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103 Editorial Comment
104 Waveguide Transmission
106 World of Wireless
109 European Television Stations
117 Time Past-Beam and Broadcast By P. P. Eckersley
121 Letters to the Editor
126 Physical Society's Exhibition
135 Relativity-2
139 Evaluating Aerial Performance-2
144 Short-wave Conditions
145 The Bifilar-T Circuit-2
149 News from the Industry
150 March Meetings
152 Random Radiations
154 Unbiased

By " Cathode Ray"
By L. S. Moxan

By Thomas Roddam

By " Diallist"
By "Free Grid"

[^0]
## Transistors

## A Simple Temperature-control System

This system involves the use of transistors to keep a block of copper at constant temperature and thus to provide thermostatic conditions for the transistors of any experimental or industrial circuit. The collector leakage current of a transistor varies with ambient temperature. The drift which thus arises affects the stability of d.c. amplifiers and test equipment.


COPPER BLOCK (approx. full size)
A copper block was constructed to the design shown in the drawing above. The block is made circular for ease of machining and for uniform heating effect. A layer of aluminium foil covering the heater coil reflects heat into the block.

A closed-loop servoméchanism controls a power transistor, which supplies the current for heating a coil wound uniformly around the block. The sensing element, used to provide the error signal, is an a.f. transistor mounted in the block. The leakage current $I^{\prime}{ }_{c o}$ of this transistor changes markedly with temperature, so that suitably amplified changes in I' ${ }_{c o s}$ are used to control the current through the heating coil. It was found that an initial current through the coil of 1.5A gave a good heating rate without much overshoot. For a 12 V supply the resistance of the coil becomes $8 \Omega$. To provide sufficient length of wire to wind uniformly, five strands of thin enamelled constantan wire are wound in parallel.
The control circuit is shown here. The connection of Tr 3 and Tr 4 allows the OC 72 to drive the OCl 6 without danger of excessive dissipation. The gain of the combination is then the product $\alpha_{1}^{\prime} . \alpha_{2}^{\prime}$. The combined base-emitter voltage provides sufficient voltage for the collectors of Tr 1 and Tr 2 ; at
the same time it limits the maximum possible dissipation of these transistors. The variable resistance provides a path for some of the leakage current, allowing adjustment of the temperature to which the block is set.

When the circuit is switched on there is very little leakage current flowing in the sensing transistor. Accordingly the OC16 output transistor 'bottoms' and the current flowing in the heating coil is very nearly $\mathrm{V}_{\mathrm{cc}} / \mathrm{R}_{\mathrm{H}}$ ( $\mathrm{V}_{\mathrm{cc}}$ is the supply voltage and $\mathbf{R}_{\mathbf{H}}$ the heating coil resistance). The block heats rapidly to nearly $40^{\circ} \mathrm{C}$, at which temperature the leakage current of Tr 1 rises rapidly and the OCl 6 is cut-off, and remains cut-off till the block temperature has dropped to the set vaiue. There is a slight overshoot of temperature because the junction temperature of Tr 1 does not react to changes instantaneously.

Trl is the temperature-sensing transistor. Its collector voltage is limited to about 1 V , whilst the working voltage is about 0.6 V . To limit the maximum change of temperature inside the block to $0 \cdot 1^{\circ} \mathrm{C}$, the maximum variation of $\mathrm{I}^{\prime}$ o must be $0.012 \%$. The actual stability achieved over at least 24 hours satisfies this condition.

A long-tailed-pair amplifier using germanium transistors was tested in the block. A variation of ambient temperature from 20 to $35^{\circ} \mathrm{C}$ resulted in a change of $0.75^{\circ} \mathrm{C}$ in the block. The drift, referred to the input, was about $75 \mu \mathrm{~V}$, or $5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$.

[5518]

CONTROL CIRCUIT

Vol. 65 No. 3

## Growing Pains

FROM January 1st this year the Electronic Engineering Association has ceased to be a member of the Radio Industry Council. This is not the result of any hasty decision, but of a level-headed and objective study of the divergence of the predominant aims and interests of these two great bodies which has become increasingly apparent in recent years.

In announcing the change Mr. G. Darnley Smith, chairman of the R.I.C., said,
"Electronics have become of such enormous importance in the last few years, playing an essential part in many major industries and having uses in all of them, that we-meaning all the Associationshave had to recognize that ours is now a two-fold industry. One side is dealing largely with broadcasting techniques and equipment and the other with electronics for all other purposes. To keep pace with the rapid technical, industrial and commercial advances on the capital goods side of the industry, it has been agreed that the E.E.A. should pursue its own policies and objectives. Close co-operation, however, will continue with the E.E.A. on matters affecting the welfare of both sections of the industry, particularly in the technical field, by means of interassociation committees."

A parallel statement by the Electronic Engineering Association in its annual report reads as follows,
"The Association remained a constituent of the Radio Industry Council during the period under review, during which time the four constituents reached the unanimous opinion that the capital goods side of the industry had developed in size and scope to the point at which capital goods and consumer goods could be recognised with advantage to all concerned as two distinct industries, each with its own objectives and policies. As from the 1st January, 1959, therefore, the Association has given up its right to nominate representatives to the Radio Industry Council, thereby permitting the three remaining nominating bodies, the British Radio Equipment Manufacturers' Association, the British Radio Valve Manufacturers' Association and the Radio \& Electronic Component Manufacturers' Federation to devote their attention to the affairs of the domestic broadcast entertainment industry, thus enabling the Association to take full responsibility for the capital goods interests of the Industry. It will, of course, continue its close co-operation with the components and valve industries, with the domestic equipment industry, and also, on a technical level, with the other industries associated with it in the Electronic Forum for Industry."

Commenting on this statement, Mr. F. S. Mockford, chairman of the E.E.A., said,
" Two distinct and vast irdustries have grown up side by side-one, the domestic entertainment industry supplying millions of sound and vision receivers, radio-gramophones and audio equipment to the general public; the other supplying millions of
pounds' worth of capital equipment for home and overseas, much of which is telecommunication, radar and navigational aid radio equipment, but a great deal of which is closed-circuit equipment, such as industrial television, computers, machine controls, and so on.
"Each industry has many more problems than in the past, but their common problems are fewer. The performance, characteristics and siting of broadcasting transmitters and their frequency allocations is still a matter of concern to both, but the E.E.A. to-day has as much or more need for association with other industries-for example, telecommunications, office appliances, machine-tools, aircraft, transport, fuel and power.
"The capital equipment makers, of course, will continue to require and are assured of the closest co-operation of the B.V.A. and the R.E.C.M.F., and joint committees are being formed."

If there are any twinges of regret at the disturbance of an order which has remained unchanged for so long they should be dismissed in the certain knowledge that they are but the growing pains of a strong and healthy industry. The processes of association and dissociation are as fundamental to organizations as they are to living organisms, and the radio industry is no exception. First there was the association of six firms to form the British Broadcasting Company, making both the transmitters and the receivers. Then the National Association of Radio Manufacturers, later to be joined by the traders in the N.A.R.M.A.T. Next the Radio Manufacturers' Association was formed and the traders split off into various groups of wholesalers and retailers. Inside the R.M.A. the set makers and component manufacturers formed sections which developed into the Radio Component Manufacturers' Federation and the British Radio Equipment Manufacturers' Association. These autonomous bodies together with the British Radio Valve Manufacturers' Association and the Radio Communication and Electronic Engineering Association (now E.E.A.), formed the Radio Industry Council to represent their interests in negotiations with Government departments, and in fostering the post-war development of broadcasting and electrenics.

The pattern of division and growth is apparent throughout. As the scope of industry widens the "domain boundaries" indicating further potential splits will become more readily discernible in the various associations as at present constituted. What must not be lost in the process of growth is the readymade machinery for rapid consultation, and the stabilizing influence of independent minds devoted to the task of seeing whole woods rather than trees. The R.I.C., under its newly-appointed director Sir Raymund Hart, and the E.E.A. can together provide the solid foundation upon which further specialization of interest will proceed with safety.

# Waveguide Transmission 


#### Abstract

Topics discussed at the Convention on "Long Distance Transmission by Waveguide" held at the Institute of Electrical Engineers in London on 29th and 30th of fanuary.


N radio transmissions through free space even when the signal is beamed the power losses to the outside are considerable; conversely, interference from the outside is likely. Such losses and consequent possibilities of interference can be eliminated or very much reduced by guiding the signal. One type of guide which is often employed for longdistance transmissions is the coaxial cable, but this is limited by attenuation and phase distortion to use with frequencies below about $25 \mathrm{Mc} / \mathrm{s}$. At higher frequencies low enough attenuations of the order of 10 dB per mile or less are offered either by using a single-wire conductor supporting a surface wave, or by the $\mathrm{H}_{01}$ ( $\mathrm{TE}_{01}$ ) mode in circular waveguide.

## Effects of Bends and Irregularities

Although the use of the $\mathrm{H}_{01}$ mode in a circular waveguide was suggested as long ago as 1938 by G. C. Southworth, this system has apparently not yet been used commercially. The attenuation decreases with increasing guide diameter and decreasing wavelength, but low enough attenuations can only be obtained using wavelengths as short as about 6 mm in the mode giving lowest attenuation ( $\mathrm{H}_{01}$ ) simultaneously with a guide diameter of about 2 in several times the cut-off value. In these conditions about 100 other modes are capable of being propagated, and, moreover, the $\mathrm{H}_{01}$ mode is not dominant in circular waveguide. Thus conversion of the $\mathrm{H}_{01}$ to other modes occurs readily at irregularities or bends in the guide and this leads to greatly increased losses. Reconversion is also likely to occur and, since all modes but the $\mathbf{E}_{11}$ have a different velocity from the $\mathrm{H}_{01}$ mode, such reconversion results in distortion. To minimize the effects of irregularities guide manufacturing tolerances of the order of thousandths of an inch or less are necessary. These tolerances are most severe for deviations of the axis from a straight line, according to a paper given by H. E. Rowe and W. D. Waters.
$\mathrm{H}_{01}$ mode conversion effects can also be reduced and the bends which may be required for geographical reasons allowed by increasing the losses for the undesired modes. Bearing in mind the differing field patterns in the various modes (in the $H_{01}$ mode the electric field is circumferential), the losses for the undesired modes can be increased by changing the guide wall surface impedances in different directions, in particular by making the longitudinal impedance much greater than the circumferential by circumferentially corrugating the guide wall. Considerable attention was given at the Convention to the variety of corrugated guide made by helically winding enamelled or otherwise spaced wire and coating the outside of the helix
with an absorbing material to a thickness of about one-fourtieth of an inch. With such helical waveguide the attenuation of undesired modes can be multiplied by about 1,000 or more times their value in ordinary circular guide. Unfortunately the requirements for the absorbing layer to give maximum suppression of unwanted modes and at the same time minimum $\mathrm{H}_{0}$ mode loss in intentional bends are to some extent contradictory, according to a paper by H. G. Unger. Corrugated waveguide formed from flat rings a few thousandths of an inch thick spaced a few hundredths of an inch apart in air was discussed in a paper by A. W. Gent. Besides inhibiting $\mathbf{H}_{01}$ mode conversion, such waveguide offers a reduced $\mathrm{H}_{01}$ mode attenuation of about one-tenth of its value in circular guide with a diameter equal to the ring outside diameter. Another method of increasing the losses in undesired modes is to coat the inside wall of an ordinary circular guide with a layer of low-loss dielectric a few thousandths of an inch thick.
$\mathrm{H}_{01}$ mode conversion at bends can be reduced by equalizing the path lengths at various points across the guide by filling it with a suitably inhomogeneous low-loss dielectric. Automatic methods of producing the required inhomogeneity as the guide is bent by filling the initially straight guide either with dielectric discs which become appropriately spaced or, alternatively, with dielectric structures containing air cells which become appropriately deformed were described in papers by Professor H. E. M. Barlow and D. G. Rickard and P. Marié respectively.

