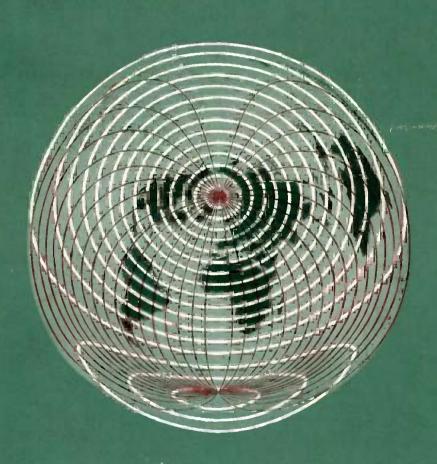
MARCH 1960

TWO SHILLINGS

Wireless World

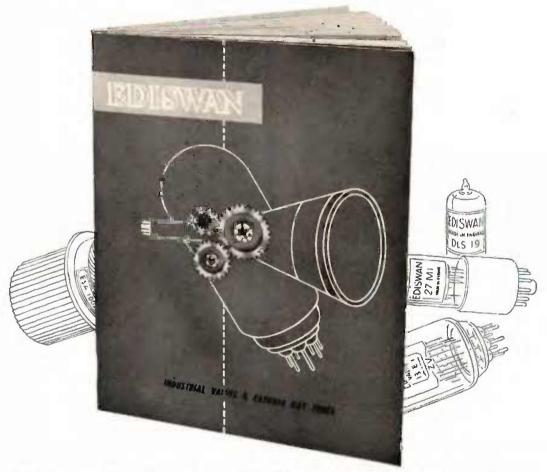
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MARCH 1960

Managing Editor:	103	Editorial Comment	
HUGH S. POCOCK, M.J.E.E.	104	Improving the Dynamic Range of	Tape Recording By L. H. Bedford
Editor;	107	Geneva Conference	
F. L. DEVEREUX, B.Sc.	108	How Long Will a Transistor Live?	By R. Brewer
	110	Transformerless Circuits for Broad By R. C. V. Macario	cast Receivers and N. E. Broadberry
Assistant Editors:	113	Manufacturers' Products	
H. W. BARNARD	114	World of Wireless	
T. E. IVALL	116	Personalities	40
	118	118 Physical Society's Exhibition	
	129	Signal-Flow Diagrams—2	By Thomas Roddam
	132	Short-Wave Conditions	
	133	Letters to the Editor	
	135	Technical Notebook	
	136	Elements of Electronic Circuits—11	By J. M. Peters
	137	The Reflectometer	By " Cathode Ray"
VOLUME 66 NO. 3	141	The Smith Chart—3	By R. A. Hickson
PRICE: TWO SHILLINGS	147	News from the Industry	
	148	March Meetings	
FORTY-NINTH YEAR	150	Random Radiations	By " Diallist "
OF PUBLICATION	152	Unbiased	By " Free Grid"

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FRAMEGRID

VALVES FOR TELEVISION

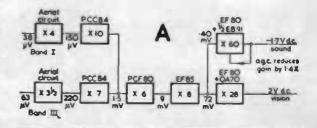
With a suitable valve line-up, receivers with two i.f. stages can be designed for various degrees of improved performance. Frame grid valves can be used throughout the tuner and i.f. stages, or some conventional valves can be retained, depending on the sensitivity required. Two classes of receiver satisfy all U.K. requirements: one for good signal strength areas, and a more sensitive receiver for difficult areas.

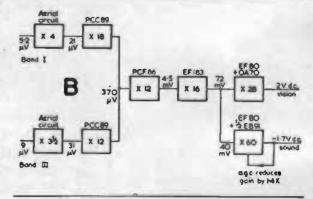
Reception is practicable at vision carrier levels down to $20\mu V$ r.m.s., with enough gain to give 2V d.c. at the detector. To ensure this sensitivity and to allow for deterioration, a fringe receiver should be designed round a nominal sensitivity of $10\mu V$.

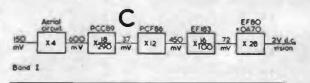
Signal levels in a conventional receiver are shown on line-up 'A'. The Band III sensitivity of $63\mu V$ falls short of the $10\mu V$ requirement by 16dB. Since frame grid valves have a gain advantage of about 6dB over conventional valves, this deficiency can be made up by the use of three frame grid valves.

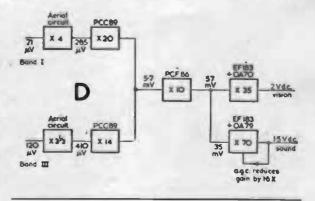
The PCC89 gives improved gain, it has a better noise factor than the PCC84, and it can handle an input signal five times greater without crossmodulation. Thus one of the frame grid valves is used in the r.f. amplifier. The other two should be in stages which are common to sound and vision, so that maximum benefit can be obtained. In line-up 'B' the PCF86 is used in the mixer and the EF183 in the common i.f. stage. The EF80 is retained in the separate i.f. stages. The most severe requirements are met, and the maximum usable gain for low signal levels is provided. In normal service areas, the gain is reduced by a.g.c. for large signals, as shown for Band I at 'C'. Rather greater signals than 150mV can be handled.

A receiver for high signal strength areas does not require 10µV sensitivity, and all requirements can be met by receiver 'A'; but frame grid valves would give the required gain with one less i.f. stage. Line-up 'D' uses the same tuner as the high-sensitivity receiver; but there is no common i.f. stage, and the EF183 is used instead of the EF80. The performance shown is for maximum sensitivity. A signal of 225V r.m.s. can be handled. All normal service area requirements are met, and the receiver is economically attractive. The EF184 can be used in place of the EF183, giving 2dB more gain; but the signal which can be handled is reduced.











MULLARD LIMITED, MULLARD HOUSE, TORRINGTON PLACE, LONDON, W.C.1

Transistor Reliability

IN theory a transistor should last for ever. There is nothing obviously expendable in it as there is in the hot cathode of a valve. The crystal lattice of the solid-state semiconductor remains, under normal operating conditions, a fixed framework through which electrons and holes circulate under the influence of applied fields to and from the external connections of the device. Although the charges associated with individual atoms may change temporarily, and the current carriers may leave with greater energy than that with which they entered, there is no net gain or loss of material, and the chemical nature and physical arrangement of the atoms remain unchanged.

We said "in theory" and "under normal operating conditions." In practice, of course, transistors sometimes fail and in the early days failures were frequent and of many kinds. Indeed there grew up a whole descriptive pathology of transistor diseases such as "sleeping sickness" leading to "slow death" or "sudden death." Some of these were peculiar to the first point-contact transistors and were accounted for by mechanical disturbance of the contact wires. In junction transistors most of the early troubles were due to contamination of the outer surface, particularly by moisture, which gained access, in spite of encapsulation in plastic, through incomplete bonding with the leads or the light bulk permeability of the best synthetic resins hen available. Silicone greases and varnishes in conjunction with low-temperature glass seals or old metal-welding techniques have since mastered the moisture problem.

Given sound mechanical construction to ensure freedom from catastrophic failure through breakages at the lead-out connections and with the flanks safely guarded from surface contamination, deterioration or failure can only occur on the main fronts of the transistor action. Could a breakdown occur as the result of penetrating radiation, for example, cosmic-ray particles? All the evidence is against this as a significant possibility. Irradiation by gamma rays has caused failure in germanium transistors in 5 seconds, and in 30 seconds in silicon, but the failure was due primarily to surface effects and not to bulk effects involving the crystal structure*.

The worst enemy of the transistor is temperature, which is in effect vibration of the crystal lattice. If this becomes too severe dislocations

may occur and the distribution of impurity atoms, so essential to the proper functioning of the junctions, may be permanently disturbed. Whatever the explanation may be, the fact remains that the performance deteriorates as temperature increases, and the deterioration may be permanent if the maker's recommended junction temperatures are exceeded. On the other hand, if the circuit conditions are such as to ensure a conservative rating, there seems no reason to doubt that the life expectation of a transistor may well be 50 to 100 times that of a valve and that it will outlive many of the components with which it is associated.

Space Experiments

NOW that argument in Parliament and in the Press about the propriety of accepting a lift in an American "satellite vehicle" has subsided, it is gratifying to learn that matters have been settled and that the first launching of a satellite from a "Scout" fourstage solid-fuel test vehicle is scheduled for the latter part of 1961. It is even more gratifying to learn that the experiments which have been planned come very close to our interests and will be concerned with the ionosphere, and in particular its outer limits where the terrestrial atmosphere merges into that of the sun.

The long tradition of British interest and achievement in ionospheric research, and especially recent experience in launching ionization experiments in "Skylark" rockets, together with the original laboratory work on ionized gases conducted by research groups under Prof. J. Sayers at Birmingham and Drs. R. L. F. Boyd and A. P. Willmore at University College, London, no doubt influenced the choice.

Experiments planned are based on the Langmuir probe (which measures the rate of change with voltage of current due to impinging ions) and on the measurement of the complex dielectric properties (at 10 Mc/s) of the gaseous medium between parallel plates. These will give information on the identity of heavy ionized particles as well as on the density and temperature of the electron population of the fully ionized plasma at high altitude.

We know that the equipment will be well made and carefully prepared and it only remains to keep our fingers crossed for a successful launch.

^{*&}quot; Transistors Can be Reliable" by C. H. Zierdt, Jr. Electronic Design, 1st April 1957.

Improving the Dynamic Range of

USE OF A PILOT TONE TO COMPENSATE FOR MANUAL COMPRESSION OF THE INPUT

By L. H. BEDFORD, C.B.E., M.A., B.Sc. (Eng.), F.C.G.I., F.I.R.E., M.I.E.E., M. Brit. I.R.E.

HE dynamic range of the domestic tape recorder is nearly but not quite good enough. A signal/noise ratio of 50 dB is frequently claimed, and more usually 40 dB is attained. This is to say that the maximum undistorted signal is 50 to 40 dB above the noise level. Even the latter figure suffices for the recording of low-contrast subject matter such as speech and some types of music. Other types of music, in particular orchestral, organ and even pianoforte, themselves extend over a dynamic range of 40 dB, and so are clearly not directly

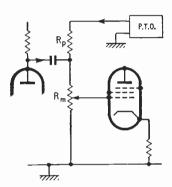


Fig. 1. I asic method of providing a pilot tone with a level proportional to the recording gain control level.

acceptable to the tape system; for this would imply that the reproduction of pianissimo passages must occur at unity signal/noise ratio.

The solution commonly adopted is precisely the same as that in any other communication system of restricted signal/noise ratio, namely that the input signal is compressed by "monitoring"; the monitor manipulates a gain control so as to hold the fortissimo passages below the distortion level and the pianissimo passages suitably above the noise level.

This process is quite justifiable from an engineering standpoint because, once the signal/noise ratio has been thus maintained, the subject matter can be "de-compressed" by manual operation of the gain control on reproduction; this process could be called "de-monitoring." However, such an operation is little more than a theoretical possibility, since not only is the gain-control manipulation an intolerable burden but the information required for correct operation is lacking. Attempts to perform this process automatically, but still in the absence of the correct information, have appeared under the descriptive title "expander circuits"; these are now quite properly in disfavour.

So long as the monitoring process remains a manual operation, and there are good reasons why this should be the case, there is no possibility

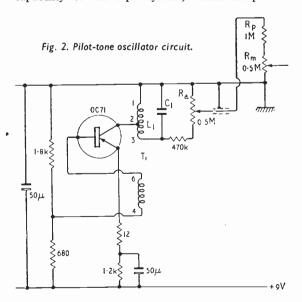
of automatic "de-monitoring" unless the monitoring gain control history can be conveyed as such to the de-monitor.

The device now to be described allows just this and so is quite distinct from an "expander." On recording, the monitoring history is written in terms of a superposed pilot tone whose amplitude is subjected to the same monitoring process as the signal. On reproduction, an a.g.c. circuit holding constant the amplitude of the reproduced pilot tone constitutes a perfect de-monitor. By this combination, the monitoring operations are precisely "unwound". The original contrast is restored and only the signal/noise advantage of the monitoring process is retained.

With this arrangement a new situation arises. The monitor, who hitherto may (or may not) have felt restricted by some artistic considerations, can now go to work to his heart's content, finally throwing aside all decent restraint, confident in the knowledge that his every effort to destroy the music will be exactly defeated.

The pilot-tone operated gain control is by no means a new idea, but it may be well to describe it basically in the present context.

In Fig. 1 is shown the part of the recorder amplifier around the monitoring gain control potentiometer R_m . The top of R_m is connected through a resistance R_p to a pilot-tone oscillator (P.T.O.). The latter provides a constant signal of a frequency outside the audio range but within the recording capability of the tape system; which in practice



Wireless World, March 1960

implies a frequency slightly

implies a frequency slightly above the required audio range,

The pilot tone thus applied is subject to exactly same monitoring variations as the signal, it understood that monitoring takes place exclusively by means of the gain-control potentiometer R_{...}. If then on reproduction the pilot tone is separated out by a frequencyselective circuit and used as input to an circuit controlling audio gain, the monitoring operations are "unwound"

to an accuracy defined only by the perfection of

Tape Recording

he a.g.c.

The following particulars relate to an experimental application of this principle to a domestic tape recorder which has worked with somewhat spectacular results. Some 15 dB stretch of the dynamic range has been obtained.

The pilot-tone frequency was selected at 13.5 kc/s, this being considered the lowest frequency that could be filtered out without detriment to the music. This implies working with a tape speed of 15in per second with a normal R/P head, but will allow 7.5in per second with a modern high-resolution head.

Only a minute power output is required from the pilot-tone oscillator, which suggests a simple form of transistor circuit. This is shown in Fig. 2. The transistor is an OC71 used in the grounded-emitter condition, but for the 12-ohms emitter-follower resistor. The circuit operates at 9 V, 6 mA total, of which the transistor current is 2 mA approx. (It will in fact oscillate and produce adequate power output down to much lower voltages.)

For play-back the same transistor and most of its circuit is used to form the selective pilot-tone amplifier. The arrangement is shown in Fig. 3. In this case, the transistor operates as a grounded-

base amplifier

Input is taken to the emitter through the $13.5 \, \mathrm{kc/s}$ series-tuned circuit L_2C_2 , which is fed from the 15-ohms output of the normal playback amplifier in the recorder. This series-tuned feed, together with the parallel-tuned collector circuit, provide adequate selectivity at $13.5 \, \mathrm{kc/s}$. The collector-tuned circuit now feeds a diode and can produce some $12 \, \mathrm{V}$ of a.g.c. linearly related to the $13.5 \, \mathrm{kc/s}$ input. In practice the a.g.c. potential will not exceed $8 \, \mathrm{V}$.

It remains only to suppress the transmission of $13.5~\rm kc/s$ pilot tone to the final amplifier, which would otherwise be objectionably audible to those of sufficiently-low age group. Series-tuned circuit L_3C_3 and resistance R_3 perform this function, it being understood that the system will invariably work through a power amplifier and not directly through its own speaker.

Finally, Fig. 4 shows the practical arrangement of

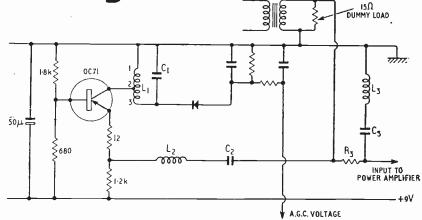


Fig. 3. Selective pilot-tone play-back amplifier for providing an a.g.c. voltage to correct the play-back level.

the complete de-monitor unit in which it only remains to invoke a relay for the purpose of changing over the circuit from its Record to its Play-back function. This relay is operated by the h.t. feed current to the normal bias oscillator. Thus no extra switching is involved and the only adjustment required is the pre-setting of the pilot tone amplitude control \mathbf{R}_q .

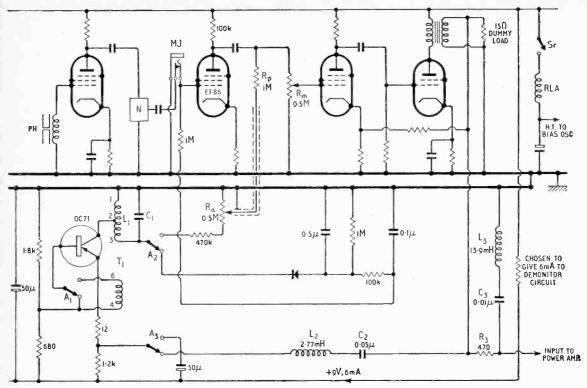
OUTPUT TRANSFORMER OF RECORDER

In Fig. 4 is shown also a skeleton circuit of the basic play-back amplifier in order to clarify the connection of the de-monitor unit to it. It will be seen that the a.g.c. voltage is applied as grid bias to a single stage. The behaviour of this circuit under this somewhat peculiar condition is shown in the following table.

Bias (V)	Input at MJ (mV)	Loss (dB)	Remarks
0 - 5.9 - 6.8 - 7.4 - 7.8 - 8.1	2.5 5 10 20 40 40 40	0 6 12 18 24 30	Distortion observable

The maximum undistorted tape signal at MJ in Fig. 4 being approximately 25 mV, we see that it is possible to operate over a range of 18 dB monitoring compression with an output range of 20 log (7.8/5.9) = 2.5 dB. This is to say that a monitoring range of 18 dB has been reduced by 15.5 dB.

After the adjustment of the three pilot-frequency tuned circuits there remains only one operational adjustment, namely the pilot-tone amplitude control R_a . The setting of this controls the dynamic range of the de-monitoring. The value 15.5 dB referred to above was obtained with the pilot voltage (viewed at the anode of the audio output valve) set to approximately one-third of the nominal distortion point audio voltage. This adjustment should be made with the gain control R_m set to its maximum intended value,



 T_1

 C_1

Fig. 4. Practical arrangement of pilot-tone oscillator and selective play-back amplifier and their connections to skeleton circuit of basic play-back amplifier.

KEY

PH	Play-back head
N	Frequency-correcting network
MJ	Microphone jack
RLA	Relay winding, 400Ω, 45 mA
A_1, A_2, A_3	Relay change-over contacts, shown in un-
	energised condition (Play-back)
S.	Switch on "Record" button (shown in
	Play-back condition)

Pilot-tone oscillator transformer Winding 1, 2, 3 0, 50, 280 turns total 1, 3 40 mH nominal (L_1) 4, 6 4 turns (32 s.w.g. Lewmex wire on LA1 Ferroxcube core) 3470 pF nominal (to tune with L_1 for 13.5 kc/s) 2.77 mH (60 turns of 22 s.w.g. on LA7) 0.05 μ F nominal (to tune with L_2) 13.9 mH (165 turns of 28 s.w.g. on LA1) 0.01 μ F nominal (to tune with L_3) Crystal diode, \star 25V p.i.v.

which will in general be its extreme maximum value. It may be remarked that when the demonitor circuit is in operation the gain control \mathbf{R}_m becomes inoperative as such in the play-back condition because the a.g.c. circuit acts to oppose it. It does, however, control the range over which the demonitor works and will therefore normally be set at maximum on play-back. If it is turned down by some 25 dB with a compensating (but far smaller) increase of gain in the power amplifier, the reproduction becomes un-demonitored. This is a useful test or demonstration procedure.

The circuit can be built up in compact form and will usually be small enough for installation *inside* an existing tape recorder. The likelihood of this is increased by the fact that anyone who is sufficiently quality-conscious to wish to try this arrangement is liable to have remoted the power pack in aid of hum reduction!

In conclusion, we may usefully examine the "economics" of the device. Assessing it in the form described, which by no means carries the idea to its limit, we see that we gain 15 dB of dynamic range for

an increase of tape speed which may for practical reasons need to be 2 to 1. This increase of tape speed by itself would theoretically offer us 3 dB, so we have picked up a net 12 dB by means of the device. For this we have paid a capital investment of one transistor, three pilot-frequency tuned circuits and a relay; surely a nice bargain for the Communication Engineer! Unfortunately on the domestic front the situation is not so good. As already pointed out the existing dynamic range is nearly good enough. Thus the domestic user may not be interested in a further 12 dB even at so trivial a capital cost.

A 16²/₃ r.p.m. record was recently released by Rank Records. This is similar to an ordinary l.p. but recorded and replayed at half speed. In addition, a bass cut and mid-frequency boost are superimposed on the standard recording characteristic, and the recording level is about 5dB lower than usual. The upper recorded frequency limit is about 5kc/s, and the total playing time for a 12in record one hour and thirty-five minutes.

Geneva Conference

FINDINGS OF THE I.T.U. MEETINGS ON FREQUENCY ALLOCATIONS

T Geneva on December 21st representatives of 89 countries signed the new International Telecommunication Convention and a new set of Radio Regulations which come into operation in May, 1961. These two documents were the outcome of two conferences-the Ordinary Administrative Radio Conference and the Plenipotentiary Conferenceorganized by the International Telecommunication Union which together lasted over four months. The main task of the Administrative Radio Conference was to apportion internationally frequencies between the various "services" and users—maritime and aeronautical mobile, maritime and aeronautical radio navigation, meteorology, amateur, land "fixed" and "mobile," broadcasting, etc. This necessitated a complete overhaul of the Atlantic City (1947) allocations and, as with most international conferences, because of the expansion of some services and the growth of the "younger" countries, it was a tug-of-war between the "haves" and the "have nots". In addition to the growing demands of expanding services provision also had to be made for the requirements of new "services" such as space communication, radio astronomy, and tropospheric and ionospheric scatter. Whereas in the Atlantic City regulations the table of frequency allocations did not go above 10,500Mc/s the new table extends to 40Gc/s (40,000Mc/s).

The demands for frequencies in some sections of the radio spectrum have been so great, and are likely to become even greater, that in order to reduce the demands on the 3-30Mc/s band and prevent interference with long-distance radio-communications "administrations are encouraged to use, whenever practicable, any other possible means of communica-It was undoubtedly this h.f. band which posed some of the biggest problems the conference had to solve. So great were they that a small panel of specially-chosen "experts" is being set up to investigate them with a view to reducing the present congestion. Also an entirely new procedure has been adopted for the registration of frequencies. The International Frequency Registration Board, which comprises eleven elected members each of a different nationality but acting as impartial advisers and custodians, has been given the task of preparing a new Master International Frequency Register. So far as broadcasting stations are concerned the broadcasting authorities have been asked to submit four operational schedules each year covering the summer, winter and the two equinoctial seasons. This information will be combined into a Tentative Schedule which will reveal incompatabilities in frequency assignments.

Whilst on the subject of broadcasting we would mention that no changes have been made in the frequency bands, but it is good to know that provision has been made just below the television Band I for the ionospheric scatter stations, some of which have been operating on frequencies in the lower channels of the band. Ionospheric scatter stations designed to operate over distances exceeding 800km

must confine their transmissions to the following bands: 32.6-33, 34.6-35, 36.2-36.8 and 39-39.4Mc/s.

The use of broadcasting stations, both television and sound, "on ships or aircraft outside national

territorial waters is prohibited."

New frequency tolerances for broadcasting stations are to come into force in January, 1966. In the 10-1605kc/s band it will be reduced from 20c/s to 10c/s, in the 1605-4000kc/s band from 50 to 20 parts in 10⁶ and in the 4-29.7Mc/s band from 30 to 15 parts in 10⁶. There have also been closer limits imposed on the television and sound broadcasting stations in the v.h.f. and u.h.f. bands.

An interesting aspect of the Regulations is the provision for communication between "space" vehicles and earth-to-space services for research purposes. The frequencies (Mc/s) allocated are 10.003-10.005, 19.990-20.010, 183.1-184.1, 1700-1710 and 2290-2300. One of the hundreds of documents circulated during the conferences dealt with telecommunications and the peaceful uses of outer-space vehicles. It covered not only the telecommunication needs of space craft both for their remote control and for transmission of information but went on to say "artificial satellites will undoubtedly be used in the near future to establish new telegraph and telephone connections . . . and . . . sound and television broadcasting services." In 1963 a special conference will be called to consider the problems of space communication.

The frequency allocation table has been called the regulation with a 1,000 footnotes for there are so many variations from and modifications and exceptions to the basic plan. Another point of interest about the plan is that it defines for the first time priorities where a number of services share a band of frequencies. Services are defined as primary, permitted or secondary. The first two have equal rights except that a primary service has a priority in the choice of frequencies. Stations in the third category must not cause harmful interference to or claim protection from either of the others.

So far as amateurs in Region I, which includes the U.K., are concerned, the only changes in their frequency allocations were a loss of 50kc/s in the 7Mc/s band and a reduction in the width of the 420-450Mc/s band to 430-440Mc/s.

A world-wide reservation of the following bands has been made "for the use and development of airborne electronic aids to air navigation":—960-1215, 1535-1660, 4200-4400, 5000-5250 and 15400-15700Mc/s. The frequency of 243Mc/s has been

allocated to survival craft.

It has been impossible within the limitations of this short article to give an exhaustive survey of the Radio Regulations, which cover some 600 pages; moreover, some of the regulations or recommendations are of interest to only a comparatively small section of the radio fraternity. However, we propose to publish from time to time items of interest to particular sections of the diverse readership of Wireless World.

How Long Will a Transistor Live?

ADVICE TO THE USER GAINED FROM RECENT EXPERIENCE

By R. BREWER*

T is natural that with many commercial products we ask the question, "What is its expected life?" Transistors belong to this class of products, and even the non-technical user is beginning to ask, "How long will a transistor live?"

This article has been prepared to give guidance in a problem that appears simple, but which is really quite complex, and the basic information is given for a broad, but realistic, approach to the general

question of transistor life.

The simple question, "How long will a transistor live?" has no simple answer, because it begs another question: "When is a transistor dead?" problem is put this way we are led to the answer of a rather different question that is probably more important to the user of transistors than the question first asked. It is also worth remembering that, as with human beings, living conditions strongly influence the length of life!

A transistor dies either when a catastrophic event, such as a short-circuit between elements, occurs, or when an electrical characteristic has deteriorated to a point that is unacceptable in the circuit in which the transistor is being used. The definition of death is thus closely linked to the conditions of use; only the catastrophic type of failure is unambiguous and valid

for all circuits.

Comparison with Valves.—It is useful to compare the life patterns of transistors with those of thermionic valves, because these also have catastrophic and deterioration type failures. Apart from the generally much lower rate of catastrophic failures in transistors, the most striking difference between the life characteristics of the two types of device is that for many valves the deterioration process, which is gradual in the early stages, may later become so rapid

* Research Laboratories, General Electric Company.

as to cause system breakdown almost regardless of differences in circuit conditions. In other words, once a valve has started to "nose dive" it is more or less unacceptable in any circuit.

By comparison, the deterioration process in many types of transistors is extremely gradual, and after an initial "settling-down" period, subsequent changes may be small. The two characteristics most likely to change during life are current gain and leakage current. Fig. 1 shows a typical record of gain in a sample of GET103 transistors which have been on life test for 20,000 hours (nearly 2½ years). These transistors are running at a junction temperature of 65°C, a figure which is fully adequate for applications such as portable personal radio receivers. It is worth noting that in apparatus of this type, where the duty may average about 5 hours a day, it would take about 11 years to build up a total operating time of 20,000

It must be remembered that transistor manufacturers are constantly striving to improve the quality of their products, including the life characteristics. Long-term life test evidence inevitably lags behind the latest production processes, and records such as that shown in Fig. 1 cannot reflect recent improvements resulting from increased production experience. Since this life test started, the introduction of better techniques has enabled the maximum junction temperature of the GET103 transistor to be raised from 65° to 85°C. Changes of this kind depend on a thorough appraisal of all the many factors involved, including extensive life test evidence, but another two years must elapse before the 20,000hours performance at the higher junction temperature has been confirmed. At present there are no reliable, well-established techniques for accelerating transistor life effects—we have no substitute for time.

Pattern of Transistor Life.—Many of the early



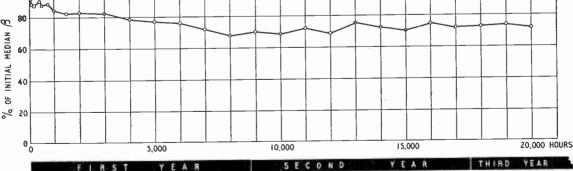


Fig. 1. Change during life of median value of β for a sample of fifteen GET103 transistors manufactured in May 1957. Junction temperature during electrical life test is approximately 65°C; while β is measured at 25°C.

Wireless World, March 1960

junction transistors exhibited the degradation type of life failure due to imperfect manufacturing techniques, but major improvements in this respect have been made in recent years. The characteristic "nose dive" seen in many valves is generally absent in transistors, but there may be fairly large differences in the rate of change among individual specimens of similar transistors. While there is little reason to expect any tendency towards major changes in the average characteristics late in life, this fact has yet to be proved.

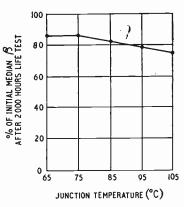
A study of the evidence from life tests such as that referred to above suggests that transistors in general may be capable of outliving the equipment in which they are incorporated. It must not be assumed, however, that all the transistors in a batch will have identical survival characteristics. Like every other mass-produced article, the transistor is subject to chance variations in manufacture that can cause an unpredictable breakdown during life, or an unusual change in gain or leakage current, and it is the incidence of troubles of this kind, rather than a general end of life having been reached, with which we are really concerned. For the operators of equipment using large numbers of transistors the question is therefore not so much "How long will a transistor live?" as "How often will chance failures occur?"

Time Between Failures.—The last question is one to which some provisional answers can be given, though they do not necessarily apply to all types of transistor, and it has yet to be shown that the failure rate is the same throughout life. Evidence from various types of transistor equipment, and from life tests involving many millions of transistor-hours, indicates that the failure rate for low-frequency germanium transistors of the GET103 type, for example, is in the region of 0.05 to 0.01% per thousand hours. With this information we can immediately work out the mean time between failures (M.T.B.F.) for an equipment, and this is something we really want to know. For example, assuming a rate of 0.05% per 1,000 hours, the M.T.B.F. in a unit using 1,000 transistors of the same type would be 2,000 hours. This order of reliability applies to germanium transistors working in ambient temperatures up to 40° to 60°C, and with junction temperatures in the region of 70° to 85°C.

Guidance to Users.—It may be held that the rapidity of technical advances in the industry, and the lack of sufficient time for long-term effects to have become manifest, make it unwise to forecast the ultimate reliability of transistors at the present time. Against this is the fact that transistors are already being used in equipment where long life is a major requirement, and it therefore seems advisable to give users some guidance on the best approach to the subject in the light of present knowledge.

The failure rate figure given above is a realistic one for several types of low-frequency germanium transistors now in production, but in considering it, a number of important reservations must be made. First, very wide differences may exist in the life characteristic of types of transistors that are superficially similar. Also it must be realized that some of the latest types of high-frequency transistors are made by advanced techniques which may produce

Fig. 2. Change in current gain of GET103 transistor after 2,000 hours electrical life test at five junction temperatures. Sample size of 20 at each temperature; β measured at 25°C.



life characteristics differing from those associated with the more well-established techniques used for low-frequency devices. Secondly, circuit characteristics and operating conditions can have far-reaching effects on the reliability achieved in practice. This cannot be stressed too strongly. Such matters are obviously absent in any general statement, and it is therefore essential to consider the details in each application. As an illustration, Fig. 2 shows how current gain can be affected by junction temperature, and it will be seen that at the higher junction temperatures greater circuit tolerance is necessary. This subject has been dealt with more fully in a paper† which is now in course of publication and should be available in the near future.

Circuit Design Factors.—The foregoing comments show how important are the details of operating conditions when considering statements about the life and reliability of transistors. Reliability information in these circumstances may easily be misconstrued, and general statements, such as have been given here, serve only as landmarks.

The designers and operators of equipment using transistors have a large part to play in determining the M.T.B.F. of the transistors used in the systems with which they are concerned, and their contribution to transistor reliability will be most effective if the following precepts are remembered:

(1) Survey the electrical and environmental conditions under which the transistors will be used, paying particular attention to the extreme values of supply voltages and temperatures likely to be encountered. The possibility of voltage transients and oscillations should also be checked.

(2) Employ tolerant circuit arrangements so that changes in transistor characteristics during life have a minimum effect on system performance.

(3) Consult the transistor manufacturer for advice if necessary.

(4) Watch for accidental misuse as a cause of transistor failure. An increasing body of information suggests that high early failure rates in transistors are due to faulty installation and maintenance procedures.

If these points are kept in mind when new equipment is being designed or installed, the chances of realizing the low failure rate of which the transistor is capable will be significantly increased.

^{†&}quot; A Reliability Appraisal of Semiconductor Devices," by R. Brewer and W. W. D. Wyatt, presented at the I.E.E. International Convention on Transistors and Associated Semiconductor Devices, May 1959.

Transformerless Circuits for Broadcast Receivers

NEW CIRCUITS

DISPENSE WITH

A VARIETY OF

CIRCUIT ELEMENTS

By R. C. V. MACARIO*, Ph.D. and N. E. BROADBERRY*, Grad.Brit.I.R.E.

AONSTRUCTION of the electronic circuit can be made simpler when transformers and other wire wound components do not form part of the circuit. This may be regarded as being due to the greater versatility of the more compact circuit elements such as the resistor, the capacitor and the transistor, which together lead to neat circuit arrangements. The modern trend of electronics towards the solid state circuit also favours the use of this type of component, to which may be added other solid state components such as the ceramic i.f. transformer.

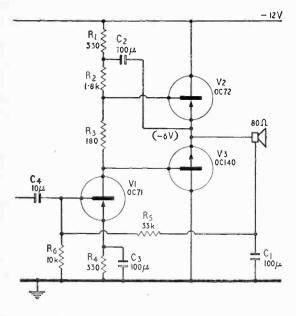


Fig. 1. Audio amplifier and 150-mW output stage for portable receiver using 12-volt two-terminal supply.

However, a review of the extensive literature on the modern transistor broadcast receiver suggests that the circuits used today are not of this type; unfortunately they make use of components other than the recommended ones. In this article alternative circuit arrangements are introduced designed to match in with the above ideas.

Portable-receiver A.F. Stage.—For the small portable receiver Class-B operation is essential unless one of the more esoteric forms of signal-controlled Class-A circuits is introduced. The complementary n-p-n/p-n-p arrangement clearly provides the simplest design without a driver trans-

former for a Class-B circuit. The criterion for a good design would appear to be a maximum power output with minimum distortion and minimum complexity. This suggests the use of a common-emitter stage driving a common-collector pair. A discussion of this type of transistor amplifier is to be found, for instance, in the *Handbook of Semiconductor Electronics*². In a final form the circuit leads to a fairly straightforward design as shown in Fig. 1.

A description of the circuit behaviour to d.c. is as follows: the forward bias of V1 is set so that its collector potential is quiescent at -6V with the d.c. collector load $R_1+R_2+R_3$. This bias depends on the current through R4, the emitter resistor, and the base potential determined by the potential divider R_δ and R_θ . The collector of V1 is coupled directly to the bases of V2 and V3; thus the connection common to the emitters of V2 and V3 assumes a potential almost identical with that of V1 collector (emitter-follower action). To prevent a flow of d.c. through the load its "earthy" connection may be taken either to a centre tap on the battery or to earth through a d.c. blocking capacitor (C1). When a blocking capacitor is used the p.d. across C1 follows V1 collector potential; therefore, if the potential divider R_5 , R_6 is fed from this point, a direct-coupled negative-feedback loop is closed, a loop which helps to maintain equal potentials across V2 and V3.

Turning now to the a.c. or signal conditions; the collector of V1 drives the bases of V2 and V3, which conduct on alternate half cycles, V2 on the negative and V3 on the positive half cycles. R_3 provides a small forward bias for the output pair of transistors, to eliminate crossover distortion. As R_5 is taken from the "earthy" end of the loudspeaker there is no signal feedback. If, on the other hand R_5 is taken from the "live" end of the loudspeaker there is feedback and the power gain of the stage falls by about 10dB. There is, nevertheless, some feedback due to the fact that the full output voltage of V1 is not applied between the base and emitter of the output transistor because of the presence of the load in the emitter circuit, but this has been corrected by including the bootstrap circuit C_2 , R_1 . The overall voltage gain of the stage is about 44dB and the input impedance is about $1k\Omega$.

In order to estimate the correct conditions for the driver transistor the following method may be used. Considering the half cycle when V2 conducts, the peak output current occurs when the driver just cuts off and at this instant all the base current i_{bV3} of V2 is supplied by R_2 . The voltage which drives this current will be equal to 6 volts which is the supply voltage across V2, and this drive will also have to overcome the input impedance R_{in} of V2 and

^{*} Plessey Co. Ltd.

the reflected load $\beta_{v\varrho}.R_L$, where β is the common-emitter current gain.

:
$$(R_2+R_{in}+\beta_{v2}, R_L)$$
. $i_{bv2}=6 \text{ V}$ (1)

 i_{1} , v_{2} in turn will be equal to $1/\beta_{v_{2}}$ times the peak load current,

 $i_{10}v_2 = (1/\beta_{v_2}) \cdot \sqrt{(2 \times \text{peak power/R}_1)} \cdot (2)$ Considering a current gain $\beta = 50$ and for 100 mW output it follows that:—

$$i_{\text{bV2}} = 1 \text{ mA}$$

 $R_2 + R_{\text{in}} = 2 \text{ k }\Omega.$

Since R_{in} is only about 100 ohms, R_2 should be slightly less than 2,000 ohms. Alternatively, if $\beta=100$ and for 150 mW output:—

$$i_{\rm bV2} = 0.61 \text{ mA}$$

 $R_2 + R_{\rm in} = 1.8 \text{ k}\Omega.$

It is clear that to increase the output power the β of the output stages should be increased. On the other hand, to increase the output power by reducing the speaker load leads, unfortunately, to excessive current consumption by the driver stage (because R_2 would have to be reduced) and this would reduce markedly the battery life.

The correct collector current for V1 is estimated

$$i_{eV1} = 6/(R_1 + R_2 + R_3/2) = 2.75 \text{ mA}$$

The linearity of the output is shown in Fig. 2 and indicates the circuit operates satisfactorily up to outputs of approximately 150 milliwatts. A 12-volt battery supply has been used as this allows a much greater degree of flexibility throughout the design of the receiver circuits; this may be slightly higher than normal but it is single-ended.

Detector Stages.—The output stage just described may be driven directly from a detector stage using the conventional series-diode circuit. However, this form of detection may not always be the most efficient type because of the series resistance of the diode. The following measurements were made to compare the performance of series-diode, shunt-diode, and transistor-detector circuits. Fig. 3 shows the three circuits that were explored and in circuits (i) and (ii) the transistor is the final i.f. amplifier and the diode detects; each circuit was adjusted for maximum performance. In (i) and (ii) a ceramic wave-filter element can be connected across the

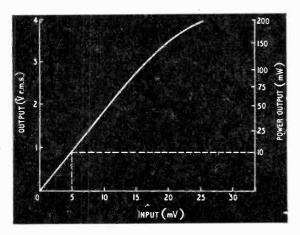


Fig. 2. Input/output choracteristic of audio-frequency amplifier and power output stage shown in Fig. 1.

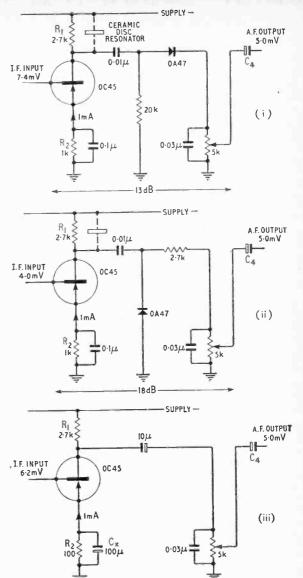
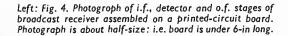


Fig. 3. Series (i)- and shunt (ii)- detector circuits and their final i.f. amplifiers compared with transistor-detector equivalent (iii). I.F. input (modulated 30% at 400c/s) shown produces 10mW a.f. output in I-k Ω following load.

14.5dB

collector load of the transistor if increased i.f. selectivity is required. However, a comparison of the voltage conversion gain from i.f. to a.f. suggests firstly, that the shunt-diode is more efficient than the series connection, and secondly, the transistor detector is more attractive as a diode is not used.

In Fig. 3(iii) the transistor acts both as the a.f. amplifier and the rectifying amplifier³. R₁ mainly determines the direct potential on the collector as the effective load is the input impedance of the next stage. The emitter current is adjusted by varying the forward bias on the base until maximum sensitivity is obtained, noting that the current required for optimum detector efficiency is usually lower



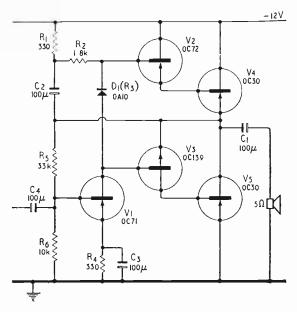
Below: Fig. 5. Medium-power (2-W) a.f. amplifier and output stage using 150-mW amplifier of Fig. 1 as basis of driver stage.

than that required for maximum β . R_2 is included to provide d.c. and thermal stability; the value chosen is such that when a signal is present it does not unduly disturb the operating conditions. The stage also produces amplified a.g.c. and the preceding i.f. stage can be so biased that an overload-protection diode is no longer necessary⁴. The emitter bypass capacitance C_X is shown as an electrolytic in Fig. 3(iii): $0.1\mu F$ may be used at the expense of slight reduction of a.f. gain, but this gives the advantage of a.f. negative feedback.

Practical Construction.—To illustrate the simplicity of construction of the type of circuit just described, Fig. 4 shows a half-scale photograph of a printed-circuit-board assembly containing the a.f., detector and i.f. stage of a standard receiver. The a.f. and detector circuits are as described in Fig. 1 and Fig. 3(iii). The two small discs at the left-hand end of the board are piezoelectric ceramic i.f. transformers¹.

Higher Power Audio Stage.—For the better quality receiver, a higher-power a.f. stage is essential and power outputs of up to 2 watts may be desired. This would include both the car radio and f.m./a.m. home use where, on certain types of programme, strong peaks that would almost certainly be distorted by the 150-milliwatt stage occur even in low-level listening.

Fig. 5 shows the circuit and component values of the complete a.f. stage that will deliver 2 watts into a 5-ohm load. It will be observed that Fig. 5 is an extension of Fig. 1 and again represents a circuit which aims at the best output for the simplest design. The circuit may be described by extending the description of the previous audio stage. The complementary pair V2 and V3, acting as phase splitters, now drive the p-n-p output transistors V4 and V5: hence the last two stages both operate in a Class-B mode. The junction of C2 and R_5 and the output terminal is again held at -6 volts by the feedback through R5 controlling the current through the transistor V1. The small forward bias for both pairs of Class-B transistors is again derived in V1 load; but R3 has been replaced by a junction diode which has a negative temperature coefficient and hence improves the temperaturestability factor of the overall circuit. R₁ and C again form the bootstrap circuit enabling the full output of the driver to be applied to both halves of the following stages. The final stages V4 and V5 are also driven from low-impedance sources and this reduces the effect of current-gain fall-off at large Consequently the proportion of drive currents. distortion components in the output is reduced.



The design procedure is exactly similar to the previous amplifier. Here, however, as the driver only "sees" the loudspeaker through V2 and V4, β_{V2} of expressions (1) and (2) becomes β_{V2} . β_{V4} , whilst the input impedance of V2 is now much higher, at about $1k\Omega$, because of the emitter load. Thus for 2 watts into a 5-ohm loudspeaker:—

$$i_{\text{bV2}} = (1/50.30) \sqrt{(4/5)} = 0.6 \text{ mA},$$

$$R_2 + R_{in} = 2.5 \text{ k}\Omega$$

$$R_2$$
=1.5 kΩ. (A higher β would increase the estimate).

A 1.8 k Ω resistor was used in the practical circuit, the correct d.c. for the driver is then again 2.75 mA. This current is set by R_5 . The bleed current through R_5 should be high relative to $I_{c0} + I_c/\beta_{v1}$ to minimize temperature drift effects and the unbalancing action of the circuit due to differences between individual transistors.

The quiescent current for the complete stage is about 5 milliamps, whilst a musical programme may average about 50 mA. The battery supply is again a single-ended 12-volt source.

Further Improvements.—Although there is no detectable distortion on a.m. reception at high listening levels using the circuit shown in Fig. 5, the amplifier may be elaborated to give even better quality if it is used, say, for f.m. receivers. The frequency response appears adequate, but for wider bandwidths the higher harmonics of distortion may become noticeable. Possible extensions of the design would be the inclusion of resistors across the base-emitter junction of V4 and V5 to reduce the variation in input

impedance of the final stages; also a resistor could be inserted in series with the emitter of V3 to balance the impedance seen by V1 on both halves of the signal cycle. The capacitance of C1 would obviously have to be increased to make full use of the bass response, but listening to the amplifier with the loudspeaker mounted on a large baffle indicates that most of the low frequencies are reproduced.

On the other hand, if cost is important the following modifications can be made. The speaker can replace R_1 , thereby dispensing with R_1 and C_1 , provided the feedback in Fig. 1, introduced by connecting R_5 to the emitters of V2 and V3 and the d.c. through the loudspeaker can be tolerated and, secondly, R₄ and C₃ may be left out if a much lower value for R₆ is acceptable.

REFERENCES

1 "Ceramic I.F. Filters Match Transistors," D. Elders and E. Gikow, Electronics, p. 59, April 25, 1958: (Vol. 31).

² Handbook of Semiconductor Electronics, Ed. L. P. Hunter. McGraw Hill, 1956.

3" Transistor A.G.C.," W. Woods-Hill, Wireless World, February, 1958: (Vol. 64, p. 94).

4" Transistor A.G.C. Circuits," Wireless World,

November 1959: (Vol. 65, p. 508).

