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## Line Sitandards

IN 1956 the Postmaster General of the day asked the Television Advisory Committee amongst other things to say whether the existing 405 -line standards were likely to remain adequate for all purposes for the next 25 years and if there was any reason why the United Kingdom should not adopt 625 lines for Bands IV and $V$ if it were recommended by the C.C.I.R. as the European standard.

To the first question the Committee in its 1960 Report ${ }^{\star}$ has given an unequivocal no, on the grounds that a "definite" improvement in picture quality is possible with 625 lines (assuming also an equivalent increase in channel width) and that with the trend towards increasing picture size the line structure will be less visible with 625 than with 405 lines. A direct answer to the second question is avoided and the response takes the form of an inversion and a recommendation that 625 lines should be adopted, not only for Bands IV and V but ultimately for Bands I and III also.

Although the evidence for a case against 625 lines is included in the Report it is scattered and un-co-ordinated, and before the growing compulsion towards a change of standard reaches the proportions of a general obsession, we think that the case for the retention of 405 lines should be made with at least equal emphasis. Briefly it is that the 405 -line standard is already capable of giving better picture quality and higher definition than is at present realizable on the viewer's screen, and that any improvement which might be obtained from 625 is marginal and not worth the 15 years or more of disruption, the cost of duplicating services and the dual standards receivers which would be necessary to effect a change. When finally Band I and III stations were converted to 625 lines and the last 405-line-only receiver had become obsolete we should finish up with nation-wide dual standards receivers in which the 405 section would be redundant, and for our trouble we should have a slightly better picture but at least one national programme less than we could have enjoyed if we had stuck to 405 lines and $5 \mathrm{Mc} / \mathrm{s}$ channel spacing.

How much better than 405 would the picture quality of 625 be in the home on the average commercial receiver? In our opinion, after seeing both British and Continental performance, not much. But if you do not accept our opinion read the report $\dagger$ of the Working Party of the Technical Sub-Committee of the T.A.C. which says (p.139) ". . . This is not very positive evidence for either the 405 -line or 625 -line pictures in Band V, but it indicates that the 625 -line pictures are not worse than their 405 -line counterparts and they are perhaps very slightly better." These tests were made with $7 \mathrm{Mc} / \mathrm{s}$ channel

[^0]and $5 \mathrm{Mc} / \mathrm{s}$ video bandwidth. With the proposed $8 \mathrm{Mc} / \mathrm{s}$ and $6.75 \mathrm{Mc} / \mathrm{s}$ respectively the difference in quality would undoubtedly be more "definite," but still, we think, marginal. Such differences as exist can be easily and much more economically accommodated by intelligent camera work and minor adjustments of viewing distance.
Visibility of lines on the larger screens is undoubtedly the strongest argument so far advanced against 405 lines, but the average assessment of observers in the field tests $\dagger$ is Grade 3 (definitely perceptible, but not disturbing) for 405 lines as against Grade 2 (just perceprible) for 625 lines. But there are less expensive ways of overcoming "lininess" than turning the whole broadcasting system into bedlam-at the receiver, for example, by the use of an elliptical scanning spot or "spot-wobble." Some people prefer to see the lines, taking them as an indication of a "sharp" picture, and have been known to switch off spot-wobble in sets in which it is provided. It is entirely a matter of personal opinion.
That a 405 -line and $5 \mathrm{Mc} / \mathrm{s}$ channel standard would put us in an invidious position in our international relations and upset European plans for the general adoption of 625 lines and $8 \mathrm{Mc} / \mathrm{s}$ channels is not necessarily true. It may be a necessary expedient on the Continent where national boundaries are contiguous, but on this side of the Channel we have the advantage of geographical isolation, particularly on Bands IV and V where, as the field test report confirms, propagation is more difficult and (p.15) "the limits of a Band V service area are somewhat more precisely defined than those of a Band I service." Mutual interference with other European stations is much more likely on Bands I and III. Why then should we have prematurely committed ourselves to $8 \mathrm{Mc} / \mathrm{s}$ spacing on the higher frequencies, while reserving the right to transmit for an indeterminate period with $5 \mathrm{Mc} / \mathrm{s}$ spacing on the lower and more easily propagating bands? It would have been much better to forgo "tidymindedness" and to take advantage of our insular position to make the best use of the available bandwidth in providing more alternative programmes, at the same time safeguarding our neighbours' interests by careful attention to aerial directivity in stations near the south and east coasts.

Summarizing the case against a change to 625 we can say that (i) the value to the viewer in terms of picture quality would be negligible; (ii) there are cheaper and easier ways of overcoming "lininess" if it is thought to be objectionable; (iii) a third programme with national coverage would no longer be possible in Band III; (iv) if Bands IV and V were eventually brought into service, more programmes could be established earlier and at less cost with 405 than with 625 lines.

# Instruments, Electronics and Automation 

THE I.E.A. EXHIBITION AS SEEN BY THE TECHNICAL STAFF OF WIRELESS WORLD


#### Abstract

AN ELECTRONICS exhibition is probably the only place where one would find accessories so remotely connected as television aerials and hydraulic valves. On this occasion the great diversity of the show at Olympia was partly due to the inclusion of many radio and electronic component manufacturers (there being no R.E.C.M.F. exhibition this year). Such variety inevitably meant a much larger exhibition than the one held in 1958, and both the Grand and National Halls were occupied. Technical interest, too, was greater-particularly in the field of automatic control, where there were more complete systems to be seen in addition to the isolated bits of apparatus used in these systems.


## INDUSTRIAL ELECTRONICS

Process Control Systems.-Most of the automatic control equipments on view were intended for continuousprocess operations, as, for example, the continuous adjustment of properties of liquids flowing in pipes. This type of application allows the general principle of the servomechanism to be used. A monitoring transducer measures the required property of the material concerned and the measured value is compared with the required value to produce an error signal, which is used to actuate a control device to correct any deviation. An example of this straightforward technique was a temperature controller for electric furnaces, shown, rather appropriately, by the Phoenix Telephone and Electric Works. The input signal is derived from a thermocouple and from a reference signal given by a potential divider connected across a Zener diode voltage source. The difference output from the thermocouple and reference source is amplified by magnetic amplifiers and used
to control a saturable reactor which in turn controls the supply of electrical power to the furnace. The feature of this system is that it avoids the use of mechanical contacts which have to be periodically serviced.

In such systems the control signal is usually directly proportional to the amplitude of the error signal. There are certain processes, however, which require the control signal to be a more complex function of the error signal. For example, in some plants a measured deviation in the form of a sudden step might, if fed back as a correcting signal, cause "hunting" in the process control system. It might be necessary, therefore, to integrate the sudden step into the form of a slowly rising control voltage. In many of the controllers on view there were, in fact, facilities for providing these special functionsusually integration and differentiation 'to give a control signal depending on the rate-of-change of the error signal). These, in addition to the normal proportional control, give
what are commonly known as "threeterm" controllers. As an example, Evershed demonstrated a three-term controller being used to control automatically the concentration of mixtures of liquids on the basis of conductivity measurements. The measuring transducer produces an electrical signal proportional to conductivity, and a resistance network gives the difference between this value and the desired value (set manually by a potentiometer from a voltage source). The difference signal then passes through an amplifier system in which its amplitude can be manually set and its waveform modified by adjustable integrating and differentiating circuits - to emerge as a signal which controls the rate of operation of one of the pumps contributing to the mixture of liquids.

An integrating circuit was used for giving a delay effect on the control signal in an interesting equipment shown by Haynes \& Haynes on the Lancashire Dynamo stand. This was for controlling the wall thickness of plastic tube during extrusion. Variations of wall thickness are used to regulate the speed of the motor which drives the "haul off" conveyor system for the plastic tube. The measuring transducer is a capacitor with its two plates not parallel but inclined to each other so that a section of the plastic tube can pass between them. The tube wall, therefore, provides part of the capacitor's dielectric and any variations in wall thickness alter the capacitance. This transducer is connected in series with another capacitor of equal value to a 3-Mc/s oscillator, and an output signal is taken from between the two


Phoenix Telephane
temperature control equipment.

De Havilland " Anatrol " analogue computer for process control.
capacitors. The result is a bridge arrangement in which any unbalance between the two capacitors (due to wall thickness variation) is indicated by the amplitude of the output signal, while the direction of unbalance (increase or decrease of thickness) is indicated by the phase of the output signal relative to a reference signal from the oscillator. These changes are detected and passed to the integrating circuit, the output of which is used to control the variable-speed drive of the plastic tube conveyer. Data Processing.-One could not go very far in the exhibition without seeing some equipment or other for the transmission or conversion of in-formation-for telemetering, com"unications, computation, or the "logging" of data from transducers in industrial plants. A particularly impressive example was a large equipment shown by Bristol Aircraft for converting tape recordings, obtained from the receiving end of a 24-channel time-multiplex telemetering system, into the form of data on punched cards. By this means the weeks of work normally required for analysing the telemetered information could be reduced to a single day. An electronic analogue-to-digital converter, also shown by Bristol, had the unusual feature of giving a digital output which was corrected for any non-linearity, drift or gain variations in the analogue transducer


Texas Instruments solid-circult binary adder, compared in size with a safety match.
system providing the input. This correction is achieved by supplying to the converter, along with the analogue input, reference voltages which are a calibration of the analogue signal and are subject to the same unknown variations. The converter works on the well-established principle of comparing the analogue input with a succession of fixed voltages; when equality is reached the action stops and a binary counter registers the number of comparison steps which have been taken. In the Bristol Aircraft system these fixed voltages are provided by the reference voltages mentioned above, so that the unwanted variations in the transducer system are automatically compensated.

Another type of error correction was a feature of many of the telegraph and digital data transmission


Evershed equipment with .three-term controller for controlling mixture of liquids.
systems on view. In general, the method consists of transmitting extra digits along with the normal information. These give a special pattern to each of the characters or numbers transmitted so that any mutilations of the signal in transit can be automatically recognized at the receiving end. Great interest was attracted by a G.P.O. demonstration of this principle in which the number of errors detected and the number of characters transmitted were "clocked-up" on counters. In general such correction techniques can reduce the normal error rate of about 1 in 40,000 characters to about 1 in $10^{6}$ characters.

Turning to the computation side of data processing, one of the most interesting exhibits relating to the design of computers was a working binary adder constructed from semiconductor solid-circuits. Demonstrated by Texas Instruments, it consisted of a group of tiny flat plates of silicon, measuring $\frac{1}{4}$ in by $\frac{1}{6}$ in, each of which was an integrated circuit element formed by diffusion, etching and deposition techniques. Four types of solid circuit elements were used: (1) voltage inverters, each consisting of a transistor, a diode and resistors, (2) diode gates for AND and OR operations, (3) a bi-stable circuit, containing transistors, diodes, resistors and capacitors, for delay purposes, and (4) a diode gate to provide the correct drive for the bistable circuit. Interconnections were made with the aid of printed conductors and the whole adder was well spread out for display purposes. In practice, however, the individual circuit elements are packed together face to face into a tiny cube-the incredibly small size of which can be

seen from the illustration on this page. The saving in volume over an equivalent adder using conventional semiconductor devices is of the order of $100: 1$, according to Texas.

Semiconductor circuit elements are now the standard thing for all new digital computing systems, and one particularly interesting example was a digital machine shown by Elliott-Automation, designed for incorporation in process control loops in industrial plant. The machine is called the "Optimat," because it is not a straightforward computer but, a device for seeking the optimum performance point within a specified regime of operation of the plant. It does this by making trial-and-error incremental variations in the control signals to the plant until the plant conditions meet the specification (which is laid down in the programme to the machine). In this way optimum performance can be obtained even against the influence of uncontrolled parameters in the operation of the plant. The logical elements, established already under the name "Minilog," are transistor and diode circuits mounted on printed circuit cards which in turn are wired in groups on to larger boards carrying plugs for insertion into a chassis.

For simpler process control applications the analogue computer is particularly suitable, since the control and monitoring signals to and from the plant are necessarily analogue signals. An example at the exhibition was the "Anatrol" anologue computer, developed by de Havilland, which was shown as a means of solving equations necessary to keep the composition of a blended product at some specified value. As


Electromethods Series 5100 miniature chart recorder, showing 2 in diameter chart.
in the "Optimat," the circuitry was based on semiconductor devices.

An analogue computer with some unusual facilities was the "Simlac Minor" shown by Short Brothers \& Harland. It has a push-button selection system which enables the d.c. amplifiers and passive circuit elements to be selected and connected to various measuring instruments and also allows the coefficient potentiometers to be automatically set up, by servo control, to an accuracy of $0.1 \%$. A novel "patch-cord" system gives flexibility of interconnecrion between the units while using the minimum number of cords and avoiding the need for the cords to cross. The computer is actually a small general-purpose machine and uses 32 d.c. amplifiers.
Information Recovery from difficult-to-get-at places is common in these days of "sputniks"; however, many examples are far nearer at hand than an orbiting artificial satellite-the inside of a continuous paint-stoving oven, for instance, or the face of a piston in a running internal-combustion engine.

Shown on the D.S.I.R. stand was a small telemetry transmitter designed to fit inside the piston of an engine. Developed by the British Internal Combustion Engine Research Association, recourse to telemetry was made because the number of slip rings and wiping contacts required could have resulted in false signals and would have been difficult to maintain. The single-transistor transmitter operates at about $2 \mathrm{Mc} / \mathrm{s}$, feeding an aerial which projects into the crankcase. A thermistor sensing element on the face of the piston modulates directly the frequency of the transmitter. To withstand the high temperatures built up, and the oil present, the unit is encapsulated. A silicon transistor is used.

Another approach was illustrated by Electromethods, with their series 5100 recorder. This a miniature ( $6 \frac{1}{2} \times 4 \frac{3}{4} \times 3 \frac{1}{2} \mathrm{in}$ ) instrument using a clockwork-driven chart of pressure sensitive paper only two inches in diameter. A self-balancing bridge


Ericsson "Rotapulse" transducer for automatic weighing.
system is used, balance being achieved between the input and a slide wire whose moving contact is coupled to the stylus marking the chart. Periodically the input signal is removed automatically and a second servo system re-balances the zero of the bridge.

For applications such as monitoring the temperature throughout a stoving oven, the whole recorder, complete with its battery pack, can be put inside a vacuum flask with leads connected to an external resistance thermometer. When used for potential recording the sensitivity is such that signals of 10 mV or so may be registered. For use with a resistance sensing element, accuracy figures quoted were $\pm 2 \%$ of f.s.d. and variation of zero and sensitivity $\pm 0.25 \%$.
Weighing Automatically materials into bins or sacks is a typical example of how automation can reduce the physical labour of a task, and at the same time cut down clerical work with its opportunities for error. The adaptation of an ordinary weighing machine, too, brings in a common problem-the sensing of rotation without the imposition of a significant load on the apparatus. The general principle is that the rotation of the scale pointer is sensed and converted to pulses, which are counted and registered on a reversible decimal stepping display. Preset circuits allow the rate of delivery of the material to be reduced as the correct weight is approached and cut off when the weight is correct. A converter may be fitted, for instance, to operate a teleprinter which would record the weight.

Microcell, in their Type 171 machine, use a radially striped disc coupled to the pointer spindle and another stationary grating. Light shining through the gratings is thus modulated by the passage of the rotating grating, and picked up by photocells. To allow for other than critical damping of the weighing machine, the sense of rotation must be discovered so that the counter can be "instructed" whether to add on or take off pulses. This is done by using two photocells with $90^{\circ}$ angular separation.

The Rotapulse transducer, shown by Ericsson, is also primarily an
optical system: it is offered as an entirely separate unit with lowfriction bearings (less than 0.050 / in torque is required). Four quadrantally positioned phototransistors are used, and with the 250 and 249 seg ments on the gratings 1,000 pulses are given for each revolution of the spindle. The pulse-forming circuits are transistorized and are built on a flexible printed-wiring board which is wrapped round the rotating mechanism.

A selsyn transmitter is coupled to the weighing-machine pointer in the E.M.I. "Emiway" equipment. This transmits the rotary motion to the remote equipment where analogue to digital conversion is achieved by a coded disc driven by the synchronous receiver motor. The control unit provides for the checking of the machine zero, two reduced rates of feed and the starting of the print-out operation. The accuracy achieved is better than one part in 2000 in $360^{\circ}$, or $0.1 \%$ of f.s.d.
Industrial Television.-When direct viewing through windows or by means of large mirrors is impracticable, stereoscopic television may well provide a solution, for instance, when handling radioactive substances behind screening. To give a true solid image two pictures must be presented, one to each eye, from the positions and angles of view that would be occupied by the eyes if they were observing the scene to be transmitted.
E.M.I. were showing a stereo TV system using two small cameras with automatic adjustment of the angle subtended to the subject by the eyes, achieved by coupling to the optical focus control. Display was on two c.r.ts, the images being combined by a half-silvered mirror and the tubes were covered with orthogan-ally-set sheets of optical polarizing medium. The wearing of similarly polarized spectacles separated the pictures at the eyes.

Pye, however, were showing stereo TV with only one camera, amplifying chain and c.r.t. This was achieved by the use of mirrors to produce the left and right views on halves of the


Stereoscopic television camera showing optical system for producing left- and right-eye views (Pye).
sensitive area on the camera tube. The monitor thus displays, side by side the two images, which are directed into the appropriate eyes by an optical system similar to that used on the camera. Of course, the solid picture produced is of only half the area of a "flat" picture, but this is of no importance for industrial purposes.
E.M.I. were demonstrating their new colour camera which uses three vidicon tubes and a novel optical system. The optical system utilizes supplementary lenses in the lens turret to give varying fields of view. Inside the camera the light is split into its red, green and blue components by mirrors and filters, and then individually focused on to the tubes by separate lenses. The smearing of moving objects so often seen with vidicon pick-up tubes has been reduced considerably by the use of new short-lag tubes.

The camera is designed to operate on 625,525 and 405 line systems. This was a general trend noted throughout the exhibition; for instance, Thorn were showing a picture monitor and waveform generator capable of being switched between these standards and Epsylon, in their range of industrial television equipment also cater for quick changing from one standard to another. The Epsylon equipment incorporates a picture monitoring circuit which adjusts automatically the potential fed to the camera tube for a $50: 1$ variation in light intensity.
Unit Construction seems to be the main trend in industrial control and automation equipment. The aim of this, of course, is to have available a number of standard blocks, such as amplifiers, counters, timers and power units, which can be fitted together to suit practically any requirement. This must reduce the "electronic" interest of the system, but it must be remembered that the prime requirement for industrial equipment is that it should, in the event of a fault, remain unserviceable for the minimum possible time. The best method of fulfilling this require"ent is the use of standard plug-in ""blocks," so that only a small number of spares need be kept. This approach was exemplified by, for instance, Mullard, with their Combielements and Norbits, Fox Yarborough and Lancashire Dynamo, who were also exhibiting a unit rack assembly developed for naval use. This uses a novel cooling method. Instead of the usual extractor fan and inlet filter drawing air through the actual apparatus, cooling is achieved by mounting heat-producing components on a spring-loaded metal plate. When the unit is forced fully home this plate is pressed into intimate contact with the machined walls of the closed "cell" in which the unit is held. The cell walls are of extruded light alloy, carrying internal finned ducts through which
air is drawn. In this way efficient air-cooling is achieved without the fire danger caused by forced-draught cooling of racks of open equipment.

Power-supply units have become more or less standard in their design, but one novelty seen on the stand of International Electronics was a tran-sistor-stabilized $300-\mathrm{V} \quad 500-\mathrm{mA}$ unit. Normally transistors are not used for h.t. stabilization because of the danger of damage should the supply
be short circuited. However, by designing the error-amplifier so that, in the event of a short circuit, the outpur transistors are switched hard on, and by including a currentlimiting resistor in the unstabilized side of the supply, the dissipation in the transistors can be kept low enough to ensure safety while the output fuse blows. Naturally, the use of transistors achieves a considerable saving in weight and bulk.

## MEASURING INSTRUMENTS

Sine Wave Oscillators.-In a transistorized oscillator shown by R.R.E., constant output (within $\pm 0.1 \mathrm{~dB}$ ), low distortion ( $<0.1 \%$ ) and a wide frequency range ( 10 to 1) are obtained by means of a basically simple circuit (see diagram) in which a single linear potentiometer controls the oscillation frequency. This potentiometer is connected between two simple CR leading and lagging phase-shift networks


Basic circuit of new R.R.E. oscillator.
so that the phase of the amplifier input can be varied. Oscillations are produced at the frequency at which the input phase is the same as the positive feedback output "phase. "(In the circuit diagram the "upper" resistor and capacitor should have a higher impedance than the corresponding "lower.") Three transistors are used in the amplifier circuit. If a linear potentiometer is used, the oscillation frequency varies nearly logarithmically with the spindle angle except near the beginning and end of the potentiometer element. A linear potentiometer also gives a more uniform scale graduation spacing than is usually obtained with a logarithmic potentiometer. This is because a logarithmic potentiometer normally approximates to the ideal logarithmic curve in three straight line segments, so that at the two joins of these segments a sudden change in the scale graduation spacing is produced.

An unusual method of sine wave generation is used in the Marconi

TF1382 low-frequency (down to $0.0033 \mathrm{c} / \mathrm{s}$ ) sine, square and ramp waveform generator. The latter (ramp) waveform is that basically produced in this generator. This waveform is then shaped in a Zener diode circuit to produce sine waves with a distortion of less than $5 \%$. Square waves are also produced from the triangular waveform via a bistable circuit.

A tone-burst generator was shown by Ferguson. When fed with a sine-wave input this produces sine waves in bursts whose individual duration and repetition rate can be varied, each burst containing an integral number of sine waves. Such a generator is useful for testing audio amplifiers, since its output provides a simple approximation to the high peak-to-mean power ratio conditions of music and speech. When an amplifier is fed with such bursts its output-stage operating point corresponds more closely with that obtained in practical use than does the operating point obtained with a highpower continuous sine-wave input. Marginal instability is also easier to detect when the high-power signal is cut off at the end of each burst. To obtain such bursts the sine-wave input is gated on and off, the gating being controlled from the sine-wave input so that only an integral number of sine waves is produced in each burst.
Oscilloscopes.-Transistorized instruments were shown by Microcell and Tektronix ${ }^{(L)}$. An unusual feature of the latter instrument is that the flyback trace is suppressed by feeding a suitable signal to a separate deflection plate system in the c.r.t. so that these plates intercept the beam. Intensity modulation of the beam is also possible.

A cathode-ray tube containing a set of deflection plates forming the capacitive elements in a lumped L-C line (travelling-wave deflection) is

Agents in U.K. for foreign instruments: (C L), Claude Lyons Ltd., Valley Works, Hoddesdon. Herts; ( $\mathbf{G} \& \mathbf{G}$ ), Griffin $\&$ George (Sales), Ltd, Ealing Road, Alperton, Wembley, Middx.; (L), Livingston Laboratories, Ltd., Retcar Street, London, N.19; (N\&'T). Nash \& Thompson Ltd., Hook Rise, Tolworth, Surbiton, Surrey; ( RHC ), R. H. Cole (Overseas), Ltd., 2 Caxton Street, London, S.W.I.
used to obtain a response up to as high as $100 \mathrm{Mc} / \mathrm{s}$ in the Tektronix ${ }^{\text {(L) }}$ Type 585.

A very small ( $5 \mathrm{in} \times 3 \frac{3}{4} \mathrm{in} \times 6 \frac{1}{4} \mathrm{in}$ ) oscilloscope was shown by Sciaky Electric Welding Machines. A single EF91 is used for the time base and a second EF91 as the d.c. Y-amplifier. The response extends to $300 \mathrm{kc} / \mathrm{s}$ at a sensitivity of $700 \mathrm{mV} /$ cm.

An unusual feature of the oscilloscopes shown by the East-German VEB Funkwerk Kopenick ( $G$ \& $G$ ) is that no internal Y-amplifiers are provided, separate units being available for this purpose.

Switching between as many as five inputs is an unusual facility possible with the Czech Krizik (N \& T) K552 oscilloscope. The inputs are switched on and off in turn either by a $100 \mathrm{kc} / \mathrm{s}$ multivibrator, or alternatively, for viewing high-frequency waveforms, in synchronism with the timebase.

The frequency range of an oscilloscope for recurrent waveforms can effectively be greatly extended by using very fast rise-time pulses to sample various portions of a recurrent waveform, the sampled pulses being stored and ampliffed in the relatively much larger time between the taking of successive samples. This method was described more fully in our review of the 1959 Physical Society Exhibition (March 1959 issue, p. 131), with reference to an oscilloscope shown by the U.K. Atomic Energy Authority. It was also used in oscilloscopes or attachments for existing oscilloscopes shown this year by Lion Electronic Developments, Hewlett-Packard ${ }^{(L)}$, Tektronix(L), and Lumatron(L)' Usually the high-speed sampling pulse is obtained from avalancheoperated transistors and a rise time of about $0.6 \mathrm{~m} \mu \mathrm{sec}$ is obtained. In the Lumatron ${ }^{(L)}$ Model 12 oscilloscope, however, a dynode secondaryemission valve is used to provide the fast-rise time sampling pulses. The pulses from the dynode are clipped and differentiated to provide a final pulse rise time of only $0.4 \mathrm{~m} \mu \mathrm{sec}$.
Transistor Testers.-An unusual meauring meter for such instru-ments-the quadrant electrometeris used in the French A.O.I.P. Tran-


Sciaky miniature oscilloscope.
sistormeter. Input and output resistances, leakage currents and gains can be measured by this instrument. All measurements are referred to measurements of collector current changes. These are carried out by first charging both pairs of quadrants to a voltage proportional to the collector current. One pair of quadrants is then kept at this potential by a high-insulation capacitor, while the other pair is brought to a new potential proportional to the changed collector current. The electrometer deflection is then proportional to the change in collector current. One advantage of this system of measurement is that the initial conditions are stored so that any change can be accurately measured after an interval of time.

Features of the new Microcell Type 156 are the use of a widerange ( $1 \mathrm{kc} / \mathrm{s}-10 \mathrm{Mc} / \mathrm{s}$ ) oscillator as signal source, and a differential valve-voltmeter as indicator. This type of indicator eliminates errors due to spurious voltages developed across the resistive networks. Cutoff frequencies, amplification factors, leakage currents and turnover voltages can be measured with this instrument.

An instrument for service depart-ments-the Type 105 C -was shown by Labgear. This can measure current gains and collector leakage currents and turnover voltages.

Collector voltage/collector current characteristics at up to five different base or emitter input currents can be simultaneously presented visually on the Dobbie McInnes Transigraph TG104 oscilloscope. Current gains, output impedances and optimum operating conditions can then be readily determined.
Grid-dip Meters were shown by Grundig Instruments and the Czech firm Kovo(N \& T) (Tesla BM342). These consist simply of a calibrated oscillator whose grid current can be measured. They can then be used as an absorbtion resonance-frequency indicator (minimum grid current) or as a signal source. Alternatively, the oscillator valve can be connected as a diode and the instrument used as an absorption wavemeter (maximum current).
Frequency Response curve tracers usually display simply the output variation so that inaccuracies and complications are produced if the input source varies. However, in the Siemens $(\mathbf{R H C})$ ratio tracing receiver Type Rel 3K217c the input is used to alter the gain of the output amplifier so that input variations of up to 10 dB are compensated for to within 0.3 dB . The amplified output traced on the c.r.t. screen is then proportional (within 0.3 dB ) to the response.
Voltage Measurement.-A precision ( $0.05 \%$ accuracy) r.m.s. decade volt-meter-the D-930-A - was shown by Muirhead. In this instrument the


Kovo Krizik (N \& T) five-channel oscilloscope.
unknown voltage is fed via the range switch to an a.c. amplifier followed by a decade attenuator and second a.c. amplifier. The output of this second a.c. amplifier is fed to a Wheatstone resistor bridge in which a lamp forms one arm. Since the resistance of this lamp depends on the electrical heating power developed in it, the bridge balance is determined by the r.m.s. value of the input. The point at which the bridge balances is first standardized against the direct voltage from three Weston reference cells. This voltage is then reduced by about $80 \%$, and the bridge rebalanced by adding an internally supplied a.c. voltage. This standardizes the a.c. voltage. This voltage is then attenuated and fed to the input of the voltmeter to standardize the a.c. amplifier gains. In the Marconi TF1377 suppressedzero voltmeter potentials are measured by balancing them off against the output of a potentiometer fed from a standard voltage-the potentiometer setting being shown on a three-digit indicator. Residual unbalance voltages and thus voltage changes can be measured by means of a differential valve voltmeter.
Both the amplitude and phase of a voltage can be measured by a simple device shown by F. C. Robinson and Partners. In this instrument current is fed from the mains, or from any $15 \mathrm{~W}, 15-\Omega$ output impedance amplifier, through a stảndard potentiometer $P_{1}$ in series with the primary of a mutual inductance. Across the secondary of this inductance is connected a second potentiometer $P_{2}$. The voltages developed across the two potentiometers $P_{1}$ and $P_{2}$ are then 90 degrees out of phase. The phase and amplitude of an unknown voltage can then be measured by balancing this voltage off against the outputs of the potentiometers $P_{1}$ and $P_{2}$ 。

Wires carrying alternating current can be detected without having to make any contact with them by picking up the electrostatic field produced by the current in the Everett Edgcumbe "Metrac" live
 pair using electrometer valves) followed by a pair of cathode-follower outputs. In use, one of the two inputs is cross-connected to one of the outputs, a different input being connected according to whether voltage or current is being measured. If, in addition, a standard resistor is connected between the other input and output, the impedance between the two inputs is then very low (voltage drop $<5 \mathrm{mV}$ ) so that direct currents are conveniently measured between the two inputs. On the other hand, with the standard resistor disconnected, the impedance between the unconnected input and output is extremely high so that direct voltages are conveniently measured between this input and output. Input voltages can also be stored on a $0.15 \mu \mathrm{~F}$ polystyrene capacitor (producing an input impedance of $10^{10} \Omega$ ) so that a reading can be taken after the probe has been removed from the test point. Resistances can also be measured with this instrument. Current Measurement.-An unusual photo-electronic chopper technique is used in the prototype Nanoammeter shown by the French firm A.O.I.P. Here the unknown d.c. input is passed through a photoresistor on which a light is shone. This light is periodically interrupted by a mechanical movement so that the photo-resistance is periodically altered. The d.c. input is thus partially converted into a.c. which is then amplified and measured.
Direct currents can be measured down to 3 mA full scale without interrupting the circuit under test in the Hewlett-Packard Model 428A
clip-on d.c. milliammeter. In this instrument the magnetic field produced by the current in a probe of magnetic material which is clipped round the current-carrying conductor is measured by a fluxgate technique. In this technique an a.c. signal passed through a coil wound on the probe drives this probe into magnetic saturation on alternate half cycles. The additional field due to the current makes the magnetization curve of the probe slightly asymmetrical with respect to zero field and induction. This results in a second harmonic output being produced in a second coil wound on the probe. This second harmonic output is detected and measured. Interfering effects produced by uniform direct fields (such as the earth's field) can be arranged to cancel out by using the fact that the fields on opposite sides of a current-carrying conductor are in opposite directions. The probe is also magnetically shielded against external a.c. or d.c. fields. An advantage of this method of measuring current is that no resistance and very little inductance ( $<0.5 \mu \mathrm{H}$ ) are introduced into the circuit being measured.
Power Measurement.-In the Burndept BE281 powers can be measured at frequencies up to $1,000 \mathrm{Mc} / \mathrm{s}$ by feeding them to a $50 \Omega$ coaxial resistor (made up of a carbon film on a cylindrical ceramic core) which is mounted along the axis of exponential cavity to provide broad-band matching. The r.f. voltage developed across a portion of the resistor is rectified and measured. This purely
electrical method of measurement gives a much faster response than is obtained when r.f. is measured by using it to produce a heating effect (as in a bolometer).
L, C, R Measurement.-When highloss reactive components are measured using bridges, the measurements of the reactive and resistive parts of the component usually influence each other so much that many adjustments are needed to obtain a balance point. Two general methods of considerably reducing the number of such adjustments required were seen. In capacitor bridges shown by Rank Cintel and Winston Electronics phase-sensitive detectors are employed so that the resistive and relative out-of-balance voltages (which differ in phase by $90^{\circ}$ ) can be separately detected and zeroed. In the General Radio (C L) Type 1650-A L, C, R bridge, the resistive and reactive adjustments are mechanically connected by friction clutches such that, when the resistive adjustment is altered. the reactive adjustment is also altered so as to keep the ratio of the resistive and reactive parts constant; but, when the reactive adjustment is altered, the resistive adjustment is not affected.
In capacitance meters shown by E. C. Robinson \& Partners and the Czech firm kovo ${ }^{\text {N \& }}$ T) (Tesla Model BM271) measurements are made by placing the unknown capacity in a resonant circuit which also contains a calibrated variable capacity. By keeping the resonant frequency fixed by adjusting the calibrated variable capacity, the effect of the unknown
capacity can be measured. In the Tesla BM271 the resonant circuit is placed across the output of a fixedfrequency oscillator and adjusted to resonate at the oscillator frequency by maximizing the voltage developed across the circuit. By also adjusting this voltage to a fixed value by means of a variable resistor placed across the unknown capacitor, the loss of the capacitor can be similarly measured. This variable resistor consists of a thermionic diode with a by-passed variable resistor in series with it; this arrangement providing a resistance which is sufficiently non-inductive for use up to $30 \mathrm{Mc} / \mathrm{s}$. In the F. C. Robinson and Partners Picofarad Meter the resonant circuit forms part of an oscillator and the resonance is adjusted to a fixed frequency by listening to the audio beats produced with another fixed-frequency oscillator in a loudspeaker. Since the oscillator frequencies are at $1.5 \mathrm{Mc} / \mathrm{s}$, such beats produce a very exact indication of correct frequency adjustment.
A very simple method of measur-

## PARTS AND MATERIALS

Strip Wiring is formed from flat copper strips a few thousandths of an inch thick, supported and insulated from each other and external contact by plastics films.

Technograph have been making flexible "printed" wiring for several years, including, for instance, resistive elements on a rubber compound (for the de-icing of aircraft control surfaces) and strain gauges on various films such as Terylene. Their latest strip-wiring cable consists of copper conductors sandwiched between two layers of 0.001 -in Melinex film. The Melinex softens sufficiently to be stripped from the copper at about $160^{\circ} \mathrm{C}$, so leaving free the ends of the conductors, for connection. T:C.C. introduced a flat-strip cable, called Flexistrip, at the exhibition: in their cable the copper strips are moulded into polythene, and an overall jacket of Melinex is then applied. Possibly the most startling thing about strip cables is their flexibility-Flexistrip, for instance, passes the DEF5000 test, which involves no fewer than 20,000 flexings, and is normally applied to tinsel-braid cables.

Connectors for strip-wiring could take practically any form; for instance, it could be soldered directly to a printed-wiring board, or be fitted with eyelets. However, any method requiring individual handling of each wire to achieve a disconnection is not likely to be acceptable where many connections are required. Thus several connectors have made their appearance, each quite different in principle.

Belling and Lee were showing an experimental moulded housing to fit on to either circuit-board connectors


Robinson \& Partners Picofarad Meter.
ing inductors and capacitors used in radio or television receivers is used in the East-German V.E.B. Werk für Fernmeldewesen ( G \& $G$ ) LCM1. An oscillator with a cathode-follower output stage provides an alternating current which is passed through the component. The voltage developed across the component is then a sufficiently accurate indication of the component reactance, since the resistive loss in components used in radio or television receivers is generally small.
or a small piece of board, so forming a socket and plug which can be attached to the cable. The housing itself has three slots through which the cable is passed, to form a cable grip. The ends of the strip are pierced and soldered onto lugs projecting from the socket or plug board and the housing, which has a snapon cover, can incorporate a fingerrelease locking clip.
N.S.F. have a modified form of their Varicon interconnection tags, which, when mounted on a printed-wiring board, form both "plug" and "socket" contacts. The modified tag carries on its rear a toothed portion under which the end of the unstripped cable is trapped. On mounting the tags, the teeth cut through the insulation and enter the conductor, so making contact. Slots punched in the board could provide a cable grip. For quick assembly Varicon can be supplied in strips of thermoplastic material, which, after fixing the tags to the board, is warmed and removed.

Continental Connectors, a division of Ultra, were showing a fitting like an ordinary circuit-board socket. However, it accepts only a very thin section-the end of a flexible strip cable, in fact, with the conductors cleaned of insulation and folded back over the cable. For very flexible cables which would not be rigid enough for direct plugging in, a thin piece of board could be used as a supporting memiber, with the stripped strip conductors folded over the edge of the board. Again, slots in the board could provide for a cable grip and retaining mechanism.
Connection to Printed-wiring boards
is made without the use of solder by a new technique shown by Belling and Lee. Known as Prestincert, it depends upon the discovery that a disc or peg can, with the aid of a die and press, be punched into sheets of insulating material or metal without first making a hole for it. Knurling the edges of the insert prevents rotation, and the die can peen over the penetrating end, so forming a strong fixing. For component connection the peg is formed into a lug with a soldering terminal at the top and an oversize collar that seats firmly on the board. A slot, cut diagonally into the peg across a diameter, takes the component lead: on punching, the lead is squeezed into intimate contact with the board and the insert, the excess wire being sheared off. Also insulation on the wire is stripped automatically.
Printed Power resistors, shown by Technograph, consist of a meandered resistive track on an insulating coating on a metal panel. This may then be fixed to another metal plate, possibly the chassis of a piece of apparatus, for dissipation of heat at ratings up to $10 \mathrm{~W} / \mathrm{in}^{2}$. The resistance values and ratings on one of the panels displayed suggested that it was intended for use as the mainsdropping resistors in a television receiver, so we arrive at the paradox of having to put back the chassis to act as a cooling fin for the contactcooled h.t. rectifier and mains dropper, after eliminating it by the use of printed-wiring panels!
Capacitor Construction has for several years remained largely unchanged, except for the entry to the field of plastics-film dielectrics. However, a development shown by Dubilier may well challenge the ubiquitous wax and paper capacitors. The Dubilier "Blue-cap" employs a paper dielectric, but instead of wax, a synthetic-resin impregnant is used. The absence of wax or oil in the manufacturing process makes possible hermetic sealing in a plastics


Flat-strip cable, by Technograph.
sleeve without danger of moisture penetration at the lead-in wires. The sleeve material has a high melting point and is designed to withstand any normal soldering operations.

Another novel construction was shown by Hunts, in their WF49 " Duolectric" capacitors. These are
housed in aluminium cans and occupy roughly a third of the volume of the equivalent rating of waxedpaper types. To achieve this reduction polyester film has been used as the dielectric; but to avoid the relatively high cost of metallized film a sandwich construction of plain film with metallized paper electrodes has been employed. The largest Duolectric capacitor $(2 \mu \mathrm{~F}, 1 \mathrm{kV})$ is only $3 \times 1 \frac{1}{4} \mathrm{in}$.
Work on tantalum electrolytic capacitors has resulted in the elimination of liquid sulphuric acid as the electrolyte: the effects of a burst or leaking capacitor containing this can only too easily be imagined. Hunts have developed a chemically inert-electrolyte for use in slug-type capacitors, which are, for extra safety, contained in two cans, one inside the other. Dubilier have a range using a solid layer of semiconductor material as the "electrolyte" and a sprayed-zinc coating is used for the second electrode. Other types (Dubilier and Hunts) utilize a construction similar to that of the common "dry" aluminium electrolytic capacitor.
Push-button TV Tuners have many advantages in simplicity of operation; for instance, it is not necessary, as it is with some rotary types, to clank through several unused channels, possibly moving inadvertently the fine tuner as well. A.B. Metal Products had on show a new fourchannel push-button unit using a frame-grid cascode triode and triode pentode in the common circuit arrangement. In place of the rotary turret, however, was a push-button mechanism for selecting two Band-I and two Band-III channels. When a button is depressed it allows the appropriate coil strip to rise under spring pressure and engage with several double-leaf contacts projecting from "busbars" joined to the rest of the circuit. A feature of the new tuner is that separate preset fine-tuning controls for each channel

are put into circuit by an extra contact. on the coil strip.
Wire-less Transformers or, more correctly, piezoelectric transformers, were shown by Brush. These depend for their operation on the mechanical excitation of ceramics such as lead titanate zirconate either by another section of the same ceramic or by the magneto-strictive effect in a ferrite carrying a winding. Demonstrated was one of bar form, used to light a small neon sign needing about 1 mA at 2 kV . The lowpotential part of the bar was excited by 10 V applied from an oscillator connected across its "thickness." Here the impedance is relatively low, but by polarizing the other half of the $\lambda / 2$-long bar along its length it can be made to oscillate in the lengthways mode, which corresponds to a high impedance between the ends of the $\lambda / 4$ section. Thus, by attaching wires to the ends of this bar a high potential at low current may be extracted.

The efficiency of the ceramic-toceramic transformer shown was of the order of $50 \%$; but, by driving, the lengthways-mode "crystal" from a magneto-strictive transducer cemented to it, greater power can be fed in to the bar, with a consequent increase in both efficiency and output, which may be made as high as 40 kV .
Semiconductor Devices.-High-frequency transistors at economical prices is the aim of the alloy diffusion method of manufacture developed by Mullard, and a whole range of new p-n-p types based on this principle was on view. Briefly, the technique uses a wafer of p-type germanium to form the collector, and on one face of this two metal pellets are placed side by side to form the emitter and base. During heat treatment n-type impurities diffuse into


Above: Four-channel pushbutton TV tuner (A.B. Metal Products). Fifth button actuates on/off switch and six-button version can switch in separate f.m. tuner.

