of $0.002 \mu \mathrm{H}$ and this range can be extended by the use of external capacitance standards. Resistance can be measured up to $100 \mathrm{k} \Omega$ in five decades. H. W. Sullivan Ltd., Murray Road, Orpington, Kent, BR 5 3QU.
WW 371 for further details

Avo universal bridge type B. 150 is batteryoperated and features a digital in-line display. Units of measurement and the decimal position are displayed automatically along with the relevant figures. Capacitance may be measured up to $1199 \mu \mathrm{~F}$, inductance up to 119.9 H and resistance up to $11.99 \mathrm{M} \Omega$ An internal oscillator allows measurements to be made at 1 kHz , but other audio frequencies may be supplied from an external source. A polarizing voltage of up to 500 V can be applied externally in testing electrolytic capacitors.


An inductance adaptor, type A151, is available to allow iron-clad inductors to be measured directly on the bridge with varying d.c. bias current and at different a.c. signals superimposed via the bridge. Avo Lid., Avocet House, Dover, Kent.
WW354 for further details

Startronic resistance measuring bridge type 100.2 is designed for the measurement of lower value resistors. The ranges are -0.05 to $5 \Omega$ with an accuracy of $1.5 \%$ and $0.5 \Omega$ to $50 \mathrm{k} \Omega \Omega$ with an accuracy of $0.5 \%$, in four switched ranges. Type 108.2 is designed for the measurement of resistance in the milliohm range and is powered by internally fitted dry cells. A special feature of this instrument is the ability to vary the resistance measuring current. Range is $0-160 \mathrm{~m} \Omega$

with accuracy $\pm 2 \%$ above 50 m ! ) and $\pm 1 \mathrm{~m}$ § below 50 m (). The measuring current can be set at $600,300,200,100$ or 50 mA , as required. Both instruments are priced \{39. Startronic Division, Stow Electronics Ltd., Ponswood Industrial Estate, St. Leonards on Sea, Sussex. WW352 for further details

Bruel and Kjaer 100 kHz deviation bridge type 1519, allows for measurements to be made over the following ranges:
$R: 10 \Omega-100 \mathrm{k} \Omega$
C: $12 \mathrm{pF}-1 \mu \mathrm{~F}$
L: $5, \mathrm{dH}-20 \mathrm{mH}$
The bridge voltage is 0.35 volt for all ranges and the voltage appearing across the test component is 0.175 V . The centre point of the bridge can be grounded. B. \& K. Laboratories, Lid., Cross Lances Road, Hounslow, Middx.
WW363 for further details

Advance BR1, is a bridge for the measurement of inductance, capacitance and resistance to an accuracy of at least $1 \%$. An internal transistor oscillator of $1 \mathrm{kHz} \pm 1 \%$ provides the bridge energizing source for the measurement of $L, C$, and $R$ together with $Q$ of an inductor and loss factor $D$ of a capacitor. Alternatively, an external signal generator of frequency between 20 Hz and 20 kHz may be substituted. A high sensitivity, low noise, bridge amplifier detector precedes the meter circuit and facilitates an accurate and rapid balance condition. The range of measurement for $L, C$ and $R$ is respectively $0.2 \mu \mathrm{H}$ to $110 \mathrm{H}, 0.5 \mathrm{pF}$ to $1100 \mu^{+}$ and $10 \mathrm{~m} \Omega$ tol1M $\Omega$ each range being covered in eight decades. Resistance measurements may be achieved using the internal d.c. supply source, a

9-V dry battery, or any externally applied potential not exceeding 500 V d.c. The BR1 is fitted with an isolating transformer enabling it to be used with a wide range of signal generators for measurements at frequencies other than 1 kHz . Advance Electronics Ltd., Roebuck Road, Hainault, Essex. WW366 for further details
'Logohm' (Baldwin) wide-range resistance bridge is a direct-reading Wheatstone portable instrument. It is accurate to within $1 \%$. A galvanometer and dry batteries are contained in the instrument. Balance is obtained by rotating a dial carrying a wire-wound logarithmic potentiometer, thus ensuring smooth and continuous adjustment. The dial carries a logarithmic scale calibrated from 5 to $500 \Omega$. A four-position switch marked XO.01, X1, X100, X 10,000 covers an overall resistance range from $0.05 \Omega$ to $5 \mathrm{M} \Omega$. Due to its logarithmic scale, the percentage accuracy is substantially uniform throughout the range of the bridge, and is, within $1 \%$, a degree of accuracy adequate for most practical requirements. A feature of the galvanometer is that its pole-pieces are specially shaped to ensure very high sensitivity at zero for accurate balancing, and low sensitivity on maximum deflection, to prevent damage to the movement when the bridge is well out of balance. The range switch is so connected in the bridge arrangement that its contact resistance does not enter into the bridge equation for balance. P.S.B. Instruments Lid., Palmerston Road, Wealdstone, Harrow, Middx.
WW364 for further details

Sprague Transfarad capacitor analyser measures capacitance between 1 pF and $2000 \mu \mathrm{~F}$ in five overlapping ranges. Normally, 25 V d.c. is applied, but for ceramic capacitors rated below 25 V , insulation resistance may be calculated from leakage current measurements at the exact rated voltage. Power factors between 0 and $50 \%$ can be measured by a Wien bridge. Leakage current can be measured between 0.6 and $600 \mu \mathrm{~A}$ in seven ranges, and is read off the meter at the exact rated d.c. voltage of the capacitor. In the a.c. bridge, only 0.5 V is distributed, and less than this appears across the capacitor under test. An electronic indicator simplifies bridge balancing for capacitance and power factor measurements. In testing electrolytics, a polarizing voltage is available continuously adjustable from 0 to 150 V . Sprague Electric (U.K.) Ltd., Trident House, Station Road, Hayes, Middx. WW 350 for further details

British Physical Laboratories' 1 kHz component comparator, model CZ $457 \mathrm{Mk} . \mathrm{V}$, is capable of fast tolerance testing of $R, C$ and $L$ components under mass production conditions. Percentage deviation, covering tolerances from $0.1 \%$ upwards, is directly read off multi-coloured scales. Semiconductors are used throughout. The measurement ranges are:
$L 10 \Omega-5 \mathrm{M} \Omega$ and $0.01 \Omega-2.2 \mathrm{M} \Omega$
$C 20 \mathrm{pF}-10 \mu \mathrm{~F}$ and $0.02 \mathrm{pF}-1.8 \mu \mathrm{~F}$
$L 2 \mathrm{mH}-100 \mathrm{H}$ and $2 \mathrm{pH}-18 \mathrm{H}$.
The bridge voltage is less than 500 mV r.m.s.


## measure



The Culton 167
Automatic Component Bridge

log


The Culton 267
Digital Comparator/Serialiser
control


The Culton 367 Programmed Limit detector
goods inwards inspection
quality control
research
development

The Culton 167 is ready within five seconds of switch-on, makes measurements to $0.1 \%$ without preliminary set-up or manual intervention, measures to as close as $0,01 \%$ with minimum of manual operation.

The Culton 167A Guarded Component Jig takes all sizes of components. even with cropped or mutilated leads, has stray capacity less than 0.01 pF ., triggers the bridge automatically when closed.
component evaluation
production test
quality control
environmental test

Log results for statistical analysis. spot trends and deviations in production processes.
Set nominal value and percentage tolerance on the Culton 267 (20, 10. 5.2 , etc. to $0,05 \%)$, and measure the batch. Numbered entries on typewriter are printed in red if out of tolerance.
The Culton 267 will drive an IBM typewriter or Addo printer. If limit detection is not needed, use the Culton 167 direct into a Kienzle printer.
advanced production techniques
component trimming processes
computer-controlled test systems

The Culton 167 can be addressed with a limit value (by simple contact closure to ground) and gives High/ Low signal within 40 milliseconds to a discrimination of 1 part in ten thousand. Set upper and lower limits on the Culton 367 and get Pass/Fail signal in less than 100 ms . Or, instead of setting switches, use plug-in programme board or tape reader.

# TINSLEY HIGH PRECISION RESISTANCE BRIDGES TYPE 5577 

A Modern Resistance Bridge in which the variable arm consists of conductance decades, built up in binary increments covering a million steps. Six resistors per decade. lowest resistor 50 ohms therefore switch contact resistance negligible. All resistors in circuit when maximum current is flowing. Built-in lead compensator.
Total range 0.0001 ohms to 100 M ohms in seven ranges. Limit of error at mid-range $0.001 \%$
For full details of this and other instruments in the
 Tinsley range, write to:-

## H.TINSLEY \& CO LTD <br> Pacemakers in Precision Measurement WERNDEE HALL. SOUTH NORWOOD. LONDON. S.E.25. Tel: 01-654 6046



Accuracy is $\pm 2 \%$ of indicated deviation or $1 \%$ of F.s.d. whichever is the greater. British Physical Laboratories, Radlett, Herts.
WW 356 for further details.

General Radio impedance bridge, type 1608-A, measures $C, R, L$ and $G$ with digital readout to an accuracy of $\pm 0.1 \%$. Six bridge circuits cover all possible phase angles so that any network can be measured, even filters, transducers and equalizers. The ranges are as follows:
$C: 0.05 \mathrm{pF}$ to $1100 \mu \mathrm{~F}$ in 7 decades
$L: 0.05 \mu \mathrm{H}$ to 1100 H in 7 decades
$R: 0.05 \mathrm{~m} \Omega$ to $1.1 \mathrm{M} \Omega$ a.c. or d.c.
$G: 0.05$ nanomho to 1.1 mhos a.c. or d.c. Also $D$ and $Q$. In a.c. resistance and conductance measurements a $Q$ adjustment for precise balancing gives phase information useful in predicting high-frequency behaviour. For production testing of components, a test jig is available. An internal $1 \mathrm{kHz} \pm 1 \%$ generator module is normally supplied but plug-in modules for other frequencies are available. The detector can be external or internal, flat response or selective at the frequency of the plug-in module. Three d.c. supplies are

included for maximum sensitivity over a wide range of frequencies. Operation is from mains supply and consumption is 10 W . General Radio Co. (U.K.) Lid., Bourne End, Bucks.
WW 372 for further details

Capacitance bridge B541C Mk. II by Wayne Kerr allows for rapid capacitor checks, and provides continuous readings of changing values. It has comprehensive faciltites built-in for comparative measurements, for off-selting any deflection (enabling the desired nominal value to be adjusted to mid-scale) and for backing-off the first digit of any reading to give increased meter resolution. The power unit incorporates a rechargeable battery giving long periods of operation independently of 110 or 240 V supplies. Seven ranges cover $0-10 \mathrm{pF}$ to $0-10 \mu \mathrm{~F}$ in decade steps. Voltage and current out puts are provided for recorders, pass/reject systems or control circuits, with a response time of only 60 milliseconds. Bridge circuits of the B541 are constantly maintained in the "balance" condition by a system of feedback. A third terminal is available for screening connections when required. On all ranges accuracy is $\pm \mathbf{0 . 2 5 \%}$ of maximum. Wayne Kerr Co. Ltd., New Malden, Surrey. WW $\mathbf{3 6 7}$ for further details

## Culton automatic component bridge, type 167,

 is a transformer-ratio arm bridge with an accuracy of better than $0.1 \%$ when measuring $L, C$ or $R$. Accuracy is controlled by transformer ratios used and by an internal standardized 10 nF capacitor and $10 \mathrm{k} \Omega$ resistor which have been aged and calibrated. The instrument is suitable for checking thick film circuits which contain a number of resistors or capacitors of differemt values on the same substrate, and an input unit can be supplied which enables the bridge to be programmed to give "pass or fail" information. Range selection and balance is auto-
matic. Alternatively, the bridge can be connected to a unit which incorporates a detector and a circuit for driving a printer. Component values can then be logged and compared against values set into the detector; any components outside the set values can be rejected. The bridge measures inductance from 1 mH to 1099.9 H , capacitance from 0.01 pF to 10.999 uF and resistance from $10 \Omega$ to $10.999 \mathrm{M} \Omega$. The component to be measured is held in a springoperated Y-shaped clamp. Transistors are used throughout and the bridge power source is a 1 kHz oscillator which becomes 1592 Hz when measuring inductance. Culton Instruments Ltd., Dorking, Surrey.
WW 370 for further details

Siemens universal bridge type M565-A1 operates over the following ranges:

## $R: 0.1 \Omega$ to $110 \mathrm{M} \Omega$ in 9 decades

L: $10 \mu \mathrm{H}$ to 1100 H in 8 decades and using a 1000 Hz signal
C: $10 \mu \mathrm{~F}$ to $1100 \mu \mathrm{~F}$ in 9 decades using a 50 Hz signal
A comparison circuit is provided for impedance measurements using an external standard in the range $0.1 \Omega$ to $1 \mathrm{M} \Omega$. The scale division is $-20 \%$ to 0 to $+20 \%$. Error is less than $1.5 \%$ of full-scale value, less than $3 \%$ of the nominal value in the ranges $10^{1}$ to $10^{7}$, less than $5 \%$ of the nominal value in the ranges 10 and $10^{8}$, and less than $5 \%$ of full scale value ( $20 \%$ ) for comparison measurements. A protection circuit is provided for the meter. An auxilliary voltage source of 9 V can be employed from either dry batteries or accumulators. Distributed by Cole Electronics Lid., Lansdowne Road, Croydon CR 9 2HB.
WW 358 for further details

Derritron digital Wheatstone bridge, is balanced in the usual way, null balance being detecred with a sensitive taut suspension galvanometer. The "unknown" resistance value is then given directly by an illuminated display. The display consists of four figures approximately 16 mm high, a decimal point and the correct unit symbol, i.e. $\Omega, \mathrm{k} \Omega$, or $\mathrm{M} \Omega$. The resistance value is given without ambiguity, the operator merely reads in a straight line without having to note pointer positions or to work out the ratio in use. The total range is $0-9.999 \Omega$ in seven sub-ranges. Resistance coils are wound in Constantan and adjusted to an accuracy of better than $\pm 0.1 \%$. The 2.5 -volt test source is from sealed nickel-cadmium cells which are continuously trickle charged whenever the instrument is switched on. These cells require no maimtenance. The 100 -volt test potential is derived from a fullwave rectifier with smoothing circuit. Regulation of this supply is such that it may be short-circuited without damage. This feature also limits the power dissipated in the unknown resistor and means that the test voltage is highest for high resistance values,
giving maximum sensitivity. Derritron Electronics Lid., Instruments Division, Sedlescombe Road North, Hastings, Sussex.
WW 365 for further details