Manufacturers' Products

NEW ELECTRONIC EQUIPMENT AND ACCESSORIES

Transistor Control Relay

IN some control systems it is necessary to use either very light or high-resistance contacts. In the first case even slight arcing can damage seriously the device and, with high-resistance contacts, it is often difficult to provide easily for satisfactory operation. The Electro Methods Control Relay Type 273B uses a maximum control-circuit power of less than 5mW, and this sensitivity is achieved by the use of a transistor which switches the current through the coil of a mechanical relay. The control circuits are completely isolated from the mains supply (200 to 250V, 50 to 60c/s) which powers the Control Relay. Indicator lamps, denoting the state of the relay, are fitted.

The load-switching capacity of the main relay is 15A at 250V a.c. (non-inductive) and the connections can be arranged so that the unit "fails safe." Manufacturers: Electro Methods, Ltd., Caxton Way, Stevenage, Hertfordshire.

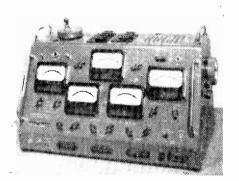
Valve Analyser and Bridge

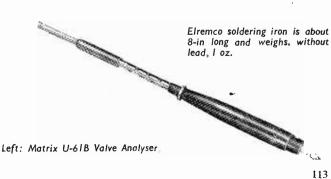
DESIGNED for laboratory investigations rather than the quick checking of valves in a servicing workshop, the main feature of the Metrix Type U-61B valve analyser is that valve-electrode potentials and currents may be indicated simultaneously and separately on five large-scale meters. Four, separate, independently-variable, stabilized power supplies feed a maximum of two screengrid, one anode and one control-grid electrodes and heater supplies between 1.1V, 3A and 117V, 0.15A are available. Due to the provision of stabilized supplies it is possible to plot manually a valve's static characteris-tics with the minimum of incidental readjustment of electrode potentials. These supplies may also be used for the energizing of apparatus external to the tester: sockets for output are provided on the front panel. An unusual point is that the multitude of seldom-used sockets fitted to most valve testers is avoided by the provision of individual plug-in panels carrying one or two sockets. With the addition of the Valve Bridge Type 661 (which can be used independently) dynamic characteristics can be measured at a variety of electrode potentials.

The Valve Analyser costs £275 and the Bridge £370: both are available from Metrix Instruments, Ltd., 59 Victoria Road, Surbiton, Surrey.

Lightweight Low-Voltage Soldering Irons

THE Elremco soldering iron Type SMS is available in several ratings between 10W and 75W at 20V. Primarily designed for use with the Elremco low-voltage bus-bar installation, the high bit temperature makes possible soldering direct to equipment chassis, large solder lugs, etc., without changing to an iron of higher rating than that used for wiring. Twenty standard bits are available with tip diameters from 18-in to 18-in and the bit and metal stem are electrically isolated from the element and it connections, so making the iron par-ticularly suitable for work on transistor equipment. The moulded-plastics handle has a hexagon-shaped section, which reduces to a minimum the chances of rolling when set down: the weight of the iron, without cable, is about loz, and the price is 11s 10d each for quantities of 100 and over. Manufacturers: Electrical Remote Control Co. Ltd., The Fairway, Bush Fair, Harlow, Essex.





Wireless World, March 1960

WORLD OF WIRELESS

E.F.F.I. Conference

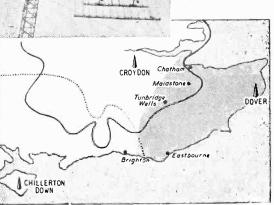
THE Electronic Forum for Industry, consisting of associations with common aims in making better known the developments and applications of electronics in industry, is to hold a three-day conference at Olympia, London, during the forthcoming Instruments, Electronics and Automation Exhibition. Details of the conference, which will be held on May 24th, 25th and 26th, will be announced later, but, broadly, the three sessions will cover practical experience of data processing, factory applications (including machine tool control) and instrumentation.

A prospectus and form of application will be available shortly from the E.F.F.I., c/o The Electronic Engineering Association, 11 Green Street, London, W.1.

Domestic Receiver Production

DESPATCHES of both television and sound radio receivers were at record levels in 1959, according to provisional figures based on returns supplied by members to the British Radio Equipment Manufacturers' Association, which gave the net figures

Estimated service area of the new I.T.A. station at Church Hougham, near Dover, is shown (shaded) in relation to the 0.5mVlm contours of the London and Isle of Wight stations. Inset, riggers are shown ascending the 750-foot mast (erected by B.I. Callender's) with the channel 10 aerials (designed by E.M.I.). The aerials are screened to reduce to a minimum radiation towards France. Transmissions are vertically polarized with a vision er., p. of 100kW in the direction of maximum radiation. The transmitters and associated equipment were installed by Marconi's and the station is operated by Southern Television in conjunction with the Chillerton Down station.



of deliveries by manufacturers to the home trade, including those to rental and relay companies.

The year's despatches of TV receivers totalled 2,745,000, which was 36% above 1958, the previous highest year. The total of 1,551,000 for sound receivers was 19% more than that for 1958 and 14% above 1957, the previous highest year. Despatches of radiogramophones at 187,000 were 14% lower than for 1958 and 30% below 1957.

1961 Computer Exhibition

ALTHOUGH the 1958 exhibition of electronic computers was intended to be a "once only" show, the joint organizers, the Electronic Engineering Association and the Office Appliance and Business Equipment Trades Association, have been encouraged to hold another. It will be held in the National Hall, Olympia, London, from October 4th to 12th next year. A business computer symposium will again be held concurrently with the exhibition.

Audio Fair

DEMONSTRATION rooms have been booked by nearly all the 70 or more exhibitors with stands at the forthcoming Audio Fair to be held in the Hotel Russell, London, W.1, from April 21st to 24th. We hope to publish a list of exhibitors in our next issue.

As in past years, free admission tickets will be available from the organizers (Audio Fairs, Ltd., 22 Orchard Street, London, W.1), exhibitors, audio dealers and Wireless World. As some of the tickets are for specific days, it would help if applicants stated their day of preference, and they are also asked to enclose a stamped addressed envelope.

P.A. Show.—Twenty exhibitors have taken space at the exhibition being arranged by the Association of Public Address Engineers at the King's Head Hotel, Harrow-on-the-Hill, Middx., on March 9th. In the morning admission is restricted to A.P.A.E. members and the Press, in the afternoon to the trade, but from 5.0 to 7.30 the public will be admitted. Free tickets are obtainable from Alex J. Walker, 394 Northolt Road, South Harrow, Middx.

The Electronic Organ Constructors' Society, proposed in a letter from A. Le Boutillier in our September, 1959, issue, has now been formed. Its objects are to "encourage and assist amateurs in the construction of organs [electronic and pipe] and to provide opportunities for the exchange of ideas and technical information." Alan Douglas is president, L. W. Roche, chairman, and A. Le Boutillier, whose address is 26 St. Catherine's Road, Chingford, London, E.4, is secretary.

B.A.R.T.G.—A change of title without altering the initials has been made by the British Amateur Radio Teletype Group since their attention has been drawn to the fact that the word "Teletype" is a registered trade mark. The word "teleprinter" has been substituted.

Transistors and transistor parts may now be imported from the Dollar Area without a Board of Trade Licence. Imports of Japanese transistors and transistor sets are still subject to control.

WIRELESS WORLD, MARCH 1960

"Rocket and Satellite Instrumentation" is the title of a one-day symposium being organized jointly by the Society of Instrument Technology and the British Interplanetary Society for September 1st in London. Subjects it is proposed to cover include a design study of a communications satellite; a digital data reduction system for use in the static testing of rocket motors; transducers for rocket motor testing; and the measurement, transmission and recording of data in the Skylark missile.

Non-Destructive Testing.—The general theme of a joint meeting between the Non-destructive Testing Group of the Institute of Physics and the Société Francaise de Métallurgie, to be held in London from May 2nd to 4th, will be the relationship between structure and physical properties of metals. Details of the programme, which will include papers on recent advances in non-destructive testing techniques, are obtainable from the Institute, 47 Belgrave Square, London, S.W.1.

Ten Million.—Last year's increase of 1;215,352 in the number of combined TV/sound licences in the U.K. brought the total at the end of the year to 10,114,419. The first million television licences was reached in 1951—five years after the introduction of television licences—and since 1953 the average annual increase has been about 1,200,000.

Italian V.H.F.—Reference was made in January (page 12) to the rapidly increasing number of v.h.f. sound broadcasting stations in Italy. The story is going around that when the European Broadcasting Union asked the Italian broadcasting organization how many v.h.f. stations they had in service, on December 31st, they asked if they wanted the morning or evening total!

B.A.T.C. Convention.—The fifth Amateur Television Convention arranged by the British Amateur Television Club will be held on September 10th, in the Conway Hall, London, W.C.1. Details will be available later from D. S. Reid, 149 Ongar Road, Brentwood, Essex.

R.S.G.B. Officers.—The new president of the Radio Society of Great Britain is W. R. Metcalfe (G3DQ) of Whitby, Yorks. The executive vice-president is H. A. Bartlett (G5QA), who was president in 1955. The four ordinary members of the council elected at the recent annual general meeting are: C. H. L. Edwards (G8TL), R. C. Hills (G3HRH), A. O. Milne (G2MI) and G. M. C. Stone (G3FZL).

Relay Services Association.—Sir Walter Womersley, Bart., president of the Relay Services Association since 1948, has again been re-elected. J. W. Kinsman (Relay Exchanges) has become chairman of the council with B. R. King (British Relay Wireless) as deputy chairman.

B & K Laboratories, who have for the past five years organized an international instruments show in London, have decided not to hold one this year. When their show was conceived all the existing major exhibitions were national rather than international in character. This is no longer true. In making this announcement **B** & K Laboratories state, "We are not exhibition organizers and we hope that specialists in this field will continue the trend towards larger and better international exhibitions."

An international festival of sound is being held in Paris from March 18th to 23rd. Organized by the Syndicat des Industries Electroniques de Reproduction et d'Enregistrement (S.I.E.R.E.), it includes an exhibition, demonstrations of equipment, conferences and demonstrations of stereo f.m. transmissions by the French broadcasting authority, R.T.F.

A Moscow exhibition of British scientific instruments is being organized by the Scientific Instrument Manufacturers' Association for June 16th to 26th.

"Marine Electrics."—Electronic and electrical equipment from the aircraft carriers *Victorious* and *Hermes* will be included in the "Marine Electrics" feature at the 9th National Electrical Engineers Exhibition at Earls Court from April 5th to 9th.



Dip. Tech. Awards.—Sir Harold Roxbee Cox, who has succeeded Lord Hives as chairman of the National Council for Technological Awards, presenting a Diploma in Technology to Ian Stanley (A.E.I. Research Laboratories), one of the first five successful candidates from the North ampton College of Advanced Technology, Loncon. Other recipients, all of whom had completed a four-year sandwich course in applied physics which is heavily biased in electronics, were: Edward Feakes (R.A.E.), Frank Jacob (Vickers Armstrong), John Swain (A.W.R.E.) and Shirley Wallis (R.A.E.), the first woman to receive the Diploma.

Education and Training.—"The changing pattern of electrical engineering education and training" is the theme of a conference to be held at the Polytechnic, Regent Street, London, W.l, on the morning of March 8th. Particulars of the conference, of which R. E. Burnett, managing director of Marconi Instruments, is chairman, can be obtained from the kegional Advisory Council for Technological Education, Tavistock House South, Tavistock Square, London, W.C.1.

Teacher Training.—Many of the 2,000 or more teachers recruited each year for technical colleges join the staffs direct from industry. Special one-term courses aimed at improving the quality of teaching are now being provided for such teachers at four colleges. Course will start after Easter at the three Technical Training Colleges—Bolton, Huddersfield and Garnett College, London—and the College of Technology, Wolverhampton.

Vacation Courses.—Sound radio and TV servicing and telecommunication engineering are among the wide variety of subjects listed in the "Programme of Short Courses" for teachers and others in the educational service arranged by the Ministry of Education for the Baster and Summer vacations. The 30-page booklet is obtainable from the Ministry, Curzon Street, London, W.1.

A two-day conference on "The training of the industrial physicist" is being held by the Institute of Physics in Birmingham on April 21st and 22nd. Detailed programmes and registration forms are available from the Institute, 47 Belgrave Square, London, S.W.1.

Analogue Computing Techniques.—A five-day introductory course on analogue computing techniques is being conducted at the Loughborough College of Technology, Leicestershire, from April 11th. The fee for the course is 10 gns, plus 5 gns for residence.

Personalities

The Hon. R. T. B. Wynn, C.B.E., M.A., M.I.E.E., Chief Engineer of the B.B.C. since 1952, is retiring on April 19th. Educated at Uppingham School and Trinity Hall, Cambridge, he received his engineering training with Siemens Bros. In 1922 he joined the staff at Marconi's experimental station at Writtle, Essex, where he was associated with the early broadcasts. He joined the B.B.C. in 1926 as head of the Technical Correspondence Department. Mr. Wynn, who is 62, has successively been head of the Operations and Maintenance Department and assistant chief engineer. F. C. McLean, C.B.E., B.Sc., M.I.E.E., Deputy Chief Engineer of the B.B.C., succeeds Mr. Wynn. The post is being redesignated Deputy Director of Engineering and the post of deputy chief engineer is being abolished. Mr. McLean, who is 56, joined Standard Telephones & Cables in 1925 after graduating at Birmingham University. He left S.T.C. to join the B.B.C.'s Planning and Installations Department in 1937. He headed various groups within the Engineering Division prior to his appointment in 1952 as Deputy Chief Engineer. He has been a member of the Radio Research Board since 1958.





F. C. McLean

K. G. Smith

K. Graham Smith, this year's vice-president of the Radio and Electronic Component Manufacturers' Federation, has been appointed deputy managing director of N.S.F. Ltd. He joined the company nearly 20 years ago as chief engineer and has been a member of the board since 1947. He has represented N.S.F. on the council of the R.E.C.M.F. for some years and was chairman in 1958/59. N.S.F., who are members of the Simms Motor and Electronics Corporation, also announce the appointment of Percy C. D. Mace as a director. He was for some years works manager of Welwyn Electrical Laboratories, and since March 1958 has been general manager of the N.S.F. works at Keighley, Yorks,

S. R. Wilkins, who, as announced in "News from the Industry," succeeds J. H. Rawlings as managing director of Avo Ltd., joined the company (then known as the Automatic Coil Winder and Electrical Equipment Company) in 1934. Six years later he was appointed chief electronic engineer and manager of the electronic instrument section in charge of design and production. He became technical director in 1956.

W. P. Rowley, M.B.E., M.Brit.I.R.E., has joined W.S. Electronics Ltd., a wholly owned subsidiary of K.G. (Holdings) Ltd., as assistant managing director. During the war he was commissioned in the Royal Corps of Signals and after a period as lecturer in radio at No. 1 Radio Mechanics School, was appointed Staff Officer

(Wireless) to the Signal Officer in Chief, G.H.Q., Home Forces. He later held a similar appointment with S.H.A.E.F.





W. P. Rowley

E. A. W. Spreadbury

E. A. W. Spreadbury, M.Brit.I.R.E., who, as announced briefly last month, has succeeded E. M. Lee, of Belling and Lee, as chairman of the Radio Trades Examination Board, has been associated with the work of the Board since 1943 when he was appointed an examiner and a member of the examinations committee. Mr. Spreadbury represents the Brit.I.R.E. on the Board's council of management. Since 1941 he has been technical editor of Wireless & Electrical Trader, which he joined in 1937 after spending 14 years in the radio industry.

Air Commodore W. E. G. Mann, C.B., C.B.E., D.F.C., M.I.E.E., R.A.F. (Ret.), Director-General of Navigational Services in the Ministry of Transport and Civil Aviation since 1950, recently retired. At one time during the war he was Chief Signals Officer, R.A.F., Middle East, and since joining the Ministry in 1945 as Senior Signals Officer and U.K. Representative, Middle East, had held several administrative telecommunications posts.

R. H. Vivian, B.Sc., A.M.I.E.E., has joined Wireless Telephone Company, Ltd. as chief engineer. For many years he was development engineer in charge of transistor investigations and applications with A. C. Cossor Ltd. and since 1957 has been resident consultant with Associated Industrial Consultants Ltd.







R. C. Parry

Roy C. Parry has joined Mullard Ltd. to take over the duties of Government Liaison Officer, covering valves, tubes, semiconductors and components. He was previously with Marconi's W/T Co., where he led a section engaged in the design of naval radar equipment. Before joining Marconi's he served for a number of years in the Royal Navy.

Consequent upon the setting up of five new divisions under the reorganization of the A.E.I. group (see January, page 16), a number of appointments have been made. Four of the new divisions—telecommunications, radio and electronic components, cable and construction—are managed by A.E.I. (Woolwich) Ltd., previously Siemens Edison Swan. The board of directors of A.E.I. (Woolwich) includes Lord Chandos (chairman), Dr. J. N. Aldington (group managing director), Dr. T. E. Allibone (director of research), L. S. Crutch (director engineering, telecommunications). G. W. Giffin and J. T. Thornhill (directors of manufacture), B. A. Hensler (director, export), J. W. Ridgeway (commercial director), A. Whitaker (director of engineering, radio), S. E. Goodall (director of engineering, cable), R. L. Basset and A. F. Street. Except for the resignation of Sir Alexander Sim and the inclusion of S. E. Goodall (formerly chief engineer of Henley's Telegraph Works), the board is unchanged from that of Siemens Edison Swan. In the Radio & Electronic Components Division C. C. McCallum, A.M.I.E.E., becomes general manger. He joined Metropolitan-Vickers radio department in 1928 and two years later transferred to Edison Swan's





A. Whitaker

C. C. McCallum

radio valve department and was for ten years service department manager at the Cosmos works, Brimsdown. A. G. Everett, divisional manufacturing manager, has been with the organization since 1920 when he joined Metro-Vick as a college apprentice. He was appointed a director of Edison Swan Electrical Co. in 1948. J. Donegan, B.Sc., A.C.G.I., D.I.C., engineering manager (development), was a valve development engineer, becoming chief engineer valves and c.r. tubes on the merger of Siemens and Edison Swan. He is vice-chairman of the engineering advisory committee of the British Radio Valve Manufacturers' Association. C. L. Hirshman, A.C.G.I., D.I.C., now engineering manager (consultative), has been at the organization's Brimsdown applications laboratories since 1931. The divisional sales manager is P. V. Lister.

The following appointments are announced in the Telecommunications Division: W. G. Patterson (general manager); J. M. Wilcox (manufacturing manager); F. G. Pheazey (chief engineer); T. J. Scudder (production engineering manager); and D. J. Green (commercial manager). In the Cable Division the general managers are J. S. A. Bunting and E. J. Vidler; manufacturing managers, V. L. J. Plascott and S. J. Wilson; chief engineers, W. G. Hawley and J. H. Savage; sales manager, H. D. Parsons; and commercial manager, F. V. Vaissiere. In the Construction Division, A. V. Burnett is general manager; W. Sim, chief engineer; and L. F. Capeling, commercial manager.

B. C. Cook has joined Wolsey Electronics Ltd. as technical consultant Vision Network Systems. He has for many years been with Belchers Ltd.

T. Kilburn, D.Sc., Ph.D., M.A., M.I.E.E., reader in electronics at Manchester University, has been appointed to the newly created post of Professor of Computer Engineering in the University. Dr. Kilburn, who was at the R.R.E., Malvern, from 1942 to 1946, co-operated with Professor F. C. Williams, head of the electrical engineering laboratories, Manchester University, in building in 1948 the University's first computer, on which was based the Ferranti Mark I. He later worked on the protetype of the Mercury computer and since 1957 has been engaged on the transistorized computer to be known as Muse.

Harry Cartwright, M.B.E., M.A., A.M.I.E.E., the new Director of Industrial Power in the Atomic Energy Authority's development and engineering group at Risley, Derbyshire, was with the English Electric Co, immediately prior to joining the Atomic Energy Division of the Ministry of Supply (forerunner of the A.E.A.) in 1949 and was previously with the Decca Navigator Co. Mr. Cartwright, who is 40, took a first clast honours degree in the mechanical sciences tripos at St. John's College, Cambridge, in 1940 after which he was a signals officer at R.A.F. ground radar stations. He has successively been chief engineer and deputy director of the group of which he is now appointed director.

OUR AUTHORS

L. H. Bedford, C.B.E., M.A., B.Sc.(Eng.), F.C.G.I., F.I.R.E., M.I.E.E., M.Brit.I.R.E., director of engineering at the Guided Weapons Division of the English Electric Co., writes in this issue on an aspect of one of his hobbies—tape recording. Mr. Bedford, whose name is associated with the elevation attachment which he produced for the early gun-laying radar equipment, started his industrial career with the Western Electric Co. and spent some months at the Bell Telephone Laboratories in America. He joined A. C. Cossor in 1931 to initiate their development and manufacture of cathode-ray tubes. He stayed with that company until 1947 when he joined the English Electric group as chief television engineer of Marconi's W/T Co. Mr. Bedford was president of the Brit.I.R.E. from 1948 to 1950 and was appointed a member of the technical sub-committee of the Government's Television Advisory Committee in 1955.

Ralph Brewer, who was for some years in charge of the valve life-testing department of the G.E.C. Research Laboratories, is now concerned with the study of the survival characteristics of transistors and related semi-conductors, and contributes an article on this subject on page 108. Mr. Brewer, who is 45, joined the Research Laboratories in 1937 and during the early part of the war worked on the development of magnetrons. He received an award for his paper on life testing of valves read at the 1958 National Symposium on Reliability and Quality Control in Electronics in the United States.

R. C. V. Macario, B.Sc., Ph.D., Grad.I.E.E., joint author of the article in this issue on transformerless broadcast receivers, graduated in 1953 at King's College, London, where in 1956 he completed post-graduate studies concerned with the propagation of very low-frequency radio waves. Subsequently he spent some time at the A.E.I. Research Laboratories on semiconductor devices, and in 1958 joined the Plessey Company, where he is working on solid-state magnetic and dielectric devices.

N. E. Broadberry, Grad.Brit.I.R.E., who with Dr. Macario contributes the article on p. 110, attended the Institute of Science and Technology, Dublin, specializing on receiver design and after spending a year with Pye (Ireland) and nearly three years as a civilian radio and radar specialist on a military aerodrome, joined Murphy Radio (Ireland) for three years. In 1953 he became a lecturer at his former college. He joined Pye (Cambridge) in 1955, and last year was appointed a senior electronic research engineer with Plessey.

Physical Society's Exhibition

NEW TECHNIQUES IN ELECTRONICS AND MEASUREMENT

HERE are very few fields today in which electronics does not play some part: it has seemed even as if electronics were becoming a hydra-headed monster, judging by the complexity of some sections of the art. However, following last year's trend, many of the exhibits suggested by their ingenious simplicity and application of first principles that the monster is being changed into a well-mannered pet. A selection of these items is described in the following report.

The exhibition itself was even more crowded than it was last year and the organizers were unable to accommodate thirteen

would-be exhibitors who had applied for space.

INDUSTRIAL ELECTRONICS

Electrical Power from Heat, without the usual intermediate mechanical stage, was the theme of a display on the stand of the United Kingdom Atomic Energy Authority.

Two working demonstrations were shown—a thermionic method using a diode, and a semiconductor thermojunction heated by steam. The diode used a bright-emitter tungsten filament and a caesium-vapour filling -this latter neutralizes the emissioninhibiting effect of the space charge which gathers round the cathode of an unsaturated diode-and produced a small but nevertheless useful current. Although the diode is primarily a d.c. generator it is possible that, by modulation of the gas discharge, a.c. may be produced. Of course, the power input to the demonstration model was far greater than its power output, but the eventual hope for this means of power generation lies in the use of hot fuel rods (at about 2,000°C) of an atomic pile as the cathodes, the physical construction being such that individual diodes are connected in series so generating high potentials. The thermojunction—really a modern version of the thermocouple—used bismuth telluride semiconducting elements. Again the power output was far smaller than the input, but this method may be economical for "mopping up" waste heat from an atomic reactor.

The magnetohydrodynamic method uses a jet of ionized gas to replace the moving conductors of a conventional generator and the power developed is "picked out" by electrodes placed in the gas flow. The practical application of this principle awaits the introduction of materials suitable for temperatures of about 2,000°C.

Associated with the display was a mechanical analogue (after Kaye of M.I.T.) of a diode valve. Delightfully clear in its presentation, a rotating ridged roller shot steel ball bearings up a plane with a double incline to represent work function and space charge. The balls with sufficient energy to reach the top of the plane ran down a shute (representing the external circuit) and rotated a small paddlewheel before returning to the "cathode." The paddlewheel formed part of a generator whose output deflected a galvanometer.

Radio Frequency Spectroscopy is employed for the routine analysis of free radicals (uncombined particles of the material)* in the Newport Instruments electron paramagnetic resonance spectrometer. As this instrument is designed for laboratory use it incorporates a large electromagnet and elaborate stabilization arrangements for the power supply to ensure a uniform (1 part in 104), constant and controllable field of between 150 and 8,000 gauss is achieved. The field is adjusted to bring the resonance being studied to 9.5Gc/s (9,500Mc/s) and elaborate precautions are taken in the waveguide assembly to ensure that accurate results are obtained-for instance, the klystron local oscillator is mounted in a constant-temperature oil bath. In operation the field is varied either at l.f. by coils on the pole pieces or at r.f. (100kc/s) by a single-turn coil round the specimen. The klystron output is thus modulated as the resonance in the sample passes through the local-oscillator frequency: but, to improve accuracy over the detection of a change of amplitude the local oscillator and "returned from sample" signals are mixed in a waveguide bridge and their relative phase is detected.

Microanalysis of Metallic Alloys by a technique originally developed in France by R. Castaing and applied by the Cambridge Instrument Company uses a 10-6 m.-diameter electron spot to scan a 0.5 × 10-3 m. square area of the sample. The beam is produced by an ordinary triode gun and focused by a double magnetic lens in the manner of an electron microscope. The etched







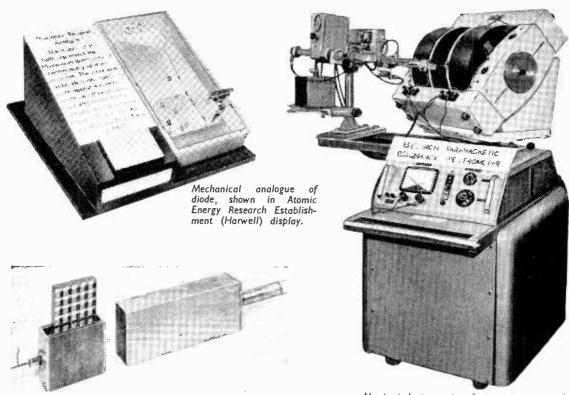






Images obtained with Cambridge Instruments Electron-probe X-ray Microanalyser during preliminary investigation of copper(70%)—nickel-tin-iron (10% each) alloy. L to R: optical, electron and X-ray(iron, nickel, copper, tin) images.

^{*} For a fuller account of electron spin and nuclear magnetic resonances see Wireless World. p. 68, February, 1960.



"Exploded" view of Penning ion pump (Mullard Ltd.).

Newport Instruments electron paramagnetic resonance spectrometer, showing magnet and power supplies; also wave-guide bridge.

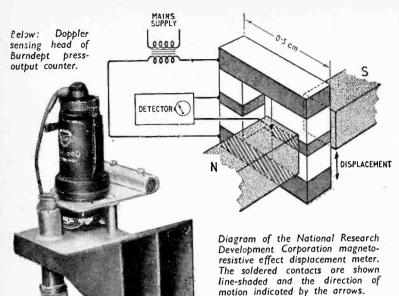
surface of the sample reflects some electrons and these cause the emission of light from a phosphor mounted near the sample; a photomultiplier produces from the light a signal which modulates the beam current of a cathode-ray tube, scanned in synchronism with the analyser beam. Thus an "electron picture" of the sample is built up. X-rays produced by the electrons that penetrate the surface of the sample strike an analysing crystal (gypsum or lithium fluoride) which "bends" them into a proportional-counter detector. The electrical signal from this is similarly used to produce an X-ray microgram. The wavelength of the Xrays produced by different metals varies, so that the angle through which the spectrometer crystal bends them also varies: thus by rotating the crystal the distribution of different metals in an alloy can be seen.

High-vacuum Maintenance is one of the problems associated with many electron devices and, whilst many ingenious approaches have been devised, few have the simplicity of the Penning ion pump. "Pump" is, perhaps, a misnomer, for the Penning gauge and the pump derived

from it bear little resemblance to any of the commoner evacuating devices. The basic gauge is a coldcathode discharge tube and the small current which flows under high potential is a measure of the pressure. The principle of operation of the pump is similar: a magnetic field causes electrons to take up a spiral path between the electrodes. This increases the volume swept out and hence the chance of ionization. Sputtering of the titanium electrodes occurs, so causing the titanium to act as a continuous getter. Two firms, Mullard and Edwards, were showing pumps based on the Penning-gauge principle. The Edwards pump was cylindrical in form with two cathode discs on either side of a wire-ring anode. A 500-gauss field and 3 to 5kV are applied, the current ranging from 0 to 15mA and depending naturally on the degree of ionization. The Mullard pumps use a horseshoe magnet producing some 1,000 gauss and the anode is made up as a rectangular honeycomb structure; also a small getter filament is included. Pumping speeds of 1 to 5 litres/sec and pressures down to 10⁻⁹ torr (mm of mercury) have been reached with Penning pumps.

Microscopic Measurements made by the use of a calibrated vertical traverse on the microscope are limited in accuracy to far below the maximum possible resolution of the microscope. However, Wayne Kerr have developed a technique using their B721 Electronic Micrometer which is a transformer-ratio bridge capable of measuring minute changes of capacitance. The microscope is fitted with a Baker interference objective (which makes exact focusing easy) and a non-contacting probe connected to the micrometer. In use the microscope is focused in turn on to the points representing the top and bottom of the "depth" to be measured, and the distance of the probe from the specimen measured by the micrometer. Repeated measurements of the depth of the etching of photogravure printing plates gave results with a standard deviation of 5×10^{-6} in on a $2 \times$ 10⁻³in mean.

Displacement Measurement—Nearly all semiconductors increase in resistance when they are placed in a magnetic field. The use of this magnetoresistive effect, as it is called, to detect displacements down



to about 10-scm was demonstrated by the National Research Development Corporation. The semiconductor used is indium antimonide (InSb), since in this the effect is relatively large. Four contacts are soldered on to InSb elements so that the resistances between these contacts form a Wheatstone bridge. Following normal practice, this bridge is energized from an a.c. supply across one pair of opposite contacts, and any out-of-balance signal detected between the other pair of opposite contacts. Two of the InSb resistors are arranged to be parttially in a 10,000-gauss field produced by a permanent magnet, so that relative motion between the InSb and magnet in one direction moves one of these resistors further into the field and the other resistor further out of the field. The magnetoresistive effect then increases the value of the resistor which moves further into the field and decreases the value of the resistor which moves further out of the field. This unbalances the bridge and the out-ofbalance signal gives a measure of the displacement of the InSb relative to the magnet.

Strain Gauges take many forms, but an unusual type was an "acoustic" gauge shown on the Acoustics Group stand. This consists of a length of ferromagnetic wire stretched between two massive supports on the body under test. Placed near the wire at its centre are two

coils, one connected to the input of an amplifier and the other to the output. The connections are phased so that positive feedback takes place which results in oscillation at the resonant frequency of the wire. If the strain on the wire is altered, its frequency alters: thus the strain is given by measuring the frequency of oscillation.

Tachometer-A method of measuring rates of rotation shown by the Research Development National Corporation utilizes the magnetic field induced when a conducting cylinder rotates in a magnetic field. This induced field is at right angles to the original field, so it may be distinguished from the original field and measured by a detector, such as the fluxgate used, which measures only the component of the total field in the direction of the induced field. This method can be used to measure rotation rates as slow as 0-1 r.p.m.

A Doppler-radar Sensing Head was demonstrated with the Burndept BE250 press-output counter. The output from a small 3-cm klystron is fed into half of a waveguide horn radiator: in the other half is a mixer crystal. When a moving object passes the horn the radiation reflected from it beats with the klystron frequency at the mixer crystal, producing an a.f. pulse at the Doppler frequency. The sensitivity is such that the movement of a 2-BA washer one foot away can trigger the counter. Although this may sound an expensive method, it should overcome most, if not all, of the disadvantages of cutting beams of light or producing pulses in coils as oil and dirt should not affect it, and it will detect any object giving a reflection.

Ultrasonics are playing an increasing part in industrial measurement and test. An example of the measurement of fluid velocity in a pipe was shown on the B.S.I.R.A. stand. In this method two transducers are mounted in line along the outside of the pipe so that the beam transmitted by one is reflected from the opposite pipe wall and received by the second. As the propagation speed depends on the flow speed, the time difference between pulses travelling in each direction will represent the flow velocity. A novel feature of the arrangement is that, as both transducers operate simultaneously, the "forward" and "backward" waves travel the same path, thus eliminating any temperature effects. Slightly different frequencies centred on 5Mc/s are used so that the two sets of information may be separated and a phase-sensitive detector (operating at a low frequency) triggers a flip-flop, or bistable, circuit which generates pulses whose area is thus equivalent to flow velocity. A moving-coil meter indicates the average area under the pulses and gives a direct flow reading.

MEDICAL ELECTRONICS

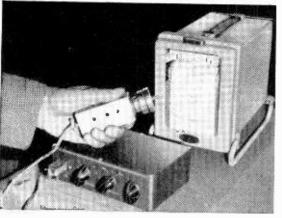
Medical Manometry.—The first British "radio pill," soon to be manufactured in quantity, was shown by the Medical Research Council. It is a subminiature tadio transmitter, encased in a cylinder L9cm long and 0.8cm in diameter, which signals pressure (or temperature) values during its passage through the gastrointestinal tract. A 400-kc/s transistor oscillator using a Gouriet circuit is modulated in frequency by the pressure transducer. This con-

sists of a pot-cored inductor, forming the tuning inductance of the oscillator, with its magnetic circuit including a small ferrite disc which is attached to a flexible diaphragm. Pressure variations applied to the diaphragm therefore change the inductance and vary the frequency of the oscillator. The whole assembly, including the sealed Mallory battery, is completely potted, and the oscil-

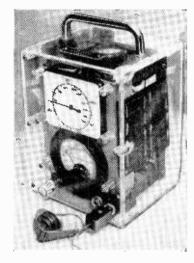
(Continued on page 121)

WIRELESS WORLD, MARCH 1960

Right: Capacitancetype transistor pressure gauge (Medical Research Council).



Below: Photocellpowered medical pulse meter shown by University of Edinburgh.



University. Pressure is applied to one arm of a flexible U-tube containing liquid. Increasing pressure therefore causes the level of the liquid in the other arm (which has a transparent glass section) to rise. By raising this transparent arm, however, the liquid in it can be brought back to its original level—the amount of movement required being proportional to the increased pressure. The object of the follower system is to detect the initial movement of the

liquid level from a reference position in the transparent arm, then to raise or lower the arm, by means of a carriage and rail system, until the level returns to the reference position. Direction of liquid movement is detected by one phototransistor in the carriage, while two others, and their associated circuitry, respond to acceleration in either direction. Appropriate signals control a motor which drives the carriage up or down in a correcting sense, and the movements of the carriage are transferred by a cord system to a pen on a recording chart which indicates the pressure.

Edinburgh University also displayed a transistorized medical pulse meter, using a carbon resistance transducer for strapping to the thumb, which was made portable and reliable by the use of light energy to provide the power via rechargeable sealed cells. A bank of eight selenium cells in series gives the charging current for the battery, which will operate the pulse meter (1-2mA drain) for 50 hours without recharging. A diode prevents discharge of the battery through the selenium cells when the instrument is left in the dark.

lator is switched on by giving a shake to the "pill" to operate an inertia mercury switch. The maximum frequency deviation is 35kc/s. At the receiver, a superheterodyne system provides an output signal which varies between 0 and 35kc/s. The cycles of this are converted into square pulses of regular width and amplitude, which are passed to a pen recorder. The integrating effect of the pen recorder mechanism on the pulses then provides a pen deflection which is proportional to their frequency, and hence to the pressure.

Another transistorized device on the same stand was a miniature capacitance-type pressure gauge using a diaphragm of aluminized Terylene as the pressure-sensitive element. It was notable for having the frequency-modulated oscillator and discriminator all built into the body of the gauge, thereby reducing drift and avoiding trailing r.f. cables.

Reduction of drift was also the object of a recording manometer based on a servo or follower principle which was shown by Edinburgh

TEST AND MEASURING GEAR

Sine-Wave Generators. Several new transistor oscillator circuits were shown by the Royal Radar Establishment. Two of these utilized socalled class-D push-pull operation. This combines the half-cycle current flow in each transistor of class-B operation, with the nearly zero voltage drop across the transistor during the whole of current-flow period of class C. This retains the advantages of both class-B and class-C operation without their respective disadvantages of finite average voltage drop during current flow and less than half current on/off ratio. The nearly zero voltage drop across the transistor results in a high efficiency, and the half-cycle conduction period in relatively low distortion. The transistors are driven by a current switched alternately into one base or the other. One of the class-D circuits shown was a tape erase and bias oscillator. This used two OC24's to provide 3.5 W output at 35kc/s with a second harmonic distortion of less than 0.1%.

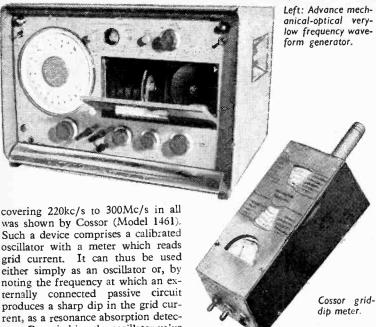
Another new transistor circuit shown by R.R.E. is the equivalent of a well-known valve circuit often used to provide very-low frequency oscillations. This circuit consists of a

phase invertor followed by two similar virtual-earth Miller integrators. This provides a total phase shift of 360° at one frequency, so that when the output is fed back to the input oscillations at this frequency are produced. The output from the first integrator is limited and used to provide positive feedback at the phaseinvertor input so as to stabilize the oscillation amplitude. Since the 360° phase shift and consequent oscillations are only produced at one frequency, the fact that the positive feedback consists of a limited and thus distorted sine wave does not increase the oscillation distortion.

A transistor b.f.o. was also shown by R.R.E. Here a 100kc/s oscillator is switched between a load and output at a frequency variable between 80 and 100kc/s. This produces a b.f.o. in the output which is variable in frequency up to 20kc/s. Since all spurious frequencies produced by the switching are at least as high as 80kc/s, they may be readily removed.

An unusual feature of the Solartron DO905 is that the cable supplied has no effect on the output level, since level-stabilizing feedback is taken from the end of this cable.

A grid-dip meter with plug-in coils



was shown by Cossor (Model 1461). Such a device comprises a calibrated oscillator with a meter which reads either simply as an oscillator or, by noting the frequency at which an externally connected passive circuit produces a sharp dip in the grid current, as a resonance absorption detector. By switching the oscillator valve to act as a diode, it can also be used as an absorption meter by determining the frequency at which a peak in the meter current is produced. The Dawe Type 1208, when fed

with a sinusoidal signal in the frequency range from 5c/s to 5kc/s, produces a frequency differing from the input by an amount which can be fixed between 0.5 and 2.5c/s. This fixed difference-frequency output is for feeding a stroboscopic lamp so as to produce a slow-motion effect with objects vibrating at the original input frequency. The input is fed to two groups of phase shifters, each of which consists of a passive CR network connected between the anode and cathode of a concertina phase splitter. The first group contains four such phase shifters connected in series and tuned to produce 45° phase shift at 5, 50, 500 and 5,000 cycles respectively: the second group contains three such phase shifters in series tuned to the geometric means of these frequencies, i.e. 15.8, 158 and The phase difference between the outputs from the two groups of phase shifters is then within 5° of 90° for any input frequency between 5c/s and 5kc/s. By the use of two additional phase invertors, four outputs spaced at 90° phase intervals can then be obtained for any input frequency between 5c/s and 5kc/s. These outputs are fed to four equally-spaced tappings on a circular potentiometer whose wiper rotates at the required difference frequency. The output between the wiper of this potentiometer and any of the four tappings then has a phase difference with respect to the original frequency which increases at a constant rate of 360° per potentiometer wiper revolution period. This variable-phase output can be shown to be equivalent to a constant phase output differing from the input frequency by the rate at which the potentiometer wiper rotates.

Pulse Group Generator-To select and observe special types of nuclear particles or radiation, elaborate types of pulse groups produced by these particles or radiation may have to be dealt with by the apparatus used. Complex pulse groups may also be needed in physiological Specialized pulse generators are thus required for testing such nuclear and physiological equipment.

An example of such a specialized generator shown by Nagard, their Type 5101, produces groups of pulses in which the group length, and group and individual pulse repetition rates are all independently variable, subject to the proviso that, for convenience, only a whole number of pulses is produced in each group. A variable-frequency multivibrator triggers a variable-length flip-flop to provide the variable-rate and length group pulse. This group pulse is specially shaped and fed to trigger on and off a second variable-frequency multivibrator working at the individual pulse rate: the group pulse shaping being such as to only trigger on and off the individual-pulse

generator at the beginning and end of a pulse.

Function Generator-The new Advance Type SG88 uses a mechanical-optical system to repeat any required waveform at any rate between 0.005 and 50 times per second. A rotating transparent disc is partially blacked out so that the remaining transparent area corresponds to the waveform to be generated expressed in polar co-ordinates. The disc is illuminated, and the light transmitted through the disc and a fixed radial slit is focused by a cylindrical lens on to a phototransistor. As the disc rotates, the fixed slit thus scans the waveform to be produced so that the output of the phototransistor reproduces this waveform. The rise time of the device is limited by the slit thickness and varies from 300msec at 0.005 waveforms/sec to 0.3msec at 50 waveforms/sec. An integral triggering device can be used to decrease these rise times to less than 3µsec for square waves at all repetition rates.

Noise Generator-A simple device for producing a broad band of noise centred on 1,500c/s for acoustic calibration purposes was shown by Dawe. This Type 1417 generator contains about 6,800, 0.06-in diameter, steel ball bearings. These fall through an aperture onto an anvil which deflects them on to a vertical mica plate where they produce the required noise output. About three-quarters of the bearings strike the mica in the useful measurement period of about 5 seconds. The noise is 3dB down at 750 and 3,000c/s and its intensity 90 + 1dB at a distance of 8 cm from the mica.

Acoustic Measurement-A method of finding the radiation impedance on a rigid piston was shown by the Admiralty Research Laboratory. An acoustically simple source is used consisting of a longitudinally excited piezoelectric crystal cemented to a glass head. Except for the end face of the glass, this source is covered by a watertight container. The whole source is immersed in water or any other medium with a high acoustic resistance but low viscosity. All parts of the source then work into air except for the face of the glass in contact with the water. The only significant acoustic loading is thus that produced by the water on this The radiation reactance on this face can then be obtained from the change in the resonant frequency of the source when it is immersed in water, and the radiation resistance from the Q of the resonance in water.

Magnetic Measurements — The measurement of susceptibilities down to about 5×10-7 c.g.s. units (about that of pure water) was demonstrated by the National Research Development Corporation. An extra softiron pole piece containing an indium-antimonide Hall-effect magnetic-field detector is placed between the two normal soft-iron pole pieces of a permanent magnet. If the magnetic reluctances of the two gaps so formed between the extra pole piece and the two normal pole pieces are equal, no flux will be produced in the central pole piece containing the magnetic-field detector. If then a sample is introduced into one of the gaps, the reluctance of this gap is altered, the two gap reluctances become unequal, and a magnetic flux is produced in the pole piece containing the magnetic field detector. The detector output can then be shown to be proportional to the susceptibility of the sample.

A vibrating reed magnetometer for measuring the saturation magnetization of ferromagnetic materials was shown by the G.E.C. Research Laboratories. A sample of the material is attached to one end of a flat non-magnetic reed, the other end of which is rigidly clamped. The sample is placed in a non-uniform magnetic field across the width of the reed. The force on the saturated sample due to the non-uniform magnetic field is then proportional to the product of the magnetic moment of the sample and the rate of change with distance of the field strength, and this force is in the direction of the stronger field. With hemispherical pole pieces, the non-uniformity is such that this rate of change is proportional to the displacement of the sample from the central position. This adds to the force produced on displacement by the elasticity of the reed so as to alter the reed's natural resonant frequency. Attached to the reed are two barium titanate transducers; one is driven by a variable frequency oscillator so as to cause the reed to vibrate, and an increase in output from the other is used to indicate the reed's resonant frequency. The change in resonant frequency with and without the magnetic field is proportional to the saturation magnetization of the sample. The instrument is calibrated using a nickel sample for which saturation magnetization is The measureaccurately known. ment can be shown to be unaffected by the size of the sample apart from a small error produced by the slight curvature of the field lines. By applying the field across the reed width, any attraction of the sample to one pole piece is counteracted by the much greater stiffness of the reed to motion in the direction of this attraction.

Microwave Measurements-An automatic complex reflection-coefficient plotter was shown by the G.E.C. Research Laboratories. Three probes spaced equally along a guide between 1/8 and 3/16 of a guide wavelength apart each feed a squarelaw bolometer detector. It can then be shown that if the differences between the outer probe detector outputs and between the middle and the mean of the two outer probe detector outputs are fed to the Y and X plates of an oscilloscope respectively, the c.r.t. spot is deflected to a point corresponding to the complex reflection coefficient of the impedance at the end of the guide. Because of the restriction on the spacing between the probes to between 1/8 and 3/16 of a guide wavelength, three fixed probes can only be used over a ±

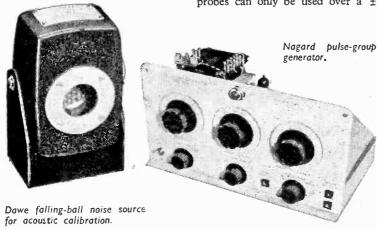
20% range of guide wavelengths. There are, however, no inherent frequency-range limitations in the display unit

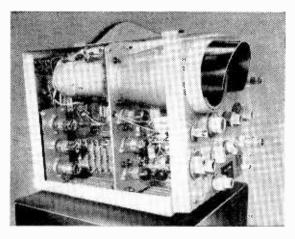
An exhibit of the A.E.I. Research Laboratories showed how large frequency differences (e.g. 1kMc/s in 35kMc/s) could be measured to a high resolution (e.g. 1Mc/s). The signal input feeds a section of waveguide which contains three phase detectors spaced at 120° intervals and which is terminated by a long length of short-circuited delay line. Any slight change in wavelength of the standing-wave pattern is added up in the long length of delay line to produce a considerable shift in the pattern, and this shift is measured by the three detectors. The three detector d.c. outputs are amplified and fed one to each of the three coils of a desyn. It can then be shown that the angular rotation of the desyn magnet pointer is proportional to the product of the frequency change and delay line length. By attaching a revolution counter to this pointer, large frequency changes can be measured.