Left: Demonstration circuit board with components connected by the Belling-Lee "Prestincert" principle.
the germanium wafer from both pellets to produce an extremely thin base layer between the emitter and base electrodes. At the same time, p-type impurities diffuse relatively slowly out of the emitter pellet only and produce a small p-type layer around this pellet, confined within the n-type base layer. The highfrequency properties are obtained mainly as a result of the extremely thin base layer (about 5 microns) and partly because the graded distribution of impurities gives an accelerating field which reduces the transit time of the current carriers through the base even further.
In addition to the established OC170 and OC171 made by this technique, Mullard showed two lownoise transistors for v.h.f. communications which give power gains of 10 dB at frequencies of $100 \mathrm{Mc} / \mathrm{s}$ and $200 \mathrm{Mc} / \mathrm{s}$ respectively. There was also a switching transistor for use in computers operating with p.r.fs up to $10 \mathrm{Mc} / \mathrm{s}$, and a type for driving ferrite core stores which was capable of producing 0.5 A output pulses with rise times of less than 40 nanoseconds. A p-n-p-n four-layer switching device made by the same technique had an impedance ratio for its "on" and "off" states of higher than 3 million to one, while an avalanche switching transistor was capable of producing 50 mA pulses with a rise time of 1 nanosecond $\left(10^{-9}\right)$.

Other manufacturers are using the " mesa" construction and the driftfield technique for their highfrequency transistors. A.E.I., for example, had two new mesa transistors, XA161 and XA162, with minimum cut-off frequencies of $25 \mathrm{Mc} / \mathrm{s}$ and $35 \mathrm{Mc} / \mathrm{s}$ respectively, and three drift types, XA141, XA142 and XA1.43, with minimum cut-off frequencies of $20 \mathrm{Mc} / \mathrm{s}, 40 \mathrm{Mc} / \mathrm{s}$ and $60 \mathrm{Mc} / \mathrm{s}$ respectively. This firm has also introduced four power transistors for industrial applications. Two of. them, XC155 and XC156, have peak current ratings of 10A and collectorbase voltage ratings of 80 and 100 volts respectively. The other two, XC141 and XC142, have peak current ratings of 3 A and collectorbase voltage ratings of 40 V and 60 V .

In the field of power control, as distinct from power amplification, the silicon-controlled rectifier is rapidly invading the domain of the industrial thyratron and other large devices. It enables several kilowatts of power to be controlled by a few milliwatts. Examples were shown by Westinghouse and International Rectifier. On the Westinghouse stand an impressive demonstration was given of a 10 kW tungsten-lamp sign being turned on and off by two silicon controlled rectifiers connected, in inverse parallel, between the a.c. supply and the load. The r.m.s. output voltage was varied by controlling the proportion of each half cycle for which the rectifiers were conducting.

## TECHNICAL FEASIBILITY OF ALTERNATIVE PLANS FOR TV DEVELOPMENT

SINCE the Television Advisory Committee was reconstructed in 1952, under the chairmanship of Admiral Sir Charles Daniel,* it has issued several reports but the most eagerly awaited was that published on June 1st. $\dagger$ Although the broad terms of reference of the committee are "To advise the Postmaster General on the development of television and sound broadcasting at frequencies above 30 megacycles per second and related matters, including competitive television services and television for public showing in cinemas and elsewhere," the committee was asked in March 1956 specifically "for advice on fundamental technical problems of television development." In particular the members were asked whether they would
(a) recommend whether the existing 405-line standards were likely to remain adequate for all purposes for the next 25 years;
(b) say whether there was any reason why the United Kingdom should not adopt 625 lines for Bands IV and V in this country, if it were recommended by the International Radio Consultative Committee (C.C.I.R.) as the European standard;
(c) make recommendations regarding the general principles of a compatible colour system for operation, initially at least only in Bands IV and V;
(d) recommend the best technical means of transmitting the colour signals associated with (c) above, bearing in mind that these need not necessarily be in the same frequency band as the monochrome signals;
(e) take note of, and report on, any proposals by the B.B.C. or I.T.A. for adding colour to transmissions within Bands I and III; and
(f) give their views as to the technical advantages to be gained from the use of higher standards in Bands I and 1II, if the possibility of extension of television into Bands IV and V were to be disregarded, and taking into account the improvement in receiver and other apparatus that may be expected in the next 25 years.
It is in answer to these specific questions that the present report was presented to the P.M.G.

Because of the widespread interest in the report Mr. Bevins promised Members of Parliament that it would be published. It is, however, stated in the foreword, and the P.M.G. has personally stressed the fact, that the Government has reached no decision as to which of the possible alternatives should be adopted. Furthermore, if any changes in line

[^1]standards were to be decided upon, they would require to be made in accordance with a long-term phased programme which should take account of the interests of the viewers, the broadcasting organizations, and the radio industry. The committee has emphasized that 405 -line services would need to be continued for many years so that there would be no question of 405 -line receivers becoming prematurely obsolescent.

It will be appreciated that although the questions posed are technical, there are political and economic factors which enter very largely into the picture. For example:
(a) the number, nature and coverage of the television programmes to be provided;
(b) the method and time-table by which the new standards should be introduced;
(c) the costs of introducing the new standards and the way in which they could be met.

These are, however, mostly outside the committee's terms of reference and the purpose of the report is solely to give the Government the technical information it needs to formulate policy.

Over the past two or three years various technical studies, including propagation tests in Bands IV and V, 625-line test transmissions from a Band V transmitter at Crystal Palace and colour transmissions on 405 lines, have been undertaken by the T.A.C.'s Technical Sub-Committee $\ddagger$ in collaboration with the D.S.I.R., the radio industry, Post Office and broadcasting organizations.

## How Many Programmes?

As is shown in Table I, Bands I and III, if fully exploited on the present 405 -line standard, could provide three programmes-two with at least $98 \%$ coverage and one with a coverage of over $95 \%$.

Tests have shown that an acceptable television service could be provided in Bands IV and V. Nevertheless, the service area of a transmitter operating in these bands would be more restricted than for the lower bands and more irregular, particularly in mountainous or hilly terrain, and to give a nation-wide service a greater number of transmitters would therefore be needed. The report states that whereas some 20 stations are required in each of the lower bands to provide upwards of
$\ddagger$ See Appendix II for list of present members.

TABLE 1

|  |  | No. of channels for operation on |  | No. of channels per national programme |  | No. of programmes which could be provided using |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Band | Range (Mc/s) | 405 lines $5 \mathrm{Mc} / \mathrm{s}$ channels | 625 lines $8 \mathrm{Mc} / \mathrm{s}$ channels | No. | for estimated population coverage | 405 lines and 5 Mc/s channels | 625 lines and 8 $\mathrm{Mc} / \mathrm{s}$ channels |
| 1 | 41-m8 | 5 | 3 |  |  |  |  |
| III | 174-216 | 8 | 5 |  |  | $1(95-98 \%)$ |  |
| IV | 470-582* | 22 | 14 | 2/13 | 95\% | 3(98\%) | 2(98\%) |
| v | 606-800* | 38 | 24 | $\int 17 / 18$ | 98\% | $\begin{gathered} 4(95 \%) \\ 1(70 \%) \end{gathered}$ | $\begin{aligned} & 2(95 \%) \& \\ & (90 \%) \end{aligned}$ |

*As amended at the Geneva Conference, 1959.
$98 \%$ population coverage, possibly four or five times as many stations would be needed in Bands IV and V to give $95 \%$ coverage.

Because of the undoubted advantages of the v.h.f. bands over the u.h.f. bands for television, the Committee sought advice whether any broadening of Band III was practicable within the foreseeable future. The Radio Industry Council, too, feel strongly that any extension of television up to four national or near-national programmes should, if at all possible, be accommodated within Band I and an extended Band III. Both the Committee and the R.I.C. were, however, informed "that the pressure in the v.h.f. portion of the spectrum is immense, and that the Government must hold a balance between desirable broadcasting development and the requirements on these frequencies for other services." That being so, at this stage no hope can be held out that additional frequencies could be made available in the v.h.f. bands for television purposes. Any extension of television must, therefore, be made in the u.h.f. bands.
It is stated in the report, although this may not have been previously generally known, that the T.A.C. advised the P.M.G. early last year that the U.K. delegation attending the C.C.I.R. meeting at Los Angeles (April, 1959) should be empowered to say that "in the interests of frequency planning the United Kingdom would adopt an $8 \mathrm{Mc} / \mathrm{s}$ channel in Bands IV and V, if Europe generally adopts this, and further that if the United Kingdom should decide to adopt 625 -line standards in those Bands a $6 \mathrm{Mc} / \mathrm{s}$ video bandwidth would be used."

## 405-line Standard Inadequate

The Committee states that good as the 405 -line picture may be for the size of screens now in general use they do not think the 405 -line system will be adequate for the next 25 years.
As will be seen from Table II the majority of European countries as well as some in the western hemisphere and Australasia have adopted 625 lines. In field trials in Band V a comparison was made of 405 -line and 625 -line pictures. The results showed that the overall assessment of the 625 -line pictures was not significantly different from that of 405 -line although in areas of comparatively high field strength the 625 -line pictures generally received a slightly higher assessment. The Technical Sub-Committee felt that the fact that there was not a significant difference in the overall assessment of picture quality was due partly to the nature of the trials and partly to the restriction of the video bandwidth of the 625 -line system to $5 \mathrm{Mc} / \mathrm{s}$. They considered, however, with one dissentient, that with further development of this system using a $6 \mathrm{Mc} / \mathrm{s}$ video bandwidth and receivers with improved noise factors 625 -line pictures, particularly on larger screens, would show a definite superiority. Following further international discussion the Sub-Committee considered that there would be technical advantages and no loss in picture quality in restricting the video bandwidth to $5.5 \mathrm{Mc} / \mathrm{s}$ and increasing the width of the vestigial side-band from $0.75 \mathrm{Mc} / \mathrm{s}$ to $1.25 \mathrm{Mc} / \mathrm{s}$.
Assuming it to be the Government's policy to develop television beyond the capacity of the present two bands, the committee points out that the introduction of Bands IV and V will provide the last opportunity of improving the standards of definition. They recommend the use of $625-l i n e$ with an

TABLE II

|  | Adopted in |  | Channel width |
| :---: | :---: | :---: | :---: |
| 405-line | United Kingdom Ireland |  | $5 \mathrm{Mc} / \mathrm{s}$ |
| 525-line | Bermuda <br> Brazil <br> Canada <br> Colombia <br> Cuba <br> Dominican Rep. <br> El Salvador <br> Guatemala <br> Iran <br> Korea | Japan <br> Mexico <br> Nicaragua <br> Panama <br> Peru <br> Philippines <br> Puerto Rico <br> Saudi Arabia <br> Thailand <br> Uruguay <br> U.S.A. | $6 \mathrm{Mc} / \mathrm{s}$ |
| 625-line | Argentine $\dagger$ <br> Australia <br> Austria <br> Belgium <br> Bulgaria* <br> Cyprus <br> Czechoslovakia* <br> Denmark <br> Egypt <br> Finland <br> E. Germany <br> W. Germany <br> Hungary* <br> Iraq <br> Italy | Lebanon <br> Netherlands <br> New Zealand <br> Nigeria <br> Norway <br> Poland* <br> Portugal <br> Rumania* <br> Spain <br> Sweden <br> Switzerland <br> Turkey <br> U.S.S.R.* <br> Venezuela $\dagger$ <br> Yugoslavia | $\begin{aligned} & 7 \mathrm{Mc} / \mathrm{s} \\ & (* 8-\mathrm{Mc} / \mathrm{s} ; \\ & +6 . \mathrm{Mc} / \mathrm{s} \\ & \text { channel) } \end{aligned}$ |
| 819-line | Algeria France Belgium | Monaco Morocco Luxembourg | $13 \mathrm{Mc} / \mathrm{s}$ <br> $7 \mathrm{Mc} / \mathrm{s}$ |

$8 \mathrm{Mc} / \mathrm{s}$ channel in these bands and ultimately in Bands I and III. It is pointed out that the maintenance of 405 -line operation here would show the United Kingdom to a disadvantage in Eurovision as standard convertors degrade picture quality, particularly for conversion to a higher standard, and this would have its effect in selling United Kingdom programme material to the rest of Europe.

On the question of colour the members of the committee state " we are of the opinion that present technical and economic limitations make it undesirable to introduce a colour television system in the near future. We will report further on the technical details of colour television standards as soon as we are in a position to do so."

## APPENDIX'I

Present members of T.A.C.
Admiral Sir Charles Daniel, chairman.
B. St. J. Trend (Treasury).
D. W. G. L. Haviland (Ministry of Aviation).
W. A. Wolverson (Post Office).
H. Carleton Greene (B.B.C.).

Sir Robert Fraser (I.T.A.).
G. Darnley Smith $\}$ (Radio Industry).
C. O. Stanley

Sir Edward Herbert.
Lord Aberconway.
Sir Walter Puckey.
J. L. Judd (Post Office), secretary.

## APPENDIX II

Present members of Technical Sub-Committee
A. H. Mumford (Post Office), chairman.

Sir Harold Bishop (B.B.C.), deputy chairman.
Capt. C. F. Booth (Post Office).
A B. Howe (B.B.C.).
P. A. T. Bevan (I.T.A.).
T. C. Macnamara (Associated Television).

Dr. L. F. Broadway (E.M.I.).
K. I. Jones (Ferguson).
E. P. Wethey (K.B.).
V. J. Cooper (Marconi's).

Dr. R. L. Smith-Rose (D.S.I.R.).
T. M. C. Lance (Rank-Cintel).
C. W. Sowton (Post Office), secretary.

# WORLD OF WIRELESS 

## I.T.A. Plans

FUTURE plans for the extension of the coverage provided by the I.T.A. stations inclúde five new transmitters to be opened next year and further stations the following year. As has already been announced, the Authority is also increasing the height of the masts at Croydon (which with its new aerial will then have an e.r.p. of 200 kW ) and at Black Hill and Lichfield.

The first three stations, to be brought into service next spring, are the dual transmitters in the south-west-Stockland Hill, Devon (channel 9) and Caradon Hill, Cornwall (12)-and Caldbeck, near Carlisle (11). Towards the end of 1961 a high-powered transmitter will be opened in Kincardineshire, Scotland, on channel 9 and a low-power transmitter at Selkirk, the channel for which has not yet been announced.

Stations planned for 1962 are for Inverness-shire, West Wales, Londonderry and the Channel Islands if the provisions of the Television Act are extended to cover the Islands.

## Receiver Production

THE sale of 17 -inch television receivers continues to dominate the home market and in 1959 represented $91 \%$ of the 2.75 M receivers sold. The previous year's figure was $83 \%$ of the 2 M sets sold. The sale of 21 -inch sets rose from about 100,000 in 1958 to 165,000 last year-the percentage of the total sales being 5 and 6 respectively. The demand for 14 -inch sets continues to decline.

Receiver exports for 1959 , valued at $£ 3,247,000$, declined by $8 \%$ compared with the previous year. Sweden, for the second successive year, heads the list of receiver importers with a total of $£ 298,796$ of which all but $£ 5,000$ was for television receivers. Nigeria is second in the list with a value of $£ 248,531$, but little of this was for television-the country's first TV stations opened last October. The second highest importer of television receivers was Eire with £101,934 worth.

These figures are given in the annual report of the British Radio Equipment Manufacturers' Association.

## Inst.P. and Phys. Soc. Amalgamate

PROPOSALS for the amalgamation of the Institute of Physics and the Physical Society have been a frequent topic of discussion and now a new. body under the name "The Institute of Physics and The Physical Society " has been incorporated.

The Physical Society was founded in 1874 and on the initiative of that Society the Institute of Physics was founded 45 years later. The original scheme for the Institute envisaged a kind of federation of societies interested in physics.

Broadly speaking, the scientific meetings and publications of the Institute were confined to applied physics, while those of the Society were concerned more with pure physics. As, however, over the past 20 years or so, the boundary between these two aspects became less defined, there has been increasing
overlap in the activities of the two bodies and in their membership.

The three Institute of Physics grades of member-ship-fellow, associate and graduate-will continue under the new organization but there will also be fellows of the Physical Society.

The first president of the amalgamated body is Sir John Cockcroft.

## Birthday Honours

AMONG the recipients of awards in the Queen's Birthday Honours list are the following:-

## Knighthood

Dr. Gordon B. B. M. Sutherland, F.R.S., Director, National Physical Laboratory.
Dr. Basil F. J. Schonland, C.B.E., F.R.S., Director, Research Group, U.K. Atomic Energy Authority. K.B.E

Dr. Robert Cockburn, C.B., O.B.E., Chief Scientist, Ministry of Aviation.

## C.B.

Major-General E. S. Cole, C.B.E., Director of Telecommunications, War Office.
Dr. J. S. McPetrie, Director-General of Electronics. Research and Development, Ministry of Aviation.
C.B.E.

Dr. T. E. Allibone, F.R.S., Director, A.E.I. Research Laboratory, Aldermaston.
Dr. R. A. Smith, Chief Scientific Officer, Royal Radar Establishment.
O.B.E.
F. W. Bates, Works Director, Kelvin and Hughes.

Dr. L. F. Broadway, Head of E.M.I. Research Laboratories.
R. J. Clayton, Manager, G.E.C. Applied Electronics Laboratories.
Dr. A. L. Cullen, Professor of Electrical Engineering, University of Sheffield.
F. J. D. Taylor, M.B.E., Staff Engineer, Post Office Research Station.
M.B.E.
F. H. Austen, General Manager, Rediffusion (South East), Ltd.
W. F. Coleman, Deputy Director of Broadcasting (Engineering), Ghana.
W. G. Dickson, Wireless Communications Superintendent, Ministry of Home Affairs for Northern Ireland.
C. A. Green, lately Communications Officer, Office of the U.K. Commissioner for Singapore and CommissionerGeneral for South-East Asia.
C. H. Pope, Radio Supervisor, War Office.
B.E.M.

Miss S. Holloway, Communications Officer, Birdlip Radio Station, Ministry of Aviation.

Jubilee Lectures.-To mark the 25th anniversary of the formation of the company, Ultra Electric (Holdings), Ltd., is organizing a series of lectures, the first of which will be in the Recital Room of the Royal Festival Hall, London, on September 14th. The speaker will be Professor Arthur Porter, Dean of Electrical Engineering at the University of Saskatchewan, whose subject will be "The evolution of instrumentation." On October 19th, also at the Royal Festival Hall, G. W. A. Dummer, Superintendent of Components Research, Development and Testing at R.R.E., will review the latest developments in components.
D.S.I.R. Grants for Radio Astronomy.-In addition to continued support for the Manchester University's Jodrell Bank experimental station with three grants totalling $£ 187,000$, it is stated in the D.S.I.R.'s 1959 Report that it has awarded nearly $£ 40,000$ to Professor M. Ryle for the development of new techniques and equipment at Cambridge University. Professor Ryle's work will be aimed at new methods of constructing and improving radio-telescopes and new automatic systems of data-recording for automatic computation.
Disc Production.-Figures issued by the Board of Trade show that in the first four months of this year the production of 45 r.p.m. records increased by $41 \%$ compared with last year ( 17.3 M against 12.3 M ) and $33 \frac{1}{3}$ r.p.m. discs by $22 \%$ ( 5.5 M against 4.5 M ). During the same period $63 \%$ fewer 78 r.p.m. discs were produced ( 1.5 M compared with 4.1 M a year ago)

Licences.-During April the number of combined television-sound licences in the U.K. increased by 98,932 bringing the total to $10,568,685$. Sound-only licences totalled $4,484,063$, including 432,790 for car radio. In West Germany (including West Berlin) the number of sound radio licences showed a decrease for the first time during the first four months of the year$15,617,338$ compared with $15,899,447$ on January 1 st. During the same period combined television-sound licences increased from $3,385,003$ to $3,883,145$.

St. Dunstan's has re-established a scientific committee to study the whole field of sensory reading and guiding devices. This is announced in the charity's latest report. Members of the committee are: Air Commodore G. Bentley Dacre (chairman); Dr. A. M. Uttley, superintendent, control mechanisms and electronics division, N.P.L.; Dr. H. B. Barlow, King's College, Cambridge; Dr. R. L. Beurle, English Electric Valve Co.; Dr. D. E. Broadbent, director of applied psychology research unit of the Medical Research Council, Cambridge, and Lord Fraser of Lonsdale.

Anglesey Radio, the new Post Office coast radio station at Amlwch opened by the P.M.G. on May 23rd, has taken over all the coast station services previously provided by Seaforth Radio, which it replaces.
"Broadcast Entertainment by Wire."-We regret that, due to a printer's error, the name Teleng, Ltd., was misspelt in the acknowledgments on page 214 of the May issue.

Potential-indicating Lamps.-Acru's fiuorescent-green indicator lamps require a minimum striking potential of 160 V , not 7160 V as stated on p. 301 of our June issue.

Biological Engineering Society is the name of a new group which was formed in June. The society has a distinct bias towards electronics and is intended to bring together doctors, physiologists, electronic engineers, mechanical engineers and physicists to further the applications of engineering to biological and medical problems. The president is Dr. R. Woolmer of the Royal College of Surgeons, and the acting secretary is Dr. A. Nightingale, Physics Laboratory, St. Thomas' Hospital, London, S.E. 1 .

Control Engineering.-A short course providing an introduction to control engineering theory and practice is being conducted by the Loughborough College of Technology from July 18th to 29 th. A leaflet, obtainable from the college, gives details of the course. The fee for the course and full residence is 35 gn .

Non-Destructive Testing. A conference on the "Theory and practice of ultrasonic inspection" is to be held at the Queens Hotel, Cheltenham, from September 22nd to 24th. The arrangements are being made jointly by the Institute of Physics' Non-Destructive Testing Group, the Society of Non-Destructive Examination and the Non-Destructive ' $e s t i n g$ Society of Great Britain. Details can be obtained from the conference secretary, I. M. Barnes, Materials Laboratory, de Havilland Propellers Ltd., Hatfield, Herts.

Air Traffic Control.-The Guild of Air Traffic Control Officers is to hold its third A.T.C. Convention at Bournemouth on October 18th and 19th. Details are obtainable from the Guild at 118, Mount Street, London, W.1.

Electronic telephone exchanges is the subject of a conference being organized by the I.E.E. for November 22nd to 24 th. It is hoped it will provide an opportunity for the interchange of information and experiences of the construction and operation of fully electronic exchanges both in this country and overseas. Further details and a form of registration may be obtained on application to the I.E.E., Savoy Place, London, W.C.2.
"Television Explained."-First published. in 1947 under the authorship of W. E. Miller, managing editor of the Wireless and Electrical Trader, this book is now in its 7th edition. It includes a new chapter on combined television and fm . sound receivers. Both this edition and the preceding ene were revised by E. A. W. Spreadbury, associate editor of the Trader. It is obtainable from our Publishers, price 12s 6 d .


FOCAL POINT of the B.B.C. Television Centre, Wood Lane, London, W.12. All sound and vision signals from the nine studios and the telecine and videotape machines are fed to this Central Apparatus Room for distribution. The first transmission from the centre is on June 29 th.

Audio Manufacturers' Group of the British Radio Equipment Manufacturers' Association has elected the following member firms (whose representatives names are in parentheses) to form the management committee: A.E.I. Sound Equipment (L. R. Metcalfe); Beam-Echo (H. M. Rahmer); E. K. Cole (J. A. Catchpole); Clarke \& Smith (Major J. F. E. Clarke); Electric Audio Reproducers (L. Stone); Gramophone Co. (H. S. Futter); Grampian Reproducers (J. E. Morley); Jason (G. G. Blundell); Lowther (D. M. Chave) and Trix (D. A. Lyons). The Committee has re-elected Major J. F. E. Clarke as chairman and elected D. M. Chave vicechairman in succession to D. A. Lyons.

Autumn Audio Fair.-The venue for this year's Autumn Audio Fair is to be the Palace Hotel, Southport, Lancs. It is being organized by Audio Fairs Ltd., 22 Orchard Street, London, W.1, and will be held on October 7th, 8th and 9th.
R.I. Club.-The report presented at the annual general meeting of the Radio Industries Club on May 31st recorded a membership of the "parent" club of 996. The membership of the seven affiliated clubs in the provinces is 1,330 . F. W. Perks, chairman of Radio Industry Exhibitions Ltd. and immediate past chairman of B.R.E.M.A., is the new president of the club.

Radio Ball.-This annual function organized by the Radio Industries Club during the National Radio Show will be held at Grosvenor House, Park Lane, London, W. 1, on August 26th.
I.E.E.-More meetings were held by the Electronics and Communications Section of the I.E.E. than all three other sections of the Institution. The Electronics and Communications Section also has the largest membership $(6,171)$ of any specialized section. During the year ended in March, the Institution membership increased by 1,678 to 46,222 . Student members increased by 592 to 4,689 and graduates by 530 to 14,545.

## Personalities

Lord Halsbury has been appointed a Governor of the B.B.C. until 1962 in succession to Sir Edward C. Benthall, who has resigned. Lord Halsbury was managing director of the National Research Development Corporation for ten years until his retirement in March, 1959. He is now chairman of International Rectifier Co. (G.B.) Ltd., which was formed by the International Rectifier Corporation of the U.S. and the Lancashire Dynamo Company, and also of L.C.E. Ltd. recently formed jointly by G. \& E. Bradley (a subsidiary of Joseph Lucas) and Collins Radio Co. of America.

Alan Wolstencroft is the new Director of Radio Services in the Post Office in succession to W. A. Wolverson, who, as announced in our last issue, has become a Deputy Director General. Mr. Wolstencroft, who is 46 , joined the Post Office in 1936. He was closely associated with the preparations for the setting-up of the Independent Television Authority in 1954 and was in fact its first secretary for a year.
M. L. Jofeh, O.B.E., A.M.I.E.E., manager of the industrial division of Sperry Gyroscope Co., which he joined in 1947, has been appointed an additional director. For several years he headed the engineering unit at the company's Stonehouse, Glos., factory, but in 1954 returned to Brentford as deputy chief engineer. He was appointed chief engineer in 1957 and, with the reorganization of the company in January last year, became manager of the industrial division in which is concentrated the company's interests in industrial control engineering. For eleven years before joining Sperry Mr. Jofeh was in the research laboratories of Cossor.

John D. Clare, M.Sc., A.M.I.E.E., has succeeded Air Commodore H. B. Wrigley, C.B.E., as Director of Guided Weapons Research and Development (Air) in the Ministry of Aviation. Born in 1920 and educated at Birmingham University, Mr. Clare was employed throughout the war in the development laboratory of the G.E.C., Coventry. From 1945 to 1950 he was a senior engineer with Sobell Industries. He then entered the Civil Service at what is now the Royal Radar Establishment, and was for five years section leader responsible for the centimetric receiver system on new fin control "radar" and low-altitude guided weapons. Since 1955 he has been superintendent of the surface-to-air guided weapons department.

R. I. Kinross, M.I.E.E., managing director of Rediffusion Research, Ltd., is the new president of the Society of Relay Engineers. He was for nine years with E.M.I. and for a year with Philco before the war. During the war he served in the Royal Corps of Signals and was seconded to Military Intelligence. He joined Rediffusion after the war as chief engineer of a region, and subsequently took charge of the company's Development Department. He has been managing director of Rediffusion Research and a director of Television Research since 1958. R. P. Gabriel, B.Sc., M.I.E.E., A.M.Brit.I.R.E., the new vice-president of the Society of Relay Engineers, is chief engineer of Rediffusion.
P. W. Faulkner, O.B.E., has joined Rank-Xerox Ltd. as deputy managing director. He joined Plessey in 1952 and for some time was general manager of the company's commercial and metallurgical division at Towcester, Northants. He has been a director of the Plessey International Co . and also an executive director of the Plessey Co. for several years.
J. W. Soulsby, chief radio officer in the British India Steam Navigation Company's vessel Uganda, has been re-elected chairman of the Radio Officers' Union for the sixth consecutive year. He joined the Marconi Marine Co. at the age of 18 in 1918. W. S. Armstrong is again vice-chairman. It is his fourth term of office. Mr. Armstrong, who is 47 , was with the Marconi Company's marine staff until 1947, when he was appointed to the staff of the Inspectors and Technical Employees' Section of the Union.
J. Sykes, M.I.E.E., M.Brit.I.R.E., M.I.N., has left the Ministry of Aviation, in which he was superintendent of the civil aviation communications centre at Croydon Airport, and is setting up as a consultant specializing in technical training and recruitment schemes. He has been with the Ministry and its predecessors for 25 years. Mr. Sykes, whose address is Red Lion Court, Stalbridge, Dorset, serves on the City and Guilds Advisory Committee on Telecommunication Engineering, and the membership committee of the Brit.I.R.E.
P. T. H. Dannahy, A.M.Brie.I.R.E., has joined Radio and Allied Industries Lid., manufacturers of Sobell and McMichael receivers, as chief radio engineer. He was chief engineer and production manager with Peto-Scott Electrical Instruments until 1945, when he joined the Ferguson Radio Corporation as chief radio engineer.
P. T. H. Dannahy

E. David Parchment, who joined Leevers-Rich Equipment last August as technical sales manager, has been appointed a director in place of G. W. Parkes, who has resigned from the directorate of the company. Mr Parchment was for many years with the Decca Record Co. and subsequently was sales director of Epsilon Sales and Services Ltd.
W. E. C. Varley, Assoc.I.E.E., A.M.Brit.I.R.E., has been appointed by the B.B.C. Superintendent Engineer, Transmitters, in succession to E. F. Wheeler, O.B.E., D.L.C., M.I.E.E., who has retired after 17 years in the post and 36 years' service with the Corporation. Mr. Varley joined the Corporation in 1933. During 1943 and 1944 he was chief broadcasting engineer at the Allied Forces Headquarters in North Africa.
M. H. Hall, M.B.E., has become Assistant Superintendent Engineer, B.B.C. London Television Studios, in succession to H. Wa'ker, O.B.E., A.M.I.E.E., who is retiring. Mr. Walker joined the B.B.C. in 1931. He was appointed engineer-in-charge of the Alexandra Palace television station in 1950 and since 1953 has been Assistant Superintendent Engineer, London Television Studios. Mr. Hall joined the Corporation in 1927. In 1950 he was appointed engineer-in-charge of the B.B.C Television Studios at Lime Grove.
W. D. Hatcher, B.Sc.(Eng.), A.M.Brit.I.R.E., who succeeds Mr. Hall as Engineer-in-Charge, London Television Studios, joined the B.B.C. in 1931. During the war he was concerned with the design and equipment of the B.B.C.'s high-power short-wave transmitting stations.
C. Powell, contributor of the article "Radio Aids to Hydrography" in this issue, has been with the Decca Navigator Co. since 1946. He is now in charge of the company's technical information deparment, but was initially concerned with applications of the Navigator for surveying and originated the two-range Decca technique. Mr. Powell's industrial career began in 1934 when he was personal assistant to $\mathbf{P}$. K. Turner, of Hartley-Turner Radio. For part of the war he was attached to the Army Operational Research Group.
J. H. Mitchell, B.Sc., Ph.D., M.I.E.E., has been appointed to succeed G. D. Christie as chairman of the board of directors of Associated Transistors Lid., which is operated jointly by Automatic Telephone \& Electric Co., English Electric, and Ericsson Telephones. Dr. Mitchell, who is director of research of Ericsson Telephones, which he joined in 1947, was at the Bawdsey Research Station in 1936 and for his radar contributions he received an award from the Royal Commission on Awards to Inventors. Mr. Christie, who is a director of A.T. \& $E_{\imath}$, remains a member of the board of Associated Transistors Ltd.
D. M. MacKay, B.Sc., Ph.D., who is in the Wheatstone Physics Laboratory, King's College, London University, has been appointed to the Research Chair in Communication founded by Granada Television in the University College of North Staffordshire.

Wing Commander A. R. Gilding, newly appointed assistant technical secretary of the Electronic Engineering Association, was until recently in charge of the branch of the Air Ministry responsible for airborne radio servicing policy and some aspects of new airborne equipments. Throughout his career in the R.A.F., which he began in 1927 as an aircraft apprentice, Wing Commander Gilding specialized in communications and radar. For two years before being posted to the Air Ministry in 1954, he was at N.A.T.O. headquarters, Oslo.

## OBITUARY

Rupert Browne, O.B.E., B.Sc., who, owing to ill-health, resigned from the secretaryship of the Radio Industry Council in 1957, died on May 21st. Born in 1897 and trained as a chemist, Rupert Browne graduated at London University. In 1924 he joined the staff of the National Association of Radio Manufacturers and then, when the Radio Manufacturers' Association was formed, he joined that organization. When in 1945 the Radio Industry Council was born out of the R.M.A. he became its secretary. He was appointed an O.B.E. for his work on a war-time committee, under the chairmanship of Lord Hankey, devising and working a scheme, in collaboration with others in the radio industry, for the training of radio engineers for the Services.
F. G. Robb, A.M.I.E.E., chief of Marconi's Test Division from 1948 until his retirement five years ago, died on May 14th. He was with the company for 36 years except for a period during the last war, when he was seconded to the Admiralty, where he became chief of radar test. He was for some years in the company's design and development section, where at one time he worked on the development of transmitters for the Marconi-Franklin short-wave beam system.

John A. J. Cooper, sales director of Leevers-Rich Equipment since 1954, died on May 16th at the age of 67. After service with the War Department, Mr. Cooper joined the B.B.C. in 1928 and became senior recording engineer in the engineering division.

Phllatellsts among our readers will be interested in this series of six stamps recently issued by the Czechoslovak Postal Authorities commemorating international pioneers of wireless. They are Tesla (Yugoslovian); Popov (Russian); Branly (French); Marconi (Itolian); Hertz (German); and Armstrong (American). Each stamp includes a portrait of the pioneer and an illustration depicting on aspect of his work.


# News from the Industry 

Relay Exchanges Ltd., record a group trading profit during 1959 of $£ 3,583,311$, compared with $£ 2,515,988$ the previous year. After deducting over $£ 2 \mathrm{M}$ for depreciation and provision for renewal of equipment and $£ 128,182$ for taxation, the year's net profit was just over £1M. Sudsidiaries of Relay Exchanges include 16 radio and television relay companies, six Rentaset renting companies, four retail concerns and three manufacturing companies including Goodmans Industries.

Simms Motor and Electronics Corporation have announced a group trading profit for the past year of $£ 1,037,690$ before taxation, compared with the previous year's £632,072. Taxation absorbed £548,200 in 1959 and $£ 366,000$ the year before. Reference is made in the directors' report to the activities of N.S.F., which in terms of output and profit ranks second in the group, and to the recent acquisition of Cawkell Research and Electronics.

Pye closed-circuit television has been installed in a new plant of the Dunlop Rubber Co. at Fort Dunlop, Birmingham, to facilitate the examination of tyres being tested at speeds of up to $500 \mathrm{~m} . \mathrm{p} . \mathrm{h}$. Each of the two cameras has two lenses. A calibration grid is electronically superimposed on the television display in the control room from which the cameras are remotely operated.
The Chloride Electrical Storage Co. has opened new central research and development laboratories at Fletcher Avenue, Clifton, Swinton, Manchester, for fundamental research into the physics of electrochemical couples, including fuel cells, and into the problems of extending life and reducing weight and cost in conventional types of cell. The laboratories have a floor space of 41,500 square feet and there is a staft of 180 , including 40 qualified specialists.

Du Mont Agents.-Aveley Electric Ltd., of Ayron Road, Aveley Industrial Estate, South Ockendon, Essex, have been appointed U.K. agents for all products of the Allen B. Du Mont Laboratories Inc., of the U.S.A. The Du Mont range of equipment includes the new 425 digital-readout high-frequency oscilloscope, oscilloscope recording cameras and photo-multipliers.

Ferranti announce that they are engaged in the development of another radar fire control system, Airpass II. The original Airpass was shown publicly for the first time at the S.B.A.C. show at Farnborough last September. Airpass II will provide the following facilities:--air-to-air radar search and automatic tracking, computer-controlled approach for blind or visual attacks, and radar-assisted attacks against surface targets.

Britec Lid., of 17 Charing Cross Road, London, W.C. 2 (Tel.: Whitehall 3070), have been appointed distributing agents for Elesta cold-cathode tubes and electronic controls manufactured in Switzerland.
Marconi's are supplying a $50-\mathrm{kW} 50-\mathrm{cm}$ airfield control radar, Type S264, with two moving-coil display units and ancillaries, for the Royal Radar Establishment airfield at Pershore, Worcestershire.
R.C.A.-Dr. H. R. L. Lamont, director of R.C.A. European technical relations, has moved his office from Pall Mall to 36 Berkeley Square, London, W. 1 (Tel.: Grosvenor 1217).
A new factory for the production of selenium rectifiers and silicon diodes was opened at Oxted, Surrey, on May 25th by the International Rectifier Company (Great Britain) Ltd., which is jointly owned by Metal Industries Ltd. (through its acquisition of Lancashire Dynamo Construction) and the International Rectifier Corporation, of Los Angeles.

Grundig in N. Ireland.-A new company, Grundig Works Ltd., is being formed in Belfast to operate a factory in Dunmurry on the outskirts of the city. The factory, which is planned to begin operation in September, will initially produce one model tape recorder, but eventually other equipment will be made. All products made at the factory will be distributed throughout the U.K. by Grundig (Great Britain) Ltd. The directors of Grundig Works include Max Grundig (founder of the organization), three others from the parent company and G. S. Taylor, chairman and managing director of Grundig (G.B.) and of Wolsey Electronics.

Nash \& Thompson Ltd., Hook Rise, Tolworth, Surbiton, Surrey, have been given approval as a Part III Test House (including testing under environmental conditions) by the Director General of Inspection for Functional and Performance Testing of Electronic Components. The company's environmental and electronic testing laboratories have been in existence for over six years and have been approved as a Test House by the Air Registration Board for some time. Any firm or organization may submit components for testing to a specification and a Certificate of Test will then be issued stating that the tests have been carried out in accordance with the requirements of the Air Registration Board or Director General of Inspection.
E.M.I. Electronics Ltd. have supplied the vision equipment and have carried out the "technical" wiring and installation at Associated Rediffusion's new Studio 5 , opened recently at Wembley, Middx. They are also supplying 15 cameras for the television studios being construcied at Teddington, Middx. for A.B.C. Television.

Marconi's have been awarded the contract for the supply and installation of the vision and sound transmitters for the new I.T.A. station at Caldbeck to serve the Carlisle area. In addition to the duplicated transmitters and ancillary equipment, Marconi's are supplying the mast and horizontally polarized sixteen-stack quadrant aerial which will give a vision e.r.p. in the direction of maximum propagation of over 100 kW .

## EXPORT NEWS

Midas range of magnetic tape data recording equipment developed by Royston Instruments Ltd.; of Byfleet, Surrey, is to be marketed in the Western Hemisphere and Australasia by Lockheed Aircraft Services Ltd., of Ontario, California.

The Italian hydrofoil, Freccia dell Adriatico, which operates a passenger service between Trieste and Venice (a distance of 80 miles) at an average speed of $40-45$ knots, carries Kelvin Hughes marine radar type 14/9. Kelvin Hughes have also received orders for radar for Italian-built hydrofoils for operation in Norway and Finland.

Poland.-Three Marconi Mark IV television cameras and ancillary equipment have been ordered for the Warsaw studios of the Katowice station which was equipped by Marconi's. Polish television operates on the $625-\mathrm{line}, 8 \mathrm{Mc} / \mathrm{s}$ standard.

Brazil.-The complete studio and transmission equipment for a new television station at Recife, Brazil, costing in the region of $£ 250,000$, has been supplied by Marconi's.

India.-G. S. Dhingra, director of Union Radio \& Appliances Private Ltd., of 72 Janpath, New Delhi, will be visiting the U.K. in July to negotiate with firms interested in collaborating in the manufacture in India of components such as fixed and variable resistors and capacitors, loudspeakers and switches.

# Wire Broadcasting in Folland 

PLANS FOR DISTRIBUTING TELEVISION OVER THE EXISTING SOUND NETWORK

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LTHOUGH Holland is a relatively small country it has always held a high reputation for its contributions to art, science and the social services. In broadcasting its radio stations led the way in Europe in the early days, and today it enjoys one of the most efficient and widely disseminated wire distribution systems in the world.

This is run by the postal telegraph and telephone authorities, the Netherlands PTT, which was made responsible during the war for control of several independent networks and has since continued to develop and extend an integrated system.

The accompanying map gives some indication of the coverage (1959) of the main trunk cables and branches. Points on the map represent one or more power amplifiers feeding local "networks" of anything up to 1,800 houses. The total number of subscribers in Holland is about 500,000 in a population of $11 \mathrm{mil}-$ lions ( 1 in 22) whereas in Great Britain the ratio is 1 in $50(1 \mathrm{M}$ in a population of 50M). All new housing estates in Holland are now wired in advance for telephone and broadcast relay services, without any obligation on the tenants to become subscribers. Terminal outlets are embedded in the wall plaster and covered by a flush-fitting cover, which is easily replaced by a control panel with stepped volume control and selector switch if the occupants of the house or flat elect to take the relay service. Nearly 70,000 dwellings were fitted in this way last year.

