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Wireless World

ELECTRONICS, TELEVISION, RADIO, AUDIO

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# Wireless World 

ELECTRONICS, TELEVISION, RADIO, AUDIO

## A Genuine Reject?

WHILE the article by T. D. Towers in this issue on transistor type numbers should assist readers in identifying a particular device and tracking down the manufacturers, one is tempted to ask how much faith can be placed in a type number? Large users of semiconductors buy direct from the manufacturers and can have every confidence in the devices they receive. The situation is rather different, however, for the home constructor who requires only a couple of AC 107s. It has become apparent that unscrupulous dealers are stamping reject transistors with well-known type numbers and selling them as genuine items. Type numbers have also bsen altered when a particular device is in short supply; an example of this would be so remove the D from OC 81D. Restamped devices often do not resemble the transistors they replace, electrically or mechanically or both. It is a well-known fact that many transistors will operate, to the detriment of the circuit, under conditions for which they were not intended, thereby facilitating the deception. The deceit is not limited to individual sales of semiconductors; complete equipments and sometimes kits are being marketed that use reject semiconductors, although no mention is made of this in the literature.
The home constructor would blame his own workmanship, lack of knowledge or the circuit he was making when it failed to operate satisfactorily rather than suspect that the semiconductors were not what they claimed to be. We do not deprecate the use of reject transistors; we only object when they masquerade as "on spec" devices.
How can the home constructor recognize these devices? The deception is not easily detected although the print used on restamped devices is usually large and untidy or the new markings are sometimes placed on a plastic sleeve slipped over the transistor. Genuine transistors nearly always carry the manufacturer's name or emblem and usually also a batch number-re-marks have neither of these. Another point to watch for is the case and lead-out configuration; sometimes the re-marks are not even in the correct encapsulation. Because the range using the old European coding "OC" is probably the most common and best known as far as the home constructor is concerned it is these devices that appear to be most often misrepresented.
Our advice is, deal only with a reputable supplier, return any devices that are not what they are claimed to be and beware of the isolated term "guaranteed." Guaranteed for what?

## By Numbers

ANOTHER aspect of numbering is raised by a correspondent whose letter is published in this issue. He pleads for a common identification or part-numbering system for components. Instead of each user of a component giving his own part-number to itdepending upon the particular piece of equipment in which it is to be used-there should be, the writer suggests, a British Standard part number that all could use. If this were done it would certainly simplify the specification of components but it would, of course, mean that every variant of the physical and electrical specification of a particular resistor, capacitor, or what have you, would have to bear a different number. Only those most closely concerned with the supply of components will fully appreciate the difficulties experienced in the present jungle, but is the correspondent's suggestion practicable? Perhaps the introduction of i.cs provides a golden opportunity for starting such a scheme.

## JANUARY

1968

# Radio Signals from the Heart of Matter 

# An old circuit-the superregenerative receiver-put to a new use in analysis of materials by nuclear quadrupole resonance spectrometry 

By D. A. TONG, b.sc., Ph.D.


#### Abstract

Only too often research into atomic scale phenomena involves large and costly electronic apparatus. Presented here is a technique that can provide useful information about the actual electron distribution within a molecule but which only costs a few pounds to set up. The article should provide enough information to enable the interested electronics experimenter or student to experience for himself the thrill of picking up radio signals from the very heart of matter, the atomic nucleus.


T'HE superregenerative receiver' was widely used by radio amateurs in the early days of v.h.f. because it combines very high sensitivity with great simplicity. Later on as v.h.f. techniques advanced, its disadvantages, i.e., poor selectivity, poor frequency stability and radiation of interference, resulted in its virtual elimination as a serious rival to the superheterodyne receiver for communications work. Since the early 1950s, however, the superregenerative detector has embarked on a new career in a totally different field, that of the branch of nuclear magnetic resonance (n.m.r.) spectrometry known as nuclear quadrupole resonance (n.q.r.)2 spectrometry. Later in this article we will describe simple circuitry with which it is possible to detect n.q.r. in suitable solids, but first it will be useful to discuss briefly the physical basis of the phenomenon itself.

## THEORY OF N.Q.R.

A spinning atomic nucleus has a magnetic moment which is colinear with the axis of spin, and therefore can be regard-


Dr. David Tong, who is 26, graduated in chemistry at Leeds University and" received his Ph.D. for research into chemical applications of N.Q.R. He has been a research fellow at the University of Warwick and has recently joined the staff in the Department of Chemistry, University of Glasgow. Dr. Tong has developed an improved superregenerative N.Q.R. spectrometer which is to be produced by Decca Radar Ltd. to which he is a consultant.
ed as a minute spinning bar magnet. Because of its submicroscopic size, however, the motion of such a magnet in a magnetic field can only be described adequately by means of quantum mechanics, and, in fact, differs somewhat from that of a spinning magnet of ordinary size.

When an ordinary bar magnet is placed in a magnetic field, it tends to take up a position of minimum potential energy; that is, with its magnetic moment aligned along the field direction as shown in Fig. 1(a). Any small displacement from this position makes the magnet oscillate as shown in (b), until its potential energy has been dissipated by friction, when it again comes to rest. If, instead of being stationary, the magnet is continually spinning about an axis coincident with its magnetic moment (c), it still tends to align along the field (shown by $H_{0}$ ), except that any displacement now results in a precession about the field direction (d), instead of a vibration. The behaviour is entirely analagous to that of a spinning gyroscope in the earth's gravitational field, and the angular frequency of the precession is proportional to both the magnetic moment, $M$, and the applied field strength, $H_{0}$. Fig. 1 (e) is a vector diagram representation of the precession.
When, on the other hand, one considers the motion of a spinning magnet of nuclear dimensions in a magnetic field, one finds that, unlike the classical magner, it can never align itself completely along the field, to do so would be to violate the Heisenberg Uncertainty Principle. In fact, the angle that the spin axis makes with the field is restricted to one of a limited number of " allowed " values and the spin axis therefore precesses continually about the field direction.

In practice one is not concerned with the nucleus of a single atom but usually with specimens containing exiremely large numbers of atoms. Under such conditions the individual nuclear magnetic moment of one atom cannot be observed and one can only detect the resultant of all the microscopic moments. Not surprisingly, perhaps, this resultant, or " macroscopic", magnetic moment behaves in many ways as if it belonged to a spinning bar magnet as shown in Fig. 1. In particular, it tends to align itself completely along the direction of an applied magnetic field. Then, if the spin-system is suddenly disturbed by suitably applying energy to the sample, the macroscopic moment temporarily begins to precess around the field direction at some particular angle and at a frequency, the "Larmor" frequency, which depends on the nuclear magnetic moment and the applied field strength.

The method usually adopted for transferring energy to the sample is to introduce a second magnetic field at rightangles to the first, but one which is oscillating at a radio

(d)

(b)

SPIN AXIS

(d)

Fig. 1. (o) Equilibrium position of a bar magnet in a uniform magnetic field. (b) shows the oscillations of the bar magnet about the equilibrium position resulting from a small displacement therefrom. Assuming the bar magnet is spinning about its long axis, (c) shows it in its equilibrium position in the uniform magnetic field. (d) shows the precession about $H_{0}$ which results from displacing the spinning magnet from its equilibrium position. In a real system the precession dies away as shown as the potential energy of the displaced magnet is dissipated Finally (e) shows in vector form the precessing magnet in the absence of frictional forces; $M$ represents the magnetic moment of the magnet.
frequency. Such a field is equivalent to two separate fields rotating in opposite directions, and if the rotation frequency is much different from the Larmor frequency, neither has any appreciable effect on the spin system. In contrast, when the two frequencies are identical the effect is considerable, because, no matter how the macroscopic moment precesses, the resultant of the steady field, $H_{0}$, and the component of the oscillating field which rotates in the same direction as the precession, $H_{1}$, will always act so as to pull it away from the $H_{0}$ direction. The system is then said to be in resonance, and energy is absorbed from the rotating field. This effect is shown in Fig. 2. In the case of the hydrogen nucleus, for example, the Larmor frequency is 42.577 MHz in a field of 1.0 tesla ( 10,000 gauss), and is therefore well within the radio-frequency range. All other nuclei have lower frequencies than this.

So far we have only discussed the phenomenon of n.m.r., but it is now only a small step to extend the discussion to explain n.q.r. (Fig. 3). Atomic nuclei with the property of spin fall into two groups according to whether the distribution of their positive charge is spherical or non-spherical. Nuclei in the latter category, in addition to having a magnetic moment also possess an electric quadrupole moment. Such a moment is equivalent to two electric dipoles placed back-to-back, and if a quadrupolar nucleus is present in a non-uniform electric field, it experiences a torque and will tend to align itself with its quadrupole moment (which is co-linear with the axis of spin and the magnetic moment) along the direction of maximum electric field gradient (e.f.g.). Morcover, as the nucleus is spinning, precession at certain " allowed" angles will again occur, but in this case it will be about the e.f.g. direction.

Notice, however, that such precession still involves a precessing magnetic moment and the resultant of a large number of such moments can still couple with a rotating magnetic field. In short, in n.q.r. the interaction energy, and hence the resonance frequency, is determined by an electrostatic field gradient acting on the nuclear quadrupole moment, whereas in n.m.r. the important interaction is that of the nuclear magnetic moment with an external magnetic field. Both effects can be detected by interaction between the macroscopic magnetic moment and a rotating magnetic field.

(a)

(b)

Fig. 2. Interaction of the macroscopic nuclear magnetic moment $M$ with a steady magnetic freld $H_{0}$ and a rotating magnetic field $H_{1}$. (a) When the frequency of rototion $\omega$ is different from the Larmor frequency $\omega_{0}, M$ remains aligned along $H_{0}$ and is unaffected by $H_{1}$. (b) When $H_{1}$ rotates at $W_{0}, M$ experiences a force which pulls it away from the $\mathrm{H}_{0}$ direction, about which it precesses.


Fig. 3. (a) In the case of n.m.r., the component $M^{\prime}$ of $M$ rototing in a plane perpendicular to the steady field direction induces an alternating voltage in the sample coil. (b) The case of n.q.r. in powdered somples is similar, except that $H_{0}$ is not present and $M$, which is now the resultant of a large number of macroscopic moments each precessing about a different direction, is accompanied by an equal and opposite moment rotating in the opposite direction.

The main difference between the two techniqu s in practice is that a large external magnetic field must be applied to the sample to detect n.m.r., and by varying the field strength the resonance frequency can be brought into a suitable range. The e.f.g. necessary for n.q.r., on the other hand, is already present in most crystaline solids and the resonance frequencies are fixed. The e.f.g. arises from the detailed electron distribution within chemical bonds, and since the n.q.r. frequency depends directly on the e.f.g. important information about these distributions can be obtained. Typical n.q.r. frequencies range from as low as 5 MHz right up to several thousand, depending on the properties of the particular nucleus and the type of compound.

## THE SUPERREGENERATIVE DETECTOR

Having looked at the magnetic resonance phenomenon, we can now return to what is probably more familiar ground and see where the superregencrative receiver enters the picture. We have mentioned that to excite a resonance it is necessary to place the sample of material in a coil which is supplied with r.f. current. This can be done very conveniently by making the coil part of the tank circuit of an oscillator. The system can then be easily tuned over the necessary wide frequency range. If the level of oscillation is arranged to be critically dependent on the tank circuit $Q$, as in the so-called marginal oscillator, the sudden absorption of energy by the sample as the oscillator passes through the resonance frequency causes a drop in r.f. level and provides a means of detecting the resonance. Such oscillators, many of them very simple, are widely used for studying n.m.r. For n.q.r., however, considerably greater values of $H_{1}$ are required, and this is where the superregenerative oscillator (s.r.o.) comes into its own, for even when a superregenerative receiver is adjusted for maximum sensitivity, the average r.f. level may be several volts. In fact, it behaves also as a low-power transmitter-as anyone knows who has tried to operate two such receivers within half a mile of each other.
The s.r.o. can successfully combine high sensitivity with high r.f. levels because it is sensitive only during a very short interval between pulses, and at this point in the quench cycle the valve anode current, and hence the gencrated shot-noise, is low. During the bursts of r.f. oscillation, however, large peak voltages can be attained without affecting the detection process. There is, though, one important difference between a s.r.o. used for n.q.r. and one used as a communications receiver. In the lastmentioned case the circuit is adjusted to operate in a com-
pletely incoherent condition because then the gain of the circuit is a maximum, whereas a certain degree of coherence is essential for resonance excitation.

The term " incoherent" refers to the random phase relationships which exist between successive bursts of oscillation when each one builds up from noise voltages only, i.e., when each burst is completely damped before the next pulse starts to build up. When the r.f. output of such an oscillator is monitored on a receiver one finds that there is no definite oscillation frequency present but only a band of noise spread over several hundred kHz . Such a signal is useless for exciting nuclear resonances because negligible power is present in the relatively narrow width (a few kHz ) of the resonance line. The situation changes, however, if the oscillations are less severcly damped berween pulses, because then the starting phase of each pulse is determined partly by the tail of the previous pulse and partly by noise. In other words the coherence is increased. The effect on the monitor receiver is to cause discrete frequencies to appear, and since the oscillator is pulsed, they are spaced at integral multiples of the quench frequency on cither side of the oscillator's fundamental frequency.

The available power is now concentrated into narrow frequency bands and many of them are sufficiently strong to excite a resonance in a sample of material placed in the s.r.o. tank coil. Such resonances are then simultancously detected by the circuit because the precessing macroscopic magnetic moment induces an r.f. voltage in the coil, and this voltage provides an input signal for the s.r.o., now acting in its role as a receiver (Fig. 3). The operation might be crudely pictured as that of a radar system: the s.r.o. sends out a pulse which excites the spin-system in the sample, making it ring like a high-Q tuned circuit, and the ringing signal is then picked up as the "response."

In practice, the " ringing" time, or to use its correct name, the "spin-phase relaxation time", is of the order of milliseconds and is considerably longer than typical quench periods ( 10 to 100 microseconds). This is why the nuclei " see" the r.f. waveform as its Fourier components (or sidebands) rather than as individual

bursts of oscillation, as it appears on a wide-bandwidth oscilloscope.

## SUITABLE APPARATUS

A very simple circuit which has been widely used to detect n.q.r. in the 15 to 50 MHz range is shown in Fig. 4. The circuit is a self-quenched Colpitts oscillator, the quench frequency and coherence depending on the grid time-constant, which can be varied by $\mathrm{VR}_{1}$. It is also self-detecting because when a signal is present the quench rate increases slightly, and this results in an increased voltage drop across the anode load resistor. After filtering out the quench frequency components the audio output signal is amplified in a conventional lownoise preamplifier, such as that of Fig. 5.

A suitable setting for $\mathrm{VR}_{1}$ can be arrived at by monitoring the r.f. output on a receiver, the correct adjustment being half-way between that giving a very sharp series of sidebands and that which results in a broad band of noise. Since the gain of the derector depends on the coherence, $\mathrm{VR}_{1}$ can also be adjusted by observing the noise level at the output of the preamplifier. In this case the correct setting is somewhere between the ones giving maximum and minimum noise amplitudes.

In order to observe a resonance line directly, some method is required of repetitively sweeping the oscillator frequency back and forth over a range of up to several hundred kHz , while simultaneously observing the output on an oscilloscope. Many oscilloscopes have a terminal which allows a connection to be made to the timebase output, and if this is the case, it is only necessary to connect this output in such a way as to always reversebias the variable capacitance diode D1. Care must be taken, however, not to exceed the maximum rated reverse voltage for the diode ( 30 V for the BA110), or to drive it into forward conduction. The depth of modulation can be adjusted by both $\mathrm{VR}_{2}$ and $\mathrm{C}_{2}$. If a sweep output is not available from the oscilloscope, a sinusoidal sweep can be obtained from a mains transformer giving say 20 V output, together with a suitable battery connected so that the diode is never forward-biased. Finally, if an oscilloscope is not available, it is still possible to detect a resonance by listening to the output on a pair of headphones or a loudspeaker.

The construction of the circuit should follow good v.h.f. wiring practice, keeping all r.f.-carrying wires as short and direct as possible. Beehive trimmers are suitable for $\mathrm{C}_{2}, \mathrm{C}_{3}$ and $\mathrm{C}_{4}$, but a good quality tuning capacitor is essential for $C_{1}$, together with a good slowmotion drive. Since the resonances to be detected are likely to be only two or three times larger than the noise level at first, careful tuning is required, and the


Decca prototype n.q.r. spectrometer. (An automatic frequency marker unit is missing from this particular example and would normally occupy the blank panel space.)


Fig. 5. A suitable audio frequency preamplifier for use with the Fig. 4 circuit.


Fig. 6. Photographically recorded c.r.o. trace showing the ${ }^{35} \mathrm{Cl}$ n.q.r. line in potassium chlorate at room temperature, obtained with a circuit similar to that of Fig. 4, The position of the centre of the resonance line was slightly different on each of the five sweeps recorded because of jitter on the sawtooth generator used for the frequency sweep. Slightly higher signal-to-noise ratios can be expected for ${ }^{35} \mathrm{Cl}$ in para-dichlorobenzene.
h.t. supply should be well smoothed to eliminate hum. The r.f. choke in the cathode circuit is a rather critical component and if the oscillator exhibits dead-spots, i.e., ranges over which oscillations cease, it is likely that the particular choke has a series resonance in the range. The solution is to try a different choke.

A suitable substance which at room temperature gives a strong signal at about 34.27 MHz and another weaker one at 27.01 MHz is para-dichlorobenzene. The signals are due to the n.q.r. of the ${ }^{35} \mathrm{Cl}$ and ${ }^{37} \mathrm{Cl}$ nuclei, respectively, and the two frequencies are in the ratio of 1.2688 to 1 , which is the ratio of the quadrupole moments of the two nuclei. In nature, the two isotopes occur with relative abundances of approximately three to one and this accounts for the different intensities of the two resonance lines. To observe the stronger, ${ }^{35} \mathrm{Cl}$, line the oscillator should be set to sweep around the 34 MHz region by monitoring on a receiver, or by temporarily injecting a signal at 34.27 MHz into the s.r.o. from a signal generator. Carcful tuning of $\mathrm{C}_{1}$ should then enable the resonance signal to be seen (or heard). Final adjustments to $\mathrm{C}_{3}, \mathrm{C}_{4}$ and $\mathrm{VR}_{1}$ should then be made in order to obtain best signal-to-noise ratios. With the para-dichlorobenzene sample contained in a glass tube of half-inch internal diameter and packed tightly to a depth of about one inch, a signal-to-noise ratio of at least ten should be attainable with a good oscillator valve. It is important to have as much sample material as possible within the coil volume and therefore the sample tube should have thin walls
and the coil should be wound tightly on the sample tube itsclf.

Because of the sidebands present in the oscillator power specirum, several responses are seen for each true resonance line, and for serious work some way of eliminating all but the fundamental is required. Usually this requires the use of the slightly more complicated externally-quenched s.r.o., but the methods used generally rely on the fact that if the quench frequency is altered, only the fundamental response will be unmoved. Unfortunately, with the self-quenched circuit the coherence, and hence the gain, varies as the quench rate is altered (using VR ${ }_{1}$ ), but the effect should still be ohservable.

Sometimes the circuit will display apparent resonances which are in fact not due to the sample. These may be recognized by removing the sample, when of course a true n.q.r. signal would disappear. A neater method, however, is to place a small magnet near to the sample, whereat any n.q.r. line will be broadened so much that it will be effectively erased. This effect arises because of the interaction between the individual nuclear magnetic moments and the field, which results in a splitting of the resonance line into several components. The extent of the splitting is dependent on the orientation of the magnetic field with respect to the internal reference axes of the crystal and, in a powdered sample with its random distribution of angles, the effect is to broaden the line. Any line which does not show this behaviour cannot be attributed to n.q.r.

Another effect which can be casily demonstrated is that of temperature. If the sample is heated, the resulting expansion causes small changes in the internal electron distributions of individual molecules and leads to a change in the n.q.r. frequency. If the sample is subject to a non-uniform temperature, caused, for example, by body heat during handling, or by a nearby soldering iron, different parts of the sample have different resonance
frequencies and the line is broadened. It is always advisable, for this reason, to wait five or ten minutes after handling the sample before tryirg to detect a resonarce. Since the n.q.r. frequency depends only on the nuclear quadrupole moment, a constant of nature, and the e.f.g., a property of the particular chemical compound, it has been suggested that n.q.r. be used as a thermometer ${ }^{3}$. Such a thermometer would be useful in situations where frequent calibration is impossible, c.g., remote weather stations or space probes. A suitable sample in such applications is potassium chlorate, whose ${ }^{35} \mathrm{Cl}$ resonance is at 28.2133 MHz at $0^{\circ} \mathrm{C}$. Its advantage lies in its low natural line-width of about 500 Hz , which means that more accurate frequency measurements are possible, and in its fairly large temperature coefficient of about -4.8 kHz per degree.

Other nuclei whose n.q.r. frequencies fall in the frequency range of the circuit of Fig. 4 are ${ }^{69} \mathrm{Ga},{ }^{[1} \mathrm{Ga}$, ${ }^{63} \mathrm{Cu},{ }^{65} \mathrm{Cu}$, and ${ }^{59} \mathrm{Co}$. Cobalt and gallium compounds are not readily available but a suitable copper compound is cupric oxide. This shows two broad resonance lines at 25.955 MHz and 24.028 MHz , at $28^{\circ} \mathrm{C}$, corresponding to the ${ }^{63} \mathrm{Cu}$ and ${ }^{65} \mathrm{Cu}$ nuclei respectively. Finaily, to give the reader an idea of the results 10 expect with the apparatus described, a photograph of the ${ }^{35} \mathrm{Cl}$ n.q.r. line in potassium chlorate at room temperature is shown in Fig. 6.

In conclusion the writer would like to point out that apparatus basically identical to that described above is being used in laboratories throughout the worll for serious research in n.q.r.

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## A Logical Bassoon

THE bassoon is notorious for the difficulty of its fingering (the pattern of raised and lowered fingers necessary to produce a particular note), and certain orchestral passages such as in Stravinsky's Rite of Spring, can daunt the most accomplished player. Dr. G. S. Brindley, F.R.S., a physiologist at Cambridge University, has attempted to ease the player's task by designing a new iype of bassoon which uses electronic logic circuits to simplify the fingering. He gave a demonstration of the new instrument, which is based on a German bassoon, at a meeting of the British Acoustical Society held at the B.B.C. Research Department on 5th December.

Acoustically the Brindley bassoon is similar to a conventional instrument except that the acoustic column, constructed from Sapele wood, is of rectangular crosssection instead of circular. The player's fingers operate a set of keys incorporating micro-switches, and the on-off signals from these are fed to diode-transistor logic circuits which control solenoids powered from a 24 V supply. The solenoids raise and lower pads over the holes in the acoustic column. "The logic circuits are arranged to separate the "holing" (patterns of open and closed holes) from the fingering, so that for each note it has become possible to choose the holing that is best acoustically and the fingering that is best for facility. (It would
be possible to use a fingering system as for the piano.) First the microswitch signals are fed into "recognition" logic circuits-a series of AND gates-where each pattern of raised and lowered fingers causes a particular "note output " terminal to be activated. The signals from these terminals then pass into "programming" logic, comprising a series of OR gates, the outputs of which directly operate the holing solenoids. All the electronic circuitry, except the solenoid power supply, is mounted on the outside of the acoustic column.

Another helpful innovation in the bassoon is a $15-\mathrm{W}$ electric heater, which is used not only to get rid of condensation but to tune the instrument. When being played the bassoun stands on the floor between the player's legs and causes no obstruction to his line of sight. Ordinary bassoon reeds are used. The timbre has a slight suggestion of saxophone quality.

Also at the B.A.S. meeting H. D. Harwood (B.B.C. Research Department) described and demonstrated a new B.B.C. monitoring loudspeaker which is outstanding in both its freedom from colouration and its repeatability of frequency characteristic. This has been achieved mainly by the use of a new cone material, a type of sheet polystyrene called Bextrene, which is shaped by a vacuum forming technique.

## R.F. Measurements and Standards

SINCE July the British Calibration Service has been in operation and is seeking to establish centres of expertise in r.f. measurements on a larger scale than has been possible hitherto. An indication of the interest in r.f. measurements and standards, which until recently suffered from lack of official support and recognition, was provided by the attendance of about 150 at a conference on the subject held at the National Physical Laboratory from November 14th to 16 th. Organized by the Institution of Electronic and Radio Engineers in collaboration with the Institution of Electrical Engineers, the conference was formally opened by Sir Leonard Atkinson, president-elect of the I.E.R.E., and a total of 18 papers was presented.
There is nothing comparable in this country to the strong central facility in the United States at the Radio Standards Laboratory of the National Bureau of Standards, although a new Division of Electrical Science has recently been formed at the National Physical Laboratory and one of its first tasks will be to establish a laboratory for r.f. mealsurements. Close co-operation with other European countries is also desirable and it might be a sensible plan to arrange some division of the work between countries if we are to attempt to reach the same standards of accuracy as the N.B.S. over the whole field of measurement.

## COMPARISON WITH ABROAD

At the conference a review of the present position in measurement capability presented the state of the art in the U.K. in comparison with the range of standards developed in Europe and the United States. The fields covered in the range 100 kHz to 3 GHz were frequency, power, impedance and reflection co-efficient, current and voltage, attenuation, noise, field strength and power density. It was clear that further effort was required in several fields to match the best of foreign standards but at least one speaker made the point that small teams in this country had achieved results in their selected areas quite as good as those, for example, of the N.B.S. The Electrical Inspection Directorate of the Ministry of Technology, has played a leading part in improving our r.f. standards and measuring procedures and standards of impedance, power, attenuation and bridge and reflectometer methods were described. Developments are in progress to extend the accuracy of impedance standardization to $0.1 \%$ over the range 200 MHz to 3 GHz and with a new piston attenuator it is hoped to achieve an accuracy of $\pm 0.01 \mathrm{~dB}$ in 100 dB . The present accuracy of power measurement is $\pm 1 \%$ between 2 and 80 mW but a new form of dry load calorimeter is under construction. This makes use of a sensitive thermal convertor of the multi-junction type developed by Wilkins at N.P.L. to detect the temperature rise in the metal-film load resistor. The frequency range is from audio frequencies to about 5 GHz and the expected accuracy is $\pm 0.2 \%$ for powers of 40 mW to 4 W .
Another approach to power measurement was described by Marconi Instruments in developing a range of commercial power meters extending from 100 mW to 1 kW . Unlike the E.I.D. instrument the thermo-element is here incorporated in the r.f. circuit, the heater of the junction forming part of the connection to the load resistor. By adopting a thin-film form of construction for the heater and by changing from co-axial to slab-line
geometry the insertion of the thernal element can be arranged without causing appreciable discontinuity. The e.m.f. generated by the thermo-junction is very nearly proportional to the (current) ${ }^{2}$ passing to the load resistor and is read on a millivoltmeter calibrated directly in total incident power to an accuracy of $\pm 5 \%$.
Thin films form the load resistors in both the E.I.D. and Marconi instruments: they also find application at the other end of the system in the source resistor of the thermal noise standard developed by Ferranti. The resistor, an alumina tube coated with pyrolitic carbon, is maintained at a temperature of $1,000^{\circ} \mathrm{C}$ in a vacuum enclosure. It is matched to a $50 \Omega$ coaxial line at the operating temperature. There are indications that the present temperature co-efficient of resistance, amounting to several parts in 10,000 , can be reduced and this will enable the standard to be used over a range of temperature, indicated by a platinum/rhodium thermo-couple, without appreciable mis-match. These sources have been examined on the noise comparator designed by the Services Valve Test Laboratory and the mean value at 300 MHz agrees closely with two different types of noise diode, giving confidence that an absolute accuracy of $\pm 0.1 \mathrm{~dB}$ has been achieved. Comparison with other sources suggests that this holds up to 1 GHz .

The measurement of attenuation is central to many r.f. procedures and several papers described methods for the comparison of attenuators. The arrangement favoured by both Marconi Instruments and E.I.D. was parallel i.f. substitution with the standard attenuator operating at a fixed frequency, usually 30 or 60 MHz . This method imposes severe requirements on the linearity of the first mixer stage and in the E.I.D. equipment the thermionic diode used is linear to better than 0.01 dB over the range -7 to -107 dBm , from a frequency of 108 MHz to more than 1 GHz . The sensitivity and stability of the equipment enables a change of 0.001 dB to be detected under good signal-to-noise conditions. The Post Office Research Station has also under development equipment which it is hoped will enable insertion gain and loss to be measured to an accuracy of $\pm 0.01$ dB in 60 dB at frequencies in the range $0-50 \mathrm{MHz}$, the accuracy falling progressively to $\pm 2 \mathrm{~dB}$ at frequencies of $1-2 \mathrm{GHz}$ (see p. 599, December issue).

An interesting paper from the University of Southampton described the problems arising in measurements on very small thin-film and monolithic circuit components, in the frequency range $100-1,000 \mathrm{MHz}$. For thin films it is permissible to scale-up the measurement and the resistance and capacitance per unit area are obtained from test pieces evaporated in concentric form which then provide convenient terminations for the coaxial lines of the admittance/impedance measuring equipment. For monolithic components the measurement must be conducted in situ and it is necessary to reduce the dimensions of the standard $50 \Omega$ coaxial lines to an area of about $0.01 \times 0.01$ inch while remaining well shielded from each other. The transition is made by the use of a micro-stripline formed by laying a gold ribbon between two layers of dielectric sheet and clamping between thick brass plates. The connection to the chip is made by a gold wire or by extending the gold ribbon but a significant reduction in the residual inductance of about 1 nH can be achieved by making the final link in the form of a uniline.


A view from the audience seating in colour studia 8 at the B.B.C. Television Centre during a rehearsal.


## B.B.C. COLOUR SERVICE

FOR the past ten months or so the B.B.C. has been gradually installing colour equipment at the Television Centre, in West London, so that when its colour service was officially inaugurated on December 2nd two production studios, a continuity studio and two mobile control rooms were fully operational. As a result, up to 25 hours of the 30 or so hours of programmes on BBC-2 each week are now in colour.

Each of the two production studios is equipped with four Marconi Mk. VII four-tube cameras and the continuity studio with three Peto-Scott three-tube Philips plumbicon cameras. Peto-Scott cameras are also used in the mobile control rooms. A third production studio, which will be brought into service in the Spring, will have four E.M.I. four-tube cameras. The new studio, which is at Alexandra Palace, is being equipped with three Marconi cameras and will be brought into service in colour in January; until then the news will be in monochrome.

All the cameras are equipped with zoom lenses and not fixed-focus lenses in a turret. The main reason for this being that it is extremely difficult to maintain in a matched condition the colour characteristics of different lenses. Cameras have to be warmed up for two hours before line-up can

Sound controller's position with tape and disc backing facilities, in the sound control room associated with one of the two colour studios.

The vision and lighing control room studio one. It is equipped with what is called a "Q-file " lighting control, made by Thorn, by means of which up to 100 lighting combinations and levels can be pre-set and brought into operation sequentially.

be undertaken and the line-up itself takes a further hour and a half. Care is neccessary to keep the colour temperature of the lighting constant. Colour studios are operating on a level of 1615 lux ( 150 ft . candles). To provide the extra and more evenly distributed illumination necessary for colour the B.B.C. has developed a dual-purpose lantern one end of which produces a spot source and the other a soft-light.

Cameras are, of course, only partalbeit a crucial part-of the colour installation. The backing-up facilities already in use at the Television Centre include four Ampex 2000 videotape colour recorders (a further two will be installed early in the new year); one mobile Ampex 2000 (a second is planned for next Spring); and one R.C.A. TR70 vision tape recorder for news at Alexandra Palace where a Pye 16 mm telecine unit using a four-tube camera will also be brought into service in January. Telecine equipment at the Television Centre includes four 16 mm and five 35 mm Cintel twin-lens units and a further three will be added in the Spring. Then, of course, one must not forget the field store convertor, developed by Rainger for the conversion of American 60 -field colour signals to 50 fields and vice-versa (described in our October 1967 issue). There is also a SECAM/PAL transcoder.

The PAL system: next page

Service areas of the 14 stations which transmitted colour television on the opening of the service on December $2 n d$ are shown with a line tint. The other stations shown are scheduled to be in operation by early 1969, in foct the first six (Nos. 15-20) are expected to be ready for use in 1968. The key to the stations gives in parentheses the chonnels for BBC-2

General view of the control desk and monitors, only two of which are for colour, in the production control room of studio 8.



| 1 | Crystal Palace Sutton Coldfield |  | . | , | $\cdots$ | (33) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2 |  |  |  | . | . | (40) |
| 3 | Wenvoe |  | . | . | , | (51) |
| 4 | Winter Hill |  | . . | . | . | (62) |
| 5 | Rowridge |  |  | . | . | (24) |
| 6 | Emley Moor |  | $\cdots$ | . | - | (51) |
| 7 | Belmont . | $\ldots$ | . | . . | . | (28) |
| 8 | Tacolneston | . | . | . . | . | (55) |
| 9 | Dover .. | . |  | . . | . . | (56) |
| 10 | Llanddona | . . | . | . . | . | (63) |
| 11 | Black Hıll | . | . | . | . | (46) |
| 12 | Pontop Pike | . | . | . | . . | (64) |
| 13 | Divis . | ., | , | . | . | (27) |
| 14 | Durris . | . | ., | . | $\cdots$ | (28) |
| 15 | Walcham | . | - | .. | $\cdots$ | (64) |
| 16 | Sudbury | $\cdots$ | - | . | . | (44) |
| 17 | Oxford | .. | . | . | - | (63) |
| 18 | Bilsdale | . | . | . | . | (26) |
| 19 | Moel-y-Parc | . | . | . | . | (45) |
| 20 | Balcalk .. |  | . | $\cdots$ | .. | (63) |
| 21 | Mendip Forest | . | . | $\cdots$ | . | (64) |
| 22 | Sandy Heath | . |  | . | . | (27) |
| 23 | Caradon Hill | . |  | . | . . | (28) |
| 24 | Craigkelly |  |  | $\cdots$ | . | (27) |
| 25 | Londonderry |  |  | . | . | (55) |
| 26 | Heathfield |  |  |  | . | (52) |
| 27 | Staffordshire |  |  | . | , |  |
| 28 | North Hampsh |  |  | $\cdots$ | - | (45) |

# THE PAL COLOUR TV SYSTEM 

## A simplified explanation of how it works

By S. C. RYDER-SMITH, b.Sc.

1TEIEVISION set giving a black-and-white picture is a fairly complex piece of equipment. With colour the complexities obviously multiply, and a host of fresh terminology is introduced into the subject. What follows is an attempt to explain, in fairly simple terms, how the PAL system operates. The explanation offered goes no further than outlining the background theory, and building on this to the point where a PAL receiver block diagram can be understood.