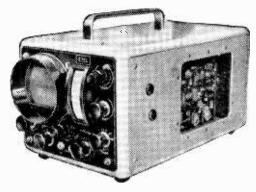
Frequency Meters-An often-used type of circuit first transforms the input to a square wave which is then used to pass a fixed charge every cycle into a capacitor. The charge current, being proportional to the input frequency and independent of the input waveform shape, thus gives a measure of the frequency. Both the Venner Type TSA501 and the Solartron "f-C" meter used this basic type of circuit. The latter instrument can also measure capacity by means of an internally-generated input of known frequency, since the charge current is also proportional to the capacity.

Phase Meter—In the Dawe transistorized Type 630, the two signals whose phase difference is to be measured are limited, the resulting square waves differentiated, and the positive going pulses so produced used to trigger on and off a bistable circuit. The mean d.c. output of this circuit is indicated by a meter. This output is proportional to the "on" over "on-plus-off" current ratio of the bi-stable circuit and this is equal to the phase difference required divided by 360°.

Incidentally, Ekco, for transistor convertors, have adopted a simple method of obtaining three phases from two single-phase supplies. Known as a Scott transformer, it entails only the connection of one supply to the centre tap of the other. As long as the phase difference be-







Solartron portable double-beam oscilloscobe.

Electronic Tubes transistor oscil-loscope designed to illustrate the use of their CR132 cathode-ray tube.

tween the single-phase supplies is held at 90° and the correct voltage ratio is achieved, a balanced, stable three-phase supply is developed between the free ends of the singlephase windings.

Logarithmic Amplifier-In the Solartron model TA965, the logarithmic characteristic is obtained simply by making use of the logarithmic forward voltage/current characteristic of a thermionic diode under lowcurrent temperature-limited conditions. The diode is used as the anode-to-grid feedback resistance in a virtual-earth type of amplifier. Since the gain of this amplifier is proportional to this feedback resistance, this gain is also proportional to the logarithm of the input current.

Clip-on D.C. Milliameter-In a transistorized instrument shown by Solartron, their model AM1002, the magnetic field produced by the current to be measured is collected by a Mumetal clip round the wire and measured using a Hall-effect device. This device, together with as much as possible of the magnetic clip circuit, is magnetically shielded from stray fields. The effect of the earth's field near the current-carrying conductor is cancelled out by extending the magnetic circuit on the other side of the Hall-effect device away from the current carrying conductor in such a way that the earth's field there induces an equal cancelling field in the Hall-effect device. The Halleffect device is energized at 40kc/s and the output amplified, rectified and indicated on a meter. A fraction of the rectified output is fed back to coils wound on the clip so as to produce a field opposing the detected field. This feedback linearizes the instrument against changes in the reluctance of the movable

parts of the magnetic clip circuit. The feedback coils are wound on copper cylinders to provide shortcircuited turns which greatly reduce the inductive reaction on the circuit being measured. The instrument is calibrated by means of a standard current.

Oscilloscopes—For illustrating the use of their CR132 c.r.t., a transistor oscilloscope was shown by Electronic Tubes. This uses 21 transistors to achieve a sensitivity of 75mV/cm from d.c. to 20Mc/s (3dB down), and an input impedance of $250k\Omega$. The total power consumption is 2.5W, of which the c.r.t. heater uses 1W. To achieve the relatively-high deflection voltages required, two transistors are used in series across the supply in what was termed a "beanstalk" circuit.

A new range of portable test and measuring instruments in attractive two-tone pastel colours introduced by Solartron included a double-beam oscilloscope (Model CD1014). This has a bandwidth from d.c. to 5Mc/s (3dB down) at a sensitivity of

100mV/cm. Time and voltage measurement to within at least ± 5% are possible.

A two-channel oscilloscope introduced by Nagard, their Model 311, has a sensitivity as high as 100 μV/ cm with a bandwidth from d.c. to 250kc/s (3dB down). An internal square-wave generator with nine alternative outputs allows voltages to be measured to within \pm 2%. Time intervals can be measured to the same accuracy using the calibrated sweep speeds available.

By changing to a slightly different type of storage tube like that described in our review of the 1958 Physical Society exhibition (May 1958 issue, p. 221), Cawkell have been able to increase the maximum writing speed on their Type S01 Remscope by a factor of about 20. The trace persistance time can also now be varied between one second and two minutes, and the trace intensity varied to allow the reproduction of half tones. The new tube is also much more difficult to damage by operating the oscilloscope incorrectly.

COMPUTERS

Piezoelectric Multiplier, a small and relatively simple device for analogue computing and other uses, was presented as an alternative to the Halleffect multiplier by Imperial College on the N.R.D.C. stand. The device produces a voltage which is proportional to the product of two input currents, by using the currents to energize electromagnetic actuators which apply mechanical forces to the piezoelectric crystal. The currents, x and y, are applied to the coils of the actuator as shown in the diagram, so that the m.m.f. in one of the C-shaped cores is proportional to

x+y and the m.m.f. in the other core is proportional to x-y. The resulting forces applied to the crystal (a bender type) through the soft-iron armature are proportional to $(x+y)^2$ and $(x-y)^2$ respectively. Since these two forces are applied so as to oppose each other, the actual force on the armature and crystal is their difference, proportional to $(x+y)^2 - (x-y)^2$. On the basis of the well-known rela- $(x+y)^2 - (x-y)^2 = 4xy$, this difference force produces an output voltage from the crystal proportional

(Continued on page 125)

Wireless World, March 1960

to xy. A displacement of the armature of less than one micron gives an output of about 1 volt. The output for x=y=5mA is 250mA. Almost perfect linearity in terms of either x or y is claimed, while the response is said to be independent of input-current frequency between 15c/s and 500c/s. Extension of the frequency range up to 50kc/s is thought to be possible, giving up to 10^s analogue multiplications per second with an accuracy of better than 0.5%.

Magnetic Thin-Film Stores for digital computers based on materials with rectangular hysteresis loops have been under development for some time (see W.W., July/August issue, p. 312) because of the possibilities they offer of high switching speeds and high-density storage of binary digits. Most of the experimental stores have used arrays of separate spots of magnetic film, each capable of being switched from one direction of magnetization to the other; but the Royal Radar Establishment were showing that the basic principle could also be applied to a continuous film of material. Their store consisted of an evaporated film of nickel-iron, 1200 Ångstroms thick, surrounded by an array of one-turn driving coils. The film has a preferred direction of magnetization (obtained by applying a magnetic field during deposition) and binary digits are stored at individual positions by the driving coils causing local reversals of magnetization. These reversals take place by rotation of the direction of magnetization in the plane of the material like a compass needle. The fact that a selected area of film could be switched with-

OUTPUT VOLTAGE XY

PIEZO
CRYSTAL

SOFT IRON
ARMATURE

X Y Y Y CURRENTS

Imperial College piezoelectric multiplier (shown on N.R.D.C. stand)

out affecting neighbouring areas was demonstrated by photographs in which domain walls had been made visible after local reversals of magnetization. Switching speeds are capable of dealing with pulses of 5mmsec duration, and a storage density of over 200 binary storage "cells" on a 3in×3in film is possible. Additional colls at right angles to the drive colls are needed to provide "bias" fields so that the initial direction of magnetization is

slightly displaced angularly from the preferred direction of the material. Currents applied to the drive coils can then rotate this "vector" to either of two directions, thereby making possible binary storage in each "cell" formed at the intersection of a drive conductor and a "bias" conductor. Mullard also showed work in this field, using separate spots of nickel-iron film and conductors formed by printed-circuit flexible layers.

COMMUNICATIONS

Hall-effect Modulator working at 5Mc/s was a feature on the A.E.I. (Metrovick) stand. Care in the alignment of the leads to the crystal, and their arrangement, helped to reduce the carrier break-through in the double-sideband suppressed-carrier modulator but a correcting coil, carrying d.c. bias and wound on the ferrite-core, allowed a fine adjustment to compensate for errors in the positioning of the connections to the crystal, resulting in a final rejection ratio of 60dB. Associated with the modulator was a wide-band transformer to match the low impedance of the indium arsenide crystal, the whole covering 100kc/s to 20Mc/s at the -3dB points. The demonstration included a c.r.o. display of the modulated output: the bow-tie pattern typical of suppressedcarrier modulation displays was produced with the good linearity expected of a Hall device.

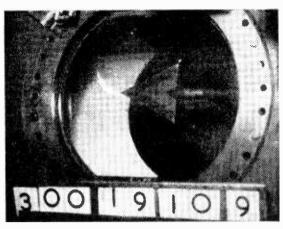
Fast-wave Electron-Beam Parametric Amplifiers.—Since in the wave of space-charge density modulation of

an electron beam which travels faster than the beam itself, the fast-wave, the mean energy is greater than the d.c. energy of the beam, prospects are offered of absorbing the excess energy corresponding to noise density modulations in this wave. In the case of the more commonly used slow-wave, the mean energy is less than the d.c. energy of the beam, so that noise could only be eliminated by adding equal noise energy to the beam, and this in practice would be impossible.

A fast-wave amplifier in which a longitudinal magnetic field causes the electrons in the beam to spiral from the inuput to the output, and in which the spiral amplitude corresponds to the signal strength was shown by English Electric (Type N1036). This valve is similar to that described in "Technical Notebook" in our November 1958 issue (p. 555). It has the valuable property of amplifying in one direction only so that feedback from the output to the input with consequent instability is eliminated. In addition, the tube protects the following stages, since it acts as a limiter for excessively large inputs, the maximum power output being 50 µW. A noise figure of 2dB at a signal frequency of 200Mc/s is quoted for this valve. Only a relatively small magnetic field of 70 gauss is required and the highest direct potential in the tube is also only 100V. Beam currents and voltages are also very low, typical figures being 30 µA and 6V respectively. A 600Mc/s tube is in development.

The G.E.C. Research Laboratories demonstrated fast-wave amplification by passing a beam through two cavities into which the signal and pump frequencies were fed so as to modulate the beam, as in a klystron. A movable cavity tuned to the signal frequency was used to show the increase in the signal level along the beam

Noise Reduction in Backward-wave Oscillators.-Due to residual gas ions, backward-wave oscillators normally exhibit peaks of noise at sideband frequencies of the order of 1Mc/s. This noise restricts their use as local oscillators to receivers with an inconveniently high inter-However, an mediate frequency. exhibit of the Mullard Research Laboratories showed that this sideband noise can be considerably reduced either by reducing the residual gas pressure below its normal value between 10-7 and 10-8mm of mercury down to about 10-0mm, or alternatively by draining off the gas ions by means of a trans-



Untouched photograph of plume discharge from tip of earthed electrode in wind tunnel (R.A.E.).

Crown Copyright Photograph

verse electric field across the beam. This transverse field need only be of the order of 10V/cm.

Corona Discharge from radio aerials on aircraft at great heights is a serious problem because not only does it result in a power loss in the discharge, but also the discharge produces a serious mismatch at the aerial, further reducing radiated power and upsetting operation of the transmitter. A combined exhibit from R.A.E. and Sheffield University showed some of the approaches to the investigation of the problem. A $\lambda/4$ radiator on a ground plane was enclosed in an evacuated bell jar and an r.f. output applied. At first the discharge started, as would be expected, round the high-potential point at the tip of the radiator; but, above a certain power level, the discharge produced such a mismatch that it moved to the lower part of the aerial, so wasting more than half the transmitter power. One detection method used is to modulate the r.f. with square waves: mounted in the tip of the radiator is a small capacitance "microphone" whose resonant frequency corresponds to the modulating frequency. At the onset of corona the potential gradient changes sharply, altering the force on the diaphragm; this results in a change of output.

A display of photographs (taken at R.A.E.) of earthed electrodes in a wind tunnel revealed an interesting phenomenon. At high speeds (Mach 1.9), low pressures (the atmospheric) and in the presence of water vapour, a plume of corona discharge takes place against the air stream from a sharp point on the leading edge of an earthed electrode. Although still under investigation it is thought that this is due to the gathering up of charges from the air.

COMPONENTS

Microwave Components.—It was surprising to find three new components produced by novel methods rather than merely by the usual adaptation of existing methods to new frequency bands. These new components were shown by Mullard for the $2\frac{1}{2}$ mm band.

The first of these components, a variable attenuator, is restricted to use at such high frequencies. It consists of an intrinsic region of semiconductor material placed across the guide between two heavily doped p- and nregions outside the guide. By applying a forward voltage between the pand n-regions, electrons and holes are injected into the intrinsic region where they absorb a fraction of any incident microwave power. For correct operation of this device, the

intrinsic region cannot be made much wider than 10^{-2} cm, so that the new waveguide must be tapered even at $2\frac{1}{2}$ mm wavelengths.

Another of the new components shown by Mullard was a variable directional coupler in circular waveguide. A long strip, one end of which is fixed, is placed centrally in the guide. If the other end of the strip is rotated, incident radiation which is plane polarized at right angles to the initial plane of the strip has its plane of polarization gradually rotated as it passes by the strip. Depending on the angular rotation of strip end, a variable fraction of this incident radiation can then be coupled out via a probe parallel to the initial plane of the strip. Any reflected radiation, since it consists of the reflected portion of the incident radiation not coupled out by the probe, has its plane of polarization at right angles to the coupling probe and is thus not coupled to this probe. Attached to the movable end of the absorbing strip in the same plane as this end is a flat tapered absorbing vane. This absorbs the same fraction of the reflected radiation as is coupled out from the incident radiation. This vane is at right angles to the rotated plane of polarization of the incident radiation so that it does not absorb any of the incident radiation.

The third new component shown by Mullard was a variable impedance, also in circular waveguide. This consists of a flat tapered absorbing vane attached to a movable metal plunger filling the guide. The component of the incident radiation whose plane of polarization is parallel to the vane is totally absorbed by the vane, whereas the component at right angles to this vane is totally reflected by the plunger. Thus, by rotating the vane and moving the plunger along the guide, any required complex reflection coefficient can be obtained. A second fixed flat tapered absorbing vane perpendicular to the plane of the incident radiation absorbs any interfering reflected power polarized at right angles to the incident radiation.

A simple type of waveguide to coaxial directional coupler was shown by Decca. This consists of a piece of strip line inclined to the guide length near one narrow guide wall, the signal being coupled out from the strip at its end furthest from the incident radiation. The strip occupies a space of one quarter of a wavelength along the guide: its width and distance from the guide wall effect the match and coupling. This type of coupler can be made with a performance which compares favourably with that of a normal two-slot waveguide coupler.

A self-calibrating coaxial line wavemeter consisting of a movable short circuit which was shown by Flann illustrates how it is possible to obtain very wide bandwidths using coaxial line: in this case the bandwidth is from 800 to 9,000Mc/s. The distance moved by the short circuit between two adjacent resonance positions is half a line wavelength.

A slot-array aerial for X-band was shown by Elliott. This consisted of eight rows of eight slots in the single metal ground plate of a strip transmission line, the whole aerial being about seven inches square. The strip transmission line is divided up rather like a family tree to provide radiators opposite the slots. This type of aerial possesses a number of advan-

WIRELESS WORLD, MARCH 1960



20th Century image-intensifier tube. Note ring accelerator electrodes on inside of glass envelope.

tages over the normal paraboloid, for it is smaller and lighter, and the feed does not obstruct the aerial radiation. Such slot aerials can be easily reproduced since the strip line and slotted ground plate can be printed on dielectric slabs of the correct thickness to provide the required line-to-plate spacing.

V.H.F. Transistors operating at over 100Mc/s are now beyond the development stage and are available as commercial products. example, Texas Instruments demonstrated a diffused-base germanium transistor of "mesa" construction operating at 200Mc/s in a 50mW power output stage of a transmitter. Another "mesa" transistor was shown operating in a decade counter of 10Mc/s p.r.f., while other similar types on view, both germanium and silicon, were intended for amplification in the 100-150Mc/s region. Mullard have a v.h.f. transistor made by the alloy diffusion process with a cut-off frequency of 100Mc/s, but they also showed an experimental type of similar construction which would amplify at this frequency. It was demonstrated in a series of 100Mc/s amplifiers, each using two transistors in push-pull giving an output of 0.5W. A silicon "mesa" transistor with a cut-off frequency of 80Mc/s (in the common base con-

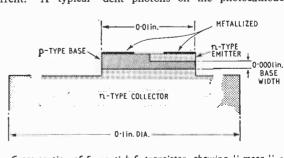
Transistor transmitter producing 50mW at 200Mc/s. Texas Instruments "mesa" transistor is used in output stage.

nection) represents the first essay of Ferranti's in the transistor field. It is an n-p-n type and the method of construction is shown in the diagram. A slice of the n-type silicon which subsequently forms the collector has a layer of p-type material diffused into it (providing the base), and then, by means of masking techniques, a small n-type layer (the emitter) is diffused into a part of the p-type base. The "mesa" or raised structure (named after a flat-topped steepsided mountain) has the object of reducing collector capacitance, and is formed by etching away the surrounding silicon. Metallizing is applied to the base and emitter for contacts, while the collector connection is made through a large metal support on which the collector is mounted.

Current Regulating Semiconductor diode demonstrated by G.E.C. Research Laboratories is a counterpart of the well-known Zener diode voltage regulator. The principle of control is the constriction of the current carrier path through the semiconductor by two space-charge regions, rather as in the Tecnetron and other field-effect devices. A somewhat complex structure is used, in which the widening space-charge regions are produced by the voltage drop sustaining the current flow through the device. The two regions actually meet at a certain applied voltage, the pinch-off" voltage. Thereafter the current is limited to a constant value with increasing voltage until a breakdown of the reverse-biased junctions in the device allow a sudden surge of breakdown current. A typical value for the "pinch-off" voltage is 5V, and the current after this is limited to 5mA until the breakdown occurs at 100V. A voltage regulating circuit giving 6 volts output was demonstrated in which the new device formed a series element while a Zener diode provided a shunt element. Variations of 50% in supply voltage produced output variations of only $\pm 0.005\%$.

Image Intensifiers normally work by the acceleration of photo-electrons from a photo-emissive layer and by the concentration of the image into a smaller area. A new intensifier tube shown by 20th Century Electronics accelerates the electrons from the photocathode in the normal way, but provides additional gain by using five electron multiplier stages of the transmission type. In these the electrons pass through a thin layer of alkali halide on an alumina support, and after multiplication are accelerated by ring electrodes (metallized on the inside of the glass envelope) to the next stage. There is a total electron gain of 2,000 in the multiplication system, while the overall light gain from photocathode to fluorescent screen (15kV potential) is of the order of 50,000. The image remains the same size throughout, and the electron rays are kept parallel by enclosing the tube in a magnetic focusing solenoid.

Last year Siemens Edison Swan showed an image intensifier for electron microscope work which used a photoconductive pick-up tube to give television-type pictures (March, 1959, issue, p. 133). This year A.E.I. (as they now are) were demonstrating a modified version for light amplifica-Electrons from the phototion. cathode are imaged by an electrostatic lens system on to the front of the selenium photoconductive layer of the pick-up tube. The resultant charge pattern at the rear of the layer is then scanned in television fashion by a low-velocity electron beam. Sensitivity is such that about 5 incident photons on the photocathode

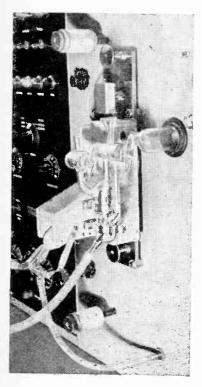


Cross-section of Ferranti h.f. transistor, showing "mesa," or plateau, form of construction.

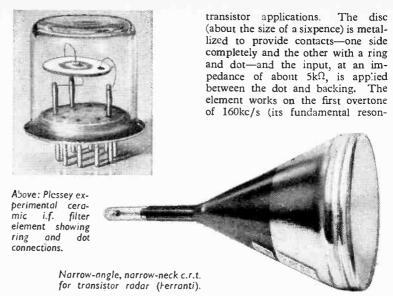
will give a visible spot on the display monitor.

Storage Tubes, which hold information as patterns of electrostatic charges, were shown in both the display and non-display types by various firms. A new non-display tube by Mullard uses a single electron gun for both "writing" and "reading" on to the magnesium fluoride storage surface and is notable for its small dimensions of about $1\frac{1}{2}$ in diameter and 7 in long. It has a resolution of 600 lines and operates with an anode voltage of about 300V. The Vidicon type of television pickup tube is often criticized because of the storage properties of its photoconductive light-sensitive layer, but E.M.I. were showing a version in which this effect had been deliberately enhanced for the purpose of holding images. Storage time was in the region of 1-3 minutes, depending on image brightness.

7-in Radar C.R. Tube developed primarily for transistor equipments was shown on the Ferranti stand. It has a narrow deflection angle of about 30° and a narrow diameter



Machine by G. V. Planer, Ltd., for the application of resistive oxide f.lms to glass fibre. Whole machine is complete in one cabinet, including instrumentation and control apparatus.



neck (23mm), both of which facilitate scanning by low-power transistor circuits. This firm was also showing a new 6-inch c.r.t. of high sensitivity and brightness specially designed for displaying characters for xerographic printing (as used for computer output data). The well-known 1CP1 c.r.t. with the 1-inch screen was shown by Electronic Tubes in a new version which requires an anode voltage of only 350V, thereby allowing it to be operated from normal h.t. supplies.

Digital Display Tube working on a new principle was shown in experimental form by G.E.C. Research Laboratories. It is basically a small c.r. tube and the numerals are formed as the shadows of shaped wire electrodes in the path of an "unfocused" electron beam which floods the whole area of the 1½-inch fluorescent screen. Ten such electrodes, shaped respectively 0-9, are provided and each one is selected for display by applying to it a negative voltage of about 60V. This repels the electrons in the beam close to the wire and so forms the shadow of the selected numeral on the screen. The shadows are very clearly defined and the only limitation appears to be in the precision and artistry with which the wire numerals can be shaped. An anode voltage of 600V is required.

Ceramic I.F. Transformers were shown by Plessey in the form of small circular discs of lead titanate zircente. This material exhibits piezoelectric properties and, when a circular disc is excited in the radial mode the performance resembles that required of an i.f. transformer for

ance) which is 455kc/s and the output, at an impedance of about $1k\Omega$, appears between the ring and backing. The characteristics of the radial-mode resonances are such that, by pairing a fundamental-frequency resonator with one operating on its first overtone, the overtones do not coincide. Consequently out-of-band responses are not troublesome; also the selectivity curve is improved.

Oxide-film Resistance Elements are used in new components by G. V. Planer Ltd. The component is a tilt-sensing device giving a continuous response. It consists of two shaped stannic-oxide films coated onto the inside of a sealed glass tube in which a drop of mercury runs, joining the two "stripes". To avoid contamination of the mercury by metallic contacts the oxide films are brought out through the glass of the seal to contacts outside.

The continuous coating of oxides on to 10⁻³-in diameter glass fibre, which can subsequently be wound on a bobbin to produce a resistor, is performed by a machine developed by Planer. This comprises precision winding gear for passing the fibre through the coating furnace and an automatic speed control system to ensure the deposition of correct film thickness. After leaving the lower bobbin the fibre is first sprayed with the mixture of salts. It then passes through a furnace to fire and oxidize the film, and on leaving for the takeup spool it passes through two mercury cups, the resistance between which governs the drive speed. Values as high as $2M\Omega$ per linear inch of fibre are achieved.

WIRELESS WORLD, MARCH 1960

Signal-Flow Diagrams

2.—Application to the Schmitt Trigger Circuit Using Valves

By THOMAS RODDAM

AST month I discussed the general principles of signal-flow diagrams, the maps which frequently show useful short-cuts to the solution of com-The rules for drawing maps plicated circuits. and for simplifying them were described and a not particularly illuminating example was used to illustrate the application of the method. It was not particularly illuminating because it was too simple: you don't draw a map when you want to say "Turn left at the end of the road and it's there on your left." This month we can consider something more ambitious which really gives the method a chance. I had thought of tackling the transistor Schmitt trigger, but I have decided to take the valve version which is used as an example by J. G. Truxal in "Automatic Feedback Control System Synthesis" (McGraw-Hill, 1955). My reason is that he goes on to discuss a whole lot of other developments, such as impedance determination and return difference evaluation, and although I do not propose to discuss these topics, this article will break the ground for any of you who want to go further by consulting this book.

The circuit is shown in Fig. 1. It is a two-stage d.c. amplifier with positive feedback produced by the common-cathode resistor. To avoid using primes in the analysis we shall assume that

$$1/R_1' + 1/R = 1/R_1$$

so that R_1 is the effective anode load of the first stage. In a practical case R is usually so much bigger than R_1 that within the usual tolerances we can take the anode resistor as the total load.

We are treating this circuit in its a.c. application as a limiter, not in its on-off large-signal application. The usual problem then is to find the conditions for just snapping over. Signal-flow diagrams are concerned with linear differential equations, so we must assume that the circuit is somehow poised at mid-travel, either because it has not enough feedback to make it unstable, or because we have caught it in mid-travel. All our terms are small signals, not the fixed biases and standing currents.

The starting point for the map is e_{in} and I have begun Fig. 2 by putting in the first move, e_{in} . The first equation which springs to mind is the obvious one:

$$e_{\rm L} = e_{\rm g1} + e_{\rm k}$$

Like this the equation would have two branches, from two new nodes, leading to the input node $e_{\rm in}$. This is not considered good practice, so we rewrite the equation as

$$e_{g1} = e_{in} - e_{k} \quad . \tag{1}$$

mark in the nodes e_{g1} and e_k and draw the two unit arrows shown in Fig. 2(a). At this point Truxal decides to put e_k down below the main horizontal line. One application of the circuit is as a phase-

splitter, the long-tailed pair, with the anode-grid coupling as an extra something—positive feed-forward one might call it—to push up the gain. This interpretation makes us think of the second stage as fundamentally grounded-grid, with e_{\parallel} as the drive, and so I shall keep e_{\parallel} on the line.

The anode current of the first valve is given in the usual way by

$$i_{1} = (\mu_{1}e_{.1} - e_{k})/(\rho + R_{1})$$

$$= \left(\frac{\mu_{1}}{\rho + R_{1}}\right)e_{g1} - \left(\frac{1}{\rho + R_{1}}\right)e_{k} ... (2)$$

This adds a new node i_1 and two new branches leading to it from $e_{\rm g1}$ and $e_{\rm k}$. These are shown in Fig. 2(b), where the new branches are the heavy lines: in each following figure the latest additions will be identified in this way.

The drive to the second valve, in my view, comes from e_{\cdot} , which is the drop across the cathode resistor due to both anode currents. Thus

$$e_{k} = i_{1}R_{k} + i_{2}R_{k} \dots \qquad (3)$$

We need to mark in the new node i_2 and then we can add the local map representing this equation to give us Fig. 2(c).

Before we can write an equation for i_2 we need to know something about the feed-torward path. Clearly

$$e_{a1} = - R_1 i_1 \ldots (4)$$

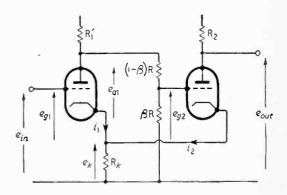
and this equation is shown by the new heavy branch in Fig. 2(d)

We now need to find out what happens in the second valve. The grid to cathode signal is quite clearly given by

$$e_{12} = \beta e_{a1} - e_{b}$$
 .. (5

and we can establish the new point e_{g2} in Fig. 2(e) by this equation.

Fig. 1. Schmitt Trigger Circuit.



ordinary valve equation for the second valve

$$i_{2} = (\mu_{2}e_{g2} - e_{k})/(\rho + R_{2})$$

$$= \left(\frac{\mu_{2}}{\rho + R_{2}}\right) e_{g2} - \left(\frac{1}{\rho + R_{2}}\right) \cdot e_{k} ... (6)$$

This equation makes its appearance in the next figure, Fig. 2(f).

Now all we need is to know eout, which is given

$$e_{\text{out}} = -i_2 \mathbf{R_2} \qquad \dots \qquad (7)$$

 $e_{\rm out}=-i_2{\rm R}_2$.. (7) and leads us to Fig. 2(g). Quite obviously the signal-flow diagram of Fig. 2(g) is a complicated affair, but the original circuit, as anyone who has ever carried out the solution by algebra knows, is by no means as simple as one might expect from five resistors and two valves. In the form of Fig. 2(g) the diagram is not of much use to anyone, so we must settle down to the task of simplifying it.

We take a new sheet of paper and begin Fig. 3(a). A first step is to notice that we do not really care about the point e_{a1} and we can go straight from i_1 to $e_{\rm g2}$ by a branch — $R_1\beta$. Sheer laziness makes me leave out the diagram with this reduction in it, because I can also see that the signal which flows from e_k through e_{g2} to i_2 can be written as a contribution directly from e_1 to i_2 of size $-1 \times$ $\mu_2/(\rho + R_2)$ and then I can get e_{e_2} out of the picture as well. These steps are all combined to give Fig. 3(b).

The two parallel branches in the same direction from e_k to i_2 can be added together, but before we do this in Fig. 3(c) let us notice also that we have another easily seen step in the reduction. There is a flow of signal from e_k to i_1 by way of e_{g1} which we can replace by a direct path. This direct path, which will have transmittance $-1 \times \mu_1/(\rho + R_1)$ is in parallel with the existing direct path $-1/(\rho + R_1)$ $(\rho + \hat{R}_1)$ and these two can be combined together. The result of these various operations is shown in

To make the figure look a little neater it is rearranged in Fig. 3(d) and the point e_{g1} is dropped altogether.

The next stages in the reduction require a certain amount of care to avoid throwing the baby out with

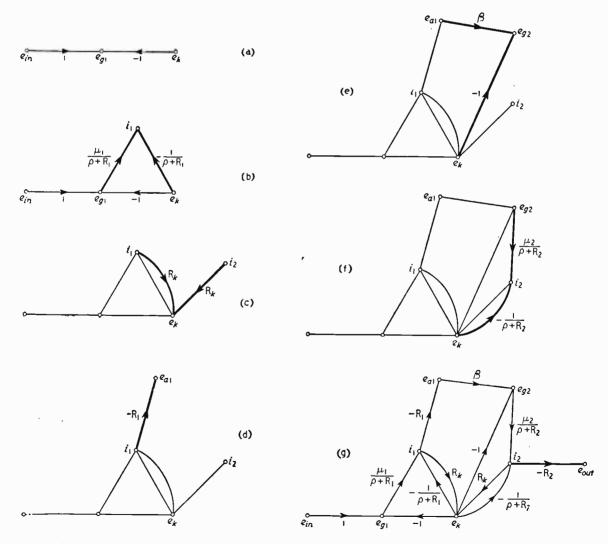


Fig. 2. Signal-flow diagram for the circuit of Fig. I obtained step by step. Heavy lines indicate the additions corresponding to successive equations.

130

Wireless World, March 1960

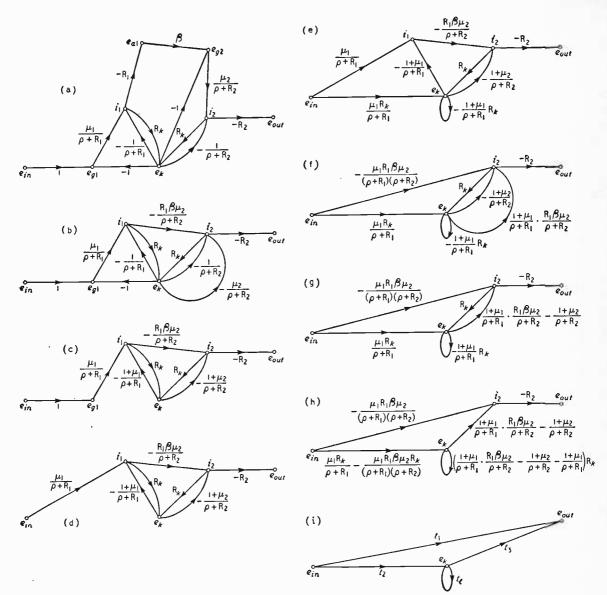


Fig. 3. Stages in the reduction of the signal-flow diagram for the circuit of Fig. 1.

the bath-water. Let us focus our attention on the the bath-water. Let us focus our attention on the flow of signal from e_{in} to e_k . There is a contribution through i_1 which is clearly of total transmittance $[\mu_1/(\rho + R_1)] \times R_k$. We can therefore construct a direct branch from e_{in} to e_k and cross out the i_1 to e_k branch of transmittance R_k . When we do this, however, we are breaking the loop e_k to i_1 to e_k , of transmittance $[-(1 + \mu_1)/(\rho + R_1)] \times R_k$, and we must put in the corresponding self-alone. But the we must put in the corresponding self-loop. But the signal from e_k to i_1 also contributes to the path i_1 to i_2 , and it is not legitimate to drop the e_k to i_1 branch. The reduction therefore gives us Fig. 3(e).

Now we consider the contribution made by e_k to i_2 by the route e_k to i_1 to i_2 . The transmittance is $[(1 + \mu_1)/(\rho + R_1)] \times [R_1\beta\mu_2/(\rho + R_2)]$, with two minus signs disappearing together. We can make this a direct contribution, when we no longer need the point i_1 . This leads us to Fig. 3(f) and, immediately,

to Fig. 3(g).

Again we consider the flow of signal from e_{i_1} to e_{k} and we see there is a path via i_2 with a transmittance

$$\frac{-\mu_1 R_1 \beta \mu_2}{(\rho + R_1) (\rho + R_2)} \times R_k$$

and that this will introduce another self-loop at e_k of transmittance

$$\left(\frac{1+\mu_1}{\rho+R_1}\cdot\frac{R_1\beta\;\mu_2}{\rho+R_2}-\frac{1+\mu_2}{\rho+R_2}\right)\times R_{\mathbf{k}}$$

Fig. 3(h) shows the effect of this change on the diagram.

In the final diagram, Fig. 3(i), the point i_2 has been eliminated in the obvious way, to leave us with what is called an essential diagram of order one. Once you get the trick of it the steps described above can be carried out very quickly. In the same way you can go on to reduce the essential diagram to a single branch joining input to output and having a transmittance

$$t_1 + t_2 t_3 / (1 - t_1)$$

Each of these t's is an expression as long as your arm, which is why I have not written them out in full. For specific problems you may not want to do this anyway. As an example, the determination of the conditions in which the circuit will just turn over, implying infinite gain, are seen to be that the term t_1 should be unity. We know, in any practical case, the values of μ_1 , μ_2 , ρ , R_1 , R_2 . We may wish to make β unity and find R_k , or to fix R_k and find β . But the expression for t_1 is greatly simplified by putting in numbers. In fact, as you will see if you consult

Truxal (loc. cit.) you can carry numbers all the way through to considerable advantage.

The only way in which you can get the full benefit of this method of handling circuits is by practice. Once you get the swing of it you will find that it really does help enormously to be able to concentrate on one limited aspect of your circuit equations and to ease that bit a step or two towards the final solution. This discussion is only an introduction to the method, which can be extended to the determination of input and output impedances, to the effects of reactances, and to the whole field of linear circuit problems. For complex problems it seems to be an extremely useful method for those who want to keep their eye on what is happening as they solve their equations.

BOOKS RECEIVED

Stereo Handbook, by G. A. Briggs. Wide survey of fact and opinion on the methods and results of stereophonic sound reproduction. Takes the form of an inquiry in which awkward questions were posed to a number of experts and from which some sound sense has been distilled by the author. Topics considered include pickups, record wear, loudspeakers, room acoustics and broadcasting. Pp. 146; Figs. 88. Price 10s 6d. Wharfedale Wireless Works, Ltd., Idle, Bradford.

Stereo Record Guide, by Edward Greenfield, Ivan March and Denis Stevens. Collection of critical assessments of existing stereo gramophone records, as regards both musical content and recording quality. Arranged under composers' names and with entries up to the end of 1959. (Supplements are to be issued.) Pp. 320. Price 21s. The Long Playing Record Library, Ltd., Squires Gate Station Approach, Blackpool, Lancs.

Stereo and Hi-Fi as a Pastime, by Douglas Gardner. Layman's introduction to the technicalities of disc, tape and broadcast sound reproduction. Pp. 147; Figs. 15, and illustrations of typical commercial installations. Price 15s. Souvenir Press, Ltd., 94, Charlotte Street, London, W.1.

Mullard Circuits for Audio Amplifiers. Designs for power amplifiers and pre-amplifiers for mono and stereo

sound reproduction from microphone, tape, gramophone disc and radio signals. Includes dimensions of suggested chassis and general notes on construction and assembly. Pp. 136; Fi3s. 160. Price 8s 6d. Mullard, Ltd., Mullard House, Torrington Place, London, W.C.1.

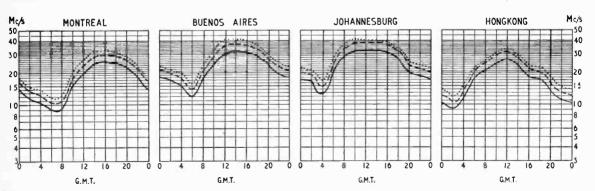
The Practical Hi-Fi Handbook, by Gordon J. King, Assoc. Brit.I.R.E. Introduction to high-quality sound reproducing equipment written with the needs of the service technician in mind. Pp. 224; Figs. 158. Price 25s. Odhams Press, Ltd., Long Acre, London, W.C.2.

The Conversion of Ionospheric Virtual Height/Frequency Curves to Electron Density/Height Profiles, by J. O. Thomas, M.A., Ph.D. (Cavendish Laboratory, Cambridge) and M. D. Vickers, B.Sc. (D.S.I.R., Radio Research Station, Slough). D.S.I.R. Special Report No. 28 on a digital computer programme and the basis of its formulation, with appendices including an extensive bibliography. Pp. 48; Figs. 10. Price 3s 6d. H.M. Stationery Office, Kingsway, London, W.C.2.

From Microphone to Ear, by G. Slot. Second edition of a Philips Technical Library review of modern sound recording and reproducing techniques, now including a long chapter on stereophony. Pp. 268; Figs. 110. Price 21s. Cleaver-Hume Press, Ltd., 31, Wright's Lane, London, W.8.

SHORT-WAVE CONDITIONS

Prediction for March



THE full-line curves indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths from this country during March.

Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.

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FREQUENCY BELOW WHICH COMMUNICATION SHOULD BE POSSIBLE
ON ALL UNDISTURBED DAYS

WIRELESS WORLD, MARCH 1960

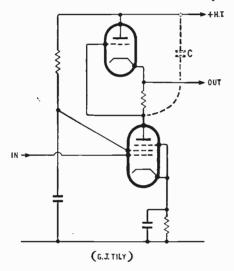
LETTERS TO THE EDITOR

The Editor does not necessarily endorse the opinions expressed by his correspondents

"Economical High Gain A.F. Amplification"

MUCH as I admire the ingenuity of your contributor Arthur R. Bailey in devising the circuit described in the article in the January issue, I must point out that he has omitted mention of the most important features of this circuit. In respect of gain, hum and noise level, and low output impedance, two R-C coupled pentodes with voltage negative feedback can compare favourably with this circuit. In my opinion the principal advantages of the circuit described are that it lends itself readily to direct coupling, and that this circuit has effectively only one stray capacitance.

At the low anode current mentioned by the author (0.1mA) the cathode bias resistor for the triode can be of the same order as the anode load of the pentode,



and one resistor can be used for both functions as in the accompanying diagram. A possible use for this arrangement is deflection amplification for a cathode

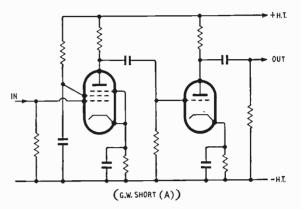
ray tube.

The single stray capacitance can be an important factor if this circuit is used as the amplifying stage in a feedback amplifier, as it restricts the phase shift to 90°, which ensures stability in such circuits.

Hounslow, Middlesex. G. J. TILY.

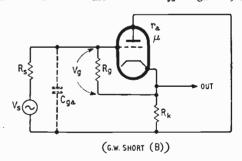
FEW readers will fail to sympathise with Mr. Bailey's aim "to obtain the maximum amplification from the minimum number of components" (January issue, p. 25). But it is to be doubted whether his circuit represents a marked advance in this direction. Seven resistors and four capacitors are employed in the basic circuit (Fig. 2 of the article) and the gain of a practical version is given as 3500. If we take as a figure of merit the gain divided by the number of components, we get 3500/11 = 318. The completely conventional circuit shown here in diagram (A) uses 13 components, and taking Mr. Bailey's own figures for the gain of the pentode without any fancy business (200) and the gain of the triode (40), it provides, by cascading a gain of 8000. So its figure of merit is 8000/13 = 615, which is nearly twice as good as that of the unconventional circuit. If valves, valveholders and decoupling components are counted, the conventional circuit shows up rather better.

Moreover, some of Mr. Bailey's statements do not



bear examination. He rightly gives the effective firststage output resistance for a response 3dB down at 10kc/s, with 5pF strays, as $10/\pi$ $M\Omega$, and this value is indeed attained with an 8-M Ω anode load in parallel with a 5-M Ω anode impedance. But taking the captions of the contraction of the contracti acitance as 10pF (which is a reasonable figure allowing for wiring strays), the required effective output resistance becomes $10/2\pi~M\Omega$, or 1.59 $M\Omega$. To obtain the required value of 1.59 M Ω , the anode load resistance must be reduced to 2.3 M Ω in contrast with Mr. Bailey's value of $4M\Omega$. The gain is then reduced to about three-quarters of the original amount.

Mr. Bailey refers to the output stage as a cathode-follower, but this is not correct. Disregarding supply voltages and biasing arrangements, the circuit reduces to diagram (B), where V_o , R_s represent the signal source, R_o is the physical grid-cathode resistance, and R_k is the load resistance. If the effect of C_{oa} is ignored, and if



R, approaches infinity, then R, can be removed without raterially affecting the output impedance, which is then R_k in parallel with r_a . In other words, it is the same as that of a normal triode amplifier stage with an anode load equal to R_k . The circuit is then a "bootstrap" amplifier. To make it into a cathode-follower R, must be small, but this is not so in Mr. Bailey's circuit, where it is $5M\Omega$, the anode resistance of the preceding pentode stage. The output impedance is actually $r_a (R_s + R_g)$ $[r_a + R_s + R_g (1 + \mu)]$, in parallel with R_k , which comes

to roughly $\frac{\mathbf{r}_a}{2}$ in the practical circuit.

In so far as the circuit behaves as a bootstrap amplifier, the effective size of C_{ga} is increased as a result of Miller effect. The voltage acting upon $C_{\sigma a}$ is $V_{\sigma} + V_{\sigma u}$, which is $V_{\sigma}(1 + A)$ as in a normal triode stage.

Finally, it is perhaps worth remembering that, with a.f. amplifiers as with h.f. amplifiers, the product of gain and bandwidth obtainable with a given valve is a constant. If a gain of 3500 and a bandwidth of 10kc/s are to be attained, then the gain-bandwidth product must be 35Mc/s. While good h.f. pentodes may be expected to have gain-bandwidth products running into hundreds of Mc/s, it must be remembered that in h.f. amplifiers the full rated mutual conductance can be achieved, whereas in an audio stage, operating with low current into a resistance load, the actual g, is only a small fraction of the normal value, and the gain-bandwidth product is correspondingly reduced.

Croydon.

G. W. SHORT.

A. R. BAILEY'S article on high-gain a.f. amplification reminded me of some experiments I carried out on the Jeffery circuit. The final modification to the Jeffery circuit (Wireless World, Aug., 1947, p. 274) used direct coupling (inspired by D. T. N. Williamson's phase splitter) of the pentode anode to the triode grid, so saving three resistors and three capacitors. The pentode was biased back so that it drew only a small anode current, and this current, flowing through the triode grid leak (50 to $100 k\Omega$), provided the correct grid bias for the triode. However, this was in the days of Government-surplus EF36 and 6J5 valves, and the circuit was never applied successfully because it suffered from a high hum level and excessive h.f. losses. reduce the hum to a tolerable level much negative feedback was necessary; but the application of this was precluded by the combined phase-shifts of the phase splitter and the particular amplifier used. Also the heater-cathode insulation of the triode was too highly stressed (at under half h.t.) unless a separate supply was used; this offsets the saving of a valve. As most of these disadvantages can be overcome with modern valves and techniques this circuit might bear re-examination.

But what happens if we apply these ideas to Bailey's circuit? A saving of two resistors and one capacitor is made, and the result looks very much like a straightforward bootstrap. Another point to be considered is heater-to-cathode potential limits for the triode: just under h.t. potential is applied. The writer's (sad) experiences indicate that it is unwise to exceed 100V or so for long-term reliability. In fact, the only superiority of Bailey's circuit may lie in the tape pre-amplifier, where the very high impedances make it easy to obtain a low l.f. boost roll-off point.

Surbiton, Surrey.