An ingenious method of reducing mode conversion by making the $\mathrm{H}_{01}$ mode effectively dominant by using an anisotropic guide wall to decrease the wavelengths of the other initially longer wavelength modes, either directly or after conversion to other modes, was described in a paper by Professor H. E. M. Barlow. The other modes can be eliminated in this case by operating closer to the $\mathrm{H}_{0,1}$ mode cut-off, but since this considerably increases the $\mathrm{H}_{01}$ mode loss, this method will only be usable with shost lengths of guide.

Even after mode conversion effects have been minimized, residual distortion will probably impose the use of pulse modulation so that signals can be exactly reconstituted at intervals and distortions cannot accumulate.

## $\mathrm{H}_{01}$ Mode Launching

Launching the $\mathrm{H}_{01}$ mode is less difficult than bend negotiation, although to secure a sufficient bandwidth and to avoid generating unwanted modes the launching section must be longer than usualup to about 30 wavelengths. However, at the short wavelengths used this length can easily be accommodated. The signal is generally first produced as the dominant $\mathrm{H}_{0}$, mode in rectangular waveguide, and this can be transformed over a broad frequency band into the $\mathrm{H}_{0}$, mode in circular waveguide by two basic methods. Either the waveguide cross-sectional shape can be gradually changed from rectangular to circular in such a way as to change
the field from the rectangular to the circular $\mathrm{H}_{\mathrm{n}}$ mode, or alternatively the rectangular and circular waveguides can be directly coupled through holes aiong a common wall. Direct coupling is usually through a common external wall, although a new type of coupler in which the rectangular guide is centrally placed inside the circular was described in a paper by B. Oguchi and K. Yamaguchi. One way of gradually altering the waveguide cross-section is to decrease the width of one narrow waveguide wall and at the same time to increase the width of the other narrow wall until the crosssection becomes a sector of a circle. The electric field in the $H_{01}$ mode in rectanglar guide from one broad wall to the opposite is thus converted into arcs from one boundary radius of the sector to the other. These arcs are gradually converted to the circular electric field lines of the $\mathrm{H}_{0}$, mode in circular guide by increasing the arc cross-section angle until a circle is formed. Narrow-band resonantslot methods of exciting the $\mathrm{H}_{01}$ mode can also be used.
Measurements of the large number of unwanted modes and of the low losses in such guides present a number of problems. The obvious way to measure such a small loss is, of course, to increase it by repeatedly reflecting the input signal to and fro in a comparatively short sample length of guide. Unfortunately, since in this method any guide irregularities give rise to periodically repeated effects, large spurious attenuations may be produced, as was pointed out in a paper by A. E. Karbowiak. Another standard way to measure the loss is from the Q of a short length of guide short-circuited at both ends to form a resonator. Since this Q will be of the order of $10^{6}$, relatively small frequency differences must be measured to obtain the width of the resonance. Such differences can be measured using a waveguide many wavelengths long and short-circuited at its far end, as described in a paper by D. G. Keith-Walker. In this case the small wavelength changes in the standing-wave pattern will add up so as to give a large phase change with changing frequency at the waveguide input. Frequency stability problems are avoided by sweeping the source frequency.

## Mode Identification and Measurement

Modes are usually identified and measured either directly from the actual electric and magnetic field pattern, or alternatively from their group velocity obtained from the transmission time through a known length of guide. The transmission time can be measured directly using short pulses of about $10 \mathrm{~m} \mu \mathrm{sec}$ duration. An alternative method of measuring the transmission time which was described in a paper by H. G. Effemey is to beat a sawtooth f.m. signal with the same signal delayed in transmission, when the principal beat frequency produced will be proportional to the transmission time.

Electric and magnetic fields at the guide walls can be measured by coupling them out through a small hole (containing a probe for electric field measurement) into a section of $\mathrm{H}_{01}$ mode rectangular guide with a crystal at its end appropriately oriented to detect the various field components. To measure the field at various points the hole is made in a sleeve which slides between and overlaps into two spaced sections of waveguide. The usual longi-
tudinal slot for field measurement would seriously disturb the propagation of many of the modes. Internal fields are more difficult to measure because of the disturbance which would be produced by the connection required to a normal conducting probe. Two methods have been developed which do not need such a link. One of these methods depends on measuring the change in the Q of a cavity as a small piece of metallic, dielectric or ferromagnetic material suspended by a nylon thread is moved about inside the cavity. Another method which was described in a paper by Professor H. E. M. Barlow uses a similarly suspended small dipole rotating about an axis through its centre and perpendicular to its length. Some of the electric field incident on this dipole is scattered, and this scattered signal can be readily identified, since it is modulated at twice the spinning frequency of the dipole. By measuring the phase and amplitude of the scattered signal the electric field at the dipole can be obtained. Very similar results were obtained using somewhat different lengths of dipole, thus showing that no serious perturbation of the field is produced by the dipole.

## Single-wire Surface-wave System

Less attention was devoted at the Convention to the alternative single-wire surface-wave system. With such a surface wave the spread of the field beyond the surface of the conductor and consequently interference from outside increases with decreasing frequency, whereas the attenuation increases with increasing frequency. To compromise between these two effects an operating frequency of about $200 \mathrm{Mc} / \mathrm{s}$ is generally adopted. The roughly $100 \mathrm{Mc} / \mathrm{s}$ bandwidth of this system is much less than the potential one or more $\mathrm{kMc} / \mathrm{s}$ bandwidth of the circular waveguide system. Moreover, the potential attenuation using circular waveguide is also somewhat less than that of a surface wave on a single wire. However, the single-wire system has already becen utilized commercially as a television link over a 14 -mile distance in America.

Bends are also a problem in the surface-wave system. However, since only one mode can exist on the surface, mode conversion and consequent distortion cannot occur at bends. Instead radiation takes place which increases the losses and makes interference from outside more likely. However, radiation at bends can easily be minimized by confining the wave more closely to the surface of the conductor (which generally has a diameter of about half an inch) either by corrugating this surface, or alternatively by covering it with a polythene sleeve of roughly one inch outer diameter.
Surface waves can be launched over a broad frequency band on a continuation of the inner conductor of a coaxial line input by enlarging the outer conductor to form a conical horn. Narrow-band launching using a radiating circular hole concentric with the conductor is also possible.

The conductor can ideally be supported by nylon cords from the poles of an existing telephone system. The telephone wires do not have any effect provided they are half a wavelength or more away, and even closer spacing only increases the transmission loss without producing any distortion, according to a paper by G. Goubau. Even severe weather conditions have little effect on such a line.

# WORLID OF WIRELESS 

## Radio Show Organization

THE announcement, discussed in greater detail in this month's leader, that the Radio Industry Council and the Electronic Engineering Association will in future pursue separate objectives and policies has been followed by the news that B.R.E.M.A. will now undertake responsibility for radio and television shows which are primarily of broadcast techniques and equipment. The component and valve manufacturers will continue to support the National Radio Exhibition and to share in its organization.

## Record Radio Exports

PROVISIONAL figures for last year's exports of radio and electronic equipment, issued by the Radio Industry Council, show an increase of over $£ 1.5 \mathrm{M}$ on the previous year. The provisional total is $£ 45 \mathrm{M}$ -nearly four times the 1948 figure.
As can be seen from the table the highest proportionate rise was for valves and cathode-ray tubes. Last year's exports of sound reproducing equipment were more than double the value for 1955.
Exports of components, which had reached a peak value in 1957, were slightly lower in value although maintaining their volume. Their growth over recent years, together with that for valves, reflects the fact that several British manufacturers are assembling receivers in the consumer countries. This, in turn, affects the figures for the direct exports of receivers. Last year's figure does, however, include $£ 1.1 \mathrm{M}$ for television receivers which reached a significant level for the first time.
A reduction in the export of capital goods (transmitters, navigational aids, industrial electronics, etc.) is partiy due to the continued fall in Defence orders.

| Capital goods | $\ldots$ | $\ldots$ | $\begin{gathered} 1958 \\ \mathrm{f} \mid 5.45 M \end{gathered}$ | $\begin{gathered} 1957 \\ C 16.04 M^{*} \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: |
| Sound and television receivers | $\ldots$ | $\ldots$ | 3.54 | 3.56 |
| Sound reproducing equipment | $\ldots$ | ... | 11.28 | 9.86 |
| Components .. | $\cdots$ | $\cdots$ | 9.56 | 10.10* |
| Valves and tubes .. | $\cdots$ | $\ldots$ | 5.30 | 3.90 |
|  |  |  | 645.13M | 643.46M |

* Includes items not covered in provisional 1958 totals.


## National Scientific Libraries

THE proposed National Lending. Library for Science and Technology-the nucleus of which already exists in the D.S.I.R. Lending Library Unit at Chester Terrace, Regents Park, London-will be housed at Thorp Arch, near Boston Spa, Yorks. It will take over the responsibility for the lending service now provided by the Science Museum Library, which will then concentrate on serving the enlarged Imperial College of Science and Technology. The second scientific library under discussion is the National Reference Library of Science and Invention, which it is proposed to establish in London as the successor to the Patent Office Library.

## New Coast Station

ILFRACOMBE RADIO, the new Post Office Coast Radio Station at Mulacott Cross, North Devon, was opened on January 29th by T. A. Davies, O.B.E., Inspector of Wireless Telegraphy. For the past three years the short-range radio-telephone service for the Bristol Channel area has been operated from a temporary station at Iffracombe Head Post Office, and this, as well as the short-range radio-telegraph service previously given through Burnham Radio, will be taken over by Ilfracombe Radio.

This year is the jubilee of the Post Office shipshore radio for on September 29th, 1909, the Post Office took over services which had previously been operated by Marconi's and Lloyds.