Four programmes are available-the Dutch Hilversum I and II and a choice of two foreign programmes or of recorded items depending on propagating conditions and on the available material. By
mutual arrangement with the Belgian authorities, any of their programmes may be selected by remote control and passed direct from Brussels to Rotterdam by land line. Programmes from other countries are picked up by special receivers placed at favourable sites on the borders of Holland. The receivers for the three B.B.C. services are installed in a water tower at Domburg on the island of Walcheren and normally give an acceptable signal/ noise ratio from Wrotham ( 130 miles) or Norwich.
Special care is taken to provide high quality of reproduction and a frequency response of $40 \mathrm{c} / \mathrm{s}$ to $10 \mathrm{kc} / \mathrm{s} \pm 2 \mathrm{~dB}$ is guaranteed in all parts of the system. Local distribution amplifiers are housed

in kiosks which are visited regularly each month by technical staff in a van specially fitted with the instruments necessary for a thorough check of performance. The audio power available at the subscribers' outputs is of the order of $\frac{1}{2}$ watt which provides an acoustic level sufficient for most people's needs if a loudspeaker of reasonable efficiency is used. Although the authorities do not supply more complicated reproducing equipment, guidance is given for those who may wish to use existing receivers or add high-quality power amplifiers.



At present there is no regular wire distribution of television in Holland, but a thorough investigation has been made* of the feasibility of using the existing sound broadcasting cables at high frequencies. It has been found that the polythene-insulated, unscreened cables with four pairs, each twisted with different pitches have a good performance at frequencies up to $10 \mathrm{Mc} / \mathrm{s}$ and that the older paper insulated cables can also be used satisfactorily. The polythene cables used in Holland appear to suffer less from increased attenuation-about 10 to $15 \%$ under adverse conditions compared with over $30 \%$ for similar types used elsewhere. The average loss per kilometre is between 23 and 57 dB at the chosen vision carrier frequency of $7 \mathrm{Mc} / \mathrm{s}$ depending on the

[^2]Above: The receiving station for English programmes is situoted at the top of the water tower at Domburg in the island of Walcheren.

Right: Control centre at Rotterdam, where programmes are selected and monitored. Level indicators for each of the four programmes are at the back of the control desk.



Above: Distribution box for subscribers' lines. Television and/or sound signals can be applied to any subscriber by the insertion of the appropriate resistive or inductive elements which are encapsulated in foamed polyster resin.

Right: A mobile laboratory visits each network amplifier once a month for a thorough check of performance.

type of cable under consideration. The cross-view between pairs at a distance of 2 km is better than 40 dB . With a sending level of 3 volts peak-to-peak the maximum radiated field at a distance of 3 metres is $200 \mu \mathrm{~V} / \mathrm{m}$. Interference picked up on the unscreened cable would be of longitudinal character and experience has shown that its magnitude is unlikely to cause trouble unless the signal level falls below 20 mV .

The characteristic impedance of the cables at r.f. is of the order of 100 ohms and careful matching at all junctions is necessary to avoid reflections and "ghosts." The ratio of the special transformers used at these points varies according to the number of subscribers which may be up to as many as 16. The impedance "seen" at the secondary is about 1.5 ohm and as each subscriber's loading is of the order 5,000 to 10,000 ohms decoupling is very effective. Separation of sound and vision takes place after the subscriber's selector switch by simple series inductors and capacitors.

By using a vision carrier of $7 \mathrm{Mc} / \mathrm{s}$ with a complete upper sideband and vestigial lower sideband it is possible to legislate for the use of existing receivers by changing the oscillator frequency and to keep the f.m. sound signal at the normal spacing of $5.5 \mathrm{Mc} / \mathrm{s}$, i.e., at $12.5 \mathrm{Mc} / \mathrm{s}$. Alternatively, if special simplified television receivers become available for use with the wire system it is proposed to


Spectrum of the proposed television transmissions with possible alternatives for the sound accompaniment.
provide the sound accompaniment as a double sideband a.m. signal on a $50 \mathrm{kc} / \mathrm{s}$ carrier. This is preferable to sending the sound at audio frequency as it enables the original four sound-only progranımes still to be distributed. Although the initiation of a regular wire television service in Holland still awaits official sanction and possibly the backing of private enterprise to provide the recommended special receiving equipment, the technical problems of transmission have been solved.

Through the courtesy of the Netherlands PTT Wireless World was able recently to witness test transmissions of alternative television programmes over standard four-pair cables carrying the regular sound programmes. The two television signals were using synchronized carriers on adjacent pairs of an experimental circuit installed at the Leidschendam research laboratories. Excellent picture quality was obtained and there was no trace of interference from either sound or the alternative vision programme, even when the modulation of one vision carrier was switched off and the raster examined at full gain.

Similar tests carried out over a period of a year with the collaboration of subscribers in a district of the Hague have proved the reliability of the system in the field. Only one live television programme is at present available in Holland, but satisfactory tests of cross-view were made with the alternative programme supplied by a pattern generator. It was also confirmed that the effect of any external interference picked up by the unscreened feeders is negligible.

## European Television Stations

Europe's 680 or more television stations in Bands I, III and IV are tabulated geographically and in order of frequency in the 5 th edition of the list of TV stations produced by the Technical Centre of the European Broadcasting Union. A chart showing the stations in each of the channels is included in the publication, which is obtainable from the E.B.U., 32 avenue Albert Lancaster, Brussels 18, Belgium. The list and its supplements costs 50 Belgian francs.

# High-gain Video Amplifier 

FURTHER NOTES ON VERSIONS FOR GRID OR CATHODE MODULÁTION

By R. G. YOUNG

I$T$ is very gratifying to learn from the Editor that there has been a widespread interest in, and requests for, further practical details of a circuit which I should have thought would have been of merely limited interest. Evidently many folk still prefer to make their own television receivers, even in these mass-produced days and in spite of the trouble in getting parts. You almost need a licence to get $110^{\circ}$ scanning coils!

The circuit was evolved after many efforts to overcome i.f. instability, bearing in mind that "fringearea" operation was required. The instability manifested itself partly by bad streaking after bright cbjects-this was particularly so in scenes containing venetian blinds or staircases. The other effect was a black edge on bright objects, due to overshoot; this was not quite so bad, but looked terrible on Test Card C.

It seemed to me that the trouble arose from excessive signal levels appearing at the detector diode. The only answer could be to get more videofrequency amplification, and this circuit was the result after much trial-and-error experimentation with various arrangements.

It seems a little complicated, but it goes into a
Fig. I. Top and under-side views of $5 \times 2 \frac{1}{2}$-in chassis layout used for cathode-modulating version of amplifier. Leads must be kept short and stray coupling avoided.

space 5 in $\times 2 \frac{1}{2}$ in quite comfortably (see Fig. 1). I put both cathode-bypass capacitors on top of the chassis where they were easy to get at, as a change of $20 \%$ in value produces an appreciable difference in the picture, and can make or mar results. It appears that if the first cathode-bypass capacitor is too high in value, a "ring" at about $2 \mathrm{Mc} / \mathrm{s}$ is caused and too high a capacitance in the second amplifier cathode circuit causes a less-severe ring at about $1 \mathrm{Mc} / \mathrm{s}$.

In the original version definition was quite satisfactory with a wideband i.f. amplifier; but it is possible that the circuit might be used with an i.f. amplifier having a narrower pass-band, in extreme fringe conditions. In this case a choke in the second v.f. stage anode circuit can improve the picture.

Fig. 2 is the final result. Shown here is the use of a thermionic diode for the detector-some readers may prefer it-and it does make the polarity quite clear.

If space is very limited a triode-pentode of the ECF82 type can be used for the second amplifier and cathode follower with, however, a loss both of gain and peak-to-peak output (about $30 \%$ ). Some may wish to eliminate the cathode follower altogether. To do this would mean using a lower-value anode resistor in the second v.f. stage, as the stray capacitance of the c.r.t. and synchronizing separator would then become important. The effect would then be to reduce the available output (peak-topeak). It could be done, but I do not recommend such a radical change, just to save a single triode; better to use a triodepentode. Further information about the use of cathode followers in v.f. amplifiers was given in a Mullard technical advertisement (Wireless World for August, 1955, p. 90†).

It may be desired to use the amplifier in a receiver designed for grid modulation of the c.r.t. Several changes are needed in addition to the

[^3]Fig. 2. High-gain amplifier giving neg-ative-going picture for cathode modulation of c.r.t. Note

changes in polarity. The d.c. restorer becomes superfluous as the second v.f. stage grid will do the same job, by reducing the cathode bias. The purpose of the "safety circuit" round the cathode follower in Fig. 2 is to ensure that, in the event of a valve failure, the c.r.t. beam is cut off. For grid modulation of the tube this is achieved when the feed is taken directly from the cathode follower, so this safety circuit is not necessary. The modified circuit is shown in Fig. 3.
There now remains to be supplied data on choke winding. The coils of $70 \mu \mathrm{H}$ and $30 \mu \mathrm{H}$ are closewound solenoids, and the $130 \mu \mathrm{H}$ and $100 \mu \mathrm{H}$ inductors employ pile-wound sections. Details are shown in Fig. 4. All the coils were checked on an inductance (audio-frequency) bridge: it would be wise to adopt this procedure because surprising variations in inductance can occur with hand-wound coils.
It must be remembered, too, that when one of these circuits is used to replace the existing video amplifier in a receiver, some adjustment to the bril-liance-control network may be necessary to achieve proper range of control.
In conclusion, one unexpected bonus from the use of this circuit is the apparent reduction of "snow"

5 SECTIONS PILE WOUND. EACH 40 TURNS (200 TOTAL) 34 D.S.C.


$30 \mu \mathrm{H}$
90 TURNS
42 D.S.C.


4 SECTIONS PILE WOUND EACH 40 TURNS (I60TOTAL) 34 D.S.C.

Fig. 4. Coil-winding details (wire gauges given are s.w.g.).

Fig. 3. Alternative video amplifier for grid modulation of c.r.t. Sync pulses taken from cathode-folluwer output are negative-going: for sync separators needing positive pulses feed may be taken from 12AU7 anode.

on the picture; just why that is, I do not attempt to explain, but just mention it in passing.

Several readers have asked for recommendations of diode type. This is quite unimportant; I have tried Mullard OA10, G.E.C. GEX34 and "unknown" (surplus) types and could perceive no difference in the image obtained.

## APPENDIX

For use at i.f. below $35-38 \mathrm{Mc} / \mathrm{s}$ the filtering is not really satisfactory. As the circuit stands, a small amount of an i.f. below say, $20 \mathrm{Mc} / \mathrm{s}$, could appear in the output. This might not cause trouble in all cases, but it would be advisable to connect a $70-\mu \mathrm{H}$ choke in series with the cathode follower output when using a low i.f.

Where $\frac{1}{2}$-in diameter formers are not available 0.3 -in diameter can be used. For the $130-\mu \mathrm{H}$ and $100-\mu \mathrm{H}$ coils use wire gauge and spacing as shown in Fig. 4 but increase each section to 65 turns. The $70-\mu \mathrm{H}$ coil would be sectionally wound with similar dimensions to the $100-\mu \mathrm{H}$ choke, but each section would contain 50 turns.

## Birmingham-London TV Link

SO that programmes, rehearsals and advertisements originating at A.TV Alpha Studios, Aston Road, Birmingham, can be seen at Associated Television's headquarters in London, Pye Telecommunications Ltd. have installed for A.TV a $7 \mathrm{Gc} / \mathrm{s}$, 135 -mile-long link for both sound and vision. Three automatic repeater stations are used at Meriden, Cold Ashby and Barkway and the terminals are on the C.M.L. building in Birmingham and at Highgate, London.

A feature of the Barkway repeater, which has a tower over 200 ft high, is the use of passive reflectors on the tower with the microwave transmitter and receiver aerial "dishes" mounted horizontally only a few feet above the ground. Normally both the transmitter and receiver would have to be placed at the top of the tower, or long waveguides would be necessary to feed the aerials, so a considerable saving in both initial and


Plane passive reflector on mast reflects to next repeater beem from aerial at ground level.
maintenance costs has been made possible by the use of this technique of "mirroring" at the top of the tower the aerials at ground level.

Faults occurring at any station are automatically indicated on the London control board by telemetry circuits operating over a $450-\mathrm{Mc} / \mathrm{s}$ control link. Authority to install and operate the system was granted by the General Post Office, who have recently made available a band of microwave frequencies for such operations.

SHORT-WAVE CONDITIONS


Prediction for July

-***.... FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE FOR $25 \%$ OF THE TOTAL TIME

-     - PREDICTED MEDIAN STANDARD MAXIMUM USABLE FREQUENCY
—— FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE
ON ALL UNDISTURBED DAYS


General view of the four huge reflectors at Thule, Greenland

## BMEWS

American Long-range Radar Warning System

ALTHOUGH primarily a defence project for N . America the ballistic missile early warning system (BMEWS for short) is of more than passing interest to us in the U.K. first for its technical features and secondly because one of the three "forward bases" is to be in this country. Under an agreement between the U.S.A. and the U.K. we are co-operating in setting up and operating a radar tracking station on Fylingdales Moor, Yorkshire. The technical equipment for the station is being provided by the U.S. but the station will be commanded and operated by the R.A.F. The other two bases are at Thule, Greenland, and Clear, Alaska, and all three will be linked by duplicated communication channels, using cables, tropospheric scatter and microwave radio links, to the control centre of the North American Air Defence Command (NORAD) at Colorado Springs, Colorado.

The first base to be completed is at Thule which is planned to come into operation later this year. At this site there are four large rectangular reflectors each measuring 400 feet long and 165 feet high. These are for the pulsed-Doppler detection radar and together they will give a total azimuth coverage of 150 degrees.

Pulsed transmissions, fed into each reflector from horns on its nearby transmitter building, form two stationary horizontal fans at different elevations. Prediction of the probable land fall of a missile will be obtained by extrapolating its path from the range, azimuth, bearing and time sequence
data recorded as it passes through the fans. Three similar reflectors, giving a coverage of over 100 degrees are being installed at Clear, Alaska, which is scheduled to come into operation some time next year.

At Thule there will also be a dual purpose tracking radar the paraboloid of which will be housed in a specially treated plastic sphere 140 feet in diameter. It is this type of radar which will be installed in this country. At Fylingdales there will be three of these radomes capable of both detecting and tracking missiles and they will provide azimuth coverage of over 100 degrees.
The paraboloid and its pedestal weigh over 150 tons. The radome of the prototype tracking radar at the R.C.A. establishment at Moorestown, New Jersey, was assembled from 1,646 hexagonal sections

Communication routes linking the three radar bases with the BMEWS control centre in Colorado. It will be seen that all links are duplicated.

each section consisting of a 6 in thick "biscuit" of resin-impregnated paper between fibre glass walls.

At Thule station staff are protected against possible radiation hazards by the provision of screened passages linking all buildings on the one mile-square site.

The three stations when completed will have an overall range of some 3,000 nautical miles, which is ten times that of the DEW (Distant Early Warning) line which was established across the North American continent some time ago.

Reliability of the whole system is of paramount importance. To ensure continuous operation all equipment is duplicated and there is an elaborate system of checking and monitoring installed.

Some idea of the magnitude of the whole BMEWS project can be gained from the following statistics given by the American Department of Defense. Although R.C.A. is the "prime contractor" for the project with Western Electric responsible for the communications network, there are 2,900 sub-contractors. The permanent staff at Thule when it becomes operational will be about 1,000 and at Clear about 600 .

The estimated cost of the whole project is over $\$ 950 \mathrm{M}$; about half this sum being for the Thule site.



High-power klystrons for the BMEWS project.
The cost to the U.S. Air Force of the Fylingdales site is $\$ 114 \mathrm{M}$.

At the Thule site power during the construction of the station and subsequently for the operation of the system is supplied by a U.S. Navy generator ship-incidentally, the heat dissipated by the generating equipment keeps ice from forming in the basin in which the vessel is anchored.

Scanner of the tracking radar inside its 140 -foot radome. Note the sectional construction of the sphere. On the right is the prototype radome housing the tracking radar atop its transmitter-computer building at the R.C.A. experimental establishment at Moorestown, New Jersey. The buildings and radome to be built ot Fylingdales, Yorkshire, will be of similar construction.


SHOULD WE ADOPT CONTINENTAL CONVENTIONS?

ASK any service or maintenance engineer which part of a service manual is the most important: infallibly the answer will be "the circuit diagram." Without it the engineer is like a sailor without a chart; with it-no matter how little other information may be available-the skilled man will be prepared to tackle the most complex unit. But do British equipment manufacturers always make their main circuit diagrams as clear and as informative as possible? Here the answer will be no less certainin many, too many, cases an unequivocal "no."


Fig. 1. The "common or garden" fixed resistor, (a) as cultivated in Britain, (b) the less elegant but equally distinctive Continental variety, (c) a Continental bunch, complete with wattage coding.

For more than a decade, drawing offices have had as their guide British Standard 530:1948, "Graphical Symbols for Telecommunications," together with a series of supplements, now six in number. Even the most casual inspection of the circuit diagrams actually used in maintenance manuals by British electronic firms will show, by the great diversity of styles, that so far this Standard is accepted fully by only a small percentage of draughtsmen-or at least of those who determine the house styles. But are the circuit diagrams, even when drawn in accordance with B.S. 530, as informative as they might be? And do they not represent far too much wasted time and effort in the drawing offices? It is illuminating to compare traditional British circuit conventions and symbols with those now coming generally into favour on the Continent, more especially in Austria and Germany. The best European circuit diagrams (and it must be admitted immediately that there are many very bad ones) may look strange and unattractive at first to British eyes, but they offer substantial advantages both to those who must pay for their production and to the ultimate user, once he has familiarized himself with their codes and circuit symbols.

In British practice, the resistor-possibly the most common of all components-requires upwards of nine separate lines (see Fig. 1 (a))-and overenthusiastic draughtsmen, proud of their ability to produce a mathematically correct zig-zag, not infrequently extend them to as many as seventeen or eighteen bends. If the wattage reading of the
resistor is to be indicated, it must be written out alongside the symbol, adding time and clutter to the diagram. The unadorned rectangular box of the Continental diagram, Fig. 1 (b), is not only simpler to produce (particularly with a stencil), but opens the way to providing wattage information with a minimum of effort. Fig. 1 (c) shows a commonly used code which can quickly be added to the box, reproduces well and requires no extra space.

Fixed capacitors or condensers (for, despite all the efforts of the powers that be, the old term still marches gaily on) are drawn basically similar the world over (Fig. 2 (a)); but the overseas draughtsman seems much more inclined to throw in additional information for good measure. In the United States, it is common practice to indicate the correct connection for the "earthy" side (outer foil) by using a curved line on one side: Fig. 2 (b). On the Continent, the correct voltage rating is often indicated by means of simple codes; a representative code is shown in Fig. 2 (c). The objection may be made that, when reproduced by printing processes, small dots may appear or disappear accidentally; in practice this would seem to be no more bothersome than the many other potential sources of error.

The widespread Continental adoption of the nanofarad unit of capacitance ( $\operatorname{lnF}=1000 \mathrm{pF}$ or $10^{-9}$ farad) is yet another valuable aid in cutting down


Fig. 2. How the familiar two-line capacitor symbol can be adapted to provide additional information: (a) standard; (b) indication of outer foil connection; (c) a representative voltage rating code as used on the Continent.


Fig. 3. Coils may look like (a), but to draw this symbol accurately is not simple-one way is shown in (b) which is not so quickly accomplished as the straightforward half-circles (c) now widely used on the Continent and in tre United States
drawing time and circuit clutter when showing component values on circuit diagrams rather than relegating them to a separate component list-a practice which, though leading sometimes to overcrowding of the diagram, is generally popular with service and maintenance engineers, especially where the complete wattage or voltage specification can be given. More and more drawings for service manuals, British and foreign, now include on the main diagram the valve pin numbers and typical check voltages, as well as component values. With the nanofarad, such values as $0.003 \mu \mathrm{~F}, 0.02 \mu \mathrm{~F}$ need be shown only as $3 n, 20 n$, retaining $\mu$ for values of $0.1 \mu \mathrm{~F}$ and above. Incidentally, why do so many firms persist-despite paragraph 36 of B.S.530-in adding $\mu, \mathrm{F}$ or H after every value? It is a long time now since even the Royal Navy abandoned its beloved jars of capacitance.

Then there is the question of inductive components. Here Continental and American practice scores on production time, if for no other reason Fig. 3 (a) shows the traditional British symbol for all forms of inductors. To produce this accurately on the drawing board requires a procedure along the lines indicated in Fig. 3(b). Compare this with the symbol now finding increasing favour on the Continent and in the United States-Fig. 3(c)which is just a series of half circles but which is every bit as distinctive as the traditional symbol (often, in fact, clearer when reproduced on a small printing block which may cause the small loops of the British symbol to fill in).

Audio-frequency and mains transformers take a considerable time to draw accurately with the loop system of Fig. 4(a), particularly by draughtsmen whose ambition it appears to be to include almost as many symbolic turns as there are real ones. The Continental style, Fig. 4(b) may look a little uglier at first-but how much easier to produce! And immediately distinguishable from other types of transformers.

To come down to earth (or chassis) on the Continent requires only one small line-Fig. 5(a)compared with our four (Fig. 5 (b)). Indeed the main chassis line is frequently omitted altogether, again simplifying production, though-to some British eyes-often resulting in a rather untidy and difficult to follow circuit diagram.

The sum.total of the time spent in British drawing

Fig. 4. Will export orders wait while your drawing office painstakingly produces the mains transformer (a)? Your European rival may beat you to it with the simply produced (b).

1
(a)

(a)

(b)


Fig. 5 The single line of (a) takes less time to drow than the four lines of (b).
(b)


Fig. 6. Sketch (a) may seem a distinctive cathode, but (b) is quicker to drow, and in the form (c) provides additional information.
offices in producing carefully shaped valve cathodes (Fig. 6(a)) in accordance with British Standardssensibly not done, one notices, by Wireless World -would surely make any organization and methods man come down heavily in favour of the Continental short-cut (Fig. 6(b)). A useful additional feature of this symbol is that it can easily be adapted to indicate that a particular cathode is common to other parts of a multiple valve (Fig. 6(c)).

One should perhaps not be too hard on British traditional conventions: in the best examples the circuit diagrams look good, are easy to follow, and keep many draughtsmen contentedly exercising their skill - even if so many of the lines they draw are unnecessary and add nothing to the information conveyed to the user. But keep those Time and Motion Study people out of the drawing offices, or many firms may finish up-despite B.S. 530 -adopting some at least of the Continental practices.

## CLUB NEWS

A mobile rally is being organized jointly by a number of clubs in the southern counties for July 17th. It will be held at Beaulieu Motor Museum, near Southampton Control stations G31VP/A (1980kc/s) and G2HIF/ $\mathrm{A} /(144.13 \mathrm{Mc} / \mathrm{s})$ will be operating from $10.30 \mathrm{a} . \mathrm{m}$. Programmes of the Southern Counties Mobile Rally, costing 6 d , are available from R. Bassett, 42 Northam Avenue, Shirley, Southampton.

Prestatyn.-Meetings' of the Flintshire Radio Society are now held at the Ffrith Hotel, Ffrith, Prestatyn, at 7.30. At the meeting on July 4th, T. A. P. Colledge, of the G.P.O., will talk on subscriber trunk dialling. On the 25 th the club is holding a $160-$ metre d.f. hunt.

Tumbridge Wells.-The second of a series of talks on 2-metre operation will be given to members of the West Kent Amateur Radio Society by the president, W. H. Allen (G2UJ), on July 22nd. The club meets on alternate Fridavs at 7.30 at Culverden House, St. John's, Tunbridge Wells.

## "The Eyes of the Few"

THE many Wireless World readers who served in the wartime R.D.F. system (radar to newcomers) would be well advised to take this book* on holiday with them. If, like the author-Daphne Carne, née Griffiths-they spent 1940 in the active Kent and Sussex sector, they may find it almost unbearably nostalgic. Unpretentiously told, this account of the experiences of a W.A.A.F. R.D.F. Operator is vivid and exciting. Those who were not there to see for themselves should find it not only entertaining and at times moving but also informative concerning an essential and none-too-well publicized part in the saving of the world from Nazi domination. M. G. S.

[^4]
# B.B.C. SATELLITE STATIONS 

THE coverage of the present 23 B.B.C. television stations is about $98.8 \%$ of the population, though this figure includes people in some areas where reception is, at times, subject to severe interference. When the 14 stations (marked stage 1 on the map below) announced last year, come into service by March 1962, a further 200,000 people will come within the service area and about a million will have an improved service. The P.M.G. has now given "approval in principle" to the second stage of the B.B.C.'s plans for extending and improving


# Pickup Design 

## By J. WALTON*

## relative importance of various factors

VARIOUS excellent articles have appeared from time to time on one or more of the factors mentioned here, so that it is not proposed to enter deeply into these factors in themselves, but rather to try to indicate their relative importance.

This is attempted as a result of observing how such articles sometimes cause those who do not have the facilities or time to make quantitative comparisons to attend to a particular feature in a manner that is disproportionate to its overall effect on the performance of the gramophone.

## Pickup Arm Length and Stylus Tip Radius

I am starting with "tracking error" since this seems to be a common source of care and of a desire for longer pickup arms in systems with a far greater source of distortion, namely the size of stylus tip employed.
Now if we accept the maxim that a chain is as strong as its weakest link, and we also wish to play the last ten minutes or so of our precious 1.p. records, then we must consider the fact that the size of the stylus tip is the factor causing the greatest distortion in any type of gramophone pickup.
The distortion due to tracking error is given by H. G. Baerwald ${ }^{1}$ as

$$
\eta_{2}=\frac{\mathrm{V} \alpha}{u} \times 100
$$

where $\eta_{2}$ is the percentage second harmonic distortion, V the peak recorded velocity, $\alpha$ the tracking error in radians and $u$ the groove speed.

Let us then consider a moderately large signal at the inside of the record where tracking error will do its worst: say $5 \mathrm{~cm} / \mathrm{sec}$ r.m.s. lateral velocity when the diameter is $4 \frac{3}{4}$ in on a record rotating at $33 \frac{1}{3}$ r.p.m., and a tracking error of 4 degrees. Then the groove velocity

$$
\begin{aligned}
u & =\frac{4.75 \times 3.14 \times 2.54 \times 100}{3 \times 60} \\
& =21 \mathrm{~cm} / \mathrm{sec} \\
\text { so that } \eta_{2} & =\frac{5 \times 1.41}{21} \times \frac{4}{57} \times 100 \\
= & 2.3 \%
\end{aligned}
$$

which, after correcting for the recording characteristic,

$$
\approx 1 \cdot 3 \% \text { (second harmonic) }
$$

Now, according to H. E. Roys ${ }^{2}$, the main component of the lateral tracing distortion is given by

$$
\mathrm{V}_{\mathrm{D} 3}=\frac{3(\pi r f)^{2} \mathrm{~V}^{3}}{4 u^{4}}
$$

where $\mathrm{V}_{\mathrm{p} 3}$ is the third harmonic distortion velocity, $r$ the stylus tip radius and $f$ the frequency.

And, if $\mathrm{V}_{\mathrm{D} 3}$ is small, this is approximately equivalent to

$$
\eta_{3}=\frac{3(\pi r f \mathrm{~V})^{2} \times 100}{4 u^{4}}
$$

where $\eta_{3}$ is the percentage third harmonic distortion.
Take the same signal of $5 \mathrm{~cm} / \mathrm{sec}$ and the smallest available stylus radius of 0.0005 in (which according to usual manufacturing tolerances is likely to be nearer 0.0006 in). Since the distortion here depends on frequency, we will take a moderate $3 \mathrm{kc} / \mathrm{s}$.

Then $\eta_{3}=75 \pi^{2} \times$

$$
\begin{aligned}
& \frac{(0.0006 \times 2.54 \times 3000 \times 5 \times 1.41)^{2}}{21^{4}} \\
& =4 \%
\end{aligned}
$$

which, after correcting for the recording characteristic,

$$
=1.6 \% \text { (third harmonic) }
$$

i.e. with $5 \mathrm{~cm} / \mathrm{sec}$ r.m.s. velocity at $3 \mathrm{kc} / \mathrm{s}$, the distortions from a 4 -degree tracking error and from a nominally $\frac{1}{2}$-thou stylus are about the same.
Moreover, we may say that below such figures where tracking and tracing distortions are equal they are both of little importance for practical hi-fi purposes.

Distortion figures above this point, however, are attained at an enormously greater rate in the case of tracing distortion than in the case of tracking error distortion.
For instance by the time one "reaches" even


Fig. 1. "Effective stylus tip radius" (r) for lateral modulation.
$10 \mathrm{~cm} / \mathrm{sec}$ at $4 \mathrm{kc} / \mathrm{s}$, we get $12 \%$ tracing distortion and $2 \cdot 6 \%$ tracking-error distortion, even for a nominally $\frac{1}{2}$-thou tip, and a nominally 0.0007 -in tip gives $20 \%$ tracing distortion. However, I must say that where the values calculated are extremely large, they are not so large by measurement.

Since an 8 -in arm can give a $2 \frac{1}{2}$-degree tracking accuracy, and certainly better than 4 degrees even with manufacturing tolerances, there would appear to be no advantage in exceeding 8in unless the stylus tip radius can also be reduced to well below 0.0005 in .

Whilst considering tracing distortion, I would like clarification upon whether or not the change in the effective stylus tip radius with modulation adds another variable to the equations. Lateral modulation motion would appear to me to give

[^5] motion.
significance to the dimension $r$ (see Fig. 1) which I call the " effective stylus tip radius." This is $1 / \sqrt{ } 2$ times the actual radius R for a 90 degree unmodulated groove, but is modified to a higher value by the pinch effect.

Thus the effective stylus tip radius would appear to me to be an extra variable in the lateral case, but not in the stereo case of a 45 degree movement. However, the effective tip radius is the larger (equalling the actual radius) for stereo, and this must be a factor giving greater distortion in stereo reproduction, unless, of course, the correspondingly smaller actual tip radius is used.

I would conclude this section by saying that since only lighter tracking-weight pickups allow of smaller stylus radii on account of their effect on wear, a gramophone pickup must primarily be assessed on its genuine tracking weight and tip radius.

## Needle Trail, Vertical Tracking and Stereo

Again, if the effect of tracking error was serious, we should be in greater trouble with vertical tracking errors in stereo pickups than has actually appeared to happen, since a difference of 23 degrees of "vertical" tracking angle occurs between the cutterheads of one manufacturer and another, to say nothing of differences between different pickups, and a further 15 degrees change between the top and bottom records in a changer.

It is not my intention to belittle these discrepancies. On the contrary, I am at present trying to correct the vertical motion of a flexible transmission arm similar to that described in the April 1959 issue of Wireless World (p. 182). Since the whole stylus arm flexes here, an effective centre of rotation can be found which determines the instantaneous direction of motion of the stylus tip.

For the arrangement shown in Fig. 2; as used in a typical mono head, the "vertical" motion is at 30 degrees to the vertical. This is unimportant in the mono case since the actual forward movement due to the lift produced by pinch effect is too minute to add any significant further tracing errors, but the departure from vertical motion must be very much reduced in a stereo pickup.

Let me take this opportunity of discussing the effective centre of rotation in relation to cantilever stylus arms in general, and needle trail in particular.

Any normal cantilever arm will have an effective centre for the stylus motion, and not only will this govern the direction of "vertical " motion of the stylus, but it will also give the effective trail of that stylus.

Now it has long been considered advantageous to have a small amount of needle trail, and some cantilever styli have been criticised on this basis when
 negative rake.
the rondel has been mounted with an apparently "negative rake" (as in Fig. 3). But surely the original purpose of " positive rake" as it was applied to the older type of needle (as in Fig. 4(a)) was to ensure that any forward drag at the stylus tip did not cause the stylus to dig further into the groove (as in Fig. 4(b)) with a cumulative effect upon this drag and its resultant digging.

The cantilever shown in Fig. 3 is, however, completely exonerated if its flexing centre is at C as shown, for it then has in fact an effective positive rake of about 70 degrees. This does not, however, exonerate the cantilever from having a 20 degrees to vertical "vertical" motion if it is to be used for stereo.

## Vertical Compliance and Tracking

It could be imagined that the "hill and dale" aspect of stereo where there is no restriction in the vertical downzvard direction could lead to a condition where the groove receded from the stylus at a more rapid rate than that at which the compliance could cause the stylus mass to follow. It might be thought also that mechanical resistance to motion of the stylus arm could give an even more stringent tracking condition, and that all these problems were something not encountered in lateral recording in which two groove walls "direct" the stylus.

It can be shown, however, that there is no essential difference between the lateral and vertical cases, since not only are the groove walls at 45 degrees to both the vertical and horizontal, but also the friction, particularly at low tracking weights, is too low to make any appreciable difference between the lateral force required to make the stylus ride up out of the groove and the vertical force required to make the stylus leave the groove in a frictionless manner as in the "hill and dale" case.

Consider Fig. 5. For static equilibrium the upward reaction force $F$ on the stylus tip of mass $m$ must equal the tracking weight, and $m$ must also be subject to an equal downward force $F$. If the groove is suddenly lowered then the acceleration $\alpha$ of the


Fig. 4. Older-type styli with positive (a) and negative (b) rakes.
stylus tip mass $m$ is given by $\alpha=\mathrm{F} / m$. Thus the acceleration is not dependent upon the compliance but on the tracking weight and stylus mass, provided that the displacement of the stylus is too small to appreciably alter $F$.

In fact the groove displacement at $.8 \mathrm{kc} / \mathrm{s}$ and a peak recorded velocity of $22 \mathrm{~cm} / \mathrm{sec}$ is $22 \div(2 \pi$ $\times 8000)=0.00044 \mathrm{~cm}$. Now consider a $1-\mathrm{gm}$ pickup with an effective stylus tip mass of $\frac{1}{2} \mathrm{mgm}$ and a compliance of $25 \times 10^{-6} \mathrm{~cm} /$ dyne. The static displacement of the stylus as the pickup is placed on the record is 0.025 cm . Thus the above condition that the groove displacement be too small to appreciably alter the 1 gm force on the stylus is fulfilled. The stylus acceleration is thus 2000 g , which is nearly twice that of the groove $(2 \pi \times 8000 \times$ $22 \mathrm{~cm} / \mathrm{sec}^{2} \approx 1100 \mathrm{~g}$, and the tracking possibilities thus remains as for lateral modulation.

Mr. R. W. Bayliff in as yet unpublished work relating to the Decca stereo pickup has pointed out that a stiffer vertical movement can give a greater tracking capability, for a given effective stylus tip mass and tracking weight, in that portion of the upper middle register where the vertical compliance


Fig. 5. Stylus tip mass (m) in static equilibrium. $F$ is equal to the tracking weight.
resonates with the effective tip mass. But this does not alter the main argument, nor that a greater compliance gives a greater tracking possibility at all frequencies below this resonance and that a lower tip mass gives better tracking above this resonance.

If resistance is now introduced into the stylus arm, this will have its greatest effect at maximum velocity (zero acceleration with a sine wave) and have no effect at zero velocity (maximum acceleration). Thus the vertical tracking condition is as before, and provided that the maximum force due to resistance is not greater than that due to stylus inertia or cantilever stiffness then no additional tracking weight is required.

It might be said, however, that to cope with a peculiar case where the effects of maximum velocity, acceleration and displacement all occurred together, then a tracking force corresponding to the sum total would be required.

There would also appear to be the possibility of an integration of upward signal impulses by the momentum of the head so as to require extra tracking weight thereabouts. This of course applies also in lateral recording in the form of integration of impulses not only of vertical pinch effect movements but also of extra forward drag due to modulation, this integration being converted into side thrust that tends to push the stylus out of the groove so as also to require extra tracking weight. However, due to


Fig. 6. Simplified mechanical impedance anologue of a flexible cantilever arm pickup. Typical impedance values at $10 \mathrm{kc} / \mathrm{s}$ are shown.
the pickup arm geometry, the side thrust here will only be about one-fifth of the forward drag. Thus although the stylus arm resistance may be limited in magnitude to, and entirely out of phase with, the other factors of mechanical impedance for sine wave motion, then even in that motion, the resistance and impedance could conceivably have a combined effect in their integrated sum.

And we could probably continue " in ever decreasing circles" discussing this and that smaller and smaller points, but the problems of producing a high-fidelity stereo pickup do not, I think, yet warrant this.

## Groove Speed and Record Wear

Since most record deformation normally takes place at points of maximum acceleration or displacement, it would appear therefore that the inclusion of a substantial amount of resistance in the stylus arm will not decrease the life of a record.

I might mention that the increase with age of the resistance of p.v.c. must be taken into account in most of the pickups in use. This increasing resistance may also alter the effective stylus impedance at high frequencies as shown by Fig. 6.

When considering record wear in a theoretical way we have usually started from Hertz's equations for a static indenter in the elastic range, and a conception of record wear is then evolved around the " mean pressure" under the indenter (thus the pressure is assumed to be inversely proportional to the square of the stylus radius). Now not only does this assume a direct relation between mean pressure and wear, and so does not take into account any wear due to the greater impact of a larger stylus radius in a smaller groove curvature (i.e. under conditions of high tracing distortion) but it does not even take into account the linear speed of travel of the stylus in an unmodulated groove.

I hope to be able to present some experimental results in the near future on these last aspects and their general relation to the gramophone record. Results so far obtained indicate that recourse to lower turntable speeds would be severely detrimental not only to quality (distortion varying inversely as the fourth power of the groove speed) but also to record life, and that other means of gaining playing time are both possible and preferable.

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2 R.C.A. Revierv, Vol. 10, June 1949, p. 254.

# Ohm's Law and Negative Resistance 

-AND THE LAW THAT KIRCHHOFF FORGOT TO INVENT

By "CATHODE RAY"

SPPECTATORS of the recent duel between D. L. Clay and myself in the correspondence columns no doubt noted with interest Mr. Clay's shrewd thrust with my own weapon dated August 1953. To alter the metaphor, he used my own voice to pronounce me dimmer than a beginner. Must I accept this unflattering assessment, or alternatively eat my 1953 words? The dilemma is unattractive. Being a generous opponent, however, Mr. Clay invited further explanation. This course (assuming the Editor is equally generous with space) I elect to adopt. Quite apart from the obvious possibilities of a verbal smokescreen for evasive action, further explanation appears to be justified for the following reasons. With youth now at the helm, what Cathode Ray said in 1953 must seem almost as far-off as what Gladstone said in 1888. Next, in a misguided attempt to be brief, what I said (or didn't say) in last February's issue evidently left room for Mr. Clay-and maybe others-to find obscurities and contradictions. And the whole thing has convinced me that Ohm's law is even more treacherous than I thought, and that is saying a lot.

First of all (summarizing the 1953 contribution) we must say what we mean by "Ohm's law." My guess is that what most people mean is

$$
I=\frac{E}{R}
$$

(or its equivalent, $\mathrm{E}=\mathrm{IR}$ or $\mathrm{R}=\mathrm{E} / \mathrm{I}$ ) in which E is the e.m.f. in volts required to drive a current I amperes through a resistance R ohms. Though undoubtedly a useful piece of information, this would have looked very strange indeed to Dr . Ohm, who would have been at a loss to account for his name being attached to it, since volts and amperes


Fig. I. When a current/ voltage graph is a straight line passing through the origin, it refers to a resistance of the kind known to Ohm .
had not been invented in his lifetime and the ohm was a unit of wine, equal to about 40 gallons and therefore presumably beyond the means of a struggling teacher. Even the concepts of e.m.f. and resistance would have been novel to him. So much so that it is not easy for us, saturated in "Ohm's law," to follow just what it was that Ohm discovered. Put into modern terms, it seems to have been that the ratio of e.m.f across a conductor to current through it (i.e., its resistance) does not vary with the amount of current, provided the temperature is constant.
Note that our "Ohm's law"- $\mathrm{R}=\mathrm{E} / \mathrm{I}-$ says nothing of the kind. For all it knows, R may be variable. In these days of semiconductors it often is. But Ohm's experiments were carried out on metal wires, and the constancy of their resistance has been confirmed within very much closer limits than were possible with his crude apparatus.
It is probably too late to make "Ohm's law" mean what Ohm meant, which is true for metals but not for semiconductors or insulators. We call metallic resistances "ohmic" or "linear" (because their current/voltage graphs are straight lines passing through the origin). But "Ohm's law" in present-day usage is simply a convenient formula, relying on a system of units Ohm never knew, and true for any kind of conductor, linear or otherwise. It can also be regarded as a definition of resistance.
However, the thing is not quite as simple as that, because sometimes "Ohm's law" must be understood to imply the "law of Ohm" (as we may call what Ohm meant). For instance, an elementary exam paper might say "The current through a $500-\Omega$ resistor is 0.3 A ; use Ohm's law to find the voltage across it." This obviously means the well-known formula, and it would make no difference to the answer if the resistor disobeyed the "law of Ohm." But if the question were "When 150 V is applied to a resistor the current is 0.3 A , what is with 40 V ?" one would have to assume the resistor obeyed the "law of Ohm" to be able to answer it at all. A knowledge of the system of units used is unnecessary, whereas the answer to the first question would depend on the units (e.g., it could be $150 \mathrm{~V}, 0.15 \mathrm{kV}$, $15,000,000,000$ e.m.c.g.s. units, or 0.5004 e.s.c.g.s. unit).
There are some other circumstances to be understood. Ohm's experiments were carried out with d.c., and that is generally taken for granted in connection with "Ohm's law" too. Fig 1 shows current/ voltage graphs for two resistances, which can both be recognized as ohmic by their straightness. It doesn't matter how much e.m.f. is applied (so long as the temperature is not altered appreciably); its ratio to the current gives the same resistance every time. Reversing E (i.e., multiplying it by -1 ) reverses I too, so we have the continuations towards the bottom left.
Fig. 2, which is the sort of result we might get with


Fig. 2. Many circuit elements now in common use have resistances that do not follow the pattern of Fig. I. Though non-linear to d.c., they may be at least approximately linear to a.c. within limits, os for example $P$ to $Q$ here.
an ohmic resistor in series with a germanium rectifier, is quite different. $E_{1} / I_{1}$ gives one value of resistance, which anyone ignorant of non-ohmic resistances would assume to be representable by the dotted line OP . Increasing the voltage to $\mathrm{E}_{2}$ causes a disproportionate current increase to $I_{2}$, so the calculated resistance $E_{2} / I_{2}$, represented by $O Q$, is less. Neither of them is the resistance most likely to interest us in this more sophisticated era, when E might consist of a steady voltage half-way between $\mathrm{E}_{1}$ and $\mathrm{E}_{2}$, plus an alternating voltage with peak value $\left(E_{2}-E_{1}\right) / 2$. The latter would alternate between $\mathrm{E}_{1}$ and $\mathrm{E}_{2}$, and the corresponding part of the current would alternate between $I_{1}$ and $I_{2}$. So far as this current is concerned, the resistance (being represented by the nearly straight line PQ passing through the a.c. origin) is almost ohmic, and less than either of the d.c. values. It is known as the a.c. or incremental resistance. The poor beginner is usually left to guess from the context whether "resistance" means the d.c. or the a.c. kind.
The next complication is that the circuit may contain reactance. In point of fact, it is bound to; but what I mean is that the reactance may be enough to have an appreciable effect on the amount of a.c. flowing. Now there are various ways of handling this. A common one in elementary textbooks is to produce the following (or something equivalent) as "Ohm's law for A.C.":

$$
I=\frac{E}{\sqrt{R^{2}+\left(\omega L-\frac{1}{\omega C}\right)^{2}}}
$$

Having now gone on record as accepting, however reluctantly, "Ohm's law" as meaning "I = E/R", I will no doubt be exposing myself to attack from Mr. Clay if I object to the above on the ground of its being outside the experience of Dr . Ohm. In self-defence I might claim that it is wildly remote from his experience. But my main objection is that it brings in a new principle. To teach beginners that reactance is just a special kind of resistance that applies only to a.c. (as some books do) may make it look easier at first but is likely to make things more difficult later on. I would be inclined to emphasize that reactance is something quite different
from resistance, even though it is reckoned in ohms and mixed up with it in impedance. The reason is that resistance is essentially something that takes any electrical energy it can lay its hand on as an outright gift, whereas reactance accepts it only as a short-term loan, scrupulously paying it back in full within a single a.c. cycle. (Idea for a chapter title in Cathode Ray's Monster Nursery Book of Electrical Theory-"The Story of the Bad Mr. R and the Good Mr. X." Including, of course, "Pat-acake, pat-a-cake, Vector's Man; Turn it through half pi and mark it with ${ }^{\prime} \mathrm{j}$ "). Anyway, just now resistance is enough to be thinking about, without complicating the issue with reactance.

