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# Radiation at Very Low Frequencies 

IN the present state of the art there are often times when the study and practice of radio may seem to be a hard grind. "Under the sun there's nothing new" and work consists of mastering known principles and taking advantage of improved materials to refine existing designs. But our subject is never dull for long. The field is wide, the labourers are many and new finds in old workings are constantly turning up to fire the imagination and shift the centres of interest. Hertz worked with centimetre waves, and electromagnetic waves at frequencies which would be audible if air were the propagating medium were widely used in the early days.

From these extremes of frequency interest tended later to converge to the intermediate values which gave greater bandwidth and traffic-handling capacity, lower capital cost for transmitters and greater running economy. In recent years the interest, as far as frequencies are concerned, has again become divergent; to centimetric waves for narrower beams and higher discrimination in radar, and to very long waves for special navigation and communication systems and for geophysical research.

In passing, it is of interest to recall the criticism which was levelled at the Post Office for continuing in their policy of building a high-power long-wave station at Rugby, when the new short-wave beam stations were showing every promise of making it obsolete. A similar situation is now arising between the fact of the Commonwealth telecommunications project and the promise of satellite communications.

Although the bulk of the world's communication traffic took to the new short-wave bypass, we have reason to be grateful, now more than ever before, that the Post Office, and authorities in other countries, did not allow the old low-frequency roads to fall into disrepair. The services of Rugby GBR ( $16 \mathrm{kc} / \mathrm{s}$ ) and Criggion GBZ ( $19.6 \mathrm{kc} / \mathrm{s}$ ) are much in demand all over the world not only for communication but for time and frequency comparison and, because of their wide dissemination and high stability, for navigational purposes. New stations, working on similar frequencies with powers of 1 MW , have been built in the U.S. for communication with submarines.

At " extremely low" frequencies" there is worldwide interest and much international co-operation in the investigation of electromagnetic phenomena of terrestrial and extra-terrestrial origin. Lightning flashes produce transients with e.l.f. components of considerable magnitude, and correlation has been established with perturbations of the geomagnetic field. But by far the most interesting development in the e.l.f. and v.l.f. ranges has, in our opinion, been the investigation of "whistlers," which can be heard under favourable conditions and in high latitudes by connecting an aerial directly to the input of an audio amplifier. These have been
shown to have their origin in lightning flashes and to owe their peculiar sound to the different rates of propagation with frequency of the components of the original pulse waveform in following the lines of force of the earth's field through a dispersive medium. This gave the first intimation, later confirmed by rocket exploration, of the existence of belts of ionized particles extending to heights of many earth radii.

While much new knowledge has been collected by patient observation and analysis of natural phenomena, more rapid progress would be made if controlled impulses and burst of radiation at specific frequencies could be launched, particularly from the polar regions. But the resources of conventional radio engineering find themselves hard pressed in trying to emulate lightning as far as power and radiator dimensions are concerned. Many ingenious suggestions have been put forward, for example, that v.l.f. signals might be superimposed on the electricity grid system. Unfortunately, grid systems are seldom to be found where the scientists want them. An experiment, reported on page 614 of this issue, makes use of a $10,000 \mathrm{ft}$. aerial wire suspended from a helicopter in what must surely be the biggest mobile-marine transmitting station ever devised. Even so, the aerial is no longer than that used at the Marconi transatlantic receiving station at Towyn in the early 1920s.

To our way of thinking, the most ingenious suggestion put forward so far is that due to M. G. Morgan ${ }^{2}$, of the Thayer School of Engineering, Dartmouth College, Hanover, New Hampshire, U.S.A. This is to use the whole of Deception Island in the South Shetland Islands as a slot aerial, using the surrounding sea as the ground plane conductor. The island is the rim of a submerged volcano and has one gap leading to an inner lagoon. An expedition of the Falkland Islands Dependencies Survey is now on its way with equipment to check that the rock formation is a good enough insulator and to measure the impedance of the "slot," which has a calculated resonant frequency of $5 \mathrm{kc} / \mathrm{s}$. If the answers are favourable it is hoped to try to energize the "aerial" at points opposite to the lagoon entrance by current entering the sea on each side of the saddle, and thus to launch waves upwards into space.

We wish every success to this enterprise, which has shown that the older arts of radio are still as capable of stirring the pulse as those of its younger offspring, electronics.

[^1]
# Practical Aerial Measurements 

INVESTIGATING PERFORMANCE OF AERIALS BY MEANS OF SCALE MODELS

CUALCULATING the radiating and/or pick-up properties of any aerial is a tedious process and does not take into account its environment once in situ. For this reason alone many engineers and organizations who specialize in aerials use scale models operated at very high frequencies for determining performance under working conditions. Practical experiments of this nature are fascinating to say the least, and may be carried out by anyone with a few simple items of equipment plus a little knowledge of basic principles of radiation.
The aerial "table" described in this article was constructed mainly for investigating the performance of aerials used for amateur radio transmitting and for live demonstrations during lectures.

For this article the reader is asked to assume that all reference to the performance of an aerial is in the radiating sense. The characteristics and performance of an aerial during reception are for all practical purposes reciprocal and identical. It is easier to think and talk in terms of radiation.

In general the object of the aerial designer is to produce standing waves on the radiating elements of an aerial so that radiation takes place in a known manner and where required. For example, resonance may be obtained with a quarter wavelength of wire if one end is on a theoretically infinite conducting sheet (e.g. copper) placed perpendicular to the wire. The reflection from the sheet then produces an image of the quarter-wave wire, as, say, a pencil standing on a mirror appears to be carried on into the mirror. The system resonates as freely as a halfwave length of wire but with halved radiation resistance.

The diagram of Fig. 1 shows how reflection from the earth takes place and how the angle of reflection


Fig. 1. Direct and reflected radiation from aerials adjacent to earth.
is determined by either the height or the length of the aerial above ground. As the height or length is increased the relative phase angle between the direct and reflected wave will vary, thus modifying the strength of radiation in different directions.


Fig. 2. Radiation field from vertical aerial; (a) in horizontal plane, (b) in vertical plane.

Radiation from any aerial system is determined by resonant length, or by the physical assembly of a number of radiating and/or reflecting and directing elements (beam aerials). For example, the radiation from a single vertical $\lambda / 4$ radiating element is uniform for all directions around the aerial (horizontal plane). In the vertical plane the field strength varies as the angle from the ground increases, being maximum at ground level and zero in the true vertical direction as shown in Fig. 2.
In free space, clear of any conducting or semiconducting obstacles, an aerial generally behaves according to accepted theories, but in the presence of other conductors and earth, the true radiation pattern may become very much modified. Since the height of an aerial above ground will never be very great (compared with an imaginery position in free space) the earth and its conductivity always play a great part in forming the actual radiation pattern. The term "earth" or "ground" applies also to aerials on aircraft since the body of an aircraft may be regarded as "earth" and usually has considerable effect on radiation.
The Theory of Similar Aerials.-The difficulties involved in plotting, other than the ground radiation

[^2]field of aerials operating in the lower portion of the radio frequency spectrum, make it almost impracticable to attack the problem. It becomes necessary, therefore, to devise some other system, other than calculation, for investigating true radiation patterns. The use of scale models is not an entirely new idea, for the method was used before the war by the B.B.C. and others, although the scale reduction of the models was limited owing to lack of suitable sources of r.f. power for very high frequencies'.
From a practical point of view a half-wavelength aerial resonant at say 10 centimetres $(3,000 \mathrm{Mc} / \mathrm{s})$ would behave in the same way as an aerial of similar type operating at 10 metres ( $30 \mathrm{Mc} / \mathrm{s}$ ) or even at 100 metres ( $3 \mathrm{Mc} / \mathrm{s}$ ), provided all other conditions were maintained, i.e. height above ground, conductivity of ground, dielectric constants and/or conductivity of materials used for the manufacture of the aerial and the effects of surrounding objects. Most of these can be simulated in scale models and the results of small differences ignored, although at certain very high frequencies the construction of an accurate model does become more difficult.

The problem of suitable methods of feeding the aerial also arises and at frequencies above about $200 \mathrm{Mc} / \mathrm{s}$ the radiation field, particularly from balanced aerials such as the dipole, may become asymmetrical owing to inaccurate matching of aerial to feed system. (Some asymmetry may be noticed in the dipole pattern of Fig. 9.)

The method does, however, provide a means of establishing the pattern of the radiation strength in


Fig. 3. The $3,000-\mathrm{Mc} / \mathrm{s}$ model aerial testing table (scale $4 \mathrm{~mm} / \mathrm{ft}$ ) used by the author. The aerial shown in lower centre is a scaled-down 6 -element, stacked array for a normal frequency of $30 \mathrm{Mc} / \mathrm{s}$.


Fig. 4. Typical $3,000-\mathrm{Mc} / \mathrm{s}$ Klystron oscillator, the Type CV67.
both the vertical and horizontal planes, as well as the effect of surrounding buildings and similar structures. Measurement of gain in beam aerials may be carried out and the effect of physical modifications rapidly ascertained. Many commercial concerns use the scale model method for determining radiation patterns, etc., the U.S.A. Bureau of Standards for example have a "Model Antenna Testing Range"; many aircraft companies use scale model aerials in conjunction with model aircraft to the same scale ${ }^{3.3}$.
Suitable Frequencies.-The writer's earliest experiments were carried out with an operating (model) frequency of $144 \mathrm{Mc} / \mathrm{s}$ and many types of aerials commonly used for the radio amateur bands of 1.8 , $3.5,7$ and $14 \mathrm{Mc} / \mathrm{s}$ were scaled down for measurement of gain and radiation. More than 30 model aerials were constructed, most of the tests being carried out over normal earth. Attempts were later made to construct a "table" which could be used indoors but even at a frequency of $800 \mathrm{Mc} / \mathrm{s}$ reflection from walls, etc., was found to be troublesome and imposed considerable restriction on accuracy.
A Practical $\mathbf{3 , 0 0 0 - M c} / \mathrm{s}$ Model Aerial Table.-To avoid inaccuracies due to unwanted reflection and to allow time for the complete formation of the electric field from a radiating aerial, the field strength meter and its aerial should be at least five wavelengths distant from a radiating aerial under test. The higher the frequency the shorter becomes the free space wavelength and at $3,000 \mathrm{Mc} / \mathrm{s}$ it becomes possible to construct equipment suitable for indoor demonstration, as well as for reliable estimation of field patterns.

The aerial table shown in Fig. 3 was constructed in three sections to form a total area 4 ft 6 in square. Plywood, backed with thin aluminium sheet was used for this and forms an effective ground plane or artificial earth. The table contains a section of a typical residential area modelled at $4 \mathrm{~mm} / \mathrm{ft}$ and is constructed mainly from materials that simulate as near as possible those normally used for full-size structures.

Mounted directly under the centre of the table is a $3,000-\mathrm{Mc} / \mathrm{s}$ Klystron oscillator (CV67), complete with power supplies and modulator, which provides r.f. power for the model aerials, the r.f. output being taken to an aerial via a specially designed rotating coupling flush with the top of the table. The Klystron and its associated assembly can be turned through $360^{\circ}$ by means of a slow-


Fig. 5. Klystron oscillator, power supply and modulator used by the author. $V 1=250 \mathrm{~V}$ rectifier; $V 2=1,250 \mathrm{~V}$ rectifier; $V 3=E F 91 ;$ Klystron (see text); $T 1=$ modulation transformer, ratio $/$ to $\mid ; T 2=$ tapped a.f. transformer.
motion drive and hand wheel located at the side of the table.

Klystron oscillators, such as the CV67 shown in the photograph of Fig. 4, may be obtained from Government surplus supply sources. Other Klystrons suitable for this work include the CV36 ( $2,850 \mathrm{Mc} / \mathrm{s}$ ), CV116 ( $3,540 \mathrm{Mc} / \mathrm{s}$ ), CV23 ( $3,300 \mathrm{Mc} / \mathrm{s}$ ) and the CV238 $(3,050 \mathrm{Mc} / \mathrm{s})$. The latter two are low voltage types and require 250 volts between anode and cathode and 100 to 200 volts for grid bias and the target anode. The others operate with up to 1,500 volts on the target anode and take an average current of $5 \mathrm{~m} / \mathrm{A}$. These Klystrons have fine and coarse tuning adjustment plugs fitted around the resonant cavity and provide an r.f. output of 100 to 150 milliwatts.

The Klystron oscillator and modulator circuit used by the writer is shown as Fig. 5. Component values may have to be modified slightly, although those given should be close enough to enable the circuit to function reasonably efficiently.

On either side of the table and on a line through the axis of the rotating coupling, two supports carry an inverted "V" frame. This is based on an arrangement used by the U.S.A. Bureau of Standards, which permits the field-strength meter and its associated receiving aerial, which are mounted at the apex of the frame, to move through an arc of $180^{\circ}$. This arc extends from ground level on one side of the aerial to ground level on the other.

The frame may be locked in any position so


Fig. 6. Receiver ond field-strength meter used for oerial measurements at $3,000 \mathrm{Mc} / \mathrm{s} . \quad V I=E F 9 I, V 2=E L 9 I ; T I=1$ oudspeaker transformer; T2 = I to I ratio o.f. transformer; M1 meter rectifier or diode.
that a horizontal radiation pattern can be plotted for any vertical angle with respect to earth.
The Receiver and Field Strength Meter.-The $3,000-\mathrm{Mc} / \mathrm{s}$ receiver consists of a tunable resonant cavity and a crystal diode from which the demodulated signal is fed via an audio amplifier to a variable attenuator. Various surplus $3,000-\mathrm{Mc} / \mathrm{s}$ radar tuning units may be adapted for this purpose and 120 used in conjunction with an amplifier circuit similar to that of Fig. 6. The audio signal level is indicated by a sensitive meter calibrated in 15 equal divisions, this being equivalent to the ruling of Chartwell Polar Co-ordinate paper, Sheet No. 4001. The attenuator is directly calibrated in decibels. The Klystron modulating signal is about $1,000 \mathrm{c} / \mathrm{s}$ so that a loudspeaker may be connected in place of the meter for practical demonstration purposes.

As an alternative to the Klystron for a source of r.f. power there are many conventional valves that will oscillate at a $1,000 \mathrm{Mc} / \mathrm{s}$ or more. If lower frequencies are used, however, the model aerials will be correspondingly larger but still suitable for openair operations. Only a few milliwatts are needed to drive the aerials and ensure full-scale meter readings, so that radiation will be quickly absorbed and will not travel very far if experiments have to be made out of doors.
Plotting Radiation Patterns.-Clear surroundings are very necessary and the equipment should be carefully sited several wavelengths (at operating frequency) away from other conductors such as electrical wiring, metal window frames and pipes. Gain measurements may be made with respect to a "s standard dipole," the radiated power from this being taken as a reference and regarded as 0 dB . The importance of clear surroundings will be appreciated when attempts are made to simulate by means of models the conditions imposed by the presence of buildings and other structures. Twodimensional "solid" patterns of a radiation field are more interesting than the conventional polar diagrams, especially for demonstration and from these a very clear conception of radiation directivity may be obtained. These two-dimensional shapes are produced by taking patterns from an aerial every few degrees in the vertical or horizontal plane, whichever is applicable, each pattern being drawn on thin cardboard, cut out and glued together in the correct order as shown by the example of Fig. 7. A selection of patterns taken from model aerials on the testing table described in this article are shown in Figs. 8, 9, 10 and 11. The model aerials are shown in the photograph of Fig. 12.

Fig 7. Solid radiation pattern of two vertical half-wave aerials spaced one wavelength opart and fed in phase.



Fig. 8. Radiation potterns of two vertical half-wave aerials as described in Fig. 7 and measured at three different vertical angles.


Fig. 9. Gains of 3- and 8-element Yagi aerials compared with a half-wave dipole measured at $3,000 \mathrm{Mc} / \mathrm{s}$.


Fig. IC. Radiation patterns and gains compared with halfwave dipole of single-turn helix with ground plane and ten-turn helix with ground plane.


Fig. 11. Vertical radiation pattern of a scale model 2element " Quad" aerial.


Fig. 12. Some of the scale model aerials used for the measurements described in the text.

Models provide a unique system for testing and estimating the performance of aerials for all the lower radio frequencies up to at least $30 \mathrm{Mc} / \mathrm{s}$ and offer an excellent method of demonstration during lectures. The writer has used over 100 different models and several aerials suitable for amateur radio transmitting have been developed with the aid of the testing table. Using scale models, a number of enlightening experiments were recently carried out to determine the radiating properties of "loaded aerials" suitable for mobile operation on the $1.8-$ and $3.5-\mathrm{Mc} / \mathrm{s}$ amateur radio bands and for certain applications at frequencies around $30 \mathrm{Mc} / \mathrm{s}$.

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## BOOKS RECEIVED

The Radio Amateur's Handbook, 1960 edition, compiled by the American Radio Relay League, West Hartford, Connecticut, U.S.A. First published in 1926 and now widely recognized as the standard manual of amateur radio communication. Chapters devoted to construction of equipment have been extensively revised to include the latest s.s.b. and radio-teleprinter techniques for amateur use. Several chapters are devoted to basic principles clearly explained. Pp. 584 plus 32 pp . of tabulated data on valves and semiconductors with over 1,300 illustrations. Obtainable from leading technical booksellers, also from Radio Society of Great Britain, New Ruskin House, Little Russell Street, London, W.C.1; price 32s 6 d ( 34 s by post).

Mullard Reference Manual of Transistor Circuits. More than 60 circuits of audio amplifiers (including p.a.), hearing aids, simple radio receivers, oscillators, switching circuits, stabilized power supplies, etc., with component values, coil winding data and constructional hints. Prepared by the staffs of Mullard Research Laboratories, Applications Research Laboratory and Semiconductor Measurement and Application Laboratory. There are nine introductory chapters on transistor characteristics, equivalent circuits, bias and stabilization and limiting values. Pp. 308; Figs. 241. Mullard, Ltd., Torrington Place, London, W.C.1. Price 12 s 6 d ( 13 s 6 d by post).

British Semiconductor Guide-1960. Characteristics of 963 semiconductor devices including transistors, diodes and rectifiers and photosensitive devices, with introductory articles on considerations determining the use of valves and transistors, and on transistor parameters and specifications. Pp. 66. Heywood \& Co., Ltd., Drury House, Russell Street, London, W.C.2.

Einführung in die Microwellen-Electronik. Teil II Lauffeld-röhren by Drs. Kerner Kleen and Klaus Pöschl. Mathematical treatise on the principles and operation of travelling-wave tubes in general and special types, including backward-wave oscillators and amplifiers. Pp. 192; Figs. 125. S. Hirzel Verlag, Birkenwaldstrasse 185, Stuttgart N.

## 100-kc/s Crystal Marker

EMBODYING A NEON MODULATOR

## FOR HARMONIC RECOGNITION

By<br>F. G. RAYER*

THE use of a $100 \mathrm{kc} / \mathrm{s}$ crystal marker, to provide frequency checking harmonics at $100 \mathrm{kc} / \mathrm{s}$ intervals, is quite well known. Such a marker is most often employed in conjunction with a communicationstype receiver to give exact calibration, or check points at $100-\mathrm{kc} / \mathrm{s}$ intervals. Amateur transmitters can employ a marker of this kind to locate the band edges, or for calibrating or adjusting a transmitter VFO. When used with a reasonably accurately calibrated receiver, this method is an acceptable one for frequency determination with amateur transmitting equipment.

The usual simple type of crystal marker has the disadvantage that it is not always possible to distinguish its signal from other signals, especially when tuning to higher harmonics. Even with the receiver aerial removed, other carriers are heard on the crowded short wave bands, so that the crystal marker's signal cannot always be located immediately and without possibility of error. To avoid this, the circuit shown in Fig. 1 is provided with a

- Amateur Radio Station G3OGR.


The completed crystal marker showing the top plate screwed to the chassis box.
neon " modulator." This strikes two or three times a second, pulsing the 6BA6 control grid negative, so that the marker signal can always be identified at once. Such a "modulator" does not produce sidebands which would confuse tuning. In addition, the 6BA6 acts as an harmonic amplifier, so that the marker can be used up to about $30 \mathrm{Mc} / \mathrm{s}$.

The crystal oscillator stage consists of a 6AM6, with a $100-\mathrm{pF}$ pre-set capacitor for frequency adjusiment. The degree of adjustment is quite small, but it allows the crystal frequency to be checked against the B.B.C. $200 \mathrm{kc} / \mathrm{s}$ Light Programme, or the National Physical Laboratory standard frequency transmissions on $2.5 \mathrm{Mc} / \mathrm{s}, 5 \mathrm{Mc} / \mathrm{s}$, and $10 \mathrm{Mc} / \mathrm{s}$. Other valves, such as the CV138 or EF91, etc., may be used in this stage. The type of neon is of little importance, provided it will operate with about 150 V to 250 V . Various other valves could also be utilized instead of the 6BA6.

The unit illustrated was built into a universal chassis $4 \mathrm{in} \times 7 \mathrm{in} \times 2 \mathrm{in}$, with back plate, obtainable from Home Radio (Mitcham) Ltd. The layout employed is shown in Fig. 2. To simplify work the


Fig. i. Circuit diagram for the $100-\mathrm{kc} / \mathrm{s}$ crystal marker.
marker was built on the $4 \mathrm{in} \times 7$ in back plate and when completed it was fitted on to the chassis box. For lining up with other equipment it was most convenient to have the heater switch, supply leads, and output at one end of the chassis box. If the marker is connected to its supply only when used, no switch is required, but if it is permanently connected, the h.t. circuit has to be switched, as in Fig. 1, to stop the neon striking.

It is essential that the $100-\mathrm{pF}$ pre-set capacitor is isolated from the chassis. It is also necessary that the $0.1-\mu \mathrm{F}$ capacitor, in parallel with the neon, should be without appreciable internal leakage, or the neon will not strike. For the same reason the associated $1,000-\mathrm{pF}$ capacitor should be a mica type.

The crystal connections shown in Fig. 2 are for a Quartz Crystal Co's. octal-based Q5/100 crystal unit. This has so low a temperature coefficient that for normal purposes no temperature control is needed, and the crystal stage of the marker is a modified Colpitts intended for use with this crystal. With the component values given, the accuracy of the marker would usually be sufficient even without
checking the frequency against the B.B.C. or MSF (Rugby) transmissions.

Power for the unit can often be drawn from the receiver; if not current may be available from a small power pack, or from some other item of equipment. Coupling between the marker and receiver will be sufficient, for lower frequencies, if the marker output lead is placed near the receiver aerial terminal. For the high harmonics, the output lead is connected directly to the receiver.

To check against the B.B.C. or MSF transmissions, tune in the signal, and adjust the marker coupling to the receiver so that its signal is of similar strength to that of the station. In some parts of the country this may mean that no aerial, or a very short length of wire, will suffice. The receiver BFO should not be used, and if the MSF transmissions are chosen they should be at those periods (of the hour) when no modulation is present.

When the $100-\mathrm{pF}$ capacitor is far from its correct setting, the difference in frequency between the crystal marker and the station transmission will be audible as a low pitched flutter. As the capacitor


Fig. 2. Layout of components and wiring on the underside of the "top " plate of the chassis box.


This underside view of the top plate should be studied in conjunction with Fig. 2 for identification of the components.
is adjusted towards correct capacity, the rapidity of flutter will decrease. If a tuning meter or indicator is fitted to the receiver, this will rise and fall as the marker and station signals fall in and out of step, and the best setting for the capacitor can easily be located. The B.B.C. $200-\mathrm{kc} / \mathrm{s}$ transmissions are easily received in many areas, frequency is maintained to a very high order of accuracy, and adjusting the marker against this frequency is particularly simple.

For checking receiver calibration, the receiver

BFO should be switched on, and the aerial removed. The gain controls will need to be quite low for harmonics such as those used to mark the edges of the $1.8 \mathrm{Mc} / \mathrm{s}$ to $2 \mathrm{Mc} / \mathrm{s}$ or $3.5 \mathrm{Mc} / \mathrm{s}$ to $3.8 \mathrm{Mc} / \mathrm{s}$ bands. For higher harmonics, such as are required to indicate the edges of the $28 \mathrm{Mc} / \mathrm{s}$ to $30 \mathrm{Mc} / \mathrm{s}$ band, the receiver gain controls will need to be near maximum.

For signal generator or VFO calibration or adjusting, the BFO should be off, so that the generator or VFO signal can beat with the crystal marker's signal.

# Amateurs' National Convention 

VARIED R.S.G.B. PROGRAMME AT CAMBRIDGE

THE fourth post-war National Convention of the Radio Society of Great Britain-held at Cambridge from September 15th-17th and attended by some 200 members and guests-was notable for the very full programme of technical lectures and discussions, in addition to visits to places of technical interest and those social functions which are traditional at amateur radio conventions.

The links between radio amateurs and radio astronomers have always been close; indeed the first steerable radio telescope was built by an American amateur and a British amateur played an important rôle in establishing the radiation of radio signals by the sun. The Cambridge convention fittingly included talks by Prof. Martin Ryle (who holds the call G3CY) and Dr. F. G. Smith, of the Mullard Radio Astronomy Observatory. The programme also included visits to the Observatory to see the unique development of aperture synthesis and EDSAC II, the University's computer on which the observations are analysed. One of the radio telescopes now records signals automatically in digital numbers for feeding direct to EDSAC.

Prof. Ryle described the two main aerial arrays which would be capable of detecting a one milliwatt transmitter on the moon and which in June identified radio sources $4,500,000,000$ light-years dis-tant-the world's one-way DX record! Present investigations are concentrated on 38 and $178 \mathrm{Mc} / \mathrm{s}$, though work is also being carried out on 81.5 and $408 \mathrm{Mc} / \mathrm{s}$. By aperture synthesis, the $3,200 \mathrm{ft}$, $38 \mathrm{Mc} / \mathrm{s}$ fixed array used in conjunction with the relatively small moving section gives results equivalent to a paraboloidal reflector of $1,500 \mathrm{ft}$ diameter.

The low-noise maser, parametric amplifier and electron beam amplifier are not exactly standard amateur practice yet (though some interesting work on parametric amplifiers has been carried out by amateurs in the United States and also, on a more limited scale, in this country) but A. H. W. Beck and Dr. F. G. Smith graphically summarised recent research in these developing fields. The Adler tube, with its remarkable ability to separate signal from noise and then to lose the noise, appears particularly promising in view of its relative simplicity of adjustment, wide bandwidth and 25 dB or so gain to 3,000 $\mathrm{Mc} / \mathrm{s}$-if only prices fall to suit amateur pockets. At the other end of the radio spectrum, Dr. B. H. Briggs, of the Cavendish Laboratory (formerly

G2FJD) described investigations into the strange phenomena of the "whistling atmospherics."
The rapidly increasing interest in single-sideband transmission ensured a lively audience for a survey of modern amateur practice by G. C. Bagley (G3FHL)-one of the few British amateurs who can already claim more than ten years' experience of this type of operation. Another topic attracting increasing attention stems from the results now being achieved by members of the British Amateur Radio Teleprinting Group which sprung into being after the first British amateur teleprinter contacts were made last year in the $3.5 \mathrm{Mc} / \mathrm{s}$ band.
A most effective demonstration of the present advanced state of amateur television transmission in the U.K.-which appears to lead the world in this specialized branch of amateur radio-was presented by members of the British Amateur Television Club. Two 27 in receivers displayed the 405 -line transmissions from G3NOX/T at Duddenhoe End, some 15 miles from Cambridge. This station, one of an active network of stations reaching well down into Essex, transmits on $436.3 \mathrm{Mc} / \mathrm{s}$ with an e.r.p. of 5.7 kW .
The pictures were of impressively high standard; so incidentally was the programme technique displayed by the team of operators who had pooled resources to provide three camera chains. Transmissions from this group of stations are made regularly on Saturday evenings and can be received over distances of up to about 30 miles on standard television receivers fitted with $70-\mathrm{cm}$ converters. At the end of the demonstration, viewers at Cambridge were able to question the operator (I. M. Waters, G3KKD) at the television transmitter via a duplex radio link on 70 cm and 2 metres.
Throughout the Convention transmissions were made on amateur bands from $1.8 \mathrm{Mc} / \mathrm{s}$ to $144 \mathrm{Mc} / \mathrm{s}$ under the special call-sign GB2CAM.
The convention was rounded off by a dinner at the Guildhall with guests including the Mayor of Cambridge, who had assisted the event in many ways, Dr. F. G. Smith, Brigadier E. J. H. Moppett and Major-General E. S. Cole, now Director of Telecommunications at the War Office, but formerly better known to amateurs as G2EC, SU1EC and SVIEC. General Cole spoke of the effective work of the R.S.G.B. in preserving amateur frequency allocations and of the value of radio communication in encouraging international friendships.-J. P. H.



INTERKAMA Z 1960

DUSSELDORF EXHIBITION OF AUTOMATION AND INSTRUMENTS

OCCUPYING thirteen exhibition halls of various sizes having a total floor area of nearly 10 acres, the Interkama exhibition and congress was held from 19 to 26 October, 470 exhibitors, 188 from countries other than Germany, displayed equipment for test, measurement, process control, data logging and the myriad other functions required by industry today. There were sections devoted to French, East German and British firms (this latter arranged by S.I.M.A.), but these were not the only displays because many "foreigners" showed their products on the stands of their German agents, or had stands of their own elsewhere in the exhibition.

## Industrial Control and Automation

The simplest type of control system uses a man, who makes adjustments by reference to a continuous-reading instrument, as part of a servo loop. The next step is, naturally, the closing of the loop without the man, so that automatic control is achieved. A practical illustration of a closed-loop control system was given by Eckhardt, with a liquid-flow meter and control valve. The sensor uses the Faraday effect of the generation of a potential on a conductor moving in a magnetic field: an electromagnet energized from the mains creates a field across the pipe carrying the liquid. Two electrodes pick out the potential generated by the flow, and this potential is "bucked" or backed off by a servodriven potentiometer whose motor is energized, via amplifiers, by the difference of potential between the potentiometer and electrodes.

For quality control of a product the recording of the many parameters of a plant must be carried out constantly. The registering, collating and working out of results from these readings can, in a complex plant, consume much time and effort, but this "watchdog" facility can be provided by a central data-processing machine, for instance, the Beckman " 123 " system. This can accept up to 100 electrical inputs representing plant parameters, scanning each one by a stepping switch in 60 sec , processing each input as selected by the position of plugs in a peg board-a zero offset might be required for one input, say, whilst another channel may require a logarithmic amplifier characteristic. The measurements taken can be written out on an electric typewriter or punched-tape system at preset intervals or on demand, and a digital electrical output is available.
"Automation" of processes using small closed loops can have the disadvantage of inflexibility when, say, a chemical process using raw materials of variable quality
or where fairly frequent changes in output are required. Each control loop must be reset to a new value each time a change is required, and all the possible combinations must have been worked out beforehand. With more than a very small number of variables it can be seen that the amount of work required could be uneconomical. However, the necessary flexibility can be achieved by opening the individual closed loops and interposing a "communal" computer. Then, by giving the computer the reaction equation and data on raw


Macq Electronique drift meter for stabilized Dower supplies
material purity and price changes the programme can be set to operate the plant automatically for maximum profit.

A system of this type using a Ramo-Wooldridge RW 300 computer to control the manufacture of dodecylbenzene was illustrated on part of the French stand by the Compagnie Européenne d'Automatisme Electronique.
Another point noticed was the widespread use of small-package solid-state logic (such as "and," "or," "not") and instrumentation (timers, amplifiers, etc.) units. There would seem to be a good case here for the various manufacturers to standardise practical details such as mounting, connections and power supplies so that a Siemens "Simatic," say, is directly interchangeable with an Ateliers des Constructions Electronique de Charleroi "Logacec," a Telefunken "Logistat," or a Mullard " Norbit" or "Combi-element."

## Instruments

A major feature of the exhibition was the display of instruments for test and measurement, as distinct from automatic control, in the laboratory or the production line. A complete floor of one exhibition hall was divided into "classrooms" in which 56 instrument manufac-
turers gave 249 courses of instruction on their products.
The Wandel and Goltermann WZL-181 test set is designed for the testing of transformer windings. Basically the instrument itself consists of a bridge which can be preset to check coil resistance, turns ratio, etc., a "good" coil returning a zero or "in balance" reading. In the case of unbalance an indication of direction, i.e., high or low, is given. The coil undergoing test is placed in a test jig and connected up; a standard coil induces power, and the tests (including a phase-reverse facility in case the leads have been reversed on connection) can be carried out by the use of push buttons on the jig.

On the General Radio stand small variable delay lines were found. These, having maximum delays of 0.5 and $1 \mu \mathrm{sec}$ are made rather after the fashion of a wirewound potentiometer, but the winding is skewed on the former to enable a greater length of wire to be in close proximity to a "screen" which forms the capacitive element of the line. This type of variable line is used in the GR 1392A delay generator which can be used for measurements of time intervals as short as $1 / 100 \mu \mathrm{sec}$. Another instrument seen here was a transfer-function and immittance (impedance and admittance) bridge for the range 25 to $1500 \mathrm{Mc} / \mathrm{s}$, enabling measurements to be made on two, three and "four" (one input and one output earthed) terminal networks. Two calibrated indicator heads are supplied with this bridge, one for each use, and these have small rotatable loops as variable coupling elements to the lines connected to the network under test. The loops are energized from an external generator and a null detector is connected to the junction of the output from the network, a known susceptance and a known conductance. By adjustment of the couplings from the generator to the network input and the resistive and reactive arms, a null can be achieved, when the prameters of the tested object can be read from the calibrated scales of the indicator heads.

For the testing of many circuits frequency-sweep techniques have immense advantages but until the Carcinotron*, whose frequency can be varied widely electrically, was produced by the C.S.F. the widest sweep available in the s.h.f. bands was somewhat limited. Several manufacturers have produced swept signal generators using Carcinotrons, notably Ferisol (Geffroy et Cie ). A generator making its debut at the exhibition was the Rhode and Schwarz Type ZWC. This has a choice of plug-in Carcinotron modules covering the range 1 to $18 \mathrm{Gc} / \mathrm{s}$ in (roughly) octave sweeps and constant output is achieved by sampling the signal at a directional coupler at the entry to the test piece so that any anomolous effects due to the cable or waveguide coupling and the three " marker-producing" wavemeters are neutralized. These wavemeters are mounted below the display unit for ease of use, and the display itself offers double-trace working on a $20-\mathrm{cm}$ ( $\approx 7 \frac{3}{4}-\mathrm{in}$ ) square cathode-ray tube.

The short-term drift, or random variations, of a stabilized power supply are usually so small that they are insignificant. However, if it is desired to know what is happening to one's "stable" supply, the diffculty of measurement has to be overcome. Macq Electronique had on display a drift-meter for this purpose; accepting inputs in the range 1 V to 2.5 kV , a resistive network is used to drop the applied potential to that of a standard Weston cell contained within the instrument. A mechancial chopper switches between the cell and applied potential, and the resulting a.c. is amplified, detected and applied to a centre-zero meter. The sensitivity of the amplifier can be adjusted so that variations of $1 / 100,000$ or $1 / 10,000$ of the input produce half-scale deflection; also sense (i.e., fall or rise) is indicated.

[^3]

Beckman Instruments ": 123" datalogging system. "Peg board" is used for setting up.

Wandel u. Golterman transformer test set showing the test jigs.


Swept signal generator using Corcinatron (Rhode u. Schwarz).


# WORLID OF WIREIESS 

## Commonwealth Telecoms

WITHIN the 62 pages of the 9 th General Report of the Commonwealth Telecommunications Board, covering 1959, are reports on the activities in each of the eight partner countries which have nationalized telecommunications systems. The report as a whole is concerned mainly with traffic statistics but there are also some interesting items covering research and developments.

In this country experimental tropospheric scatter transmissions from a transmitter near Bodmin, Cornwall, to receivers near Leafield, Oxon (173 miles) and at West Beckham, Norfolk ( 300 miles) continued with the aim of obtaining a year's cycle of measurements for analysis of fading characteristics. The tests have included comparisons between multi-channel telephony with f.m. and narrow-band s.s.b. with a.m. and also television with f.m. Tropospheric scatter transmissions from Holland to West Beckham ( 150 miles) are also being conducted.
The C.T.B., of which Sir Ben Barnett is chairman, consists of a representative from each of the partner governments. Its primary function is the formulation of joint telecommunications policies.

## For C.A.R.A.C. read C.C.E.C.A.

WHAT was originally known as the Civil Aviation Radio Advisory Committee has been renamed the Consultative Committee on Electronics for Civil Aviation. Broadly its terms of reference are to provide a forum for discussing U.K. policy in the field of electronics for civil aviation in relation to industrial research and development.
The representatives of the Electronic Engineering Association on the committee are Dr. B. J. O'Kane (Marconi's), K. E. Harris (Cossor) and Air Comdre. C. A. Bell (G.E.C.). The Ministry of Aviation, Admiralty, Air Ministry, B.O.A.C., B.E.A., Post Office, Society of British Aircraft Constructors and the British Independent Air Transport Association are also represented.

## Student Exchange

SINCE the scheme for the international exchange of students for technical experience was introduced in 1947, over 52,000 students have participated. During the past year 6,430 took advantage of the scheme, which is operated by 24 countries. The 1960 report of the International Association for the Exchange of Students for Technical Experience records that Germany received the largest number of students $(1,195)$ and also sent abroad the largest number ( 1,105 ), with the U.K. second, receiving 909 and sending 883.

The report issued by the U.K. branch shows the distribution by country of the students sent from and received in this country and also by the subjects which they were studying. About $20 \%$ (178) of U.K. students went to Sweden, from which the largest number of students (171) came to this country. The address of the London office is 178 Queen's Gate, S.W.7.

## B.B.C. Plans

CONTRACTS have recently been placed by the B.B.C. for the building of a number of low-power transmitters to extend and improve its television and v.h.f. sound services. Combined stations are to be erected at Oxford, Llandrindod Wells (Radnor) and Redruth (Cornwall) and a TV station at Manningtree (Essex). Also the site has been chosen for a combined station near Galashiels (Selkirk).

Approval in principle has been given by the P.M.G. for a low-power station near Ballachulish (Argyll). This station is additional to those originally planned for Western Scotland. A point-to-point link was originally to be installed for this part of the route feeding two new stations. It has been decided, however, to build a small auxiliary station which will serve the dual purpose of linking the two stations and providing direct television reception for the people in the area.

## Technical Writing

AUTHORS of technical articles on British electronic developments, published during 1960, who wish to compete for the six premiums of 25 guineas each awarded by the E.E.A. and R.I.C., must submit their entries by December 31st. They should be sent to the Electronic Engineering Association, 11 Green Street, London, W.1, and should consist of a copy of the journal* and the relevant pages, proof or reprint. Entries must be accompanied by a signed declaration of eligibility, showing that the entrant is not paid a salary solely for writing and not earning $25 \%$ or more of his income from articles or book royalties.

* Wireless World is sent by arrangement to each of the judges

Receiver Despatches.-Figures issued by B.R.E.M.A. show that manufacturers' despatches of television receivers during the first nine months of this year were $30 \%$ below the same period last year ( $1,257,000$ compared with $1,787,000$ ). Domestic sound receivers and car radio despatches during the same period rose by $23 \%$ ( $1,346,000$ against $1,091,000$ ). Radiogramophones, too rose from 114,000 to 130,000 .

Recording Studios.-The revised topographical list of members of the Association of Professional Recording Studios shows that the Association now has a membership of nearly 100 "private professional studios" in this country. It operates a scheme under licence from the Mechanical-Copyright Protection Society, permitting its members to make recordings for private customers of works controlled by the M.P.S.

## CHRISTMAS PRESENTS

Why not give your technical friends a copy of the Wireless World Diary as a Christmas present? It has an 80 -page reference section and costs 6 s 9 d (leather) or 4s 9d (Rexine). Friends who seem quite incapable of adjusting correctly the controls of their television sets might appreciate a copy of "Improve Your Television Reception" (5s) or "Correcting Television Picture Faults" (4s).

Satellite Communications.-Technical discussions between American and British officials on matters associated with the possible use of communications systems via earth satellites are being held in the U.S.A. The U.K. team is headed by Major General L. de M. Thuillier, of the Cabinet Office, and includes Captain C. F. Booth, W. J. Bray, H. Leigh and F. J. D. Taylor, of the Post Office; J. R. U. Page, of the Office of the Minister for Science; F. E. J. Girling, A. G. Earl and C. Williams (Ministry of Aviation); F. A. Kitchen (Admiralty) and Group Captain A. Foden and C. F. Sutton (Ministry of Defence).
B.C.A.C.-When the British Conference on Automation and Computation, which is a consortium of associations and institutes interested in the field of computation, was formed in 1957 it was divided into three autonomous groups. One dealt with engineering, another control and the third sociological interests. It has now been decided to reconstitute the association with a central council. There are now about 30 societies in the reconstituted B.C.A.C. of which Sir Walter Puckey is chairman with L. T. Blakeman, J. F. Coales and H. G. Conway as vice-chairmen. The honorary secretary is W. K. Brasher, secretary of the I.E.E. which is providing secretarial services.

Local Broadcasting.-A detailed survey of their plans for 100 or more local broadcasting stations in this country has been issued as a 13 -page brochure by Pye Ltd. Comparisons are drawn with the situation in the U.S.A. where extensive use is made during daylight of common medium-frequencies by a large number of stations with geographical spacing of 90 miles or less.

Fleming Memorial Lecture of the Television Society will be given by Professor D. M. MacKay on January 26th at the Royal Institution, Albemarle Street, London, W.1. His subject will be "Behind the Eye". Admission is by ticket obtainable from the society, 166 Shaftesbury Avenue, London, W.C.2. Dr. MacKay was recently appointed to the research chair in communication founded by Granada Television at the University College of North Staffordshire.


MARY E. CHAMBERS, of the E.M.I. Research Laboratories, is the first woman to be awarded the Dip. Tech. with first class honours. She has taken a four-year sandwich course at the Brunel College of Technology and E.M.I., where she has been concerned with the preparation of photo-conductors for television camera tubes and phosphors for c.r. tubes.

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\text { Publication date. - Owing to the Christmas } \\
\text { Holidays our January issue will not be published } \\
\text { until December } 3 \text { lst. }
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The National Radio Show, 1961, will be held at Earls Court from August 23rd to September 2nd with a preview for overseas and invited guests on August 22nd. This, the 28th National Radio Exhibition, will again be organized by Radio Industry Exhibitions Ltd., 59 Russell Square, London, W.C.1.

Aeronautical Electrics and Electronics.-A joint committee, with Professor G. A. Whitfield, of the College of Aeronautics, Cranfield, as chairman, was recently formed by the I.E.E. and the Royal Aeronautical Society to consider ways and means of stimulating discussion of problems of common interest. As a result, joint meetings of the two organizations are to be held in London and plans are being made for the formation of joint groups in the provinces. At the inaugural meeting in London on November 11th, Viscount Caldecote spoke on the future of "electrics" and "electronics" in aircraft and guided missiles.

Transducers.-The Electronic Engineering Association is seeking the help of manufacturers and users in compiling a list of transducers. It is collating information not only on current types but also on those under development. Those interested in helping the association in compiling this list, which will be available to nonmembers, are invited to send for the questionnaire obtainable from J. F. Richardson, Technical Secretary, E.E.A., 11 Green Street, London, W.l.

Gramophone records produced during the third quarter of the year totalled just over 16 M , about $5 \%$ more than in the corresponding period last year. There were $59 \%$ fewer 78 r.p.m. records produced ( 688,000 compared with $1,691,000$ ) and $2 \%$ fewer $33 \frac{1}{3}$ r.p.m. discs ( $3,488,000$ against $3,546,000$ ) but $18 \%$ more 45 's and medium-play ( $11,860,000$ against $10,092,000$ ).

Vacation Courses.-" Introduction to Industry " training courses are again available at the Chippenham Works of Westinghouse Brake and Signal Co. Ltd., for boys who are taking G.C.E. "A" level scientific subjects. Two courses are arranged during the Christmas vacation, January 2nd to 6th, and 9th to 13 th. Applications should be made before December 1st.

Engineering aspects of the plastics industry will be covered by a new monthly journal, International Plastics Engineering, to be published in association with British Plastics. The first issue will appear on February lst.

## INDUSTRIFI GROUPS-3

E.M.I.-Since going to press with the section of this issue in which the third of our series of surveys of industrial groups appears (p. 611), E.M.I. has supplied additional information which brings the total number of companies within the group to nearly 80. The additional companies are:-
E.M.I. (Australia) and its subsidiaries, Television Patents, Marconiphone Co., E.M.I. Finance Corp., Hayes Properties, and Castle Music; the following subsidiaries of Capitol Records Inc.: Ardmore Music Corp., Beechwood Music Corp., Capitol Record Club Inc., Capitol Records Distributing Corp, Capitol Records International Corp., Capitol Publications Inc. Electric \& Musical Industries (U.S.), Electric \& Musical Industries (Canada), Capitol Records of Canada, Ardmore \& Beechwood (U.K.), and Capitol Royal Club of Canada; Edition Accord, Kristall Schalllpatten, and Odeon Musikhaus, subsidiaries of Carl Lindstroem; and these ten additional subsidiaries of Morphy-Richards, Morphy-Richards (Cray), Morphy-Richards (Australia), Morphy-Richards (Canada), Morphy-Richards, S. Africa, Morphy-Richards Inc., U.S.A., Morphy-Richards' (Nederland), M-R Salgsorganisation A/S, Denmark, Eureka Electric, Astral Equipment, and Yelsen.

## Personalities

Major-General E. S. Cole, C.B., C.B.E., has accepted the invitation to become president of the Radio Society of Great Britain for 1961. He has been a member of the sociery for nearly thirty years and is well known among amateurs by his call G2EC. General Cole was commissioned in the Royal Signals in 1925 and from 1941 to 1946 was Corps Chief Signal Officer. In 1952 he became Chief Signal Officer, Headquarters, Southern Command, and in 1954 was appointed Deputy Chief Signal Officer, Signals Division, S.H.A.P.E. Since 1958 he has been Director of Telecommunications, War Office. General Cole was at one time chairman of the British Joint Communications Board.