There are 12 stations in the Post Office maritime service. The largest is Burnham Radio and it serves ships at sea in any part of the world. The remaining 11 provide communication up to about 300 miles.


#### Abstract

"What's in a Name? "-The word " radio" no longer appears in the title of what was, at one time, the Radio Section of the I.E.E. because "it is felt that this word, with its modern connotation of certain limited applications, is now insufficiently comprehensive." Some years ago the word "telecommunication" was added to the title, but this, too, has now been omitted and the word "communications" (to be "interpreted broadly within the context of electrical engineering") has been added. Announcing the new title-Electronics and Communications Section-the Council of the I.E.E. states that the incorporation of the word "electronics" in the title "is intended to emphasize to the profession that the Institution is manifestly the learned society for those properly qualified electrical engineers who regard themselves as 'electronic' engineers."


Noise and Vibration.-The Acoustics Group of the Physical Society is sponsoring two symposia dealing with the problems of noise and vibration. The first is at 2.15 on March 24th in the Physics Department of Imperial College, London, S.W.7, under the title "Recent Studies of Noise Problems." The second is a one-day meeting on April 7th in the Physics Department of Southampton University, entitled "New Techniques in the Analysis of Noise and Vibration"; it will include contributions on both practical and theoretical aspects of the use of correlation techniques, of digital computers for data processing and analysis, and on general applications of statistical communication theory. Further details of this joint meeting with the Institute of Physics may be obtained from D. M. A. Mercer, Physics Department, The University, Southampton.

Nine Million.-During January combined televisionsound licences in the U.K. passed the nine-million mark, for at the end of the year the total was $8,899,067$. Sound-only licences at December 31st totalled 5,853,549 including 371,391 for car radio. The year's increase in TV/sound licences was 1.13 M compared with 1.19 M during 1957 and 1.17 M during 1956.
C. \& G. Radio Society.-L. H. Bedford, C.B.E., chief engineer of the English Electric Company's guided weapons division, has been elected president of the City and Guilds College Radio Society in succession to Dr. J. D. McGee, O.B.E., professor of instrument technology at Imperial College.

Telemetry Symposium.-Since going to press with the "March Meetings" (page 150 ) we have received amended details from the Brit.I.R.E. of the radio telemetry symposium which is being held on March 25th at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1. The afternoon session (3-5.30) opens with an introductory paper by A. Cowie, of R.A.E., which is followed by three papers describing a 24 -channel time-division multiplex f.m./a.m. system. Papers to be read at the evening session (6.30-8.30) cover a six-channel frequency-division multiplex f.m./ a.m. system and a pulse position modulation system. Non-members should apply to the Institution, 9, Bedford Square, London, W.C.1, for tickets.

Valve design techniques will be featured at a five-day London exhibition being organized by the English Electric Valve Company. It will be held at Kensington Palace Hotel, De Vere Gardens, W.8, from March 17th to 21 st and will be open each day from 10.0 to 7.0 but on the first day admission will be limited to the Press until 3.0.
$3 \frac{3}{4}-\mathrm{in} /$ sec Tape Records.-Since "Free Grid" prepared his copy for this issue in which he mentions one supplier of " $3 \frac{3}{4}$ " tape records we have been advised by Guildford Sound Recordings, of Birmingham, that Music on Tape Ltd., of Laurence Pountney Hill, London, E.C.4, also produces these tapes.
" Hi-Fi-A Guide to Good Listening" is the theme of the exhibition of high quality radio and audio equipment being staged by the Council of Industrial Design at the Design Centre, 28 Haymarket, London, S.W.1. It will continue until March 14th. The Centre is now open until 7.0 on Wednesdays as well as Thursdays.
"Do-It-Yourself."-Gilbert Davey, who in 1957 showed viewers to the B.B.C. Television Children's Hour how to make a one-valve regenerative detector receiver, is starting a series of instruction on building a transistor pocket receiver. It will begin on March 23rd in the Children's Hour programme "Focus."

## FROM ABROAD

Kelly Award.-In honour of Dr. Mervin J. Kelly, chairman of the board of Bell Telephone Laboratories, the American I.E.E., in collaboration with the Laboratories, is establishing an annual award "for achievement in the field of telecommunications." It will consist of a bronze medal and $\$ 1,000$. Dr. Kelly, who retires on March 1st after 41 years with Bell Telephones, was closely associated with Sir Gordon Radley, directorgeneral of the Post Office, in planning the first transatlantic telephone cable.

Microwave Tubes.-An international congress on microwave tubes on the lines of those held in Paris in 1956 and in London last year is being planned by the Verband Deutscher Elektrotechniker for 1960. It will be held in Munich from June 7th to 11th.

New Zealand has adopted the 625-line standard for its experimental television transmissions starting in Auckland. These tests, using channel 3 ( $55.25 \mathrm{Mc} / \mathrm{s}$ vision and $60.75 \mathrm{Mc} / \mathrm{s}$ sound), are being conducted by the New Zealand Broadcasting Service from its mediumwave station building in the capital. According to one manufacturer 17 in sets will cost about $£ 100$.
West Berlin is to have a new $100-\mathrm{kW}$ transmitter to replace the $20-\mathrm{kW}$ equipment at present used at the Sender Freies Berlin siation which radiates on $566 \mathrm{kc} / \mathrm{s}$. It is being provided by Telefunken, the cost being met by a grant of $950,000 \mathrm{DM}$ from the association of broadcasting organizations (Arbeitsgemeinschaft der Rundfunkanstalten). The transmitter will be switchable to one third of the power on any frequency in the mediumwave band and is anode modulated through a pushpull Class B amplifier. It is planned to come into service in the Autumn.

## Personalities

Air Marshal Sir Raymund Hart, K.B.E., who succeeds Vice-Admiral J. W. S. Dorling as director of the Radio Industry Council was, until the end of January, Controller of Engineering and Equipment in the Air Ministry. Sir Raymund, who is 60 , qualified as a signals officer in 1928 and was for three years prior to 1939 employed at the radar research establishment at Bawdsey on the development and operation of ground-based radar systems. During this period he was concerned with the training of radar operating and servicing staff for the chain of radar stations and was responsible for developing the radar reporting system. In 1939 he went to the headquarters of Fighter Command, of which he later became Chief Signals Officer. In 1944 he was appointed Chief Air Signals Officer at S.H.A.E.F. Among the posts he has filled since the war are Air Officer Commanding No. 27 Signals Training Group; A.O.C. No. 90 Signals Group, DirectorGeneral of Engineering at the Air Ministry and, since October 1956, Controller of Engineering and Equipment. Air Marshal Hart was knighted in 1957.


Sir RAYMUND HART

F. C. LUNNON
F. C. Lunnon, assistant engineer-in-chief of Marconi's W /T Co., has retired after 47 years' service with the company. Mr. Lunnon's early years with the company were spent at the radio stations at Clifden, Ireland and Glace Bay, Nova Scotia, the two stations which provided the first commercial transatlantic wireless circuit. In 1926 he was given charge of the Writtle development establishment and remained there until 1946 when he was appointed development manager. He became assistant engineer-in-chief in 1951.
W. A. S. Butement, O.B.E., who was Assistant Director of Scientific Research in the Ministry of Supply during the latter part of the war and since the war has been chief scientist in the Australian Department of Supply, was promoted Commander of the Order of the British Empire in the New Year Honours. On the recommendation of the Royal Commission on Awards to Inventors he received an award for his "contribution to the development of radar installations" which included the "split" method of d.f., and a fire control system using echoes from shell splashes.

Major General W. A. Scott, C.B., C.B.E., Director of Communications in the Foreign Office, was appointed a Knight Commander of the Order of St. Michael and St. George (K.C.M.G.) in the New Year Honours.
S. F. Follett.-We omitted to announce in our February note (page 58) on the appointment of S. F. Follett as Deputy Director of the Royal Aircraft Establishment that he was appointed Commander of the Order of St. Michael and St. George in the New Year Honours.

Sir John Cockcroft, K.C.B., F.R.S., who has accepted the invitation to become the first Master of Churchill College to be built in Cambridge, recently received the honorary degree of doctor of technical sciences at the Delft technical university. Sir John, who was Cbief Superintendent, Air Defence and Research Establishment throughout the war and was for some time director of the Atomic Energy Research Establishment, is now a member of the U.K. Atomic Energy Authority.
T. S. England, B.Sc., Ph.D., F.Inst.P., A.M.I.E.E., has been appointed head of airborne radar at the Royal Radar Establishment, Malvern. Dr. England, who graduated at the University of Durham in 1937 was for two years working on radar in the Ministry of Aircraft Production before going to T.R.E. (now R.R.E.) in 1942. In 1948 he returned to Durham University where, as a result of two years research work in medical physics, he received his Ph.D. degree. He rejoined T.R.E. in 1950 becoming Superintendent, Circuits and Electronics, in 1954 and since 1956 has been Superintendent, Radar Ballistics.

Sir Robert Fraser, O.B.E., B.A., B.Sc., DirectorGeneral of the Independent Television Authority, has been awarded the Fellowship of the Television Society.

Dr. C. S. Szegho, who was for seven years head of cathode-ray tube research in Baird Television and since 1942 has been director of research with Rauland Corporation, of Chicago, has been awarded the Fellowship of the Television Society. Dr. Szegho, who was born in Hungary, received his doctorate of engineering in Germany.
E. K. Cole, C.B.E., chairman and managing director of the well-known firm bearing his name, which he founded in 1926, has been elected an Honorary Member of the Brit.I.R.E., "in recognition of his services to the radio and electronics industry and profession."
E. L. E. Pawley, O.B.E., M.Sc., M.I.E.E., head of the B.B.C. Engineering Services Group, has been re-elected chairman of the E.B.U. Technical Committee. He is also chairman of the committee's working party concerned with television and sound broadcasting on v.h.f. and u.h.f. Another B.B.C. representative on the Committee, M. J. L. Pulling, C.B.E., M.A., M.I.E.E., is chairman of the working party covering international television relays. Mr. Pulling, who has been with the B.B.C. since 1934, was for some years Superintendent Engineer (Recording) and is now Controller, Television Service Engineering.

Ralph Brewer, who as mentioned briefly last month (page 56), received the National Reliability Award for his paper entitled "Life Tests of Electron Tubes and the Analysis of Failure Causes" read at last year's American National Symposium on Reliability and Quality Control in Elestronics, has been in the G.E.C. Research Laboratories since 1937. During the war he worked on early magnetrons for radar and after the war took charge of valve reliability studies. His work has now extended to cover the study of the survival characteristics of transistors and related semiconductor devices. He is 44.
R. W. Stobbs, F.R.I.C., F.I.M., has been appointed general manager of Preformations Limited, the company recently formed by The Plessey Company and the Arnold Engineering Company, of Illinois, for the manufacture of "Magloy" permanent magnets at Swindon. Mr. Stobbs joined Plessey five years ago as principal metallurgist at the Company's Ilford factory.