For we are coming to the bit that was turned against me. I said (in 1953) "that it is no good trying to apply Ohm's law to a circuit containing an e.m.f.-at least, not without allowing for the e.m.f. If $\mathrm{E}=\mathrm{IR}$ were applied to find the voltage between the terminals [in Fig: 3 here] the answer would be $0.5 \times 20=10$. But the measured voltage would be 14. The reason for this discrepancy is obvious, and even a beginner would have to be rather dim to fall into the trap." So I suppose it really was asking for trouble to offer (in 1960) the diagram reproduced here as Fig. 4, and, having requested the audience to look between the terminals eastwards to see a positive resistance $R$, proceed to turn them around facing the battery and suggest that it appears to be a negative resistance-even with the proviso "not an ohmic one in this case!". Being more modest than you might think, I had not imagined the possibility of anyone following my utterances with such attention as to be able instantly to quote any of them made within at least the last seven years. I hope it will be a lesson to me.

Mr. Clay-for none other than he is the prodigy alluded to-skilfully turned aside my defensive stroke (which was to ask what gave him the idea that I was applying "Ohm's law" to Fig. 4) by a thrust cunningly disguised as an apology. He said he was sorry he had incorrectly assumed I had used "Ohm's law" to show that a generator is equivalent to a negative resistance, but it was the only way he knew of obtaining a value of resistance from a voltage and current. And to increase my discomfiture he added that he also assumed "non-ohmic" meant "non-linear," but, as the current did not affect the voltage in any way, perhaps "nonresistive" would have been better.
"Touché," I believe, is the appropriate expression. Not that I concede having said anything actually wrong, but an explanation which left room for such

comments can't have been a very good (i.e., foolproof) explanation of negative resistance. The fault was not in what I said but in what in my haste I didn't say to guard readers against the treacheries of "Ohm's law."

So let us begin with Fig. 4 again. Looking at what we have between the terminals on the righthand side, we recognize exactly the situation we met in Fig. 1. We measure the voltage between the terminals, and find it to be equal to E . We also find a current I flowing from the positive to the negative terminal, which is conventionally the same direction. So the current has the same sign ( + or -, depending on which terminal we choose as our reference) as the voltage. We therefore have the data for plotting the two points $P$ and $\mathbf{P}^{\prime}$ in Fig. 5. If we had the opportunity to vary E we would be able to plot more points, which (assuming R to be "ohmic," in the sense already defined) would all fall on the sloping straight line marked "R."
Now let us about-turn and consider the same pair of terminals from the opposite direction. Performing the same two measurements, we would find the


Fig. 5. $P$ and $P^{\prime}$ are points on a current/ voltage graph of $R$ in Fig. 4. $Q$ and $Q^{\prime}$ are points measured in the same way on $E$.
direction of the current reversed relative to the voltage, so the results would have to be represented in Fig. 5 by $Q$ and $Q^{\prime}$. Since $P$ and $P^{\prime}$ represented $R, Q$ and $Q^{\prime}$ must represent $-R$. That is what I meant by saying that the battery was equivalent to a negative resistance. The logic of the conclusion seems inescapable. Moreover, the positive resistance $R$ is a dissipator of power; its opposite or negative should therefore be a source of an equal amount of power, and that is precisely what the battery is. And, just as according to Kirchhoff's Second Law the total voltage around a circuit (reckoning a current-carrying resistance as a negative voltage source) must be zero, in an alternative view


Fig. 6. Two types of current/voltage curve showing negotive resistance.

the total resistance around a circuit (reckoning a voltage source as a negative resistance) must be zero. And, just as the negative voltage source in Kirchhoff's law is a rather peculiar one, depending entirely on the current, so my negative resistance is (as Mr. Clay pointed out) an analogously peculiar one. If we had the opportunity to vary $R$, the additional points would not lie on the straight line joining $\mathrm{Q}^{\prime}$ to Q . They would, of course, be on vertical lines through $\mathrm{Q}^{\prime}$ and Q . That was why I warned readers not to expect an ohmic resistance. And since this is only an extreme case of the sort of thing we found in Fig. 2, I am sticking to that, rather than Mr. Clay's "non-resistive." For the equivalent negative resistance of the battery is found by "Ohm's law," in just the way he is used to. But I ought to have warned him (and less knowledgeable readers) that an equivalent resistance is one calculated by applying "Ohm's law" without regard for the rule (needed for finding real resistances) about excluding sources of e.m.f. For example, the equivalent or apparent resistance between the terminals in Fig. 3 is $14 / 0.5=28 \Omega$. The $4-\mathrm{V}$ battery absorbs the same power and has the same voltage across its terminals as an $8-\Omega$ resistor at that current. In fact, unless one was allowed to vary the current one could not distinguish between the battery and a resistor by any electrical test.

Just as the source of current-opposing e.m.f. in Fig. 3 is thus equivalent to a positive resistance, the source of current-assisting e.m.f. in Fig. 4 is equivalent to a negative resistance, as I hope all can now see. The point I was trying to make in this way is that a negative resistance is in effect a power source. I did not intend to convey that all negative resistances vary with current in the same way as the battery in Fig. 4 does. Practical negative resistances are negative over only a limited range of current and voltage, and fall into the two classes shown in Fig. 6, distinguished for obvious reasons as " $N$ " and " $S$ " types. Because of this limited range they are of interest chiefly as regards a.c. (Compare Fig. 2.) Certain valves and semiconductor devices* have characteristic curves of one or other of these types. Similar effects can be produced artificially by positive feedback in an amplifier. For instance, an a.c. negative resistance is found between the terminals in Fig. 7, and a tuned circuit connected to them is set into oscillation thereby, provided that its positive resistance isn't enough to make the total resistance positive.

[^6]Mr. Clay and perhaps spectators around the ring may be thinking-even openly saying-that in order to get myself out of an awkward position I have been making up the rules as I go along. In 1953 I said any fool can see that one mustn't include e.m.fs when applying "Ohm's law," and in 1960 I said of course that is only for "real" resistances; for "equivalent" resistances one can. Where is there a rule about there being different sorts of resistance?

The British Standard definition of resistance (B.S.205:1943, No. 1276) says "That property of a body by virtue of which it resists the flow of electricity through it, causing a dissipation of electrical energy as heat. It is equal to the constant difference of potential applied to the ends of the body divided by the current which it produces when the body has no e.m.f. acting therein."

That is not only resistance in the strictly real sense, but is confined to d.c.-note the constant d. of p. That obviously doesn't get us very far in these days, but lower down (No. 1283) there is a definition of "effective resistance"
"Of a circuit element with alternating current. The component of the terminal voltage in phase with the current divided by the current. The power dissipated in heat divided by the square of the current."

This is really two definitions, defining quite different things. The second one agrees with No. 1276, and in fact could be substituted for it as an a.c./ d.c. definition. For the power $P$ in a d.c. resistance is given by EI , and dividing that by $\mathrm{I}^{2}$ leaves $\mathrm{E} / \mathrm{I}$, which is "Ohm's law" again. I am assuming that the definers meant strictly the power dissipated as


Above: Fig. 8. What is the a.c. resistance between the terminals?

Right: Fig. 9. And what is the resistance between these terminals?
heat in the circuit element referred to. That is rnther an imbortant point, as we see if we consider Fig. 8. What is the "circuit element" to which the terminas are connected? The primary winding? The whole transformer? Or the transformer and its load regarded as one element?
With any reasonably efficient transformer, the heat dissipated in the primary, or even the whole transformer, should be small. Most of the power fed in at the terminals would be dissipated in the load resistance. But if that is reckoned as a separate circuit element, its heat mustn't be counted. Its effect on the circuit between the terminals is solely as an e.m.f.-the e.m.f. induced in the primary by current flowing in the secondary circuit, which depends mainly on the load resistance. However, if the whole of Fig. 8 is deemed to be "a circuit element," then this e.m.f. is a purely internal arrangement for distributing the dissipation, and the definition is of the strict kind.

Note that heat is also dissipated in the iron core,
because currents are induced therein, and in the insulation because of charging currents between wires, and extra "skin-effect" heat in the wires because of e.m.fs induced in them. It would be hopeless to try to find the a.c. resistance of the whole thing accurately by counting up the ohms due to all these different effects; hence the idea of doing it in one by measuring the power and dividing it by the current squared. This gives the single resistance that would run away with the same power when the same current was flowing.

How do we measure this power? Since the definition refers to power dissipated as heat, we are committed to a calorimeter measurement, which is a messy, time-consuming and (except perhaps in the N.P.L.) inaccurate business. So we are strongly attracted to the first No. 1283 definition, which allows us to measure the electrical power fed in. In this circuit it would give the right answer. But not everywhere. Consider Fig. 9. Here, if the thing is doing its stuff, most of the power is radiated. To make the two definitions agree it would then be necessary to stretch the "circuit element" to include the entire universe, which seems rather far-fetched. Or, in case the heavy old electrical engineers who composed this definition begin to murmur that of course they didn't mean this new-fangled wireless telegraphy, we can quickly silence them by passing on to the example of an a.c. circuit feeding an electric motor. The second No. 1283 definition would include only the power dissipated as heat in the motor, which should be quite small; the second would include also all the mechanical power developed by the said motor, which could be many times greater.

These are only a few of the recognized varieties of resistance, more of which are defined in B.S.204.
To sum up: It seems that in its strictest sense the resistance of any circuit or part of a circuit concerns the power dissipated as heat therein. Measuring amounts of heat is something we definitely don't want to do. With d.c. we can avoid it by using "Ohm's law," provided we carry out a preliminary frisking to make sure the thing being measured has no concealed e.m.fs. With a.c. we measure the current in phase with the applied e.m.f., but if we exclude all e.m.fs we will find ourselves excluding almost everything but standard resistors. Most often, we want to include at least all the losses, whether they come into our circuit as e.m.fs or not. We may want to include all permanent departures of power, such as radio waves, light, sound, etc.
No doubt it was naughty of me-even with the praiseworthy object of reducing the thing to ultimate simplicity-to begin with a d.c. circuit, Fig. 4, without explaining unmistakably that this was just as a step towards the a.c. circuits of commerce, in which "equivalent" and "effective" resistances are established conventions. Another point is that even with a.c. resistance, counting effects brought in by e.m.fs, these e.m.fs usually bear some relationship to the input current, whereas E in Fig 4 is completely independent, like the battery in Fig. 3 (apart from its internal resistance). As I have shown, the principle involved here is not fundamental, but it would have been better to have pointed out that the battery didn't behave like a practical negative resistor if the current was varied. I hope, however, that it has been justified by the bringing to light of Kirchhoff's hitherto unknown Third Law.

The Editor does not necessarily endorse the opinions expressed by his correspondents

## Demonstrating Electron-spin Resonance

FOLLOWING the description ${ }^{1}$ by G. B. Clayton of a demonstration apparatus for electron-spin magnetic resonance absorption, readers may like to know of a simple modification to the circuit which enables a satisfactory estimate of the magnetogyric ratio, $\gamma$, to be made. The time required for the observation is such that it may easily be carried out in the course of a lecture, thus adding greatly to the value of the demonstration.
$\gamma$ is given by the relation $2 \pi f=\omega=\gamma \mathrm{B}$ and it is the ratio of the magnetic moment to the angular momentum of the electron, not the reciprocal of this ratio as given in the article. $f$ is the frequency of the radio-frequency oscillator, B the magnetic flux density at the sample. If $f$ is expressed in $\mathrm{Mc} / \mathrm{s}$ and B in weber $/ \mathrm{m}^{2}$ we have $\gamma=(2 \pi f / \mathrm{B}) \times 10^{6}$ coulombs $/ \mathrm{kgm}$
The frequency, $f$, may be measured in a convincing manner by means of a short-wave wireless receiver. A characteristic purring tone will be heard, representing a pair of absorption pulses repeated at $100 \mathrm{c} / \mathrm{s}$.
For the measurement of the resonance value, B, of the magnetic flux density an ammeter (reading r.m.s. current) is inserted in series with the Helmholtz coils and some provision is made for varying the current continuously. This can most conveniently be done by use of a "Variac." but failing that a rheostat will do. As the current is reduced the trace on the oscilloscope screen shrinks until, for a certain value, $i$, of the r.m.s. current, the extremes of the trace coincide with the peaks of the absorption line. Then the field, $\mathbf{B}$, is given by
$\mathrm{B}=4 \pi \times 10^{-7} \times(8 \mathrm{~N} / 5 \sqrt{ } 5 \mathrm{R}) \times \sqrt{ } 2 i$ weber $/ \mathrm{m}^{2}$ where $N$ is the number of turns on each coil and $R$ is the radius of the coils in metres.
The resulting value of $\gamma$ may be compared with the known value of the charge-to-mass ratio, $e / m$, for the free electron, namely $1.76 \times 10^{11}$ coulomb $/ \mathrm{kgm}$, since for the unpaired electrons in diphenyl picryl hydrazyl at frequencies above $5 \mathrm{Mc} / \mathrm{s}$,

$$
\gamma=e l m
$$

to a very good approximation ${ }^{3}$.
In an actual demonstration, with an oscillator frequency of about $24 \mathrm{Mc} / \mathrm{s}, \mathrm{I}$ obtained

$$
\gamma=1.7 \times 10^{11} \text { coulomb } / \mathrm{kgm} .
$$

The oscillator ${ }^{2}$ was found to work very well at this frequency without the capacitors $\mathrm{C}_{1}$ and $\mathrm{C}_{2}$, adjustment of the amplitude being made by varying $\mathrm{C}_{0}$.
The University, Edinburgh.
A. G. A. RAE
${ }^{1}$ Wireless World, Feb., 1960, p. 68.
${ }^{2}$ Wireless World, Feb., 1960, p. 70, Fig. 8.

- Garstens, Singer and Ryan, Phys, Rev. Vol. 96 (1954), p. 53.


## Transistor Tape Recorder Amplifier

I HAVE just read Mr. Blick's article in your April issue and I congratulate him on an elegant solution to the problem of the recording amplifier.
Although the leakage of modern electrolytic capacitors is reasonably low, the small residual magnetization of the head would be enough to introduce noise in the recording process and harmonic distortion in the playback process. A solution which avoids this is to use a doublewound transformer in place of the reactor $\mathrm{L}_{\mathrm{r}}$ in Mr . Blick's Fig. 1. The recording head directly connected across the secondary eliminates the leakage current although not, of course, switching transients.

It is probably worth while for high-quality work including a variable direct e.m.f. in the secondary circuit

(see accompanying diagram) and to adjust this for minimum noise when playing a virgin tape. This is also the condition for minimum distortion.

Salisbury Polytechnic, PHILIP F. RIDLER S. Rhodesia.

## Deeper Amplitude Modulation

MR. BLANCHARD (June issue) is no doubt correct in his facts and I have myself noticed all the effects, including the distortion. Increased percentage modulation and clipping also reduce the dynamic range, which is also observable. Furthermore, the operation of the a.g.c. system in normal sets will be affected by the increased modulation and is likely to reduce even further the dynamic range.

The practical result, in my view, would be unfortunate as it would be the fashioning of another nail ready for the coffin for medium- and long-wave amplitude modulation.
No medium- or long-wave service hinders international contacts and aids the spread of extremist nationalism but on 3 purely personal level I, and presumably Mr. Blanchard, find the Continental broadcasts more entertaining than those of the B.B.C. or else why roam the ether? If we really wans entertainment why surrender to the evidently ever-selfish motorists?
London, N.W.II.
L. STREATFIELD

## V.R.F./E.M. Car Radio

R. V. TAYLOR'S article in the June issue on v.h.f./f.m. reception in a car was interesting, but having experimented extensively with this myself, and decided that it was not the answer unless operating only in small areas relatively near to a transmitter, perhaps my remarks would be of interest.

I first started some eighteen months ago, exactly as Mr. Taylor suggests, by using a standard commercial tuner feeding into the a.f. and power pack unit of an ordinary car radio. The alterations necessary to do this were very simple, consisting merely of re-wiring the valve heater. The adjustable car-radio whip aerial was set at $2 \frac{1}{2}$ feet, and was in what was then the normal position for such an aerial-on the offside front wing.

One of the first things I discovered was the need for far better interference suppression on the car than was ever necessary for good medium- and long-wave reception, the main trouble coming from the voltage regulation. After quite a lot of trouble, this was cured, as
were other sundry crackles coming from unbonded parts of the car, although generator whine was never quite eliminated. Surprisingly enough, the ignition gave no trouble at all! When all this had been done, I found that the level of interference from other traffic was too high for comfort, but, of course, nothing could be done about it. However, the receiver worked fairly well, and was used for some time to find out whether the service area of the v.h.f. transmissions was large enough to warrant making up a receiver with better amplitude limiting. It was found with this tuner that the area within which the v.h.f. programme was of use (i.e., little fading or interference) was only a circle of about 20 miles radius. Outside this radius, in towns, flutter and distortion were much in evidence, although in the country reception was naturally better, but outside 30 mules reception was pretty poor anywhere, except, of course, on the tops of hills. Modifications were made to the tuner to improve limiting, but this only gave an increase of the good usable range to 25 miles, and the maximum usable range to 40 miles. If circles of these radii are drawn round the major transmitters on a map, the large gaps in coverage will be noted. Of course, these radii take no account of the topography, so that ranges in some directions may be better than others, but this will not materially affect the overall result.

The sensitivity of the tuner was not in question, it being possible to hear Sutton Coldfield on the South Coast, although the signal was quite useless for entertainment.
Another difficulty was that of tuning accurately, which had to be done if maximum a.m. rejection was to be achieved. Since it was obviously impossible to use visual indicators of any sort, tuning had to be done by ear, and with the slight fading experienced even in the best areas of coverage, this was quite difficult. In fact, I usually found it best to stop the car and then tune accurately.

Several Continental journeys were made with the v.h.f. tuner installed, and the situation there was even worse, due to the widespread use of low-power repeater stations. These had an effective range of about only 10 miles and when travelling on a motor-road this meant that it was necessary to re-tune (if a new station were avanlable, that is!) every 15-20 minutes.

All these factors contrast very unfavourably with medium- and long-wave reception, about the only advantage v.h.f. having over these for car use being that when a good signal is obtainable, it is free from interference from other transmitters. The fact that better quality is to be obtained from v.h.f. is not significant for mobile reception, unless listening is confined to periods when the car is stationary-hardly mobile reception! Fading on medium waves is not a problem until ranges of $100-$ 150 miles are reached, and then only at night, and so there are no gaps in the B.B.C. coverage of the U.K. If one can tolerate a little noise, it is possible to listen to Home or Light programmes while holidaying on the Continent, which certainly cannot be done with v.h.f. And tuning is far easier with a.m. than with f.m. if it has to be done aurally.

For these reasons I took the v.h.f. tuner out of the car some months ago, and reverted to the usual frequencies, with a certain amount of relief. Experiment with v.h.f./f.m. in a car by all means, but I don't think it's nearly as much use there as it is at home!

Tangmere, Sussex.
W. BLANCHARD

WHILST many of the considerations involved have been dealt with adequately in Mr. R. V. Taylor's article, this cannot be said of the treatment of interference generated by the vehicle itself. However effective the a.m. interference rejection of the sets may be, the remainder of the set must be very prone to interference conveyed by the power supply lead and probably interference radiated from the ignition system and other electrical components of the vehicle. Mr. Taylor makes no men-
tion of this and from long experience of this I can say that it is a very difficult problem.

Mr. Taylor is also, probably unintentionally, misleading regarding the B.B.C. v.h.f./f.m. coverage. At chimney-pot height what he says regarding the field strength laid down by the B.B.C. transmitters is no doubt true, but it is certainly not' true of the field strength at car radio aerial height. It is, for instance, quite impossible to get an adequate signal on the road between London and Birmingham, and the same applies in many other places, and a thorough survey of this has shown that for a satisfactory car radio set, it is necessary, however effective may be the v.h.f./f.m. car radio, for medium- and long-wave a.m. reception facilities to be provided as well.

Pinner, Middlesex.
W. CROSSLAND

## Power Transformer Design

IN his article on small power transformer design (June issue) Mr. Saull seems to have made an error in his example of a practical design. In step (c) he uses a figure of $216 \mathrm{~T} / \mathrm{in}^{2}$ for the LT1 winding whereas Table II shows that $17 \mathrm{~s} . \mathrm{w}^{\mathrm{w} . g \text {. enamelled copper wire winds } 280}$ $\mathrm{T} / \mathrm{in}^{2}$. Consequently the remainder of his calculations for the example are wrong.

Using the figure shown in Table II the LT1 winding will occupy 0.125 sq in , making the total for the two l.t. windings $0.24 \mathrm{sq} \mathrm{in} .\mathrm{The} \mathrm{remaining} \mathrm{space} \mathrm{for} \mathrm{the} \mathrm{h.t}$. winding is then 0.39 sq in, from which it is found in step (e) that the $\mathrm{T} / \mathrm{in}^{2}$ is 9,168 . The nearest wire gauge from Table II is therefore $34 \mathrm{~s} . w . g$., having a current carrying capacity of 66.5 mA . In consequence this winding will have a slightly lower copper loss, resulting in a small improvement in regulation and efficiency.
St. Leonards-on-Sea, Sussex. W'. E. THOMPSON
MAY I comment upon a statement by Mr. D. Saull in his article (June issue) "Power Transformer Design"?

Whilst not disputing that for mechanical reasons it may be preferable to wind the heavier gauge windings of a transformer first, an improved voltage regulation is not necessarily obtained for this winding arrangement.

The voltage regulation is a function of the sum of the primary resistance; $\mathrm{R}_{\rho}$, referred to the secondary terminals, and the secondary resistance, $\mathrm{R}_{3}$, that is, the effective secondary voltage drop $=I_{s}\left(\mathbf{R}_{p} / n^{2}+\mathbf{R}_{s}\right)$ where $n=$ $\mathrm{I}_{s} / \mathrm{I}_{\rho}=$ turns ratio.

The shorter mean length of turn of the secondary winding will reduce $\mathrm{R}_{s}$, however the resultant longer mean length of turn of the primary winding will increase $\mathrm{R}_{\rho}$, and, for equal primary and secondary winding areas, the effective secondary voltage drop will be unchanged.

Weymouth. A.D. WAITE

## The author replies:

Mr. W. E. Thompson is perfectly correct; I apologise for the error quoted by him which came about through misreading the comprehensive wire tables that I normally use when transformer designing. These tables include single and double silk and cotton covered wire, in addition to enamelled wire. The turns/in ${ }^{2}$ I quoted for $17 \mathrm{~s} . \mathrm{w} . g$. wire was 216 which is, of course, for double-cottoncovered wires. As Mr. Thompson points out, I should have used 289 turns $/ \mathrm{in}^{2}$ in my example.

It is refreshing to see that Mr. Thompson has read my article with a discriminating mind and eye and is not prepared to accept all he sees in print without first agreeing. I like to see this quality in a reader-it keeps up the standard of technical literature. Once again, sorry for my error Mr. Thompson!

The point brought up by Mr. A. D. Waite is, I believe, a controversial one amongst transformer designers. Some say the winding sequence should be h.t. winding, primary winding, and l.t. windings on the outside; others say a winding sequence of primary winding first, followed by
the h.t. windings, and finally the l.t. windings on the outside; a third group, which I support, believe that the best results are obtained by winding the 1.t. windings on first, the h.t. windings last, and positioning the primary in the centre of the transformer.

In support of the windings sequence I have chosen I will make these points-it is up to the reader to make his final choice.
(1) The purpose of my article was to produce a quick method of design; one that contained a minimum of variables; one that the design engineer could " get into " without too much preliminary digesting-otherwise he might just as well read a standard text book on transformer design.
(2) Positioning the primary between the l.t. and the h.t. windings, and consequently more to the centre of the lamination window, gives a smaller flux gradient across the window area. This results in a lower leakage induct-ance-consistent with the growing requirements of power supplies for transistorized, or partly transistorized, equipment.
(3) For design simplicity, the I.t. windings are required to carry current densities up to $2,000 \mathrm{~A} / \mathrm{sq}$ in. A winding on the extreme inside is approximately half the length of one of the extreme outside, hence approximately half the resistance. However, if the primary is in the centre its resistance will not alter a great deal if moved slightly inwards or outwards. Primary current is a function of secondary wattage and, with small power transformers, the l.t. windings are responsible for the lion's share.
H.t. windings are operated at lower current densities, and in the smaller variety of power transformers, their resultant primary wattage is the lamb's share.

Heating effects are also a consideration: $I^{2} R$, if $R$ is small, is half the answer that would have resulted if $R$ became 2 R. L.t. windings placed near to the core will lose heat, through the large surface area of the laminations, quicker than directly from a smaller surface area in contact with air, had the l.t. windings been placed on the extreme outside.

Now reflected resistance, $n=\sqrt{\mathbf{R}_{1} / \mathrm{R}_{2}}$, only holds good at $100 \%$ efficiency. With $86 \%$ efficiency, which is an average for small power transformers, not all of the secondary resistance is reflected into the primary winding!

We haven't discussed leakage currents between windings when the transformer is connected in circuit. These currents produce a chemical effect resulting in the enamel coating on the wire and other insulation deteriorating, an important factor in long-term installations such as instrumentation equipment used in atomic power stations.

My article was based on simplicity and therefore generalized. It contained a little error-perhaps inexcusable, but I have, I hope, tried to lead the engineer to the meat, cutting away the fat; and I say, "Brother, put these turns on and you will produce a transformer that is reasonable. You might be able to improve on it-but if time is money to you-I think that you will find it acceptable."
D. SAULL

## Wire Broadcasting

IN your article in the May issue describing the TV relay equipment you mentioned Capt. P. P. Eckersley's experimental sound relay where the electric light supply mains were used for programme transmission. Some years before Capt. Eckersley's experiments I applied to the Post Office, in association with Charles Melhuish, at that time proprietor of the Crapstone and Yelverton Electricity Supply Co., for permission to install on the company's supply a Multi-Programme Broadcast Relay using an h.f. carrier modulated by a number of lower frequency sub-carriers each accommodating a separate radio programme. Our plans were well advanced including the special mains spacing and h.f.c. arrangements required, and one of the receivers even included a form of a.g.c. which varied the gain by altering the h.t. voltage. It was not very effective!

Although we had the support of the late Sir Arthur de Freece then living at Dunster, our plans were turned down by the Post Office as they were held to infringe the B.B.C. monopoly. I bring this to your notice as, although there may have been previous proposals along these lines, of which I am not aware, ours did precede Capt. Eckersley's.
Haverfordwest. K. F. PONTTING-BAKER

## Radio Telescope

"UNBIASED" by "Free Grid," in your May 1960 issue, finds the name radio telescope inadequate and confusing and suggests that an entirely new word should be given to this instrument. I quite agree with him and since he has asked your readers to help in this "matter, I would suggest "electronic cosmoscope" or "radio cosmoscope."

Mombasa, Kenya.
K. C. FOLEY

## Medical Electronics

THE third International Conference on Medical Electronics opens at Olympia on July 21st for a week. Some 100 papers will be presented at this conference which is being organized by the Electronics and Communications Section of the I.E.E. in association with the International Federation for Medical Electronics.

About 80 exhibitors are participating in the exhibition which is being held at Olympia throughout the conference. In the following list of exhibitors we have indicated the country of origin of those from overseas. The exhibition will be open daily from 9.30 to 6.0 ; admission is 3 s 6 d .
A.E.I.

Air-Shields (U.S.A.)
Allen \& Hanburys
Alvar-Electronic (France)
Amplivox
Antares (France)
Ateliers de Construction
Beaudouin (France)
Atlas-Werke (W. Germany)
Atomic Weapons Res. Estab.
Bailey, I. G., \& Co.
Barr \& Stroud
Becker, J. (Holland)
Belling \& Lee
Bird Oxygen Co. (U.S.A.)
Chiba Electric Works (fapan)
Coulter Electronics
Cox, Stanley
Dawe Instruments
Disa Elektronic (Denmark)
Ekco Electronics
Electronic \& X-Ray Applications
Elect-onic Industries Assoc. (fapan)
Electronic Machine Co.
Elekt-n'aboratoriet (Denmark)
Elema-Schonander (Sweden)
Elga Products
Endomer-ics
English Electric Valve Co.
Etudes et Constrmetions
Electro-Medicales (France)
Faraday Electronic Insts.
Frieseke \& Hoepfner (W.
Germany)
Fukuda Electro Co. (Gapan)
Fukuda Medical Electric Co. (7a力an)
G.H.S. Electronics

Godart \& Minhardt
Godart-Miinhardt (Holland)
Heiwa Electronic Institute (7aかan)
Hellige \& Co. (W. Germany)
Heywood \& Co.
Hilger \& Watts
Hitachi (fapan)

Infra Red Development Co.
Japan Radio Co. (fapan)
Leitz, E. (Instruments)
Leitz, E. (W. Germany)
Leland Instruments
Marconi Instruments
Medische Apparaten (Holland)
Mullard Equipment
Multitone Electric Co.
Nagard
New Electronic Products
Nikkoh Electronic Instrument Co. (fapan)
Nippon Electric Co. (Tapan)
Nnc'ear Enterprises (G.B.)
Officine Toscane Elettromeccaniche (Ital $\psi$ )
Offner Electronics (U.S.A.)
Ossa (Switzerland)
Picker International Corp. (U.S.A.)

Purtschert, M. J., \& Co. (Switzerland)
R.C.A. Great Britain
S.S. Electronics

Sanborn Co. (U.S.A.)
San'ei Instrument Co. (fapan)
Sanyei Manfg. Co. (fapan)
Saunders-Roe \& Nuclear Enterprises
Schwarzer, F. (W. Germany)
Selig Electromagnetics
Shimazu Seisakusho (fapan)
Siemens-Reiniger-Werke (W). Germany)
Sierex
South London Electrical Equipment Co.
Telco (France)
Telefunken (W. Germany)
Tinsley \& Co.
Tokvo Shibaura Electric Co. (fapan)
Dr. Ing. J. F. Tonnies (W. Germany)
Townson \& Mercer
Winston Electronics


# Electronics in Israel 

STEADY GROWTH EXEMPLIFIED AT THE RECENT EXHIBITION IN TEL AVIV

By R. DANZIGER, M.Brit.I.R.E.

THE first Electronics Exhibition was held recently in Tel Aviv and was organized by the IEMA (Israel Electronics Manufacturers Association) and the trade schools. Since its modest start in 1950 the electronics industry has increased the value of its annual output from 350,000 to $8,000,000$ IL (Israel Pounds).

The 1960 exhibition was the first of what is intended to be a series of annual events. The main part of the show was devoted to domestic radio receivers which ranged from tiny transistorized sets through car radios to elaborate "hi-fi" equipment for rooms with special acoustic arrangements. Wood was still the favourite material for the Continental styled cabinets of most makers with plastic materials being the exception. Some firms showed prototypes of their future production programme including TV sets, stereo amplifiers, tape recorders and d.c. record players.

An important part of the exhibition was the components show. Except for valves, transistors and certain types of condensers and resistors all parts are manufactured in Israel. A wide range of components was shown including piano-type waveband switches, polystyrene condensers, several types of variable condensers, many types of loudspeakers and tweeters and a wide range of all types of transformers and coils.

Perhaps the most interesting part of the show was that devoted to the professional exhibits. Whilst the manufacture of professional electronics equipment is still at its beginning, it is backed by considerable know-how. Outstanding in this section were electronic fire alarms and counters made by ELCO (Ramat Gan), printed circuits, epoxy castings and pulse equipment by Israel Electronics Co. (Rishon Lezion) and quartz crystals by Tadir
(Holon). All being up to a high standard both in respect of specification and finish. Other items in the professional section included timers, miniature pulse transformers, measuring gear, electro-medical equipment, loud hailers and of course a large variety of office and industrial intercommunication systems.

In addition to the commercial exhibits one wing of the show was devoted to training aids and the work of the students of the trade schools. There are seven major trade schools in Israel with 4 -year courses in electronics. They have a combined capacity for 500 students and turn out about 200 qualified technicians per year. The demand for places at those schools is considerable and stiff entrance examinations are held, resulting eventually in a high standard of training. The exhibits in this section included many student-built equipments such as oscillators, amplifiers, transmitters, etc., but also more advanced equipment such as u.h.f. waveguides, pulse-shaping circuits nicely demonstrated by means of several oscilloscopes, and even a radiocontrolled model aeroplane.

The most sophisticated equipments were shown by the two leading academic teaching institutions, The Hebrew University, Jerusalem, and the Technion, Haifa. They exhibited an analogue computor, pulse generators, coincidence plug-in units and even a 24-channel pulse height analyser.

One leaves the exhibition with the definite impression that the local electronics industry has left the " music box" stage for good and is embarking on a serious domestic and professional production programme in the electronics field.

The exhibition, with a floor space of $500 \mathrm{~m}^{2}$, comprised 27 stands and was open for four days ( 4 hours per day) during which time the number of visitors was of the order of 2,000 .

# Radio Aids to Rydrography 

DECCA TWO-RANGE AND LAMBDA POSITION-FIXING SYSTEMS

By C. POWELL*

This article describes the latest version of one of the radio aids to survey at sea. The aid is used extensively throughout the British Commonwealth and in ships of the Royal Navy

"THE science of Hydrography", says the Admiralty textbook on that subject, "originated in the need for the production of maps specially designed for the use of the mariner. ... During the nineteenth century nearly every specialised maritime country founded a department for the sole purposes of dealing with the issue and publication of charts and for the coordination and execution of marine surveys, and immense progress has been achieved in the production of charts ${ }^{\prime \prime}$. An important contribution to that progress, since the Second World War, has been made by radio position-fixing systems, and these are now accepted as standard surveying equipment in hydrographic work.

A radio position-fixing system can be used to fix the position of the survey ship, so that observations can be assigned to their correct geographical positions; to hold the ship on the survey tracks, and to navigate the ship to the survey area with the minimum delay and wasted steaming. It can be used when haze or mist prevent shore marks from being used for fixing and running lines by the classical methods; also at distances such that shore marks are beyond the limit of visibility. This has brought radio aids into the field of oceanographic surveying, involving ship-to-shore distances of several hundred miles. A widely-used radio aid to surveying at sea is Two-range Decca; as this is derived from the Decca Navigator ${ }^{2,3}$ some aspects of the latter should be noted.

Decca Navigator.-Decca employs unmodulated c.w. transmissions occupying spot frequencies in the $70-130 \mathrm{kc} / \mathrm{s}$ band. At these frequencies, the attenuation of the surface wave is sufficiently small to permit use of the system by ships at distances greatly exceeding that of the "radio horizon" and the stability of propagation condition during the hours of daylight, coupled with the potentialities of the phase-comparison method of obtaining a position line, make possible a degree of accuracy appropriate to most hydrographic-survey requirements. A chain of Decca stations normally comprises a central master with two or three outlying slave stations disposed about it. The slave transmissions are phase-locked to the master and the basic function of the receiver carried by each user of the system is to compare the phase of the master transmission with that of each slave. The phase differences are displayed on three pointer-type phasemeters, known as Decometers (one for each master/slave combination) and each reading locates the user on one

[^7]of a family of hyperbolic constant phase-difference lines focussed on the master station and the appropriate slave. A position fix is given by the intersection of two such position-lines: on board ship the readings are usually plotted manually on a chart overprinted with correspondingly-numbered Decca-grid lines. In aircraft, and in certain specialduty ships, the fix is continuously displayed on an automatic plotter (Flight Log ${ }^{4}$ or Track Piotter) which provides a pictorial presentation of position by a pen, representing the user vehicle, moving across a map.

Two-range Decca.-Soon after its commercial introduction in 1946, the Decca Navigator became established as an aid to hydrographic surveying by use of the permanent European coverage as it then existed, supplemented by mobile chains in Greenland, Sweden and the Persian Gulf. In 1950, however, there arose an operational requirement which the conventional Decca station layout could not meet; a survey organization operating in the Antarctic required a radio position-fixing system having the characteristics of Decca; but using only two shore stations instead of three because three suitably-disposed and habitable sites could not be found in the area of the proposed survey. As a solution the survey department of the Decca Navigator Company proposed a layout in which the two slave stations would remain on the shore, the master being installed on the survey ship together with the receiver. While confining the use of the chain to a single ship at a time, this arrangement overcame the siting difficulty and also introduced other advantages, notably the fact that the Decometers now indicated the direct distances to the shore stations rather than the distance-differences of the conventional system.

The ensuing development of the Two-range system into an operational hydrographic-survey tool was undertaken in close co-operation between the Hydrographic Department of the Royal Navy, the Admiralty Signal and Radar Establishment (now the Admiralty Surface Weapons Establishment) and the manufacturers.

## Principles of Two-range System

The layout of a Two-range chain is shown diagrammatically in Fig. 1. The master transmitter on the ship radiates a c.w. signal of frequency $12 f$ where $f$ is approximately $14 \mathrm{kc} / \mathrm{s}$. As all the radiated and phase-comparison frequencies are harmonically related, as in conventional Decca, it is more convenient to refer to them in a harmonic notation rather than in numerical terms. The "red" slave station ashore receives the master transmission and radiates a signal of frequency $8 f$, in a manner such that the slave and master signals have a constant phase relationship at the common multiplied-up frequency value of $24 f$; a stable pattern of constant-phase-difference lines is therefore generated about
the two stations (broken lines in Fig. 1) this pattern being identical with that which would be produced if signals frequency $24 f$ (about $340 \mathrm{kc} / \mathrm{s}$ ) were actually radiated from master and slave. In the shipborne receiver the signals are received and multiplied for phase comparison at $24 f$, and the phasedifference meter (Decometer) makes one rotation if the ship-to-slave distance alters by one phasedifference cycle or "lane". Along the line between master and slave, the lanes recur at uniform intervals each equal to half a wavelength at the comparison frequency, and the lane pattern that the ship uses therefore takes the form of a family of concentric circles centred on the slave station. At $24 f$, the lanewidth is roughly 420 metres and the Decometer, which can be read to less than half-ahundredth of a revolution, will therefore respond to a change of a metre or two in the ship's distance from the slave. A similar process takes place in the "green" co-ordinate ( $9 f$ slave frequency) at a common comparison frequency of $36 f$, giving in this
case a lanewidth rather less than 300 metres as shown in the accompanying table.

Each Decometer embodies a lane-counting pointer geared down from the phasemeter rotor, together with a subsidiary indicator driven through a further stage of gearing and recording groups of lanes ("zones") passed through. Assuming a value for the speed of propagation of electromagnetic radiation, these readings can be converted into distance units: this is generally done by plotting them on a chart overprinted with the two patterns of circular position lines, numbered to correspond with the Decometer readings and drawn at constant radial intervals of one or more lanewidths. A "two-range" fix of the ship's position with respect to the shore stations is given by the intersection point of the two circles (interpolating between the lines on the chart as required) indicated by the meter readings.

If a second receiver were placed close to, say, the red-slave station, the red Decometer reading of this receiver would show no change even if the ship


Fig. I. Diagram of layout of Two-range Decca Hydrographic-survey system.