Maj. Gen. E. S. Cole

J. K. Laughton
J. K. Laughton, who joined Plessey's Telecommunications Division in 1958 as sales engineer, has been appointed manager of that division. He went to Plessey after 20 years' service in the Royal Navy. He was promoted to Commander in 1954 and was for the last year of his naval career Assistant Director of Radio Equipment (Communications) in the Admiralty.
J. H. H. Merriman, O.B.E., M.Sc., A.Inst.P., M.I.E.E., was recently appointed staff engineer of the Overseas Radio Planning and Provision Branch of the Post Office. For the past four years he has been seconded to the Treasury as the assistant secretary responsible for computer work and policy in all Government Departments. A physics graduate of King's College, London, where he later worked under Professor Appleton on non-linear oscillators for his M.Sc. degree, he joined the Research Station staff at Dollis Hill in 1936 . He was closely associated with the development of MUSA-the steerable aerial system.
N. T. Atkinson, formerly head of television development in Ekco, has been appointed to the post of chief of television and radio development. He has been concerned with the development of Ekco television receivers since 1938 and in his new position he will be directly responsible to A. J. Brunker, executive director and chief engineer. E. W. Maynard, who has been in charge of export television development, succeeds Mr. Atkinson as head of television development.

Dr. I. L. Pawsey, F.R.S., assistant chief of the Division of Radiophysics of the Commonwealth Scientific and Industrial Research Organization, Australia, has been awarded the Hughes Medal of the Royal Society for his distinguished contributions to radio astronomy both in the study of solar and of cosmic radio emission.

Terence B. Tomlinson, Ph.D., B.Sc., A.M.I.E.E., A.M.Brit.I.R.E., joined Southern Instruments earlier this year as chief engineer. He is responsible for design and development for the newly formed company, Drayton-Southern, Ltd. After graduating from London University he became a laboratory demonstrator in electronics at the Cavendish Laboratory, Cambridge. He spent two years with Sobell Industries, returning to the university to take a higher degree-his research being concerned with low-frequency fluctuations of emission from oxide-coated cathodes. Dr. Tomlinson subsequently became a lecturer in electronics at Southampton University, where he stayed for six years before joining the G.E.C. Research Laboratories for work on solid-state physics. He transferred to the G.E.C./I.C.T. subsidiary, Computer Developments, Ltd., where he stayed until joining Southern Instruments.

Dudley Saward, O.B.E., who joined Texas Instruments Ltd. on its formation in 1956, has resigned from the managing directorship in order to devote more time to other activities, but will remain a director of the company. Mr. Saward was chief radar officer, Bomber Command, from 1942 to 1945 and was for some time after the war controller of navigation and telecommunications with British European Airways. He is succeeded as managing director by A. N. Provost, a former U.S. naval officer and an engineering graduate of Tufts and New York Universities. Mr. Provost was for some time with Sylvania Electric Products before joining the semiconductor components division of Texas Instruments, Inc., in Dallas, U.S.A.
J. D. Dale-Lace, A.M.I.E.E., D.F.H., the new sales manager of the radio components and special products department of A.E.I.'s Radio and Electronic Components Division, was a technical officer in the Royal Navy from 1940 until 1946 when he joined Pye, where he worked on the development of television transmission equipment. A year later he went to South Africa as technical representative of Sperry's South African agents. He rejoined Pye in 1951 as a telecommunications sales engineer, and since 1956 has been with de Havilland Propellers as guided weapons service manager at Hatfield.

T. J. Murphy, A.M.I.E.E., A.M.Brit.I.R.E., of the engineering division of Radio Eireann, has been given the City and Guilds Insignia Award in Technology (C.G.I.A.) for "his competence in the field of electrical industries (broadcasting-sound and television) and in respect of a thesis on the design of modern sound broadcasting studios." Mr. Murphy had been three years in Pye's final test department before joining the Engineering Division of the B.B.C. in 1943. He left the Corporation in 1951 to join Radio Eireann, Dublin, and for the past five years has been in charge of the studio department of the engineering division.

Colonel H. A. Lewis, O.B.E., T.D., B.Sc.(Eng.), M.I.E.E., A.C.G.I., has relinquished his appointment as managing director of E.M.I. Sales and Service, Ltd., to become managing director of Newmark (London) Distributors, Ltd., precision instrument manufacturers, of Great Portland Street, London, W.1. Col. Lewis has been with E.M.I. since 1956. For the previous six years he was in the Broadcasting Division of Marconi's. From 1933 to 1948, except for the war years when he was in R.E.M.E., Col. Lewis was in the Engineering Division of the B.B.C.
F. W. Perks, sales director of British Radio Corporation Ltd., is retiring. His career in the radio industry extends from 1914 when he joined Marconi's W/T Co. Freddie Perks, as he is affectionately known in the industry, has held many important appointments including the chairmanship of B.R.E.M.A. and B.R.V.M.A. and is at present chairman of Radio Industry Exhibitions Ltd. (organizers of the National Radio Exhibition) and president of the Radio Industries Club. He is succeeded as B.R.C. sales director by T. B. (Jock) Henderson, general sales manager of Philco (Great Britain) Ltd., an associate company.
P. A. M. Curry, M.A., B.Sc., has been appointed deputy to H. St. A. Malleson, head of Mullard's Government and Industrial Valve Division. After graduating with a first in engineering science at Oxford, Mr. Curry studied for a research degree and was awarded a Henry Fellowship providing a two-year course at Harvard where he obtained the degree of Master of Business Administration.
L. J. M. Knotts has joined Mullard Equipment Ltd. as technical adviser on the application of electronics to railway signalling and communication systems. He was formerly signal engineer to the Scottish Region of British Railways.
F. S. Barton, C.B.E., M.A., B.Sc., M.I.E.E., who, as announced in October, has retired from the Civil Service, has joined the board of Painton \& Co., Ltd., of Northampton.

Keith Lockyer, B.Sc., A.M.I.E.E., A.M.Brit.I.R.E., has been appointed general manager of Donvin Instruments Ltd., and Dumar Optics Ltd., the instrument, camera and tape recorder repair firms which are part of the R. B. Pullin group.

Dr. T. W. Straker is deputy manager of Marconi's Radar Division and not manager as was inadvertently announced in our last issue. The manager of the division is E. N. Elford, O.B.E., A.M.I.E.E.
H. M. Palmer has retired from Marconi International Marine after nearly 50 years' service. He has been in charge of the company's Fleetwood service depot for the past 25 years. He is succeeded by J. E. Dallaston who joined the company in 1931 as a sea-going radio officer, as did Mr. Palmer.
G. H. W. Johnson, Assoc.Brit.I.R.E., for the past three years marine director of Marconi (South Africa) Ltd., has relinquished his seat on the board and has joined Norsk Marconikompani A/S in a similar capacity. Mr. Johnson joined the Marconi organization as a radio officer in 1938.
G. A. Dwyer, manager of the Marconi Marine Co.'s Bombay depot from 1952 to 1954, has been appointed the company's representative in India and Ceylon. He joined Marconi's sea-going staff in 1929. In 1941 he was appointed to the shore technical staff at Glasgow and three years later was seconded to the Naval base at Ardrossan.

Angela M. Lane has been appointed administrative secretary of the Television Society. She joined the secretariat of the Society in 1955 and has been assistant secretary for some time. Geoffrey Parr, the honorary secretary of the Society since 1945, who recently entered hospital for heart trouble, is now convalescing.

## OUR AUTIIORS

G. W. H. Larkby, author of the article describing a remotely controlled TV camera designed by the B.B.C., has been with the Corporation since 1936, except for the war period when he served with the R.A.F. in 60 Group (Radar). On his return to the B.B.C., Mr. Larkby was attached to the outside broadcast department of the Television Service. He later transferred to the Designs Department where he has been chiefly concerned with the many complex electro-mechanical and optical problems in the television field.
J. W. Bunn, joint author of the article on ceramic permanent magnets, is powder metallurgist with Swift Levick and Sons, with whom he has been associated for 13 years. In 1954 he gained the associateship in metallurgy from the University of Sheffield where he had undertaken part-time study. He is a licentiate of the Institute of Metallurgists. J. Harrison, his co-author, is magnet design engineer with Swift Levick. He was previously assistant metallurgist at the central research laboratory of the Permanent Magnet Association.
F. C. Judd, contributor of the article on aerial measurements, is a free-lance technical writer. Until making writing his full-time occupation, he was in the radar research laboratory of Kelvin \& Hughes.

## OBITUARY

Dr. G. W. O. Howe, Emeritus Professor of Electrical Engineering of Glasgow University, died on November 7th, aged 84. Professor Howe was successively technical editor of and consultant to our sister journal Wireless Engineer (now Electronic Technology) for over 30 years. In 1956 he was awarded the Faraday Medal of the I.E.E. for "his pioneering work in the study and analysis of high-frequency oscillations and on the theory of radio propagation; and for his outstanding contributions to engineering eclucation." Dr. Howe, who was apprenticed to Siemens at Woolwich, went to Imperial College as a lecturer in 1905 and in 1921 was appointed James Watt Professor of Electrical Engineering at Glasgow University, the position he held until his retirement in 1946 when he was appointed an Emeritus Professor. Dr. Howe was a great teacher and his editorials in Wireless Engineer over very many years were an outstanding contribution to electrical and radio theory. In recognition of "his contributions to the advancement of radio science and in particular for his long, series of notable editorial articles in Wireless Engineer'", he was, in 1956, elected an honorary member of the Brit.I.R.E.
Professor Bernard Hague, D.Sc., Ph.D., F.C.G.I., D.I.C., M.I.E.E., who had occupied the James Watt Chair of Electrical Engineering at Glasgow University since 1946, has died aged 67. His academic career began in 1920 when he became a lecturer at the City and Guilds College, prior to which he was for four years at R.A.E., Farnborough. Dr. Hague was a lecturer at Glasgow University from 1923 until 1929 when he went to the Brooklyn Polytechnic Institute, N.Y. He returned to Glasgow as senior lecturer in 1930.

Reginald Lawler, manager of the Electronics Division of Ferguson Radio Corporation, which he joined in 1938 as a radio development engineer, died on November 2nd, aged 59 . He was at one time in charge of the company's factory at Hirwaun, Glamorganshire, but returned to Enfield in 1952.

# News from Industry 

Pye-Ekco Merger. - The boards of Pye Ltd. and E. K. Cole Ltd. have recommended to their shareholders a merger of the two companies through a holding company, as yet unnamed. C. O. Stanley would be chairman and E. K. Cole deputy chairman. Pye will have a $75 \%$ holding in the new company and Ekco $25 \%$. Their combined assets are nearly $£ 25 \mathrm{M}$.

Plessey-Garrard.-Agreement has been reached in principle under which the Plessey Company would acquire all the issued share capital of the Garrard Enginecring and Manufacturing Co. If the offer is accepted Garrard will continue as a separate entity under its own board of directors with Hector V. Slade as managing director.

Cossor.-Major-General Sir Miles Graham, who succeeded the Marquess of Exeter as chairman of A. C. Cossor Ltd. last March, reported at the annual general meeting that for the year 1959;60 the group trading profit was $£ 127,699$, compared with $£ 534,260$ the previous year. There was, however, a net loss of 197,977 . The group includes Cossor Radar \& Electronics, Sterling Cable Co., Cossor Communications, Cossor Instruments, Lea Bridge Cabinet Works and Best Products.

Baird.-Radio Rentals Ltd. have entered into an agreement to purchase Baird Television Ltd., and will market television and sound receivers bearing the name "Baird." All the company's sets are made in Bradford by its Bradford manufacturing subsidiary, Mains Radio Gramophones, and will in future carry the new trade name.

BMEWS.-R.C.A. Great Britain Ltd., the U.K. asso"ciate of the Radio Corporation of America, is "weapon system contractor" for the Fylingdales, Yorks., base for the American ballistic missile early warning system. It is estimated that the cost of the whole Fylingdales project will be $\$ 115 \mathrm{M}$. Work has already begun on this, the third site for the BMEWS chain (see Wireless World, July, 1960, p. 335).

Eimac.-In association with the development of new Eimac klystron valves for high-power radar and communications, Eitel-McCullough are building a "superpower" supply unit drawing 12 amps at 282 kV (approximately 3 MW ). Agents for Eimac products in the U.K. are Walmore Electronics Ltd., Phœnix House, 19/23 Oxford Street, London, W.1.

Trav-ler Tape Recorder.-This transistor battery/ mains portable tape recorder introduced at the National Radio Show by Casian Ltd., is being produced by Aerialite Ltd., of Hargreaves Works, Congleton, Cheshire, who have acquired the manufacturing and distributing rights.

Balzers High Vacuum Ltd. has been formed, with offices at 1 Mornington Terrace. Regent's Park, London, N.W.1, to handle in this country the products of Balzers A.G., of Liechtenstein, including equipment for high-vacuum metallurgy and coating, vacuum pumping gear and measuring instruments.

Grundig.-To mark the sale of the 100,000 th Grundig TK20 tape recorder in the U.K., G. S. Taylor, managing director, presented the 100,001 st model to the Royal National Institute for the Blind at a recent ceremony in London. It is understood that it has already been assigned to the Chorley Wood College for Girls.

Air-to-Ground Telemetry.-A consortium of five British manufacturers has been formed on the recommendation of the Ministry of Aviation to exploit the overseas and U.K. markets for a major air-to-ground multi-channel telemetry system. This has been developed by the member companies in collaboration with the Royal Aircraft Establishment, Farnborough. The members are: Elliot Brothers (London) Ltd., McMichael Radio Ltd., Rank Cintel Ltd., Southern Instruments (Contracts) Ltd., and W. S. Electronics Ltd., who have been appointed to act on behalf of the consortium.

Thermionic Products (Electronics) Ltd., of Hythe, Southampton, have ceased manufacturing domestic tape recorders and dictating machines. A limited supply of spares will be available until the end of next year. The company is concentrating on the production of multichannel recorders for airport communications logging and other fields of data logging.

Marconi's have received a further order valued at over $£ 93,000$ for the supply of translators, amplifiers and associated equipment for the B.B.C. The equipment will be used in the first stage of the Corporation's programme for the extension of its television and v.h.f. sound services by satellite stations.

Vactric (Precision Tools) Ltd., and Vactric (Control Equipment) Ltd., subsidiary companies of Vactric Ltd., which was recently put in the hands of a receiver, are continuing to operate as separate entities.

Texas Instruments France, with headquarters in Nice, has been set up by the Dallas company to manufacture semiconductor devices and components for the European Common Market.

## EXPORTS

I.L.S.-Through their Canadian associates, Pye Telecommunications Ltd. have received a contract from the Royal Canadian Air Force for the supply of an instrument landing system for installation at the R.C.A.F. Airfield, Trenton, Ontario. This is said to be the first installation of British I.L.S. in the western hemisphere.

Montreal's two new commercial television stations are to be equipped by Marconi's with duplicate $18-\mathrm{kW}$ vision and $9-\mathrm{kW}$ sound transmitters. Each station will transmit a bilingual service (English and French).

Broadcasting Transmitters.-British Sarozal, of 22 Berners Street, London, W.1, have recently supplied a number of short-wave and medium-wave broadcasting transmitters to Africa. They include a $5-\mathrm{kW}$ transmitter for Radio Pax, in Beira, Mozambique, which will radiate on 3.952 and $7.205 \mathrm{Mc} / \mathrm{s}$ and another for Praia, Cape Verde Islands, for operation on $3.960 \mathrm{Mc} / \mathrm{s}$.

Communications Equipment.-Racal, of Bracknell, Berks., have received through their agents, Instronics Ltd., of Stittsville, Ontario, three major orders for communications equipment from the Canadian Department of Defence Production. They have a total value of $£ 560,000$.

Multi-lingual Catalogue.-Alfred Imhof Ltd., of 112/116 New Oxford Street, London, W.C.I, have produced an international version of their catalogue of racks, cases, consoles and accessories, every item being described in ten languages (English, French, German, Spanish, Italian, Portuguese, Dutch, Swedish, Norwegian and Danish).

# Ceramc Permanent Magnets 

By J. W. BUNN*, A.Met., L.I.M., and J. HARRISON*

Arelatively recent addition to the range of magnetic materials is the barium ferrite ceramic permanent magnet ${ }^{1}$. This material, commercially available under a variety of trade names, e.g. Feroba, Magnadur, Caslox, Indox, Oxit, etc., is basically an oxide compound having the formula $\mathrm{BaFe}_{12} \mathrm{O}_{19}$. It is non-metallic and, in common with other ceramics, is hard, brittle, a non-conductor, and relatively light in weight when compared with alloy magnets. These properties together with its unique magnetic characteristics have opened up a new field in magnet applications.

Barium ferrite, although of similar appearance to the more commonly known "soft" magnetic


Fig. 1. Demagnetization curves for barium ferrite.
core materials, is distinct from this class inasmuch as magnetically soft materials are characterized by high permeability and low losses when used in an a.c. circuit, whereas barium ferrite is a magnetically hard material possessing a high coercivity, remanence and energy product $(\mathrm{BH})_{\text {max }}$. The compound $\mathrm{BaFe}_{12} \mathrm{O}_{19}$ has a hexagonal crystal structure with one axis of easy magnetization parallel to the hexagonal axis. This uniaxial crystal anisotropy together with a small crystal size (approximately 1 micron) is largely responsible for the high coercivity of this material compared with a " soft" magnetic ferrite which has a cubic crystal structure, i.e. three easy directions of magnetization, resulting in low coercivity. Magnetic Properties.-An essential feature of a permanent magnet is its ability to withstand demagnetization, and this is directly related to the coercivity of the material. Barium ferrite has an

[^4]exceptionally high coercivity of the order of 1600 oersted, compared with 600-650 oersted for Alcomax alloys, which is responsible for its ability to withstand high demagnetization effects such as a.c. fields, repulsion by other magnets, and excessive mishandling. Its high coercivity coupled with low permeability makes it possible to use ceramic magnets to minimize magnetic leakage fields of magnetic circuits and so increase the overall efficiency of the assembly.

Of the two further properties, remanence $\mathrm{B}_{r}$ and energy product ( BH$)_{m a x}$ the former is a measure of the magnetization remaining in a material when the applied field is reduced to zero, and the latter is the maximum product of $\mathrm{B} \times \mathrm{H}$ on the demagnetization curve, shown in Fig. 1.
The low saturation magnetization ${ }^{2}$ of barium ferrite leads to comparatively low $\mathrm{B}_{r}$ and $(\mathrm{BH})_{\text {max }}$, typical values for the isotropic variety being shown in Table 1.
Additional properties of permanent magnets which are sometimes required are the incremental or reversible permeability, and the maximum useful recoil energy ${ }^{3}$. In isotropic barium ferrite these values are 1.2 and 0.7 m.g.o. (mega gauss-oersted) respectively, from which it is seen that the recoil

TABLE 1

| PROPERTIES OF BARIUM FERRITE COMPARED WITH ALCOMAX III |  |  |  |
| :---: | :---: | :---: | :---: |
|  | Isotropic <br> $\mathrm{BaFe}_{12} \mathrm{O}_{18}$ | Anisotropic $\mathrm{BaFe}_{12} \mathrm{O}_{19}$ | Alcomax III |
| Remanence $\mathrm{B}_{r}$ (gauss) | 2100 | 3700 | 12,800 |
| $\underset{\text { (oersted) }}{\text { Coercivity }} \mathrm{H}_{\mathrm{c}}$ | 1700 | 1750 | 675 |
| $\begin{aligned} & \text { Maximum energy } \\ & \text { product } \\ & (\mathrm{BH})_{\text {max }}(\mathrm{m} . \mathrm{g.o.}) \end{aligned}$ | 0.9 | 3.0 | 5.5 |
| $\mathrm{B}_{w}(\mathrm{BH})_{\text {max }}$ (gauss) | 1000 | 2400 | 10,200 |
| $\left.\mathrm{H}_{w} \mathrm{BH}^{\text {at }}\right)_{\text {max }}$ (oersted) | 900 | 1250 | 540 |
| Recoil permea- | 1.25 | 1.15 | 3.5 |
| Maximum <br> recoil <br> (m.g.o.) <br> energy <br>  | 0.7 | 1.0 | 1.5 |
| Temperature coefficient of magnetization ( $\% /{ }^{\circ} \mathrm{C}$ ) | -0.2 | -0.19 | -0.02 |
| $\begin{aligned} & \text { Curie tempera- } \\ & \text { ture }\left({ }^{\circ} \mathrm{C}\right) \end{aligned}$ | 450 | 450 | 860 |
| Resistivity (ohmcm) | $1 \times 10^{6}$ | $1 \times 10^{6}$ | $55 \times 10^{-6}$ |
| Density (gm/cc) | 4.8 | 5.0 | 7.35 |
| Magnetizing field for saturation (oersted) | 10,000 | 8,000 | 3,000 |

energy is $47 \%$ of that of Alcomax III alloy magnet material, whereas the total energy $(\mathrm{BH})_{m a x}$ is only $18 \%$ of that of Alcomax III. In view of the comparatively low cost of barium ferrite a definite economic advantage is gained by using this material under recoil conditions. Other features of the recoil properties are dealt with in the section on design and applications.

The Curie point of this material, i.c. the temperature at which for practical purposes the material ceases to be magnetic, is $450^{\circ} \mathrm{C}$, so that all traces of magnetism can be removed by heating to above this temperature. For barium ferrite, this is the best method of demagnetizing, but this method cannot be used for normal alloy magnet materials.

A serious disadvantage of barium ferrite is its high temperature coefficient of magnetization which


Fig. 2. Isotropic ceramic magnets for d.c. motor stators. Continuous cylinder (left) segments (right).
at $-0.2 \% /^{\circ} \mathrm{C}$ is some ten times greater than that of normal alloy magnets. This excludes the use of barium ferrite for, say, measuring instruments, unless elaborate temperature compensating devices are included.

The resistivity of barium ferrite is of the order of $10^{6}$ ohm-cm, and this is of great importance in h.f. applications.

A density of $4.8 \mathrm{gm} / \mathrm{cc}$ is usual, which can be advantageous where weight considerations are paramount.

In 1952, Rathenau and others ${ }^{4}$ found that by aligning the crystallites of this material when in powder form by means of a strong magnetic field, an anisotropic effect was observed in the direction of the aligning field. Following this, intensive work by different workers ${ }^{5}$ in various parts of the world has led to the anisotropic variety which has a higher remanence and a $(\mathrm{BH})_{m a x}$ some three times greater than isotropic barium ferrite.

At the time of writing, only the isotropic form is commercially available in large quantities in this country. Table 1 gives typical properties of both isotropic and anisotropic barium ferrite, together with the more commonly known alloy material Alcomax III. Demagnetization curves for both varieties of barium ferrite are shown in Fig. 1. Manufacture.-The basic ingredients are iron oxide $\mathrm{Fe}_{2} \mathrm{O}_{3}$ and barium carbonate $\mathrm{BaCO}_{3}$ which are mixed in the correct proportions and then heated in air at $1000^{\circ} \mathrm{C}$ approximately to form the required compound $\mathrm{BaFe}_{12} \mathrm{O}_{19}$. This is followed by ballmilling to a very small particle size, after which a binder is added and the powder pressed to the required shape by normal powder metallurgy techniques. The compacts so formed are then fired by heating to approximately $1200^{\circ} \mathrm{C}$ in an oxidizing atmosphere. The density of the fired compact is about $90 \%$ of its solid density, and volume
shrinkage occurs, the final compact being approximately five-eighths by volume of the pressed compact.

The anisotropic differs mainly from the above isotropic variety in that the ball-milled powder is magnetically aligned in the required direction during the pressing process. This alignment is achieved by applying a high magnetic field to the powder, which can be either dry or in the form of a slurry, whilst the powder is in the die. To obtain precise dimensions, the finished magnet may be ground, but normal drilling and machining methods are impracticable due to the brittle nature of the ceramic material.

After mechanical inspection and magnetic testing for quality, final magnctizing is carried out, when a force of 7000 ampere-turns per centimetre length of material must be applied to achieve saturation. Design.-In the design of permanent magnets or magnetic circuits incorporating permanent magners, the arm is to produce the required performance in the most economical manner. In other words, the most efficient use of the magnetic material must be obtained.

Reference has already been made to the demagnetization curve and the three propertics $\mathrm{B}_{r}, \mathrm{H}_{c}$, and ( BH$)_{m_{m a x}}$ of a magnetic material, and these are the main features of the material used by the design engineer.

In the majority of permanent magnet applications it is desired to achieve a given magnetic field strength within the confines of a finite gap. In a gap of length $l_{g}$ and area $a_{y}$, where $\mathrm{H}_{g}$ is the field required in the gap,

$$
\text { where } \begin{align*}
& \mathrm{H}_{u} l_{l} k=\mathrm{H}_{w} \mathrm{~L}_{m} \ldots \ldots .  \tag{1}\\
\mathrm{L}_{m} & =\text { length of magnet material } \\
\mathrm{H}_{w} & =\text { demagnetizing force } \\
k & =\text { reluctance factor } .
\end{align*}
$$

The demagnetizing force $\mathrm{H}_{w}$ is the value at the particular operating point selected for the design on the demagnctization curve, and $k$ is a factor to allow for the total reluctance of the magnetic circuit and usually has a value of between 1.1 and 1.3.

If $\mathrm{B}_{w}$ is the simultaneous induction value at point $\mathrm{H}_{w}$ on the demagnetization curve, then the total flux required in the gap is:-

$$
\begin{equation*}
\mathbf{H}_{g} a_{v} \mathbf{K}=\mathbf{B}_{w} \mathbf{A}_{m} \tag{2}
\end{equation*}
$$

where $\begin{aligned} \mathrm{A}_{m} & =\text { magnet cross-sectional area } \\ \mathrm{K} & =\text { leakage factor }\end{aligned}$
Combining (1) and (2) we have:-

$$
\begin{equation*}
\mathrm{V}_{m}=\frac{\mathrm{H}_{g}{ }^{2} \mathrm{v}_{g} \mathrm{~K} k}{\mathrm{~B}_{w} \mathrm{H}_{w}} \tag{3}
\end{equation*}
$$

where $\mathrm{V}_{m}=$ volume of magnet material

$$
\mathbf{v}_{g}=\text { gap volume }
$$

from which it is scen that the minimum volume of magnet material required to produce a given gap energy is obtained when the product $\mathrm{B}_{w} \mathrm{H}_{w}$ is a maximum. Hence it follows that the most efficient design will be obtained when the magnet geometry is constructed so that the point $\mathrm{B}_{n} \mathrm{H}_{u}$ coincides with the $(\mathrm{BH})_{m a x}$ point of the particular magnet material used in the design. For isotropic barium ferrite operating at $(\mathrm{BH})_{\text {max }}$ the value of $\mathrm{B}_{w}$ is 1000 gauss and $\mathrm{H}_{w}$ is 900 oersted.

The above holds for all designs and magnet materials used, but in the use of ceramic barium ferrite the high value of $\mathrm{H}_{w}$ makes practical much shorter lengths than is normal for other materials,


Fig. 3. Three typical methods of using anisotropic ceramic magnets as d.c. motor stators.
although greater magnet areas are required due to the lower value of $\mathrm{B}_{w}$.
Dynamic Applications - Generators and Motors.-Despite the low value of $\mathrm{B}_{w}$, the high cocrcivity of barium ferrite makes it greatly superior to other commercial permanent magnet materials in its ability to reproduce substantially the same flux after being subjected to adverse conditions. Demagnetizing fields of the order of the coercive force of the material may be applied and, upon removal, approximately $97 \%$ of the flux is recovered. Subsequently applied demagnetizing fields equal to or less than the coercive force will have no further effect on the resultant flux after removal of the demagnetizing field. From this it is seen that a magnet working at any point on the demagnetization curve after having been subjected to a demagnetizing force will return to its original working point. This unique recoil property opens up new fields in permanent magnet applications.

As outlined above, the most efficient design is achieved by operating the magnet at $(\mathrm{BH})_{\text {max }}$. With conventional alloys this is in the main achieved by incorporating the magnet in a flux carrying or mild stecl assembly and to obtain optimum efficiency magnetization must be carricd out after assembly. Owing to the favourable recoil properties of isotropic barium ferrite, magnetization may be carried out prior to assembly. This obviates the necessity for complicated and often costly magnetizing jigs and fixtures.

In bicycle dynamos and generators using a permanent magnet rotor, it is common practice to magnetize an alloy magner betore assembly in a jig, from which it is pushed directly into a mild steel ring to act as a keeper to maintain a closed magnetic circuit. It has then to be transferred in a similar manner into the stator. When barium ferrite is used, nonc of these precautions is necessary as the rotor may be open-circuited after magnetization without impairing the performance of the generator. Furthermore, the rotor of such a generator or any barium ferrite magnet in any assembly may be totally removed from the circuit and replaced without the necessity to remagnetize the magnet to maintain the performance after reassembly. This allows for the assembly to be dismantled and reassembled if required.

A further property of barium ferrite arises from the combination of high coercivity and low permeability which makes it possible to place poles of opposite polarity in close proximity without progressive self-demagnetization effects occurring. As many as 24 poles can be impressed on the outer surfaces of a plain cylinder of lin diameter. The limiting factor to the number of possible poles
is the difficulty of producing a magnetizing fixture to provide sufficient field to obtain magnetic saturation.

Ceramic permanent magnets may also be used in the construction of d.c. motors to provide direct current fields. Here again the high coercivity and low recoil permeability play an important part. Owing to the low flux density of this material it is necessary to provide the maximum possible magnet area in the stator, and at the same time the high coercivity enables short radial lengths to be employed. In all magnetic circuits greater efficiency is obtained by placing the magnet material as close as possible to the gap. With these considerations in mind, the simplest and most efficient method of using isotropic barium ferrite for d.c. motor fields is to have the magnet in the form of plain cylinders or segments which in themselves constitute the pole pieces (Fig. 2). In view of the low permeability of this material it is necessary to provide a low reluctance return path to complete the magnetic circuit, conveniently in the form of a mild steel ring which also serves as the motor housing. The magnetic characteristics of the material allow air gaps between three and four times those used in normal motor construction, and also allow for variations in air gap. This means that magnets in the " as fired " condition may be used, so obviating a costly internal grinding operation. For larger motors where the flux provided by isotropic barium ferrite would be insufficient, the anisotropic variety can be employed, but in this case pole pieces would normally be required in order to provide a radial field. Three typical constructions are shown in Fig. 3 using anisotropic magnet b ocks. Similarly, isotropic and anisotropic ceramic magnets can be


Fig. 4. (a) Ring type loudspeaker assembly (b) Pot type assembly.
used to provide the field energy for flywheel magnetos such as are commonly used in small two-stroke engines.
Radio and Television Applications.-In the domestic radio and TV receiver field, the main use of permanent magnet is in moving coil loudspeaker asscmblies. The main advantage of the ceramic magnet over the alloy magnet in this field is an cconomic one due to the very low cost of the raw materials uscd. In loudspeaker design, only the anisotropic variety is considered because the
lower $\mathrm{B}_{w}$ of the isotropic material leads to designs having unwieldy and bulky dimensions.

The magnetic properties of anisotropic ceramic material lead to designs which, although having large magnet area, have exceptionally short axial length (Fig. 4(a)). This is important where space considerations are vital, as in midget portable receivers, tape recorders, etc. Two main constructions are used for loudspeaker magnet assemblies incorporating anisotropic barium ferrite, namely ring type and pot type, as shown in Fig. 4. The latter may only be used for low gap field strength assemblies, but has the advantage of having little or no external field. This factor assumes importance in television receivers where stray magnetic fields adversely affect the c.r.t. performance. This pot assembly requires a dual magnet construction in order to utilize barium ferrite to its best advantage dimensionally ${ }^{6}$. A similar magnet construction to loudspeaker assemblies is used for vibrators, moving-coil microphones, and for punch card actuators.

Extensive use has been made of ceramic magnets for television tube focusing units, the high coercivity enabling two isotropic rings axially magnetized to be used in opposition. Changing the


Fig. 5. Holding assemblies using ceramic magnets.
axial distance between the two rings alters the axial field strength distribution and also the working points of the two magnets, thereby changing the field strength of the magnetic lens.

Ceramic magnets are also used in conjunction with television picture tubes as ion trap and picture shift devices and as correction magnets for wide angle tubes.
Holding Applications.-Magnetic holding force is proportional to $\mathrm{B}^{2} \mathrm{~A}$, where B is the flux density at the contact faces, and A is the contact area. Taking for example a thin rectangular block of isotropic barium ferrite, there are several methods of utilizing this as a holding device, as shown in Fig. 5. The simplest method of magnetizing is shown in Fig. 5(a), where two poles are impressed on one contact face. Since the pull of a magnetic device is proportional to the product $\mathrm{B}^{2} \mathrm{~A}$, a greater pull is obtained by
impressing a number of poles on one face, as in Fig. 5(b). This is possible with barium ferrite due to the high coercivity which reduces the effects of selfdemagnetization. In order to achieve a higher $B$, advantage can be taken of the possible high magnetic saturation values of soft iron or mild steel. Typical arrangements are shown in Fig. 5(c), (d), and (e). Here it is seen that the mild steel is suitably proportioned to collect and concentrate the low density flux from a large area of magnet to provide a high flux density at the contact areas. Using a construction similar to Fig. 5(c), with an isotropic disc magnet inside a mild steel cup, an assembly 3 in diameter and $7 / 16$ in deep would lift approximately 70 lb , i.e. over one hundred times the assembly weight. Using an anisotropic magnet, the assembly would be 3 in diameter and $3 / 4$ in deep, and should lift over 1001 b . Construction Fig. 5(d) is widely used for applications like magnetic door catches as it has the advantage of occupying very little space. When used for such applications as magnetic separators, chucks and jigs, multi-assemblies as in Fig. 5(e) have the advantage of providing a greater number of poles in any given area than can be obtained using any other magnet material.
Magnets in the Presence of A.C. Fields.-The high resistance to demagnetization combined with a high resistivity allow ceramic magnets to be used in conjunction with fairly strong a.c. fields. Typical applications are vibrating armatures where the armature consists solely of ceramic magnet material. Alternatively, a mild steel armature inside an a.c. actuated coil may be used. This is made a close fit inside a ceramic magnet ring having mild steel pole pieces which polarize the central armature, so imparting a vibrationary motion. This principle is made use of in bells, small pumps, etc.

A similar principle applies to polarized relays. Other Applications.-There are many other uses for which ceramic magnets may be used with advantage due to their unusual magnetic characteristics, low weight, or for economic considerations. Amongst these may be cited: remote control services, magnetomechanical coupling possibly operating through gas-tight seals, arc blowouts, snap contact switches, travelling wave tubes, synchronized artificial eyes, toys and novelties. Barium ferrite may be embodied in plastic or rubber to form a flexible magnet material for such purposes as magnetic seals, etc.
Conclusion.-From the foregoing, it is seen that the advent of barium ferrite magnets has broadened the field of applications of permanent magnets in addition to providing an economically advantageous alternative, in some cases, to conventional alloy magnets. The unique properties of this material provide a stimulus for further investigation of its many and varied possibilities.

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2.-ACTIVE NETWORKS: VALVES AND TRANSISTORS

IN the first part of this article we introduced nodal analysis to new students, showing how it is complementary to mesh analysis. In mesh analysis we use constant-voltage generators and equate known voltages to find unknown current magnitudes. In nodal analysis we employ constant-current generators and equate known currents to find unknown voltage magnitudes.

In this second part we propose to indicate a method by which "active" networks (i.e. those which contain a generator such as a thermionic valve or transistor) may be solved. The method will be identical with that used for passive networks, except for one additional convention.
Network Theory with Valves.-Every student knows that although the anode characteristics of a valve are far from straight, it is still possible to work the valve in a linear mode by avoiding the lowest parts of these curves, where the curvature is worst, and by carefully selecting the position of the load line, so as to work in Class A1. If, in addition, the part of the characteristic worked over is small, we can almost certainly assume linearity with negligible error. Idealized triode and pentode characteristics are shown in Fig. 1(a) and (b).
Valve Admittance Parameters.-The primal equation for a valve is

$$
i_{a}=\mathrm{f}\left(v_{g}, v_{a}\right)
$$

This may be immediately developed into

$$
\delta i_{a}=\frac{\partial i_{a}}{\partial v_{g}} \delta v_{g}+\frac{\partial i_{a}}{\partial v_{a}} \delta v_{a}
$$

Both the partial differential coefficients have the dimensions of an admittance, hence the equation may be written
$g_{m} v_{g}+g_{a} v_{a}=i_{a}$
where the currents and voltages are all incremental: for our case we may say they represent the instantaneous values of the sinusoidal currents and voltages.
Conventional Development.-It is usual to develop the primal equation by specifying that a signal voltage be applied between the grid and cathode only, and that a load be placed in the anode circuit. The equation then becomes

$$
g_{m} v_{g}-g_{a}\left(i_{a} \mathrm{R}_{l}\right)=i_{a}
$$

where $-i_{a} \quad \mathrm{R}_{l}$ is the back e.m.f. across the load. $g_{a}$ is then inverted and expressed as a resistance $\left(r_{a}=1 / g_{a}\right)$ and the equation easily resolves to

$$
i_{a}=\frac{e_{g} g_{m} r_{a}}{r_{a}+\mathrm{R}_{l}}=\frac{\mu e_{g}}{r_{a}+\mathrm{R}_{l}}
$$

where $\mu$ may or may not have a positive sign, according to the convention adopted.

The next step shows how this equation may be represented by the circuit shown in Fig. 2(a), where $i_{a}$ is obviously the current flowing through the generator and load.

This, however, is only one way of interpreting the equation. Suppose we apply Norton's Theorem,


Fig. I. Idealized triode ( $a$ ) and pentode (b) characteristics.


Fig. 2. Voltage (a) and current (b) generator equivalent circuits of the valve amplifier.
as we did in the first half of this article, and transform, the valve from a "constant-voltage generator" to a constant-current generator. We obtain the circuit shown in Fig. 2 (b).

In this circuit $i_{a}$ is the current flowing in from the generator; it has a constant value for all loads from short circuit to open circuit; in the latter case it all flows out through $g_{a}$.
Nodal Development.-Returning now to the primal valve equation, we see that it has a nodal form since it gives current in terms of voltage multiplied by admittance. Since there are two voltages there must be two significant nodes and hence two equations, one for the input and one for the output. We may immediately write down the general set of nodal equations:-

$$
\begin{aligned}
& Y_{11} V_{1}+Y_{12} V_{2}=I_{1} \text { (node 1, input) } \\
& Y_{21} V_{1}+Y_{22} V_{2}=I_{2} \text { (node 2, output) }
\end{aligned}
$$

Now a valve is an active network: it extracts unidirectional energy from the power supply unit and transforms it into alternating energy. Thus we would hardly expect its nodal equations to fall out exactly as for a passive network. The differences, however, are not large.

Fig. 3 shows the basic ideal valve amplifier (no load resistor and no grid resistor) with the input and output nodes $N_{1}$ and $N_{2}$ marked in. A cursory glance at this figure will show us that $\mathrm{V}_{1}=v_{g}$ and $\mathrm{V}_{2}=v_{a}$, but I hope we will not now fall into the trap of writing $I_{2}=i_{a} . \quad I_{2}$ is the current supplied from an external generator to node 2. In this case

[^5]there is no such generator since we have specified one generator only, feeding a signal to the grid. $\mathrm{I}_{2}$ is therefore zero. However, it is important to note that although $\mathrm{I}_{2}$ is zero, this does not mean that no current flows into node 2 at all. Actually the anode current, $i_{a}$, flows from node 1 (the grid) to node 2 (the anode) and then to earth again, via $g_{a}$.

Now what of the coefficients $Y_{21}$ and $Y_{22}$ in the second nodal equation? $Y_{21}$ corresponds to current flowing into node 2 (the anode) due to a voltage ai node 1 (the grid). This, by definition, is $g_{n}$. $\mathrm{Y}_{22}$ is the admittance in the output circuit. For the case we are considering this, by definition, is $g_{a}$. In a practical amplifier $Y_{22}$ would include all the admittances in the output circuit, i.e. the load resistor, stray capacities etc. The second nodal equation for a valve is thus

$$
g_{m} \mathrm{~V}_{1}+g_{a} \mathrm{~V}_{2}=0(\text { node } 2)
$$

This is our primal equation in nodal form. We might have expected the current flowing into node 2 , $g_{n} \mathrm{~V}_{1}$, to have a negative sign, so as to agree with our selected conventions. It must be remembered, however, that this current is itself negative, since increasing $V_{1}$ positively increases the How of electrons or negative current, and a negative current flowing in a negative direction wil have a positive sign.

What of the equation for node l? The derivation of this will be quite straightforward. For the case where we have no grid resistor, and when we work in Class Al (no grid current), $\mathrm{I}_{1}$ must be zero. Hence, if $I_{1}$ is to be expressed in terms of $V_{1}$ and $V_{2}$, these must both be prefixed with a nought, which gives us

$$
0 V_{1}+0 V_{2}=I_{1}
$$

This just means that the valve itself draws no current from the generator driving it. Looking in detail at the coefficients $\mathrm{Y}_{11}$ and $\mathrm{Y}_{12}$ in this equation, $\mathrm{Y}_{11}$ is the admittance in the input circuit. With no grid resistor this will be zero. In a practical ampliner, however, the grid ad nittance $\mathrm{Y}_{g}$ will appear in $\mathrm{Y}_{11}$ and then $I_{1}$ will have a definite value. $\mathrm{Y}_{12}$ corresponds to current flowing into node 1 (the grid) due to a voltage at node 2 (the anode). For the perfect valve we are considering $Y_{12}$ must thus also be zero. (We will consider the practical case-"Miller Effect" -later.)
If we assemble our results for the ideal valve we obtain

$$
0 V_{1}+0 V_{2}=I_{1}(\text { node } 1)
$$

$g_{m} \mathrm{~V}_{1}+g_{a} \mathrm{~V}_{2}=0$ (node 2)
These equations show clearly the differences between active and passive networks.
(1) In a passive network $\mathrm{Y}_{12}$ always has the same magnitude and sign as $\mathrm{Y}_{21}$. In an active network this is not so. (Here $\mathrm{Y}_{12}=0$ and $\mathrm{Y}_{21}=g_{m}$.)


Fig. 3. Basic ideal valve amplifier.


Fig, 4. Simple two-node passive network with its associated nojal equations.


Fig. 5. Equivalent nodal circuits of a valve amplifier (a) in the ideal form of Fig. 3 and (b) with the addition of a grid admittance $Y_{g}$ and load admittance $Y_{l}$.
(2) In a passive network the linking admittance joining nodes 1 and 2 appears in $Y_{12}$ and $Y_{21}$ prefixed with a negative sign, and also in $Y_{11}$ and $Y_{22}$ with its normal sign. (An easy example of this is shown in Fig. 4, together with its set of equations.) In the active network this is not so. ( $g_{m}$ appears in $\mathrm{Y}_{21}$ only, and is not included in the other three parameters.)
Hence a separate sign is necessary to indicate the unilateral admittances of an active network. We could make use of the appropriate arrow symbol for a oneway street, pointing from $N_{1}$ to $N_{2}$ say for $Y_{21}$ and from $\mathrm{N}_{2}$ to $\mathrm{N}_{1}$ for $\mathrm{Y}_{12}$ in cases where it has a value other than zero (as in a transistor). The convention would be that where a unilateral admittance appears it is only shown once, in either $Y_{12}$ or $\mathrm{Y}_{21}$, according to the arrows, and with the sign shown. For example, the equivalent diagram of a valve only, in the ideal condition, would be as shown in Fig. 5(a). The valve with a grid admittance $\mathrm{Y}_{g}$ and a load admittance $\mathrm{Y}_{t}$ would have an equivalent circuit as in Fig. 5(b).

It only remains to state that any passive elements added to an active network behave as they normally do in a passive network, even if connected across nodes one and two, when they appear in all four parameters.

We can now proceed to apply these equations, and I think you will find this much easier than the derivation of the equations!

General Case of an Active Network.-We can now obtain some standard formulae for gain, input admittance and output admittance, which will apply to all such networks, passive or active, valve or transistor.

The general nodal equations for the case when there is no external generator connected to the output (node 2), i.e. with $I_{2}=0$, are

$$
\begin{aligned}
& Y_{11} V_{1}+Y_{12} V_{2}=I_{1} \\
& Y_{21} V_{1}+Y_{22} V_{2}=0
\end{aligned}
$$



Fig. 6. Valve with interelectrode capacities (a) and its nodal equivalent circuit (b).

It can immediately be seen from the second equation that the voltage gain

$$
\frac{\mathrm{V}_{2}}{\mathrm{~V}_{1}}=-\frac{\mathrm{Y}_{21}}{\mathrm{Y}_{22}}=\mathrm{M}_{v}
$$

Similarly the short-circuited (output) current gain will be

$$
\frac{I_{2}}{\mathrm{~V}_{1}} \times \frac{\mathrm{V}_{1}}{\mathrm{I}_{1}}=\frac{\mathrm{Y}_{21}}{\mathrm{Y}_{11}}=\mathrm{M}_{i}
$$

The input admittance of the network, which can include the load, may be found from the equations as follows. From the second equation $\mathrm{V}_{2}=-\mathbf{Y}_{21} \mathrm{~V}_{1} /$ $\mathrm{Y}_{22}$ and substituting this into the first equation gives

$$
\mathrm{Y}_{11} \mathrm{~V}_{1}-\frac{\mathrm{Y}_{12} \mathrm{Y}_{21}}{\mathrm{Y}_{22}} \mathrm{~V}_{1}=\mathrm{I}_{1}
$$

Thus $\mathrm{Y}_{i n}=\frac{\mathrm{I}_{1}}{\mathrm{~V}_{1}}=\mathrm{Y}_{11}-\frac{\mathrm{Y}_{12} \mathrm{Y}_{21}}{\mathrm{Y}_{22}}=\mathrm{Y}_{11}+\mathrm{M}_{v} \mathrm{Y}_{12}$
Conversely, the " looking-back " admittance at the output will be

$$
Y_{0}=Y_{22}-\frac{Y_{12} Y_{21}}{Y_{11}}=Y_{22}-M_{i} Y_{12}
$$

From these equations we can see that the load only affects the source admittance, and vice versa, if both $\mathrm{Y}_{12}$ and $\mathrm{Y}_{21}$ have valucs (i.e. neither arc zero).
Simple Valve Amplifier.-The nodal equations for the valve under ideal conditions are

$$
\begin{aligned}
& 0 \mathrm{~V}_{1}+0 \mathrm{~V}_{2}=\mathrm{I}_{1} \\
& g_{m} \mathrm{~V}_{1}+g_{a} \mathrm{~V}_{2}=0
\end{aligned}
$$

The voltage gain $=-g_{m} / g_{a}=-g_{m} r_{a}=-\mu$.
The nodal cquation brings out clearly to the new student the fact that the valve amplification factor is merely the ratio of two admittances.