Professor A. L. Cullen, Ph.D., B.Sc., who occupics the chair of electrical engineering in the University of Sheffield, has been awarded a grant of $£ 3,835$ by the Paul Instrument Fund Committee for the construction of a detector in which radiation pressure is used to convert a microwave signal to an audio or intermediate frequency.

Vice-Admiral Sir John Eaton, K.B.E., C.B., R.N. (Retd.) has joined Marconi's as Chicf of Administration at the Research and Development Laboratories at Great Baddow, Essex. He will be responsible for all administrative matters to the Chief of Research, Dr. E. Eastwood.

Air Commodore C. A. Bell, formerly Director of Electronics Research and Development (Air) in the Ministry of Supply (see December 1958, page 576) has joined the staff of G.E.C.'s Electronics Division. During the war, as a member of the British Air Commission in Washington, he was responsible for the radio equipment of American aircraft for the R.A.F. and prior to joining the Ministry in 1954 held several R.A.F. appointments in the research and development field.

William T. Frost, who after 10 years with the B.B.C. went to the U.S.A. last year and joined Ampex Corporation's video development unit, has been promoted to staff engineer. He is at present in charge of an advanced development investigation of basic head/tape phenomena in instrumentation wide-band recording.
A. J. Gray, B.Sc., A.M.I.E.E., who has been with Ferranti since 1935, has been appointed general works manager of the company following the retirement of W. Hunt, M.B.E.

## OBITUARY

Ronald Keen, M.B.E., B.Eng., M.I.E.E., the direc-tion-finding specialist and author of the textbook "Wireless Direction Finding", died at Umtali, Southern Rhodesia, a few months ago. He joined Marconi's in 1912 and during the first World War was with the Admiralty serving for the most part overseas on d.f. installations. He returned to Marconi's after the war and together with Capt. H. J. Round and the late G. M. Wright was closely associated with the design and construction in 1923 of the Land's End d.f. station-the first coastal station specifically designed as a service to shipping. He transferred to the company's Traffic Services, which eventually became part of what is now Cable and Wireless, and from 1924 to 1939 was in charge of the Brentwood receiving station. During the war he was a Major in the Army Special Communications Unit and was responsible for the installation and technical operation of a network of high-frequency d.f. stations throughout the British Isles.

Hans Bredow, who died on January 9th this year, aged 79, joined the Telefunken Company in 1904. He was associated with Graf von Arco in the early development of spark telegraphy. In 1919 he joined the German Post Office and was responsible for the preparations which led to the commencement of broadcasting in Germany in Octcber, 1923. In 1926 he became Commissar for Broadcasting. He was called on as a consultant when German broadcasting was reorganized after the war.
Stanley T. Cope, Marconi's technical librarian died on January 31st aged 52. He joined the company's research department in 1933 and in 1947 transferred to the technical information division where his work lay in the writing and editing of technical handbooks. He became technical librarian seven years ago.

## "Wireless World" Index

As stated last month, the index to Volume 64 (1958) is now available price is (postage 3 d ). Our publishers will undertake the binding of readers' issues, the cost being 25 s per volume, including binding case, index and return postage. Copies should be sent to lliffe \& Sons, Ltd., Binding Department, c/o 4 lliffe Yard, London. S.E.17, with a note of the sender's name and address. A separate note, confirming despatch, together with remitrance shouid be sent to the Publishing Department, Dorset House, Stamford Street, London, S.E.I.

# European Television Stations 

survey of the continent's NeTwork

DURING the recent abnormal propagation conditions reception of, and interference from, foreign television stations has been frequent, and we have received a number of requests for help in identifying stations. It is thought, therefore, that it would be of more than passing interest to bring together in one survey the operating characteristics of the 500 or more television stations now operating on the Continent.

We have limited this survey to the European Broadcasting Area-it is hoped to cover other parts of the world at a later date. This area is bounded on the South by parallel $30^{\circ}$ North (bringing in parts of North Africa) on the East by the meridian $40^{\circ}$ East (thus including only Western U.S.S.R.) and on the West by coastlines. Most of the information given has been obtained from the broadcasting organizations in the countries concerned, and this is supplemented by data published from time to time by the European Broadcasting Union.

The map on the following two pages is based on information prepared by the E.B.U. It is impracticable on a map of this size to show all the available information regarding radio and cable links; the only differentiation shown, therefore, is between links for the main transmitters and those for satellites which in most cases depend for their input on direct reception of a main transmitter. We also give on the map details of the standards conversion available for Eurovision links. Incidentally, it is worth recording that secondary television circuits to bypass the national networks are now provided extensively on the Continent to facilitate the unilateral, bilateral or multilateral interchange of programmes.
Although basically there are three television systems in use on the Continent there are in fact a number of variants. For instance, the Belgium version of the French 819-line system is accommodated in a $7-\mathrm{Mc} / \mathrm{s}$ channel with a vision bandwidth of
$5 \mathrm{Mc} / \mathrm{s}$ compared with 14 and $10.5 \mathrm{Mc} / \mathrm{s}$, respectively, in France. Similarly, there are major differences between the 625 -line service used in the majority of European countries and those employed in the U.S.S.R. and Belgium. The characteristics of the world's television standards are given in Table 1.

## Channels : E.R.P.: Polarization

Details of the channels employed in the European systems are given in Table 2. These channel numbers are used after the name of the station in the particulars of each country in the following pages.
An asterisk has been inserted against the e.r.p. of some transmitters. This indicates that the figure given is the maximum for a directional aerial. Transmitters employing vertical polarization are marked (V), the others being horizontally polarized.

A reproduction of the test card or
(Continued on page 111)

TABLE I:WORLD'S TELEVISION STANDARDS

|  | 405 | 525 | 625 (C.C.I.R.) | 625 (O.I.R.) | 819 | 819 (Belgian) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Vision bandwidth (Mc/s) Channel width (Mc/s) .... ... | 5 | 4 6 | 5 7 | 8 | 10.4 | 5 7 |
| Sound carrier relative to vision carrier (Mc/s) $\ldots$... | -3.5 | +4.5 | $+5.5$ | $+6.5$ | -11.15§ | $+5.5$ |
| Sound carrier relative to edge of channel ( $\mathrm{Mc} / \mathrm{s}$ ) | +0.25 +10.25 | -0.25 | 15,625-0.25 | -0.25 $15,625+0.05 \%$ | +0.10§ | $20,475 \pm 0.1 \%$ |
| $\begin{array}{llll}\text { Line frequency (c/s) } \\ \text { Lramefrequency (c/s) } & \ldots & \ldots & \ldots \\ \text { Fin }\end{array}$ | 10,125 50 | 15,750 60 | 15.625 $50.1 \%$ | 15,625+0.05\% | 20.475 50 | $\begin{aligned} & 20,4 / 5 \pm 0.1 \% \\ & 50 \end{aligned}$ |
| $\begin{array}{llll}\text { Frame frequency (c/s) } \\ \text { Picture frequency }(\mathrm{c} / \mathrm{s}) & \ldots . . & \ldots & \ldots\end{array}$ | 25 | 30 | 25 | 25 | 25 | 25 |
| Sense of vision modulation $\quad \cdots \quad . .$. | positive | negative | negi.tive* | negative | positive | positive |
| Blanking level as \% of peak carrier ... ... | 30 0 | 75 | $7{ }^{\text {7 }}$ |  | 25 | $0 \xrightarrow{25}$ |
| Minimum level of carrier as \% of peak carrier... | 0 | $\leqslant 15 \dagger$ | $10^{*}$ | 10 min . | $\leqslant 3$ | a.m. ${ }^{\text {a }}$ |
| Sound modulation Deviation (ke/s) | E.m. | f.m. | 1.m.* | f.m. |  |  |
| Deviation (kc/s) Pre-emphasis ( $\mu \mathrm{sec}$ ) $)$$\ldots$ |  | ${ }_{75}$ | $\pm 50$ | 50 |  | 50 |

$\dagger$ In the Japanese 525 -line system the figure is $10-15 \%$.

* In the Belgian 625-line system positive vision modulation is used; the blanking level is $25 \%$; minimum level of carrier is 0 - $3 \%$; and sound is a.m. with $50 \mu \mathrm{sec}$, pre-emphasis.
§ In some of the French channels the vision and sound carriers are reversed-the vision carrier being the lower.
TABLE 2: EUROPEAN TV CHANNELS IN BANDS I \& III

| U.K. 405 lines ( $5 \mathrm{Mc} / \mathrm{s}$ channels) |  |  | C.C.I.R. 625 lines and Belgian 819 lines ( $7 \mathrm{Mc} / \mathrm{s}$ channels) |  |  | O.I.R. 625 lines ( $8 \mathrm{Mc} / \mathrm{s}$ channels) |  |  | French | 819 lines channels) | (14 Mc/s |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| B1 | 45.00 | 41.50 | E2 | 48.25 | 53.75 | $\cdot{ }^{\circ} \mathrm{Ol}$ | 49.75 | 5625 | F2 | 52.40 | 41.25 |
| B2 | 51.75 | 48.25 | E3 | 55.25 | 60.75 | $\mathrm{O}^{2}$ | 59.25 | 65.75 | F3 | 56.15 | 67.30 |
| B3 | 56.75 | 53.25 | E4 | 62.25 | 67.75 | $\mathrm{OH}^{+}$ | 77.25 | 83.75 | F4 | 65.55 | 54.40 175.15 |
| B4 | 61.75 | 58.25 | E5 | 175.25 | 180.75 | O4t | 85.25 | 91.75 | F5 | 164.00 | 175.15 |
| B5 | 66.75 | 63.25 | E6 | 182.25 | 187.75 | O5t | 93.25 | 99.75 | F6 | 173.40 | 162.25 |
| B6 | 179.75 | 176.25 | E7 | 189.25 | 194.75 | O6 | 175.25 | 181.75 | ${ }_{\text {F }} \mathrm{F}$ | 177.15 | 18830 |
| B7 | 184.75 | 181.25 | E8 | 196.25 | 201.75 | O7 | 183.25 | 189.75 | ${ }_{\text {F8A }}$ | 185.25 | 174.10 175.40 |
| B8 | 189.75 | 186.25 | E9 | 203.25 | 208.75 | O8 | 191.25 199.25 | 197.75 205.75 | F88 | 186.55 190.30 | 175.40 |
| ${ }^{89}$ | 194.75 | 191.25 | E10 | 210.25 | 215.75 222.75 |  | 199.25 | 205.75 213.75 | F10 | 199.70 | 188.55 |
| B10 B11 | 199.75 204.75 | 196.25 201.25 | El1* | 217.25 | 222.75 | O11* | 215.25 | 221.75 | FII | 203.45 | 214.60 |
| ${ }_{\text {B12 }}$ | 209.75 | 206.25 |  |  |  | O12* | 22325 | 22975 | FI2 | 212.85 | 201.70 |

The vision carrier precedes the sound carrier in this list.
$\dagger$ These channels are outside the limits of Band I ( $41-68 \mathrm{Mc} / \mathrm{s}$ ).