E. MANSFIELD.

"Subjective Colour Tests"

IN your article on this subject in the January issue it was reported that a Land-colour rendering of the scene was obtained when the two photographs were viewed by displaying one to each eye with the appropriate filter in the light path to one eye. I find it difficult to under-stand this statement as I have found a very different effect. If the two photographs are viewed in a stereoscope with a red filter over the appropriate eyepiece, there is almost a complete absence of the colours seen When the during projection in the normal manner. two transparencies are projected through Polaroid filters and viewed through Polaroids, so arranged that each eye sees one of the images, there is a very marked decrease in coloration, although more colour may be seen than when using a stereoscope. This may be due to imperfections of the projection system and screen which results in incomplete channel separation, for it can be shown that even a small difference in the wavelengthintensity distributions of the two lights used for projecting the transparencies can give rise to Land colours. Thus, when using a 300 watt and a 1000 watt projector, without any filters, Land colours may be observed owing to differences in the colour temperatures of the two light sources.

When photographs of a test chart containing 24 areas of different colours against a neutral background are viewed in a stereoscope with a red filter over the appro-

priate eveniece, most subjects report absence of colour other than reds, whites and pinks. Some subjects, however (including myself), observe that one of the test areas has a bluish-red or purple appearance. This area is seen as green in the original display and is reported as blue-green when viewed in the conventional Land manner. With continued viewing in the stereoscope, retinal rivalry is observed. In this case there is a slow oscillation between seeing mainly what is fed to the left eye or mainly what is fed to the right eye.

Thus it appears likely, to me, that the Land colours are due mainly to processes in the retina and not in the brain as suggested in your article. The effects due to retinal rivalry are, on the other hand, due to activity in the brain, as they can be controlled by attending to one eye or the other. This effect is similar to that encountered when using a microscope when it is possible to keep both eyes open without interference from the eye which is not looking through the microscope.

Whilst not wishing to over-emphasize the importance of Land's work in relation to colour television, it should be pointed out that all colours are, in a sense, subjective. The objective stimulus is light of a certain wavelength-intensity distribution. It may be seen differently in terms of colour by different observers and the colours may appear different to the two eyes of the same observer if he is colour defective in one eye. In colour television it is unlikely that the amount of light leaving the screen will ever approximate to the amount of light being reflected from the objects which are being recorded by the television camera. In order to make the picture appear like the display, changes are necessary in the tricolour specifications of individual areas to allow for the increases in saturation, and possibly changes in colour, due to the decreased level of illumination of the The Land process tends to desaturate colours (whilst the trichromatic system tends to oversaturate them), and to some extent this is a desirable characteristic. As the display which is televised has other attributes than that of colour which are at present relatively imperfectly reproduced, it may well be that, with a fixed available bandwidth, a two-channel system would give a satisfactory compromise between definition, clarity of movement, and colour.

Manchester, 13.

C. E. M. HANSEL, Department of Psychology, University of Manchester.

Early Public Address

IT may interest Mr. Haydon G. Warren (Dec. 1959 issue) to know that in 1919 or 1920 when I was with the Western Electric Company at North Woolwich we received from America a p.a. equipment of no mean performance. Associated with the rather enormous amplifier was a stretched diaphragm carbon microphone. A battery of balanced-armature loudspeakers with corrugated Bakelized linen diaphragms provided a considerable volume of good quality speech. I toted this lot about the country and had a lot of fun with it, for it was new and exciting. I even took it to sea for trials of gun control but, alas, the first round blasted the diaphragms beyond repair. On the mess-decks the matelots had other means of damping the output!

On another occasion I remember we stopped the traffic in Birmingham while doing tests prior to a visit by the then Prince of Wales, later Edward VIII.

Walmer, Kent.

S. G. KNIGHT.

Editors and Editing

Those were the days!

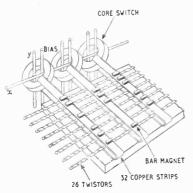
I TAKE it that Mr. Waldron, whose letters you published in the January and February issues, will not be contributing to the Indian radio journal whose editor frankly warns intending authors of his right "to suppress, revise, alter or mend each or any para of the article"!

London, S.E.22.

I. P. HAWKER.

Television Aerial Design is a field in which many specialized problems have to be solved—for instance, the making of an aerial which responds efficiently to one channel in Band I and one in Band III. One of the earliest methods of doing this was by the addition to the Band-I dipole of $\lambda/4$ (at the Band-III frequency) rods, or "twigs" as they became known, in such a way that, when excited by the Band-III signal, they acted as "metallic insulators" or short-circuited $\lambda/4$ stubs. These reflect a high impedance at the ends of part of the dipole, so forming a $3\lambda/2$ or $3\lambda/2$ Band-III radiator. J. D. Burke has discovered that the function of the "twigs" may be carried out by ferrite beads or sleeves round the dipole elements (Prov. Pat. Spec. 17109/59) in much the same way that small ferrite beads same way that small territe beaus are used for decoupling in place of wound r.f. chokes in valve heater circuits, etc. The sleeves may be placed either $\lambda/4$ or $3\lambda/4$ from the feeder connections to the dipole and they appear to act in much the same manner as the "twigs" in that Band-I performance is not materially altered. Burke points out that a similar technique may be useful for "tuning" the metal legs of a television receiver, so that they can act as a resonant aerial.

Magnetic "Punched Cards", or permanent information stores, from which programme data can be electrically fed into computers much faster than from conventional punched cards, have been developed by Bell Telephone Laboratories in the U.S.A. and prepared for mass-production by the associated Western Electric Company. Whereas a punched card stores information by the positions of the punched holes, the new medium stores it by the positions of tiny "bar magnets" etched from a continuous magnetic film on a plastic card. The new units are not read by being fed through a machine but are kept stationary in close proximity to wiregrid sensing planes, and a complete store consists of a whole stack of alternate magnet-cards and sensing grids. In each sensing grid the wires





running in one direction are "Twistors" (see January, 1958, Twistors" issue, p. 32), while those in the other direction are ordinary conductors. The Twistors have a spiral wrapping of magnetic tape round a copper wire, providing a helical magnetic path, and are embedded in a strip of Mylar film. The other co-ordinate of the grid is made up of strips of copper foil 0.060in wide, also embedded in a Mylar film. This film is folded over the Twistor strip so that one complete loop is formed, each loop of copper strip providing a single-turn coil at right-angles to the Twistors. For access to the store, each copper-strip coil, or "word coil", is connected to a biased square-loop ferrite-core switch. This provides a means of selecting a par-ticular "word coil" out of a matrix of such coils. Reading out from the store is accomplished by pulsing the copper-strip "word coils" by means of the ferrite cores, and the action of the "bar magnets" is to magnetically inhibit the effect of this pulsing on the Twistors. If there is no magnet at a given intersection of "word coil" and Twistor, a pulse is generated in the Twistor and is read out. If a magnet is present, its field prevents the production of the Twistor pulse, and no signal is read out. The pulse from the Twistor is about 6mV and 1µsec in duration. Speed of reading has not been stated precisely but is presumably of the order of 10^a binary digits per second. The magnet cards can be changed easily when a new programme is required.

Irasers, Lasers and Rasers are similar to masers except that they operate at infra-red, light and radio frequencies respectively, rather than at microwave frequencies as does the maser. Since the first letter of the word maser stands for its operating frequency, this letter has been changed in the names of these three varieties of maser to correspond to their different operating frequencies. According to *Electronic News* for June 22, 1959 (p. 5) the operation of irasers will depend on transitions between electron spin quantum energy levels in the same way as masers. Rasers, on the other hand, because of their lower operating frequency, must use smaller energy differences than those which can be obtained between electron spin levels. These required smaller energy differences can be obtained between nuclear spin quantum levels. It is proposed to induce transitions between suitable nuclear spin quantum levels by means of electron spin transitions.

Variable Capacitance Diodes are used in parametric amplifiers for operation in the u.h.f. and s.h.f. region because they have the necessary speed of response for variable-reactance elements at these frequencies. A diode with the required properties has now been put on the market in Britain by G.E.C.



Suitable for use in radar and communications systems, it will operate at frequencies up to about 4 Gc/s. It is mounted in a coaxial structure for direct insertion into coaxial and waveguide circuits, and has the very low series inductance of 0.5muH. Because of its very low forward impedance and very high reverse impedance, the device can be used as a microwave switch. Another application is as a frequency multiplier—an important aid to the design of microwave receivers based entirely on solid state devices.

Figure of Merit for audio output valves suggested by R. M. Mitchell in Audio for November, 1959 (p. 40), is power output x damping factor x efficiency x maximum possible grid circuit resistance ÷ (input voltage x harmonic distortion x price), the harmonic distortion and input voltage being taken at the stated power output. A high maximum possible grid resistor and low input voltage are desirable so as to make the valve easier to drive. The effect of feedback and qualifications to this figure of merit (such as weighting individual items) required for particular types of circuit are discussed in the article.

Elements of Electronic Circuits

11.—TRIGGERED TWO-STATE CIRCUITS

By J. M. PETERS, B.Sc. (Eng.), A.M.I.E.E., A.M.Brit.I.R.E.

A COMMONLY used two-state circuit which requires only one valve is the transitron, shown in Fig. 1. It consists of a pentode with suppressor and screen grids linked by a capacitor, and it behaves as a single-stage amplifier with feedback from screen to suppressor. It possesses only one stable state followed by one unstable one; does not freely run (as in the case of the multivibrator); and requires a trigger or initiating signal to maintain the action. The action will be described with reference to Fig. 2 and will be divided into three main intervals.

Interval 1:

(a) Initially the circuit is in a stable state with I_a and I_{g2} flowing. $V_{g3}=0$ and C is charged.

(b) A positive-going sync voltage applied to V_{g1} causes I_a and I_{g2} to rise. V_{g2} and V_a therefore fall. As g_3 is connected to g_2 by C the instantaneous potential change is conveyed directly to g_3 , which consequently follows and V_{g3} goes negative. I_a is therefore reduced and I_{g2} increases. A further reduction in V_{g2} results in a corresponding fall in V_{g3} . This action is cumulative and ends when I_a is cut off and all the space current flows to the screen.

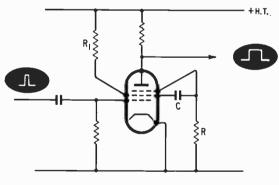


Fig. 1

(c) After the initial drop, V_a very quickly rises to h.t. potential.

(d) The first stage terminates when V_{g1} drops to zero. I_{g2} drops and V_{g2} and V_{g3} rise, but I_a remains cut-off. This is because g_3 was driven sufficiently negative during the first stage and is still well beyond suppressor cut-off voltage.

Interval 2:

(e) During the next interval C discharges through V and R with time constant CR ($R \gg R_1$), and the voltage on g_3 rises towards suppressor cut-off.

(f) I_a begins to flow, V_a drops and I_{g2} is reduced. V_{g2} increases, a further rise in V_{g3} is caused; I_a increases, causing a further drop in V_a . This action

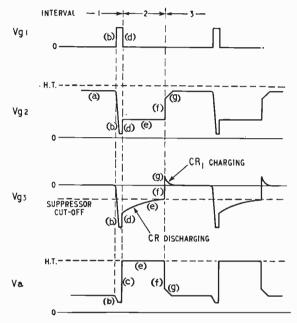


Fig. 2

is cumulative and the circuit returns to its stable condition.

Interval 3:

(g) V_{g3} follows the rise in V_{g2} and, due to its positive excursion, I_{g3} flows, causing C to charge rapidly through R_1 . This flow of current through R_1 produces an irregularity in the anode and screen grid waveforms at the trailing edge of the output pulse.

The duration of the output pulse from the anode depends on the time constant CR, and since the anode voltage is only indirectly affected by the charging and discharging of C, the waveform is square. Instead of applying a positive-going sync pulse to g_1 it is possible to trigger the circuit by a negative-going pulse either at g_2 or g_3 . The action is similar, although the sync pulse must be of larger amplitude.

It may also be noted that, under certain conditions, provided that the sync pulse is of sufficient amplitude and duration, the cumulative actions described above can be initiated by a negative-going sync pulse applied to the control grid.

The control of I_a by V_{g3} requires the creation of a space charge ("virtual cathode") between g_2 and g_3 . Under these conditions the pentode may be regarded as a pair of triodes with cathode, g_1 and g_2 as one valve; and with "virtual cathode", g_3 and anode as the other.

Principles of a Measuring Device Used on Lines and Waveguides

A READER, sharing my dislike for hazy or uncertain ideas, has cited the reflectometer as an example. He has seen a number of treatises on it, all of which failed to convey to him how it works. On the assumption—which I, knowing him, regard as most reasonable—that if he is puzzled others will be, he has urged me to do something about it.

Students of Q.S.T. and the A.R.R.L. Handbook (for 1957 thru 1959, as they say over there) will know the reflectometer better as the Monimatch. It has also been called the Directional Coupler.

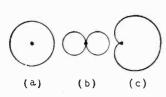


Fig. 1. Polar diagrams showing the directivity of (a) an ordinary vertico! aerial, (b) a loop aerial, and (c) the well-known direction-finding aerial in which (a) is added to (b)—one loop of (b) being negative.

Whatever the name, its purpose is to enable r.f. power travelling along a line or waveguide to be sorted out according to direction of flow, so that the direct and reflected power can be measured separately. This obviously enables the standingwave ratio* to be calculated. In turn this indicates the ratio of mismatch at the far end of the line,

which is what one wants to know when setting up a transmitter for maximum efficiency or when measur-

ing v.h.f. impedances.

The more familiar method of measuring s.w.r. is to have a slot cut along more than half a wavelength of the line, through which a suitable voltmeter probe can be slid. The maximum and minimum readings give the ratio directly. The sliding process is not always very convenient; and if the source of power is a magnetron, which is apt to generate undesired frequencies when badly adjusted, it can happen that the best s.w.r. is indicated when the power is divided up amongst the greatest number of such frequencies -quite the reverse of the general intention. The reflectometer, on the other hand, can be fixed at any convenient point in the line, and indicates the reflected power directly-either in total, or frequency by frequency, according to the type of detector. It is also very easy and cheap to make. Between them, published designs of reflectometer cover frequencies at least from 2 to 3,300 Mc/s, but the most usual applications seem to be in the v.h.f. band.

At first thought it may appear rather a difficult thing to tell how much of the r.f. power is going each way when it is going both ways at once. Come to that, it's not altogether obvious how to tell which way it is going even when it is all going one way. The ordinary loop aerial can show the line of travel but not the direction along it. Readers familiar with

radio direction finding, however, will remember the old dodge of combining a loop aerial with an ordinary vertical aerial giving the same amount of output. With the combination aerial facing one way, these two outputs are out of phase and cancel one another; facing the opposite way, they add up to give a maximum. Plotted on a polar diagram, the combined output yields a cardioid or heart-shaped diagram, compared with the ambiguous figure-eight of the loop alone and the omnidirectional result with the vertical aerial alone—Fig. 1.

Essentially the same principle is used in the reflectometer. Whether the line has a central conductor (coaxial) or not (waveguide), the space inside is being swept by the electric and magnetic fields which together make up the electromagnetic wavetrain conveying the power along it. Now we know (if we don't, we shall in a moment) that the directions of these two component fields and the direction in which the waves are travelling are all three at right angles to one another. This is a thing that fairly shouts to be illustrated by an animated diagram in colour, but with a little imagination Fig. 2 should convey the essential facts. The invisible fields are here represented in the usual way by "lines of force," which are fair enough so long as they don't give anyone the idea that the lines really exist or that the fields act only along the lines and not in the spaces between. The electric field is represented by continuous lines and the magnetic field by broken ones. And, of course, the directions marked are those established arbitrarily by convention. The direction of wave motion is at right angles to both sets of lines, (a) towards or (b) away from you.

The novice might ask why it should be the magnetic field that has been reversed in (b). The answer is that it would be equally correct to show reversed electric field. At any given place along the line, these two alternatives alternate at the frequency of the waves. And at any given instant of time, the same two alternatives alternate at

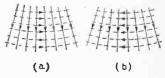


Fig. 2. The continuous lines represent the electric field, and the broken lines the magnetic field, t gether making an electromagnetic wave travelling at right angles to them, (a) towards you, and (b) away from you.

half-wavelength distances along the line. If the directions of *both* fields in either (a) or (b) were reversed it would not affect the direction of wave motion.

Another fundamental point is that in so far as power is being conveyed either way the two fields are in phase with one another.

What the problem boils down to, then, is to find which way across the cable or waveguide the magnetic field is directed, relative to the direction of electric field. If one places oneself so that the electric field

^{*} Incidentally, why is the term "v.s.w.r." (voltage standing-wave ratio) so often used where there is no point in emphasizing voltage particularly? The current ratio is the same, so why not just

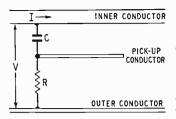


Fig. 3. The basic principle of a reflectometer is the coupling of a pick-up conductor to both electric and magnetic fields in such a way that the two effects cancel out for either direct or reflected wave as desired.

at any instant is downward, then if the magnetic field at the same instant is from left to right the waves are coming towards one; if from right to left, they are going away.

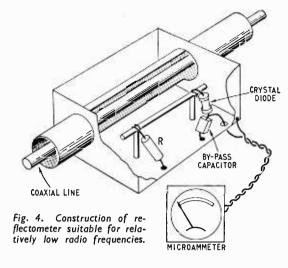
A reflectometer must be made so that it responds to both fields equally. Then if these responses add up to give a double measure from a wave going wholly one way, they will cancel out and give no response at all to a wave going wholly the opposite

The essential feature of all reflectometers, then, is a device for responding simultaneously to electric and magnetic fields. It has appeared in two main forms. In one (which includes the Monimatch) there is a short length of rod or wire fixed parallel to the inner conductor of the coaxial line, so that it is magnetically coupled to the said inner conductor and a.c. flowing therein generates an e.m.f. in it. Being located in the electric field between inner and outer conductors, it also has an e.m.f. between it and the outer conductor. This pick-up rod is dimensioned and connected so that the two e.m.fs equally operate on a suitable indicator, either in phase or 180° out of phase.

Constructional details vary, but they have to take account of the fact that the e.m.f. induced by magnetic coupling is proportional to the mutual inductance and the rate of change of current flowing in the "primary." So its peak is displaced 90° from the current peak. And because the voltage between the two line conductors is in phase with the current through them, the pick-up circuit must be arranged to give a 90° phase shift between the line voltage and the resulting voltage fed to the indicator. The usual way of doing this is shown in Fig. 3. The pick-up conductor is connected to the inner conductor through a small capacitance C-its selfcapacitance is usually enough—and to the outer conductor through a resistance R which is very low in comparison with the reactance of C. The phase of the current driven by V through C and R in series is therefore determined almost entirely by the reactance, so it leads V by nearly 90°. So it is nearly in phase, or 180° out, with the e.m.f. induced in the pick-up conductor by I. Provided that the length and spacing of the pick-up are right, these two voltages are equal, so the voltage between the pick-up and outer conductor is a measure of the power travelling along the line in one direction only. Power in the other direction makes the voltages cancel out, to give no reading.

It will be obvious that Fig. 3 is rather too theoretical. For one thing, if C is self-capacitance it will be distributed all along the pick-up. And what about the indicator?

Fig. 4 shows a practical design for 4–15 Mc/s, described by O. Norgorden in U.S. Naval Research Laboratory Report No. 3538 of 1949. The coaxial line has half of the outer conductor cut away



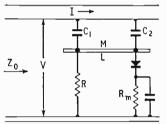
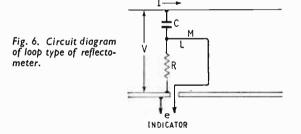


Fig. 5. Approximate equivalent circuit diagram of the Fig. 4 type of reflectometer.



for a distance which is a small fraction of the wavelength. The screening is maintained by a surrounding metal box, inside which is installed the pick-up unit shown, differing from Fig. 3 only in C being distributed over the whole length of the conductor, and the addition of a crystal diode and microammeter as an indicator. R is of the order of 100Ω , and the resistance R_m of the meter is chosen to give it a suitable range in relation to the r.f. power used. For purposes of analysis the distributed capacitance between pick-up and inner conductor is assumed to be concentrated at the points where the connections are made, as in Fig. 5. This assumption seems to be justified in practice. So as not to interrupt ourselves with a lot of algebra at this point, the working has been exiled to an appendix. The upshot of it all is that the condition for no meter reading, when all the power is flowing in the direction causing the inductive and capacitive responses to oppose one another, is $M \simeq RCZ_o$

where M is the mutual inductance between the

For instance, over a 2:1 frequency band the lowest frequency gives a response 6 dB tess than the highest. And there is loss in the pick-up. Allowing, say, a total of 75 dB, the reflected-wave voltage picked up from 10 watts with a s.w.r. of 1.05 in a 70- Ω line comes out at 4.7mV. This is really too small for a

crystal and microammeter.

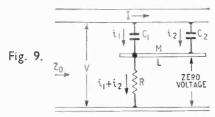
A sensitive measuring gear using a radio receiver is shown by W. H. Elkin in Marconi Instrumentation, Dec. 1956. Fig. 8 is fairly self-explanatory. It can be worked with an oscillator uncalibrated for output level, but does call for a calibrated piston attenuator with which to measure the ratio of direct to reflected power giving equal receiver output. Alternatively if an output-calibrated signal generator is available the input can be varied by its attenuator to get equal receiver response with the reflectometer loop in its two set positions.

Anyone whose thirst for information on reflectometers, especially their practical details, has not yet been slaked should refer to this and other literature I have mentioned, and perhaps also are earlier paper by H. R. Allan and C. D. Curling, in *Proc. I.E.E.*, Jan. 1949, which deals particularly with 10cm.

waveguide technique.

APPENDIX

If the reflectometer is properly made it will give zero reading with power flowing along the line in one direction. In that case the voltage across the detector in Fig. 5 is zero and that arm of the circuit can be omitted, as in Fig. 9. The voltages ac vss C1 and R must add up to V;



the voltage across C2 must be the same; and the voltage across the rod and that across R must add up to zero:

$$\frac{i_1}{j\omega C_1} + (i_1 + i_2)R = V . . . (1)$$

$$\frac{i_2}{j\omega C_2} = V (2)$$

$$(i_1 + i_2)R + i_2j\omega L - Ij\omega M = 0 . . . (3)$$

$$V = IZ_0 (4)$$
If is found from (2) is substituted in (1),

$$\frac{i_2}{i_W C_0} = V \dots \dots (2)$$

$$(i_1 + i_2)R + i_2j\omega L - Ij\omega M = 0..$$
(3)

$$V = IZ_0 (4)$$

The value of i_2 found from (2) is substituted in (1), from which i_1 is found, and both are substituted in (3), where I is replaced by V/Z_0 from (4). After a bit of from which I_1 is replaced by V/Z_0 from where I is replaced by V/Z_0 from X_0 , manipulation this yields $\frac{M}{Z_0} = R(C_1 + C_2 - C_1\omega^2 LC) + j\omega \left(C_2 L - C_1 R \frac{M}{Z_0}\right) ... (5)$ To make this possible, the "j" term must be zero; i.e., $C_2 L = C_1 R \frac{M}{Z_0}$ or $\frac{M}{Z_0} = \frac{C_2 L}{C_1 R}$ The ratio of the reactance of I.

$$\frac{M}{Z_0} = R(C_1 + C_2 - C_1 \omega^2 LC) + j\omega \left(C_2 L - C_1 R \frac{M}{Z_0}\right) ... (5)$$

$$C_2 \mathbf{L} = C_1 \mathbf{R} \frac{M}{Z_0}$$
or
$$\frac{M}{Z_0} = \frac{C_2 \mathbf{L}}{C_1 \mathbf{R}}$$

to the reactance of C_2 , and in practice this is much less than 1. (In other words, the pick-up is much too small to resonate at the working frequency). So $C_1\omega^2LC_2$ can be neglected in comparison with $C_1 + C_2$. With these amendments, (5) boils down to $\frac{M}{Z_0} = \frac{C_2 L}{C_1 R} \simeq R(C_1 + C_2)$

$$\frac{M}{Z_0} = \frac{C_2 L}{C_1 R} \simeq R(C_1 + C_2) \qquad .. \quad (6)$$

This states how the circuit must be proportioned if the

reflectometer is to ignore waves travelling in one of the two directions through it. When the position of the detector, or the loop itself, is reversed, M in (3) is reversed in sign and a reading is given. The current in the detector circuit is then proportional to the current in the line (1) and therefore to the square root of the r.f. power. (There is, of course, mutual inductance between the pick-up and outer conductor, so M is really the coupling to the inner conductor minus that to the outer.)

The type of instrument shown in Fig. 4 is reasonably well covered by (6) if $C_1 = C_2$ and $C_1 + C_2 = C_3$

simplifying it to

$$\frac{M}{Z_0} {=} \frac{L}{R} {\simeq} RC$$

In the loop type with capacitance concentrated mainly at one end (Fig. 6), on the other hand, $C_2=0$ and $C_1=C$; adapting (5) accordingly gives

$$\frac{M}{Z_0} = \frac{RC}{1 + j\omega CR}$$

and since R is made much smaller than the reactance of C, jωCR ≤1 aird

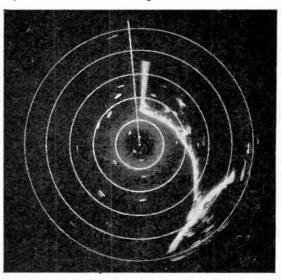
 $\frac{M}{Z_0} \simeq RC$

which is the same as with Fig. 5 except that there is no stipulation about L. Note that within the limits of the approximations the balar, ce condition is independent of frequency, so the setting up is effective over a wide frequency band.

Secondary Radar For Marine Use

THE correlation of radar paints and a navigation chart is eased by the use of secondary radar, as transponder-carrying "targets" will give a strong identified response. RACON is one such transponder beacon for marine use: it produces, when triggered by a ship's radar transmission, a chain of impulses corresponding to 25 to 20 dots extending over the equivalent of four miles of the p.p.i. screen on the bearing, and starting at the range, of the beacon-equipped object. The transponder sweeps the allocated band of frequencies once in 75 sec during 5-min operating periods between 5-min intervals. RACON is made by Kelvin & Hughes (Marine) Ltd. to a Trinity House specification and is, at the moment, undergoing trials.

Response (to right of radial marker) from RACON transponder at about two-miles range.



WIRELESS WORLD, MARCH 1960

inner and pick-up conductors, $C=C_1+C_2$, and Z_0 is the characteristic impedance (=V/I) of the line.

A procedure for achieving this condition is to terminate the line with an accurately matched load and feed it with power. The spacing between conductors (which determines M and C), or R, is then varied to give zero reading.

The instrument having been set up correctly, any reading indicates reflected power. To indicate direct power for comparison, the mutual inductance must be reversed in sign. This could obviously, but inconveniently, be done by reversing the whole reflectometer in the line. The same effect is produced by interchanging R and the indicator. In the Monimatch it is done by having two reflectometer units and switching the microammeter from one to the other. Details of a construction in which the pick-up wires are mounted end-to-end, with a common resistor R in the middle, are given by L. G. McCoy in Q.S.T., Oct. 1'956, and in the 1957 A.R.R.L. Handbook, p.51%. A more compact version, with separate wires lying head-to-tail on opposite sides of the inner conductor, is described in Q.S.T., Feb. 1957, and in the 1958 Handbook, p.530. A still more compact version, in which the first arrangement is adapted to a length of flexible coaxial cable wound into a hank, and balance is obtained by varying R, is shown in the 1959 Handbook, p.530.

The other main variety is the rotatable loop type, which is tending; to supersede the foregoing, presumably because it is more suitable for higher

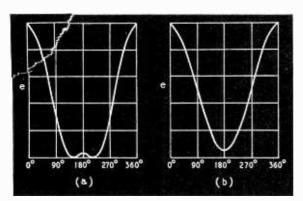


Fig. 7. Variation of response as the loop in Fig. 6 is rotated through one whole turn, with the magnetic coupling (a) too tight, (b) too loose, compared with the electric.

Fig. 8. Diagram of apparatus using a reflectometer for measuring standing-wave ratio and hence the impedance of line terminations.

frequencies and is applicable to waveguides. It is also easier to adjust. But the differences are more mechanical than electrical. Fig. 6 shows that electrically it is essentially the same as Fig. 5, except that the loop may be shaped to put extra capacitance at one end. The algebra used for Fig. 5 can easily be adapted for Fig. 6 by putting $C_2=0$ and $C_1=C_2$, and the result is the same, apart from quantities small enough to neglect. Mechanically, the loop is arranged so that it can be rotated from outside through 180°. So its kinship with the d.f. aerial is more obvious.

This continuous rotability makes it easier to tell whether the capacitive and inductive couplings are equal. If C is too small or too large in relation to M, either there will be two zero readings each side of the position where the loop lies along the line, as in Fig. 7(a), or no zero at all, (b). These diagrams, incidentally, are just cartesian versions of what come out as cardioids in the polar form (Fig. 1).

Besides being proportioned for equal capacitive and inductive e.m.is, a reflectometer pick-up device must exten a along the line for only a small fraction of a wave-length—which means a very small loop in celtin letre waveguides—and not be large enough to cause appreciable reflection or absorption of the transmitted power itself. At the same time it must be sensitive enough to indicate small amounts of reflected power.

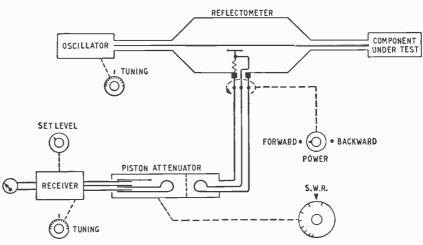
Suppose, for example, that standing-wave ratios at least down to 1.05 are to be measurable. That

$$\rho = 1.05 = \frac{V + v}{V - v}$$

where ρ is the s.w.r. and V and v the voltages of the direct and reflected waves. From this we get

$$\frac{v}{V} = \frac{\rho - 1}{\rho + 1} = 0.0244$$

The ratio of reflected to direct power is equal to the square of this, 0.0006. The power taken by the reflectometer ought not to be more than a like fraction of this power, or 0.00000036; in other words, a loss of at least 65 dB between the lowest power to be monitored and that available for the indicator. If the reflectometer is to be used over a band of frequency there is a further loss, because the voltages picked up are proportional to the rate of change, and hence the frequency, of the wave voltage and current.



THE SMITH CHART

By R. A. HICKSON*

3.—Matching Transmission Lines to Aerials and Uses of Stubs

(Concluded from page 85 in the February issue)

CONTINUING with our study of the applications of the Smith Chart we will now consider one of the most common requirements, namely matching transmission lines to aerials and how mismatch can be corrected by means of stubs.

Matching a Resistive and a Reactive Impedance.—Normally this means matching a load such as an aerial, to a transmission line, and the problem will be discussed in these terms. Various matching elements may be used, such as: —

(1) Matching stub connected in parallel with the load.

(2) Matching stub which can be located at any point on the line.

(3) Matching stubs which are located at fixed points on the line.

(4) L-networks.

(5) Series elements, such as quarter-wavelength transformers, in conjunction with phasing sections. A phasing section is a length of line between the load and the transformer, used to convert the load impedance to a resistive value. A more practical embodiment of this idea is the slug tuner, which is a quarter-wavelength long metal or dielectric sleeve, sliding on the inner conductor of a coaxial line. The double-slug tuner has two independent sleeves and will match a wide range of impedances. A full description can be found elsewhere?

In order to avoid losses in the stubs, they are normally terminated in a short circuit and their length is adjusted to obtain the required reactance. Capacitive terminations are used on occasion, for example, when the transmission line is also used to supply low-frequency de-icing current to an aerial.

A useful discussion of various matching methods in relation to their performance over a wide band of

frequencies is contained in reference 8.

Stub in Parallel With the Load.—This is best treated on an admittance basis, since two admittances add when connected in parallel. Consider the admittance curve of a simple dipole, resonant at 56Mc/s, Fig. 18. If the tolerable v.s.w.r. is 2, then the bandwidth of the dipole is 5Mc/s. The effect of adding susceptance is to shift each point on the curve along the line of constant conductance. Thus point A, at the intersection with the G=0.5 circle can be moved at 0.5+j0 by the addition of +j0.48. Similarly point B can be moved to 0.5+j0 by the addition of -j0.88. Intermediate points will require correspondingly smaller susceptances to bring them within the "v.s.w.r. = 2" circle.

The susceptance of a short-circuited quarter-wavelength stub, resonant at a frequency between 52Mc/s and 60Mc/s, will provide the required

compensation. In general, the susceptance variation of a simple dipole is such that it can be compensated by a stub of this sort. As is well known, the two elements of a simple dipole may be bent over to form a Vee, imparting some directional properties to the aerial. If the elements are bent further than usual, so that they tend to become parallel, the similarity of the aerial to the quarter-wavelength open-circuited section of transmission line can be readily seen. Its susceptance variations are accordingly such as to be compensated by a quarter wavelength short-circuited section.

The resonant frequency of the stub is found by an approximate method which depends on the fact that the variation of susceptance with frequency is very nearly linear in the region of the resonant frequency. For a frequency change of 60 - 52 = 8Mc/s the normalized susceptance changes by 0.88 + 0.48 = 1.36. The change in frequency for a change in susceptance of 0.48 is therefore

 $8 \times 0.48/1.36 = 2.825$ Mc/s.

The resonant frequency of the stub is, 60 - 2.825 = 57.175Mc/s and the electrical length of this stub at 60Mc/s is $0.25 \times 60/57.175 = 0.2625$ wavelength.

The susceptance of a stub of the same characteristic impedance as the transmission line at 60Mc/s is given by moving 0.2625 wavelength towards the load from the point of infinite susceptance. This brings us to +j0.082. We need +j0.48, therefore the impedance of the stub must be $(0.082/0.48)Z_o$ or $0.171Z_o$. As the chart is normalized to 75 ohms, the stub impedance required is 12.8 ohms. (In Fig. 18 the movement is in the direction "Towards Generator" because the short-circuited end of the stub is electrically further from the generator than the actual load which is to be matched.)

Calculations for the other frequencies may now be carried out on this basis. The procedure is:—

 Calculate the stub length in wavelengths at the required frequency.

(2) Determine from the Smith chart the susceptance of the stub if its characteristic impedance is equal to Z_o .

(3) Multiply this susceptance by 75/12.8 to obtain the susceptance of the 12.8-ohm stub.

(4) Add this susceptance to the aerial by moving the point the required distance along the circle of constant conductance.

The result is as shown in Fig. 18. The bandwidth for a v.s.w.r. of 2 is now 8Mc/s, an increase of 60%. A greater improvement could be obtained by selecting a 50-ohm feeder. This would shift the whole curve downwards, so that a greater proportion of it could be folded into the "v.s.w.r. = 2" circle by adding reactance. This is of course not practical for the television receiving installation but has applications elsewhere.

The only admittances which can be matched by

^{*} Belling and Lee Ltd.

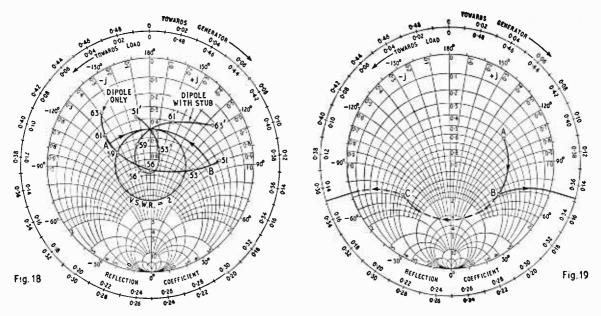


Fig. 18. Use of a stub in parallel with an aerial to increase the useful frequency range.

Fig. 19. Use of a sliding stub to match the load at a single frequency.

this technique are those lying within the crescent defined by the circles of constant conductance which are tangent to the circle of constant v.s.w.r. within which the admittance is required to fall. Such admittances can always be matched at one frequency, but matching over a band may not always be possible. (See Ref. 8.)

Stub at any Point on the Line.—This technique makes use of the fact that the impedance of a mismatched line varies along its length. It can be applied to any load and is in principle carried out in two steps:—

(1) Move along the line away from the load to reach a point at which the conductance is unity.

(2) Add a stub to cancel the susceptance present at this point.

On the Smith chart we plot the normalized load admittance and draw the arc of a circle centred on the centre of the chart, moving clockwise through the load to intercept the circle of unity conductance. There will be two interceptions and normally the one nearer the load is chosen. However, the other may lead to a shorter stub, which may be preferable in some applications. The required stub susceptance will be equal and opposite to that of the line at the point of intersection with the circle of unity conductance.

As an example we may consider a load of 0.4 + j0.6, point A, Fig. 19. Moving 0.062 wavelength along the line brings the admittance to 1 + j1.33. The length of a short-circuited stub having a susceptance of -j1.33 is found by moving along the circle of constant susceptance -j1.33 to the edge of the chart and thence along the edge, i.e. the circle of pure susceptance, to the short circuit, which is the point of infinite susceptance. The stub length is 0.102 wavelength.

The same load may also be matched by moving 0.234 wavelength along the line to bring the admittance to 1-j1.33. The stub length required now is 0.398 wavelength.

The technique is suitable for single-frequency

operation with a constant load, or for experimental work with open-wire lines. For most other purposes, the disadvantage of providing a sliding contact on the transmission line is found to be excessive. The variation in matching with frequency is therefore of little interest.

Matching Stubs at Fixed Positions on the Line.—This system allows a greater range of load impedances to be matched than does the single stub in parallel with the load, and it does not require the provision of sliding or movable contacts on the line. Two stubs will allow most loads to be matched, and the two-stub system will be described in detail. Three stubs will allow all loads to be matched; they are commonly spaced \(\frac{1}{8}\)-wavelength apart and their operation is the same in principle as that to be described for the two-stub system.

One stub is connected in parallel with the load, the other at a distance of { or } wavelength along the line. So far as the remainder of the line is concerned the effective load is formed by the two stubs, the section of line between them, and the load, all taken together. The stub, nearest the generator, therefore, can only be used to cancel susceptance. The stub at the load is adjusted so that the resultant admittance produced by this stub in combination with the load, when transformed by the intermediate section of line, falls on the circle of unity conductance. addition of susceptance by the second stub will then result in a pure conductance of the required value. The admittance of the actual load in parallel with the load-end stub must lie on a point which can be reached by travelling \(\frac{1}{8} \) (or \(\frac{3}{8} \) of a wavelength towards the load from the circle of unity conductance. In other words, it must lie on a circle of the same diameter as the circle of unity conductance, rotated bodily through $\frac{1}{8}$ (or $\frac{3}{8}$) of a wavelength about the centre of the chart. These circles are shown in centre of the chart. Fig. 20(a).

The load admittance, therefore, must have a conductance component which will allow the point

representing it on the chart to be moved on to the transformed unity-conductance circle by the addition of the susceptance of the load-end stub. An examination of the chart shows that this condition can be met for all loads except those within the circle of "conductance = 2," which is shaded. Such loads may be inverted by a quarter-wave section line, i.e. the load-end stub may be placed a quarter-wavelength from the load. In fact, the circle representing the loads which cannot be matched may be placed at any position round the edge of the chart by choosing the appropriate length of line between the load and the load-end stub.

The choice of spacing between the two stubs arises from the facts that (a) any closer spacing than $\frac{1}{8}$ wavelength would lead to field distortion and would effectively place the two stubs in parallel: (b) the use of spacings between $\frac{1}{8}$ and $\frac{3}{8}$ wavelength would increase the area on the chart representing loads which could not be matched. At $\frac{1}{4}$ -wavelength spacing, loads within the circle of unity conductance could not be matched. A spacing of $\frac{3}{8}$ wavelength may be used where $\frac{1}{8}$ wavelength is mechanically inconvenient. Wider spacings than $\frac{3}{8}$ wavelength would offer no electrical advantage; a spacing of $\frac{1}{2}$ wavelength, for example, would be equivalent to placing the two stubs in parallel.

As an example, we will match the load 0.6 + j0.6, point A on Fig. 20(b), using one stub in parallel with the load and the other stub $\frac{3}{8}$ wavelength along the line. The first stage is to select the susceptance of the load-end stub so that the resultant load lies on the rotated circle. There are two choices, points B and C, where the "conductance = 0.6" circle intersects the rotated circle. When transformed through the $\frac{3}{8}$ -wavelength section of line, the admittances become points D and E. From either of these points, the addition of susceptance by the generator-end stub will result in a completely matched system.

L-Network Design.—Matching with an L-network,

using the Smith chart, follows similar lines to matching with a stub of adjustable position. In this case, however, the impedance-transforming property of a length of line is not used, a reactive element being placed in series with the line instead, and accordingly the chart must represent impedance as well as admittance.

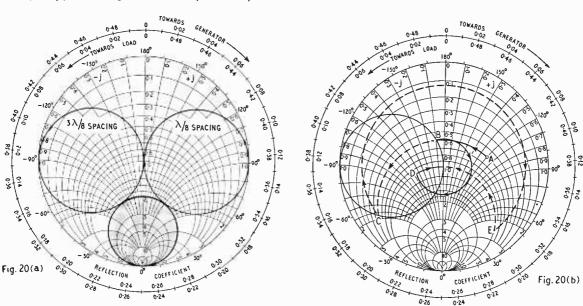
There are two general types of L-network, as shown in Fig. 21. Considering first the type A network, the resultant admittance of $X_2 + Z_L$ must lie on the circle of unity conductance, so that susceptance only need be added to complete the matching process, exactly as in the case of single-stub matching. In other words the resultant impedance of $X_2 + Z_L$ must lie on the circle surrounding area B. To be matched by a type-A network, therefore, the load impedance must be such that, by adding or subtracting reactance, we can reach the circle surrounding area B. The shaded area A on Fig. 21 defines the loads which cannot be matched by a type-A network. Similar reasoning shows that the loads which cannot be matched by a type-B network are those defined by the shaded area B on Fig. 21.

As an example of the design of an L-network, we will match an impedance of 20+j50 ohms to a 50-ohm line with a loss-free type-A network. The procedure can be followed on Fig. 22. The normalized impedance is 0.4+j1.0 (point A). The series element will be a capacitor of reactance equal to $1.49Z_o$ at the operating frequency. This produces a resultant impedance as shown at point B, corresponding to the admittance (relative to 20 millimhos) shown at point C. The addition of inductance of susceptance equal to $1.25Y_o$ at the operating frequency results in complete matching.

If losses in the inductor or capacitor are not negligible an allowance must be made for them. Considering the inductor first, let G = -B/Q so that a susceptance of -1.0 will involve a conductive component of +1/Q. The resultant admittance of load and capacitor combined must now lie on the line

Fig. 20 (a). Limits within which the double-stub matching method can be applied.

Fig. 20 (b). Matching with two stubs spaced at $3\lambda/8$.



Wireless World, March 1960

1-B/Q+jB, where B is the required susceptance. For example, if Q=10, this line will pass through the points 1+j0, 0.9+j1, $0.8+j2.\ldots 0+j10$; it is, in fact, a circle, the centre of which can be found by laying off arcs from three or more points such as those calculated above. The corresponding circle for the resultant impedance of load and capacitor combined is now readily drawn. Any losses in the capacitor may be taken into account without a geometrical construction by remembering that for every unit of reactance (movement along a line of constant

resistance) we must add 1/Q units of resistance. As 1/Q, which is virtually equal to the power factors, is about 0.0005 for a ruby-mica dielectric and about 0.0001 for air dielectric, correction is not often needed for capacitor losses.

The same load (point A) can be matched with an all-capacitor network. The series capacitor should now have a capacitance equal to $0.51Z_o$ at the operating frequency. This produces the resultant impedance shown at point B', corresponding to the admittance shown at point C'. The addition of a

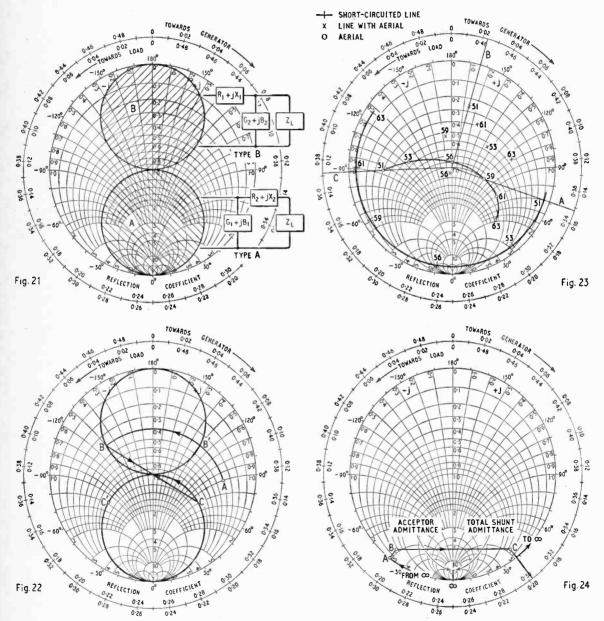


Fig. 21. The two basic types of L-network, showing the load impedances (shaded) for which each is not applicable.

- Fig. 22. Matching with a loss-free type A L-network.
- Fig. 23. Allowing for the effect of the line on the apparent impedance of a dipole at various frequencies.
- Fig. 24. Single-frequency acceptor-circuit wavetrap, with compensation.

shunt capacitor of susceptance equal to $1.25 Y_o$ at the operating frequency results in complete matching (Fig. 22).

Effect of Line Length in Measurements.—The effect of line length on the apparent impedance of an aerial has been discussed by Strafford⁹ whose example, a Channel-3 dipole, will be used here. The

principle applies to any type of load.

The first step is to measure the input impedance of the line with a short circuit in place of the unknown load, at the frequency of interest. The result might be, for example, 0.1-j1.4. As we started with a short circuit, 0+i0, the effect of the line has been to move the point on the chart 0.151 wavelength clockwise and 0.32dB radially inwards. Therefore, when the short circuit is replaced by the unknown load, the apparent impedance must be moved 0.151 wavelength counter-clockwise and 0.32dB radially outwards, to obtain the true load impedance. This procedure must be repeated at each frequency. It must be borne in mind that, while 0.151 wavelength always corresponds to the same angle (109°) on the chart, the radial distance represented by 0.32dB depends on the point at which the "Effect of Line Attenuation" scale is entered and must be evaluated afresh with this scale for each new point.