Now take a simple amplifier, with anode load $\mathrm{Y}_{l}$. Since the valve is being treated as a current generator the load will appear in parallel with the generator admittance. If we include the grid resistor ( $1 / \mathbf{Y}_{g}$ ) our equivalent diagram will be as in Fig. 5(b), and our equations will become

$$
\begin{aligned}
& \mathrm{Y}_{g} \mathrm{~V}_{1}+0 \mathrm{~V}_{2}=\mathrm{I}_{1} \\
& g_{m} \mathrm{~V}_{1}+\left(g_{a}+\mathrm{Y}_{i}\right) \mathrm{V}_{2}=0
\end{aligned}
$$

The voltage gain

$$
\mathrm{M}_{v}=-\frac{\mathrm{Y}_{21}}{\mathrm{Y}_{22}}=\frac{-g_{m}}{g_{a}+\mathrm{Y}_{i}}
$$

Unfamiliar? Change the admittances to impedances and

$$
\begin{aligned}
\mathrm{M}_{v} & =\frac{-g_{m}}{\frac{1}{r_{a}}+\frac{1}{\mathrm{R}_{l}}}=\frac{-g_{m}}{\frac{\mathrm{R}_{l}+r_{a}}{\mathrm{R}_{l} r_{a}}} \\
& =\frac{-g_{m} r_{a} \mathrm{R}_{l}}{\mathrm{R}_{l}+r_{a}}=\frac{-\mu \mathrm{R}_{l}}{\mathrm{R}_{l}+r_{a}}
\end{aligned}
$$

which is familiar to everyone-yet not as simple. If $\mathrm{Y}_{l}$ should be complex, say $g_{l}+j \mathrm{~B}_{l}$, then

$$
\mathrm{M}_{v}=\left|\frac{g_{i n}}{\sqrt{\left(g_{a}+g_{l}\right)^{2}+\mathrm{B}_{i}{ }^{2}}}\right| 180^{\circ}-\theta
$$

where $\theta=\arctan \frac{\mathrm{B}_{l}}{g_{a}+g_{l}}$
Calculation of Miller Effect.-Due to interelectrode capacitance a valve, particularly a triode, is not truly unilateral at high frequencies, but does transfer some energy from the output to the input, and so modify the input admittance of the valve. The basic circuit and its equivalent are shown in Fig. 6(a) and (b) and the pair of nodal equations will be

$$
\begin{aligned}
& {\left[\mathrm{Y}_{g}+\mathrm{j} \omega\left(\mathrm{C}_{g c}+\mathrm{C}_{g a}\right)\right] \mathrm{V}_{1}-\mathrm{j} \omega \mathrm{C}_{g a} \mathrm{~V}_{2}=\mathrm{I}_{1}} \\
& \begin{array}{l}
{\left[g_{m}-\mathrm{j} \omega \mathrm{C}_{s a}\right] \mathrm{V}_{1}+} \\
{\left[g_{a}+\mathrm{Y}_{l}+\mathrm{j} \omega\left(\mathrm{C}_{a c}+\mathrm{C}_{a}\right)\right] \mathrm{V}_{\mathbf{2}}=0}
\end{array}
\end{aligned}
$$

The voltage gain

$$
\mathrm{M}_{v}=-\frac{\mathrm{Y}_{21}}{\mathrm{Y}_{22}}=-\frac{g_{m z}-\mathrm{j} \omega \mathrm{C}_{g a}}{g_{a}+\mathbf{Y}_{l}+\mathrm{j} \omega\left(\mathrm{C}_{a c}+\mathrm{C}_{g a}\right)}
$$

The input admittance

$$
\begin{aligned}
& \mathrm{Y}_{i n}=\mathrm{Y}_{11}+\mathrm{M}_{v} \mathrm{Y}_{12} \\
&=\mathrm{Y}_{g}+\mathrm{j} \omega\left(\mathrm{C}_{g c}+\mathrm{C}_{g a}\right)-\mathrm{j} \omega \mathrm{C}_{g a}\left|\mathrm{M}_{b}\right| \\
&=\mathrm{Y}_{g}+\mathrm{j} \omega \mathrm{C}_{g c}+\omega \mathrm{C}_{g a}\left(1 / 90^{\circ}+\underline{\mathrm{M}_{v}}+\underline{190^{\circ}+\phi}\right.
\end{aligned}
$$

This is the normal equation for the input admittance of a triode amplifier at radio frequencies; we have found it, together with the gain, in a few lines of working. Other examples of valves working with grounded cathode can be solved in the same manner.

## Grounded-Anode Stage (Cathode Follower).-

Let us now examine the grounded-anode stage. Here, due to the valve being inverted, the nodal equations must be modified. The primal equation, $i_{a}=$ $g_{m} v_{g}+g_{a} v_{a}$ always applies, whatever the configuration. Referring to Fig. 7 we see that $v_{a}=-V_{2}$, and since $v_{a c}=v_{a g}+v_{g c}$, then $v_{q c}=v_{a c} \overline{V_{v}} v_{a g}$ $=-\mathrm{V}_{2}-\left(-\mathrm{V}_{1}\right)$. Hence $i_{a}=-g_{m}\left(\mathrm{~V}_{2}-\mathrm{V}_{1}\right)-$ $g_{a} \mathrm{~V}_{2}$. Allowing for the reversed direction of $i_{a}$, the nodal equations then become

$$
\begin{aligned}
& 0 \mathrm{~V}_{1}+0 \mathrm{~V}_{2}=\mathrm{I}_{1} \\
& -g_{m} \mathrm{~V}_{1}+\left(g_{m}+g_{a}\right) \mathrm{V}_{2}=0
\end{aligned}
$$

These equations are easily remembered, whilst the procedure for finding the gain etc., is exactly as before. Let us draw the circuit of a typical cathode follower and its equivalent, and then write down the nodal equations. The circuits are shown in Fig. 8(a) and (b) and the nodal equations are

$$
\begin{aligned}
& \mathbf{Y}_{-} \mathbf{V}_{1}+0 \mathbf{V}_{2}=\mathrm{I}_{1} \\
& -g_{m} \mathbf{V}_{1}+\left(g_{m}+\mathbf{Y}_{a}\right) \mathbf{V}_{2}=0
\end{aligned}
$$



Fig. 7. Basic ideal grounded-anode valve (cathode follower).
The voltage gain

$$
\mathbf{M}_{v}=-\frac{\mathrm{Y}_{21}}{\mathrm{Y}_{22}}=\frac{g_{m}}{g_{m}+g_{a}+\mathrm{Y}_{k}}
$$

This is obviously less than one. Typical values may be, $g_{m}=2.5$ milliamps per volt, $g_{a}=100$ micromhos and $\mathrm{Y}_{k}=100$ micromhos. The gain is then 0.925 .

If the gain equation is transformed into impedances we obtain for the voltage gain
$\mathrm{M}_{v}=\frac{\left(\frac{\mu}{1+\mu}\right) \mathrm{R}_{k}}{\left(\frac{r_{a}}{1+\mu}\right)+\mathrm{R}_{k}}$
This is the familiar equation, which shows voltage negative feedback.

Since $Y_{12}$ is zero, the input admittance is just $\mathrm{Y}_{g}$ and the output admittance $g_{m}+g_{a}+\mathrm{Y}_{k}$. However, things become quite different if $\mathrm{Y}_{g}$ is connected between grid and cathode (i.e. between nodes 1 and 2). The input admittance then falls to about one-tenth of $\mathrm{Y}_{g}$ and the output admittance is now no longer independent of the input admittance. Why not work it out for yourself? There are some good notes by J. McG. Sowerby in the Wireless World for September, 1948.

The behaviour of the cathode follower at high frequencies can be studied by including the interelectrode and other capacitances.

Grounded-Grid Amplifier.-Since the procedure followed is identical with that for the cathode follower, the explanation will be curtailed. Referring to Fig. 9 we see that $v_{g}=-\mathrm{V}_{1}$ and $v_{a}=\mathrm{V}_{2}-\mathrm{V}_{1}$. Substituting we find that $i_{a}=-\left(g_{m}+g_{a}\right) V_{1}+g_{a}$ $V_{2}$. Now $I_{1}$ must have the same magnitude as $i_{a}$ (there being no grid current). Therefore it should have the same coefficients as $i_{a}$, but with the signs changed, since current flowing out of node 1 is positive, whereas the same current flowing into node 2 is negative. Hence our nodal equations are

$$
\begin{aligned}
& \left(g_{m}+g_{a}\right) \mathrm{V}_{1}-g_{a} \mathrm{~V}_{2}=\mathrm{I}_{1} \\
& -\left(g_{m}+g_{a}\right) \mathrm{V}_{1}+g_{a} \mathrm{~V}_{2}=0
\end{aligned}
$$

A simple circuit and its equivalent are shown in Fig. 10(a) and (b). Its equations are

$$
\begin{aligned}
& \left(\mathrm{Y}_{k}+g_{m}+g_{a}\right) \mathrm{V}_{1}-g_{a} \mathrm{~V}_{2}=\mathrm{I}_{1} \\
& -\left(g_{m}+g_{a}\right) \mathrm{V}_{1}+\left(g_{a}+\mathrm{Y}_{l}\right) \mathrm{V}_{2}=0
\end{aligned}
$$

The voltage gain $=-\frac{\mathrm{Y}_{21}}{\mathrm{Y}_{22}}=\frac{g_{m}+g_{a}}{g_{a}+\mathrm{Y}_{l}}$
If $Y_{l}$ and $g_{m}$ are much larger than $g_{a}$, this gain approximates to $g_{m} / \mathrm{Y}_{l}$ or $g_{m} \mathrm{Z}_{l}$.

The input admittance will be $\mathrm{Y}_{11}+\mathrm{M}_{v} \mathrm{Y}_{12}$ or approximately $\mathrm{Y}_{k}+g_{m}$, for the same conditions as the gain.

Indefinite Admittance Matrix.-All the basic valve configurations have now been dealt with. We will conclude this section by giving a mnemonic to enable you to remember the three sets of equations.

First write down the admittance parameters of the common-cathode valve configuration in the position they normally occupy in the equations. We get

0
$g_{m}$
0
Now add a third column (on the right) and a third row (underneath) in such a way as to make the sum of every column or row zero. Lastly mark the rows and columns $G, A$ and $C$, as shown.

|  | G | A | C |
| :---: | :---: | :---: | :---: |
| G | 0 | 0 | 0 |
| A | $g_{m}$ | $g_{a}$ | $-\left(g_{m}+g_{a}\right)$ |
| C | $-g_{m}$ | $-g_{a}$ | $\left(g_{m}+g_{a}\right)$ |

For grounded-cathode working strike out the third (C) row and column. This brings us back to where we started, viz.
$0 \quad 0$
$g_{m} \quad g_{a}$

For grounded-anode working strike out the second (A) column and row, which gives

$$
\left.\begin{array}{cc}
0 & 0 \\
-g_{m} & .
\end{array} g_{m}+g_{a}\right)
$$

This agrees with the values already found.


Fig. 8. Typical cothode-follower (a) and its nodal equivalent circuit (b).


Fig. 9. Basic ideal grounded-grid valve.
Lastly, for grounded grid, we cross out the first (G) row and column to obtain

|  | A | C |
| :---: | :---: | :---: |
| A | $g_{a}$ | $-\left(g_{m}+g_{a}\right)$ |
| C | $-g_{a}$ | $\left(g_{m}+g_{a}\right)$ |

This places the anode in the top left-hand corner, which is input, and the cathode in the bottom righthand corner, or output. We therefore make a diagonal changeover of all four parameters to get

$$
\begin{array}{lc}
\left(g_{m}+g_{a}\right) & -g_{a} \\
-\left(g_{m}+g_{a}\right) & g_{a}
\end{array}
$$

This again agrees with the values previously found.
You may protest that these equations are easy enough to remember without this mnemonic; if so I would agree. The real reason they were brought in here was to show that the method gave the right
(Continued on page 603)
answers with a valve. We will soon use the same method to convert a common-emitter configuration transistor to common collector or common base. The real justification for this method lies in matrix theory, and is due to J. Shekel ${ }^{1}$. This subject, however, is outside the scope of the present article.
Transistor as an Active Network. - The transistor, like the thermionic valve, is an active two-node network and as such its manipulation and basic equations for gain and input and output admittance are the same as the valve. Once we have obtained the nodal equations for the transistor the rest is merely a repetition of what we did with the valve.
Derivation of Transistor Parameters.-Following common practice, we will first consider the transistor in common base. The transistor collector curves, shown in Fig. 11(a) strongly resemble the pentode anode characteristics shown in Fig. 1(b), with one subtle difference. Whereas each separate curve on the valve characteristic represents a certain grid voltage, on the transistor collector characteristic each curve represents a certain emitter current. This makes quite a difference to the resulting equations. We must now say

$$
i_{c}=\mathrm{f}\left(i_{e}, v_{c}\right)
$$

Our other parameter, the emitter voltage, must then also be expressed in terms of the same variables.

Once again we get equations of the form

$$
\begin{aligned}
& \delta v_{e}=\frac{\partial v_{e}}{\partial i_{e}} \delta i_{e}+\frac{\partial v_{e}}{\partial v_{c}} \delta v_{c} \\
& \delta i_{c}=\frac{\partial i_{e}}{\partial i_{e}} \delta i_{e}+\frac{\partial i_{c}}{\partial v_{c}} \delta v_{c}
\end{aligned}
$$

$\partial v_{e} / \partial i_{e}$ is dimensionally an impedance and we may replace it with $r_{11}$. Similarly $\partial i_{c} / \partial v_{c}$ is dimensionally a conductance and may be replaced by $g_{22}$. On the other hand $\partial v_{e} / \partial v_{c}$ is a mere ratio between two voltages, let us use $\mu_{\tau}$ for this, where the subscript means "reverse" (voltage fed back to input). In the same way $\partial i_{c} / \partial i_{e}$ is the forward current gain and its symbol is usually $\alpha$.

Our set of equations thus becomes

$$
\begin{aligned}
& \mathrm{r}_{11} i_{e}+\mu_{r} v_{c}=v_{e} \\
& \alpha i_{e}+\mathrm{g}_{22} v_{c}=i_{c}
\end{aligned}
$$

with the signs depending on the convention employed.
By the reasoning employed in the first article, $r_{11}$ and $\alpha$ are measured with the output short circuited to a.c. whereas $\mu_{r}$ and $g_{22}$ are measured with the input open circuited to a.c.

Now it happens that in matrix theory there is a set of equations of the form

$$
\begin{aligned}
& \mathrm{H}_{11} \mathrm{I}_{1}+\mathrm{H}_{12} \mathrm{~V}_{2}=\mathrm{V}_{1} \\
& \mathrm{H}_{21} \mathrm{I}_{1}+\mathrm{H}_{22} \mathrm{~V}_{2}=\mathrm{I}_{2}
\end{aligned}
$$

which is of the same basic form as those we have just derived. Thus the next step was to substitute $\mathrm{H}_{11}$ for $r_{11}, \mathrm{H}_{12}$ for $\mu_{r}, \mathrm{H}_{21}$ for $\alpha$ and $\mathrm{H}_{22}$ for $g_{22}$. It wasn't really necessiry, but you must admit it looks good and is calculated to baffle the most pertinacious student. Indeed we have now reached a dead end and will have to transform these "hybrid" parameters (from the H and the mixed dimensions) into admittances, similar to the valve, if we wish to attack transistor problems without the aid of a course in memory training.

We can easily transform the H parameters into $Y$ by the aid of simple algebra, but if the transistor manufacturers would quote these values it would save us the trouble.
'Wireless Engineer, Jan. 1954, p. 6.


Fig. 10. Simple grounded-grid amplifier (a) and its nodal equivalent circuit (b).

(a)

(b)

Fig. 11. Common-base characteristics of the transistor (a) at its output and (b) at its input.

From the first equation we can quickly obtain

$$
\mathrm{I}_{1}=\frac{1}{\mathrm{H}_{11}} \mathrm{~V}_{1}-\frac{\mathrm{H}_{12}}{\mathrm{H}_{11}} \mathrm{~V}_{2}
$$

Substituting this into the second equation we get

$$
\mathrm{I}_{2}=\frac{\mathrm{H}_{21}}{\mathrm{H}_{11}} \mathrm{~V}_{1}+\frac{|\mathrm{H}|}{\mathrm{H}_{11}} \mathrm{~V}_{2}
$$

Where $|\mathbf{H}|=\mathrm{H}_{11} \mathrm{H}_{22}-\mathrm{H}_{12} \mathrm{H}_{21}$
Hence

$$
Y_{11}=\frac{1}{H_{11}}, Y_{12}=-\frac{H_{12}}{H_{11}}, Y_{21}=\frac{H_{21}}{H_{11}}, Y_{22}=\frac{|H|}{H_{11}}
$$

Transistor manufacturers usually publish a set of H parameters for common-emitter, common-collector and common-base working-three sets-for each transistor.

It is common for teachers and lecturers to approach transistors from the common-base viewpoint, but in practice the common-emitter configuration is far more common. We will therefore give our example on the common-emitter configuration. Commonemitter parameters may be found in the same way as common base, and they are always given by transistor manufacturers. In any case it is always possible to transform from one configuration to another.
(a)

(b)


Fig. 12. Common-emitter transistor current amplifier (a) and its nodal equivalent circuit (b) ( $火 \equiv$ mho).

## Common-Emitter Transistor Amplifier. -

Mullards give the hybrid parameters of their OC 71 transistor in common emitter as

$$
\begin{array}{ll}
\mathrm{H}_{11}=800 \text { ohms, } \mathrm{H}_{12}=5.4 \times 10^{-4} \\
\mathrm{H}_{21}=47, & \mathrm{H}_{22}=80 \times 10^{-6} \mathrm{mhos}
\end{array}
$$

The admittance parameters work out to
$\left.\mathrm{Y}_{11}=1,250, \mathrm{Y}_{12}=-0.675\right\}$ all figures in

$$
\left.\mathrm{Y}_{21}=58,750, \mathrm{Y}_{22}=48.275\right\} \text { micromhos }
$$

Consider the amplifier shown in Fig. 12(a).
A little consideration will show that both $Y_{1}$ and $Y_{2}$ are in parallel across the input, whilst $Y_{4}$ is obviously across the output. The equivalent circuit is then as in Fig. 12(b).
The voltage gain is still

$$
M_{v}=-\frac{Y_{21}}{Y_{22}}=-\frac{58,750}{148.275}=-396
$$

The current gain, with output short circuited to a.c.

$$
M_{i}=\frac{Y_{21}}{Y_{11}}=\frac{58,750}{1,470}=40
$$

The input admittance
$Y_{i n}=Y_{11}-\frac{Y_{12} Y_{21}}{Y_{22}}=Y_{11}+M_{v} Y_{12}=1,737 \mu$ mhos
The "looking-back" admittance at the output

$$
Y_{0}=Y_{22}-\frac{Y_{12} Y_{21}}{Y_{11}}=Y_{22}-M_{i} Y_{12}=175.3 \mu \mathrm{mhos}
$$

Perhaps you will agree that once the admittance parameters have been worked out, or obtained from manufacturer's data, the actual work involved in finding the gain, etc., is small.

Transistor in Common Collector and Common Base.-If one wishes to obtain the performance of a transistor in common collector (near equivalent to the cathode follower) or common base (equivalent of grounded grid), the admittance parameters for these configurations can easily be obtained from those
of the grounded-emitter admittances, without recourse to the hybrid parameters again. We employ the indefinite admittance matrix, as we did for the valve, using a modification of the method devised by Shekel ${ }^{2}$.

First write down the admittance parameters of the common-emitter configuration, in the positions which they normally occupy in the equations. Now add a third column (on the right) and a third row (underneath) in such a way as to make the sum of every column or row zero. Lastly mark the rows and columns, B, C and E. I have done it for the OC 71.

|  | B | C | E |  |
| :---: | :---: | :---: | :---: | :---: |
| B |  | $-0.675$ | - 1,249.325 |  |
| C | 58,7 | 48.275 | $-58,798.275\}$ | romhos |
| E | -60,0 | -47.600 | 60,047.6 | micromhos |

To obtain common-emitter parameters strike out the E row and column. (This brings us back to our starting point, of course.)

To obtain the common-collector parameters strike out the $C$ row and column. This leaves

$$
\left.\begin{array}{cc}
1,250 & -1,249.325 \\
-60,000 & 60,047.6
\end{array}\right\} \begin{aligned}
& \text { all in } \\
& \text { micromhos }
\end{aligned}
$$

which are the admittance parameters for the OC 71 when used in grounded collector. They are used exactly as were the valve admittance parameters in the cathode follower.

For the common-base configuration we strike out the first (B) row and column. This gives the collector in the top left-hand corner, which is input, and the emitter in the bottom right corner, or output: our transistor is connected the wrong way round. It is therefore necessary to make a diagonal switch of all four parameters which gives

$$
\left.\begin{array}{cr}
60,047.6 & -47.600 \\
-58,798.275 & 48.275
\end{array}\right\} \text { micromhos }
$$

Again, these common-base parameters may be used to find gain, input admittance and output admittance exactly as shown for the common emitter, or for that matter, the valve.
Conclusion and the Shape of Things to Come.This then is nodal analysis as applied to passive and active networks. The basic procedure is the same for passive, valve or transistor circuits, which should be of considerable help to students. Only a few examples of its use are shown, it really is a powerful method: however, it does not supersede mesh analysis, but is complementary to it.

And where do we go from here? The next step would be a study of four-terminal networks, using matrices, assisted by determinants. If you should be interested in the transient case, nodal analysis may still be very helpful, in conjunction with Laplace transforms, and will tell you whether a network is stable or unstable, and what sort of response (current) will occur as a result of an externally applied stimulus (which is commonly non-sinusoidal) ${ }^{3}$.

However, leave these latest methods until you are quite familiar with mesh and nodal analysis: between them they will take you a long way, especially the latter!

This article is published with the permission of Lt. Col. J. Harris, C.O., 3 Trg. Bn., R.E.M.E. Arborfield.

[^6]
# Camera Remote Control 

USED IN B.B.C. REPORTING STUDIOS

By G. W. H. LARKBY*

IN recent years the stability of vidicon cameras has improved to the degree that it is now possible to use them without the attention of technical operators. The first example of the use of an unattended camera in this fashion was, in fact, in the B.B.C.'s Parliamentary Reporting Studio at Westminster which was put into service about a year ago.
This studio is as simple as it is possible to conceive a television studio to be, consisting of a small room furnished with a table and chair, lights, microphone and a single television camera in a fixed position. The studio is brought into action by remote control from the Television News Studio at Alexandra Palace where the received signal is "mixed" into the programme. The operational success of this facility, simple as it is, stimulated a demand to be able to control the main positional functions of the remote camera in order to give variety to the presentation. In turn, the development of the apparatus for this positional control has led to the provision of a facility whereby the camera can be made to take up any one of a number of pre-set "shots" by the operation of a push-button switch, which may, if desired, be operated by the "subject" himself.

The camera functions which require to be controlled are horizontal angle (pan), vertical angle (tilt), angle of view (focal length of lens), camera focus and lens aperture. In order to avoid a hiatus in the picture while changing the focal length of the lens, advantage can be taken of a zoom lens.
As a field experiment, a vidicon camera has been fitted with a zoom lens and is installed in the B.B.C.'s studio in All Souls Hall, near Broadcasting House, and the camera can be controlled in all its main functions, either from the local control room or from the control room at Alexandra Palace, some seven miles away.

The camera is a standard vidicon, of a type in general use in interview studios where it is normally fitted with a four-lens turret. For the purpose of this experiment, the turret was removed and replaced by a cast plate on which is mounted the zoom lens together with the motor mechanisms for the zoom, focus and iris movements.

The servo system employed is shown in Fig. 1. The two potentiometers, A and B , form a bridge circuit of which the potential difference, at d.c.,
between points $A$ and $B$, is fed to a modulator, which detects the potential difference between these two points and provides a correcting signal which, in turn, operates a motor so that the two points are brought as nearly as possible to the same potential. The modulator circuit employed is the well-known "ring" connection of four diodes to which, in addition to the d.c. control voltage, a constant $50 \mathrm{c} / \mathrm{s}$ e.m.f. is also connected. In order to keep harmonics in the output voltage to a reasonably low level the amplitude of the 50 c ,'s applied to the modulator is relatively low. It can be seen from the symmetry of the circuit that at balance, i.e., when A and B are at the same potential, there will be no $50 \mathrm{c} / \mathrm{s}$ output from the bridge into the control amplifier. On the other hand, when potentiometer A is moved, say to give a current flowing in the direction from B to A, diodes M1 and M3 conduct, causing the $50 \mathrm{c} / \mathrm{s}$ e.m.f. in transformer T 1 to send a current through the centre-tapped winding of the transformer T 2 , producing an output in a particular phase. Movement of potentiometer A in the other direction, so as to cause a current to flow from A to $B$, makes diodes M2 and M4 conduct and, as compared with the previous condition, reverses the connection of the e.m.f. in transformer T1 to the terminals of transformer T2. Thus the phase of the output of the voltage from T2 changes by $180^{\circ}$ as the direction of current between A and B reverses. Accordingly, the motor can run in either direction, depending upon the position of potentiometer B relative to potentiometer A. The a.c. output from the bridge is amplified and drives the servo control motor, which will always attempt to keep the bridge

* B.B.C. Designs Department.


Fig. 2. Servo motors, zoom lens, etc., mounted in place of the normal four-lens turret. These are enclosed in a cover which olso forms a lens hood.
in balance. Thus, any movement of the control knob results in an equal movement of the controlled potentiometers as the motor drives the bridge into balance. Coupled to the motor shaft is a generator which produces an output of voltage proportional


Fig. 3. Remote control console. The two large knobs control focus and zoom (left) and tilt and pan (right).
to the speed which is fed into the servo amplifier and so adjusted as to make the whole system critically damped. The action is very nearly dead beat giving the operator the impression that the knob is directly coupled to the mechanism which it operates and he has little or no feeling that it may well be many miles away.
The use of d.c. for the control current has the advantage that for short distances, say of the order of 10 miles, low quality circuits may be used. The system as installed between Alexandra Palace and All Souls Studio makes use of phantom circuits superimposed on existing speech and music circuits. For longer distances on which a d.c. loop is not available, it would be necessary to translate the control signals to audio frequencies, at which they can be conveniently transmitted.
The control amplifiers are four identical transistorized units capable of giving a continuous output of 20 watts each. They are driven from a stabilized 24 -volt power supply. The amplifiers are straightforward audio amplifiers with emitter follower driven Class B output stages, the transistors being OC71, OC71, OC72+OC72 driving $2 \mathrm{~N} 268+2 \mathrm{~N} 268$. There are two controls, i.e., gain and velocity feedback, thus allowing the performance of any function to be varied at will.
The lens fitted has a minimum focus of 4 feet with a maximum aperture of $f / 2.8$. The movement of the knurled operating rings required to go from minimum to maximum of all three functions of the lens is approximately $180^{\circ}$, and the preliminary design called for an operational time of 5 seconds end to end. This sets the reduction gear ratio for the servo motor, the stalled torque necessary at motor shaft being sufficient to overcome the friction of the mechanism. In view of the experimental nature of this camera only two positional servos were used on the lens, i.e., focus and zoom. The iris is controlled by a non-servo type of small reversible motor of low torque, end stops being used to limit travel. The servo balance potentiometers are mounted on the penultimate drive shaft; the final drive to the lens is by neoprene toothed timing belts. These belts have built-in flexible steel wires which prevent stretch, the teeth giving positive drive so that a minimum of side loading is required and they are silent in operation.

All the connections to the motors, etc., are brought out to a multi-way plug on the side of the plate. A cover with acoustic damping fits over the whole assembly and also operates as a lens hood.
The camera and lens assembly is mounted on a panning tilt head which has a tilting centre above the camera mounting platform so that the centre of gravity of the camera can be placed at this point, thereby making the whole assembly inherently balanced. The camera tilt mechanism consists of a sector of a circle whose centre is above the camera mounting surface. Gear teeth are cut in the curved underside of this sector and mesh with a transverse driving shaft. This mechanism can be seen in Fig. 2.
A servo motor is fitted to both pan and tilt mechanisms. The error detector in the tilt servo is a carbon track potentiometer geared to the transverse shaft, but that in the pan is a three-turn potentiometer to give adequate resolution. Velocity feedback is achieved by using a generator on the same shaft as the motor, as in the focus and zoom unit.

The pan motion uses an 8 mm roller chain with a spring-loaded jockey pulley. In this case the driven chain wheel is fixed to the tripod head and the motor and driving pulley drive themselves round the head carrying the camera with them. The use of this method permits a change of gear ratio to be used without the use of large, expensive, special gears.
For continuous operation of the camera controls at the remote point, the operator sits at a console on the front of which are two knobs; each of these is moved in two modes: (a) rotation, (b) fore and aft like a lever. On the right-hand knob the rotation corresponds to pan and the fore and aft motion to tilt; on the left-hand knob the rotation corresponds to focus and the fore and aft to zoom (see Fig. 3). With a little practice, by the use of both hands the operator can control all four functions of the camera smoothly and comfortably.

It was soon apparent that to give smooth control, some form of damping on the motion is required, and this is provided by discs immersed in a silicone fluid rotating in a close-fitting chamber. The physical operation of each of these controls turns a potentiometer which forms part of the bridge of the appropriate servo mechanism.

Because the setting of any of the camera functions can be represented by the position of a potentiometer, by arranging to switch to groups of potentiometers, it is possible to give the camera a number of pre-set "shots." For example, Shot No. 1 might consist of a general view showing two people in conversation, Shot No. 2 would then be a close-up of the interviewer and Shot No. 3, perhaps, a close-up of the person being interviewed. On the operation of a push button, the camera can


Fig. 4. "Shot-box" for the control of a camera by an interviewer.
be made to take up any of these shots at will. This leads to the possibility in studios where the programme is of a pre-set nature of using say, two remotely controlled cameras which look in turn at the appropriate "shots" under the control of the vision operators in the control room without having to have studio cameramen at all.

To pre-set any particular shot, it is only necessary to adjust the corresponding potentiometers for pan, tilt, focus and zoom, there being a group of four such potentiometers for each shot required. On the experimental camera the four positions on the "shot-box" control have proved adequate and useful. Visitors to one of the B.B.C. stands at the Earls Court Radio Show had the opportunty of manipulating a camera in this manner.

# SHOHT=WAVE CONIDITIONS 

## Prediction for December



G.M.T.

G.M.T.
-•••••• FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE FOR $25 \%$ OF THE TOTAL TIME

-     - predicted median standard maximum usable frequency
- fREQuency below which communication should be possible

ON ALL UNDISTURBED DAYS

## The Editor does not necessarily endorse the opinions expressed by his correspondents

## Transistorized Wien Bridge Oscillator

ON PAGE 390 of the August issue Mr. Butler says, "There is little point in using RC oscillators at frequències much higher than $100 \mathrm{kc} / \mathrm{s}$, since it is simpler and cheaper to use switched tuned circuits."

Perhaps Mr. Butler would care to expand on this statement, since (1) there is some merit in being able to have sweep or continuous tuning. The usual Wien network oscillator has a continuous 10:1 frequency sweep, and this is not normally achieved in an LC tuned system and, (2) there are several RC oscillators commercially available extending beyond $100 \mathrm{kc} / \mathrm{s}$.

Dorking, Surrey.
D. J. COLLINS

The author replies:
I AGREE with Mr. D. J. Collins that a normal LC oscillator will not cover the $10: 1$ frequency range of an RC oscillator and that there are occasions when this wide coverage is of value in experimental work. Nevertheless the chief merit of the RC oscillator lies in the lower frequency part of the spectrum where LC oscillators require large and inconvenient tuning elements. At higher frequencies LC oscillators with continuously variable tuning elements have these advantages:-
(a) Low distortion due to the use of high $Q$ components.
(b) High efficiency as compared with RC oscillators.
(c) Satisfactory operation from low voltage unregulated power supplies.
(d) High stability of frequency can be obtained in a simple way, e.g., by the use of capacitors having a negative temperature coefficient to offset the positive coefficient of the associated inductance.
(e) Distortionless amplitude control is easy to apply.
(f) Pulsed or modulated operation is possible.
(g) For a given scale length, the frequency of an LC oscillator covering a $3: 1$ range can be read more accurately than that of an RC oscillator covering a $10: 1$ range.

As stated in my paper, there is no difficulty in adding an extra range covering up to $200 \mathrm{kc} / \mathrm{s}$ if this is required. For distortionless operation at such high frequencies it is desirable to use v.h.f. transistors. This is because of difficulties due to hole-storage effects in transistors which have relatively low alpha cut-off frequencies. Moreover, v.h.f. transistors have a much lower collector capacitance. This parameter is dependent on temperature and on the collector voltage and is responsible for a variable amplifier phase shift which affects the frequency.

Mr. Alan Carpenter's fixed frequency $100 \mathrm{c} / \mathrm{s}$ oscillator (October issue) embodies some interesting design features and is outstandingly good in respect of frequency stability. At this low frequency, variation of transistor characteristics (iike input and output capacitance) can be swamped by the much larger capacitances in the bridge elements. Even so, to achieve a stability better than 1 part in 1,000 is an excellent result.

My only criticism of the circuit is perhaps a trivial one. It concerns the method used for amplitude stabilization. As Mr. Carpenter states, the low heater resistance of the indirectly-heated thermistor shunts the emitter load of V3 and absorbs considerable signal power, at the same time introducing a rather long time constant into the control circuit operation. Although a $100 \mu \mathrm{~F}$ capacitor is used in the feedback circuit there
is a phase shift of about 10 degrees in the feedback voltage due to the associated 100 -ohm heater element. In my experience, the actual capacitance of large electrolytics is sometimes well below the nominal value and it tends to vary with time and temperature. This would cause minor frequency fluctuations in the oscillator output of the kind Mr. Carpenter has mentioned. I am almost certain that the use of an $\mathrm{R}_{53}$ or, perhaps, better, an $R_{14}$, directly heated thermistor would cure this trouble and also give a more effective control of amplitude.

There is, however, an entirely different way of using an indirectly heated thermistor for control purposes if the long time constant is acceptable. It involves the use of an extra transistor stage as a Class B amplifier in which the thermistor heater forms the collector load. A suggested circuit is shown in Fig. 1 in which V3 is the last stage of Mr. Carpenter's amplifier and V4 is the power amplifier stage used to energize the thermistor heater. The virtue of this arrangement is that V4 is completely outside the main amplifier and feedback loop.
The function of V4 is merely to supply heater power proportional to the main oscillator output power. A Class $\mathbf{B}$ amplifier is suggested because in this case there is no initial heating of the thermistor by a standing collector current in V4. Use of a Class C connection might be still better, for then there would be no thermistor heating until the oscillator output reached a definite threshold level sufficient to overcome the initial reverse base bias of V4. Where low distortion is important it is just possible that the non-linear input impedance of V4 might prove objectionable but this effect could be minimized by including a moderately high series resistance in the base circuit of V4.

When fixed frequency operation of an oscillator is required there is a great deal to be said in favour of


Fig. 1. Use of separate power amplifier for thermistor heater.

Fig. 2. Transistorized twin-T resistance-capocitance

the twin-T RC circuit. This is in effect an equivalent of the Wien bridge which has a common input and output terminal. Reference (2) of my original article concerns a valve oscillator of this type, described by A. R. Bailey. The writer has developed a transistor version of Bailey's circuit, shown in Fig. 2. Space limitations preclude a full description of this arrangement but its salient features are the same as for the valve-operated version. Due to the use of high amplifier gain and a large amount of negative feedback the performance is outstanding in respect of low distortion and high frequency stability. For variable frequency operation the circuit is less attractive since it calls for the use of three ganged variable resistances one of which differs in value from the other two. Coarse changes of frequency are made by changing the capacitors C, C and 2C shown on the diagram. Mr. Carpenter may find the circuit useful for his timing standard. Regarding the super-alpha circuit used in my two designs, readers may care to note a brief reference to this compound-connected pair in the book "Transistor Circuit Engineering," (pp. 131-133), edited by R. F. Shea and published by Wiley.

In conclusion it may be of interest to mention a few points concerning the performance of the original Wien bridge oscillator. As regards its output level, measurements made on a version using a Standard Telephones Type $\mathrm{R}_{14}$ thermistor as TH. 2 and a Type A as TH. 1 show that the output is exactly the same over the two low-frequency ranges. On the highest frequency range this remains so up to a frequency of $13 \mathrm{kc} / \mathrm{s}$. Between 13 and $20 \mathrm{kc} / \mathrm{s}$ the output falls slowly, dropping by 4 per cent (about one third of a decibel) at the highest frequency. Harmonic distortion is 1.2 per cent for 1 volt output at $1000 \mathrm{c} / \mathrm{s}$.

There is a small change of frequency if a reactive load is connected across the output terminals. To remove this it would be necessary to use an emitter follower as a final output stage. Another OC 72 would be suitable for use in this position.

Finally, an earlier letter from Messrs. Evans and Williams, to which I replied in the October issue of Wireless World, cast some doubt on the performance of the super-alpha circuit at temperatures in excess of $30^{\circ} \mathrm{C}$. This point has been checked by the crude expedient of holding a hot soldering iron near the casing of the first transistor V1 until it became uncomfortably hot to touch. Under these circumstances the total supply current dropped by about 10 per cent (due to the use of three d.c. coupled stages) but there was no noticeable change in the output level of waveform and only an insignificant change in frequency. No trouble from this source need be anticipated, using germanium transistors, at normal room temperatures. For use at very
high temperatures it might be desirable to use silicon devices in this er any other comparable circuit.

## Cheltenham.

F. BUTLER

THERE is an alternative method of using RC tuning networks which avoids the difficulties arising from the low input impedance of a transistor which were pointed out by Mr. Butler in the August issue. It is to connect the networks as current attenuators rather than voltage attenuators. (See D. E. Hooper and H. E. Jackets, "Current Derived Resistance Capacitance Oscillators using Junction Transistors," Electronic Engineering, August 1956.)

To operate a Wien type network in this way, an input current is applied to what are normally the output terminals, and an output current is taken from what are normally the input terminals. The frequency at which there is no phase shift is unaltered, and the current attenuation is the same as the voltage attenuation of the usual connection.

When the network is used in this way, the input

current must be supplied by a high-impedance source, such as a common-base transistor, and the output current must be supplied to a low-impedance load: again, a common-base transistor is indicated, though a common-emitter stage can be made to serve. Under these conditions, the effect of the maintaining amplifier on the frequency of oscillation is small, just as it is in voltage-operated circuits when the input voltage source has a low impedance and the output voltage is supplied to a high-impedance load.

A difficulty of current operation is that the resistance element of the parallel arm forms the d.c. collector load of a transistor. The value of resistance which can be used is then limited by considerations of supply voltage. This problem can, however, be avoided by using an RL network instead of an RC network. The collector current then flows through an inductor, as shown in the diagram, and there is only a small voltage drop. With resistances and inductances of equal value the frequency of oscillation is $\mathrm{R} / 2 \pi \mathrm{~L}$. A danger in RL circuits is that the self-resonance of the chokes might provoke oscilla-

tion at the wrong frequency. In general, the risk can be reduced by making R as small as possible. When this ! is done, the resistance of the load, which comes in series with one of the R's, may not be negligible at the lowfrequency end of a tuning range. It will affect both the frequency and the attenuation. One method of compensation is to add an extra resistance to the parallel arm ( $\mathbf{R}_{3}$ in the diagram). This trick is particularly useful in lamp-stabilized circuits, where variations in network attenuation are apt to have a bigger effect on the amplitude of oscillation than in thermistor-stabilized circuits.
The circuit shown is that of a "lash-up" constructed to see how and RL oscillator behaves in practice. It is not put forward as a proper design though, in fact, it exhibited reasonably good waveform and amplitude stability over the tuning range $100 \mathrm{c} / \mathrm{s}-2 \mathrm{kc} / \mathrm{s}$.

## Croydon.

G. W. SHORT

FURTHER to the recent correspondence on transistorized Wien bridge oscillators, the accompanying circuit may be of interest.
By the use of n-p-n and p-n-p transistors in the
amplifier it is possible to maintain correct bias conditions very easily, merely by biasing the first stage directly and the last stage by means of negative feedback.

The optimum bias voltage at the point $A$ is clearly half the supply potential and that at point $B$ one-sixth, since the attenuation through the bridge is three times. $R_{\text {, }}$ and $R_{8}$ bias point $B ; R_{1}$ and $R_{2}$ supply one-third of the amplifier output d.c. potential (i.e., one-sixth of the supply potential) as negative feedback to the emitter of the first transistor. The small potentials developed across $R_{4}$ and $R_{8}$ do not materially affect the biasing arrangements. With this arrangement $R_{5}$ and $R_{0}$ can be $100 \mathrm{k} \Omega$ if desired

The association of a $\mathrm{p}-\mathrm{n}-\mathrm{p}$ and a $\mathrm{n}-\mathrm{p}-\mathrm{n}$ transistor for the second pair provides for handling a wide variation in load impedance without output amplitude variation.

As shown the circuit will oscillate from 2 to $20 \mathrm{kc} / \mathrm{s}$. For supply potential variation from 13 to 22 volts, or with the load impedance varying from $\infty$ to 100 ohms, there is negligible waveform distortion, or variation in output amplitude and frequency.

## Chertsey.

W. R. NAYLAND,
Roe and Co., Ltd.

I FEEL that Mr. Carpenter (October issue) has chosen a circuit which is not really suitable for a time standard. In the first place three active elements are used to pro-

duce the necessary loop gain and phase shift to maintain oscillat on; these are three items of variation and trouble. Secondly, the symmetrical case of the $W_{i e n}$ bridge has poor selectivity and a strong dislike to a shunt impedance across the parallel RC in the network. However, the effect of this shunt impedance and the changing
 circuit parameters may be considerably offset by the introduction of the virtual earth system shown in the block diagram.

This refinement is still insufficient if a constant amplitude of output is required. Of course, this is due in the main to the thermistor.

With the preceding arguments I have been endeavouring to pave the way for a simple oscillating circuit briefly outlined in a paper in the section of the Proceedings of the I.E.E., at the Convention held on 25th May, 1959, and entitled "Discussion on Application: Linear Amplification and Oscillators." This oscillator consists of only one transistor and a selective network taking the form of the well-known parallel-T network.


When this oscillator was subjected to temperature variation of $20^{\circ}-80^{\circ} \mathrm{C}$., it is stated that the frequency shifted by less than $0.1 \%$ and the output change was less than $1 \%$.
From these details it will be realized that the circuit has at least equalled the performance of the Wien bridge oscillator with considerable saving in component cost and greater reliability.
Tolworth, Surrey.
W. WOODMAN,

## Standing Wave Ratio

IN your June issue Mr. J. E. Robson describes a method of calculating the s.w.r. with the help of a quadratic
$O B=1$

equation which on the face of it, appears to be frightening. Technicians possessing a pair of compass and a foot rule may find the following geometrical method
much more instructional and useful in practical problems.
After normalizing the terminating impedance to $Z^{\prime}=(\mathrm{R}+j X) / \mathrm{Z}_{\mathrm{o}}$ a line OB equal to a unity is drawn to represent the normalized characteristic impedance. The line $O A$ is then drawn to represent the normalized terminating impedance $Z^{\prime}$ with OB as the unity of scale. With AB as diameter the "reflection" circle is drawn as shown. Finally, O is joined to the centre of C of this circle, cutting the latter at P and Q . The s.w.r. is then given by

$$
S=\frac{O Q}{O P}
$$

This can easily be proved and, if required, Mr. Robson's equation for $S$ obtained from the above construction without resort to complex variables and coordinate geometry.

Calcutta.
Q. C. GUPTA

## Signal-flow Diagrams

IN the November issue S. R. Deards asks Thomas Roddam to try to justify the use of "signal-flow method" diagrams as superior to other methods. Thomas Roddam gracefully declines.
I would like to draw their attention to a book called "Intreduction to Electronic Analog Computers" (Prentice Hall), signal-flow methods are described in application to computers, but it states that this method is useful not so much for analysis, as S. R. Deards thinks is its prime function, but for synthesis. It goes on to state that usually there is just not one good way, but several and this method allows one to decide which will suit the requirements. Where in the previous sentence I have used the word way, this applies to design, not s.f. methods.

Liverpool.
H. BAYLISS

PLEASE get it in your noddle
That matrix was nodal, not model.
THOMAS RODDAM
[We apologize to Thomas Roddam for the unlucky misspelling in line 13, right-hand column of page 562 of the previous issue. -ED.]

## INDUSTRIAL GROUPS-3

TWO or three times within the last few months the initials E.M.I. have been in the news because of takeovers or rumours of takeovers by the Group, whose latest acquisition is Ardente and its subsidiary Ardente Acoustic Laboratories for the sum of over $£ 600,000$.

Electric and Musical Industries, Ltd., was formed in 1931 to merge the interests of the Gramophone Co. and Columbia Graphophone Co., and today the Group embraces nearly 50 companies in this country and abroad. Among the trade names used by the Group are His Master's Voice and Marconiphone (both also used under licence for certain products by the Thorn Group),

Alpha Accessories
Ardente
Ardente Acoustlc Laboratories
Capitol Records Inc.
Columbia Graphophone Co.
Columbia Graphophone Co. of Greece
Comersial del Gramofono Odeon, S.A., Spain

Compania del Gramofono Odeon, S.A.E., Spain

Davey \& Moore
Electric \& Musical Industries
Electric Musical Industries (China)
Electric \& Musical Industries (Ireland)
Electrola G.m.b.H., Germany
E.M.I.-Cossor Electronics
E.M.I. Electronics
E.M.1. Records
E.M.I. Sales \& Service
E.M.I. (South Africa)
E.M.I. Suppliers (Sud America)

EMIAG. Verkaufs A.G. der E.M.i., Switzerland Gramóón Limitet Sirketi, Turkey Gramophone Co.
Gramophone Co., Pakistan
His Master's Voice (N.Z.)
Home Maintenance
Imudico A/S, Denmark
Industrias Electricas y Musicales Odeon, S.A., Argentine
Industrias Electricas y Musicales Odustrias Electricas ${ }^{\text {Y }}$
Odeon, S.A., Chile
Industrias Electricas e Musicais ndustrias Electricas e Musicail
Fabrica Odeon, S.A., Brazil

Capitol, Columbia, Parlophone, M.G.M., Mercury, Emarcy and, of course, many beginning with "Emi."

The Group's interests in the radio and electronics field range from sound and television broadcasting transmitters and ancillary equipment and electronic computers to domestic receivers and sound-reproducing equipment. In the latter field the Group markets the Emisonic Orthotone products jointly with Clarke \& Smith, of Wallington, Surrey.