Muirhead type D-30-A is a portable, multipurpose measuring set suitable for general resistance testing and for applications in the field, particularly in the communications industry. It embodies a Wheatstone network of resistors comprising a pair of variable ratios controlled by a single switch and a $11110 \Omega$ four-decade resistance variable in steps of $1 \Omega$. An internal battery is provided and balance is detected on a pointer-type galvanometer. Multiplying powers for the ratios provided are: $\times 0.001, \times 0.01, \times 0.1, \times 1, \times 10$, $\times 100$, and $\times 1000$ and accuracy between $1 \Omega$ and $1 \mathrm{M} \Omega$ is $\pm 0.15 \%$, above $1 \mathrm{M} \Omega \pm 1 \%$ at $10 \mathrm{M} \Omega$. and below $1 \Omega \pm 10 \mathrm{~m} \Omega$; Non-reactive card-wound resistors are used adjusted to within $0.1 \%$. The galvanometer is a centre-zero moving-coil instrument calibrated $50-0-50 \mu \mathrm{~A}$. Power is provided by two leakproof dry batteries. Muirhead \& Co. Ltd., Beckenham, Kent.
WW 369 for further details

Heathkit capacitor checker, model IT28, provides complete analysis of all capacitor types, with direct reading scales. It features a low bridge voltage for safe testing of miniature electrolytics. There are 16 leakage-testing voltages. The unit measures capacitance from 10 pF to $1,000 \mathrm{t} \mathbf{F}$, and resistance from $5 \Omega$ to $50 \mathrm{M} \Omega \Omega$. A comparator circuit measures $L, C$ or $R$ with an external standard. There is a calibrated power factor control and an electronic null and leakage indicator. it measures leakage in three sensitivity ranges: 2 mA for electrolytics; $15 u A$ for miniature transistor type electrolytics and $2 u A$ for paper, mica, ceramic,

etc., with 18 switch-selected leakage voltagesfrom 3 to 600 V d.c. External generator provisions are included to allow measurements at frequencies up to 10 kHz . Daystrom Lid., Bristol Road, Gloucester.
WW 351 for further details

Marconi Instruments' universal bridge model TF2700 uses the conventional bridge configurations but provision has been made for the connection of a large number of external facilities, so that a wide-range general-purpose instrument can be rapidly converted for a specialized measurement, without need of modification or special accessories. The internal battery-powered transistor oscillator provides a bridge source for measurements of $L, C$, and $R$ at 1 kHz , or an external source can be used between 20 Hz and 20 kHz . The internal aperiodic detector also uses battery-powered transistors, and may be used with both the internal and external bridge drive: an external detector can be used instead with either source. Resistance can also be measured with d.c. using the internal battery and galvanometer, or with either item replaced by external equivalents. Finally, mixed a.c. and d.c. can be applied to the bridge when

measuring components that require polarization, or for the determination of incremental properties. The measurement ranges are $0.1 \mu \mathrm{H}$ to 110 H , and 0.5 pF to 1.100 uF , each in eight decades, with phase defect value, at 1 kHz from internal source, or 20 Hz to 20 kHz from external source: $0.01 \Omega$ to 11 M ! ) in eight decades, at d.c. or 1 kHz from internal sources, or at d.c. or 20 Hz to 20 kHz from external sources. Accuracy is within $\pm 1$ or $2 \%$ depending on range. The bridge sources are 1 kHz $( \pm 5 \%$ ) from internal oscillator (or 20 Hz to 20 kHz from external source, for $L, C$ or $R$ measurement) and 9 volts from internal battery, or an external supply for greater discrimination. Power is supplied from an internal 9-volt battery (consumption approx. 7 mA ). Price 285. Marconi Instruments Lid., Long Acre, St. Albans, Herts.
WW 349 for further details

Philips type PM6300 universal measuring bridge features a large scale display, an electronic indicator, and facilities for $\tan S$ and $Q$ compensation. The bridge is balanced by means of a range selector (coarse control) and a high precision wire-wound reostat (fine control). During measurement one pole of the component being measured is at earth potential, making it possible to measure components already mounted on a chassis. For resistance measurements the measuring range is $5 \Omega-10.5 \mathrm{M} \Omega$ in six sub-ranges. The error is less than $1 \%$ of the measured value $\pm 0.5 \%$ of full scale. Loading is 0.05 W maximum. For capacitance measurements the measuring range is $5 \mathrm{pF}-105 \mu \mathrm{~F}$ in seven sub-ranges. The error here is less than $1.2 \%$ of the measured value $\pm 0.5 \%$ of full scale, and the measuring frequency is 50 Hz . Inductance measurements are possible over the range 0.5 mH 105 H in five sub-ranges. The measuring frequency is 50 Hz and the error the same as that for capaci-
tance measurements. Factor $Q$ can be compensated between 1 and 65. Pye Unicam Ltd., York Street, Cambridge.
WW 361 for further details

Tinsley general utility bridge, type 4551, is capable of measuring inductance, time constant $L / R$, capacitance, power factor, and effective resistance. The addition of a galvanometer and an accumulator enables it to be used as a Wheatstone bridge. The a.c. source is a battery-operated 800 Hz oscillator fixed in the case. Detection is by headphones. Accuracy is $1 \%$ over the following ranges: $L(5 \mathrm{mH}$ to 10 H$), C\left(10\left(31^{*}\right.\right.$ to $\left.111^{\circ} \mu 1^{\circ}\right)$, a.c. resistance $(0.1 \Omega$ to $100 \mathrm{~K} \Omega$, and d.c. resistance $(0.1 \Omega$ to $100 \mathrm{k} \Omega$. $L$ and $C$ measurement is also possible trom 1 mH to 100 H and from 10 pF to $10 \mu \mathrm{~J}$, but the accuracy of the measurement is reduced at the extremities of the ranges. A portable galvanometer and accumulator are available as accessories for Wheatstone bridge operation. H. Tinsley \& Co. Ltd., Werndee Hall, South Norwood, London, S.E. 25 .

WW 373 for further details
T.I.A. model LCR20 universal bridge is designed to measure inductance, capacitance and resistance, generally to an accuracy of $\pm 1 \%$. It is self-contained, and transistorized, being powered by internal batteries. Provision is made to excite the bridge from its internal oscillator at $100 \mathrm{~Hz}, 1 \mathrm{kHz}$ and 10 kHz . For the measurement of resistance of reactive components (for instance, iron cored transformers), the bridge may be excited at d.c. from internal batteries. The detector is a moving coil meter movement, preceded by a high gain amplifier. Automatic gain control is incorporated to facilitate balancing the bridge with inductors and capacitors of low power factor. At d.c. a transistor chopper is switched into the circuit to convert the amplifier to a sensitive d.c. detector. The bridge operates over the following ranges:-
$R: 1.9 \Omega$ to $1.9 \mathrm{M} \Omega$ for d.c. (to $19 \mathrm{M} \Omega$ for 100 Hz a.c.)

C: 19 pF to $190 \mu \mathrm{~F}$
L: 190 cH to 1900 H
$Q$ indication is: $0.01-1$ at 100 Hz
$0.1-10$ at 1 kHz
$1-100$ at 10 kHz
Tans indication is: $0.001-0.1$ at 1 kHz $0.01-1$ at 10 kHz
An internal 4.5 volt polarizing voltage may be applied to electrolytic capacitors during testing. Thomas Industrial Automation Lid., Electronic Centre, Deansgate Lane, Altrincham, Cheshire. WW 360 for further details



Inductance meter type LRT from Rohde \& Schwarz measures inductances between $0.1 \mathrm{\mu H}$ and 1 H in seven ranges to an accuracy of $\pm 1 \%$ $\pm 0.01 \mu \mathrm{H}$. The test voltage of this transistor instrument never exceeds 80 mV , not even in the case of very high- $Q$ coils. Thus, the field strength is kept to within 5 and $20 \mathrm{~mA} / \mathrm{cm}$ permitting measurements to be made also on highly permeable ferrite or laminated core coils without introducing an additional measuring error. Moreover. direct readings of the coil $Q$ from 2 to $1000(L>1 \mu \mathrm{H})$ can be obtained with the LRT. It provides also a simple means for measuring self-capacitances of coils between 0 and $200 \mathrm{pF}(Q>20, L>42 \mu \mathrm{H})$ and the resonant frequency of parallel-resonant circuits, and to make very precise $L$ comparison measurements. Capacitance meter type KRT, is fully transistored and measures capacitance between 1 pF and $100 \mu \mathrm{~F}$ in seven ranges. The maximum test voltage is less than 25 mV permitting accurate measurements on voltage-sensitive, high-dielectric constant, and semiconductor capacitors. With this new instrument it is possible to determine capacitances at an adjustable polarizing voltage. Moreover, a built-in bias-voltage source is provided to investigate the voltage-dependence of the capacitance of varactors. Distributed by Aveley Electric Lid., South Ockendon, Essex.
WW $\mathbf{3 6 2}$ for further details

Nombrex transistor capacitance-resistance bridge, model 32, measures a wide range of resistance and capacitance, and has provision for indication of leakage and power factor in the larger values of capacitors. The ranges are: $5 \Omega$ $10 \mathrm{k} \Omega, \quad 100 \Omega-1 \mathrm{M} \Omega, \quad 10 \mathrm{k} \Omega-100 \mathrm{M} \Omega, \quad 5 \mathrm{pF}-$ $0.01 \mu \mathrm{~F}, \quad 100 \mathrm{pF}-1 \mu \mathrm{~F}$ and $0.01 \mu \mathrm{~F}-100 \mu \mathrm{~F}$. Accuracy at centre scale is $2 \frac{1}{2} \%$, maintained to within $5 \%$ except on extreme high and low values of $R$ and $C$. The instrument employs an electronic indicator to observe the measurement balance point. Separate scales are provided for resistance and capacitance, clearly marked for accurate read-off. Visual discrimination is generally within $2-4 \%$ of indicated value but as both $C$ and $R$ ranges overlap considerably, a read-off within $2 \frac{1}{2} \%$ ( $\pm$ accuracy tolerance factor) can usually be achieved by choice of suitable range. Power factor measurements up to $70 \%$ can be made. A sensitivity control permits rapid initial assessment of component value, adjustable to attain optimum critical balance indications. Nominal indications of capacitor leakage are provided by a neon indicator circuit, for capacitors of any value or voltage rating. Price 10 gn . Nombrex Ltd., Instruments Division, Exmouth, Devon.
WW 357 for further details

Radiometer R,L,C component comparator, type TRB11, is a solid-state, line-operated precision measuring instrument. It is primarily intended for use at the end of production lines to provide a rapid and accurate comparison with, or a deviation from, a selected standard component. Measurements can be performed at 1 kHz on
resistors (10!? to 10M !), capacitors ( 20 pF to 20 HF ), and inductors ( 1 mH to 10 H ) by means of two sets of terminals: one pair plus guard for the unknown, and one pair plus guard for the standard. Magnitude deviation is indicated directly in percentage within the four ranges: -1.5 to $+1.5,-6$ to $+6,-25$ to +35 , and $-5010+100$. The phase-angle deviation is indicated directly in radians within the four ranges: -0.015 to $+0.015,-0.06$ to +0.06 , -0.3 to +0.3 , and -0.6 to +0.6 . The comparator is well suited for high-speed production testing, since one measurement can be performed every second by using an optional component jig, and since no balancing is required during operation. There are output terminals for connection to limit sensing devices such as the limit sensor, type LMS1, which can operate sorting-machines, control mechanisms, etc. Instrument marketed by Omega Laboratories Lid., 57 Union Sireet, London S.E.1.
WW353 for further details

Hewlett Packard universal bridge, model 4260A, is designed for case of operation, and measures $C, L, R, D$ (dissipation of capacitors), and $Q$. The readout for $C, R$ and $L$ is digital with the decimal point automatically positioned. Units of measurement (e.g., pF and $\mu \mathrm{F}$ ) and the equivalent circuit automatically "pop up" with a twist of the function switch. There are no multipliers and no dials that need interpolation. Operation is simple. Set the function knob for the parameter to be measured, adjust the range switch for an on-scale indication, and obtain a null with the $C R L$ control. There are no interacting controls to adjust and readjust. Components with low or high $Q$ are claimed to be as easy to measure as those without loss. Five bridge circuits are incorporated in the 4260A. An internal 1 kHz oscillator drives the bridge for $C, L, D, Q$ measurements; an internal d.c. supply is used for $R$ measurements. Components may be biased by connecting a battery to the rear terminals.

The measurement range is:
C: 1 pF to $1000 \mu^{+}$
L: $1 \mu \mathrm{~A}$ to 1000 H
$R: 10 \mathrm{~m} \Omega$ to $10 \mathrm{M} \Omega$
Error is within $\pm 1$ or $2 \%$ for these measurements. Hewlett Packard L.td., 224 Bath Road, Slough, Bucks.
WW368 for further details
Danbridge Denmark universal bridge type UB1 may be used in conjunction with a number of accessories for most of the usual d.c. and a.c. measurements. By means of the various switches. and shorting straps provided the bridge circuit may be adjusted for the measurement of resistance, capacitance and power factor, self inductance and effective resistance, mutual inductance, $Q$ factor, frequency, turns ratio, resonant resistance, etc.

Measurement ranges are:-
R: $1 \mathrm{~m} \Omega-10 \mathrm{M} \Omega$
L: $1 \mu \mathrm{H}-1 \mathrm{H}$
C: $1 \mathrm{pF}-100 \mu \mathrm{~F}$
The frequency range of operation is $0-20 \mathrm{kHz}$. The basic error is within $1 \%$, but the actual accuracy obtained on a.c. measurements will depend on the various earth-admittances of the generator, detector, measuring object and the bridge circuit itself. Thus it is not possible to give exact figures for the total accuracy. The ratio resistors are adjustable: 0-1-10-100-1000-1000 ohms, accuracy $0.2 \%$. Maximum dissipation is 1 watt. The 4 -decade resistor $(10 \times 0.1-10 \times 100$ ohms) has an accuracy of $0.2 \%$ except the $10 \times 0.1$ ohms decade with an accuracy of $3 \%$. Maximum dissipation is 1 watt per resistor. The standard capacitor has a value of $0.1 \mu \mathrm{~F}$, accuracy $0.2 \%$, dissipation factor about $10^{-4}$. Price $\{69$. Distributed by Dawe Instruments Lid., Concord Road, Western Avenue, London, W. 3.
WW359 for further details

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5 mA
10 mA
20 mA


## $37 / 6$ $37 / 6$ $37 / 6$ $37 / 6$ $37 / 6$ $37 / 6$ 3816 $59 / 8$ $37 / 6$ $37 / 6$ $37 / 6$ $37 / 6$ $37 / 8$

Type M.R.85P. 4 lin. $\times 4$ in. fronts.
$50 \mu \mathrm{~A}$

|  |  | 15 |
| :---: | :---: | :---: |
| $50.0-50 \mu \mathrm{~A}$ | 59/6 | 30 amp . |
| $100 \mu \mathrm{~A}$ | 59/6 | 20 V . D.C |
| 100-0.100 2 A | 59/6 | sov. D.C |
| $200 \mu$ A | 55/. | 150 V . D.C |
| $500 \mu \mathrm{~A}$ | 52/6 | 300 V . D.C |
| $500-0.500 \mu \mathrm{~A}$ | 49/6 | 15 V . A.C. |
| 1 mA | 49/8 | 300 V . A.C |
| 1-0.1mA | 498 | 8 Meter 1m |
| 5 mA | 4918 | VU meter |
| 10 mA | 49/6 | 1 mmp A.C. |
| 50 ma | 49/6 | s amp. A.C.* |
| 100 mA | 48/6 | 10 amp . A.C. |
| 500 mA | $49 / 8$ | 20 amp . A.C |
| 1 amp | 4818 | 30 amp . A.C. |

## 

## Type MR.65P.

$50 \mu \mathrm{~A}$
$50.0-50 \mu \mathrm{~A}$
$100 \mu \mathrm{~A}$.
$100-0-10$
$500 \mu \mathrm{~A}$


## 

BAKELITE PANEL METERS


## 

ECHO HS-606 STEREO


Wonderfully com-
$\begin{aligned} & \text { cortable. } \\ & \text { Light- } \\ & \text { weight }\end{aligned}$
adjustable weight adjustable vingl headhand, 6 ft.
cable and stereo fack pluk. 25-17,000 jace pluk. $25-17,00$
eps., 80 imp. $87 / 8$. еря., 80 imp. $67 / 8$.
P. \& P. $2 / 6$.