Point A in Fig. 23 represents the input impedance of the short-circuited line. Point B represents the input impedance of the line with an aerial connected in place of the short circuit. Point C is the actual impedance of the aerial, obtained as described above. The construction is shown for one frequency only, 51Mc/s. At this frequency the effective length of the line is 0.151 wavelength. The point B, representing the input impedance of the line with the aerial at 51Mc/s must therefore be moved 0.151 wavelength counter-clockwise, from 0.477 to 0.128 on the "Wavelengths towards Load" scale. At the same time the loss of 0.32dB in the line is allowed for by moving the point the correct radial distance out-

wards, as described above.

Compensated Coaxial-Stub Acceptor Circuit.— In cases where two frequencies are present on a transmission line, one of which is undesired, this frequency can be considerably attenuated by a coaxial stub which is one half-wavelength long at the undesired frequency and is short circuited at the free end. This stub will "repeat the load" and will short circuit the line at its resonant frequency. It will also present a susceptance across the line at the signal frequency, and this susceptance may be cancelled by a second stub of suitable length.

Certain special cases must be distinguished. If the desired frequency is an integral multiple of the undesired frequency, the stub will short circuit both signals impartially. If, on the other hand, the undesired frequency is exactly twice the desired frequency, the acceptor stub will be a quarterwavelength stub at the desired frequency and no

compensation will be needed.

Suppose that the two frequencies are F_D and F_u . The electrical length of F_D of a stub which is one half-wavelength at F_u will be $0.5F_D/F_u$ wavelengths. The problem is best handled on an admittance basis so we start at the infinity point on the chart and move clockwise round the edge of the chart for the distance $0.5 F_D/F_u$ wavelengths. The point so reached is the susceptance placed across the line at the desired frequency. To reduce this

susceptance to zero, we must add an equal and opposite susceptance. We therefore locate the required point on the chart and move counterclockwise round the edge of the chart to the infinity point. The distance travelled in wavelengths is the length of the required compensating stub. The compensating stub, of course, presents a certain susceptance at the undesired frequency, but this is of no importance, as it is in parallel with a much greater susceptance. The effect of losses in the stubs may be taken into account by using the "Effect of Line Attenuation" scale.

As an example (Fig. 24) we will consider interference at 38Mc/s to a signal at 42Mc/s. The electrical length of the acceptor stub at the signal frequency will be $38/2 \times 42 = 0.5527$ wavelength. Moving this distance from the infinity point in a clockwise direction round the edge of the chart we reach point A, 0 - j2.9. Assuming that the stub has an attenuation of 0.4dB, we move radially inwards by the distance indicated on the "Effect of Line Attenuation" scale to point B, 0.5 - j2.85. The compensating stub must have a susceptance of +j2.85 to cancel this, and will have very nearly the same attenuation. Its admittance is therefore located at point C, which is the mirror image of point B in the pure conductance scale, and its length is found by moving 0.4dB radially outwards to the edge of the chart and thence counter-clockwise to the short-circuit (infinity susceptance) point. The length is 0.447 wavelength.

The admittance of the acceptor stub at the undesired frequency is not infinite, but is reduced by the attenuation of the cable used to make the stub. The actual admittance is found by moving radially inwards by the distance indicated on the "Effect of Line Attenuation" scale. Therefore, to the admittance thus found, 20 + j0, must be added the admittance 0.5 + j2.85, corresponding to point C, making the total shunt admittance due to the pair of stubs 20.5 + j2.85 at the undesired frequency. It will be seen that the effect of the compensating stub at this frequency is negligible. The ratio of the admittance of the cable following the acceptor circuit to the combined admittance of the stubs is 1 to 20.7. The interfering signal power is therefore divided between the two paths in the ratio 428 to 1, corresponding to an attenuation of 26.3dB.

SYMBOLS

 ε Effective relative permittivity (dielectric constant) of the dielectric.

F = Frequency Mc/s.

L = Physical length of line in metres. l = Length of line in wavelengths = $FL\sqrt{e/300} = FL/300v$.

v. = Velocity factor of line = $1/\sqrt{e}$.

Y =Load admittance. $Y_0 =$ Line admittance.

 $y = Normalized load admittance, = Y/Y_a$

Z =Load impedance. $Z_0 =$ Line impedance.

 $z = Normalized load impedance, = Z/Z_o$.

REFERENCES

7 "Transmission and Propagation." Volume 5 of the Services Textbook of Radio, H.M.S.O., 1958, pp. 117-119.

117-119.

8 C. H. Westcott and F. K. Goward. "The Design of Wide-Band Aerial Elements for 500-600 Mc/s Ground Radar." I.E.E. Paper No. 732, Radio Section, Proc.

I.E.B. Volume 96, Part III, No. 39, January 1949, pp.

⁹ F. R. W. Strafford. "Measuring TV Aerial Performance; Part 3, Impedance Measurements," Wireless World, Volume 64, No. 6, pp. 295, 296, June 1958.

ADDENDUM

Since publication of the first two parts of the article it has been pointed out that certain state-

ments and equations need clarification.

On page 7 second column of the January issue the statement that the v.s.w.r. is simply related to the load does not apply where the load comprises both reactance and resistance; S should then equal r not z.

As derived
$$S = \frac{1 + |K|}{1 - |K|}$$

Now
$$K = \frac{z-1}{z+1}$$
 and if $z = r + jx$

$$K = \frac{r - 1 + jx}{r + 1 + jx}$$

Since the modulus of a complex number is equal to the square root of the sum of the squares of the real and imaginary parts:-

$$|K| = \left[\frac{(r-1)^2 + x^2}{(r+1)^2 + x^2} \right]^{\frac{1}{2}}$$

If the load impedance is resistive (x = 0) this simplifies to:-

$$|K| = \frac{r-1}{r+1}$$

So that:
$$-S = \frac{1 + \frac{r-1}{r+1}}{1 - \frac{r-1}{r+1}} = \frac{r+1+r-1}{r+1-r+1} = r$$

The statement at the end of this section that the decibel may be used only when . . . "voltages are developed across identical impedances" should read ... "across impedances having the same resistive component."

In Appendix 1, page 8, January issue, the fourth line from the bottom of the second column should

read:-

$$u^2 - \frac{2ru}{1+r} + \frac{r}{1+r} = \frac{1}{1+r} - v^2$$

while the penultimate line should read:-

$$u^2 - \frac{2ru}{1+r} + \frac{r^2}{(1+r)^2} = \frac{1}{(1+r)^2} - v^2$$

In the February issue page 83, second column, ne 12 the equation "attenuation (dB) = $k\sqrt{\text{frequency}}$ " is given without specifying that k is a constant. It could be re-written "attenuation dBa frequency."

On page 84, first column, line 10, the derivation of the equation for velocity factor is as follows:-The electrical length (l) of the line is given by:—

$$l = \frac{L \dots \text{(wavelengths)}}{\lambda}$$

where:- L is the physical length in metres,

 λ is the wavelength in metres, given by $\lambda =$ 300v/F metres,

300 is the (approximate) velocity of electromagnetic waves in free space in megametres/ second.

v is the velocity factor of the line, F is the frequency in Mc/s.

$$\therefore l = \frac{\mathrm{FL}}{300v} \dots$$
 (wavelengths) and $v = \frac{\mathrm{FL}}{300l}$

Let the electrical length be l_1 at F_1 and l_2 at F_2 .

$$v = \frac{F_1 L}{300l_1} = \frac{F_2 L}{300l_2} = \frac{(F_1 - F_2)L}{300(l_1 - l_2)} = \frac{\delta FL}{300\delta l}$$

In Fig. 15 on page 84 the direction of movement from the short-circuit should be towards the generator as stated in the text (page 85 column 1), indicating a load impedance of 1 + i0.6.

"ELECTRONIC COMPUTERS"

SINCE the first edition of this Wireless World book came out just over three years ago it has been reprinted once and also translated into Russian and republished by the Soviet authorities. Technical developments in the computer field have been so rapid, meanwhile, that a second edition of the book has become necessary, and this is now available from our publishers. The opportunity has been taken to improve the exposition of basic principles which is the main purpose of the book, and it has, in fact, been largely rewritten. Three new chapters, on analogue computer circuits, programming of digital computers and the future development of "intelligent" machines, have been added, together with new illustra-tions throughout the book, which now contains 263 pages, including 32 plates. As before, the new edition covers the principles and applications of both analogue and digital machines, and is suitable as an introduction to the subject for students, technicians or laymen with some knowledge of radio and electronic techniques. "Electronic Computers," by T. E. Ivall, 2nd edition, can be obtained from any bookseller, price 25s, or direct from the publishers, Iliffe & Sons Ltd., at 26s by post.

CLUB NEWS

Birmingham.—A 160-m mobile rally has been organized by the South Birmingham Radio Society for 10.30 a.m. on March 6th, at Lickey Beacon, Rednal. At the club meeting at 9.30 on March 17th at Friends Meeting House, 220 Moseley Road, Birmingham, 12, G. E. Simonite (G3JAO) will speak about the electronic equipment at Birmingham

University.

The March meetings of the Slade Radio Society include a Mullard film show in the Bennett Hall, Y.M.C.A., Snow Hill at 7.45 on the 4th. At the meetings on the 11th and 25th, which will be held at 7.45 at Church House, High Street, Erdington, the subjects of v.h.f./f.m. reception and electronic digital computers will be dealt with respectively.

At the meeting of the Midland Amateur Radio Society on March 3rd D. Edwards (G3DO) will talk on DX working, and at the meeting on the 15th, R. Rew (G3HAZ) will deal with v.h.f. reception. The society meets at 7.0 at the Midland Institute. Paradise Street.

Midland Institute, Paradise Street.

Bradford.—J. C. Belcher (G3FCS) will discuss interference in relation to sound and vision reception at the March 8th meeting of the Bradford Amateur Radio Society. The club meets on alternate Tuesdays at 7.30 at Cambridge House, 66 Little Horton Lane.

Cleckheaton.—Dr. N. H. Chamberlain, of Leeds University, will speak on electronics in industrial research at the meeting of the Spen Valley Amateur Radio Society at 7.30 on March 30th at the George Hotel.

Leeds.—The month's meetings of the Leeds Amateur Radio Society include lecture-demonstrations of photo-electric devices by E. Sollitt, the president (2nd), of hi-fi equipment by Fane Acoustics (16th), and a home-built table-top transmitter by W. Ripley (23rd). Meetings are held at 7.45 at Swarthmore Education Centre, 4, Wood-bours Scurre house Square.

News from the Industry

J. Langham Thompson Group, which includes Datum Metal Products Ltd. and Automation Systems & Controls Ltd., was recently acquired from the Camp Bird Group by Ether Ltd. of Erdington, Birmingham. The title of the company holding the entire share capital of Ether Ltd., its wholly owned subsidiary Electro Methods Ltd., and the J. Langham Thompson Group is now Ether Langham Thompson Ltd. The chairman and managing director is F. B. Duncan. Other members of the board are C. E. Blunt and F. Coulling (directors of Ether Ltd.), and J. Langham Thompson and Rear Admiral Sir Philip Clarke (directors of the J.L.T. Group).

The Decca Record Company, the parent company of the Decca group, records a trading profit of £3,305,313—an increase of £352,774 over the previous year. The net profit of £1,031,205 was £100,221 up on last year. Exports during the year reached the record total of £5,860,000 including £1,380,000 to the U.S.A. and Canada. The chairman, E. R. Lewis, stated that during the ten years to last March, the company had produced 41M long-playing records, 150M 78 r.p.m. records and 33M 45 r.p.m. records.

I.C.T.—The trading profit of International Computers and Tabulators Group up to September 30th, which includes a year's operation of B.T.M., but only nine months of Powers-Samas, was £2,328,000. The joint figure for the previous year's operation of the two former groups was £2,020,000. Taxation absorbs £1,025,000 of last year's profit.

S.T.C. are developing tunnel diodes and in order that circuit designers may familiarize themselves with the new properties and obtain early experience with the circuit performances achievable, the company is offering sample devices. Details of the samples and of the proposed range of diodes for use at v.h.f. are obtainable from Standard Telephones and Cables, Ltd., Transistor Division, Footscray, Kent.

B.T.H. Sound Equipment Ltd.—Under the general reorganization scheme of the A.E.I. Group the name of this company has been changed to A.E.I. Sound Equipment, Ltd. The company's office is still at Crown House, Aldwych, London, W.C.2.

Radio Resistor Co., wholesale distributors of Morganite products, are assuming exclusive wholesale distribution of Electrosil resistors, made by James A. Jobling & Co., of Sunderland. The company's new address is 9-13 Palmerston Road, Wealdstone, Middx. (Tel.: Harrow 6347.)

KOVO of Czechoslovakia.—Nash and Thompson were recently appointed agents in the U.K. and certain other countries for KOVO, the organization for marketing Czechoslovakian instruments. In their announcement (page 490 of our November issue), Nash and Thompson stated in error that they were agents for the Commonwealth, whereas in Australia, Jacoby, Mitchell & Co., of Sydney, have, in fact, been the exclusive agents for some years.

Multisignals Ltd., the recently formed television relay company, announces that Ultra have joined the enterprise, with which Ekco and Thorn are already associated. A. V. Edwards, managing director of Ultra, has been appointed to the board of Multisignals.

Hacker Radio, Ltd., formerly a sales company, is being developed by R. H. Hacker and A. G. Hacker, until recently joint managing directors of Dynatron Radio, Ltd., to manufacture high-quality radio and electronic equipment. A new factory is being built at Norreys Drive, Cox Green, Maidenhead.

Non-destructive Test Equipment.—An agreement has been concluded between Kelvin & Hughes Ltd., and the Curtiss-Wright Corp., of America, for the exclusive right to manufacture and sell the non-destructive testing equipment of the other partner. This includes instruments for the inspection and measurement of materials, and industrial ultrasonic equipment. Kelvin Hughes will represent both interests in the U.K., the Commonwealth (except Canada) and Europe, and Curtiss-Wright in North America.

Avo and Taylor.—Changes in the boards of Avo and Taylor Electrical Instruments, both members of the Metal Industries Group, have been announced. Because of indisposition J. H. Rawlings has relinquished the position of managing director of Avo Ltd., but he is remaining on the board as deputy chairman, and also on the board of Brookhirst Igranic Ltd. Succeeding him as managing director of Avo is S. R. Wilkins who also joins the board of Taylor Electrical Instruments. E. Strauss has been appointed director and general manager of Taylor's and also joins the board of Avo.

EXPORT NEWS

U.S.-Bahamas forward-scatter link, providing 72 trunk telephone circuits, which came into operation in January, employs at the Nassau terminal equipment manufactured by Standard Telephones & Cables. At Nassau two 10-kW transmitters, operating in the 2,000-Mc/s band, feed into two 30-ft diameter paraboloids. Two dual-diversity receivers provide quadruple diversity reception. S.T.C. also supplied line-of-sight microwave equipment linking the forward-scatter station to the telecommunications centre in Nassau City.

Sweden.—The first of three Decca D.A.S.R.1 air surveillance radar installations for use at major Swedish airports has been installed at Arlanda, Stockholm's new airport. The D.A.S.R.1 employs duplicate 800-kW transmitters feeding into separate aerial reflectors which are mounted back-to-back on top of a 90-foot cylindrical tower. The beams are staggered in elevation. The transmitters operate on about 3,000 Mc/s in the S-band.

Colour television equipment, including the latest E.M.I. camera (using three vidicon tubes) and control equipment and a Rank-Cintel large-screen colour projector, has been exported to China. The equipment forms a complete closed-circuit installation.

Computers.—International Computers and Tabulators Ltd. have received orders from three commercial concerns in France for their medium-sized general-purpose computer, the I.C.T. Type 1202. The capital value of the machines and ancillary equipment is of the order of £60,000 in each case. Last October SAMAS, previously the French subsidiary of Powers-Samas Accounting Machines, became I.C.T.—France.

V.H.F. equipment manufactured by Pye Telecommunications, is to be installed by the Suez Canal Authority for communication with its vessels in the Canal.

Cyprus.—Equipment for the sound reinforcing and language interpretation systems in the new House of Representatives in Cyprus is being installed by Tannoy.

India.—Sundeep Electronics Corporation, of New Delhi, wish to get in touch with British manufacturers of components and equipment with a view to representing them in India and possibly undertaking the manufacture of certain items under licence. Their technical representative is visiting this country and can be contacted at 8, Crescent Road, London, N.8. The firm's headquarters are at Gurdwara Road Crossing, Karol Bagh, New Delhi, 5.

East Pakistan's port of Chittagong is to be equipped by Marconi's with v.h.f. radio-telephone transmitters and receivers at two base stations. Also mobile installations are being supplied for harbour tugs and a pilot launch. The order was placed by the Port Commissioners with International Industries Ltd., of Karachi, Marconi's agents.

A linear accelerator designed for X-ray treatment of deep-seated tumours, has been ordered from Mullard by the Soviet Union for installation in Moscow.

Western Germany.—A. B. Metal Products Ltd., manufacturers of Clarostat controls and resistors, whose new address is Walkden House, Melton Street, Euston Square, London, N.W.1, have appointed Heinz Michalski, Myliusstrasse 54, Leverkusen 3/Rhld, to represent them in Western Germany.

Anglo-Canadian Agreement.—Television broadcasting equipment manufactured by E.M.I. will be distributed and serviced in Canada by the Canadian General Electric Co.

MARCH MEETINGS

Tickets are required for some meetings: readers are advised therefore to communicate with the secretary of the society concerned.

LONDON

1st. I.E.E.—"Digital computer developments at Manchester University" by Dr. T. Kilburn (with supporting papers) at 5.30 at Savoy Place, W.C.2.

papers) at 5.30 at Savoy Place, W.C.2.

1st. Association of Supervising Electrical Engineers.—"Silicone rectifiers"
by D. R. Coleman (S.T.C.) at 7.30 at Windsor Castle Hotel, 134 King Street, Hammersmith, W.6.

3rd. Brit.I.R.E.—"Time sharing in on-line computer applications" by A. St. Johnston at 6.30 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1.

4th. I.E.E.—Medical Electronics Group discussion on "Direct writing

Group discussion on "Direct writing oscillographs" opened by A. J. Smale and S. N. Pocock at 6.0 at Savoy Place,

-" An introduction to the 7th. I.E.E.theory of masers with particular reference to the travelling-wave maser" by Dr. P. N. Butcher at 5.30 at Savoy Place, W.C.2.

7th. Society of Instrument Technology.—"Data reduction for guided weapon trials at Aberporth" by A. S. Younger, G. C. Morgan and E. S. Mallett at 7.0 at Manson House, 26 Portland Place, W.1.

9th. Association of Supervising Electrical Engineers.—"Cathode-ray tubes" by C. H. Gardner (Mullard) at 7.45 at Eltham Green School, Queenscroft Road, Eltham, S.E.9.

10th. Physical Society.—"Electronic music" by Dr. J. Bowsher at 5.30 at Imperial College, Prince Consort Road,

11th. Television Society.—"Television receiver production" by S. T. Palmer (G.E.C.) at 7.0 at the Cinematograph Exhibitors' Association, 164 Shaftesbury Avenue, W.C.2.

14th. I.E.E.—Discussion on "Should engineers be encouraged to take up administrative positions early in their careers?" opened by K. E. Greene at 6.0 at Savoy Place, W.C.2.

14th. Brit.I.R.E.—Discussion on

"Short-range navigational aids" at 6.30 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1.

15th. British Computer Society.-"Axis transformation using analogu analogue equipment" by N. Doveton (E.M.I.) at 2.30 at the Northampton College of Advanced Technology, St. John Street,

I.E.E .- "Fast-response sistor chopper-type amplifier with low carrier frequency" by I. C. Hutcheon and D. Summers at 5.30 at Savoy Place, 18th. B.S.R.A.—"Problems of reviewing tuners" by R. S. Roberts at 7.15 at the Royal Society of Arts, John Adam Street, W.C.2.

23rd. I.E.E.—"The challenge of the propagation medium to the radio engineer" by G. Millington at 5.30 at Savoy Place, W.C.2.

24th. Brit.I.R.E.—"The continuous recording of heart activity" by Dr. I. Boyd and W. R. Eadie at 6.30 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1.

24th. Television Society.—" Operating a commercial television station in the U.S.A." by J. H. Battison (Associated U.S.A." by J. H. Battison (Associated Rediffusion) at 7.0 at 164 Shaftesbury Avenue, W.C.2.

25th. R.S.G.B.—"High-fidelity sound reproduction for the amateur" by H. A. M. Clark at 6.30 at the I.E.E., Savoy Place, W.C.2.

28th. Radar & Electronics Association.—"Computers: some design problems" by Peter D. Hall (Ferranti) at 7.30 at the Royal Society of Arts, John Adam Street, W.C.2.

29th. Society of Instrument Technology.—"Application of transistors in instrumentation" by G. G. Bloodworth at 7.0 at 26 Portland Place, W.1.

Brit.I.R.E.—" Silicon photovoltaic cells for instrumentation and control applications" by Dr. V. Magee and Dr. A. A. Shepherd at 6.30 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1.

31st. I.E.E.—The Fifty-first Kelvin Lecture on "Cosmic radiation" by Professor C. F. Powel at 5.30 at Savoy Place, W.C.2.

ARBORFIELD

21st. I.E.E. Graduate and Student Section.—"Inertial navigation systems" by Wing Commander E. W. Anderson at 7.0 in the R.E.M.E. Assembly Hall.

16th. I.E.E.—"The reliability and life of impregnated-paper capacitors" by J. P. Pitts at 7.0 at the Arcadian Restaurant.

8th. I.E.E.—"The recognition of moving vehicles by electronic means" by T. S. Pick and A. Readman at 6.30 at the David Keir Building, Queen's University, Stranmillis Road.

BIRMINGHAM

8th. Brit.I.R.E.—"Transistor power amplifiers" by F. Butler at 7.15 at the Matthew Boulton Technical College.

11th. Society of Instrument Technology.—"Reading with electronics" by I. Merry at 7.0 in the Lecture Theatre of the Byng Kendrick Suite at The Gosta Green College of Technology, Aston Street.

28th. I.E.E.—"Television on tape" by W. Silvie at 6.0 at the James Watt Memorial Institute.

BRISTOL

Brit.I.R.E.—" Training operating and maintaining television studio broadcasting equipment" by Dr. K. R. Sturley and A. E. Robertson at 7.0 at the School of Management Studies, Unity Street.

BRADFORD

10th. Association of Supervising Electrical Engineers.—"H.F. heating" by M. R. Padget (Radyne) at 7.30 at The Midland Hotel.

16th. Brit.I.R.E.—" Recent developments in printed and potted circuits" by H. G. Manfield at 6.30 at the Welsh College of Advanced Technology.

CATTERICK

29th. I.E.E.—"Progress on prob-lems in ionospheric propagation during the International Geophysical Year" by W. R. Piggott at 6.15 at Headquarters I.E.E.—" Progress on prob-Mess, School of Signals, Catterick Camp.

CHELTENHAM

4th. Brit.I.R.E.—"The use of radio aids in the control of modern transport aircraft" by K. Fearnside at 7.0 at the North Gloucestershire Technical College.

COLCHESTER

11th. Institution of Production Engineers.—"Numerical control of machine tools from the production engineer's point of view" by O. S. Puckle at 7.30 at Britannia Works Canteen, Davey Paxman & Co., Ltd.

DUBLIN

10th. I.E.E.—"Aviation communications and navigational systems" by G. E. Enright and G. Jones at 6.0 at the Physical Laboratory, Trinity College.

EDINBURGH

25th. Brit.I.R.E.—"Radio guidance in the automatic landing of aircraft" by J. Shayler at 7.0 at the Department of Natural Philosophy, the University, Drummond Street.

4th. Society of Instrument Technology.—"Transistors as applied to control equipment" by R. J. Miles at 5.30 at the Administration Building, Esso Refinery.

GLASGOW

24th. Brit.I.R.E.—" Radio guidance in the automatic landing of aircraft" by J. Shayler at 7.0 at the Institution of

Wireless World, March 1960

Engineers and Shipbuilders, 39 Elmbank Crescent.

8th. I.E.E.—Faraday Lecture on "Electrical machines" by Professor M. G. Say at 7.30 at the Town Hall.

31st. I.E.E.—"Recent developments in colour television" by I. J. P. James at 6.30 at the Y.E.B. Offices, Ferensway.

LEEDS

1st. I.E.E.—"The transmission of news film over the transatlantic cable" by C. B. B. Wood and I. J. Shelley at 6.30 at the Leeds and County Conservative Club, South Parade.

7th. Association of Supervising Electrical Engineers.—"Two-channel stereo sound systems" by D. Humphries and F. H. Brittain (G.E.C.) at 7.30 at the Great Northern Hotel.

Great Northern Hotel.

LIVERPOOL

10th. I.E.E.—Faraday Lecture on "Electrical machines" by Professor M. G. Say at 6.45 at the Philharmonic

MALVERN

MALVERN
7th. I.E.E.—"Masers" by Dr. M. H.
Oliver at 7.30 at the Winter Gardens.
29th. Brit.I.R.E.—"Microwave propagation" by M. W. Gough at 7.0 at
the Winter Gardens.

MANCHESTER

3rd. Brit.I.R.E.—"Electronics in oceanography" by M. J. Tucker at 6.30 at Reynolds Hall. College of Technology, Sackville Street.

16th. I.E.E.—"Data processing" by J. C. Gladman at 6.15 at the Engineers' Club, Albert Square.

21st. Institute of Physics.—"Astronomical and atomic time" by Dr. L. Essen (N.P.L., Teddington) at 7.0 at the University.

NEWCASTLE-ON-TYNE

Brit.I.R.E.—" Silicon 9th. photovoltaic cells in instrumentation and control" by Dr. A. A. Shepherd at 6.0 at the Institution of Mining and Mechanical Engineers, Neville Hall, Westgate Road.

14th. I.E.E.—Discussion on "Component reliability" at 6.15 at the Rutherford College of Technology.

21st. Institution of Production Engi-eers.—"Electronic control mechanism" neers.—"Electronic control mechanism" by O. S. Puckle at 6.45 at the Devonshire Room, Grand Hotel.

WORKINGTON

8th. I.E.E.—"The reliability and liber of impregnated-paper capacitors" by J. P. Pitts at 7.0 at Workington College of Further Education.









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RANDOM RADIATIONS

By "DIALLIST"

The Colour Mix-up

QUITE a bit more evidence has come my way about different kinds of apparatus imported from West Germany, whose triple flex leads don't conform to our standards. One kind reader reports electric clocks with the red wire connected to the metal cases. From others and from dealers whom I've visited have come tales of spin-driers, food-mixers and other domestic gear with red earth wires. I therefore make no apology for mentioning the matter again. The risk is great, for if any of this apparatus were connected according to British conventions to a 3-pin plug, a severe shock would inevitably be received by anyone who touched the casing. And that's just what people would do when the thing didn't function—as, of course, it wouldn't with its "works" connected across neutral and earth. It's almost instinctive when nothing happens after switching on to get hold of whatever it may be to see what's amiss. I do beg anyone who contemplates buying apparatus of foreign origin to make sure that it's not potentially lethal by getting the dealer to test for continuity between the green wire and the casing. If possible, he should insist on seeing it tried out in the shop.

A Spectacle Question

AN optician, I read, has suggested that special glasses are advisable for watching TV. If he's correctly reported, he holds that the normal viewing distance is about six feet and that that doesn't fit in with either reading or distance glasses. The former normally focuses at about 14 inches; the latter's name is selfexplanatory. I suppose that the screen most widely used today is the 17-inch and the minimum viewing distance for that is a good bit over six feet. I've worn glasses for some years now and my experience is that so long as you don't sit too close to the screen, distant lenses give you as good a picture as you could wish for from a 405-line system. Have your eyes too close to the screen and you get undue lininess. Anyone who wants to watch a 17-inch screen with

glasses focusing sharply at six feet needs to have not his eyes, but his head examined!

Piped TV

THE G.E.C., I see, is to take a financial and technical interest in General Piped Television, which intends to provide services in places where poor reception is obtained in the ordinary way. One of the strong points of this and similar schemes is that any make of receiver can be used. "The new service," it was said, "finishes at the plug-in point." I've always been in favour of piped TV for several good reasons. First, it does away with the necessity for aerial arrays, which can't be claimed by anyone to add to the good looks of houses. Then, it means that good reception is assured. It's a great thing to be sure at all times of a good picture, free entirely from the effects of interference. And I do like the idea that you can choose any kind of receiver you care for, for that means a free hand not only for set owners, but also for manufacturers and dealers. Piped TV has expanded rapidly since it first came along, though there has been some opposition to systems requiring special receivers. I forecast that "any-set" piped services will spread over much of this country in the next few years.

We Did It First

THE Germans are a little late with their claim that the Medizinische Tonbandzeitung (Tape - recorded Medical Journal) conducted by the Kongressgesellschaft für Arztliche Fortbildung, or congressional society for medical advanced studies, is the first of its kind in this field. Our British College of General Practitioners has been running a taperecorded service for its members for some time now. The college organizes lectures and discussions to keep G.P.s up to date with the latest developments. These are normally held in London, and any country doctors who are unable to attend them can obtain the tape-recordings on loan. It would be no bad thing if some other institutions adopted a similar system, for many of their members can't always manage to get to meetings that they'd like to attend and might appreciate such a service.

Gale Casualties

ON the East Coast this winter we've had even more than our normal ration of gales and casualties among television and v.h.f. aerial arrays, especially those of the less-robust makes, have been many. From the window of the room where I'm writing I can see three which have suffered. One looks like a com-

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plete write-off, another has lost one half of its director and the opposite half of its reflector; the third is pretty bady buckled. That sort of thing is, I suppose, only to be expected, for some of the gusts have been terrificstrong enough to make it difficult to keep on your feet when you are out and about. What I'd like to have is a ferrite aerial mounted in the attic. One knows how effective built-in ferrite aerials can be; but the trouble with them, if you use your set in a ground-floor or first-floor room is that they pick up a good deal of interference which you don't want as well as the signals which you do want. So far as I know, no maker has produced a compact ferrite aerial working at high enough frequencies, though I don't see why in time it couldn't be done. I'm sure there would be a good demand for them in exposed, windy places if they were available.

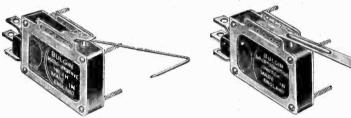
Well Equipped

WHEN the fitting out of the recently launched Orient Line's s.s. Oriana is completed a bit later in the year she will be the most fully equipped passenger ship in the world from the television point of view. Her Marconi installation will enable her to receive programmes on Band I and III from transmitters in almost any part of the world when she's within their range, for it can handle 405-, 525- and 625-line signals. I write "almost any part of the world" because it won't handle the French 819-line pictures-still, one can't expect everything! Dual standard receivers for 525 and 625 lines will be installed in public rooms and firstclass cabins. Incoming 405-line signals will be converted to 625 lines in the ship's central TV control room. Where alternative programmes are available passengers will be able to switch from one station to another. There'll be 60-odd sets to begin with, but the number can later be increased to nearly 400 without any And there will basic alteration. also be TV available for those who want it when the ship is beyond the range of transmitters ashore, for she will have a comprehensive closedcircuit installation. This will enable films chosen from the ship's extensive library of reels to be used and it will also make live TV broadcasts possible -simple studio sequences, interviews and so on. Dare I say that I hope there'll also be TV-free rooms, to which those who want a holiday from the daily ration of television at home can retire?



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UNBIASED

By FREE GRID

Pædiatric Peepboxes

IN the Building Exhibition which was held at Olympia last December, one of the exhibiting firms provided what was called "a glimpse into the future"—in which various domestic applications of television and telearchics were shown. I must confess, however, that I did not see a great deal beyond an elaboration of things that have already been done in simpler form.

Thus the famous and now ubiquitous Baby Alarm, which first appeared in the Readers' Problems section of this journal well over thirty years ago, has now been elaborated into a closed-circuit television outfit; but I award full marks for it.

TV in the bathroom is, of course, something I installed long ago and described in these columns. I seldom use it as I always feel uncomfortable when a female television announcer gazes at me from the screen as I lie full length in the bath. Perhaps a psychiatrist could explain this.

The photograph with which I illustrate this note is of the actual bathroom television installation shown at Olympia. To forestall enquiries I would hasten to add that the YL demonstrating the apparatus is unknown to me. The photograph came from the exhibitors, Laconite, Ltd., manufacturers of the wall surfaces, with whom you should communicate if you want further information about her or the unit.

This electronic baby-watching craze seems to have spread to the world of bowling alleys, for in the one opened recently in North London a closed-circuit pædiatric peepbox has been installed by Pye, Ltd., to enable parents to keep an eye on their children parked in an adjoining room. The 625-line camera is remotely controlled so that general views of the nursery or close-ups of individual children can be seen in

the bowling alley where there are two 21-in monitors.

Not Dead Yet

A FEW weeks ago, a B.B.C. speaker warned listeners of the danger of going to sleep with an electric blanket switched on, especially after taking sleeping tablets. I have done both these things without coming to any harm for over thirty years, but, be that as it may, my reason for men-tioning the matter is because of the interference caused by the older type of blanket.

In publicizing this warning the B.B.C. really fouls its own nest since, by inference, it is obviously intended that the blankets should be switched on early in the evening; in other words during the peak listening hours when the clicking on and off of the unsuppressed thermostats of the older type of blanket will provide an irritating background to the programmes.

Of course, this clicking interference is not nearly as bad as it was 24 years ago when I made my first protest about it in these columns. It was so bad during the war years that I always thought that, despite the blackout, Hitler's bombers would know when they were over a large city because of the ambient cloud of clicks which "blanketed" each large centre of population. Doubtless the bombers' radio operators were briefed to keep a special look out for these betraying signals. As so many of our pre-war blankets were of German origin, it is probable that the far-seeing Führer gave instructions for special interference-producing thermostats to be fitted to all those intended for export.

Be that as it may, it is clear that the B.B.C. speaker regards the heat-regulatory thermostats of modern electric blankets as so unreliable that they will not cut out at the pre-set temperature, with the result that the

sleeping user will roast in his bed. During 30 years I have tried out many types and will admit that on one occasion I did find that a thermostat let me down. But despite my sleeping tablet, the undue warmth soon roused me.

Nowadays, of course, blankets are far more reliable. The one I use at present is not fitted with a thermostat, it being of a type supplied for every bed in a well-known hospital for the very purpose of being left

switched on continuously.

The B.B.C. speaker may, of course, have envisaged a lethal shock being received by a sleepy wife upsetting the early-morning cup of tea brought up to her by a dutiful husband, which could effectively reduce the contact resistance between her torso and any faulty insulation in the blanket. What a suggestion for a henpecked husband toying with the idea of lawproof uxoricide! However, the heater element of a modern blanket is waterproof as well as fireproof; one firm even invites you to test their product by dropping it in the bath. But nervous people do exist, and I think the B.B.C. might have pointed out that it is possible to buy a 24-volt blanket for use with a transformer on a couple of car batteries.

Babelissimo

JUST forty years ago I recollect standing on the summit of what was then America's tallest structure. namely the 792-foot-high Woolworth building. One of the officials of the building pointed out to me with pride how puny the 612-foot Singer building and other neighbouring sky-scrapers looked beside the structure which its owner had fancifully called "The Cathedral of Commerce."

I tried to deflate his ego by point-ing out that America was still nearly 200 feet behind Europe with its 984foot Eiffel Tower which, apart from its other functions, was then the tallest radio mast in the world. I was, however, reckoning without the resiliency of Americans as he was in no way abashed, but promptly prophecied that the U.S.A. would soon have a radio mast far higher than any in Europe.

It was a wild boast that came true not so many years afterwards when a radio mast was erected on the Empire State Building making the overall height 1,250-feet above the

sidewalk.

When the height of the mast at the recently opened I.T.A. station at Mendlesham, Suffolk, was announced (1,000ft) it was thought to be the tallest in Europe. However, Sweden justifiably lays claim to this "honour," for the mast carrying the television and v.h.f. sound aerials for the new Hörby station is 320 metres (c.1,050 ft.).

Closed-circuit TV from the front and back doors (with intercom. facilities) and a B.B.C./I.T.A. receiver are incorporated in this bathroom outfit.





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Type No.	Frequency Range (Ge/1)	Pewer Bulget ever Frequency Range (mW)	Structure Voltage Range (V)	Sensitivity at Mid-frequency (Me.s per Y)	Cathoda Current max (mA)
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BA9-20	7.0 to 11.5	20 to 180	250 to 1400	5.0	28
BA16-10	11 to 18	10 to 70	500 to 2500	3.5	13

Mullard



GOVERNMENT AND
INDUSTRIAL VALVE DIVISION

Where modulation of the valve output is required this can be readily achieved by modulating the appropriate electrode. The output connection is isolated from the delay structure so that the valve may be operated with its cathode earthed and consequently high modulation frequencies may be used. These specialised microwave valves make possible the design of wide frequency range microwave instruments, microwave search receivers and f.m. carrier systems. Write to the address below for full details of these and other Muliard microwave valves.



MULLARD LIMITED

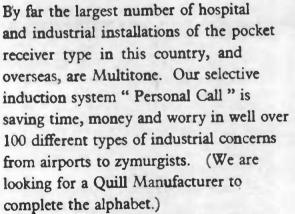
Mullard House, Torrington Place, London, W.C.1

Telephone: LANgham 6633

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Multitone leads in pocket staff location



The New MULTI-CHANNEL equipment provides over 400 individual channels using the new flat receiver (as illustrated).

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Additional Pacilities

ELECTRONIC TRUNCHEON

The Electronic Truncheon is no bigger than standard equipment carried by guards and serves the same purpose, but inside there is a transmitter which, when the button is pressed, sends out a signal. This is picked up by the loop of wire around the area to be protected. The pulse is used to operate a small receiver, which automatically switches on any form of electrical alarm. It can be operated from any point in the area.

INTERNAL TRANSPORT COMMUNICATION

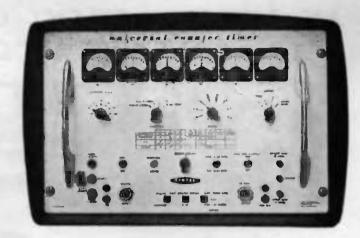
The Multitone "Personal Call" loudspeakerreceiver has been designed to solve the problem of conveying verbal instructions to transport vehicles used for handling loads inside a given area. Messages can be conveyed to all or selected vehicles from the central transmitter.

MULTITONE INDUCTION SYSTEMS CAN SOLVE YOUR STAFF LOCATION PROBLEMS:

- ★ Equally suitable for large and small areas or concerns
- ★ Low rental terms
- ★ Virtually no internal wiring

(the 'peep-peep' in the pocket), the only staff location system worth installing Write or 'phone for further particulars. We can be found in 10 seconds.

Transistorized UNIVERSAL COUNTER TIMER



Frequency Measurement

Random Counting

Frequency Division

Time Measurement

Frequency Standard

This fully transistorized portable equipment provides for a wide range of time and frequency measurement as well as facilities for counting, frequency division and the provision of standard frequencies. The facilities available are briefly listed below:

TIME/UNIT EVENT (1 LINE): For the measurement of the time interval between two occurrences in a continuously varying electrical function in the range 3µsec to 1 sec. The time for 1, 10 or 100 such events can be measured.

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AUDIOTAPE, manufactured in the U.S.A. by Audio Devices, Inc., is known the world over for its flawless perfection of sound reproduction throughout the entire audio range, and its consistent, uniform quality from reel to reel and from one type to another. Available in eight different types, with a base material and reel footage to meet every recording requirement, AUDIOTAPE has only one standard of quality-the finest obtainable. This sterling performance reflects Audio Devices' more than 10 years of experience in magnetic tape manufacture, and more than two decades of practical experience in the sound recording art.

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So easy to build



7 New Kits-Now available

STEREO CONTROL AMPLIFIER Model USC-I
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DECADE CAPACITOR UNIT , DC-I
12W. MONAURAL AMPLIFIER MA-12
MULTI-RANGE METER , MM-I

Also TAPE DECKS now available as "package deals" with other equipment.

Details and prices will be sent on application.



Sin. OSCILLOSCOPE KIT

Model O-I2U

Laboratory quality at utility oscilloscope price and ease of assembly make this kit o outstanding value. Vertical frequency response 3 c/s to 5 Mc/s., + 1.5 dB. — 5 dB., sensitivity 10 mV. per cm. st 1 kc. Horizontal frequency 1 c/s. to over 400 k/c. (±1 dB. up to 200 kc.). The Heath patented sweep circuit functions from 10 c/s. to 500 kc. In five ranges giving five times the usual sweep of other scopes. In addition it has exceedingly short re-trace and rise times and electronically stabilised power supply. Included is a 40-page.

[834.15.0]



ELECTRONIC SWITCH KIT (Oscilloscope Trace Doubler)

Model S-3U

This extremely useful, low priced device will extend the use of your single-beam oscilloscope for duties otherwise only in the province of the double-beam tube.

In short, at a nominal cost, the Heathkit model S-3U will give you the advantages of a double (or other multiple) beam "scope, while retaining all the advantages of your present single-beam instrument.

Hitherto an electronic switch of this nature, permitting the simultaneous observation of two signals on the screen of a single-beam C.R.T. oscilloscope, has cost nearly as much as the scope Itsalf.



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Measures capacity 10pF, to 1,000 μF, resistance 100Ω to 5 megohms and power factor. 5-450 v. tast voltages. Salety switch provided. £7.19.6

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TRANSISTOR PORTABLE KIT



Prerented in elegant real hide case with tasteful gold relies. Can be assembled in 4 to 6 hours and you have a set in the top flight of the 2C-33 guines class. Prealigned I.F. transformers, printed circuit and a 7in, x 4in, high-flux speaker. £15.18.6

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Moc'el AW-IU Up to 25W £13.18.6

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Gives 16 w, output (8 per channel with 0,1 per cent, distortion at 6 w, per channel), it has ganged controls, STEREO MONAURAL gram, radio and tape recorder input and push-button refection as well as many other first cass leasures well above its price range, in two-tone grey metal exhibite with a golden surround and fittings. Also ultra-linear push-pull output. \$25.5.6 Basic sensitivity 10mV, (2mV, svailable, 20/extrs).

VARIABLE FREQUENCY OSCILLATOR KIT Model VF-IU



For all Amateur Bands, 160-10 metres, Ideal for Heathkit DX-40U and similar transmitters.

Price less valvas 68/19/6 £10.12.0

DUAL-WAVE TRANSISTOR RADIO KIT Model UJR-I

This sensitive headphone set is a fine introduction to electronics for any youngster. (Not illustrated) £2.16.6

 Deferred Torms available on all orders above £10.

DAYSTROM LTD.

DEPT. W.W.J. GLOUCESTER, ENGLAND

A member of the Daystrom Group. MANUFACTURERS OF THE WORLD'S LARGEST-SELLING ELECTRONIC KIT-SETS

Technically



excellent



6 WATTS STEREO AMPLIFIER KIT Model S-33

A versatile high-quality self-contained STEREO/MONAURAL Amplifier with adequate output for a living room—or with which to convert a favourite (monaural) radiogram Into a stereo-radiogram. 3 watts per channel; 0.3% distortion at 2.5 w/chnl.; 20 dB N.F.B., inputs for Radio (or Tape) and Gram., Stereo—or Monaural; Ganged controls.

Sensitivity 100 mV.



VALVE VOLTMETER KIT Model V-7A

The world's most popular valve woltmeter with printed circuit and I per cene, precision resistors to ensure consistent laboratory performance, it has 7 voltage ranges measuring respectively d.c. volts to 1,500 and a.c. to 1,500 r.m.,s. and 4,000 peak to peak. Resistance measurements from 0.1 ohm to 1,000 M ohms with internal battery. D.C. input impedance is II Megohms and dismeasurement has a centre-zero scale. Complete with test prods, leads and standardising battery... 213.0.0

R.F. PROBE KIT Model 309-CU

This complete probe kit will extend the frequency range of the V-7A Valve Voltmeter to 100 Mc/s, and will enable useful voltage indication to be obtained up to \$1.5.6



AMATEUR TRANSMITTER KIT Model DX-100U

The world's most popular "Ham"

TX Kit

- Completely self-contained, compact "Ham" Transmitter,
- · Built-in, highly stable VFO and all Power Supplies.
- TVI: Careful design has reduced TVI to a minimum by use of effectively screened frequency-generating stages and pi tuned circuits at the input and output of the PA stage, and by III chokes and pi network filters to all outlets from the cabinet. No fewer than 35 disc ceramic- by-pass capacitors help to achieve the exceptional stability and high-performance for which this Transmitter is noted.
- eThe KT88 high-level anode and screen modulator stage gives over 100 watts of audio from less than 1,5 mV, linput.
- Adjustable drive and clamp control ensure that valves are only driven sufficiently to maintain the required output.
- Keying on CW is via the VFO and buffer amplifier cathodes; the other RF valves are biased beyond cutoff. When zero-beating the TX with incoming signals, the exciter stages only may be run without the final amplifier being switched on.
- · Provision has been made for remote control operation.
- VFO slow-motion drive is very smooth and back-lash free.
- Covers all Amateur bands up to 30 Mc/s, phone or CW,
- VFO or Crystal control. £78.10.0

MATCHED HI-FI STEREO KIT

4-speed Transcription Record Player

FIGGEL KY-IV	EIZ	10	v
6 w, Hi-Fi Amplifier, Model S-33	411	8	0
Twin Stereo Speaker Systems Model SSU-I	(20	11	0
Total cost if purchased separately	644	9	0
YOURS for £42/10 - if all ordered together deposit and 9 monthly payments of £4/3/-, speaker legs £2/14/- optional extra.			