The following companies are within the group which during the last financial year made a profit of $£ 5,348,000$ -some $£ 400,000$ more than the previous year's figure:-

La Voce del Padrone-ColumbiaMarconiphone S.p. A., Italy
Les Editions et Productions Musicales Pathe-Marconi, S.A., France
Lindstroem, Carl, G.m.b.H., Germany
Marconiphone Co.
Morphy-Richards
Morphy-Richards (Astral)
Morphy-Richards (Sales Services)
Musica Argentina e Internacional, S.R.L.

Nordisk Beka Record $A / S$, Denmark
Oesterreichische Columbia Graphophone G.m.b.H., Austria

Parlophone Co.
S.A. Gramophone N.V., Belgium Skandinavisk Grammophon A/S, Denmark
Skandinaviska Grammophon $A / B$, Sweden
Skandinavisk Odeon A/S, Denmark
Skandinaviska Odeon $A / B$ Sweden
Ardmore 8 Beechwood (Belgique) $S . A$.
Ardmore \& Beechwood (\$. Aírica)
Les Industries Musicales et Electriques Pathe-Marconi, France


## NEW DECCA RADARS

ANALYSIS by Decca of more than 350 cases of collision at sea during the past five years has shown that more than $95 \%$ occur in congested shipping lanes and in pilotage waters, and the conclusion is drawn that the possession of radar is not in itself a guarantee of safety. Simple radar systems must be supplemented by facilities which will reduce the time and effort necessary for the interpretation of target movements on the p.p.i. display. A major step forward was taken with the introduction by Decca in 1956 of True Motion displays in which returns from buoys and coastlines remained stationary while "own ship" and other vessels under way were shown by moving spots leaving persistence trails to show their true motion. Even so the officers of the watch are often hard pressed in keeping under observation the changes of range and bearing of all the vessels in the vicinity which at any time might change to potential collision courses.

A most useful additional feature is now incorporated in the new Decca TM969 and TM-S2400 in the form of an interscan marker originating from "own ship" and variable in length and inclination, with direct reading of range and bearing from scales coupled to the controls. Any number of targets can be checked in rapid succession, and the use of a longer-persistence screen helps in showing change or-more important-lack of change of bearing with the minimum of effort.

In all, four new models have been added to the Decca range giving a choice of 12 types for varying requirements. One of these operates on 10 cm (S-band) giving a high target/clutter ratio and better performance in the presence of precipitation. Two models have 16 -in displays, and in all the new consoles particular attention has been given to the grouping of controls according to function and frequency of use. The display units hinge forward,
giving instant access to circuits for ease of maintenance.

All the sets make use of narrow-beam slotted waveguide scanners with lower top weight and windage. The rotation rate has been increased to 20 r.p.m.

## All-Transistor Car Radio

The new Murphy " Voxson" fully-transitorized, longand medium-wave, twelve-volt car radio is made up of two separate units. One of these consists of a small cabinet cum rear-view car mirror (see illustration) which is attached via an adjustable ball and socket joint and suction cup to the car windscreen. This unit contains the volume, tuning and wave-change controls, the internal ferrite aerial, and the mixer, two i.f. and audio driver transistors. The $2 \frac{1}{2}$-watt audio output transistor, together with its associated heat sink, input and output transformers and 5 -in loudspeaker form the other unit. This may be sited in any convenient position in the car, for example, under the dashboard. The total cost of both units is 20 guineas (including purchase tax).


Transformer with a number of secondaries wound on separate parallel limbs is being developed by the G.P.O. With this arrangement, which is electrically equivalent to a number of ordinary two-winding transformers with their primaries connected in series, the parallel magnetomotive forces across each secondary are equal. The ampere turns are thus equal for each secondary so that the number of winding turns determines the secondary current rather than, as in a normal transformer, the secondary voltage. This new type of transformer is thus useful for feeding a number of current-operated devices such as transistors.

Peak Voltmeter in which the peak voltage storage capacitor is not, as is usual, charged directly from the input but rather from a flip-flop which is triggered from the input is described by R. P. MacKenzie in the June 17, 1960, issue of Electronics. In conventional peak voltmeters the indicated voltage may be too low at low input duty cycles because insufficient input current is available to charge up the storage capacitor to the full input voltage. In the circuit described by MacKenzie, as long as the voltage on the storage capacitor is less than the input voltage, their difference triggers a flip-flop which then for its unstable period charges the storage capacitor. Input voltage peaks thus continue to trigger the flip-flop to charge up the storage capacitor until the voltage on this capacitor reaches the input voltage and their difference is no longer capable of triggering the flip-flop.

Torque and Speed Control (TASC) units introduced by the Lowestoft Division of Pye Electric, Ltd., utilize inductive coupling to produce the control torque. The two cross-sectional diagrams show the construction of a typical unit. This consists of two independently rotating shafts and a stationary magnetizing control coil. The control input shaft is driven at a fixed speed and carries a poled rotor. This rotor is surrounded by an aluminium or copper torque tube attached to the control output shaft. (The larger

"Change-of-State" Crystal Oven was described by D. J. Fewings (of Marconi's Wireless Telegraph Co., Ltd.) at the Brit.I.R.E. symposium on "Stable Frequency Generation" held last May. This type of oven utilizes the constancy of the melting point temperature of a substance to produce the required constant crystal temperature. Ideally such an oven would consist of the crystal imbedded at the centre of a sphere of a suitable solid substance which is in turn surrounded by a uniformly thick layer of the same substance in liquid form. The crystal and inner solid portion of the sphere will then eventually take up the melt-ing-point temperature of the substance, and external temperature changes will then merely increase or decrease the amount of melted material, leaving the temperature nearly unchanged.

The action of the oven depends on there being both solid and liquid layers of material present. The simplest practical way of ensuring this is to use a substance whose volume changes substantially on melting, together with an expansible closed container for the whole oven. The change in the volume of the container as the substance melts can then be used to operate the on/off switch for the oven heater, and this must uniformly heat the whole outer surface of the sphere. In practice the temperature control cannot be perfect because the necessary supports and connections to the crystal through the sphere provide paths along which heat can flow. However, a similar limitation applies to any type of temperature control. A suitable material to use is naphthalene, which has a melting point of $79.5^{\circ} \mathrm{C}$ and expands by rather more than $10 \%$ in volume on melting. In a practical case using such an oven an external temperature change of $14.5^{\circ} \mathrm{C}$ produced a crystal temperature change of only $0.084^{\circ} \mathrm{C}$. For maximum crystal frequency stability it may be desirable to operate the crystal at the temperature at which its rate of change of frequency with temperature is zero. Crystals can be cut so that this temperature is within $\pm 3^{\circ} \mathrm{C}$ of the melting point temperature of the substance used. Any remaining temperature difference could then be taken up by altering the melting point temperature by adding a suitable impurity to the substance. In the case of naphthalene anthracene is a suitable impurity: by adding up to $10 \%$ by weight of anthracene to naphthalene the melting point of the resultant mixture can be varied from $79.5^{\circ} \mathrm{C}$ to $76.4^{\circ} \mathrm{C}$.

Low-frequency Noise can be obtained from high-frequency noise by sampling it at regular intervals and storing each sample until the next is obtained so as to produce a stepped noise output in which the step heights are equal to the sample amplitudes and in which steps occur whenever a sample is made. It can be shown that, if $T$ is the time interval between successive samples, then $90 \%$ of the total power in the stepped noise output lies in the frequency range from 0 to $1 / \mathrm{T}$. Although in this frequency range the
noise output decreases as the frequency increases from zero, the noise spectrum shape can be calculated and allowed for. A description of a low-frequency noise generator based on these principles is given by N. T. Slater in the August 1960 issue of Electronic Engineering.
Very-Low Frequency Propagation over long distances is being studied with the aid of a ship-borne 30 MW (peak) pulse generator and 3,000metre long helicopter-supported aerial according to an article by M. M. Newman et al in the July 22 issue of the American journal Electronics. With a transmission frequency of $20 \mathrm{kc} / \mathrm{s}$ a signal strength of the order of $100 \mu \mathrm{~V} /$ metre can be produced at a distance of $10,000 \mathrm{~km}$ ( 6,200 miles) by means of this generator and aerial.

Optical Maser oscillator utilizing the fluorescence of synthetic ruby has been developed by the Bell Tele"phone Laboratories. The ruby is "pumped" with green light to produce "stimulated emission" at the frequency of the fluorescent colour (red). The ruby, which is the form of a rod $1 \frac{1}{2}$ in long with a diameter of $\frac{1}{5} \mathrm{in}$, is held in the centre of a spiral photoflash lamp. The green component of the light from this lamp excites the chromium impurity atoms in the ruby. The excited chromium atoms at first rapidly relax from the excited quantum energy level to a metastable intermediate quantum energy level without emitting any radiation. The chromium atoms then rather more slowly relax from this metastable level to their original level,
emitting red fluorescent light in the process. The fluorescent light first emitted stimulates the chromium atoms still in the metastable state to radiate at the fluorescent frequency, provided that the fluorescent light stays long enough in the ruby rod to provide sufficient stimulation. This proviso is satisfied by polishing the ends of the rod until they are exactly flat and parallel, and then partially silvering them. Some of the fluorescent light which is emitted within a small angle about the rod's axis is then reflected back and forth between the silvered rod ends for long enough to stimulate emission. It is this stimulated emission which distinguishes the action of this maser from ordinary spontaneous fluorescence. This stimulation process ensures both that the stimulated radiation is all emitted within a small angle about the rod's axis (an angle of about one-tenth of a degree was actually measured), and that the stimulated radiation is coherent (i.e. has the same phase at all points on either end of the rod). Coherence was proved by observing interference between the light from two fine parallel slits in the silver. Coherence has not hitherto been achievable with a light source. As with radio waves, it makes possible the communication processes of modulation, amplification and detection. Another property of the emitted light resulting from stimulation is that its bandwidth is only about one-sixtieth of that of normal fluorescent light, and it is hoped to achieve an even narrower bandwidth. This narrow-bandwidth property, as well as the fact that the light is all emitted within a very
narrow angle, gives rise to a very bright source, in fact more than a million times brighter than the sun. This brightness should also be useful for communication purposes. The light from this maser has in fact been seen twenty-five miles away. With the present device, maser action can be sustained for about a millisecond at a time. During this time several hundred bursts of radiation spaced a few microseconds apart are produced, each lasting for about a microsecond. The stimulated emission is produced in pulses probably because, when stimulated emission occurs, it decreases the number of atoms in the metastable state so rapidly that the flash lamp is not bright enough to maintain a sufficient number of atoms in the metastable state to allow maser action to continue. Maser action then ceases, but commences again when the flash lamp has had time to produce a sufficient number of atoms in the metastable state. This alternate decrease and increase in the number of atoms in the metastable state then produces pulses of light.
Crystal-Frequency Variation with temperature can be compensated for (according to a letter from E. A. Gerber in the February, 1960 issue of Proc. I.R.E.) by applying a tem-perature-dependent force to the crystal so as to produce an opposite frequency change. A bimetal strip bearing on the crystal can be made to provide a suitable temperaturedependent force and, according to the direction of this force, a positive or negative compensating change in the crystal frequency can be produced.


## Frame Grid

## Valve Manufacture

THE mechanical precision necessary for the handling of the fine wires ( 10 microns in diameter) and the assembly of frame grids with grid-cathode spacings of the order of 50 microns has called for the development of a completely new range of electrode winding and welding machines, and of assembly methods comparable with those in the watch-making industry. The accompanying photographs show a typical frame grid, and the assembly fig for the two grids of the PCC89 cascode r.f. amplifier, one of the four new high-slope television valves now in quantity production at the Mitcham Works of Mullard Ltd.

# Manufacturers' Products 

## NEW ELECTRONIC EQUIPMENT AND ACCESSORIES

## Precision V.II.F. Frequency Standard

A COMPACT and portable (but mains-operated) frequency standard for accurately checking the operating frequencies of mobile, as well as of fixed, v.h.f. radio transmitters has been introduced by Wayne Kerr Laboratories, Ltd., Chessington, Surrey. Known as Model FS100 it consists of a crystal-controlled oscillator, three


Wayne Kerr Model FS100 crystol-controlled frequency standard for testing mobile v.h.f. transmitters.
frequency multiplier units, wide-band detector and a.c. mains power unit for $110-130 \mathrm{~V}$ or $190-240 \mathrm{~V}, 40-60 \mathrm{c} / \mathrm{s}$.

The oscillator can be switched to operate with any one of 12 quartz crystals mounted in thermostatically controlled, plug-in ovens, two crystals being mounted in a single oven. Crystals for operation on fundamental frequencies within the range 7.5 to $9.7 \mathrm{Mc} / \mathrm{s}$ can be used and each multiplier unit covers a specific part only of the total frequency coverage. For example, the Band I multiplier covers crystal frequencies of 7.5 to $8.17 \mathrm{Mc} / \mathrm{s}$ giving final outputs on 3,9 and 18 times as required of the frequency of the crystal selected. Between them the 3 multipliers will provide spot output frequencies of from 22.5 to $175 \mathrm{Mc} / \mathrm{s}$ with an accuracy of 3 parts in $10^{\circ}$.

Crystals are selected by a rotary switch mounted in the centre of the instrument panel, while multipliers are selected by means of push-buttons.

Signals from the transmitter to be tested are fed into the test set and the crystal and multiplier switches set to the transmitter's operating frequency and the output then checked for zero beat. The test set measures $13 \frac{1}{4}$ in wide, 9 in high and $8 \frac{1}{4}$ in deep and weighs 16 lb complete.

## New Plessey 21 in Loudspeaker

A NEW $2 \frac{1}{4}$ in diameter loudspeaker for use in modern miniature transistorized receivers has been introduced by The Plessey Company Ltd., Ilford, Essex.

The magnetic circuit uses a new constructional technique to enable the overall depth to be reduced to a

Plessey $2 \frac{1}{4}$ in loud. speaker hos a depth of only 0.83 in .

minimum-a rod magnet being inserted into a split aluminium ring which is fitted into an accurately machined yoke. This design arrangement gives a compact, robust motor unit without the use of magnet assembly bolts.

The $2 \frac{1}{4}$ in diameter loudspeaker is available with a low impedance voice coil ( $\delta / 10 \mathrm{ohms}$ ) for use with a matching transformer or alternatively with high impedance coils for direct connection into transistor circuits.

## 20W Transistor Amplifier

THE Type TR/ 2 transistor amplifier is a compact unit measuring $8 \mathrm{in} \times 6 \mathrm{in} \times 3$ in (high) and capable of delivering 20 W peak a.f. output to $3.5-\Omega$ or $15-\Omega$ loudspeakers. It is intended primarily for portable and mobile PA applications and.operates from a 12 V d.c. supply, imposing a no-speech load of 0.25 A and a mean working load of 1A.

It employs 5 transistors, one protective diode and a lightweight moving-coil microphone fitted with a press-to-talk switch, thereby ensuring maximum economy in l.t. consumption during intervals in operation. The price is $£ 3110$ s (od including microphone.

Although a non-reversible 1.t. input socket is fitted, a safety circuit to prevent damage to the transistors,

R.E.E. 20W transistor PA amplifier. The key gives a good indication of its size
should the battery polarity be reversed, can be included at a small extra charge.
Further details can be obtained from R.E.E. Telecommunications Ltd., Telecomm Works, Market Square, Crewkerne, Somerset.

## V.H.F. Marine Radio-telephone

SHOWN in the illustration is the latest Pye marine v.h.f. radio-telephone for shipborne use. It is suitable for use in large or small vessels having $100-150 \mathrm{~V}$ or $190-240 \mathrm{~V}, 40-60 \mathrm{c} / \mathrm{s}$ a.c. supplies and operates ${ }_{s}$ on the frequencies allocated for inter-ship, ship-to-shore, harbour and docking telephone communications.
The equipment allows for operation on 28 channels in the international marine v.h.f. band, 9 channels being for single frequency simplex and 17 for duplex working. The 28 channels are provided by 18 crystals.

Frequency modulation is employed and the transmitter power output is 20W. The Type PTC8306 equipment complies with the Hague and British G.P.O.


Pye marine V.H.F./F.M. radio-telephone Type PTC8306. The remote control uns: is on the left.
specifications for this type of equipment so it is suitable for international use as well as for use in home waters. Provision is made for full remote control over lines up to 200 ft in length.
The makers are Pye Telecommunications Ltd., Newmarket Road, Cambridge, from whom further details can be obtained.

## Oscilloscope Kit

THE Heathkit Model OS-1 enables an oscilloscope to be built with a Y-amplifier sensitivity of $14 \mathrm{mV} / \mathrm{cm}$ and a response within $\pm 3 \mathrm{~dB}$ from $10 \mathrm{c} / \mathrm{s}$ to $2.5 \mathrm{Mc} / \mathrm{s}$. The X -amplifier sensitivity is $1.4 \mathrm{~V} / \mathrm{cm}$ and its response within $\pm 3 \mathrm{~dB}$ from $15 \mathrm{z} / \mathrm{s}$ to $500 \mathrm{kc} / \mathrm{s}$ : the time-base range is from $15 \mathrm{z} / \mathrm{s}$ to $150 \mathrm{kc} / \mathrm{s}$ approximately. A $50 \mathrm{c} / \mathrm{s}$ calibration source provides a peak-to-peak potential of 1,10 or 50 V to an accuracy of $\pm 10 \%$. A printed circuit board helps to ensure that all kits are wired exactly as required. The manual provided includes step-by-step wiring instrustions, pictorial diagrams, and a circuit description and diagram. Operating voltages are also given. The Kit costs $£ 18$ 19s 6 d and is available from Daystrom Ltd., of Gloucester.

## Transistor Oscillators

WITH the new Levell Electronics Types TG150 and TG150M battery oscillators sine waves can be produced

Riọht.-Levell Flectronics Transi ir Batery Oscillator. Type TGISM.


Left. Assembled Heathkit Model OS-I oscilloscope
at any frequency between $1.5 \mathrm{c} / \mathrm{s}$ and $150 \mathrm{kc} / \mathrm{s}$ and with any amplutude up to 2.5 V into $600 \Omega$. The harmonic distortion is $<0.3 \%$ between $50 \mathrm{c} / \mathrm{s}$ and $15 \mathrm{kc} / \mathrm{s}$ and $<1.5 \%$ below $50 \mathrm{c} / \mathrm{s}$ and above $15 \mathrm{kc} / \mathrm{s}$. The frequency changes by $<0.3 \%$ at $1 \mathrm{kc} / \mathrm{s}$ for a $30 \%$ fall in the supply voltage and by $<0.05 \%$ per ${ }^{\circ} \mathrm{C}$ temperature change. The output level varies by $<0.1 \mathrm{~dB}$ over the whole frequency range, by $<0.05 \mathrm{~dB}$ for a $30 \%$ fall in the supply voltage, and by about -0.02 dB per ${ }^{\circ} \mathrm{C}$ temperature change. A thermistor stabilized Wien bridge circuit is used in these oscillators. The Type TG150M is fitted with a calibrated output meter and costs $£ 36$, a version of this oscillator without the output meter-the TG150-costs £27. These instruments are manufactured by Levell Electronics Ltd., of $10-12 \mathrm{St}$. Albans Road, Barnet, Herts.

## New Plastic Foil Capacitors

TROPYFOL capacitor is the name given to a new type of miniature capacitor which utilizes an aluminiumcoated polyester foil dielectric rolled and impregnated in a plastic resin with axial end wires. They are available in 125 V and in 400 V types, the former with capacitance ranging from $0.01 \mu \mathrm{~F}$ to $1 \mu \mathrm{~F}$ and the latter


Tropyfol $0.1 \mu \mathrm{~F}, 40 \mathrm{CV}$ miniature copacitor.
from $0.001 \mu \mathrm{~F}$ to $1 \mu \mathrm{~F}$. A $0.001 \mu \mathrm{~F}$ capacitor in the 125 V type measures 4.5 mm in diameter and 12 mm long and a $1 \mu \mathrm{~F}$ size for 400 V working is 13 mm in diameter and 24 mm long ( $0.5 \mathrm{in} \times 0.9 \mathrm{in}$ approx.).

Insulation resistance is said to be not less than 2,500 $\sec (\mathrm{M} \Omega \times \mu \mathrm{F})$ while in some cases it may reach 50,000 sec .

Supplies are available to the radio and electronics industries and further details can be obtained from Waycom Ltd., Empire Buildings, Duke Street Hill, London, S.E.l.

## Very Low Temperature Coefficient Resistors

BY combining two wires of different temperature characteristics Alma Components have produced a range of precision wire-wound resistors with the remarkably low temperature coefficient of less than 4 parts per million per degree C. The wires are wound on non-hygroscopic ceramic formers encapsulated in a sealing resin. Three sizes are available rated at lw (Type C), $\frac{1}{2} w$ (Type C2) and $\frac{1}{4} w$ (Type C4) with resistance values of $100 \Omega$ to $300 \mathrm{k} \Omega, 100 \Omega$ to $120 \mathrm{k} \Omega$ and $100 \Omega$ to $80 \mathrm{k} \Omega$ respectively. The winding is sectionalized with adjacent sections wound in opposite directions. Standard tolerances are $0.1 \%, 0.25 \%, 0.5 \%$ and $1 \%$.

Further details can be obtained from Alma Components L.td., 551, Holloway Road, London, N.19.

## Principles of

# Digital Computers 

By D. S. WILDE,<br>B.Sc., A.M.I.E.E., Grad. Inst. P.

## 2.-ARITHMETIC AND CONTROL CIRCUITS

$N$ the previous article methods of storing binary information in a digital computer were outlined. These digit patterns represent numbers and instructions and the function of the arithmetic unit (sometimes called the accumulator-Babbage called it the " mill") is to perform additions, subtractions etc. with these numbers, as dictated by the instructions.

A group of digits may be delivered from a store one after the other on a single channel (serially) or simultaneously with a separate channel for each digit (in parallel) and the choice must be made in the design of the computer between increased


| $a$ | $b$ | $a$ AND $b$ |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 1 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 1 | 1 |

Fig. 3 (a) Symbolic AND gate. (b) Truth table for the AND gate.
speed and economy of equipment. For example, a serial adder will take $n t+t$ periods to add two $n-$ digit binary numbers occurring as pulse trains of period $t$ (the final $t$ is for the last carry). Parallel addition will only take $2 t$ periods but there must be one adder per channel i.e. $n$ adders.

It is possible to effect a compromise between both extremes and the Ferranti Mercury computer is an example of this. Access to the storage elements is parallel, the parallel digits are turned into serial form and dealt with by fast serial circuits.

Fig. 4 (a) Circuit of an AND gote for positive-going inputs. (b) Input and output pulses for the AND gate for positivegoing. inputs of Fig. 4(a). (c) Circuit of on AND gote for negotive-going inputs.

(a)

Whilst the present tendency is for fully parallel machines, it is possible that with new types of storage operating extremely rapidly-perhaps up to 100 $\mathrm{Mc} / \mathrm{s}$-we may see a return to serial techniques and consequently smaller computers.
It has already been shown that the fundamental arithmetic operation is subtraction and that all the other common functions can be derived from it. Subtractions only occur about one-third as frequently as addition in most programmes, and since an addition operation performed in a subtracter would require two stages $[0-b, a-(0-b)]$ a subtracter is an uneconomical component. An adder is always provided and a complementer then enables subtractions to be performed by complement addition.
Binary addition can be reduced to a set of logical (or Boolean) operations and these operations are performed by extremely simple electronic circuits. Given the necessary reduction details a binary adder can be easily synthesized, so can other functional units.

There are a variety of logical operations although only two are essential and the others can be derived from them. It is much more convenient however to use three basic logical functions, these are AND, OR, and NOT, and they are almost self explanatory.

The AND gate is shown symbolically in Fig. 3(a) and both input $a$ AND input $b$ must be present ${ }^{8}$ to cause an output from the gate. There are four possible combinations for the conditions of $a$ and $b$ and these are shown in the " truth table" of Fig. 3(b) ( 1 signifies presence, 0 signifies absence). Only when $a=b=1$ do $a^{r}$ and $b=1$ and hence the AND operation is equivalent to the algebraic product $a b$.

The circuit for such an AND gate is shown in Fig. 4(a) and in this case the inputs must be positive going and the supply rail is a positive rail. Suppose, in a practical case, that the supply is 100 volts and $a$ and $b$


Fig. 5 (a) Symbolic OR gate. (b) Truth table for the OR gate. (c) Circuit of an OR gate for positive-going inputs. (d) Circuit of on OR gate for negative-going inputs.
earth and will be cut-off when the grid goes negative. The anode potential will be low when the valve is bottomed and high when cut off. The anode excursions are applied to the potentiometer chain $\mathrm{R}_{1}$ and $\mathrm{R}_{2}$ which can be chosen in conjunction with $\mathrm{V}_{2}$ to ensure that as the input $a$ moves from earth negatively the output NOT a moves from a negative potential to earth. $D_{1}$ merely serves to prevent the output going above earth potential.

In early computers the diodes in AND or OR gates were thermionic and valves were used for inversion circuits. The diodes are now semi-conductor types and the inversion circuits are transistors.

Using only these three circuits, AND, OR and NOT, it is possible to construct a binary adder. There is an additional circuit required for a serial adder however, this is a delay.

First it is necessary to establish the " truth table"


Fig. 6 Voltage inverter circuit of a NOT gate.

| $a$ | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $b$ | 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 |
| $c$ | 0 | 1 | 0 | 1 | 0 | 1 | 0 | 1 |
| ANSWER | 0 | 1 | 1 | 0 | 1 | 0 | 0 | 1 |
| CARRY | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 1 |
| $a O R D O R C$ | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |

Fig. 7 Truth table for binary addition.
for binary addition (Fig. 7). There will be three digits to add; these will be from the addend, the augend and the carry digit resulting from the addition of the previous digits. With three inputs, each of which can be either 1 or 0 , there are eight possible combinations of inputs to the adder. These are shown in the "truth table" together with the answer ( 1 or 0 ) and carry ( 1 or 0 ) which must result.

Suppose that the three inputs are applied to an OR gate. Whenever there is a 1 among the inputs there must be a 1 in the output. If all eight combinations are tried the output agrees with the required answer except when two of the inputs are l's together. (The output will be 1 , the correct answer should be 0 .) There must be an additional set of circuits to inhibit the output of the OR gates under these conditions. In other words the required answer is ( $a \mathrm{OR} b$ OR $c$ ) AND NOT I $(a$ AND $b$ ) OR ( $b$ AND $c$ ) OR (a AND c)]. The logical diagram now looks like Fig. 8.

The circuit combination responsible for inhibiting the output when only two 1 's occur can also be used to generate the carry pulse since there must be a carry developed when two (or three) ones are present. Unfortunately the inhibit gates will inhibit the output when three 1 's are present at the input and no answer digit will appear though one is required. A three-input AND gate is included therefore and its output is applied to a two-input OR gate whose second input is the previous output.

This logical configuration now satisfies all eight conditions of input giving the correct answer digit and the correct carry (Fig. 9). The carry digit is applied to a delay of one digit period and is fed back to the input, arriving in the next most significant place.

The same logical considerations apply to adders for a computer working in the parallel mode. The carry digits in this case are not delayed to a more significant time period but are applied as third inputs to a more significant adder.

(Above) Fig. 8 Stoge in the construction of a logical diagram of a binary adder.

There are other forms of binary adders and a binary subtracter can be developed in a similar way. The final step in the design is the substitution of circuit elements for logical symbols, and these have already been shown.

Whatever alternative solutions are found to the design of a specific logical circuit, Boolean algebra can provide a simple check that the circuit performs the function required. In the example chosen of the adder circuit the logical function is written as
$\begin{aligned} \text { Answer }= & (a \text { OR } b \text { OR } c) \text { AND NOT [( } a \text { AND } \\ & b) \text { OR }(a \text { AND } c) \text { OR }(b \text { AND } c)] \\ & \text { OR ( } a \text { AND } b \text { AND } c) \\ \text { Carry }= & (a \text { AND } b) \text { OR }(b \text { AND } c) \text { OR }(a\end{aligned}$
Previously it has been shown that $a$ AND $b=$ $a b, a$ OR $b=a+b-a b$ and NOT $a=1-a$. Algebraic substitutions can thus be made. Working the final expressions is rather tedious but a glorious simplification is available because $a^{n}=a$ (since $a=1$ or $a=0$ ). If the above substitutions are made then

Answer $=\left(\begin{array}{l}a+b+c)-2(a b+a c+b c)+ \\ 4 a b c\end{array}\right.$
Carry $=a b+a c+b c-2 a b c$
and any binary adder must satisfy these conditions. Control Unit.-The control part of the computer is responsible for reading and sequentially obeying instructions which are held in the stores. Generally an instruction is divided into two parts, a function section (add, subtract, multiply, input, output, etc.) and address section (locating in the store the number that the function is to be performed on).

In the case of a serial computer the instruction will emerge from the storage as a series of pulses. The pulse pattern is a code and must be decoded to operate various gates in the computer which allow numbers to be routed to the various parts of the computer (the arithmetic units, the output units, another store location, etc.). Since only one pulse occurs at a time a pulse pattern can only be decoded if the individual pulses are staticised (" remembered ") and all compared together after the last one has occurred. The staticisor most commonly used is the well-known bi-stable element (other names used are Kipp Relay, flip-flop, multivibrator, Eccles-Jordan circuit, and toggle). The circuit consists of a pair of valves (or transistors) cross connected from anode to grid (or collector to base). A pulse input applied to one grid causes one of the valves to conduct, the second to cut
(Right) Fig. 9 Logical diagram of a binary adder giving both onsner and carry digits.


(a)

(b)

Fig. 10 (a) Staticisor (store) for serial input train of three binary pulses. (b) Serial input train and clock pulses for storing 101 in the staticisor of Fig. 10 (a).
off and this state is maintained until a trigger pulse is applied to the other grid when the circuit changes state.

A set of three bistable elements is shown symbolically in Fig. 10 and the serial pulse train is applied to them via two-input AND gates. The second inputs to the gates are single clock pulses; they behave in the manner of a strobe pulse and $p_{1}$ is coincident in time with the first pulse of the pulse train, $p_{2}$ with the second, and $p_{3}$ with the third ${ }^{10}$. In the case of a pulse train of 101 the first bi-stable element will receive a trigger since $p_{1}$ and the first 1 of the pulse train coincide and there will be an output from the AND gate. $p_{2}$ coincides with a 0 (no pulse) and there is no output from the gate, hence the second bi-stable element is not triggered. The third one is triggered, $p_{3}$ coinciding with the third place. The pulse pattern 101 has now been stored (or staticised) and the bi-stable elements are said to be in the 1,0 , and 1 state respectively.

With three bi-stable elements there are eight possible combinations in which they can be set. These are, of course, determined by the input pulse train and could represent eight possible functions.

It is necessary to decode these eight possible states to uniquely define each of them. Fortunately this is an easy matter for a decoder is merely an AND gate ; eight three-input AND gates are necessary to decode three binary digits. These are shown in Fig. 11. The first digit has its inputs connected to the 0 side of the bistable elements. For this type of AND gate, positive inputs are required and the 0 side of a bi-stable element is defined as that side being at a positive potential when no input trigger has been received. The 1


Fig. II Set of eight AND gate decoders for the eight possible states of three bi-stable elements.

Instructions are therefore five digits long.

The cycle of operations can be defined by four time periods $\mathrm{T}_{1}, \mathrm{~T}_{2}, \mathrm{~T}_{3}$, $\mathrm{T}_{4}$ and one instruction is selected and obeyed during each cycle of four periods. What happens in each time period is detailed below.
$T_{1}$ : add unity to the top line of the control store. This control address initially has zero stored in it and its job is to read instructions from the main store one by one. During $T_{1}$ the top line is read from the control store into the control adder through the STOP gate, which is open unless a STOP instruction is being obeyed. The number which emerges from the adder is used to set up the address and function staticisors and it also replaces the old number in the control store top line.
$\mathrm{T}_{2}$ : in this period the contents of the store address specified by the control unit in $\mathrm{T}_{1}$ will be read out and transferred to the second line in the control store. This is the instruction which is to be obeyed.
$\mathrm{T}_{3}$ : the instruction in the second line of control is now read out to the function and address staticisors which are set up.
$T_{4}$ : the number specified by the setting of the address staticisors is read out of the main store and is operated on in the manner determined by the function staticisors.
To show how this sequence of operations works in practice the simple programme shown in Fig. 12 will be followed through. The computer, which has four functions and an 8 -address store, initially starts with the programme in the main store and the control store and arithmetic unit store empty. In the first $T_{1}$ period 00000 is read from the control store, 1 is added to it, and the address staticisors are set to 001 . The first two (function) digits (00) are not set up since the function staticisors are only set in period $\mathrm{T}_{3}$. During $\mathrm{T}_{2}$ line 001 is read from the store; its contents $(00,101)$ pass through the gate marked $\mathrm{T}_{2}$ into the control store. This is the first instruction and the code 00,101 is interpreted by the computer as "take the number stored in line 5 (101) and add it (00) to the arithmetic store contents".

This instruction is staticised on the function and address staticisors in $\mathrm{T}_{3}$ and the function staticisor decoders open the gate marked + and the address staticisor decoders select line 5 .

The contents of line 5 (the number 11111) ${ }^{\star}$ is routed to the arithmetic unit adder and arrives simultaneously with the read-out contents of the

[^7]

Fig. 12 Simple computer.
store. Addition is performed and the answer is stored. The first instruction has now been completed and the cycle repeats for the second instruction ( 01,110 ). This can be interpreted as "subtract from the arithmetic unit total (01) the number stored on line 6 (110)". In $\mathrm{T}_{4}$ of the second cycle the number on line 6 (01111) is fed to one input of the subtracter through the gate marked -. The number in the arithmetic store (11111) is read into the subtracter at the same time and the answer ( 10000 ) is returned to the arithmetic store.
The third instruction (10111) reads " transfer the number in the arithmetic store (10) to line 7 (111) of the store".
The fourth instruction merely says STOP. The decoded stop instruction closes the stop gate and no further adding of unity takes place in the top control line. Hence no subsequent instructions can be read and the programme has been finished with the final answer (10000) now in the main store in line 7.

This programme is an extremely simple one but it does show the way in which the basic elements work in conjunction with one another in doing a simple calculation. It illustrates also an important point from von Neumann's researches. This is that numbers and instructions were of exactly the same form. They both had five binary digits and were differentiated by the layout in the store which would be decided by the programmer. There is considerable flexibility to be gained since some programmes use few instructions but have a very large amount of numbers stored, whilst others use few numbers but a considerable number of instructions. The computer is therefore much more compact than if sepa-
rate stores were provided for numbers and instructions.

An additional and extremely important consequence is that instructions can be modified in the course of the programme by using them as if they were numbers and processing them in the arithmetic units to give new instructions.

The simple programme did not exploit the "decision facility" which was recognized by Babbage as essential to an automatic computer. A short series of instructions will illustrate the importance of this "decision facility". Suppose it is required to divide 9 by 3. This can be done by simple subtraction, counting the number of subtractions performed before the answer goes negative, and this is the required quotient. The programme would read
(1) Subtract 3 from the number in the arithmetic store (9).
(2) Test the answer to see if it is positive. If it is, proceed to instruction (3), if it is not, STOP. (For this purpose 0 is a positive number.)
(3) Count 1 and store it.
(4) Repeat from instruction (1).

The test facility is not difficult to achieve in practice since it has been shown that negative binary numbers are recognized by having a 1 in the most significant place. Detection of this 1 is the deciding factor in whether the computer carries on its programme or whether it stops. Far more sophisticated uses are made of this property but they cannot be dealt with here.

The principles of digital computers are simple, but a full scale computer is obviously a much larger collection of elements than those discussed in this article. There are many additional circuits necessary which are concentrated on connecting the computer to the outside world, but the aim of these articles has been to emphasise the basic simplicity at the heart of even the most mighty machines.

## REFERENCES

8 "Electronic Computers" (2nd edition.) T. E. Ivall, Iliffe, 1960, p. 100.
${ }^{10}$ Ibid. p. 120.


## High-power Loudspeaker System

By michael lorant

FOR STUDYING THE PHYSIOLOGICAL EFFECTS
OF HIGH-INTENSITY SOUND

RESEARCH technicians of the Acoustical Engineering section of Stromberg-Carlson have recently developed for the Aerospace Medical Division of the Wright Air Development Centre, Ohio, a new highintensity acoustic system for studies of the physiological effects of high-intensity sound.

The most unusual feature of the new system is the "high-fidelity" quality of its performance-the generation of undistorted sound throughout the full 10 -octave range of normal audibility:

The "business end" of this system is a huge assembly of loudspeakers- 480 in all. These are mounted in 32 separate baffies for maximum flexıbility in arrangement and control. Each baffle contains three low-frequency "woofers" and twelve high-frequency "tweeters". All the loudspeakers are especially designed to deliver "high-fidelity" sound
at high acoustic powers for sustained periods.
The system is controlled from a console providing four possible inputs-sine wave, electronic "white noise", tape recordings (of jet engine, missile, etc.), or an external source.

The main audio power equipment consists of two pairs of audio amplifiers. One pair of these is for " low-power" use only, with each amplifier providing an output of 200 watts. Each of the pair of highpower amplifiers has an output of 7,000 watts. The frequency response of the system is "flat" over the entire audible range, from 20 to $20,000 \mathrm{c} / \mathrm{s}$.

To avoid unintentional exposure of subjects to high-intensity sound, a safety precaution makes it necessary for the operator to put the controls first into the low-power position before energy can be supplied to the high-power amplifiers.

AMAIN ambition of power engineers is to generate electricity direct from heat, instead of going through the tiresome sequence of using the heat to boil water and making the steam impinge against blades in a turbine, causing it to move an array of conductors in a strong magnetic field. They may or may not be encouraged by the fact that direct conversion of heat into electrical energy does take place in absolutely every circuit. The evidence is before our very eyes-and ears-in every TV screen and loudspeaker receiving weak transmissions. The annoying thing is that this electricity is random, and therefore not only useless but detrimental.

Last month we considered the constant factor connecting temperature with random electrical power (" noise"): Boltzmann's constant, denoted by $k$. Since this form of electrical power is worse than useless, the fact that we get a very poor rate of exchange is a point in favour rather than otherwise. In our ordinary units, $k$ is only $1.38 \times 10^{-23}$ ioule per ${ }^{\circ} \mathrm{K}$. The resulting noise voltage across a resistor or tuned circuit is of the order of one or two microvolts. But of course that is enough to be a serious nuisance if the signal we are trying to receive is of the same order. When amplified 120 dB , microvolts become volts. The most significant source of this universal circuit noise (called thermal or Johnson noise) is obviously the circuit followed by the greatest amplification. In fact, if the first stage of amplification is doing well, the contributions from circuits other than its input circuit may be neglected. But Johnson noise is not the only kind that causes the escaping-steam sound in the loudspeaker and the animated graininess on the screen.

If the subject of valves is not too passé, démodé, suranné, vieux jeu, radio à la vapeur, or any of the other contemptuous French epithets we may choose to bestow on it in these transistorizing days, let us consider how they too contribute their quota of noise. Even the transistor hounds may find it worth their valuable time to do so, for at bottom the principles of transistor noise are similar.
One thing had probably better be made clear at the start: so far as the end product is concerned, valve noise is the same as Johnson noise. There is nothing to give a clue as to how much is due to valves and how much to circuits. (To cover myself against argumentative readers I had better declare a possible exception: two ingredients of valve noise to be mentioned might be recognizable by their tendency to occur increasingly at one end or other of the frequency scale.)

As has just been ninted, valve noise can be subdivided. The basic ingredient is what is usually called shot noise or shot effect. Sometimes writers (for the sake of effect?) name it in German, as Schroteffekt, which sounds even more penetrating. The air of mystery and confusion is further heightened by others who call it (wrongly, as it happens) Schottky effect, because it was first described by one Schottky. All these names are even more than usually subject to individual variations in spelling.

The thing was called shot noise (to come back to plain English), so we are told, because of its supposed resemblance to the sound of shot peppering a target. Since the sound in question, even after enormous amplification, is much more like a gentle hiss of steam, I prefer to think that the reference to shot is more to remind us (or even inform us, if we are beginners) that the flow of electric current through a valve is not perfectly smooth like the flow of treacle from spoon to plate, but more like the flow of water from cloud to earth, as rain.

The merit of likening it to shot or even rain is, I think, confined to the simple idea of current being made up of discrete particles and therefore subject to random variations from an average rate of flow. In

other ways it is misleading. A fairly obvious discrepancy is that what we get is not the magnified sound of electrons striking the anode, but a result due to random fluctuations of current in the anode circuit. A less obvious but most important difference is that shot has no effect on the target before actual impact. To be quite correct I suppose we must say that the first effect is produced by the extra pressure of air pushed against the target by the shot, but by that time the shot is virtually home. With electrons in a valve it is quite different. Electrically, they affect the anode circuit all the time they are travelling there from the cathode, and arrival marks the end of this influence, not the beginning. The fact that the movement of an electric charge near a conductor induces a current in the conductor used to be demonstrated by one of the stock gold-leaf-electroscope experiments of the nineteenth century. Nowadays I suppose one relies on Conservation of Energy. Q in Fig. 1 is the negative charge on a particle moving in response to the constant electric field between a pair of electrodes having a difference of potential V . (The field strength is V divided by the distance between the electrodes.) As its potential is being raised, work is being done on it, and the requisite energy can come only from the battery, through which a corresponding charge must therefore pass. The faster Q moves, the greater the rate of charge flow through the circuit; i.e., the greater the current. Assuming an electron starts from rest at the cathode, and the field is uniform, its acceleration is constant, so its velocity steadily increases, and the current waveform is as in Fig. 2.

For the sake of a nice round number, let us suppose
the anode current in a valve is 16 mA . The rate at which electrons cross the vacuum must then be $1 U^{17}$ per second. If they were released at exactly equal intervals the anode current would therefore be modulated at a fundamental frequency of $10^{17} \mathrm{c} / \mathrm{s}$. That is in the X-ray band, so we are safe in assuming that it would be smoothed out into imperceptibility in any practical anode circuit. But although the average rate of flight is $10^{17}$ per second, owing to the nature of thermionic emission this rate fluctuates in a random manner around its average, rather like the day-to-day fluctuation in actual births compared with a steady long-term average. At the foot of Fig. 3 the ideal waveform of Fig. 2 is repeated at intervals representing an average of eight electrons in flight at any one instant, but varying irregularly. Above is the result of adding all these pulses, and we see that it is mostly d.c., with a very high-frequency ripple-the one we are assuming is smoothed out, like the residual r.f. in a detector circuit-and a comparatively slow and quite irregular variation of the d.c. from its average. This last is the shot noise current.

It should be clear by now that the name " shot noise " is misleading in another respect: any idea that we are hearing even an indirect effect of the separate electrons pattering on the anode is completely unfounded; their frequency is far too high.


Fig. 2. Assuming o uniform field in Fig. 1, the charge is accelerated uniformly, and since the current in the circuit is proportional to the charge's velocity it increases steadily throughout the flight.

Just to make sure, and for the fun of the thing, I calculated what sort of a current Fig. 3 really represented, assuming the fairly typical transit time of $10^{-9} \mathrm{sec}$. The average rate of arrival of electrons is than $8 \times 10^{9}$ per sec., and as the frequency of the sawtooth ripple is the same- $8 \times 10^{9} \mathrm{c} / \mathrm{s}$ or $8 \mathrm{Gc} / \mathrm{s}-$ it is actually a microwave frequency, with a wavelength of $3 \frac{3}{3} \mathrm{~cm}$. We may perhaps be surprised to find the electron frequency inside a commercially used radio waveband. But lest we imagine a great peak of noise at this particular frequency, let us remember two things. The first is that in this case one cycle is only one-eighth of the transit time, so is right outside the working frequency of the kind of valves we are considering. The second is that the anode current is only $0.001 \mu \mathrm{~A}$, which wouldn't do much anyway. But it is easily measurable, even though as few as eight electrons at a time are in transit between cathode and anode; which makes one think.

In practical valves the anode current is millions of times larger, and the sawtooth ripple (which is
" shot noise" in the most literal sense) is correspondingly higher in frequency and can be left right out of account. But the slower ups and downs, due entirely to the fact that electrons are not emitted at regular intervals, are only too evident in high-gain amplifiers. Their power is distributed evenly over the frequency band. While this fact may seem natural for a purely random effect, so that we might have guessed it, it does need to be proved. Anyone sufficiently interested can wade through (e.g.) 15 pages of difficult mathematics in Davenport and Roots "Random Signals and Noise." Don't ask me to put it into a few lines of elementary algebra. I have to be content that the authorities agree on

$$
\mathrm{I}_{\mathrm{N}}=2 e \mathrm{I}_{\mathrm{a}} \mathrm{~B}
$$

where $I_{V}$ is the r.m.s. shot noise current, $I_{a}$ the anode current, $e$ the electric charge of one electron ( $1.6 \times$ $10^{-19}$ coulomb), and B the frequency bandwidth. We may note in passing that $I_{N}$ is proportional to the square root of $I_{4}$, so it doesn't do to starve the first valve in a high-gain amplifier. When the anode current is on the small side the gain is almost directly proportional to $I_{3}$, so signal/noise ratio would get worse if $I_{4}$ were reduced too much.
Actually it is rather too soon to start drawing such conclusions, for the formula just quoted is limited to such exceptional cases that we are unlikely to find any use for it. One of the rare exceptions is a brightemitter diode run with such low filament voltage and such high anode voltage that every electron emitted is immediately pulled across the vacuum. In other words, the anode current depends entirely on cathode temperature, not anode voltage. Its technical description is a temperature-limited diode. Another exception is a photoemissive cell.

Valves as normally used are run at a cathode temperature that emits more electrons than are required. Imagine a great crowd of delegates to a conference moving out after the first session into a room with a buffet at the far end. The first arrivals spot it and stream across, but the main body can't even see it for the crush, and mill around just inside the entrance. They form a "space charge "s screening the attractions of the buffet from the late entrants. This is roughly what happens in a valve (Fig. 4). Suppose the emission is at the rate of 1 amp. That is 6.4 billion (American trillion) electrons per microsecond. Even before the cathode is fully hot after switching on, enough electrons have emerged to build up a negative charge near the cathode that more than counterbalances the positive attraction of the comparatively distant anode. So nearly all the electrons emitted from then on are repelled back to the cathode. Only the few within sight, as it were, of the anode feel its attraction and stream across, forming a current of perhaps a few milliamps. If none at all left the cathode permanently, the space charge would soon be used up, but of course the slightest reduction in its numbers reduces its repelling power and allows some of the emitred electrons to join. And so a balance is reached between the attraction of the anode and the repulsion of the space charge, resulting in a certain rate of anode current, depending on the anode voltage.