[^1]

tuning signal used for the country's television service is also given in most of the summaries.

## ALBANIA

There is at present no television service in Albania. The Stockholm Plan of 1952 provided for one station in Band I and three in Band III using the O.I.R. 625-line system.

## ALGERIA

Although outside the natural boundaries of Europe, Algeria does come within the European Broadcasting Area and is therefore covered by the provisions of the Stockholm Plan. The television service is provided by the French broadcasting authority, Radiodiffusion-Télévision Francaise; the standards employed ( 819 lines) and the test card are, therefore, the same as in France. Under the Stockholm Plan five transmitters are provided for but at present only two, at Algiers (Cap Matifou) and Oran, are in operation. The service is government operated and is financed from licence revenue - 2,000 francs sound, and 6,000 francs television. The number of television receivers in use is about 28,000 .
Algiers
Oran channel
FJI
e.r.p.
20 kW
20

## AUSTRIA

Two years ago the Austrian Broadcasting System, Österreichischer Rundfunk, introduced a regular television service. There are now 8 main transmitters in Bands I and III, which are listed below, and 5 satellites. There is also a second station in Vienna which radiates in Band IV.


The television service is financed by a bank credit covered by a guarantee of the Austrian Federal Government. In January this year advertising programmes were introduced to supplement the income. The annual licence fee is 600 schillings (excluding sound

|  | channel | e. |
| :---: | :---: | :---: |
| Gaisberg (Salzburg) | E8 | 60 kW |
| Jauerling (St. Pölten) ... | $\dagger$ | 60 |
| Kahienberg (Vienna) ... | 5 | 60 |
| Linz, Upper Austria ... | 6 | 3 |
| Patscherkofel (Innsbruck) | 4 | 30 |
| Pyramidenkogel (Klagenfurs) | 10 | 30 |
| Schöckl (Graz) ... ... | 7 | 60 |
| Sonnwendsiein, Lower |  | 1.5 |
| Austria | 10 | . 5 |

†Vision $49.75 \mathrm{Mc} / \mathrm{s}$, sound $55.25 \mathrm{Mc} / \mathrm{s}$.
radio) which is about £8. The present number of licences is approx. 60,000.

## BELGIUM

Because its neighbouring countries operate on different standards (625 and 819 lines), Belgium operates a two-standard television service using a modified version of the Gerber

(C.C.I.R.) 625-line system for its Flemish transmissions and a modified version of the French 819-line system (with a $5-\mathrm{Mc} / \mathrm{s}$ video bandwidth) for its French transmissions. All sets in Belgium are designed to receive four standards: both the national services, the C.C.I.R. 625line standard and the French 819 lines.

The country's television service is operated by the Institut National Belge de Radiodiffusion (I.N.R.) and is financed by the government. Since January, 1958, television set owners have had to pay an annual tax of 840 Belgian francs (£6). The number of television receivers is about 300,000.

| Antwerp (625) |  | channel |  | ${ }_{\text {e.r.p. }}^{6+}$ (V) |
| :---: | :---: | :---: | :---: | :---: |
|  |  | ... | E2 |  |
| Liege (819) |  | ... | 3 | 100 |
| Ruiselede (625) |  | ... | 2 | 100 |
| Wavre (819) | $\ldots$ | ... | 8 | 100 |
| Wavre (625) | .. |  | 10 | 100 |

## BULGARIA

So far only experimental television transmissions using a low-power station at Sofia have been made by Radiodiffusion Bulgare, which has adopted the O.I.R. 625-line standard. For these tests channel O3 has been used. Regular transmissions from a new station in the capital are due to begin on May 1st.

## CYPRUS

An experimental television service was introduced on the island by the Cyprus Broadcasting Service just over a year ago. The 625 -line transmitter in Nicosia radiates in channel E2 with an e.r.p. of 1.5 kW . Television licences, costing £1 per annum, totalled 193 at the end of November.

## CZECHOSLOVAKIA

Five stations are now ised in the Czechoslovak television chain which
operates on the O.I.R. 625-line standard. The service is state financed and viewers pay a fee of 180 crowns

|  |  | channel |  |  | e.r.p. |
| :--- | :---: | :---: | :---: | :---: | :---: |
|  |  | $\ldots$ | 03 | $12 k W$ |  |
| Bratislava... | $\ldots$ | $\ldots$ | 9 | 10 |  |
| Brno | $\ldots$ | $\ldots$ | $\ldots$ | 10 | 0.6 |
| Karlovy Vary | $\ldots$ | $\ldots$ | 10 | 12 |  |
| Ostrava | $\ldots$ | $\ldots$ | $\ldots$ | 2 | 5 |
| Prague $\ldots$ | $\ldots$ | $\ldots$ | 2 | 5 |  |


(£9) excluding sound radio. At the end of last June chere were about 200,000 television licences in force.

## DENMARK

Two high-power transmitters and four medium-power stations employing the $625-1$ ine standard are operated by the national broadcasting service, Statsradiofonien, which is financed by the revenue from licence fees. Television set owners pay an annual licence fee of 55 kroner

(about £3). There are about 200,000 televisión licences in force.

|  |  | channel |  | e.r.p. |
| :--- | :---: | :---: | :---: | :---: |
| Aarhus | $\ldots$ | $\ldots$ | $\ldots$ | E8 |
| Aalborg | 10kW |  |  |  |
| Copenhagen | $\ldots$ | $\ldots$ | 5 | 10 |
| Fyn | $\ldots$ | $\ldots$ | 4 | 10 |
| Sönderiyliand | $\ldots$ | $\ldots$ | 7 | 10 |
| Vestyylland | $\ldots$ | $\ldots$ | 10 | 60 |
|  |  |  |  |  |

## EIRE

There is no television service in Eire, but the Minister for Posts and Telegraphs recently appointed a twenty-one-man Television Commission "to consider and make recommendations on the question of establishing a television service." The committee is to base its recommendations on the assumption that the cost of the service will not fall on the Government. The Stockholm Plan provides for 5 stations in Band III for Eire and these are shown as operating on the British 405-line standard.

## FINLAND

Three main stations, with two provisional stations, form the television network of the Finnish broadcasting authority-Oy. Yleisradio Ab --which employs the C.C.I.R. 625-line system. Since January, 1958, viewers have had to pay a licence fee of $6,000 \mathrm{Mk}$ (£7), excluding sound radio which is a

further $1,200 \mathrm{Mk}$. Television licences totalled 7,750 at the end of 1958.

|  |  |  | channel |  | e.r.p. |
| :--- | ---: | :---: | :---: | :---: | :---: |
| Helsinki | $\ldots$ | $\ldots$ | $\ldots$ | E6 | IOkW |
| Kotka | $\ldots$ | $\ldots$ | $\ldots$ | 5 | $\dagger$ |
| Lahti | $\ldots$ | $\ldots$ | $\ldots$ | 9 | 15 |
| Tampere | $\ldots$ | $\ldots$ | $\ldots$ | 8 | $\dagger$ |
| Turku | $\ldots$ | $\ldots$ | $\ldots$ | 7 | 25 |

†Provisional transmitters.

## FRANCE

Regular television transmissions have been radiated in France since 1938 when a 455-line system was used by the Eiffel Tower station. It

radiated in Channel F1 with a $7.6-\mathrm{Mc} / \mathrm{s}$ bandwidth. The scanning rate was subsequently changed to 441 lines and the transmissions from Eiffel Tower continued for over five years after the introduction of the present 819 -line standard in 1950. Since the cessation of the 441-line transmissions the French Channel 1 ( $46 \mathrm{Mc} / \mathrm{s}$ vision, $42 \mathrm{Mc} / \mathrm{s}$ sound) has not been used.

In order to accommodate the maximum number of stations in the few $14-\mathrm{Mc} / \mathrm{s}$ channels available in Bands I and III, the R.T.F. (Radio-diffusion-Télévision Française) has adopted a scheme whereby they accommodate two channels in one. This is done by reversing the position of the vision carrier relative to the sound carrier in alternate channels so that four carriers come within a $14-\mathrm{Mc} / \mathrm{s}$ band.

The French television service is financed from licence fees and from
government grants. The fee for a home television set is 6,000 francs ( $£ 410 \mathrm{~s}$ ), but is being increased to 7,500 francs in July. Where a television set is used in public places the fee is four times as much.

The present chain includes the 22 main stations listed below and 15 satellites. Under an agreement recently concluded between the Monacan and French governments, the television station at Monte Carlo will receive the major part of its programmes from the R.T.F.
An unusual feature of the French television service is that in addition to the R.T.F. network a number of satellite stations have been erected by private enterprise to improve local reception. The number of these stations increased so rapidly and so indiscriminately that recently regulations were drawn up prohibiting the erection of such stations except by local authorities. The power of the satellites varies from 0.1 W to 5 W .

At the end of December there were 988,594 television licences in force.

| Bordeaux... | channel |  |  | e.r.p. $25 \mathrm{k} \mathrm{W}^{*}$ |
| :---: | :---: | :---: | :---: | :---: |
|  | $\ldots$ | ... | F10 | $25 \mathrm{kW*}$ |
| Bourges ... | ... | $\ldots$ | 9 | 200 |
| Caen | ... | ... | 2 | 50 |
| Cherbourg | ... | ... | 12 |  |
| Côte d'Azur |  | $\ldots$ | 6 | 10 (V) |
| Diion | ... | ... | 10 | 30 (V) |
| Grenoble | ... | $\cdots$ | 10 | 20 |
| Lille ... | ... | ... | 8 A | 200 |
| Luttange ... | ... | $\ldots$ | 6 | 200* |
| Lyon ... | ... | ... | 5 | 0.1 |
| Marseille ... | ... | $\ldots$ | 8 | 300* |
| Mont-Pilat | ... | $\ldots$ | 12 | 200 |
| Mulhouse... | ... | $\ldots$ | 8 | 200* |
| Nancy | ... | ... | 7 | $1.6^{*}(\mathrm{~V})$ |
| Nantes | ... | ... | 4 | 0.5 (V) |
| Paris | $\ldots$ | ... | 8 A | 180 |
| Pic du Midi |  | ... | 5 | 20 |
| Puy de Dốne | $\ldots$ | ... | 6 | 160* (V) |
| Reims | $\ldots$ | ... | 5 | 0.1 (V) |
| Rennes ... | $\ldots$ |  | 5 | 0.5 |
| Rouen ${ }_{\text {Strasbourg }}$ | $\cdots$ | $\cdots$ | 10 | 50** |

## GERMANY, EAST

Television in the German Democratic Republic is State controlled and is operated by the Deutscher Demokratischer Rundfunk. The 625 -line system is employed but the channel numbering differs from either the C.C.I.R. or O.I.R. channels in table 2. The vision and sound carriers (in $\mathrm{Mc} / \mathrm{s}$ ) of the channels at present in use are: 1 (59.25/64.75); 2 (145.25/150.75); 3 (55.25/60.75); 5 (175.25/180.75);

D.D.R. EAST GERMANY

6 (182.25/187.75); 8 (196.25/201.75); 11 (217.25/222.75).