TAble
Typical values for frequency and lane width


Note: All frequencies are harmonically related to a non-transmitted fundamental value $f$.
altered her position by several hundred miles; this is because the slave station's function is to radiate a signal having a constant phase relationship with the incoming master transmission. The shipborne Decometer indicating the master/slave phase difference is, however, sensitive to any change in the ship-to-shore distance, as this alters the length of the transmission paths from the master to the slave and back without a corresponding change in the direct path from the master to the ship receiver.

If maximum accuracy is to be secured from the system, corrections have to be applied for the nonuniform speed of propagation of radio waves in the groundwave mode and for fixed phase shifts. The full expression for measurement of the distance between the ship and a slave station therefore becomes:

$$
d=(\lambda c f / 2)(\phi-\alpha-\psi)
$$

where $d$ is the distance from the "electrical centre" of the ship to the mid-point between the receiving and transmitting aerials at the slave station, $\lambda c f / 2$ is the lanewidth in metres for the appropriate pat-1 tern, assuming free-space velocity, $\phi$ is the observed Decometer reading (whole lane number plus fraction), $\alpha$ is the "locking constant," and $\psi$ is a correction to the free-space value of the speed of propagation.

The exact location of the electrical centre of the ship in the above formula is found by calibration at a known distance and on a number of different headings. The locking constant is the name given to the overall phase shift due to the close proximity of the receiver to the master transmitter (placing the former in the "induction field") and, at the slave station, a possible fixed displacement from the nominal zero phase-difference condition that is assumed to exist between the received master signal and the outgoing slave transmission. The value of the locking constant for each pattern is found at the start of a survey by observations at exactly known distances from the slaves, and is thereafter subtracted from all observed Decometer readings.

The quantity $\psi$ refers to the dependence of the effective speed of propagation upon the nature of the medium over which the signals are transmittedan aspect of the groundwave mode of propagation which is of fundamental importance when the transit time of low-frequency signals forms the basis of position or distance determination. This phase lag with distance results from absorption of energy by an imperfectly conducting earth ${ }^{\star}$ and Fig. 2 shows a practical set of phase-lag correction curves for the red and green patterns. The increase in the cor-
rection value at short ranges is the result of the complex field existing around the transmitter, and the increase beyond 100 km or thereabouts is the effect of the phase lag. The mean speed of propagation resulting from the phase lag varies widely with the electrical characteristics of the medium over which the signals are transmitted; e.g., the sum of experience so far with Two-range Decca points to a mean velocity of $299,650 \mathrm{~km} / \mathrm{sec}$ over seawater transmission paths, while a corresponding figure for land paths of the lowest soil conductivity yet encountered (of the order of $\sigma=5 \times 10^{-15}$ e.m.u.) amounts to about $298,400 \mathrm{~km} / \mathrm{sec}$. If no steps were taken to correct for this variation, an uncertainty of 1 part in 240 could exist in the range determination (assuming the possibility of a Two-range chain being operated over land as well as over sea) which would render the system completely unacceptable as a survey instrument. Fortunately it is possible to apply corrections for different path conductivities; also for paths of mixed conductivity such as the case where a large island or promontory intervenes between the ship and the shore.


Fig. 2. Phase correction (expressed in lanes) for seawater transmission poth.

Application of the corrections shown in Fig. 2 for transmissions over seawater leaves, with present knowledge, a residual uncertainty amounting only to one or two parts in ten thousand. The curves were constructed on a theoretical basis, but have been confirmed by observations with the conventional Decca system as well as with the Tworange version at distances up to about 200 km : beyond this distance it is hoped to obtain practical

[^8]confirmation from trials specially designed for this purpose. Here we encounter a familiar problem in the practical deployment of modern radio aids to surveying and navigation: to check the radio aid satisfactorily it is necessary to know distances and positions with an accuracy several times greater than that of which the aid itself is capable, and this is liable to tax present survey resources to the utmost. *For example, if the correct value of $\psi$ over seawater could be determined experimentally without error, the Two-range technique should then be capable of measuring, say, the distance from a point on the north coast of East Anglia to a point near Aberdeen with an accuracy of one or two parts in 50,000 , yet this is not far short of the accuracy with which the actual distance between the two points in question can be stated from present survey knowledge.

Random Errors.-The phase errors so far described can be partly or wholly corrected. Rather more important are errors of a random character, which may be due to instability associated with wave propagation or with instrumental variations, or both. From sunset until sunrise at all seasons, and also during daylight in winter, random variations due to skywave interference start to become detectable at ship-to-shore distances of about 40 miles, and thereafter increase in magnitude with range. The actual survey operations requiring accurate fixing are, therefore, generally confined to daytime, but this does not apply to incidental manouvres such as journeys between different survey areas, which call for a lower degree of accuracy than the survey itself. Typical random-error contours for the use of the system by day are indicated in Fig. 3. In practice, the overall accuracy of Two-range Decca has been such as to permit plotting of the results at a chart scale of $1: 70,000$ (about one inch per nautical mile) without the errors or variations due to the system itself being detectable at this map scale.

A geometrical characteristic of the Two-range layout is the relatively large proportion of the coverage, compared with a hyperbolic system having the same distance between the slave stations, within which a high fix accuracy can be obtained. The angle of cut between the two circular position-line patterns is good (i.e., near $90^{\circ}$ ) over a wide area, and the layout is favourable in that there is no lane expansion such as occurs when similar equipment is operated as a hyperbolic chain.

Referencing.-A potential source of error in the receiver itself arises from the differential phase shifts between the master and slave channels. To check and correct these, the receiver incorporates a reference source, whose output is a $0.5 \mu \mathrm{sec}$ pulse having a recurrence frequency equal to the fundamental value $f$, to which all the transmitted and phase-comparison frequencies in the system are harmonically related. Applying this pulse to the input of the receiver, each channel extracts its harmonic 'frequency. As the harmonics are related to a common fundamental, the two Decometers would read zero if there were no differential phase shifts in the channels. If a reading other than zero is observed, a compensating phase shift is applied so that a zero reading is restored.

Frequencies.-The transmissions are of the pure continuous-wave type with no modulation. This characteristic enables receiver bandwidths of a few
$\mathrm{c} / \mathrm{s}$ to be employed, which in turn secures the required performance from transmissions of low radiated power. The shipborne master transmitter installation radiates approximately two watts and the slave stations (in the standard version of the system) approximately four watts. The use of low-power transmissions having no modulation sidebands minimises the problems of frequency allocation and of mutual interference with other services.
Those familiar with conventional Decca will have noticed from the table that a different harmonic relationship for the transmitted frequencies is used


Fig. 3. Fixing-accuracy contours in feet for summer daylight operation. Contours are drawn for a 0.01 lane deviation.
in the Two-range version of the system, the master frequency having twice the normal harmonic value. This is in order to secure maximum strength of radiated signal from a shipborne aerial mast which is necessarily of restricted size. On the frigate-type survey ships of the R.N. Hydrographic Service, for example, the height of the mast is limited by the available staying radius to about 45 feet. Early Tworange chains used a master frequency $6 f$, i.e., a spot value between 85 and $90 \mathrm{kc} / \mathrm{s}$, which resulted in a radiated power of approximately one-third of a watt for an input to the aerial of 350 watts. By doubling the frequency an approximately eight-fold gain in radiated power for the same input is achieved, and this has permitted operation in tropical regions, where the noise level is high, at distances well in excess of 150 miles.
Equipment.-Briefly reviewing the items of equipment comprising a Two-range Decca chain, the shipborne master installation consists of a duplicated unit containing a stable crystal oscillator which provides the source of the master signal, feeding a 350 -watt c.w. transmitter. The 45 -foot tubular transmitting aerial mast (Fig. 4) is base-insulated and is supported at three heights by stays insulated at their lower ends.
The shipborne receiver is generally installed in or near the chart room and uses a standard Decca receiving aerial which is a vertical fibreglass tube containing a length of insulated wire. The receiver
(Continued on page 355)
is of the Decca "survey" type which is capable of use with either the hyperbolic or the Two-range layout, the latter necessitating two small adjustments: a reduction in the gain of the master channel because of the proximity of the master transmitter, and a reversal of the sense of Decometer rotation so that the readings increase instead of decrease with distance from the slave station. The Decometers are supplemented, when necessary, by the Track Plotter which is the marine counterpart of the Flight Log. Ashore, each slave station comprises, in duplicate, a " control unit" which contains the oscillator forming the signal source, together with the equipment for phase-locking the outgoing slave transmission to the received master signal. The second "standby" control unit, as it compares the phase of the master and slave transmissions and displays their phase difference, acts as an independent monitor of the phase pattern. The transmitter is similar to that used at the master, and a similar transmitting aerial system is employed except that, in general, its height is approximately twice that of the shipborne mast.

Lane Ambiguity.-From a practical point of view, Two-range Decca as so far described has a serious limitation in the form of a high degree of pattern ambiguity. From the above table it will be seen that typical lanewidths for the two co-ordinates are about 420 and 280 metres respectively, which means that before starting work, or after an interruption, the user must know his distance from the red shore station to better than $\pm 210$ metres and from the green to better than $\pm 140$ metres. At moderate ranges this has not given rise to any serious difficulty; the ship usually starts from a known point in any case, appropriate plotting procedures can reveal the development of a whole-lane error should this occur and, if the journey to a check point where the lane values are known should have to be made this may not be a major undertaking when surveying within a small area. At distances greater than about 100 miles from the stations, however, the lane ambiguity becomes an increasingly serious problem, and some form of lane identification, such as is provided on all the permanent Decca navigational chains, would greatly improve the system. The conventional Decca lane-identification method does not lend itself to use with mobile transmitting equipment where light weight and compactness are paramount, but a modified technique known as the "Lambda" method (Low-AMBiguity DeccA) which overcomes this difficulty has recently been evolved and is incorporated in a new Decca survey system based on the Two-range principle.

## Lane Identification in the Lambda System

In essence, any lane-identification system consists in superimposing uipon the ambiguous lanes a coarse pattern in which one "lane" or phase-difference cycle embraces a number of the fine lanes. Thus, if a pattern resulting from phase comparison at a frequency $1 f$, using the previous notation, is superimposed upon the $24 f$ red pattern, a phasemeter responding to the coarse pattern would make one revolution for 24 red lanes passed through, and would indicate the correct red lane number for the ship's position in a group of 24 . Similarly, a $1 f$ pattern will identify the correct green (36f) lane out of a group of 36. As there will be many more than 24


Fig. 4. British Noval Survey Ship equipped with Tworange Decca showing tubular aerial mast for transmission of the master signal.
or 36 lanes in the total fine patterns, some ambiguity will still remain; but, remembering that the "1f" lanes are about 10 km wide measured along the master-to-slave line, this requires only that the user should know his distance from the slave station initially to $\pm 5 \mathrm{~km}$ in order to be able to make use of the coarse patterns for setting correctly the Decometers. In practice, this remaining degree of ambiguity causes no difficulty since the ship's position can be found accurately enough by an astronomical fix or other standard practice.
Since it is out of the question actually to transmit If ( $14 \mathrm{kc} / \mathrm{s}$ ) from the stations, this frequency must be extracted from them by other means. In the Lambda system, the shipborne receiver obtains a if master signal direct from the appropriate circuit in the master-transmitter control unit. For the receiver to obtain a $1 f$ signal froms the slave stations, which normally transmit $8 f$ and $9 f$ for red and green respectively, the slave frequencies are momentarily counter-changed so that, given a means of "memorising" the original phases of the signals, a lf beat note can be derived from each station. Comparing the If transmission thus obtained from the slave with the $1 f$ signal from the master results in the generation of the if phase pattern necessary for lane identification.
Lane identification is initiated in the Lambda system by stopping the $12 f$ transmission from the ship and replacing it, for about 1.3 sec , by an 11 f signal. This triggers the changeover of frequencies at the slave stations and provides the slave with a "notching" datum, to be described later.

A locked-oscillator technique is used at the shipborne receiver to extract the required $1 f$ beat note from each slave station. (Fig. 6(b)). At the receiver on board, the $8 f$ and $9 f$ locked oscillators preserve the phase of the signals that normally control them, and $1 f$ beat notes are extracted for each slave by mixing the oscillator outputs with the signals received during lane identification. The phase difference between the master and each slave at the frequency $1 f$ is displayed by a sector-shaped pointer on the coarse Decometers (lower dials in Fig. 5). If the lane-identification pointer moves so as to enclose the lane-counting pointer, the latter is reading correctly. If this does not occur, the lane-counting pointer is reset manually to the position indicated. After the one-second lane identification transmissions the normal transmissions are resumed until the user again decides to check or identify the lanes. The actual indications are "frozen" for several seconds, for ease of reading.

Lambda Lane Patterns.-In a system designed for oceanographic surveying several hundred miles from land, every possible precaution must be taken to ensure continuity of operation. The ambiguity which the lane-identification system is designed to resolve should be made less severe, if this is feasible, and the possibility of a whole lane error occurring should be reduced to the absolute minimum. Accordingly, in the Lambda system the basic patterns are produced by comparing phase at the frequency of the slave transmission without multiplication. This results in a greater discrimination against noise interference than when frequency multiplication takes place, leading to a greater range for a given probability of lane loss. At the same time the lanes thus generated by phase comparison at the relatively low slave frequency are correspondingly wider and less ambiguous.
The basic Decometers (the two lower dials on the display unit shown in Fig. 5) respond to these slavefrequency patterns and the movements are so geared to the lane-counting pointers that the latter make one revolution per zone, i.e., one revolution for 8 red slave-frequency lanes and 9 green slave-frequency lanes. On turning manually the red meter with the reset button, therefore, it will be found that the lane pointer can take up any one of eight equally spaced positions around the dial. The lane-identification pointer is coaxial with the lanecounting pointer, so that the basic function of the former is to indicate which of these eight positions is the correct one, i.e., to identify the correct slave-frequency lane within a zone of 8 or 9 such lanes for red and green respectively. As already mentioned, a rough fix serves to tell the user which zone of each pattern he is in. The individual zones are counted, as in normal Decca practice by additional dials appropriately geared to the lane-counting pointers.

Unhappily, technological achievements do not include the extraction of something from nothing, and owing to its wide lanes the system as described
so far would be three times less sensitive to a cnange in the distance from the ship to the red slave station than the earlier version of Two-range Decca, and four times less sensitive in the greefi co-ordinate. To remedy this, narrower lanes are interpolated between the slave-frequency lanes simply by carrying out a further phase-comparison at a higher frequency. This requires further frequency-multiplication in the receiver, together with additional discriminators and a pair of fine Decometers, as shown in the block diagram of Fig. 6 (a). The fine meters, which are the upper pair on the display unit, make one revolution per $24 f$ (red) and $36 f$ (green) lanes, and thus restore the pattern sensitivity to the same level as that of the previous system. Except in the strictly arithmetical sense, no extra ambiguity is introduced by this measure; the lower Decometers have scales marked in fine-lane units and can be read to a fraction of a fine lane, and a glance at the upper meters which are calibrated in hundredths furnishes the second digit. In effect, therefore, the upper meters operate simply as expanded scales for the coarse meters, and present no individual setting-up problem apart from the necessity that they should periodically be " referenced" to zero together with all the other meter movements in the system.
"Notching."-Reverting to the generation of slave-frequency lanes, it will be realized that these contain an additional ambiguity of their own, which is true of any Decca pattern in which phase is compared at a frequency lower than one or both of the radio transmissions involved; this is the result of frequency-division which, even if it does not take place literally in a dividing circuit, is nonetheless carried out in effect when slave-frequency lanes are produced. Here we need consider only the phase relationship between a slave station such as the red (frequency $8 f$ ) and the $12 f$ master transmission to which it is phase-locked. Granted that the phasecontrol circuit at the red slave holds the outgoing $8 f$ signal at zero phase difference with an $8 f$ signal derived from the master, it can do this for three different relationships between the $8 f$ and $12 f$ master signals. In other words, the $4 f$ frequency corre-


Fig. 5. Display unit of Lambda-type installation. Upper meters indicate fine pattern readings (comparison frequencies $24 f$ and $36 f$ ). Whole lanes and lane identification are given by lower dials.

(a)

OISCRIMINATOR

the same total number of phase-difference ambiguities as a $24 f$ pattern, despite the relatively wide slavefrequency lanes. Further, if there is a potential notch error in counting down $12 f$ to $8 f$, there must also be a greater uncertainty in extracting $1 f$ signals from the various stations for laneidentification purposes in the manner already described. In practice, all such ambiguities are dealt with by the single notchcorrecting facility embodied in the equipment.
Ship Installation.-Fig. 6(a) is a rudimentary block diagram of the receiver in its normal state (i.e., in the absence of lane-identification transmissions). The master transmitter circuits, other than the output stages, are incorporated in the same box as the receiver, and the $1 f$ oscillator which forms the basis of
sponding to division of the 12 f by 3 can start on any $12 f$ cycle it likes without affecting the phase comparison at $8 f$, but for every one such cycle giving the correct pattern phasing, there are two which produce pattern errors of either one- or two-thirds of an $8 f$ lane. These "notch" errors, as they are called, are equivalent to shifts of one or two fine (24f) red lanes respectively, so that if nothing were done about them the $8 f$ pattern that we have gone to so much trouble to generate would have exactly
the master transmission also provides the receiver with its master phase datum. For the latter reason, there is no notch ambiguity problem at the receiver, since the slave-frequency master signals required by the receiver are derlved direct by selecting the eighth and ninth harmonic from the pulse output of the $1 f$ oscillator, and no uncertainty therefore exists as to the relationship of these signals with the $12 f$ transmitted signal which is also derived directly from the same oscillator.

The functioning of the two receiving channels by which the shipborne equipment obtains the red and green slave signals depends upon the use of locked oscillators; that is to say, stable crystal oscillators in temperature-controlled "ovens" whose outputs are compared in frequency and phase with the received slave signals and locked to them by servo loops involving phase discriminators and reactance circuits. The oscillator outputs therefore provide noise-free replicas of the incoming signals, an arrangement which decreases the effective bandwidth of the receiver, as well as furnishing the phase memory required by the method of lane identification shown in Fig. 6(b).

Slave Stations.-Figure 7 shows the essentials of a slave station. The signal source is an oscillator of frequency $1 f$ which is locked at its twelfth harmonic to the incoming $12 f$ master signal. The equipment is identical at the two slave stations; if we assume that it is switched to operate as a "red" slave, the eighth harmonic of the basic $1 f$ oscillator is amplified and transmitted continuously, except during lane identification. The locking of the slave to the master, upon which the generation of a stable pattern of position line circles depends, is carried out in two stages. The incoming master signal, having been passed through a two-stage crystal iilter also in the "oven," is amplified and phase-compared with the twelfth harmonic of the slave's oscillator. The discriminator output controls the phase of the oscillator to keep the $12 f$ signals phase-locked. The second stage of phase-locking (not shown in the diagram) is between the radiated output from the aerial and the input to the transmitter. The $8 f$ output of the oscillator passes to the transmitter through a reactance stage. This is controlled, in turn, from an $8 f$ phase discriminator, in which the drive and radiated $8 f$ signals are phase-compared, to keep the radiated signal locked to the drive signal, irrespective of capacity changes in the transmitting aerial. This whole arrangement is duplicated at $9 f$ so that the slave transmission at that frequency during lane identification shall be similarly phase-stable with respect to the master.

Elimination of the notch ambiguities is simply a matter of ensuring that the slave If oscillator has the correct phase relationship with the master if oscillator and this relationship is displayed to the slavestation operator every time the ship initiates a laneidentification transmission. The "notch meter" indicates the phase-difference between the $11 f$ signal from the master (i.e., the eleventh harmonic of the master oscillator) and the eleventh harmonic of the slave oscillator. The latter is sufficiently stable for it still to be considered as locked to the interrupted $12 f$ master signal, so that the twelfth harmonics of the two oscillators are already in phase; if the eleventh harmonics are seen by the notch meter to be also in phase the master and slave oscillator outputs will have the correct (i.e., zero) phase relationship at their fundamental frequency. If one of the eleven other possible readings is observed, a 12 -position notch control, which operates a phase-shifting network, is turned a sufficient number of clicks in the appropriate direction to bring the eleventh harmonic of the oscillator into the right phase. The use of a click device on the phase control simplifies the setting up process as correction can then be made after the lane-identification period ( $\approx 1 \mathrm{sec}$ ) has finished.

When the meter reads zero, the whole system is correctly notched, i.e., the slave transmissions have the correct phase relationship with the master at the fundamental frequency if and hence also at the transmitted harmonic frequencies. The probability of a notch error developing during a day's work has been shown to be extremely remote. However, to enable the slave operator to check the notch in the event of an interruption in transmission, provision is made for him to request the ship to initiate a laneidentification transmission. The signal takes the form of a momentary phase shift in the slave transmission which is too rapid to introduce a Decometer error, but which serves to trigger a "slave-call" lamp on the receiver display unit.

On the survey ship, the receiver and the input sections of the master transmitter are now housed in a bulkhead-mounted container similar in size to that of the familiar Decca Navigator Mark 5 shipborne

set; the transmitter itself and the aerial system are disposed in the same way as before (Figs. 1 and 4). The slave stations are stowed on board the ship in transit, and are then taken ashore in small boats or helicopters and set up for the duration of the survey. To assist stowage and transport, the units are housed in standardised boxes of the type used for Army signals equipment: hydrographic surveys are in the nature of military operations, and every possible effort has to be made to ensure mobility as well as reliability in the equipment that the modern surveyor employs.

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# Elements of Electronic Circuits 

15.-THE "BOOTSTRAP" TIMEBASE

By J. M. PETERS, B.Sc. (Eng.), A.M.I.E.E., A.M.Brit.I.R.E.

1TIMEBASE circuit which uses both positive and negative feedback is shown in Fig. 1. Known commonly as a "bootstrap" timebase, it com-prises:-(i) A switch, consisting of a triode V1 operated by a negative-going square pulse, shunting the timebase capacitor C . (ii) A charging resistor, R , across which a constant potential is maintained. (iii) A triode, V2, connected as a cathode-follower amplifier, i.e., with current negative feedback, and developing unity gain. (iv) A capacitor, $\mathrm{C}_{1}$, providing positive feedback to $\cdot$ 2, the feedback voltage being developed across the timebase charging resistor R.
The operation of the circuit is as follows. An initial quiescent state is assumed with V1 conducting and V1 anode at a low potential. A negative-going square wave cuts off V1, and the timebase capacitor C starts to charge via R . If point X in Fig. 1 had been connected to a fixed h.t. potential C would have charged in an exponential fashion. To provide linear charging we endeavour to keep the potential across R constant so that the current through it is constant also. In other words, the waveform generated at $Y$ must also appear at X . (Hence the description "bootstrap" timebase-derived from the notion of pulling oneself up by one's bootstraps.)
This type of action is achieved in the following way. The V2 stage is arranged to have approximately unity gain by operating it as a cathode follower. The timebase waveform developed at Y , if applied to V2 grid, therefore appears in phase at V2 cathode. We now connect the cathode of V2 via C, to point X, which (by cathode follower action) will follow the potential of Y. Thus, by positive feedback via $\mathrm{C}_{\mathrm{i}}$, a constant potential can be maintained across $R$, through which the timebase capacitor C is being charged. As a result C charges linearly.


The potential levels during the quiescent period are maintained by $C_{1}$ and $R_{1}$. Capacitor $C_{1}$ does not discharge appreciably during the period of the sweep provided that:
(a) the value of $C_{1}$ is very much greater than $C$;
(b) the time constant $C_{1} R_{1}$ is large compared with the duration of the sweep. Voltage changes across $\mathrm{C}_{1}$ during the sweep must be small, otherwise the voltage across R will not be constant, thus causing non-linearity;
(c) the time constant $C_{1} R$ is also large compared with the duration of the sweep.
Ideally the voltage across $R$ should be exactly constant, but this cannot be attained in practice on account of the gain of V2 departing from unity. With a gain of less than 1 the voltage at X lags slightly on that at $Y$, the current through $R$ falls slightly and the timebase sweep voltage becomes non-linear.
During the quiescent period (i.e., absence of input voltage at V1 grid) current flows through $R_{1}, R$ and V1 to earth. During the period of the sweep, the current through R remains constant but the decrease in current through $R_{1}$ is made up by that flowing through $\mathrm{C}_{1}$.

## B.S.R.A. Constructors' Competition

THIS year's competition, held in conjunction with the annual dinner of the British Sound Recording Association, was notable for the exceptionally high standard reached by all the competitors, not only in design but in the workmanship and finish of the sound recording and reproducing equipment shown.
The President's Trophy was awarded to L. Widger, A.M.I.E.E., for a fully automatic system of sound accompaniment for a cine film in which commentary from tape and background music from discs are blended through mechanically-operated faders by notch cues on the film and conducting strips on the magnetic tape. The system includes means of precise speed control of the film and facilities for recording the combined sound effects on a single tape if required.
A neat v.h.f./f.m. receiver with pulse-counter-type discriminator won the Wireless World prize. This was designed by R. N. Baldock, B.Sc., in cylindrical form (2in diameter $\times 8$ in long) for mounting behind existing panels where space may be limited. Another tuner, with a similar discriminator, also of very compact design was entered by A. Robinson and was runner-up in the section for non-members of the Association.
A. J. Harper was awarded the Guy Fountain Prize with a neat turntable and pickup mounting incorporating an unusually smooth pickup lowering mechanism, most of which is below the motor board.
A prize for non-members of B.S.R.A. has been donated by Hi Fi News and was won this year by J. T. Gilbert for a stereo tape recorder using a modified commercial mono deck and with an unusually wide range of facilities, including dual level indicators and a built-in oscillator for balance adjustments.

## Manufacturers' Products

## NEW ELECTRONIC EQUIPMENT AND ACCESSORIES

## Screened Jack Plug

A FULLY screened telephone or microphone plug with standard $\frac{1}{4}$-in-diameter shank, has been introduced by A. F. Bulgin and Co. Ltd., Bye-Pass Road, Barking, Essex, primarily for use with tape recorders, but it has also applications wherever screened concentric or coaxial cables are terminated in a plug of this kind. It takes cables up to $\frac{7}{32}$-in outside diameter and the screwon metal cover, which is electrically connected to the shank electrode of the plug, is available with either polished chrome (P538) or 22ct gold plate (P539) fimish. The price is 4 s 6 d in either finish.

No soldering is required, the centre conductor of the cable is secured by a grub screw to a terminal block and

the screened sheath is gripped in a claw-like clamp, which serves also to take any strain on the cable.

A companion $6-\mathrm{mm}$ model, for use with "Continental " equipment, is available also.

## Flexible Wood-veneer Strips

THE processed forms of wood-from plywood to veneered chipboard-have very definite advantages for the construction of equipment and loudspeaker cabinets; but, if it is desired to "finish" as polished wood, inferior or complicated construction of the edge joints often has to be adopted to cover the multi-layer or chip nature of the material. To avoid the need for difficult construction Flexible Veneers Ltd. offer the $\frac{3}{3}$-in-wide Agastrip and Agatape paper-backed, flexible-veneer edging strips in light oak, mahogany and walnut. Agastrips are supplied in four- and ten-yard-long coils with the grain across the strip and with along-the-strip grain in $3-\mathrm{ft}$ lengths. These have to be glued on; but the selfadhesive Agatape ( $19 \frac{1}{2}$-in lengthwise-grain strips) has only to be pressed on to the clean, dust-free surface.

The address of Flexible Veneers Ltd is: Cobbs Court Buildings, Carter. Lane, London, E.C.4.

## Transistorized Public-address Amplifier

WITH the new WS Electronics "Lilliput" amplifier a maximum output power of 12 W r.m.s. into an impedance of 3 or $15 \Omega$ may be obtained from an input of 2 mV r.m.s. at $30 \Omega$ impedance. The average current consumption when amplifying speech to maximum power output is about 1.5 A (at 12 V d.c.). The amplifier may be operated from 12, 24 or 28 V d.c. supplies. Fuses in the amplifier protect the battery supply against being shorted out and also prevent the transistors being


WS Electronics "Lilliput" transistor public-address amplifier.
damaged by the application of a reversed voltage. The weight of the "Lilliput" amplifier is 51 b and its size $6 \frac{1}{2}$ in by 4 in by 4 in . It costs $£ 1910 \mathrm{~s}$ and is manufactured by W.S. Electronics Ltd., of Brunel Road, East Acton, London, W.3.

## High-resistance Kilovoltmeter

THE E.I.R. Instruments kilovoltmeter has a sensitivity of $250 \mathrm{k} \Omega / \mathrm{V}$ and indicates over the ranges 0 to $100 \mathrm{~V}, 0$ to 20 kV (direct only) and 0 to $300 \mathrm{M} \Omega$ on a 4 -in scale. Using a $100-\mu \mathrm{A}$ basic movement, the $4-\mu \mathrm{A}$ f.s.d. sensitivity is achieved by the use of a cathode-follower current amplifier. This has a high-value cathode resistor, thus stabilizing the meter against mains-supply variations and the effective grid-to-earth resistance is $25 \mathrm{M} \Omega$, corresponding to the $100-\mathrm{V}$ range. To change to the 20 kV range a $5-\mathrm{G} \Omega(5,000 \mathrm{M} \Omega)$ series resistor mounted in an insulated tube is used: this resistor is a single unit of a special type rated at 30 kV . To measure resistance the unknown resistor forms a potential divider from h.t. with the input resistance of the meter.

Capable of operating from 110 or 200 to $250-\mathrm{V}$ a.c. supplies the meter is available in several forms (skeleton for incorporation in equipment, wooden or metal case, with or without megohm range) and it costs between $£ 20$ and $£ 23$.

Manufacturers : E.I.R. Instruments Ltd., 329 Kilburn Lane, London, W.9.
$100-\mathrm{V}$ and $25-\mathrm{kV}$ meter has sensitivity of $250 \mathrm{k} \Omega / \mathrm{V}$.


## Technical Notebook

Industrial R.F. Generators usually employ large tank circuits in which the losses have to be kept to a minimum. This implies the use of a high-Q circuit with a small value of tuning capacitance; but then the variable reactance imposed by the work may "pull" their frequency outside the set limits. A large tankcircuit capacitance can overcome this effect by swamping the imposed variations, but the losses in a circuit of normal construction are then increased. In Mullard Technical Communications, Vol. 5, No. 41, F. Dittrich describes the design of "laminated circuits," the aim of which is to combine the inductive and capacitive components of the tuned circuit, at the same time avoiding joints and sharp corners in the path of current flow. The tuned circuit is made up from a set of plates of the form shown in the sketch, stacked and spaced, alternate plates being reversed to make the crosshatched portions overlap from alternate edges and form the capacitance. This forms a parallel-plate capacitor with an integral single-turn inductor

-the inductor being split into many sections so that each section carries only the circulating current associated with its own capacitor plate. Another advantage is that, outside the crosshatched area on the sketch, adjacent plates are at the same r.f. potential so that metallic spacers and bolts may be used for assembly.

Adjacent-Channel Interference due to multiplex signals may not be so great in practice as simple theoretical considerations would indicate, according to a letter from L. B. Arguimbau published in Proc. I.R.E. for August 1959. Measurements on two commercial v.h.f./f.m. receivers were made-one nominally broad-band and the other nominally narrowband. In both cases there was little change in adja-cent-channel interference when an ordinary f.m. signal was replaced by an f.m. signal which included a sub-carrier such that the maximum total deviation was increased (although at the same time the main-carrier de-
viation was reduced and the amplitude of the added sub-carrier was made less than that of the main carrier). The reason for this small change in adjacent-channel interference with increasing deviation is probably due to the fact that in any practical receiver the response outside the passband does not fall off immediately to zero. In this latter simple theoretical case where the response outside the passband does immediately fal off to zero, if the deviation of a signal normally just outside the passband is increased, the interference, of course, increases from zero to some finite value, so that the proportionate increase in adjacent-channel interference is theoretically infinite.
Drop-outs and Noise due to imperfections in the tape are problems besetting any type of magnetic-tape recording, but for digital work either can result in false information. Many precautions are thus taken to reduce the effect of these distortions. The Telegraph Construction \& Maintenance Company have attacked the root of the problem by, surprisingly enough, returning to the Blattnerphone idea, replacing the oxide-coated tape by 0.001 -in thick Vicalloy. This material is a malleable and ductile permanent-magnet alloy with a saturation flux density of about 12,000 gauss. It is claimed that, due to the lower incidence of drop-outs and noise, much more information can be packed on the tape, so saving space in compact instrumentation.
Pulse Shortener developed by the Admiralty Surface Weapons Establishment uses variable-capacitance diodes as the capacitors in a lumpedconstant transmission line. To shorten a pulse the bias on the diodes is altered so as to decrease their capacitance as the pulse is travelling down the line. This decrease in the diode capacitance produces an increase in the velocity of propagation down the line. Since the pulse continues to occupy the same physical length of line, this increase in the pulse velocity decreases the pulsc period. Decreasing the diode capacitances also increases the pulse energy stored.

## random radialions

## By " DIALLIST

## It's Worked in New Zealand

 A READER in New Zealand tells me that the term radiotrician, which I mentioned in the April issue, is in fairly common use in his country, but the official word is serviceman. Official? Yes, in New Zéaland it is illegal for anyone who has not served a specified term as a trainee and subsequently passed an official examination to undertake the servicing of any mains-operated electronic equipment. Having served his time and passed the exam, he is registered as a serviceman and receives his certificate of competence. To become a trainee, a man must obtain a "permit to assist" a certificated tradesman for so many years, after which he can present himself as a candidate for the exam. We have, of course, in this country the R.T.R.A. exams and certificates in both sound radio and television servicing; but it isn't agin the law to undertake such work if you haven't got these certificates. Whether or not it should be is a moot point. I wonder what readers think about it?
## Wire or Wireless?

WHICH, I wonder, will win the race to provide better television and v.h.f. sound reception in places where they're not now too good? Stage 2 of the B.B.C.'s expansion programme
is scheduled for completion by March 1964 and the additional 21 satellites will bring in an important number of new viewers and listeners, as well as improving reception for many thousands more. But the piped services people are getting on fast with the job of providing strong "clean" signals in places where they're now weak, or interferenceridden, or both. Myself, I'm rather inclined to believe that piping is the only certain way of providing good services in built-up areas in which interference, ghosting and so on are bad. But it probably would be uneconomic to extend it to remote country districts and it's the villages and isolated houses that satellite transmitters and translators can do most to help.

## V.H.F. DX

FROM Aylesbury comes further news of successful long-distance reception on the very high frequencies. The reader who sends it tells me that in the four years in which he's been at it he has logged all bar one of the B.B.C.'s v.h.f. stations. He has also logged nearly 60 European stations, including thirty-six in Italy. He points out how heavily the v.h.f. DX-er scores over his medium-wave opposite number. On the very high frequencies good, clear reception of


Continental stations can be obtained; but that's too often far from being the case on the crowded mediumand long-wave bands.

## The Dry Cell Problem

IN the May issue of Wireless World I mentioned the possibility of serious damage through the puncturing of one or more cells of a partly run down dry battery left in a transistor receiver, a hearing aid, or an ohmmeter. Several suggestions for making things safe have come along; wrap the battery with Sellotape or with insulating tape, or place it in a polythene bag in the set and should it spring a leak, throw it away bag and all. So long as there's room for a wrapped battery and provided you can contrive watertight exits for the connecting leads, any of these should answer. But the real answer is surely that dry batteries meant for use in such apparatus should be made up of leakproof cells. They cost a bit more, but in my view it's jolly well worth it, for I have never known a leakproof cell of good make to belie its name, even though badly treated. Just to see what would happen, I once kept a trio of run-down leakproof cells on the shelf for a whole year. They weren'r leaking when I threw them away.

## V.H.F. and Polarization

AS you know, the B.B.C.'s v.h.f. sound transmissions are horizontally polarized and I have been surprised to find that with an horizontal dipole I have often picked up vertically polarized signals sent out by nonbroadcasting stations. This so intrigued me that one day I tried the experiment of changing my aerial from the horizontal to the vertical position. Reception from the local broadcasting station wasn't so good; but I did get a signal of some kind with the dipole in any position between the horizontal and the vertical. A friend who often listens to European v.h.f. stations tells me that with his horizontal Band III dipole he is frequently able to receive vertically polarized transmissions-or at any rate, transmissions emanating from vertical aerials-from other countries. It must be, I suppose,
that as they journey these very short waves tend to twist a little.

## Electron Beam Welding

SINCE writing the note in the June issue on electron beam welding and cutting I have learned that the originator of the technique in Europe was J. A. Stohr, of the French Atomic Energy Authority. I am interested to learn also that Edwards High Vacuum Ltd. have obtained a licence to manufacture in this country vacuum welding equipment using an electron bombardment heat source. Edwards also kindly sent me a reprint of a paper on the subject presented by two members of their research staff, M. E. Harper and E. G. Nunn, at a recent meeting of the Institute of Welding.

## Getting Down to It

THE striking and animated picture entitled "TV," which attracted so much attention at this year's Royal Academy Exhibition, might well have had a sub-title "How not to watch it," for nearly every method of obtaining a poor picture and straining the eyes is being practised by the viewers. To begin with, the receiver is on the floor, which means that except for the children and the dog, who are themselves on the carpet, everyone must look right down at the screen-a most uncomfortable business and hard on the eyesight. The children, poor mites, have their eyes within inches of the screen. Despite the fact that they're already so close to the set all the grown-up viewers are leaning forward. It's a question of artistic licence, I suppose.

## Without the Book

AMAZING-isn't it?-what a mess a ham-handed fellow who tries to adjust a television receiver can make of things. I don't mean by taking the back off and poking about inside but just by messing with the external controls. A friend of mine recently acquired a first-rate set, capable of showing an excellent picture, which was put in and adjusted by his dealer. When I dropped in a few days later to see how it was doing I found that it was showing just about as bad a picture as you can imagine. Surely the dealer hadn't left it like that, I suggested. No, I was told, it wasn't quite like that: but the owner thought that he could make it just a tiny bit better, so he tried his hand. No, he hadn't bothered about the instruction book; just tried altering one knob's setting after another. Afraid he hadn't made much of a job of it; could I be so kind . . .?


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## Radar in 1896?

IF I were to state that radar was invented before radio communication, it would probably result in strong letters of protest being sent to the Editor. Yet this is stated almost daily in the popular Press and even, I'm sorry to say, in some so-called technical journals.

The writers do not make this statement directly but they certainly do in effect when they tell us, for instance, that radar was used at Jodrell Bank to switch on and off the transmitter of one of the American space probes. We also often hear that "radar waves" are used to control guided missiles, and to destroy American rockets in mid-flight when it has been found necessary owing to something having failed to work out according to plan.
This grievous perversion of the word radar is obviously due to the fact that these offending writers imagine it to mean control at a distance or, in other words, telearchics. To some extent this is the fault of people in responsible positions who coined the expression "secondary radar" to describe what was really a special application of telearchics. No doubt the expression secondary radar is highly convenient for those who know what they are talking about but it was bound to cause confusion and trouble among the less well-informed who at once proceeded to apply the term radar to any other application of telearchics.
Now if we are going to allow the word radar to be used as a synonym for telearchics we must, to be logical, be prepared to admit that radar was invented before radio communication. A moment's thought will make this clear.
Everybody will admit that before it was possible to establish radio communication, Marconi and others had to find some means of causing incoming radio waves to waggle the diaphragms of a pair of headphones or trigger off a local source of power to operate the armature of a morse inker. In other words, telearchics, which is loosely and falsely called radarcontrol by certain writers, had to precede radio communication, as is made abundantly clear in Marconi's patent specification of June, 1896.
However, telearchics-and therefore wrongly so-called radar-was established long before 1896. I have been reading a book published in 1917 and written by B. F. Meissner, of the U.S. Navy. It is mainly about the control of torpedoes by telearchicswhich the author calls teledynamics or radiodynamics.
In this book Meissner stresses that
the first instance of the electrical control of mechanisms from a distance was when Le Sage, of Geneva, established his electrostatic telegraph nearly two centuries-ago in 1774. Its wireless counterpart-using the word wireless in its literal sense-was in 1838, when Steinheil obtained deflections of a galvanometer needle over a distance of fifty feet by means of earth currents.

I will cap this by pointing out that in 600 B.C., Thales of Miletus demonstrated wireless control at a distance when he caused amber to attract pieces of paper. Obviously, even in pre-electrical days, telearchics always preceded communication, for no matter whether we receive a message aurally or visually, it is first necessary for the incoming signals to waggle our eardrums or agitate our retinæ.

## Music Hath Charms

EVERBODY has heard of Congreve's famous words:-
"Music hath charms to soothe a savage breast,
To soften rocks, or bend a knotted oak,",
and it is for this reason that in days of old, young men used to serenade their lady loves, and try to soften their hard hearts by fiddling beneath their bedroom windows. This technique is, of course, exactly the same as that employed by a snake charmer who can, by a suitable tune, bend the most savage serpent to his will.

In the jargon of present - day psychological science, this musical mesmerizing of a maiden would be called conditioning her to accept a proposal.
I hear that this old technique has been adopted by some go-ahead dentists to soothe their patients and "condition" them to accept pain. So far I have only heard of it being employed in the case of a young lady of my acquaintance, but it may be used on some men also, for certain of our sex are undoubtedly as susceptible to the strains of Orpheus' lute as were any of the other beasts of the field.
The lady who supplied me with the information, told me her dentist used the very latest technique in
supplying the mesmerizing music. He had installed a modern stereophonic system which ground out a scherzo as his drill ground into her carious cavity.

I am wondering if the choice of music is always left to the dentist's professional judgment or whether the patient can have a say in the matter. A scherzo is, of course, a light, quick and animated movement; in fact, just the sort of movement the patient herself would make when the drill lighted on a nerve.