A space-charge-limited diode (for that is what we are talking about) is therefore like a triode in which the part of the negative grid, controlling the flow of electrons to the anode, is played by the space charge. It is really like automatic grid bias, as produced by a cathode resistor. Any increase in cathode current

Fig. 3. Here there are, on the average, 8 electrons in fight all the time, but they start at irregular intervals. The resulting total current therefore has a slow irregular fluctuation as well as the fast saw-tooth ripple.

increases such bias and checks the increase in current. Similarly an increase in emission, due perhaps to mains voltage rising, increases the space charge and repels more electrons back to the cathode, so the current remains fairly constant.

The purpose of the negative grid in a triode is to supplement the space charge, reducing the anode current as required, even down to cutting it off altogether. So long as the potential of the grid is not allowed to rise high enough to attract electrons to itself, a triode behaves similarly to a diode as regards noise.
And how is that? Well, suppose the delegates were released from the conference room at short irregular intervals at such a rate that no crowd could form; the rate of flow across to the buffet would be equally irregular, as in a temperaturelimited valve. A crowd, on the other hand, would tend to even out such irregularities and ensure a relatively steady flow. This is much more certain with electrons, which, being free from individualities such as varying degrees of hunger, politeness or sociability, work according to simple and predictable laws. That is not to say that the quantitative working out is not even more difficult than in the temperaturelimited case. But having just seen that the space charge works rather like automatic grid bias, counteracting changes in anode current, we should not find it hard to understand that it has a cushioning effect on irregular emission, as indeed experience confirms.

In our ignorance of the extent of this noise-reducing or smoothing effect it can be brought into the formula for the temperature-limited valve as a factor, say S:

$$
\mathrm{I}_{\mathrm{N}}{ }^{2}=2 \mathrm{SeI}_{\mathrm{a}} \mathrm{~B}
$$

It is found that $S$ can be as small as $1 / 50$, signifying a 50 -fold reduction in noise, but something like $1 / 5$ is more usual. Note that this is in a formula for $I_{N}$ squared, which is proportional to noise power; don't confuse it with an equivalent factor $(=\sqrt{ } S$ ) that is sometimes used in the formula for $\mathbf{I}_{\mathrm{N}}$.
The question of noise in diodes seldom arisesexcept when they are used as frequency changers in microwave receivers-so we pass on to amplifying valves, especially those that might be considered for first-stage positions. One of the things designers want to do is compare valve noise with circuit noise. As we have seen, the only circuit that matters very much, usually, is the grid circuit of the first stage. The amount of effort worth expending on reducing this depends on how much noise the valve contributes. If circuit noise is given in microvolts at the grid, and valve noise in microamps in the anode circuit, one has to do calculations to find out which is greater,
and by how much. It is of course possible to find how many microvolts at the grid would cause that number of microamps of noise, but direct comparison with circuit noise microamps would be vitiated by the fact that both figures depend on the respective frequency bandwidths, and it would be a coincidence if they were equal.

So the convenient custom is to express valve noise in terms of the resistance which, if included in the grid circuir of a noiseless but otherwise similar valve, would cause the same amount of noise. Let us call this imaginary resistance $R_{s}$. According to the formula we already have for the noise voltage at the terminals of a resistor,

$$
\mathrm{V}_{\mathrm{N}^{2}}=4 k \mathrm{TBR}_{\mathrm{N}}
$$

$\mathrm{I}_{\mathrm{N}}$ is related to $\mathrm{V}_{\mathrm{N}}$ by the mutual conductance, $g_{\mathrm{m}}$ :

$$
\mathrm{I}_{\mathrm{N}}=g_{\mathrm{m}} \mathrm{~V}_{\mathrm{N}}
$$

So substituting $2 \operatorname{SeI} \mathrm{I}_{\mathrm{a}} \mathrm{B} / g_{\mathrm{m}}{ }^{2}$ for $\mathrm{V}_{\mathrm{N}}{ }^{2}$ in the previous equation we get

$$
\begin{aligned}
& \frac{2 S_{e I}, \mathrm{~B}}{\mathrm{~g}_{\mathrm{m}}{ }^{2}}=4 k \mathrm{TBR}_{\mathrm{N}} \\
& \therefore \mathrm{R}_{\mathrm{N}}=\frac{\mathrm{SeI}_{\mathrm{a}}}{2 g_{\mathrm{m}}{ }^{2} k \mathrm{~T}}
\end{aligned}
$$

This doesn't get us where we want, unless we know S. The rather difficult working-out I mentioned gives

$$
\mathrm{S}=\frac{1.28 k \mathrm{~T}_{\mathrm{c}} \mathrm{~g}_{\mathrm{m}}}{e \mathrm{I}_{\mathrm{a}}}
$$

where $T$. is the temperature of the cathode. Substituting this in the equation for $R_{N}$ we get

$$
\mathrm{R}_{\mathrm{x}}=\frac{0.64 \mathrm{~T}_{\mathrm{c}}}{\mathrm{~g}_{\mathrm{m}} \mathrm{~T}}
$$



Fig. 4. The many electrons emerging from the cathode, on the left, form a negative space charge that neutralizes the distant positive charge on the anode and restricts further movement towards it. The surplus electrons are forced back to the cothode.

The cathode temperature $\mathrm{T}_{\mathrm{c}}$ is usually about $1130^{\circ} \mathrm{K}$ and the room temperature T about $290^{\circ} \mathrm{K}$, so

$$
\mathrm{R}_{\mathrm{X}}=\frac{2.5}{\mathrm{~g}_{\mathrm{n}}}
$$

Because of the various assumptions in the smoothing theory, as well as the temperatures, this conveniently simple formula is far from precise, but it does at least give a rough idea. A moderate $g_{\mathrm{m}}$ for a triode is 2.5 mA , which makes $\mathrm{R}_{\mathrm{y}}=1 \mathrm{k} \Omega$.

At frequencies less than say $30 \mathrm{Mc} / \mathrm{s}$, noise resistance of this order is hardly worth bothering about as a noise source, compared with the usual grid-circuit resistance, to say nothing of manmade noise and atmospherics, both of which are worse at the lower frequencies. At very high frequencies, however, all these other noises are so much reduced that shot noise is appreciable and choice of first valve is correspondingly important. And at these same frequencies another effect adds to random noise. It concerns the electrons streaming through the grid.

We are still assuming that the grid is kept sufficiently negative to prevent any electrons landing on


Fig. 5. An electron flying from cothode to anode through a control grid induces a grid current as shown.
it. But during the first step of an electron's journey to the anode it is moving towards the grid, so it induces a small current in the grid circuit, just as was explained in connection with Fig. 2 for the anode circuit-though the waveform is unlikely to be quite the same. After it has passed through the grid it induces a current of opposite polarity; see Fig. 5. The duration of this cycle is of course the transit time, which we have already noted is usually of the order of $10^{-9} \mathrm{sec}$. At moderate signal frequencies this is too quick for the grid potential to change appreciably, so the positive and negative portions of Fig. 5 represent equal amounts of work, and cancel out. But at very high signal frequencies the grid potential changes appreciably during the transit time and the resulting inequality of positive and negative work causes a loss in the grid circuit. This loss can be represented by a resistance. Being in parallel with the grid circuit, it is one of the things that make the net resistance low at very high frequencies. At the moment, we are not directly concerned with that; what does concern us is that this resistance depends on the rate at which electrons pass the grid, and as that varies randomly the resistance varies and causes noise.
Even at low radio frequencies one naturally
uses a pentode rather than a triode, mainly for the sake of the screening it provides between grid and anode. The need for this is even greater at v.h.f. But it was soon found that at v.h.f., where noise is the limiting factor, a pentode is much noisier than a triode. This was traced to the fact that the second grid intercepts a fraction of the anode current. A row of waitresses with trays, widely spaced across the conference refreshment room, would waylay some of the delegates; each one thus intercepted would reduce the number arriving at the buffet, and thereby modulate that stream in a random manner.

It is easy to see that any current-taking grid introduces an additional random element and hence noise; but since the grid current is seldom as much as one-fifth of the whole, we might expect its noise contribution to be in that proportion and therefore not enough to worry about. But whereas the shot noise due to random emission, common to triodes and pentodes, is reduced by space charge to perhaps a tenth, there is no such luck with partition noise (as it is called). So it is usually at least double the cmission noise.
I haven't left myself space to go into details about its derivation, but the rough formula for pentodes, corresponding to $2.5 / \mathrm{g}_{\mathrm{m}}$ for triodes, is

$$
\mathrm{R}_{\mathrm{s}}=\frac{\mathrm{I}_{\mathrm{a}}}{\mathrm{I}_{\mathrm{n}}+\mathrm{I}_{\mathrm{g} 2}}\left(\frac{2.5}{\mathrm{~g}_{\mathrm{n}}}+\frac{20 \mathrm{I}_{\mathrm{g} 2}}{\mathrm{~g}_{\mathrm{m}}{ }^{2}}\right)
$$

Here the first term in the bracket is obviously the emission noise-the triode term reduced to the extent that the space current is diverted from the anode-and the second is the partition noise. Filling in values taken from data for a pentode designed for v.h.f. amplification:

$$
\begin{aligned}
\mathrm{R}_{\mathrm{N}} & =\frac{10}{10+2.5}\left(\frac{2.5}{7.4}+\frac{20 \times 2.5}{7.4^{2}}\right) \\
& =0.8(0.34+0.92) \mathrm{k} \Omega \\
& =270 \Omega+730 \Omega=1000 \Omega
\end{aligned}
$$

So the total is about 3 times what it would be as a triode ( $340 \Omega$ ). That accounts for the urge to use triodes for the first stages of sensitive v.h.f. receivers, however unsuitable they appeared for the task. The first expedient to make them work there was to arrange them with earthed grid as a screen between input and output. This method has now been generally superseded by the cascode arrangement, in which a double triode secures the advantages of a pentode without its disadvantages.

At the lower frequencies it is common practice to make the first valve the frequency changer. The higher the frequency, the greater the temptation to do so, since it becomes difficult to coax amplification at the original frequency. But the signal/noise ratio of frequency changers is several times worse than that of amplifiers, so even a little r.f. amplification is not to be despised. The reason is that the effective amplification of a frequency changer is related to $g_{c}$, the conversion conductance, which is only about $\frac{1}{3}$ of $g_{\text {m... }}$, whereas the noise is at least
(Continued on page 627)
as great; in fact, slightly greater, owing to the oscillator. The rough formula for a triode is

$$
\mathrm{R}_{\mathrm{N}}=\frac{4}{\mathrm{~g}_{\mathrm{c}}}
$$

and for a pentode

$$
\mathrm{R}_{\mathrm{N}}=\frac{\mathrm{I}_{\mathrm{a}}}{\mathrm{I}_{\mathrm{a}}+\mathrm{I}_{\mathrm{g} 2}}\left(\frac{4}{\mathrm{~g}_{\mathrm{c}}}+\frac{20 \mathrm{I}_{\mathrm{g} 2}}{\mathrm{~g}_{\mathrm{c}}{ }^{2}}\right)
$$

These speak for themselves, which is just as well, for the closing bell may sound any minute.
Lastly, there are the heptodes, hexodes, etc., which can be lumped together as multi-grid frequency changers. The more current-carrying electrodes, the more noise is introduced by random partition; so valves in this class are the worst of all, and would never be chosen for the first stage of a sensitive v.h.f. receiver. The formula is

$$
\mathrm{R}_{\mathrm{N}}=\frac{20 \mathrm{I}_{\mathrm{a}}\left(\mathrm{I} k-\mathrm{I}_{\mathrm{a}}\right)}{\mathrm{I} k \mathrm{~g}_{\mathrm{c}}^{2}}
$$

where $\mathrm{I} k$ is the cathode current. I suggest you look up a valve data list and work out a few values of $R_{N}$ in the various classes. Remember, of course, to use consistent units-mA and $k \Omega$, or $A$ and $\Omega$.

We have been concentrating rather heavily on the high end of the frequency scale, because that is where random noise is the thing that limits reception, and also the transit-time ingredient increases with frequency. For the sake of completeness mention should be made of flicker effect,
which seems to be due to comparatively slow changes at the cathode surtace causing random variations in emission. At any rate, it is important only at low a.f. and below, where it increases steeply. In badly made valves there is yet another cause of random noise: ionization, due to inadequate vacuum. Of course it inevitably occurs in gas-filled valves even when (in fact, because) they are properly made, and such valves are sometimes used as sources of random noise for testing. In case any reader is not up in the jargon, perhaps I had better mention that random noise, in so far as it is distributed equally over the whole frequency spectrum, like white light in the visible band, is often called white noise.

In brief: all valves are subject to shot noise because of random emission. In nearly all, this effect is considerably muffled by the space charge. So diodes and triodes on the whole are relatively quiet. Current carrying grids introduce further noise (partition), which is not muffled. So pentodes are usually several times worse than triodes. Valves used as frequency changers are relatively noisier than as amplifiers, because they amplify the signal about $\frac{2}{3}$ less. The order of preference of the different classes of valve in this role is the same as in the straight amplifying role. Beside shot and partition noise, which are very " white," there are gridadmittance noise which increases steeply at the v.h.f. end, and flicker noise which increases steeply at the a.f. end.

# Elements of Electronic Circuits 

20.-ADDITION AND SUBTRACTION CIRCUITS.

By J. M. PETERS, B.Sc. (Eng.), A.M.I.E.E., A.M.Brit.I.R.E.

MATHEMATICAL operations can be performed on waveforms, especially the square wave, with reasonably accurate results and use is made of this in the field of computation.

It is possible to carry out the operations of addition, subtraction, differentiation and integration without having to resort to valves or other non-linear devices, but accuracy is increased with the aid of the feedback amplifier. The operations of multiplication, division, squaring and the extraction of square roots (which will be dealt with later) are very difficult to perform using only linear elements such as resistor networks. Depending on whether accuracy or speed is the governing factor a choice can be made from several different methods. For example, in multiplication, electromechanical methods may give a high degree of
accuracy, but if speed is desired it will be necessary to use circuits with multi-electrode valves.

## Addition and Subtraction

Addition can take a number of forms and the process is met for example in (i) heterodyning in radio receivers, (ii) feedback, where a portion of the output of an amplifier is "added" to the input, (iii) a.g.c. in radio receivers, where the a.g.c. voltage is added to the bias of the earlier variable- $\mu$ stages. The simplest form of adding circuit involves linear passive elements such as the parallel-resistor adding circuit of Fig. 1.

$$
\begin{aligned}
v_{o u t} & =\left(v_{1} / \mathrm{R}+v_{2} / \mathrm{R}+v_{3} / \mathrm{R}\right)(\mathrm{R} / 3) \\
\therefore v_{o u t} & =\left(v_{1}+v_{2}+v_{3}\right) / 3
\end{aligned}
$$

Inductors and capacitors can also be used in simple adding networks of this sort.

The simplest form of subtraction circuit involves a phase-reversing device such as a valve or a transformer. One of the principal drawbacks to the simple addition circuit of Fig. 1 is that at high frequencies stray capacitance effects may distort


$$
\text { Fig. } 2 .
$$

the waveforms. This is overcome by inserting a small capacitor in parallel with each resistive branch and a resistor in parallel with the stray capacitances. The time constants (CR) are arranged to be the same. Series addition circuits are also used for the addition of voltage, e.g. Fig. 2 where transformers are used.

$$
v_{\text {out }}=v_{1}+v_{2}+v_{3}
$$

A balanced equal-arm bridge is a useful method of eliminating mutual coupling between the inputs. An example is shown in Fig. 3. Solving the mesh equations by Kirchhoff's laws it can be shown that

$$
v_{o u t}=(\mathrm{Z} / 2)\left[v_{1} /\left(\mathrm{Z}+\mathrm{Z}_{1}\right)+v_{2} /\left(\mathrm{Z}+\mathrm{Z}_{2}\right)\right]
$$

Difficulties are often encountered when it is desired to add voltages with short rise times such as rectangular pulses; but the method of overcoming the stray-capacitance effect described above is usually successful. Transients which produce overshooting and shock-excited oscillations can be minimized by damping and compensating networks. Waveform

shaping (described in an earlier instalment) can also be performed at the same time as addition and helps in compensating for waveform slow rise times.

## Use of Valves

The use of valves in addition and subtraction circuits provides a means of isolating the network from the source voltages and for changing the


Fiq. 4. impedance levels between network and source. Reasonable accuracy can be obtained by using multi - grid valves as shown in the subtraction circuit of Fig. 4.

$$
v_{\text {out }}=v_{1}-v_{2}
$$

Note that the low input impedance at the cathode ( $\approx 1 / g_{m}$ ) makes it necessary to supply the cathode from a lowimpedance waveform source.
Alternatively the application of signals to control and screen grids results in the output being the sum of these voltages. Sum and difference voltages


Fig. 5.


Fig. 0.
may be obtained from the cathode-coupled paraphase amplifier or long-tailed pair (described in No. 5, September, 1959) (Fig. 5). The difference in voltages appearing at the anodes ( $v_{o u t}$ ) is proportional to the difference between the inputs at the grids ( $v_{2}-v_{1}$ ). On the other hand, the cathode voltage, $v_{k}$, is proportional to the sum of the inputs to the grids $\left(v_{2}+v_{1}\right)$.

One of the most useful addition circuits is perhaps the parallel-addition circuit of Fig. 1 followed by stages of amplification with negative feedback (Fig. 6). This gives an output of inverted polarity, compared with the passive network. The advantages are that the amplitude of the output can exceed the input, also that with this circuit only one amplifier is necessary regardless of the number of inputs to be added.

$$
v_{o u t}=\left(\mathrm{R}_{\mathrm{i}} / \mathrm{R}\right)\left(v_{1}+v_{2}+v_{3}\right),
$$

as long as the gain within the feedback loop is high. This presupposes a high value for $A$, the amplifier gain, and a fairly low numerical value for $R_{j} / R$, which is the overall gain of the circuit with feedback.

## Commercial Literature

Thirty Transistor Circuits, covering a range of functions including a.f. amplifiers, regulated power supplies, d.c. to d.c. and d.c. to a.c. converters, oscillators and a photographic elec-tronic-flash unit are contained in "Newmarket TransistorsApplication Notes " from Newmarket Transistors Ltd., Exning Road, Newmarket, Suffolk.
Molybdenum Disulphide as a lubricant and additive to libricating solutions can bring about reductions in wear. Booklet on molybdenum disulphide, giving data on applications, from Rocol, Ltd., Rocol House, Swillington, near Leeds.
Silvered-Mica Capacitors made by Johnson Matthey have the electrodes screen-printed, instead of sprayed, on to the mica. The makers call these "Silver Star" and claim that the sharply defined edges of the electrodes aid stability. Booklet describing the process and listing the capacitors from the manufacturers at 73-83, Hatton Garden, London, E.C. 1 .

Radio and Television Interference. Advice on the tracing and suppression of interference is given in a booklet called "Evęrÿnan's Guide to Electrical Interference" published by Belling and Lee, Ltd., Gt. Cambridge Road, Enfield, Middlesex.

## DECEMBER MEFITNGS

Tickets are required for some meetings; readers are advised, therefore, to communicate with the secretary of the society concerned.

## LONDON

2nd. I.E.E.-Discussion on "Telemetering biological data" opened by Dr. G. H. Byford and Dr. K. G. Williams at 6.0 at Savoy Place, W.C.2. 2nd. Brit.I.R.E.-" Progress in microminiature circuit techniques for digital computers" by D. Roberts, Dr. D. S. Campbell and P. M. Thompson at 6.30 at the London School of Hygiene, Keppel Street, W.C.1.
7th. Institute of Physics and Physical Society.-" The international conference on physics education" by N. Clarke at 6.0 at 47 Belgrave Square, S.W.I.

7th. Brit.I.R.E.-" Flight evaluation of airborne electronic equipment" by H. G. Hinckley at 6.30 at the London School of Hygiene, Keppel Street, W.C.1.

8th. Television Society.-" The all transistor receiver: where do we stand?" by A. Landman (Plessey) at 7.0 at the Cinematograph Exhibitors' Association, 164 Shaftesbury Avenue, W.C.2.
8th. Radar \& Electronics Associa-tion.-" True motion radar" by J. H. Beattie at 7.30 at the Royal Society of Arts, John Adam Suree1, W.C.2.

12th. I.E.E.-"Topology concepts in network theory" by Dr. P. R. Bryant at 5.30 at Savoy Place, W.C. 2 .
13th. Brit.I.R.E.-" Objective and subjective requirements for loudspeakers" by F. H. Brittain at 6.30 at the London School of Hygiene, Keppel Street, W.C.1.

13th. I.E.E. Graduate and Student Section.-"Electrical engineering in the Naval Service" by Sir Hamish MacLaren at 6.30 at Savoy Place, W.C.2.

16th. B.S.R.A.-" Magnetic recording in the home cinema" by D. O'C. Roe at 7.15 at the Royal Society of Arts, John Adam Street, W.C.2.

20th. I.E.E.-Discussion on " Digital transducers" opened by L. Airey at 5.30 at Savoy Place, W.C.2.

21st. I.E.E. - "Microminiaturization" by L. J. Ward at 5.30 at Savoy Place, W.C.2.

## BELFAST

13th. I.E.E.-"Wave guides" by Dr. P. J. B. Clarricoats at 6.30 at the Civil Enginecring Department, Queen's Üniversity, Stranmillis Road.

## BIRMINGHAM

5th. I.E.E.--" Radiocommunication in the power industry" by E. H. Cox and R. E. Martin at 6.30 at the College of Technology.

## BRADFORD

6th. I.E.E.-" Thermistors-their theory, manufacture and application" by Dr. R. W. A. Scarr and R. A. Setterington at 6.30 at the College of Technology.

## BRISTOL

13th. Television Society.-"Ferrites in television receivers and aerials" by E. C. Snelling at 7.30 at the Hawthorns Hotel, Woodland Road, Clifton.

## CHESTER

12th. I.E.E.-" The potentialities of artificial earth satellites for radiocommunication" by W. J. Bray at 6.30 at the Town Hall.

## EDINBURGH

5th. Institute of Physics and Physical Socicty.-"Low noise amplifiers" by

Dr. R. A. Smith at 7.15 at the University.

## GLASGOW

6th. Institute of Physics and Physical Society.-"Low noise amplifiers" by Dr. R. A. Smith at 7.15 at the University.

## HOVE

I.E.E.-" Radiocommunication in the power industry" by E. H. Cox and R. E. Martin at 6.30 at 10 Queens Gardens.

## LIVERPOOL

14th. Brit.I.R.E.-"V.H.F. sound broadcasting" by L. G. Dive at 7.0 at the Adelphi Hotel.

## MALVERN

5th. I.E.E.-" Parametric amplifiers" by R. V. R. Carter and I. A. Bagnall at 7.30 at the Winter Gardens.

## MANCHESTER

1st. Brit.I.R.E.--" Industrial television" by I. M. Waters at 7.0 at the Reynolds Hall, College of Technology.

7th. I.E.E.-Discussion on "New semiconductor devices" opened by A. A. Shepherd at 6.15 at the Engineers' Club.
21st. I.P.R.E.-" Control systems" by J. Pyke at 7.30 at Central Hall, Oldham Street.

## NEWCASTLE-UPON-TYNE

5th. I.E.E.-" Radar observations of birds and 'angels'" by Dr. E. Eastwood at 6.15 at the Rutherford College of Technology, Northumberland Road. 7th. Brit.I.R.E.-" Human engineering " by S. G. Ramsay at 6.0 at the Institution of Mining and Mechanical Engineers, Neville Hall, Westgate Road.

## NEWPORT, I.O.W.

9th. I.E.E.-Discussion and demonstration of stereophonic sound at 6.30 at S.E.B. Showrooms.

## NORWICH

6th. I.E.E.-"Engineering aspects of commercial television programme presentation" by T. C. Macnamara at 7.30 at the Assembly House.

## READING

15th. Institution of Production Engineers.-" Electronic data processing allied to production engineering" by F. W. Purchall at 7.30 at the George Hotel, King Street.

## SOUTHAMPTON

6th. I.E.E.-"F.M./A.M. v.h.f. portable transistor receivers" by L. E. Jansson at 6.30 at the University.

14th. Brit.I.R.E.-"Some new piezoelectric devices" by A. E. Crawford at 7.0 at the University.

## STONE

12th. I.E.E.-" Recent developments in colour television " by I. J. P. James at 7.0 at Duncan Hall.

## SWANSEA

13th. I.E.E.-Faraday Lecture on "Transistors and all that" by L. J. Davies at 6.30 at the Brangwyn Hall.

## WORKINGTON

13th. I.E.E.-" The transmission of news film over the transatlantic cable" by C. B. B. Wood and I. J. Shelley at 7.0 at the College of Further Education.

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## RANDOM RADIETIIONS

## By " DIALLIST"

## Far-East Colour TV

SO Japan, or at least Tokyo and Osaka, has a colour television service. The man who after a threeyear struggle forced a reluctant government to make up its mind on the subject is said to be Mr. Matsutaro Shoriki, president of Nippon Television (NTV), who got the country's 525-line black-and-white service started in 1953. Ordinary television has become so popular in Japan that all but the very poorest now have receivers and the total is over six million. Mr. Shoriki is firmly convinced that there will be a similar boom in colour TV. To get things going he put up colour receivers in various public places. The system used is the N.T.S.C. and it is reported that the Japanese have improved upon the results obtained in the United States. They have apparently not succeeded in developing low-priced domestic receivers; but even though the earnings of factory and other workers are much less than they are in America, Mr. Shoriki is convinced that colour sets will sell. Japan, it appears, is well provided with "never-never" facilities.

## What About Us?

After five years of experimental colour transmissions the B.B.C. has announced that, technically at any
rate, it is ready to start a regular service, if and when permission is given. The radio industry doesn't feel that the time is yet ripe, for no manufacturer in his senses-and ours are very much in theirs-wants to go into the production of colour receivers unless there is likely to be a good demand for his wares. The big snag is the cost of the receiver, governed largely by the cost of the three-colour tube, and, frankly, I don't think that our people would go for them it: a big way unless this can be greatly reduced.

## DX Printing

THE setting of the type of the New York and Paris editions of the $N \in w$ York Times simultaneously is a very remarkable feat. The Paris edition now goes to press at the same time as the early edition in New York. The setting is done by a tape perforated by means of wireless signalsthe transatlantic cable will be used if radio conditions are bad-which is fed into linotype machines. Naturally some of the type-setting will be done in the ordinary way in Paris, for purely local American news won't be carried in the French edition, the space being devoted to happenings on this side of the Atlan:ic. Many of the advertisements will be different. Owing to the time difference-Paris is six hours ahead of New York-the

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Paris issue must derive from a very early American edition. But, of course, the U.S.A. is a large country and very early editions are needed if the paper is to appear on distant breakfast tables.

## Anti-mortar Radar

IN collaboration with the Royal Radar Establishment, E.M.I. have developed a new type of radar equipment for which substantial contracts have been placed by the War Office and by the Swedish army. This equipment enables mortar shells or bombs to be tracked in flight and makes it possible to calculate accurately and quickly the exact position from which they were fired. It is small and highly mobile and is said to be very easy to operate. Something of the kind was done towards the end of the last war with ordinary radar sets. They were used to spot shells as they left an enemy battery and proved very usefal in Italy for pin-pointing hostile battery positions with considerable accuracy.

## I.G.Y. No. 2

THERE are plans afoot, I see, for holding a second International Geophysical Year on a smaller scale at the time of the next sunspot minimum, which is due to come along in about four years' time. The idea is that we can get to know a good deal more if we compare the happenings recorded during a period of maximum solar activity with those which occur when such activity is at a minimum. An excellent scheme and may it have the success it deserves. Completely satisfactory explanations of some of the phenomena of I.G.Y. No. 1 haven't so far been found and the second series of observations may help to clear things up.

## TV Progress in France

AT long last television seems to be really getting into its stride in France. At the end of June this year the total of "declared" sets had risen to $1,700,000$. And there's pretty sure to be a largish number of undeclared, or pirate receivers, as undoubtedly there is in this country. The French transmitter, booster and relay net-
work is not yet complete and where transmissions have only recently begun the increase in the number of viewers has so far proved surprisingly slow. And that is unfortunately true of certain other regions which have had a service for some time. Alsace, for instance, can boast only 28,900 receivers, though the Strasbourg transmitter has been at work ever since TV began in France.

## Bigger and Bigger

THEY like big things in the U.S.A. and whatever sized TV screen comes along there's always a demand for something a bit larger. Having introduced 27 -inch sets comparatively recently some of the manufacturers now have plans for 29 -inch sets. I'd have thought that the best way of getting a really big picture would be to make use of projection techniques But that doesn't seem to have caught on over there any better than it has done here. Talking of screen sizes reminds me of a couversation between two fishermen that I overheard on a boat not long ago:
"How big's your TV set?"
"Fourteen inches."
"Oh, you want something bigger than that."
" "Does me all right."
"But don't you see you're missing a lot of the picture."
"How do you mean?"
"Stands to reason. A screen of that size isn't big enough to take the whole picture that's sent out. The same sized picture is sent to everyone; but if your screen is small, some of it laps over the edges and you can't see those bits."
"Oh, I hadn't thought of that."
Frankly, I hadn't either!


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## UNBIASED

## Unlicensed Sets

HOW many unlicensed sets are there in this country? I am told that a recent industrial survey covering car radio showed that only about $50 \%$ of the sets were licensed. It does not surprise me in the least to hear this opinion, and I think the proportion in the case of domestic sets is fully as bad.

Look for example, at the numerous houses which are let out in what arc euphemistically called one-room flats. In these days of miniaturized transistor sets, I wonder how many of the tenarits-some of whom are very transient-bother about a licence? I would go even further and say that in a great many cases the owners of these sets think that the licence-if any-of the landlord covers them.

When a householder changes his address he can fill in a form for letters to be redirected and one of the questions asked on the form is for particulars of his wireless licence, if any. If he has no licence he is asked to write "None." The nonlicence holder can write " None" with a perfectly clear conscience; at any rate that part of his conscience which deals with truth-telling will be clear, if, indeed, a conscience can be compartmented, as so many people obviously think, judging by their behaviour. It would seem to be more useful to ask if the form-filler had a receiver, but there would, of course, be little point in it unless it were made an offence to give a false answer.
But all the same, the very fact of a person having to fill in particulars of his licence or else write "None" might have the same psychological effect on a guilty conscience as does the presence of a G.P.O. wireless detector van in the neighbourhood.
Even so, only the householder is affected because it is clearly stated in the Post Office Guide (July, 1960 edition, page 67) that the P.M.G. will not undertake to re-direct the correspondence of those in lodgings, etc., and therefore, the people whom I regard as the real offenders, namely sub-tenants or lodgers, do not receive even the psychological stimulus of answering the questions about licences on the mail redirection form.

I am quite aware a G.P.O. inspector has authority to call at any time and inspect the wireless installation of a licence holder. But he certainly can't demand entrance to the home of a non-licence holder but must apply to a Justice of the Peace for a search warrant, after first satisfying the T.P. that he has reasonable grounds for suspicion. But Section 15 of the Wireless Telegraphy Act of

1959 only mentions a search warrant for premises. It makes no mention of a search of the person of a suspect which would, therefore, be ultra vires. In these days of tiny transistor sets, a radio receiver can easily be concealed in a man's pocket, or even in the corsage of a lady's dress. If an over-zealous Post Office official dared to search in the latter place, he would find himself on very dangerous ground indeed. Supposing the lady in her haste to conceal the set had inadvertently knocked the switch to the " on " position, it would be very galling but singularly apt if, as the perspiring official searched the room, the strains of "Music While you Work" burst forth from the depths of her person. He could only gaze at her in baffled impotence, knowing full well that he dared not search her. It would be truly an apoplectogenic situation.

At one time the outdoor aerial made it obvious when a TV set was installed but in these days more and more indoor aerials are in use, and before many years have gone by, we shall all be using sets with built-in aerials.

Could not a question about wireless licences be included in the census form which we shall all have to fill in on April 23 rd next, or better still, could not such a question be put in the form which we all receive annually for the compilation of the electoral roll.
In my opinion, if every person who ought to have a licence did, in fact, have one, the B.B.C.'s revenue would be increased considerably.

## Cymatology

RECENTLY, when reading an interesting article by a learned professor who is also a Fellow of the Royal Society, I was a little surprised to find that when he mentioned electromagnetic waves, he appended in parentheses the explanatory words, "light waves or radar waves." It was at once clear to me that he was not writing for his peers, who would naturally not need to be told what electromagnetic waves do for a living; nor, I realized, could he be writing for people having a slightly lesser standard of knowledge, possessed, for instance, by those who have only just matriculated in Wireless World readership.

It was only after further thought that it suddenly dawned on me that the writer (Professor J. L. Synge) had chosen the words " radar waves" with that careful and precise thought which one associates with an F.R.S., and that he may have intended it as an indirect rebuke to all those of us who use the vague and unsatisfactory
word "radio" in such phrases as "radio receiver" and "r.f. amplifier."
When all is said and done, the true definition of a radio wave is one which is radiated in any medium. There is nothing to restrict the word radiation (and, therefore, radio) to the electromagnetic spectrum except customary usage or, more correctly, misusage; if that be allowed to influence us we cannot cavil at the great majority of our fellow citizens who misuse the lovely old English abbreviation "ain't," and pervert it from its true meaning.
It is clear that electromagnetic waves are merely a subdivision of radiated waves (or, in other words, of radio waves). Sound waves also are undoubtedly radiated and, in like manner, if we drop a stone into a pond, do we not bring about the radiation of surface waves in the water in ever-widening circles? The O.E.D. lends support to my views as it tells me, inter alia, that radiation is "the manner in which the energy of a vibrating body is transmitted in all directions by a surrounding medium "; it does not define the medium.

The word "radar" which the professor uses has a very clear and precise meaning, for it was synthesized


Searching for my stud with e.m. waves.
from the initial letters of words defining a particular application of e.m. waves, and has nothing to do with the altosonic or supersonic waves which bats use for a similar purpose.
We ought, therefore, to rename our radio receivers, and also cease to talk of an r.f. amplifier. But we must be careful not to adopt the name e.m.w. receivers for, obviously, that could equally well mean a camera, and it would certainly lead to confusion if we spoke of an e.m.f. amplifier even though the set-up which we call an r.f. amplifier is, in fact, a voltage amplifier. I do feel that only a person with the precise mind of an F.R.S. could help us straighten out our nomenclature; but probably some of you may have some good ideas?

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At 28 gns., you'd be missing a great deal to pay less and there's no need to pay more. <br> \section*{Abridged specification <br> \section*{Abridged specification <br> 3 INDEPENDENT MO <br> 3 INDEPENDENT MOTORS}

4 RECORDING SPEEDS
FAST REWIND in either direction. $1,200 \mathrm{ft}$. reel rewound in 45 seconds.

WOW AND FLUTTER FREQUENCY RANGE:
Below $.05 \%$ at 25 ips $15 \mathrm{ips}: 50 / 16,000 \mathrm{c} / \mathrm{s} \pm 3 \mathrm{db}$
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$4304 \frac{1}{2}$ WATT-RESISTANCE $10 \Omega$ to $5 \cdot 1 \mathrm{M} \Omega \pm 1 \%$ (approved to RC2.D of RCS. 112 issue 2 from $10 \Omega$ to $1 M \Omega$ )
$4305 \frac{1}{4}$ WATT-RESISTANCE $4.3 \Omega$ to $4 \cdot 7 M \Omega \pm 1 \%$ (approved to RC2.E of RCS. 112 issue 2 from $10 \Omega$ to $1 \mathrm{M} \Omega$ )
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[^8]

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| $6 \frac{1}{2}$ in. | 6G | 6500 g | $21 / 6$ | $6 / 11$ |
| $7 \times 4 \mathrm{in}$. | 47G | 6500 g | $20 / 6$ | $6 / 7$ |
| $7 \times 3 \mathrm{in}$. | 37 G | 6500 g | $20 / 6$ | $6 / 7$ |
| $8 \times 3 \mathrm{in}$. | 38 G | 6500 g | $20 / 6$ | $6 / 7$ |
| $8 \times 5 \mathrm{in}$. | 58 C | 8500 g | $24 / 6$ | $7 / 10$ |
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R.F. LEAKAGE: Less than I microvolt.

# RUGGEDNESS 

"Yes it's a straightforward enough valve characteristic, but it is special because it will be maintained in spite of the punishment-shock and vibration-it will get in the equipment where it is installed.
You see, right the way through Mullard Special Quality rugged types are designed and built to the tightest specifications-tolerances are finely controlled, assembly methods are extremely accurate and welding techniques are of the highest standards. Indeed all the staff who work on them are carefully selected and specially trained-even completely new manufacturing techniques have been devised to make sure that the end product is as rugged as possible.
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 Ruggedness is only one feature of the Mullard Special Quality range; there are types available with extra long life expectation and others with closely controlled characteristics, some even have a combination of these features." "So far as information is concerned, Mullard can let you have all you need on any particular valve type. What's more, the people at Mullard House, who are pretty knowledgeable on these things, will always help in selecting the best valve for a particular job."

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Please write on your company letterhead for data sheets and any further information on the $\mathrm{OC}_{2} 8$.

## Abridged Data



| $\mathrm{V}_{\mathrm{cb}} \max .\left(\mathrm{I}_{\mathrm{e}}=0\right)$ | -80V |
| :---: | :---: |
| $\mathrm{V}_{\mathrm{ce}} \mathrm{max}$. (cut-off). | $-60 \mathrm{~V}$ |
| $\mathrm{V}_{\text {ce }} \max .\left(\mathrm{I}_{\mathrm{c}}=0.5 \mathrm{~A}\right)$. | -60V |
| $\mathrm{V}_{\mathrm{ce}} \max$ ( $\mathrm{I}_{\mathrm{c}}=6 \mathrm{~A}$ ) | $-60 \mathrm{~V}$ |
| $\mathrm{I}_{\mathrm{c}}$ max. | 6 A |
| $\mathrm{p}_{\text {tot }} \mathrm{max}_{\text {temperature }}$ at $4{ }^{\circ} \mathrm{C}$ mounting base..... | 30 W |
| $T_{j}$ max. (continuous rating) | $90^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{j}}$ max. (intermittent rating, 200 max.) | $100^{\circ} \mathrm{C}$ |
| $\bar{\alpha}^{\prime}\left(\mathrm{at} \mathrm{V}_{\mathrm{c}}=-\mathrm{IV}, \mathrm{I}_{0}=I \mathrm{~A}\right)$ | 20 to 55 |
| $\alpha^{\prime}\left(\right.$ at $\left.V_{\text {c }}=-I V, I_{c}=6 \mathrm{~A}\right)$ | 15 to 30 |

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TAPE AMPLIFIER UNIT KITS. Models TA-IM and TA-IS. This Combined Tape Record/ Replay Amplifier is available in both monophonic and Stereophonic models. Model. TA-IM can be modified to the stereo version with modification kit TA-IC.
TA-IM El6.14.0; TA-IS, E22.4.0; TA-IC

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E4.9.0
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Ell.8.6
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611.11.0

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| :--- |
| $\begin{array}{l}\text { El3.i } 8.6\end{array}$ |

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| (1) TYPE | DESCRIPTION | MAX. VSWR. |  |
| :---: | :---: | :---: | :---: |
| PP 4020 Q | Straight Waveguide Section | $\overline{7}$ | Length 5, 10 or 20 cm |
| PP 4025 Q | E.plane Bend | 1.07 | Radius of curvature 35 mm |
| PP 4030 Q | H-plane Bend | 1.07 | Radius of curvature 37 mm |
| PP 4035 Q | Twist | 1.07 | Length 50 mm |
| PP 4050 Q | Hybrid Tee | 15 | Decoupling $>35 \mathrm{~dB}$ |
| PP 4080 Q | Horn | I. 15 | Directivity: E-plane $15^{\circ}, \mathrm{H}$-plane $16^{\circ}$ |
| PP 4130 Q | Variable Flap Attenuator | 1.15 | Max. attenuation $>20 \mathrm{~dB}$ <br> max. mean power 200 mW |
| PP 4150 Q | Variable Rotary Attenuator | 1.15 | Max. attenuation 50 dB , accuracy $\pm 3 \%$ |
| PP 4170 Q | Low-power Matched Load | 1:05 | Max. mean power I W |
| PP 4200 Q | Klystron Mount | - | When using klystron 55,335 the output power is 100 mW |
| PP 4222 Q | Adjustable X-tal Mount | 1.25 | $50 \Omega$ BNC-connector |
| PP 4260 PP 4270 | Calibrated Short Circuit | from $100^{50}$ | Accuracy of the displacement 0.02 mm |
| PP 4270 Q | Sliding Screw Tuner | from 10 to 1.03 | Insertion loss for a VSWR of 10 is $\geqslant 2 \mathrm{~dB}$ |
| PP 4300 PP 4382 | Broadband Wavemeter | 1.20 1.03 | Relative accuracy $5.10^{-4}$-loaded $Q>3000$ Accuracy of the displacement 0.01 mm |
| PP 4382 Q | Standing Wave Detector | 1.03 | Accuracy of the displacement 0.01 mm max. probe penetration $1 \mathrm{~mm}, 50 \Omega$ BNC-connector |
| PP 4420 Q | Ferrite Isolator | 1.15 | Freq. range $33-36 \mathrm{kMc} / \mathrm{s}$, forward att. $<1 \mathrm{~dB}$ reverse att. $13-26 \mathrm{~dB}$, max. mean power 200 mW |

Additional Instruments; Klystron Supply, type 4485 - D.C. Microvoltmeter, type GM 6020

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Heater
$\mathrm{Vh}=6.3 \mathrm{~V} \quad \mathrm{Ih}=0.55 \mathrm{~A}$

## Capacitances

Cx＇ーx＇．．．．．1．6pF．
Cy＇ージ ．．．．．．1．5pF．
One X plate to all other electrodes
less other X plate－3．0pF．
One $\mathbf{Y}$ plate to all other electrodes
less other Y plate－2．7pF．

## Typical Operation

|  |  | 5BUP31 |
| :--- | :---: | :---: |$\quad$| 5BVP31A |
| :--- |

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\begin{array}{r}
5 \mathrm{~V} \\
2.5 \mathrm{~V} \\
4 \mathrm{~V} \\
2 \mathrm{~V} \\
3.5 \mathrm{~V} \\
2 \mathrm{~V} \\
10 \mathrm{~V} \\
10 \mathrm{~V} \\
10 \mathrm{~V} \\
10 \mathrm{~V} \\
10 \mathrm{~V} \\
10 \mathrm{~V} \\
20 \mathrm{~V} \\
20 \mathrm{~V} \\
5 \mathrm{~V} \\
22 \mathrm{~V} \\
28 \mathrm{~V}
\end{array}
$$



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| :---: | :---: |
| 6 V | 15 A |
| 6 V | 50 A |
| 6 V | 100 A |
| 12 V | 10 A |
| 12 V | 20 A |
| 12 V | 30 A |
| 12 V | 60 A |
| 12 V | 105 A |
| 12 V | 210 A |
| 12 V | 1,000 A |
| 24 V | 12 A |
| 24 V | 20 A |
| 24 V | 30 A |
| 24 V | 60 A |
| 24 V | 105 A |
| 24 V | 200 A |
| 24 V | 750 A |


| $\underset{\substack{£ \\ f \\ \hline}}{ }$ |
| :---: |
| £66 |
| ¢15 |
| 622 |
| £28 |
| ¢45 |
| 662 |
| ${ }_{\text {¢ }}^{685}$ |
| ¢185 |
| $\underline{63}$ |
| E27 |
| 633 |
| ¢41 |
| 670 |
| $\pm 86$ |
| ¢262 |


| 36 V | 10 A | 626 |
| :---: | :---: | :---: |
| 36 V | 20 A | ¢ 62 |
| 36 V | 40 A | 442 |
| 36 V | 60 A | 655 |
| 110 V | 5 A | 632 |
| 110 V | 10 A | 642 |
| 110 V | 15 A | 653 |
| 110 V | 20 A | 667 |
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## DECEMBER 1960

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# "BELLING-LEE" NOTES 

## No. 23 of a Series <br> Fusing, part 7

A convenient way of expressing the degree of protection afforded by a fuse is by stating the "Fusing Factor'"; this is the ratio of the minimum value of current which will blow it to the current rating. Thus a 2 amp . fuse having a fusing factor of $2: 1$ will not operate normally at any current below 4 amp ., although it should be appreciated that if it is subjected to currents between 2 and 4 amp . there will be an excessive temperature rise (possibly beyond what is permitted) and, depending on the material of which its element is made, embrittlement may result, followed ultimately by failure as we have seen earlier.

Now fuselinks having different types of construction and performance can have similar fusing factors, and so, useful as this characteristic is, it gives no indication of the mode of operation, which is a function of current and time. The readiest way of expressing this is pictorially, in the form of graph. In its conventional form, this is a curve
current (r.m.s. value) would give the same result as D.C.

A typical set of characteristic curves is shown for four different types of fuses having the same rating and fusing factor. The usefulness of these curves can be considerably extended by plotting them as envelopes containing the spread of results obtained in practice due to random variations between different specimens, as we indicated in part 3 of this series. The upper boundary then provides information on the absolute protection given by that particular type of fuse, while the lower boundary defines the maximum isolated surge that the fuse will withstand.

Curves of this type are indispensable when considering problems of discrimination. In any system which includes two or more circuit protection devices arranged in series, it is obviously undesirable that a fault should operate more than one of them. For instance, in an ordinary domestic electrical installation there are a number of individually fused outlets (sub-circuits) fed through a main circuit fuse, and it would be most inconvenient if a fault arising in one of the sub-circuits blew the

COMPARISON OF FUSE CHARACTERISTICS

drawn through a minimum of six points plotted on Cartesian coordinates whose axes are prospective current, and pre-arcing time, shown on logarithmic scales. The current source which is used to obtain the results so plotted is usually D.C., since an alternating supply would give misleading results at very short pre-arcing times because they would depend on the instantaneous value of the current at the time of applying the overload; at fairly long prearcing times, of course an alternating
main fuse thus putting all the other sub-circuits out of action; in some situations it could even be dangerous. The characteristics of the main fuse must therefore be such that it is able to discriminate between a main circuit fault and an overload in a sub-circuit and the simplest way of ensuring this is by seeing that its performance curve nowhere intercepts those of the sub-circuit fuses.

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The method is also useful for testing for discontinuities in joints, e.g. crimped joints, in conjunction with continuity resistance measurements made before and after exposure. It is complementary to normal salt spray tests, and somewhat easier to evaluate.