At the end of August there were 257,000 television licences issued in the Republic.

|  | channel |  | e.r.p. 100 kW |
| :---: | :---: | :---: | :---: |
| Berlin ... | ... | 5 |  |
| Brocken |  | 6 | - |
| Katzenstein (Dresden) | .. | 2 | 100 |
| Helpterberg | ... | 3 | - |
| Inselsberg | ... | 5 | - |
| Karl-Marx-Stadt | ... | 8 | - |
| Leipzig | ... | 1 | 100 |
| Marlow | ... | 8 | - |
| Schwerin | ... | 11 | - |

## GERMANY, WEST

Although the television stations in the German Federal Republic are operated by a number of authorities - each one covering a zone of the immediate post-war period-there is a common television programme known as Deutches Fernsehen, to which each of these organizations contributes. There are 26 main stations and these are listed on page 114 with the initials of the operating authority against each-B.R. (Bayerjscher Rundfunk); H.R. (Hessischer Rundfunk); N.D.R. (Norddeutscher Rundfunk); S.D.R. (Süddeutscher Rundfunk); S.F.B. (Sender Freies Berlin); S.W.F. (Südwestfunk); and


BAYERISCHER RUNDFUNK


HESSISCHER RUNDFUNK

N.D.R., W.D.R. and S.F.B.
W.D.R. (Westdeutscher Rundfunk). In addition to the main stations listed there are over 80 satellites in use. A few experimental transmitters operating in Band IV have also been built but these are not listed.

West German television, which employs the C.C.I.R. 625-line system, was, until recently, financed entirely from licence fees, but this is now supplemented by commercial programmes. A combined televisionsound licence costs 84DM (£7) a year, of which the postal authorities retain about $27 \%$.

With the incorporation of the Saar in the Federal Republic there has arisen the problem of the commercial sound and television stations in this territory. The French 819-line standard was employed by the commercial television stations, but these are now closed down and the present station at Saarbrücken employs the $625-$ line standard. Its e.r.p. is being increased to 100 kW .

In addition to the national network there are also a few Band IV stations (not listed) being operated for the American Forces. These employ the U.S.A. 525 -line standard.

The number of television receivers


SÜDDEUTSCHER RUNDFUNK


SÜDWESTFUNK


SAARLAND
in the Federal Republic was $1,765,410$ at the end of September.

|  | channe. |  |
| :---: | :---: | :---: |
| galen (S.D.R.) ... | E8 | $20 \mathrm{~kW}{ }^{*}$ (V) |
| Berlin (S.F.B.) | 7 |  |
| Biedenkopf (H.R.) | 2 | 20* (V) |
| Bremen-Oldenburg <br> (N.D.R.) | 2 | 100 |
| Cologne (W.D.R.) | 11 | 5 |
| Dillberg/Nürnberg (B.R.) | 6 | 100 |
| Feldberg/Schwarzwald (S.W.F) | 8 | 100 |
| Fridberg/Taunus (H.R.) | 8 | 100 |
| Flensburg (N.D.R.) | 4 | 50* |
| Grünten (B.R.) | 2 | 100* |
| tramburg (N.D.R.) | 9 | 100 |
| Hannover (N.D.R.) | 8 | 5 |
| Harz-West (N.D.R.) | 10 | 100 |
| Hoher Meissner (H.R.) | 7 | 100 |
| Hornisgrinde (S.W.F.) | 9 | 100* |
| Kiel (N.D.R.) | 5 | 5 |
| Koblenz (S.W.F.) ... | 6 | 50 |
| Kreuzberg/Rhön (B.R.)... | 3 | $100^{*}$ (V) |
| Langenberg (W.D.R.) | 9 | 100 |
| Raichberg (S.W.F.) ... | 4 | 40 |
| Saarbrücken ... ... | 2 | 10* (V) |
| Stuttgart-Degerloch (S.D.R.) | 11 | 100 |
| Teutoburger Wald (W.D.R.) | 11 | 100 |
| $\checkmark$ deinbiet (S.W.F.) | 10 | 50* |
| Wendelstein (B.R.) | 10 | 100* |
| Würzburg (B.R.) | 10 | 1 |

## GREECE

No provision was made in the 1952 Stockholm Plan for television stations in Greece because the Government " had not yet finalized its plans for y.h.f.; sound and television broadcasting." The delegation to the conference did, however, state that initially three stations would be erected at Athens, Salonika and Patras. No announcement has been made of the implementation of these plans.

## HUNGARY

Following a series of experimental transmissions in channel O2 from a low-power transmitter on the outskirts of Budapest, a new high-power station was brought into service in the capital in January last year and recently a second station, at Pécs, was opened. The television service, which employs the O.I.R. standard, is State financed and the licence fee

is 600 Forints (£19) a year excluding sound radio. The present number of receivers is approximately 24,000 .
Budapest
Pécs
channel

## ITALY

By far the biggest concentration of television stations in Europe is in Italy where, at the beginning of the year, there were 270-
approximately one half of the Continent's total. Of this number only 24 are major stations (they are listed below), the remainder being satellites which radiate a main station's programme received by radio. One of these main stations (M. Penice) has as many as 52 satellites. With a total of some $1,100,000$ television receivers in the country, there is an average of about 4,000 sets to each transmitter.
Although Italy adopted the 625line standard when the national service was started by Radiotelevisione Italiana (RAI) in 1954, the channels used vary somewhat from those generally employed on the Continent. Moreover, under a protocol

to the Stockholm Plan (1952), Italy is permitted to use an additional channel ( 81 to $88 \mathrm{Mc} / \mathrm{s}$ ). Italy's television channels are designated by the following letters with which we give in brackets the vision and sound carriers: A (53.75/59.25); B (62.25/ 67.75); C (82.25/87.75); D (175.25/ 180.75); E (183.75/189.25); F (192.25/197.75); G (201.25/206.75); H (210.25/215.75).

The television service is financed both from licence fees- 14,000 lire (£8) a year for a combined sound and television licence-and from advertising which was introduced when the number of sets in use exceeded 150,000 .

| Gambarie | ... |  | channel D | e.r.p. 19 ch |
| :---: | :---: | :---: | :---: | :---: |
| Martina Franca | $\ldots$ | $\ldots$ | D | 220 |
| Mi.an .. | $\ldots$ | $\ldots$ | G | 24 |
| M. Argentario | ... | ... | E | 2.5 |
| M. Caccia... | -. |  | A | 53 |
| M. Cammarata | ... | ... | A | 29 |
| M. Conero | ... | ... | E | 24 |
| M. Faito ... | *** | ... | B | 53 |
| M. Lauro | ... | ... | F | 100 |
| M. Limbara | ... | ... | H | 3 |
| M. Nerone | ... | . | A | 29 |
| M. Peglia | ... | ... | H | 34 |
| M. Peltegrino | ... | $\ldots$ | H | 8 |
| M. Penice | ... | $\ldots$ | B | 100 |
| M. Sambuco | ... | . | H | 35 |
| M. Scuro... | ... | . | G | 5 |
| M. Serra | -•• | ... | D | 270 |
| M. Soro | ... | ... | E | 5 |
| M. Venda | ... | $\ldots$ | D | 190 |
| M. Vergine | ... | ... | D | 1 |
| Portofino |  |  | H | 127 |
| P. Badde Urbara |  | $\ldots$ | D | 145 |
| Rome | ... | ... | G | 36 |
| Turin ... | ... | $\cdots$ | C | 16 |

## LUXEMBOURG

Tt 8 819-line standard, but with a bandwidth of $7 \mathrm{Mc} / \mathrm{s}$ as in Belgium, was adopted by the Compagnie Luxer, ,ourgeoise de Télédiffusion wher it added television to its com-

mercial sound broadcasting service in January, 1955. The station, which is built on the top of the Ginsterberg ( $1,460 \mathrm{ft}$ ), radiates in channel E7 with a vision e.r.p. of 100 kW . Transmissions are horizontally polarized. The service is financed by advertisements so that no licence fee is paid by the 4,000 set owners in the Grand Duchy.

## MONACO

Since early 1955 a commercial television station, Tele-Monte-Carlo, has been operated in the principality by the company which owned the commercial stations in the Saar. Early last year the station, which radiates on 819 lines in channel F10 with an e.r.p. of 50 kW , was taken over by the Monacan government. It is now leased to R.T.F., the French broadcasting authority, which provides the major part of the programmes.

## NETHERLANDS

Sound broadcasting in the Netherlands is conducted by five societies (representing different political and religious parties) whose activities

since 1947 have been co-ordinated through the Nederlandsche Radin Unie. These same societies in $195^{\prime}$ formed a co-ordinating body, which is known as Nederlandse Teievisic Stichting, for television. Experimental transmissions were onductit from the end of that year $7 \mathrm{t.i}$ regular service using 625 lines w. . started in 1953. Viewers F : annual licence fee of 30 Dutcin florms (about £3), excluding radio, and th-

|  |  |  |  | channel | e.r.p. |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Goes | $\ldots$ | $\ldots$ | $\ldots$ | $E 7$ | $5 k W$ |
| lrnsum | $\ldots$ | $\ldots$ | $\ldots$ | 6 | 25 |
| Lopik | $\ldots$ | $\ldots$ | $\ldots$ | 4 | 20 |
| Markelo | $\ldots$ | $\ldots$ | $\ldots$ | 7 | 50 |
| Roermond | $\ldots$ | $\ldots$ | 5 | 50 |  |

service is government subsidized. Television licences totalled 374,738 at the end of November, 1958.

## NORWAY

An experimental television service has been radiated from a low-power transmitter in Oslo for the past four years.

The 625-line standard has been employed for these tests conducted by the Norwegian broadcasting organization, Norsk Rikskringkasting. As a result of these tests, the Norwegian government has drawn up plans for a national television network and regular transmissions are scheduled to begin in 1960. As with sound broadcasting the provision of

the technical facilities (transmitters and links) comes under the country's telecommunications administration and the programme technical operations are carried out by Norsk Rikskringkasting. Under the Stockholm Plan there is provision for 10 stations in Band I and 23 in Band III, but this number is considered to be insufficient to cover the country satisfactorily. A revised scheme providing for 28 main stations ( 18 of which will be ligh-power) and 19 satellites has, therefore, been drawn up. It is planned to have most of the proposed stations constructed for unattended operation.