AVOMETERS Supplied in excellent
condition fully tested and checked. Com-
plete with prods leads and instrucModel $7 \quad$ \&13/10/0 P. \& P. $7 / 6$.
B.F.O. Aerial trimmer, internal speaker and

## SINCLAIR EQUIPMENT




[^11]
## 








MODEL TE-90 50.000 OP. M . Mirror scale overload protec-



 LAFAYETTE DE-LUXE
100 XBFOLT "LAB


MODEL TE-12. 20,000 O.P.V. $0 / 0.6 / 6 / 30 / 120 / 600 / 1.200 /$
$3.000 / 6,000 /$
r. D.C. $0 / 6 / 30 / 120 /$ $3,000 / 6,006 \mathrm{~V} . \mathrm{D.C.0/6/30/129}$
$600 / 1,200 \mathrm{v} . \mathrm{AC} .0 / 60 \mu \mathrm{~A} / 8$
$60 / 600 \mathrm{~mA} 0 / 6 \mathrm{~K} / 600 \mathrm{~K} / 6 \mathrm{Meg}$. $60 / 600 \mathrm{~mA} .0 / 6 \mathrm{~K} / 60 \mathrm{~K} / \mathrm{fMeg}$.
$60 \mathrm{Meg}. \mathrm{\Omega} .50 \mathrm{PP} .2 \mathrm{MFI}$ $60 \mathrm{Meg}. \Omega$
$25 / 18 / 6$.


TE-900 20.000 $/$ /VOLT
GLANT MULTMETER
 Mirror scale and over full prow meter. ${ }^{6} \mathrm{ln}$ colour scale. 0/2.5/1.2
$250 / 1,000 / 5,000$ v. A.C. $25011,000 / 5.000 \mathrm{~V} .4 . \mathrm{C}$ $250 / 1.000 / 5,000$ v. D.C 10 mmp D.C. $02 \mathrm{~K} /$
$200 \mathrm{~K} / 20 \mathrm{MEG}$.
OHM
 $0 / 10 \mu \mathrm{~A} / 6 / 60 / 300 \mathrm{MA} / 12$
$0 / 2 \mathrm{~K} / 200 \mathrm{~K} / 2 \mathrm{M} / 200 \mathrm{M} \mathrm{M}$.
to $\operatorname{to}_{3 / 6}+17 \mathrm{~dB} . £ 12 / 10 / \mathrm{m} . \mathrm{P}$. \&
 MODEL TE-70, 30,100 D.C. $0 / 6 / 30 / 120 / 800 / 1,200$ A.C. $0 / 30 \mu \mathrm{~A} / 3 / 30 / 300 \mathrm{~mA}$
$0 / 16 \mathrm{~K} / 160 \mathrm{~K} / 1.6 \mathrm{M} / 16 \mathrm{Meg} \mathrm{Q}$ $\mathbf{~} 5 / 10 /=$ F. $\Delta$ P. $3 /-$.

TE-51. NEW $20,000 \mathrm{n} /$ VOLT MULTMETER, with
overloal protection and mirror scale. of/6/60/120.
1.200 v. A.C. $0 / 3 / 30 / 80 / 300 /$ (i00//3,000v. D.C. $0 / 60 \mu \mathrm{~A} / 12$


MODEL TE-10A. $200 \mathrm{k} \Omega$ Volc $5 / 25 / 50 / 250 / 500 / 2,500$ w.
D.C. $10 / 50 / 100 / 500 / 4,000 \quad$ v.

 69/8. P. \& P. $2 / 6$


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High quality ceramice conat ructlon. Windinge embedded In witreous enamel.
Heavy duty bruah wiper. Continuous rating. Wide range a a ailable ex-alock.

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$\begin{aligned} & \text { Accurate wide range sig- } \\ & \text { nal kenerator coveritog } \\ & 120 \mathrm{Kc} / \mathrm{s}-500 \mathrm{Mc}\end{aligned}$
$\begin{aligned} & 120 \mathrm{Kc} / \mathrm{s} \text { - } 500 \mathrm{Mc} / \mathrm{Mron} \text { on } \\ & 6 \text { 1mands. Directly call. }\end{aligned}$
$\begin{aligned} & \text { brated. Variable RF. } \\ & \text { atcenuator, audo output. }\end{aligned}$
$\begin{aligned} & \text { atlenuator, audio output. } \\ & \text { Xtal mocket for callora. } \\ & \text { tion. } 220 / 240 \mathrm{~V} \text {. A.C. }\end{aligned}$
$\begin{aligned} & \text { iion. } 22 w / 20 \mathrm{~V} \text {. A.C. } \\ & \text { Brand new with inatruc- } \\ & \text { tions. } \mathbf{£ 1 5 . ~ C u r r . ~ 7 / 6 . ~}\end{aligned}$
8ire $140 \times 215 \times 170 \mathrm{~mm}$.

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 $2 \%$ C. 10 MFYt $\pm 2 \%$ TURNS RATIO $1: 1 / 1000-1: 11100$ Opanted fom 9 volts. $100 \mu \mathrm{~A}$. Meter indication. Attractive 2 tone metal case. Size $71^{\circ}$


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wave $20-30.000 \mathrm{cps} . \mathrm{O} / \mathrm{P}$
$\mathrm{HIOH} \mathrm{IMP} .21 \mathrm{v}, \mathrm{P} / \mathrm{P}$
 aniable R.F. attenua:
tlon. Int./Ext. Modulas Aton. Incorporates dual purpose meter to monilar.
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bandwidth $1.5 \quad \mathrm{cpm}-800$ bundw. Input limp. 2 meg 0
KHZ 20 PFF . Time base. 5 range
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Weight $15 \mu \mathrm{Ba}, 220 / 240 \mathrm{~V}$
C.
output to feed mod
anpliters. Operates
on 9 volt battery. Cowerage $8 \mathrm{~s}-108 \mathrm{Mc}$. Retads
bullt ready for une. alue for mone
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The nost rellable way of teating for a dry joint in to meanure circult boart. Our klt for doing this comprises a large-scale (3in.) moving-coll meter, a gariabie resirtatice for adjusting
zero setting, and a wiring diagram with instructions. The only additional itema you will need are a battery, mone wire, suair of test roda. Price 18/6. Postage and lnsuratice $2 / 0$ REED-SWITCH
suftable for dozens of diflesent applichtiona, such as burglar alamas, conveyor-bell switching. These ure cimply glakg-
encaned wwitchen which can be operated by a pasaing permanent magnet coll. A special buy enablea no to offer these at $2 / 6$ each, or $24 /-$-a dozen. 'Sultable uaghets

BLANKET SIMMERSTAT
Although looking like, and fltted an an ordinary blunket ewitch, this it in fact a device for switiching the blanket on
or varying time periode, thurk giving a complete control rom of to full heat. Also auitable for controlling the
cemperature of any other applances uming up to 1 amp. temperature of any other applianceg uning up to 1 amp.
Linted at $27 / 6$ each, we offer these while our stock lasts at ouly $18 / 6$ each.
ade If R.P.M. GEARED MOTOR
ut are very powerfiche. thene are almost silent runalng māius a nd the final-shaft speeds 16 R.P.M. 15/-. P. \& 1 ln . $2 / 9$ THERMOSTATS
Type "A" 15 anap. for controlling room heaters, kreenhouke. aring cupboard. Has apladle for poslnter lyob. Quickly
adjuatable from $30-80$ deg. F. $8 / 6$ plus $1 /$ jost. Buitable box for wall mounting $5 /-$ P. P. P. 1/-.
 the famous Bunvic Co. 8pladle salduats thly from 50 to 350 deg. F. Internal acrew alter the setting so thls could be adjustable over 30 deg. to 1000 dex. F 30 deg. to 1060 der. F .
8uitable for controling immersion heater or to furnace, oven klhn . immertion heater or make flame-start or flre alarm. 8/6 plus $2 / 6$ post and
insurance.; We cell thin the tee-stat as it cuts in and out
Type " D ". When taround treezing point. $2 / 3$ amps. Rat many unes, one of wlich would be to keep the loft pipes from treezing. If
 djustments cos nor reingerstor temperature $7 / 6$. plus $1 /$ post.
Type
ur". Glank encased
for controlling the temp. of ilquid-particularly those tin glata cuakk, vatse or sitiks. Chermontar in held (hoalf nubmerged) by rubber aucker or wre clip deal for hin tanks -developers and chemical 150 deg. F. Price $18 / \mathrm{F}$, plus $2 /$ post and insurance.

ELECTRIC CLOCK WITH 25 AMP, SWITCH Made by Smith's these unitsare as tited to many top quallity cookers tr control the oven. The clock is mains driven and frequency conon and off times to be accurately get. Ideal for awitching on cape
recorders. Ofered at only a fraction of the regular price-new and recorders. Offered a only a faction of the regular price-new and
unumed only 39/6. leas than the calue of the clock alone-pust and insurance $2 / 9$.

## INFRA-RED <br> HEATERS

Make up one of these lateat type heaters.


Ideal for bathroom. etc. They are simple

to make from nur easy-h-follow instructions ins indered wave length ( 3 microns)
useb slites enclosed elememts denigned mor the concet inlustrated. 18/6. plus $4 / 6$ post and Pricè for 850 watts elemb
ins. Pull switch $3 /$ extri.

## THIS MONTH'S SNIP

DEAC RECHARGEABLE BATTERIES
$6 \mathrm{v}, 500 \mathrm{~mA} / \mathrm{hr}$. size $2 \frac{1}{2}{ }^{\prime \prime} \times 1 \frac{1^{\prime \prime}}{}$ dia. Really powerful, will deliver 1 amp for $\frac{1}{2}$ hour. Regular price $65 /$-our price 29/6. New and unissued.


ELECTRIC CLOCK whi 3 amp switch mavie by truiths for Dreamland. These are mains driven and frequency controlled so are extremely accurate. The dial
enablen ""witch-on" time to be accuratels set. 8 witch on in 3 hours hater or by manual control. Intended for switching electric blankets thie needn only oue setting for the weamon. In neat plantle case with maine lead and two outlet pluge In new and unused, $30 / 6$, post and hasurance $3 /(\theta$.

## VARYLITE



Will dim fivoreacent or tacandeacent lighting up to 000 W . from full briliance to out. Fitted on M.K. Fluah plate, same size and flixing an standard wali withe be ilted in place of thise, or mount on surface. Price complete switch so may be tited in place of that, or mount
in heavy plastic box with control knob. $£ 3.19 .6$.

These infre red binoculars when fed from a high voltage source will enable objects to be seen in the dark, provided the objects are in the rayn of an infa-red beam. Each ere infra-red ceil. Theseoptical systems can be used an lensea for TV cameras-light celis. etc. (detnils mapplied). The equipment). They are unued and belleved to be in good working order but eold without a guarantee. Price $£ 3 / 17 / 6_{\text {. }}$ MINIATURE WAFER SWITCHES


4 pole, 2 way-3 pole, 3 way- 4 pole, 3 way
-2 pole, 4 way- 3 pole. 4 way- 2 pole -2 pole, 4 way- 3 pole. \& was-2 pole
6 way -1 pole, 12 way. All at $3 / 6$ each, $36 /-$ dozen, your asoortment.


BLANKET SWITCH Double pole with neon let into side so lumse with waterproof element-new plastic ame. $5 / 6$ each. 3 hent model $7 / 8$.
THERMOSTAT WITH
 PROBE

This has a sensor attached to a 15 AA s witch by a 141 m . tubiag-control range is 2odeg. F. to 150 leg.P. F o it is suituhle to control soll eapecially when in buckets enpectatly when in bucketo
or portabe veusels the the
sensor can be raised out and lowered into the ressel. This sound a bell or otber alarm When critical temp. is reached in stack or heap subject to pontaneous combuation or if liguld la being heated by gas of other means not controllable by the switch. Made by
tamous Teddingtun Co., we offer these at $12 / 6$ each.

Where postage is not scated then orders
over $\mathbb{C}$ are post free. Below 63 add $2 / 9$.
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post free. S.A.E. with enquiries please.

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New Science Projects combine fascination of Optics with Electronics.

## INFRA-RED TRANSMITTERS \& RECEIVERS

Unique devices in a brand new electronic field that can be exploited in a wide range of applications. Miniaturized construction and solid state circuit design is combined with outstanding modulation and switching cepabilities to provide infinite possibilities as short distónce speech and data links, remote relay controls, safety devices. burglar alarms, batch counters, level detectors, etc., etc

INFRA-RED PHOTO RECEIVER - MSP3
Ulira sensitive detector/amplifier for infra-red (Gallium Arsenide) or visible light optical links eception. Spectral response 9500 A. Robust. Cylindrical package is coaxial with incident light to facilitate optical alignment and heat sinking.

## 85/-

MAX RATINGS
Total dissipation (in tree air, Tamb $=25^{\circ} \mathrm{C}$.) $\quad . \quad 100 \mathrm{~mW}$. Derating factor......... $2 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$.

Suppled complete with suitable lenses, full Technical Data and Application Sheets.

GALLIUM ARSENIDE LIGHT SOURCE-MGA 100 alignment and heat stnking
 35/-

MAX RATINGS
Forward current If max.* D.C._.... 400 mA . Forward peak current if max. (pk) 6A Power dissipation
Reverse voltage Va max 1.0 V .
Supplied complete with suitable lenses, full Technical Data and Application Sheets, including Line of Sight Speech Link.