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# Aspects of design 



This is No. 29 in the series of articles dealing with advanced problems in circuit design published by The Ediswan Mazda Applications Laboratory. No. 30 will appear next month. We shall be pleased to answer queries arising from this or other articles.
Reprints of the first twenty-four articles, in booklet form, are available on request.

The operation of line deflection output stages of television receivers was dealt with in "Aspects of Design" numbers 8 and 26. It was pointed out in "Aspects of Design" number 26 that an output stage can be operated more efficiently if the ratio of the number of turns on the efficiency diode winding of the transformer to the number of turns on the pentode winding is kept as high as possible. This results in a high value of boosted HT voltage which, although assisting in producing good efficiency, subjects the pentode and efficiency diode to higher peak voltages during the line flyback interval. The Maximum Peak Voltage ratings of diode and pentode often set a limit to which this course can be pursued. By suitable proportioning of transformer leakage inductance and self-capacitance a condition of tuning can be achieved which allows these peak voltages to be considerably reduced. This process is often termed 3rd Harmonic Tuning.

This technique can best be understood by considering the equivalent circuit of a line output stage.

Fig. 1 shows the equivalent circuit of a typical line output stage in which a winding or "overwind "is included to step up the peak voltages during flyback for the purpose of generating the cathode ray tube extra high voltage supply. The various elements of the circuit are shown as referred to the primary of the transformer. $\mathrm{L}_{1}$ represents the parallel combination of deflector coil load inductance and primary inductance. $\mathrm{L}_{2}$ represents the leakage inductance between dentode anode connection and EHT rectifier connection

$\mathrm{C}_{2}$ and $\mathrm{C}_{3}$ are the small capacitances associated with this inductance. $C_{1}$ which is large provides most of the tuning of the inductance load. The switch represents the combination of pentode and efficiency diode. At the end of the deflection stroke $\mathrm{L}_{1}$ carries a high current which is rising linearly at a rate slow compared with the natural resonance periods of the circuit. The switch is opened and the circuit commences a free oscillation. At the end of the flyback interval the current in $\mathrm{L}_{1}$ has reversed and has reached a maximum in the opposite direction. Very shortly after this time there comes a moment when the rate of change of current in $\mathrm{L}_{1}$ and therefore the voltage across $\mathrm{C}_{1}$ have the same value as they had during the previous deflection stroke. The switch is now closed once more and this current is constrained io collapse linearly to zero at a rate $L_{1} \frac{d_{1}}{d t}=V_{B}$. This linearly decaying current flows into the battery returning energy in doing so.

If there is current in $L_{2}$ or charge on $\mathrm{C}_{2}$ at the moment of closing the switch the circuit $\mathrm{L}_{2} \mathrm{C}_{2} \mathrm{C}_{3}$ will continue to oscillate during the next stroke. In any practical circuit the switch comprising the pentode and efficiency diode has finite resistance which serves to couple oscillations of $\mathrm{L}_{2} \mathrm{C}_{2} \mathrm{C}_{3}$ into the load inductance $\mathrm{L}_{1}$ where they produce velocity modulation of the trace. In addition the energy of this circuit $\mathrm{L}_{2} \mathrm{C}_{2} \mathrm{C}_{3}$ will be dissipated in its losses and will not be available for recovery by the energy recovery process. The aim of third harmonic tuning is that at the end of the flyback interval there shall be no current in $\mathrm{L}_{2}$ and no charge on $\mathrm{C}_{2}$. This implies that at the end of the flyback interval the current in $\mathrm{L}_{2}$ is zero and the rate of change of current is also zero. During the free flyback oscillation $\mathrm{L}_{2}$ can be supposed to carry components of current at the two natural frequencies of the circuit. These frequencies are $f_{1}$, a low frequency manly governed by $\mathrm{LI}_{1} \mathrm{C}_{1}$ and $\mathrm{f}_{2}$ a higher frequency governed mainly by $\mathrm{L}_{2}$ in parallel with $\mathrm{C}_{2}$ and $\mathrm{C}_{3}, \mathrm{C}_{1}$ appearing as a very low impedance.

Fig. 2(a) shows a plot against time of these two components of current ' $\mathrm{n} \mathrm{L}_{2}$ during the flyback interval with the circuit constants critically adjusted for the " 3 rd Harmonic" tuning condition. This occurs when $f_{2}$ is slightly less than $3 f_{1}$. The total current in $\mathrm{L}_{2}$ is the sum of these components and this is shown in Fig. 2(b). It is seen that this condition achieves the requirement of current in $L_{2}$ and rate of change of current being simultaneously zero at the beginning and end of the flyback interval. The voltage waveform across $L_{2}$ is shown by the dotted curve in Fig. 2(b) (voltage of point $P$ with respect to pont $\mathbb{Q}$ ). The effect of this is to depress the voltage peak at $P$, the pentode anode, and increase the voltage peak at Q , the EHT rectifier point.
These waveforms are shown in Fig. 3. The waveform at Q , of course, is of greater amplitude than that at $\mathbf{P}$ by the step up action of the transformer ratio which, for simplicity, has been omitted from the equivalent circuit.




In practice $\mathrm{L}_{2} \mathrm{C}_{2}$ and $\mathrm{C}_{3}$ are governed by the transformer winding configuration and only $C_{1}$ is under control by adding capacity across the transformer primary. The procedure in design is to adjust $C_{1}$ for the tuning condition by observing cancellation of ringing on say the pentode anode waveform and then measuring flyback time. If, for instance, flyback time is too short then the winding must be redesigned for greater leakage inductance $\mathrm{L}_{2}$ or greater self capacitance of the EHT overwind to reduce $f_{2}$ and the transformer re-tuned for the " 3rd Harmonic" condition.

The Ediswan Mazda 30P4 and 30P19 are processed to have good high voltage performance to give the engineer scope for achieving very efficient designs using this procedure.

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$$
\begin{array}{lll}
\text { Heater Current (amps) } & \mathrm{I}_{\mathrm{h}} & 0.3 \\
\text { Heater Voltage (volts) } & \mathrm{V}_{\mathrm{h}} & 16
\end{array}
$$

## TENTATIVE RATINGS AND DATA

 Maximum Design Centre Ratings| Anode Dissipation (watts) |  | Triode | etrod |
| :---: | :---: | :---: | :---: |
|  | $\mathrm{pa}_{\text {(max) }}$ |  |  |
| Screen Dissipation (watts) | $\mathrm{Pg}_{2}($ |  |  |
| Peak Anode Voltage (Pulse positive) |  |  |  |
| Peak Anode Voltage (Pulse negative) (kV) |  |  | 0.5 |
|  |  | 25 | 250 |
| Screen Voltage (volts) <br> Heater to Cathode |  |  | 250 |
|  |  |  |  |
| Voltage (volts rms) | $\mathrm{V}_{\mathrm{h}-\mathrm{k}(\mathrm{max}) \mathrm{rms}}$ | $150 \dagger$ | 50† |
| Mean Cathode Current (mA) |  |  | 75 |
| Resistance Grid 1 to Cathode |  |  |  |
| Self Bias (M $\Omega$ ) |  |  |  |
| Fixed Bias ( $\mathrm{M} \Omega$ ) |  |  |  |

*Maximum pulse duration $5 \%$ of one cycle with a maximum of 1 msec .
$\dagger$ Measured with respect to the higher potential heater pin.

## TRIODE CHARACTERISTICS

| Anode Voltage (volts) | $V_{\text {a }}$ | 200 |
| :---: | :---: | :---: |
| Anode Current (mA) | $\mathrm{I}_{\text {a }}$ | 10 |
| Grid Voltage (volts) | $\mathrm{V}_{81}$ | -7.7 |
| Mutual Conductance ( $\mathrm{mA} / \mathrm{V}$ ) | gm | . 4 |
| Amplification Factor |  | 18 |



TETRODE OPERATION IN FRAME TIME BASE
Allowance must be made in circuit design, not only for component variation, but for valve spread and deterioration during life. Values of total tetrode peak anode current, for an average valve when new and at the assumed end of life point for any valve, are as follows:-

|  | $\mathrm{V}_{\mathrm{g}}$ | $\mathrm{V}_{\mathrm{g} 2}$ | $\mathrm{~V}_{\mathrm{g1}}$ | $\mathrm{I}_{\mathrm{g}}$ |
| :--- | :--- | :--- | :--- | ---: |
|  | $(\mathrm{V})$ | N | (V) | $(\mathrm{mA})$ |
| Average New Valve | 55 | 170 | -1 | 210 |
| Asumed End of Life Condition | 50 | 170 | -1 | 135 |
| Average New Valve | 55 | 185 | -1 | 235 |
| Assumed End of Life Condition 50 | 185 | -1 | 151 |  |

Mounting Position: Unrestricted
Base: B9A (Noval)

## Tentative Characteristic Curves of Ediswan Mazda Transistors Type 30PL14





VIEW OF FREE END


Maximum Dimensions (mm)
Overall Length
78.5

Seated Height
Diameter
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# - Voitción quality equipment 

The W.V.B. recorder has an additional amplifier and head with provision for "before" and "after" record monitoring while the recording is in progress, and this also has echo facilities.
The W.V.A. recorder has provision for a plug in stereo head and can be supplied with this and stereo playback pre-amplifiers with equalisation each having an output of I volt from a cathode follower. This is type W.V.A/S.

A heavy mumetal shielded microphone transformer is bullt in for $15-30$ ohms balanced and screened line, and requires only 7 micro-volts approximately to fully load. This is equivalent to 20 ft . from a ribbon microphone and the cable may be extended to 440 yds . without appreciable loss.
The 0.5 megohm input is fully loaded by 18 millivolts and is suitable for crystal, P.U.' 's, microphone or radio inputs.

The playback amplifier may be used as a micro-
phone or gramophone amplifier separately or whilst recording is being made.

The meter fitted for reading signal level will also read bias voltage to enable a level response to be obtained under all circumstances. A control is provided for bias adjustment to compensate low mains or ageing valves.

The power output is 4 watts heavily damped by negative feedback and an oval internal speaker is built in for monitoring purposes.

## THE VORTEXION W.V.B. or W.ViA/S

are eminently suitable for making a high quality recording almost indistinguishable from the original since these models have facilities for monitoring the recording actually put on the tape with only a fraction of a second delay.

By this means, when for any reason the signal is distorted or not as required, the result of the recording on the tape can be heard almost instantly, and adjustments can be made until the results are as required.

Many types of music today have the treble boosted considerably, and may result in greater power being recorded at high frequencies than at
middle frequencies, an overload of the tape at high frequencies gives a mushy quality with lots of hiss and background noise.

Adjustment to the bias level while listening to the result is useful in this connection especially where the brand of tape and the bias setting for it are not exactly known.
Again if clean treble recordings at $3 \frac{3}{4} \mathrm{in}$. are of prime importance it is now recognised that no other method is quite so effective in achieving this as reducing the bias slightly while listening to the results. The meter reading of the new bias setting for the particular tape used may be noted for future use.

# Vortevion quality equipment 

Will deliver 120 watts continuous signal and over 200 watts peak Audio. It is completely stable with any type of load and may be used to drive motors or other devicess to over 120 watts at frequencies from 20,000 . down to 30 cps in standard form or other frequencies to order. The distortion is less than $0.2 \%$ and the noise level -95 dB . A floating series parallel output is provided for $100-120 \mathrm{~V}$. or $200-250 \mathrm{~V}$. and this cool running amplifier occupies $12 \frac{1}{4}$ inches of standard rack space by 11 inches deep. Weight 60 lb .

## $30 / 50$ WATT AMPLIFIER

Gives 30 watts continuous signal and 50 watts peak Audio. With voice coil feedback distortion is under 0.1\% and when arranged for tertiary feedback and 100 volt line it is under $0.15 \%$. The hum and noise is better than -85 dB referred to 30 watt.
It is available in our standard steel case with
 Baxendale tone controls and up to 4 mixed inputs, which may be balanced line 30 ohm microphones or equalised P.U.s to choice.
$120 / 200$ WATT AMPLIFIER


## ELECTRONIC MIXER/AMPLIFIER

This high fidelity $10 / 15$ watt Ultra Linear Amplifier has a built-in mixer and Baxendale tone controls. The standard model has 4 inputs, two for balanced 30 ohm microphones, one for pick-up C.C.I.R. compensated and one for tape or radio input. Alternative or additional inputs are available to special order. A feed direct out from the mixer is standard and output impedances of $4-8$ - 16 ohms or 100 volt line are to choice. All inputs and outputs are at the rear and it has been designed for cool continuous operation either on $19 \times 7$ in. rack panel form or in standard ventilated steel case.
Size $18 \times 7 \frac{1}{2} \times 9 \frac{1}{2}$ in. deep.
Price of standard model $£ 49$.

The 12-way electronic mixer has facilities for mixing 12 balanced line microphones. Each of the 12 lines has its own potted mumetal shielded microphone transformer and input valve, each control is hermetically sealed. Muting switches are normally fitted on each channel and the unit is fed from its own mumetal shielded mains transformer and metal rectifier.

Also 3-way mixers and Peak Programme Meters. 4 -way mixers and $2 \times 5$-way stereo mixers with outputs for echo chambers, etc. Details on request.

## 12-WAY ELECTRONIC MIXER



Full details and prices of the above on request


## Evidence in Camera



Of interest not only for its story, this picture has provided (quite unintentionally) striking evidence of the reputation enjoyed by LEAK. It is a typical incident of the use of LEAK equipment by professional audio engineers in broadcasting and recording studios throughout the world, who choose LEAK for quality of performance and reliability Does your installation measure up to these standards? If it does not, your LEAK Dealer can help you. The prices of LEAK studio quality equipment are made possible only by world-wide sales.


The new LEAK Varislope Stereo pre-amplifler (illustrated above) Incorporates faclities which make it the most comprehensive pre-ampllifer presently avallable.

PRICE $£ 25$
We shall be pleased to send you a copy of Thomas Helnitz' review of thls "Remarkable new control unit for stereo" reprinted from "Records and Recording:"
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## Monaural

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## Television Noise Limiting



The current November issue of ELECTRONIC TECHNOLOGY includes among other interesting articles one in which methods of noise limiting in the sound channel of television receivers are discussed. The author considers various well-known noise-limiting circuits, draws attention to some of their defects, and shows how they may be overcome to provide improved performance.

Design criteria and practical circuit details are given for a number of noise-limiting circuits and it is shown that a considerable improvement in noise suppression of an indifferent receiver can often be made without resorting to expensive and complicated circuitry.

(b)

## ARTICLES

## IN THE DECEMBER ISSUE INCLUDE:

## SWITCHING TRANSISTORS

Various methods used for predicting the performance of switching transistors are described in this article and their validity under conditions of large collector current is discussed. Different types of transistor are included and it is shown that, at high levels, evaluation of switching times in terms of the charge control is to be recommended with certain reservations.

TWO-SIGNAL BRIDGES
In this article the author shows that by a simple modificam tion to the usual bridge-circuit arrangement it is possible for the input and output of the bridge to have a common terminal. This arrangement can also be used to reduce stray admittances and some arrangements enable the bridge to be remotely situated.

Electronic Technology retains all the familiar features of Electronic \& Radio Engineer including the well-known Abstracts and References section. Regular readership will keep you in constant touch with progress in the entire field.

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"COMMANDER"' Transistorised.

Power Supply: $12 / 24 \mathrm{v}$. D.C. at 2.85A.

Finished in grey dimple stove enamel.
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## brief specification

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Vollage (120-140 kc/s) across unknown
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Load Resislance Limits. The accuracy is
maintained with the following loads:
Inductance: $20 \mathrm{k} \Omega$ shunt, $10 \Omega$ serles
Capacitance: 0.1 Ma shunt
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 transistor scaleddown for ultra-


The OC123 is a ferrite core driver transistor similar to the OC23, but scaled down into a small metal can for those applications where space saving is of cardinal importance--such as in closely packed printed circuit boardsand where the dissipation rating need not be so high as with the OC23.

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The OC123 is particularly suitable for gating current pulses generated by its larger companion type OC23 for driving ferrite cores. In such an application, a. 350 mA 2 microsecond pulse with a leading edge rise time of 0.4 microsecond can be passed by the transistor when it is fully bottomed.

Abridged details are given below-for full data please write to Mullard House.

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| $V_{c b} \max .\left(1 e^{\prime}=0\right)$ |  | -50 V |
| :---: | :---: | :---: |
| $\mathrm{V}_{\text {ce }}$ max. $\left(V_{\text {be }}>+0.5 \mathrm{~V}\right)$ | - | -50 V |
| $\mathrm{V}_{\text {ce }} \max .\left(\mathrm{I}_{\mathrm{C}}=0.5 \mathrm{~A}\right)$ |  | -25 V |
| $\mathrm{i}_{\mathrm{c}}(\mathrm{pk})$ max. |  | 0 A |
| $\mathrm{I}_{\mathrm{C}}$ (av) max. (averaging time 20 ms ) |  | 0.5 A |
| $p_{\text {tot }}$ at $45^{\circ} \mathrm{C}$ ambient |  | 200 mW |
| $\mathrm{T}_{\mathrm{j}}$ max. |  | $90^{\circ} \mathrm{C}$ |
| Junction temp. rise above |  |  |
| Junction temp. rise above cas | ase | $0.06{ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $V_{C}(\mathrm{knee})\left(1 c^{\prime}=400 \mathrm{~mA}\right)$ |  | -350mV |
| $\mathrm{f}_{1}\left(\mathrm{~V}_{\mathrm{c}}=-2 \mathrm{~V}, \mathrm{I}_{\mathrm{c}}=100 \mathrm{~mA}\right.$, | , $\left.\mathrm{T}_{\mathrm{j}}=25^{\circ} \mathrm{C}\right)$ | $1.5 \mathrm{Mc} / \mathrm{s}$ |

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## 29

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 Primary $200 / 250$ v. 50 cycles. Secondary 2,000/1,500/0/1,500/2,000 V. 500 milliamps. Supplied brand new and boxed, $£ 6 / 10 /=$ ea. P/P 10/-.PARMEKO TABLE TOP TRANSFORMERS. Input $230 \mathrm{v}, 50 \mathrm{c} / \mathrm{s}$. Output 6201550 $375 / 0 / 375 / 550 / 620 \mathrm{v} .250 \mathrm{~mA}, 5 \mathrm{v} .3 \mathrm{amp}$ 5 v .3 amp . Slze $6 \frac{1}{} \times 6 \frac{1}{} \times 5 \frac{1}{2} \mathrm{in}$. Brand new, boxed, 45/- ea. P/P 5/-
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| :---: | :---: | :---: | :---: |
| 2 tin . | Perdio | 3 ohm | 17/6 |
| 2tin. | Perdio | 15 ohm | 17/6 |
| 3 in . | Plessey | 5 ohm | 15/6 |
| 3 in . | Rola | 3 ohm | 17/6 |
| $4 \frac{1}{2} \mathrm{in}$. | Plessey | 3 ohm | $15 / 6$ |
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| 6 tin. | Plessey | 3 ohm | 1716 |
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| 10 in . | R.A. | 3 ohm | $27 / 6$ |
| $6 \times 4 \mathrm{in}$. | Plessey water | 3 ohm | 12/6 |
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| $8 \times 6 \mathrm{in}$. | Rola | 3 ohm | $17 / 6$ |
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Complete equipment comprises search head amplifier, headset, control box, telescopic rods for search head, search head test unit, test measure and haversack.
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 These special units contain a 1 microamp. Tinsley mirror galvonometer and a double selenium photo electric cell. Brand new, 69/19/6 еа., P/P. 7/6.



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Large type Bin, $\times 2$ inn. Supplied brand new boxed
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METER BARGAINS
20 microamp D.C. M/C flush rd. $2 \frac{1}{2} i$ 25 microamp D.C. M/C proj. rd. $2 \frac{1}{2}$ in. 50 microamp. D.C. M/C proj, rd. $2 \frac{1}{2}$ in. 100 microamp D.C. M/C flush rd. 2 in. 100 microamp D.C. M/C flush rd. $3 \frac{1}{2} \mathrm{in}$. 200 microamp D.C. M/C proj. rd. $2 \frac{1}{2}$ in. 300 microamp D.C. M/C flush rd. $2 \frac{1}{2}$ in. I milliamp. D.C. M/C flush sq. 2 in.... milliamp. D.C. M/C flush rd. $2 \frac{1}{2}$ in. 1 milliamp. D.C. M/C flush sq. 4 in. 30/0/30 milliamp. D.C. M/C flush $2 \frac{1}{2}$ in.
15 rd. . D.C. M/C. proj. rd. 2 in.
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components supplied. Valve line-up: $2-6 \mathrm{SG} 7^{\circ}$ s $6 \times 5 \mathrm{GT}$ and 6 V 6 GT . Chassi ready drilled. Cabinet size $10 \frac{1}{2}$ in. $x$ IOin. wide. Maximum depth at base 5 in . tapering to $3 \frac{1}{2} \mathrm{in}$, at top. Sloping front. Very attractively finished in light walnut and peach Each component brand new and tested prior to packing. Complete instruction booklet with practica and theoretical diagrams is provided. Booklet available at $1 / 6$ post free. Our price complete post free. Our price complete
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To build yourself Medium and Long waves-Push-Pull Superhet A.V.C. Perfect Car Radio reception. Size Join. $x 6 \frac{3}{4} i n . x$ $4 \frac{1}{2}$ in. at base tapering to 4 in. at top.
Very attractive two-tone grey Vynide covered cabinet with black and gold printed escutcheon plate, eream and gold knobs, handle and cabinet fittings. $t$ Weight-complete with long-life $7 \frac{1}{2}$ volt battery- $4 \frac{1}{2} \mathrm{lb}$. + Mazda high-grade transistors throughout. * High-Flux Tin. $x$ 4in. Elliptical Speaker. t Slow motion tuning. t Co-axial socket at rear for direct connection to Car Radio Aerial. * Improved reception by use of seven-section plated telescopic aerial disappearing into Cabinet when closed, 34in. above Cabinet
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Construction simplified by Bakelite chassis board with the following components already mounted I.F. Transformers (3). Oscillator Coil, Trimmer Bank, Output Transformer, Interstage Trans former, Aerial Brackets and Earth Bar. SPECIAL INCLUSIVE PRICE for all required components, full assembly instructions-nothing more to buy-is $£ 10 / 19 / 6$ plus $3 / 6$ P. \& P. Alignment service available. Full assembly instructions and individually priced parts list, all of which are available available. Full assembly
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(12) JASON "ARGONAUT" AM/FM Tuner.
(13) F.M. Power Pack (suitable for most tuners)
$\left\{\begin{array}{l}14 \\ 15 \\ \text { R.C. } 3 / 4 \text { watt Amplifier (with Bass, Middle and Treble controls) }\end{array}\right.$
(15) 2-amp. Battery Charger
(I6) R.C. Transistor/Crystal Receiver ('phones extra)
(17) R.C. Super Transistor/Crystal Rec. (ditto)
(18) R.E.P. 1-valve Battery Receiver.
(1) CRY-BABY" ALARM (Baby Alarm)
(21) MULLARD 510 Amplifier (printed circuit) Ultra Linear Version

MULLARD 510 as above plus input selector and spare power supplies.
(22) "DE-LUXE " Printed Circuit Superhet
(23) "DE-LUXE" with New Look Cabinet
(24) JASON J.T.V. 2 Tuner
(25) RADIO JACK
(26) MULLARD TYPE "C" Tape pre-amp.
(27) JASON WII Wobbulator
(28) JASON Valve Voltmeter EM 10 ( 23 ranges)
(29) NEW JASON F.M. TUNER FMT2 with built-in power supplies and cabinet
(30) NEW JASON FRINGE F.M. TUNER FMT3, as above
(31) PULLIN Series 90 TEST METER
(32) R.C. Super Personal Portable I-valve (phone extra)
(33) R.C. Super Personal Portable 2 -valve (phone extra)
(34) R.C. TRANSETTE 2-Transistor Personal Portable
(35) JASON EVEREST 6-Transistor 2-wave Portable
(36) JASON EVEREST 7-Transistor 2-wave Portable
(37) CLYNE Cathode Ray Oscilloscope
(38) Compact Multi-range Test Merer
(39) CAR RADIO, Printed Circuit, 5 -valve Sher. NEW LOW
(40) JASON Audio Generator AG 10
(41) JASON Oscilloscope OG 10
(42) Super SHORT WAVE RADIO., I valve
(43) "WAVEMASTER " 7 -Transistor Luxury Portabl
(44) "GOLD STAR" De-Luxe I-valve Portable

Instruction Books whilch contaln full deacription, enay-to-follow practical wiring diagrams theoretical diagrams 1 -
fsed price lists, etc., are free of charge with all parcels but may be purchased separately as shown above

All required components at special.
Inclusive Inclusive
price

Instruction Book and itemised price list
available separately

PLEASE NOTE:-A selection of the above items are described more fully in this advertisement!!

PRINTED CIRCUIT CAR RADIO (for Home Construction). We are proud to be able to offer this New sype Car to be able to offer this New rype Car cuitry, special 12 volt valves and transistorised out-put stage. The highest degree of sensitivity is assured by the
 R.F. Stage. Covers Medium and Long Wavebands.

## - MORE

## CLYNE RADIO BARGAINS

## CABY UNIVERSAL TEST METERS

 These pocket-size multi-range test meters are of excellent quality and cover all the most useful ranges (A.C. Volts, D.C. Volts, resistance and current). Supplied complete with test prods, instruction book and batteries Model A. 10 ( 2,000 ohms per voltModel B. 20 ( 10,000 ohms per volt) E . $6 / 10 /=$.
Plus P. \& P. 3/6 on each
Fully detailed and illustrated leaflet available on request

## RECORD PLAYERS

Full range at usual competitive prices. Interesting H.P. facilities E.M.I. 4-SPEED STEREO/MONAURAL SINGLE RECORD UNIT. Complete with Stereo Head and Sapphire Styti. Brand New and Fully Guaranteed

## plus $3 / 6$ P \& \& P ARRIVED

LATEST GARRARD MODEL 210. Four-speed manual or automatic. 10 in . and 12 in , records of same speed can be mixed in any order, wired for stereo, attractive
white colour scheme. Price $10 \frac{1}{2}$ gns. white colour sc
LATEST B.S.R. UAI4. 4-speed. Attractive appearance. Wired for stereo. Fully guaranteed. $£ 7 / 19 / 6$. plus $3 / 6$ P. \& P.
B.S.R. UAB STEREO/MONAURAL. Few only at $£ 7 / 19 / 6$, plus 3/6 P. \& P. Brand new Guaranteed. COLLARO CONQUEST. speed. Wired for stereo. Brand new. $\& 7 / 19 / 6$, plus $3 / 6$.P. \& $P$ No. 38 AFV WALKIE-TALKIE. A wonderful offer. This famous trans-receiver unit, with relay operated $7.4-9 \mathrm{Mc} / \mathrm{s}$ band, range approx. 5 miles. Good condition. Opprox $22 / 6$ plus $2 / 6 \mathrm{P}$. \& P . per ONLY (less accessories). Quancity
unit export inquiries welcomed.
AERIAL TUNING UNIT ZA.0841. This well made ex-W.D unit contains a host of useful components including: I mA. 2 in . flush round M/C meter, I MA. Westinghouse full-wave meter rectifier, 5 -pole 5 -way heavy-duty silver plated wavechange switch. 3 in.
dia. silver plated rotary tuning in. dicator. 350 pF tuning condenser with insulated coupler and 3 tin calibrated dial ( $0-180$ deg) etc. etc. Contained in strong metal carrying case 9 in . $x$ 9in. $\times 8 \mathrm{in}$ with hinged lid. ONLY $27 / 6$ plus A CONSTRUCTOR'S MUST The latest "Pifco" Instrument The latest "Pifco" Instrument Bit Soldering Iron
With integral Stand and built-in Spot-light for illuminating work
$200 / 250 \mathrm{v}$. ON LY $22 / 6$. P. \& P. I/6. 200/250 Y. ONLY 22/6. P. \& P. I/6.
SOLDER. New boxed I Ib reels, 16 S.W.G. 50/50 at $8 / 6$ only, plus 1/- P. \& P
VIBRATOR PACKS. Limited quantity of both 6 volt and 12 volt types available. Output 300 volt. $100 \mathrm{~m} / \mathrm{a}$. Fully smoothed. Brand new ex-Gove. surplus. Price $35 / \mathrm{l}$ new ex-Govt. surplus. Price 35/-
ea., plus $2 / 6$ P. \& P. Please specify input voltage required.
12 VOLT VIBRATOR PACK. (Mallory). Output 150 v . @) 40 mA . Complete with synchronous vi-
brator. Brand new. ONLY $12 / 6$, brator. Brand n
plus $2 /$ P. \& P.
12 in . BAKERS SELHURST LOUDSPEAKER. 15 ohms, 15 watt $3 C-14,000$ cps. Brand new,
E4/10/-. P. \& P. 3/6.
I2in. RICHARD ALLAN P.M. LOUDSPEAKER. 3 ohm speech coil. Brand new. Only $32 / 6$ plus 2/6 P. \& P.
DEAF AID TYPE EARPIECES. Ardente Standard magnetic type complete with lead and plug. Only 12/6. P. \& P. 1/-.


SUPER MAGNETIC RECORDING TAPE SPECIAL!!! Famous American Ferrodynamics "BRAND FIVE" An enthusiast's " must." Brand new (NOT
SUB-STANDARD) High grade Acecate Base, 5 in . $600 \mathrm{ft} .16 / \mathrm{-}, 5 \mathrm{in}$. 900 ft . $18 / 6$, 5 z in. $1,200 \mathrm{ft} .23 / 6$, 7 in . 1,200ft. 25/-, 7in. 1,800fr. 35/-. Extra quality Mylar Dupont. 5in. 1,200ft. 37/6. 7in. 1,800ft. 44/-. 7in. 2,400ft. $60 /$. Each on plastic spool. P. Iree. Trade enquiries invited.

## EXTRA SPECIAL OFFER !

 A small three-valve PORTABLE RE: CORD PLAYER AMPLIFIER mounted on baffle $12 \times 7$ in, with High Flux $6 \frac{1}{2}$ in. Loudspeaker. Valve line-up ECC83, EL84, EZ80. Incorporates separate bass and treble cóntrols. Max. output 3 watts. Will match all types of high impedance pick-up. Ready to use, $65 / 12 / 6$. P. \& P, 3/6. NEW STYLE CABINET finished in cwo-tone Leatherette. Will accommodate above Amplifier and Baffle without modification, also most types of Ancillary Equipment. Overall size $18 \times 13 \frac{1}{2} \times 8$ in. Fitted with carrying handle, $63 / 9 / 6$ plus $5 /-$ P. \& P.NOTE. If both items puechased together they will be supplied at a special inclusive price $88 / 7 / 6$ plus $6 / 6 \mathrm{P}$. \& P.


## * TAPE RECORDER CONSTRUCTORS *

COLLARO TAPE TRANSCRIPTOR Mk. 4. $\begin{aligned} & \text { 3-speeds, fitted with }\end{aligned}$ digital counter. List price C25. Our Price £I7/10/-. P. \& P. $5 / \%$.
COLLARO TAPE PRE-AMPLIFIER AND BIAS OSCILLATOR. Complete with power pack for use with the above deck. 4 valve plus EM8I magic eye. $110-240 \mathrm{v}$. A.C. Input sensitivity: microphone socket $5 \mathrm{~m} / \mathrm{y}$., a uxillary socket $500 \mathrm{~m} / \mathrm{v}$. Speed equalisation switch gives com$5 \mathrm{~m} / \mathrm{y}$, auxilary socket $500 \mathrm{~m} / \mathrm{y}$. Speed equalisation switch gives com-
pensation at all 3 speeds. Full wiring instructions included. List price C2 pensation at all speeds. Full wiring instructio
Limited quantity only at LSI $19 / 0^{\circ}$ PA\& \& PRANSCRIPTOR. 3 motors,
3 speeds, $17,3 \frac{1}{3}, 7 \frac{1}{2}$ i.p.s. takes 7 in . spools. Push-button controls, $\mathrm{f} 12 / 19 / 6$ 3 speeds, 17, $\frac{3 \pi}{2}, 7$ i.p.s. takes
plus $5 /-P$. $\&$ P. Usual H.P. facilities.
LATEST B.S.R. "MONARDECK." Single speed Tape Deck. Takes 54 in spools- $3 \frac{3}{4}$ i.p.s. At only $£ 8 / 19 / 6$ plus $5 /-$ P. \& P.
TAPE RECORDER AMPLIFIER-MANUFACTURER'S SURPLUS: Suitable for use with either of the above Tape Decks, and mose other types. For A.C. mains, 4 watts ouspur. $40-12,000 \mathrm{CPS}$ at $7 \frac{1}{2}$ i.p.s. $\pm 3 \mathrm{db}$. Facilities for superimpose. Valves: ECL82, I2AX7, EM84, and contact cooled metal rectifier. Radiogram input, microphone input, volume control and separate treble and bass controls. Chassis measurement $\| \frac{1}{2} \times 3 \times 4 \frac{2}{5} \mathrm{in}$. Supplied complete with attractive grey/blue escutcheon plate finished in black and gold. Circuit diagram and eonnecting instructions included. Price $£ 10 / 10 /$ only, plus $3 / 6$ P, \& P. Limited quancicy. If purchased with either of the above decks, both items post

ATTRACTIVE TWO-TONE PORTABLE CARRYING CASE Suitable for above amplifier and Collaro, Studio deck. Limited quantity only at $72 / 6$ plus $3 / 6$ P. \& P
MIC 45-1 Acos latest flat pistol-grip crystal microphone. Attractive black and gold finish. OUR PRICE 29/6 plus I/-P. \& P
ACOS MIC 39-1. Crystal stick mierophone. List price 5 gns. Our price $39 / 6$ plus $1 / 6$ P. \& P.
MIC 40. General purpose erystal microphone with desk stand. Our price $25 /$ - only plus $1 / 6$ P. \& P.
M.C.24. Imported, crystal, attractive streamlined polished metal case, incorporates muting swith. List price 64/-. OUR PRICE 42/- only 1/-P \& P.
ANOTHER HAND MIKE BARGAIN! Lightweight crystal with built-in desk stand at only 19/6. P. \& P. 1/-


## LIMITED IMPORT QUOTA ONLYI OUTSTANDING BUY!

MODEL U. 1 MULTI-RANGE TESTMETER. Ideal for amateurs and service engineers. Incorporates 3 inch rectangular meter. Ranges: A.C. and D.C. voltage. $0-10-50-250-500$-1,000 v. D.C. current $0-100-$ $500 \mathrm{~m} / \mathrm{a} 0-1 \mathrm{~m} / \mathrm{a}$. (used at $0-10 \mathrm{v}$. range). Resistance $1-2,000$ ohms (centre 24 ohms). $100-200,000$ ohms (centre 2.4 k .). Sensitivity 1,000 ohms per volt A.C. and D.C. Size: $5 \mathrm{in} . \times 3$ in. $\times 2$ tin. Weight 22 ozs. only. Fully guaranteed. Supplied complete, and ready to $\begin{array}{llll}\begin{array}{l}\text { use with test leads, at the } \\ \text { very low price of only }\end{array} & 79 / 6 & \text { P. \& P P. } \\ 3 / 6 .\end{array}$ very low price of only
(ed)


BERNARDS Latest Manual No 167. "Eleven Tested Transistor Circuits, using Pre-Fabricated Circuit Units." Price $2 / 6$, post free Covers the use of the new items listed below TSL LP45F MINIATURE LOUDSPEAKER, diameter 1 lin., depth fin. High flux 9,500 guass, 150-15,000 eyeles, $25 /$, plus $1 / 6$ P. \& P.
TOROTOR TRANSISTORISED FM FRONT END ( $80-100 \mathrm{~m} / \mathrm{cs}$.$) )$
Completed with $2-0 \mathrm{Cl} 171 \mathrm{VH} F$ Completed with Transistors. Ready assembled, 79/6. Plus 26/9 P.T.
Suitable set of $10.7 \mathrm{me} / \mathrm{s}$. sub-miniacure IF TRANSISTOR Transformers
(2 IF-I RDT) $27 / 6$ per set, plus $1 / 6$ P. \& P.

SUB-MINIATURE DRIVER and Output Transistor Transformer 12/6 per pair, plus 9 d . P. \& P.

## MINIATURE

Twin Gangs 196/87 pF in. $x$ lin. $x$ in. 17/6, plus 9d. $P$ \& $P$ Set of Miniature Transistor IF and OSC Coils ( $455 \mathrm{kc} / \mathrm{s}$.). ( 3 IF and I OSC), $21 /$, plus $1 /-$ P. \& P.
3 WAVEBAND CONVERTER TA-12401. With OC170 transistor in emitter injected autodyne con verter circuit. S.W., 5.9-13 Mc/s. M.W., $510-1,620 \mathrm{ke} / \mathrm{s}$.; L.W., 150 $275 \mathrm{kc} / \mathrm{s}$. Wavechange and on-off by four pushbuttons. Overall size $3.4 \times 2.1 \times 1.8 \mathrm{in} .67 / 6$. Plus $22 / 9$ P.T.

## TRANSISTORS!!!

SURPLUS P.N.P.
RED SPOT (Audio/Experimental Application) WHITE SPOT
R.F. up to $2.5 \mathrm{M} / \mathrm{c}$ OCl69 VHF PNP TRANSISTOR. PNP JUNCTION
 Atcractive discounts for bulk pur chases. The above is a selection only Full range in stock by all leading manufacturers. Let us ha
FRUSTRATED EXPORT
repeatablel L. M. and SW. Not repeatable! L., M. and S.W. SUPERHET RECEIVER. Manufactured by McCarthy for export. At present for operation on 6 volts, but conversion details supplied free.


Valve line-up: $6 \mathrm{~K} 8 \mathrm{G}, 6 \mathrm{~K} 7 \mathrm{G}, 6 \mathrm{Q} 7 \mathrm{C}$ $6 F 6 G, 6 \times 5 G$ and 6 volt 4 -pin non synchronous vibrator. Bin. P.M Speaker, 4 watts output, P.U. socke Ext. L.S. socket, etc. Tone control Fitred in polished wood cabinet size $2 l \frac{1}{2} i n . \times 10 \frac{1}{i n}, \times 10 \frac{1}{4} \mathrm{in}$. These cabinets are slightly soiled owing to storage, but each is guaranteed unused, in serviceable condition, tested prior to despatch. Price [5/19/6 only plus P. \& P. 7/6, plus $27 / 6$ for A.C Mains Conversion Components required. OUTSTANDING ACOS GP73-2A: Turnover cartridge for Stereo and Monaural Standard and L.P. Few only at 29/6, also GP67-3 Mono at 18/.. Both plus 9d. P. \& P. LOUDSPEAKERS. EX. CHASSIS. As new guaranteed perfect, by leading manufacturers. 5 in. $9 / 6 ; 6 \frac{1}{5}$ in. $10 / 6$ former ( 5,000 ohms), $17 / 6$.
All 3 ohm speech coil, also 8 in . available, in attractive cloth covered cabinet, ideal for extension speaker, 22/6. Each item plus $1 / 6$ P. \& P. Complete list of new speakers on request.


AVOMETER MODEL D \&8.19.6 (P. \& P. 3/6)

| D.c. Volta | A.C. Volts | D.C. Current | A.c. Current |
| :---: | :---: | :---: | :---: |
| 105 mv . | 7.5 V | $10 \mathrm{~mm} / \mathrm{A}$. | $75 \mathrm{~m} / \mathrm{A}$. |
| 300 mv . | 15 V. | $30 \mathrm{~m} / \mathrm{A}$. | $150 \mathrm{~m} / \mathrm{S}$. |
| 1.5 V . | 75 V . | . $160 \mathrm{~m} / \mathrm{A}$. | $750 \mathrm{~mm} / \mathrm{A}$. |
| 3 V | 150 V. | $300 \mathrm{~m} / \mathrm{A}$. | 1.6 Ampe. |
| 15 \%. | 300 V . | 1.5 Amps. | 7.5 Ainps. |
| ${ }^{30} \mathrm{~V}$. | 600 \%. | 3 Ampa. | 15 Amps. |
| 150 | 750 V | 15 Amps. |  |
| 300 V . | 1.5 KV . | 30 smpe. | Reaistan |
| 1.6 KV . $0-10 \mathrm{~K}$ ohms |  |  |  |
|  |  |  |  |
| Thoroughly overbauled. Complete with instructions. An extremely robust meter at a |  |  |  |
|  |  |  |  |

## MOVING COIL PHONES. Finest

 quality Canadian with chamois ear-muffs and leather-covered headband. With lead and jack plug. Noise excluding and supremely comfortable. 19/6. Post 1/6. MATCHING TRANSFORMER (for Hi impedance), i.e. for HRO, CR100, ete. with standard jack plug, 4/6.SELENIUM BRIDGE RECTIFIERS. Funnel cooled. A.C. input 45 v. RMS. D.C., output 30 v .10 amps . BRAND NEW. Boxed. 45/. Post $3 / 6$.

## MARCONI IMPEDANCE, BRIDGE.

 Type Tf373. Measures, $L, C \& R$ at Type1.000 Cycles. Accuracy $1 \%$. $0-100 \mathrm{H}$; $0-100 \mu \mathrm{~F} ; 0-1 \mathrm{M} \Omega$ each in 5 ranges. Power Factor and "Q." First-class condition. E35. Carr. paid.
HALLICRAFTER VIBRAPACK. Input 6 v . output 300 v . at 170 mA . Designed for $5 \times 28$ or 527 . Size $6 \frac{1}{2} \times 7 \times 7$ in. BRAND NEW, BOXED. A real bargain at $29 / 6$. Carr. $3 / 6$.

## ADMIRALTY HT TRANSFORMERS

 Pri, $230 \vee .50 \mathrm{c} / \mathrm{s}$. Sees. $620-550-375$ - 0 $375-550-620 \quad \mathrm{v} . \quad(620$ and 550 v. 200 two 5 v. ${ }^{3}{ }^{3}$ amp. rectifier windings.
Total rating 278 VA. Upright meg. Wt. 25 lb . Made 1953. BRAND NEW. Original boxes. 45/-. Carr. 5/-.

INSTRUMENT TRANSFORMERS. 230 v . A.C. input. Outputs $0-65-130-195 \mathrm{v}$. $85 \mathrm{~m} / \mathrm{amps} ., 6.3$ v. $5 \mathrm{amps} ., 6.3$ v. 0.3 amps . Shrouded., Size. $34 \times 3 \frac{1}{2} \times 3 \frac{1 i n}{}$. high. $15 / \mathrm{m}$. Shrouded.
Post FREE.



## TRIPLETT METER

## MOVEMENT

This article consists of a basic 400 microamp meter movement mounted on a Bakelite panel $5 \frac{3}{2} \times 2 \frac{7}{8}$. The dial is scaled as a 15 range Testmeter. A circuit and parts list of the original instrument is supplied..
BRAND NEW. Boxed, 35/a, post paid.

## QQVO6-40 37/6

PV1-35 32/6, 2 D21 7/6, OC3 6/\%, PTIS 12/6, CV51(Y65) 5/-, 6F33 5/-. BRAND NEW in individual cartons. Bulk enquiries invited.

## RECEIVERS R-1155B

A first-class 10 -valve Communications receiver, covering $75 \mathrm{Kc} / \mathrm{s}$. $1018 \mathrm{Mc} / \mathrm{s}$. ( $16.2-4,000 \mathrm{~m}$.) in 5 bands. The large seale and superior dual ratio slow-motion drive make tuning easy and the R.F. stage and
reception. All the receivers we sell have been thoroughly overhauled, completely realigned and are in first-ciass overhauled, completely realign
working order. ONLY $£ 9 / 19 / 6$.
A.C. MAINS POWER PACK OUTPUT STAGE, In handsome black erackled steel cabinet to mateh the R-1155 Fitted with RCA 8in. speaker. Just PLUG IN and switch on Only the finest quality components are used and we guarantee OUR power packs for 6 months. ONLY $£ 6 / 10 /$-. Deduct 101- when purchasing receiver and power unit rogether Send S.A.E. for lurther details or $1 / 3$ for 10 -page illustrated booklet giving technical data and circuits etc. (FREE with each
receiver). Add $10 / 6$ carriage for receiver, 5/- for power unit.

## RCA AR-88 SPEAKERS

A high quality 3 ohm unit fitted into heavy gauge black crackled steel cabinet, size $10 \frac{1}{2} \times 11 \frac{1}{2} \times 6$ in. Fitted with rubber feet and 6ft. lead. Ideal for extension speaker. CR 100, etc. In original cartons. BRAND NEW. 45/-. Post $3 / 6$.

MINIATURE 373 IF STRIPS. For FM tuner described in "Practical Wireless." Complete with 3 of EF91, 2 of EF92 and I of EB91. A Iresh release enables us to offer these once again. BRANO NEW. Complete reprint of conversion instructions and circuit supplied Iree. 35/-. OR less valves, 12/6. Post, either, $2 / 6$.

HRO SENIOR RECEIVERS


Complete with ALL NINE general coverage plug-in coilsets for $50 \mathrm{Kc} / \mathrm{s}$. to $30 \mathrm{Mc} / \mathrm{s}$. Instruction booklet, and circuit, but less external power supply unit. Table models, as new condition, 21 GNS. Packing and carriage 22/- extra. Send S.A.E. for further details.
S.A.E. for further details.
HRO POWER PACKS.

HRO POWER $115 / 230$ v. A.C. malns input. Tested, and in good condition. 59/6. Post 3/6.

## LOUD-HAILER EQUIPMENT

IDEAL FOR CROWD CONTROL, FACTORIES, FETES, ETC. CONSISTS OF ${ }^{4}$ SPEAKER UNITS AND CONTROL ONTT. COMPLETE WITH M1OROPHONE, HEADPHONES AND SPARES, OPERATES FROM 22 VOLTS D.C. (OR 6 YOLTSA.C. WTH SLIGHTLY REDUCED ALL TESTED AND WORKING, BOT SLIGHTLY SOILED. A GENUINE BARGAIN. S4/19/6. CARRIAGE 26/6.
T.C.C. VISCONOL CONDENSERS. 8 mfd .800 v D.C. wkg. at 71 deg . C. CPI 52 V . Size $3 \times 1 \frac{7}{8} \times 5 \mathrm{in}$. high. BRAND NEW. Boxed. 8/6 each, post paid.
4 mfd .600 v . wikg. CP i30T. $4 / 6$ each, post paid.
MINIATURE RELAYS (ALL BRAND NEW and BOXED) G.E.C., sealed, wire ends, 670 2M2B H/D M1095......... 8/6 G.E.C., sealed, wire ends, 670〔ц, 2 H/D makes, M1099... 15/G.E.C., sealed, wire ends, $670 \Omega, 4$ clovers, platinum, M1092 $19 / 6$ G.E.C., sealed, wire ends, $5,000 \Omega, 2$ elovers, platinum, G.E.C., sealed, wire ends, 5,000s, 2 elovers, platinum, $17 / 6$ Siemens High speed, ik $+i K \Omega, 1$ clover...................... 1016

## GIANT COMPONENT PARCEL

Contains 1001 and 1 watt resistors, 50 Ri Stab resistors, wire wound resistors, carbon and W/W pots. 100 capacitors (mica. paper, Bprague, eta. All components are unused. GUARANTEED VALUE, 25/- plus $2 / 6$ pnet.