## POLAND

In 1952, after an experimental jeriod with two transmitters operating on different standards (441 and 625 lines), the Polish Ministry of Posts and Telecommunications adopted the O.I.R. $625-$ line system. The present chain of 8 stations is operated by the Central Radio and Television Administration of the government. A tax of 480 zloty's

(£7) a year is paid on each television receiver. The number of sets in use is now about 80,000 .

| Gdansk |  |  | channe. | e.r.p |
| :---: | :---: | :---: | :---: | :---: |
|  | $\cdots$ | $\ldots$ | 03 |  |
| Katowice | ... | $\cdots$ | 8 | -26kW |
| Lódz | ... | ... | 6 | 2.5 |
| Poznan | ... | ... | 7 | 3.5 |
| Stettin | $\cdots$ | $\cdots$ | - |  |
| Warsaw | ... | ... | 11 | 95 |
| Warsaw | ... | ... | 2 | 7 |
| Wroclaw | $\ldots$ | $\cdots$ | 12 | 123 |

## PORTUGAL

Three years ago the Portuguese government granted a concession to the Radiotelevisĩo Portuguesa, S.A.R.L., to organize the country's television service. Its licence permits the transmission of commercial programmes and the company is also

allowed to sell and rent television sets and accessories. R.T.P., as it is known, is now operating five stations, the service areas of which cover a large part of the country. 'The service employs the 625-line standard. To encourage the purchase of receivers, licences are not being collected for the first two years of this service. The estimated number of receivers in use is 22,000 .

| Lisbon |  | channe | e.r.p. 100 kW |
| :---: | :---: | :---: | :---: |
| Lousã, Coimbra | ... | 3 | 50 |
| Monchique | ... | 5 | 6.5 |
| Montejunto | ... | S | 1 |
| Oporto |  | 9 | 100 |

## RUMANIA

Under the Stockholm Finn Rumania is to have three stations in Band I and eight in Band III. At present, however, only one transmitter, in Bucharest, is operating. It employs the O.I.R. 625-line standard and radiates in Channel O2 with an e.r.p. of 7.5 kW . The service is government financed.

## SPAIN

Since June, 1956, a regular television service on 625 lines has been broadcast from the low-power transmitter installed experimentally at Madrid by the broadcasting cepartment of the Ministry of Information. A second station at Barcelona was recently opened and plans has been made for additional stations : Santiago and Zaragoza and for a new high-power station at Navacerrada to serve the capital. There are tr:proximately 10,000 television rece1 rs in use. The annual lic-

pesetas ( $£ 210 \mathrm{~s}$ ) for sets with a 14 -in tube and 500 pesetas (£4) for larger sizes.

3arcelona $\begin{array}{cccc} & & \text { channer er.p. } \\ \ldots & \ldots & \text { E3 } & 20 \mathrm{k} v / \\ \ldots & \ldots & 3 & 2\end{array}$

## SWEDEN

Regular television transmissions using the C.C.I.R. 625-line standard have been radiated in Sweden since June, 1956, following a long series of test transmissions. The stations are built and maintained by the Board of Swedish Telecommunications and the programmes provided by the Swedish broadcasting organization, Sveriges Radio AB. The television service is intended to be financed from licence fees, but the government is making grants towards initial development costs. The fee is 100 Swedish crowns (about £7) a year, which is payable quarterly. Incidentally, in areas where "reasonably good reception" is not obtainable set owners pay a registration fee of only 10 Swedish crowns. There are, at

present, about a quarter of a million receivers in use.

|  |  |  | channel | e.r.p. |
| :---: | :---: | :---: | :---: | :---: |
| Gothenburg | ... | $\ldots$ | E9 | 15 k |
| Malmö ... | ... | ... | 10 |  |
| Norrköping | ... | ... | 5 | 15 |
| Stockholm | ... | ... | 4 | 60 |

## SWITZERLAND

Switzerland's tri-lingual television service, which operates on the C.C.I.R. 6.25-line standard, is radiated by the seven stations listed. The stations are provided by the postal administration and the programmes by the Swiss Broadcasting Corporation. The S.B.C. receives $70 \%$ of the proceeds from licence fees and the P.T.T. $30 \%$. The annual fee is 84 Swiss francs, about $£ 7$ for home receivers, and 168 francs for public reception. The s also sub-

sidized by the Swiss Newspaper Publishers Association which is providing 2 M francs a year for 10 years on condition that advertisements and sponsored programmes are not broadcast. The number of licences at the end of the year was 50,300 . In addition to the seven main transmitters there are five privately owned low-power booster stations.

|  |  |  | channel | e.r.p. |
| :--- | :---: | :---: | :---: | :---: |
| Bantiger ... | $\ldots$ | $\ldots$ | E2 | 30 kW |
| Chrischona, | Basel | $\ldots$ | 10 | 10 |
| La Dôle $\ldots . .$. | $\ldots$ | $\ldots$ | 4 | 100 |
| Monte Ceneri | $\ldots$ | $\ldots$ | 5 | 10 |
| San Salvatore | $\ldots$ | $\ldots$ | 10 | 10 |
| Säntis $\ldots$. | $\ldots$ | $\ldots$ | 7 | $30^{*}$ |
| Utliberg | $\ldots$ | $\ldots$ | 3 | 20 |

## TURKEY

Provision is made in the Stockholm Plan for Turkey to have 43 television stations-11 in Band I and 32 in Band III. Tests using 625 lines have been conducted for some time but there is no immediate prospect of a service being introduced. An experimental station in the Technical University, Istanbul, has been radiating in channel E4.

## UNITED KINGDOM

Being the only country to employ the 405 -line system, it is often said that the U.K. is odd-man-out in television, but it must not be forgotten that it was the first country in the world to have a regular television service-in 1936. There is strong feeling in many quarters of industry and research that a change should be made to 625 lines, but, be that as it may, we are here concerned with the present television service which, after the wartime shut-down, was restarted on the same standards in 1946. The service was provided by the B.B.C. until 1955 when, under the provisions of the Television Act, the Independent Television Author-


When used by the I.T.A. this test card carries these initials.
ity started an alternative service. Whereas the B.B.C.'s income is derived from receiving licence fees, the I.T.A.'s comes indirectly from advertising. For the sake of our overseas readers it should be made clear that the I.T.A.'s programmes are not sponsored by advertisers. They are provided by the programme contractor licensed to operate a station, the advertisements being inserted in "natural breaks" in the programmes.
The present combined televisionsound licence fee is £3, plus £1 excise duty which is retained by the Treasury. The B.B.C. receives $87.5 \%$ of the revenue from all receiving licence fees after the Post Office has deducted an amount for its services in collecting fees and investigating complaints of electrical interference. The remaining $12.5 \%$ passes to the Treasury.

There were 8.9 M television licences in force at the end of December.

|  | channel | e.r.p. |
| :---: | :---: | :---: |
| Black Hill (I.T.A.) | 10 | $475 \mathrm{~kW}{ }^{\text {( }}$ (V) |
| Blaen Plwy (B.B.C.) | 3 | 1 |
| Burnhope (I.T.A.) | 8 | 100* |
| Chillerton Down (I.T.A.) | 11 | 100* (V) |
| Croydon (I.T.A.) ... | 9 | 120 (V) |
| Crystal Palace (B.B.C.) | 1 | 200(V) |
| Divis (B.B.C.) ... | 1 | 12 |
| Douglas (B,B.C.) | 5 | 2.5* (V) |
| Dover (B.B.C.) | 2 | 4* (V) |
| Emley Moor (I.T.A.) | 10 | 200* (V) |
| Folkestone (B.B.C.) | 4 | 0.1 |
| Holme Moss (B.B.C.) | 2 | 100(V) |
| Kirk o'Shotts (B.B.C.) | 3 | $100(\mathrm{~V})$ |
| Lichfield (I.T.A.) | 8 | 200(V) |
| Les Platons (B.B.C.) | 4 | 1 |
| Londonderry (B.B.C.) | 2 | 1 |
| Meldrum (B.B.C.) | 4 | 17* |
| N. Hessary Tor (B.B.C.) | 2 | 15* (V) |
| Norwich (B.B.C.) | 3 | 10* |
| Orkneys (B.B.C.) $\dagger$ | 5 | 17 (V) |
| Pontop Pike (B.B.C.) | 5 | 12 |
| Rosemarkie (B.B.C.) | 2 | I |
| Rowridge (B,B,C.) | 3 | 32* (V) |
| St. Hilary (I.T.A.) | 10 | 200(V) |
| Sandale (B.B.C.) | 4 | 16 |
| Sutton Coldfield (B.B.C.) | 4 | $100(\mathrm{~V})$ |
| Wenvoe (B.B.C.) | 5 | 100 (V) |
| Wick (E.B.C.) $\dagger$ | 1 | 4* (V) |
| Winter Hill (I.T.A.) | 9 | 100(V) |

†Temporary stations at present in use.

## U.S.S.R.

As mentioned in the introduction, only that part of the U.S.S.R. within the European Broadcasting Area is dealt with in this survey. Regular transmissions using 343 lines and 240 lines were started respectively from Moscow and Leningrad in 1938. A few years later the Moscow transmitter was modified for a scanning rate of 441 lines, and this standard was used until the introduction in 1948 of its present 625-line standard approved by the O.I.R. (International Broadcasting Organization), representing the Eastern European broadcasting authorities.

The major problem in providing a television service for so vast a country is the provision of links between stations. One method recently employed for a relay from Moscow to Leningrad-about 400 miles-was to use two aircraft as relay stations with an intermediate ground station between them.

We have been unable to obtain from the Soviet authorities a list of stations with the frequencies and powers employed. However, from announcements made from time to time we have been able to prepare a list of some towns west of meridian $40^{\circ} \mathrm{E}$ in which stations are said to be operating. This list, which has been supplemented by information kindly supplied by the Society for Cultural Relations with the U.S.S.R., is given below. Where known the

channel number is given in brackets:-

Dniepropetrovsk; Gomel; Kharkov (O1); Kherson; Kiev (O3); Kishinev; Leningrad (O2); Lugansk; Lwow; Minsk; Moscow (O1); Moscow (O2); Novgorod; Odessa (O1); Petrosavodsk; Riga (O2); Stalino; Stalinogorsk; Tallinn (O2); Vilna (O5); Yaroslavl.

In addition to these main television centres there are also a number of relay or satellite stations. It was recently reported that 30 new stations were brought into operation last year; the northernmost on the Taimyr Peninsular, well within the Arctic Circle. There are about 2.5 M television sets in use in the Union. The "subscription fee" for a television set is 10 roubles a month (about £10 a year).

## VATICAN CITY

Although residents in the Vatican City are able to receive transmissions from the Italian television service, provision was made in the Stockholm Plan for the Papal authorities to have their own television station. Two frequencies-one in each band-were allocated, but so far no decision has been made on the erection of a station.