MICRO-MINIATURE INFRA-RED DETECTOR - 31 F2 Silicon NPN photo-dioded of passivized planar construction, suitable for punched cay
counters. fim sound track. elc.
Infra-red devices (except 31F2) are supplied complete with
suitable lenses, technicaldata and typical application information.

## PHOTOCONDUCTIVE CELLS

CADMIUM SULPHIDE CELLS (Cds)
inexpensive light sensitive resistors which require only simple circuitry to work as ight riggering units in a wide range of devices. such as: llashing or breakdown conscious - use with A.C. or D.C. Spectral response covers whole visible light range.

## MKY251

Epoxy seated $1 \frac{1}{2}$ in. diam. $x \frac{1}{2}$ in. thick. Resistance at 100 Lux -700 to 3.000 ohms. Maximum voliage 200 A.C. or D.C. Maximum curren MKY101-C
Epoxy sealed. $\frac{8}{\text { in in. diam. } x}$ in. thick. Resistance at 100 Lux -500 to 150 mW . Maxmum voltage 150 A.C. or D.C. Maximum curren
$10 / 6$ post free

## MKY71

Glass sealed with M.E.S. base, Glass envelope th in diam overall length 1 in. Resistance at $100 \mathrm{Lux}-50$ Kohms to $150 \mathrm{Kohms}$. Maximum
voltage 150 A.C. or D.C. Maximum current 75 mW . $8 / 6$ post free
CADMIUM SELENIDÊ CELLS (Cdse)
These have a higher dark resistance in a given period than Cadmium Sulphide Cells. indicating much faster response. Surtable for all Cds applications plus applications in Time response showní in megohms is dark resistance measured 10 secs. after 400 Lux light intensity is intercepted.


## MKB5H

Hermerically metal sealed. $\mathbb{t}$ in: diam. $x$ in. thick. Time response Resistance at 10 Lux -50 Kohms to 1 megohm. Maximum voltage 50 A.C. or D.C. Maximum current 10 mW . Continuous current 5 mW . 16/6 post free

MKB12H
Hermetically metal sealed 8 in. diam. x $\frac{7}{1} \mathrm{in}$. thick. Time response 100 megohms. Resistance at 1,000 Lux - 100 ohms to 1,000 ohms Resistance at 10 Lux - 1 Kohm to 10 Kohms. Maximum voltage $16 / 6$ post free

## PHOTOGENERATIVE CELLS



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Highly flexible light guides that transmit light to inaccessible places as easily as electricity is conducted by copper wires. Fibre opties make it possible to control. miniaturize. split, reflect or transfer light from one source to many places at once sible. Proops offer both glass fibre optics or inexpensive Crofon plastic fibres for hundreds of experiments or serious applications in a fascinating new science.

RANK TAYLOR-HOBSON ENGINEERS KITS


All the basic components needed to demonstrate new ways to use light in serious applications With glass fibre optics consisting of housands ferruled, optically polished ends. Kit includes 12 18, and 24 inch standard light guides in 1.5 and 6 mm widths. 24 inch iwin exit guide yith $2 \times 1 \mathrm{~mm}$. outputs. Non-random ' $Y$ gulde wit $2 \times 3 \mathrm{~mm}$. outputs, adaptors and battery operated liaht source Supplied complete with card illustrated applications.

LOW-COST CROFON FLEXIBLE LIGHT GUIDE
Newly developed plastic light transmitting media made by Ou Pont and consisting of 64 special plastic fibres, each sheath. diam. and bundied logether in a lough, fiexible pensive prototype work Ends sern be pround to and mexcapped with Epoxy resin. Temp, range - $40^{\circ}$ to dyed or No loss of light through bending. 12-page data and polications booklet supplied

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## Other advanced Solid-State devices

RCA INTEGRATED CIRCUIT - CA3020
Complete Audio or Servo Ampllfier in one tiny package!
Preamp, phase invertor, driver and power output function in a single package gives maximum. and $\frac{2}{2}$ in. high. Operates from single D.C. supply of 3 to 9 volts high gain is coupled with built-in temperature compensation (-55 $10123^{\circ} \mathrm{C}$ ) and wideband operation. Complete with data and clrcuit applications.

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Suitable for light dimming and motor control circuits
Gate-controlled, full-wave, A.C. silicon switch with integral trigger that blocks or conducts instanily by applying reverse polarity voltage. Scitable for A diam. $x$ 훙 in. high. Complete with heat sink, data and applications information. 45/-post free

45/-post free

# Wilkinsons FOR RELAYS <br>  <br> P.O. TYPE 3000 AND 600 BUILT TO YOUR REQUIREMENTS-QUICK DELIVERY COMPETITIVE PRICES - VARIOUS CONTACTS DUST COVERS - QUOTATIONS BY RETURN Large stocks held of miniature sealed relays 



PHOTOGRAPHIC EQUIPMENT
Dullmeyer Prolection Lens $\mathbf{F}=65 \mathrm{mmi}, 35 \mathrm{~mm}$ mount $70 /=$ each, post $2 / 8$. Condenser Lenses Plano-Convex optically ground and notished $14^{\circ}$ dia. $21^{\circ}$ focus $7 / 6$ each. nost $2 / 6$ $21^{\prime \prime}$ dia $3^{\prime \prime}$ focus $10 /$ each. post $2 / 6.6^{\circ}$ dia. $10^{\prime \prime}$ focus $35 / 6$ post 4
LAMP HOUSES 230 volts 1,000 watls $10 /$ - ea., Thost $7 / 6$. 9 * square case. Ideal spotlight $70 /$ - each. 1post $10 /$. HIGH SPEED COUNTERS
$31 \times 1$ in.. 10
counts second with 4 flures. The
following DC following D.C.
voltaces
are voltages are
avallable. 6 ar
v..



SUB-MINIATURE Microswltch Honeywell S.P.D.T type 11 SM1 TN 13 size $8^{\circ} \times 1^{\circ} \times 1^{\circ} 6 / 6$ ea. or mounted in fives for $22 / 6$ post free.
DIGITAL INDICATOR. KGMM5 28 vt. 0 to $9,50 /-\mathrm{ea}$ SPEAKERS ELAC Sin. ROUND. 9700 Gauss. $3 \Omega$ JACK PLUGS. screw-on cover, $2 / 6$, point with Po 201 on head phone cord $3 /=$, most $1 / 6$. PLUG-IN RELAYS. Jondex 4 change-over HD contacta 28 V. D.C. or $240 V$. A.C. with base and cover, $35 /$ eea RELAYS, 24 voit DC, 4 make, of break heavy duty TRANSISTORS DIODES SCR'S ZENNERS VALVES ${ }^{\text {A82 } 200} 4 / 6 \quad 2 N 6985 /-\quad$ SX68 4/6 $181312 /-$

 LOWGLEY HOUSE LONGLEY RD. CROYDON SURREY

ROBUST AIRCRAFT PUSH 5C/898 of bakelite barrel tyne construction, With $18^{*}$ square 4 hole
fixing top with actual push below the level of a 1 " hakelite circle to prevent it being used accidentally Samples $5 / 6$ each lange nuantities available.
MAGNETIC COUNTERS Veeder Root with zero reset. 800 counts ner minute. counting to 099.999 zero METERS Volts AC or 110 volts DC. $65 /-$ each. prost $3 /$.
 Microamps $0 / 10024 \mathrm{in}$. MC $40 /-$ $\begin{array}{lll}\text { Microamps } & 0 / 500 & 2 \mathrm{in} . \text { MC } \\ \text { Microamps } 0 / 500 & 25 /- \\ 37 / 6\end{array}$ Milliamps $0 / 5002$ in. MC $37 / 6$ Millamps $0 / 5024 \mathrm{in}, \mathrm{MC}$. $35 /-$ Milliamps $0 / 500$ 3 4 in . MC Amps $5000-502 \operatorname{in}$.
Volts $5 / 0 / 521$ in. M Volts $0 / 202 \operatorname{in}$. MC 25/6 MICROAMPS Volts MICROAMPS $0 / 50$ gcaled in Rontgens 21 in . MC $45 /$ LEAK DETECTOR A.E.I, malns powered 635 er. PORTABLE YOLTMETERS 0/250 Moving Iron AC

ONE HOLE FIXING SWITCHES SINGLE POLE. Double Throw, 3 amp. 250 v. A.C. can be used as on/ OFF or CHANGE-OVER switch.
$18 /-$ per dozen, $130 / \mathrm{L}$ per 100 .

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\text { post } 2 /-. \quad \text { nost } 5 /-
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Precision made Contacta making and breaking twice per second in soundproof case with thermostat controlled heating. 12 or 24 v 8/6 post 6/
VISCONOL-CATHODRAY" CONDENSERS. $001 \mathrm{mfd} .10 \mathrm{kV}, 5 /-.002 \mathrm{mf}, 15 \mathrm{kV}, 9 /-: .02 \mathrm{mf}$. 10 kV . $1-; 6 \mathrm{kV}, 17 / 6: 0.5 \mathrm{mf} .2 .5 \mathrm{kV} .17 / 6: 1 \mathrm{mfd} .2 \mathrm{kV}$. $17 / 6$. RESISTORS, wire wound or carbon, yotentiometers, condensers, quantities ex-stock at low prices. BRIDGE MEGGERS SERIES I. With resistance

## LATEST RELEASE OF

## RCA COMMUNICATION RECEIVERS AR88



BRAND NEW and in original cases-A.C. mains input. 110V or 250 V . Freq. in 6 bands $535 \mathrm{Kc} / \mathrm{s}-32 \mathrm{Mc} / \mathrm{s}$. Output impedance 2.5-600 ohms. Complete with crystal filter, noise limiter, B.F.O., H.F. tone control, R.F. \& A.F. variable controls. Price $£ 87 / 10 /-$ each, carr. £2.
Same model as above in secondhand cond. (guaranteed working order), from $£ 45$ to $£ 60$, carr. $£ 2$.
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## MARCONI SIGNAL GENERATORS

## TYPE TF-I44G

Freq. $85 \mathrm{Kc} / \mathrm{s}-25 \mathrm{Mc} / \mathrm{s}$ in 8 ranges. Incremental: $+/-1 \%$ at $1 \mathrm{Mc} / \mathrm{s}$. Output: continuously variable 1 microvolt to 1 volt. Output Impedance: 1 microvolt to 100 millivolts, 10 ohms $100 \mathrm{mV}-1$ volt52.5 ohms. Internal Modulation: $400 \mathrm{c} / \mathrm{s}$ sinewave $75 \%$ depth. External Modulation: Direct or via internal amplifier. A.C. mains $200 / 250 \mathrm{~V}, 40-100 \mathrm{c} / \mathrm{s}$. Consumption approx. 40 watts. Measurements: $19 \ddagger \times 12 \ddagger \times 10 \mathrm{in}$. The above come complete with Mains Leads, Dummy Aerial with screened lead, and plugs. As New, in Manufacturer's cases, $£ 40$ each. Carr. $30 /$. DISCOUNT OF $10 \%$ FOR SCHOOLS, TECHNICAL COLLEGES, etc.
HRO RECEIVER. Model 5T. This is a famous American High Frequency superhet, suitable for CW, and MCW, reception crystal filter, with phasing control. AVC and signal strength meter. Freq. range $50 \mathrm{kc} / \mathrm{s}$. to $30 \mathrm{mc} / \mathrm{s}$. with set of nine coils. Complete HRO 5 T SET (Receiver, Coils and Power Unit) for $\mathbf{~} \mathbf{3 0}$, plus $\mathbf{3 0 / -}$ - carr.
COMMAND RECEIVERS; Model $6-9 \mathrm{Mc} / \mathrm{s}$., as new, price $£ 5 / 10 / \cdot \mathrm{cach}$, post 5/-.
COMMAND TRANSMITTERS, BC-458: 5.3-7 Mc/s, approx. 25 W output, directly calibrated. Valves $2 \times 1625$ PA; $1 \times 1626$ osc.; $1 \times 1629$ Tuning Indicator; Crystal $6,200 \mathrm{Kc} / \mathrm{s}$. New condition- $\mathrm{E} 3 / 10 /-$ each, $10 /$ post. (Conversion as per "Surplus Radio Conversion Manual, Vol. No. 2," by R. C. Evenson and O. R. Beach.)
BC-433G COMPASS RECEIVER; Freq. $200-1,750 \mathrm{Kc} / \mathrm{s}$. in 3 bands, suitable for aircraft, boats, etc. Complete with 15 valves, power supply input 24 v. D.C. at 2 amps. Receiver only $\& 5$ each, carr. 15/-.

ROTARY CONVERTERS: Type 8a, 24 v D.C., 115 v A.C.@ 1.8 amps , $400 \mathrm{c} / \mathrm{s} 3$ phase, $\mathbf{6} / 10 / \mathrm{e}$ each, $8 /=$ post. Converter 12 v D.C. input, 110 v A.C., $60 \mathrm{c} / \mathrm{s} 92.73 \mathrm{amps} .0 .300 \mathrm{Kva}, £ 15$ each, carr. £1. Converter 230 V D.C. input,
115 v output $60 \mathrm{c} / \mathrm{s}$ @ $2.73 \mathrm{amps} .0 .300 \mathrm{Kva}, £ 15$ each, carr. $£ 1.24 \mathrm{v}$ D.C. input, 115 v . output $60 \mathrm{c} / \mathrm{s} @ 2.73 \mathrm{amps} .0 .300 \mathrm{Kva}$, 15 ca
175 v D.C. @ 40 mA output, $25 /-$ each, post $2 / \mathrm{l}$.
CONDENSERS: $150 \mathrm{mfd}, 300 \mathrm{v}$ A.C., £7/10/- each, carr. $15 /-40 \mathrm{mfd}, 440 \mathrm{y}$ A.C. wkg., £5 each, $10 /-$ post. $30 \mathrm{mfd}, 600 \mathrm{v} w \mathrm{~kg}$. D.C., $£ 3 / 10 /-$ each, post $10 /-$ $15 \mathrm{mfd}, 330 \mathrm{v}$ A.C. wkg., $15 /-$ cach, post $5 /-.10 \mathrm{mfd}, 1000 \mathrm{v}, 12 / 6 \mathrm{each}$, post $2 / 6$. $10 \mathrm{mfd}, 600 \mathrm{v}, 8 / 6$ each, post $5 /-.8 \mathrm{mfd}, 1200 \mathrm{v}, 12 / 6$ each, post $3 /-.8 \mathrm{mfd}, 600 \mathrm{v}$,
$8 / 6$ each, post $2 / 6.4 \mathrm{mfd}, 3000 \mathrm{v}$ wkg., 33 each, post $7 / 6$. $2 \mathrm{mfd}, 3000 \mathrm{v}$ wkg., f2 each, post $7 / 6.0 .25 \mathrm{mfd}, 32,000 \mathrm{v}, 87 / 10 /-$ each, carr. $15 / \mathrm{-} .0 .25 \mathrm{mfd}, 2 \mathrm{Kv}, 4 /$ each, $1 / 6$ post.
AER1AL MASTS: 40 ft., complete with base, $£ 10$ each. Carr. £2.
RACK CABINETS: 6 ft . by 19 in ., and 16 in . depth, with rear door and safety switch, $\mathbf{~ 5}$, cars. £2.
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COSSAR 1035 OSCILLOSCOPE, £ 30 each, $30 /$ - carr. RELAYS: Relay Unit (with 9 American relays) 24 v. D.C., 250 ohm coils.
heavy duty, M. \& B. $30 /-$ each, $4 /$ post. GPO Type 600,10 relays (a) 300 heavy duty, M. \& B. $30 /$ - each, 51 - post. GPO with 2 M and 10 relays @ 50 ohms with 1 M ., 22 each, $6 /-$ post. 12 Small American Relays, mixed types £2, post $4 /=$.