The whole thing is thus reminiscent of the technique of the ancient Chinese dentist who, so Marco Polo would have us believe, used to employ thumbscrews on the patient when extracting a tooth. The idea was, of course, that the pain of the thumbscrews acted as a counterirritant to the oral pain inflicted by the dentist. The thumbscrew agony was so excruciating that the dental pain paled into insignificance.

That being so, I think that if I had my choice, I would pick as my counter irritant something by Bartok or Hindemith as I cannot think of anything more painful than being forced to listen to their efforts. Modernists will naturally not agree with me and would probably cloose Bach or Beethoven as their counter irritants. But for a tooth extraction,


Counter-irritant to oral poin of course, nothing would be so effective as the Eton boating song with the swinging melody which accompanies the words "we all pull together."

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## UNUSUAL SINGLE SPEAKER SYSTEM

## A

## CONCRETE COLUMN

An inexpensive column speaker constructed from a concrete pipe and fitted with an $8 / 145$ unit attracted much interest and admiration at the London Audio Fair because of its clean open sound.
The absence of cabinet resonance produces bass of unusual crispness normally associated with larger and more expensive speakers.
A complete kit of wooden fixtures including absorbent wadding and diffusing cone are available at a price of $£ 315 \mathrm{~s}$. 0d. Full details free on request. Suitable concrete pipes can be purchased from builders' merchants at about $12 / 6$. Recommended unit type $8 / 145$ £6 19s. 11d. inc. P.T.

## W3

 THREE SPEAKER SYSTEMWhere a ready-to-use complete speaker system is required the W. 3 in its handsome cabinet is always a popular choice.


BASS L.F. output is produced by a special 12 in . unit type WLS/ 12 fitted with a heavy cone and a new type of suspension which permits large linear excursions and gives a low fundamental resonance of $25-30 \mathrm{c} / \mathrm{s}$.
TREBLE The upper registers are handled by 5 in . and 3 in . units connected in parallel via a quarter section $1 \mathrm{kc} / \mathrm{s}$. dividing network, with an extra series capacitor to protect the small speaker. The volume controls permit adjustment of midrange and treble to give tone control and facilitate balancing speakers on stereo.

Cabinet size $28 \mathrm{in} . \times 14 \mathrm{in} . \times 12 \mathrm{in}$.
Weight 48 lb . complete. Impedance 15 ohms. Maximum input 15 watts.
Effective frequency range $30-20,000 \mathrm{c} / \mathrm{s}$.
Descriptive leaflet free on request.

The elegant cabinet is available in a choice of walnut, oak or mahogany veneers. Also available in whitewood, price $£ 36 / 10 /$ - Tropical model made with resin-bonded plywood can be supplied at $£ 2$ extra.



R-C Oscillator Type TF 1101
A compact general-purpose oscillator giving a low-distortion output of $\frac{8}{3}$ watt max. into 600 ohms. Frequency range $20 \mathrm{c} / \mathrm{s}$ to $200 \mathrm{kc} / \mathrm{s}$. Output level indicated by panel-meter and calibrated attenuator. A band-pass filter provides for a $1-\mathrm{kc} / \mathrm{s}$ output of very low distortion.


## Audio Tester Type TF 894A

The TF 894A covers from $50 \mathrm{c} / \mathrm{s}$ to 27 $\mathrm{kc} / \mathrm{s}$, It comprises a heterodyne oscillator and $0-$ to $50-\mathrm{dB}, 600-\mathrm{ohm}$ attenuator combined with a three-range a.c. voltmeter which is available for external use. Output: 2 watts maximum at 600,15 , and 3 ohms. Voltmeter ranges: 80, 8, and 4 volts full-scale.


## A.F. Power Meter Type TF 1347

A sensitive, accurate, direct-reading instrument. Its ten power ranges, covering $10 \mu \mathrm{~W}$ to 6 W , and $5 \frac{1}{2}$-inch meter provide excellent discrimination. Impedance range: $2.5 \Omega$ to $20 \mathrm{k} \Omega$ in 11 steps. Frequency characteristic substantially flat from $50 \mathrm{c} / \mathrm{s}$ to, $20 \mathrm{kc} / \mathrm{s}$.

A.F. Power Meter Type TF 893A

A wide-range absorption-type power meter for use in the frequency range 20 $\mathrm{c} / \mathrm{s}$ to $35 \mathrm{kc} / \mathrm{s}$. The power measurement range is $20 \mu \mathrm{~W}$ to 10 watts and the input impedance can be set to any of 48 different values between 2.5 ohms and $20 \mathrm{k} \Omega$.


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## Pye at Dounreay

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The Pye Instrument Group has supplied all the equipment to the U.K. Atomic Energy Authority for the irradiated fuel element laboratory at Dounreay. In addition to supplying equipment, Pye Ltd. acted as consultants and designers on all matters in that laboratory relating to instrumentation and remote handling. The illustration above shows manipulators working in conjunction with a television camera to handle and measure a sample from the fast reactor.

# 27th June, 1925 <br> "Hello Cuckoos" 


... 1925. Another of the famous Hendon Air Displays took place in the presence of the King. For the first time wireless telephony was used in the air.

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## type FST1/4 for television receivers

The FSTI/4 Silicon Power Diode has been specially designed for domestic television receiver H.T. power supplies and is of particular interest to circuit designers planning receivers with $110^{\circ}$ scanning, 625 line receivers and colour television receivers. Two diodes may be used in series to provide capacitor smoothed H.T., direct from 250 volts A.C. mains.
SenTerCel FST1/4 slif́con rectifiers are miniature wire ended devices which can be speedily mounted to tag panels, no heat sink being required. Typical performance curves and design procedure. are included in leaflet MF/109.

Important advantages of the FSTI/4 sillcon rectifiers are:-

Large Power Output for Small size

35 Amp Surge Current Rating ( $5 \mathrm{~m} / \mathrm{secs}$.)

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No Heat Sink Required
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Selected Frequency Output: Iv open circuit at output terminal for full scale meter deflection. Output level control provided. Frequency response $\pm 1 \mathrm{db} .20 \mathrm{cps}$ to 50 KC . Output impedance approximately 600 ohms.
B.F.O. Output: Iv open circuit at output terminals. Output level control provided. Frequency response $\pm 1 \mathrm{db}, 20 \mathrm{eps}$ to 50 KC . Output impedance approximately 600 ohms.
Automatic Frequency Control: Range of frequency holding is $\pm 100$ cycles minimum.
Power: $115 / 230 \mathrm{v} \pm 10 \%, 50 / 1600$ cycles, 3 watts (approx.) Terminals provided for powering instrument from external battery source. Battery supply range 28 v to 18 v .
Weight: Net 43 lbs . Shipping 63 lbs . (cabinet mount). Net 35 lbs. Shipping 55 lbs. (rack mount).
Dimensions: Cabinet. Mount: 203" wide; $12 \frac{1_{2}^{* *}}{3}$ high; $14 \frac{1}{2}{ }^{*}$ deep. Rack Mount: $19^{\prime \prime}$ wide; $10 \frac{1^{\prime \prime}}{}{ }^{3}$ high; $13 \frac{1}{2}^{\prime \prime}$ deep.
Price: Delivered U.K. and exclusive of duty-where payable £718 (cabinet mount)
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Continuous progress in design may affect the above specification which is therefore subject to change without notice.

Input Impedance: Determined by setting of input attenuator: 100,000 ohms on 4 most sensitive ranges, I megohm on remaining ranges.


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|  | Permalloy ' ${ }^{\text {' }}$ | Permalloy 'c' | Permalloy 'D' | Permallay 'F' | v.Permendur |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Specific Gravify | 8.3 | 8.8 | 8.15 | 8.4 | 8.2 |
| Electricat resistivity- |  |  |  |  |  |
| microhms jer cm cube | $5^{55}$ | 60 | 90 | 26 | 26 |
| Initial permeability $\mu_{0}$ Mavimum permeabitify | 2000106000 | 15000 to 40000 | 1800103000 | 400 to 1000 | 700101000 |
| $\mu_{\text {max }}$ | 150001040000 | 50000 to 150000 | 120001020000 | 200000 to 400000 | 3000 to 6000 |
| Magnetislng force for $\mu_{\text {max-orsteds }}$ | 0.20 to 0.40 | 0.025 to 0.04 | 0.2 to 0.5 | 0.03 to 0.10 | 2.0108 .0 |
| Waximum fiux densily-oauss | 16000 | 8000 | 13000 | 14000 | 24000 |
| Coerclue force in oersteds for $B_{\text {max }}=5000$ gauss | 0.15 | 0.03 | 0.15 | $0.05 *$ | 2.3 † |
| Remanence in gauss for $\mathrm{B}_{\text {max }}=5000$ gauss | 4000 | 3500 | 3500 | $13000 *$ | $16000 \dagger$ |
| Hysteresis loss in ergs/ce/ cycle for $\mathrm{B}_{\text {max }}=5000$ gauss | 160 | 40 | 200 | 220 * | 125001 |
| Total loss in watts/lo for $B_{\text {max }}=5000$ gauss $50 \mathrm{c} / \mathrm{s}$ 0.015 in . sheet | 0.11 | 0.04 | 0.2 | 0.3* | 4 |
| * for $\mathrm{B}_{\text {max }}=14000$ gaiss |  |  | + for $\mathrm{B}_{\text {max }}=20000$ gauss |  |  |

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GM 6012


## Broadband Millivoltmeter, type. GM 6012

Frequency range: $2 \mathrm{c} / \mathrm{s} \cdot 1 \mathrm{Mc} / \mathrm{s}$
Measuring ranges 1 mV (f.s.d.) • 300 V in 12 steps dB scale: 80 dB up to $+52 \mathrm{~dB}(0 \mathrm{~dB}=1 \mathrm{~mW}$ into $600 \Omega)$.
Imput impedances $4 \mathrm{M} \Omega$ in parallel with $20 \mu \mu \mathrm{~F}$ (up to 3 V ) $10 \mathrm{M} \Omega$ in parallel with $10 \mu \mu \mathrm{~F}$ (in the other ranges).
! Overall aceuracy with respect to full seale:
${ }^{\circ}$ within $\pm 2,5 \%, 5 \mathrm{c} / \mathrm{s} \quad 100 \mathrm{kc} / \mathrm{s}$
within $\pm 5 \% ; 2 \mathrm{c} / \mathrm{s} \cdot: 1 \mathrm{Mc} / \mathrm{s}$
Pre-deflection : $<100 \mu \mathrm{~V}$
High Frequency Millivoltmeter, type GM 6014

Withōut
preatienuator
Frequency range: $1 \mathrm{kc} / \mathrm{s} \cdot 30 \mathrm{Mc} / \mathrm{s}$ Measuring rauge, 1 mV (f.s.d.) =

6 steps $\cdot 30 \mathrm{~V}$ in 6 steps dB seale: $\quad .80 \mathrm{~dB} \mathrm{np}$ to $.8 \mathrm{~dB} \quad .40 \mathrm{~dB}$ up to +32 dB Damping at 1, ke/s. 1 M $\Omega \quad 50 \mathrm{M} \Omega$ $1 \mathrm{Me} / \mathrm{s}: \quad 700 \mathrm{k} \Omega \quad 10 \mathrm{M} \Omega$ so Mels: $\quad 50 \mathrm{k} \Omega \quad 2 \mathrm{M} \Omega$ Jnput eapacitancé: $7 \mu \mu \mathrm{~F} \quad 2 \mu \mu \mathrm{~F}$
Pre-deflection: Compensated by electrical zero setting Variations of the frequeney characteristics:
$<5 \%$ over the whole range, with respect to the :response at the frequency of the calibration voltages.
Orerall aceuracy: $<3 \%$ with respect to full scäle and with reference to the frequency characteristic.

## (C Microvoltméter, type GM 6020

Input I.
Measuring rangei, $100 \mu \mathrm{~V}$ (f.i.d.)
10 V in-l1 steps
Input II

MO (
in parallel with
$-20 \mu \mu \mathrm{~F} \quad 10 \mu \mu \mathrm{~F}$ 10 mV (f.s.d.) 1000 V in 11 steps
$100 \mathrm{M} \Omega( \pm 1.5 \%)$ in parallel with
verall accuracy wilh respectito full scale, $3 \%$ re-deflection i< $5 \mu \mathrm{~V}$
Drift $:<1 \dot{\mu} \mathrm{~V}$ per hour after 1 h hour of warming-up tutomatic polarity indication doubles the effective cale length with respect to centre-zero instraments. C currents may be measured directly with this instrument te to the high accuracy of the input resistance.
Measuring ranget $100 \mu \mu \mathrm{~A}$ (f.s.d.) $\cdot 10 \mu \mathrm{~A}$
feeuracy $:<3.5 \%$

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Transcription Pick-Up Arm designed for monaural and stereophonic record reproduction. It is an instrument of the highest quality with its modern styling finished in Ivory, Chrome and Red. Fitted with M.P.M. 4 Plug-in moulding which accepts most cartridges, it is the companion to the Model 301 Transcription Motor.

## .... and always

## GOODMEMS

## PIONEERS OF COMPACT

## LOUDSPEAKER

## SYSTEMS

There was once a time when Full Range High Fidelity reproduction from a Loudspeaker housed in a small enclosure was considered impracticable-the text books said so and this appeared to be confirmed by experimental work.
The first real break-through came before the war-from GOODMANS-with the introduction of a high compliance twin-cone unit mounted in a totally enclosed $18^{\prime \prime}$ cube. After the war, development was taken up again and complete multiple Loudspeaker Systems were developed for use specifically in very small enclosure volumes. Again GOODMANS led the market. Then the research and development effort was directed to overcome the remaining disadvantages ; complexity, low efficiency, high cost. The result was Model A.L/120-incorporating all the valuable experience gained over many years as well as the latest developments in enclosure loading, diaphragm design, high frequency radiation, magnet design, to say nothing of advanced methods of precision manufacture.
This achievement is best judged and appreciated by ear ; the actual description of the A.L/120 is as follows :Frequency range $35 \mathrm{c} / \mathrm{s}$ to $20,000 \mathrm{c} / \mathrm{s}$ with a maximum power handling capacity of 15 Watts. Overall enclosure size- $24^{\prime \prime} \times 11_{2}^{1 \prime} \times 14_{4}^{1 \prime \prime}$. Enclosure loading-Acoustical Resistance (GOODMANS Patent No. 790997 [British]). Drive unit: $12^{\prime \prime}$ Triaxial unit comprising three concentrically mounted radiating elements, eack designed to specialise in low distortion reproduction of one part of the overall scale; bass, middle, treble; and integrated on to a common axis to approach the ideal of the "point source" radiator with its freedom from phase interference between the separate units. Bass radiation is from a large diaphragm with plastic treated high compliance suspension, with mechanical crossover to a moulded high stability mid-range radiator ; and finally electrical crossover (twin $\frac{1}{2}$-section L.C. network $12 \mathrm{db} /$ octave) to a high precision horn loaded high frequency pressure unit, with separate L-pad balance control.


Model A.L/100 also follows these lines in most respects, except that it employs a two element drive unit and provides smooth coverage from $35 \mathrm{c} / \mathrm{s}$ to $15,000 \mathrm{c} / \mathrm{s}$., with a power handling capacity of 12 Watts.
THESE LOUDSPEAKER SYSTEMS ARE DESIGNED AND BUILT $\dot{W} I T H$ GREAT CARE TO BRING TRUE HIGH FIDELITY INTO YOUR HOMECOMPACTLY, ELEGANTLY, EXCITINGLY.
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## how many ohms



Fanciful? No, because as our photograph proves, a shoelace can be made electrically conducting by impregnating it with "dag"* colloidal graphite.
Unrealistic? As a practical proposition the shoelace test is just that; as a simple illustration of a fundamental idea we think it is effective enough.
Over-simplified? Perhaps, yet thousands of cars are fitted with a non-metallic conducting braid which has much in common with a shoelace impregnated with colloidal graphite.
A non-conducting material can be made conductive by treating it with "dag" colloidal graphite, either by impregnation (surface coating or dipping after manufacture) or incorporation (addition during manufacture). Furthermore, you can impart any or all of the many other characteristics of colloidal graphite: low friction and "parting" properties, resistance to heat and wear, chemical inertness.

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## Aches on

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Ediswan valve
with a CV4000 specification


The EDISWAN S2P20 (CV4097)
Filamentary Beam Tetrode

Here is a new special quality Filamentary Beam Tetrode with a really low anode voltage, for use as an RF Power Amplifier at frequencies up to $100 \mathrm{Mc} / \mathrm{s}$.
Instantaneous filament heating enables the valve to be switched off during non-duty periods, which makes it particularly suitable for use in battery operated portable equipment. Its specially rugged construction enables the valve to withstand continuous vibration at 2.5 g and a short duration shock of 500 g .

MAIN PARAMETERS ARE AS FOLLOWS:-

| $V^{\text {f }}$ | Filament Voltage (volts) | 2.5 or 5.0 |
| :---: | :---: | :---: |
| $\mathrm{P}^{\text {f }}$ | Filament Power (watts) | 1.15 |
| $V_{a(\max )}$ | Anode Voltage, maximum (volts) | 150 |
| $V_{g_{2}(\text { max }}$ | Screen Voitage, maximum (volts) | 150 |
| $\mathrm{g}_{\mathrm{m}}$ | Mutual Conductance (mA/V) | 4.3 |
| $\mathrm{P}_{\mathrm{a}(\mathrm{max})}$ | Anode Dissipation, maximum (watts) | 5.0 |

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Mounting
Finish
Dimensions (overall)

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Zero resistance
Accuracy
Maximum current

Terminals
Mounting
Finish
Dimensions (overall)

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Depth $3 \frac{3}{4}$ ins. ( 9.5 cms .)
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## Tunnel Diodes

The current June issue of Electronic Technology includes an article on one of the latest and most promising solidstate devices-the tunnel diode.

In this, the author gives complete details of the principles of operation and electrical characteristics and discusses the desirable and undesirable features of the diode. Applications of the device are considered and are illustrated with a number of practical circuits including an oscillator, a free-running multivibrator and a divide-by-two circuit.


## ARTICLES IN THE JULY ISSUE INCLUDE:

THYRATRON PULSE GENERATOR
A pulse generator which uses a thyratron as an electronic switch in a novel circuit is described in this article. The generator produces pulses of exponential shape at reasonable output stability within repetition rates from $1 \mathrm{c} / \mathrm{s}$ to $12 \mathrm{kc} / \mathrm{s}$. The principles of operation are discussed and complete circuit details are given.
PULSE MEASURING EQUIPMENT
This article describes a system which was developed to investigate the phase difference between the signals from a single very low frequency transmitter received at two sites simultaneously.
In addition to general design considerations, the authors discuss equipment details and performance.

Electronic Technology covers all technical interests in electronics, using this word in its widest possible sense. All the familiar features of Electronic \& Radio Engineer are retained, including, of course, the well-known Abstracts and References section. Regular readership will keep you in constant touch with progress in the entire field.

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RADICAL DEVELOPMENTS IN MULLARD ALLOY DIFFUSED TRANSISTORS

## V.H.F. Communications Transistors

Here is the most direct development of the highly successful OCi 70 and $\mathrm{OCrinl}_{1-a}$ new Mullard low noise transistor for $100 \mathrm{Mc} / \mathrm{s}$ r.f. amplifier stages in professional communications equipments. Power gain is greater than ro dB and noise less than 6 dB at roo $\mathrm{Mc} / \mathrm{s}$. This is being followed by another transistor with the same performance at twice the frequency which is also intended for use in V.H.F. communications receivers.

## Extremely High Speed Logic Transistors

The alloy diffusion technique can be used to provide logic transistors for operation at a p.r.f. of $10 \mathrm{Mc} / \mathrm{s}$ which will fully 'bottom'. An especially sophisticated transistor is being made available to meet all the requirements of the logic circuits for the next generation of high speed computers. This transistor will have a $f_{1}$ greater than $300 \mathrm{Mc} / \mathrm{s}$ at iomA, a low bottoming voltage and a very rapid turn-off time. Already available is a favourably priced logic computing transistor with a frequency cut-off greater than $40 \mathrm{Mc} / \mathrm{s}$, a maximum collector current of 25 mA and a maximum voltage of 40 V .

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A new core driver transistor is to be introduced shortly for high speed storage in conjunction with the new high performance logic transistors. This has a collector current of 750 mA and a very fast rise time- $f_{1}$ is greater than $60 \mathrm{Mc} / \mathrm{s}$.
The high frequency and power characteristics of this transistor also suit it for carrier telephony and transmitting equipment where an output of three watts at $8 \mathrm{Mc} / \mathrm{s}$ can be obtained, or up to one watt at higher frequencies.

## * PNPN Switching Transistor

A wider view of the potentialities of the alloy diffusion technique has led to the development of a four layer pnpn device. This is a three terminal germanium transistor with a negative resistance characteristic which can be operated as a high speed electronic switch. After being switched "on", the transistor automatically remains bottomed until it is switched off. This characteristic makes the device particularly suited for use as a speech path cross-point in automatic telephone exchanges.
Extremely simple arrangements can be devised for astable, monostable and bistable circuits. Maximum ratings are in the region of 35 volts and 25 mA . The on/off (or series : parallel) impedance ratio is three million to one.

## * Special Avalanche Transistor

A more specialised alloy diffusion development, but one which is invaluable for some applications such as high speed sampling oscilloscopes, is a device which does not function by normal transistor action but which utilises the "avalanche" phenomenon.
This transisotor works over a limited current region, but has a high gain and an exceptionally fast rise time of the order of a millimicrosecond. Its pulse current is as high as 50 mA which represents a remarkable current rate of rise of 50 amps per microsecond.

The germanium alloy diffusion techniqueproved by Mullard in the quantity production of OCifo and OCipr r.f. tran-sistors-has been intensively developed to provide transistors for a much broader range of applications. In many new industrial fields equipment designers will be able to specify Mullard "alloy diffused" when they need high performance transistors and when economies must be borne in mind.

Specific types of industrial transistors outlined here are successively being put into production. Sample quantities are becoming available - watch Mullard announcements in the coming months for full details.

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| ary of another transformer to | 60 A | $\mathbf{6 6 7}$ | $\mathbf{£ 7 9}$ |
| give very fine changes of | 100 A | $\mathbf{6 9 9}$ | $\mathbf{£ 1 1 9}$ |
| output. Overvoltage avail-  |  |  |  |
| able as extra. |  |  |  |



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| 1,200 | $\checkmark 225$ | 630 |



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Emitter to base voltage (volts).................................. - 2
Collector current (mA)................................................... -100
Collector dissipation $\mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}(\mathrm{mW}) \ldots \ldots \ldots \ldots \ldots \ldots \ldots . . . . . . . .$.
Collector dissipation $\mathrm{T}_{\mathrm{amb}}=71^{\circ} \mathrm{C}(\mathrm{mW})$.as............... 10

| PARAMETER CHARACTERISTICS* ( $\mathrm{T}_{\mathrm{smb}}=25^{\circ} \mathrm{C}$ |  | XA141 | XA142 | XA143 |
| :---: | :---: | :---: | :---: | :---: |
| Static current amplification |  |  |  |  |
| at $\mathrm{V}_{\mathrm{ce}}=-7 \mathrm{~V}, \mathrm{I}_{\mathrm{c}}=-5 \mathrm{~mA}(\mathrm{hFE})$ maxamin.......is | Minimum | 20 | 20 | 20 |
|  | Average | 45 | 45 | 45 |
| Collector to base capacity ( pF ) | Average | 2 | 2 | 2 |
|  | Maximum | 5 | 5 | 5 |
| Gain/bandwidth product (frequency for current gain=1) at $\mathrm{V}_{\mathrm{cc}}=-7 \mathrm{~V}, \mathrm{I}_{\mathrm{c}}=-5 \mathrm{~mA}(\mathrm{Mc} / \mathrm{s}) \ldots$ | Minimum | 20 | 40 | 60 |
|  | Average | 30 | 50 | 75 |

*Typical production spreads

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Radio and Electronic Components Division
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Tel: GERrard 8660 Telegrams: Sleswan Westcent London CRC 15/67

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## JULY 1960

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## AND COMPONENTS

## COMPRRHENSIVE IEEHNIGAL HANDBOOKSERVIEE

The Mullard Technical Handbook has long been established as the comprehensive reference work for all those needing full data on Mullard Valves, Tubes and Semiconductors.
It has now been replanned : a volume on Electronic and Magnetic Components has been added and Volume 1A incorporated in enlarged Volumes 1 and 3.
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VOLUME 2 Older Receiving and Amplifying Valves and Cathode Ray Tubes, available for maintenance of existing equipment.

VOLUME 3 Power Valves and Rectifiers, Gasfilled Valves and Tubes, Cathode Ray Tubes and Microwave Devices for Industrial and Transmitting Equipment.

VOLUME 4 Semiconductor and Photoelectric Devices.

## VOLUME 5 Electrical

and Magnetic Components.
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MVM 4 :6A

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* Stability of colour registration
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* Stabilised gain of amplifiers

E.M.l. brings in the colour camera

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# "BELLING-LEE" NOTES 

No. 18 of a series
Fusing, part 2

A Fuse is a device for opening a circuit by means of a conductor which is designed to melt when an excessive current flows. Properly speaking, the term covers the complete device including any housing. The active part of the fuse, which breaks the circuit, is the Fuseelement. This may be an open wire bridging two terminals, or it may be supported on a carrier or enclosed in a cartridge, the element and its immediate container being called a Fuse-link.

One must never lose sight of the fact that a fuse is a thermal device; melting of the element occurs when its temperature is raised to the characteristic level of the material of which it is made. The melting point is a precise physical property of a metal or alloy and the amount of heat energy required to raise a given mass of it to the melting point is therefore a precise quantity under any specific set of conditions. However, any variation of these conditions such as alteration of the temperature differential, or of the heat losses, will alter the quantity of heat required to produce melting, and in relation to a fuse-element this means that its performance will be affected. Thus, a fuse-element made of a material which melts at $250^{\circ}$ C. will require appreciably less applied heat to blow it in the tropics than at the north pole or, put another way, if a fuse-element is chosen to give adequate protection to an equipment at the equator, it may not satisfactorily protect the equipment in mid-winter in northern Alaska where it might take appreciably longer to operate under the same degree of fault current. Equally, a fuse-element running in a zone of unusually high temperature, e.g., inside an enclosure where considerable heat is developed, may interrupt the circuit without any electrical fault having occurred.

There are other practical considerations associated with thermal effects. The element in an unfilled cartridge fuse-link is usually designed to be supported at the ends only, clear of the walls of the tube. However, if such a fuse-link is badly made so that the element rests on the inside of the bore, some heat will be conducted away by the tube wall, which means that additional energy will be needed to replace
this; i.e., protection will be less satisfactory than the designer intended. A similar effect can occur if the element is slack, even if it is not actually touching the inside of the tube under normal running conditions. As soon as excess current commences to flow, the expansion of the element as the temperature rises may cause it to sag against the cartridge wall, and this will slow down the rate of temperature climb, and delay fusion and opening of the circuit. This is one of the reasons why "Belling-Lee" do not enclose identity labels inside glass cartridge fuse-links, since there is a risk of them touching the elements and altering the blowing characteristic. (Another practical reason is that the very action of blowing usually so disfigures the label as to render it unidentifiable! We think it preferable to stamp the rating indelibly on the end caps, since even marking the glass is rendered ineffectual if the tube becomes broken.)

The position of mounting, and the connections to a fuse also introduce thermal effects which can affect performance. When mounted vertically, heat generated at the bottom of the element helps to raise the temperature at the upper end, and accelerates blowing. Poor contact resistance at the ends of a fuselink can lead to the development of considerable heat, and it is not uncommon for a high rating fuselink mounted in a faulty carrier to blow prematurely, or for the soldered connection between element and caps to melt. The use of a fuselink of higher rating than its carrier can produce a similar failure due to overheating of the carrier, and the fact that a carrier accepts a fuselink does not necessarily mean that they are intended to work together. The connecting links between a fuse and the circuit can also have a bearing on the performance, for together with the terminals they form a heat sink, and excessive cooling due to the use of too generous conductors can retard the rate of action of the element and impair the protection.

For all these reasons the design and testing of fuse-links are related to closely specified operational conditions which are well known to circuit designers, and fuse-link types and ratings in an installation should not be altered indiscriminately.
(To be continued)
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L. 1370/Au or Ag.

12-pole, 0.15 " Module L. $1372 / \mathrm{Au}$ or Ag.

18-pole, $0.15^{\prime \prime}$ Module

## L.I380. Guide for printed

circuit panels

## L.I381. Polarising Block

These connectors employ unique spring contacts, curved in two planes, which ensure constant, low contact resistance with minimum wear on board and connector.
They can be used with conventional or printed wiring at the solder spills. When used with printed wiring the base printed circuit should be drilled or punched with holes on a 0.05 in . grid. The plug-in board (single- or double-sided) should have a thickness of 0.0625 in . $\pm 0.005 \mathrm{in}$.
The mouldings have open ends to obviate machining of wide boards, and permit end-on grouping of connectors to accommodate broad rows of contacts.
The insertion of a board can be controlled by guides, mounted on the connector or separately. Correct polarity can be achieved by replacing one of the contacts by a Polarising Block, and slotting the board to engage this.
Other printed circuit components are available to cater for screened (coaxial) and unscreened connections, and the fitting of fuse-links. Please write for details, indicating your particular interest.

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| Type | Maximum Continuous PI.V. (V) | Maximum Transient P.I.V. ( 5 ms )* (v) | Maximum average current $\dagger$ (A) | Peak one-cycle surge current <br> (A) | Maximum average gate power (W) | Average gate current to fire (mA) | $\begin{gathered} \text { Typical } \\ \text { turn-on time } \ddagger \\ (\mu \mathrm{s}) \end{gathered}$ | $\begin{gathered} \text { Typical } \\ \text { turn-offtime } \\ (\mu s) \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SCR961 | 25 | 35 |  |  |  |  |  |  |
| SCR962 | 50 | 75 |  |  |  |  |  |  |
| 5CR963 | 100 | 150 | 10 | 120 | 0.5 | 20 | 2 | 15 |
| SCR964 | 150 | 225 |  |  |  |  |  |  |
| SCR965 | 200 | 300 |  |  |  |  |  |  |

## SEMICONDUCTORS

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## Aspects of design

This is the twenty-fourth of a series of special features dealing with advanced problems in television and radio circuit design to be published by The Ediswan Mazda Applications Laboratory. We will be pleased to deal with any questions arising from this or other articles, the twenty-fifth of which will appear in the August 1960 issue.

The ratio detector is a circuit used for the detectlon of frequency modulated signals, and, by suitable design, it can be made insensitive to unwanted amplitude modulation of the signal. The frequency modulation is detected by making use of the phase differences between the voltages across the primary and across each half of the centre tapped secondary circuit of a double tuned transformer as in the normal phase discriminator. Typical detector circuits are shown in Figs. 1 and 2.

The rectified currents in the diodes D1 and D2 are varied by the deviation of the signal frequency from its central value and the difference between these currents, flowing through the output audio load RL, produces the required audio output signal. The diodes conduct at the peaks of the input signal, allowing the stabilising capacitor $\mathrm{C}_{1}$ to be charged to a voltage dependent on the signal strength. The stabilising capacitor and its associated parallel resistance $R_{1}$ have a sufficiently long time constant to resist any rapid change of potential. Hence, if the input signal increases, the current flowing into the capacitor rapidly increases, and this adds very considerably to the load on the tuned circuits,' consequently reducing their gain. Similarly, a reduction of signal causes a reduction of loading and an increase in gain. By this means, rapid variations in the signal strength (i.e. amplitude modulation) are considerably reduced.
There is a limit to the amount of downward modulation (i.e. temporary reduction of signal) which can be handled by a ratio detector. At some value of modulation, depending on the particular circuit used, both diodes are biased beyond cut-off and the circuit fails to function as an FM detector. From tests carried out on FM receivers, under widely varied conditions of reception, it is generally considered that an acceptable design has an AM-FM rejection ratio not less than 35 dB (see below for method of measurement), and that the FM detector should handle a downward amplitude modulation of $40 \%$.

The stabilising capacitor in itself cannot entirely eliminate the effects of amplitude modulation, and various methods are used to reduce the unwanted output by producing an antiphase component. Some methods principally affect the balanced component, i.e. the AM component which is dependent on the frequency deviation, and others the unbalanced component, i.e. the AM component which is independent of frequency deviation. To obtain good AM rejection a combination of methods may be necessary.

A well designed ratio detector should meet the following requirements:-

1. The load resistance across the diodes must be low enough to reduce the average working $Q$ to a value at which the circuit can handle the maximum likely amount of downward modulation. In general the working Q should be about a quarter of the unloaded Q .
2. The coupling between primary and secondary should be less than critical coupling at the working $Q$.
3. The ratio of tertiary voltage $\mathrm{V}_{3}$ to the half secondary voltage $V_{2}$ should be nearly unity.
4. The FM detector characteristic must be linear over a frequency band which will handle a deviation of $\pm 75 \mathrm{kc} / \mathrm{s}$ plus any drift in the receiver oscillator.
5. The stabilising capacitor value must be large enough to maintain limiting action at the lowest audio frequency. This is particularly important in an unbalanced circuit where one side of the load resistance is "earthy" and the voltage across the stabilising capacitor appears in the audio outpur.
6. The AM rejection ratio should not be unduly dependent on the level of the signal input, and any method of improving the performance figure must cater for the spread in production tolerances of all the circuit components, including valves or crystal diodes.

## MEASUREMENT OF AM REJECTION RATIO

The unwanted AM audio signal depends in general on the amount by which the carrier frequency differs from the centre frequency of the discriminator curve. Thus, if the amplitude and frequency modulated signals are applied individually, the measured ratio will be very dependent on the alignment of the discriminator circuit. For this reason a better assess-
ment of the ability to discriminate against AM is made by using a signal which is simultaneously amplitude and frequency modulated, but using different modulation frequencies to enable the signals to be separated after detection. As the output from the AM may contain appreciable harmonics, it is advisable to use a lower frequency for the frequency modulation, and to separate the audio modulation output by means of a high pass audio filter.
Suitable conditions are achieved by using $30 \%$ frequency modulation (i.e. $\pm 22.5 \mathrm{kc} / \mathrm{s}$ deviation) at $50 \mathrm{c} . \mathrm{p} . \mathrm{s}$. and $30 \%$ amplitude modulation at 400 c.p.s.

The rejection ratio is then given by
rms audio output produced by FM
expressed in dB where suitable correction is made for the insertion loss of the audio filter network used in the measurement.
If the unwanted AM output is known to be mainly due to an unbalanced component, the rejection ratio can be satisfactorily obtained by measuring the audio output using amplitude and frequency modulation separately applied.

## TYPES OF RATIO DETECTORS

Two basic types of ratio detector circuits are shown in the illustrations, and obviously there are a number of ways in which their individual features may be combined. Fig. 1 uses a separate tertiary winding tightly coupled to the primary circuit, and a bifilar secondary winding to obtain a good electrical centre tap on the secondary circuit. The diode load is balanced to earth. AM rejection is improved by a suitable choice of values for resistors $\mathrm{R}_{2}$ and Rs. Fig. 2 uses a tap on the primary winding instead of a separate tertiary winding, and a capacitor tap on the secondary winding. The use of a capacity tap avoids the necessity of using a bifilar winding in the secondary circuit. The circuit, in addition, uses an unbalanced diode load, and AM rejection can be improved by a suitable choice for $R_{4}$ and $\mathrm{C}_{2}$. The unbalanced circuit is essential if the diodes in a triple diode triode valve are used in an FM detector.
In the next issue, details will be given of a ratio detector circuit based on Fig. 2 which is designed to give a consistently good AM rejection performance.