The first question to be considered is: how can we set about analysing the colour content of any scene, and then reproduce the scene so that the full range of colours is preserved? Fortunately, the solution to this problem has already been discovered in colour photography, and is fairly familiar. A colour may be analysed into its red, green and blue components, and then reconstructed by adding red, green and blue light in the same proportions as discovered in the original. This is illustrated in the simple colour television system shown in Fig. 1.

In this system three television cameras view a scene simultaneously. One, by looking through a red filter, transmits the red component only, the next, with a green filter, the green component only, and the last with a blue filter, the blue component only. Each camera output drives a cathode ray tube monitor. The monitor receiving the " red " camera output has a red filter in front of it, and therefore gives a red image, which is focused by a lens on to a viewing screen. The monitors receiving the "green" and "blue" outputs similarly give green and blue images on the viewing screen, and so the original scene is reconstructed in full colour.

The major difficulty with this scheme is the impossibility of aligning the three separate colour-component pictures, red, green and blue, as each is taken from a slightly different viewpoint. The answer, at the camera end, is to use a single camera lens system, and, with suitable mirrors and filters behind the lens, separate out the red, green and blue parts of the image, and project each on to a separate camera tube. (See front cover.)

S. C. Ryder-Smith graduated from Queen Mary College, University of London, with a degree in Electrical Engineering in 1956. After initial training as a graduate apprentice with S.T.C. he joined the staff of their transistor division applications laboratory. Here, besides general circuit design work, he made a special study of voltage breakdown in transistors, and published various works on this subject. He now heads the market developments group in the S.T.C. component marketing division.

At the receiver end of the chain, there is also the problem of aligning the three scparate colour pictures, and presenting them on a single screen. This may be overcome by depositing three different phosphors, in some pre-determined pattern, on the screen of a c.r.t. which is equipped with three separate electron guns. One phosphor emits red light when excited, another green, and the last blue. It is arranged that the three electron beams coming from the guns scan together under the influence of a single set of scan coils, but that the beam from one gun can excite only the red phosphor, the beam from the next gun only the green, and the beam from the last gun only the blue. The way in which this is achieved in the shadowmask tube has been described in detail in the March 1967 Wireless World but a diagram from this article is repeated here as Fig. 2 to show the basic principle. A tube of this sort is capable of producing three superimposed pictures, one red, one green and one blue, in which the strengths of the red, green or blue components can be independently varied by changing the grid voltages on the appropriate electron guns.

A more practical form of the colour system shown in Fig. 1 can now be devised. This is shown in Fig. 3.
The system arrived at in Fig. 3 would make an excellent basis for a colour service, if it weren't for two drawbacks. In the first place, three separate transmission paths are needed, and hence three times the bandwidth. Secondly, any normal black-and-white receiver could receive only one of these colour signals, and would get a picture with grossly distorted tonal values (equivalent to looking at a scene through a strong red, green or blue filter).

The problem, then, is to find a way of transmitting the $R$ (red,) $G$ (green) and $B$ (bluc) information in such a way that a black-and-white set, with no modifications, will display a good picture with no tonal distortion. In addition, the total bandwidth used for the transmission must be no greater than that allocared for normal black and white, and yet a colour receiver must be able to recover from this signal the $R, G$ and $B$ information.

The way in which the $R, G$ and $B$ signals are coded to form a single combined signal for transmission is ingenious. First, a new signal, $Y$, is formed, by adding portions of the $R, G$ and $B$ signals:

$$
Y=0.30 R+0.59 G+0.11 B
$$

In this equation, it is assumed that a maximum red output is represented by $R=1$, and a zero red output by $R=0$. A similar assumption is made for $G$ and $B$.

By adding together the red, green and blue picture signals in this way, what results is a signal representing the blaek-and-white view of the scene. A normal monochrome set can therefore receive the $Y$ signal and reproduce the correct black-and-white picture. The reason why only 0.11 of the blue signal is used, whereas 0.59 of the green is used, is a matter of human physiology. The human eye is much less sensitive to blue than to green. A bright green appears to the human eye lighter


Fig. 1. Simple colour television system.
> (Below) Fig. 2. Principle of shadow mask c.r.t. : (a) beams converging on mask and diverging on to screen; (b) close-up of mask and screen.
than a bright bluc. Therefore, when a brilliant green is being televised, $G=1, R=B=0$, and $Y=0.59$ (a light grey). When a brilliant blue is being televised, $B=1$, $R=G=0$, and $Y=0.11$ (a darker grey). Producing $Y$ according to the equation given above therefore results in a black-and-white picture with a tonal range acceptable to the human eye.

So far, the encoding described has merely reduced the colour signals to a black-and-white signal. How does a colour set separate out the original $R, G$ and $B$ signals?

First, for the sake of compatibility with black-and-white sets, it has been necessary to produce the $Y$ signal. Further independent signals must now be provided so that a colour set can use them in conjunction with the $Y$ signal to produce the original $R, G$ and $B$ information. There are, in fact, two additional signals:

$$
(R-Y) \text { and }(B-Y)
$$

and these are called colour-difference signals because, as can be seen, they result from subtracting the $Y$ signal from colour component signals.

Adding the $Y$ signal to the two colour difference signals gives

$$
\begin{aligned}
& (R-Y)+Y=R \\
& (B-Y)+Y=B
\end{aligned}
$$

Therefore a colour receiver can use the incoming $Y$, ( $R-Y$ ) and ( $B-Y$ ) signals to produce the original $R$ and $B$ signals. There is still the problem of obtaining the $G$ signal in the receiver. This can be done by making use of the following mathematical relationship.
$0.30(R-Y)+0.59(G-Y)+0.11(B-Y)$

$$
\begin{aligned}
& =0.30 R+0.59 G-0.11 B \\
& \quad-0.30 Y-0.59 Y-0.11 Y \\
& =0.30 R+0.59 G+0.11 B \\
& -Y(0.30+0.59+0.11) \\
& =0.30 R+0.59 G-0.11 B-Y
\end{aligned}
$$

But $Y=0.30 R+0.59 G+0.11 B$
$\therefore 0.30(R-Y)+0.59(G-Y)+0.11(B-Y)=0$
It follows from this that

$$
-(G-Y)=\frac{0.30}{0.59}(R-Y)+\frac{0.11}{0.59}(B-Y)
$$

In other words, if the two incoming colour difference signals $(R-Y)$ and $(B-Y)$ are added together in the correct proportion, and the sign of the resulting signal is changed, a signal equal to ( $G-Y$ ) can be produced.

A simplified schematic of the decoding in the receiver is shown in Fig. 4.


Fig. 3. More practical form of Fig. 1 system.
Note that the final comparison between the colour difference signals and $Y$ is achieved by feeding a negativegoing voltage proportional to $Y$ (indicated as $-Y$ ) to the cathodes of all three electron guns, while voltages proportional to the colour difference signals are fed to the grids of the appropriate guns. The beam current in any gun is determined by the difference between the cathode and grid voltages. Thus, in the red gun the beam


SUPPRESSED CARRIER MODULATION


NORMAL AMPLITUDE MODULATED SIGNAL
Fig. 5. Suppressed corrier modulation compared with a.m.
oulated signal


CHANGE-OVER SWITCH POSITIONS

MODULATED IMPUT



Fig. 6. Suppressed corrier demodulation in the receiver.


Fig. 7. Demodulator oscillator $90^{\circ}$ out of phase.
current and hence the amount of excitation of the red phosphor, is proportional to:-

$$
R-\bar{Y})-(-Y)=R
$$

Similariy, for the green and blue guns:

$$
\begin{aligned}
& (G-Y)-(-Y)=G \\
& (B-Y)-(-Y)=B
\end{aligned}
$$

The same effect could be produced by adding the colour difference signals to the $Y$ signal before reaching the colour tube.
It has already been noted that the $Y$ signal gives a good black-and-white representation of the scene being televised. What do the colour difference signals represent? First, assume that a scene containing no colouronly black, white, and the intermediate greys-is being televised. Pure white may simply be defined as having equal quantitics of red, green and bluc. Thercfore $R=G=B=n$, where $n=1$ for full white, intermediate values for greys, and zero for black.

$$
\begin{aligned}
Y & =0.30 n+0.59 n+0.11 n
\end{aligned}
$$

The colour difference signals become

$$
\begin{aligned}
& R-Y=n-n=0 \\
& G-Y=n-n=0 \\
& B-Y=n-n=0
\end{aligned}
$$

Therefore, when a black and white picture is being transmitted, $Y$ continues to have a valuc representing the tonal value, or luminance of the scene, but the colour difference signals disappear. The colour difference signals only have a value once colour is introduced into the scene. It becomes obvious therefore, that the function of the colour difference signals is simply to provide information as to the colour of a scene, while the brilliance, or luminance, of the scene is conveyed in the $Y$ signal. For this reason, the $Y$ signal is called the luminance signal and the $(R-Y),(G-Y),(B-Y)$ signals are called the chrominance signals.

Experiment has shown that while the human eye is sensitive to detail arising from differences in luminosity, it is relatively insensitive to details arising from colour changes only. A benefit of this is that while the $Y$ signal must be transmitted at full bandwidth to get good definition, the chrominance signals can be transmitted with a considerably reduced bandwidth.
The remaining problem in constructing a practical colour television system is how to transmit the ( $R-Y$ ) and ( $B-Y$ ) signals without (a) increasing the overall bandwidth of the system, and (b) interfering significantly with the operation of a norınal black-and-white set displaying the picture due to the $Y$ signal.
The methods described so far are common to all colour systems. Where
N.T.S.C., PAL, and SECAM differ is in the methods adopted in transmitting the $(R-Y)$ and $(B-Y)$ signals.

Most of the credit for making colour television possible must go to the developers of the N.T.S.C. system. PAL is basically N.T.S.C. with modifications based on the now extensive experience of the problems and operation of N.T.S.C. in the U.S.A.

## N.T.S.C. TRANSMISSION SYSTEM

The basic problem has been outlined above: How to transmit the $(R-Y)$ and ( $B-Y$ ) signals in addition to the $Y$ signal without increasing the transmission bandwidth, or interfering unduly with the reception of the $Y$ signal by a normal black-and-white receiver. The problem is complicated by the fact that the $(R-Y)$ and $(B-Y)$ signals can have either a positive or a negative value. Normal methods of modulation deal only in magnitude and not with sign.

The solution adopted in N.T.S.C. has been to use suppressed carrier modulation. A simple way of looking at this type of modulation is to assume that the modulating waveform is chopped by the carrier. The waveforms resulting from this operation are shown in Fig. 5. For comparison, a normal a.m. signal is also shown.

Note that when the modulating signal is zero, with suppressed carrier modulation the output is also zero. With amplitude modulation, on the other hand, a zero modulating signal is represented by a carrier of constart amplitude. Demodulating an amplitude modulated signal is simple: a normal diode detector will do the job. With suppressed carrier modulation, however, demodulation is a major difficulty. The method normally employed is to make use of a second electronic change-over switch operated in exact synch ronism with the modulating switch. The demodulation process is illustrated in Fig. 6.

For this sort of demodulation to work successfully, there must exist within the receiver an oscillator which is not only precisely locked in frequency to the carrier oscillator at the transmitter, but is also closely in phase with the transmitter oscillator. Fig. 7 shows what happens when the demodulating oscillator is $90^{\circ}$ out of phase with the incoming signals.

In this case, when the high trequency elements of the output are filtered out, the net output is zero.

Although there is obviously a drawback in the fact that the local oscillator in the receiver must be phase as well as frequency locked to the carrier oscillator in the transmitter, advantage can be taken of this phase sensitivity. It has been shown that if the carrier modulating the signal is $90^{\circ}$ out of the phase with the receiver oscillator, then demodulation produces zero output (after filtering the high frequency components). Take the case where the modulated signal and the local demodulating oscillator are exactly in phase, and a correctly demodulated output is being obtained. If a second signal is added to the original modulated signal, having an identical carrier frequency but being $90^{\circ}$ out of phase, then this second signal will not produce any changes in the demodulated output, just because it is $90^{\circ}$ out of phase. However, if a second demodulator is used, driven from the same local carrier oscillator, but with a 90 phase change infroduced, then this demodulator will produce an output


Fig. 8. Technique for conveying two independent sets of information.
from the second signal, and the original signal will give a zero output. This is illustrated in Fig. 8.

Thus it is possible for a single signal to carry two independent channels of information.
In the N.T.S.C. colour system advantage is taken of this by using a single suppressed carrier modulated signal to convey both the $(R-Y)$ and ( $B-Y$ ) information. The carrier frequency selected is in the region of 4.4 MHz . The $Y$ signal is, of course, transmitted in the normal amplitude modulation mode used for black-and-white transmissions. The suppressed carrier chrominance signal, centred on 4.4 MHz , and containing both the $(R-Y)$ and $(B-Y)$ signals, is then added to the $Y$ waveform, and treated as normal video information.

Although this method neatly solves the problem of transmitted $(R-Y)$ and $(B-Y)$ information with no increase in the overall bandwidth, two questions immediately spring to mind. Surely the chrominance signal will appear on the screen as normal high frequency video? Secondly, will high frequency video arising from the picture content be interpreted as chrominance information, and affect the colour? In other words, the luminance, or $Y$, signal can interfere with the chrominance signals $(R-Y)$ and $(B-Y)$ and vice-versa.

This cross coupling does in fact occur. But by a careful choice of chrominance carrier frequency-in PAL it is 4.43361875 MHz - the effects can be minimised. The chrominance signal produces a fine and unobtrusive pattern of dots across the screen, and fine detail in the picture content can produce a small distortion in the colour. However, both of these effects are small.

There still remains the problem of ensuring that the local carrier oscillator in the receiver is in frequency and phase lock with the transmitted carrier. This is done by

choosing a part of the transmitted waveform where video information is not present-i.e., during the sync pulse and tly-back period, and transmitting a short burst of carrier. A gate in the receiver separates this from the rest of the video wave-form, and feeds it to the local oscillator to synchronize it.

Although there are, of course, a number of sophistications to the N.T.S.C. system not described here, the main outline of the method has been covered.

## PAL

The major shortcoming of the N.T.S.C. system has proved to be its sensitivity to phase errors in the chrominance channel. Fairly exact phase relationships must be kept if proper separation between the $(R-Y)$ and $(B-Y)$ channels is to be achieved. Once phase errors do occur, then false $(R-Y)$ and $(B-Y)$ information is given, and colour reproduction deteriorates. A particularly sensitive area for phase errors to occur is, of course, the transmission path between the transmitter and receiver. N.T.S.C. receivers must therefore be equipped with a " hue" control to correct for these phase errors, and under adverse conditions fairly frequent adjustments to this control are necessary.

The purpose of PAL is to take the N.T.S.C. system, and modify it to make it less sensitive to phase errors in the chrominance channel. This is donc by inverting the carrier phase of the $(R-Y)$ signal on alternate lines. This is why the system is called PAL-Phase Alternation Line. Fig. 9 shows how the phase inversion is obtained at the transmitter by an electronic switch. In the receiver,
a corresponding switch is operated on alternate lines, which restores the $(R-Y)$ signal to its correct phase. The consequence of this phase alternation is that any phase error which occurs during one line is balanced by an equal phase error in the opposite sense in the following line. (Originally, of course, the phase error is always in the same sense on each line. But alternate lines are phase reversed in the receiver to correct the phase alternation of the $(R-Y)$ signal. The phase error is therefore also inverted on alternate lines, and the average phase error is reduced to approaching zero).

It is of course necessary for the receiver to identify what line is being transmitted-one with $(R-Y)$ normal or phase inverted. This is done by phase inverving the burst of colour carrier on aliernate lines in synchronism with the phase inversion of the $(R-Y)$ signal.

In PAL, phase errors in one line are balanced by equal and opposite phase errors in the following line. In a simple PAL receiver, PAL-S, averaging of these crrors is left to the human eye. Where the errors are small, this can be quite satisfactory. However, large errors lead to a coarse line structure, sometimes referred to as the Hanover blind effect. A more satisfactory solution is to perform the averaging electronically. This is done in a PAL-D receiver by means of a delay line which delays the chrominance signal for the exact duration of one line. Each sine of chrominance information, as well as being directly fed to the c.r.t., is also fed into the delay line, and added to the following line, where the phase errors cancel.
The block diagram of a complete PAL-D receiver is shown in Fig. 12. This looks at first rather terrifying.


Fig. 10. Schematic of PAL-D colour receiver.

Taken bit by bit, and in the right. order, however, it turns out to be relatively harmless.

The signal from the aerial is fed into a normal tuner, identical with the tuner used in a b!ack-and-white set. Tlee output of the tuner, at i.f., is fed to the vision i.f. amplifier, from which it goes into the vision detector. The sound signal is picked off from the vision i.f. amplifier, or alternatively, the vision detector, and goes through a normal sound charnel to the loudspeaker.

The video output from the vision detector is fed through a video pre amplifier, a delay line giving a delay of approximately $0.6 \mu \mathrm{~s}$ and the luminance amplifier (basically a normal video output stage). The output signal consists of the luminance signal $Y$, plus the unwanted, but unavoidable, encoded chrominance information. This output is fed to the cathodes of the three c.r.t. guns. The delay circuit of $0.6 \mu \mathrm{~s}$ is not the main PAL delay line. Its function is to compensate for the short delay which the chrominance signals undergo in passing through the decoding circuits.

Before going through the $0.6 \mu \mathrm{~s}$ delay line, the video signal is also fed to two other circuits. The first is the a.g.c. sync separator. This circuit provides (a) the required sync pulses, which are taken to the line and irame timebases, and (b) an a.g.c. signal which is used to control the gains of the tuner and vision i.f. amplifier. The second circuit to receive the video signal is the chrominance amplifier. Here, that part of the video signal which contains the chrominance information is filtered out and amplified prior to demodulation.

## PHASE LOCKING SYSTEM

At this point it is as well to leave the direct chrominance signal path, and follow the parts of the circuit used to provide a correct phase locked carrier for the chrominance demodulation. The burst gate is connected to the output of the first chrominance amplifier. This gate is opened for a short period during the start of each line scan by a signal derived from the line timebase. The colour burst, transmitted to phase lock the local oscillator of the receiver, occurs during the period when the burst gate is open. The burst amplifier therefore receives the colour burst, but no other part of the video waveform. The output of the burst amplifier is compared in phase with the output of the local oscillator, which is crystal controlled.

It will be remembered that the phase of the colour burst alternates from line to line, and that the phase of the colour burst on any one line provides information on whether the $(R-Y)$ signal has its normal phase, or is phase inverted. In fact it is arranged that the colour burst phase changes back and forth by $90^{\circ}$. On one line it leads the required local oscillator phase by $45^{\circ}$, and on the following line it lags the required phase by $45^{\circ}$. The output of the phase detector is, therefore, a signal varying positive and negative at half line frequency. It is arranged that the circuit containing a reactance (variable capacitance) diode used to control the phase of the crystal oscillator is much ton slow to follow the line to line variations in the output of the phase detector. Instead, it takes up a mean position, which is, of course, the required phase.

Meanwhile, the 7.5 kHz (half line frequency) signal at the phase detector is used for two purposes. A bistable circuit is driven from the output of the line oscillator and changes state at the start of each line. Its output is used to phase invert the drive to the ( $R-Y$ ) demodulator on each alternate line, in order to correct for the phase inversion given to the $(R-Y)$ signal on alternate
lines at the transmitter. However, it is obviously necessary to phase invert the $(R-Y)$ demodulator drive on the same lines as which the $(R \sim Y)$ signal is phase inverted. Information on which lines have the phase inverted $\left(R-Y^{\prime}\right)$ signal is contained, as already explained, in the phase of the colour burst. As the alternation in phase of the colour burst from line to line gives rise to the 7.5 kHz signal at the phase detector, this 7.5 kHz signal can be used to identify the line with $(R-Y)$ phase inverted. For this reason, this signal is referred to as the "ident " signal. It is fed to the bistable which is constrained to operate in phase with the ident signal. In this way, the phase inversion of the drive to the $(R-Y)$ demodulator is made to occur always on the alternate lines on which the $(R-Y)$ signal is phase inverted.

A second function of the ident signal is this. When a black-and-white picture only is being transmitted, it is important that no luminance information gets through the chrominance chamnel. If it did, parts of the picture where fine detail were present might appear coloured, and this is obviously very undesirable in a black-andwhite transmission. This problem is solved quite simply. When a black-and-white picture is being transmitted, no colour bursts are included in the video waveform. The 7.5 kHz signal therefore does not appear at the phase detector. In its absence, the colour killer circuit comes into operation, and turns the second chrominance amplifier off. It follows that when the colour bursts are absent, no information at all can get through the chrominance channel.

Let us return now to the chrominance signal at the output of the chrominance amplifier. This is fed into the PAL delay line, and also into a circuit which adds it to the output of the delay line, and another which subtracts the output of the delay line. Remembering that the output of the delay line represents information from the preceding line, in which the $(R-Y)$ information will have an opposite phase, the result of adding and subtracting adjacent lines of information can readily be calculated.

If the signal emerging from the delay line is $\pm(R-Y)$ $+(B-Y)$ then the signal coming from the chrominance amplifier output, representing the following line of information, will be

$$
\mp(R-Y)+(B-Y)
$$

Adding these two lines gives

$$
\pm(R-Y) \mp(R-Y)+2(B-Y)=2\left(B-Y^{\prime}\right)
$$

Subtracting gives

$$
\begin{aligned}
& \pm(\mathrm{R}-\mathrm{Y}) \frac{ \pm}{2(R-Y}(R-\mathrm{Y})+(\mathrm{B}-\mathrm{Y})-(\mathrm{B}-\mathrm{Y}) \\
&= \pm \frac{1}{2(R-}
\end{aligned}
$$

This part of the circuit, the delay line, adder and subtractor, therefore carries out two functions: it provides the phase error correction by averaging between succeeding lines, which is a basic feature of PAL, and it also separates the $(R-Y)$ and $(B-Y)$ signals. Both signals are fed into synchronous detectors (these are the demodulators described earlier). The ( $B-Y$ ) demodulator is driven direct from the phase locked crystal oscillator: The $(R-Y)$ demodulator derives its drive from the crystat oscillator, after it has first passed through (a) a 90 phase shift circuit, and (b) the phase inverter switch described above.

The two demodutators produce at their outputs the original $(B-Y)$ and $(R-Y)$ signals. These then go to a matrix where they are added in the correct proportions to produce the $(G-Y)$ signal. Finally, all the colour difference signals are taken to the grids of the appropriate guns in the shadow-mask colour cathode-ray tube.

# Emitter-coupled, Emittertimed Multivibrators 

## 1: Astable Circuits

ASTABLE and monostable multivibrators are well known and widely used pulse circuits. The astable multivibrator switches repetitively between two quasi-stable states generating a series of rectangular pulses. The monostable circuit has one stabie state in which it remains until a suitable trigger pulse is applied, causing it to switch rapidly to a quasi-stable state, in which it remains for a period of time, before returning to its original state; thereby generating a single rectangular pulse for each trigger pulse. The characteristics of these circuits that are normally considered to be of importance are: stability of pulse amplitude and width with respect to changes in supply voltages, temperature, spread in transistor parameters and switching time between states.

The most common forms of the multivibrator circuits are the collector base coupled versions shown in Figs 1 and 2 , in which the timing function is performed in the base circuit. The transistors are normally saturated in order to stabilize pulse amplitude against changes in transistor parameters, but the pulse amplitude is still dependent on supply voltage and pulse durations are affected by changes in temperature. The less well known emitter-coupled, emitter-timed forms of the circuits possess definite advantages in that the timing operation is performed in the emitter circuit resulting in a pulse


Fig. 1. (left) A conventional collector-base coupled astable multivibrator.

Fig. 2. (right) A conventionalbase coupled monostable multivibrator.


Fig. 3. (left) Basic emitter coupled, emitter timed, multivibrotor.

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Fig. 4. Waveforms of an emitter timed multivibrator.


Fig. 5. Equivalent circuit of Fig. 4. (c) Trl on-Tr2 off; (b) Tr2 onTrl off.
duration which is far less dependent on transistor parameters and, therefore, orr temperature variations. In addition the pulse duration and amplitude can be made less dependent on power supply variations and the minimum switching times for a particular transistor type can be realized. Also a square waveform can be produced and the circuits have a completely " free " collector.

The circuit of an emitter-timed astable multivibrator is given in Fig. 3 and its idealised waveforms in Fig. 4. The circuit loop between the emitter of Tr1collector Tr1-base Tr2-emitter Tr2 and the emitter of Trl is regenerative, so that both transistors conduct together only during the rapid switching between states. As Trl switches on the potential at its collector, and at the base of $\operatorname{Tr} 2$, falls rapidly causing $\operatorname{Tr} 2$ to cut off. The emitter current of $\operatorname{Tr} 1$ is then made up of two components (Fig. 5a), $I_{1}$ flows through $R_{3}$ and $I_{2}$ charges capacitor $C$ causing the porential at the emitter of $\operatorname{Tr} 2$ to fall. After a time $t_{1}, \mathrm{Tr} 2$ comes into conduction again and a regenerative action takes place causing the emitter

Fig. 6. Practical emitter timed astable requiring only a single power supply.

of $\operatorname{Tr} 2$ to be driven suddenly positive. This change is communicated by capacitor $C$ to the emitter of $\operatorname{Tr}$, cutting it off. The emitter current of Tr 2 is then also made up of two components (Fig. 5b), $I_{4}$ through $R_{1}$ and $I_{3}$ charging capacitor $C$ and causing the potential at the
cmitter of Trl to run down for a period $\mathrm{t}_{2}$. Then Trl comes into conduction again and the cycle repeats.
An approximate analysis of the circuit may be carried out if it is assumed that the negative step at the collector of Trl is small compared with the negative supply, the charging currents will then be taken as being constant during the timing periods. The effect of lakage currents will be ignored for silicon transistors.

The negative step at the base of $\operatorname{Tr} 2$ is $x_{r b}\left(I_{1}+I_{2}\right) R_{3}$ which is approximately equal to $\left(I_{1}+I_{2}\right) R_{1}$ since $\alpha_{\alpha b} \approx$ to unity.

Capacitor $C$ must charge by an amount :
$J V=\left(I_{1}+I_{2}\right) R_{1}-\left(\delta V_{B}+V_{b e 2}\right)$
Where $\delta V_{e}$ is the step at the emitter of $\operatorname{Tr} 2$ and $V_{\text {be } 2}$ is the difference between the base emitter voltage of $\operatorname{Tr} 2$ when switching occurs and the base emitter voltage when the current $I_{4}$ is flowing.
Current $I_{3} \approx I_{1}=\frac{V_{c e}-V_{b e 2}}{R_{3}} \quad \ldots \quad . \quad$.
Continued on page 636


Fig. 7. Woveforms present in the circuit of Fig. 6. (o) taken at the collector of Trl; (b) emitter Trl; (c) the collector at Tr2; (d) emitter of Tr2; all taken with $R_{2}$ at $220 \Omega$. The slope at the top and battom of the waveforms is due to the charging currents not remaining constant. (e) Waveform at the emitter and ( $f$ ) at the collector of Trl where the timing copacitor $=1,000 \mathrm{pF}$. (g) Collector of Tr 2 when the timing capacitor is reduced to 100 pF , the smallest rise time for o particular tronsistor type is realized. (h) Ubper-collector and lower-emitter of Trl (i) Tr 2 when the timing capacitor $=0.01 \mu \mathrm{~F}$ and $\mathrm{R}: \mathrm{g}=470$ §. The effect of allowing $\operatorname{Tr} 2$ to saturate can be clearly observed.


Fig. 8. Variation of frequency plotted against supply voltage for the circuit shown in Fig. 6 with $C=0.1 \mu F, R_{2}=470 \Omega$.


Fig. 9. Improved astable multivibrotor.


Fig. 10. Showing charging paths in the circuit of Fig. 9: (a) Tr2 onTrl off; (b) Tri on-Tr 2 off.
and

$$
\begin{equation*}
I_{4} \approx I_{2}=\frac{V_{c 1}+V_{e e}-V_{b e 2}}{R_{1}} \tag{3}
\end{equation*}
$$

The time period $t_{1} \approx \frac{\Delta V C}{I_{2}} \approx \frac{\left(I_{1}+I_{2}\right)}{I_{2}} R_{1} C$

$$
t_{2} \approx \frac{\Delta V C}{I_{3}} \approx \frac{\left(I_{1}+I_{2}\right)}{I_{1}} R_{1} C
$$

The term ( $\delta V+\delta V_{b e 2}$ ) is neglected. Substituting for $I_{1}$ and $I_{2}$ gives:
$t_{1} \approx\left[1+\frac{R_{4}}{R_{3}} \frac{1}{\left.1+\frac{V_{c 1}}{V_{e e}-V_{\text {ee1 }}}\right)}\right] C R_{1}$
$\boldsymbol{t}_{2} \approx\left[1+\frac{R_{3}}{R_{4}}\left(1+\frac{V_{c 1}}{V_{e e}-V_{b e 1}}\right)\right] C R_{1}$
$V_{b e 1}$ is assumed equal to $V_{b e 2}$.
If the currents $I_{1}$ and $I_{2}$ are made equal, $t_{1}=t_{2}$, and the frequency is approximately equal to $1 /\left(4 C R_{1}\right)$
Both timing periods are seen to depend on the ratio $V_{c 1} /\left(V_{e e}-V_{b e 1}\right)$ which indicates the possibility of obtaining a multivibrator with very good frequency stability against changes in supply voltage. Increasing the supply voltages, with this ratio held constant, causes an increase in the charging currents, but it also causes the same fractional increase in the voltage step through which the capacitor has to charge. The constancy of the ratio can be assured by using only one power supply and a resistive divider ( $R_{6}, R_{7}$ ). The need for the second positive supply may be removed by including the resistor $R_{3}$; the circuit is shown in Fig. 6.

In the above equations we can now replace $V_{c 1}$ by:

$$
V_{c}^{\prime}=V_{c c} \frac{R_{5}}{R_{1}+R_{5}} \quad \text { and } R_{1} \text { by } R_{1}^{\prime}=\frac{R_{1} R_{5}}{R_{1}+R_{5}}
$$

An emitter-coupled emitter-timed multivibrator is required, operating frequency around $100 \mathrm{kc} / \mathrm{s}$ and a markspace ratio not far from unity. The design procedure is as follows. Using $f=1 /\left(4 C R_{1}{ }^{\prime}\right)$, if $C$ is made $0.01 \mu F$ then $R_{1}^{\prime}$ must be $250 \Omega$.
But $R_{1}^{\prime}=R_{1} R_{5} /\left(R_{1}+R_{5}\right)$ so we make $R_{1}=R_{5}$ $=470 \Omega$.
With a nominal supply voltage of 20 V the resistive divider was chosen to give an effective emitter supply of -15 V . This makes $V_{c}^{\prime}=2.5 \mathrm{~V}$. The approximate values of the charging currents are determined from equations (2) and (3). $R_{3}=3.3 \mathrm{k} \Omega . R_{4}=4.7 \mathrm{k} \Omega$. makes $I_{1} \approx I_{2}=3.5 \mathrm{~mA}$. The value of $R_{2}$ determines the amplitude of the signal at the collector of $\operatorname{Tr} 2$. Two different values were tried, $220 \Omega$ and $470 \Omega$. It was found that the $470 \Omega$ resistor caused Tr2 to saturate. The transistors employed were inexpensive plastic encapsulated general purpose silicon planar type made by Texas Instruments.

Fig. 7 shows the oscillographs taken from the circuit of Fig 6 and demonstrates clearly the effects of altering the values of $R_{2}$ and the timing capacitor.

The frequency dependence of the circuit on supply voltage was measured with $R_{2}=220 \Omega$ and $R_{2}=470 \Omega$. In the former case a supply change from 15 to 25 V caused the frequency to change from 132 to $124 \mathrm{kc} / \mathrm{s}$, whilst in the latter case a change from 15 to 30 V caused a much smaller change in frequency from 126 to $123 \mathrm{kc} / \mathrm{s}$. The frequency dependence of the saturating circuit was also measured with a timing capacitor of $0.1 \mu F$. The results are indicated graphically in Fig. 8. A change of


Fig. 11. Oscillograms taken in the circuit of Fig. 9. (a) collector Tr1; (b) emitter Trl; (c) collector Tr2; (d) emitter Tr2. It can be seen that the mark-space ratio is very close to unity and the top and bottom of the waveforms are flatter thon those of the circuit of Fig. 3. (e) The rise time and (f) the foll time of the waveforms at the collector of Tr2 showing the ropid switching time.
supply from 15 to 30 V changed the frequency by less than $1 \%$ overall. The difference in behaviour between the saturating and the non-saturating circuits are considered to be largely due to the terms $\delta V_{e}$ and $V_{b e z}$ as these alter with changes in charging currents. However, these changes are smaller when Tr 2 is allowed to saturate.

A modified circuit was designed which has the rapid switching and sharply defined waveforms associated with non-saturating operation but whose frequency stability against changes in power supply voltage is superior to the saturating circuit discussed above. The mark space ratio of the waveforms is very close to unity and the pulse height varies little with changes in supply voltage. The circuit is shown in Fig. 9; the emitter resistors are replaced by Tr 4 which acts as a constant current source. Diodes D1, D2 and transistor $\operatorname{Tr} 3$ cause the whole of this current to charge capacitor $C$ during the timing periods.

Assume that a regenerative action has just resulted in Trl being driven into cut off. The constant current supplied by Tr 4 charges capacitor $C$, the charging path being through Tr2 and D1 (Fig. 10a). D2 and the emitter base junction of $\operatorname{Tr} 3$ are reverse biased. The potential at the emitter of Trl falls at a uniform rate, and, after a period of time $t_{2}, \mathrm{Trl}$ comes into conduction and a regenerative action switches off Tr 2 . The forward bias across the emitter base junction of Trl and the voltage drop across $R^{*}$, prevents D1 from conducting and the constant current charges $C$ through $\operatorname{Tr} 1, \mathrm{Tr} 3$ and D2 (Fig. 10b). The emitter of 'Tr2 falls at the same uniform rate at which the emitter of Trl fell (assuming the $\alpha_{c o}$ of transistor $\operatorname{Tr} 3$ is close to unity), for a time $t_{1}$ until $\operatorname{Tr} 2$ comes into conduction again and the regenerative action switches off Trl repeating the cycle. If we neglect the step in the emitter voltage of $\operatorname{Tr} 2, \delta V$ and the small term $\delta V_{b c 2}$ (eq. 1) capacitor $C$ has to charge through a voltage $\Delta V=I_{c} R_{1}^{\prime}$. Where $I_{s}$ is the constant current


Fig. 12. Variation of frequency with supply voltage for the circuit of Fig. 9. $\mathrm{C}=1,000 \mathrm{pF}$.
supplied by $\operatorname{Tr} 4, R_{1}{ }^{\prime}$ is the effective collector load resistance of Trl. The rate of charging is the same in both
cases $\frac{I_{c}}{C} V /$ sec.
so: $t_{1}=t_{2}=\frac{\Delta V \cdot C}{I_{c}}=C \cdot R^{\prime}$
The frequency of oscillation, $f=1 /\left(2 C \cdot R^{\prime}\right)$. The performance of the modified multivibrator is illustrated by the oscillograms of Fig. 11. The frequency dependence on the supply voltage was measured, a change of supply from 15 to 30 V caused a change of frequency of $0.5 \%$. With a timing capacitor of 1000 pF stability was good and is illustrated in Fig. 12. A change in supply from 15 to 30 V altered the frequency by $0.14 \%$ ard the pulse height, at the collector of $\operatorname{Tr} 2$, by less than $10 \%$.
(Next month: monostable circuits)

## REFERENCES

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2. Mullard Technical Communications. April' '62.