CALIBRATION TACHOMETER Mk. II: Maxwell Bridge Type 6C/869 25 each, £. 2 carr
ROTAX VARIAC \& METER UNIT: Type 5G.3281. Reading 0-40 v., 0-40 mA and 0.5 amps., all on 275 deg. scales, $£ 30$ each, $£ 2$ carr.
HEWLETT PACKARD TYPE $400 \mathrm{C}: 115 \mathrm{v} \cdot 230 \mathrm{v}$. input $50 / 60 \mathrm{c} / \mathrm{s}$. Freq. range $20 \mathrm{c} / \mathrm{s}-2 \mathrm{Mc} / \mathrm{s}$. Voltage range: $1 \mathrm{mV}-300 \mathrm{v}$. in 12 ranges. Input impedance 10 megohms. Designed for rack mounting, $£ 30$ each, carr. $15 /-$.
TCS MODUI.ATION TRANSFORMERS, 20 watts, pr. 6,000 C.T., sec. 6,000 ohms. Price 25/-, post 5/-.
AUTOMATIC PILOT UNIT Mk. 2. This complex unit of diodes and valves, relays, magnetic clutches, motors and plug-in amplifiers, with many other items, relays, magnetic clutches,
price $£ 7 / 10 /-$, $£ 1$ carriage.

FOR EXPORT ONLY: B. 44 Trans-ceiver Mk. IIl. Crystal control, 60$95 \mathrm{Mc} / \mathrm{s}$. AMERICAN EQUIPMENT: 5C-640 Transmitter, 100 , 156 $\mathrm{Mc} / \mathrm{s}$., 50 watt output. For 110 or 230 v . operation. ARC 27 trans-ceivers,
28 v , D. input. Also have associated equipment. BC-375 Transmitter. 28 v. D.C. input. Also have associated equipment. BC-375 Transmitter.
BC-778 Dinghy transmitter. SCR-522 trans-ceiver. Power supply, PP893 BC-778 Dinghy transmitter. SCR-522 trans-ceiver. Power supply, PP893/
GRC 32A; Filter D.C. Power Supply F-170/GRC 32A: Cabinet Electrical GRC 32A; Filter D.C. Power Supply F-170/GRC 32A: Cabinet Electrical
CY 1288/GRC 32A; Antenna Box Base and Cables CY 728/GRC; Mast Erection Kits, 1186/GRC; Directional Antenna CRD.6; Comparator Unit, CM.23; Directional Control CRD.6, $567 / \mathrm{CRD}$ and $568 / \mathrm{CRD}$; Azimuth TS.622/U.

VARIABLE POWER UNIT: complete with Zenith variac 0-230 v.9 9 amps.;
2 tin. scale meter reading $0-250 \mathrm{v}$. Unit is mounted in 19in, rack, £16/10/- each, 2 tin .8 scal
$30 / \mathrm{carr}$.
SOLENOID UNIT: 230 v. A.C. input, 2 pole, 15 amp contacts, £2/10/- each post $6 /$-.
CONTROLPANEL: 230 v. A.C., 24 v. D.C. @ $2 \mathrm{amps} .$, £ $2 / 10 /=$ cach, carr. $12 / 6$. AUTO TRANSFORMER: 230-115 v.; 1,000 w. \&5 each, carr. 12/6. 230-115 v.; 300VA, \&3 each, carr. 10/-
OHMITE VARIABLE RESISTOR: $5 \mathrm{ohms}, 5 \frac{1}{2} \mathrm{amps}$; or 2.6 ohms at 4 amps . Price (either type) $\& 2$ each, $4 / 6$ post each.
POWER SUPPLY UNIT PN-12B: 230 v. A.C. input, 395-0-395 v. output @ 300 mA . Complete with two $\times 9 \mathrm{H}$ chokes and 10 mfd . oil filled capacitors. Mounted in 19 in . panel, $£ 6 / 10 /-$ each, $£ 1$ carr.
TX DRIVER UNIT: Freq. $100-156 \mathrm{Mc} / \mathrm{s}$. Valves $3 \times 3 \mathrm{C} 24$ 's; complete wit filament transformer 230 v . A.C. Mounted in 19 in . pantl, £4/10/- ea $: \mathrm{h}, 15 /$ - carr POWER UNIT: 110 v . or 230 v . input switched; 28 v . @ 45 amps . D.C. output. Wt. approx. 100 lbs , $£ 17 / 10 /$ - each, $30 /$-carr. SMOOTHING UNITS suitable

## Signal generators:

MARCONI TF-144G : freq. $85 \mathrm{Kc} / \mathrm{s}-25 \mathrm{Mc} / \mathrm{s}$, internal and external modulation, power supplies $200 / 250 \mathrm{v}$. A.C. (secondhand cond), price $£ 25$ ea.; carr. 30/-.
CT53. Freq, range 8.9-300 Mc/s. with Calibration chart. Output $1 \mu \mathrm{~V}-100 \mathrm{mV}$. internal square wave and sinewave modulation at $100 \mathrm{c} / \mathrm{s}$., external modulation $50 \mathrm{c} / \mathrm{s}-10 \mathrm{Kc} / \mathrm{s}$, , 230 v . A.C. Complete with chart, etc., price $£ 27 / 10 /$ - ea., carr. £1.

MARCONI CT. 480 and 478: 1.3-4.2 Mc/s., F.M. or A.M., price £75 each, carr. 30/-.

NIFE BATTERIES: 6 v .75 amps ., new, in cases, £15 each, £1 carr.; 4 v .160 amps, new, in cases, $£ 20$ each, $£ 110 /$ carr. L.R. 7 Cells, only 1.2 v .75 amps. new, $£ 3$ each, $12 /=$ carr. The above batteries are low resistance designed to give a heavy surge for starting and can be stored for long periods without any effect to their performance.

FUEL INDICATOR Type 113R: 24 v . complete with 2 magnetic counters $0-9999$, with locking and reset controls mounted in a 3 in . diameter case. Price 0-9999, with locking a
$30 /-$ each, postage $5 /-$.

UNISELECTORS (ex equipment): 5 Bank, $50 \mathrm{Way}, 75 \mathrm{ohm}$ Coil, alternate wipe, £2/5/- each, post 4/

FREQUENCY METERS: LM13 or BC-221; $125-20,000 \mathrm{Kc} / \mathrm{s}$., £25 each., carr. 15/-. TS.175/U, £75 each, carr. £1. TS323/UR, 20-450 Mic/s., £75 each, carr. 15/-. FR-67/U: This instrument is direct reading and the results are presented directly in digital form. Counting rate: $20-100,000$ events per sec. Time Base
Crystal Freq.: $100 \mathrm{Kc} / \mathrm{s}$. per sec. Power supply: 115 v ., $50 / 60 \mathrm{c} / \mathrm{s}$., $\mathbf{i} 100$ each, Crystal
carr E 1.
CT. 49 ABSORPTION AUDIO FREQUENCY METER: freq. range $450 \mathrm{c} / \mathrm{s}-$ $22 \mathrm{Kc} / \mathrm{s}$., directly calibrated. Power supply $1.5 \mathrm{v} .-22 \mathrm{v} . \mathrm{D} . \mathrm{C} . \mathrm{£} 12 / 10 /-\mathrm{cach}$, carr. 15/-.
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APNI ALTIMETER TRANS./REC., suitable for conversion $420 \mathrm{Mc} / \mathrm{s}$, complete with all valves 28 v. D.C. 3 relays, 11 valves, price $£ 3$ each, carr. $10 /-$.

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CONTROLS: Selector Switch. Tape
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| $2 \mu \mathrm{~F}$ | 350 volt | $8 \mu \mathrm{~F}$ | 12 volt | $50 \mu \mathrm{~F}$ | 6 volt | All at $1 /-$ each. |  |
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$1 \mu \mathrm{~F}, 2 / 8$.
$400 \mathrm{~V}=1,000,1,500,2,200,3,300,4,700 \mathrm{pF}, 6 \mathrm{~d}$.
$0,800 \mathrm{pF}, 0.01,0.015,0.022 \mu \mathrm{~F}, 7 \mathrm{~d}$. $0.033 \mu \mathrm{~F}, 8 \mathrm{~d}$. $0.047 \mu \mathrm{~F}, 9 \mathrm{~d}$. $0.068,0.1 \mu \mathrm{~F}$, $11 \mathrm{~d} .0 .15 \mu \mathrm{~F}, 1 / 2.0 .22 \mu \mathrm{~F}, 1 / 6.0 .33 \mu \mathrm{~F}$. 2/3. $0.47 \mu$ F, $2 / 8$.
Modular, metallised, P.C. mounting, $20 \%, 250 \mathrm{~V}: 0.01,0.015,0.022 \mu \mathrm{~F}$, 7d. $0.033,0.047 \mu \mathrm{~F}, 8 \mathrm{~d}$. $0.068,0.1 \mu \mathrm{~F}, 9 \mathrm{~d}$. $0.15 \mu \mathrm{~F}$, $11 \mathrm{~d} .0 .22 \mu \mathrm{~F}, 1 /-.0 .33 \mu \mathrm{~F}, \mathrm{I} / 5$. .47 $\mathrm{F}, 1 / 8.0 .68 \mu \mathrm{~F}, 2 / 3$. $\mathrm{I}_{\mu \mathrm{F}, 2 / 9,}$

POLYSTYRENE CAPACITORS: $5 \%$, 160 V (unencapsulazed): $10,12,15$. 18, 22, 27, 33, 39, 47, 56, 68, 82, 100, 120, 150, 180, 220, 270, 330, 390, 470, 560, $680,820 \mathrm{pF}$, 5 d . $1,000,1,500,2,200 \mathrm{pF}$, $6 \mathrm{~d} .3,300,4,700,5,600 \mathrm{pF}, 7 \mathrm{~d} .6,800$ $8,200,10,000 \mathrm{pF}$, $8 \mathrm{~d} .15,000,22,000 \mathrm{pF}$, 9 d .
$1 \%$. 100 V (encapsulaced): $100,120,150,180,220,270,330,390,470,500,560$, $680,820 \mathrm{pF}$. $1 /$, $1,000,1.200,1,500,1,800,2,200,2,700,3,300,3,900 \mathrm{pF}, 1 / 3$. $4.700,5,000,5,600,6.800,8,200,10,000,12,000,15,000 \mathrm{pF}$. $1 / 6.18,000,22.000$ $27,000,33,000,39,000 \mathrm{pF}, 1 / 9.0 .047,5,000,0.056 \mu \mathrm{~F}, 21.00 .068,0.082,0.1 \mu \mathrm{~F}$, $2 / 3.0 .12 \mu \mathrm{~F}, 2 / \mathrm{i}, 0.15,0.18 \mu \mathrm{~F}, 3 /=0.22 \mu \mathrm{~F}, 4 /=0.27,0.33 \mu \mathrm{~F}, 5 /-0.39 \mu \mathrm{~F}, 5 / 9$.

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Miniature: 0.3 W at $70^{\circ} \mathrm{C} . \pm 20 \% \leqq\left\{\mathrm{M}_{\mathrm{l}} \pm 30 \%>\{\mathrm{M}\right.$. Horizontal $(0.7 \mathrm{mn} . \times$
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RESISTORS (Carbon film), very low noise. Range: $5 \%, 4.7 \Omega$ to IM $\Omega$ (E24 Series) $10 \%$ i $10 \Omega$ to IOM $\Omega$ (EI2 Series)
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 800 P.IIV., 3/3. (6A): 200 P.I.V., 3/-, 400 P.I.V., 4/-. 600 P.I.V., 5/-. 800 P.I.V.. $6 /-$. THYRISTORS (5A): 100 P.I.V., 8/-. 200 P.I.V., $10 /-, 400$ P.IVV., is/-.
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Send S.A.E. for January, 1969 Cafalogue

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BOONTONSTANDARD SIGNAL GENERATOR MODEL Bo. Frequency ${ }^{2}$ $400 \mathrm{Mc} / \mathrm{s}$. in 6 ranges. AM. 400 and $1,000 \mathrm{c} / \mathrm{s}$. and external modulation. Provision for pype atrenuzer $0.1,-100 \mathrm{mV}$ separate meter for modu separate meter for modu Precision flywheel tuning 117 v. A.C. input. With in. seruction manual, E95. Car riage 30/-.


OOLARAD UHF SIGNAL GENERATOR. Frequency $950 \mathrm{mc} / \mathrm{s} /$ $2.400 \mathrm{mc} / \mathrm{s}$ in one range. Acsenuator 0.1 $\mathrm{mV}-200 \mathrm{mV}$. Sync. selector internal square wave, sin $\ddot{\text { pos }}$ positive and negative race mulsiplyer
$30-420 \mathrm{c} / \mathrm{s}$. Pulse delay $2.5-350 \mathrm{u} / \mathrm{sec}$. Pulse width 5 microsec (incorporating square wave switch). Modulation: positive and negative, fllo. Carriage posit
As above but frequency $3,830-11,050$ $\mathrm{mc} / \mathrm{s}$, counser read our, pulse delay XI , $\times 10$ and $X 100$ at 2.20 microsecs. Pulse rate $\times 10, \times 100, \times 1,000$ at $1-10 \mathrm{c} / \mathrm{s}$. Cl65. Carriage 30/.

MARCONI SIGNAL GENERATOR TYPE TF $144 \mathrm{G} .85 \mathrm{kc} / \mathrm{s} .-25 \mathrm{Mc} / \mathrm{s}$. Excellent laboratory tested condition. with all necessary accessories with instruction manual, \&45. P. \& P. $15 / \%$.

MARCONI SIGNAL GENERATOR TF 801/A/I. $10-300 \mathrm{Mc} / \mathrm{s}$. in 4 bands. Internal at $400 \mathrm{c} / \mathrm{s}$. I ke/s. External $50 \mathrm{c} / \mathrm{s}$ to $10 \mathrm{kc} / \mathrm{s}$. Output $0-100 \mathrm{db}$ below 200 mV from 75 ohms source. C85. P. \& P. 20/-, including necessary connectors, plugs, and in. scruccion manual.