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6.3
Heater Current (amps) $I_{n}$
0.3

## TENTATIVE RATINGS AND DATA

## Maximum Design Centre Ratings

Anode Dissipation (watts)
Screen Dissipation (watts)
Anode Voltage (volts)
Screen Voltage (volts)
Heater to Cathode Voltage (volts rms)
Control Grid to Cathode
Resistance (megohms)

| $\mathrm{p}_{\mathrm{a}(\text { max })}$ | 2.5* |
| :---: | :---: |
| Pgr(max) | 8* |
| $\mathrm{V}_{\mathrm{ar} \text { (max) }}$ | 250 |
| $\mathrm{V}_{\mathrm{g} 2 \text { (max) }}$ | 250 |
| $\mathrm{V}_{\mathrm{h}}^{\mathrm{k}} \mathrm{k}(\mathrm{max}) \mathrm{rms}$ | $150 \dagger$ |

$\mathrm{R}_{\mathrm{g} 1-\mathrm{K}(\max )}$
*With grid to cathode resistance not exceeding $10 \mathrm{k} \Omega$.
$\dagger$ From cathode to higher potential heater pin.
$\ddagger$ With $\mathrm{p}_{\mathrm{a}(\max )}=2 \mathrm{~W} ; \mathrm{p}_{\mathrm{E} 2(\max )}=0.5 \mathrm{~W}$; and assuming a common anode and screen decoupling resistance $<2.2 \mathrm{k} \Omega \pm 10 \%$.
Inter-Electrode Capacitances (pF)§
Input Capacitance
Output Capacitance
Grid 1 to Anode
Grid 1 to Grid 3
Grid 1 to Grid 2
Grid 1 to Cathode
Grid 2 to Anode
$\mathrm{C}_{\mathrm{c} \text { in }}$
Cout
$\mathrm{C}_{\mathrm{BL}}-\mathrm{s}$
Cg1-83
C51-82
$\mathrm{C}_{\mathrm{gl}}-\mathrm{K}$
$\mathrm{Cg2}_{\mathrm{g}} \mathrm{E}$
Grid 3 to Anode


Tentative Characteristic Curves of Ediswan MazdaValve Type $6 \mathbf{F} 24$




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0.75 AMP AT $25^{\circ} \mathrm{C}-0.5$ AMP AT $100^{\circ} \mathrm{C}$

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| :---: | :---: | :---: |
| $C V 7030$ | $800 V$ | $8 G 7$ |
| $C V 7029$ | $600 V$ | $6 G 8$ |
| $C V 7028$ | $400 V$ | $4 G 8$ |
| $C V 7027$ | $200 V$ | $2 G 8$ |
| $C V 7026$ | $100 V$ | 168 |

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Model 125A．Both an A．C．voltmeter covering 3 to 600 kc ．in one band and an A．C．VTVM with flat response（ $\pm 0.2, \mathrm{db}$ ）from 1 to 600 kc ．Selectivity settings of 250 cps and 2.5 kc ．Measures voltages from -90 to +32 dbm within $\pm 1 \mathrm{db}$ ；frequencies $\pm 1 \mathrm{kc}$ ．to 100 kc ．，and $\pm 2 \mathrm{kc}$ between 100 and 600 kc ．As a flat A．C．VTVM，it has a range of -30 to +32 dbm ． 40 in ．precision frequency scale．

## BENTLEY

EXPRESS POSTAL SERVICE ：ALL ORDERS DESPATCHED SAME DAY AS RECEIVED．


| OA2 ．．17／6 | 6B8G ．．4／6 | 6SQ7GT 9\％－ | 1207GT 51－ | 85A2． $15 /-$ | DM170 7／8 | ${ }_{\text {EF36 }}^{\text {EF37 }}$ |  | $\begin{array}{lr} \text { GZ33 } & 19 / 11 \\ \text { GZ34 } & 14 /- \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OB2 ．．17／6 | 6BA6 ．．7／8 | BU4GT 12／6 | 128 A7 8／8 | 150B2 16／－ | EA50 ${ }^{\text {a／a }}$ | ${ }_{\text {EF3 }}$ EF3 3 | 8；－ |  |
| 024 ．． $51-$ | 08E6．．81－ | 6U5G 7／8 | 128 K 78 － | 185BT 38／2 | EA76 9／6 | EF39 | $5 / 6$ | Habcso |
| $145 \cdots 81-$ | 6BG6G 23／3 | 6V6G 7／－ | 12sQ7 11／6 | 185 BTA | EABCSO 9 － | EF40 | 15／－ | $\mathrm{H}^{13 / 6}$ |
| 1A7GT 12／－ | ${ }_{6} \mathrm{BHH6}^{81} 8$ | 6V60TG 8／－ | 12807816 | 33／2 | EAC91 $4 / 8$ | EF41 | 10／8 | $\begin{array}{ll}\text { HL23 } & 15 / 3 \\ \text { HVR2 } & \text { 20／－}\end{array}$ |
| $1 \mathrm{CJ}^{\text {a }}$ ． $12 / 6$ | 6BJ6 6／－ | $6 \times 4$ 5／－ | $145787 / 10$ | 807 ．－7／8 |  | EF42 | 10／6 | $\begin{aligned} & \text { HVR2 } \\ & \text { HVR2A } \\ & 60 /- \end{aligned}$ |
| 1D5 ．． $91-$ | 6BQ7A 151－ | $6 \times 50 \mathrm{~T}$ 8／－ | 19AQ5 10／6 | 4033L 12／日 | EB34 ${ }_{\text {E }}$ | EF50 ${ }^{\text {E }}$（ | （）） $7 /-$ |  |
| 1D8 $. .10 / 8$ | fibR7 15／－ | 8／30L2 10／－ | 19B66G23／3 | $5763 \ldots 12 / 6$ | EB41－． $8 / 6$ | EFSOOL | ） $51-$ | KT2 ${ }^{\text {KT3 }}$－ 101 － |
| 1HSGT 10／6 | 6BW6 $8 / 6$ | $7 \mathrm{B6} \quad . .21 / 3$ | 20D1 ．．15／3 | AC6PEN7／日 | EB91 4／－ | EF54 | 51－ | KT33C 10／－ |
| 1L4 ．．4／8 | 6BW7 7／－ | 787 ．．8／6 | 20F22．． $28 / 6$ | ATP4 ．．5／－ | EBC33 5／－ | EF73 | 10.6 | KT36 29／10 |
| 1LD5 ．． $6 /-$ | $6 \mathrm{BX6}$ 71－ | 708 ．．81－ | 20L1 ．． $26 / 6$ | AZ31 10／－ | EBC41 8／8 | EF80 | 7／－ | KT41 $12 / 6$ <br> KT44 $12 / 6$ |
| 1LN5．．${ }^{\text {b }}$ | $6 \mathrm{Cl}_{4}$ ．．．3／－ | 706 ．．8／－ | 20P1 ．．28／6 | A241 13／11 | EBC81 8／－ | EF85 | $7{ }^{1 /}$ | KT44 $12 / 6$ |
| 1N5GT 10／8 | 6050 6／6 | 7106 ．．10／6 | 20P3－．28／3 | B36 15／－ | EBF80 9／－ | EF86 | 10／8 | $\begin{array}{ll}\text { KT61 } & 12 / 6 \\ 7 /-\end{array}$ |
| 1R5 ．． $8 / 6$ | 6CD6G 36／8 | 7177.8 | 20P5－．23／3 | BL63 7／6 | EBF83 | EF89 | $\stackrel{9 /-}{ }$ | KT63 7\％－ |
| 184 ．． $91-$ | 6CE6 $\%$ | 787 ．．9／8 | 25 A6C 10／6 | CBL31 23／3 | 13／11 | EF91 | 4／6 | KT66 151－ |
| 185 ．．61－ | $6 \mathrm{E} 5 . .12 / 6$ | 7Y4 ．．7／8 | 25L6GT10／－ | CCE35 23／3 | EBF89 9／6 | EF92 | $4 / 6$ | KTW81 6／8 |
| 1T4 ．．${ }^{1 / 6}$ | 6 Fl ．．28／8 | 8D3 ．．4／6 | $25 \mathrm{Z4G}$ 9／6 | CL33 19／3 | ELB21 23／3 | EF97 | 13／3 | KTW62 7／6 |
| 1U4．． $12 / 8$ | 6F6G ．．7／－ | 9BW6 15／3 | $25 \mathrm{Z5}$ 9／6 | CV63 10／6 | EBL31 23／3 | EK 32 | 8／6 | KTW63 8／8 |
| 1U5 ．． $6 /-$ | $6 \mathrm{Fl2}$ 4／6 | 10Cl ．．12／－ | 2526G 10／－ | CY1 18／7 | EC52 5／6 | EL32 | $5 .-$ | KTZ41 8／－ |
| 2x2 ．． $4 / 8$ | 6F13 ．．11／8 | $10 \mathrm{C2}$ 26／6 | 27SU 19／11 | CY31 16／7 | EC54 6／－ | ELS3 | 12／6 | KTZ63 7／6 |
| 344 ．． 81 － | $6 \mathrm{FF}^{23} \quad 10 / 6$ | 10ヶ＇1 26／8 | 2807 ．．7／－ | D1 ．．3／－ | EC70 12／6 | EL34 | 15／－ | L63 ． 6 6／－ |
| 3 AS ．． $10 / 6$ | 6F33 ．．7／6 | 10F9 ．．10／8 | 30 Cl ．．8／－ | D15 ．．10／6 | EC92 13／3 | EL38 | 28／6 | MU14 8／－ |
| $3 \mathrm{B7}$ ．．12／6 | 6G6 ． 916 | 10LD3 8／8 | 30 F 5 7－ | D43 ．．17／3 | ECC31 15／－ | EL41 | $91-$ | MX40 15／－ |
| 3D6 ．．51－ | 6H6GT 3／－ | 10LD11 | 30 FL 1 10／－ | D77 ．．4／－ | Ecce32 5／6 | ELA2 | 10／8 | N37 19／11 |
| $3 \mathrm{Q4}$ ． 776 | 6．J5G ． 5 \％ | 15／11 | 30 L 1 ．．8i－ | DAF91 8／－ | ECC33 8／6 | EL81 | 12／6 | N78 19／11 |
| 3Q5GT $9 / 6$ | 6 J 6 ．． 56 | 10 P 13 15／－ | $30 \mathrm{Pl2}$ ．．7／6 | DAF96 8／8 | ECC34 24／7 | EL84 | 7／8 | N108 19／11 |
| 384 ．．71－ | 6J7G ．．B／－ | 10P14 19／3 | $30 \mathrm{P16}$／78 | DD41 13／11 | ECC35 8／6 | EL85 | 13／11 | N308 20／7 |
| $3 \mathrm{~V}^{4}$ ． $7 / 8$ | 6K6GT 8／－ | 12A6 5／－ | $30 \mathrm{PL1} 11 / 6$ | DF66 15／－ | ECC81 6／－ | EL91 | 5－ | N339 15／－ |
| 5R4GY 17／6 | 6K7G 5／－ | 12AC6 15／3 | 35 A5 21／3 | DF70 15］ | 12CC82 6／6 | EL95 | $10 / 6$ | PABC80 |
| 5U4G 6／6 | 6 K 8 G 6／6 | 12AD6 $17 / 3$ | 35L6GT 9／6 | DF91 4／6 | ECC83 7／日 | EM34 | 9／6 | 13／11 |
| 5VAG 10／－ | 6 K 251811 | 12AE6 13／11 | $35 W 4$ 7／6 | DF96 $8 / 6$ | ECC84 9／－ | EM71 | $23 / 3$ | PCC84 $81 /$ |
| 5 EBGT 6／8 | 6 LI ．．23／3 | 12AH8 12／8 | $35 \mathrm{Z3} 10 / 6$ | DF97 ．－9／－ | ECC85 8／6 | EM80 | 81－ | PCCSE 9／6 |
| 8Z3 ．．12／6 | $6 \mathrm{L6G}$ 8／－ | 12AT6 7／6 | 35Z5GT 9／－ | DH63 6／8 | ECC8823／11 | EM81 | $9 \mathrm{j}-$ | PGC88 $23 / 11$ |
| 574G 9／－ | $617 \mathrm{GT} 7 / 6$ | 12AT7 B／－ | 43 ．．．．10／－ | DF76 5／－ | ECF80 10／6 | EMR4 | 10／6 | PCC89 11／6 |
| 6489 9／－ | 6 L 18 13／－ | 12AU6 23／3 | 50C5 12／6 | DH77 7／－ | ECCF82 10／6 | EN31 | 3\％－ | PGF80 8／－ |
| 6 6AB8 9／－ | $61 / 19$ 23／3 | 12AU7 8／6 | 50CD6G | DK40 21／3 | FCH21 23／3 | EY51 | 9／－ | PCF82 $10 / 6$ |
| 6AC7 4／－ | 6LD20 15／11 | 12AV6 12／8 | 38／6 | DK91 6／6 | ECE35 9］6 | EY83 | 18／7 | PCL82 10／－ |
| 6ags 5／6 | 6N7 ．．8／－ | 12AX7 7／8 | 35Z4GT 8／－ | DK92 9／－ | ECR42 9／－ | EY80 | 9／－ | PCL8s 11／6 |
| 6AKS 8／－ | $6 \mathrm{P} 2512 / 8$ | 12BA6 8j－ | 50L6GT $8 / 6$ | DK98 8／6 | ECH81 9／－ | EZ40 | $7 /$ | PCL84 $12 / 8$ |
| 6 6L5 4／－ | 6 P 28 28／6 | 12BE6 9／－ | 53 KU 19／11 | DL66 151－ | ECE83 | E241 | 71－ | PEN45 19／6 |
| 6 6M6 4／6 | 6Q7G 8／6 | 12BH7 21／3 | $72 . . .4 y^{8}$ | DL68 15／－ | 13／11 | EZ80 | 7－ | PEN $467 / 6$ |
| 6AQ5－7／6 | 6R7G 10／－ | 12J7GF 9／8 | 78 ．．．．6／6 | DL92 \％ 7 － | ECL80 9／－ | ER81 | 710 | PEN383 |
| ＂AT6 7\％－ | 68A7CT 8j8 | 12K5 17／11 | 80 ．．．．9／－ | DL94 7／6 | ECL82 10／6 | FC4 | 15／－ | $23 / 3$ |
| 6aUf 101－ | 6SL7GT 8／6 | $12 \mathrm{~K} 7 \mathrm{GT} 5 / 6$ | 83 ．．．．．151－ | DL96 $8 / 6$ | ECL83 19／3 | G730 | 91 | PL33 11／3 |
| Av6 12／8 | 68N7GT 5／6 | 12K8GT14／－ | 83 V ．．12／8 | DL810 $10 / 6$ | EF22 14i－ | G732 | 10 | PL36 12／－ |

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| :---: | :---: |
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| Q8150／15 | UBC41 8／8 |
| 10／6 | URC81 11／4 |
| 212 ．．9／－ | UBr80 ${ }^{\text {¢ }}$ 9－ |
| R18 ． $14 /$－ | UBF89 9／6 |
| R19 ．19／11 | UBL21 23／3 |
| SD6 ．．．12／－ | UCC84 $14 / 7$ |
| 8P41 ．．3／6 | UCC85 9／－ |
| 9P61 ．．3／8 | UCF\％ $18 / 7$ |
| SU25 ．．28／8 | UCH42 916 |
| sU61 9／－ | UCH81 9／6 |
| T41 ．．23／3 | UCL82 11／6 |
| TDD4 $12 / 6$ | UCL83 19／3 |
| TP22 ． 15 ／－ | UY41 9／－ |
| TP25 15／－ | $\begin{array}{lll}\text { UF42 } & 12 / 6\end{array}$ |
| TY86F 13／3 | UF80 $10 / 6$ |
| U12／14 8／6 | UF85 9／－ |
| U16 ．．10／－ | UF86 17／11 |
| U18／20 8／6 | UF89 9／－ |
| U19 ．．38／－ | UL41 90－ |
| U22 ．． 81 | TL44 26／6 |
| U24 ． $89 / 10$ | UL46 1416 |
| U25 17／11 | UL84 816 |
| U26 ．．10！－ | UM4．17／3 |
| U31 ．${ }^{9 / 8}$ | UM80 15／3 |
| U33 ． 2818 |  |
| $\begin{array}{ll}\text { U35 } & . .28 / 8 \\ \text { U37 } & \text { ．} 26 / 6\end{array}$ | $\begin{array}{ll}\text { URIC } & 9 /- \\ \text { UU8 } & 26 / 8\end{array}$ |

3．30 P．M．

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AVOMETER MODEL D. 88.19 .6 (P. \& P. 3/6)
D.C. Volts A.C. Volts D.C. Current A.C.Current $150 \mathrm{mV} . \quad 7.5 \mathrm{~V} . \quad 15 \mathrm{~m} / \mathrm{A} . \quad 75 \mathrm{~m} / \mathrm{A}$. $\begin{array}{llll}5100 \mathrm{mV} . & 15 \mathrm{~V} . & 30 \mathrm{~m} / \mathrm{A} & 150 \mathrm{~m} / \mathrm{A} .\end{array}$ $\begin{array}{llll}1.5 \mathrm{~V} . & 75 \mathrm{~V} . & 150 \mathrm{~m} / \mathrm{A} & 750 \mathrm{~m} / \mathrm{A} .\end{array}$ $\begin{array}{llll}3 \mathrm{~V} . & 150 \mathrm{~V} . & 300 \mathrm{~m} / \mathrm{A} . & 1.5 \mathrm{Amps.} \\ 15 \mathrm{~V} . & 300 \mathrm{~V} . & 1.5 \mathrm{Amps} . & 7.5 \mathrm{Amps.}\end{array}$ $\begin{array}{llll}30 \mathrm{~V}_{.} & 600 \mathrm{~V} . & 3 \text { Amps. } & \\ 150 \mathrm{~V} . & 750 \mathrm{~V} . & 15 \text { Amps. } \\ 300 \mathrm{~V} . & 1.5 \mathrm{KV} . & 30 \text { Amps. } & \\ 750 \mathrm{~V} & & \text { Resistance }\end{array}$ $\begin{array}{ll}730 \mathrm{~V} . & 0-1000 \text { ohms. } \\ 1.5 \mathrm{KV} . & 0-10 \mathrm{~K} \text { ohms. }\end{array}$
Thoroughly overhauled. Complete with batteries and inatructions. An extremely robust meter at a very reasonable price.

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 Funnel cooled. A.C. input 45 v. RM5. D.C. outpur 30 v. 10 amps. BRAND NEW. Boxed. 45/-. Post $3 / 6$.MARCONI IMPEDANCE BRIDGE Type TF373. Measures, $L, C$ \& $R$ at 1,000 cycles. Aecuracy $1 \%$. $0-100 \mathrm{H}$; 0-100uF; $0-1 M \Omega$ each in 5 ranges. Power Factor and "Q." First-class condition, 435, carr. paid.

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| 300 Volts A.C. M/T 6 in . | Fiush Circular. Mide 1955 | 7916 |
| 300 Volts A.C. M/T 21 ln . | Flush Circular . . . . . . . . . . | 25 |
| b00 Volta A.C. M/I 2 ) in . | Fluab Circular | 25/- |
| 40 Auperes D.C. M/C 2 in . | Flush Circular | $7 / 6$ |
| METAL RECTIFIERS. Fult w $1 \mathrm{~mA} .8 / 6,5 \mathrm{~mA} .8 / 6 . \quad 8 T C 2 \mathrm{~m}$ | bridge. BRAND NEW. 5/6. | liond |

SIGNAL GENERATOR TF390G. These are complete with A.C. mains power pack and incorporate a $2 \frac{1}{2}$ in. 50 mieroamp. meter for carrier level mounting. There are 4 ranges, 1.6-2.4 Mc/s., $8-16 \mathrm{Mc} / \mathrm{s}$., $16-32 \mathrm{Mc} / \mathrm{s}$., and $32-60 \mathrm{Mc} / \mathrm{s}$. They are less their original calibration charts so guidance readings only are provided. In first class mechanical and working only are provided. In first class
condition. $54 / 19 / 6$. Carriage $10 / 6$.

> CHARLRS BRITAIN (Radio) LTD. II UPPER SAINT MARTIN'S LANE LONDON, W.C. 2 TEMple Bar 0545
One minute from Leicester Sq. Station. (Up Cranbourne Sr.) Shop Hours: 9-6 p.m. (9-1 p.m. Thursday.) Open all day Saturday

## G.E.C. SELECTEST DIII



This testmeter has exactly the same ranges as the Avo "D." The scale is even larger. Those we offer are in first class condition, completely overhauled and carefully tested prior to despatch. Complete with battery, prior to despatch. Complete with battery,
test leads and insrructions. $£ 7 / 10 / \%$. test leads
P. \& P. $3 / 6$.

ELECTROSTATIC METER. Dia. $6 \frac{1}{2} \mathrm{in}$. ELECTROSTATIC METER. Dia. $6 \frac{1}{2} \mathrm{in}$.
reads $5-18.5 \mathrm{Kv} Manufactured 1953.$. reads $5-18.5 \mathrm{Kv}$. Manufactured 1953.
Contained in wooden case $10 \times 10 \times 9$ 9in. Contained in wooden case 10
high. $£ 9 / 19 / 6$. Post paid.

SANGAMO.WESTON ANALYSER E772. A useful multi-range meter Thoroughly overhauled and in perfect working order. For full derails see previous adverts. $£ 7 / 10 /$. Carr. $4 / 6$.
MARCONI TF987/I NOISE GENE= RATORS. Range $100 \mathrm{Kc} / \mathrm{s}$. to $200 \mathrm{Mc} / \mathrm{s}$. Determines noise factor of $A M$ and $F M$ receivers. Fully stabilised H.T. supply A.E. mains operation. Brand new and in original boxes. $£ 15$. Carr. $7 / 6$.
MARCONI TF. 340 OUTPUT MET. ERS. Perfect working order, $£ 9 / 19 / 6$. Carr. 7/6.

HEAVY DUTY SLIDER RESISTORS $1.25 \Omega 20$ A. $12 / 6$, post $3 / 6$. I $\Omega 12$ A., $8 / 6$. ZENITH ÁDJUSTABLE $25 \Omega 4$ A., $8 / 6$ PEN 2/6.
PRECISION RESISTORS. I Megohm 1\% I watt wire wound, Ex-U.S.A. BRAND NEW. $10 / 6$ per dozen.
D.C./A.C. CONVERTERS. Input 12 v . D.C. Output 230 v. $50 \mathrm{c} / \mathrm{s}$. A.C. at 135 warts. Fitted with $0-300$ v. A.C. $2 \frac{1}{2}$ in. meter and slider resistor for voltage adjustment. In stout wooden carrying case with lid. Perfect working order. f9/19/6. Carr. 10/6.
24 v. Input 230 v. A.C. $50 \mathrm{c} / \mathrm{s}$. 100 watts output. In grey metal case. BRAND NEW. $92 / 6$. ${ }^{\text {output. }} \mathbf{~ C a r r . 6 . ~}$
RADIATION METERS. Pórtable dose rate meter, containing modern type rectangular 50 micro-amp. meter, CVX494 electrometer valve, etc. BRAND NEW. In canvas carrying case, $£ 3 / 19 / 6$. Post $2 / 6$. for details of other equipment, see our previous adverts.

## MOVING COIL PHONES. Finest

 quality Canadian with chamois ear-muffs and leather-covered headband. With lead and jack plug. Noise excluding and supremely comfortable. 19/6. Post $1 / 6$. supremely comfortan . Hi impedance) i.e. for HRO, CRIOO, ete., with standard jack plug, $4 / 6$.MIKE/HEADSET, all moving coil. As used on 19 set. BRAND NEW. $12 / 6$. Post $2 /$-.

## FERRANTI

 YOLTMETERS N5. 0-300 voles, 25 $100 \mathrm{c} / \mathrm{s}$. Moving iron, 6in. scale. Fl. meg. Hers metically sealed. meticaly sealed. grade IN. Made NEW. BRAND NEW. Boxed.$79 / 6$. Pose $3 / 6$.


## 제 Smpir



WESTON MODEL 772 TESTMETER


| A.C. VOLTS | D.C. | A.C. CUR |
| :--- | :--- | :--- |
| 2.5 v | CURRENT | RENT |
| 10 v. | $100 \mathrm{micro} / \mathrm{a}$. | 500 ma. |
| 50 v. | 1 ma. | 1 amp. |
| 250 v. | 10 ma. | 5 amp. |
| $1,000 \mathrm{v}$. | 50 ma. | RESIST. |
| D.C. VOLTS | 100 ma. | ANCE |
| 2.5 v. | 500 ma | 100 ohms |
| 10 v. | OUTPUT | 1,000 ohms |
| 50 v. | METER | 100 k. ohms |
| 250 v. |  | 10 megohms |

Supplied in perfect working order complete with internal batteries. $\mathbf{E 7 / 1 0 / - . ~ P / P . ~ 4 / - ~}$


FIELD TELEPHONES TYPE F. Generator bell ringing. Supplied complete with batteries fully tested and complete with wooden carrying case 59/6 each. P/P. 3/6. 5/- pr.


DON Mk. 5 FIELD TELEPHONES Ideal for all inter-communication. Buzzer calling. Supplied fully tested, complete with batteries and instructions. 39/6 each, P/P. 3/6 ea., 5/- pr.


BRAND NEW MEDRESCO HEARING AIDS


Fully tested, complete with earpiece. all necessary leads and battery pouch. incorporates three sub-miniature valves and sensitive crystal microphone. Price only 32/6 each, plus 1/- P. \& P. Batteries 5-/ extra.


## COSSOR 339 double beam OSCILLOSCOPES

Operation $110 / 200 / 250$ volts A.C. Ten position time base, 6 cps . to $250,000 \mathrm{cps}$. Amplifier 10 cps , to $2,000,000 \mathrm{cps}$. Periect working order,

## only £15 Each

Carriage 10/-


MARCONI TF. 373 UNIVERSAL IMPE: DANCE BRIDGE. Reconditioned to makers spec. $1,000 \mathrm{c} / \mathrm{s}$. Ranges: 100 H .100 mid . I MEG. 100 Q. $200 / 250$ v. A.C. operation. E35 each.

PADDED MOVING COIL HEADPHONES Good quality complece with moving coil hand mike. Brand new. $12 / 6$ per set. P. \& P. 2/-

PHOTO VOLTAGE AMPLIFIERS. These special units contain a I microamp. Tinsley mirror galvo and a double selenium photo cell. Brand new, $£ 9 / 19 / 6$ each. P/P. 7/6.
MARCONI TF-329 " $Q$ " METERS. Range 0 to 500 Q . Frequency $50 \mathrm{kc} / \mathrm{s}$. $1050 \mathrm{Mc} / \mathrm{s}$. $200 / 250$ volts A.C. operation. Reconditioned to maker's spec. 265 each.

## MARCONI TF. 428 B/I. VALVE VOLT.

 METERS 5 ranges A.C. and D.C. I.5, 5 , 15,50 and 150 volts. Complete with internal H.F. wrobe. Operation 200/250 volts A.C. Brand new, \&17/10/- each. P/P. 10/-.MINE DETECTORS No. 4a Complete equipment comprises Search Head Amplifier Headset, Control Box, Telescopic Rods for Search Head, Search Head Test Unit and Test Depth Measure and Haversack. Opseration is from a standard $60 \mathrm{~V} . / 1.5 \mathrm{v}$. combined dry battery. The unit will detect lerrous or nonferrous metals to a depth, of 24 in . giving maximum signal but can be used at g'reater depths giving lower output. ideal for tracing underground pipes or cables and any hidden metallic obiects.
Complete equipment supplied brand new in original transit cases complete with circuit and operating instructions.

AMERICAN SUPER LIGHTWEIGHT HEADSETS. Res. 50 ohms. Brand new, $15 /$-. P/P. I/6.

SOUND-POWERED TELEPHONE HANDSETS. No batteries required. 15/- each. P/P, I/6.

LEACH 12 VOLT AERIAL C/OVERRELAYS. Double pole, 7/6 each. P/P. 9d.

MUIRHEAD PRECISION STUD SWIT. CHES. 4 bank, 4 pole 24 positions. New, boxed 17/6 each. P/F. 1/3.

CR. 100 SPARES KITS. Contains 15 valves, resistors, pots, condensers, output trans., etc All brand new, $59 / 6$ set. P/P. 3/6.

24 AMP. VARIAC TRANSFORMERS. 230 v . input. Variable output 185 to 250 volts. Can be used reversely giving 230 volts out with variable input. \&/2/10/-. P/P. 10/-.

1,000 WATT MAINS ISOLATION TRANSFORMERS. 230 to 230 volts. Heavy duty, exAdmiralty. New, boxed, 65 each. P/P. 10/-

750 WATT AUTO TRANSFORMERS. Tapped from 110 to 230 volts. Fine heavy duty type, 69/6 each. P/P. 5/-

AR. 88 WAVECHANGE SWITCH ASSEMBLY. Brand new with screens, $17 / 6$ each. P/P. $2 / 6$.

MARCONI TF-5I7 SIGNAL GENERATORS. $10-18 \mathrm{Mc} / \mathrm{s}$; $33-58 \mathrm{Mc} / \mathrm{s}$ : $150-300 \mathrm{Mc} / \mathrm{s}$. $200 / 250 \mathrm{v}$. A.C. operation. 65/- each FOR CALLERS ONLY.

24 VOLT ROTARY CONVERTERS Input 24 volts D.C. Output 230 volts A.C. 50 cycles, 100 watts. Housed in metal carrying case with inlet/ outlet plugs. Brand new, 92/6 each. P/P. 7/6.


## R. 1155 RECEIVERS

Standard Model B with improved geared drive, perfect order, $£ 8 / 19 / 6$ each, $7 / 6 \mathrm{P} / \mathrm{P}$. Trawler Band Model L. or N, $£ 12 / 19 / 6$ each. PiP. 7/6. Combined Power Pack and Audio Output Stage suit either model, 85/-extra.

## ROTARY CONVERTERS



12 v. D.C. input 230 volt A.C. 150 watts 50 cycles output. Housed in wooden case and fitted with voltage control slider resistance swireh plugs resistance switch, plugs and A.C. mains voltmeter. Supplied in perfect condition, individually tested $69 / 19 / 6$ each. P/P. 10/-.

## EC 221 HETERODYNE FREQUENCY METERS

$125 \mathrm{kc} / \mathrm{s}$ to $20 \mathrm{mc} / \mathrm{s}$
Complete with all vaives, crystal, headset and instruction book, but less calibration charts. 100\% condition.
$\underset{\substack{\text { special } \\ \text { SRice }}}{ } \mathbf{£ 1 4 - 1 0 - 0}$
Curisectiox axt.


Carriage 10/6.


## Portable/Mobile

V.H.F RADIO TELEPHONE


A modern 14 -valve superhet receiver and AM transmitter using current series of B7g valves. Valve line-up: 2-CV136/7D9, 1-CV137/EAC91, 7-CV138/EF91, 4-CV416/ 6F17. Robust cast aluminium case includes 5 in . loudspeaker. Internal vibrator pack (synchronous type) provides operation from 12 -volt accumulators or vehicle or boat 12 -volt supply, in fixed or mobile use. Available, less crystals and accessories, but with connecting plugs, ex-stock. Accessories and crystals for specified frequencies in the range $60-95 \mathrm{Mc} / \mathrm{s}$ can be supplied to order at extra cost.

Each unit is fully tested and in good condition. Price (including packing FOB London), £20 each. Special quotation for quantities up to 500 sets.

## 50 MICRO AMP MOVING COIL METERS

(Brand New \& Boxed)

Made on Government Contract by Famous British Maker 3 g in . Square- 800 ohms resistance. 4 Scales operated by lever "Set-zero"-"0-3"-" $0-30$ "-_" $0-300$." Easily coupled to rotary range switch by cord or lever. Ideally suitable for transistor tester, output meter, volt-milliameter.

## A RANGE OF METER BOXES

Completely finished and enamelled, with all screws, sockets, etc., designed to take one or two meters and with provision for controls, caters for all kinds of applications of this meter.
Boxes to take one Meter, small $5 / 6$, medium 7/6, large 10/6. Two Meters, small 9/6, large 15/6.
Circuits for many applications-free.


Complete with data and circuits


and NOW - WE INTRODUCE
tWO UNITS METICULOUSLY MATCHED TO CORRECTLY OPERATE
THE NEW GARRARD "MAGAZINE" TAPE DECK

| - MODEL | HF/G2P TAPE PREAMPLIFIER |
| :--- | :--- | :--- | :--- |
| MODEL | HF/G2A |
| TAPE AMPLIFIER |  |

Model Based on the very suecessful MULLARD tape DESIGNS, incorporating HF/GR2 only HIGH GRADE COMPONENTS and MULLARD VALVES.

Both Units form an entirely new "Easy to handle " presentation, each is completely seif contained with power supply, Loudspeaker (Amplifier HF/G2A only) and all INPUT and OUTPUT sockets being incorporated on the chassin, whleh liselif $1^{s}$ conatructed to allow for direct attachment to the tape deck (as shown in illustration). Thus the tape deck with the Amplifer (or Preamplifer) fxed to It form ONE COMPLETELY EELF-CONTAINED WORKING UNIT whtch requires only screwfing into a Cablnet and Connecting to the Mains supply.

Model EF/G2A Amplifler
A Complete Tape Amplifer-Incorporating

- Magic Eye Level Indicator
- Volume Control.
- Superimpose switch
- Effective Tone Control.
- Monitortag Facilities.
- Extension Loudspeaker Socket
- Inputs for recordlng from Mike, Gram. and Radio Tuner.
- Incorporates Loudspeaker and Power Supply on Chassie.

Model HF/GRP Preampliter
Forms the Ideal " Luk " to add Migh Quality Tape Recording facilities to exletling Audio Installations, such an our suitable to operate through the Pick-up Sockets of most Radio Recelvers.
It incorporates:

- Magic Eye Level Indicator and Control.
- Supertmpose switch.
- Inputs for recording from Mike, Gram. and Radio.

W) OFFER AS FOLLOWS (a) MODEL HF/G2R PORTABLE $\mathbf{T A P E}$ RECORDER, Includes $\mathbf{~} 33.0$日pool of L.P. tape and crystal microphone.
H.P. TERMS: Deposit $£ 6 / 12 /-, 12$ month. payments $£ 2 / 8 / 5$. AMPLIFIER and TAPE DECK. £27.10.0 Includes spool of L.P. tafle and loudspen H.P. TERMS: Deposit $£ 5 / 101-, 12$ monthly
(o) ASSERBLEED and TESTED AMPLIFIER MODEL HF/GRA. E.P. TERMS: Deposit £3, 12
(d) MODEL MFG/GPP PORTABLE PREAMPLIFIER Compiete in
portable case (Ilke $H F / G 2 R$ ). H.P. TERMB: Deposit $£ 6.12$ ments $£ 2 / 14 /$ (e) MODEL HF/GRP-D comprising PREAMPLIFIER Rnd TAPE DECK. Includes syool of L.P. tap paymente $£ 1 / 18 / 2$,
(f) ASSEMBLED ASSEMBLED \& MESTED PREH.P. TERMB: Deproit \&2/16.
£15.0.0
onthly pay-
£30.0.0
£26.0.0
2 monthly
£14.0.0 12 monthis


## ! RADIOGRAM CHASSIS ! !

## ARMSTRONG MODEL A F 208

Complete AM/FM chassis, sepsrate Bass and Treble controls ARMSTRONG "STEREO" TWELVE.
most complete $/ \mathrm{FM}$ atereo chassis yet produced. ARMSTRONG" JUBILEE" provlding chassi
ARMSTRONG AM/FM "STEREO 44*
Provision is made for Btereo and Monaural playback from pick.up
RADIO TUNING UNITS
The JASON "MERCURY" mk. If Switehed FM Tuner
Choice of 4 stations plus 1.T,V, and B.B.C. T.V, Bound transmissions, PRICEA: KIT OF PARTB $£ 10 / 10 /-$. ABSEMBLED.
DULCR MODEL FMT/2
 A complete seif "powered FM Tuner incorporatin
ARMSTRONG " $\$ . T .3$." AM/FM Tuning Unita. A self-powered therer covering VHF, medium and long wave bands with nutomatic frequency control on VHF.
DULCI "F4/T" AM/FM Tuning Units.
A 4owavebsad seli-powered tuner covering the FM traismission plus the long, medlum and short wavebands.
NEW RIRE PUROHASE TERMS are available on all above. Ilustrated leaflers available send B.A.E. (Carr, and Ins, 5/- extra).

STERN'S MK. II "fidelity" F.M. TUNING UNIT (Pius 5/- carriani ina) FIRE PURCKASE: IDenosit PRICE 214.5.0 f2/17/-and 13 mionths at f1/O/11. Incorporates
the Litent. JULLARD PERMEABILITY TUNTNG FEAIET tond the corresponding MULLARD VALVE FEAtRT bind the corresponding MULLARD VALVI
LINE UF compralig ECC85, 2 type EFSS: (or EF893). EMSt, Tuning Indieator, plus 2 type O.A. 7is Germanium Diodes. A really frat-clans Tuner very attractively presented and comparable to many of fered at much higher prices. Power consumption is only 1.5 amps at 0.3 voits and $25 \mathrm{~m} . \mathrm{a}$. at 250 volts.
! HOME CONSTRUCTORS


## THE "ADD-A-DECK

incorporating the
NEW B.S.R. ©MONARDECK* and MATCHED PREAMPLIFIER 817.17.0 Deposit $£ 3 / 12 /-$ E/6/2 (plus $7 / 6$ carr. and ins.) Sockets of the standard RADIO RECEIVER hrough which first-class results are obtained consists of a single speed Twin Track Tape Deck incorporating matched Preamplifier, and operates at 3 lin./sec. speed. lt uses 5 in . Tape Spools, thus providing up to $1 \frac{1}{2}$ hours' playing time on L.P. Tapes or I hour on the standard 6 in . Tape Spools.

The equipment is supplied fully rested and completely assembled on an The equipment is supplied fully tested and completely assembled on an an existing cabinet and only requires connections to the mains supply and the Pick-up Sockets, for which purposes "floating" leads are incorporated on the Preamplifier.

## STERN'S 12 VOLT CAR <br> RADIO incorporating

PRINTED CIRCUIT and POWER TRANSISTOR
A versatile design covering both LONG and MEDIUM WAVEBANDS, incorporating 12 volt car battery. We offer it on the UNIT ABBEMBLIY BASIS ...consisting of 12 vol car battery. We offer it on the UNIT ABSEMBLY BAASI THict consisting of Only 12 solder jofnts are required to finish the complete receiver.
Send $1 / 6$ for mantual containing complete data,

£23.2.0
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£28.7.0
£14.10.0
£24.13.4
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£25.15.2
£25.15.2 $-$

# Stern'S "fidelity" TAPE EQUIPMENT 

## THE FINEST RANGE OF TAPE EQUIPMENT FOR THE home constructor

## A SELECTION OF HIGH FIDELITY

## PORTABLE TAPE

 PRE-AMPLIFIERSAdds "HI-FI" Tape Recording to your existing Audio Installation.
IN ALL MODELS WE INCORPORATE THE

## TYPE "C" PRE-AMPLIFIER

and offer it complete in portable case with (a) The new "COLLARO" STUDIO Deposit: $\leq 7 / 6 /-12$ months $\leq 2 / 13 / 6$.
(b) The COLLARO Mk. IV "Transcriptor" 3 Speed Deck. Deposit: $: 88 / 6 /-12$ months $\varepsilon 3 / 0 / 11$
(c) The new TRUVOX Mk. VI Tape Deck. Deposit:
(d)/14/-. 12 months $E 3 / 3 / 10$
(d) The BRENELL Mk. $\vee 3$ Speed Deck.
(d) $\varepsilon 10 / 6 / 612$ manths $\varepsilon 3 / 15 / 7$
(e) The WEARITE MODEL AA Tape Deck. Deposit:

TAPE PRE-AMPLIFIER-ERASE UNIT INCORPORATING

E THE
NEW FERROXCUBE POT CORE PUSH-PULL OSCIL LATOR and 3 SPEED TREBLE EQUALISATION by means of the latest FERROXCUBE POT CORE INDUCTOR.
PRICES ... INCLUDING SEPARATE SMALL POWER SUPPLY UNIT COMPLETE KIT £14.0.0 ASSEMBLED AND £17.0.0 OFPARTS 14.0 .0 TESTED
Deposit $£ 3 / 8 /$ and 12 months of $£ 1 / 4 / 11$. Assembled unit only.
£11.15.0 and $£ 14.10 .0$ $\qquad$ (Carr. and Ins. 5/- extra) Send S.A.E. for leaflet or $2 / 6$ for Complete Assembly Manual. Send S.A.E. for leaflet or $2 / 6$ for complete Assembly Manual We present this "Hi-FF "" Pre-amplifier strictly to Mullard's specification ete., incorporating ONLY NEW HIGH GRADE COMPONENTS and che SPECIFIFD NEW MULLARD VALVES. It comprises a COMPLETELY SELFCONTAINED UNIT, all components and vatves being contained in a well ventilated Box-Chassis neatly finished in Hammered gold with a very attractively engraved PERSPEX FRONT PANEL.
For permanent high fidelity installations WE ALSO OFFER (excluding Case) the following
(a) The COLLARO "STUDIO TAPE DECK and our Mullard Type " C ". PE-AMPLIFIER and Power Unit Assembed and Pested
(b) As above but TYPE " C " PRE-AMPLIFIER supplied as complete Kit of Parrs.
(c) The COLLARO Mk. IV TAPE DECK and the MUILLARD Type "C" Pre-amplifier and Power Unit assembled, tested
£32.10.0
£29.0.0
H.P. Deposit $£ 7$ and iz months $£ 2 / 1 i 1 / 4$.
(d) As in (a) above but the Type "C" supplied as COMPLETE KIT OF PARTS
(e) The TRUVOX Mk. VI TAPE DECK and the assembled Trpe " C " Pre-amplifier and Power Unit
(f) As above but the Type " $\mathbf{C}$ " supplied as complete

KIT OF PARTS
(g) The, RRENELL MK. V Deck and the assembled Type "C" PRE-AMPLIFIER and POWER UNIT
(h) As above, but the Type " $C$ " supplied as complete KIT OF PARTS
(i) The WEARITE $4 A$ DECK with Type "C $C$ assembled

(Carriage and Insurance on above quotes $10 /$ - extra)
 you can bulld A COMPLETE HIGH QUALITY tape recorder for $£ 36.0 .0$ $\begin{array}{ll}\text { H.P. TERMS } & \text { Deposit } \\ \text { E7/4/., } 12 \text { months E2/12/10 }\end{array}$ E7/4/-, 12 months $\varepsilon 2 / 12 / 10$ FOR THIS WE SUPPLY:-
E KIT OF PARTS TO BUILD COMPLETE KIT OF PARTS
THE HF/TR 3 TAPE AMPLIFIER. THE HF/TR TAPE AMPLIFIER.
THE NEW COLLARO "STUDIO "TAPE DECK. PORTABLE CARRYING CASE (as illustrated). ROLA/CELESTION IOin. x 6 in . P.M. LOUDSPEAKER. ACOS CRYSTAL MICROPHONE 1200 fL . SPOOL E.M.I. TAPE. Alternatively for those who prefer another type of TAPE DECK we will supply precisely as above-but IN PLACE of the COLLARO "STUDIO " DECK-WE INCLUDE:-
(a) The MK. N COLLARO "TRANSCRIPTOR" DECK $£ 39.15 .0$ H.P. TERMS.. Deposit $£ 8,12$ monthly payments of e2/18/2 (E) extra if we are required to wire up the
Transeriptor Switch Banks).
(b) The new TRUVOX Mk. VI DECK.
£45.0.0
and Ins. on all above is $12 / 6$ extra).
For constructors with their own Cabinet-WE OFFER:-
COMPLETE KIT to build the HF/TR3 Amplifier,
£28.0.0
(b) As above but HF/TR3 ASSEMBLED and TESTED...
H.P. TERMS: Deposit $£ 6 / 6 / \mathrm{e}, 12$ months of $£ 2 / 6 / 2$.
(c) COMPLETE KIT to build the HF/TR3 together with
the MK. IV COLLARO "TRANSCRIPTOR" DECK
(EI extra if we are required to wire up Deck Banks)
e31.10.0
(d) As above but HF/TR3 ASSEMBLED and TESTED.
H.P. Terms: Deposit $£ 7,12$ months at $£ 2 / 10 / 5$.
£34.10.0
(e) COMPLETE KIT to build the HF/TR3 together with
the NEW TRUVOX Mk. VI TAPE DECK .......
(f) As above but HF/TR3 ASSEMBLED and TESTED, $\mathbf{H}$.
(g) COMPLETE KII to build the HF/TR3 AMPLIFIER with
the BRENELL Mk. VTAPE DECK
(h) As above but HF/TR3 ASSEMBLED and TESTED
H.P Terms: Deposit e9, 12 months of E3/6/-. TESTED HF/TR AMPLIFIER
with the WEARITE MODEL 4A DECK, incorporates
Wearite Head Lift Transformer, etc...
H.P. TERMS: Deposit $£ 11,12$ months of $£ 4 / 0 / 8$.
(Carriage and Insurance on each above is 10/- extra).
Attractive PORTABLE CASE is available to accommodate the TRUVOX or COLLARO TAPE DECKS and we offer it together with ROLA/CELESTION $10 \times 6 \mathrm{in}$. LOUDSPEAKER-ACOS CRYSTAL MICROPHONEand 1200 ft . SPOOL E.M.I. TAPE-ALL FOR
(Carriage and Insurance 5/- extra).
£9.0.0
WE HAVE THE NEW 2-SPEED TWIN TRACK
TRUVOX Mk. VI Tape Deck in stock 826.5 .0 Deposit $£ 5 / 5 /-$ It facorpmales PRECISION REV. COUNTER and PAUBE CONTROL and fully malntains the reneral high standard of all Truvox equipment. The very popular Colla io Tape Decks and the BRENELL Mk. V Decks are alao available.

## THE MODEL HF/TRJ TAPE AMPLIFIER

Incorporating by means of the latest FERROXCUBE POT CORE INDUCTOR PRICE for COMPLETE
KIT OF PARTS
E12/15/-
 AIRE PORCHASE: Deposit £3/6/6 and
 quailty amplifier hased on the very succeskull Mape LABOIgATORIEB. ONLY NEW HIGH-GRADE COMPONENTE are incorporited Including MULIARD VALVES and a GILSON OUTPUT TRANB FORMER...other features are: Magic Eye Recording Head Indleator-Effective Tone
Control-Monitoring and Extension Eneaker Bockets-has own Power Supply and caf Control-Monltoring and Extension Gneaker Bockets-hat own Power Supply and cat
be used as independent Amplifer for direct reproduction of Gram. Records or from Radic be used as independent A mplifier for direct reproduction of Gram. Records or from Radic
Tuner. Overall size $11 \times 6 \times 6 i n$. Truvox-Collaro-or Brenell-please specify which Tuner. Overall size $11 \times 6 \times$ in. C Truvox-Collar
gend 8 . A .E. for leaflet or $2 / 6$ for Assembly Manual.
$\overline{\text { PLEASE ENCLOSE S.A.E. WITH ALL CORRESPONDENCE }}$

STERN'S MULLARD DESIGNS
COMPLETE KIT OF PARTS Designed by MULLARD-presented by MULLARD " $5-10$ " MAIN AMPLIFIER For use with the MULLARD 2-stage pre-amplitier with which an undistorted power output of up to 10 watts IE obtajned. We supply 8PRCIFIEL COM. PONENTS AND NEW MULLARD VALVES including PARMEKO MAINS TRANSFORMER and cholce of the lateat Ulta-linear PARMEKO or the
PARTRIDGE Output Transformer. PARTRIDGE Output Transformer Price: COMPLETE KIT (Parmeko O/put Trans.). £11.10.0

## above incorporating partridge output trangFormer elyble extra. MULLARD'S

## PRE-AMPLIFIER TONE CONTROL UNIT

 Employing two EF86 valves and designed to operate with the Mullard MAIN AMPLIFIER, but also periectly suitable for other makes - Epualfastion for the latest R.I.A. characteriatics. incorporating Input for Crystal Pick-ups and varlable reluctance magnetic types.Input (a) Direct from High Imp. Tape Head. (b) From a Tape Amplifier or Pre-Amplifier - Sensltive Microphone Chanmel Wide range BAas and TRFBIL Price: COMPLETE KIT 86.6 .0 Alteraiatvely we supply

## COMPLETE MULLARD 5-10. AMPLIFIER

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A VERY HIGH QUALTTY AMPLIEIER DEVELOPED FROM THE VERY POPULAR 3-VALVE
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Dimenslons 15$\}$ wide, 12 in . high, 4 itin. deep.