## CHARLES BRITAIN (Radio) LTD.

I upper saint martin's lane LONDON, W.C. 2

TEMple Bar 0545
Near Leicester Sq. Station,
(Opposite Thorn House)
Shop Hours: 9-6 p.m. (9-1 p.m. Thursdays). Open all day Saturday.


CRYSTAL CALIBRATOR No. 10 A crystal controlled heterodyne wavemeter covering $500 \mathrm{Kc} / \mathrm{s}$. to $10 \mathrm{Me} / \mathrm{s}$ (Harmonics up to $30 \mathrm{Mc} / \mathrm{s}$.) Requires 300 V 15 mA . and 12 V .0 .3 a . D.C., but can be easily modified for 120 V . and I. 4 V . working. Size $7 \times 7 \frac{1}{2} \times 4 i n$. Good condition, complete with valves, erystal, instruction manual and circuit. ONLY 59/6. Post $3 / 6$. This item available complete as above. BRAND NEW and with spare set of valves. $\mathrm{E} / \mathrm{/l0/-}$, post $3 / 6$.

CANADIAN CRYSTAL CALIBRATOR. Uses double crystal and multivibrator circuit to give "" pips " at I $\mathrm{Mc} / \mathrm{s} ., \mathrm{l}$
Modutor. He/s. and book supplied. $79 / 6$. Modulator. Handbook supplied.
Post 2/6.
ELECTROSTATIC METER. D a. $6 \frac{1}{2} \mathrm{in}$. reads 5-18.5 Kv. Manufactured 1953. Contained in wooden case $10 \times 10 \times 9 \mathrm{in}$. high. E9/19/6. Post paid.
SANGAMO WESTON ANALYSER E772. A useful multi-range meter. Thoroughly overhauled and in perfect working order. For full details see previous adverts 67/10/-. Carr. 4/6.
MARCON: TF987/L NOISE GENERATORS. Range $100 \mathrm{Kc} / \mathrm{s}$. to $200 \mathrm{Mc} / \mathrm{s}$. Determines noise factor of AM and FM receivers. Fully stabilised H.T. supply A.C. mains operation. Brand new and

HEAVY DUTY SLIDER RESISTORS. $1.25 \Omega 20$ A., $12 / 6$, post $3 / 6$. $1 \Omega 12$ A., $8 / 6$ PRECISION RESISTORS. I Megohm $1 \% 1$ watt wire wound. Ex-U.S.A. BRAND NEW. $10 / 6$ per dozen.

## D.C./A.C. CONVERTERS. Input 12 v .

 D.C. Output 230 v. $50 \mathrm{e} / \mathrm{s}$. A.C. at $2 \frac{1}{2} \mathrm{in}$. meter and slider resistor for voltage adjustment. In stout wooden carrying adjustment. wid. Perfect working order case with lid. Per24 v. Input $230 \mathrm{v} . \mathrm{A}, \mathrm{C} .50 \mathrm{c} / \mathrm{s}$. 100 watts
 outpue. ${ }^{\text {In }}$ grey metal
NEW. $92 / 6$. Carr. $7 / 6$.

RADIATION METERS. Portable doserate meter, containing modern type rectangular 50 micro-amp. meter, CVX494 electrometer valve, etc. BRAND NEW. In canvas carrying case. $£ 3 / 19 / 6$. Post $2 / 6$. For details of other equipment, see our previous adverts.

## MICROAMMETERS

R.C.A. $0-500$ microamps. 2 tin . circular flush panel mounting. Dials are engraved $0-15,0-600$ volts. As used in the American version of the No. 19 set. BRAND NEW. Boxed. $15 \%$.
American 0- 100 microamps, $2 \frac{1}{2}$ in. square flush panel mounting. BRAND NEW. Boxed. 42/6.

FERRANTI
VOLTMETERS N5.
0-300 volts, 25 $100 \mathrm{c} / \mathrm{s}$. Moving iron, 6 in . scale. iron, bin. scale.
Fl. mtg. Hermetically sealed. grade IN. Made NEW. Boxed. 79/6. Post 3/6.



## TAPE AMPLIFIERS and PREAMPLIFIERS presented from MULLARD DESIGNS




## MODEL HF/G2P-D

## THE IDEAL" LNK TO ADD PULL TAPE

 RECORDING FACILITIES TO HIOH QUALITY HOME INSTALLATIONS, Radiograms, ete. Comprises the HF/G2P Tape Pre-amplinier tited to the carrara Nupe into the tape input or piek-up sockets of existing ampliter or Radio Chassis.COMPLETE WORKING UNIT contalning \$in. apooi of Long Plasy Tape. $\quad \mathbf{~ P r i c e}$ 23.15.0
Hire Purahuse,Termas: Deposit, $84 / 15 /$ - and 12 monthly payments of $£ 1 / 14 / 10$.
Aternatively we offer-Complete Kit of Parts to build the HF/G2P Pre-amplitier with the TESTED $£ 20.15 .0$ Deposit. £4/3/- and 12 months at $£ 1 / 10 / 5$
The Preampliter Model HF/GRP is a vailable separately lor:
(a) Complete kit of parts
(b) Asssmbled .
89.10.0
\&11.5.0


Model HE/GLA-D UNIT (described opposite). A small robust recorder with outstanding performance. Truly portable, weigbs only 22 l bs . Twin Track operates on 3 itin./sec. speed. Price 329.15 .0 H.P. Term
$22 / 3 / 7$.

MULLARD TYPE "C

## TAPE-PREAMPLIFIER

## ERASE UNIT

The "Hi-Fy " link to add full tape
recording facillites to High Fidellity
home installatious. Incorporates
FERROXGUBE POT CORE PUSE-PULL OBCILLATOR and 3 -speed treble equalsatio by FERROXCUBE POT CORE INDUUTOR. FOR WEARITE - COLLARO Unit. - Si40 Dincludes separate Power Suppiy kit of parts .......... £14.0.0 or Assembled ........ £17.0.0 (Excludung Power Unit $£ 11 / 15 /-$ and E14/1 and 12 months a


## MODEL HF/TR3 TAPE AMPLIFIER

(Mnllard Type "A" desugn)
A very hish quality Amplifer incorporatlig 3-gueed treble equalisation, uring the TOR FOR COLLARO-TRUVOX-BRENELL WEARITE or MOTEKK Tape Deeks, has GILsEN Outpat Transformer. Includes separate Power supply Uuit


KIT OF PARTS
£12.15.0
or ASSEMBLED
£16.10.0

## FOR THE HOME CONSTRUCTOR SPECIAL "COMBINED ORDER" PRICES

(a) The COLLARO " STUDIO " TAPE DECK and our Mullard Type "C" PRE-AMPLIFIER and Power Unit assembled and tested
H.P. Terms: Deposit $£ 5 / 18 /-$ and 12 months at $£ 2 / 3 / 3$.
(b) As above but Type "C PRE-AMPLIFIER supplied as complete Kit of Parts
(c) The COLLARO Mk. IV TAPE DECK and the MULLARD Type "C" PRE-AMPLIFIER and Power Unit assembled and tested.
H.P. Deposit $£ 7$ and 12 months $£ 2 / 11 / 4$.
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(e) The TRUVOX Mk. VITAPE DECK and the assembled Type " C" PRE-AMPLIFIER and Power Unl H.P. Deposit $£ 8$ and 12 months $£ 2 / 18 / 8$.
(f) As above but the Type " C " supplied as complete

Kit of Parts
(g) The BRENELL MK. V Deck and the assembled Type H.P. Deposit $\in 9 / 4 /$ -
(h) As above, but the Type "C" supplied as complete

Kit of Parts
(i) The WEARITE $4 A$ DECK with TYpe " $C$ " assembled H.P. Deposit $£ 1 / / 4 /$ and 12 months $£ 4 / 2 / 11$.
(Carriage and Insurance on above quotes $10 \%$ extra)
THE ABOVE SUPPLIED IN PORTABLE CASE FOR E5/IO\% extra, THUS
FORMING A COMPLETE PORTABLE PRE-AMPLIFIER.
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£26.10.0
\&35.0.0
£32.0.0
£40.0.0
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£46.0.0
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£56.0.0
(a) COMPLETE KIT to build the HF/TR3 Amplifier
(b) together with the COLLARO ". STUDIO ". DECK
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(E) extra is we are to wire up Deck Switch Banks
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(f) As above but HFITR3 ASSEMBLED and TESTED H.P. Terms: Deposit $\in 7 / 18 /-12$ months of $£ 2 / 17 / 11$.
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£29.0.0
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(i) THE ASSEMBLED and TESTED HF/TR3 AMPLIFIER with the WEARITE MODEL 4A DECK, incorporates Wearite Head Lift Transformer, etc.
\&36.0.0
£39.10.0
£41.10.0
£45.0.0

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\text { H.P. TERMS: Deposit } £ 11 \text {, } 12 \text { months of } £ 4 / 0 / \mathrm{B} \text {. }
$$ (Carriage and Insurance on each above is $10 /$ extra)

Ateractive PORTABLE CASE is available to accommodate the TRUVOX or COLLARO TAPE DECKS and we offer it together with ROLA/ CELESTION $10 \times 6 \mathrm{~B}$. LOUDSPEAKER-ACOS CRYSTAL MICROPHONE -and $1,200 \mathrm{ft}$. SPOOL E.M.I. TAPE-ALL FOR.
£9.10.0

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COMPLETE KIT OF PARTS Designed by MULLARD-presented by MULLARD " 5-10" MAIN AMPLIF: ARR
For use with the MULLARD 2-stage pre-ampilfer with which an undtatorted power output of up to 10 wates is obtalned. We enpply RPFCIFIED COM power output of up to 10 watts is obtalned. We atpply 8PECIEIED COMTRANSFORMER and choice of the latest Ultra-liuear PARMEKO or the PARTRIDGE Output Transformer.
Price: COMPLETE KIT (Parmeko O/fut Trans.)
£10.0.0
Alternatvely we supply ABSEMBLED AND TESTED $\quad £ 11.10 .0$ AROVE INCOAPORATING PARTRIDGE OUTPUT TRANGFORMER E $1 / 6 /$ ext MULLARD'S 2-VALVE

## PRE-AMPLIFIER TONE CONTROL UNIT

amploying two whso valves and designea to operate with the Mullard MAIN AMPLIFIBR, but almo perfectly suitable for ot her makes.
pqua stictiy to Melerg in incorating - Equalisation for the latest R.I.A.A. characteristies.


Input for Crystal Pick -upa and val iable reluctance magnetle types.
Input (a) Direct from High Imp. Tape Head. (b) From t Tape Amptither or Pre-Amplifier
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£6.6.0
assembled and tegted
COMPLETE MULLARD 5-10 AMPLIFIER

## The popular and very successful complete " $5 \cdot 10$ " providing up to 10 watta high quasity revruduction.

specified componenta and new MULLARD VALVES are suppled facluding PARMEKO MAINS TRANGYORMERS and choice of the latert PARMEKO Price: COLDGE UlJTRA Linest Ontput Transformers. Hte COMPLETE KIT. Parmeko Transfrmer....... Aiternatively we supply AB8EMBLED AND TESTED. 213.10.0 HIre Purchase (Assembled Amp. only). Deporit e2/14/-, 12 months at 10/10 ABOVE incorporating PARTRIDGK OUTPUT TRANSFORMER E1/6/-exta


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MULLARD FOUR CHANNEL MIXING UNIT


COMPLETE MULLARD 3-3 A VERY HIGH QUALITY AMPLIFIER DEVEL OPED FROM THE VERY POPULAR 3-VALVE
3-WATT AMPLIFIER DESIONED IN TEE AMPLIFIER DESIGNED
Price for COMPLETE KIT Price for COM
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(Plus 6/6 carriage snd line £7.10.0 (Plus $6 / 6$ carriage and insurance). Aleratively suppled A88EMBLKD AND
FULLY TEBTED (Plus $/ 6$ FULLY TEBTED (Plus
carriage and insurance). £8.19.6 H.P. TERMS: Deposit $\mathcal{E} 2$ and 8 monthly payments of $£ 1$.

Our kit is eomplete to the MULT,ARD specification including supply of specifled cominputs for 78 and L. P records.plus a Radio position. Extra power to drive a Radio Tuniu Unit is riso available.

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Comprises two MULLARU 3-3 Main Amplifiers on one chissis. Oyerates with MULLARD
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or Assembled.
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Mk. II "Fidelity" FM TUNING UNIT An atractively presented Unit incorporating MULLARD PERMEABLLLTY TUNING Mullard Amplifiers.
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SPECIAL CASH ONLY OFFER!!
This very attractive PORTABLE AMPLEFER CASE toget her with a good quality GRAM AMPLIFIER and a
matched P. M. BPEAKER. ALL FOR ONLY 88.7.6 (phasists of the 3 modern BVA valves gud has separate BABS and TREBLL FoNTHOTAS The Portable Case $11 \Gamma$ also acenumodate almost any make ofatochanger and is attractively BUPPLY gEPARATELY:-
a) The 2 -atage (plus Rectifer) AMPLIFIER (b) The PORTABLE CARRYING CASE (c) 64 in . P.M. SPEAKER. $\qquad$

"Hi-Fi" LCUDSPEAKERS WE HAVE IN STOCK A GOODMANS - WHARFEDALE-W.B. ILLUSTRATED AND PRICED LEAFLETS ON REQUEST
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## COMPLETE STEREO AMPLIFIER

Meetg the many requests for a how priceet but good quality steremphonic Ampllifer. Out-

KIT OF PARTS........... OC ASSEMBLED.... $\mathbf{2 1 0 . 1 0 . 0}$

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Thia model theorporates two 2-valve Pie-Amplifera (described atove) comblned into Blagle Unit
enabing ft to be ueed for both ETEREOYHONIC enabling it to be used for both sTerfor and MONAURAL oneration. It is designied primarlly and MONAU UAL, operation. It is degigned primarif
to operate with our range of MULDARD MANN to operate uit olir range of
AMPLFIER8 but will uleo operate equally well
with any make of with any make of Amplifers lequrmg at mput

Price: COMPLETE $£ 12.10 .0$
Atematively A8sembled $£ 15.0 .0$ H.P. Tems on anzembled unit: 53 veronte nad 12 monthe of $81 / 2 /$.
! ! RECORD PLAYERS !
The LATEST MODELS are in Slock. Many at REDUCED PRICES ! !
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The colLaro 4 -speed single Record Player, incorpmating ine studio THE NEW P. B.S.R. Model UAi2 is in tock. A 4-"SPEED" MXER AUTOCHANUER
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TT HOME CONSTRUCTORS ! ! A RANGE OF "EASY TO ASSEMBLE" PREFABRICATED CABINETS Dexigued by the W.B. "BTENTOHIAN"LUMPANY for "Hi-Fi" Loudppeaker aymtems

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##  F. \& S. PROOPS send SINCERE XMAS GREETINGS to the hundreds of customers they never meet-and promise them that Proops advertising will stay reliable and Proops guarantees will hold good throughout 1961. <br>  <br> RECEIVER TYPE 88 (R1475) <br> Highly stable, specially accurately calibrated, Marconi design, RAF

 communications receiver covering $2-20 \mathrm{Mc} / \mathrm{s}$ in 4 bands with built-in 600 kcs Xtal reference oscillator for checking dial which can be reset by special panel trimmer control. 11 Valves: $3 \times 6 \mathrm{~K} 7,6 \mathrm{~K} 8$, 6J5, 3x6Q7, 6H6, Y63 tuning indicator and VR150-30 voltage regulator. Two stage IF with 8 tuned circuits, Xtal controlled B.F.O. Four position selectivity with audio filters for narrow bandwidth C.W. Fast and slow AVC, high and low noise suppression, A plug-in unit with additional mixer provides a "listening through " guard channel of either $2-4$ or $4-7$, Mc/s. Receiver 1 vil AC (or 12 v . Power pack 8 . $\times 9 \times 11 \mathrm{in}$. Complete with 200-250 voit AC (or 12 v . Used, but in very good condition. Guaranteed serviceable. A sound buy indeed at 3.10 .0 carriage paid.

## POST FREE SNIPS

Double pole knife changeover switch on porcelain base. 2 for $5 /$ G.P.O. 230 volt mains, twin six inch gong, ourdoor bells 33/6 Siemens high-speed relays. $1,000=0-1,000 \mathrm{ohm}$ coils. Pyrex Aerial Insulators. Four 3 in. OR two 8 in. Uyrex Aerial $8 / 6$
$7 / 6$
$5 / 6$ U.S.A./British co-ax adaptors. Four for $7 / 6$万, Six 80 volt for $7 / 6$
$7 / 6$

This is the attractive lightweight, American Radio Altimeter that superseded the British version. A complete 14 valve radar set covering $420-460 \mathrm{Mc} / \mathrm{s}$ it is ideal for conversion to radio control of models or 70 cm . work. It embodies three selfcontained sub-units in separate detachable aluminium cases, as follows:

TRANSMITTER/RECEIVER


TRANSMITTER

A push-pull feed-back oscillator tuneable either side of $445 \mathrm{Mc} / \mathrm{s}$ frequency modulated at $100 \mathrm{c} / \mathrm{s}$ by a particularly robust moving coil transducer. Two 955 high frequency acron valves. Case size only $3 \frac{1}{2} \times 6 \frac{3}{2} \times 2 \mathrm{in}$. plus $2 \times 2 \frac{2}{2} \mathrm{in}$. dia. for transducer.
RECEIVER
Tuneable to transmitter frequency. Size $3 \frac{1}{2} \times 6 \frac{1}{2} \times 2 i n$. Two 9004 acorn valves.

## AUDIO AMPLIFIER

Self-contained RC coupled 12SH7, 12SH7 and 12SJ7. Size $3 \times 5 \times$ $1 \frac{\mathrm{k}}{\mathrm{in}}$. Amplifies the received signal which is passed to detector circuit giving a D.C. voltage proportional to the difference between the transmitted and received (reflected) signal to operdte internal relays which pass appropriate correction signals to autopilot and supply external indicator ( 5 mA meter).

## MAIN CHASSIS

The main chassis carries the 3 sub-units and has a further three 12 SH7 one 12SJ7, two 12 H 6 and one VR150 regulator, three $1 \%$ wire-wound resistors, one 4 -pole changeover relay, two SPCO relays, three twinganged pre-set potentiometers, trimmers, fuses, etc.
Power supply is derived from a 27 -volt dynamotor (charging rate for 24 v supply) delivering 285 volts at 75 mA .
BRAND NEW, a very useful buy indeed at only $\{2$ plus $7 / 6$ carriage.
MEGiSTORS $125,1,000$ or 10,000 MEGohms
Glass encapsulated $10 \%$ tolerance high value resistors for minute grid current applications. Ideal for extending the range of sensitive meters or using in probes to provide a really high impedance input lor VTVM's or 'Scopes. One of each value plus any chosen two, the 5 for $10 /-$ post free by return.

## HIGH QUALITY POWER PACK

Admiralry Rectifier Unit Design 95, totally enclosed in heavy gauge attractive light grey case size $11 \frac{1}{4}$ in. high $\times 6$ in. wide $\times 14 \mathrm{in}$. deep. Admiralty ratings: transformer $400-0-400$ at $50 \mathrm{~mA}, 6.3 \mathrm{v}$ at 1 Amp , 5 v at 3 Amp for 5 U 4 G . Insulation tested to 3 kV . Two 350 ohm 20 henry 80 mA chokes; Two $4 \mu \mathrm{~F}$ at 600 v . ceramic terminal square canned paper smoothing capacitors. Double pole mains switch, two 2A fuses and two spares all in screw-in holders on front panel. 3 pin $250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$ mains input, and 3 pin output with matching plug on short screened cable providing 650 volt D.C. and 6.3 AC with common earth. An unusuall". neat attractive, high quality unit, Brand New, still boxed
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carriage paid.

## Twin tube CRT indicator

Attractive, lightweight, black crackle box $11 \times 7 \times 13 \frac{1}{2}$ in. deep with $4 \times 2 \frac{1}{4}$. and $3 \frac{1}{2} \times 3 \mathrm{in}$. square windows on front panel for twin 5FP7 tubes. Neat arrangement of appropriate (independent) controls and variable scale illumination. Totaily enclosed detachable magnetic focussing colls. All connections to rear sockets. Ideal TV monitoring unit as used by many amateurs. Used, but in very good condition, tubes guaranteed $0 . K . ~ £ 1 / 10 / 0$ carriage paid.


ANTENNA INDICATOR
Remote indication to within $1^{\circ}$ on precision instrument type flush fitting black crackle indicator with 3 in . dial calibrated in $2^{\circ}$ steps plus the four cardinals. Simple D.C. wiring ( $6-30$ volt) from specially wound potentiometer in sealed diepotentiometer in sealed die-
cast housing with drilled spindle transmits accurate signal of horizontal or vertical bearing Brand New, Post Free,

SCR. 522 RECEIVER \& TRANSMITTER CHASSIS
Receiver chassis 10/-. Transmitter chassis (with modulation trans former) 15/-. Post paid

## 200 AMP D.C. GENERATORS

These relatively small but really heavy duty genera tors were designed for a continuous output of 200 amps at 29 volts and are very successfully employed as a portable welding plant when driven from at ractor take-off pulley or separate engine as required. Guaranteed fully serviceable.
ony $£ 6$ . 15 carriage paid.
 LOW INERTIA INTEGRATNG MOTORS Linear Voltage/Speed Curve Precision servo system tool comprising Electro Methods high efficiency 24 volt D.C. motor unit mounted to robust aluminium face plate. Light weight fibre and brass reduction gear to double ball bearing spindle terminating in miniature bevel gear. Adjustment for varying speed. Jwelled bearings. Gold brushes. Brand \& New, post paid.

## PRECISION SIGNAL GENERATOR CT53. A modern laboratory standard instrument still in current use



## FEATURES

- Vernier tuned, Triple screened, 6-Band coll turret covering 8.9 to $300 \mathrm{Mc} / \mathrm{s}$ with 72 ohm output from 100 mV down to $1 \mu \mathrm{~V}$.
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- CW or modulated $30 \%$ by $1,000 \mathrm{c} / \mathrm{s}$ Sine or Square wave (variable mark $/ \mathrm{space}$ ratio)
- External mod. by sine wave from $50 \mathrm{c} / \mathrm{s}$ to $10 \mathrm{kc} / \mathrm{s}$. or pulses down to ${ }^{3} \mu \mathrm{Sec}$.
- Seven B7G Valves, Potted "C" core transformers, Paper capacitors, Stabilised HT
- Selected spare oscillator, pre-aged spare monitor, $100 \mu \mathrm{~A}$ meter.
- Mains, HT, Bias and Filament supplies fully RF filtered.
- Combined cabinet/rack mounting case, Pressure sealed, Desicator, Panel Mains voltage adjustment, Triple fused, in fact, "the lot"
Offered straight from Service use, complete with calibration book, cables, circuit diagram and principal technical information, checked serviceable and fully guaranteed
Plus 15/- for careful packing and carriage
E17.10.0


## I-30A SIGNAL GENERATOR $100-156 \mathrm{Mc} / \mathrm{s}$

Modern, portable, battery operated, 5 valve Signal Generator with alternative crystal or master oscillator, either optionally modulated by $1,000 \mathrm{c} / \mathrm{s}$ Hartley oscillator. Large directly calibrated dial with precision slow motion drive. Five step and variable attentuator. Supplied with matching black crackle carrying case for 6 and 135 volt batteries with 10ft. supply cable, and metal cased 1 mA . test meter for checking crystal resonance, etc. Brand new. £2/17/6 plus 7/6 packing and carriage.
I-95-A FIELD STRENGTH METER $100-156 \mathrm{Mc} / \mathrm{s}$
Self-contained, tunable-input, valve-voltmeter with telescopic aerial and battery-fed diode rectifier and pentode amplifier for measuring field strength, presence of modulation, and approximate frequency of transmitter. Compensating circuit for state of $1 \frac{1}{2}$ and 45 volt batteries. In attractive black crackle case. Brand new.
£2/5/-plus 5/-packing and carriage.

## D.C. GYRO \& SERVO MOTOR

Beautifully engineered Minneapolis-Honeywell precision gyro, totally enclosed in sealed light-alloy housing about $8 \frac{1}{2} \frac{1}{2}$ in. cube. Automatic erection and precession correction. Large diameter Dessyn type transmitting potentiometers provide signals corresponding to the magnitude of the deviation of gimbal arms. Powerful D.C. motor coupled through a differential reduction gear to a 4 in . spur driving gear integral with a 3 in. dia. spiral groove cable driving drum. Two powerful solenoid clutches and corresponding brakes hold drum rigidly in position or clutches and corresponding brakes hold drum rigidy in pusition or set free for "neutral." Nom 12 volts. Size $10 \times 6 \times 8 \mathrm{in}$.
$\Sigma 10$ each unit or $£ 17 / 10 /=$ pair, carriage paid

## £IO GEIGER COUNTER

Circuit embodes U.K.A.E.A. patent. Specially moulded case. Currently being supplied throughout the world. Three ranges-highly sensitive-light-portable-visual and audible response-plus output socket. Ideal for introduction to radiation measurement and nucleonic circuitry. Specially written 40 -page instruction manual supplied. Batteries $£ 2 / 15 / 3$ extra.

## KIT OF PARTS E4/I7/6

Idéntical parts. Guaranteed performance* Manual and printed circuit plates for battery pack supplied (assembled pack £2/15/3 extra). Fully illustrated assembly instructions. Spares and service permanently available.

## BC. 929 SCOPE UNIT

Neat, modern indicator unit especially suitable for quick conversion to attractive general servicing scope: (Suitable circuit diagram and all component values supplied.) Contains fully mumetal screened 3BP1 tube, intensity and focus controls, 3 -position rotary switch and 8 pre-set, potentiometers, plus $2 \times 6 \mathrm{SH} 7,2 \times 6 \mathrm{H} 6,6 \mathrm{G} 6,6 \mathrm{X} 5$ and 2 X 2 valves. Designed for 24 v. D.C. or $400 \mathrm{c} / \mathrm{s}$ A.C. input. Size $14 \times 8 \frac{1}{2} \mathrm{in}$. square. Well known and deservedly popular buy. Offered new, less (unwanted) motor driven aerial switching unit, for post paid.

50/-

200/220v D.C. to $200 / 250 \mathrm{v}$. A.C. $50 \mathrm{c} / \mathrm{s}$.
New 200 -watt D.C. to A.C. rotary inverters in sound-proof cabinet. $£ 9 / 10 /=$ carriage paid.

ROTARY RELAY. Superb, fast acting, brand new precision unit made by Price for RCA. Nominally 12 volt, but mighty lively on 6 -volt supply. Two heavy duty single pole changeover contacts and one low current for external circuits, plus one break set that extends coil winding to seduce initial energising current to 50 mA . (at 6 v .) for holding. Solid milled armature, laminated steel frame, $2 \frac{1}{4} \times 2 \frac{1}{2}$ in. $\frac{3}{1}$ in. thick, moulded inset dielectric block.
A highly recommended spares box buy at $7 / 6$ each, posi free.

## CHROMEL/ALUMEL THERMOCOUPLE LEADS

7, 10 and 21 foot lengths of flat twin lead- $90 \%$ nickel, $10 \%$ chrome and $95 \%$ nickel, $5 \%$ aluminium. Total resistance 0.875 ohms. For use with millivoltmeter to read 0 to $1,000^{\circ} \mathrm{C}$. Sheathed in heat resisting silicon rubber that will stand $200^{\circ} \mathrm{C}$. Price, post paid, $7 / 6,10 / 6$ and 21/-respectively.

## VARIABLE SPEED HYDRAULIC GEARBOX

This specially made oil-filled caseing houses a hydraulic torque conversion unit originally precision made by Westinghouse from high quality materials for the U.S. Government at an acquisition cost exceeding $£ 150$ each. Highly suitable for lathe head drive, workshop variable speed power take-off, etc.
Basically the unit is a back-to-back mounted, oil submerged, variable displacement hydraulic pump (input shaft) feeding a reversible hydraulic motor (ourput shaft) so that variation of the pump displacement by manual control gives very fine selection of output speed from zeio up to $6 \%$ below input speed while a changeover valve in the supply lines to the motor provides instantaneous reverse at any speed. Recommended intput speed $500-1,000$ r.p.m., maximum power $1 \frac{1}{2}$ h.p. Both shafts in. dia. with Woodruff key.
Tested and fully guaranteed, supplied complete with technical data and performance curves for the remarkable price of $£ 16$ only, carriage paid.

## NOMOTRON DECADE COUNTER TUBES

STC Type G10/241 latest type cold cathode, gas-filled, single pulse, uni-directional decade counter which illuminates numerals on tube face. Operating range- $20 \mathrm{kc} / \mathrm{s}$. Cathode output 40 volts, 3.7 mA . HT supply 310 v . plus. Applications include: tachometers, counting and batching, frequency and time measurement, direct operation of electro-magnetic relays, sequential monitoring of up to 10 different waveforms, etc. Brand New, complete $32 / 6$
with special base and instructions.


#### Abstract

post paid.


## EVERETT EDGCUMBE SYNCLOCKS

Grade 1 industrial process timer with 3 inch dial covering ${ }^{3}$ to 10 minutes in one tenth divisions. Driven by a 16 volt $50 \mathrm{c} / \mathrm{s}$ synchronous clock. A metal rectifier provides D.C. to pull in a relay that engages the drive until the time set on the dial elapses. "As it reaches zero heavy duty contacts snap shut to close the external circuit (at the same time other contacts break to arrest the clock). Swisching off the power trips the relay and the spring loaded dial returns to the time set ready for the next cycle. Whole totally enclosed in a heavy cast wall mounting case, stove enamelled black, size $6 \frac{1}{2} \times 7 \times 4$ inches. Brand new in original packings $55 / \mathrm{m}$ post paid.

## VENNER TIME-SWITCHES

Type T.S.2, first grade precision time switches as supplied to G.P.O. Comprises absolutely silent self starting 250 volt $50 \mathrm{c} / \mathrm{s}$ synchronous clock, mechanism totally enclosed in heavy gauge brass case. Central drive takes detachable dial that revolves to operate sensitive on and off trips for external mains operated circuit.. Self contained clock is easily detachable from rear mounting panel (self starting down to 80 v . and keeps running down to 15 v.).
Brand new, in original packings, and with dial and adjustable stops,
post paid.
$37 / 6$


BRAND NEW VARIABLE VOLTAGE TRANSFORMER． 230 volt A．C．input． Fitted in steel hammer finish case complete with 0－300 volt M．C．A．C．Meter，fuse and neon Indicator light．Output constantly variable from 0－270 volt A．C．Type 1． 2.2 amp．Price £ $8 / 10 /$－carriage $10 /$ ．Type 2． 5 amp ．Price女12．carriage 101 m ．

> BRAND NEW VARIABLE VOLTAGE TRANSFORMER. For 230 voll A.C. input. In cases exactly as above with meter, fuse and indicator light. Output constantly variable from 0.230 volt A.C. Type 15 . 15 amp. Price $£ 22 / 10 /-$ Carr. $15 /$.-.

W．W．RHEOSTAT．New． 3.5 K or 5 K .25 watts．Price 7／6．P．\＆P． $1 / 6$ ．

NEW WIRE WOUND RHEOSTAT ON CERAMIC． 58 ohm． 50 watt，complere with instrument knob．Price $8 / 6$ ．P．\＆P． $1 / 6$ ．

EXP．O．MAGNETIC COUNTER． 500 ohm type for 24 volt also 3 ohm type for 6 volt
D．C．operation．Price $6 / 6$ each．P．\＆P．I／－．

AUTO TRANSFORMERS．Step up，step down．110－200－220－240 $\mathrm{y}_{0}$ Fully shrouded． New． 300 watt type $£ 2 / 2 /-$ each．P．\＆P． $2 / 6$ 500 watt type $€ 3 / 3 /-$ each．P．\＆P． $3 / 9$ ．1，000 watt type $44 / 4 /-$ each．P．\＆P． $6 / 6$ ．
HEAVÝ DUTY L．T．TRANSFORMER． Very conservatively rated for continuous duty． New．In manufacturer＇s cases．Input $110-260$ volt multi－tapped． 50 cycles，single phase． Output 28－29－30－31 volts at 21 ampere．Price £6／15／－，carriage 10／－

NEW GALVA． NOMETERS， Solid brass，3in．dial， in polished wooden scale， 35 mA either side． 100 ohm coil． Price $12 / 6$ each．P．\＆ P． $1 / 6$ ．

EX R．A．F．AIR POSITION IN－ DICATOR，con－ plate infinitely varia－

ble resolving gears，miniature spur bevel and worm gear drives，also toggle，push button and rotary switehes，repeater motor， 4 mechanical counters，miniature lamp holders and lamps etc．As new．Illustration below．Price 22／6． P．\＆P．3／6．



EVERSHED AND VIGNOLES＂WEE MEGGER．＂ 500 volt in brand new leather case．Guaranteed perfect．Price $£ 13 / 15 /$－ P．\＆P．2／6．
A．R．B．U．S．A．RECEIVERS． 24 volt．Covering 195－9，050 k／c．in 4 bands．As new，suitable for use on boats，etc．Price $£ 6$ including car－ riage．
HIGH SPEED RELAY．Siemens．Two bob－ bins 1,000 ohms each．New，10／6 each．P．\＆ P．I／－．
18－WAY P．V．C．COVERED 14／36 WIRE， screened overall，covered with P．V．C．，all colour coded， $3 / 6$ per ${ }^{\circ} \mathrm{yd}$ ．；$£ 15$ reel of 100 yds ． Carriage paid．

## BRAND NEW FREQUENCY

 METERS manufactured by Nalder \＆Thompson Ltd． Calibrated 45 cycles to 55 cycles per second．6in．dial． Panel mounting type．In original manufacturer＇s boxes． PRICE $£ 10 / 15 /-$ ea．Postage $3 / 6$ ．

20 WAr strite containing standard Post Office telephone Jack Sockers，overall size $11 x$ $3 \frac{1}{2} \times \frac{1}{2}$ in．New．Price $15 /-$ each．P．\＆P． $1 / 6$ ． 10 WAY STRIP standard Post Office tele－ phone Jack Sockets，spacing allowing Igranic Jack Plugs．New．Price $10 /$ ．P，\＆P． $1 / 6$ ．
19－INCH RACK MOUNTING 20－WAY P．O．JACK STRIPS with 40 terminals at rear． Price 25／－．P．\＆P．3／6．
19－INCH RACK MOUNTING 20－WAY P．O．LAMP STRIPS．Price $25 /$－．P．\＆P． $2 / 6$ ． LATEST MOST MODERN TYPE OF EX W．D．MINIATURE HEADPHONES．
As illustrated．Brand As illustrated．Brand Price： $10 / 6$ plus P．\＆ Price：
NEW MOVING
COILHEAD． SETS．Complete with Tannoy carbon

hand microphone，with plug suitable for No． 19 set．Price： $12 / 6$ each，plus P．\＆P． $2 /$ ．
12 v．D．C．AMP－ LIFIER，as new， for operation on 12 v ．car battery， 10 v．car battery， torted output， torted output，
with 6 L 6 valves in with 6L6 valves in
push－pull．Mike／
 Gram．input，tap－ $15,100,250$ or 500 ohms． ped output $7 \frac{1}{2}, 15,62,1$
$\varepsilon 9 / 17 / 6$ each．Carr． $15 /-$ ．
PYE LEVER OPERATING MICRO SWITCHES．Single pole change over．Brand SWITCHES．Single pole change ove
new．4／－each or $42 /$ dozen，$p$ ．paid．

## VARIABLE YOL．

 TAGE TRANS： FORMER． ＇BERCO．＂Brand new in manufactur－ 110 volt A．C．Input． Constandy ．input． constancly variable 22 mp volts． 2.2 amp．type．$\ell 4$ ． P．\＆P．3／0\％． 5 amp．type $6 / 10 \%$ ．$\&$ P． 3／6．


ELLIOTT SWITCHBOARD MOUNTING PEN RECORDER． ${ }_{2} \not{ }_{4} \mathrm{in}$ ．chart． 1 mA ． movement． 2 speed mechanism．Com－ plete with pen，and charts．Recondi－ tioned as new and guaranteed．Limited quantity．Price $\mathrm{E55}$ ， carriage 10／－．

PLATE TRANS－ FORMER of very FORMER of very brand new，origi－ nal manufacturer＇s cases．Input tapped at 190／210／230／250 v．Output 2250－0． 2250，centre tap－ ped 400 mA ．Nett weight 761 b ．，size 13 in ．$\times 9 \mathrm{in}$ ．$\times 6 \frac{1}{2} \mathrm{in}$ ． Price $66 / 10 /$ each， plus carr．10／－．


NEW UNCHARGED UNFILLED 12 VOLT ACCUMULATOR 9 ampere in unspillable plastic ases．Comprises $6 \times 2 \mathrm{v}$ ．separate cells connected by terminal strips． 6 $\times 5 \frac{1}{2} \times 4 \frac{1}{2}$ ．over terminals．Price 19／－，plus P．\＆P． 2／9．


245 AMP． 2 VOLT ACCUMULATOR． Admiralty type in wooden casing．Size $15 \times$ $7 \frac{1}{2} \times 7 \frac{1}{2}$ in．Weight 601b．Unfilled，uncharged． New．Price £4．Carriage 10／－．


MiniAture P．M．MOTOR． 12／24 volt，re－ versible．I⿳亠丷厂彡⿱丆贝⿴囗⿱一一犬in dia．New．Price $10 / 6$ each．P．\＆ P． $1 /$－

AIRCRAFT CINE CAMERA G45B Mk．III．
 Fully modified，fitted with $\mathrm{f} / 3.5$ triple an－ astigmatic lens，takes 25 ft ．of 16 mm ．film fitted with 24 motor． 16 exposures per sec．Brand new，original packing，$£ 4 / 10 /$ each．P．\＆P．paid．


SOLENOID OPERATED MAGNETIC RELAY. Type 5CW/3945, 4 pole changeover, 10 A contacts 24 v . operation. Brand new 13/6. P. \& P. 1/6.

CARPENTER'STYPE POLARISED RELAYS. $2 \times$ 9,500 turns at 1,685 ohms. Price 22/6 each. P. \& P. $1 /=$ Carpenter's similar to above, but type 5A48. Coils $\mid \times 3200$ turns at 100 ohms and $1 \times 2000$ turns at 145 ohms, $22 / 6$ each. P. \& P. $1 /-$ - Bases for same $2 / 6$.


SIEMENS H.S. RELAY. Very latest type, sealed. H96E. I,700 ohms plus 1,700 ohms, single C.O. contacts. Brand new with fixing clip. In maker's cartons.
Price $16 / 6$ each, plus I/- P. \& P.
Siemens sealed similar relay to above, but 2.2 ohms plus 2.2 ohms. Minus clips, $12 / 6$ each. Plus I/- P. \& P.

SUPERIOR BRAND NEW RELAY. 7,000 ohms coil. Will pull in at 750 microamp, and out at 450 microamp. Change-over, platinum contacts. Vacuum sealed, will therefore not be affected by oil, moisture or water and never needs adjusting. Weight $2 \frac{1}{2}$ oz. Price $18 / 6$. P. \& P. I/-.

MINIATURE MOVING COIL DIFFERENTIAL RELAY. Two coils 350 ohms each
 Operating current minimum 140 microamp., nominal 400 8 milliamp. One pole two way, or, centre stable. Two way contact current 100 mA at 50 V . A.C. or D.C. Size If $x$ 查 $\times \frac{3}{4}$ in. Price $22 / 6$ each.
G.E.C. SEALED RELAY. Type M. 1090. 180 ohms coil. $6 / 12$ vole. $4 \mathrm{C} / \mathrm{O}$. Brand new. 18/-. P. \& P. I/-.
G.E.C. SEALED RELAY.

Type M. 1092. 670 ohms coil. $12 / 24$ volt. 4 ClO . Ex new equipment. Unused. $10 /-$. P. \& P. $1 /-$.
G.P.O. 600 TYPE RELAY. 400 ohms coil. 24 volt. 2 C/O plus 2 M. New 7/6. P. \& P. I/-. MINIATURE OPEN TYPE RELAY. 700 ohms coil. 24 volt. $2 \mathrm{C} / \mathrm{O}$. Ex new equipment. Unused. 7/6. P. \& P. I/-
ROTARY RELAY. 12 vole. Heavy duty change-over contacts and one low current for external circuit, plus one break set. Price $7 / 6$. P. \& P. $1 / 6$.


MINIATURE UNISELECTOR SWITCH. Two banks of ten plus home contacts one bank continuous of normal. 30 ohm coil for 24 volt operation. Brand new, manufacturer's packing. Price 22/6 each. P. \& P. 2/6. As illustrated.


PRECISION MADE GEARED MOTOR BY MRAYTON CO., for 230 volt 50 cycles A.C. TYPE R.Q.R., reversible. 37 r.p.m., overall size 5 in . $x$ $4 \mathrm{in} . \times 5 \frac{1}{2}$ in. Weight $4 \frac{1}{2} \mathrm{lb}$. lb Ex brand new equipment. Unnew equipment. used. Price

SOLAR OIL FILLED CONDENSER. 240 mid . for 230 V.A.C. or 600 volt D.C.
Overall size 14 in . $x$ 9 in . $\times 5 \frac{1}{2} \mathrm{in}$. plus feet. Weight 46 lb . Brand new. Guaranteed perfect. Manufacturer's packing. Price $67 / 10 /$, carriage 10/-.


DIAL THERMOMETER. Made by Short \& Mason, Calibrated 0-160 degrees Fahrenheit. $4 \frac{1}{2} \mathrm{in}$. dial. Gin. rim for flush mounting with 6 in . long rod protruding at the back. Brand new. Manufacturer's packing. Price 22/6. P. \& P. 3/:
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CATHODE RAY OSCILLOSCOPE
The famous model $160-$ B. C.R. 'Scope.. manufactured by R.C.A. of U.S.A. Best general purpose instrument of its kind, complete with 6 in. cathode ray tube. Unused, guaranteed perfect. For operation on 110 v. A.C. Price $\mathbf{£ 2 2} / 10 /-$. Carr. $10 /$ -
Step-down transformer to enable the above to operate on 230 v . Price $19 / 6$.

BRAND NEW SOUND POWER OPERATED EXADMIRALTY HEAD AND BREAST SETS. Two such sets connected up will provide perfect intercom.; no batteries required. Will operate up to $\frac{1}{2}$ mile. Original manufacturer's boxes. Price $17 / 6$ each, plus P. \& P. 2/-; or $32 / 6$ per pair P. \& P. 3/-


MUIRHEAD PRECISION 1 bank, I pole, 25 position Stud Switch. Brand new. Price 12/6. P. \& P. I/-.

MIDGET ROTARY ROTARY FORMERS
$2 \frac{1}{4} \mathrm{in}$. dia. $\times 4 \frac{1}{3} \mathrm{in}$ Input 11.5 volt Output $310 / 365$ volts at 30 mA . Brand new. 12/6 each. P. \& P. 1/6.
MINIATURE INSTRUMENT RECTIFIERS. Bridge Type 1 milliamp. Guaranteed perfect, 7/6 each.
S.T.C. RECTIFIER. 36 plates by 120 mm . Bridge connected. Maximum A.C. input 60 volt, D.C. output 15 amp . New, perfect. Price 60/-. P. \& P. 3/6.

8-day clockwork Time Switch. Contacts $2 \frac{1}{2}$ amp. 230 vole, 24 hour phase, $\frac{1}{\frac{1}{2}}$ hour divisions, allow setting for one make and one break to be made every 24 hours, complete with key. Used but guaranteed perfect. Price $27 / 6$ each. P. \& P. I/6

## METERS GUARANTEED PERFECT

Charging Types
${ }_{5} \frac{1}{4}$ amp. D.C. M.I. 2 in . fl. rnd................ $7 / 6$ 5 amp. D.C. M.I. $2 \frac{2}{2}$ in. fl. rnd............. $11 / 6$
$7 \frac{1}{2}$ amp. D.C. M.I. $3 \frac{1}{\frac{1}{2} \mathrm{in} . ~ p r o j . ~ r n d . . . . . . . . . ~} 12 / 6$ 9 amp.D.C. Hot Wire W.R. $2 \frac{1}{2} \mathrm{in}$. fl. rnd. $6 / 6$ 15 amp D.C. M.C. 2 in . rnd................ 30 amp D.C. M.C. 2 in . fl. sq. 100 amp . A.C. M.I. $4 \frac{1}{4} \mathrm{in}$. fl. rnd.
Voltmeters
12 v. D.C. M.C. $2 \frac{1}{2}$ in. proj. rnd.. 20 v. D.C. M.C. $2 i n$. fl. sq....
25 v. D.C. M.C. $2 i n$ f. rnd. 25 v. D.C. M.C. 2in. fl. rnd.
30 v. M.I. 3in. proj. rnd...... 30 v. M.l. 3in. proj. rnd.
40 v. M.C. 2 in. fl. sq.... 300 v. A.C. M.C. 2 in. fl. rnd 300 v. A.C. M.I. 2 tin. fl. rnd 400 v. A.C. M.I. 4tin. rnd...
Milliammeters
1 mA . M.C. $2 \frac{1}{2} \mathrm{in}$. fl. rnd... 200 mA . M.C. $2 \frac{1 \mathrm{in} .}{\mathrm{f}}$ f. rid. 500 mA . M.C. $2 \frac{1}{2} \mathrm{in}$. fl. rnd.


## Microamp

50 mieroamp., scaled $0-100$, M.C. $2 \frac{1}{2} \mathrm{in}$. fl. rnd. square, side fittin 50 scales scales
500 microamp., M.C. 2in. rnd. F.L. scaled $15 / 600$ volt. NEW .. Postage on all meters I/- each

Miniature latest type moving coil $0-5$ milliamp meter $1 \frac{1}{3} \mathrm{in}$. diameter, flush fitting, complete with fixing clip. Price 17/6. P. \& P. 1/-.