BROADBENT MICROWAVE SIGNAL GENERATOR TYPE 903. Frequency range $6.800-11,000 \mathrm{mc} / \mathrm{s}$, $\mathrm{c} / \mathrm{s}$ and $\times 10$ multiplyer, delay $3-300$ $\mathrm{U} / \mathrm{sec}$. Width .05 to $10 \mathrm{U} / \mathrm{sec}$. Input for external syncronisation and modulation. Output delayed and undelayed syncronised directly calbraced attenuator. C85. Carraige 30/

DAWE VALVE VOLT METER TYPE 613B. Range $0.03 v$ to 300 v in nine ranges. Frequency $20 \mathrm{c} / \mathrm{s}$ to 2 $50 \mathrm{c} / \mathrm{s}$ i $17 / 10 /$ - Carriage $30 /$

SOLATRON LABORATORY REGULATED POWER UNIT MODEL SRS 15I A. Variable voltage, positive output: 20-250v; 250/500v $0-170 \mathrm{y}$ (metered). Nogative output $0-170 \mathrm{y}$
(unmetered). Fixed negative ourput 170 v . Two separate 6.3 v and 5 amp outputs. Volts -mA meter switch. H.T. Safety cutout. 200/250v A.C. 50 c/s. 645, Carriage 30/.

MARCONI VIDEO OSCILLATOR TF 885A. Sine wave output $25 \mathrm{c} / \mathrm{s}$ co $5 \mathrm{Mc} / \mathrm{s}$ in 2 bands. Squarewave output $50 \mathrm{c} / \mathrm{s}$ to $150 \mathrm{c} / \mathrm{s}$ in 2 bands. Freq. accur. $\pm 2 \% \pm 2 \mathrm{c} / \mathrm{s}$. Power supply $100 / 125 /$ 200/250 v. A.C. 655. (Ditto but 25/12 me/s in 3 bands/885A/I). C85. Carriage

PRECISION VHF FREQUENCY METER TYPE 183. 20-300 Mc/s wish wath accuracy $0.3 \%$. Additional band on harmonics $5.0 .6 .25 \mathrm{Mc} / \mathrm{s}$ with accuracy $+-2 \times 10^{-4}$. Incorporating calibrating quarta $100 \mathrm{kc} / \mathrm{s}+-5 \times$
$10-120 / 220$ v. A.C. mains.
C85. 10- ${ }^{-120 / 220}$
Carriage $E 2$.

AIRMEC FREQUENCY STAND. ARD METER TYPE 761. 10c. 100c, lokc, lookc. IMc. E80. Carriage 30\%

COSSOR OSCILLOSCOPE TYPE 1049. 645. Carrlage 30\%

Fuller descriptions of the following upon request.

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COMPLETE WITH P.S.U. HEWLETT-PACKARD ELEC. TRONIC COUNTING UNIT.

MICROWAVE SPECTRUM ANA LYZER TYPE SA 18 MANUFAC. TURED BY RACAL.

OAWE STORAGE OSCILLO. SCOPE TOGETHER WITH TRACE SHIFTER.

SIGNAL GENERATOR CT 218 (FM/AM) MARCONI TF 937. $85 \mathrm{kc} / \mathrm{s}$ to $30 \mathrm{mc} / \mathrm{s}$ in 8 ranges. Output level variable in I db steps from $1 \mu \mathrm{~V}$ to 100 mV into 75 ohms. Also 1 vole outputs down to $0.1 \mu \mathrm{~V}$ into 7.5 ohms. and $3 \mathrm{kc} / \mathrm{s}$. Varlable mod. depths and andiation. Crystal calibrator $200 \mathrm{kc} / \mathrm{s}$ and $2 \mathrm{mc} / \mathrm{s}$. F.M. at frequencies above $394 \mathrm{ke} / \mathrm{s}$. Monitor speaker for beat detection. Panclimatic. 100 to 150,200 to 250 V A.C. 45 to $100 \mathrm{e} / \mathrm{s}$. Weight $117^{\prime \prime} \mathrm{lbs}$. Measurements $17^{\prime \prime} \times 201^{\circ} \times$ 171". E85. Carriage 30/.

WINOSOR MODEL I50A OUT. PUT POWER METER. 5 mW to 5 W F.S.D., 2.5 to 20.000 ohms. $\mathrm{C} 15 / 10 / \mathrm{l}$ Fost and packing i5/.

BOONTON " Q " METER TYPE 150 A . Frequency range $50 \mathrm{kc} / \mathrm{s}$ so $50 \mathrm{me} / \mathrm{s} . "$ Q" range $0-250$ with mulciplier of 2.5 . Main tuning capacitor
$30-500 \mathrm{pF}$ with separate $\pm 3 \mathrm{pF}$ incer. polating capacitor. Power supply $220 / 250 \mathrm{vAC}$, 675 . Carrlage $30 \%$.

AVO VALVE TESTER MODEL 3. Measurement of mutual conductance $0-100 \mathrm{~mA} / \mathrm{V}$ in four ranges. Screen $0-300 \mathrm{v}$., panelled $0-400 \mathrm{v}$., grid $0 /-100 \mathrm{v}$, filament 0/126v. Insulation 0/10m ohms. Rectifying valves and signal diodes can be tested under load con disions, short circuiting of electrode and cachode insulation can also be measured. Complete with data book (aty. Carrize 301.

FURZEHILL SENSITIVE VALVE VOLTMETERTYPE 378 B/2. Accurate measuring AF and MF voltazes up to to 100 v . (full scale). Logarithmetically divided. A db scale provided for $0-20$ $\mathrm{db}, 0 \mathrm{db}$ being 1 mV . Automatically set zero for every range. A jack is provided for monitoring the input signal If required. 220/250v. A.C. $627 / 10 /$. Post and packing 10/.

END OF RANGE: VOLTMETER Type T.F.428. E9. Carriage 10/.

SIGNAL GENERATOR. Type C.T.53. Without chart 610 , with chart C.T.53. Without ch

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SHEpherd's Bush 4946

VALVES





| $8 / 6$ |
| :---: |
| $7 /-$ |
| $9 /-$ |
| $9 /-$ |
| $3 \cdot 10$ |
| $19 /-$ |
| $6-40$ |
| $85 /-$ |
| $18 \cdot 40 \mathrm{~A}$ |
| $100 /-$ |
| $0-15$ |
| $7 / 6$ |
| $108 /-$ |
| $010 /-$ |
| $8 /-$ |
| $8 /-$ |
| $4-16$ |
| $57 / 6$ |
| $17 / 6$ |
| $8 / 6$ |
| $7 / 8$ |


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## TRANSISTORS, ZENER

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tee valve with a GUARANTEE
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D.C. MOVING COIL METERS
$100 \mu \mathrm{~A}$. 3 in. square panel colib $0-2 w \ldots$ $200 \mu \mathrm{~A}$. 2 in . round panel, sealed calibro-30 22/6 $200 \mu \mathrm{~A}$. $2 \frac{1}{\mathrm{f}} \mathrm{in}$. round panel.
$500, \perp$ A. 2 in. round proi.
$500 \mu \mathrm{~A}$. 2 in . round panel
$750-0-750 \mu \mathrm{~A} .2 \mathrm{in}$. round plug-in
$1 \mathrm{~mA} .2 \frac{1}{\mathrm{in}}$. round panel.
1 mA .2 in . round panel sealed
5 mA .2 in . round clip-fix panel or proj.
$10-0-10 \mathrm{~mA}$. 2 l in . round panel
0 $-30 \mathrm{~mA} .2 \frac{1}{\mathrm{in}}$. round panel
$75 \mathrm{~mA} .2 \frac{1}{1} \mathrm{in}$. plug in
100 mA . Itin. proj.
100 mA . $1 \frac{1}{2} \mathrm{in}$. round panel
$100 \mathrm{~mA} .2 \frac{1}{2} \mathrm{in}$. round panel ................... 500 mA .21 in . round panel
2 amp .2 in . round panel
$25 \mathrm{amp} .3 \frac{1}{2} \mathrm{in}$. round proj.
$50 \mathrm{amp} .2 \frac{1}{2} \mathrm{in}$. round panel
$0-1.5 \mathrm{~V} \& 0-150 \mathrm{~V} 3$ eerminals round panel
20 VDC $2 i n$. square panel
100 V 4 in . round panel
150 VDC 4 in . round panel
150-0-1500 mA. $3 \frac{1}{2} \mathrm{in}$. round panel
1.5 KV with res. 2 in . round panel.

## MOVING IRON METERS

15 VAC 2 itin. round panel
500 VAC $2 \frac{1}{2}$ in. round clip fix
miniature meters.
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\begin{aligned}
& 65 \mathrm{~mA} . \text { D.C. } 18 / \mathrm{l} .3 \% \\
& 150 \mathrm{~mA} \text {. D.C., } 15 /=
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"S" METER FOR H.R.O. RE-
PURE. MINIATURE SUB - MINIATURE "PENNY SIZE" METERS. lin. round, flush ring nut mounted $500 \mu \mathrm{~A}$ FSD, cal NULL METER (unscaled 25-0-25 micro-amps $3!n$ square panel. 45/-.
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LOSCOPE 2 isin. tube $220 / 250$ CIL-
E22 l0s. Carriage 30/*
HEWLETT-PACKARD VTVM
MODEL 410B. A compact instrument for A.C., D.C. and resistance measure-
ments with A.C. meas. in excess of ments with A.C. meas. in excess of
$100 \mathrm{mc} / \mathrm{s}$; D.C. polarity reverse facility and 7 resistance ranges. Large easily and 7 resistance ranges. Large easily
readable meter. A.C. range $1-300 \mathrm{~V}$ in 6 ranges. D.C. range $1,1,000 \mathrm{~V}$ in 7 ranges. Resistant 0-50 Mohms. E22/10/-. Carriage 15/
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Precision engineered light source
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| P. $\&$ P. |

VAN DE GRAAF ELECTROSTATIC GENERATOR, fitted with motor drive for 230 v. A.C. giving a potential of approx.
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number of incerescing experiments, and full | instructions. This inserument is |
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200/250 v. AC HORSTMAN 20AMP TIME SWITCH

2 on/off every 24 hrs . at any pre-set cime
Fitced in metal case 36 hr . \$pring reserve Used bue fully tested. Fraction of maker' price. $\mathbf{6 3 . 1 9 . 6}$ plus 4/6d. post and pack.
 COPPER LAMINATE PRINTED CIRCUIT BOARD. Large sheet $151 \times 51 \mathrm{in}$. Price $3 / 9,3$ for $10 /$ post paid.

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220 v ., $110 \mathrm{v} ., 115 \mathrm{v}$. $220 \mathrm{v.},, 10 \mathrm{v},. 115 \mathrm{v}$.
6 . 12,36 , 110 v
A.C. These coils are also used for D.C. \& 19 plus 15/-carr. Leaflet on request.


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AUTO TRANSFORMERS. Step up, step down. $110-200-220-240 \mathrm{~F}$. Fully shrouded. New. 300 wat P. \& P. 6/6. 1,000 watt type, $65 / 5 / \mathrm{e}$ each. P. \& P. $7 / 6$. LEVER MICRO SWITCH Brand new lever operated micro switch. 5 for $\& 1$ post paid.

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220/240v. A.C. COOLING UNIT
2,300 r.p.m. Gin. blade size. Smooth powerful motor. All metal constructested. Offered at fraction of maker's price, 2/15/-. P. \& P. 7/6.
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25 WATT
O/25/50/100/250/500/1,000/1,500/2,500 ohm, 14/6, P. \& P. $1 / 6$.
Black Silver Skirted knob calibrated in Nos. 1-9. 1/ $\frac{1}{4}$
in. dia. brass bush. Ideal for above Rheostats, $3 / 6$ each.

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Type 2. 28 r.p.m. corque 20 lb . In rever-
sible $1 / 80$ th h.p. 50 cycle $\cdot 28$

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& 28 \text { r.p.m. corque } 20 \mathrm{lb} \text {. in rever- } \\
& \text { sible } 1 / 80 \text { th h.p. } 50 \text { cycle } 28
\end{aligned}
$$



The above two precision made U.S.A. motors are offered in 'as new' condition. Input voltage of motor 115 v A.C. Supplied
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Price, either type $£ 2.17 .6$ plus $6 / 6$ P. \& P. or less crans ormer $\mathbf{£ 2 . 2 . 6}$ plus $4 / 6 \mathrm{~d}$. P. \& $P$.
$230 / 250$ V. A.C. SOLENOID Heavy ducy eype. Approx. 31b. pull $7 / 6$ plus $2 / 6$ P. \& POLENOID
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$4 \times .5$ volt unit series con at 20 mA . in sunlight,
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Earch Sacellites, 451-. P. \& P. P. $1 / 60$
PRECISION INTERVAL TIMER From 0-30 seconds (repecitive). Jewelled
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Latest American. New. Plastic THYRISTOR LATEST TYPE SELENIUM BRIDGE RECTIFIERS | 30 vole 3 amp. $11 / /$, plus $2 / 1 / \mathrm{P}$. \& P |
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Sintered Cadmium Type 1.2 v . 7AH. Size: height . Weight: approx 13 ozs. A.C. AMMETERS 0.1, 0-5, 0.10, 0-15, 0-20 amp. F.R 2 in . dia. All at 21 each. $\quad$. C . 0 - 50 v., $0-150 \mathrm{v}$. M. 2 in . Flush round all at $21 /-$ each. P. \& P. extram

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> 25 ohm coil, 24 v. D.C. $E 5 / 17 / 6$, plus $2 / 6$. P. \&.

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| 280 | $6-12$ | 2 co | $14 / 6$ |
| 280 | 9-18 | 4 co | 15/6 |
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## Radiomobile <br> BRITAIN'S CAR RADIO SPECIALISTS <br> Have a vacancy for a fully experienced <br> SERVICE ENGINEER <br> The successful applicant will be employed in our Main Service Workshop repairing :- <br> Transistorised \& Valve Operated Car Radios. Car Tape Recorders. <br> Coach Radio \& P.A. Equipment <br> and also in our Service Garage on installation work and the servicing of equipment already fitted to vehicles. <br> After gaining considerable knowledge of our products. it is possible that an opportunity might arise in the future for duties to be extended to include Field Service Work. <br> Applications should be made in writing 10: The Personnel Manager, <br> Radiomobile Limited, <br> Goodwood Works, <br> North Circular Road, <br> London, N.W 2. <br> GLA 0171. Ext. 4335 <br> A Subsidiary of SMITHS INDUSTRIES LTO.