## Post Office Receiving Station Refurbished

THE TRANSITION from Nissen huts and manually operated equipment to brick buildings and automatically tuned radio receivers is now complete at the G.P.O. high-frequency ( 4 to 27 MHz ) radio receiving station at Bearley in Warwicks. The new installations cost about $£ 0.5 \mathrm{M}$. This station will combine efficient and reliable reception of long-distance radiotelephone and radiotelegraph communications with the maximum possible economy. Atthough much of the future transoceanic signal traffic will be carried by submarine cables and Earth satellites, h.f. radio can still play a useful role in world communications in lightly loaded routes for communicating with ships and for auxiliary and standby purposes alongside cable and satellite systems.

An outstanding feature of Bearley is the frequency generating equipment which controls the accuracy of the receiver synthesizers. It consists of three 100 kHz crystal controlled oscillators sunk into 30 feet deep boreholes where the temperature remains within about $0.5^{\circ} \mathrm{C}$ of $10^{\circ} \mathrm{C}$ without any artificial control. The accuracy of this master frequency can be maintained to within one part in ten million, with adjustment at about yearly intervals, or, if required, to 1 in $10^{8}$ with adjustments about once a month. This central master frequency source provides, by synthesis, the extremely accurate beat oscillator frequencies. The majority of the 60 receivers at this station are solid-state i.s.b. types suitable for the reception of telephony or multi-channel telegraphy and were designed by Plessey Electronics Group to a Post

Office specification. The PVR 800 , as it is called, is a quadruple superheterodyne receiver capable of remote control for tuning either to any one of the six predetermined frequencies or by fully synthesized control selecting any one of the 200,000 discrete channels available (in increments of 125 Hz ). Because of the accuracy of the synthesized frequencics the receiver can carry out an automatic carrier search process for, and identify, a wanted carrier signal. When the wanted transmission is found, the receiver can automatically maintan correct tuning, providing the transmitter frequency variations do not exceed internationally agreed limits.
The original aerial system has been retained more or less unchanged. A ring of rhombics ( 70 ft high), efficient over the important band of frequencies above 8 MHz , combines global coverage with facilities for special aerial diversity reception.. Diversity operation is necessary to achieve efficient reception of telegraph transmissions. In this case, two similar aerials spaced several wavelengths apart feed two separate receivers whose outputs are combined. This method of space-diversity reception (compared with single aerial reception) is said to be equivalent to increasing the power of the distant transmitter by upwards of 30 times. All rhombics at Bearley are bi-directional, each rhombic end being terminated at the internal aerial distribution board, where, by means of a wideband passive hybrid network, it can serve up to four receivers simultancously

## New Earlh Satellite Station in Australia

WITH work well up to schedule, the new Earth station being built at Morec, in northwest New South Wales, by the Australian Overseas Telecommunications Commission, is expected to be in service by the beginning of the year. The total cost of the project is more than \$A4 million. It is the eighth space communications establishment built or in the planning stage in Australia. The Moree satellite communications station will be employed to link Australia into the Intelsat II satellite system, providing commercial communications and television transmission and reception with North America and major points in the Pacific. Countries which will be served will include the U.S., Canada, Japan, the Philippines, Hong Kong and other countries of Eastern Asia. It will supplement the $\$ A 250$ million broad band coaxial cable system which Australia and other Commonweath partners have built across the Pacific and Atlantic Oceans. The new station will send and reccive signals via the Intelsat satellite positioned directly over Fiji. Intelsat II was launched
from Cape Kennedy for the International Satellite Consortium of which Australia is a foundation.

A high degree of accuracy was required in siting the structure of the Moree Earth station. It had to run precisely due north and south. Margin for error was only $0.000008 \%$ or 10 ft in 23,000 miles. The station has been built on a 257 -acre site and it includes a 90 ft parabolic antenna weighing 200 tons, mounted on a four-storey operations building. Australia's other space communications establishments are at Cooby Creek, Queensland, three stations near Canberra, in the Australian Capital Territory, two stations associated with the Woomera Rocket Range installations, and two in Western Australia, at Muchea and Carnarvon. NASA is reported to be considering establishment of a further station in the Canberra area, but no official announcement has yet been made about this project. Altogether more than \$A100 million has been spent in Australia on these projects in the past six years.

## Changes in Marilime Radio Regulations

SUBSTANTIAL amendments have been made to those parts of the 1959 Radio Regulations and Additional Radio Regulations which apply to the maritime mobile service. This is a result of the World Administrative Radio Conference which was convened in Geneva on the 18th September by the International Telecommunication Union and which completed its work on 3rd November with the signing of the Final Acts. These will come into force on the 1st April 1969. The amendments have been determined substantially by the fact that since the last revision of the regulations in 1959, there has been a significant drop in the number of passenger ships owing to the growth of air travel, and a notable increase in the number of cargo ships. There has also been a rapid expansion in fishing fleets and other craft.

Thus requirements for radiotelephone and radiotelegraph channels have increased considerably

Among the decisions of the Conference are the following: the gradual introduction up to 1st January 1982 of s.s.b. radiotelephony in the bands allocated to the maritime service between 1605 and 4000 kHz ; the gradual introduction up to Ist January 1978 of s.s.b. radiotelephony in the bands between 4 and 23 MHz ; allocation of frequencies for narrowband direct printing telegraph systems (teleprinters) and data transmission systems; assignment of frequencies for the transmission of oceanographic data; and in general measures to increase safety at sea (signal code, watch on distress signals, etc.), including conditions governing the use of emergency position-indicating radio beacons.

## Subscriplion Television

IT APPEARS from American press reports that a decision authorizing subscription television on a regular basis has been delayed for yet another year by Congress. Ever since the Zenith Radio Corporation first demonstrated the technical feasibility of this form of television viewing twenty years ago there has been controversy. The Federal Communications Commission is faced with the problem of putting pay television into operation while providing adequate protection for existing commercial stations who, with theatre owners, strongly oppose such a system, which they say, would ruin "free" television and the theatre. At the same time the F.C.C. has to consider the right of the public to choose a system where they were willing to pay for programmes uninterrupted by commercials.
Here in Great Britain a subscription television system has been operating experimentally in London and Sheffield by Pay-TV and during the past year the programmes have included feature films which were screened six months after general release. An indication of the prospects for subscription television in this country is expected soon from the Postmaster-General.

## Communicalions Experiments

THE THIRD in a series of five applications technology satellites, ATS-C, was launched from Cape Kennedy on 3rd November. Among the nine experiments carried on board is one concerned with communications. This will be conducted using two microwave repeaters (receiver/transmitter) which constitute the spacecraft's s.h.f. communications subsystem. Both repeaters operate in three modes, the first two (multiple access and frequency translation) are used in a microwave communications test, and the third mode, wideband data, is used for transmitting television pictures from the spacecraft's spin scan cloud cover camera to the ground. The basic objective of operating the repeaters in the first mode is to evaluate the s.s.b. technique for muitiple access communications. This technique is a promising approach to the development of a multiple access system where two or more ground stations use the spacecraft simultancously, since it affords a maximum number of voice channels in the minimum bandwidth of the overcrowded radio frequencies. The repeaters are operated in the third mode for evaluating a high quality f.m. system for relaying wideband data such as colour television, digital and facsimile signals. The f.m. system used for these tests is a refinement of those installed in the Telstar and Syncom communications satellites.

## Sophisticated Surplus

A NEW generation of surplus electronic equipment is now becoming generally available on the open market as computer after computer ends its days at the breaker's yard. For instance, at the recent R.S.G.B. exhibition one could buy a bank of 26 unused thermionic digital indicators neatly mounted on a paxolin printed circuit strip and marked with all power supply voltages for the princely sum of 10 s (less than the cost of one of the indicators). Clearly there are many bargains about provided the reader is prepared to search for them and separate the wheat from the chaff. Some boards are coated with a thin layer of epoxy resin rendering it extremely difficult, though not impossible, to salvage any usable components. A range of boards, ex I.B.M. computers, do not suffer from this defect and in many cases are usable more or less as they stand for the original purpose they were intended. These boards contain gates, bistables, differential amplifiers, etc,, and cost in the region of a couple of shillings each. One example contained four two-input NAND gates that operated quite happily from a $6-\mathrm{V}$
supply; such boards should prove very useful to schools and colleges. Buying these items is semething of a lucky dip and a good deal of time mast be syent in tracing out individual circuits to discover what one actually has. The I.B.M. boards mav be obtained from Pattrick \& Kinnic or L.S.T. Components.

Information Services provided by the I.E.E. in the fields of physics, electrorechnology, and control are known collectively as INSPEC. Exploitation and development of this facility is to be assisted by a grarit from the Office for Scientific and Technical Irformation, and by collaboration with the Institute of Electrical and Electronics Engineers, and the American Institute of Physics. The object of this expansion programme in 1968 is to extend the present service (limited to publication of Science Abstructs and Current Papers) to include a service of selective dissemination of information S.D.I. The above facilities will be changed to a computerbased service, and the present publications will be produced by computer methods from the January issues in 1969. From the same date S.D.I. will come into operation, and magnetic tapes containing data concerning all literature processed by INSPEC will be available.

The possibility of a nationally recognized qualification and title for technician engineers was discussed by 31 engineering institutions and societies and members of the Council of Engincering Institutions on the lst December. The result was that those organizations who are outside the C.E.I. would group into like interests, and each group prepare and submit recommendations for a joint meeting with C.E.I. in February nest.

## ANvoUVCEMENTS

Ten weekly lectures on studio audio control equipment begin at the Northern Polytechnic, Holloway, London N.7, on January 11th. The fee is 21 s and application forms can be obtained from the Head of Department of Electronic and Communications Enginecring.

A series of 12 weckly lectures on piezo-electric devices and their applications will be held at Southall College of Technology, Beaconsficld Road, Southall, Middx., commencing January 17th. The course fee is 6 gn .

A course of six lectures on u.h.f./s.h.f. techniques will be held at Nor wood Teclinical College, Knight's Hill, London, S.E.27, commencing 20th February. The lectures will take place each Tuesday evening. Fee is 15 s .

Mr. E. W. Weaver, Director of the London Telecommunications Region of the G.P.O., formally opened London's first p.c.m. telephone link (between Sunbury, Middlesex and Central London) on November 27 th.

A new company, Electronic Brokers Lid., has been formed to collect and offer prompt payment for electronic couinment and components at present lying unused in many British companies. The head office of this company is at $8 \mathrm{Broad}-$ fields Avenue, Edgware, Middx.

A series of one-week courses on vacnum te hnology will be held during 1968 at Edwarde. High Vacuum Ltd., Vanor Royal, Crawley, Sussex. Details are available from the Customer Training Officer.

AEI-Thorn Semiconductors, Lincoln, are providing a maskmaking service for industrial, a ademic and research establishments. Sample masks within ten days are offered. Plates of up to two-inches square can have a registration to within 40 in.

Film strips and slide sets produced by Mullard will now be distributed by Educational Systems Lid., ESL Heuce, Imperial Drive, North Harrow, Middlesex. (Tel: 01-868 4400.)

## PERSONALITIES

R. I. Walker has been appointed chief engineer of the Semiconductor Division of the Ferranti Electronics Department at Gem Mill, Oldham, Lancs. Mr. Walker, who has been with the company for seven years, occupying the position of deputy chief engineer, was formerly with the Services Electronics Research Laboratory, at Baldock, where he was responsible in the late 1950s for much of the early development work on silicon mesa transistors. Ferranti also announce the appointment of Alan Williamson as product marketing manager, discrete components, and Brian Down as product marketing manager, intergrated circuits. Mr. Williamson has been with the company for seven years latterly as senior field sales support engineer, and Mr. Down, who was formerly in the application group of the Ferranti Semiconductor Division, has rejoined the company after two years with Texas Instruments.

John S. Walker, M.Sc., F.I.E.E., who recently joined De La Rue Instruments Ltd., as managing director, has for the past 10 years been with Texa Instruments Ltd. where he was latterly manager of the Research and Development Department. From 1949 to 1953 Mr . Walker was at Manchester University where he took a course in physics, which

J. S. Walker
he followed by an M.Sc. course in electrical engineering in 1952/53. He then spent two years with Standard Telephones and Cables. In 1955 Mr. Walker joined International Computers and Tabulators and then the British Tabulating Machine Co. before going to Texas Instruments. Mr. Walker is a member of the I.E.E. Panels on Semiconductor Devices and Integrated Circuits.
G. H. Stearman, B.Sc.(Eng.), M.I.E.E., for the past ten years lecturer at the Col-
lege of Aeronautics, Cranfield, where he specialized in electronics and digital techniques, has joined Feedback Ltd., of Crowborough, Sussex, as development department manager. He obtained his degree at Brighton Technical College and was with Cable \& Wircless Ltd., for

G. H. Stearman
two years before joining Southern Instruments Ltd. in 1951 where he stayed for six years. In 1964 Mr. Stearman was seconded for a year to the National Aeronautical Laboratory at Bangalore.
D. G. Smee, M.B.E., A.M.I.E.E., commercial director of the Marconi Company since 1965, has been appointed chairman of the board of directors of Elliott-Automation Microelectronics Ltd., which forms part of the ElliottAutomation group of companies recently acquired by English Electric (parent company of Marconi). In this new position he will be responsible for coordinating the activities of Marconi and E.A.M. in the field of microelectronics. Mr. Smee, who is 50 , joined the Marconi Company in 1933, working at the Research Laboratories until the outbreak of war in 1939, when he joined the Royal Signals. He returned to Marconi in 1946, and in 1950 became assistant commercial manager. Six years later he was appointed manager of the Company's Broadcasting Division.
D. H. Roberts, B.Sc., M.I.E.E., F.Inst.P., for some time head of solidstate research at Plessey's Allen Clark Research Centre at Caswell, Northants, has become general manager of the company's Semiconductor Division at Swindon in succession to Brigadier J. D. Haig who is appointed general manager of overseas plant operations. Mr. Roberts joined the Plessey laboratories at Caswell in 1953 after graduating in physics at Manchester University. Also transferred from Caswell to the Swindon
production team are: W. Holt, B.Sc., A.R.C.S., aged 34, who joined Plessey in 1961 from Marconi's Research Laboratories, and has been chief development engineer at the Allen Clark Research Centre; R. C. Foss, B.Sc., Ph.D., M.I.E.E., aged 31, principal engineer, integrated circuit development, at the Centre, who joined Plessey in 1964 from E.M.I. Electronics; and M. J. G. Gay, A.M.I.E.E., aged 30, who joined Plessey from the Mullard Thin Film Unit in 1964 and has been in charge of circuit techniques research at the Caswell Rescarch Centre.
S. N. Ray, M.Sc., B.Sc.(Eng.), M.I.E.E., F.Inst.P., acting head of the Department of Electrical and Electronic Engineering, Borough Polytechnic, London, for the past year, has retired. Born in Calcutta in 1902, Mr. Ray came to this country after receiving his M.Sc. degree from Calcutta University in 1925 and continued his studies for his B.Sc. (London) and the Diploma of Faraday House. For 11 years he was in the radio industry and joined the staff at the Polytechnic in 1939. He was senior lecturer in radio engineering until he was appointed principal lecturer in applied electronics in 1961. He has been acting head of the Department of Electrical and Electronic Engineering since V. Pereira-Mendoza, M.Sc.Tech., F.I.E.E., became principal in 1966. The new head of the Department is Kenneth W. E. Gravett, M.Sc.(Eng.), M.I.E.E., A.M.I.E.R.E., who has been senior lecturer in electrical measurements at

K. W. E. Gravett
the Brighton College of Technology. After graduating at King's College, University of London (where he also obtained his master's degree), he served an apprenticeship with the British Thomson-Houston Company at Rugby. He subsequently held appointments at the Post Office Research Station and on the staff of the Battersea College of Technology.

# Semiconductor Type Numbering 

Some guidelines through the chaos of type code numbers that face you nowadays

By T. D. TOWERS,*

M.B.E., M.A., C.Eng.

THERE is a lovely old proverb that runs: " Who buys has need of a hundred eyes." How true this is when you set out to select a transistor or a diode nowadays from the host of different kinds of type numbers used, either from one of several "standard" sytems in operation, or from the non-standard systems used by individual manufacturers. In Great Britain you can come across transistors or diodes of almost any nationality. If you are going to find your way confidently among them, you have to know something of the basic numbering systems used, and these are discussed individually below.

JEDEC system.-Probably the oldest standard numbering system in current common use is the American " JEDEC." $\dagger$ In this, the Electronic Industries Association (E.I.A.), in the United States, registers devices from specifications put up by manufacturers. It uses a numbering system in which the first numeral shows how many diode junctions the device has, with a " 1 " for a diode, a " 2 " for a triode transistor and a " 3 " for a tetrode. After this initial numeral comes an "N," and then the number in serial order under which the device was registered with the authority. As an example, the " 2 N 914 " is the 914 th triode transistor registered.

By the end of 1967, both 1N (diode) and 2N (triode) numbers registered had passed the 5,000 mark.

Any manufacturer, provided he meets the specification as registered by the original manufacturer with E.I.A., can supply devices to JEDEC numbers. The full details of any individual registered device can be obtained from E.I.A., 2001 Eye St., N.W., Washington, D.C., 20006. Unfortunately, they do not publish an easily available comprehensive authoritative list of JEDEC devices and their characteristics.

PRO ELECTRON system.-Although the JEDEC standard numbering has come into fairly common use in Europe, there is a European standard system, known as "PRO ELECTRON," which is also widely used here in parallel with JEDEC. The organizing authority is the Association Internationale PRO ELECTRON, of 10, Avenue Hamoir, Brussels.

As with the JEDEC system, the manufacturer registers with PRO ELECTRON a device he has developed. Any other manufacturer can then supply devices marked with the same registered number, provided his version also meets the electrical and mechanical specification registered with PRO ELECTRON.

The PRO ELECTRON system has one big advantage over JEDEC. All you can tell from a JEDEC number is whether the device is a diode, triode, etc., and some indication of the time of registration, since low numbers mean the device was registered years ago. With PRO ELECTRON, the letters and numbers used are much more significant.

The PRO ELECTRON type number always has five places: cither two letters and three numerals (as in $\mathrm{BC107)}$ or three letters and two numerals (as in BCY72). The first letter indicates the bulk semiconductor material used: $\mathrm{A}=$ germanium; $\mathrm{B}=$ silicon; $\mathrm{C}=$ gallium arsenide; and $\mathbf{R}=$ compound photo-conductive material.
The second letter indicates the circuit type of the device: $\mathrm{A}=$ signal diode, non-power; $\mathrm{B}=$ variable capacitance diode; $\mathrm{C}=$ transistor, 1.f., non-power; $\mathrm{D}=$ transistor, l.f., power; $\mathrm{E}=$ tunnel diode; $\mathrm{F}=$ transistor, h.f., nonpower; $\mathrm{G}=$ multiple device; $\mathrm{H}=$ field probe; $\mathrm{K}=$ Hall generator; $L=$ transistor, h.f., power; $M=$ Hall modulator or multiplier; $\mathrm{P}=$ radiation sensitive device (photodiode, photo-transistor or photo-conductive device); $\mathrm{Q}=$ radiation generating device; $\mathrm{R}=$ specialized breakdown device; $\mathrm{S}=$ transistor, switching, non-power; $\mathrm{T}=$ controlling and switching device with breakdown characteristics, power (s.c.r. or thyristor, etc.); $\mathbf{U}=$ transistor, switching, power; $\mathrm{X}=$ multiple diode; $\mathrm{Y}=$ rectifier, power; and $Z=Z e n e r$ diode (voltage reference or regulator).
The final three places of the PRO ELECTRON five-place registration number give an indication of the general area of use and a serial number. Where three numerals are used (e.g., BC107) this indicates a device for "entertainment" or "consumer" use; i.e., for radio or television receivers, audio amplifiers, tape recorders, etc. The three numbers run from 100 to 999 . Where a letter and two numerals are in the last three places (c.g., BCY72), this indicates a device for use in industrial and professional equipment. The letters (which bear no significance) in this case start from $Z$ back through $Y$, X , etc. The accompanying serial numbers run from 10 to 99 only.
Sub-classifications are permitted in certain devices such as Zener diodes, power diodes and thyristors (s.c.rs) in the PRO ELECTRON system. These are indicated by further codings added after a hyphen at the end of the five-place basic number according to a significant system.

For Zeners, the code addition gives information on the nominal voltage and its tolerance. The tolerance appears first as a single letter: $\mathrm{A}=1 ; \mathrm{B}=2 ; \mathrm{C}=5 \%$; $\mathrm{D}=10 \%$; and $\mathrm{E}=15 \%$. The nominal voltage follows as a numeral plus the letter V in the position of the decimal point where necessary. Thus BZY88-C9V1 represents a silicon Zener for industrial use, with registration number Y88, tolerance $5 \%$ and nominal voltage 9.1 V

For rectifiers and thyristors, the additional FRO ELECTRON code numbers signify the repetitive peak reverse voltage in volts. Thus BYX35-100 indizares a silicon rectifier for industrial use with registration number X36 and a $100-\mathrm{V}$ rating, while the BTY99-100 represents a silicon thyristor for industrial use with registration number Y99 and a $100-\mathrm{V}$ rating. With

[^1]power rectifiers and thyristors, the cathode is normally connected to the stud mounting. Where the anode is connected to the stud ("reverse polarity"), a final letter R is added. By this a BTY99-100R signifies a reverse-polarity BTY99-100.

Recently supplementary codings have arisen for ordinary transistors, too. You may come across the well-known BC108 in versions coded BC108A, B and C. The final letter suffix in this case denotes a narrowspread selection of current gain within the wider spread limits of the basic BCl 08 device.
Old European coding system.-The PRO ELECTRON system has become widely accepted in Europe during the 1960 s , and is often referred to as the "new" European system. It has replaced the old European system under which semiconductors were indicated by an initial " O " (standing for zero heater volts in the then existing valve coding). After the initial O came a letter in the coding with $\mathrm{A}=$ diode, $\mathrm{C}=$ triode, etc., and a registration number. Many readers will remember with nostalgia such codings as the OC71 transistor and the OA81diode. Devices are still being marketed under this old system, but it is to be expected that they will ultimately disappear.
Japanese system.-Japanese transistors appearing for sale and in equipment in Britain over the last decade have faced engineers with a new set of numbers according to a standard widely used in Japan. The first two symbols of the code are " $2 S$ " for triode transistor, and the third gives an indication of the general characteristics of the transistor according to the following code: $A=p-n-p$, r.f.; $B=p-n-p$, a.f.; $C=n-p-n$, r.f.; and $\mathrm{D}=\mathrm{n}-\mathrm{p}-\mathrm{n}$, a.f. As an illustration, the 2 SA 49 is a $\mathrm{p}-\mathrm{n}-\mathrm{p}$, r.f. transistor with registration number 49.
"Services" standard systems.-On the British market, the user will occasionally come across devices bearing type numbers according to some Government standard.

The commonest of these are the "CV" types, where the type designation consists of the letters CV followed by a four- (and recently five-) digit number. In the future this is likely to be supplemented by a separate British Standard (BS9000) series arising out of the work of the celebrated Burghard Committee.

The British Post Office, too, has in the past issued its own series of semiconductor specifications and users may come across these in a self-evident numbering series, $\mathrm{PO} 1, \mathrm{PO} 2$, ctc.

The only other Government numbering system the

ordinary user is likely to meet is the American "Mil. Spec." series corresponding to the British "CV" system. Under this coding, devices are normally specified as the corresponding commercial JEDEC number with the preficx "JAN" added; e.g., JAN 2N3093 is the Mil. Spec. version of the 2N3093. This is the current procedure, but Mil. Spec. devices may also be found coded under the previous system, where the prefix indicated the branch of the services sponsoring the device. The single JAN prefix now used replaces the separate prefixes USA, USAF and USN formerly used. The "Mil. Spec." jargon name for these devices arose because they were related to a specification document numbered Mil-S19500, where the individual devices were distinguished by a suffix number; for example, the 2 N 914 has the designation Mil-S-19500/373 in its military version.

House Codes.-Most manufacturers sell semiconductor devices under their own special serics of "house" numbers, as well as under numbers according to one of the standard systems. Some of these house codes have woven themselves firmly into the structure of the British market, and it will be long before they disappear

Some guide to the transistor house codings is given in Table I, which shows the more common initial letters used by semiconductor manufacturers in the U.K. Diode house codes tend to be much more numerous and less distinctive than transistor codes and are not therefore included.

Apart from the house numbers put out in published data, semiconductor manufacturers sell a considerable portion of their output under special or "private" house numbers. Little guidance can be given on this to the general user, but, if he comes across a device the characteristics of which he cannot trace, he can always write to the manufacturer (whose name should appear on the device along with the type number).

A final mystifying feature of transistor numbers is that large users frequently lay down their own "in-house" specifications with their own code numbers, and manufacturers mark the devices they supply with these "inhouse" numbers. When you come across one of these, it is, I fear, not easy to find out details of its specification.

## INFORMATION ON SEMICONDUCTOR DEVICE TYPES

Having discussed the many different methods of coding a semiconductor device which may be met with in practice, the reader can be forgiven if he thinks: "That is all very well, but where can I go to find out the characteristics of any particular device?" In the case of a device in a standard numbering system such as JEDEC or PRO ELECTRON, he could write direct to the registration authority, but this can be a long and expensive procedure. The ordinary engineer-in-the-street usually turns to one of the commercial publications described below.

The most complete current commercial tabulations of data on semiconductor device types are published by Derivation and Tabulation Associates, Inc., of 32 Lincoln Avenue, Orange, New Jersey, 17050, U.S.A. Three of their publications circulate world wide among semiconductor users.

Transistor D.A.T.A. Book.-This is a characteristics tabulation for virtually every transisior (about 13,000 types at the time of writing) commercially available in the world. It is completely revised biennially in Spring and Autumn, with separate updating supplements in

Summer and Winter. The annual subscription is currently $\$ 30.50$ in the U.K. It does not include obsolete transistors, but there is a separate publication for these.

Discontinued Transistor Yearbook and Replacement D.A.T.A. Book.-This is an annual edition issued each Summer and is a compilation of all discontinued types since 1956. Each edition costs $\$ 15.25$ in the U.K. Diodes are covered by a third publication.

Semiconductor Diode and SCR D.A.T.A. Book.-This covers virtually every type of available diodes and already runs to some 66,000 entries. It is issued in complete revisions in Spring and Autumn and the annual subscription is $\$ 39.50$ in the U.K. These three "D.A.T.A." books give sufficiently detailed tabulation of characteristics for most uses of the devices, and in addition give mechanical outlines. For the user of many semiconductor types, they have become almost "bibles." But they are expensive, and less ambitious students have to turn to more modest publications.

Iliffe's Radio Valve Data.-This data tabulation (covering transistors and diodes as well as thermionic valves) is the successor to the well-known Wireless World Valve Data Manual and still costs only a modest 9s 6 d . Even so, it is probably the best easily available data tabulation for British semiconductor devices, and it has the useful feature of being brought up-to-date regularly

Avo's International Transistor Data Manual.-This rransistor tabulation, issued by Avo Ltd. for use with their commercial transistor tester, is also marketed separately at 45 s . It, too, is a most useful general data tabulation, with many features not commonly found. For example, it contains listings of CV and Russian transistors.

Other commercial tabulations.-There are a number of other commercial listings of transistors published, but they are generally less useful than those described above, either because they tend to go out of date or are aimed primarily at a non-British market. For completeness, however, some of the more easily available are listed helow:
(i) "Techpress" Transistor Specifications and Substitution Handbook, 1967, by Techpress Inc., Brownsburg, Indiana 46112.
(ii) Transistor Specifications Manual, by Foulsham-Sams Technical Books, W. Foulsham and Co., Ltd., Slough, Bucks.
(iii) "Datadex" Transistor Reference Book by M. W'.
table I
INITIALS OF TRANSISTOR HOUSE CODES IN COMMON USE BY MANUFACTURERS IN THE U.K.

| C,CP | SGS-Fairchild |
| :--- | :--- |
| DT | Lucas Semiconductors |
| FI, FK, FM, | SGS-Fairchild |
| FSP, FT, FV |  |
| GET | Mulfard-G.E.C. (Assoc. Semiconductors) |
| GM | Texas Instruments |
| HT | Emihus |
| M | Motorola |
| NKT | Newmarket Transistors |
| P | SGS-Fairchild |
| PEP | A.E.I. Semiconductors |
| V | Newmarket Transistors |
| SE | SGS-Fairchild |
| TH. TM | Texas Instruments |
| TK | S.T.C. |
| ZDT, ZT, ZTX | Ferranti Semiconductors |
| 2G, $2 S$ | Texas Instruments |



Fig. 1. Simplified mechanical details of the moie common standard JEDEC "TO" transistor ouilines (Typical dimensions only).

Lads Publishing Co., Philadelphia, P.A.
(iv) British Transisior Directory, by E. N. Bradley, Norman Price (Publishers) Ltd., London.
(v) Guide Mondialc des Transistors, by Société des Editions Radio, 9 rue Jacob, 75, Paris, 6.

In all this, it should not be overlooked that if you write to any semiconductor manufacturer he will be pleased to send you information on his devices.

## INFORMATION ON SEMICONDUCTOR OUTLINES

In the early days of transistors, fifteen years ago, manufacturers invented their own device shapes and lead configurations, but of recent years there has been considerable standardization.

JEDEC outlines.-As in device numbering, the E.I.A. in the U.S.A. led the way in outlines. It registered the dimensions of certain preferred cases or encapsulations for semiconductor devices under "TO" (transistor) and "DO" (diode) outline standard numbers. Full details of the outlines so registered can be found in the JEDEC publication 12E, "Registered Outhines and Gauges for Semiconductor Devices." You can also find the JEDEC ourlines at the end of the D.A.T.A. publications described earlier.

Some of the registered JEDEC outlines have virtually dropped out of use with time, but certain "standard" ones have been adopied by most manufacturers. In transistors, the commoner outlines in use are TO1, TO3, TO5, TO8, TO18, TO46, TO66, and TO72. Simplified drawings of these are given in Fig. 1.

VASCA outlines.-Over here some moves towards outline standardization have been made. A "Record of Semiconductor Outlines" from the Electronic Valve and Semiconductor Manufacturers' Association, Mappin House, 156/162 Oxford Street, London, W.1, gives details of the VASCA system, in which outlines are registered under an "SO" (semiconductor outline) number related to the American "TO" JEDEC numbers. VASCA also registers semiconductor lead configurations under an "SB" (semiconductor base) series.
I.E.C. outlines.-A separate standard numbering system for registered outlines has been developed by the International Electrotechnical Commission (I.E.C.), 1 Rue de Varembe, Geneva, Switzerland, and issued in their Publication 191-2, "Mechanical Standardizataion of Semiconductor Devices." Both this and the VASCA publication relate their standard outline numbers to JEDEC and to other standards.

CV outlines.-In the numbering of semiconductor outlines, you may come across the British Government CV system which typifies various outlines according to an appendix number to a semiconductor code popularly known as K1007. Thus probably the commonest encapsulation for silicon small-signal transistors appears as K1007/A1-D14, as well as JEDEC TO18, I.E.C. 2-106, and VASCA SO12A.

## CONCLUSION

Although we have examined most of the multifarious type and outline coding systems used by manufacturers, it would seem at long last standardization is beginning to take hold. The bulk of semiconductor devices used in the British market in the future are likely to be coded on either the PRO ELECTRON or the JEDEC numbering systems (with a few house codes sprinkled around), and outlines will generally be described by the JEDEC "TO" system.

# Units and their Abbreviations 

READERS may have noticed that we have been gradually introducing the hertz $(\mathrm{Hz})$ as the name for the unit of frequency in place of $\mathrm{c} / \mathrm{s}$ over the past few months. Much' was said in support of both of these names in the course of an argument in our correspondence columns early in 1967, but there is no question that the hertz is now being widely adopted and is here to stay. Wireless World therefore intends to standardize on Hz , together with its multiples, $\mathrm{kHz}, \mathrm{MHz}, \mathrm{GHz}$ and THz.

Since the hertz is an internationally recommended name for one of the derived SI (Systeme Internationalc) units,* this seems an appropriate time for $W . W$. to standardize on SI units generally. In practice this means that there are no changes to the most common electrical units and their symbols (V, A, $\Omega, W, C, J, F, H$, etc.). Since, however, SI is really a development of the m.k.s. (metre-kilogramme-second) system and therefore brings in metric units for length and mass in place of British measures, some of the other SI units appropriate to electronics and communications may be rather unfamiliar. A selection of these is listed (right) with comments. With frequency it has only been necessary to change the name of the unit -its value has not been affected. The SI unit names in the table, however, represent units of different size from the older-established units, and so one has to use conversion factors to change the older units into SI units or vice versa.
Although the basic unit of length in the SI system is the metre, it would obviously be impracticable, at the present juncture, to abandon the British inch, foot, yard and other units of length completely. These will still be widely used in physical dimensions, for example chassis and cabinet sizes. We shall therefore adopt a policy of introducing the metric units of length gradually,

[^2]| Quantity | Unit and Abbreviati | on Remarks $\dagger$ |
| :---: | :---: | :---: |
| Short wavelengths (as in light) | micron ( $\mu \mathrm{m}$ ) | Replacing angstrom unit (A) |
| Forse (as in transducers) | newton (N) | $=\mathrm{kg} \mathrm{m} / \mathrm{s}^{2}$. Replacing pound-force (lbf), poundal (pdl), dyne (dyn). |
| Pressure (e.g. acoustics, transducers) | newton per square metre ( $\mathrm{N} / \mathrm{m}^{2}$ ) | Replacing $\mathrm{lb} / / \mathrm{in}^{2}$, dyn/ $\mathrm{cm}^{2}$, in $\mathrm{H}_{2} \mathrm{O}, \mathrm{mmHg}$, torr, bar, atm. etc. |
| Magnetic flux | weber (Wb) | $=\mathrm{V}$ 5. Replacing lines, Maxwell. |
| Magneric flux density | tesla (T) | $=\mathrm{Wb} / \mathrm{m}^{2}$. Replacing gauss, lines $/ \mathrm{cm}^{\text {b }}$, Maxwells $/ \mathrm{cm}^{2}$. |
| Magnetic field strength | ampere permetre $(A / m)$ | Replacing oersted. |
| Illumination (e.g. selevision, opto-electronics) | lux (lx) | $=1 \mathrm{~m} / \mathrm{m}^{2}$. Replasing footcandle, lumen per square foot ( $1 \mathrm{~m} / \mathrm{ft}^{2}$ ). |
| Luminance | candela persquare metre ( $\mathrm{cd} / \mathrm{m}^{2}$ ) | Replacing foot-lambert, $\mathrm{cd} / \mathrm{ft}{ }^{*}, \mathrm{~cd} / \mathrm{in}^{2}$ |

Conversion factors between SI and other units are given in the N.P.L. booklet "Changing to the Metric System
in some cases using them alone, in others printing them alongside the British units. A similar method of gradual introduction will be adopted with other physical quantities for which the present, non-SI, units are widely used and familiar to our readers.