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 Latest BSR Monardeck. Single speed $3 \frac{3}{4}$ i.p.s. Will take $5 \frac{3}{4}$ in, spools, $49 / 19 / 6$. P. \& P. Collaro Studio Tape Transcriptor. 3 speeds 17. 37, $7 \frac{1}{2}$ i.p.s. 3 motors. Push button controls, Will take7 in. spools. 15 gns. P. \& P 7 in.
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OUR PRICE 19 gns .
AVANTIC SP A2I Stereophonic Amplifier. Special feature of this Amplifier is high sensitivity permitting direct operation from magnetic Pick-up and high impedance tape replay Head, power output each channel 25 watts peak, rumble filter, L.S. impedance 4,8 and 16 ohms., 6 -position input selector, bass, treble, volume on/off controls. Dimensions $14 \frac{1}{2} \times 14 \times 4 \mathrm{in}$.
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AVANTIC DL7/35 Power Amplifier, designed to the highest possible standard ${ }^{5}$ to meet present day demands and when used in conjunction with the SP2I Pre-amp. Control Unit perfection in stereo reproduction is achieved. Specifications: power output 54 watts peak; L.S. impedance, 4,8 or 16 ohms, power inputs $105-250 \mathrm{v}$ Valve line-up GZ34, 2-EL34, ECC83, EF86. Dimensions $14 \frac{1}{2} \times 9 \times 8 \frac{1}{2} \mathrm{in}$.

AVANTIC SLI2-2I Speaker System employing 12 in . dia. P.M. L.S. and high frequency pressure Unit amounted in an acoustically designed enclosure, impedance 15 ohms, dimensions $38 \times 187 \times 15 \frac{7}{8} \mathrm{in}$. Finish matt medium walnut with front and sides covered with fawn fabric, standing on small contemporary legs. Original
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Garrard RCl2I Mk. 24 speed, wired for stereo and with plugrin Head ... elo 196

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| VOLTA | volTs | Current | Current |
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| 300 mV . | 15 v. | 30 mA . | 150 mA . |
| 1.5 v. | 75 v. | 150 mA . | 750 mA . |
| 3 v. | 150 v. | 300 mA . | 1.5 amp . |
| 15 v. | 300 \%. | 1.5 amp . | 7.5 amp. |
| 30 v . | 600 v . | 3 amp . | 15 amp . |
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| 750 \%. |  |  | 1,000 $\Omega$ |
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Type BMI. An all-dry battery eliminator. Size $5 \frac{1}{2} \times 2 i n$. approx. Completely replaces batteries supply 1.4 v . and 90 v . where A.C. mains $200-250 \mathrm{v}$. $50 \mathrm{o} / \mathrm{s}$. is avallable. Suitable for all battery portable receivers requiring 1.4 \%. diagram 39/9 or ready for use $46 / 9$
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wailve
$200-250$ A.C. pains L. and M. wave T.R.F.
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## ASSEMBLED

 CHARGERS6 v .1 a.
$0 . \mathrm{V}_{12} \mathrm{va}$
0.1 a
$6 / 12 \mathrm{v} .2$ a.
Above ready for -.. 56/9
Above ready for use with mains and output leads. Cases well ventilated and finished in stoved blue hammer. Carr. \& pkg. 3/6.

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200-230-250 v. $50 \mathrm{c} / \mathrm{s}$ | $0.9-15$ | v .1 l a. | $12 / 9$ |
| :--- | :--- | :--- | $0-9 \cdot 15$ v. $2 \frac{1}{2}$ a. ...... 15/9 $0-9-15$ v. 3 a. 0-9-15 v. 5 a. ...... 19/9

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6 v. or 12 v. 2 amps . Fitted Ammeter and selector plug for 6 v . or 12 v . Louvred metal case, finished attractive hammer blue. Ready for use with mains and output leads. Double Fused. Only $49 / 9$ As above, but for 3 amp. charging. Only 59/6. Carr. $3 / 8$

ASSEMBLED 6 v . or 12 v .

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Fitted Ammeter and variable charge selector. Also selector plug for $6 \mathbf{v}$. or 12 v . charging. Double fused. Well ventilated steel case with blue hammer finish. Ready for use with $69 / 9$ output leads. Carr. $5 /$-. Or Deposit $13 / 3$ and 5 monthly payments of $\mathbf{1 3 / 3}$.
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2.v. 16 A. $\mathrm{H}, \mathrm{EX}$, GOVT. ACCUMULATORS. New Boxed. Only $5 / 6$ each, 3 for $15 /$, plus $3 / 6$ carr

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All $200-250$ ๒. $50 \mathrm{c} / \mathrm{s}$. input.
Pr. $0-110 \cdot 200-230-250$ v., $275 \cdot 0-275$ v. 100 mA ., 6.3
7 A., 5 v. 3 ~.
250 v. $60 \mathrm{~mA} ., ~$
6.3 v. 2 a.
$300-0-300$ v. 60 mA .6 .3 v. 2
$265,0-265$ v. 150 mA ., 6.3 v. 11 a. .... $10 / 11$
$0 \cdot 24-26-28$ マ. 15 amps . A.C. conservative Govt. rating (rark ed with D.C. rating after rectifcation) 69/ $\theta$. Carr. $1 \overline{\mathrm{E}} /-$ $0-10-20-25$ v. 24 a . (Govt. rating) 79/6. Carr. 15/. Carr. $7 / 6$. $\quad 50$ watts, $0-110 / 120 / 230 / 250$ v.
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D.C. SUPPLI KITS. Sultable for electric trains. Consists of mains trans. 200-250 r. B0 c.p.A.; 12 v. 1 atap, selenium rwitch, variable speed regulator, partially drilled steel case and circuit. Very limited number, $33 / 9$.

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[^12]
## R．S．C．A10 ULTRA LINEAR 30 WATT AMPLIFIER


inputs such as＂mike＂and WITH PLUG IS DNGLUDED FOR SUPPLY OF 300 v． 20 mA ．and $6 \cdot 3$ ₹． 1.5 A．FOR A RADIO FEEDER UNIT．Price in kit form with ensy－to－tollow wring diagranis．
ONLY 1 Or Factory built with 12 months＇guarantee Elisil9／6．TERMS ONLY 1 AITS．ON ASSEMBLED UNTTS．DEPOSIT $31 / 8$ and 9 monthly pay－ Carr，10／－ments of 31／9．
Cover as Illustrated Type 807 output valres are naed With High Quality Sectionally
$18 / 9$ extra． $18 / 9$ extra． operation．Negative teedback of 20 D．B．In main loop．CERTIFIED PERFORMANCE FIGURES ARE EQUAL TO MOST EXPENSIVE UNITS AVAILABLE．Frequency response
$\pm 3 \mathrm{D} . \mathrm{B} .30-20,000 \mathrm{c} / \mathrm{cs}$ ．Tone Controls $\pm 12 \mathrm{D} . \mathrm{B}$. at 50 c／cs．$\pm 12 \mathrm{D} . \mathrm{B}$. to $-6 \mathrm{D} . \mathrm{B}$.
 Chassis finished blue hammer．Overall size $12 \times 9 \times 9$ in．approx．Power consumption 150 SUITABLE FOR THE CONNOISSEUR OR FOR LARGE HALLS CLUBS OR OUTSIDE FUNCTIONS，IDEAL FOR USE WITH MUSIOAL INSTRUMENTS SUCR AS STRING BASS． ELECTRONIC ORGAN，GUITAR，etc．FOR DANCE BANDS，GARRISON THEATRES， oto．，otc．We can supply Miorophones，Spenkers，日tc．，at keen assh prices or on terms with
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LINEAR L45 MINIATURE $4 / 5 \mathrm{~W}$ ．QUALITY AMPLIFIER．Sultable for use with any record playing unit and most microphones．Negative feedback 12 D．B．Bass and Treble controls，For A．C．mains input oi $200-250$ 区． $60 \mathrm{c} . \mathrm{p.s.s}$ ．Out put for $2 / 3$ ohm speaker．Three Quaranteed 12 months Only $\mathrm{C5} / 19 / 6^{\text {Or Deposit } 22 /- \text { and } 5 \text { roonthly paymente．}}$
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$250-0-250 \mathrm{v} .100 \mathrm{~mA}, ~$
20.3 v． 4 a．， 5 v .8 a. $300-0-300$ v． $100 \mathrm{~mA}, 6.3$ v． 4 a．，$\delta$ v． 3 a ． $350-0-350$ v． $100 \mathrm{~mA}, 6.3$ v． 4 a．， 8 v． 3 a．
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Standard Pentode $5,000 \Omega$
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Push－pall $16-18$ watts，sectionally wound，ois．
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Push－pull 20 watt high－quallty sectlonally wound， 6L6，KT66，etc．，to 3 or $15 \Omega$ ．
SMOOTHING CROKEs SMOK COKKEs $250 \mathrm{~mA}, 5 \mathrm{H}, \mathrm{s} 100 \Omega \mathrm{~S} 11 / 9$ 150 mA ． $7-10 \mathrm{H}, 250 \Omega 11 / 9$ $100 \mathrm{~mA}, 10 \mathrm{H} ., 200 \Omega 8 / 8$ PARMEKO MAINS TRANS 450－0－450 ₹． $120 \mathrm{~m} A ., 6.3$ マ 500－0－500 จ 120 mA .6 .3 マ． 4 а．， 5 マ． 3

30 mA .10 H．， $350 \Omega$
$60 \mathrm{mA},$.10 H．， 400 D 4／11 60 mA．， $10 \mathrm{H}, 400 \Omega 4 / 11$
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Two hnput sociseta with assocl－ of controls sallow mizing A．10．Hiph nensitivity．Includes
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JUNIOR 5 WATT．High Quality Output，Separate Buss and TreDle＂cut＂and＂bocist＂controls Sensitivity 15 mv ．Eigh Flux 8in．1／speaker．Input ockets or Badio／Tape or Gram Pick－up and Mike cabinet（uize approx． $14 \times 14 \times 7 \mathrm{in}$ ．）．Finisbed in atin walnut and fitted carrying haudle． £8／19／6 Carr．7／6．or Deposit $£ 1$ and 9 Send S．A．E．for leatlet．

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time．Two Loudspeakers are incorporated，a 12 hn ． time．Two houdspeakers are incorporated，a 12 hn ．
P．M．for Bass notes，and $17 \times 4$ in，elliptlcal for P．M．for Bass notes，and $17 \times 4 \mathrm{ia}$ elliptleal for walnut．Size approz． $18 \times 18 \times 8 \mathrm{in}$ ． 15 Gns，Plus 10／－carr．TERMS．DEPOSIT． $34 / 9$. and 9 monthly payments $34 / 9$ ．Both models for $200-250$ v．A．C．mains．

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LOUDSPEAKERS IN FOLISHED WALNUT FINISHED CABINET．Gauss 12,000 unes Speech coll 3 ohms or 15 obms．Only g 9／19／6．
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Rin． 20 WATT 15,000 line $1 /$ speakers 15 ohms，in Cabinet fintshed as above，Size $18 \times$ $8 x$ gin． $87 / 19 / 6$ or Deposit $17 / 9$ and 9 monthly paymenta of $17 / 9$ ．

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（ 15 obms），consisting of a high quallity 12 in．speaker of orthodox design support－ lag a gmall elliptical speak－ er ready wifed with choke and condensers to act as tweeter．This high fidellty uult ls bighly recommended for use with our All or any 10 watts．Gauss 12000 llnes．Price only E5／17／6
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For Cathodu Kay Tubes having Feater/Cathode short circuit and for C.R.
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RESISTORS. AH preferied ralues. $20 \% 10$ ohms to 10
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$\left.\begin{array}{r}5 \text { watt } \\ 10 \text { watt }\end{array}\right\}$ WIRE-WOUND RESISTORS
15 watt $\}$ 25 ohms- 10,000 ohms.
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WIFE-WOUND POTS, 3 w:
$\underset{\text { Kre-set Min. T.V. type }}{\text { Min }}$ Knarled Blotted knob,
All values 25 ohtns to 25 K ., 3/. ea, $30 \mathrm{~K} .50 \mathrm{~K}, 4 /=$ $\begin{array}{ll}3 /- \text { eas, } & 30 \mathrm{~K} ., 50 \mathrm{~K}, 4 /- \\ \text { Ditto, }\end{array}$ 30 K. To 2 Meg., $3 /$. ratio push-pull, $/ 6 / 6$. Miniature duty 50 mA ., $4 / 6$. Multi Push pull 10 watts, 15/6. MULLARD " 510 " $4 / 6$. Hygrade L.F. GHOKES 15/10H $60 / 65 \mathrm{~mA}$, $5 /-$. $10 \mathrm{H} 85 \mathrm{~mA}, 10 / 6$ $10 \mathrm{H} 150 \mathrm{~mA} ., 14 /$

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 2 p. 6-way, or 4 p. 2 -way, or 4 p. 3 -way, long spindle Wave change "MAKITS, 1 wafer $8 / 6 ; 2$ wafer, $12 / 6$ 3 wafer 16/e; 4 wafer $19 / 6 ; 5$ wafer $23 /-; 6$ wafer $26 / 6$,
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Medium and long wave. Powerful output from 6in. high Flux Speaker. T.C.C. Printed circuit and condensers. All components of finest quality clearly identified for assembly with full instructions. Osmor Ferrite Aerial and Coils, Rexine covered attache case type cabinet. Size $12 \mathrm{in} . \times 8 \mathrm{in} . \times 4 \mathrm{in}$. Batteries used BI26 (L5512) and AD35 (L5040), 10/extra. Details and instructions $1 / 9$ (free with kit). Mains Unit ready made for above $39 / 6$. Same size as batteries, sold separately.

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three wavebands
S.W. $16 \mathrm{~m},-50 \mathrm{~m}$.

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diode.

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Long spindle. Guaranteed $\frac{1}{5}$ year. All values. 5 K . ohms up to 2 Meg .
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To build yourself Medium and Long Waves-Push-Pull Superhet A.V.C. Perfect Car Radio reception. Slze $10 \mathrm{in} . \times 6$ 行 $\times \mathbb{4} \mathrm{in}$, at base tapering to 4in. at top.
Very attractive two-tone grey Vynide covered cabinet witb black and gold printed escutcheon plate, cream and gold knobs, handle and cabinet fittings. $\star$ Weight--complete with long-lite 7 t voit battery- $4 \frac{\mathrm{l}}{\mathrm{lb}}$. $\star$ Mazda High-grade transistore throughout. $\star$ High-Fiux 7in. $\times 4 \mathrm{in}$. Elliptleal speaker. $\star$ Slow motion tuning. $\star$ Co-axial socket at rear for direct connection to Car Radio Aerial. $\star$ Improved reception by use of seven. section plated telescopic aerial disappearing lato Cabinet when closed, 34in. above Cabinet when fully extended.
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A smal! three-valve PORTABLE RE CORD-PLAYER AMPLIFIER mounted on baffle $12 \times$ Tin., with High Flux $6 \frac{1}{2} \mathrm{in}$. Loudspeaker. Valye line-up ECC83, EL84, EZ80. Incorporates separate bass and treble controls. Max. output 3 watts. Will match alf types of hlgh impedance ick-up. Ready to use, $25 / 12 / 6$. P. \& P. 3/6 NEW STYLEE CABINET finished in wo-tone Leatherette. Will accommodate above Amplifier and Baffle without modification, also most types f Ancillary Equipment. Overall size $18 \times 13 \frac{1}{2} \times 8 \frac{1}{\frac{1}{2}} \mathrm{in}$. Fitted with carrying handle, $\varepsilon 3 / 9 / 6$ plus $5 /-\mathrm{P}$. \& P . NOTE. If both items purchased together they will be supplied at a special inclusive price $£ 8 / 7 / 6$ plus $6 / 6$ P. \& P.
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 3 kV . wkg. $4 /-$ ench. $0.025 \mu \mathrm{~F}, 2.5 \mathrm{kV}$. Whg., 416 each. $0.0025 \mu \mathrm{~F} ., 4.6 \mathrm{kV}$. wkg.
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$100 \cdot 2$ so volt A.C. input, 24 v . at 3 amps or 12 v., twice at 3 amp . each winding. Continuous tropical ratling awitched and fused, ete., in metal case that fits 19 in . rack, size $19 \times 7 \times 7 \mathrm{in}$. Brand nem £8/15/\%, carr. 7/6 (with eltcuit).

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## HOME RADIO <br> 79/6



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$2 / 9$ each. 6F1, $6 \mathrm{~F} 12,6 \mathrm{~F} 13,6 \mathrm{~F} 14,6 \mathrm{~F} 15,6 \mathrm{~K} 7,6 \mathrm{LD} 20,10 \mathrm{~F} 1, \mathrm{PEN} 45$, PEN46, U22, UF41. $7 / 9$ each. $5 U 4,5 Y 3,6 A 8,6 \mathrm{~K} 8,6 \mathrm{Q} 7$, 6U6, EABC80, EBC33, ECC81, EL38, EZ40, KT36, EZ80.
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50 Microamps 100 Microamps 100 Microamps 500 Microamps 500 Microamps

1 Milliamp
1 Milliamp 30 Milliamps 100 Milliamps 200 Milliamps 500 Milliamps 5 Amperes
15 Amperes
25 Amperes D.C.
50 Amperes 30-0-30 Amp 50-0-50 Amp 10 Volts $5: 0$ Volts 300 Volts

| Size | Type |
| :---: | :---: |
| 2 2in. | MC/FR |
| 2 l in. | MC/FR |
| 34 in . | MC/FR |
| 2 in . | MC/FR |
| 2 in . | MC/FR |
| 2 in . | MC/FS |
| 2 2in. | MC/FR |
| 2 tin . | $\mathrm{MC} / \mathrm{FR}$ |
| $2 \frac{1}{2} \mathrm{in}$. | MC/FR |
| 2lin. | MC/FR |
| $3+\mathrm{in}$. | MI/FR |
| 2 in . | MC/FS |
| 2 in . | MC/FR |
| 21in. | MI/FR |
| 4 in . | MI/F/PR |
| 2 in . | MC/FR |
| 2 in . | MC/FS |
| 2 in . | MCR/FS |
| 34.10 | MC/FS |




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TRANSFORMERS，potted，＂C＂core；Input 230 V ．Outputs：－ 6.3 v ．， 0.3 A．（1 A．actual），twice，8／6；Outputs：$-510-0-510$ v． 275 mA $375-0-375$ v． $83 \mathrm{~mA}, 5$ v． 3 A．， 6.3 v． 7 times（ 17 A.$), 45 /-\mathrm{CON}$ DENSERS，block，paper； 8 mid． $250 \mathrm{Vw} .4 /-; 600 \mathrm{Vw} .6 /-; 4$ mfd． 2 kVw $7 / 6 ; 600 \mathrm{Vw} .3 / 6$ ．Switch fuse splitter，DP． 15 A ． $15 /$－．Panel fuse holders， $1 / 3$ ；Panel Lampholders（indicators）， $1 / 6$ ．POWER UNITS Input A．C． $115 / 250 \mathrm{v}$ ．Outputs：D．C． $330 \mathrm{v}, 120 \mathrm{~mA}$ ．and 6.3 V ．A．C． twice．Potted trans．and LF choke，new（post 3／6），30／－，MONITOR 56 ， triggered oscilloscope，comprising Indicator 248 and Power Unit 675. Valves VCR138a，3／EF50，2／ECC33，5／EF55，EF37A，6V6，3／EA50 and $2 / 5 \mathrm{U} 4 \mathrm{G}, \mathrm{VU} 120 \mathrm{~A}$ ．Two units each $12 \times 9 \times 18 \mathrm{in}$ ，black finish． 230 v．A．C．input，with 18 way cable and mains cable and circuit Cathode probe unit extra，17／6，£8／10／－（Rail 20／－）＇INDICATORS， Type 101 with VCR530 and 2／EB91，2／EF91，2／R10，new cond．30／－ （post 7／－）．Type 1 with VCRX263，2／EF52，5／6J6，1／6V6，1／EY51， 2／EB91，3／EF91，RF EHT Generator and $28 \mathrm{kc} / \mathrm{s}$ ．xtal，45／－（Rail 7／6） HEADPHONES，CLR，7／6．CR100 Noise Limiter assemblies，with valve， $3 / 6$ ．NEW M．C．METERS， $3 \frac{1}{2} \mathrm{in}$ ．round fiush， $50 \mu \mathrm{~A}, 70 /-; 200 \mu \mathrm{~A}$ ， centre zero， $50 /-$ ； 1 mA ，centre zero， $45 /-; 1 \mathrm{~mA}_{1} 55 /-2 \frac{1}{2} \mathrm{~m} .1 \mathrm{~mA}$ ， $2 \frac{1}{2}$ in．， $15 /$－，VIBRATORS，Mallory G634C 12 v． 4 －pin， $7 / 6 ; 6$ v． 5 －pin reversible，7／6．R1155B，good condition，tested with handbook． s7／10／－（Rail 10／－）．DRIVES：slow－motion Admiralty 200：1 ratio， scaled $0-100,5 / 6$ ．R1155 S．M．＂N＂type，new，10／6．VIBRAPAK，
 $21 /-(p . p, 3 / 6)$ DYNAMOTORS（post 3／6）． 12 v. to $250 \mathrm{v} .65 \mathrm{~mA}, 11 / 6 ;$ 6 v ．to $250 \mathrm{v} .60 \mathrm{~mA}, 11 / 6$ ．ROTARY GONVERTERS．Imput $24 \mathrm{v} . \mathrm{D} . \mathrm{C}$ ． Output 50 v．， 43 a．， $50 \mathrm{c} / \mathrm{s}$ ．， $40 / \mathrm{m}$（Rail 7／6）．CATHODE RAY TUBES． New：VCR 139A or VCR 138，each 30／－；Potentiometers，miniature wirewound，$B \Omega, 100 \Omega, 600 \Omega, 1 \mathrm{k}$ and 2 k ，each， $1 /=$ ．CHOKES，LF 10 H $200 \mathrm{~mA}, 8 / 6 ; 100 \mathrm{H} 60 \mathrm{~mA}$ ． $8 / 6,9 \mathrm{H} 100 \mathrm{~mA}, 5 / 6 ;$ Potted 10 H 100 mA ， $7 / 6$ ；＂${ }^{\text {C }}$＂${ }^{\text {＂}} 5 \mathrm{H} 400 \mathrm{~mA}, 10 / 6,10 \mathrm{H} 250 \mathrm{~mA}, 12 / 6 ; 10 \mathrm{H} 50 \mathrm{~mA}, 6 /-$ ．R．F．27， good cond．，18／－（p．p．3／6）．METAL．RECTIFIERS， $240 \mathrm{~N}, 100 \mathrm{mA}$. ． $4 /-$ ； 240 v． $30 \mathrm{~mA} .3 / 6 ; 600$ v． $30 \mathrm{~mA} .5 / 6 ; 240$ v． $80 \mathrm{~mA} .5 / 6 ; 1,000 \mathrm{v} .30 \mathrm{~mA}$ ， 7／6， 350 v．2：a， $20 /$ ．Mic．Inserts，G．P．O．carbin，2／6．CONTROLS camera Type 35；a timing device，new 10／6．（post 3／6），COMMAND Receivers，medium－wave（ $520-1,500 \mathrm{kc} / \mathrm{s}$ ）， 6 valves；new， $97 / 6$ ；used 82／6．Converslon data for above to CAR RADIO， 12 v. ，with circuit， $1 / 6$. RELAYS，potted，small， $1,700+1,700 \Omega$ ，hi－speed SP c／o，10／6．2，500 2 make， $6 /$－．TEST SET $263,9280-9480 \mathrm{mc} / \mathrm{s}$ ．with waveguides， $50 \mu \mathrm{~A}$ and 1.5 mA ，meters，£5／10／－（Carr．10／－）．
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| 6 | 63 | 127 | 303 | 522 | 639 | 709 | 977 | 1052 | 1135 | 1196 | 1289 | $143{ }^{\circ}$ | 1765 | 1945 | 2742 |
| 8 | 64 | 133 | 312 | 525 | 640 | 716 | 980 | 1033 | 1136 | 1207 | 1293 | 1442 | 1788 | 1951 | 2769 |
| 12 | 66 | 138 | 315 | 540 | 641 | 720 | 982 | 1054 | 1137 | 1216 | 1295 | 1443 | 1794 | 1963 | 2810 |
| 13 | 67 | 140 | 338 | 543 | 644 | 721 | 984 | 1062 | 1141 | 1221 | 1307 | 1444 | 1805 | 1988 | 2840 |
| 15 | 72 | 173 | 360 | 555 | 646 | 728 | 991 | 1069 | 1144 | 1223 | 1322 | 1453 | 1812 | 1990 | 2874 |
| 18 | 82 | 185 | 368 | 571 | 647 | 723 | 992 | 1072 | 1145 | 1234 | 1324 | 1458 | 1835 | 1991 | 2967 |
| 21 | 85 | 188 | 392 | 577 | 648 | 727 | 993 | 1076 | 1152 | 1236 | 1325 | 1480 | 1841 | 1993 | 3579 |
| 22 | 86 | 199 | 395 | 581 | 649 | 797 | 1001 | 1082 | 1155 | 1240 | 1327 | 1488 | 1850 | 2101 | 3684 |
| 25 | 87 | 202 | 405 | 594 | 656 | 801 | 1019 | 1088 | 1160 | 1241 | 1331 | 1491 | 1832 | 2215 | 3595 |
| 34 | 90 | 208 | 416 | 604 | 659 | 802 | 1020 | 1090 | 1163 | 1249 | 1335 | 1501 | 1853 | 2216 | 3601 |
| 35 | 94 | 209 | 424 | 608 | 663 | 819 | 1022 | 1091 | 1166 | 1250 | 1355 | 1510 | 1837 | 2312 | 3621 |
| 43 | 100 | 216 | 428 | 610 | 664 | 848 | 1025 | 1103 | 1167 | 1260 | 1401 | 1569 | 1859 | 2547 | 3693 |
| 45 | 101 | 222 | 454 | 611 | 667 | 849 | 1028 | 1106 | 1171 | 1267 | 1402 | 1574 | 1878 | 2568 | 3747 |
| 50 | 102 | 223 | 465 | 612 | 683 | 897 | 1035 | 1107 | 1172 | 1272 | 1403 | 1577 | 1885 | 2663 | 3797 |
| 52 | 103 | 233 | 487 | 615 | 690 | 916 | 1037 | 1113 | 1174 | 1274 | 1408 | 1582 | 1896 | 2666 | 3799 |
| 54 | 112 | 248 | 490 | 616 | 697 | 922 | 1043 | 1118 | 1177 | 1278 | 1409 | 1582 | 1896 | 2666 | 3799 |
| 56 B | 114 | 265 | 493 | 621 | 700 | 988 | 1044 | 1119 | 1180 | 1283 | 1429 | 1587 | 1931 | 2701 | 3830 |
| 57 | 121 | 278 | 509 | 623 | 702 | 970 | 1045 | 1127 | 1183 | 1286 | 1432 | 1756 | 1935 | 2706 | 3886 |

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Tin. reels of $1,200 \mathrm{ft}$. P.V.C. base tape, $21 /-$, plus $1 / 6$ post and pkg.
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 $\mathrm{MPC}, 1,500 \mathrm{pF}, \mathrm{m}, 000 \mathrm{pF}$ ． 8 g ．each．Wkg．， $470 \mathrm{pF}, 10 \%, 820$
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| 4 ．．．．．．．7／－ | 12AH7GT 6／6 | EF41 ．．．8／3 |
| U4G ．．．51－ | 12K7 ．．．5／3 | EF42 ．．．7／3 |
| 6AG5 ．．．．．．4／－ | 12 Q 7 ．．．5／3 | EF86 ．．．10／－ |
| $6 A 15$ ．．．3／9 | 25Z4G ．．．7／6 | EF91 ．．．3／6 |
| 6AM6 ．．．3／6 | 35W4 ．．．6／6 | EL41 ．．．8／3 |
| 6AT6 ．．．6／11 | $35 Z 4 \ldots 5 / 3$ | EL42 ．．．9／3 |
| 6B8G ．．．2／11 | 8078 ．．．3／6 | EL84 ．．．6／6 |
| 6BA6 ．．．5／11 | 954 ．．．1／6 | EY51 ．．．8／6 |
| 6BE6 ．．．5／11 | 955 ．．．．．．3／6 | EY86 ．．．7／II |
| 6BG6G 12／3 | 956 ．．．．．．2／6 | EZ80 6 |
| 6BJ6 ．．．5／6 | 9001 ．．．．．．3／11 | GTIC ．．．6／11 |
| 6BW6 ．．．8／6 | 9004 ．．．3／11 | P61 ．．．．．． 21 |
| $6 \mathrm{C}_{4} \ldots 3 / 3$ | 9006 ．．．3／11 | PCC84 ．．．7／3 |
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| $6 \mathrm{CH6}$ ．．．9／－ | DAF96 7／9 | PCL82 ．．．11／3 |
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| 46 ．．．J／II | DL96 ．．． 719 | PL83 ．．．7／3 |
| 6J5G ．．．2／6 | DM70 ．．．7／6 | PY80 ．．．6／ll |
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I4in. $\times 6$ in., plus
handle $6 \mathrm{in.}, \mathrm{27/6}$.
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SOLENOID OPERATED MAG NETIC RELAY.
Type 5CW/3945, 4 pole ehangeover. 10 A contacts $24 v$. new 13/6. P \& $P$ ।

NEW CARPENTER'S TYPE POLARISED RELAYS. $2 \times$ 9,500 turns at 1,685 ohms. Price $22 / 6$ each. P. \& P. $1 /$-.

Carpenter's, similar to above but type 5A48. Coils $1 \times 3200$ turns at 100 ohms and $1 \times 2000$ turns at 145 ohms, 22/6 each. P. \& P. 1/-. Bases for same 2/6


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8-day clockwork Time Switch. Contacts $2 \frac{1}{2}$ amp. 230 vole, 24 hour phase, $t$ hour divisions, allow setting for one make and one break to be made every 24 hours, complete with key. Used but guaranteed perfect. Price 27/6 each. P. \& P. 1/6.


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$2 \frac{1}{4}$ amp. D.C. M.1. 2 in . fl. rnd,

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116
$7 \frac{1}{2} 2 \mathrm{mp}$. D.C. M.I. $3 \frac{1}{2} \mathrm{in}$. proj. and
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Microamp
50 microamp., scaled $0-100$, M.C.
$2 \frac{1}{2}$ in. fl. rnd.
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Miniature latest type mov. ing coil $0-5$ milliamp meter, $1 \frac{1}{2} \mathrm{in}$. diameter, flush fitting, complete with fixing clip. Price $17 / 6$. P. \& P. $1 /=$


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 crystal controlled 4 -valve high-grade instrument in the same category as the famous B.C. 221. Directly calibrated, does not require cross reference or charts - functions as follows: (I) A crystal controlled oscillator which provides fixed frequency signals of 500 KC and all harmonics of 500 KC to beyond 10 Meg. and up to 30 Meg .
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 RECTIFIER mounted on 200/250 volt A.C. input transformer. Output $36 / 40$ vole D.C. at 1.2 amps. New, perfect. Price 16/6. P. \& P. 3/6.

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ASSISTANT AUTHORS to assist in preparation of engineering manuals and Test Specifications for Factory and Field Testing. Candidates should have electronic background; previous experience as technical author is desirable though not essential. These appointments might suit ex Radar N.C.O.s who have Installation experience of complex equipment.
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We have vacancies for men with electronic experience for testing. Radar and radio technicians with fault finding experience would be suitable.
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Write to:
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Men aged 19 or over or interesting work providing and maintaining aeronautical telecommunications and electronic navigational aids at aerodromes and radio stations in the U.K. Fundamental knowledge of radio or radar with some practical experience essenradar with some practical experience essen-
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Southampton County Borough Education Committee
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## Instrument Mechanics (Physical and Electronic) and Instrument Electricians

We have vacancies for men experienced in fault diagnosis, repair and calibration of a wide range of instruments used in nuclear reactors, radiation laboratories and chemical plant operation. The work is interesting and involves working with instruments using pulse techniques, wide band low noise amplifiers, pulse amplitude analysers, counting circuits, television, and industrlal instruments for the measurement of flow, pressure and temperature.

Men with appropriate experience in H.M. Forces or with industrial experience of radar, television, radio or industrial instrumentation are invited to write for further information.
The rate of pay is $£ 12.12 .0 \mathrm{~d}$, for a 44 -hour, 5 -day week. Housing will be available for married men. Promotion prospects are good and there is a superannuation scheme.

Application forms and further information can be obtained from?
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Appointments with a progressive organisation operating in a rapidly expanding field are available for Equipment Designers in the following categories:

1. Engineering of Standard Products, including Mechanical Design, circuit
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Positions are available at Junior andSenior levels and your application should state the preferred category together with details of education and previous experience.
All applications will be treated in strict confidence and should be addressed for the attention of the Chief Designer.
LANCASHIRE DYNAMO ELECTRONIC PRODUCTS LIMITED, Rugeley, Staffs.

## AEI

## SENIOR ENGINEER

required to work in Applications section of Valve Development Laboratory. Minimum qualification H.N.C. Previous experience in testing and designing test gear for medium and high power radio transmitting valves, and experience with R.F. heating.

## Apply to Personnel Deportment Associated Electrical Industries Ltd.

Cosmos Works, Brimsdown, Enfield, Middx.

## TEST ENGINEERS

Required for varied and interesting work (partly experimental), on high quality communications equipment. Good experience in the use of measuring instruments is essential. Starting salary instruments is essential. Starting salary qualifications and experience.
The company operates a Pension Scheme and Sports/Social Club.
Holiday commitments will be honoured. Write, in confidence, to:

The Manager
Central Engineering Depariment,
British Relay Wireless Limited
1-7, Croft Street, Bermondsey, London, S.E.8.

VACANGIES IN GOVERNMENT SERVICE
A number of male vacancies offering good career prospects, exist for:-

RADIO OPERATORS
Write giving detals of Education, Qualifications and Experience to:-
Personnel Officer (3/R),
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53, Clarence Street,
Cheltenham, Glos.

## EMI

Interesting vacancies exist at the Feltham Laboratories of E.M.I. Electronics Ltd., for the following:

## TECHNICAL SPECIFICATION WRITER

There is a vacancy at our Feltham Laboratories for a Technical Specification Writer. Candidates must have a background of electronics and be able to write clearly and concisely. The post involves the preparation of technical reports for publications and entails close liaison with engineering teams. Initial salary will be determined by qualifications and experience and it is Company practice to review salaries annually on the basis of ability and potential,

Ref. S/10/1

## FIELD ENGINEERS

Engineers are required by the Field Services Division of the Company to engage in trials in the field of complex prototype electronic equipment developed by E.M.I. Electronics Ltd. Sound practical knowledge of the operations and maintenance of radar or communication equipment is necessary. The posts may involve periods away from base and a willingness to live away from home is essential. Starting salaries are based on qualifications and experience and it is Company practice to review salaries annually on the basis of ability and potential.

Ref. P/8/22

## INTERMEDIATE ENGINEER

An Intermediate Engineer is required to take engineering responsibility for the development programme of a telemetry sender. The post involves circuit design, engineering to extreme environmental conditions and technical liaison with customers. A degree with considerable relevant experience is essential. Preferred age 26-31.

Ref. P/6/15

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A Technical Assistant is required to assist in the development, construction and testing of a wide range of electronic equipment. Some experience of the use of Electronic Test Gear and bench tools and an ability to understand circuit diagrams is essential.

Ref. P/4/235
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Personnel Manager<br>E.M.I. ELECTRONICS LTD.<br>HAYES, MIDDLESEX

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## AEI

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We need new test and callibration engineers to help us therease the output of our very wide range of telecommunications meaburing instrumiente.
The work requires the understandlog of the raost modern and varied circult tochniques and embraces all frequencies up to $10.000 \mathrm{Me} / \mathrm{s}$.

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Call and talk th over if you live close to us at st. Abbans. Alternatively, write to Depi. C.P.E. Marconl House, $\mathrm{E}^{3317}$ 8trand, London, W.C.2, giving fuill details of your education and experience asd quouing reference WW 297011.

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## ARE YOU A YOUNG AND PROGRESSIVE ELECTRONICS ENGINEER

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## O.N.C. Electrical

Experience in the use of digital techniques for counting and data processing.
Have some years of industrial experience, both development and production.
There is a unique position for such an engineer in a new and expanding electronics division which a progressive Company has brought into being to handle a large number of problems involved in the handling of paper and associated documents. This Company is the leading manufacturer of paper handling equipment and can offer a young engineer a bright and secure future.
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SENIOR age 28-40 to lead T.V. Development Projects. B.Sc. or A.M.I.E.E. and practical experience in current receiver techniques in a senior position. Knowledge of transistor applications an advantage.
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Applicants to assist in aerial development work-a minimum of 3 years experience of R.F. measurement techniques preferably associated with aerial systems-suitably academically qualified.
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To undertake the initial systems design of new projects and to co-ordinate the design work through to the production stage.

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To be responsible for sections concerned with the design of major items of electronic equipment.

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To be responsible for the design of electronic equipment, especially in the following fields:-

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3.2 Microwave components and receiver/transmitters
3.3 Data processing

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Experience in the design of electronic equipment for use by the Services is essential.
Engineers with the necessary experience, and qualifications are invited to write to:

The Personnel Manager, Cossor Radar \& Electronics Limited, Elizabeth Way, Harlow, Essex.

Housing arrangements can be made for selected applicants.

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required in the Rig Design Section to be located at Guildford Power Station.
A knowledge of electronics circuitry is of advantage, but it is essential that the engineer appointed be lully competent to carry out the engineering design of electronic equipment to given circuit designs and provide detailed information to draughtsmen preparing manufacturing drawings. The equipment to be designed includes that required for electronic instrumentation of experimental apparatus of many kinds, but the principal responsibility will be for equipment to be developed for automatic control applied to power stations and the grid distribution system. This equipment must be designed to the high standards appropriate to power station practice.
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The starting salary will be within the range $£ 1,195-£ 1,775$ p.a., or $£ 1,090-$
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Applications stating age, qualifications, experience, present position and salary, to the Personnel Officer, 24-30, Holborn, London, E.C.1, as soon as possible. Envelopes should be marked "Confidential Ref. WW/196."

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The F．B．C．is inaugurating this service in November 1960 in conjunction with Rhodesia Television（Privoze）Lid．who will be responsible for studio engineering and programmes production side．
Applicants should have technical qualifications of a high standard and appropriate recent and extensive practical television engineering experience．

The stcressful applicant will be stationed at Salisbury，Southern Rhodesia，but will be requirad 10 iravel from time to time in connection with television installation in other centres．

There is a contributory pension scheme，medical aid and leave is at rate of three days for each month of service．Free air passage provided for the successful applicant，and assisted passages for his family．The scale of the post will be $£ 2,000 \times £ 60$ rising to £2，300 per annum，and entry point，which may be the maximum，will be according to qualifications and experience．

Write to the Crown Agents， 4 Millbank，London，S．W．1．State age，name in block letters， qualifications and experience and quote M2A／50827／WF．

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[^0]:    * Report of the Television Advisory Committee 1960, H.M. Stationery Office. Price 1 s .
    $t$ "Television Field Trials of 405 -line and 625-line Systems in the U.H.F. and V.H.F. Bands 1957-1958." Published by the B.B.C. Price 20 .

[^1]:    * See Appendix I for list of present members.
    + Report of the Television Advisory Committee, 1960, Stationery Office. 1s.

[^2]:    * See for example " Some Problems Concerning the Experimental WIre Television Service as Realised in the Netherlands, by A. P. Bolle, Het PTT Bedrijf, Vol. IX, No. 3, November 1959.

[^3]:    *See " Letters to the Editor," p. 294, Wireless World, June 1960.

    + This is an abridged version of a paper published in Mullard Technical Communications No. Technical Communicat
    12 (May 1955), p. 42.

[^4]:    *P. R. Macmillan Ltd. 15/-. pp. 238. 14 illustrations.

[^5]:    * Decca Record Co. Ltd.

[^6]:    * A new example is coming up for attention next month.

[^7]:    *Decca Navigator Co. Ltd.

[^8]:    * This has been the subject of theoretical work, notably by Sommerfeld, Bremmer and Norton, which is described in detail in a paper by A. B. Schneider ${ }^{3}$. Extensive work has also been carried out by Dr, B. G. Pressey and his associates at the Radio Research Station (D.S.I.R.).

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