## HAWKER SIDDELEY

 AVIATION LIMITED
## at DUNSFOLD AERODROME TECHNICIANS

are required for the testing and maintenance of aircraft navigation equipment.
Experience of electronic/electro-mechanical servo mechanisms essential.
Contributory Pension and Life Assurance Scheme, Sports and Social Club, Staff Canteen.
Please telephone or write quoting P7MW/05 to:-
Personnel Officer,
Hawker Siddeley Aviation Limited,
Dunsfold Aerodrome,
Nr. Godalming, Surrey.
Telephone Cranleigh 2121
BBO Transmitter Engineers
30 engineers are required for appointment between December, 1968 and April, 1969, at transmitting stations in various parts of the country. Their duties will mainly be concerned with UHF television transmission in colour and the VHF/FM radio services. Training in transmitter engineering techniques will be provided and previous experience in this field is, therefore, not essential.

## ESSENTIAL QUALIFICATIONS:

Higher National Certificate in Electrical Engineering (Light Current) or
City and Guilds Full Technological Certificate (Telecommunication)
and the ability to demonstrate a good practical knowledge of electronic principles. Applicants must be British subjects, and have normal colour vision. Starting salary $£ 1,130-£ 1,475$ p.a. depending upon experience. These posts are permanent and pensionable. The transfer of pension rights from another scheme can usually be arranged.

For further details and application form write to:
The Engineering Recruitment Officer, BRITISH BROADCASTING CORPORATION, Broadcasting House, London, W1A 1AA.

enjoy exciting new scope now in Air Traffic Control

There are opportunities in the National Air Traffic Control Service, a Department of the Board of Trade, for you to play a vital part in the safety of Civil Aviation. You'll work on the latest equipment including Computers, Radar and DataExtraction, Automatic Landing Systems and Closed-Circuit Television, at Civil Airports, Air Traffic Control Centres, Radar Stations and other engineering establishments, including Heathrow, Gatwick and Stansted.

If you are 19 or over, with practical experience in at least one of the main branches of telecommunications, fill in the coupon now. Your starting salary would be £869 (at 19) to £1,130 (at 25 or over); scale maximum $£ 1,304$ (rates are higher at Heathrow). Non-contributory pensions for established staff.

Career Prospects. Your prospects are excellent, with opportunities to study for higher qualifications in this expanding field.

Apply today, for full details and application form.


The Division designs, manufactures and installs cables to meet expanding worldwide demands for telephone systems. New types of cable and associated manufacturing techniques are under constant development for local distribution networks and trunk circuits at home and overseas. There are immediate vacancies in the Engineering organisation at Prescot, Lancs.

## SYSTEMS ENGINEERS

Experienced Telecommunications Engineers are required for the Systems Engineering Department. Systems Engineers are responsible for the interpretation of customers' system requirements and the preparation of engineering submissions for tenders. The posts are of senior status and require a high standard of individual responsibility and initiative. Some travelling in the U.K. and abroad is involved.

Dur Systems Engineers must possess the qualifications of a chartered engineer. In addition we are looking for people who are really interested in tele. communication engineering, are personable in their relations with others and can write clear technical reports. Previous experience of the cable aspects of telecomm. systems is not necessary, as long as a sound basic knowledge exists. Training in our own field can be given after appointment, if necessary.

These posts provide a good route into management level appointments on the engineering or commercial sides of the Company at large.

## INSTRUMENT DEVELOPMDNT GNGINEER

An engineer is required by the Instrumentation and Control Department to join a small team engaged on the design and development of instrumentation for monitoring cable quality in its production and test stages. The work also involves the design of plant control equipment embodying closed loop control techniques, and offers plenty of scope for imaginative thinking and purposeful application.

We should prefer this engineer to be a graduate in electrical engineering but shall be glad to consider anyone with H.N.C. Previous experience of plant instrumentation design would be valuable, especially in the field of closed loop control systems.

Telephone Cables Division is located in modern factory premises with its own laboratories at Prescot, Lancs. As part of the BICC Group we provide the conditions of service of a large progressive Company. including a Profits Participation Scheme.

Please write, giving a usefully broad outline of your career to date, to:-
G. F. Turner (Ref. 38/86)

Telephone Cables Division
British Insulated Callender's Cables Ltd.,
PRESCOT, Lancs. L34 5WO.

## APPOINTMENTS

SOUTHERN
require a SENIOR ENGINEER (VIDEO TAPE RECORDING)

The successful applicant will be based at the Southampton Studios and will be required to work shift duties

Applicants must have conisiderable experience with Ampex VR-2000 video tape equipment and have proved operational ability. Technical qualifications of at least Higher National Certificate are required and candidates must be in good health and have normal colour vision.

The post commands a salary of $£ 2.166$ per annum (A.C.T.T. Grade B) and the Company operates a first class contributory pension scheme.

## Applications in writing to:

The Personnel Officer, Southern Independent Television, Northam, Southampton, SO9 4Y0

# SVSTEMS TEST ENGINEERS RADAR and TV 

THE JOB
Systems Test Engineering on advanced training aids for aircraft including simulation of Radar and using Closed Circuit Colour Television.

THE MEN
Electronic Engineers preferably with O.N.C. or H.N.C. having had practical experience of Radar or Television equipment who have a keen desire to learn new techniques and applications.

## THE REWARDS

Competitive salaries will be paid. High job interest. The opportunity to work on complex systems incorporating digital and analogue computers and associated peripherals, as a member of a team.
Opportunity to fly and operate simulated aircraft and other equipments. High quality training will be given.

## OTHER BENEFITS

Our terms and conditions of employment are good and include contributory Pension Scheme and free Life Assurance. Good welfare benefits. We offer long term careers, not short term jobs. Opportunities for limited travel since we export $80 \%$ of our products.

Apply, quoting reference $W W / 269$, to:
Personnel Manager. REDIFON LIMITED,
FLIGHT SIMULATOR DIVISION.
Gatwick Road, Crawley, Sussex.
Telephone: Crawley 28811


## SERVICE TECHNICIANS

Experienced electronic engineers, minimum qualifications ONC/City \& Guilds, to service and repair a wide range of electro-acoustic instruments. Driving experience essential.
Excellent salary and opportunities for advancement. Write or telephone for immediate interview.
Personnel Department, Amplivox Ltd., Beresford Avenue, Wembley. Tel.: 01-902 8991

UNIVERSITY INSTITUTE OF TECHNOLOGY
A constituent college of the University of Wales DEPARTMENT OF APPLIED PHYSICS
M.Sc./DIPLOMA COURSE
IN ELECTRONICS

Applications are invited for places in the full-time one-year M.Sc./Diploma course in Electronics, commencing 29th Seplember, 1969.

Application forms (together with further details) can be oblained from, and must be returned to the Registrar, University Institute of Technology, Cardiff, CFI 3 NU as soon as possible.

## ENTHUSIASTS

Have you considered a career in Technical Authorship? If you have sound experience in electronics or communications and ability to write clear concise English we would train applicants as Technical Authors. The commencing saláries range from $£ 1,450$ to £1,800 depending on experience with the prospects of high future rewards and earnings. Box No. 5052.

## TECHNICIANS

The Margaret McMillan College of Education require the following:
Audio Visual Aids Technician-Grade 2 -to be responsible for the maintainance and repair of audio visual aids equipment. An interest in general craftwork and model making would be an advantage.
Junior Technician Grade I-to assist in the maintainance of equipment and making of teaching aids. An interest in craft, e.g. woodwork, would be an advantage and candidates should preferably have 3 G.C.E. ('O' level) passes in appropriate subjects.

Salaries are in accordance with the N.J.C. Technicians Grades namely:

Grade 2- $\mathbf{E 7 6 5}$ to $£ 895$
Grade I-£330 to $£ 765$
Plus a qualification allowance of either $£ 30$ or $£ 50$ per annum where appropriate.

Application forms and further details of the posts may be obtained from the Principal, Margaret McMillan College of Education, Trinity Road, Bradford 5, and should be returned within 14 days of the appearance of the advertisement.

## KODE LTD.

Data Processing Equipment
KODE LTD. require
ELECTRONIC SERVICE ENGINEERS
for the Greater London Area.
The applicants with semiconductor and/or valve experience, capable of organising and fulfilling service loading independently, will secure excellent salaries, company vehicles (for business and private use), together with very real opportunities for advancement.

Apply with summary of career to :-
The Service Manager
Kode Limited, Calne, Wiltshire.
Tel. Calne 3771

# Computer Engineering 

NCR requires additional ELECTRONIC, ELECTRO-MECHANICAL ENGINEERS and TECHNICIANS to maintain medium to large scale digital computing systems in London and provincial towns.
Training courses will be arranged for successful applicants, 21 years of age and over, who have a good technical background to ONC/HNC level, City and Guilds or radio/radar experience in the Forces.
Starting salary will be in the range of $£ 900 / £ 1150$ per annum, plus bonus. Shift allowances are payable, after training, where
applicable. Opportunities also exist for Trainees, not less than 19 years of age, with a good standard of education, an aptitude towards and an interest in, mechanics, electronics and computers.
Excellent holiday, pension and sick pay arrangements. Please write for Application Form to Assistant Personnel Officer
NCR, 1,000 North Circular Road, London, N.W.2, quoting publication and month of issue.

## Electronic Technicians


#### Abstract

Ampex Quality Control Department now has vacancies for electronics technicians. Successful applicants will be responsible for fault finding and testing a complete range of sophisticated magnetic recording equipment.


Experience gained in the electronic industry or radio or television servicing would be an advantage or a qualification of O.N.C. standard.
Attractive salary based on qualifications and experience will be paid and the company operates an excellent range of Life Assurance and Pension Schemes, etc.

Please write or telephone for application form to the Personnel Officer, Ampex Electronics Limited, Acre Road, Reading,
(Tel.: Reading 84411).
AMPEX

## 0.1 <br> <br> We have vacancies for

 <br> <br> We have vacancies for}
## FOUR EXPERIENCED TEST ENGINEERS

in our Production Test Department. Applicants are preferred who have experience of Fault Finding and Testing of VHF and UHF Mobile Equipment. Excellent opportunities for promotion due to expansion programme.

Please apply to Personnel Manager
PYE TELECOMMUNICATIONS LTD., Cambridge Works, Haig Rd., Cambridge. Tel: Cambridge 5135:, ext. 355

## IRELAND RADIO TELEFIS ÉREANN ENGINEERS

Opportunities exist in Ireland's national broadcasting service for Engineers with initiative and Imagination.

Experience in broadcasting, telecommunications or related fields is desirable but not essential. Of equal importance is the ability to relate specific technologies to the wide range of problems that occur in radio and television.

Applicants should have a Degree or equivalent qualification in Electrical Engineering or Experimental Physics.

The salary scale extends to $\mathbf{2 2 , 0 7 0}$ per annum and excellent promotional opportunities exist. The commencing salary will depend on experlence.

Age limit- 35 years.
Applications, giving particulars of age, qualifications and experience should reach the
Personnel Administration Manager, Radio Telefis Eireann, Donnybrook, Dublin, 4. IRELAND, not later than 7th February, 1969.

Envelopes should be marked "ENGINEER (W.W.)."

## SYSTEMS ENGINEER

Racal Communications is now producing new sophisticated systems which embrace Radar and Communications techniques. It requires a man possessing sound knowledge of basic communications biased towards U.H.F. and Radar applications.
He will establish, control and guide a team of Test Engineers performing comprehensive Test and Inspection duties to rigid engineering specifications.
Opportunities in the Test Engineering field are excellent throughout the RACAL Group and the successful candidate will join a dynamic team:
Salary will be attractive and location at Bracknell. possessing plenty of private building plus facilities for rented accommodation in the New Town, which is adjacent to open country.
Please submit details of experience and present salary to
Mr. P. Cousins,


Group Personnel Manager, Racal Electronics Limited, Western Road, Bracknell,
Berks. Tel: Bracknell 3244

## ELECTRONC ENGNEERS

Are you interested in applying your knowledge? Our high powered electronic flash units and light sources with associated measuring accessories are being used in many fields ranging from photography to heavy industry. If you are qualified to construct and service such equipment, we can offer you interesting opportunities. Send details of experience and qualifications to Strobe Equipment Limited, 56 Turnmill Street, London E.C.1. CLERKENWELL 9268.

THE GENERAL POST OFFICE has vacancies for
RADIO OPERATORS II at its COAST RADIO STATIONS
Applications are invited from men between 21 and 35 years of age who must hold either the Postmaster General's First or Second Class Certificate of Competence in Radiotelegraphy or an equivalent certificate issued by a Commonwealth'Administration or the Irish Republic.

The posts which will be temporary in the first instance, carry a salary scale of $£ 765-£ 1,129$, depending on age at entry, but successful applicants will be eligible to enter the open competitive selection for permanent appointment to be held in the spring and autumn of 1969.

Applicants should write to: The Inspector of Wireless Telegraphy, Union House, St. Martin's-le-Grand, London E.C.1, or telephone 01-432 5628 for further information.

## THE UNVERSTITY OF ASTON IN BIRMINEHAM Electrical engineering dept. <br> M.Sc. COURSES October 1969 to September 1970

Graduate courses, of one year duration, leading to
a Master's Degree are offered in Electrical Engineer-
ing and in Precision Measurement and Inserumentation.

## M.Sc. in ELECTRICAL ENGINEERING (Ref. M.Sc.8)

One-third of the lecture work will cover mathematics and electrical engineering materials. The remaining sime will be devored to one specialise option selected from the following:
(d) Control and System
b) Power Systems
(d) Communication Systems
(d) Design and Pulse and Digital Circuits and Systems
The Science Research Council has accepted this course as suitable for tenure of its advanced course studentships.
M.Sc. in PRECISION MEASUREMENT AND INSTRUMENTATION (Ref. M.Sc.27)
This course is un by an interdepartmental group comprising Electrical Engineering, Mathematics, Mechanical Engineering, Physics and Production Engineering departments.
Both courses are open to applicants who have graduated in science or engineering or who hold equivalent professional qualifications.
Suitably qualified persons who wish to attend for part of either course (without examination) may do so by arrangement.
Application forms and further particulars (quoting ref. no.) may be obtained from:
THE HEAD OF THE DEPARTMENT OF ELECTRICAL ENGINEERING。
THE UNIVERSITY OF ASTON IN BIRMINGHAM,
GOSTA GREEN.
BIRMINGHAM 4.

## EEECTROMC TCEHMCAMIS

## Marconi

Can offer you
NON-TIED HOUSING IN A NEW TOWN ATTRACTIVE SALARY ANNUAL SALARY REVIEWS GOOD WORKING CONDITIONS 37-HOUR WORKING WEEK
At Basildon we have a number of vacancies for technical test staff to work on advanced aeronautical electronic systems, maintenance and building of test equipment and other major projects. These positions will be of particular interest to men with experience of transmitters, receivers, aerials, closed circuit T.V. or digital systems.