GLOUCESTER

a superbly designed stereo cabinet kit





TRANSCRIPTION RECORD

With 4-speed A.C. motor unit and Stereophonic Pick-up completely assembled on plinth,

High performance at low cost.

This attractive Transcription Record Player incorporates many new features which make it suitable for all types of recordings on discs. It has the new Collaro RP.594 unit with the Ronette Stereo Pickieup and gives excellent results on stereo or mono (33, 45 L.P. or 78 r.p.m.) gramophone records.



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Covers all smateur bands from 80 to 10 metres. Power input 75 watts C,W, 60 watts peak controlled carrier phone. Output 40 watts to aerial. Provision for V.F.O. Filters minimise T.V. Interference. \$29.10.0

AUDIO VALVE MILLIVOLTMETER KIT

Model AV-JU, ImV to 300V £13.18.6

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dependable



COTSWOLD SPEAKER SYSTEM KIT

This acoustically designed enclosure measures 26 x 23 x 154in, and houses a special 12in, bass speaker with 21in, speech coil, elliptical middle speaker together with a pressure unit to cover the full frequency range of 30-20,000 c/s. Its polar distribution makes it Ideal for really Hi-Fi Stereo. Delivered complete with speakers, cross-over unit, level control. Tygan grille cloth, etc., Left "in the white "for finish to personal taste, all parts are precut and drilled for ease of assembly.



HI-FI F.M. TUNER

Tuning range 88-108 Mc/s. Flywheel tuning. Attractive Plastic Front Panel in two-tone gray with golden trim, surround and motif. Thermometer type visual tuning indicator. Pre-aligned I.F. transformers (eliminate adjustment). Three I.F. Stages, Wide-band low distortion. Ratio Detector. Complete R.F. Unit, wired, tested and pre-aligned (ready for mounting to chassis). Printed Circuit for I.F. Ampfifiers and Ratio Detector, for ease of assembly. No alignment necessary after assembling, Built-in power supply, Output tockets for stereophonic adaptor (for stereo transmission when available). TUNET UNIT Model FMT-4U (incl. 16/11 P.T.) 43 2 0 with 10.7 Mc/s. I.F. output.

I.F. AMPLIFIER with case and			610	10	6
Sold separately	 	Total	(13	12	6

-* FREE ON REQUEST *.

A copy of our (British) Heathkit Catalogue, Prices include free delivery in U.K.



CAPACITANCE METER KIT Model CM-IU

This Diroct-Reading Capacitance Meter is a very low priced, time-saving instrument which is so useful that it should be part of the general equipment of every electronic laboratory and production line. Easily built in a few hours. 0-100 µF, 0-1,000 µF, 0-0,01 µF, 0-0,1 µF. The meter has 4fin, scale and can be used by an unskilled operator after a few minutes' instruct. £14.10.0 tion.



URAGUUM

STEREO-HEAD BOOSTER KIT

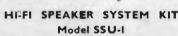
Hi-Fi Stereo Pre-Amplifier for low-output Hi-Fi P.U.s. Input 2 mV, to 20 mV, Output adjustable from 20 mV, to 2 V, 40-20,000 c/s, Also suitable as low-noise R.C.-Coupled high-gain monaural ampli-



AUDIO SIGNAL GENERATOR KIT Model AG-9U

10 c/s. to 100 Kc/s., switch selected. Distortion less than 0.1%. 10 v, sine wave output metered in volts and d8's.

• Deferred Terms available on orders over £10.



Please tick the items in which you are interested and we will send you full details.

\$88	Hi-Fi Stereo Amplifier Kit	UJR-1	Dual-Wave Transistor Radio Kit
5-33	6-Watts Stereo Amplifier Kit	S-3U	Electronic Switch Kit
UXR-I	Transistor Portable Kit	309-CU	R.F. Probe Kit
DX-40U	"Ham "Transmitter Kit	DX-100U	Amateur Transmitter Kit
O-12U	Sin, Oscilloscope Kit	-	Matched Hi-Fi Stereo Kit
V-7A	Valve Voltmeter Kit	R/NUI	Transcription Record Player
SSU-I	Hi-Fi Speaker System Kit	-	Gloucester Stareo Cabinet Kit
AG-9U	Audio Signal Generator Kit	_	Cotswold Speaker System Kit
C-3U	Resistance-Capacitance Bridge Kit	-	HI-Fi F.M. Tuner
VF-IU	Variable Frequency Oscillator Kit	CM-IU	Capacitance Meter Kit
USP-I	Stereo-Head Booster Kit	MA-12	12 w. Monaural Amplifier Kit
USC-I	Stereo Control Amplifier Kit	MM-I	Multi-range Meter Kit
OS-1	3in, Service Oscilloscope Kit	-	Tape Deck " Package deals "
B-IU	Balun Coil Unit Kit	AW-IU	Audio Wattmeter Kit
DC-I	Decade Capacitor Unit Kit	AV-3U	Audio Valve Hillivoltmeter Kit

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Ohmmeter 100 - 10,000M2 £52.10.0



Flash Tester 500 - 2000 V A.C. £26.0.0



R.C.C. Bridge 5 pf - Vf 50 - 500 MQ £30.0.0



Sensitive Value Voltmeter 3mV-300V. 40 cls-500 kcls £49.50.0



D.C. Value Voltmeter 0.1 - 1000 volts 0.01 - 1004A £88.10.0

NASHTON

precision measuring equipment means accurate economic buying

These six models are a selection from the wide range of NASHTON miniaturized electronic test instruments available for immediate delivery. Designed by Nash & Thompson to meet your everyday needs accurately, efficiently and economically, NASHTON units are compact, modern, streamlined, easy to read and operate. NASHTON give you what you need most — functional efficiency, not frills.

Others in the range include the A.C./D.C. Valve Voltmeter, Universal D.C. Meter, L.F. Quadrature Oscillator, 0.5 Amp. Stabilised D.C. Power Supply, Transistor Tester, Shorted Turn Detector, Digital Read-Out Meter, Preferred Value Resistance and Capacitance Boxes.

Nash and Thompson ...

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Four first-class performers

These four Cossor Oscillographs, each designed for an important range of applications, offer first-class performance backed by rigid adherence to published specifications.





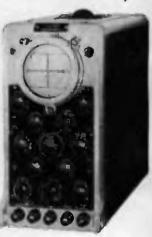
MODEL 1065 PULSE OSCILLOGRAPH

Tube. single-beam, P.D.A. Bandwidth: d.c. to 15 Me/s (—50%).
Sensitivity: 250 mV/cm.
Overshoot: less than 3%.
Time-base: triggered or repetitive over range 40 cm/sec to 5 cm/isec.
X Amplifier: gain 5, continuously variable.
Time-base delay:
2 ranges, continuously variable.
Calibration: voltage and

time, by calibrated shifts Probe: 1.5 MQ, 12 pF

MODEL 1058 FOR THE TV & RADIO ENGINEER

Tube: single-beam
Bandwidth: d.c. to
6 Mc/s (—50%).
Sensitivity: 250 mV/cm.
Time-Base: triggered or
repetitive, over range
30 cm/sec to 1.5 cm/usec.
Special facilities for
triggering from TV line or
Frame pulses on IV.D.A.P.
composite video waveform.
X Amplifier: gain 5,
continuously variable.
Calibration: time and voltage
calibration facilities.





MODEL 1049 INDUSTRIAL DOUBLE-BEAM OSCILLOGRAPH

Y Amplifier: Al: d.c. to 200 ke/s (—30%) at gain 900: A2: d.c. to 400 ke/s (—30%) at gain 30.

Time-Base: repetitive or triggered in 18 ranges, down to 7.5 sec/sweep. Intensity modulation: three modes including beam bright-up.

Calibration: time and voltage, by calibrated shift (X and Y1) and multiplier (Y2).

MODEL 1035 MK III GENERAL PURPOSE DOUBLE-BEAM OSCILLOGRAPH

Y Amplifiers: A1: 5 e/s to 5 Mc/s (—30%), Maximum gain 3,000. A2: 5 c/s to 250 ke/s (—30%) at gain 30, with trace Inversion facility. Time-base: repetitive or triggered in 9 sweep ranges from 100 msec to 10 usec. Time-base delay and pulse bright-up facilities. X Amplifier: gain 5, continuously variable. Calibration: voltage and time, by calibrated shifts.

Let us send full details of Cossor Instruments or arrange for a representative to discuss your special needs.

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Transistorized Power Units



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MODEL 1328 LABORATORY POWER UNIT

Model 1328 provides a continuously variable output monitored by two front-panel, mirror-scale, voltage and current meters. It is eminently suitable for use in the design stage of transistor circuits where a power supply of high purity is essential.

Output: 0-30V at 1A.





MODELS 1326 & 1329 LOW VOLTAGE POWER UNITS

The 6V Transistorized Power Unit Model 1326 is an ideal supply for transistor d.c. amplifiers, transistor pulse-technique circuits and filaments of thermionic valve amplifiers, particularly in low-level microphone stages.

MODEL 1326 Output: 6V at 0-2A. MODEL 1329
Output Voltage: Continuously variable 5V-10V.
Output Current: 0-1A in range 5V-9V
0-0.5A at 10V.

(chassis mounting)

MODEL 1327 BATTERY ELIMINATOR

The Battery Eliminator Model 1327 has been designed primarily to power the Cossor Pre-amplifiers, Models 1430, 1434 and 1440, but it can be used in many other applications requiring a power unit to provide high and low tension supplies.

L.T. Supply: Output 6V-6.5V at 1.7A. H.T. Supply: Output 120V at 15mA.



Please send for the latest Cossor Catalogue or ask for a representative to call and discuss your special requirements.

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Mullard Alloy

brings you high frequency

The need for the new technique

Demands for devices which will operate at higher and higher frequencies have been made ever since the transistor was first introduced.

So far as the normal p-n-p germanium alloy transistor is concerned, the OC44 with its average frequency cut-off of 15 Mc/s probably represents the upper limit of frequency attainable in quantity production. A major factor limiting the high frequency performance is the transit time of the minority carriers between the emitter and collector junctions. It is possible, by adding an "accelerating" or "drift" field to the OC44, to increase its high frequency performance by a factor of about 2 or 3, but the thickness of the base layer prevents a more fundamental reduction in transit time.

Using the alloy diffusion technique it is possible to make on a large scale, transistors with a base of only a few microns thickness and with a correspondingly high frequency performance. The diffusion produces a drift field which also contributes to the favourable high frequency characteristics.

The new technique in practice

The alloy diffused transistor is built up on a wafer of p-type germanium which forms the collector. Two metal pellets for the base (B) and emitter (E) are placed close together on one side of the wafer. Pellet B contains only n-type impurities, while pellet E contains both n and p types.

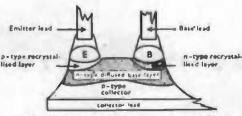
Pellet with notype impurities only to form ohmit contact with notype impurities to form ohmit contact with notype (base) which is made to extend from junction

E

B

Collector material (potype)

before heating (NOT TO SCALE)

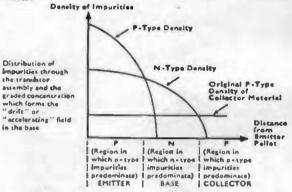


Simplified translator assembly after alloying and etching (NOT TO SCALE)

When this assembly is heated to an appropriate temperature in a gaseous atmosphere, germanium dissolves into the metal pellets until saturation is reached.

If the temperature is maintained, the impurities in the pellets B and E diffuse into the germanium wafer.

However, the p-type impurities in pellet E diffuse very slowly and they penetrate only a negligible distance into the wafer.



On the other hand, the n-type impurities in both pellets E and B have a fast speed of diffusion. These penetrate the wafer and form an n-type layer, and since diffusion also takes place via the gaseous atmosphere the layer extends to the exposed surface of the germanium between the two pellets.

The diffused n layer forms the base of the transistor, which by specialised manufacturing techniques can be controlled to a thickness of only 5 microns. The concentration of impurities in this layer is graduated between the emitter and collector junctions, and it is this gradient that produces the "accelerating" field.

When the assembly is cooled, a layer of germanium recrystallises from the pellets as with normal alloy techniques.

The recrystallised layer beneath pellet E is p-type as the p-type element chosen has a greater solubility in the recrystallised germanium than the n-type impurity which is also present in this pellet. The p-type material forms a p-n-p junction with the diffused n-type layer and the original p-type collector wafer.

The recrystallised layer under pellet B is n-type, and forms an n-n (non-rectifying) junction with the n-type base layer which extends from the p-n-p junction between emitter and collector. Pellet B is thus used for making the ohmic contact to the extremely thin base.

After etching away the base layer where it is not part of the junction and where it does not form the ohmic contact with the base pellet, the assembly is ready for final processing and encapsulation.

Diffusion Technique

transistors at economic prices

One of the most important recent advances in transistor technology is the alloy diffusion technique used by Mullard. This technique not only provides transistors with uniform high frequency characteristics—it provides a method of producing transistors at prices economic to the user.

First in the

Abridged Data

P-N-P Germanium alloy diffused transistor suitable for use as mixer-oscillator in short wave receivers and as i.f. amplifier in f.m. and a.m./f.m. receivers.

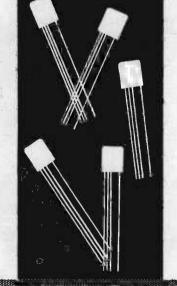
f. min	40 Mc/s
f ₁ average	70 Mc/s
$V_c \; max. , , , , , , , , , , , , , , , , , , $	-20 V
Is max:	10mA

Typical Operating Conditions

Acertanoranga to graterination	-6.0 V
I _e	LQ-mA
Ideal unilateralised power gal	in at
f= 10.7 Mc/s	32 db



Mullard
high frequency
range



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industrial semiconductors

Complete Coverage from 7.5 Mc/s to 230 Mc/s

fm mmmmm



am www.www.mm

TYPE S.G.63A FM/AM SIGNAL GENERATOR

Here, within this one modestly priced instrument, are the essential facilities and the essential accuracy demanded for servicing radio and television receivers on bands I, II and III. The S.G.63A provides:

- Frequency range 7.5 Mc/s to 230 Mc/s.
- Crystal Calibrator checks at 5 Mcis points (accuracy ± 0.03%).
- F.M. at 1,000 c/s fixed ± 22.5 kc/s and ± 75 kc/s.
- Wide deviation at line frequency variable 0 to ± 200 kc s.

• A.M. at 1,000 c/s fixed 30%

Full sechnical details in Leastet W61 available on request.

NETT PRICE IN U.K. £90

Advance

- to be sure!

THIS INSTRUMENT IS ALSO AVAILABLE WITH OUTPUT IMPEDANCE OF 50 OHMS

Advance

COMPONENTS LIMITED

INSTRUMENTS DIVISION

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GD 87

CELESTION MODEL G44/1300 FULI WIDTH STEREOPHONIC

LOUDSPEAKER SYSTEM

The complete answer to the demand for a reasonably priced high quality speaker system for the reproduction of stereophonic recordings in the home.

Utilising two very small pressure type direct radiator units for the higher frequencies and two 12in. reproducers for the lower, the equipment covers a response which is substantially level from 35-15,000 c/s. with true stereophonic effect.



The price for two HF,1300 and two G.44 units, complete with enclosure details is:-

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The two HF.1300 units should then be placed near the corners of the room, one on each side of the enclosure and approximately 4 feet above the floor. The width of sound will very nearly correspond to the distance between these two units. No elaborate cross-over networks are required and the system is completed by a 12 Mfd. capacitor in series with each high note unit.

Designed and developed by CELFSION

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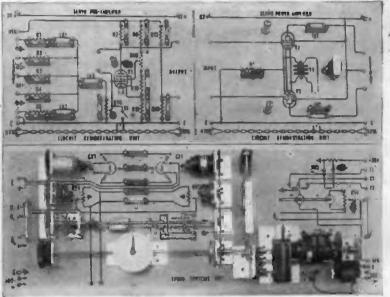
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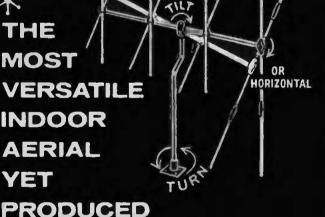
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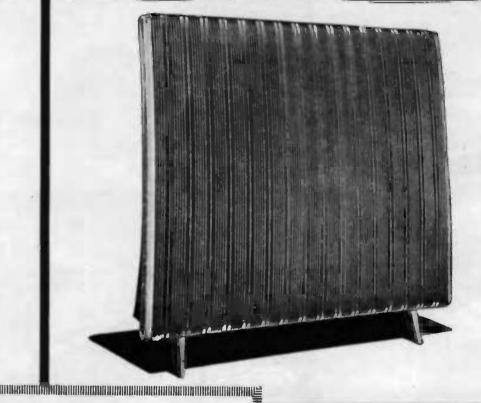
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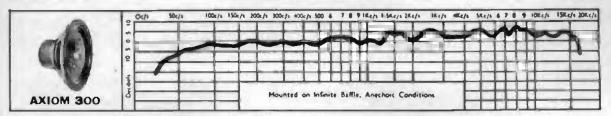
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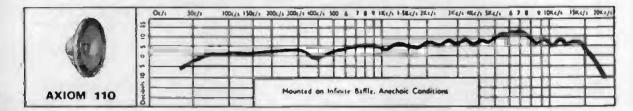


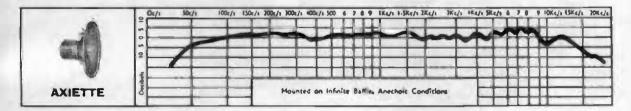
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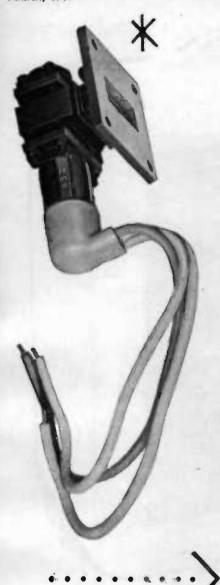


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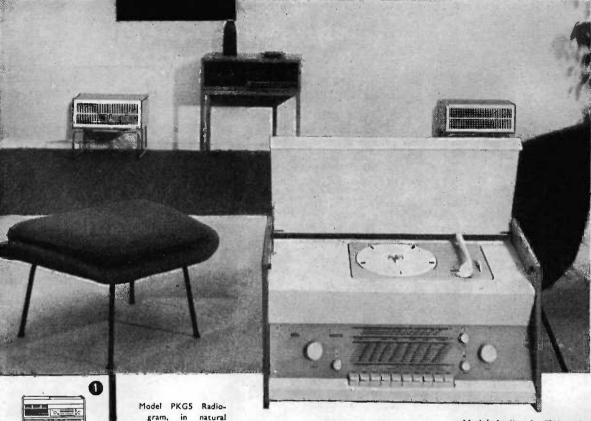
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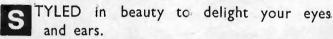
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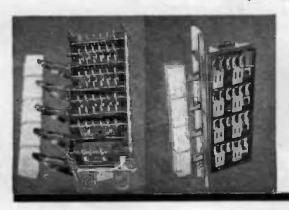
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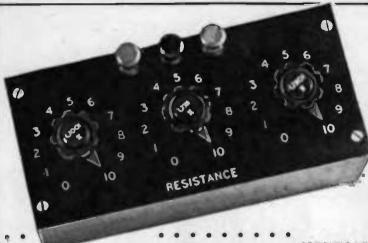
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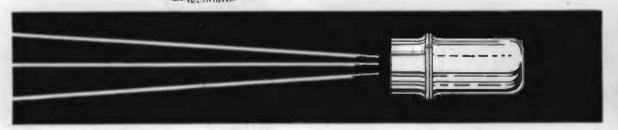
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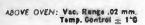
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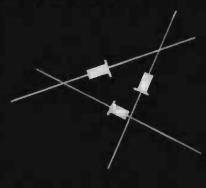


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R.F. Output:

5 watts

Channel Spacing:

50 or 60 Kc/s.

Modulation:

F.M.

Service:

F 3 telephony

Operation:

Single or double

frequency simplex

or duplex.

Power Supply:

100-150 and

200-250 volts

A.C. 40-60 c/s.

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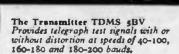
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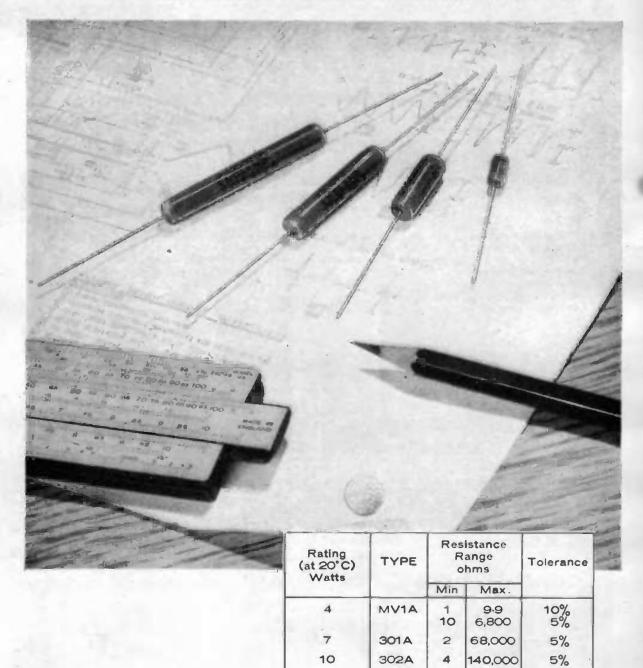
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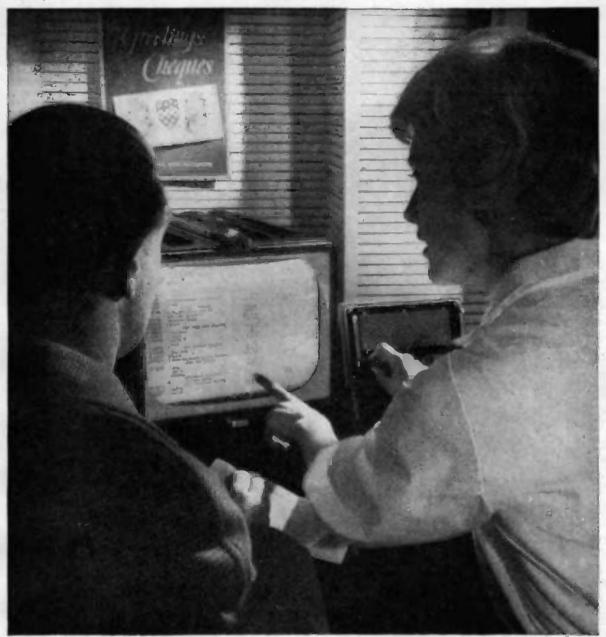
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					Last	Lead	Load	Lase
1	628	Epif-Wave	125	61	135	115	135	120
1	C311	н н	125	120	128	85	130	120
1	C28	D 46	250	- 11	275	245	200	255
1	638	ps 10	250	120	275	245	29.0	275
1	C28	Voll-Benbler	125	60	275	245	280	255
1	630		125	120	260	295	296	765

hty. Type		Cleçuil	Marimem Input Volta Ir.m.s.)	Max. Output Corrent mA (meas)	Typical Dig. 0 16 wfd. Hesvr. Cap. Balf Full Lord Lead		otaut Vollag 37 mfd. Besor, Cap. Bail Full Lean Lea	
1	CZV	Pash-Pull	125-0-125	128	148	128	148	330
2	C79	D4 10	258-9-250	120	275	250	200	255
1	C37	50 20	125-0-125	240	130	115	140	130
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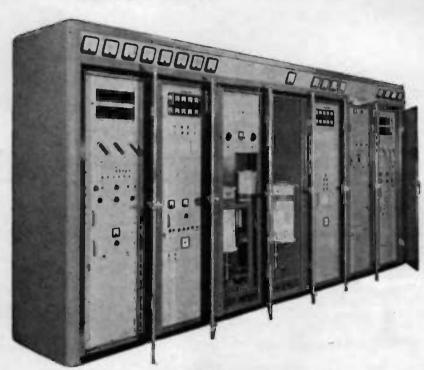
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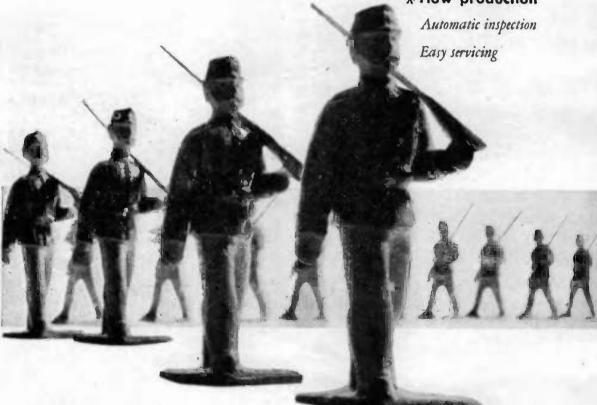
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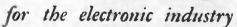


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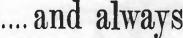
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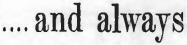
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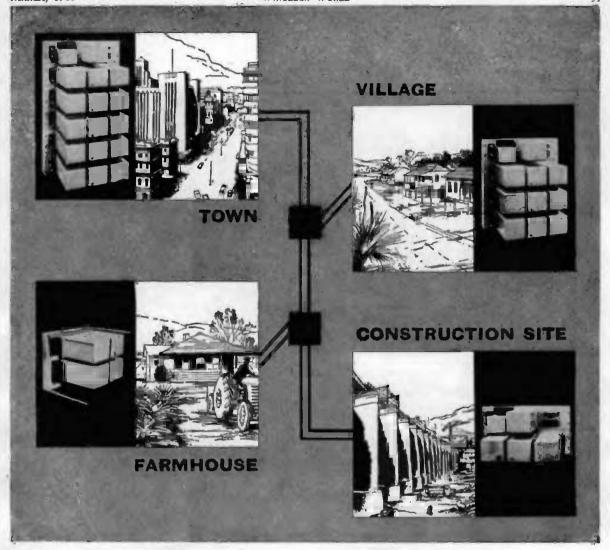
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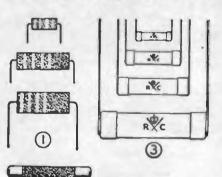
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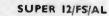
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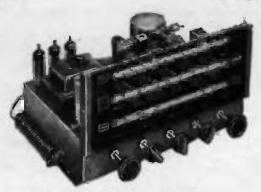
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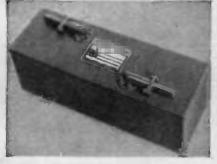


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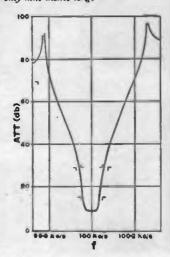
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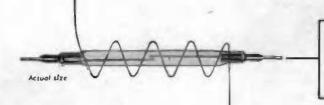
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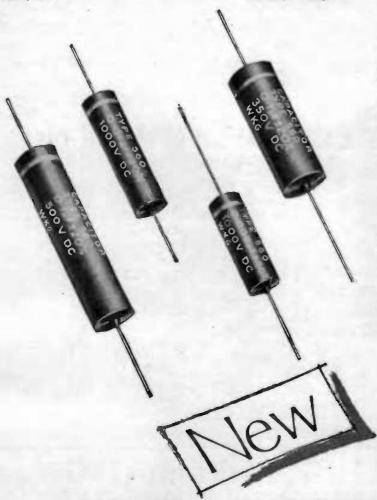
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0.002	1,000	2,500	250	1	1
0.005	1,000	2,500	250	1	1
10.0	1,000	2,500	250	1	18
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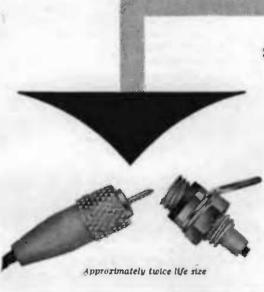
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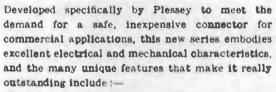


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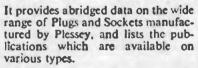
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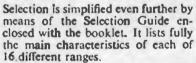


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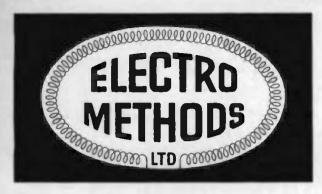
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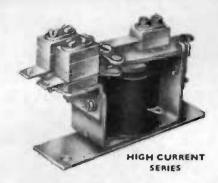
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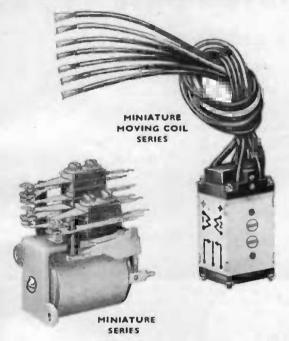
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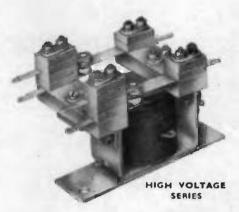
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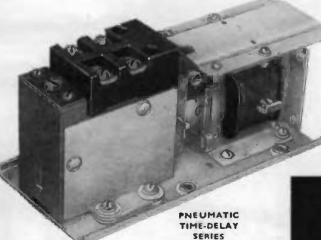
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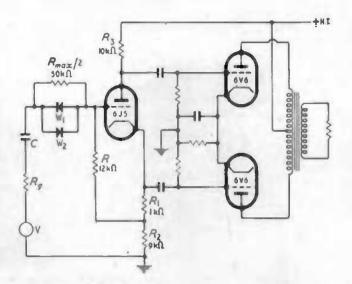
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ARTICLES IN THE FEBRUARY ISSUE INCLUDE:

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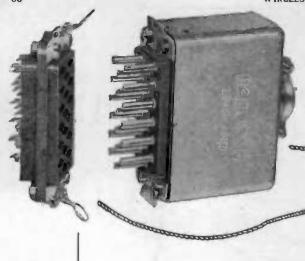
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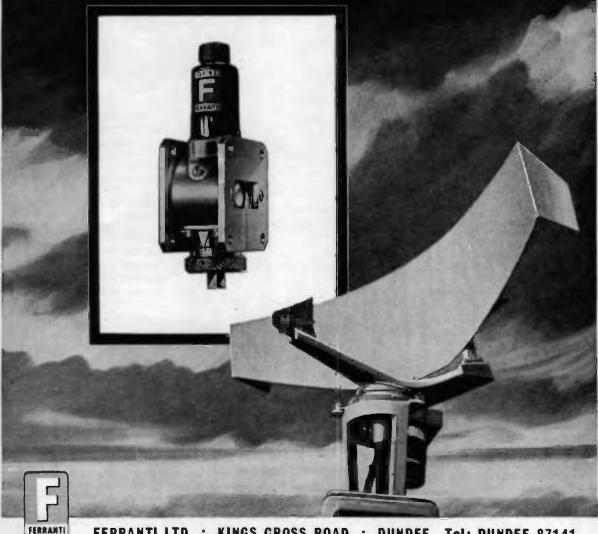
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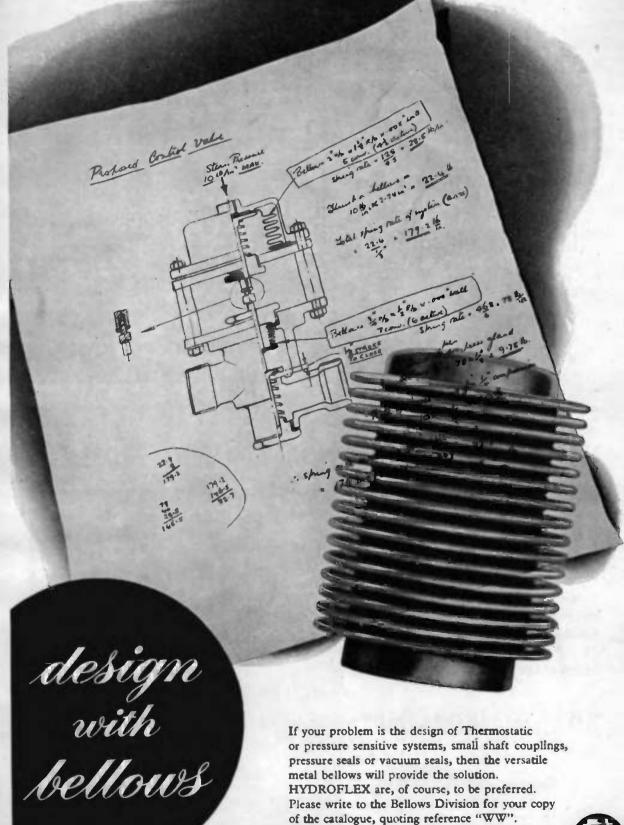
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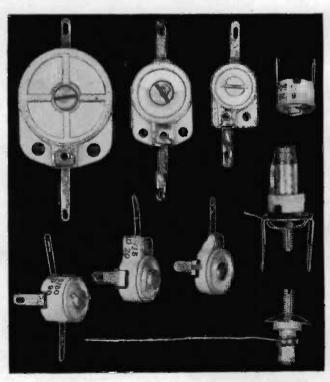
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Voltage Range: 3µv to 300v.

Ranges provided by an input attenuator switch and a meter range switch in steps of 1:3 or 10 db. Meter range is indicated by a dial mechanically linked to input attenuator. An absolute-relative switch, in conjunction with a variable 10 db control is provided for adjustment of intermediate values.

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Voltage Accuracy: ±5% of full scale value.

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Intermediate Frequency Rejections Intermediate frequency present in input signal rejected by at least 75 db.

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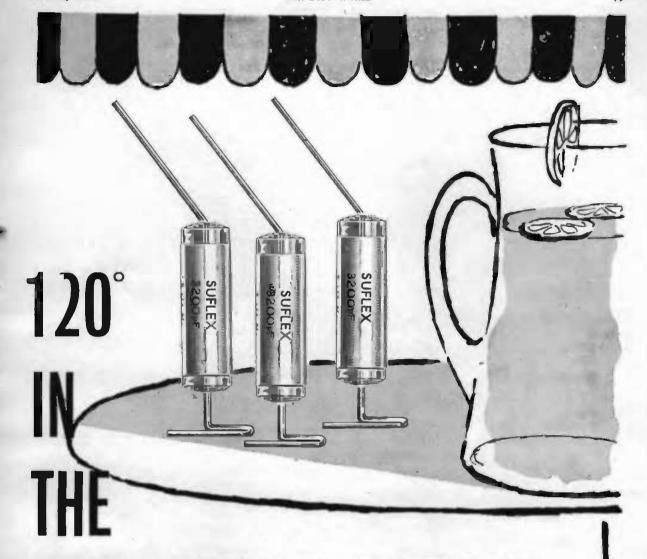
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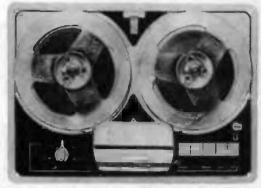
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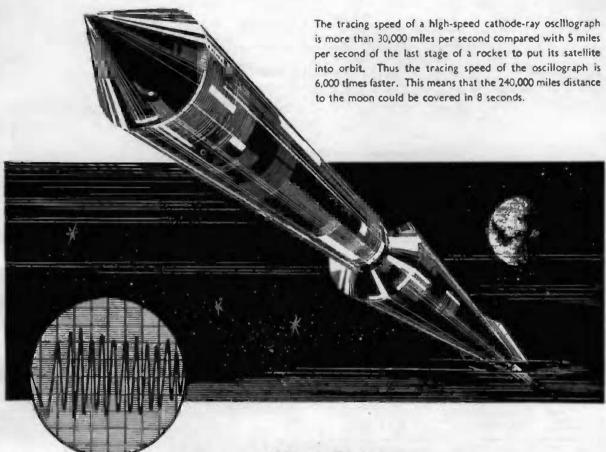
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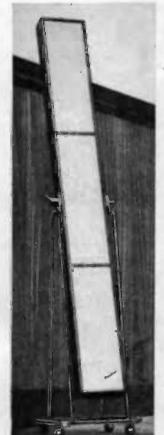
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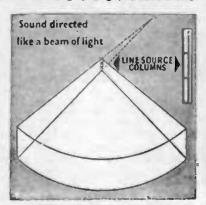
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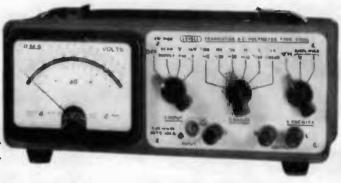
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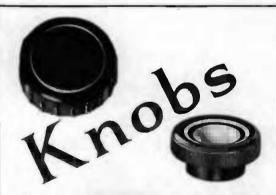
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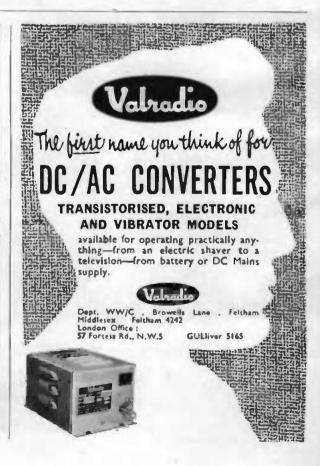
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V ₈ 44.(kV)	0.5	1.0	1,5	2.0	1.5	1.4	2.0	4.0	1:5	1.5
V _{a3} (kV)	0.5	1.0	1.5	2.0	1.5	1.8	2.0	4.0	4.0	1.5
V ₂₄ (kV)	-	_	_	4.0	3.0	4.0	4.0	8.0	8.0	-15
V _a 5 (kV)	-	_	_	-	-	10	_	_	_	15
Y scan (mm)	28	55	70	80	75	60	95	95	95	60
Y sensitivity (V/cm)	45	11,5	16	23	27	12.5	17.5	36	36	2.7
X scan (mm)	28	55	90	90	90	95	115	115	115	100
X sensitivity (V/cm)	53	20	23	36	27	26.5	29	- 60	60	11.2
Screen diameter (mm)	30	71	94	108	108	137	137	137	137	137
SCREEN TYPES:						- 1				7/4
Medium persistence	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes	Yes
Long afterglow	No	Yes	Yes	Yes	Yes	Yes	To order	To order	To order	To order
Blue photographic	To order	To order	Yes	Yes	Yes	To order	To order	To order	To order	To order
Short persistence	To order	No	No	To order	No	No	No	To order	To order	No

Data is given for each gun.

types ticked
ICPI SBKPI
3AFPI SBUPI
3AZPI SBYPI
4EPI SBYPIA
4LPI SCLPI



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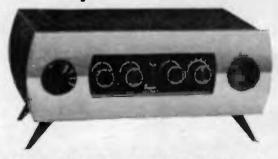
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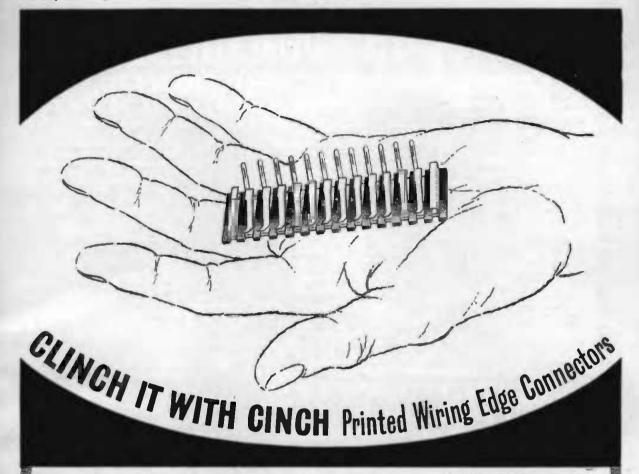
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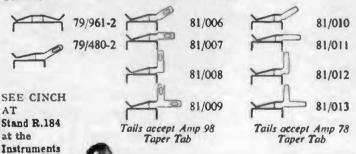
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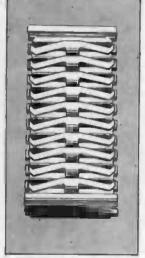
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Quad Amplifier C22 10 0 \$44			
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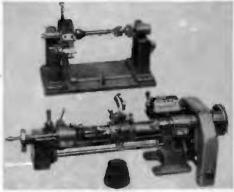
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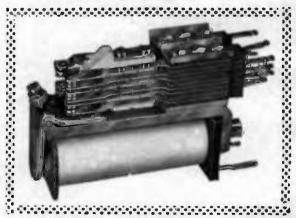
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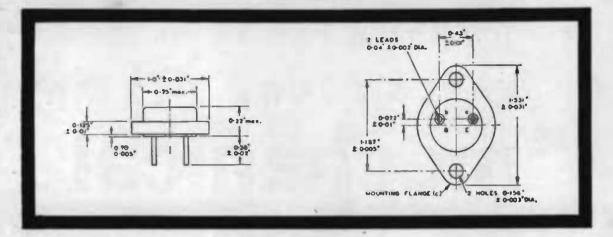
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Peak collector to emitter voltage, base open circuit (volts)	-50	- 65
Peak collector to emitter voltage, base and emitter joined or with		
an external base/emitter circuit resistance less than 40 ohms (volts)	-65	- 80
Peak emitter to base voltage (volts)	- 60	- 60
Peak collector current (amps)	-10	- 10
D.C. Collector current (amps)	- 5	- 5
Collector dissipation (mounting flange temperature 85°C) (watts)	10	10
Switching Characteristics (Common Emitter) (Typical production spread	's)	
D.C. Current gain (V _{ce} = -1.5v, I _c = -4A) minimum	20	20
average	26	26
maximum	50	50
D.C. Collector to emitter saturation		120
voltage ($I_c = -4A$, d.c. $I_b = -400$ mA) (volts) average	-0.4	-0.4
maximum	-0.8	-0.8

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ELECTRONICS, RADFO, TELEVISION

MARCH 1960

Managing Editor:	103	03 Editorial Comment							
HUGH S. POCOCK, M.J.E.E.	104	4 Improving the Dynamic Range of Tape Recording By L. H. Bed							
Editor:	107	Geneva Conference	, ,						
F. L. DEVEREUX, B.Sc.	108	How Long Will a Transistor Live?	By R. Brewer						
•	110	Transformerless Circuits for Broadcast Receivers By R. C. V. Macario and N. E. Broadberry							
Assistant Editors:	113	Manufacturers' Products							
H. W. BARNARD	114	World of Wireless							
T. E. IVALL	116	Personalities .	**						
	118	Physical Society's Exhibition							
·	1 2 9	Signal-Flow Diagrams—2	By Thomas Roddam						
	132	Short-Wave Conditions							
	133	Letters to the Editor							
·	135	Technical Notebook							
	136	Elements of Electronic Circuits—11	By J. M. Peters						
	137	The Reflectometer	By " Cathode Ray"						
VOLUME 66 NO. 3	141	The Smith Chart—3	By R. A. Hickson						
PRICE: TWO SHILLINGS	147	News from the Industry							
TRICE, TWO SHIELINGS	148	March Meetings							
FORTY-NINTH YEAR	150	Random Radiations	By " Diallist "						
OF PUBLICATION	152	Unbiased	By " Free Grid"						

Offices: Dorset House, Stamford Street, London, S.E.1

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FRAME GRID

VALVES FOR TELEVISION

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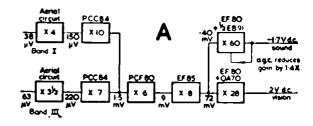
With a suitable valve line-up, receivers with two i.f. stages can be designed for various degrees of improved performance. Frame grid valves can be used throughout the tuner and i.f. stages, or some conventional valves can be retained, depending on the sensitivity required. Two classes of receiver satisfy all U.K. requirements: one for good signal strength areas, and a more sensitive receiver for difficult areas.

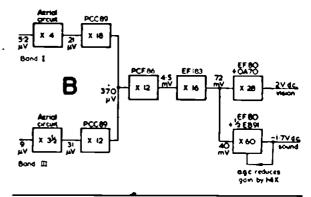
Reception is practicable at vision carrier levels down to $20\mu V$ r.m.s., with enough gain to give 2V d.c. at the detector. To ensure this sensitivity and to allow for deterioration, a fringe receiver should be designed round a nominal sensitivity of $10\mu V$.

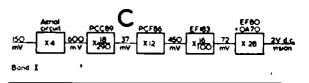
Signal levels in a conventional receiver are shown on line-up 'A'. The Band III sensitivity of $63\mu V$ falls short of the $10\mu V$ requirement by 16dB. Since frame grid valves have a gain advantage of about 6dB over conventional valves, this deficiency can be made up by the use of three frame grid valves.

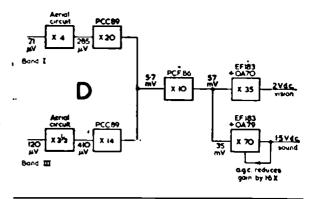
The PCC89 gives improved gain, it has a better noise factor than the PCC84, and it can handle an input signal five times greater without crossmodulation. Thus one of the frame grid valves is used in the r.f. amplifier. The other two should be in stages which are common to sound and vision, so that maximum benefit can be obtained. In line-up 'B' the PCF86 is used in the mixer and the EF183 in the common i.f. stage. The EF80 is retained in the separate i.f. stages. The most severe requirements are met, and the maximum usable gain for low signal levels is provided. In normal service areas, the gain is reduced by a.g.c. for large signals, as shown for Band I at 'C'. Rather greater signals than 150mV can be handled.

A receiver for high signal strength areas does not require 10µV sensitivity, and all requirements can be met by receiver 'Al; but frame grid valves would give the required gain with one less i.f. stage. Line-up 'D' uses the same tuner as the high-sensitivity receiver; but there is no common i.f. stage, and the EF183 is used instead of the EF80. The performance shown is for maximum sensitivity. A signal of 225V r.m.s. can be handled. All normal service area requirements are met, and the receiver is economically attractive. The EF184 can be used in place of the EF183, giving 2dB more gain; but the signal which can be handled is reduced.