## OUR COVER

THE theme of colour television is portrayed by the dichroic prismatic separation system employed in the Philips three-Plumbicon camera. Several of these cameras, which are marketed in the United Kingdom by Peto Scott, are being used by the B.B.C. for its colour service which opened on December 2nd.

## A Critique of Class D Amplifiers for A.F.

# 2: THE DESIGN OF A CIRCUIT 

By K. C. JOHNSON, M.A.

THE first article in this series considered the advantages and disadvantages of the class $D$ principle of operation for power amplifiers in general and for transistor audio circuits in particular. The conclusion formed was that the class D principle does not lead to any overwhelming advantages and that such circuits are not likely to displace the conventional class $\mathbf{B}$ type on any large scale. Nevertheless, they do have considerable intrinsic interest and readers may like to see a circuit that the author has developed which attempts to exploit as many of the special features of the class $D$ principle as possible. This circuit uses the simple feedback form of modulator for generating the switching wave form, despite its comparatively poor distortion characteristics, since any improvement requires unjustifiable extra complexity. The last two stages work in class $D$; three might have been so employed to give a lower standing current in exchange for a. lower maximum output amplitude and lower efficiency at the larger output levels. The top-cut filter at the output is a simple choke, although some small improvement in performance could be gained by using a more complicated network.

## OUTPUT STAGE DESIGN

The circuit diagram of the complete amplifier appears in Fig. 4. It will be seen that the two final transistors Tr6 and $\operatorname{Tr} 7$ are employed as switches to provide a powerful square-wave voltage source from which current is drawn through an audio band filter to the loudspeaker; essentially as shown in Fig. 2 last month. The diodes D2 and D3 are included because the relatively low frequency current required for the loudspeaker in such an arrangement will often be flowing "backwards" with respect to the voltage being generated by the switching action and this backwards current is carried by these diodes. It cannot be carried by the transistors unless they are made to meet severe "symmetrical" ratings in addition to the other difficult requirements, since the currents involved are substantially equal to the peak currents that the devices must be able to carry in the forward direction. Notice that this reverse current is carrying power back from the reactive components in the filter network to the power supply, and that it is directly because of this returning of unwanted power that the class $D$ system is potentially so highly efficient.

The two transistors thus work in conjunction with the diodes opposite to them, as is indicated in the drawing; when large amplitudes are being handled only one such pair is switching the real current at any one time while the other carries little or nothing. Because the variations in the audio signal are comparatively slow it is possible to connect the centre-tapped inductor $L_{1}$ as shown on the diagram without any significant effect on the basic switching action. When, however, real transistors are
being switched at a speed approaching the limit of their capabilities it will always in practice be found difficult to ensure that a perfect "break-before-make" action is obtained, and this inductor helps to prevent any serious build-up of unwanted current due to this transient overlap of transistor conduction. In this circuit the strapping of the bases of the drive stage, $\operatorname{Tr} 4$ and $\operatorname{Tr} 5$, ensures that such overlaps will never be very serious, but the extra inductor costs little and enables the transistors to be switched that tiny bit faster with better standing current. The detailed design of this stage, and indeed of the whole circuit depends on the characteristics of the transistors selected for the positions Tr6 and Tr7. As has been said already this choice is difficult; it it were not so, a complementary pair of devices would be employed and several advantages realized, but at present it is difficult to find a single adequate type and hence an arrangement of the form familiar in conventional circuits is used, where only the drivers Tr4 and Tr5 need be complementary. Accordingly a single transistor type serves for both positions, so that reasonable matching is easily achieved. The device chosen is the Fairchild BC 119 , this allows a maximum current of 1 A , with guaranteed saturation to 1.5 V when the base drive is 100 mA , it has a cut-off frequency of at least $40 \mathrm{Mc} / \mathrm{s}$ and a maximum voltage, at any allowable current level, of 30 V without avalanche breakdown. It is an $n-p-n$ silicon planar epitaxial transistor in a TOS case.

The use of this device, to within its ratangs, fixes the power supply voltage at a maximum value of 30 V . If, as proposed last month, the modulation level is restricted to $60 \%$ of the ultimate value on account of the sideband distortion effects, then the available output amplitude at the loudspeaker cannot possibly be guaranteed to exceed $\pm 7.5 \mathrm{~V}$ because of the allowances that must be made for ohmic losses in both the transistors and the filter network. If, moreover, the current is also held within the allowable limit then the maximum useful value will be about $\pm 0.8 \mathrm{~A}$ after taking into account the ripple in the filter. Therefore, the maximum output power that can be guaranteed is 3 W average into a loudspeaker system that has been adjusted to present a load of exactly $9.5 \Omega$. Into a speaker of different impedance the power limit will be lowered, since either the voltage or the current will be unable to reach its full value.

Needless to say any pair of transistors of this type will almost certainly be found to function perfectly well at twice this current and at larger voltages as well, so that more power will in fact be obtainable, but there can be no guarantee of this and neither the manufacturers nor the author can be blamed if devices fail. In a conventional circuit these same transistors can be used up to 60 V where they are always cut off before voltages above 30 are applied. Since the full 1 A current can also be used the power limit for a pair in class B working
is about 14 W into a load of 28 !? over four times more than with class D! For this output to be maintained for any length of time heat-sinks are mandatory, but this presents no real difficulty, while the very high frequency cut-off allows a large factor of feedback to be applied without any serious stabilization problem.

The diodes for the positions of D2 and D3 must be able to carry the same peak voliages and currents as the power transistors and here the selection of a suitable type is even more difficult since the forward voltage drop must be small to avoid unnecessary turning on of the opposite transistor ${ }^{5}$ and the switching speed has to be fast. The Fairchild type EB 383 can stand the reverse voltage, but the published limit values of stored charge and capacitance are barely adequate while the forward characteristic is not specified at all for currents exceeding 50 mA . To be able to guarantee the performance of the circuit a more exotic diode type ought have been employed; unfortunately none are readily available. However, the specimens of EB 383 that the author tested have proved to be entirely satisfactory. The diodes for use with these transistors should really be able to carry a forward current of 1 A at less than 1.5 V , a capacitance of not more than 10 pF and a charge-storage characteristic equivalent to a time of perhaps 10 ns .

The driver stage, comprising Tr 4 and Tr5, requires a pair of complementary devices each capable of delivering a peak current of 100 mA or more to the final transistors, with approximately matched speed and saturation characteristics. Notice that in this form of circuit the drive current is not delivered unless the load demands it, and that the final transistors are not turned on at all unless the output current exceeds perhaps 25 mA in the appropriate direction (due to the low value used for the resistors $R_{18}$ and $R_{18}$ ). This low value for the base resistors also ensures that the final transistors are switched off rapidly when required. The driver transistors must be able to stand the full supply voltage without breakdown and have adequate speed, but the current levels
are so much lower that the selection of suitable types is comparatively casy. From the Fairchild range type BC 125 will serve in $n-p-n$ position while BC 126 is the matching p-ri-p device. Both these types are TO5 size but are encapsulated in plastic. The bases of the drive transistors are connected directly together sinçe there is no critical adjustment of cross-over current needed and a bit of "slack" is indeed desirable to reduce the effects of both the top and bottom devices being turned on at the same time.

The centre-tap of the inductor $L_{1}$ provides a symmetrical output for the switching stage and it is from here that the "bootstrap" capacitor $C_{s}$, the feedback network $R_{3}$, and the main filter inductor $L_{2}$ are all driven. The value of $L_{2}$ is chosen to be $250 \mu \mathrm{H}$ and represents a compromise between the need to keep down the ripple current at the switching frequency, which causes inefficiency and reduces the available output current, and the requirement that the high audio frequencies must not be restricted. Clearly a more complicated filter network with a sharper cut-off could have been used, but the design of such an arrangement would involve nothing new and, moreover, it is rather doubtful whether the improvement would justify the trouble. Remember that these inductors must be able to carry the full loudspeaker current without magnetic saturation effects being significant, while the resistance of the windings is a major contribution to inefficiency at high power levels and must be kept low. Thus these components must inevitably be comparatively expensive and bulky.

Fig. 5 shows how both of the inductors $L_{1}$ and $L_{2}$ can be made as a single unit using two pieces of standard ferrite aerial rod for the magnetic circuit. If the reader can obtain properly designed ferrite "pot" cores then a conventional winding for each inductor can of course be used, but remember that an adequate air-gap is essential and that the capacitance between the ends of $L_{2}$ must be kept particularly small to avoid high frequencies reaching the loudspeaker leads.


Fig. 4. The circuit diagram of the amplifier.

To construct the coils wind a layer of thin insulating tape round two pieces of ferrite rod 0.3 inches in diameter and 1.5 inches long. $L_{\text {is }}$ is made up of 50 complete figure-of-eight turns of 32 s.w.g. enamel covered copper wire, care being taken to ensure tight packing at the cross-over point. $L_{1}$ consists of $10+10$ turns wound in a single layer, round both rods, using the same wire. The assembly is completed by winding with a layer of insulating tape to hold the turns in place. Using this form of construction the measured values of the coils were: $L_{1}=5.6 \mu \mathrm{H}$. 0.16 ! (each half). $L_{2}=$ $250 \mu \mathrm{H} . \quad 0.8 \Omega$. The 50 or so complete turns required are not difficult to wind by hand and form an inductor with a not too inefficient arrangement of copper, ferrite and air-gap which has a very low capacitance and doesn't require a specially made core assembly. Notice that the mutual inductance of the two coils wound in this way is comparatively negligible so that there is no question of having to connect $L_{1}$ the proper way round, and also that no exact balance between the two halves is necessary.

The loudspeaker is connected directly to this inductor while the d.c. blocking capacitors, $C_{9}$ and $C_{19}$, in parallel are between the loudspeaker and the power supply. This is a transposition of the arrangement shown in Fig. 2 last month, but there is of course no difference in the method of operation. The change is made partly to avoid the appearance of signal voltages on components which will inevitably be large, but mainly so that the "bootstrap" circuit can draw its current from these capacitors and so get it for "half price." This rather surprising possibility comes from the fact that the voltage on $C_{0}$ remains substantially constant at about half the supply voltage and that the switching circuit maintains this value by an efficient transformation action. If a current averaging 10 mA is drawn from $C_{9}$ then a current of this magnitude will flow in the inductor $L_{2}$, but the transistors $\operatorname{Tr} 5$ and $\operatorname{Tr} 7$ will carry this current, on average, for only half the time, so that the steady drain at the power supply is only 5 mA or thereabouts instead of 10 . A further power saving in the "bootstrap" action is obtained by using a diode DI, rather than the usual resistor, to draw the current for charging the capacitor $C_{8}$. This becomes possible in the class D circuit since the regular switching action ensures that this capacitor is fully recharged every cycle of the carrier frequency. The capacitor $C_{*}$ is, therefore, maintained at almost half the voltage of the power supply, when the circuit is in operation, and provides a source of extra voltage to ensure that $\operatorname{Tr} 5$ is adequately saturated when $\operatorname{Tr} 3$ is cut off, in the usual way.
It will be noticed that $C_{0}$ and $C_{11}$ are shown to be connected in parallel on the circuit diagram and also that $C_{10}$ and $C_{12}$ are similarly arranged. This is done simply to emphasize the point that return capacitors for the fast switching currents must be mounted within one inch or two of the transistors to reduce radiation. It will not be practical to mount the whole of the large canacitor that is required at this position. Accordingly it is suggested that a relatively small bart of the capacitance (even $100 \mu \mathrm{~F}$ is only $0.5 \%$ of the total) should be mounted close to the transistors while the remainder may perhaps be a few feet away as convenient. It will be seen that these capacitors are not connected as a bridge, but that $C_{10}$ and $C_{12}$ are across the whole voltage while the half-way rail is only bypassed downwards. This is done to reduce the effect on the signal of ripple on the supply due to the use of a simple cheap rectification circuit. If a good smooth power source is available then the capacitors can be used more economically if $C_{12}$


Fig. 5. Showing how to wind the inductors.
has its negative end transferred to the half-way rail, smaller values will then serve for the same low frequency performance. With this circuit the switching action may start appreciably sooner after switching on if the capacitors are connected as a bridge, but this is not a very important consideration for most applications.

## INTERMEDIATE STAGES

The two stages comprising $\operatorname{Tr} 2$ and $\operatorname{Tr} 3$ together form the hysteresis circuit shown in Fig. 3 last month. Complementary transistors are used in these positions but the requirements are not as severe as for the more powerful stages, except that $\operatorname{Tr} 3$ has to be able to carry a slightly greater peak voltage due to the "bootstrap" circuit arrangement. The current level is so very much lower that the same types can be used as in the driver stage without any worry about the ratings being exceeded.

It has already been described how the capacitor $C_{w}$ carries a " bootstrap" voltage generated from the switching square-wave by the action of the diode D1. The resistor $R_{1,5}$ is included solely to limit the diode current to a safe value during transients; note that $R_{14}$, below it, is the main load resistor for $\operatorname{Tr} 3$. The drive voltage developed by this resistor is transmitted to $\operatorname{Tr} 4$ and $\operatorname{Tr} 5$ through the capacitor $C_{i}$, and the a.c. coupling action of this capacitor ensures that the drive is substantially balanced in the two directions. Thus both $\operatorname{Tr} 4$ and $\operatorname{Tr} 5$ receive adequate currents to make certain that they saturate properly and that the minimum voltage is dropped in the final transistors so as to give both high efficiency and to avoid the generation of second and other even harmonics that would result from unbalance in the action. Once the proper operation of the circuit is established the few microamps carried by $R_{16}$ and $R_{1}$. become negligible in comparison with the base currents that the driver stage receives through $\boldsymbol{C}_{7}$. These resistors must be included, however to ensure that there is a sufficient amount of d.c. coupling between these stages to give a satisfactory "self starter" action. If at any time the circuit is not self-oscillating, there is a feedback action which automatically brings the various voltages towards their correct values, since with these resistors in circuit there is a d.c. coupling at every stage round the loop. If, for example, the voltage on the capacitor $C_{11}$ is too low, then this feedback cuts off $\operatorname{Tr} 1$, so that $\operatorname{Tr} 2$ and

Tr3 are also cut off. $R_{17}$ causes $\operatorname{Tr} 5$ to conduct, and the voltage across $C_{11}$ is made to increase. Conversely, if $C_{1}$, has too high a voltage it is "pulled down" by both $\operatorname{Tr} 3$ and $\operatorname{Tr} 4$. This action may be expected to take a few seconds whenever the amplifier is turned on. As it comes to an end the trigger circuit, $\operatorname{Tr} 2$ and $\operatorname{Tr} 3$, will switch, and when this happens a relatively large current flows through $C_{7}$ so that a powerful action occurs and the correct voltages for the proper working of the amplifier are set up within a few milliseconds.

The coupling from the collector of Tr 2 to the base of Tr 3 is designed to transmit switching edges as effectively as possible. When $\operatorname{Tr} 2$ is turned on the small capacitor C. injects a "shot" of charge into the base of Tr3 so that is comes on very quickly. When turning off is required, the relatively large capacitor $C_{6}$ is available to provide a reverse bias equai to the maximum permitted base-emitter voltage and the comparatively low value of $R_{11}$ allows a considerable reverse current to flow, so that again the action is very rapid. $R_{10}$ fixes the steady base current in the on condition while $R_{12}$ is used to set the voltage on $C_{6}$. The changes of mark-space ratio that are an essential feature of the action cause small variations in the voltage on the capacitor $C_{i}$, but these have no serious effect on the working.

The reverse coupling through $R_{\text {, }}$ and $C_{6}$ causes the required trigger effect by contributing a positive feedback current to the base of $\operatorname{Tr} 2$. Again there is a small capacitor to deliver a "shot" of charge to the base, it works both ways this time, and a resistor to give a d.c. action. It will be noticed that this feedback is taken directly from the collector of $\operatorname{Tr} 3$ whereas the resistor $R_{1}$ is included in series with the capacitor $C_{3}$ which transmits the main current to the output stages. This resistor serves two purposes. First, it evens out the quantity of current sent to $\operatorname{Tr} 4$ as $C_{7}$ discharges and its volrages gets less during loud low notes when the markspace ratio may differ from 50:50 by a considerable amount for a comparatively long time. Secondly, it ensures that each action of the trigger circuit is irrevocably started before any significant slackening of the drive to the output stages is allowed to occur.

## THE INPUT STAGE

It will be remembered that in the simple feedback modulation arrangement, which was explained last month (Fig. 3) and which is used in this amplifier, the first part of the circuit serves to integrate the error of the system. Notice that this error signal is not small, as in most ordinary feedback systems, since the output from which the feedback is taken is the full size switching square wave without any smoothing from the filter choke $L_{\text {. }}$. It is an essential feature of this system that the low frequency components in this error are kept small by the operation of the circuit. They cannor, however, be made to be exactly zero in this simple arrangement, as will be seen in the final article, and it is this finite error which causes most of the distortion in this form of amplifier.

In Fig. 4 the crror is obtained as a current resulting from unbalance in the negative feedback network formed by $R_{1}, R_{3}, R_{3}$ and $R_{1}$, and it is made to flow into the base of iransistor $\operatorname{Tr} 1$. The integration action comes from the familiar Miller effect, using the capacitance between collector and base within the transistor itself together with the voltage swing developed at the lower end of $R$. No extra capacitance is added to ensure that the contribution to this voltage from the resistor $R_{3}$ is made as large as possible in comparison with the swing at the
base of $\operatorname{Tr} 2$. This is because the latter will contain a component due to the positive feedback current from $\boldsymbol{R}_{9}$ and perhaps also some non-linearity which will both introduce inaccuracy into the integration action and hence possible extra distortion at the output. Observe that the value of $R_{\mathrm{s}}$ can be altered if an adjustment of the carrier frequency of the finished circuit is required for any reason. The current in Trl thus essentially slides smoothly up and down, with the integration of the error, between limits at which it causes the trigger pair, formed by $\operatorname{Tr} 2$ and $\operatorname{Tr} 3$, to switch by overcoming the positive feedback current from $\boldsymbol{R}_{9}$.

The level of current chosen for Tr 1 is a compromise between the requirement that the transistor must carry enough to accommodate the necessary swing without serious non-linearity at the emitter, and the need for it to still have sufficient collector voltage, even when $R_{\text {s }}$ is made relatively large, for saturation to be avoided and for the collector-base capacitance to be reasonably constant. The choice of $R_{t i}$ determines this current, since this resistor must develop the right voltage to keep Tr 2 near its switching point, the average value in this circuit is made to be about $120 \mu \mathrm{~A}$. A transistor type must thus be used which has a good performance at low levels of operation, for this the Fairchild BC 114 is very suitable. It has a typical current gain of over 200 at this current as well as both low noise and adequate ratings for voltage and speed. This device again is packaged in plastic but is of the small TO 18 size.

The emitter of this stage is held at almost half the supply voltage by the resistor chain $R_{7}$ and $R_{k}$, while $C_{3}$ provides a bypass path to the negarive rail. Since the feedback through $R_{i,}$ and $R_{2}$ is fully effective at very low frequencies, due to the inclusion of $C_{1}$ and $C_{2}$, this arrangement automatically holds the average voltage at the output of the final switching pair at the centre of the available range. This also means that the markspace ratio of the switching square wave is made to have an average value of $50: 50$. Ordinary tolerance resistors will normally serve adequately for the positions $R_{;}$and $R_{\mathrm{g}}$, but their values may be adjusted if more exact fixing of the average level is needed.

The use of a split attenuator for the feedback, with $C_{2}$ at its centre returned direct to the emitter, allows Trl to draw an appreciable amount of steady base current without upsetting this d.c. action, while at the same time it permits a high value of gain to be obtained in the audio band where $C_{2}$ has a low impedance and the attenuator has its full effect. The feedback then sets the overall voltage gain at a value which in this circuit is about 45 times. The input impedance is determined directly by the resistor $R_{1}$, since the base of Trl is a " virtual earth," and the value chosen, 220 k ! , is a compromise between the gain obtained in the amplifier and the distortions introduced by the inaccuracies in the feedback action due to the current and the voltage swing at the base of Trl.

## FEEDBACK ERRORS

An estimate of the magnitude of these inaccuracies can be obtained by considering the working conditions of the first transistor. Its mean collector current is around $120 \mu \mathrm{~A}$ and the variations necessary to give switching of the trigger arrangement will be perhaps $\pm 20 \mu \mathrm{~A}$. Thus the voltage swing at the base-emitter junction needed by the mutual conductance, will be roughly $\pm 10 \mathrm{mV}$, while the base current changes required by the current gain will typically be $\pm 0.1 \mu \mathrm{~A}$. Now these two effects are essentially similar, and as the impedance of
the feedback network as seen by the base is about $100 \mathrm{k} \Omega$, assuming a high impedance at the amplifier input terminals, the current swing is just equivalent to a further $\pm 10 \mathrm{mV}$ so that the two effects can be combined as a single effective voltage of $\pm 20 \mathrm{mV}$ at the base. However, we can if we wish consider this voltage as if it were an extra unwanted input added to the normal input, and its effective value is then $\pm 64 \mathrm{mV}$ as we must allow for the action of $R_{1}$ and $R_{2}$. The waveform of this voltage corresponds to the integral of the error of the overall feedback loop, by virtue of its derivation. That is to say that it is approximately triangular in shape with the peaks at the well defined constant levels quoted above but with the sloping parts changing with the input waveform. But since the error of the modulation system we are using is known from the theory to be given next month, its integral is also known. Each component of the error will be multiplied at the output by the factor $1-j(2 / \pi)\left(f_{c} / f_{t}\right)(64 / 300)$ where $j$ and the frequency ratio are the direct result of the integration, $f_{c}$ being the carrier frequency and $f_{E}$ the frequency of the error component under consideration; $2 / \pi$ is a constant and the $64 / 300$ is a measure of the magnitude of this effect compared with the input required to give a fully modulated square wave at the output. It will be noticed that this distortion effect appears to be most serious at low frequencies, but as we shall see next month this is just where the basic modulation distortion is least, so that the results are not necessarily so catastrophic as they seem.

A further inaccuracy in the action of the feedback arises from the fact that when the trigger circuit, Tr2 $\operatorname{Tr} 3$, switches there is a small step in the voltage at the base of $\operatorname{Tr} 2$, apart from the quick kick due to the action of $C_{3}$. This causes a corresponding step in the current through Tr1, due to the action of the integration capacitance in holding a constant voltage at the collector. It has already been pointed out that the use of a relatively large resistor for $R_{5}$ reduces this effect, but even with this circuit the voltage step will be perhaps 200 mV , so that the current will jump about $\pm 5 \mu \mathrm{~A}$. This means that in addition to its smooth integration current change the transistor is carrying a further $\pm 5 \mu \mathrm{~A}$ of current swing which follows the square wave switching action. In exactly the same way as before this can be represented as an additional signal at the input terminals, and its effective value is then +16 mV . There is no integration involved here and the effect is to increase not only all the distortion components by a uniform factor of $1+16 / 300$ but the main signal as well, so that there is no practical effect on the distortion at all. The ratio $16 / 300$ comes from the effective amplitude at the input due to this effect and the input required to fully modulate the square wave as before.

Notice that both these imperfections only introduce distortion in proportion to that which has already been generated by the failing of the basic modulation process itself. If this could be reduced these effects could become less important. Clearly, however, the design of this stage could be altered fairly easily so as to reduce them directly at the expense either of a loss of overall amplifier gain or a need for additional transistors. In this circuit the gain and economy have been preferred to the relatively small advantage that would result from their reduction. It is interesting to observe that it is the second of these two effects, the one that increases the gain by a more or less constant factor, that governs the success of the feedback in eliminating the distortions caused by errors in the edge timings and the amplitude of the square wave at the output. The factor $16 / 300$
indeed also represents the amount to which these effects are reduced by the feedback action. An apparently dramatic improvement might perhaps be gained here by the simple addition of a resistor of about 3 MO 2 directly between the collectors of Tr 2 and Tr 1 . This could be adjusted so as to exactly compensate the effect of the voltage step, but the author has not investigated this.

## CONSTRUCTION AND TESTING

In constructing this circuit it must not be forgoten that switching edges of duration shorter than $1 \mu \mathrm{~s}$ are essential in its working, so that the layout must be neat and compact with no signal leads more than an inch or so in length. All the components, except the two large capacitors, can easily be mounted on a plastic board about 4 in $\times 3$ in, and there are no special heat sink arrangements required for the final transistors. The power supply must be able to provide about 300 mA maximum current at 30 V , usual input and loudspeaker arrangements being made.

When switched on a circuit of this type should begin to function within a few seconds, but a brief pause must be expected as the voltages on the capacitors are brought to the correct values and then a faint "tick" will be heard as the self-oscillation commences. When switching on for the first time it may save needless expense if resistors of about 100 ? are put in series with both the loudspeaker and the power supply. This form of circuit is not worse than class B in this respect, indeed it is rather better, but these resistors may prevent serious damage to the expensive semiconductors in the event of faulty components or wiring errors. With them included in the circuit low amplitude signals should be reproduced reasonably well and the various voltages and currents may be checked before they are removed.

If the circuit is not functioning correctly then a fault has to be found, and as the reader may be perhaps unfamiliar with this type of circuit some guidance will be given. If the circuit is not oscillating then each stage round the loop must have its d.c. state examined until a point is found where the output is not as would be expected bearing in mind the present d.c. input conditions (regardless of the a.c. input). When this is done the fault is usually found quickly and correct functioning obtained. If on the other hand, the circuit is already oscillating then there is little difficulty in finding a break in the signal path in the usual way.

The circuit as shown in Fig. 4 has more than enough sensitivity to give a good output when driven directly from a normal crystal pickup or microphone, but there is, of course, no objection to the use of any of the usual forms of pre-amplifier if more gain or tone control facilities are required. As explained already, no claims for outstanding quality of reproduction can be made for this circuit, but it is hoped that some contributions have been made towards the exploration of the possibilities. To obtain more bass response simply increase the value of all the electrolytic capacitors; but for almost any other improvement, more power, less distortion, more gain or higher efficiency, it will almost certainly be necessary to find a superior type of transistor for the final stage and modify the design along lines that have been suggested.

Next month's article will discuss in more detail just what the errors introduced by pulse width modulation are, and how they could in principle be reduced.

## REFERENCE

5. Letters to the Editor, M. D. Salmain, Wireless World, June 1965.

## LETTERS TO THE EDITOR

The Editor does not necessarily endorse the opinions expressed by his correspondents

## Burghard Committee and Common Standards for Components

WITH great enthusiasm many are engaged in preparing or awaiting the publication of the new British Standards for electronic parts in the B.S 9000 series-the common standards recommended by the Burghard Committee. Common standards they may be in some respects but they will be lacking in one important detail-a common system of identification or part numbering.
Very soon now tens of part makers will be busy allocating their own identification numbers and sales codes to all the many styles, values, tolerances, wattages, etc., covered by the new specifications: and early next year a hundred companies intending to use these parts will be busy preparing their schedules of part numbers for use by their drawing offices for purchasing or stock control purposes. And then later, each parts manufacturer will need to prepare a cross reference list showing the equivalence between his many customers' part numbers and his own.
The Services, too, will be allocating their N.A.T.O. stock numbers to the items they intend to purchase.

What a waste of national effort! What an opportunity missed-to have a British Standard part number that all could use.

Time is slipping by and it is now too late to grasp this nettle in the first specifications to be published: but there are more to come. Can nothing be done?
E. P. Stanton
(Quality Control Manager)
Plessey Components Group, Swindon

## "Honour to whom Honour"

E. AISBERG Director of our Paris contemporary Electronique Industrielle, has written commenting on the origin of the term "class $D$ " given by K. C. Johnson in his article last month. He writes:-

L'auteur attribue l'invention de ce montage et l'appellation "classe D" à P. J. Baxandall. Celui-ci a en effet consacré aux amplificateurs classe $D$ un article dans Proceedings I.E.E. on 1959.

Cependant, l'amplificateur classe D a été inventé par l'ingénieur francais Roger Charbonnier, à l'époque directeur de "Rochar Electronique". La brevet correspondant a été déposé au nom de cette maison le 6 janvier 1954. Et la première description a paru sous la signature de J. P. OEhmichen dans le numéro 1 (marsavril 1955) de notre Revue Electronique Industrielle.

J'ai tenu à préciser ce petit point d'histoire afin de rendre à César ce qui lui est dû.

## Buy British

I SUPPOSE that most of your readers will agrec with your editorial in the December number of your journal, but I must say that I think that the industry must bear part of the blame for the situation. I will not use your space to recount, in detail, my attempts to get data or
products out of British firms. It may be some consolation to Mr. Thompson that I would not rate the chances of a small buyer of getting an answer as better than one in five. On the other hand, my only letter to an American firm was answered by return, and the goods were despatched on receipt of my firm order and cheque.

I would be only to happy to buy British i.cs if I knew that they were available. So far, I can recall seeing only American i.cs offered on the retail market, and I have had some of them. I expect to buy more i.cs, but, on their past performance, I am reluctant to spend time and money on fruitless enquiries to British firms. It therefore seems that my choice is between buying foreign, and buying nothing. While I should prefer to "Buy British," I have no intention of going without these fascinating devices. Will any British firm, perhaps by the appropriate advertisement, giving price and channels of availability, in your journal, prove me wrong?
J. B. G. Parker (G3SOL)

London, E.6.
I DOUBT very much if "any" young engineer, as you suggest in your December Editorial, would be allowed to buy American at will, if only because of import duty. Certainly this is not so in my establishment. We buy American usually when the item is not made here, or when the American article is obviously superior-one might add, there is often little difference in the price, and delivery has so far been good.

You may be interested to know that a British instrument advertised in $W . W$. at the end of 1966, and ordered by me near the beginning of this year, had still not been delivered at the end of November when I cancelled it as it was not yet in production!

British makers seem to think they get a raw dealperhaps some of them do. Undoubledly, however, there are a number who get what they deserve. And if certain foreign firms can do incomparably better, as they can in some fields, we have no right to play the hurt, misunderstood British routine. If British makers can produce, the profession will gladly buy.
"Engineer"

## Bailey Amplifier Mod.

I HAVE received one or two queries regarding the cut-off frequency of the treble filter in the pre-amplifier circuit I described in the December 1966 issue. I have looked into this and have discovered that the capacitors used were about $50 \%$ greater in capacitance than their marked value. In order to obtain the correct performance this means that all the capacitors should be uprated by $50 \%$ in the treble filter. The new values will therefore be 0.015 and $0.0075 \mu \mathrm{~F}$ or as near as possible. The large tolerance on capacitors had been overlooked in this instance so it is important that capacitors of at least $10 \%$ tolerance should be used. If a slightly lower cut-off frequency is desired there is no reason why the values cannot be increased to 0.02 and $0.01 \mu \mathrm{~F}$, there being more convenient values to obtain.

Arthur R. Bailey

# Sub-surface Propagation 

Some points from an I.E.E./I.E.R.E. conference on m.f., I.f. and v.l.f. radio

IT has been known from the early days of radio that in round-the-world transmission the energy is confined between the earth and the ionosphere, thus overcoming the diffraction losses round the curve of the earth. On v.l.f. the height of the lower boundary of the ionosphere is no longer large compared with the wavelength and the ray method of studying the propagation characteristics, so useful at h.f., is only practicable for use at short distances. For long distances it is necessary to treat the region between the earth and the ionosphere as a waveguide and to study the propagation in terms of mode theory.

In a survey paper at the recent I.E.E./I.E.R.E. conference on propagation J. R. Wait, himself a leading expert in this field, referred to the fundamental researches of K. G. Budden giving the full wave treatment of the modes, including the effects of the curvature and of the magnetic field of the earth. He treated the problem in a severely mathematical way that many engineers must find difficult to appreciate, but the basic results emerging from this study are proving most valuable as a means of interpreting v.l.f. field-strength measurements in terms of possible electron distributions in the $\mathbf{D}$ region of the ionosphere.

## geological waveguide

A further interesting development of the waveguide concept is the proposed application to long-distance propagation in sub-surface geological strata. It is suggested that at depths of several miles there may exist extensive strata of very low conductivity between regions of much greater conductivity, constituting a waveguide with very low attenuation. While much has been written on the theoretical side, based on highly idealised models, and communication has been established over several miles, the technical and economic problems are immense and considerable doubt has been expressed about finding strata of sufficiently low conductivity of the required extent in the desired places.

There is evidence that such communication between subterranean points may sometimes be achicved by the "up-over-and-down" mode whereby energy from the transmitter travels up to the surface, escaping into the air and travelling, possibly with the help of the ionosphere, along the surface of the earth, some of it then being refracted into the earth to the receiving point below.

The attenuation of radio waves through sea water is very great, but it decreases with decreasing frequency and the use of v.l.f. for submarine communication is being actively pursued. The rigid mathematical theory is exceedingly difficult, but simple physical principles show that contact between a base above ground and submarines anywhere on the earth can be achieved by using v.l.f. The wave travelling over the earth is vertically polarised and is refracted vertically downwards and is receivable on a suitably oriented horizontal dipole on a submarine that is sufficiently near to the surface.

It follows similarly that communication between submarines, too far apart for direct propagation through the
water, must be by an "up-over-and-down" mode with the implied limitation in depth below the surface, and that using electric dipoles they should be horizontal and end-on to one another. Very little practical information is available, but the theoretical analysis makes reference to magnetic dipoles even though the available size of a loop regarded as a single turn would be very inefficient compared with an electric dipole at these frequencies.
For communication purposes the use of v.l.f. is inevitably restricted by the limited bandwidth available, but the advent of extremely accurate reference clocks and frequency-stabilized v.l.f. transmissions has prompted their use for time signals and navigational aids with a world-wide coverage. The latter application depends for its success upon the high stability of the D region of the ionosphere as a reflector of v.l.f. waves, the height of reflection by day being nearly constant at about 70 km and changing at sunset in a well-predictable way to about 85 km at night and back again at sunrise.