Please telephone or write for an application form to :-
Mrs. B. Bridgen, Personnel Officer, The Personnel Dept., The Marconi Company Limited, Christopher Martin Road, Basildon, Essex. Phone: Basildon 22822.

## INSTRUMENT SYSTEMS ENGINEER

## The Job

Designing and commissioning electronic aircraft simulator instrument systems in association with analogue and digital computer equipment.

## The Man

Qualified and/or experienced engineer who has a knowledge of aircraft instrument design theory and a sound understanding of servo and synchro techniques. Applicant should also have a knowledge of analogue and digital computing techniques.

## The Rewards

Long term career. High job interest in association with the airline industry. Good working conditions. Contributory pension scheme coupled with free Life Assurance. Good welfare benefits. Excellent salary.

Apply giving brfef detalls of experience and qualifications, quoting reference $W$.W. 2269, to
$\qquad$

## EXPERIENCED IN INSTRUMENTATION? LOOKING FOR A CHANGE? WITHIN REACH OF SLOUGH? OR POOLE?

We require a number of first-class engineers for the repair and calibration of a wide range of instrumentation.
If you are experienced in the maintenance of C.R.Os, D.V.Ms, V.Vs, Sig. Gens., etc., we would like to hear from you. Please write or phone.

G. R. QUIRK, Chief of Test<br>TECHNIVISION SERVICES<br>812/813 Weston Road, Slough, Bucks<br>Telephone No. Slough 29091

# ASSISTANT ENGINEER <br> (Radio Communication/Broadcasting) Required by THE CROWN AGENTS <br> for their London Office 

Candidates should preferably have a Degree and/or be Corporate Members of the Institution of Electrical Engineers. Applications will, however, be considered from those holding an HNC (Telecommunications) or equivalent qualification. The latter would be appointed as Technical Officers, the grade depending on age and experience.

Candidates should have received their training with an established manufacturer of Broadcasting or Radio Communications equipment or with a Broadcasting or Radio Communications Authority, and have had subsequent experience in one or other of these fields. Television experience, including telecine, would be an advantage, as would previous contract experience.

Candidates must be resident in the U.K., or anticipate being so in the near future, and be prepared to undertake short assignments overseas.

The Crown Agents is not a Department of the British Government; nor are its staff Civil Servants, although their salaries and conditions of service are based on those of the United Kingdom Civil Service.

## SALARIES

Assistant Engineer: $£ 1,429$ (age 25)- $£ 2,114$
Technical Officers: Grade I-£1,690-£2,059
Grade II-£1,472-£1,690
Further details of the post and an application form may be obtained by writing to: Crown Agents, "M" Department, 4 Millbank, London, S.W.1. Please quote reference M2S/OFFICE/VI and title of the post.

```
    UNIVERSITY OF BRISTOL
        Dept. of Entra-Mural Studies
Weekend Courses in Colour
    Television-Spring 1969
        Feb. 28th, Mar. 1st, 2nd
        Lecturer:H. V. Sims (B.B.C.)
            Mar. 28th, 29th, 30th
        Lecturers: Dr. G. B. Townsend
            (Thames Television)
    M. H. Cox (M.H. Cox Electronics Ltd.)
        Both courses will be held in Bristol
Full details from D. S. Wilde, 20A Berkeley
        Square, Bristol BS8 1HR
```


## Electro-Medical Service Department requires

## ENGINEERS

for testing and servicing electronic apparatus. Applicants should be aged 23-30, and should be of H.N.C. standard. Apply in first instance in writing to:

## SIEREX LTD.,

Electro-Medical Dept., Heron House, Wembley Hill Road, Wembley, Middx.

## Technical Authors

for important new projects
Applications are invited from authors with established ability and experience for positions in the following fields:-

## Data processing

Servo systems
Navigational aids
Sonar systems
Solid state radar
Radio communications
Electronic instrumentation
Electro-mechanical systems
These are positions of responsibility with an expanding company. Opportunities exist at the Company's London and Portsmouth offices. Also, on-site authors are required in counties to the north and south of London, and on the south coast. Formal qualifications to H.N.C. standard, and a minimum of five years in the engineering industry, will be an advantage.

Generous salaries according to experience and qualifications.
Please apply in writing to:-
The Technical Publications Manager (A.D.R. Houchin),

Irwin Technical Limited,
109/123, Clifton Street,
London, E.C. 2.

## ELECTRONIC TEST ENGINEERS AND TECHNICIANS

As a result of expansion, additional opportunities are offered to work on a full range of professional T.V. equipment for a world wide market.

Engineers and Technicians are required for the testing to specification, setting up, fault finding, etc., of T.V. transmitters, outside broadcast vehicles, colour cameras and monitors, sync. pulse generators, video tape recorders, and associated broadcast studio equipment.

Previous experience of professional T.V. equipment is not essential but applicants must have a
sound functional understanding of transistorised pulse circuitry and experience of equipment testing. Possession of H.N.C. or H.N.D. will be a definite advantage.

These vacancies will also be of special interest to members of H.M. forces shorly leaving with experience of transmitters, radar or communication equipment.

Appointments are at Cambridge and Weybridge and offer a career opportunity of exceptional value together with a continuing demand within the company for technical staff with a thorough grasp of the company's products.

Conditions of employment and working environment are attractive. A good starting salary depending on previous experience will be offered.

## REת

This company is currently setting up a new division to manufacture and market its gramophone records in the United Kingdom for the first time. The Technical Recording Department will be situated in West London and we are currently seeking the services of experienced men for the two positions outlined below:

## TECHNICIAN <br> (Disc Cutting)

Applicants should be fully experienced in the operation of Neumann
 half speed. We shall require the successful applicant to work to high standards of quality and to show from his past record that he is capable of doing so. A general background in audio engineering would be an asset.

## MAINTENANCE ENGINEER

(Sound Recording Equipment)
Applicants should be experienced in the electro-mechanical and electronic maintenance of studio tape recording and disc cutting equipment. Familiarity with up-to-date testing techniques and equipment will also be required. Academic qualifications in electrical and electronic engineering are desirable, but experience will be considered to be of primary importance.

First-class salaries will be offered in both cases and prospects of advancement are excellent. All applications will be treated as confidential.

Please write or telephone for an Application Form to Mr. A. Fremantle,

[^12]
## GLOBE TROTTERS

> We need Engineers with a yen for travel to commission our HF Communications Equipment which is selling in ever expanding world markets.
> If you have experience on high power HF Equipment and would like to see the world at our expense, then we want to hear from you. You will be responsible for carrying out trials, and handing equipment over to the customer in good working order. You may also be required to instruct the customers' engineers and take charge of teams of local labour.
> You should be between the ages of 25 and 45, highly mobile and preferably single (most jobs are unaccompanied).
> An HNC in electronics is desirable, but practical experience combined with a comprehensive understanding of modern circuit theory will not be discounted.
> In return you will receive an excellent salary with generous allowances for overseas travel.
> Please write or phone:
> Tom Anderson, Personnel Officer,
> Standard Telephones and Cables Ltd.,
> STC
> Oakleigh Road, New Southgate, N. 11

01-368 1234. Ext. 2578.

## B/ACALD <br> Communications

Applications are invited for the following positions:

## TEST EQUIPMENT REPAIR \& CALIBRATION ENGINEERS

To carry out repair and calibration of high quality proprietary test equipment including spectrum analysers, oscilloscopes. signal generators. etc. Previous experience essential and it is expected that the successful applicants will be qualified to at least ONC level. Attractive salaries will be discussed at interviews.

## ELECTRONICS TEST PERSONNEL

Progressive position for electronic Test Engineers and Testers engaged on a wide range of communications equipment, including transmitters and receivers. Applicants should have technical knowledge equivalent to City \& Guilds with previous experience of testing commercial equipment. Attractive salaries plus productivity payment.
Applications in writing, please, to:
Mr. P. Cousins, Group Personnel Manager,
Racal Electronics Ltd.,
Western Road, Bracknell, Berkshire.

OPPORTUNITIES exist for Radio Technicians to undertake interesting work involved with the maintenance and installation of equipment at airfields, Inland and marine mobile networks and on North Sea Drilling Rigs.
APPLICANTS should have experience in one or more of the following classes of equipment. VHF and UHF base station and mobile equipment employing both AM and FM techniques. HF Receivers and Transmitters up to 1 kw with SSB. ISB and FSK techniques.
Remote control systems for Transmitters and Receivers operating over GPO landlines.
Teleprinters and Telegraph error correction equipment.
City and Guilds Certificate or equivalent level qualification is desirable.
Applicants must have a valid U. K. driving licence and be willing to work outside normal working hours on a call-out roster basis.
THE POSTS offer starting salaries in the range $£ 1.200-£ 1.500$ commensurate with experience and excellent career prospects and will be based at the Company's Head Office between Hayes and Heston which is situated in close access to the M4 Motorway. Benefits include membership of an excellent Contributory Pension and Life Assurance Scheme and concessions on holiday air fares. IAL are a fast expanding company engaged in the field of communications. aviation services and engineering.
Please write stating brief details of age and career to date 10 :
Personnel Officer (R)
InTERNATIONAL AERADIO LIMITED
aeradio house - hayes road - southall - middlesex


## 5

## RADIO \& TELEVISION SERVICING RADAR THEORY \& MAINTENANCE

This private College provides efficient theoretical and practical training in the courses for men who have had previous training.
Write for details to: The Secretary, London Electronics College, 20 Penywern Road, Earls Court, London, S.W.5. Tel.: 01-373 8721.

## TECHNICAL BOOKS EDITOR (ELECTRONICS)

Large book publishing company based in London requires an experienced EDITOR to head a section producing books on electronics subjects, including radio and television servicing. Applicants (age 28-40) should have a recognised technical qualification in this field and experience in technical publishing. Industrial experience also will be an advantage.

Apply in writing giving full details of experience and qualifications to: The Staff Appointments Officer, Butterworth \& Co. (Publishers) Ltd., 88 Kingsway, London, W.C.2.

## CIVILIAN RADIO TECHNICIANS AIR FORCE DEPARTMENT

Are you

* INTERESTED IN doing vital work on raf radar and wireless equipment ?
* Aged 19 and over, of good educational standard with at least 3 years training and practical experience in radio/radar servicing.
If so, we offer
* Good pay. Salaries start at up to $£ 1130$ pa (according to age) and rise to £1304 by annual increments.
* Good prospects of promotion (top posts in excess of $£ 2000 \mathrm{pa}$ ).
* Excellent prospects of a good pension or a gratuity after 5 years service.
$\star 5$ day week. 3 weeks 3 days annual leave rising to 6 weeks, plus public holidays.
Vacancies exist at
RAF Sealand near Chester, RAF Henlow in Bedfordshire and periodically at other RAF stations.
Write to : MINISTRY OF DEFENCE, CE $3(H)$ (AIR), SENTINEL HOUSE SOUTHAMPTON ROW, LONDON, W.C. 1
or call at No. $30 \mathrm{MU}, \mathrm{RAF}$ SEALAND, between the following times :
Monday-Friday 8.30-4, Saturday 8.30-12.30


## REDIFFUSION

## COLOUR TELEVISION FAULTFINDERS \& TESTERS

We have a number of vacancies in our Production Test Departments for experienced faultfinders and testers.
Knowledge of transistor circuitry and experience with Colour Receivers together with R.T.E.B. Final Certificate or equivalent qualifications required.
These will be staff appointments with all the expected benefits.
Applications to:
Works Manager,
Rediffusion Vision Service Ltd.,
Fullers Way South,
Chessington, Surrey (near Ace of Spades).
Phone: 01-397 54II

## SITUATIONS VACANT

A full-TIME technical experienced salesman reprevious experience. salary required to-The Manager. Henry's Radio, Ltd., 303 Edgware Rd., London, W.2.
DESIGN DEVELOPMENT ENGINEER for laboratory 1 work in the design of audlo amplifiers, V.H.F. tuners and quality tape recorders. Only persons holding a stmilar position מeed apply. Salary according to
experience. Apply to Elizabethan Electrontcs Ltd., Ref: W.W.1. Crow Lane, Romford, Fssex. Tel. Romford 64101 .
[2108

ELECTRONIC ENGINEER required for development/ and professional recording equipment. Qualifications ONC/HNC Electronics or C. \& G. Saiary £ 1,000 £1.500, according to age, quailfications and experience. -Apply with details of qualifications and experience to Box werirnced TV Engine
EXPERIENCED TV Engineer required. Permanent quired. This is an addition to staff to cope with expanding TV service. REM RADIO, 79 Church Road, Ashford.
Tel. Ashford 5336 (Middlesex).
[79
CuATEMALA: Small radio station requires volunteer Oradio technician to assist in establishing relay stations and radlo schools. Interesting post concerning the development of remote areas. Volunteer terms: board, lodging, pocket-money, fares, allowances.-
Write: CIIR/OV, 38 King St., London, W.C.2. PRODUCTION TEST ENGINEER wanted to join our senior staff with experience of valve and transistor
audio equipments. 40 hour week. Salary E 1.200 per audio equipments. ${ }^{40-h o u r}$ week. Salary $\mathrm{E}^{1,200 \text { per }}$
annum. North West London grea. Box W.W. 2124 Wireless World.
R ADIO AND TAPE RECORDER TESTERS AND R ADIO AND TAPE RECORDER THOOTERS required Excellent rates of pay; $8 \mathrm{a} . \mathrm{m}$. to $5 \mathrm{p} . \mathrm{m}$. Five-day week. Ellzabethan Electronits Ltd., Crow Lane. Romford, Essex. Tel.: Romford 64101.
$\mathbf{R}^{\text {EDIFON LTD. require fully experienced TELE- }}$ R COMMUNICATIONS TEST ENGINEERS, Good commencling salaries. We would particularly welcome
enguiries from ex-Service personnel or personnel enquiries from the Services. Please write giving fult details to-The Personnel Manager, Redifon Lid.. Broomhlll Rd., Wandsworth, S.W.18.
TESTERS and Trouble-Shooters required by manufactarers of car radios. tape recorders, record players, etc. Good rates of pay. Apply to Elizabethan Electrontes Lid. Ref: W.W.2, Crow Lane, Romford,
Essex. Tel. Romford 64101 .
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[^4]:    * Liverpool College of Technology

[^5]:    1. Motorola Information Note AN-403.
[^6]:    $\ddagger$ See "The Diode-transistor Pump" by D. E. O'N. Waddington. Wireless W'orld, July 1966 for a full explanation.
    ** Adapted from a formula given in the above article.

[^7]:    * Cable and Wireless Lid.
    + See graphs opposite

[^8]:    $\pm$ See our report on the exhibition, November 1968, p.405-ED.

[^9]:    * The original article appeared in Funk-technik No. 20 (1967) pp. 783-786

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