MULLARD LIMITED, MULLARD HOUSE, TORRINGTON PLACE, LONDON, W.C.1



The proved reliability of the Brimar 'T' range of valves has been "bullt-in" as the result of experience gained from a programme of examination and testing. One of these tests is illustrated. Valves are placed in a rotor inside a chamber which is later evacuated. The rotor is then accelerated at high speed to simulate the effects experienced by super high speed aircraft flying at high aititudes. The information derived from this and other tests on valves for special applications is used to improve manufacturing techniques on commercial types: which makes Brimar the obvious choice when the demand is for a reliable valve.





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VALVE DIVISION: FOOTSCRAY - SIDCUP - KENT - FOOTSCRAY 3333



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USE AGOS x500 STYLI







ARE DOING THINGS IN STYLI

"BELLING-LEE" NOTES No. 14 of a Series Some aspects of G

In the last issue we wrote that this rather big subject could best be dealt with under four headings, (1) Steady acceleration, (2) Vibration, (3) Shock, (4) Bumping and we dealt superficially with Steady acceleration which is generally considered to be in the region of say 20G.

Let us now look at Vibration which is a periodic phenomenon which may be expressed in cycles per second. In an aircraft the vibration of the structure varies considerably in different parts. It is likely to be of a complex nature due to variation of response to the many frequency components fed in e.g. four engines and buffeting of the control surfaces by the airstreams. It is similar to the diaphragm of a loudspeaker, except that in an aircraft there are many sources (or moving coils) feeding in vibrational energy. This energy contains many frequencies. Any one portion of the aircraft may respond to a few of the many frequencies giving appreciable vibration. For example, the larger amplitude vibration in the wing-tips or tail sections will have quite a different period from those experienced in the cockpit. You have only to move a cup of coffee to different parts of a table to see the different vibration patterns appear on the surface of the liquid.

Now these vibrations may affect any part of any equipment free to move and let us say here and now that the movement in an assumed rigidly constructed equipment, vibrated and viewed under stroboscopic light, has to be seen to be believed. But as component manufacturers we are more concerned with the behaviour of components. First of all, there are those whose design calls for resilience and these include plugs and/or sockets. Yes, a resilient plug can be made to vibrate violently in a socket. So the design must be such that this cannot happen at the mechanical frequencies at which the plug or socket is expected to operate, otherwise intermittent contact will result. Relays, of any kind, thermal delay switches and any component containing springs must be checked to ensure that they will not cease to function under the influence of vibration. Other hazards are fuses: where the fuse wire might have become embrittled through an overload, the element will sometimes fracture " under false alarm " through continued vibration at the right (or wrong) frequency,

We have mentioned vibrations set up in the frame of an aircraft, but of course we have also to consider vibrations peculiar to tanks, fighting vehicles and ordinary motor cars or lorries. Vibration tests are carried out on a piece of apparatus which is virtually a large moving coil loud-speaker energized by a low frequency amplifier capable of producing vibrations from 10 to 10,000 cycles per second.

Shock Test

Under this heading come the conditions brought about say in a crash landing of an aircraft, or a collision, or the firing of a gun. It is a very severe condition and may in unusual cases amount to something like 500/ 1,000G. Such a shock may be considered as an acceleration of the duration of a few milliseconds. Unless an equipment is designed to take the shock, the effect will most likely be catastrophic, as comparatively heavy components such as transformers would come adrift and carry all before them. Some anti-aircraft shells have a radar transmitter receiver incorporated in the nose cap, and this must survive the shock that drives the shell up to 50,000 feet.

Under the influence of shock, unretained fuses or valves will fly out of their holders and delicate mechanisms would be destroyed. The test rig for shocks is a device rather like a plie driver. A very heavy weight is used as the vehicle, this is lifted up to a predetermined height and allowed to fall freely. To ensure controlled results, the bottom of the weight or vehicle has a specially designed spike which plunges into a pot of solid lead which takes the part of the hard rubber in the bump test.

Bump Test

This is usually described as a succession of shocks of lesser magnitude. A bump test rig is generally a cam-operated device which lifts a "vehicle" of predetermined weight and allows it to drop freely on to a bed of hard rubber of known resistance. The frequency of operation can be controlled so that secondary shocks caused by bouncing are eliminated. It is an essential feature of such controlled tests, that the weight of the "vehicle" is heavy compared with the weight of the articles being tested.

The standard test delivers a series of bumps having a peak value of 38-42G.

The three articles on the Smith Chart by Mr. R. A. Hickson of this Company will be available in reprint form, and a copy will gladly be sent to readers of this column on request to Belling & Lee Ltd.

Advertisement of BELING & LEE LTD. Great Cambridge Road, Enfield, Midds.

"BELLING-LEE" MINIATURE COAXIAL

Plugs & Sockets



L.1417/FP/Au or Ag.
MINIATURE FREE PLUG
.L.1417/FS/Au or Ag.
MINIATURE FREE SOCKET

This is a miniature version of the coaxial connectors L.734/P and L.734/J, for use with cables having an outside diameter of 0.16 in. This connector has a robust cable clamp intended for use in miniaturized equipment where reliability is a prime consideration. Available with contact surfaces either gold-plated (/Au) or silver-plated (/Ag). They can be mated with the coaxial inserts in the miniature unitor range L.1391.

Max. Working Voltage: 200 V. d.c. Voltage proofs 1,400 V. d.c. Cable Size:

Outside dia. 0.16 In.
Max. dia. over outer conductor in In.
Max. dia. over inner conductor
0.033 In.

Finish

Inner conductor, gold-plated brass. Outer conductor, aluminium alloy. Dielectric, P.T.F.E. Circlip (L.1417/FS), Nylon.

Weight: 1.3 gm. (0.05 oz.).

LI417/CS. MINIATURE CHASSIS SOCKET

This socket accepts the plug L.1417/FP described above. It has a nylon circlip, insulated body, and the socket is available gold-plated (/Au) or silver-plated (/Ag). Weight 0.7 gm. (0.02 oz.).

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MARCONI

COMPLETE SOUND BROADCASTING SYSTEMS

Aspects of design

This is the Twentieth of a series of special features dealing with advanced problems in television and radio circuit design to be published by The Ediswan Mazda Applications Laboratory. We will be pleased to deal with any questions arising from this or other articles, the twenty-first of which will appear in the April 1960 issue.

During the line flyback interval the deflector coil current in a television line output stage executes slightly more than half a cycle of a cosine oscillation from a high value in one direction to a similar high value in the opposite direction. The purpose of the efficiency diode is to terminate the flyback interval by coming heavily into conduction shortly after the reverse peak of current is established. This conduction returns the energy in the coil to a storage capacitor where it serves to boost the HT supply and in so doing it provides the first part of the follow-ing scanning stroke.

In the course of the Ayback oscillation the diode is subjected

to a high peak voltage in the reverse direction. This, together with the heavy conduction at the beginning of the scanning stroke represents a very exacting condition of operation and every effort must be made to see that the limiting ratings are not exceeded.

In domestic television receivers the diode is generally used in the auto-transformer circuit of which the relevant part is shown in Fig. 1(a) (See also Aspects of Design No. 8). Occasionally, in special equipment, the diode may be connected to a separate winding on the line transformer, Fig. 2(a). The limiting ratings of the valve are published to cover both uses.

to the HT supply and during the flyback interval the cathode receives the high voltage positive pulse. During the whole of the scanning stroke the cathode usually remains at or near HT potential. In the auto-transformer circuit, the diode anode is connected

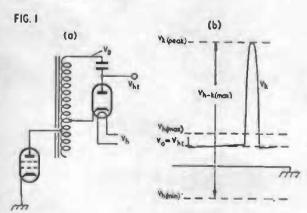
In most receivers the diode heater is connected near the high potential end of an a.c./d.c. series heater chain and in consequence the whole heater may swing as much as 300 volts positive and negative to chassis during the a.e. supply voltage cycle. These conditions are shown diagrammatically in Fig. 1(b).

conditions are shown diagrammatically in Fig. 1(b).

If the diode is connected to a separate winding the working conditions are different. The diode cathode is connected to the boosted HT potential and during the flyback interval the diode anode receives a high voltage negative pulse. During the scanning stroke the diode anode is at or near boosted HT potential. Again the heater swings positive and negative of chassis potential during the supply voltage cycle. These conditions are shown diagrammatically in Fig. 2(b).

In order to explain the limiting ratings applying to this class of valve, consider the Ediswan Mazda U191. Here the relevant ratings are:

Max Peak Inverse Anode Voltage	(kV)	5.0
Max Anode Current (mean)	(mA) 120	150
Max Anode Current (peak)	(mA) 600	450
Max Heater-Cathode Voltage		
(heater - ve, d.c. + peak a.c.)	(volts)	900
Max. Heater-Cathode Voltage		
(heater - ve, pulse)	(kV)	5.0
Max Heater-Anode Voltage		
(heater + ve, pulse)	(kV)	3.0
These ratings should be considered	from the poi	int of view of
1) Current and 2) Voltage.		



EFFICIENCY DIODES FOR TV LINE **OUTPUT STAGES**

I. Current Ratings.

First consider the current ratings. These two ratings must be used together. If the working mode of the timebase is such that the peak to mean ratio of the diode current is 5.0 then a mean current of 120 mA and a peak current of 600 mA must not be exceeded. If however we are using a timebase with long diode conduction, as is usually the case in present-day European practice, the peak to mean ratio may be only 3-0 and a mean current of 150 mA and a peak current of 450 mA must not be exceeded.

2. Voltage Ratings.

Next consider the voltage ratings. A Maximum Peak Inverse Voltage of 5-0 kV is quoted. This is a Design Centre Rating and refers to Maximum Peak Voltage between anode and cathode in the reverse direction. An Absolute Maximum figure is also given for this rating. This must not be exceeded even under the most adverse conditions, i.e. high mains voltage, limit components, no synchronising signal and hold control at the most adverse and of its purchage reases.

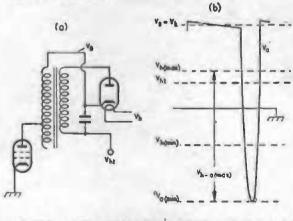
adverse end of its working range.

The Peak Inverse Anode-Cathode Rating is a fundamental feature of the valve but in efficiency diode circuits, unlike conventional rectifier circuits, it is rarely the rating which first imposes a limitation on the designer.

In auto-transformer connection the Maximum Heater-Cathode Voltage (pulse) with heater negative is more likely to set the limitation. This refers to potential difference between cathode when at the positive maximum of the flyback voltage and the heater when at the negative minimum of the supply voltage cycle (Fig. 1b). The table gives an absolute maximum value for this rating.

With the diode on a separate winding the most probable limiting factor is the Maximum Heater to Anode Voltage. This refers to the potential difference between the anode when at the negative peak of voltage during flyback and the heater when at the positive maximum of the supply voltage cycle. Now the valve is primarily designed with auto transformer operation in mind and the cathode is therefore connected to the top cap. This allows a higher voltage rating between cathode and heater than between anode and heater (Fig. 2b).

The rating Maximum Heater-Cathode Voltage (heater negative, d.c.+peak a.c.) is of greater consequence in the circuit of Fig. 2(a). As the cathode is connected to the boosted HT supply voltage, there is a high d.c. plus low frequency a.c. voltage between heater and cathode. The insulation requirements for this voltage stressing differ from those for pulse voltages, so a separate rating is given. In an auto-transformer circuit this rating refers to voltage between cathode and heater during the long forward stroke of the timebase but this rarely represents a limitation.



Associated Electrical Industries Ltd Radio and Electronic Components Division Technical Service Department 155 Charing Gross Road, London, W.C.2 Tel: GERrard 8860. Grams: Sleswan, Westcont, London

EDISWAN MAZDA 10L14

The 10L14 is a VHF Twin Triode with separate cathodes, for use as RF Amplifier and Self Oscillating Mixer in FM broadcast a.c./d.c. receivers. The RF amplifier ensures a low noise input stage and reduces the radiation from the local oscillator. Great care has been taken in the design of this valve to provide effective internal screening between the two sections.

internal screening between the two sections.

Henter Current (amps) Ib 0.1

Henter Voltage (volts) Va 26

Tentative Ratings and Characteristics MAXIMUM DESIGN CENTRE RATINGS

Harriston Debien Charlette Itt		
Anode Dissipation (either section) (watts)	Pu(max)	2.5
Total Anode Dissipation (watts)	P(a"+a") max	4.5
Anode Voltage (volts)	Va(max)	250
Heater to Cathode Voltage (volts rms)	Va-k (max) rms	90
Cathode Current (mA)	Ik(max)	15

INTER-ELECTRODE CAPACITANCES (pF)

3.0
1.5
1.2
0.18
< 0.04
socket.

CHARACTERISTICS (EACH SECTION)

Anode Voltage (volts)	V _a	170
Anode Current (mA)	1.	10
Grid Voltage (volts)	V _e	-1.5
Mutual Conductance (mA/V)	K on	6,2
Amplification Factor	ja .	50

TYPICAL OPERATION AS RF AMPLIFIER

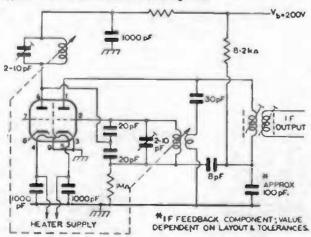
Supply Voltage (volts)	V.	170	170
Anode Load (kΩ)	R.	1.5	1.3
Anode Voltage (volts)	V.	155	160
Anode Current (mA)	Le	8.7	6
Grid Blas Voltage (volts)	V	-1.4	-2
Mutual Conductance (mA/V)	g _m	6	4.7
Valve Anode Resistance (δν./δί.) (kΩ)	Ta*	8.4	10.5
Equivalent Grid Noise Resistance (Ω)	Rey	500	650

TYPICAL OPERATION AS SELF OSCILLATING MIXER

Supply Voltage (volta)	Vb	170	200
Anode Load (kΩ)	Re"	4.7	8.2
Grid to Cathode Resistance* (MΩ)	Rende	1	1
Anode Current (mA)	In-	4.8	5.2
Peak Heterodyne Voltage (volta)	Vhet(pk)	4	4
Conversion Conductance (mA/V)	Be .	2.2	2.3
Valve Anode Resistance (δv./δia) (kΩ)	Fa"	16	15

*IF feedback voltage tends to stabilise oscillator performance and permits this relatively high grid leak.

Typical Circuit of IOLI4 as Self-Oscillating Mixer



Associated Electrical Industries Ltd

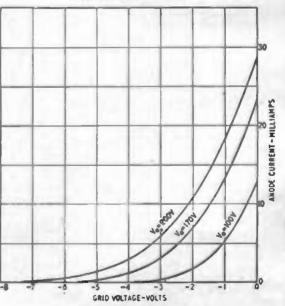
Radio and Electronic Components Division Technical Service Department 155 Charlog Cross Road, London, W.C.2 Tel: GERrard 8660. Grams: Sieswan, Westcent, London



Note: It is recommended that the triode on pins 6, 7 and 8 should be used as the RF amplifier and that on pins 1, 2 and 3 as the self oscillating mixer.

MAXIMUM DIMENSIONS (mm)
Overall Length 56
Scated Height 49
Diameter 22.2

Characteristic curves of Ediswan Mazda Valve Type 10L14 (each section)



EDISWAN

MAZDA

CRC 15/53

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Temperature compensating range

P 100, 3-30 pF N 750 6-100 pF Capacitance tolerance: ±0.5 to ±20% Working voltage: 500 V d.c. Diameter: 0.2 in. to 0.5 in. Thickness: 0.1 in. to 0.25 in.

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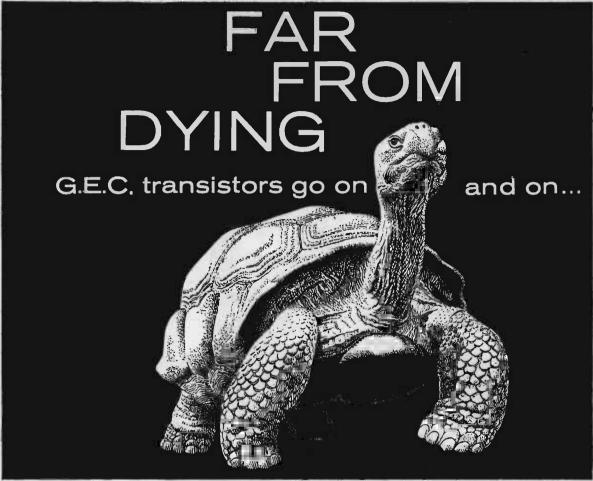
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- * Five position turret including one special position providing facilities for: easy withdrawal of pick up tube through turret without opening camera sides; and mounting of diascope or non standard type of lens.

E.M.I.'s new 203

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retained with additions and Improvements. Our high standard which has made these recorders famous has been maintained, resulting in their being chosen for the foremost musical centre in this country.

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coll feedback distortion is under 0.1% and when arranged for tertiary feedback and 100 volt line it is under 0.15%. The hum and noise is better than—85 dB referred to 30 watt.

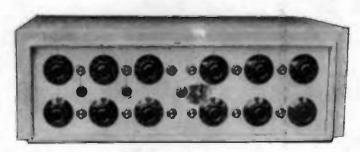
It is available in our standard steel case with Baxendale tone controls and up

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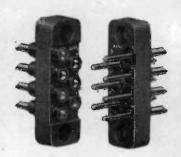
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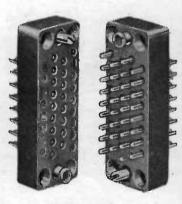
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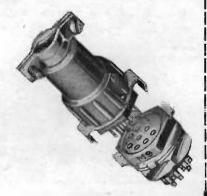


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throughout the world*

Extract from Test Report by J. C. G. Gilbert reprinted from the Music Trades Review, also reprinted in our advertisement in the October issue of this magazine. The full two-page Test Report and an illustrated brochure on the amplifiers will be sent you on request. be sent you on request.

The "Point-One Stereo" pre-amplifier is designed so that it can be used with any Leak monaural power amplifier or a combination of any two Leak monaural power amplifiers additionally to its more normal use with the "Stereo 20" or "Stereo 50."

"The Point-One Stereo "pre-amplifier is probably the most comprehensive unit in existence covering every requirement for stereo tape, disc and radio plus monaural amplification for any form of input signal... it is difficult to think of any additional requirement that one would ever wish. The equipment performs with the high performance always associated with the tradition of Leak equipment. It is a fine example of design and construction, and the pre-amplifier can be used with any other Leak main amplifiers. How the pre-amplifier can be sold for a little as £21 can be answered only by Harold Leak...

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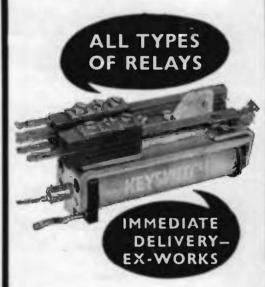
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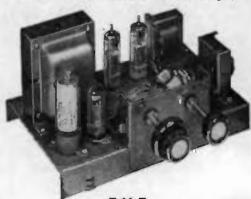
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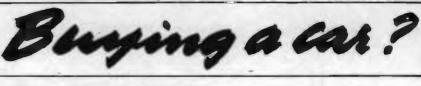
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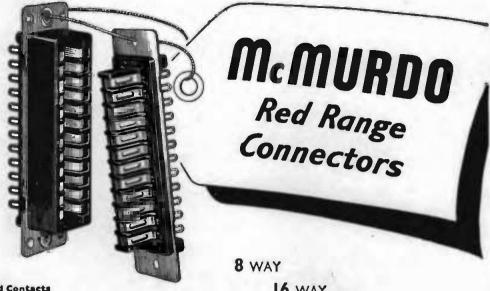
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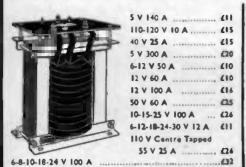
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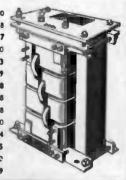
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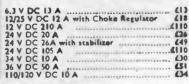
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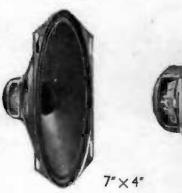
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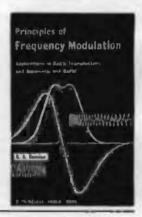
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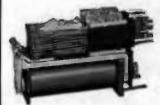
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4. C/O 4/6 14	V		5.000		6/6	8/4
6. C/O 6/6 24	1-	14	10,000		91-	14/6
8 C/O 8/6 37	Ý- "		20.000		14/-	17/6
Other build ups to order		-			14/-	20/-
types of relays built to		100				_
specification.	,	7	80,000 Slugged	coile	AVICE	

SIEMEN'S HIGH SPEED CIO RELAYS

250 + 250	ohm	Twin	Coils	6/6	1,000+1,000	ohm	Twin	Coils	10/4
850 - 850			44	8/6	1,700 + 1,700		-	**	17/6

G.E.C. MINIATURE SEALED RELAYS

No.	Ohme.	Build Upe	Voltage	Price
Z.530002	180	4C	12	41 2 6
Z.530005	2	2C	ï.a	12 6
Z.\$30006	40	2C	6	15 0
Z.530008	670	20	24	17 4
Z.530010	40	2C 2K	6	17 4
Z.530011	180	2C 2K	12	61 2 6
Z.530014	2	IC	1.3	10 6
Z.530015	40	IC	6	12 6
2.530016	180	IC	12	19 6
Z.530018	2,500	IC	48	41 2 4
2.530019	2	2C 2K	1.3	14 4
Z.530020	2	4C	1.3	16 6
Z.530022	2	M.B.	1.3	12 4
2.530023	2	28 2H	1.3	12 6
Z.\$30024	40	2M	6	12 6
Z.530025	40	M.B.	6	12 6
2.530027	180	12M	12	17 4
Z.530028	180	M.B.	12	17 6
Z.530031	670	M.B.	24	17 6

S.T.C. MINIATURE SEALED RELAY

4184GD 700 2C 1/6 Post & Packing on all relays. Send for lists



ROTARY TRANSFORMERS

Delivery ex stock. Quotations on application.

H.T. 31 Input 11.5 v. Output 250 v. at 125 mA.

H.T. 32 nput 11.5 v. Output 490 v. at 65 mA.

AS SUPPLIED TO GOVERNMENT DEPARTMENTS AND LEADING MANUFACTURERS. NEW AND BOXED.

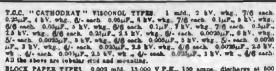
ROTARY **TRANSFORMERS**

Made by DELCO TYPE 1 27/6. P. 8 P. 3/6. TYPE 2, 37/8. P. 8 P. 3/6. TYPE 1. Dual voltage 12 or 24 v., input 265 v., 120 mA. output; 500 v., 26 mA. out-

Type 2. 12 v. input 275 v. 110 mA. output; 500 v., 50

mA. output.
Both types dual output.
MADE IN U.S.A.

OTHER DYNAMOTORS IN STOCK, SEND FOR LIST



no use spore are towars sive and mounting.

BLOCK PAPER TFESS, 0.000 mild, 15.000 V.P.E., 100 amps., discharge at 500 times per second, six 165 s 5 s 5 in., veramic used, 25/6 each, 3/* post, 0.40 mild, 16 W.* wig at 71 deg. C., cernante innuit, size 16 to 123 miles, 36% sach, 5/* post, 10 mild, 1,500 v. wig, 15/* each, 6 mild, 1,500 v. wig, 15/* each, 6 mild, 500 v. wig, 5/6 each, 6 mild, 5 to V. wig, 6/6 each, 6 mild, 6 to V. wi

25FF, AERIALS. Heavily galvanised steel tubes, four 5ft., 2in. dis. steel tubes fit into the base, these are then paged to the ground. Must four sections, as advertised Jan. Issue, £12/10/-, carr. £l.

POWER UNITS 100-230 v. A.O., input, 24 v., at 3 amps. or 12 v. twice at 3 amps. each. Tropical rating, switched and fused, etc. Pits 19ts., rack £3/15/-, carr. 7/6. SHOOTHING UNIT, Yor the above power supply, £2, tarr. 7/6. (As nivertised in the Jan. issue with Photo.)

FREL RACES. Standard type U channel steel 6ft. 19tn., 24, 107- post. Non-standard type 3ft. 19tn., 22, core. 7ft.

RELAY BOXES with 10 (600 type Relays) and 5 rectifiers to operate the relays in mest metal box, 8 × 6 x 50c., 307-, each. 267 post.

WHEAT-TONE BRIDGE in a beautiful oak case croire zero galvanometer 2.5 mA-4 stud switches 0-10, 0-100 ohms, 0-1af. aise 16 s 7g s 6a., 30f. sech, 37 post. EFF DRIVER DNIT. Freq. 100-155 encis., salves 2, 450-405/c; C. CWIOTS, B. CWIOTS, B. CWIOTS, D. CWIOTS, B. CWIOTS, D. CWIOTS, B. C et 10/m

VALVES. 6K70, 2:6; ARP12, 2:6; VR54, 1/-, GLACCO, 15/-, CV73, 2:6; CV121, 12/6; PK96, 10/-, CV23, 15/-, V8110, 3/- sach. All valves are new boxed.

13/6. FARM, 10%, CVES. 15/c. VS110, 3/c acca. All varies are saw board.
AMERICAN L.T. TRANSPORMERS. Potted type, finished in black erackle and
very conservatively rated. (1) 230 v. input. 3x5.v. C.T., at 3 amps. each and 4v. at 3x
output. 13/6 such. (3) 230 v. input. 3x5.v. C.T., and 3a., and 6.3 v. C.T., at 3 a. output.
13/6 such. (3) 230 v. input. 28 v. volts at 3 a., and 6.3 v. C.T., at 3 a. output.
13/6 such. (4) 230 v. input. 28 v. at 2a., and 3 v. at 1 a., 12/6 each. (5) 230 v. input.
3 v. 5 v. 5 a. C.T., 1.6.3 v. 3a., 20/6 each. (All those transformers are new and
bored, please include postage 2/6 each.)

RANGE CONVERTORS used with the R204 Freq. 113-900 Kc/a, valvas ARTE2. EFTS. Mulrhead slow motion drive, 32/6, carr. 7/4.

AR 83 FILTER PACE. 3 H 4 mid., 500 v., 25/- each, P.P. 2/4.

PLEASE INCLUDE POSTAGE ON GOODS

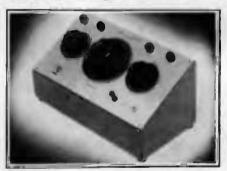
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"The above, in conjunction with a Signal Generator and Valve Voltmeter, is capable of making a variety of measurements by the resonance method in the approximate frequency range of 50 Ke/s to 20 Mc/s." PRICE £16-10-0d.

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LONDON'S GREATEST DEALERS IN RADIO AND ELECTRONIC EQUIPMENT

HEAVY DUTY SLIDING RESISTORS. 26 Ω 6 A, double tube slider control 45/~ 73 Ω 1-3 A. Completely enclosed single tube slider control 35/~ 120Ω 1.75-0.9 A., completely to anclosed single tube slider control 32/6. ly enclosed single tube slider control 12, 1,25 Q 25 A, double tube slider control 27, 1.2 & 25 A., gened drive control 17/6. 1.2 Ω 14 A., gened drive control 18/-. 3 Ω 10 A. 12/6. 1.2 Ω 15 A. 10/6. 1 Ω 12 A. 8/6. 11 Ω 4.5 A. 12/6, All single tube slider control 5.3 Ω 8 A., fixed 10/-. 605 Ω 2.8-0.4 A., fixed 10/-. Carr, on all resistors 2/6.

HEAVY DUTY OHMITE RHEOSTATS. 2Ω7 A, 18/-.. 15 Ω 2.24 A. 12/6. 25 Ω 0.75 A. 5/6. 350 Ω 25 watt 3/6. 1,280 0.14 A. 6/6. 25 Ω 2 A. 15/-.. 58 Ω 0.6 Q, 8/6. All types ±in. spindle. P.P. on all rheostats 2/-.

G.E.C. XPELAIR. A.C. 200-240 volt. 9in. window lans. Brand new in maker's carcons. 19/19/6. Care, 4/-.

SPERRY H.T. TRANSFORMERS. Tapped Pri. 110-250 v. Sec. 450-0-450 v. 106 M.A. 6.3 v. 2 A., 6.3 v. 1.5 A., 5 v. 3 A. Potted type. Brand new. 33/-. Carr. 5/-.

BRAND NEW TELEPHONE CABLE. Type D3 twin. 500-yard drums. 35/-. Carr. 7/6.

Also available brand new D3 single on 1/3 or one mile drums. De twin on heavy wood drums. Let us know your requirements.

WESTINGHOUSE DOUBLE WOUND STEP DOWN TRANSPORMERS. 250-230-210 v-110 v. Tropically rated at 400 watts. But guaranteed to give 600 watts. Brand new. 65/10/- Carr. 7/6.

HEAVY DUTY AUTO TRANSFORMERS. Tropically rated at 5 kVA. Tapped 250, 240, 230, 220, 120, 115, 110, 105 volts. Completely anclosed in metal case, Size 23 x 14 x 11 inches. Weight approx. 2 cwt. Brand new 215. Ex

We have London's largest selection of Auto Transformers from 40 watts to IS kVA. Available from stock. Let us

ADMIRALTY INTEGRA-TORS. Type AS91, Incorporat-ing a very fine galvanometer movement, coil 40 ohms. Centre zero to F.S.D.I microamp. Small meror one metre radius. A very useful laboratory instru-ment, Guaranteed in perfect condition. 59%. Carr. 4/-,

WHEATSTONE BRIDGE. Built-in polished wood case size 16 x 7½ x 6in, with four stud switch controls, and centre zero 2.5 M.A. galvanometer. 2.5 M.A. galvanometer. In perfect condition, 37/6. Carr. S/-,

A.C. CHECK METERS. 200-250 v. 20 amp. 22/6. 10 amp. 19/6. Carr. 3/6. Reconditioned and guarantoed.

SANGAMO SYNCHRONOUS MOTORS. A.C. 200-250 v. 1§in. dia., 7/6. P.P. 1/6. Assached to Gear Train Unit con-taining over 30 Gear Wheels. 10/m. P.P. 1/6.

YENNER EIGHT-DAY CLOCK-WORK TIME SWITCHES. One make and one break every 24 hours. 5 amp. 230 volt switch con-tacts. Complete with Key and Mounting Bracket in perfect condition. 29/6. P.P. 1/6.

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Brand new, 44,6. P.P. 31-.

SPECIAL OFFER OF B.A. SCREWS,
STEEL. 4 BA Jin. C.S., 15 gross 27/6. P.P. 31-.

Or 2/6 per gross. Post free. 4 BA Jin. steel
C.S., 20 gross cartoni, 32/6. P.P. 31-. Or 2/6 per
gross, post free, 2 BA brass Idn. C.S., 5 gross
cartons 15/-. P.P. 2/-. 4 BA steel R.M., 1 Jin.
S gross cartons 15/-. P.P. 2/-. NUTS, BOLTS,
WASHERS: Special bargain offer, 5/- carton
2, 4, 6 BA nuts, bolts and washers.

SPECIAL OFFER LATEST M.O.S. LT SUPPLY UNITS THE CHEAPEST CHARGER ON THE MARKET THAT GIVES THESE SPECI-FICATIONS

S.T.C. BATTERY CHARGER **TYPE ZB 10234**



put 100-260volts 45-65c/s. D.C. 24 volts 10 amps at max ambient tem ture of FURFAR-5 . e d m a x output of 20

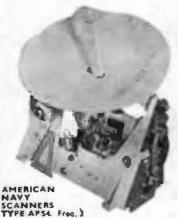
A.C. In-

amps. All components are rated by manufecturers at this current. The charger is fitted with 20 amp. fuses on the D.C. output, 10 amp. fuses on the A.C. input, 24in, 0-20 M.C. D.C. ammeter. On/off full charge/trickie charge switch. Heavy duty output terminals and mains neon indicator lamp. Behind control panel are mounted full charge ballast and trickle charge reistances.

These units are designed to charge all 24 volt lead-acid battery combinations. That is two These units are designed to charge all 24 volt lead-acid battery combinations. That is two 12 volt or four 6 volt batteries in series at a 20 amp, max, rate. Can also be used for trickle charging 24 volt batteries at 125, 350 and 700 m.a.; are ideal for the electronic industry, research laboratories, schools, etc., as a general purpose L.T. supply unit.

Supplied brand new at a fraction of maker's price. Size: 2ft, x lft, 3\(\frac{1}{2}\)in. x 2ft, 8in. Weight: 34llb. £22/10/-. Ex, warehouse.

HUGE STOCKS OF LABORATORY GLASS WARE. Flasks, beakers, burettes, cylinders, measurers, distilling tubes and jackets. U and Y tubes. Offered at a fraction of maker's price. Callers only.



centimetres, operate on 12-24 v. Contain geared motors, relay awitch equipment, heavy goar trains, etc. Supplied new in original packing cases at a fraction of maker's price. £3/19/6. Carr, 10/-.

HEAVY DUTY LT. TRANSFORMERS.
All ratings Tropical and in perfect condition. No.

1. Pri. 200-240 v. Sec. 12 v. 40 a., completely enclosed in metal case. 52/6 Carr. 4/- No. 2. Pri.
230 v. Sec. tapped 4-6-11 v., 200 a., E##10/Carr. 7/6. No. 3. Pri. 200-250 v., Sec. 50 v.,
30 a., £6/10/-. Carr. 7/6. No. 4. Pri. 200-240 v.,
Sec. 50 v., 20 a., £6/10/-. Carr. 7/6. No. 5.
Pri. 200-250 v., Sec. tapped 28-29-30-31 v.,
21 a., £4/17/6. Carr. 7/6. No. 6. Pri. 100-250
v., Sec. two separate windings, tapped 15-1617 v., 4 a., 35/-. Carr. 4/-. No. 7. Pri. 220-240
v., Sec. two separate windings, 65 v., 50 a.,
6 v. C.T. 15 a., 6 v. C.T. 2.5 a., £6/19/6. Carr.
7/6. No. 8. Pri. 220-240 v., Sec. 6.3 v. 15 a.,
25/-. P.P. 3/6. No. 9. Pri. 220-240 v., Sec.
6 v. C.T. 4 a., 5 v.,
C.T. 4 a., 5 v., C.T. 4 a., 4 v. 4 a., Potted Type,
21/6. P.P. 3/6. No. 10. Pri. 220-240 v., Sec.
C.T. 4 a., 5 v., C.T. 4 a., 4 v. 4 a., Potted Type,
21/6. P.P. 3/6. No. 10. Pri. 220-240 v., Sec.
C.T. 4 a., 6.3 v. 4 a., Potted Type,
21/6. No. 18. Pri. 220-240 v., Sec.
C.T. 4 a., 6.3 v. 4 p. Potted Type,
21/6. No. 11. Pri. 115-230 v., Sec. 5 v. 15 a., 15 kV.
Insulation, 37/6. Carr. 5/-. No. 12. Pri. 220-240 v.,
20-240 v., Sec. 45 v. 2 a., 17/6. P.P. 3/6. No. 11.
Pri. 200-240 v., Sec. tapped 9-15 v. 4 a., 12/6.
P.P. 2/6. No. 14. Pri. 220-240 v., Sec. tapped
10-17-18 v. 10 a., 52/6. Carr. 4/-.
OIL FILLED HEAVY DUTY L.T. TRANS-

OIL FILLED HEAVY DUTY L.T. TRANS-FORMERS. Pri. 420 v., 400 v., 380 v. Single phase. Sec. 19 v. 150 smps. Weight 141 lbs. Supplied dry 610. Carr. 15;-

ADMIRALTY THREE-PHASE TRANS-FORMERS. Pri. 400-440 v. 50 cycles. Sec. 50 v. 6 amp. Completely tropicalised. Size 78 x 14 x Sin. Weight approx. 601b., 85/-, Carr. 7/6, Brand new in maker's cases.

A.M. L.T. SMOOTHING CHOKES. Resistance & ohm. Ideal for smoothing 12-24 volts D.C. 5 amps. Tropically rated. Unused, 15/-,

LT. SUPPLY UNIT No. 19 YA 8087.
A.C. Input 100-250 v. D.C. output tapped 12/24 volts, continuous tropical rating, 3 amps.
Built-in mets1 case 17 x 7 x 6\(\frac{1}{2}\)in, with fuses and switch, An ideal LT. supply unit for operating relays, contactors, battery charging, etc. In perfect condition, £3/17/6. Carr. 7/6.

G.E.C. L.T. SUPPLY UNITS TYPE O.S. 1773 G.A. A.C. input 200-240 v. D.C. output 24 volts 10 amps. Tropically rated. Built-inmetal case size 20 x 15½ x 10in. Supplied brand new in maker's cases, £13/10/-, ex warehouse.

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ohms 2 CO 8/6. F.F. 1/6 each.

BRITISH 3000 TYPE. RELAYS NEW
M.O.S. 6.000 ohms, 4M 28 12/6; 6,000 ohms
2M 28 10/6; 1,000 ohms 2 CO 8/6; 250 ohms
4M 4B 10/6; 10,000 ohms 3M 8/6; 2,000 ohms 18
6/6; 200 ohms 1M 1B 7/6; 600 Type 750 ohms
1M, 600 ohms-2 CO, 150 ohms 1CO 5/6 each.
All Relays guaranteed and checked before
despatch. Please Include aufficient for postage.

SPECIAL OFFER. H.D. TWO CIRCUIT TYPE AUTO WOUND TRANSFORMERS A.M. TYPE No. 2773. Input 225-230 v, with switch in run position. Output 225, 230, 235, 240, 245, 250, 255, 260, 270 volts, 75 amps. With either switch in start position. Output No. I or 2, 270 v., 290 v., 310 v., 100 amps. I minute in two hours. Other outputs at in two hours. Other outputs at in two hours. r minute in two nours. Other outputs as in run position but at 50 amps. The transformers are built in heavy metal cabinets approx, weight 3½ cwt, with sloping deit front on which the startfrun switches and 3.0–100 MJ, ammeters are mounted. With an alteration of the exterior wiring 100 volts at 35 amps can be obtained. £18/10/- ex warehouse.

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NEW-The "CONTINENTAL-6" "For Style, Quality Performance and Value for Money" COMBINED TRANSISTOR PORTABLEICAR RADIO SUPERHET

SPECIFICATION

- ★ 195,to \$60 metres on medium wave. 1,150 and 1,800 metres on long wave.
- 400 mW. push-pull output.
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 HI-FI SPEAKER: 5 inch large magnet.

 Double tuned IF's.

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 Resistor and Condenser leads pretrimmed.
- rinted circuit board marked with component numbers.

XA102, 2-XA101, XB103, 2-XC101, 2-DIODES.

TRANSISTOR "8" STILL AVAILABLE AT (10-19-6 (pp. 2/6) FREE BOOKLET ON REQUEST

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components available separately. Send for descriptive leaflet and prices.

A highly sensitive and selective portable fully tuneable on medium and long waves. Performs equally well as a car radio. Low running costs, good looks and case of construction combine to produce a radio equal to any commercial receiver In the 20 gn. class.

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2-WATT POWER STAGE For use with "Continental," Works from 12-volt supply. Overall size 49 x 32 x 22in. All parts with Power transitor, less speaker, \$2/6. P.P. 2/-.

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MAJOR-2 * 4-stage reflex * Medium wave; (2-Transistor Pocket Radio)



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GOOD RECEPTION ANYWHERE!

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* 3-stage Reflex * Medium wave * Ferrite aerial

★ Size 3 x 2 x

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High sensitivity and ealectivity combine to give excellent reception on both medium and long waves, this set is recommended by us as being one of the easiest-to-build printed circuit transistor sets ever offered.

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20 Mc/s 25 Mc/s 30 Mc/s 50 Mc/s 6 Mc/s 6mW H.Freq. 10/-10mW H.F. " 15/-10mW H.F. " 22/6 10mW H.F. " 30/-120mW Audio 16/-90mW Osc 18/-\$830\$ \$8231 SB231R X8104 XA IO 165mW Power pr. 32/-90mW RF, IF 15/-200mW Power 17/-XC101
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15 Mc/s
60mW 0c. 26/OC45
6 Mc/s
60mW Rf, IF 18/OC71
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D.C. ELECTRONIC VOLTMETER.

6-Ranges. 0-3-10-30-100-300 and 1,000 volts. Input res: 11-meg. constant on all ranges. Sensitivity: 3,666,666 ohms per volt on 3 v. scale.

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ELECTRONIC OHMMETER.
6-Ranges, from 0.1 ohms to 1,000 megohms.
Movement. 200 microamperss. D.C. accuracy ±2%.

COMPLETE WITH INSTRUCTION BOOK AND TEST PRODS, BRAND NEW.

Input 110-250 volts A.C.

ONLY £12/10/0 P.P. 3/6

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T/X	TYPES AND	SPECIAL PL	PRPOSE VAL	VES
EF91 \$/-	705A 15/-	807 7/6	1625 \$/-	\$800 45/-
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ABOVE INCOM-PORATING PARTHIDGE OUTPUT TRANSPORMER 210- 0000-

MULLARD'S PRE-AMPLIFIER TONE CONTROL UNIT

Employing two EF90 valves, and designed to operate with
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The 20-Valve Pre-Amplifers
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Adds "HI-FI" Tope Recording to your existing Audio Installation.

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£43.10.0 £51,10.0

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TAPE PRE-AMPILINGORPORATING THE NEW FERROXCUBE POT CORE PUSH-PULL OSCILLATOR and 3 SPEED TREBLE EQUALISATION by means of the lacest FERROXCUBE POT CORE INDUCTOR.



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THE NEW COLLARO "STUDIO" TAPE DECK.
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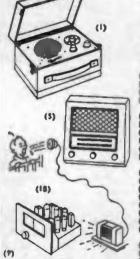
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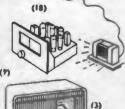
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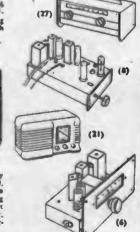
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(24) MULLARD TTPE "C" Tape pre-amp			0	3/6	2/6	11.
(25) JASON J.3-3 Stereo pre-amp.	(15		0	26	2/6	di
(26) JASON J.4-4 Stereo pre-amp	-	14	V	40	3/-	1 (
	C)	19	4	3/6	2/6	113
(26) NEW JASON FRINGE AREA F.M. TUNER as above	610	19	ě	3/6	2/6	- 1
(28) NEW JASON FRINGE AREA F.M. TUNER as above (29) PULLIN Script 90 TEST METER	(5	19	ě	2/6	1/6	,
(30) R.C. Super Personal Portable, I-valve (phone extra)	61	15	0	2/6	2/-	
(31) R.C. Super Personal Portable, 2-valve (phone extra)	a	1	o	2/6	2/-	
(32) R.C. TRANSETTE 2-Transistor Personal Portable	0		Á	2/-	2/-	-
(33) JASON EVEREST 6-Transistor 2-wave Portable	(1)		•	3/6	3/6	K
(34) JASON EVEREST 7-Transistor 2-wave Portable	€15		9	3/6	3/6	1
(35) CLYNE Cathode Ray Oscilloscope	412		4	S/-	10/-	1
(36) Compact Multi-Range Test Meter		17	4	1/6	1/6	
(37) CAR RADIO, Printed circuit, 5-valve Superhet			6	3/6	3/6	
(38) JASON Audio Generator AG10		5	0	3/6	2/-	
(39) JASON Oscilloscope OGIO			0	5/-	3/6	
(40) Super SHORT WAVE RADIO. I valve	61	15	0	2/-	2/-	
Instruction Books which contain full description, easy-to	o-follo	w p		tical	wiring	
diagrams, theoretical diagrams, itemised price lists, etc.,	are f	100	of e	charg	e with	
all parcels but may be purchased separately as shown a	bove.					

NEW! NEW! THE "WAVEMASTER" 7-TRANSISTOR LUXURY PERTABLE

I-IRANDIDIUR LUAUTI FURIABLE
To build yourself! Medium and Long Waves
Push-Pull Superhet A.V.C. Perfect Car
Radio reception, Size 10in. x 64in. x 64in.
at base tapering to din, at top. Very attractive two-tone Vynide covered cabinet with
cream and gold printed escutchoon plate
cream and gold knobs, handle and cabinet
fittings. Weight complete with long-life
7½ voit battery—4½ lb. Haad transistors
throughout, High-Flux 7in. x 4in. Elliptical
Speaker, slow motion tuning, co-axial

throughout, High-flux 7In. x 4In. Elliptical Speaker, slow motion tuning, co-axial socket at rear for direct connection to Car Radio Aerial, Improved reception by use of seven-section plated telescopic serial disappearing into Cabinet when closed, 34in, above Cabinet when fully extended. Construction simplified by bakelite chassis board with the following components already mounted:—Tuning Condenser—I.F. Transformers (3). Oscillator Coll, Trimmer Bank, Output Transformer, Interessage Transformer, Ferrite Rod Aerial, Brackets and Earth Bar. SPECIAL INCLUSIVE PRICE for all required components, full assembly instructions—nothing more to buy—is 410/19/6 plus 3/6 P. B. P.
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(4)

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TAPE SPECIAL III
Trade enquiries imrited
First delivery Famous American Ferrodynamics Acetate Base High Quality Recording Tape.
An enthusiast's "must," Brand new (NOT SUB-STANDARD), 7nn. 1,200ft, on plastic spool, 25f-3 7in. 1,800 ft, on plastic spool, 25f-3 7in. 1,800 ft, on plastic Spool, 25f-3 7in. 1,900 ft, on plastic Spool, 25f-3 7in. 1,900 ft, on plastic Spool, 25f-3 7in. 1,900 ft, on, 5in. 1,900 ft, on, 5in. 1,900 ft, on, 5in. 1,900 ft, or, 5in. 1,900 ft, or



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The latest addition to our comprehensive stocks quality equipment for the constructor. This is an exception sound and robust instrum of the most versatile type, that will be a boon to the seriously minded amsteur, seriously minded amsteur, ser-



A.M. GRAM CHASSIS SPECIAL! (By famous manufacturer). This special offer chassis is being offered for a limited period only and represents the best possible value for money. Spec.; 3 wavebands, Long. Medium and Short. S miniature valves—6.07, 6FIS, 6LD20, NIO8, UIO7. Attractive vertical glass dial (13in, x 3\(\frac{1}{2}\)in,) in red, green and gold on black background. Two-speed dial drive. Full range tone control. Output approx. 4 watts to match 3 ohm speaker, For A.C., mains 110/250 v. Overall size 13in. x 6\(\frac{1}{2}\)in, x 6\(\frac{1}{2}\)in, high, WHILST STOCKS LAST, 67/19/6 ONLY, plus 7/6 P. & P.