This stability in relation to a phase-comparison navigational aid is much greater than for the corresponding use of the $E$ and $F$ layers at higher frequencies, but much work is still needed to take account of sudden phase anomalies due to ionospheric disturbances, especially in the polar regions. A suggestion has been made for the automatic suppression of errors due to diurnal and seasonal variations in the ionosphere by working at two frequencies symmetrically displaced about 12 kHz .

In his opening address at the conference on m.f., l.f. and v.l.f. propagation, J. A. Ratcliffe deplored, as a scientist, the very limited use that had been made of v.l.f. transmissions for the study of the lowest regions of the ionosphere during the period when ionospheric sounding at high frequencies had been developed for the study of the $E$ and $F$ layers and the prediction of the propagation characteristics of high-frequency communication. In this he was perhaps over-modest in view of the work of the team that he directed for so many years at the Cavendish Laboratory using the transmissions from Rugby GBR.

## SCIENTIFIC RESEARCH

It was notable that the recent conference was mainly concerned with the use of v.l.f., not as a means of communication but as a tool for scientific fesearch. The advent of rockets and satellites has given an immense impetus with the possibility of receiving signals in the ionosphere from terrestrial v.l.f. transmitters and of transmiting v.l.f. signals to earth. The study of the wave forms of atmospherics from lightning flashes is greatly advancing our knowledge of the earth-ionosphere waveguide and of resonance effects at e.l.f. The associated phenomenon of whistlers with their large frequency dispersion in the audio band has been explained in terms of the magneto-ionic theory of propagation in the ionosphere, but the observations made in the ionosphere have revealed that the v.l.f. ionograms are as complicated as those being obtained by sounding at h.f. from the original satellite Alouette I which is now in its sixth year of operation.

## 'ELECTRONIC CAM'-THE BEST OF BOTH WORLDS?

FILM offers teicvision companies a medium by w'Mich programme material can be interchanged between countries without regard to line standards or the colour system in usc. Producing a film using motion picture methods is an expensive and time-consuming process and it has long been considered desirable it devise a system for exposing film using television multi-camera techniques. The film camera is not "interested" in whether black and white or colour stock is being used and does not suffer from the degradation in picture quality associated with telecine machines. The basic idea of marrying a television camera to a film camera to enable the scene to be monitored remotely is not a new one, this latest system" Electronic Cam "was devised by Arnold and Richter of Munich and has been developed by engineers from Rediffusion Television Lid. over the past two years

Basically the system consists of the marriage of an Arriflex 35 mm camera and a plumbicon camera tube; light from the scene is reflected by a mirrored segment on the shutter to the plumbicon during the film pull-down period. The output of the plumbicon drives a smal! television monitor that acts as the camera view-finder and also drives other monitors throughout the studio. In the complete installation three such cameras are employed, the film motor of each
canera being cortrolled remotely by means of switches on a central production control console. Four monitors are employed on this console, one for each camera and, in addition, one for the camera that has been selected. Switching, or cutsing, between cameras can be carried out in about one-third of a second, this being the time taken for the camera mechanism to reach operating speed or, if desired, all cameras can be left running, only one being used for the "take," allowing instantancous "cutting" between cameras but wasting large amounts of film stock. Rehearsals can be carried out without film in the cameras and in this case tootage counters on the control console make it possible to predict the amount of film required in each camera for the actual take, eliminating wastage. Identification and synchronizing marks are recorded 011 both the film and the magnetic tape used for the sound track indicating which cancra it came from and facilitatin ${ }^{-}$the assembly of the film and sound track. In a pilot production film, taken to assess the performance of the system, a fifreen minute film was made in approximately one hour on the studio floor. The film was divided into three sections, each being filmed as a continuous take, the director cutting between cameras as required, achieving a film utilization ratio of 1.52:1

## CODE OF PRACTICE FOR AERIAL INSTALLATION

WITH the advent of colour television the question of aerial installation has become of greater importance and it is fe't strongly botl by the Radio and Television Retailers Association (R.T.R.A.) and the Radio and Electronic Component Manufacturers Federation (R.C.E.M.F.) that high standards will have to be adhered to. To this end a code of practice for aerial installation has been agreed by the two bodies and in future ali members of the R.T.R.A. will be ex-
pected to conform to these standards. Any serious departure from them may result in disciplinary action being taken by the Association. It is also suggested that any member that does not erect his own aerials should forward a copy of the code to the company concerned in order that an undertaking may be given that installations will be made in accordance with the code. Copies of the code are available free from the R.T.R.A., 19-21 Conway St., London, W.1.

## NUCLEONIC INSTRUMENT FIRM EXPANDS

THE largest company in the nucleonics field in Europe will be formed as a result of a major rationalization of the nuclear instrument industry in the U.K. Nuclear Enterprises of Edinburgh, founded as recently as 1956, has taken over the nucleonics interests of E.M.I at both Hayes, $M$ iddlesex and Wells, Somerset, as the first stage in a triple acquisition. Nuclear Enterprises is also, subject to necessary consents, acquiring Isotope Developments Lid. and the

Baldwin Instrument Company, both members of the Elliott-Automation Group situated near Aldermaston. The Nuclear Enterprises range of radiation detecters and instruments will be supplemented with medical, physics, and data handling equipment from E.M.I., low cost laboratory and medical instrumentation from I.D.L. and indusirial nucleonics instrumentation for gauging, analysis and process control from Baldwin Instruments.

Cable and Wireless Ltd. have placed contracts with Submarine Cables Ltd (an A.E.I. company) for a deep sea submarine telephone cable that will provide a maximum of 640 telephone circuits between Canada and Bermuda. The project, which is known as CANBER, requires 800 nautical miles of cable, 81 submersible repeaters and five submersible equalizers worth a total of $£ 3.5 \mathrm{M}$. The cable will be jointly owned by the Canadian Overseas Telecommunications Corporation and Cable \& Wireless Lid. Some of the new materials needed for fabricating the cable will come from Canadian sources. CANBER is due to be laid in 1969 by the 8,960 -ton cable ship Mercury from the Cable and Wireless fleet. CANBER will land in Canada in the vicinity of Mill Village, Nova Scotia, permitting connections with the Canadian satellite earth stations

The information services of the Gcvernment of Hong Kong have announced that steps are being taken to prevent manufacturers wrongly describing radio receivers by incorporating into them non-functioning transistors. In talks with the manufacturers the Colony's Commerce and Industry Department found that the manufacturers were opposed to this practice and that the dummy transistors had been included at the request of overseas buyers! As from January 1st the Commerce and Industry Department will institute checks to determine whether any local transistor recciver factory is incorporating nonfunctioning transistors and legal action will be considered against any that are continuing with the practice.

Orders for four harbour radar systems worth a total of $£ 128,000$ to be installed at Montreal, Brisbane, Rostock (Easi Germany), and Wallasey (Cheshire), have been received by Decca Radar Litd. The installation for the port of Montreal is to be completed in two phases. The first of these consist of installing a two-channel radar and two 16 -inch displays that will provide a traffic control service. In the second phase the radar coverage will be increased by a remote scanner, controlled by a u.h.f. link relaying its information back to the control room via a microwave system. The other three systems will not have the remote scanner and differ from the Montreal system only in acrial and display sizes.

The G.P.O. has placed an order with Standard Telephones and Cables Lid., for a 6 GHz microwave system to link the Post Office tower in London with Norwich. The equipment, type RL6D, will provide six broad-band radio channels berween London and Stoke Holy Cross (Norwich), with repeater stations at Kelvedon Hatch and Sibleys in Essex and at Wickhambrook and Mendlesham in Suffolk. The RL6D provides a $10-\mathrm{W}$ power output; the aerials used will be of the cassegrain type for single or bipolar operation.

## 1968 CONFERENCES AND EXHIBITIONS

## LONDON

Mar. 11-14 Physics Exhibitior
Apr. $8 \& 9$
Thick Film Technology
Apr. 18-21
Audio Festival \& Fair
Apr. 22-24 Systems
May 13-18
Instruments, Electronics and Automation Exhibition
May 20-31
Royal Lancaster Horel Communication-Satellite Earth Stations
July 29-Aug. 2
Olympia
Ships' Gear International Exhibition
Sept. $9-12$
Elementary Particles
Scpt. 9-13 I.E.E., Savoy PI. International Television Conference
Scpt. 20-Oct. 2
Tropospheric
Oct. 2-5 R.H.S. New Hall
R.S.G.B. Radio Communications Exhibition

BELFAST
Apr. 1-3
Queen's University
Heavy Particle Collisions
BRIGHTON
Oct. 8-10
Hotel Metropole
National Electronics Packaging Conference \& Exhibition
BRISTOL
Jan. 2-4
The University
Integrated Circuits Symposium and Exhibition
Mar. 27-29 The University
Thermodynamics and Fluid Mechanics
CARDIFF
Apr. 18 \& 19 Cathays Park Audio-Visual Aids Conference and Exhibition

CRANFIELD
Mar. 25-28
College of Acronautics Aerospace Instrumentation Symposium

DURHAM
Apr. 2 \& 3
The University
Semimetals and Narrow Gap Semiconductors
EDINBURGH
Aug. 5-10
I.F.I.P. Data Processing Congress \& Exhibition

FARNBOROUGH
Sept. 16-22
R.A.E.

Electronics and Air Show
GLASGOW
Mar. 8-16
Kelvin Hall NORBEX-North British Engineering Exhibition

## HARROW

Mar. 12-14
Public Address Show
King's Head Hotel

HARWELL
May 9 \& 10
A.E.R.E.

Low Energy Electron Diffraction
LOUGHBOROUGH
Apr. 16-19 University of Technology
Modular Education for Industry
MANCHESTER
Jan. 3-6
Inst, of Science and Technology Solid State Physics Conference

| Sept. 24-28 |
| :---: |
| Electronics, |
| Exhibition |

Instruments, Control and | Bellc Vuc |
| ---: |
| Components |

Electronic lnstruments Exhibition

## NOTTINGHAM

Sepi. 11-13
The University
Physical Aspects of Noise in Electronic Devices
PAISLEY
Apr. 17-19
College of Technology
Automation Techniques in Industry
SOUTHAMPTON
Jan. 9 \& 10 The University Materials for Acoustic \& Vibration Damping
Apr. 22-24 The University
Nucleation, Growth and Structure of Thin Films
SWANSEA
July 15-18 University College
Electrical Contact Phenomena

## TEDDINGTON



Wireless World, January 1968

## BOOKS RECEIVED

Microwave Valves by C. H. Dix and W. H. Aldous, presents an account of the basic physical processes and the operation of micrewave valves. Although only the essential mathematics is included the book is intended for readers with H.N.C. or a degrec. The approach to the subject begins with a description of the the motion of electrons and the properties of the various types of r.f. circuits and transmission lines that are used in the devices. Although microwave triodes are discussed, the emphasis is on beam devices both linear and crossed field, and in describing these the spacecharge wave approach is used. Further chapters cover the formation and fecusing of electron beams, the noise properties of microwave devices, construction and applications. Pp. 269. Price 55s. Iliffe Books Lid., Dorset House, Stamford Street, London, S.E.1.

The Practical Aerial Hardbook, by Gordon J. King. The introductory chapters provide a grounding in the principles of radio propagation and acrial design. Different types of aerial are discussed and guidance is given on choosing the best acrial system for a particular need together with practical installation information. The remainder of the book covers methods of combining signals received by separate aerial systerns; methods of supplying several receivers from one acrial; improving reception using acrial booster amplifiers; shared aerial systems as used in blocks of flats, etc.; and combating interference. Appendices give information on aerials for colour television, aerials for stereo radio; the distant reception of v.h.f. and u.h.f. signals. Pp. 224. Price 35s. Odhams Books Ltd., 40 Long Acre, London, W.C.2.

Introduction to Vector Analysis by W. D. Day. Suitable for self-tuition, because of the numerous worked examples and graded exercises, this book presents the theory of vector analysis in a form suitable for the electronics engineer. Starting from basic definitions and notation, the concept of scalar and vector products of two vectors, triple products, diferentiation, line and surface integral is established. The differential equations of electron motion in uniform magnetic and electric fields at right angles are considered in some detail. The scalar point, scalar potential, divergence, curl, cartesian, cylindrical and spherical co-ordinates are all examined. The remaining chapters are devoted to more general vector fields in particular to the time varving electromagnetic field governed by Maxwell's equations. Pp. 260. Price 42s. Iliffe Books Ltd., Dorset House, Stamford Strcet, London, S.E. 1 .

Techniques of Pulse-Code Madulation in Communication Networks by G. C. Hartley, P. Mornet, F. Ralph and D. J. Tarran. This book, from the I.E.E. Monegraph series, is published at an opportune moment with the recent opening of London's first p.c.m. link. After the introduction and an historical review, the principles of p.c.m. are outlined and such topics as time sampling, signal reconstruction, quantization, companding, eic., are discussed in some detail. The remainder of the work is devoted to the application of p.c.m. communication principles, basic system clements and factors affecting system design, a glossary of terms is also included. Pp. 110. Price 30s. Cambridge University Press, Bentley House, 200 Euston Road, London, N.W.1.

## H. F. PREDICTIONS - JANUARY



The maximum usable frequency is, by definition, that at which communication should be possible for over $50 \%$ of the time. Satisfactory communication will, of course, be possible above the MUF but for decreasing percentages of time as frequency is raised.

Operation below the MUF will give improved probability of communication ( $90 \%$ at FOT) but worsening signal/ noise ratio. LUF is the frequency at which a specified signal/noise ratio will be obtained for a given percentage of time. The curves were drawn by Cable \& Wireless Lid. for reception of point-to-point telegraph circuits in the U.K. When evaluating, an assumption is made that communication probability will be $100 \%$ at the LUF, this is not always valid as shown by the Hong Kong curve.

## LETTER FROM AMERICA

SOME controversy has been caused among electronics engineers over here during the past few months by S.I.As. What are S.I.As.? Well, they are basically very small stub antennas which have built-in transistors to give extra gain. They are usually $1 / 25$ th wavelength and the inventor, Edwin Turner, of the Air Force Avionics Laboratory in Dayton, Ohio, claims that S.I.As could be built into TV sets and "would out perform aerials many times the size." However, there are doubters. Harry Greenberg, chief electronic engineer of Channelmaster Corporation, says categorically, "In our opinion, they would not perform as well as ordinary rabbit ears aerials, let alone replace ro $=$-top aerials." It is well known that the smaller the aerial length, the lower the signal strength received. Hence, the signal-to-noise ratio tends to get worse as the pick-up aerial gets shorter. However, this is offset to some extent by the fact that atmospheric noise is very high at high frequencies and so the signal-to-noise ratio is less dependent on the aerial or receiver. So a smaller aerial will not necessarily mean a small signal-tonoise ratio although the signal itself will be less. In Time magazine Turner states, "We have in effect substituted a short aerial carrying a large current for a long aerial carrying a small current." He went on to say, "A S.I.A. at $1 / 16$ th wavelength instead of $\frac{1}{4}$ is about equal in signal-tonoise ratio to a dipole aerial or tunable rabbit ears, even at $1 / 25$ th the noise characteristic is comparable with a dipole when mismatches in the dipole were considered." Turner claims that S.I.As provide a wide impedance match and quotes ratios of up to 50 to 1 . In one version the transistor d.c. current is controlled to move the optimum bandwidth matching range. Considerable work in this field has been carried out by Hans Meinke at the Munich Technical University and a circuit was published in Electronics last July. But so far from silencing the critics, it provoked more oppositien. One, Wilfred Carson, said, "It was obvious that one stage was about to break into oscillation and so the stage gain would be abnormally high." At the Canadian International Electronics Conference Dr. Flachenecker said "from v.l.f. up to 30 MHz S.I.As show field strength sensitivities nearly equal to the external noise-field
strength if the aerial height is around 1 metre." So the debate continues.

The first commercial colour television receiver was introduced back in 1954 by R.C.A. This was a $15-\mathrm{in}$ model costing about $\$ 1,000$. Some 10,000 sets were sold that year, but by January 1967, the total number of colour sets in the U.S.A. had soared to $9,750,000$. Early in 1967 experts forecast total sales of colour sets at six million and at the end of October the sales had reached $4,086,521$ for colour, compared with $4,394,857$ for monochrome. At a recent E.I.A.* merchandising seminar, a speaker caused a stir of interest when he said his firm was already campaigning for that second colour TV in every home. Support for this expression of "Gracious Living" came from R.C.A., who are now pioneering a low cost 14 -in portable. Says the Sales President, "In pioneering the new 14 -in diagonal screen size, we are counting heavily on a second set market for colour that will appear much sooner than it did in black-and-white TV. The colour set viewer who is spoiled by colour in the living room won't accept a monochrome substitute in the den or bedroom." (I could add from experience -neither will the children!) How about prices? Well, they range from about $\$ 199$ for a portable to around $\$ 500$ for a console with a 23 -in screen. Tube sizes are a little confusing as some manufacturers use diagonal measurements while others stick to the tube face area. Philco have just released a large screen portable ( 267 sq in) at $\$ 299.95$. They ask, "Why should the least expensive large screen colour set cost a working man a month's pay?" Why indeed! This price is certainly very reasonable but it is possible that some of the Japanese imporss will be cheaper still. The modern colour sets are very easy to operate but ingenious devices are fitted to some models to ensure accurate tuning. For instance, Motorola uses an automatically switched indicator lamp to show "on the nose" tuning and Westinghouse sets have a tuning bar. When this bar is depressed vertical black bars appear on the screen and the trick is to turn the fine tuning control until only one bar is seen. The circuit is quite complicated and it employs two multivibra-

[^3]tors with a gating valve and variable relay. Incidentally, some Motorola models use transistors throughout but most other designers prefer hybrid circuits. At the moment, integrated circuits are widely used for audio stages or i.f. sections, etc. They are also employed in f.m. tuners, receivers and amplifiers-R.C.A. even has an i.c. pre-amplifier built in a pickup cartridge. Westinghouse has just released a single i.c. audio amplifier which can replace nearly all the components in low power record players or tape recorders. The input is high impedance so a ceramic pickup can be used and the output is rated at 1 W for $5 \%$ distortion.

Amperex have an interesting i.c. called a "Bifet" which consists of a mosfet coupled to a transistor emit-ter-follower plus biasing resistors all housed in a normal three-lead TO-18 can. Input impedance is very high, being of the order of $10^{12}$. Noise is exceptionally low-the total voltage measuring less than 25 microvolts! This is comparable with the best valve amplifiers and so the "Bifet" will be particularly useful for lowlevel microphone pre-amplifiers, photo-cell head amplifiers, etc.

NEW TRANSISTOR devices are appearing almost every day but one of the, most interesting is called a "Pitran" transducer. This is a silicon planar transistor that has the emitter-base junction mechanically coupled to a tiny diaphragm located at the top of the can. When a pressure is applied to the diaphragm a large reversible change is produced in the transistor characteristics. Sensitivity is quoted as 4 V per gramme point force and linearity is said to be better than $1 \%$.

The vietnam war has given a tremendous imperus to electronic research and development-particularly in the communications field. Probably one of the most bizarre inventions concerns enemy-or rather people-detection. It consists of a pump that pipes in air to a colony of bedbugs. The presence of human beings causes the bugs to become agitated so modulating a r.f. field to give audible or visual indication. Sensitivity is said to be very high but it is not stated whether the bugs discrimate between Viet Cong and Americans!
G. W. Tillett.

## WORLD OF AMATEUR RADIO

## World-wide Nelwork of

 Amateur Radio Beacons?PROPOSALS for the establishment of a world-wide network of amateur radio beacon transmitters to operate on frequencies in the amateur 21 - and $28-\mathrm{MHz}$ bands have been put forward by a scientific ionospheric observation group within the German national amateur radio society. The group, which has 100 regular observers, is continuing work done in Germany during the International Geophysical Year (I.G.Y.) and in the subsequent International Quiet Sun Years (I.Q.S.Y.), and its proposals visualize the setting up of one beacon in each of the five continents to operate in the $21-\mathrm{MHz}$ band and at least two beacons in each continent to operate in the $28-\mathrm{MHz}$ band. Each beacon will use a main and a secondary frequency, the main frequency being common to all beacons in a particular band. Secondary frequencies will be spaced in an arrangement of channels of $2.5-\mathrm{kHz}$ wide below the main frequency. The secondary frequency assigned to a particular station will be transmitted when the main frequency is not being used.
Transmissions on the main frequency will be arranged in accordance with a time-shared world-wide schedule, which will enable radio amateurs and scientific institutes to monitor automatically, or by means of pen-recorders, etc., the actual world-wide propagation conditions on the band concerned. Transmissions on the secondary frequencies will supplement observations on the main frequency and will allow permanent checks on conditions for a certain general path direction and provide a means to monitor the effect of sudden solar events, and of band openings, which cannot be covered by the main frequency transmissions because of time sharing.
It is also hoped to provide a similar world service on a frequency in the $50-\mathrm{MHz}$ (six-metre) band but unfortunately this band is not generally available to amateurs in Europe and Asia. Special facilities, however, are visualized for this scientific project. The proposals for a world-wide network of amateur radio beacon transmitters are to be submitted to the International Amateur Radio Union for consideration by the 75 national societies that form the Union.

European Fox Hunting Championships.-Teams from the Soviet Union, Yugoslavia and Hungary were respectively placed 1st, 2nd and 3rd in the 80 -metre section of the European Fox Hunting Championships held recently in Czechoslovakia. The individual winner (a Russian) located the four hidden transmitters ("foxes ") in 49 mins 6 secs and the time of the winning team was 118 mins 26 secs. The 2 -metre section was won by a team from Hungary with teams from Bulgaria and the Soviet Union in the 2nd and 3rd places. The time of the winning team was 89 mins 53 secs (locating six "foxes") and the individual winner (another Russian) located three hidden transmitters in 37 mins 30 secs.
R.N.A.R.S. Code Proficiency Transmissions.-Morse code proficiency transmissions arranged by the Royal Naval Amateur Radio Society, now take place on the first Tuesday of each month at speeds of $15,20,25,30$ and 35 werds per minute. Transmissions commence at 20.00 G.M.T. on $3,520 \mathrm{kHz}$ and perfect ( $100 \%$ ) copy at a particular speed is required to qualify for the appropriate Code Certificate. Completed $\operatorname{logs}$, together with five 3 d stamps, should be sent to R.N.A.R.S., 27, Oxted Rise, Oadby, Leicester.

QSL Cards for R.A.F.A.R.S. Members.-Cards depicting six Royal Air Force aircraft spanning 25 years of R.A.F. history are now available to the 450 members of the Royal Air Force Amateur Radio Society, to confirm contacts.

Slow-Scan Television.-The United States Federal Communications Commission has recently proposed that slowscan television shall be authorized in certain parts of the amateur high-frequency bands, namely, 3.8-3.9, 7.2-7.25, 14.2-14.275 and $21.25-21.35 \mathrm{MHz}$ as well as in the telephony bands at 10, 6 and 2 metres. The bandwidth will be that of a normal single sideband signal, nominally 3 kHz . It is not yet known whether similar proposals have been put forward by any licensing authority in Europe or Asia. Slow-scan television (although permitted in the United Kingdom for those holding an amateur television licence) has not, so far, attracted a great deal of interest.

Nigerian Award.-The 5N2 Award is available to any radio amateur or short-wave listener who can produce evidence of having worked or heard five Nigerian amateur stations (5N2 calls) on two or more amateur bands. (For example, four stations can be worked or heard on 21 MHz and one on 28 MHz .) The Award will be issued in three classes: telephony (including single sideband), telegraphy and mixed. Applications, together with a certified copy of the log (or QSL cards in the case of short-wave listeners) and accompanied by five international reply coupons should be sent to the Awards Manager, N.A.R.S., P.O. Box 2873, Lagos, Nigeria.

Amateur Radio in India.-New rules for amateur radio licences, drafted by the Indian Department of Communications, came into force on September 1st, 1967. They permit the issue of licences to young people aged 14 years and upwards. For some reason, which the Amateur Radio Society of India has not been able to discover, no new licences have been issued in India since the beginning of 1967 when the membership of the society was around 350 .

Simulated Emergency Test.-In late January during a simulated emergency test, organized by the American Radio Relay League, the opportunity will present itself for all United States radio amateurs to take part in a nation-wide demonstration of amateur radio public-service facilities. The emergency test will take place during the weekend January 27 th- 28 th, and will include all Amateur Radio Public Service Corps members in local and national exercises for the Red Cross, Civil Defence and similar organizations.

Championship of France. - The annual contests organized by the French national amateur society (R.E.F.), to decide the champions of France for 1968, will be held on January 27th/28th (telegraphy) and February 24th/25th (telephony). Both contests will commence at 14.00 on the Saturday and finish at 21.00 on the Sunday.

Monaco Amateurs to join I.A.R.U.-The Association des Radios Amateurs de Monaco is seeking membership in the International Amateur Radio Union. Formed in 1953, the Association now has 18 licensed transmitting membersthe total number of licensed stations in the Principality. Licences are issued to visitors to Monaco who submit proof of being bona fide licensed amateurs in their own country.
V.H.F. Licences in Germany.-Call signs in a new series, DC6 followed by two letters, are now being issued to German amateurs who wish to operate on frequencies above 144 MHz . Holders of these calls have passed a technical examination but not a Morse code test. In the United Kingdom call signs in the series $\mathbf{G 8}$ followed by three letters are issued to amateurs who wish to operate on frequencies above 420 MHz . john Clarricoats, G6CL.

## JANUARY MEETINGS

Tickets are required for some meerings: readers are advised, therefore, to communicate with the society concerned

## LONDON

2nd. I.E.E.-Colloquium on "The economical collection of meteorological data" at 2.30 at Savoy Pl., W.C.2.
4th. I.E.E.-Hunter Memorial Lecture "Changing patterns in communications" by J. H. H. Merriman at 5.30 at Savoy Pl., W.C. 2.

Sth. I.E.E.-"The practical use of radar and d.f. techniques in locating earth satellites" by Dr. H. G. Hopkins and W. A. S. Murray at 5.30 at Savoy Pl., W.C.2.

8th. I.E.E.-Discussion on "Logarithmically periodic acrials" at 5.30 at Savoy Pl., W.C. 2.

8th. I.E.E.-Discussion on "Domain originated functional integrated circuits (DOFICS)" at 5.30 at Savoy Pl., W.C.2.
9th. I.E.E.-Discussion on "Electrical signals for data acquisition and transmis-sion-what form should they take?" at 5.30 at Savoy Pl., W.C. 2.

10th. I.E.R.E.-"Some aspects of electrostatic loudspeakers" by Prof. J. Merhaut at 6.0 at $8-9$ Bedford Sq., W.C. 1

10th. S.E.R.T.-"Digital voltmeters" by J. R. Pearce at 7.0 at London School of Hygiene \& Tropical Medicine, Keppel St., w.C.1.

11th. Inst. of Electronics-" Modern semiconductor devices" by D. F. Dunster at 6.45 at the School of Hygiene and Tropical Medicine, Keppel St., W.C.1.

16th. I.E.E. \& I.P.P.S.-Colloquium on "MOST devices" at 2.30 at Savoy Pl. W.C. 2.

17th. Inst. Navigation-" Sub-surface navigation" by Dr. W. P. Williams at 2.15 at the Royal Inst. of Naval Architects, 10 Upper Belgrave St., S.W.I.

17th. I.E.E.-"Colour television receiver design" by B. J. Rogers at 5.30 at Savoy P1., W.C. 2

18th. R.T.S.-"London schools television service" by W. Kemp and P. W. Lines at 7.0 at the I.T.A., 70 Brompton Rd., S.W. 3.

22nd. I.E.E.-Colloquium on "Microwave integrated circuits" and "Microwave solid state sources" at $10 \mathrm{a} . \mathrm{m}$. at Savoy P1., W.C. 2

22nd. I.E.E.-Discussion on "Alicrowave electrostatic wattmeter" at 5.30 at Sswoy P1. W.C. 2 .
2tih. I.E.R.E.-"Studio colour equipment" by G. Parker at 6.0 at 8-9 Bedford Sq., W.C. 1 .

26th. R.T.S.-"Television aids to film production" at 7.0 at the 1.T.A., 70 Brompton Rd., S.W. 3.
29th. I.E.E. \& I.E.R.E.-Discussion on "Diathermy" at 5.30 at Savoy Pl., W.C.2. 31 st. R.S.G.B. -"The development of a u.h.f. television service" by R. C. Hills at 6.30 at the I.E.E., Savoy PI., W.C.2.

## ABERDEEN

10th. I.E.E.-."The engineer and the law " by H. B. Morton at 7.30 at Robert Gordon's Institute of Technology.

## BIRMINGHAM

24th. R.T.S.-"The transmission of colour television signals over Post Office links" by E. Howorth at 7.0 at the Medical Institure, Harborne Rd., Edgbaston.

29th. I.E.E. \& I.P.O.E.E.-" Design considerations in microwave links" by $G$. Wanless and E. Jamieson at 6.0 at M.E.B., Summer Lane.

## BOURNEMOUTH

31st. I.E.R.E.-" Some circuit aspects of M.O.S. transistors" by N. E. Broadberry and L. N. M. Edward at 7.30 at the College of Technology.

## BRISTOL

16th. S. Inst. Tech.-" Instrumentation in medicine" by D. H. Follett at 7.30 at the Dept of Physics, the University.
18th. I.E.R.E.I.E.E. \& R.Ae.S.- Concord " by H. Hill at 7.0 ar the University. 25th. I.E.E.-" The best method of educating engineers-full time or sandwich?" by D. M. Dummer and P. L. Arlett at 6.0 at the Technical College.

## CARDIFF

10th. R.T.S.-"The philosophy of colour camera design" by C. B. B. Wood at 7.30 at the Angel Hotel.

24th. I.E.R.E.-"The development of satellite communications" by J. K. S. Jowett as 6.30 at the $\mathbf{X}$ 'elsh College of Advanced Technology.

## DUNDEE

11th. I.E.E.-"The engineer and the law " by H. B. Morton at 7.0 at Robert Gordon's Institure of Technology.

## EDINBURGH

10th. I.E.R.E.-" Microwave and optical communication systems of high traffic capacity" by R. W. White at 7.0 at the Dept. of Natural Philosophy, the University.

## GLASGOW

11th. I.E.R.E.-" Microwave and optical communication systems of high traffic capacity " by R. W. White at 6.0 at the University of Strathclyde.

## HAMBLE

17th. I.E.E. \& R.Ac.S.-" Telecommunications in aviation" by W. P. Nicol at 8.0 at the College of Air Training.

## HUDDERSFIELD

30th. I.E.E- "The role of the systems engineer" by Dr. Wilson at the College of Technology.

## HULL

25th. I.E.E.-"The engineer and the law" by H. B. Morton at 6.30 at Y.E.B. Offices.

## ISLE OF WIGHT

26th. I.E.E.-"The work of the Engineering Institutions Training Board " at 6.30 at the Technical College.

## LEEDS

23 rd . I.E.E.-" The future use of solidstate devices in the microwave field" by Dr. J. E. Carroll at 6.30 at the University.

## LIVERPOOL

8th. I.E.E.-" Electromagnetic levitation" by H. R. Bolton at 6.30 at the University.
17th. I.E.R.E.-"Manufacturing aspects of the shadowmask tube" by P. T. Funnell at 7.0 at the Regional College of Technology, Byrom St.

22nd. I.E.E.-"Jodrell Bank radio telescope" by R. Lascelles at 6.30 at the University.

## MALVERN

22nd. I.E.E.-" Electronic telephone exchange" by L. R. F. Harris at 7.30 at the Abbey Hotel.

## MANCHESTER

22nd. I.E.E.-Faraday Lecture "Medical electronics" by Dr. D. W. Hill at 7.30 at the Free Trade Hall.

23rd. I.E.E.-Faraday Lecture "Medical Electronics" by Dr. D. W. Hill at 2.30 (Schools) and 7.30 at the Free Trade Hall.

31 st. I.E.E. Grads.-" Superconductivity" by Dr. A. C. Rose-Innes at 7.0 at U.M.I.S.T.

## MIDDLESBROUGH

11th. I.E.E. \& S. Inst. T'ech.-"System reliability and safety assessment "by G. Hensley at 6.30 at the Cleveland Scientific Inst.

## NEWCASTLE-UPON-TYNE

10th. I.E.R.E.-" Some applications of eiectronics to oceanographic sensors" by A. M. East at 6.0 at the Inst. of Mining and Mechanical Engrs., Westgate Rd.

15th. I.E.E.-"Microelectronics" by Dr. S. S. Forte at 6.30 at Rutherford College of Technology.

## NOTTINGHAM

16th. I.E.E.-Faraday Lecture "Medical electronics" by Dr. D. W. Hill at 7.15 at the Albert Hall.

## OXFORD

10th. 1.E.E.-"The future of the Institution of Electrical Engineers" by Dr. G. F. Gainsborough at 7.0 at S.E.B., 37 George St.

## PLYMOUTH

3rd. R.T.S.-" Graphics in television" by Don Baker at 7.30 at the Studios of Westward Television Ltd.

## PORTSMOUTH

171h. I.E.E.-" The problems of digital s.s.b. systems" by R. T. A. Standford at 6.30 at the College of Technology.

24th. I.E.E.-" Project control-critical path analysis" by E. H. Harrhy at 6.30 at the College of "rechnology.

## READING

11th. I.E.R.E.-" Parametric amplifiers" by Prof. D. P. Howson at 7.30 at the J. J. Thomson Physical Lab., the University.

## RUGBY

3rd. I.E.E.-"Fabrication uses of the electron beam" by H. N. G. King at 6.15 at the College of Engincering Technology.
10th. I.E.R.E. \& I.P.P.S. "Atomic measurement of time" by Dr. L. Essen at 6.30 at the Col. of Advanced Technology.

## SOUTHAMPTON

16th I.E.R.E-"Microwave ultrasonics" by Dr. R. W. B. Stephens at 6.30 in the Lanchester Theatre, the University.

23 rd . I.E.E.-"The introduction of direct digital control" by Dr. V. Latham at 6.30 at the Lanchester Theatre, the University.

## STEVENAGE

15th. I.E.E. "Post Office Tower" by D. G. Jones at 7.0 at the College of Further Education.

## WOLVERHAMPTON

31st. I.E.R.E.-" The use of a computer tc conirol an industrial process" by D. G. Leak at 7.15 at the College of Technology, Wulfruna St.