CRYSTAL CALIBRATOR No. 10. A
 crystal controlled 4-valve high-grade instrument in the same category as the
famous B.C. 221. famous B.C. 221. does not require cross reference or chares - functions as follows: (1) A crystal controlled osciliator which provides fixed frequency signals of 500 KC and all harmonics of 500 KC to beyond 10 Meg. and up to 30 Meg .
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Supplied complete with 3 spare valves, all leads and maker.'s instruction book in carrying haversack. The complete outfit is brand newrepeat NEW. Price: $£ 4 / 19 / 6$. Carr. $3 /$-.

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RECTIFIER mounted on 200/250 volt A.C. input transformer. Output $36 / 40$ volt D.C. at 1.2 amps. New, perfect. Price 16/6. P. \& P. 3/6.


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Assemble it yourself and SAVE £ £ £'s
COMPACT GRAM. AMPLIFIER 2-valve printed circuit type for use on L.C. or
D. $0.200 / 250$ v. matins incorporating modern D.O. 200/250 V. mains incorporating modern mhatature valves. Output 2 watts, overall dimen-
sions fit $\times 2 \times 3$ In. sions $6 \frac{1}{2} \times 2 \times 3$ in.
Price 59/6, plus P. \& P. 2/6.
Amplifer Cabinet, 22/19/6, plus 5/- P, AT-
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If ali the above items are purchased at the anme time they can be supplied at $£ 13 / 15 /=$, plus



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AVANTIC SPAll Stereophonic Amplifler. Technical details: power output (each channel) 10 watts peak, L.S. impedance, 4, 8 and 16 ohms 6 -position input selector, bass, treble, volume on/off controls, stereo reverse switch, phase reverse switch, stereo balance control, P.U. balance control. Dimensions $14 \frac{1}{2} \times 8 \frac{1}{2} \times 4 \mathrm{in}$. Original price 28 Gns . P. \& P. 7/6. OUR PRICE 19 Gins.

AVANTIC PL621 20 -watt monaural Amplifier, frequency response $10 \mathrm{c} / \mathrm{s}-30 \mathrm{Kc} / \mathrm{s}$. IdB. L.S. impedance 4, 8 or 16 ohms. Dimensions 14 in . $x$


AVANTIC STEP II. Stereophonic Magnetic Pick-up Amplifier Unit. Price $64 / 4 /$-.

All this equipment is Brand New and in manufactúrer's original sealed cartons. Full descriptive literature available.


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- High Q itrame aerials.
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- Instruction book 16 .
- Bize only $8 \times 8 \times 4 \xi \mathrm{in}$.
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These two receivers use the latest type circuitry and are fitted into attractive cabinets $12 \times 6 \frac{1}{2} \times$ $5 \frac{1}{2}$ in., in either walnut or ivory Bakelite or wood 1/- extra. Individual instruction books 1/each, post free.

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T.v. Valves, Fuses and Dial Bulbs. Will also give an accurate circust continuity test and also has builtolin
7 sud 9 ivive straighteners. 7 sud ${ }^{9}$ gize 5 ive straighteners.


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$\star$ Completely transistorised circuit.
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* Recordings interchangeable with other recorders. * Remarkable reproduction on both speech and music. Price complete with Microphone 25 GNS. plus
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-and
PREMIER BATTERY ELIMINATOR
Housed in two containers which are to replace AD 35 and B126 batteries.
KIT $37 / 6$ plus $2 /-$ post and packing. Only suitable for use with DK 96 Series valves.
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A NEW DESIGN $4 \frac{1}{2}$ WATT AMPLIFIER KIT MAY BE BUILT FOR

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A new circult for the home constructor requiring a good quality medium-powered Amplifier for reproduction of Records or F.M. Broaricasts. Technical 8pecifications: EL84 E7.80, Voltage adjustment for A.C. mains from 300.250 volt, 3 or 15 ohms impedance. Negative feedback Size $7 \times 5 \times 2 \mathrm{in}$., overall beight 5in. Bilver-hammered finished Chassis.


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CABY MODEL B2O MULTI-METER
DC/V $0.0 .5 \quad$ v. $0.2 .5 \quad$ ₹. $(2 \mathrm{~K} .0 \mathrm{hma} / \mathrm{V})$
DC/V 10-50-250-500-1000 + . ( $4 \mathrm{~K} \cdot \mathrm{obms} / \mathrm{V}$ )
AC/V 10-50-250-1000 $\quad$. ( $4 \mathrm{~K} . \mathrm{ohms} / \mathrm{V}$ ). DCimA $0-100$ microsmps $(500 \mathrm{mV})$
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OHMB. $2 \mathrm{~K}-20$ meg.
COMPLETE WITH TEST LEADSPRJCE $£ 6 / 10 / 0$ plus $2 \%$ P. \& P.

CABY MULTI-METER A-10
$\mathrm{DC} / \mathrm{V} 10-50-250-500 \mathrm{k} \Omega(2 \mathrm{~kg} / \mathrm{V})$ $\mathrm{AC} / \mathrm{V}$ 10-50-250-500-1k $\Omega(2 \mathrm{k} \Omega / \mathrm{V})$ Ranges: DC/mA $0.5-25-250(250 \mathrm{mV}$ онм $0-10 \mathrm{k} \Omega-1 \mathrm{M} \Omega$.

Complete with test leads
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 c1919/6. P. : P. $7 / 6.6$ TRANSCRIPTOR Twin track operation, 3 apeeds, st. 7!, 1 l.p.s. Will take 7in. spoois. £17/19/6. TAPE RECORDER AMPLIFIER specially designed to match the Collaro Studio Tape Deck. £10/19/6. P. \& P. 4/-: size $11+x$ cooled metal rectifer. Incorporates mike gram/radio mputs, ext. 1.8, Jack, superimposing switub with matching knobs.

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By well-known manufacturers, brand By well-kiown manufacture 1.800ft. on 7 in , spool
$1,200 \mathrm{it}$. on $5 \nmid \mathrm{fn}$. spool
$1,200 \mathrm{ti}$. on 5 tin . spool
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AMERICAN C.B.S.
RECORDING TAPE
Brand new, fuily guaranteed and eader Tape:-
600ft. on 5 in . 8 ppool
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1,800ft. on 7in. Spool L.'.
Plus 1 - per Spool, P. \& P.


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May be bullt for $29 / 6$ plus $1 / 6 \mathrm{p}$. and $\overline{\mathrm{p}}$. Tape Recorder Plug Extre.
Improve the quality of your recordings with the manst inexpensive Radio Jack avallable, suitable for any type of Tape Recorder, only a short extertial Aerlal required for full medium Waveband coverage. Phono Plugs--9d. Jack Pluge-3/-

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Two input sockets with associa of "mike" and gram. as in A. 10 Eigh sensitlvity. Includes
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$17 / 9$
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TRADE ENQUIRIES INVITED ON ALL ITEMS
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## Wilkínsons <br> METERS GUARANTEED F.S.D. <br> 100 Microamp 50 Microamp 250 Microamp 500 Microamp 1 Milliamp 2 Milliamp 30 Milliamp 100 Milliamp 200 Milliamp <br> 1 Ampere <br> 3 Ampere <br> 5 Ampere <br> 10 Ampere <br> 20 Volts <br> 40 Volts <br> 500 Microamp <br> 1 Milliamp <br> 5 Milliamp <br> 10 Milliamps <br> 20 Volts <br> 30 Volts <br> 40 Volts <br> 15 Amps <br> ${ }_{5}{ }^{5}$ Amps <br> 30-0-30 Araps <br> 50-0.50 Amps <br> 500 Milliamps <br> 25 Amps D.C. <br> 50 Amps A.C <br> METER RECTIFIERS $250 \mu$ A 1 M.A., 5 M.A., F.W. bridge, 8 ;'6, post 6 d .

 CROSS POINTER METERS. 2 separate 100 microamp movements, $22 / 6$. MICROAMMETER. 250 F.S.D. $3 \frac{1}{2} \mathrm{in}$. F.R. Sangamo Mod. S37. Scaled for valve voltmeter. Circuit available free, $55 /-$, post $1 / 6$.UNI-PIVOT GALVANOMETER, by Cambridge Instruments, $50-0.50$ microamps, dia. 4in. Knife pointer, mirror scale. Complete with leather carrying case. Ideal for laboratory use, $£ 10$, carriage $3 /-$ -
PORTABLE VOLTMETER. $0-160$ volts A.C./D.C. accuracy within $2 \%$, 8in. mirror scale, knife pointer, in polished case. A precision moving iron instrument at a very low price, $\varepsilon 4 / 19 / 6$, post $3 / 6$.
WHEATSTONE BRIDGE 1 to 10,000 ohms, plug type, 85 , carriage $7 / 6$. AVO TEST BRIDGES. $220 / 240$ volt A.C. Measure capacities from 5 pf . to 50 mfd . and resistances from 5 ohms to 50 megohms. Valve voltmeter range 0.1 to 15 volts and condenser leakage test. Full working instructions supplied with instrument. $\varepsilon 9 / 19 / 6$, post $3 /-$
SLOW MOTION VERNIER DRIVE. Scaled $0-180^{\circ}$. Ratio $38: 1$, diam. 3 in . 15/6, post $1 / 6$.
FREQUENCY METERS. $45-55$ cycles per second 230 volts, $6 i n$. dia., Flush Round. Brand new in maker's box, $£ 10 / 10 /$-, post $3 / 6$.


MINIATURE PRECISION MOTOR, 12 v. D.C. Size $18 \times 1+i n$. diam. Latest development. Extremely powerful with low consumption. Weighs as little as two ounces and totally enclosed in polythene protective case. Three position switch; forward, reverse and stop. $\quad 7,000$ r.p.m., self lubricating and long life sintered bronze bearing; 15/6, post 9d. Ask for free length of polythene flexible drive.

RELAYS P.O. TYPE 3000


## Built to your own

 specificationKeen Prices
Quick Delivery
Contacts up to
8-Changeover

MINIATURE RELAYS:

Siemens High Speed Sealed.

| mens High Speed Se |  |  |  | S.T.C. and G.E.C. Sealed. |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $2.2 \Omega+2.2 \Omega$ | H96A | 15/6 | $2 \Omega$ | 2 C .0. | 4184 GA | 18/6 |
| $145 \Omega+145 \Omega$ | H96C | 19/6 | $700 \Omega$ | 2 C .0. | 4184GD | 19 |
| $500 \Omega+500 \Omega$ | H96D | 22/6 | $2500 \Omega$ | 1 makeHD | 4186 EE | 22 |
| $1700 \Omega+1700 \Omega$ | H96E | 25/- | $2700 \Omega$ | $2 \mathrm{C} . \mathrm{O}$. | 4184 GE | 21/6 |
| Siemens High S | ed Open |  | $180 \Omega$ | 2 m 2 b | M1087 | 1916 |
| $100 \Omega+10082$ | H85N | 15/- | $670 \Omega$ | 4 C.O. | M1092 | 21/6 |
| $850 \Omega+850 \Omega$ | H85W | 15/- | $2500 \Omega$ | 1 C .0. | M1022 | 22/6 |
| $000 \Omega+1000 \Omega$ | H95A | 17/6 | $5000 \Omega$ | 2 C .0 | M1052 | 25/ |

ERICSSON SEALED. Highly sensitive, 700001 e 024 ve 25/ Comprehensive range available from stock
SWITCHES, 1 hole fixing, 3 amp .250 volt.

## 1/6 each, $12 /$ - doz.

RACKS-POST OFFIGE STANDARD. 6ft. high with U-channel sides drilled for 10 in. panels, heavy angle base, 4 ft . 10 in . in stock.


TELEPHONE SET TYPE "A." Kinging aind Speaking both ways on a four core cable. Carries the voice loudly and clearly over any distance. Two handsets are supplied as illustrated and the set is complete with Pushes, Buzzers, Battery, Plugs and Sockets. We can supply 4-core PVC cable at 10d. per yard or 2-core at 3d, per yard extra. Price 75/-set, post $3 / 6$. TELEPHONE SET "TELE-F." This is the best known portable telephone ever made, it has a built-in generator for ringing the other instrument and requires only twin wire between the sets. The set of two instruments and batteries in carrying case, $87 / 10 /-$, post $7 / 6$. Twin flat P.V.C. wire 3 d . yard. TELEPHONE SET TYPE"'K." The most compact telephone set available as the $4 \frac{1}{2} \mathrm{~V}$. flat battery and buzzer is built-in to the hand instrument. Ringing and speaking both ways on fwin wire, instrument is complete with 5 ft . flex. Easily hangs on the wall. Set of two instruments, $85 / 10 /=$, post $3 / 6$. JACK PLUGS. Cylindrical bakelite screw-on cover, 2 contact $2 / 6$, post 6 d
SOCKETS. One hole fixing for above, $3 / 6$, post 6 d .
RESISTORS EX STOCK, IN QUANTITY WIRE WOUNO, HIGH STABILITY GARBON ETC., BEST MAKES AT LOWES' PRICE.


## MAGNETIC COUNTERS

## Counting to 9999

$2-6$ v. D.C., 15 - each, post $1 / 6$ $75-230$ v. D.C., $15 /$-each, post $1 / 6$. HIGH SPEED TYPE No. 100 c $35 /-$, post $1 / 6$.
VEEDER-ROOT MAGNETIC COUNTER. General purpose type with zero reset. 800 counts per minute up to 999,999 . 48 volt D.C. $55 /-$, post $2 / 6$ SATCHWELL THERMOSTAT8 adjustable between $70^{\circ}-180^{\circ}$ fahrenheit Operates $0-440$ A.C., 20 amps., 11 in. stem. Fitted cover. $25 /-$, post $2 / 6$. ROOM THERMOSTAT. Adjustable between 45 and 75 deg. Fahr., 250 v 10 amp. A.C. Ideal for greenhouses, etc., 35/-, post $2 /-$.
THIS MONTH'S SPECIAL OFFER: $\frac{1}{3}$ H.P. CAPACITOR MOTORS. $230 / 240$ volts, 50 cycles, 1,420 r.p.m. $\frac{1}{8}$ in. shaft, resilient mounting Or with ini. shaft on standard foot mounting.

$$
\text { Either type } 5.10 \text { : } 0 \text {. carriage } 10 / \text {. }
$$

ROTARY CONVERTERS. Input 12 D.C. Output 230 A.C. 50 cy. 135 watts. In fitted case with variable resistance, $0 / 300$ voltmeter. The ideal job for T.V. and tape recorders where A.C. mains are not available. £10, carr. 15/-. Special connectors, one fitted with 6ft. heavy duty flex and clips for D.C. side, $10 /-$ set, post $1 /$-. ROTARY CONVERTER, input 12 v . or 24 v D.C., output 230 v. A.C., 135 watts, $£ 8 / 10 /-$, casriage $7 / 6$.

ROTARY CONVERTER. Input 24 V. D.C. Output 220 V. A.C. 250 watts Pedestal typeuth D.P. Ironclad switch. BRANDNEW. $817 / 10 /-$. Cge. $15 /-$ BATTERIES. Portable Lead Acid type, 6 yolts 125 ampere hours. In metal case $16 \mathrm{in} . \times 8 \mathrm{in}$. $\times 11 \mathrm{in}$. (Two will make an ideal power supply for our 12 volt Rotary Converters.) Uncharged 86/10/- each, carriage 15/-. 24 volts 85 amperes, 144 each, carriage 15 -
15 AMP. BATTERY CHARGER (Westinghouse Type B.C.3) will charge three lead acid cells at 15 amps. Input 200/250 volts, 50 cycles A.C. Charging current is regulated by four-position switch and variable resistance for fine control. Fitted with $0 / 20$ ammeter, rotary on/off switch and rewirable fuses. This first-class instrument at the bargain price of $£ 15 / 0 / 0, \mathrm{cge} .15 /$ GEARED GAPACITOR MOTOR8. $220-240 \mathrm{v} .50 \mathrm{cy}$., 30 watts, $300 \mathrm{r} . \mathrm{p} . \mathrm{m}$ also spindle for 1425 r.p.m. A powerful and useful motor $75 / \mathrm{F}$, post $\$ / 6$ TERMINAL BLOCKS. 2-way $4 /-$ doz., or box of 50 for $15 /=, 3$-way $6 /$ - doz., 50 for $22 / 6$, post $1 / 6$. T.C.C. CONDENSERS. $0.1 \mathrm{Mfd} .31 \mathrm{kV} .75 /-$ each, 1 Mfd. 10 kV . $45 /$ - each
SOLENOIDS suitable for remote control, mechanical indicators, etc. 12 v. D.C., 400 M.A., $80 \Omega, 3 \frac{1}{2} \mathrm{in}$. arm, $\frac{1}{3} \mathrm{in}$. movement, $5 /-$ each, post $1 / 6$.
NIFE BATTERIES, Nickel Cadmium 12 volt 18 ampere hours crated and connected alkaline filled Brand New \&4 each, carriage 10/-. Also available 2.4 volt 10 ampere hours, $20 /$ each, post $3 / 6$.

TRANSFORMER. Single Phase $250-115$ volts 50 cycles 5 KVA double wound, $\mathbf{2} 30$, carriage extra.
PUMP Electrically Driven by a 24 v . D.C. motor. Works efficiently on 12 v . Totally enclosed, self lubricating driven through 4 to 1 reduction gearbox delivering 60 g.p.h./301b./sq. in. Inlet and outlet unions + BSP $37 / 6$, post $2 / 6$. OSCILLOSCOPE No. 11 with high-class amplifier, 230 volts $£ 12 / 10 /-$, cge. 15/

## ELAC LOUDSPEAKERS P.M.

5 in . Round, 3 ohms, $12 / 6$, post $1 / 6$.


Get Finest Value from IRONGATE-England's Leading Equipment Wholesalers Bulk Buying means LOWEST PRICES. All Equipment is in TIP-TOP condition


8UPPLIED
WITH ORYSTALS. Accessories can be supplied to order at extra cost.


Only 510 3 v. (i.e. Ruben Mallory Type 1). 14 of the current series of B7G valves are employed: 1-3A4, 6-IL4, 4-IT4, 1-IS5. each. 2-IA3. Each set is in first class condition. Special quorations for quantities up to 3000 sets. "22"' SETS ALSO- 300 available only. New condition $£ 10$ each.
carr.
The best portable telephone ever
made. Original cost \&40! Range
up co 5 miles. idealfor FACTORIES,
BUILDING SITES, FARMS,
OFFICES, 2 perfect case sets with
batteries, IOOft. cable, etc.
D3 STRANDED TELEPHONE CABLE.
D3 STRANDED TELEPHONE CABLE. New Mlle Drum 85!-. Carr. 1716.

## HEAVY DUTY <br> 20 AMP. L.T. SUPPLY UNIT


s.T.c.

## Normal

 cost over £ 100equipment
for Electronic Engineering. research laboratories, schools. Ideal for battery charging etc. Guaranteed for 20 amps.
Output: D.C. Variable up to 20 amps. and 24 v . or trickle charge 125/350/700 ampere hours.
Input: A.C. $100 / 260$ volts $45 / 65$ cycles. Size: $16 \times 24 \times 32 \mathrm{in}$. high.
In attractive
Grey Cabinet.
Gey
ex Warehouse
(Circ. diags, and instru. loaned for 10\%-deposit.)

## 750,000 YARDS!! SCREENED WIRE FLEX FOR ONLY 2d. per yd.



For Immediate Delivery-far below cost. Specifloation: Close braided 14/.0048in. Covered specifioation: Pinned Copper. Screened. Assorted colours. Applications: Microphone leads, pick-up beads, ete. ON MAKER'S REELS 220 yd. REELS (min. quantity) 36/8. P. \& P.
TEN HEELS \&17. Carr. Paid. TEN HEELS £17. Carr. Paid. 30/8. $5 / \%$
 260 volts, 5 amps. INPUT 230 volts, $50 / 60 \sim$ A SHROUDED FULLY VARIABLE TRANSFORMER COR BENCH OR PANEL MOUNTING. SIZE: Approximately 210 inches Cube. WEIGHT: Approximately I3!b. B9.0.0
PRICE: RIDICULOUS, ONLY
Plus $12 / 6$ cari.. supplied and boxed new. Also available 10 amps. Transformers $£ 18 / 5 /-$ and 20 amps. $632 / 10 / \%$.

## CONSTANT VOLTAGE

 TRANSFORMERSFERRANTI 7 H/KVA MOVING COIL
Stabilized outpui voltake fin the range Stabilized outpat voltage in the range 200.250 Plug-voard tappings. The selected output roitage is constant witb $+\frac{1}{2} \%$ at all luade 0 to $30 / 37$ t amps.
wben the supply when the supply voltage is varying over the range
$+8 \%$ to $-12 \%$. Krequency compensated $45-55$ and
स Excellent output wave-form.
\& Can be used as a variable transformer. \$ Can be used as a variable traneform bouk at a fraction of the normal cort.
$\star$ A.C. MAINS STABLLIZER. ONLY


AUTO TRANSFORMERS 3 KVA Air Cooled ( $100 \%$ under-rated). GOARANTEED $230 / 250$ tapped, 12 amps. 6 KVA $105 / 120$ tapped 28.5 amps Made by well-knowin manufacturer and boused on strong metal case. Weight: 2 cwt
PRICE \&15.0.0
Carr. 25/-
FAMOUS TAPE REGORDER
LISN

## PLESSEY T.V. CHASSIS FOR SPARES <br> 9'6

56 resistances. 54 condensers. 13 valve holders. 4 transformers. Chokes 250 ma. Metal rectifers 300 volts @ 250 ma . Fuse panel. Focus magnets. Plugs. Sockets. Carr. 7/6.

REPLACEMENT, REBUILT T.V. TUBES 12 namint Carr, \& Ins. 15/6.
21 in . tube $88 / 10 /-$. 17 in . tube £7/10/-. £2 allowed on old tube. $12,14,15 \mathrm{in}$., tubes $\mathbf{5 5 / 1 0 / - , ~ £ 1 ~}$ allowed on old tube. Terms available over 20 weeks.

CONTEMPORARY
EXTENSION
SPEAKERS
19/9

Ideal for extra stereophonic speaker. covered in smart two-tone leatherette. Beautifully made. Including 8 in . | Beautifully |
| :--- |
| speaker. |
| P. made. |

## EXTENSION

SPEAKERS
19/9


Bin. P.M. Speakers fitted into polished cabinets. Standard matching to any receiver. (Complete.) Switch and flex included. P. \& P. 3/9.

## BARGAIN SPEAKER

5/9
8in. P.M. Repaired cone defect not affecting reproduction quality.
gin. P.M. SPEAKER
6/9
As above but with output transformer
8in. P.M. SPEAKER
Perfect quality. Fitted output transformer.
ELLIPTICAL SPEAKERS
15/8
$8 \times$ 3in. and $7 \times 4$ in. Brand new Also $9 \times 4 \mathrm{in}$ @ $19 / 9$.
P. \& P. on each $2 / 9$.

TELEPHONE SETS 7/9
X.W.D. Wireless remote control unit. E.M.K.II (ZA11954). Including morse tapper, switched, jack plugs, etc. Less phone. Ins. and carr, $3 / 6$.


Beautifully styled rexine covered cabinet in Red/Beige with carrying handle. in Red/Beige with cartying handle. Size: $14 \frac{1}{2} \times 13 \times 9 \frac{1}{2}$ in. Storage comp. in lid for tapes and mike. Speed 3 3in. per second. Compact set using latest 5 valve amplifier with 4 -stage amplification and separate valve for Bias osc. 2 controls. Contatms $7 \times 4 \mathrm{in}$. elliptical speaker and incorporating B.S.R. Tape deck. 5 in. standard tape. 3 months' guarantee. Ins. \& Carr. 12/6. Deposit \&8 plus Ins. Carr. and 20 payments of $17 / \%$.
 Cabinet. Size: $13 \times 101 \times 7$ Tape. Recording in two-tone coloured rexine cloth. Stylish design. Carrying handle with detachable lid. Easily adapted to Record Player Cabinet. Exceptional value at this very low price. P. \& P. 4/6.

## 17" T.V.'s $17 \frac{1}{2}$ gns.

Modern CHASsIS, modified
Complete 17 in . TUBE. VALVES SPEAKER - KNOBS. Tuned ITV/BBC. Ready to use, fully guaranteed, TUBE 12 months, CHASSIS and VALVES 3
months. Cabinet to fit $\varepsilon 1 / 11 / 6$ if ordered with set. Salvage. Set-tube-cabinet despatched separately. Carr. \& Ins, on set $£ 1 / 5 /$-; on Cabinet $8 / 6$.

## SOLO

SOLDERING TOOL 12'6

$110 \mathrm{v} ., 6 \mathrm{v}$. or 12 v . (special adaptor for $200 / 250$ v., $10 /$ - extra). Automatic solder feed including 20ft. reel of Ersin 60/40 solder and spare parts. It is a tool for electronic soldering or car wiring. Revolutlonary, in design. Instantly ready for use and cannot burn. In light metal case
with full instructions for use. Post $3 / 6$.

## RECORD PLAYER CABINET R.P. 9

19/6
Exceptional offer. A lightweight portable player Cabinet in two-tone Rust and Cream. Famous manufacturer. Size: $14 \frac{1}{2} \times 11 \frac{1}{8} \times$ 6in. Complete with moulded deck board of attractive design. Takes B.S.R. TU9 single player; 2 control Amplifier; 5 in. round Speaker, Post, Packing \& Ins., 4/6.

## "STUDIO TWIN" 30 gns <br> Now 29 gns. <br> Quality Twin Speaker Tape Recorder at this amazing reduced price. 7 in. 1,200rt. 3 -speed Deck, 17, 33, 7 I I.P.S. Includes Twin Tracks, Reverse counter, Pause control and magic eye recording indicator. Volume and tone control, superimpose switch. Two matched speakers. 3 watts output. Attractive design cabinet in beige Size 19 in. $x$ 13in. $x$ 8in. Ins. \& Carr. $12 / 6$. <br> EXTRAS: Microphone 27/6. Tape 25/-.

## B.S.R. TAPE "MONARDECK" £9.9.6

Single speed. $3 \frac{3}{4} \mathrm{in}$ per sec. Uses $5 \$ \mathrm{in}$. spools. Complete with Record-Playback and erase heads. Simple control for R/P.B. and fast rewind with safety switch for record. Size; $13 \times 8 \frac{3}{4} \times$ $2 \frac{1}{2}$ in. Colours: two-tone grey. P. \& P. 4/6. Deposit $60 /$ plus P. \& P. and 20 weeks at $8 /$.

## TAPE RECORDER AMPLIFIER

£7.19.6
Compact, well designed 5 valve amplifier. Output 3.5 watts Valve line up-ECC83, double triode first audio amplifiers ECL 82 triode pentode further audio amplifier and output valve 6BW6 bias and erase oscillator, EM84 record level indicator EZ80 H.T. rectifier. Input for mike, radio and gram. Controls record playback volume and on/ofi playback tone. Dia $8 \frac{1}{2} \times 3 \times 47 \mathrm{in}$. ins. and carr. $4 / 6$. Terms
EXTRAS: Knobs $2 / 6$ per set. Beautiful Perspex dial plate, Com plete with sockets for mike, radio and superimposed switch 3/6.

AMPLIFIERS. All portable. 12 months guarantee MK. D5.

## 39/6

Simple circuit employing ECL80 triode pentode output valve giving 3 watts output. A.C. only. Mains isolated. Single control for volume and on/off switch with knob. P. \& P. $3 / 6$. MK. D2A.
Latest design incorporating negative feedback, giving 4 watts undistorted output. Valves: ECL82, triode pentode and contact cooled metal rectifier. Tone and volume control panel on flying leads with compact amplifier chassis suitable to mount under modern autochangers to give easy mounting in small cabinet. A.C. only. Mains isolated. Output for 2.3 ohms. MK. D3A.

89/6
As above de luxe model, with separate tone controls for treble and bass.
PORTABLE AMPLIFIER (Salvage) 69/7
3 valves. 4.5 watts output. Size $7 \times 5 \times 4 i n$. An ideal amplifier for stereo record players, Tape Recorders, Microphone, Baby Alarm, etc. Volume and tone controls. 200 250 v. A.C./D.C. P. \& P. $4 / 6$.

## NEW LOOK AMPLIFIER

99/6
Ideal stereo attachment. Brown/Ivory cabinet, carrying handle. Contains high flux 8 in . speaker. A.C. mains. halves-10F3, 10P14, U404. Max. output 4.5 watts. $14 \times$ $11 \times 6 \frac{1}{3} \mathrm{in}$. Completely converted. Ideal stereo, Mike, Guitar, Records. 2 units give world of amusement and speclal stereo effects. 12 months' guarantee. P. \& P. 5/6. Deposit 30/plus P. \& P., 10 weeks at $8 /-$

## RECORD PLAYER CABINET R.P. 2 FOR ONLY 59'6

Made by famous manufacturer. In polka dot cloth with clipped lid and carrying handle. Size: $16 \times 14 \frac{x}{2}$ $8 \frac{1}{2 n}$. deep. Carr. \& Ins, 4/6 Will take B.S.R. Monarch Autochanger, $26 / 19 / 6 ; 7 \mathrm{in}$. $\times$
4in. Elliptical Speaker 15/9; 4 in . Elliptical Speaker 15/9; Amplifier 79/6.


## C.R.T. ISOLATION TRANSFORMERS

For Cathode Kay I'ubes having Elater/Gathode shor' eircuit and for C.R. Tubes with falling etaission. Full instructions supplied.
and $\mathbf{5 0 \%}$. Tapped mains primari. Optlonal Boost $25 \%$
2 volt
4 volt
6.3 volt
6.3 volt
10.8 volt
10.8 volt
$12 / 6$ each
126 each
$12 / 6$ each
ODR LATEST SUPERIOR PRODUCT. Type A2
High Quality. Low capacity, $10 / 15$ pt. $16 / 6$
Optional boost $25 \%, 50 \% .75 \%$.
Type B. Mains input. Low capacity. Mults Output $2,4,6.3,10$ and 13 volts. Optional boost 25
$50 \%$. Buitable for all Cathode Ray Tubes 21/-
RESLsTOLs. All prerenged vaiues. $20 \% 10$ ohan to 10
 10 ohms to 10 meg., Ditto $5 \%$ gd., 100 D to 0 meg . $\left.\begin{array}{l}5 \text { watt } \\ 10 \text { watt }\end{array}\right\} \quad$ WLRE-WOUND RESISTORS 12,500 ohms $-50,000$ ohmes. 10 W.
WR,500 ohms. $-50,000$ ohme
WREE OUND POTS, $\$ w$ Pre-set Min. T.V. type AII vuluch 25 ohms to 25 K $3 /-$ ea., $30 \mathrm{~K} .50 \mathrm{~K} ., 4 /$ -3i- ea., 30 K. 50 K ., $4 /$-.
 O/P TRANBFORMERS, Heavy duty 60 ta $4.4 / 6$. Mult ratio pisb-pull 7/6. Miniatire 3V4. pete. $4 / 6$. Hygiade
 $10 \mathrm{H} 120 \mathrm{mA}$. . $12 / 6.10 \mathrm{H} 150 \mathrm{mA} ..14 / \%$ 。

$$
\begin{aligned}
& \text { THANS TRANSFORMERS 800/250 } \\
& \begin{array}{l}
\text { tandard } 250-0-250,80 \mathrm{~mA} .16 .3 \mathrm{v} .3 .5 \mathrm{a}, \\
\text { tapped } 4 \mathrm{v} .4 \mathrm{a} \text {. Bectifier } 6.3 \mathrm{v} .1 \mathrm{a} \text {, tapped } 5 \mathrm{v} \text {, }
\end{array} \\
& \text { Or } 4 \text { V. } 2 \mathrm{a} \text {, DItto } 35010-0-350 \\
& \text { GIDGET, } 220 \text { ₹ } 45 \mathrm{~mA} 8,6.3 \text { ₹. } 1 \text { a } \\
& \text { MALL, } 200 \cdot 0-20050 \mathrm{~mA} \text {. } 6.3 \text { v. } 2 \text { A } \\
& \begin{array}{l}
\text { STANDARD, } 250-4-250,65 \mathrm{~mA} . .6 .3 \text { ₹. } 3.5 \\
\text { HEATER TRANS., } 6.3 \mathrm{v} .14 \mathrm{a} ., 7 / 6: 3 \text { amp }
\end{array} \\
& \text { GEATER TRANS., 6.3 V. } 14 \text { a., } 716 ; 3 \text { am } \\
& \begin{array}{l}
\text { GENERAL PURPUSE LOW VOLTAGE, Outputs } 3 \\
\text { 6. 8, 8, 10, 12. } 15,1824 \text { and } 10 \mathrm{v} \text {. at } 2 \mathrm{~A} \text {. }
\end{array}
\end{aligned}
$$

 0.3 in. FORMERS 5437 or 8 aud Cans TVI or 2 I in sq

SLOW MOTION DRIVES, Epleyclicerstio $6: 1,2 / 3$.
 KEMPLOY LNSTRUMENT IRON. 220/40 v. $25{ }^{\circ} \cdot 17 / 6$. $1,000 \mathrm{ohms} 4 / 3$. 2 a amps. 4/3. I Ad. mp. 2,060 ohme, $5 / \mathrm{m}$ 1,000 ohms
LINE CORD. $3 / 3$ a amp. 60 ohms per foot $2,2,060$ ohmp. 100 ohme per foot, 8 -way, 6 d . per toot, 3 way 7d. per Inot.

## CRYSTAL MIKE INSERT by Acos 6/6

 Precision engineered. saise oniy $\frac{6}{} \times 1 \frac{3}{26} \mathrm{in}$ ACOS CRYSTAL STICK MIKE 39-1. Bargain 35/-. MIKE TRANSF. 50:1, $3 / 9$ ea.; $100: 1$ Potted 10/6 LOUDSPEAKERS PM.$6 \mathrm{in} . \times 4 \mathrm{in}$. Rola, 18 i $6 \mathrm{in} . \times 4 \mathrm{in}$. Rola, $18 /=$
in. $\times$ in. Ples. $A, 21 /-$ 6 in . Rola, 18/6. 8in. Rols, 21/e 10 in . R.A $30 /$ HI-FI TWEETERS, 4 in . 25/-. 15 12in. Plessey, $30 /$ 121 n . Baker $15 \mathrm{wt}$.3 ohm, and 15 ohm moders, $90 /$
 12 in .15 ohm Plessey $10 \mathrm{wt}, 45,.3.31 .14 \times 8 \mathrm{in} .45 /$
I.F. TRANSFORMERS $7 / 6$ pair $465 \mathrm{ko} / \mathrm{s}$. Slag tuntag miniature can $21 \times 1 \times 1 \mathrm{lu}$. Bigh Wearite M800 L.F. Miniature $466 \mathrm{ko} / \mathrm{s} ., 12 / 6$ pair. Weymouth I.F. Standard size $463 \mathrm{kols}, 12 / 6$ pair

CRYSTAL DIODE G.E.C., 2/-. GEX $34,4 /-.40$ Circuits 3/H.R. H8ADPHONES, 4,000 ohms, brand new, 15/- pair SWITCE CLEANER Fluid, equirt apout, $4 / 3 \mathrm{tin}$. TWIN GANG CONDENSERS. 365 pf, Miniature, $1 i$ in $\times 1 \frac{1}{2}$ in. $\times 1$ igin.. $10 /=$. Miduet 76 . Sinuie 50 if $2 / 6$ 100 pf ., $150 \mathrm{pt} ., 7 / \mathrm{F}$. Bolld dielectric $100,300,500 \mathrm{pf}, .3 / 6$ VALVE HOLDERS. Pax. int. Oct. 4d. EFto, EAb0 6d R12A, CRT, $1 / 3$. Eng. and Amer. $4,5,6,7$ pln, $1 / \cdot 1$
MOULDED Mazta or fint. Oet. 6d. B7G, B8A, B8O. B9A 9d. B7G with can, 1/6; B12A, 1/3. B9A with can, $1 / 8$ CERAMIC. EF50, B7G, B9A, Oct., $1 /-$ B79, B9A Cans, $1 /-$
 Bamples, 8.A.E.
WAVECHANGE SWITCEES
2p. 2-way, or 3 p. 2.way; biort spindle
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IA $33 / 38$ megs, complcte with frame-grid valves With


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 30 mA . $7 / 6 ; 60 \mathrm{~mA}$. $816 ; 85 \mathrm{~mA} .916$ : 200 mA. $300 \mathrm{~mA} ., 27 / 6$; Fuli Wave $250 \mathrm{v} 120 \mathrm{~mA},$.151 COILS. Wearite " $\boldsymbol{P}$ " type 3/- each. Ominor Midg type adj. dust core from 4/- each. All ranges. FERRITE BOD AER1ALS. M.W. 8/9; M. \& L., $12 / 6$. $\therefore$ R.F, COILS. A/HF, 7/F pair. H.F. chokes, 2/6. JASON F.M. TUNER COIL SET, EA/-. H.F. coll aerial
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| 18 | 7/6 6N7M | $6 /$ | EABC |  |  |  |
|  | 61-6Q7G | 716 | EB91 |  | HaBC | 1216 |
| 2 X 2 | 3/6 68A7M |  | EBC33 | 816 | HVR2A |  |
|  | $7 / 6$ 68J7M | $6 / 6$ | EBC41 | $8 / 6$ | MU1 |  |
| 3 V 4 | 7/6 68N7 | 616 | EBF80 |  |  |  |
| $5 \mathrm{U4}$ | 7/6.6V6G | $6 / 6$ | ECC84 |  | PCC |  |
| 5 Y 3 | 7/6\|6x4 | 7 | ECFP80 |  | PCF80 |  |
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| 606 | 6/- 12BA5 | 816 | EF50 | 516 | PY82 |  |
| 61 | $7 / 6$ 12BE6 |  | EF80 | 8/- | ${ }^{\text {SP81 }}$ |  |
| 6 | 6 12K7 | . 616 | EF86 | 4/6 | UBC4 |  |
| 6.5 | $12 \mathrm{Q7}$ |  | EF92 | 516 | UCH4 |  |
|  | $6{ }^{3516}$ |  | EL32 | $51 /$ | UF41 |  |
| 6J7a | $6{ }^{3574}$ | 6 | EL4I | 816 | UL41 |  |
| 6K6at | 680 |  | EL84 | 81 | UY41 |  |
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$314,3 V 47 / 6$ \& ECF80 \& 9/6 \& PCC84 \& $9 / 6$ <br>
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$6 K 7$ \& $5 /-$ \& ECL80 \& 10/6 \& PCL83 \& 126 <br>
6 K 8 \& $7 / 6$ \& EF80 \& $8 /-$ \& PL81 \& 126 <br>
607 \& $7 / 6$ \& EF86 \& $12 / 6$ \& PL82 \& $0 / 6$

 

697 \& $7 / 6$ \& EF86 \& $12 / 6$ \& PL82 \& $9 / 6$ <br>
$6 V 6$ \& $8 / 6$ \& EF91 \& $5 /-$ \& PL83 \& $10 / 6$ <br>
DAF96 \& $9 /-$ \& EL41 \& $9 / 6$ \& PY80 \& $7 / 6$

 

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DL96 \& O/- \& EY51 \& 9/6 \& PY82 \& O/6 <br>
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10/- part exchange allowance on old tube Carr. and ins. 10/-. Comprehensive stocks-quick delivery.

CONDENSERS-Silver Mica. All pref values, 2 pl . to $1,000 \mathrm{pl} ., 6 \mathrm{~d}$. each. Ditto ceramics 9 d. esch. Tubulars 450 \%. T.C.C. ete., 001 mid. -01 and, $1 / 350 \mathrm{v}$., 9 d . each.
 RESISTORS-FUL NGE 10 ohme
 Midget type modern rating), 1 w. 6d., 2 w $9 \mathrm{~d}, 10 \%$ Hi-stab, $\frac{1}{4}$ W., $5 \mathrm{~d} .$, \& $\mathbf{w}, 7 \mathrm{~d} ., 5 \%$ w., 9d.,

PRE-SET W/W POTS. T/V type, 25 obme50 K ohms $3 /-\quad 50 \mathrm{~K}-2 \mathrm{Mez}$. (Carbon $3 /-$ ) SPEAKER FRET-Expanded Bronze ano dised metal $8 \times 8 i n, 2 / 3 ; 12 \times 8 i n .3 /-$
$12 \times 121 n, 4 / 6 ; 12 \times 16 i n, 6 /-24 \times 12 \mathrm{in}$. $9 /-36 \times 12 \mathrm{in}$., $13 / 6$, etc., etc.
IYGAN FRET (Contemporary pat.). $12 \times$ 2 in . 2/-; $12 \times 18 \mathrm{in}$., $3 /-12 \times 24 \mathrm{in} .4 /-$ etc
 Elac, 17/6; 3 inn. Goormans $18 / 6$; Bin mans Elliputical, 18/6; 8in. Rola, $20 /=; 10 \mathrm{in}$ R. and A., $25 /-$; 10 in . W.B.-HF1012, $95^{\prime}-$ 12in. Plessey 15 ohms with 6/4in. Tweeter and Cross Over Filter, 87/6.
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|  | $4 / 6$ |  |  |  |  | $\begin{array}{lllll}1 / 450 & \text { v. } 4 / 350 & \text { v. } 2 / 3 & 52+30 / 30 & \text { v. } \\ 50 & 5 / 6\end{array}$ $\begin{array}{llll}16+16 / 450 \\ 32+32 / 450 & \text { \%. } & 5 / 6 & 60+250 / 275 \nabla .12 / 5\end{array}$ Comprehensive range in stock.

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Three Waveband and Switobêd Gram positions. Med. 200. 500 m ., Long 1,000-2,000 m., VHFF/FM 88-95. Mc/s. Philips Cont nental Tuning insert with permeability tuning on FM and combined AM/FM IF transformers, $460 \mathrm{Kc} / \mathrm{s}$.
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| INPUT | OUTPUT | OUTPUT | OUTPUT |
| :---: | :---: | :---: | :---: |
| VOLTAGE | VOLTAGE | FREQUENCY | POWER | TYPE A


| 12 VDC | $200-250 \mathrm{~V}$ | $50 \mathrm{c} / \mathrm{s}$ <br> squarewave | 75 VA <br> to 750 VA |
| :---: | :---: | :---: | :---: |

For use where a squarewave output waveform Is acceptable.
TYPE B

| 24 VDC <br> or 28 VDC | $200-250 \mathrm{~V}$ <br> 115 V | $50 \mathrm{c} / \mathrm{s}$ sinewave <br> $400 \mathrm{c} / \mathrm{s}$ sinewave | 75 VA <br> to 750 VA |
| :---: | :---: | :---: | :---: |

For use where a strictly sinusoidal output waveform is essential; total harmonic content less than $5 \%$.

TYPE C

| 24 VDC | 240 V | $2,000 \mathrm{c} / \mathrm{s}$ | 10 W or 40 W |
| :---: | :---: | :---: | :---: |
| For battery operated fluorescent lighting ; no starters or ballasts <br> required. |  |  |  |



| 12 VDC <br> 24 VDC | As required | DC | Up to 600 W |
| :--- | :--- | :--- | :--- |
| For H.T. DC supplies. |  |  |  |

## TYPE E(1)

| 12 VDC | 275 V | DC | 30 W |
| :---: | :---: | :---: | :---: |
| For receiver power supply; prlce $£ 15.0 .0 \mathrm{~d}$ |  |  |  |

## TYPE E (2)

| 12 VDC | 310 V | DC | 80 W |
| :---: | :---: | :---: | :---: |
| For transmitter power supply; price $£ 17.10 .0 \mathrm{~d}$ |  |  |  |

Types E(1) and E(2) have been specially designed to replace the rotary generators or vibrator packs used in mobile VHF transmitter-receivers. They are compact, cheap, require no maintenance and effectively reduce battery drain. Other voltages are available, as required.
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ATP4 316kT33C $A Z 1$
AZ31
B36
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D42

## DF96

 $\begin{array}{ll}\text { DHIOI } & \text { Bj- KTIOI } \\ \text { KTW6I }\end{array}$

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SMD25 Ferguson 992T, 994T, 9961
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SMD28 Regentone Double 2 Portable SMD29 Regentone Double 2 Portable SMD30 KB Rhapsody, MP151/I
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Deccamatic, fitted with valve ECL82, contact cooled rectifier, cone and volume on/off, chassis size $8 \frac{1}{2} \mathrm{in} . \times 3 \mathrm{in}$. complete with Celestion speaker Bin. $\times 6$ in., $64 / 5 /=$.

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[^16]

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MORSE KEY with buzzer, on board, wired for $4 \frac{1}{2} v$, battery, 8/6 (p.p1/6). Keys, morse, small, brass, 8 A., $3 /-$; covered with plug, $4 / 6$; large $7 / 6$. TELE ${ }^{\prime} F^{\prime} "$ intercom. sets, good condition, pair $65 /=$, post free Gt. B. METERS ( $100 / 4$ ) with two centre-zero movements, 600 and $400 u \mathrm{~A}, 8 / 6$. TRANSFORMERS. Open, upright, input 200/250 v. and $400 u \mathrm{~A}, 8 / 6$. TRANSFORMERS. Open, upright, input $200 / 250 \mathrm{~V}$.
Outputs:- $250-0.250 \mathrm{v} ., 150 \mathrm{~mA} .5 \mathrm{v} .3 \mathrm{~A}$. and $6.3 \mathrm{v} .5 \mathrm{~A} ., 25 / \mathrm{m}$. In-Outputs:- $250-0-250 \mathrm{v}, 150 \mathrm{~mA} ., 5 \mathrm{v} .3 \mathrm{~A}$. and $6.3 \mathrm{v} .6 \mathrm{~A} ., 25 /-$. In-
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    1960 Conference on Propagation of E.L. Flectomagnetic Waves, Boulder, Colorado, U.S.A., and reported in the National Bureau of Standards Technical News Bulletin, Voi. 44, No, 5, May 1960.
    $=$ I.R.E. Trans. on Antennas and Propagation, Sept. 1960.

[^2]:    * Amateur Radio Station G2BCX

[^3]:    *P.476. W'ireless World November 1959.

[^4]:    *Swift, Levick \& Sons, Ltd.

[^5]:    *R.E.M.E. Electronic Engineering School, Arborfield.

[^6]:    ${ }^{2}$ Proc. I.R.E., Nov. 1952, p. 1493.
    'See, for example, " Linear Feedback Analysis", by J. G. Thomason (Pergamon Press).

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