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Owing to favourable purchase we can offer strictly limited quantity of these handsome chassis. A.C./D.C. 200/250 v. for Medium and Long Wave plus gram position, incorporates own frame aerial. Valve line upt UI07, NI08, DHI07, WI07 and XI09, Overall chassis size 12 m 54 m 74in, high, Attractive bronze dial wich gold and cream lettering. Dial size II4 x 44in, Scale length 74in, Logging scale provided, Price 67/19/6 only, tax paid, plus 3/6 P. & P. H.P. terms 64 deposit, plus four monthly payments of 22/-



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Both 3 watts output, printed circuit construction, valve line-up EFSC. ELBA, ECCB3, EZBO and EMS4 recording indicator. Leriest Brad High Flux Speaker. Completa with Tape and empty Spool, and Accs 33-1 stick mike with stand. Attractive two-fone Cabinet. Supplied with latest COLLARO Cabinet. Supplied with latest COLLARO. Studio 3-speed deck. Total price 25 guineas. Supplied with B.S.R. single-speed deck, fotal 20 guineas. N.B. Thase are Kits, amplifier supplied unassembled. Full assembly instructions are included. Please add 7/6 for parking and carriage. All parts available separately. Full details on anglication. application.

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Full range of changers, single players, transcription units at usual competitive prices. Interesting H.P. facilities, E.M.I. 4-SPEED STEREO SINGLE RECORD UNIT. Complete with Stereo Head and Sapphire Styli, Brand New and Fully G'teed. ONLY (4)19/6 plus 3/6 P, & P, whilst stocks

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GARRARD RC.121D MK. II STEREO
MONAURAL 4SPEED AUTOCMANGER.
Complete with GCB
plug-in Crystal Head and Sapphire Styll
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Limited stocks. ONLY £11/8/6, plus
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THE LATEST COLLARO "CON-THE L THE LATEST COLLARO "CON-QUEST" 4-speed autochanger in cream with Studio "O" insect. Brand-new, fully guaranteed. 67/19/6, plus P. & P. 3/6, COLLARO "CONQUEST" STE-REO/MONAURAL. Latest type— full guarantee. Brand new. 68/19/6, plus 3/6 P. & P.

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CABINET in two-tone rexine covering for accommodating the above items and ancillary equipment. 75/-, plus 5/- P. & P.

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Nineteen ranges D.C.J.A.C., Current and resistance. Designed and produced for use by the famous Pullin Company. All necessary components at Special Inclusive Price of only 45/19/6, plus 2/6 Pr. & P. Illustrated leaflet with full description available on request.

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Goodmans Bin, x 24In., 3 ohms, 24I-plus 1/6 P. S. P. 10in, Elac High Flux 3 ohm, 39/6 plus 2/6 P. & P. Bin. Celestion High Flux 3 ohm, 39/6 pus 2/6 P. & P. Bin. Celestion High Flux 3 ohm, 32/6 plus 2/- P. & P. Hin. Plessey Tweeter, 15/-plus 1/6 P. & P. R. & A. Type 9120. Hk, II, 12In., 10-12 watts, 3 ohm, 12.000 gauss, 55/- plus 3/6 P. & P. R. & A. Type 8120, Mk, II, 12in., 10-12 watts, 3 ohm, 10.000 gauss, 39/6 plus 3/6 P. & P. 12in. Bakers Selhurst, 15 ohms, 15 watts, 30-14,000 C.p.s., 4/10/- plus 3/6 P. & P. All the above brand new and fully gusranteed.
Special 1 Special 1 Latest E.M.I. SPEAKER BARGAINS

Special! Special! Latest E.M.I. full frequency speaker. Size 13tin. x 8tin., 3 ohm sprech coil. Double cone. Unrepeatable at 39/6 each only.

Plus 3/6 P, & P.

AERIAL TUNING UNIT
ZA9941. This well made ex-ZA001. This well made ex-W,D unit contains a hort of useful components including: I mA. Zin, flush round M/C meter, I mA. Westing-house full-wave meser certifier. round M/C meter, I mA. Westing-house full-wave meter rectifier, 5-pole 5-way heavy-duty silver plated wavechange switch. 3m, dia, silver plated rotary tuning indicator, 350 pF tuning condenser with insulated coupler and 34in, calibrated dial (0–180 deg.), etc., etc., Contained in strong metal carrying case 9in, x 9in, x 8in, with hinged lid. ONLY 27/6 plus 5)- C. & P.

No. 38 AFV WALKIE-TALKIE
A wonderful offer. This famous transceiver unit, with relay operated
SEND/RECEIVE switch, covering 7.4-Mc/s band, range approx. 5 miles. ood condition. ONLY 22/6 plus 2/6 Good condition. P. & P. per unit (less accessories).
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*** ROLEX *** SPECIAL HEAVY
DUTY MAINS/BATTERY AMPLIFIER., Very amart unit housed in grey
crackle finish case with chrome and
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200/250 v. or 6 v. D.C., battery, Valve
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Size: 13½in, x 8½in, x 7½in. Mike and
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HIGH IMPEDANCE LIGHT WEIGHT HEADPHONES LIGHT. HIGH WEIGHT HEADPHONES.
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America America and a seed in up-to-date ships, aircraft, etc. Excellent quality super lightweight low impedance magnetic headphones complete with button microphone attached and plastic are moulder. Absolutably brand new. ear moulds. Absolutely brand new 45/- pair, Plus 1/6 Ps & P.

EVERSHED AND VIGNOLES BRIDGE MEGGERS. Series 2, 250 v. Perfect and complete with leather carrying case. FEW ONLY at carrying case. FEW £19/19/-, plus 7/6 P. & P

EXTRA SPECIAL OFFER!

A small three-valve PORTABLE RECORD-PLAYER AMPLIFIER mounted on baffle 12 x7 in., with High Flux 64 in. Loudspeaker, Valve line-up ECC83. EL84. EZ80. Incorporates separate bass and treble controls. Mast. output 3 watta. Will match all types of high impedance pick-up, Ready to use. £5/12/6 plus 3/6 pl. 8 P. NEW STYLE CABINET NEW STYLE CABINET Inished in two-tone Lestherette, Will scommodate above Amplifier and Baffle without modification, also most types of Ancillary Equipment, Overall size 18 x 13 x 8 jin. Fitted with carrying handle, £79/6 plus 5/- P. 8 P. NOTE. If both items pur-NOTE. If both items pur-chased together they will be supplied at a special inclusive price of \$8/7/6 plus 6/6 P. & P.



JUST ARRIVED!! CABINET SPECIAL!!!



Leading manufacturers' special equipment cabinet— surpose, Will house all tape equipment cabinet—
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up to 8in, or 10in, x 6in,
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Dark green regine covered,
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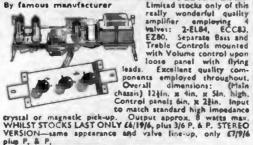
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A SPECIAL HIGH QUALITY PUSH-PULL AMPLIFIER
By famous manufacturer Limited stocks only of this



Limited stocks only of this really wonderful quality amplifier employing 4 valves: 2-EL84, ECC83, EZ80, Separate Bass and Treble Controls mounted with Volume control woon.

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★ BARGAIN CORNER ★ 12 CHANNEL T,V. TURRET TUNER (by famous manufacturer). Brand new, NOT sur-

plus or ex-equipment, 35 Mc/s. I.F. PCC B4 and PCF 80 valves, Comvalves Complete with coils; Band I Channels I to 5, Band III Channels B to 11. In manufacturer's original

carton. Fully guaranteed at only 39/6 plus 2/6 P. & P.
ACOS MIC 39-1. Crystal stick microphone with stand. List price 5 gns. Our price 39/6 plus 1/6 P. & P.

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MIC40. General purpose crystal
microphone with desk stand. Our
price 25/- only plus 1/6 P, & P.
DEAF AID TYPE EARPIECES.
Standard magnetic type complete with
lead and plug. As new. ONLY 12/6
plus 1/- P, & P.
ANOTHER PORTABLE CABINET
BARGAINI Exteading manufacturer's
battery portable actache type case.
Attractive two-tone gray rescine finish.
Size closed 13½in. x 9½in. x 1½in.
Complete with fittings and handle.
Including Medium and Long Wave
frame aerial which fits in Ild. Limited
quantity only at bargain price of quantity only at bargain price of 19/6 plus 2/- P, & P, Brand new, TRANSPORMER SPECIAL. Sup-

19/6 plus 2/- P. & P. Brand new.
TRANSFORMER SPECIAL. Supplied quality half shrouded drop thro'
Mains Transformer. Input 200/250 v.
Output 350-0-350 v. 80 mA.; 6.3 v.
3 amps. S v. 2 amps. Extequipment but
quaranteed O.K. ONLY9/6 plust/I-P. & P.
8in. LOUDSPEAKER. Extequip. as
new. Less transformer. 3 ohm speech/
coil. In attractive cloth covered cabinet.
Ideal for extension speaker. 22/6 plus
I/6 P. & P. Speaker only, less cabinet
at 13/6 plus I/6 P. & P.
BARGAIN I REPLACEMENT
PICK-UP INSERTS. All brand new
and fully guaranteed. Complete with
Sapphire Styll. FONOFLUID 21/each. B.S.R. TGB (less bracket) IS/each. B.S.R. Hi-G with bracket. I8/each. B.S.R. Hi-G with bracket. I8/each. B.S.R. Hi-G with bracket. I8/each. B.S.R. Hi-G liess bracket).
IS/- each. E. V. POWER POINT
in Garrard plug-in shell, I8/6 each.
GARRARD GC2 16/- each. E. V.
CARTRIDGE only 11/6 each. All
plus 9d, P. & P.

VALVES. We have perhaps the most up-to-date valve stocks in the trade. New imported valve types fully guaranteed and P.T. paid and all the usual surplus types at apecial prices. We also carry a comprehensive stock of all B.V.A. types at current list prices. Send stamp for NEW list now available. Note: Certain American special purpose types can be supplied. Enquiries invited.

RE-GUNNED CATHODE RAY TUBES. (As new.) Guaranteed 12 months. 12in., 14in... and 15in., 65/10j-; 17in., 66; 21in., 67/19/6; plus 10/- c, and p.



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ALSO SEE PAGES 136 137

MOBILE RADIO TELEPHONES MODEL HP II

Technical Specification:

TRANSMITTER

R.F. power output; 5 watts. Frequency stability: 3 parts in 106 per degree centigrade.

Carrier noise level: 52dB below 30 per cent, depth of modulation.

Spurious omissions; 2nd harmonic attenuated 54dB, all others at 70dB.

Modulator output: 4.5 watts. Modulation capability: 90 per cent.

Modulator response: Flat to within IdB from 150-3000 c/s.

Public address: Audio output 4.5 watts. PANEL CONTROLS

Aerial socket, pllot lamp, extension speaker, selector switch, volume control.



range 75-100 Mc/s.
Complete Transmitter Receiver Assembly, and power supply, measuring only 8 in. x 8in. x 4in. Weight

14 lb.
Each set comprises one V.N.F.
Trans/Receiver, one 12v. power supply
and one hand microphone.
Original cost over £100; supplied in
used condition but in working order at

ONLY £18 PER SET

Specification (contd.):

RECEIVER

Sensitivity: Less than I microvolt. Frequency stability: 2 parts in 106 per degree centigrade.

Selectivity:—6dB at ± 25 kc/s. —30dB at ± 50 kc/s

-50dB at ± 65 kc/s off tune Signal to noise ratio: 12dB or better for lav Input, A.F. Output; I watt. Speaker Impedance: 3 ohms.

POWER PACK

Power input: 12 v.

H.T. Output: 300 volts at 105 mA. for Transmitter. 250 volts at 54 mA. for Receiver.

L.T. Output: 6.3v. at 1.6 amps.; 6.3v. at 2.6 amps.

POCKET MULTI - METER
Brand New. 2,500 o.p.v. Multi
range. 6/30/120/3009]
1,200 v. A.C., ditto
D.C. 0-1k., 0-8 megohm; 400 micro-A.,
12 m.A., 300 m.A.;
-00 to +65 db,
5 ranges, 3 x 4½.
½in. Large clear
dial, Leads supplied.
(Likt price £6/19/6).
OUR PRICE £4/7/6.
P. & P. 21/6.

OUR PRICE LATTIS.

P. B. P. 2/6.

SIGNAL GENERATOR. 100 kc/s.-100 Mc/s.
and 100 Mc/s.-200 Mc/s. Internal Mod. 400 c.p.s.
to a depth of 30%, modulated or unmodulated
R.F. output continuously variable 100 milli-volts.
C.W. and mod. switch, variable A.F. output.
Magic eye. Metal case 10 x 6½ x 5½n. A.C.200/
250 v. Brand new. Only 66/19/6. Carr. 5/-.



1.5 to 12 Mc/s. 7-valve superhet, built like a

7-valve superhet, built like a dream. 125K7-RF, 125A7 Mixer. 12A6 Oscillator, 12A6 Oscillator, 12A6 Oscillator, 12A6 Oscillator, 12A7 Detector, AVC - BFO - 1st AF, 125K7-IF.'s.

puts 1.4 watts into 500 ohms with an input modulated only 30%. Panel controls: R.F. Gain, A.F. Gain, C.W. Pitch, bandswitch, mod.-C.W, switch, power switch, ground and aerial posts, M.O. or crystal frequency switch, speaker jack, card holder to log 30 stations, hand vernier tuning knob turning a large etched calibrated plate behind hale lined window, anti-backlash gears used. Used condition. 48/10/-, carriage 15/-. T.C.S. TRANSMITTER available at 49/18/-, carriage 15/-. Note.—If both items purchased together, 417. Cerriage 25/-. D.M. 34. America's finest little dynamotor offering 12 v, in with 220 m. St. Little dynamotor of the control of the co

cogesner, £17. Cerriage 25/-.
D.M. 34. America's finest little dynamotor offering 12 v, in with 220 v, out at 80 mA. With suppression and smoothing mounting base, 5tre 4½ x 2½ x 2½n, Original packing, ONLY 35/-.
P. & P. 3/6.



Transmitter-Receiver incor-porates "A" Set -TX/RX cover-ing 2-8 Mc/s. (37.5-150

VHF TX/RX covering 230-240 Mc/s, (1.2-1.3 netres) and intercom, amplifier. Complete with 15 valves, 500 micro-amp, check and tuning meter, dreults, and instruction book (American manufacture). In used condition, 65/m, Carr. 10/m,



to X plates switched 120. and direct, requires only suitable power pack for use as oscilloscope, 70/-, carr.

ALIGNMENT ANALYSER
Type MC12
A.C. MAINS 200/250 volts. Provides:—
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W12 V. variable charge rate up to 6 amps. Consisting of Mains Trans., F.W. (Bridge) Selenium Rectifier, 0-7 amp. meter, multiposition switch with knob, fuses, fuseholders, panels, plugs and circuit. Only 59/6 Post 4/6.

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IASSEMBLED CH	ARGERS
6 v. 1 a	18/8
6 v. 2 a	29/8
6/12 v. 1 a	28/8
6/12 v. 2 m	38/8
6/12 v. 4 a	66/8
Above ready for	use with
mains and outp	
Cases well ventil	ated and
anished in stor	
hammer. Carr. &	pkg. 3/6.

CHARGER TRANSFORMERS							
200-23						•	
0-9-15	٧.	14	a.	5 Do 1		18/1	
0-9-15							
0-9-15							
0-9-15	٧.	6 a.			07774	23/0	

s according to quantity.	
BATTERY CHARGER K	ITE
Consisting of Mains Transle	
F.W. Bridge, Metal Rec	
well ventilated steel case.	
fuse-bolders, grommets, p	
and circuit. Carr. 2/9	nulea
6 v. or 12 v. 1 amp	94/6
As above, with ammeter	
6 v. 2 amps	
6 v. or 12 v. 2 amps	
6 v. or 12 v. 2 amps	42/8
(unclusive of ammeter)	
6 v. or 12 v. 4 amps	
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variable charge rate selector	
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CHARGER AMMETERS 0-1.5 amp., 0-3 amp., 0-4 amp., 0-25 amp., 0-60 amp. 8/6 0-7 amp., 0-25 amp., 0-60 amp. 8/6 Only 59/8. Carr. 3/9

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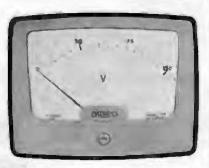
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	First Coll	Second Coll	Both coils	50 Volts

	First Coll	Second Coll	Both coils 50 Volts
Minimum F.S.D.	50 mic/A 1 Volt D.C.	100 mic/A 1 Volt D.C.	D.C. Self-contained. Tag boards and resistance boxes ex-
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Contemporary of yeal, raine covered cabined in terrotone (awn and brown, or subtled red with white puths odt. Bin 188 m modern ampifiers and antichangers, ota. Uncut recent player mounting bush 14 x 13in. supplied. Cabinet Price 23.3.0. Carr. and Inn. 3/6.

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Twin stage BCLF3 with vot. and arg. feedback. Tone controls AC. 200-250 to with double-wound finine trans. Complete with husbs. when ready sired to fit above

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Laiest Push-Pull, 4 Transistor cirruit giving
full 1 watt Output into viandard 3 ohm
speaker. Good sensitivity and improved
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ard Volume Coatron. Chamis then 61in. 3
jin. if Jin. Currot consumption 10 and
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Quality built to Mutheri's operation-

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Valve Line-up; ECCB5, ECH81, EF89, EABC80 EL84, EM81, EX80

Three Waveband and Switched Gram poritions. Med. 200-500 m., Long 1,000-2 000 m., VHP/FM 50-60 Me/s. Pashpy Constituental. Turior, insert

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Civil in recent of the company
250 v. operation. O Carr. & Ins. 5/-

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Physical Length 103 inches (40.32 cms.) Raight 23 inches (6.41 cms.) Width 23 inches (6.03 cms.) Centre of Sams to stylus (1p 12 inches (30.72 cms.). Approx. overall fartes

(30,72 cms.). Approx. overall flyfts.

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A diamond stylus is fitted to the 33 if 45 r.p.m. head supplied.

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I ohm. (measured is 1,000 c.p.s.). Frequency Response

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30 mV of secondary of transformer
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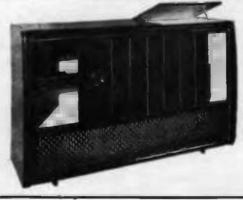
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Size I Ain, dia., Zin, spindles, PRICE 2/11 ea.

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pole 4 way,
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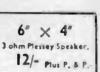
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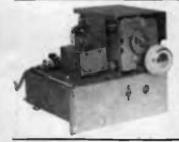
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A BARGAIN at 27/6. Plus P. & P.

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WIRE. Twin padded, grey, with maroon tracer Mains lead — usually 10d, per yd, — Our Price 25:- per 100 yd.

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Max. A.C. in 125 v. D.C. Out, 80 mA.... 4/
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CABY MULTI-RANGE TEST METER. Freshly Imported.

Guaranceed Model A-10. A.C./ Model A-10. A.C.;
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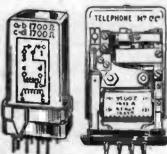
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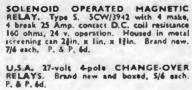
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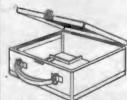
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500 ohms primary. 18 ohms secondary. P. &

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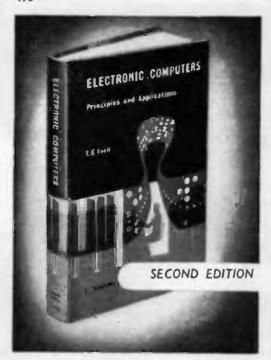
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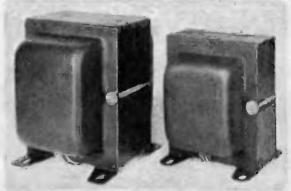
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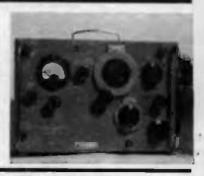
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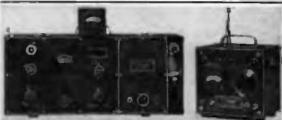
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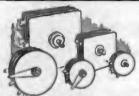
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RECTIFIERS FOR BATTERY CHARGERS

12	٧.	1	amp	4/3
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Make	F.S.D.	Res.	Scale	Size	Price
Weston (U.S.A.)	I m/a.	150 ohms	0-100		27/6
				Jain.	
Weston (U.K.)	l m/s.	75 ohms	0-100	3jin.	22/6
W.D. (U.K.)	5 m/a.	_0037 ohms	0-20	Zin.	9/-
W.D. (U.S.A.)	500 µ/a.	_	0-600	2}in.	15/6

CHARGER **TRANSFORMERS**

	primary 200/240 v.,	secon-
	V., 9 V. and 17 V.	
	type	
2 amp.	CYP4	15/4
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Acos Mic 39/1, Crystal Stick
Microphones for use as handdesk or floor stand units, for
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Milde.	Volts Size Type Price	75	12	1231	C	104.	16+ 16	180	2×11	T/6	1/-	100+200	275	1124	ă	
1	278 x 11 W/m 1/-	80	450	10×3	W'm	2/6	16+ 16	275	1 8 2	11	2/-	100 + 250	273		č	8-
2	12 13/39 x 1 M 1/4	100		VIII	M	1/4	16 - 16	380	1 × 2	C/B	20-	190 + 300	273.	11 × 41	ä	2/-
2	275 la H W 1/-	100	12	fix 11	197/8	3/-	164 16	430	1 = 2	WEAR	Ale	100 + 100	273		ă	2-
4	150 (x1) T/0 U-	100	23	× 11	T	1/-	20 + 10	450	1 × 3	C	\$ -	150 + 30	330	2 × 45	C	- W
4	150 ixli W	100	2.5	161	T/8	1/8	20 + 20	275	1×2	ľ	8/-	1304 30	300	14×6	0	8/-
	25 KI W/B I/-	100	25	La it	W	1/2	20 + 20	4.54)	1 11 3	W.	36					
6	250 x 1 WorW/B 1 2	100	270	1 2 2	C	170	20+ 20	450	1 × 3	P	3/-	TO	IDIE	S Etc		
	160 x11 T 104	100	278	1×3	Č	1/6	244 24	330	11 × 2	G	0/-	100	IFEE	2 566	-	
8	200 X W 1/-	100	275	1 × 3	19/6	9/-	25 + 25	200	1×2	F	1/6	8+8+8	330	1 = 2	P	4/-
	250 x 1 WorW 8	100	350	1 × 31	PC	3/-	30 + 30	130	12011	W//8	1/-	16+8 4	273	1×2	ľ	2.6
	275 x 1 W 1/3	150	25	1 × 1 1	12.	1/3	32 4 32	130	1 × 2	C	1/-	16+16+4	273	3 20 3	C	2.6
	350 1x2 P 1/6	150	150	1×3	141/18	1/-	32 + 32	150	1 × 2	W//B	J-	16+16+16	273	1 x 2	G	2/9
8	450 [x1] W/N 1/11	200	- 6	A M I	34	114	BB + 12	130	1 × 3	1.	10d.	20+15+15	450	1 × 3	11/8	3.6
8	750 1(× 4) C 5/6	200	1.2	1×11	18.	- T/6	22 + 22	230	1 x 21	PC	1/6	20420420	250	11×2	P	1/-
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12	25 13/38×1 30/R 1/8	250	23	E × 16	T	11-	40 + 20	130	1 x 2	1'	104	32+32+6	275	18×2	G	2.6
16	150 [x1] T/0 1/-	250	23	2 × 1	11.	1/8	40+ 40	130	1×3	1.	104.	32+32+8	234)	10×8	Q	3/-
16	275 H2 T 104.	280	An .	1.89	C	1/8	40+ 40	275	11×2	C	1/6	32 + 32 + 25	275/25	3×3	0	2/-
16	250 4×2 P 1/-	250	120	11×3	P	1.4-	40+ 40	300	1 10 24	PC/S	21-	32 + 32 + 32	350	lint	0	3-
20		400	6	1×2	F/R	84.	40 + 40	430	11 m 3	M.	3/-	32+200+50	278	18×4	a	3/-
20	12 13/32×1 M 1/4	500	- 6	[H1]	T	104.	80+ 50	130	1×2	C	1/-	32 + 300 + 70	27.3	18 × 44	Q	3/-
50	150 fw1 T 100.	500		1 × 2	C	84.	50 + 50	200	1×3	P.	1/-	40+30+20	130	1 1 2	G	1/-
50	450 Ex2 W/6 1.9 12 13/25x1 M/8 1/6	500	1.3	[m 1]	T	1/-	№0 + 50	3770	Hxx	P	1/-	40+40+12	275	11×2	G	2/6
25	12 13/22×1 M/R 1/6	500	12	9 × 1 }	11.	13	50+ 50	275	14×3	+	1/0	40+40+20	278	11×3	2	3!-
28	23 12/30 × 1 H 1/4	500	12	1×2	C	84.	50 + 50	275	14×3	C	1/6	40+40+20	300	14×9		3/-
25		500	25	1 1 2	C	16	50+ 50	275	1×3	PC	1/9	40 + 40 + 32	275	1 x 2	PO	3/-
25	50 xi) T 1/-	3000	,0	1 × 2	C	3 .	50 + 50	300	10×2	C	2 -	40+80+20	430	1 11 3	1.	8/-
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322	273 1×8 1 1/6				_		60 + 100	350	14 × 44	C	5/8	50+50+10	150	11 112	G	1/-
40	\$50 112 W/B 6d.		DO	JBLE:	5		604200	278	10 20 4	C	3/6	50+50+50	350	14 × 3	P	3/6
10	330 1x3 F 1/0						60 + 250	278	10 m 46	C	3:6	60 + 300 + 30	278	34 × 4	a	3.6
50	6 13/39 x 1 M 1/4 12 13/39 x 1 M/R 1/6	84-8	350	1×2	a	2/3	B0+300	278	15 × 4	C	36	100+40+40	130/50	11 × 41	2	4.6
50		9 + 8	450	1×2	W.	2/9	100 + 68	250	14 × 3	F	2-	100+100+50		11 × 3	1,	4/6
50	18 1×1 W 1/6	8+ 8	450	1x1		3'	100+100	18	1x2	C	1/-	100+100+20		14 × 44	C	4.5
50		8+ 16	450	1×10	W/10	3/6	100+100	25/12	1×2	P	1/-	100+250+25		2×4	C	8/-
50	50 1×11 T 1/6	10+ 10	450	1 x 2	14.14	2/6	100+100	273	18×3	C	2/8	100+400+16		14×4	Q	44-
80	275 1×3 W 1/0	12+ 12	275	1 × 2	-	1/6	100+100	300	lix3	ľ	3 -	100+400+32		11×4	Q	41-
60	250 14 x2 T/8 2/-	12+ 12	278	1×8	C	2/-	100+100	300	10 ×3	7	3/-	200+250+25		18 × 48	0	6/-
64	275 in 3 F 1/6	18+ 24	273	1 × 2	C	1/6	100+200	25	1×2	8	1/-	40+20+10+	10 350	11×2	G	3/6

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ULTRA

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Stonefield Way, South Ruislip, Middlesex.

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These posts should attract men already in the £1,000-£1,500 Salary bracket.

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SENIOR DESIGNER DRAUGHTSMEN with experience of television receivers, or similar mass-produced equipment, are required to deal with the increasing volume of mechanical design caused by the all-round expansion

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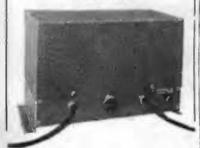
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INDEX TO ADVERTISERS

	MAR I
Acoustical Mfg. Co., Ltd. Adeola Products, Ltd. Adeola Products, Ltd. Adeola Products, Ltd. Adeola Products, Ltd. Cover Advance Components, Ltd. Alirmec, Ltd. Alirmec, Ltd. Anders Electronics, Ltd. Antiference, Ltd. Appointments Vacant 166, 167, 168, 169, 170, 171, 172, 175, 176, Arcolectric Switches, Ltd.	20 111 16 24 177 170 157 54 76 48
Arcolectric Switches, Ltd. Ardente Acoustic Laboratories, Ltd. Armstrong Wireless & Twevision Co. Ltd. Army Reserve Associated Electrical Industries Ltd. Cover ii, 2, 96, 103, 104.	88 52 183 173
Arcolectric Switches, Ltd. 175, 176. Ardente Acoustic Laboratories, Ltd Armantong Wireless & Television Co., Ltd Ltd Associated Electrical Endustries Ltd Audito Pair Audits B. B. Ltd. Austin Motor Co., Ltd. Automatic Telephone & Electric Co., Ltd. Automatic Telephone & Electric Co., Ltd. Awo, Ltd 1,	66 116 166 40 92 114
Batey, W., & Co. Belling & Lee, Ltd. Benson, W. A. Bentley Acoustic Corporation, Ltd. Berry's Radio Bradford Institute British Chas. (Radio), Ltd. British Broadcasting Corporation British Perrograph Recorder Co., Ltd. British Perrograph Recorder Co., Ltd. British Relay Wireless, Ltd. Brooks Crystais, Ltd. Broom & Wade, Ltd. Broom & Wade, Ltd. Builfin, A. P., & Co., Ltd. Builfin, A. P., & Co., Ltd. Builfin, A. P., & Co., Ltd. Builfins, Ltd.	180 116 142 116 1172 1172 172 26 171 176 80 94 151 56
Candler System Co. Catr Pastener Co. Ltd. Catrodeon Crystais, Ltd. Channel Electronic Industries, Ltd. Channel Electronic Industries, Ltd. Clyne Radio, Ltd. Cosmocord, Ltd. Costage Laboratories, Ltd. Cossor Instruments, Ltd. Coventry Radio Crawshay, P. B.	188 89 82 150 88 138 100 170 43 164
Daly (Condensers), Ltd. Davies, A. & Co. Davies, A. & Co. Davies, A. & Co. Davies, A. & Co. Davies, Dack Relays), Ltd. Davisrom, Ltd. Daystrom, Ltd. Decenser, Ltd. Dixon, L. & Co. Drayton Regulator & Instrument Co., Ltd. Dukle & Co. Duck &	62 186 94 116 . 9 177 177 154 126 118 183 73 59 149 162
E.I.R. Instruments, Ltd. E.K. Electronics, Ltd. E.K. Electronics, Ltd. Electric Acoustic Developments Electric Acoustic Industries. Ltd. Electro-Methods, Ltd. Electro-Methods, Ltd. Electronics (Punsburger Ltd. Electronics (Punsburger Ltd. Electronics (Punsburger Ltd. Electronics (Punsburger Ltd. Electronics (Ruislip), Ltd. Electronics (Ruislip), Ltd. Electronic Tubes, Ltd. Elliott Bros. (London), Ltd. E.M.J. Electronics, Ltd. Empire Rubber Co., Ltd. English Electronics, Ltd. Eta Tool Co., Ltd.	100 112 1184 178 119 110 118 113 113 113 113 113 113 113 113 113
	86 173 92 68 62

Carrard Eng. & Mfg. Co. Ltd., The 41, 70	Premier Radio Co. 153
General Electric Co., Ltd. 49, 70, 77, 108, 187 Gilson, R. F., Ltd.	Premier Radio Co
Glaser, L., & Co., Ltd. 189 Gladatone Radio	
Carrard Eng. & Mfg. Co. Ltd The 41, 70	Quarta Crystal Co., Ltd. granter and 184
Oramdeck	Radford Electronics. Ltd
Grey & Marten, Ltd. 68 Griffiths Hansen (Recordings), Ltd. 186, 189	Radford Electronics, Ltd. 94 Radio & Electrical Mart 164 Radio & T.V. Components (Acton), Ltd. 160 Radio Clearanca Ltd. 161 Radio Component Specialists 121 Radio Exchange Co. The 185 Radio Exchange Co. The 185 Radio Malling, Ltd. 165 Radio Reastor, Ltd. 52 Radio Reastor, Ltd. 52 Radio Supply Co. (Leeds), Ltd. 140, 141, 142 RadioSupply Co. (Leeds), Ltd. 140, 141, 142 RadioStructor 106
	Radio Component Specialists
Hall Electric, Ltd. 53 Harmsworth, Townley & Co. 117	Radio Experimental Products Co
Harring Supplies, Ltd. 187 Harris Electronics (London), Ltd. 181	Radio Resistor, Ltd. 52 Radiospares, Ltd. 84
Harverson Surplus Co., Ltd. 144, 145 Henry's (Radio), Ltd. 128, 129	Radio Supply Co. (Leeds), Ltd 140, 141, 142 Radiostructor
Hivac, Ltd	Rank Cintel, Ltd. 5
Hall Electric, Ltd	Radio Supply Co. (Levda). Ltd. 140, 141, 142 Radio Structor Radio Traders, Ltd. 150 Radio Traders, Ltd. 150 Radio Traders, Ltd. 150 Record Housing 64 Record Housing 64 Redifico. Ltd. 85 Redida Radio, Ltd. 139 R.F. Development Co. 189 Robitson, D. & Sona 120 Robitson, D. & Sona 120 Robitson P. & Sona 120 Robitson P. & Robitson, Ltd. 17 Rollet, H. & Co. Ltd. 189 Royal Air Force 167, 175
Hycon-Page 168	R.F. Development Co. 189 Robinson, D., & Sona 120
LB.M., United Kingdom, Ltd. 169	Rola-Celestion, Ltd. 17
Imperial Chemical Industries, Ltd 174, 176 Independent Television Authority 170	Rollet, H., & Co., Ltd
L.B.M., United Kingdom, Ltd. 122, 152, 154 Hiffe & Sona, Ltd. 124, 122, 152, 154 Imperial Chemical Industries, Ltd. 174, 176 Independent Television Authority	Cultural Charles Indiana Tad
International Computers & Tabulators, Ltd. 172 173 175 177 178	Salford Electrical Instrumenta, Ltd. 71 Samsons Surplus Stores 127 Savage Transformers, Ltd. 134 Semiconductors, Ltd. 34 Service Trading Co. 146, 147 Sifam Electrical Instruments, Ltd. 189 Simmonds, L. Z. Ltd. 90 Smith, A. K. & L. O. Ltd. 189 Smith, G. W. (Ra,30), Ltd. 124, 125 Smith, M. L. & Co., Ltd. 122, 135 Smith, M. L. & Co., Ltd. 174 South Supplies (Electrical), Ltd. 162 Sound Sales, Ltd. 186 Southern Radio Supply, Ltd. 164 Southern Technical Supplies 152 Specialist Switches 82 Spencer-West, Ltd. 185 Stamford, A. L. 185 Stamdard Telephones & Cabies, Ltd. 44, 91, 99 Steatite insulstions, Ltd. 133, 134, 135 Stone, J. L. Ltd. 175, 177 Sulfer, Ltd. 79 Sugden, & R., & Co. (Engineers), Ltd. 78
tiongere (at.O.) Co	Semiconductors, Ltd. 34 Service Trading Co. 146, 147
Kenure, Rolt Electronics, Ltd	Sifam Electrical Instruments, Ltd 169 Simmonds, L. E., Ltd
Kolectric, Ltd 60	Smith, A. K. & L. O., Ltd
Lasky's Radio, Ltd	Smith, M. L., & Co., Itd 120, 154 Smith, S. & Sons, Ltd
Ledon Instruments	South Supplies (Electrical), Ltd
Levell Electronics	Southern Technical Supplies
Light Soldering Developments, Ltd 184 Linear Products, Ltd	Spencer-West, Ltd
Lockwood & Co. (Woodworkers), Ltd. 188	Standard Telephones & Cables, Ltd. 44, 91, 99 Steatite insulations, Ltd
Lyona Radio, Ltd. 188	Stone, J. L. P., Ltd. 176, 177
McMurdo Instrument Co. Lid. 68. 115 Malvyn Engineering Works 189 Marconi Instruments, 5td. 83, 166 Marconi's Wreeks Telegraph Co. Ltd. Marriott, P. A., & Co. Ltd. 102, 172, 176 Marriott, P. A., & Co. Ltd. 104, 170 Mechanical Handling Exhibition 95 Midland Instrument Co. 178 Mills, W. 126 Mills, W. 126 Ministry of Avistion 176 Minnesota Mining & Mfg. Co. Ltd. 31 Modern Book Co. 180 Modern Electrics (Retail), Ltd. 20 M. C. Valve Co. Ltd. 170 M. R. Supplies, Ltd. 20 M. S. S. Recording Co. Ltd. 20 Millard, Ltd. 3, 14, 15, 51, 57, 38, 174 Multitore Solders, Ltd. 182, Cover 19 Multitore Solders, Ltd. 176	Sugden, A. R., & Co. (Engineers). Ltd., 78
Malvyn Engineering Works 189 Marconi Instrumenta, btd. 83, 166	Tannoy, Ltd. 189 Technical Trading Co. 120 Telecraft, Ltd. 19 Telemechanics, Let. 76 Telemechanics, Let. 72 Teletron Co., The 188 Thompson, A. 188 Thompson, A. 188 Try Trading Co., Ltd. 180 Trix Electrical Co., Ltd. Edit, 149 T.R.B. Radio 143 Truvox, Ltd. 50
Marriott P A & Co Ltd. 102, 172, 176	Technical Trading Co
Mechanical Handling Exhibition 95 Midland Instrument Co. 178	Teleng, Ltd. 60, 174
Milia, W	Tele-Radio (1943), Ltd. 72
Minnesota Mining & Mfg. Co., Ltd. 31 Modern Book Co. 183	Thompson, A. J
Modern Techniques BO	Trix Electrical Co., Ltd
M. R. Supplies, Ltd. 80 M. S. Racording Co., Ltd. 22	Truvox, Ltd 50
Mullard, Ltd 3, 14, 15, 51, 57, 98, 174 Multicore Solders, Ltd 182, Cover iv	U.K.A.E.A 175, 176
Multitone Electric Co., Ltd	U.K.A.F.A. 175, 176 Ultra Electric, Ltd. 168 Uncles, Blias & Co., Ltd. 86 Unicam Instruments, Ltd. 175 United Components Co. 168 Universal Book Co. 168
	United Components Co. 168
Nash & Thomson, Ltd. 10, 11 National Cash Register Co., Ltd. 175 Newmarket Transistor Co., Ltd. 38 Newnes, George, Ltd. 32a, 32a	Different and on the section of the
Newnes, George, Ltd 32a, 32s	Vacwell Engineering Co., atd,
Neo Mall Order 189 Northern Electric Co.—Cariada 166, 174	Ventier Electronics, Ltd
Northern Radio Services 115	Vortexion, Ltd
Oddie Bradbury & Cull, Ltd. at 186	
	Walmore Electronics, Ltd. 82 Walts, Cecil E 180 Webber, R. A., Ltd. 180 Winston L. (Croydon), Ltd. 180 Winston Electronics, Ltd. 33 Winster Trading Co., Ltd. 33 Winter Trading Co., Ltd. 36 Wolf Electric Tools Ltd. 56
Pembridge College of Electronics	Webber, R. A., Ltd. 118 Webb's Radio
Parker, A. B. 112	Weymouth Radio Mig Co., Ltd., The 72 Wharfedale Wireless Works
P.C. Radio 135 Philoo International, Ltd. 18	Wilkinson, L. (Croydon), Ltd. 158 Winston Electronics, Ltd. 33 Winter Trading Co., Ltd. 30 Wolf Electric Tools, Ltd. 66
Pitman, Sir Isaac, & Son, Ltd. 180	Wolf Electric Tools, Ltd
Painton & Co., Ltd. 42 Perinbridge College of Electronics 171 Pamphonic Reproducers, Ltd. 84 Parker A. B. 112 Parker A. B. 181 P. C. Radio 181 P. C. Radio 181 Phillop International, Ltd. 18 Phillips. N. V. 16 Pleaser Co. Ltd. 16 Pleaser Co. Ltd. 18 Portland Electronica 30 64 Post Radio Supplies 186	Wolf Electric Tools, Ltd

Premier Radio Co. Proops Bros. Ltd. Pye. Ltd. Pye Telecommunications, Ltd	PACE 153 156 172 39
Quarts Crystal Co., Ltd. appropriate	184
Radford Electronics. Ltd. Radio & Electrical Mart Radio & T.V. Components (Acton), Lld. Radio Clearance. Ltd. Radio Component Specialists Radio Exchange Co. The Radio Exchange Co. The Radio Exchange Co. The Radio Experimental Products Co. Radio Mailing. Ltd. Radio Supply Co. (Leeds), Ltd. 140, 131, Radio Supply Co. (Leeds), Ltd. 140, 131, Radio Trains, Ltd. Radio Reastor, Ltd. Radio Reastor, Ltd. Radio Radio Ltd. Radio Ltd. Radio Ltd. Record Rousing Rediffon Ltd. Referent Radio, Ltd. R.F. Development Co. Robinson, D., & Sona Robinson, D., & Sona Roisrenwerke Rois-Celestion, Ltd. Rojel, H., & Co., Ltd. Roys Air Force 167.	94 164 160 161 185 188 185 188 185 199 199 189 189 189 189
Salford Electrical Instruments, Ltd. Samsons Surplus Stores Savage Transformers, Ltd. Semiconductors, Ltd. Semiconductors, Ltd. Service Trading Co. Silam Electrical Instruments, Ltd. Silam Electrical Instruments, Ltd. Silam Electrical Instruments, Ltd. Silam Co. W. (Ra,200), Ltd. Smith, A. K. & L. O., Ltd. Smith, E. & Sons, Ltd. South Supplies (Electrical), Ltd. South Supplies (Electrical), Ltd. South Supplies (Electrical), Ltd. South Supplies Supplies Specialist Switches Specialist Switches Specialist Switches Specialist Switches Stamford, A. L. Standard Telephones & Cables, Ltd. 44, 91 Steatite Insulstions, Ltd. Stenes, J. L. P., Ltd. Sugden, A. R., & Co. (Engineers), Ltd. Sugden, A. R., & Co. (Engineers), Ltd.	71 127 184 147 169 90 189 125 154 174 162 186 186 188
Standard Telephones & Cables, Ltd. 44, 91 Steatite insulations, Ltd. Stern Radio, Ltd. 133, 134, Stone, J. L. P., Ltd. 176, Suffex, Ltd. Sugden, A. R., & Co. (Engineers). Ltd.	135 177 79 78
Standard Telephones & Cables, Ltd. 44, 91 Steathte insulations, Ltd. Stern Radio, Ltd. 133, 134, Stone, J. L. P., Ltd. 135, 134, Stone, J. L. P., Ltd. 176, Sulfer, Ltd. Sugden, & R., & Co. (Engineers). Ltd. Tannoy, Ltd. Technical Trading Co. Telecraft, Ltd. Telemertanics, Ltd. 60, Telecraft, Ltd. Telemertanics, Ltd. 60, Telecraft, Ctd. 60, Telecraft, Ltd. 76, Teleguipment, Ltd. 76, Teleg	189 120 174 120 120 174 174 188 188 188 180 149 143 50
Standard Telephones & Cables, Ltd. 44, 91 Steathte insulations, Ltd. Stern Radio, Ltd. 133, 134, Stone, J. L. P., Ltd. 135, 134, Stone, J. L. P., Ltd. 176, Sulfex, Ltd. 176, Sulfex, Ltd. 176, Sulfex, Ltd. 176, Sulfex, Ltd. 176, Technical Trading Co. 176, Technical Trading Co. 176, Telegraphical Ltd. 176, Trix Electrical Co., Ltd. 177, Trix Siectrical Co., Ltd. 177, U.K.A.F.A. 175, Ultra Electric, Ltd. 175, United Components, Ltd. 175, United Components Co. Universal Book Co. 177, Ultra Components Co. 177, Universal Book Co. 177, Ultra Components Co. 177, Ultra Co. 177, Ultra Components Co. 177, Ultra Compon	189 120 13 78 174 29 72 188 188 180 149 145 50
Tannoy, Ltd. Technical Trading Co. Telegraft, Ltd. Telemechanics, Ltd. Teleng, Ltd. Teleng, Ltd. Teleng, Ltd. Telestron Co. The Thompson, A. Toyo Trading Co., Ltd. Trix Electrical Co., Ltd. T.R.S. Radio Truvox, Ltd.	189 129 178 174 29 72 188 188 189 143 50 168 86 86 875 168 186
Tannoy, Ltd. Technical Trading Co. Telecraft, Ltd. Telemechanics, Ltd. Telemeghanics, Ltd. Telemeghanics, Ltd. Telemeghanics, Ltd. Televelipment, Ltd. Televelipment, Ltd. Televelipment, Ltd. Televelipment, Ltd. Televelipment, Ltd. Televelipment, Ltd. Trix Electrical Co., Ltd. Unitra Electrical Co., Ltd. Unitra Electrical Co., Ltd. Unitra Electrical Co. Universal Book Co. Vacwell Engineering Co., Ltd. Venner Electronics, Ltd. Vitality Bulbs, Ltd. V.Z. Electrical Service Walmore Electronics, Ltd. Waltis, Cecil E. Webber, R. A., Ltd.	189 120 19 78 124 129 722 188 188 180 145 50 156 86 175 86 86 86 175 86 86 86 175 86 86 86 175 86 86 86 175 86 86 86 175 86 86 86 175 86 86 86 86 86 86 86 86 86 86 86 86 86



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