# NEW <br> PRODUCTS 

## Stereo Tape Deck

A COMPACT stereo tape deck offering off-tape monitoring, sound-on-sound, sound-with-sound, echo and duet effects, is available from Ampex Great Britain Ltd. Smaller than previous models in the Ampex line, the model 753 measures 15 fin wide $\times 13$ in deep $\times 6 \frac{1}{2}$ in high. This deck has three heads-record, playback and crase-permitting precise monitorin $\varepsilon$ and sound-on-sound recording and playback, eliminating the possibility of crosstalk often found on models with a single record/playback head. As with all Ampex open reel audio recorders, the magnetic heads are of the deep-gap design. Sound mixing features of Model 753 make it possible to mix narration with music tracks, add sound and musical cffects to home-produced programmes, and to conduct language pronunciation studies. The sound-onsound facility permits that while listening to recorded material, new material may be recorded on the same sound track without erasing the original material. With sound-with-sound, recorded material on one track can be played and new material recorded on the second track permitting playback in stereo. Off-tape monitoring permits material to be played back as it is recorded, allowing instant adjustment for best recording fidelity. By switching a control knob to an "echo" position material may be recorded with echo effect. With "duet effect," material being recorded on one track may also be recorded on the second track, but with slight delay, achieving a special

depth of stereo sound on playback. This is said to be especially useful in recording and playing back musical soloists. Model 753 has pre-amplifiers and offers a unique line-jack that permits the recorder to be connected to any type amplifier with consistent performance. It also features twin vu-meters, all solidstate electronics, automatic shut-off, and dual capstan drive, which reduces flutter and wow to a minimum. Overall record/ reproduce frequency response measured at the pre-amplifier output is $\pm 3 \mathrm{~dB}$ at $40 \mathrm{~Hz}-15 \mathrm{kHz}$ at $7 \frac{1}{2}$ i.p.s. and $\pm 4 \mathrm{~dB}$ at $50 \mathrm{~Hz}-7.5 \mathrm{kHz}$ at $3 \frac{3}{3}$ i.p.s. Signal-tonoise ratio from peak record level to broad band noise at the pre-amp output is 46 dB (unweighted) at $7 \frac{1}{2}$ i.p.s. and 43 dB (unweighted) at $3 \frac{1}{4}$ i.p.s. The deck model weighs 23 pounds. The price is 79 gns. Ampex Great Britain Ltd., Acre Road, Reading, Berks.
ww 301 tor further detaits

## GUNN DIODE OSCLLLATOR

THE oscillators in the CL8 series by Mullard are intended primarily for experimental purposes and for performance assessment in microwave systems. They can also be used in speed checking equipment and bench-top microwave demonstration systems for schools and colleges. Each oscillator has a gallium arsenide device fitted in a small cavity which can be mechanically tuned over 1 GHz . Self-contained, a typical continuous output of 5 mW can be obtained with a supply of 7 V d.c. There are four oscillators, the CL8 $360,370,380$ and 390 , covering the range 8 to 12 GHz . Mullard Letd., Mullard House, Torrington Place, London, W.C. 1.
WW 102 for further detalis


## Operational Amplifier

FLEXIBILITY of the Westinghouse WC306 operational amplifier (dual inline package) is indicated by the fact that (a) a choice of inputs is a a ailable, a high impedance $300 \mathrm{k} \Omega$ pair of differential inputs and a low impedance $3 \mathrm{k} \Omega$ pair, (b) outputs are, a differential pair of output terminals as well as the regular single-ended output, and (c) bandwidths can be selected up to the 30 MHz unity gain of this device, using the low input impedance terminals; 40 dB of gain can be achieved at 10 MHz . Most of this amplifier's 1,100 to 4,400 open loop gain can be used without exceeding $0.2 \%$ distortion. Only sufficient feedback to maintain d.c. operating point stability is necessary. With a worst case situation of $5 \mathrm{k} \Omega$ source impedance, and a 150 kHz bandwidth the noise is only $12 \mu \mathrm{~V}$ r.m.s. In many low-frequency instrumentation applications where noise is serious, lower source impedances and restricted bandpass will substantially lower this figure. Common mode rejection ( 83 dB ) retained at high frequencies, low thermal drift ( $5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$ ) and an output voltage swing of $\pm 7 \mathrm{~V}$ are additional features of this op-amp. Applications
include high-frequency video amplifier for driving push-pull loads such as c.r.t. deflection plates, servo motors and speaker coils. The differential outputs could also be used for driving balanced transmission lines. It can also be used as a high-fidelity pre-amplifier for audio work. Westinghouse Electric Corporation, Molecular Electronics Division, Box 7377, Elkridge, Md. 21227.
ww 303 for further detaits

## Coaxial Attenuator

A Kay Electric (U.S.A.) in-line attenuator, the Series 110 has a frequency range of d.c. to 4 GHz attenuated over 132 dB in 1 dB steps. It employs a segmented intrinsic cavity switch assembly. A simple slide switch operation controls individual steps of $1,2,3,6,10,20,30$, $30,30 \mathrm{~dB}$. The overall accuracy is $2 \%$ up to a maximum of 1 dB at 1 GHz and $5 \%$ up to a maximum of 5 dB at 4 GHz . The U.K. agents are Wessex Electronics Lid., Royal London Buildings, Baldwin Street, Bristol, 1
WW 304 for further details

## TRACKING FILTER

IN order to reject harmonic and rattling distortion from the control accelerometer signal of an electro-magnetic vibration system, Derritron have introduced the solid-state tracking filter TF1. This provides precise tracking at high sweep rates. Automatic bandwidth switching between five crystal filters ( $3,10,30,100$ or 300 Hz bandwidth) may be selected by three separate programmes derived from the compressor functions of the Derritron vibrator control oscillator, VCO1, with which the TF1 is specifically designed to operate. A band reject output is available in addition to bandpass and d.c. analogue outputs. The tuning signal can be derived from any audio oscillator provided the amplitude variation does not exceed 40 dB . It is primarily designed for use in sine, random, sinerandom and swept-random vibration testing or for precise analysis and measurement of audio signals. Derritron Electronic Vibrators Lid., Sedlescombe Road North, St. Leonards-on-Sea, Sussex.
WW 305 for further details

## Silver Bearing Solders

SOLDERING of silver-coated glass and ceramic surfaces and silver-plated components presents difficulties when using conventional tin/lead solders. Enthoven Solders Ltd. have produced a range of silver-bearing solders for such work, since the solubility of silver in tin is said to be greatly reduced by using a solder alloy already bearing a specific silver content. This does not affect the inherent solderability of the tin-rich alloys. The melting ranges of the silver bearing solders lie within the normal soft solder range. Enthoven Solders Ltd., Dominion Buildings, South Place, London, E.C. 2.
WW 306 for further detalls

## Resin Disintegrator

A disintegration solvent for use with epoxy and polyester resins is available from Oxley-Developments Co. Lrd., Priory Park, Ulverston, North Lancashire. De Solv 8090, as it is called, is non-corrosive to metals (in normal use), non-inflammable and is of low toxicity. It is for use in recovering embedded electronic circuits, and for any applications where the removal of resins, paints and lacquers of the epoxy and polyester type is desirable. The disintegrator can be recovered for further use by filtering out the sediment.
wW 307 for further octails

## I.C. Breadboard

AN integrated circuit breadboard for dual-in-line i.cs is manufactured by Spectrum Electronics Ltd. The first unit to be offered is the ICB. 707 which has provision for interconnecting twelve dual-in-line circuits. From the wide range of digital and linear dual-in-line circuits available, complex systems can be quickly built, demonstrated and proved. This device features solderless interconnections throughout, and reduces to a minimum the damage to i.cs. Each pin of the twelve high grade four-teen-lead i.c. sockets is brought out to a five-way socket which may be connected to any desired adjacent socket by colour coded leads. I.C. sockets and five-way sockets are coded for easy identification. Common power and earth points are available at each i.c. socket and are terminated in 2 mm binding posts. Two coaxial sockets and binding posts with interconnection sockets are available for input and output signals. Sections of prototype systems can easily be isolated and monitored for demonstration or circuit optimization. Easy insertion and removal reduces the stockholding by making the i.cs immediately available for other experimental designs. Systems

connections can be changed quickly to investigate a new design or evaluate an alternative supplier's product. The unit is ideal for prototype educational and feasibility studies. A logic handbook is provided for i.c. interconnection leads, and an i.c. extractor tool. The breadboard is $6 \frac{1}{4} \times 6 \frac{1}{4} \mathrm{in}$ and weighs $1 \frac{1}{2} \mathrm{lb}$. Spectrum Electronics Lid., Deneway House, Potters Bar, Herts.
WW 308 for further netalls

## High-frequency Sampling Adaptor

IN the AIM sampling adaptor the gate is open for only 350 picoseconds. This refinement of circuit technology can be better understood if expressed in more practical terms; during the time the gate is open for each sample, a beam of light would travel only five inches. This waveform sampling adaptor type WSA114 gives a 1 GHz bandwidth to $x-y$ recorders, oscilloscopes and audio spectrum analysers. It permits the examination of complex repetitive waveforms containing frequencies up to $1,000 \mathrm{MHz}$. It has four modes of operation, each with good sensitivity and linearity. In "auto" mode it may be coupled to an $x-y$ recorder or lowfrequency oscilloscope. Both $x$ and $y$ inputs are provided independently. In "coherent" mode, the unit provides a very slow representation of the original waveform. This may be fed to an $x-y$ recorder, or a spectrum analyser and the original waveform can thus be recorded, analysed or recovered from noise. "Manual " mode provides the facility of scanning point by point through the original waveform. Morcover, in all these modes it is possible to magnify a single section of the waveform by a factor of up to 50 . Finally, the "in-
coherent" mode permits the measurement of peak, average or r.m.s. voltage of r.f. waveforms, without the necessity of adjusting synchronization. The technique used in the WSA114 is essentially a stroboscopic one, where samples of the high-frequency waveform are taken at successive intervals, stored for a time in an integral memory, and then assembled into a low-frequency representation of the original waveform. This technique permits recovery of signals which are obscured by up to 20 dB of noise. Moreover, the accuracy of representation of the waveform is said to be better than that of high-speed sampling oscilloscopes. In practice, the input signal is applied to an electronic gate which opens a little later in each scan of the high-frequency waveform so that over, say, 1,000 gate cycles, sufficient samples of the high-frequency trace are collected to permit faithful reconstruction at low speed. The stroboscopic sampler is locked to the incoming signal, and no delicate adjustment is needed to find and then follow a signal whose frequency is drifting. AIM Electronics Ltd., 71 Fitzroy Street, Cambridge.

WW 309 for further detalls

## Counting Modules

A RANGE of plug-in integrated circuit counting modules is being marketed by Darang Electronics under the trade name DIGIC. Each counting stage is encapsulated in an anodized aluminium can, the range being originally designed fic use in Darang's Digicron digital clocks, counters and tachometers. The modules are available either with the display tube as an integrated part of the module or with a flying lead up to sis feet in length between the display tube and the module. Counting ranges between 0 and $2,3,4,5,6,7,8$ or 9 and intermediate ranges 2-4, 3-8, 1-7 are available. The module provides a n.b.c.d. (8421) output and in addition a slave display may be driven from the display tube termination point. The display tube is a standard Mullard type

(ZM 1040) giving a 30 mm digit height; the power requirements are +6 V at 50 mA and +240 V at 5 mA . The inpui pulse requirements are amplitude +1.5 to 4 V , duration 150 ns minimum and the rise time must not exceed $1 / \mathrm{s} / \mathrm{V}$. The devices will operate in the temperature range $0-60^{\circ} \mathrm{C}$. Darang Electronics Lid., Restmor W'ay, Hackbridge Road, Hackbridge, Surrey
Ww 310 for further details

## Function Generators

TRIGGER, phase lock and tone burst capabilities are now available in two portable function generators by Wavetek, U.S.A. The Model 115 offers triggered or gated operation as well as phase lock capability. In the trigger mode, a manual or external voltage of $\pm 0.5 \mathrm{~V}$ will generate one cycle. In the gated mode, a discrete number of cycles can be generated by applying a $\pm 0.5 \mathrm{~V}$ gate The unit will phase lock to the fundamental of the dial frequency with specified accuracy. Model 116 has all the capabilities of the model 115 with the addition of tone burst operation, which may be generated automatically in the rrigger mode. Selectable from a front panel control the 116 will generate from 1 to 256 discrete cycles. Both models will also generate sine, square, triangle, ramp and sync pulse waveforms. Nine simultaneous outputs are available over a frequency range of $0.00015 \mathrm{c} / \mathrm{s}$ to $1 \mathrm{Mc} / \mathrm{s}$. Additionally, both units are voltage controlled, allowing a $20: 1$ frequency sweep over the full dial spread.

These two portable instruments have the following common specifications: dial accuracy- $0.5 \%$ of reading, frequency response amplitude change with frequency less than 0.1 dB , amplitude stability is $0.1 \%$ of maximum peak-topeak values for 30 minutes, and d.c. offset stability is $0.1 \%$ of maximum peak-to-peak values for 30 minutes short term. Sine wave distortion is less than $0.5 \%$ Triangle and ramp linearity greater than $99 \%$ to $100 \mathrm{kc} / \mathrm{s}$. The square wave tise and fall time is less than 5 nsec . Both models use silicen semiconductors throughout and have individually calibrated dials; each is avail-
able in both a.c. and battery-powered versions. General Test Instruments, Gloucester Trading Estate, Hugglecote, Glos.
WW 311 for furfher detalls

## LOGIC SYSTEMS

IN the Farnell logic system simulator, stepped progress can be made from simple logic functions to more complex logic techniques. Binary arithmetic is also introduced and the accompanying manual has a section on Boolean algebra and De Morgan's theorems. The simulator consists of a plinth unit to support the logic modules, a power supply and a range of modules including NOR, AND, nand units, lamp, switch and binary units. Additional medules available are shaper, generator, photocell and proximity units, and 150 mA and 500 mA driver units.

WW 312 for further details

## Variable Delay Line

A VARIABLE delay line adjustable between 10 and $18 . \mu \mathrm{sec}$ with a dynamic signal-to-noise ratio of $7: 1 \mathrm{minimum}$ is available from Sealectro. Deltime LG14 produces 40 mV minimum output across 4.7 k !? when driven with 10 V at 60 mA peak current. This model is soldersealed for military applications.
WW 313 for further detalls

## Linear Motor

AN electric actuator with linear movement is the description given to the linear motor produced by AB Lineara of Sweden. This type of motor has a stator fixed to a solid member, and a metal guide complete with two end stops. When current is applied to the stator, the guide moves at a speed of 1.2 metres ( 4 feet )/ sec in 20 ms with horizontal mounting and unloaded armature. Direction of movement is changed by reversing the current. Although in principle the length of the stroke is infinite, the standard motor has an armature giving a 250 mm (10in) stroke length, while the force along the stroke length is constant. For use in moving doors, valves, rejecting packages, moving items in packaging systems, this motor is said to permit a great deal of freedom in design. It can be mounted in any position, alihough minimum friction between stator and guide occurs when both are vertical. If the guide is fixed, then the stator moves and this method of operation is useful where a mounted or suspended item has to be moved over a definite length. This type of motor is intended for single phase 220 to $240 \mathrm{~V} 50 \mathrm{c} / \mathrm{s}$ operation, thus a phase-shift capacitor is required and is delivered as a standard accessory. The notor control and regulation can
be achieved with conventional electronic equipment. The armature is made from copper as standard, but the material cart be any electrically conductive nonmagnetic material. The length of the armature is not limited, and its movement can be slopped mechanically without causing motor damage. Motors can be mechanically coupled in series or parallel. The price is $\mathbf{2} 25$ 10s. The general agent and distributor in Great Britain is Hird-Brown Ltd., Bolton, Lancashire.
WW 314 tor lurther detalls

## Portable Oscilloscope

THE Cossor CDU130 solid-state portable oscilloscope has a bandwidth of d.c. 1015 MHz . Field effect transistors are used to reduce $Y$ amplifier drift, eliminate microphony and ensure accuracy at slow timebase sweep speeds. The sensitivity is $5 \mathrm{mV} /$ division at full bandwidth. The operation of this instrument is possible with external a.c. or d.c. supplies; it will also operate for five hours from the internal battery provided as a standard. A battery charger is contained within the 'scope and the battery is protected against reverse charging. Wright complete is $16 \frac{1}{2} \mathrm{l}$ b.
WW 315 for lurtier details


## OSCILLOSCOPE CAMERA ACCESSORIES

POLAROID film pack backs are now available at no extra cost as alternatives to the present roll-film backs fitted to Telford Type A oscilloscope cameras. The 3,000 ASA eight-exposure pack film is said to be quicker and easier to load and manipulate. As each exposure is precessed away from the camera, it is no longer necessary to await for the 15 s processing time betwéen successive shots, and multiple photography is greatly simplified where banks of cameras are used. Owners of Type A cameras with roll-film backs can buy pack backs separately, for $£ 30$. The pack back, when it is used with the Telford slide plate permits the taking of up to 13 exposures on one print. Also available is a high-speed $\mathrm{f} / 1.3$ lens, whose wide aperature means that rise times in the order of $10-15 \mathrm{nsec} / \mathrm{cm}$ can be photographed, using 10,000 ASA film. Telford Products Lid., 4 Wadsworth Rd., Greenford, Middx.
WW 316 for further detalls

## POWER TRANSISTORS

FOR use in radar pulse circuits as well as in high-power u.h.f. transmitters, the two transistors 2N5177-8 by TRW Semiconductors Inc. have an r.f. power output of 25 and 50 W respectively. Mounted in a grounded emitter stripline package, both devices will produce their outputs at $500 \mathrm{Mc} / \mathrm{s}$ with a $\mathrm{V}_{\text {CE }}$ of 28 V . The following parameters are common to hoth types: $\mathrm{V}_{C B O} 55 \mathrm{~V}$; $\mathrm{V}_{\text {CEO }} 35 \mathrm{~V}$, and $\mathrm{V}_{\text {EBO }} 3.5 \mathrm{~V}$. The dissipation, collector current and base current for the 2 N 5177 are $33 \mathrm{~W}, 4 \mathrm{~A}$ and 1 A respectively, and the same parameters for the 2N5177 have the following figures, $65 \mathrm{~W}, 8 \mathrm{~A}$, and 2 A respectively. M.C.P. Electronics Lid., Alperion, Wembley, Middlesex.
WW 317 tor further setalls

## Thermoelectric Generators

A STEADY and reliable electric power output, at working temperatures of up to $300^{\circ} \mathrm{C}$, is claimed for the range of thermoelectric generator modules by G. V. Planer Lid. Exploiting the Seebeck effect, these modules are intended for use in marine and aircraft nawigational aids, telecommunications systems and remote weather stations. The generators are constructed from 50 thermo-elements which in turn are produced from $p$ and $n$ type semiconductor alloys based on bismuth telluride. Although the elements are connected electrically in seriss, in order to produce the necessary "hot" and "cold" faces, they are placed in parallel thermally. The establishment of a tempera-
ture difference between the faces produces a veltage (Seebeck effect), the magnitude of which is cetermined by the temperatare gracient and the notrix configuration. The array is encapsulated to give a monelithic, mechanically strong assembly which is capzble of operation at elevated tumperature. Both types have a maximum hot sink tempersture of $300^{\circ} \mathrm{C}$ and an open circuit voltage of 3.6 V for a temperature difference of $200^{\circ} \mathrm{C}$. Typ: TPG/205 has a matched load output of 750 to 900 mW , and Type TPG/210 has a matched liad outpat of 400 to 500 mW . G. V. Planer Ltd., Windinill Road, Sunbury-onThames, Middlesex.
ww 315 fis thriner details

## Broadcast Receiver

COVERAGE of the long- and mediumwave broadcast bands and continuous coverage of the shore-wave bands down to below the popular 16 -metre band, is provided by the Eddystone EB36 solidstate broadcast receiver. It is completely self-contained, having its own audio amplifier stages and loudspeaker, but an audio output is available for an external tape recorder of hi-fi amplifier. Battery power supplies are provided within the receiver unit, to make the complete receiver independent of any external supply. In this way, it can be operated in a wide variety of portable roles, including road vehicles, small boats and even light aircraft without any additional facilities apart from an aerial. An a.c. mains power unit is available to 1eplace the battery in the receiver. The EB36 incorporates the well-known Eddystone tuning control, with a high tuning ratio to enable precise frequency settings to be obtained. The tuning control is loaded with a heavy flywheel, which makes it possible to spin the dial to cover large changes in frequency very rapidiy. Five frequency scales are pro-

vided, covering long-wave, midiumwave and three short-wave bands (from 1.5 to $22 \mathrm{Mc} / \mathrm{s}$ ). An additional scale, calibrated in arbitrary units, can be used in conjunction with a small vernier dial to provide a very precise definition of points on any of the five frequency scales. The price of the EB36 is £54 5s 7d. Eddystone Radio Lid., Eddystone Works, Alvechurch Road, Birmingham 31.
ww 319 tor further detalls

## Component Packs

FIRST two in the new range of component packs presented by Peak Sound, designed for use with the "Cir-Kit" system, are available at 15 s each. Each pack contains full building and layout instructions. Pack No. 1 contains 15 components to build any one of a range of five different circuits; a high input impedance pre-amplifier with a gain of $\times 100$, a multiple output signal injector, a multimeter high ohms range
extender, multimeter low current range extender, and a mono pre-amplifier for moving coil microphones, giving à gain of $\times 100$. Pack two contains components to build various types of pre-amplifier and multimeter range extenders. Other packs contain components for building amplifiers, pre-amplifiers, and power packs. Peak Sound (Harrow) Ltd., 10 Asher Drive, Ascot, Berks.
WW 320 for further detalls

# D.C. Amplifier 

A D.C. amplifier, the 104, providing nanovolt resolution and millisecond rise time for d.c. voltage measuring systems is offered by Keithley Instruments Inc., 28775 Aurora Road, Cleveland, Ohio 44139 , U.S.A. The gain range of 100 to 100,000 has an accuracy of $\pm 0.01 \%$ and the linearity is $\pm 5$ p.p.m. of full scale. It is particularly useful for process control and automated data handling applications where it is used with a digital voltmeter for measurement of nanovolt and microvolt signals. It has a 10 V full scale output with $10 \%$ average for all gain ranges at up to 1 mA at full scale. The peak-to-peak noise

varies from $10 \mathrm{nV}( \pm 50$ p.p.m.) at a gain setting of $10^{3}$ to $1 \mu \mathrm{~V}$ ( $\pm 5$ p.p.m.) at a gain setting of $10^{2}$. Input impedance is greater than $50 \mathrm{M} \Omega$ and the output impedance is less than one ohm. Rise times are selectable at nominal settings of $0.05,0.5$, and 5 seconds.
ww 321 for further detalls

## Ceramic Capacitors

MINIATURE ceramic capacitors covering the range 10 pF to 2700 pF are available from Erg Industrial Corporation Ltd., Luton Road, Dunstable, Beds, Although the standard tolerance is $\pm 10 \%$, other tolerances are available. The standard temperature coefficient conforms to MK-C-11015C which means that the value observed at $25^{\circ} \mathrm{C}$ will be maintained within $\pm 15 \%$ over the range -55 to $+125^{\circ} \mathrm{C}$. Working voltages are from 50 to 200 V d.c., and these capacitors can withstand a d.c. potential of $400 \%$ of rated voltage applied at $25^{\circ} \mathrm{C}$ for five seconds with

current limited to 50 mA maximum. Standard leads are tinned copper and dual purpose weldable/solderable leads of gold flashed dumet are available. ww 322 for further detalls

## Digital Integrated Circuits

A NEW series of digital integrated circuits have been introduced by the Raytheon Company of America. Designated RM2000, the series consists of a quad level translator, a current driver, and a lamp driver. The RN: 2000 quad level translator consists of four levelshifting inverters, each with two alternative inputs. Signal inputs are at 28 V for one input, 14 V for the other. The RM2001 is a monolithic high voltage ( 40 V ), high current ( 250 mA ) driver with inputs compatible with the 930 Series DTL. Because of this compatibility, the circuit offers logic flexibility in addition to its current and voltage capabilities. Intended primarily for use as a relay or lamp circuit, the RM2002 is a high voltage $(40 \mathrm{~V})$, high current $(250 \mathrm{~mA})$ unit. The inputs to the driver are compatible with 930 DTL Series circuits. A 930 DTL gate is also provided on the chip for additional logic capability.

The RM2000 Scries is guaranteed
over a temperature range of $-55^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$. The hermetic seal limits leakage to $5 \times 10^{-8} \mathrm{~cm}^{3} / \mathrm{s}$ of helium. Raytheon Overseas Lid., Lexington, Massachusetts 02173, U.S.A.
wW 323 for furtier detalls

## Frequency Doubler and Phase Shifter

THE Brookdeal DP325 frequency doubler/phase shifter has been designed particularly for use in the reference channel of a phase-detection, small signal recovery system where the information frequency is twice that of the excitation or modulating waveform. The output level of 3 V r.m.s (f.s.d.) can be monitored on the output meter, and inputs from 10 mV to 100 V (f.s.d.) ca .a be accepted. Facilities are provided for phase shifting the output with or without frequency doubling, giving greater than $180^{\circ}$ control. The frequency range is $30 \mathrm{c} / \mathrm{s}$ to $300 \mathrm{kc} / \mathrm{s}$. Input and output impedances are $100 \mathrm{k} \Omega$ and $600 \Omega$ respectively. Compensation is provided within the instrument for input sine waves of poor wave shape, and circuit stabilization is achieved through a high degree of d.c. and signal frequency feedback. Brookdeal Electronics Ltd., Myron Place, London, S.E.13.

WW 324 tor further detalls

## STEREO AMPLIFIER

AVAILABLE in kit or assembled form from Daystrom, the TSA-12 Heathkit stereo amplifier has an output of 12 W r.m.s. per channel into an $8 \Omega$ load. The output is also suitable for $15 \Omega$ loudspeakers ( 8 W r.m.s. per channel) and there are three inputs for gram, radio, and auxiliary signals. Channel separation is 45 dB or better, and the frequency response is stated as 13 Hz to 60 kHz $\pm 1 \mathrm{~dB}$ and 7 Hz to $95 \mathrm{kHz} \pm 3 \mathrm{~dB}$. Total harmonic distortion at 1 kHz at $0.5 \%$ or less at rated output; and at 20 Hz to 20 kHz it is $1 \%$ or less at the rated output. It possesses the usual complement of controls and employs 17 transistors and six diodes.
WW 325 for further details

## Tunable Pot Cores

ENCAPSULATED tunable pot core assemblies in the Plessey Alpha range have been designed with the close requirements of telecommunications work in mind. One of these requirements is that of temperature co-efficient control and t.c. gradings in these com-

ponents fall into two broad categories arising from the intrinsic characteristics of the ferrites appropriate to the frequency bands. Generally, t.c. is linear for frequencies up to $2 \mathrm{Mc} / \mathrm{s}$ and nonlinear for frequencies between 2 and $8 \mathrm{Mc} / \mathrm{s}$. This range is suitable for t.c. performance and grading from 0 to 120 p.p.m. $/{ }^{\circ} \mathrm{C}$ over a temperature range $-25^{\circ} \mathrm{C}$ to $+55^{\circ} \mathrm{C}$. The encapsulated assemblies are housed in hot tin-dipped copper screening cases, the encapsulant being flexible silicone with good dielectric properties up to $8 \mathrm{Mc} / \mathrm{s}$
wW 326 for further details

## Low-Level SCRs

A sensitive gate s.c.r. series rated at 800 mA (forward current r.m.s.) has been designed by Motorola for low-level power control circuits. The series consists of four device types-2N5060 to 2N5063 inclusive, with voltage ranges from 30 to 150 V . Gate current requirements for these units is only $200 \mu \mathrm{~A}$. The new geometry used in these devices features larger bonding areas on the die to provide better power dissipation. Additional features include low holding current ( 5.0 mA max at $25^{\circ} \mathrm{C}$ ), a 6 A peak surge for protection, a 1.7 V peak forward "on" voltage ( 1 A at $25^{\circ} \mathrm{C}$ ), and low blocking currents ( $50 \mu \mathrm{~A}$ maximum at rated voltage and $125^{\circ} \mathrm{C}$ ). Motorola Semiconductor Products Inc., York House, Empire Way, W'embley, Middlesex
ww 327 tor turther detalls

## VOLTAGE PROBE

ON the VT100 Amprobe voltage tester, voltages are shown "thermometer" style via a series of lit windows that correspond to the following voltage levels; 115, $220,277,440,550 \mathrm{~V}$ a.c. and 115,220 , $400,600,750 \mathrm{~V}$ d.c. When the probe is connected to the a.c. supply the window indicating the relevant voltage lights up and a buzz, is heard the pitch of which is determined by the supply frequency. This instrument will also indicate correct d.c. polarity. The body incorporates a sliding probe and there is another probe attached to an expanding coil cord, thus permitting measurement of points of up to three feet apart. Soss Manufacturing Company, Lynbrook, New York 11563, N.Y., U.S.A

WW 328 for further detalls



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## The Unknown Warriors

SEEING that the time to ring out the old, ring in the new, is almost upon us, it was the intention of Old Moore Vector to gaze into his crystal ball to report upon the future of the electronics industry. One preliminary sidelong peek however, and common humanity made him desist. After all, you are going to have quite enough to put up with as it is, what with forking out for mother-in-law's present, and treading on the holly which the kiddiwinks have installed by their bedsides as a Santa auto-alarm.
So instead I thought it would be nice and seasonable if we paid tribute to the forgotten men of the radio industry. The men who, long after the factorics have closed and the labs have locked up for the Christmas, will be tootling around until the small hours of the morning on behalf of those whose sets have gone up in smoke at the last moment. I mean, of course, the chaps in the little shop around the corner; the fellows who have the privilege of repairing that television receiver you bought at cut price in the Tottenham Court Road. Those Tail End Charlies of the receiver industry, the small retailer and his serviceman (often one and the same person).

As we all know to our cost, there are dealers and dealers. There is the city shark with the flashy chromium shop front who welcomes every stranger in with gently smiling jaws (although that was a crocodile wasn't it?). He lives by the late Mr. Barnum's dictum that there is one born every minute and he has never had occasion to quarrel with the sentiment. Then we have an immense variety of chain stores, furniture emporiums, bicycle shops, ironmongers and so on, who run radio as a sideline. Some have qualified personnel; many don't. There's no telling until you've been a customer.
Another phenomenon is the man who, on the strength of having built a simple kit set or two, sets up a pin-money repair business at home. The bonafide dealers love him for his artless habit of adding or subtracting components from the circuit as the mood takes him, then washing his hands of the whole business when the set refuses to work. This chap also has a mysterious source of supply from which he can offer leading makes of receivers at a "little bit off." The dealers love him for that, too.
Finally there is the genuine article; the dealer who runs an honest business and backs it with first-class service. He may be in a big way of business; he may be found in a small village store (one of the best dealer-service engineers I know sells tobacco on the other side of the shop). The tragedy is that from a superficial pavement inspection the genuine article is quite indistinguishable from the riff-raff.
Possession of well-known receiver agencies offers no criterion, for although much lip service is offered by manufacturers to the importance of an efficient service department, a commanding position in the shopping area and a good window-frontage can work wonders in the appointment of a dealer.
One truly startling feature of the radio receiver industry is that prices are actually lower today than they were 45 years ago. You don't believe it? Neither did I, until I came across some radio magazines of 1924 vintage. The price of a reputable two valve set was then roughly $£ 20$. So you can get a 10 -transistor portable today for less than the price paid for a two valve set in 1924 (off-hand I would say
that the quality of reproduction was about comparable, too, but that's another matter). Compare the price of a television receiver of 1939 vintage with one of today's models and you will see the same downward price trend.

This is in contrast to the "times six" price increase in motor cars and most other commodities over the same period. Receiver manufacturers will argue that this is a tribute to their efficient mass production techniques; in fact, it is more of a tribute to their genius for trying to slit each others' throats and thereby killing the golden goose, if I may mix my metaphors.

The effect on the poor old dealer is all too plain. It means that he is working on the same, or perhaps even a lower, profit margin than he did in 1924. Not merely in percentage but in cash. If he made $£ 5$ profit per set in the 1920's he still makes $£ 5$ today, in spite of the huge increase in overheads. Take servicing equipment for instance. In the eariy days he could get by with little more than a buzzer and battery. Today even a modest service department houses many hundreds of pounds worth of equipment. Considering the doldrums in sales which have existed these many years it's not surprising that a lot of dealers have taken a one-way ticket to Carey Street. The wonder is that any have survived at all.

Now colour is with them, like an angry dawn portending more stormy weather; more very expensive servicing equipment and a lower mean time between failures to swallow the extra profit margin. In the cities and stockbroker belts things may not be so bad, but spare a thought for the little man in a remote farm-labouring area. He can count his colour-sales prospects on the fingers of one hand and still have plenty of digits left, but if he sells one colour set he still has to provide the means of servicing it, just as surely as if he was selling them by the hundred.

There was one man, Frank Murphy, away back in the 'thirties, who came up with some ideas which were regarded by his fellow manufacturers as completely screwball. He started making receivers which were engineered to professional standards, with price a secondary consideration (oddly enough, they were only a little higher than average). He appointed his dealers very carefully, making sure that their service departments were of a high standard, but, orce appointed, they had exclusive territories of considerable size.

He made the dealer into an external arm of his manufacturing effort by requiring him to provide a monthly return of repairs effected to his sets. If, then, a component was seen to be giving trouble, a better one was substituted, even if production had to be halted temporarily.
The customers were happy because their sets kept working year in, year out. The dealers were happy because their appointment was in effect a certificate of competence and even though their discount was rather less than average, they didn't have any significant frec servicing to do. The man who dreamed the system up was happy with a modest profit. Then, surprisingly, he got out of the business altogether, to the great loss of the industry.
Meanwhile, like any junkie, the industry has relied on periodic shots in the arm to keep it going; the latest of these is colour, over which no doubt, the suicidal pricecutting policies will continue.

## Whe is an Avo meter not an Avometer?

When it gives you (a) $\pm 0.3 \%$ accuracy, (b) (c) $100 \%$ solid state, (d) (e) (f) semiconductor characteristics data, (g) valve characteristics data, or (h) digital $L / C / R$ measurements.


PRECISION AVOMETER Mearurea d.c. a voltage( $1.6-1500 \mathrm{~V}$ scales, $\pm 0.3 \%$ f.a.d. ${ }^{\circ}$ ). d.c. current $\left(1.5 \mathrm{~mA}-15 \mathrm{~A}\right.$ scal ${ }^{\mathrm{s}}$, $\pm 0.8 \%$ f.a.d."), a.c. voltake ( $3 \mathrm{~V}-1500 \mathrm{~V}$ acales, $\pm 0.75 \%$ f.s.d.), a.c. current ( $3 \mathrm{~mA}-15 \mathrm{~A}$, $\pm \mathbf{0 . 7 5 \%}$ (.s.d.). - meets B.S.S. $89 / 1954$ for precision-grade instruments.


1] MULTIMETER Hilos Battery-operated O fully-transistorised. measures a.c/d.e. voltage $(100 \mathrm{mV}-1000 \mathrm{~V}$ scales, $\pm 4 \% \pm 3 \%$ f.s.d.), a.c./d.c. current (1uA-3A scales, $\pm 4 \% \pm 3 \%$ f.s.d.), resistance ( $2 k \Omega \cdot 20 \mathrm{M} \Omega$ scales), power ( -20 to $+60 \mathrm{db}, 9$ scales), r.f. voltage ( $300 \mathrm{mV}-10 \mathrm{~V}$ scales, up to 250 MHz with external probe available separately).

$c^{M}$
MULTIMETER CT471A Battery-operC ated, fully-transistorised, sensitivity $100 \mathrm{M} \Omega / \mathrm{V}$, measures a.c./d.c. voltage ( 12 mV . 1200 V нcales, $\pm 3 \% \pm 2 \%$ f.s.d.), a.c./d.c. current ( $12 \mu \mathrm{~A}-1.2 \mathrm{~A}$ scales, $\pm 3 \% / \pm 2 \%$ f.s.d.) resistance ( $12 \Omega-120 \mathrm{M} \Omega$ scales, $\pm 3 \% \mathrm{~m} . \mathrm{s} . \mathrm{d}$ ) h.f./v.h.f./u.h.f. voltage with multiplier (4V400 V scales up to $50 \mathrm{MHz} \mathbf{4 0 \mathrm { mV } - 4 \mathrm { V } \text { up to }}$ 1000 MHz ).

d IN.CIRCUIT TRANSISTOR TESTER d TT164 Direct-reading, easy to operate. accurate measurements under static and dynamic conditions. Collector voltage: continuously variable, 0.10 V . Collector current: continuously variable $0-10 \mathrm{~mA}, 20 \mathrm{~mA}$, 30 mA . Measures beta ( $\mathbf{1 5 0 - 3 0 0}$ scblea. $\pm 5 \%$ ) and leakage current ( $300 \mathrm{nA}-1 \mathrm{~mA}$


- TRANSISTOR \& DIODE TESTER C TT537 Measurea both transistor and diode characteristics. Collector voltage: continuously variable 0-12V, stabilised. Collector 11A-50mA. Mensures hre(50-1500 scales, $\leq 3 \%$ leakage current (50, 1.5 scale s). diode forward voltage drop ( $1.5-5 \mathrm{~V}$ scalen, $0-500 \mathrm{~mA}$ forward current and breakdown
voltage ( $100-1000 \mathrm{~V}$ scales, $3 \mathrm{~mA} \& 200 \mathrm{~mA}$ voltage $\mathbf{~} 100-1000 \mathrm{~V}$ scales, $3 \mathrm{~mA} \& 200 \mathrm{~mA}$ currents

f TRANSISTOR ANALYSER MK2 Avail1 able in both mains-powered and batterypowered versions; provides accurate measurements ingrounded-emitter configuration; acremmodates high-power and switching types. Collector voltage: $0.05-12 \mathrm{~V}$ (up to 150 V external). Base current: 1.40 mA scales. Collector current: to 1 A in 5 rankes. Measures leakage current (from $2 \mu \mathrm{~A}$ ), hfe ( $25-250$ scales), saturation voltage, turnover voltage and noise factor.

$\sigma^{\circ}$ VALVE CHARACTERISTIC METER 5 VCM183 The most comprehensive instrument of its kind ever offered by Avo. Provision for testlng nuvistorn. compactrons and otherspecial types with up to 13 pin connections. No need to back off stand Ing anode current before measuring mutual conductance, which is continuously moniored under all conditions. Heater voltage: $0-119.9 \mathrm{~V}$ in 0.1 V steps. Anode and screen voltares. 12.6 V .400 V Grid voltare: $0-100 \mathrm{~V}$ continuous. Measures gm: $6 \cdot 60 \mathrm{~mA} / \mathrm{V}$ f.s.d. in 3 rangiges.


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DATA

| Service | Peak Inverse <br> voltage <br> type | CV1147 | Peak forward <br> voltage <br> max. $(\mathrm{kV})$ | Peak anode <br> current <br> max. (A) | Mean anode <br> current <br> max. (A) |
| :--- | :---: | :--- | :--- | :---: | :---: |
| BT5 | - | 1.5 | 1.0 | 12.5 | 2.5 |
| BT17 | CV1144 | 2.5 | 1.0 | 40.0 | 6.0 |
| BT19 | - | 2.0 | 2.5 | 2.0 | 0.5 |
| BT29 | - | 15.0 | 15.0 | 75.0 | 12.5 |
| BT69 | CV5141 | 15.0 | 15.0 | 75.0 | 12.5 |
| BT95 |  |  | 12.0 | 1.5 |  |

This range of Mercury Vapour Thyratrons is available from the following E.E.V. stockists. Prices are highly competitive.

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GT12A. Describes the Lilliput series of Ultra Minialure trans. formers and gives useful information and data on their application in transistor converter/inverter, wide band communication and high speed pulse circuits.

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## Voltage Stabilisers



DATA

| DATA | Service type | Operating voltage approx. (V) | Striking voltage (V) |  | Tube current range (mA) | Regulation max.$(\mathrm{V})$ | Base |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Type |  |  | 0 | - |  |  |  |
| OA2 | CV1832 | 150 | 185 | 225 | 5-30 | 6.0 | B7G |
| OA2WA $\ddagger$ | CV4020 | 150 | 165 | 225 | 5-30 | 5.0 | B7G |
| OB2 | CV1833 | 108 | 133 | 210 | 5-30 | 3.5 | B7G |
| OB2WA $\ddagger$ | CV4028 | 108 | 133 | 210 | 5-30 | 3.0 | B7G |
| OC2 | CV8766 | 75 | 115 | 145 | 5-30 | 4.5 | B7G |
| QS75/20 | CV284† | 75 | 110 | 160 | 2-20 | 6.0 | B7G |
| QS75/60 | CV434 | 75 | 117 | - | 5-60 | 5.0 | B8G |
| QS92/10 | CV188tt | 92 | 140 | - | 1-10 | 5.0 | Br.4-pin |
| QS95/10 | CV286 | 95 | 110 | - | $2 \cdot 10$ | 5.0 | B7G |
| QS108/45 | CV422 | 108 | 120 | - | 5.45 | 5.0 | B8G |
| QS150/15 | CV287 | 150 | 170 | - | 2-15 | 5.0 | B7G |
| QS150/45 | CV395 | 150 | 170 | - | 5.45 | 5.0 | B8G |
| QS1202 $\ddagger$ | CV4052 | 108 | 133 | 210 | 2-15 | 3.0 | B7G/F |
| QS1203 $\ddagger$ | CV4053 | 150 | 180 | 225 | 2-15 | 4.5 | B7G/F |
| QS1215 | CV5173 | 90 | 115 | 115 | 1-40 | 8.0 | B7G |

$\neq$ Arugged and rellable type $O$ In normal lighting in lotal darkness $\dagger \dagger$ Alsocv 1070 (operating voltage 100 V ) $\dagger$ Also CV 4083 (operating voltage 70 V )

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Gate Thyristors When be operated with shempristors are to pulses andion seeply rising current care must to operating exdised mestablishing the device of condions and selecting the reactor may be ined．A sel－saturating cuit to limit the rate－of－rise of current （ $\mathrm{d} / \mathrm{d} \mathrm{d}$ ）；this will permit a conventional high－power thyristor to carry heavy load currents which exhibit high $\mathrm{d} / \mathrm{dt}$ ．Where it is not practical to use such a reactor， which is often bulky and expensive，a thyristor with enhanced turn－on action must be used．Such action can be obtained by providing the thyristor with multiple gates．
IR multi－gate thyristors exhibit reduced turn－on voltage at any given instant during the turn－on period and shorter time for equalization of current flow throughout the entire semiconductor wafer．The consequent reduction in turn－ on power losses will permit increased load current to be carried and the device will exhibit faster turn－off time．It will also be able to withstand greater rates of rise of reapplied off－state voltages be－ cause of the lower junction temperature at the instant of current commutation．


MIM．Protection IR＇s epitaxial thyrist－ ors offer the exclusive feature of metal－ ion migration（M1M）protection．


During manufacture，the silicon wafer for epitaxial thyristors is contoured to improve the high－voltage characteristics of the device．This illustration shows the cross－section of a typical contoured silicon wafer．
Metal－ion migration can occut because of the electrical potential that exists at the junction interfaces at the edge of the wafer．When the device is energised， metal－ions are attracted from the metal mounting surface towards the junction interfaces．Migration may occur even though the wafer has been cleaned by etching and sealed with inert seaters or varnishers．When the minute metallic particles reach the interfaces，they can cause degradation or failure of the device． IR＇s epitaxial devices employ an exclu－ sive groove etching technique which provides needed contouring and，in addi－ tion，builds a guard－shield against metal－ion migration．
Bulk Avalanche Thyristors These de－ vices exhibit true avalanche bchaviour in the bulk of the crystal，thus avalanching at approximately the same voltage in both forward and reverse avalanche modes． Bulk avalanche devices are characterised by extremely low leakage current，which is mostly bulk leakage and which does not show any drift or instability under long－term，high－voltage blocking opera－ tion．In addition，IR＇s epitaxial thyris－ tors can be repeatedly broken over into the conduction mode without detri－ mental effects as long as the power rat－ ings and the rate－of－rise of turn－oft current（dI／dt）are kept within the listed specifications
As a result of the epitaxial construction， there is a substantial decrease in the for－ wW－008 FOR FURTHER DETAILS
ward voltage drop during turn－on．This reduces the total power loss during the turn－on action，which in turn reduces the temperature of the device．Therefore IR epitaxial thyristors are well adapted for inclusion in inverter and switching applications．
Ultra Fast Turn－Off Thyristors Early this year IR implemented a major techno－ logical breakihrough by going into quantity production at Oxted of thyristors exhibiting twn－off times below 3 micro－ seconds，fister than those yet produced by any other semiconductor manufac－ turer．To date this claim remains undis－ puted．The devices designated＂RCU＂ are offered in two current ranges of 8 and 10 amperes（full－cycle－average）with voltage ratings of $50-800$ volts PR V／PFV． The turn－olf＂times of all IR＂RCU＂ thyristors are measured at maximum． hase temperature．The maximum operat－ ing frequency of a thyristor circuit is obviously dependent on turn－off tinine， and introduction of＂RCU＂thyristors means that high－power inverter circuits may be operated at frequencies in excess of 30 kHz ．By utilizing＂RCU＂thyristors， the inverter designer may subsequently reduce the size and cost of the inverter components used in commutating circuits．
The principal applications for the ＂RCU＂hyristors also include high－fre－ quency induction heating，ultrasonic equipment and d．c．－d．c．converters． Delailed information about the world＇s leading range of thyristors and how they can solve your specific problems is yours on request from International Rectifier． Just ask．

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# This looks like a'B’size Ignitron 

## but it controls 65\% MORE POWER and saves money

## The new EEV Mini 'C' Ignitron

It's well-known that ' $B$ ' and ' $C$ ' Ignitrons are often used for applications which call for something in between. You can either overworka ' $B$ ' or underwork a 'C'. Whatever you do wastes money. To cut out this waste EEV have developed a new Mini ' C ' Ignitron which has a standard international ' B ' size envelope, but can handle $65 \%$ more KVA than the ' $B$ ' size. The new tube has a number of advantages. Take-over voltage is low to minimise misfiring at low current conditions, which in turn increases ignitor life. When used in place of a standard ' B ' size ignitron, you will find that the Mini ' $C$ ' lasts nearly twice as long. The cooling water is in direct contact with the vacuum envelope, and the inlet
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EEV's new Mini ' $C$ ' Ignitron is available from stockists throughout the country.

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Specially designed to mect the demand for the maximum possible flexibility from an amateur Transmitter which would otherwise be subjest so certain limi ations imposed
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 Amateur bands $160-10$ metres. fundamentals on 160 and 40 m . Ideal for Heathkit DX- 40 U
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A reasonably priced $Q$ Amplifier for the amateur and short-wave for the amateur and short-wave (200-250 v. $50 / 60 \mathrm{c} / \mathrm{s}$.) may be used with communications receivers to provide both additional selectivity
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Kit, either model . . . . ......... Assembled

C12.14.0
AERIAL TONER KITS. Model HT-I, HT-IG
Height 32it. sq. section 3ft. $\times 3 \mathrm{ft}$. at base (no stays required). Accessories available as extras
HT-1G Kit (galvanised) ©43.15.0
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Kit Total

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Ideal for use with valve FM Tuners.
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An inexpensire An inexpensive
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Kit $\mathbf{£ 9 . 2 . 6}$


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Funstions as oscillator or absorption wavemeter. With plug-in coils for continuous frequency coverage from $1.8 \mathrm{Mc} / \mathrm{s}$. to $230 \mathrm{Mc} / \mathrm{s}$.
Kit Ell.9.6 Assembled El4.9.6 Additional Plug-in Coils Model 341-U extend coverage down to $350 \mathrm{kc} / \mathrm{s}$. With dial correlation curves. 17/6.

## TRANSISTOR INTERCOM KITS

 Models XI-IU and XIR-IU 9 v . battery operated. Up to five remote stations can be operated with each Master. The Master unit can call any one. a combination, or all five Remote stations and any Remote station can call the Master| Kit | \&11.9.6 | Assembled |
| :---: | :---: | :---: |
|  | Mod | XIR-IU (Remote) |
| Kit | 4.9.6 | Assembl |

\& 17.9 .6
65.18 .0 See also Oscilloscope page

HI-FI MONO POWER AMPLIFIER KIT Model MA-I2


A compact Hi-fidelity power amplifier (including auxiliary power supply). 12 watts output. Wide frequency range and low distortion A variable sensitivity control is A varted enabling it to be used with an existing amplifier in a stereo phonic system. Other application includes sound reinforcement systems, transmitter modulators, for use wish tape recorders. Kit $\mathbb{C} 12.18 .0$ recorders.

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Kit $\mathbb{£ 2 5}$. $\mathbf{1 2 . 0}$
Also available assembled and finished $£ 33.4 .0$
$4 \frac{1}{2} \mathrm{in}$. VALVE VOLTMETER KIT

## Model V-7AU

The world's mose popular valve voltmeter with printed circuit and I per cent. precision resistors to ensure consistent laboratory performance. It has 7 , voltage tanges measuring respectively D.C. volts 400 aeak A.C. Re rist r.m.s. and ments from 0 peak. Resistance measurewith internal battery D.C megohms: weh internal battery. D.C. input resistance is 11 megohms and measurement has cenere-zero scale. Complete with test prod, leads and standardising bact 10 Pow H0-60 c/s. A.C. R. And Watts.
H.V. and R.F. Probes available as optional extras.

Kit $\mathbb{1}$ 13.18.6
Assembled f 19.18 .6
DECADE RESISTANCE BOX KIT
Model DR-IU. Range 1-99.9995 in in Secps. Ceramic switches throughout. Current rating from 500 mA . to 5 mA . according to decades in circuit. Polished wooden cabinet supplied complete.
Kit $\mathrm{flO} 18.0 \quad$ Assembled El 4.18 .0

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Assembled $\subset 10.18 .0$

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Kit 638.18 .0
Assembled $\mathbb{4 9}$.15.0 - Prices quosed are Mail Order Prices; retail Prices slightly higher.

WW-028 FOR FURTHER DETAILS

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 <br> <br> World-Leader}

The construction manual provided with the kit ensures successful assembly


Sin. WIDE BAND GENERAL-PURPOSE OSCILLOSCOPE, $10-12 \mathrm{U}$

- "Y" senslicivity 10 mV . r.m.s. per cm . at $1 \mathrm{kc} / \mathrm{s}$. Bandwidth $3 \mathrm{c} / \mathrm{s}-4.5 \mathrm{Mc} / \mathrm{s}$. - Frequency compensared input attenuator $\mathrm{Xl}, \times 10, \times 100$. T/B, $10 \mathrm{c} / \mathrm{s}-500 \mathrm{kc} / \mathrm{s}$, in 5 steps. - Two extra switch selected pre-ser sweep frequencies in T/B range. - T/B outpur approx. 10 v . peak ro peak. Builsin IV callbrator © Facilliey for "Z" axis modulation e Electronically stabilised power supply - Power req. 200-250 V . A.C., 40-60 c/s., 80 wates © Fused - Front panel, silver and charcoal grey - Cabinet, charcoal grey, size $81 \times 14 \times 17$ in. deep. Net weight 231b. 56-page construction and operation

Kit 635.17.6. Assembled £45.15.0
Attenuator and demodulator probes available as optional extras.

## 3in. PORTABLE GENERAL-PURPOSE

 SERVICE OSCILLOSCOPE, OS-2- Modern styling. lightweight and compact size, make this she ideal scope for service man, laboratory seehnician, amateur radio enthusiast or hobbyise " "Y" bandwidsh $2 \mathrm{c} / \mathrm{s}-3 \mathrm{Mc} / \mathrm{s} \pm 3 \mathrm{~dB}$. Sensisivicy $100 \mathrm{mV} /$ cm - Push-pull vertical and horizontal amplifiers - Wide range time-base generacor $20 \mathrm{c} / \mathrm{s}-200 \mathrm{kc} / \mathrm{s}$ ln four ranges. - Automatic lock-in synchronisation - Mu-meral c.r.t. shicid. Princed circule board construczion - Power req. $200-250 \mathrm{v} .50 .60 \mathrm{c} / \mathrm{s}$ A.C. 40 wates. Fused Frone panel silver and charcoal grey. Size 5 in . w. $\times 7 \mathrm{in}$. h. $\times 12 \mathrm{in}$. deep. Weight: 9alb.

Kit $\mathbf{2} 23.18 .0$ Assembled $£ 31.18 .0$

GENERAL-PURPOSE SERVICE RF SIGNAL GENERATOR, RF-IU - Ideal for the alignmene and erouble shooting of RF, IF and audio circuits - Large easy-ro-read dial - Pre-aligned coil and bandswitch assembly dial Pre-aligned coil and bandswitch assembly $\mathrm{Mc} / \mathrm{s}$. fundamentals up to $200 \mathrm{Me} / \mathrm{s}$ harmonics - 400 cycle audio signal with 4 V. outpue Dimensions $9 \frac{1}{2} \mathrm{in}$. wide $\times 6 \frac{1}{2} \mathrm{in}$. high $\times 5 \mathrm{in}$. deep.

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## See these and other Heathkit models in the FREE catalogue

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Versatile Recording facilities. So-easy-cobuild. Ousseanding performance for price.


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Complete your motoring pleasure with this small. compact. high outpus unit. Superb long and medium wave entertainment whenever you drive. for 12 v . positive or 12 v negative car earth systems.

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Total price kit (excl. LS) .... $\mathcal{L I 2 . 1 7 . 0}$ inc. P.T.
L/speakers and accessories availabie as extras.

NEW! PORTABLE STEREO RECORD PLAYER, SRP-I

- Compace, economical stereo and mono record playing for the whole family circuiery. Modern compact seyling - Detachable second loudspeaker gives optimum stereo effect - Automatic playing of $16,33,45$ and 78 rpm records - Suirease portabiliey - Two 8 in. $\times 5$ in. speakers - Controls: Volume. Balance and Tone. Dimensions: overall 27 in . wide $\times 14 \frac{\mathrm{in}}{} \mathrm{h}$. h gh $\times 7 \frac{1}{2} \mathrm{in}$. deep.

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500 and 1,000 Milliamperes.
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terminals.
RANGES: 1, 2.5. 5. 10 and 25 Amperes. $\mathbf{f 8 . 0 . 0}$
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T6

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Salary whilst overseas on scale $£ 690$ per annum rising to $£ 1,318$; start ing salary at age 21 years $£ 809$; at age 26 years or over, $£ 1,092$. Additiona! overseas allowances and shift pay

Apply:-The Secretary, S.R.C. Radio and Space Research Station, Ditton Park, Slough, Researc
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(1714)

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In addition to its present service, Independent Television will soon be entering the field of colour T.V. and U.H.F. Engineering; the future is therefore one of interest and development. The Authority requires young men with some experience in electronics to join its engineering staff; they will be based at Transmitting Stations throughout the United Kingdom and be engaged in the operation and maintenance of television transmitting and ancillary equipment.

SALARY-Commencing salary is in the range $\mathbf{6 9 5 0 - £ 1 , 2 2 5}$ per annum; subject to qualifications and satisfactory work, salary will progress to $£ 1,650$.
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HOURS-Duties are on a shift basis (day and evening) averaging 8 hours with a weekly average of 42 hours inclusive of meal breaks. The evening shifts finish shortly after programme close down at about midnight.

QUALIFICATIONS--Normally a Higher National Certificate or its equivalent would be required but as excellent residential training courses are provided, young men who have not reached this standard will be considered provided their background and experience is suitable.

If you would like to enter this interesting and developing field of engineering, please apply to :-

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Wireless World

ELECTRONICS, TELEVISION, RADIO, AUDIO

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- Great reduction in 'outage' time.
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Common features include magnetic deflection, electrostatic focus and aluminised screens.
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## quality equipment

The 12-way electronic mixer has facilities for mixing 12 balanced line microphones. Each of the 12 lines has its own potted mumetal shielded microphone transformer and input valve, each control is hermetically sealed. Muting switches are normally fitted on each channel and the unit is fed from its own mumetal shielded mains transformer and metal rectifier.


## FOUR-WAY ELECTRONIC MIXER

This unit provides for 4 independent channels electronically mixed without "spurious break through," microphony hum and background noise have been reduced to a minimum by careful selection of components. The standard $15-50 \mathrm{ohm}$ shielded transformers on each input are arranged for balanced line, and have screened primaries to prevent H.F. transfer when used on long lines.
The standard 5 valve unit consumes only 18.5 watts, H.T. is provided by a selenium rectifier fed by low loss, low field, transformer in screening box. The ventilated case gives negligible temperature rise with this low consumption assuring continuance of low noise figures.
20,000 ohms is the standard output impedance, but the noise pick-up on the output lines is equivalent to approximately 2,000 ohms due to the large amount of negative feedback used.
For any output impedance between 20,000 ohms and infinity half a volt output is available. Special models can be supplied for 600 ohms at equivalent voltage by an additional transformer or 1 milliwatt 600 ohms by additional transformer and valve.
The white engraved front pane! permits of temporary pencil notes being made, and these may be easily erascd when required. The standard input is balanced line by means of 3 point jack sockets at the front, or to order at the rear.


## THREE-WAY MIXER and peak programme meter, for recording and large sound installations etc.

This is similar in dimension to the 4 -Way Mixer but has an output meter indicating transient peaks by means of a valve voltmeter with a 1 second time constant in its grid circuit.
The meter is calibrated in dBs , zero dB being 1 milliwatt $-600 \mathrm{ohm}(.775 \mathrm{~V})$ and markings are provided for +10 dB and -26 dB . A switch is provided for checking the calibration. A valve is used for stabilising the gain of this unit. The output is 1 milliwatt on 600 ohms for zero level up to +12 dB maximum. An internal switch connects the outpur for balance, unbalance, or float. This output is given for an input of 40 microvolts on 15 ohms.
An additional input marked "Ext. Mxr." will accept the output of the 4 -Way Mixer converting the unit into a 7-W ay
controlled unit. This input will also accept the output of a crystal pick-up but no control of volume is available.
The standard input is balanced line by means of 3 point jack sockets at rear but alternative 2 point connectors may be obtained to order at the front or rear as desired.
The 8 valves and selenium rectifier draw a total of 25 watts.

$$
\text { P.P.M. for } 200-250 \mathrm{~V} \text { AC Mains ... .......... Price on application }
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Size $18 \frac{1}{\mathrm{t}} \mathrm{in}$. wide $\times 11 \mathrm{i} \mathrm{in}$. front to back (excluding plugs) $\times 6$ tin. high.
Weight 231b.
10/15 watt Amplifier with built-in mixers.
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Depicted here are typical examples of a range of Waterproof Connectors of unique design enabling electrical circuits to be connected or disconnected even when under water. The range caters for electrical loads of 440 volts a.c. 175 amps. down to the signal current levels associated with instrumentation and similar arrangements.

The basic design incorporates a patented principle referred to as "Watermate". Both plug and socket are moulded of a specially compounded neoprene rubber with unusually high insulation resistance and non-wetting surface. As mating occurs water, salt deposits, sand and other foreign matter are wiped from the sockets and ejected from a duct in the socket to form a leak-proof seal. The wiping action assures a dry connection at the moment of contact resulting in a leakage resistance of not less than 100 megohms WHEN MATED UNDER WATER.
They are pressure balanced and will not block up under high pressures. There are no glands or threads to seize up in water and the method of moulding to the associated neoprene jacketed cable provides an extremely robust and simple connector for both Military and Civil applications.
For full details of these Connectors and a new Underwater Reed Switch Assembly, please write or telephone to the Technical Sales Department.

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Aninstrument for specification performance in the form all Ferrograph instrumentsare guaranteed available for at least 10 years. Designed to a strict specification Built to endure
*Three operational speeds:
$633-1 \frac{7}{8}, 3 \frac{3}{4}$ and $7 \frac{1}{2}$.

| $633 \mathrm{H}-3 \frac{3}{4} .7 \frac{1}{2}$ and 15 i.p.s |
| :--- |
| $6 \frac{3}{2}$ and $7 \frac{1}{2}$ i.p.s. |

* Monitoring of recorded programme.
* Comparison by " $A-B$ " switching.
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米 3 watts undistorted output.


## Model 633 <br> 120 gns . <br> Model 633H <br> 125 gns.

Ferrograph



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SPECIAL OFFER: Complete HRO SET (Receiver, Coils \& Power Unit) for £ 30, plus 30/- carr
HRO-M-SETS available with UX type valves; secondhand cond., with 5 coil and power unit, $\mathbf{£} 20$ each, carr. 30-

CONVERTERS, Type $8 \mathrm{ar}, 24 \mathrm{v}:$ D.C., 115 v. A.C. at 1.8 amps 400 cycles' 3-phase, $86 / 10 /$ e each, post $8 /-$.

MARCON1 DEVIATION TEST SET, TF934: freq. 2.5-100 $\mathrm{Mc} / \mathrm{s}$ Can be extended to $500 \mathrm{Mc} / \mathrm{s}$. Deviation range $0-5,0-25$ and $0-75 \mathrm{Kc} / \mathrm{s}$. 835 each, carr. El .
MARCONI IMPEDANCE BRIDGE, TF-373: inductance $5 \mu h-100 \mathrm{H}$ in 5 ranges, capacity $5 \mathrm{pF}-100_{\mu} \mathrm{F}$ in 5 ranges, resistance $.05 \mathrm{meg} .-1 \mathrm{meg}$. power supply 250 v. A.C. $£ 37 / 10$ - cach, carr. $15 \%$
CT. 49 ABSORPTION AUDIO FREQUENCY METER: freq. range $450 \mathrm{c} / \mathrm{s}-22 \mathrm{Kc} / \mathrm{s}$., dircctly calibrated. Power supply $1.5 \mathrm{v} .-22 \mathrm{v} . \mathrm{D} . \mathrm{C}$. 812 10/- each, carr. 15-.
TICAN. Trans. /Receiver, same as ARN21, British made, STC, TR9171 complete with five 2C39As with associated valve-holders. As new price £25. Used condiion, $£ 15$, carriage $£ 1$.

RELAY UNITS. 2 high speed reluys $\mathrm{H} 96 \mathrm{E}, 1700+1700$ ohms, 1 changeover relay $14,000 \mathrm{ohms}, 1 \mathrm{CV} 455,100 \mathrm{ohms}$ and 1 meg . pot., etc. Mounted in box, 4 in . $\times 6 \mathrm{in}$. $\times 30 \mathrm{in}$., 30 - cach, 41 -post.

RECEIVERS. Type AR88D: freq. $540 \mathrm{Kc} / \mathrm{s}-32 \mathrm{Mc}$ s. \&45 each, carr. £2. AR88 SPEAKERS. New in cartons, metal case with black crackle finish $59 / 6 \mathrm{ca}$, post $7 / 6$.

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AIRCRAFT RECEIVEK ARR2. $235-258 \mathrm{Mc} / \mathrm{s}$. tunable, 24 v . D.C. inpur, £3 ca. 7/6 carr
HEWLETY P.ACKARD TYIE 400C: $115 \mathrm{v} . / 230 \mathrm{v}$. , input $50 / 60 \mathrm{c} / \mathrm{s}$. Freq. range $20 \mathrm{c} / \mathrm{s}-2$ Mcs. Voltage range: $1 \mathrm{mV}-300 \mathrm{v}$. in 12 ranges. Input impedar.ce 10 megohms. Designed for rack mounting, £30 each, carr. 15
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## SIGNAL GENERATORS:

MARCONI TF-144G: freq. $85 \mathrm{Kc} / \mathrm{s}-25 \mathrm{Mc} / \mathrm{s}$, internal and external modulation, power supplies $200 / 250$ V. A.C. (secondhand cond.), price £25 ea.; or available in transit case, complete with spares, in first class condition $\mathbf{£} 30$ ea., carr. on both 30 - ea.
TS155c/UP (as new) : price $\mathbf{\$ 7 5}$ each, carr. \&1
CT53. Freq. range $8.9-300 \mathrm{Mc} / \mathrm{s}$. with Calibration chart. Output $1 \mu \mathrm{~V}-100 \mathrm{mV}$. inicrnal square wave and sinewave modulation at $100 \mathrm{c} / \mathrm{s}$. external modulation $50 \mathrm{c} / \mathrm{s}-10 \mathrm{Kc} / \mathrm{s}, 230 \mathrm{v}$. A.C. Complete with chart, cte., price £27/10/• ea., carr. £1.
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MARCONI CT218: price $\mathbf{8} 65$ each, cart. 30 /
CT. 480 and $478: 1.3-4.2 \mathrm{Mc} / \mathrm{s}, \mathrm{F} . \mathrm{M}$. or A.M., price, 875 cach, carr. $30 /$

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GPO CANDLESTICK* TYPE TELEPHONE. Upright model with receiver, ideal novelty for converting to lampshade. Available any colour, $£ 5$, 10 - ea., post $7 / 6$. $^{\circ}$
TELEPHONE WIRE: 220 y ds., fi a roll, post $6 / \mathrm{F}$.
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TCS MODULATION TRANSFORMERS, 20 watts, pr, 6,000 C. T., sec. 6,000 ohms. Price 25 - post $5 /$ -

NIFE BATTERIES: 6 v .75 amps., new, in cases, $£ 15$ each, \&1 carr.; 6 v. 160 amps., new in cases, $£ 25$ each, £1/10/-cars.; 4 v. 160 amps., new, in cases, $£ 20$ each, $£ 1 / 10 /-$ carr.
L.R. 7 Cells, only 1.5 v. 75 amps., new, $\& 3$ cach, $12 /$ - cars

The above batteries are low resistance designed to give heavy surge for starting and can be stored for long periods without any effect to their performance.
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BATTERY NO. 4 (suitable for bells, etc.). $4 \frac{1}{2} \mathrm{~V}$., size $4 \frac{1}{2} \mathrm{in} . \times 6 \mathrm{in} . \times 2 \frac{1}{2} \mathrm{in}$, 5/- each. Post 3/-.
UNISELECTORS (ex equipment): 10 Bank, 50 Way, alternate wipe, £2/5/-ea. 6 Bank, 25 Way, alternate wipe, £2/2/6 ea. 8 Bank, 25 Way, £2/5/- ea. 6 Bank, 25 W ay, 22 ea. 4 Bank, $25 \mathrm{Way}, 35 /$ - ea. All the above


FREQUENCY METERS: IM-13 or BC-221; $125-20,000 \mathrm{Kc} / \mathrm{s} . ; £ 25$ each, carr. 15/-: TS174/U; 20-250 Mc/s. modulated, £45 each, carr. 15/-. TS323/UR; 20-450 Mc/s., £75 each, carr. 15/-. FR-67/U: This instrument is direct reading and the resuits are presented directly in digitalrorm. $100 \mathrm{Kc} / \mathrm{s}$. ing rate: Power Supply: $115 \mathrm{v}, 50 / 60 \mathrm{c} / \mathrm{s} ., \mathrm{E} 100 \mathrm{each}$, carr, \&1.
AMERICAN EQUIPMENT: Power supply, PP893/GRC 32A; Filter D.C. Power Supply F-170/GRC 32A: Cabinet Electrical CY 1288/GRC 32A. Antenna Box Base \& Cables CY 728/GRC, Mast Erection Kits, $1186 /$ GRC; Receiver type 278B; Directional Antenna CRD.6: Comparator Unit, CM.23; Directional Control CRD.6, 567/CRD and 568/CRD; Azimuth Control Units, 260/CRD

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MOTORISED ACTUATOR: 115 v. A.C. $400 \mathrm{c} / \mathrm{s}$. single phase, reversible, thrust approx. 3 inches complete with limit switches, etc. Price \&2/10/-each, postage 5/-(ex equipment).
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Actuator Type SR-43: 28 v. D.C. 2,000 r.p.m., output 26 watts, 5 inch screw thrust, reversible, torque approx. $25 \mathrm{lbs.}$, rating intermittent, price \&3 each, post. 5
28 v. D.C. 200 r.p.m. current consumption approximately 6 amps. Price £3 10 -, post $7 / 6$
FRACTIONAL MOTORS \&FANS: Low Inertia Motor 5UD/5361, Type 903, 24 v. input D.C., £2 10 - each, $5 /$ - post.
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A complete con:-
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## RADIO CLEARANCE（1965）LTD．

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ELECTROLYTIC CONDENSERS

| ${ }_{1 \mu} \mathrm{~F}$ | 25v |  | $1$ | ${ }_{6}$ | 50 F |  | 12v | ¢ ${ }^{\text {\％}}$ W．E． | S | $d$ 6 | 2，500 ${ }^{\text {F }}$ | 30v |  | ${ }_{7}$ | ${ }_{6}^{d}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $1 / \sim \mathrm{F}$ | 350 v | 1＂× W．E． | 2 | 0 | $50,1 \mathrm{~F}$ |  | 25 v | $1^{-} \times$W．E． | 1 | 6 | 2，500 2 F ． | 50 v |  | 10 | ${ }_{0}^{6}$ |
| $2 \mu \mathrm{~F}$ | 12 v | $\times$ ¢＂W．E． | 1 | 6 | $50 / \mathrm{F}$ |  | 50 v | 1＂x W．E． | 1 | 9 | 4，000 F | 25 v | $3^{\sim} \times 1^{\frac{1}{4}}$ T． 2. | 9 | 0 |
| $2 \mu \mathrm{~F}$ | 150 v | A＂×1＂W．E． | 1 | 6 | $50 \mu \mathrm{~F}$ |  | 275v | $11^{\prime \prime} \times 1^{\prime \prime}$ T．I． | 3 | 0 | 5，000 $\mu \mathrm{F}$ | 25v | $3^{\prime \prime} \times 1{ }^{\text {an }}$ T． 2. | 10 | 0 |
| $2{ }_{14} \mathrm{~F}$ | 275v | \％＂xt＂W．E． | 2 | 0 | $50 / \mathrm{F}$ |  | 350 v | $21^{\prime \prime} \times 1^{\prime \prime}$ T．1． | 3 | 6 | 10，000 0 F | 25v | $4^{* \prime} \times 1{ }^{\text {² }}$ T． 2. | 27 | 6 |
| $2 \mu \mathrm{~F}$ | 350 v | 1＂x W W． | 2 | 0 | $64 / \mathrm{F}$ |  | 450 v | 2＂$\times 1{ }^{\prime \prime}$ T．1． | 4 | 6 | $30,000 \mu \mathrm{~F}$ | 30 v | $4{ }^{\prime \prime} \times 2$ T． 2. | 45 | 0 |
| $2 \mu \mathrm{~F}$ | 500v |  | 2 | 6 | $100{ }_{k} \mathrm{~F}$ |  | 15 v | $1^{\prime \prime} \times$ W．E． | 1 | 6 | $32 \times 32 \mu \mathrm{~F}$ | 350v | $2{ }^{\prime \prime} \times 1$－${ }^{\prime \prime}$ | 4 | 6 |
| $4 \mu \mathrm{~F}$ | 25v | 1＂x＇WE． | 1 | 6 | 10012 F |  | 25v | $1^{\prime \prime} \times$ W．E． | 1 | 6 | $50 \times 50 \mu \mathrm{~F}$ ． | 350 v | $2^{\prime \prime} \times 11^{*+}$ T． 3. | 6 | 6 |
| $4 \mu \mathrm{~F}$ | 150v | F＂× W．E． | 1 | 6 | $100 \mu \mathrm{~F}$ |  | 50 v | 1＂× W．E． | 2 | 6 | $60 \times 100 \mu \mathrm{~F}$ | 275v | $2^{\prime \prime} \times 1{ }^{\prime \prime}$ T． 2 ． | 6 | 0 |
| $4 \mu \mathrm{~F}$ | 275 v | $1^{\prime \prime} \times$ W．E． | 2 | 0 | $100 \mu \mathrm{~F}$ |  | 100v | 1＂×年＂T．1． | 4 | 0 | $60 \times 250 \mu \mathrm{~F}$ | 350 v | $4^{\prime \prime} \times 1{ }^{\prime \prime}$ T． 2. | 12 | 6 |
| $4 \mu \mathrm{~F}$ | 350 v |  |  | 6 | 1004 F |  | 250v | 3＂×1 T．1． | 4 | 6 | $100 \times 100 \mu \mathrm{~F}$ | 150v | 3＂×1＂T． 3 ． | 4 |  |
| $4 \\| \mathrm{F}$ | 500 v | 1＂x＂＊W．E． | 3 | 0 | $100{ }_{\mu} \mathrm{F}$ |  | 350 v | $3^{\prime \prime} \times 1^{\prime \prime}$ T． 2. | 5 | 0 | $100 \times 200 \mu \mathrm{~F}$ | 275v | 4＂×13＂T． 2 | 9 | 6 |
| ${ }_{5 \mu \mathrm{~F}} \mathrm{~F} \mathrm{Rev}$ | 20 v |  | 2 | 6 | $100 / 2 \mathrm{~F}$ |  | 450 v |  | 7 | 6 | $150 \times 200 \mu \mathrm{~F}$ | 350 v | $4 " \times 1{ }^{\text {\％T } 2 .}$ | 12 | 6 |
| $5 \mu \mathrm{~F}$ | 50 v | W．E． | 1 | 6 | $125 \mu \mathrm{~F}$ |  | 500 v | $4^{\prime \prime} \times 1{ }^{\prime \prime}$ T． 2. | 9 | 0 | $250 \times 250{ }_{3} \mathrm{~F}$ | $325 v$ |  | 14 | 0 |
| $5 \mu \mathrm{~F}$ | 70 v | H＂× ${ }^{\text {an }}$ W．E． | 1 | 6 | $200 \mu \mathrm{~F}$ |  | 275v | $2^{\prime \prime} \times 1{ }^{\text {²m }}$ T． 2. | 6 | 0 |  |  |  |  |  |
| $6{ }^{\prime} \mathrm{F}$ Rev | 50v |  | 2 | 6 | $200 \mu \mathrm{~F}$ |  | 350v | $3^{\prime \prime} \times 1{ }^{\text {²F }}$ T． 2 ． | 7 | 6 |  |  |  |  |  |
| $8 \mu \mathrm{FRev}$ | 20v | 14＂× ${ }^{\text {a }}$ | 2 | 6 | $250 \mu \mathrm{~F}$ |  | 12 v | $1^{\prime \prime} \times$ W．E． | 2 | 6 | TE | MIN | TION CODING |  |  |
| $8 \mu \mathrm{~F}$ | 150v | $1^{\prime \prime} \times \frac{1}{6}$＂W．E． | 1 | 6 | $250 \mu \mathrm{~F}$ |  | 18v | $1^{\prime \prime} \times 1{ }^{\prime \prime}$ WE． | 2 | 6 | W．E．Wire | Ended |  |  |  |
| $8 \mu \mathrm{~F}$ | 275v | $1 \mathrm{~h}^{\prime \prime} \times \mathrm{t}^{\prime \prime}$ W．E． | 2 | 0 | $250 \mu \mathrm{~F}$ |  | 25v | 1＂x $\times$ W．E． | 3 | 0 | T．1．Tag e | ch en | of condenser |  |  |
| $8 \mu \mathrm{~F}$ | 350 v | 18＂× W．E． | 2 | 6 | $250 \mu \mathrm{~F}$ |  | 50w | 1＊＊${ }^{3}{ }^{\text {a }}$ W．E． | 4 | 6 | T．2．Single | end | g termination |  |  |
| $8 \mu \mathrm{~F}$ | 500 v | $1{ }^{\prime \prime} \times{ }^{\text {c／}}$ ，W．E． | 3 | 0 | $350 \mu \mathrm{~F}$ | $\ldots$ | 12v | 1＂x W．E． | 2 | 6 | T．3．Single | end | 8 termination |  |  |
| $10 \mu \mathrm{~F}$ | 6v | －W．E． | 1 | 6 | $350 \mu \mathrm{~F}$ |  | 25. | 11＂× T．1． | 3 | 0 | Tw | pros | xing |  |  |
| $10 \mu \mathrm{~F}$ | 50v | W．E． | 1 | 6 | $400 \mu \mathrm{~F}$ |  | 15vi | 1＂x W．E． | 3 | 0 |  |  |  |  |  |
| $10 \mu \mathrm{~F}$ | 150v | W．E． | 1 | 9 | $400 \mu \mathrm{~F}$ |  | 30 v |  | 3 | 6 | $8 \times 8 \mu \mathrm{~F}$ | 450v | 14＂$\times 1$＂T．1． | 4 | 0 |
| $10 \sim \mathrm{~F}$ | 300 v | 1＊＊${ }^{\text {\％}}$ W．E． | 2 | 0 | $400 \mu \mathrm{~F}$ |  | 50 v | $1{ }^{\prime \prime} \times \mathbf{1}^{\prime \prime}$ W．E． | 4 | 0 | $8 \times 16 \mu \mathrm{~F} \ldots$ | 459v | $2^{*} \times 1{ }^{\prime \prime}$ T．t． | ＋ |  |
| $16 \boldsymbol{\prime}$ F | 250 v | 14＂$\times$ W＂W．E． | 2 | 0 | $400 \mu \mathrm{~F}$ |  | $275 v$ | $4^{\prime \prime} \times 1{ }^{\prime \prime}{ }^{\prime \prime}$ T． 2 ． | 9 | 0 | $16 \times 16 \mu \mathrm{~F}$ | 275v | $2^{\prime \prime} \times{ }^{\prime \prime}{ }^{\prime \prime}$ T．1． | 4 | 0 |
| 164 F | 350 v | $1^{3 \prime \prime} \times 1$ W．E． | 3 | 0 | ${ }_{500} \mu \mathrm{~F}$ |  | 6 v | 1，W．E． | 1 | 6 | $16 \times 16 \mu \mathrm{~F}$ | 450 v | $2^{\prime \prime} \times 1^{\prime \prime}$ T．I． | 4 | 6 |
| $16 \mu \mathrm{~F}$ | 500 v | 13＂$\times 1$ W．E． | 3 | 6 | $500 \mu \mathrm{~F}$ |  | 15v | $1{ }^{\text {d }}$－X W．E． | 2 | 6 | $16 \times 32 \mu \mathrm{~F}$ ． | 275v | 13＂×1＂T．1． | 4 | 0 |
| $25 \mu \mathrm{~F}$ | 12 v | （1＂W．E． | 1 | 6 | $500 \mu \mathrm{~F}$ |  | 25 v | 14＊${ }^{3 / 3}$ W．E． | 3 | 6 | $32 \times 32 \mu \mathrm{~F}$ | 275v | $2^{*} \times 1^{*}$ T． 3. | 4 | 0 |
| $25 \mu \mathrm{~F}$ | 25v | ＊W．E． | 1 | ${ }_{9}$ | 1，000 $\mu \mathrm{F}$ |  | 15 v | $2^{*} \times{ }^{\prime \prime}$ | 3 | 9 | $50 \times 50 \mu \mathrm{~F}$ ． | 300 v | $2^{\prime \prime} \times 17{ }^{\prime \prime}$ T． 2. | 4 | 6 |
| 2514 F 30 | 50 v 6 v | $1^{\prime \prime} \times 1^{\prime \prime}{ }^{\prime \prime}$ W．E． | 1 | 9 | $1,000 \mu \mathrm{~F}$ |  | 18 v | 1＂$\times$ 尔＂W．E． | 3 | 9 | $50 \times 150 \mu \mathrm{~F}$ | 300 v | 34＂$\times 1$＂T． 2. | 6 | 6 |
| $30 \mu \mathrm{~F}$ | 10 v | 人＂WE． | 1 | 6 | $1,000 \mu \mathrm{~F}$ |  | 25 v | $1_{3^{\prime \prime}} \times 1^{\prime \prime}$ W．E． | 4 | 6 | $80 \times 250 \mu \mathrm{~F}$ | 275 V | 4 － | 9 | 0 |
| $32 \mu \mathrm{~F}$ | 150v |  | 2 | 6 | 1，500 $\mu \mathrm{F}$ |  | 25 v |  | 5 | 6 | $100 \times 100 \mu \mathrm{~F}$ | 275 v | $3^{\prime \prime} \times 1 . \mathrm{M}$ T． 2. | 12 | 6 |
| $32 \mu \mathrm{~F}$ | 350 v | $2^{\prime \prime} \times 1^{\prime \prime}$ T． 3. | 3 | 6 | 1，500 $\mu \mathrm{F}$ |  | 50 v | $3^{\prime \prime} \times 1^{* \prime}$ W．E． | 7 | 6 | $100 \times 400 \mu \mathrm{~F}$ | 275 v | $4 \frac{1}{\prime \prime}^{\times 1} \times 1{ }^{\prime \prime}$ | 13 | 6 |
| 32.4 | 450 v | 17＂× ${ }^{\text {² }}$ W．E． | 4 | 6 | 2，000 $\mu \mathrm{F}$ |  | 25 v | $3^{*} \times 1^{\prime \prime}$ T． 2. | 6 | 0 | $200 \times 200 \mu \mathrm{~F}$ | 300 v | $4^{\prime \prime} \times 1$ T． 2 ． | 12 | 6 |
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| 40 V | 82. |  |  |
| niov | $82 /-$ | － | 62. |
| 100 C | － | 48． | 62 － |
| 1501 | 82\％－ | 46. | 65／－ |
| 400 V |  | 4\％6 | 68－ |
| now | 584 | $50{ }^{\circ}$ | $7 \%$ |

ZENER DIODES


5\％10－WATT STUD MOUNTED


## ORY REED INSERTS


 two upertited by pertaineen magruet of

## BEEHIVE TRIMMERS



MICROWAVE DIODES
Cartrikse Type



 34．sec me／s．： $1 \times 3$ ：3u：（cvogsth）， 65 －

## AVALANCHE SILICON RECTIFIERS

## CATHODE RAY TUBES


 Hase Overall lemuth 7 （bl
Pねば…．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．40／＝


 Gi．Svi lean
VR10．


 PトにはF：．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．．． 55

 s12 F Jtage．Overall lerogl h lisin．
 $\qquad$
 Tubers．


## VALVES FOR EXPORT

Herw aty a few examplew tron our whock of over 2,500 items．

| OA2 | $31-$ | 2152 | $3 / 11$ | 6B4O 12／6 | 6an70T $2 / 8$ <br> 6VGGT $2 / 2$ |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 O8 | $5 / 9$ | 2182\％ | 184 | $6075-13 / 11$ |  |  |
| OH2 | $3 / 6$ | 3045 | $46 \%$ | 60w $49 / 9$ | 210A | 20.8 |
| Oc5 | $14 / 4$ | 3 V 4 | $3 / 2$ | 01084 9：2 | 311 A | 2819 |
| $0 \times 3$ | 47 | 5U4GB | 3.11 | 6.4 61 | 3\％A | 83． |
| O123 | $4 / 4$ | 5Yage | $3 / 6$ | 6．J5GT 2：6 | 2398 | 2\％ |
| 18：4t | $3 / 6$ | 5\％40T | $4 / 4$ | $6.17 \quad 512$ | 807 | 6／4 |
| 104 | 26 | GAK5 | $8 / 6$ | 61ARC 411 | 811A | $34 / 6$ |
| 1v2 | $3 / 9$ | －19L5 | 118 | $6 \times 7 \mathrm{CT} 4 / 4$ | 813A | 70／． |
| 17： | 202 | 6AQ5 | 2／8 | 6sL76T ${ }^{2} / 8$ | 8：911 | \％ |

TH：ABOVR PRICYS AHE FOR DIHELCT EXPORT 1．F．FOR DELIVERY TO OVERKFIAN ADDR KRS，OH TO THLE RUPPLEKg，FOH VALVFA TYPF，MARKLD AND EULK PACKRD，IN L．OT\＆OF 100 PER TYPE．
FULL EXPORT PRICE LIST AVAILABLE ON REQUEST
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R．S．T．VALVE MAIL ORDER CO．
146 WELLFIELD ROAD，STREATHAM，S．W．I6

| A731 |  | EY51 |  | 6AQ5 5／8 | 85. | 0ces 15／－ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| cle | 12／－ | EY81 | 1－ | 6AB6 ${ }^{\text {8／－}}$ | $90 \mathrm{AO} 45 /$ | 002 |
| Cblat | 15／－ | F．Y83 8／8 | QS150／36 | 6AB7 15／－ | 90AV 45／－ | OC23 7／6 |
| clils | 21／－ | EYrat $7 / 6$ | 20 | BAT6 4／－ | $30 C 1$ 12／－ | Ocus 16／－ |
| D． $1.14 \% 1$ | 4／＝ |  | ／45 | $6 \mathrm{B4O}$ 16／－ | $900025-$ | O＜－9 15／－ |
| 118） | 813 | 18\％41 8／－ | $20 /$ | $6 \mathrm{BA6}$ 4／6 | 500 25 － | O4：3is 11／6 |
| $1 \mathrm{CL} \times 10$ | 7／－ | E：\％80 5／－ | 48150／80 | 6 6ES $\quad 4 / 6$ | 15018．3 9／8 | OCH4 4／6 |
| D1\％01 | 3／－ | 6\％81 5／－ | $20 / 6$ | 6H16 7／－ | 15418：1 8／6 | $0 \mathrm{O} 454 / \mathrm{L}$ |
| 10／4． | 6／3 | （ITIC 17／6 | 481209 $2 / 8$ | filli $71-$ | sut 85－ | ut＇7t 4／6 |
| D113）318 | 89／ | 1：1／30 10／－ | QV03．12 | 6BK4 27／6 | 8033 35－ | 0078 8／－ |
| 11691 | 51－ | 1：732 9／6 | 10／－ | 6 BNB \％ $7 / 6$ | 807 71－ | OC74 6／－ |
| DK ${ }^{\text {che }}$ | $81-$ | 11284 10／－ | Q vol－7 12／6 | 6B97a 7／－ | 81130 ， | $0 \mathrm{Cz5}$ 6／ |
| ПК！ $\mathrm{Mi}^{\text {c }}$ | 7／－ | 17237 12／6 |  | 6ß1／7 8／6 | 813 35 | 0 06\％ |
| 13184 | 25／－ |  | $\text { Q VOK. } 20$ | 1888 $26 / 8$ | sbiai 13／6 | OC77 |
| 1）（A）${ }^{\text {a }}$ | $4 / 9$ |  |  | 6BW6 $71-$ | 5651 7／6 | OCN1 |
| 1 T 1.64 | $6 / 8$ | （1） | R10 15／－ | $\mathrm{HBW}^{6} 9 / 6$ |  |  |
| Thast | 71－ | $\begin{array}{ll}\text { KT61 } & 12 / 6 \\ \text { KTai } \\ 16 /-\end{array}$ | H17 8／－ | 6C4 $2 / 9$ | 56548 8／－ | OCXID |
| 131810 | 12／8 |  | H 18 119 | WCAt 5／－ | 3687 | Urwish 5／6 |
| DHSA10 | 301－ |  | R19 \％／－ | 1Сbm6 201 |  | OCs1 D3 |
| LLA819 | 30－ |  | RGJ／500 | $\begin{array}{rr}6 \mathrm{CLH} & 5 / 9 \\ 6 \mathrm{CL} / \mathrm{i} & 8 / 8\end{array}$ | $\begin{array}{ll}5691 & 25 / \\ 5749 & 10 /\end{array}$ | $\begin{array}{ll}\text { OCR2 } & 6 /- \\ 8 /-\end{array}$ |
| 19374 | 5－ | KT81(oLC) | 801－ |  |  |  |
| 1PVM | 6 |  | $\begin{array}{ll}\$ 130 & 25 /- \\ 8130 \mathrm{P} & 25 /-\end{array}$ | 6CW4 121－ | 376310 | 0 Cxib |
| $1 \mathrm{YPa}^{18}$ | $6{ }^{1}$ | 35／－ |  | 6104 15／－ | 58422 65／ | Own |
| Esmex | 12 － | KTW\％1 10／－ | \＄1P41 3／6 | 6DK，9／－ | 59463 10－ | OClim |
| EPrer | 176 |  |  | 6F23 13／6 | Gi05\％10／ | $0 \cdot 170$ \％ |
| F18：94？ | $22 / 6$ | $\begin{aligned} & \text { KTW62 } 10 \\ & \text { M1/ } 17 / 6 \end{aligned}$ |  | B．j54 | 13048 10／ | OC＇171 |
| E． $13 \times 0$ | 6／6 |  | STVE80／40 250 |  | ＋6059 18／－ | O＜2\％7／8 |
| 1：13：9 | $31 /$ | 878 15／－ |  |  | 60\％0． 6 | MX642 3／6 |
| E16ss3 | $1{ }^{-}$ | P（ewn 8／6 | 901－ | $\begin{aligned} & 8 J 70 \\ & 6670 \end{aligned}$ | H06\％12－ | $X$ Alol |
| 1815\％0 | 616 | Tlas 8／6 | SU2100 12／6 | 6 K 8 c | （1005：14／－ | XA111 3／8 |
| FBITRAS | 8／6 | P69\％ $7 / 6$ | $8 \mathrm{SU}^{15015} 12 / 8$ |  | 81063 | X A112 |
| Brluirl | $27 / 6$ | ［4xom 9／6 |  | 1897048867 | 6084 | X A125 |
| HCLLame |  | $\begin{array}{cc} \text { ruxs } & 5 / \\ \text { truxs } & 10 \% \end{array}$ | U19．U24／－24／－ |  |  | X A141 |
|  | 30／－ |  |  | tig．rial <br> f18LITIT |  | $\begin{array}{lll}\text { XAlat } & 8 /- \\ \text { XAlis } & 8 /-\end{array}$ |
| HCCS3 | 15／－ | 1 Cr＇mo | $\begin{aligned} & \text { U24 } \\ & \text { U23 } \end{aligned}$ |  | $\begin{aligned} & 6067 \\ & 60060 \end{aligned}$ |  |
| $131 \times 40$ | $9 / 8$ | 1＇crism | U125 |  | $\begin{array}{rrr}6 r+5 & 25 /- \\ 9003 & 9 /-\end{array}$ | 8Al43 8／－ |
| Hexst | $3 / 9$ | Pl｜at $7 /-$ | U191 13 | $\begin{array}{ll} \text { 6Y/94: } & 4 / 6 \\ \text { tX4 } & 3 / 6 \end{array}$ |  | TUBES |
| Fischer | $4 / 8$ |  | U7404 11／9 |  | \＄003 9／－ | ${ }^{1} 11^{11} 880 \%-$ |
| Suxas | 5／9 |  | （1AHC＊4） $5 / 6$ | 387 71－ | Silicon Rectifiers |  |
| Pachs | 5／－ | PCLLsa 8／6 |  |  |  | 3日P＇50，－ |
| $12 \mathrm{CO} \mathrm{S}^{8}$ | 7－ | P6：174 8／6 |  | $\begin{array}{ll}7 \mathrm{CS} & 15 \\ 768 & \text {／－}\end{array}$ | BY1 $1600^{5 / 6}$ | 20P1 |
| Ficlesal | 6／6 | 1＇5NR420／－ |  |  | Diodes | \＃EA1 |
| Ferse | 7－ | PLiN45DD |  | $\begin{array}{lr}7117 & 6 /- \\ 787 & 17 / 9\end{array}$ | ${ }_{\text {Transistors }}^{18131} 4$ | $318 P 8$3891 |
| E1315 | 11／－ |  | UCLAE 7／－ |  |  |  |
| 16CHst | 9／－ |  | $\begin{array}{ll}\text { UGLA3 } & 8 / 9 \\ \text { UL＋1 } & 8 / 9\end{array}$ | 7 71 | $\begin{array}{ll} 29102 & 4 / 3 \\ 20: 210 & 12 / 6 \end{array}$ | 5 SPl |
| स（＇14］ | 3／3 |  |  |  |  |  |
| ECLIns | 7／－ | PL，36 9／－ | $\begin{array}{ll}\mathrm{UL}+1 & 8 / 9 \\ \mathrm{ULS} 4 & 6 \%\end{array}$ | $\begin{array}{ll}111: 3 & \text { 2／－} \\ 12 A C & 101-\end{array}$ | $\begin{array}{cc} 2 n 210 & 12 / 6 \\ 203881 & 5 \end{array}$ |  |
| HCLSOO | 8／6 | P1asl $7 / 8$ | $\begin{array}{lll}\text { U184＊} & 6 /- \\ \mathrm{UY} 41 & 8 / 3\end{array}$ |  | 24338329401 | $\begin{aligned} & 881, \\ & 88 \mathrm{D} \end{aligned}$ |
|  | $6 / 3$ | PL84 816 | $\begin{array}{ll}\text { UY45 } & 5 /- \\ \text { VP4 } & 25 /-\end{array}$ | $\begin{array}{cc}\text { 12ALA } & 9 / 8 \\ 12 A H 8 & 30 /-\end{array}$ |  |  |
| ACLIma | $9 / 6$ | 1＇L500 13／6 |  |  |  |  |
| mitis | $8 / 9$ | $1 \times 14 /-$ | $\checkmark$ R105／30 5 － | $\begin{array}{ll}12 \text { ATG } & 4 / 6 \\ 12 A T 7 & 3 / 8\end{array}$ | $20414$ |  |
| ERO？ | 201－ | $\begin{array}{lll}1 \times 25 & 12 / 6\end{array}$ |  |  | 20.145209416 |  |
| krai | 7） | PYsy 8／6 | VR160／30 $5 /-$ | $\begin{array}{ll} 1 \because A \cup 7 & 4 / 9 \end{array}$ |  |  |
| EFP35 | 8／－ | 1Y33 8／6 |  | $\begin{aligned} & \text { 12AX7 } \\ & \text { l2BAA6 } \end{aligned}$ | ${ }^{2} 11417$ | CVISH7 504－Cvicar 35／－ |
| Firso |  | L．YH1 | V81 6／－ |  |  |  |
| Rivat | $6 / 3$ | 1－Y82 5／6 | Zatis 15／－ | $\begin{array}{lr} \text { I2haG } & 5 / 6 \\ 12 B E 6 & 5 / 3 \end{array}$ |  | Datisz |
| EF89 | 5－ | PY83 01－ | 78198754 |  | AC107 | $\text { DH3/41 }{ }^{90 /-}$ |
| 41 | $3 / 6$ | 1Y800 7／－ |  |  | $\begin{array}{ll}\text { AClo } & 7 / 6\end{array}$ |  |
| rirse | 2／8 | $\begin{array}{ll} \text { pysol } \\ \text { 1/230 } & 10 /- \end{array}$ |  | 12К B＇T $^{8 /-}$ |  | $\text { D } 13 / 10180 /-$ |
| 6F98 | 8 － |  | $\begin{array}{ll}\text { zROMU } & 15 /- \\ \text { OAZ } & 5 / 9\end{array}$ |  |  | E $550+/ \mathrm{B} / 16$ |
| E1P183 | 61 | $\begin{aligned} & 12 z 30 \\ & \text { QQvor/6 } \end{aligned}$ |  | $\begin{array}{ll} 20 p 4 & 19 /- \\ 2084 & 8 / 3 \end{array}$ | $\begin{array}{ll} \text { ACY19 } & 4 / 9 \\ \text { ACY } 20 & 4 / 9 \end{array}$ |  |
| RFF184 | 61－ | QQvos/io |  |  | $\begin{array}{ll} \text { ACY21 } & 4 / 9 \\ \text { ADD } 140 & 13 / 6 \end{array}$ | LClisio 35／－ |
| 1518404 | 21 － |  | O／4  <br> 1133 GT $8 / 6$ <br> 18  | $\begin{aligned} & 2584 \\ & 257504 \mathrm{~T} \\ & \hline 8 / 3 \\ & \hline \end{aligned}$ |  | WCR35 50／－M Whil $60 /-$ |
| EFFP6 | $10-$ | 30／－ | $\begin{array}{lr}2021 & 5 /- \\ 216218 & 201\end{array}$ | 25760T 8／6 | Alild $71-$ |  |
| 1：1190 | 76 | QQvo3／：0 |  | $30 \mathrm{Cl} \mathrm{S}^{\text {d }}$ 13／6 |  | $\begin{aligned} & \text { M Whi•2 } 601- \\ & 090 \quad 801- \end{aligned}$ |
| 1：L， 38 | 12／6 | QQvos/15 | $\begin{array}{ll}3.35 & 7 /- \\ 31328 & 40 /-\end{array}$ | $30 \mathrm{Cl} 7^{15} \quad 15$ | AF゙16 71－ | 4 80／－ |
| Eilat | ， |  |  | $\begin{array}{ll}30 \mathrm{~F} 5 & 15 / 8 \\ 30 \mathrm{FLI} & 18 /-\end{array}$ | OET571 5／－ |  |
| PISt | $8 / 6$ | $\begin{aligned} & \text { QQVON/ } 15 \\ & 105 /- \end{aligned}$ | 31828 3045 |  | 416T875 6／－ | －veriss |
| Plat | $8 / 6$ | QQVOAi／40 90 | ＋X1504 ${ }_{\text {95／－}}$ | $\begin{array}{ll} 30 \mathrm{L15} & 15 / 3 \\ 30 L 17 & 14 / 3 \end{array}$ | VKTश115／－ |  |
| H181 | 79 |  |  |  | NKT＊u4 4／－ | 50／－ |
| ELAL | $4 / 3$ | QQPo／10 $701-$ |  | $\begin{array}{ll}30 L 17 & 14 / 3 \\ 301919 & 13 /-\end{array}$ | NKT218 $7 / 6$ | 1138 A |
| E1s ${ }^{\text {ch }}$ | $7 / 6$ |  | 5Usia 4／－ | $30 \mathrm{PL4}$ 15／－ | NK＇2178／－ | A0 |
| F， | $7 / 6$ | Qs70／20 5／6 | 5V4G 8／－ | 30 PLI3 17－ | NKT214 6／－ | 38 A |
| Alusitio | $22 \%$ | 0875／20 5／6 | 5Y3CT 5／－ | $30 \mathrm{P}^{1,1 / 116 / 3}$ | NKTepsish | 35 |
|  | $81-$ $81-$ | QSitisi ${ }_{201-}$ | $\begin{array}{lr}51 / 46 & 8 / 9 \\ 4 / 301 / 2 & 13 /-\end{array}$ | $\begin{aligned} & 30 \mathrm{LiUT} 5 / 9 \\ & 35 \mathrm{w}_{4} \end{aligned}$ |  | ${ }^{16} 80$ |
| Fiasel | 16／－ | 0883／3 ${ }^{201-}$ | $\begin{array}{ll}\text { 4／30L2 } & 13 /- \\ \text { 6AK5 } & \text { 4／6 }\end{array}$ | $35 W 4 \quad 9 / 8$ |  | 88517A ${ }^{80}$ |
| 6LIMO | 20 － | Qs $0^{2} / 10$ 4／－ | ВАКG 8／6 | 60C\％5／8 | NKTa\％ 5 － |  |
| 1301 | 15 | Cssas／10 3／6 | 6 A1，5 3／－ | －0， | N1くT713 7／6 | 5173 |
| Fi3so |  | Q $8108 / 45$ | $6^{6 A 116} 3 / 6$ | 31／－ | 0CL6 20／－ | 46 |
| 184 |  | 151－ | 6AN8 10／＝ | 5）－ | $0 \mathrm{Cl} 1917 / 6$ | － |
| 1， 3 S2 | 25 | 8150／168／－ | 6AQ4 4i－ | ） | $00 \leq 0$ | 46／－ |
| All | $24$ | brand ne boxed Hour Expr er Service 6d．per Val | and <br> s Mail |  | 5at． 9 a．m．$=$ \＄at．1．30－2 Daily to 01.7690199 FOR LIST of | $\begin{aligned} & 5.45 \text { p.m. } \\ & 30 \text { p.m. } \\ & \text { allers } \\ & 1649 \\ & 2,000 \text { TYPE } \end{aligned}$ |

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 5 k ，etc．，per decade to 5 M ．Miniature： 0.3 W at 70 C ． $20 \%$ ： 1 M ，$+30 \%>\frac{1}{5} \mathrm{M}$ ． Horizontal（ $0.7 \mathrm{in} . \times 0.4 \mathrm{in}$ ．P．C．M．）or Vertical（ $0.4 \mathrm{in} \times 0.2 \mathrm{in}$ ．P．C．M．）mounting $1 /-$ each． Submin． $0: 1 \mathrm{IW}$ at $70^{\circ} \mathrm{C} \quad 20 \% \leq 1 \mathrm{M}, \pm 30 \%>1 \mathrm{M}$ ．Horizontal（ $0.4 \mathrm{in} . \times 0.2 \mathrm{in}$ ． P．C．M．）or Vertical（ 0.2 in ．$\times 0.1 \mathrm{in}$. P．C．M．）mounting，lod．each．
RESISTORS（Carbon film）：High stability，very low noise， 1 W at $70^{\circ} \mathrm{C}$ ． Body 1 in．$x$ in．Values in each decade： $10,11,12,13,15,16,18,20,22,24,27$ $1.2 \mathrm{M}, 1.5 \mathrm{M}, 1.8 \mathrm{M}, 2.2 \mathrm{M}, 2.7 \mathrm{M}, 3.3 \mathrm{M}, 3.9 \mathrm{M}, 4.7 \mathrm{M}, 5.6 \mathrm{M}, 6.8 \mathrm{M}, 8.2 \mathrm{M}, 10 \mathrm{M}, 10 \%$ 2d．each． SILICON RECTIFIERS： 0.5 at $70^{\circ} \mathrm{C}$ ． 400 P．I V ，2／9． 800 P．I．V．，3／3，1，250 P．I．V． 3／9． 1500 PIV 4／．
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[^8]

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#### Abstract

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of Communtcations. Untued Press Internat!onal. 8. Bouverle St.. E.C.j.
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Assistant Executive Engineers: G.C.E. (or equivalent) pass in English language, and one of the following: H.N.D., in Electrical or Mechanical Engineering or Applied Physics; pass in (or exemption from) Parts 1, 2 and 3 of the examinations of I.E.E., or I.Mech.E.; a pass in (ormprion Sections $A$ and $B$ of the IER.E. examinations; a pass in (or exemption from) Parts 1 and 2 of the examination of the Council of Engineering Institutions, in subjects acceptable to one of the Institutions named above

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Assistaries increased for officers serving in London. Non-contributory pension. Promotion prospects

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Applications for both posts from well qualified older candidates will be considered. (Reference: S 353)

APPLICATION FORMS are obtainable from the Secretary, Civil Service Commission, Savile Row, London, W.I. Please quote appropriate reference.


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Electronic

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Gearing for 1968's big expansion programme of the Micro Switch and Meter Division has meant rapid promotion for many Honeywell engineers . . and has created some gold-plated opportunities for newcomers to our sales teams. Attractive positions exist in the London, Manchester and Birmingham areas, so if you are keen, qualified and looking for the rewards that match your talents read on:

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If you have about three years selling experience, preferably in the Electro-mechanical or Electronic field, backed by an H.N.C. or equivalent you can now gain your full reward. Become an Outside Sales Engineer with our successful London or Birmingham Sales Force.

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This is a key position and the successful occupant will be treated accordingly. He will be a qualified Engineer with experience in the design and/or application of Industrial Sequence Controls, including Solid State Devices. As Senior Engineer he will assist the Product manager and be responsible for planning new products with liaison between Sales and Factory from concept to manufacture. This London Head Office based position is permanent, pensionable and the successful applicant will be given assistance with relocation costs if necessary.

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These are interesting permanent staff situations, and the salary paid will be commensurate with ability and experience.

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TEST ENGINEERS

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All applications treated in strictest confidence.

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## Electronics Engineer

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Apply with details of career to
Staff Appointments Officer, Ferranti Ltd.,
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Training Courses are arranged for suitably qualified men. H.N.C. Electronics, City \& Guilds Final or equivalent standard required. Men from Forces with radar experience welcome.
Knowledge of electronic or electro-mechanical equipment necessary. Good Pension and Bonus Plan in operation.
Please write for Application Form to The Personnel Officer.
NCR, 1000 North Circular Road, London, NW2, quoting Publication and month of issue.

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GATESHEAD \& DISTRICT HOSPITAL MANAGEMENT
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THE Royal Free Hospital requires an Elecironics En Rineer. The successfal candidate uill be part of a team of electromes engineers but will have special
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Applicants should state when they will be available for interview.


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and
A SENIOR LECTURER in

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Applicanis should hold a good honours degree in Electrical Engineering or in Physics and should be corporate members of an appropriate professional institution. They should have had relevant industrial or research experience and preferably teaching experience.

Whilst applications will be considered from candidates specialising in any branch of electronic engineering, it is particularly hoped to make one appointment in Radio Communications (Ref. E.11) and the other in Digital Systems (Ref. E.12).

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It is hoped that the persons appointed to these posis will engage in research work, for which opportunities and facilities can be made available
Further particulars and application forms may be obtained from the Clerk to the Governing Bodv, with whom applications should be returned as soon as possible.

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[^1]:    * Newmarkei Transistors Lid

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