## YUGOSLAVIA

Two stations were operating independently for a year until last November when a third station, in the capital, was built and all three are now linked and provide a service for $38 \%$ of Yugoslavia's population. The C.C.I.R. 625-line standard has been adopted for the service. About 6,000 receivers are now in use.

|  |  |  | channel | e.r.p. |  |
| :--- | :--- | :--- | :--- | :---: | :---: |
| Belgrade |  | $\ldots$ | $\ldots$ | E6 | 10 kW |
| Ljubljana $\ldots$ | $\ldots$ | $\ldots$ | 10 | 4.5 |  |
| Zagreb | $\ldots$ | $\ldots$ | $\ldots$ | 9 | 4 |

# Time Past 

By P. P. ECKERSLEY, M.I.E.E., F.I.R.E.

This is the second of a series of articles by the first Chief Engineer of the B.B.C. and is concerned with the progress of the revolution caused by the invention of the valve, $a$ progress during which he was intimately concerned with the beginnings of broadcasting. In the first article the author gave an account of how the nineteenth-century scientists established the foundations upon which pioneerng inventors built their systems. In a third article he will round off with some predictions about the future.

IN the early autumn of 1915 I stood on the tarmac of Brooklands Aerodrome next to the late C. E. Prince and heard him say into a microphone "Hullo, Ferdy!* If you are hearing me now it will be the first time that speech has been transmitted from ground to an aeroplane in flight. If you are hearing me, please dip." The lumbering "Rumpty," doing its forty-odd knots, fifteen hundred feet above us, gave an obedient lurch; Ferdy had received the speech, strength $R_{9}$. The incident gave me a particular thrill; it was the first time I had seen the Thermionic Valve in action.

Of the inventions of the twentieth century that of the valve seems to me to be the most important in the sense that it has had a greater effect upon the form of human life than any other. The valve, generator, detector and amplifier of high-frequency currents, made broadcasting practicable and broadcasting, be it of sound or vision, is of mighty consequence. I say this because I believe that broadcasting is the most powerful means of publication so far devised to influence the mind, taste and manners of mankind. I appreciate the counter-claim for the jet engine and the rocket; these assemblies could indeed be more influential in their capacity to destroy mankind, but I decline to match potential horrors against potential delights. Further, I recognize the importance of drugs-softening pain, subduing infections, restoring sanity-but I still maintain the claims of broadcasting as having a paramount influence upon communal psychology.
Still trying to match credit with those pioneers who deserve it can we find, among several, any one of them who could be allowed to say, in the face of fact, "I gave the world the thermionic valve"? With deep respect for the person who said it I maintain that neither he nor any one single person did so. The invention of the valve, like that of wireless itself, was too big to be borne of a single individual. But names are possible-in the time order, but not necessarily the order of importance of their contributions. I cite Edison, Fleming, de Forest, Langmuir and-a process not a person-the " getter."

Edison was unquestionably the first to arrange a plate near a filament and to explain unilateral conduction across the space between them (1883); Fleming, informed about the Edison effect and having at his disposal, on the shelf of a London University

[^2]laboratory, "Apparatus for Demonstrating the Edison Effect" probably used it, as a relatively stable rectifier which could be used to give a reasonably accurate measure of the value of highfrequency currents-hence, as a natural evolution, the Fleming diode (1904). Lee de Forest was certainly the one who first placed the third or grid electrode between filament and plate (1907), but the action of this grid was not very well explained or, perhaps because of the softness of the valve, was then inexplicable.

I have no precise evidence to support me but I believe that Langmuir first analysed the behaviour of the hard valve which the "getter" eventually got. In other words it was Langmuir who related $g_{m}, \mu$, and $r_{a}$, and showed the valve to be a voltageoperated device in which these three parameters played a co-ordinated role.

The foregoing, set out in such simple outline, may, by the degree of its generalizations, be unfair to those mentioned and neglectful of those not. If this be so then it is because of the difficulty of compressing into a few paragraphs what is a somewhat confusing and often unedifying story of protagonists upholding flimsy claims in terms of the polemic of vested interests rather than cool and factual analysis. Soft valves, soft thinking? But, with the tolerance of history, no hard words. Whatever may be the truth, the hard valve did appear out of a confused mist and proceeded, from 1914 onwards, to revolutionize the art and practice of electrical communication in all its forms.

It is surely fascinating to look back with wise-after-the-event eyes and watch the inventors of the past teetering on the edge of the obvious. A case in point concerns the use of the valve as a generator of oscillations; in other words the concept that, by positive feedback, the valve could be made to look like a negative resistance and thus overcome the losses in the resonant circuit it sets into oscillation. Lee de Forest described the triode in 1907 but does not claim regeneration until 1912. And simultaneously others, as we see from a famous legal action, had seen the same potentialities long after the appearance of the valve itself.

Thus "IN THE COURT OF APPEALS OF THE DISTRICT OF COLUMBIA . . . . Before Smyth, Chief Justice: Robb and Van Orsdel, Associate Justices . . . . This interference comes here on appeal by the parties Langmuir, de Forest and Meissner, from the decision of the Commis-
sioners of Patents awarding priority to Armstrong, also appeals by de Forest against Meissner and Langmuir jointly, and against Langmuir individually, for the invention set forth in the following counts"-and then ten or more thousand words in which other famous names came to the fore, notably Franklin and Round (England). A good deal of the evidence concerned "a beautiful clear tone" which de Forest claims to have produced from his Audion; it seems to have sounded as a syren voice in the ears of the legal pundits who gave judgment in de Forest's favour. Howard Armstrong, that prolific inventor, was much upset by a decision which was inclined to stress a somewhat loose description possessing a few months' priority against one which was far more concise if a little later in time.
While still thinking about these time-lags it is also strange to realize that we had to wait several years before we were given the immeasurable benefit of negative feedback, almost as important an invention in its influence upon the valve's ubiquity as the positive kind of feedback.
So much for genesis.
In spite of my presence at the Brooklands demonstration in 1915 I spent the greater part of the war in Egypt, Salonika and France looking after spark transmitters and crystal receivers. It was not until late in 1917 that I was appointed to do what was rather grandiloquently described as "Research," first for the Army at Woolwich and then for the Flying Corps at Biggin Hill. Here I came into intimate contact with the valve, its moods, potentialities, successes and failures.
Soon after the armistice I escaped out of a uniform (a "war to end all wars" had just been victoriously concluded, what point in remaining in military service?) and joined the Marconi Company. In a short time I became Head of the Experimental Section of the Designs Department, the laboratories being housed in an army hut in a field near the village of Writtle which is in turn near Chelmsford, where the Marconi factory was and is located. A claim to fame, before ever broadcasting came to increase it, lies with the fact that I played a considerable part designing both the first aircraft wireless telephone equipment, used extensively by Imperial Airways, also the Croydon ground station. Who remembers "Croydon calling" and its speech heard against a background hum as loud as any produced in his receiver by the most amateur of amateurs? The hum was purposive; it made tuning-in by the pilot all the easier.
Two important events accompanied my service at Writtle, first the setting up at Chelmsford of a powerful long-wave telephony transmitter and secondly the regular broadcasting service from " 2 Emma Toc, Writtle." I will not labour detail, the facts are well known and have been set out elsewhere; I would prefer rather to generalize than to indulge anecdotage. Suffice it to say that the initiative due to H. J. Round and W. T. Ditcham, in setting up the powerful long-wave telephony station at Chelmsford circa 1919 stimulated the wireless amateurs to petition for a regular broadcasting service, and that permission for this to be set up, on the limited scale of half an hour a week, resulted in the Writtle station and the Writtle programmes, a service which anticipated that started by the B.B.C., in November, 1922, by some eighteen months.

A diminishing number of wireless amateurs and others attracted to the hobby of building "wireless sets" will remember the programmes from Writtle as being frivolous, I would prefer the description "gay." Perhaps the more remarkable aspect of the Writtle transmissions was the staff that fostered them. This comprised in the order they joined me after I became Chief Engineer of the B.B.C., the late H. L. Kirke, C.B.E., sometime Head of the Research, and subsequently Assistant Chief Engineer; B. N. MacLarty now Engineer-in-Chief of the Marconi Company, the Hon. R. T. B. Wynn, C.B.E., now Chief-Engineer of the B.B.C., and Sir Noel Ashbridge who was for so long the Corporation's Chief Engineer and subsequently Director of Technical Services.

An accompanying group photograph recalls a collaboration which I dare to describe as unique.

In my first article I described how, when still a schoolboy, my subsequent career in wireless was largely determined by the tactile excitements of brass and ebonite: it was a similarly sensual experience which caused me to swerve from occupations concerned with the less romantic aspects of radio to one devoted to the service of broadcasting.

It must have been in the early autumn of 1922, before the formal creation of the B.B.C. in November of that year, when Station 2LO broadcast opera from Covent Garden. Up to the time when I was converted to a belief in broadcasting, the wireless telephone as such had to me done little more than intrigue my technical intellect, its applications were seemingly prosaic, while our Writtle broadcasts seemed to be no more than the aphrodisiac of a hobby ("keep your boys at home"). But the moment of revelation, the moment when I heard the opening bars of the opera, and was in two senses transported, then the potentialities of broadcasting were seen so vividly and so completely that thereafter all attempts to realize them have been to me tinged with disappointment.

The experience must be seen as mystical, as such I have unashamedly tried to describe it; its residue caused me, in prosaic contrast, to frame what I termed the B.B.C.'s technical policy; I still believe in it and I still believe it has not been fully implemented.
In sum it is my belief that "The Programme's the Thing" and that the mechanism which reveals it must be subservient to the art which creates it. To conclude from this that the policy so described does no more than demand realism in reproduction begs the question so long as the term realism is not defined. You do not have realism, as it sometimes is defined, when, for example, a single source of reproduction canalizes a widely diffused source of programme. A two-dimensional representation of a three dimensional subject, such as is seen in a painting cannot be said to demonstrate realism in one interpretation of the term. But the artist who knows his job knows how to make a virtue of necessity and uses the very limitations of a medium to make his art more realistic-in other words, to make the impact of his art upon the sensibilities of an audience more pronounced than realism, prosaically defined, could ever do.
Having said that the Programme is all important it might next be said, stressing the plural, that the Programmes are more so.
A hobby-horse cannot be ridden to death, since,

A notoble group, the members of which conducted the first regular broadcasting service, from station $2 M T$, Writtle, some eighteen montrs before the E.B.C. was formed. Left to right, standing, B. N. MacLarty, the !ate H. L. Kirke, R. T. e Wynn, H. J. Russell; seoted F. Bubb, N. Ashbridge, P. P. Eckersley, E. H. Trump and Miss B. Beeson.
lacking a rider, it is already dead. I shall now attempt, in a brief spell, to resuscitate my old nag. To do so demands an explanation why I believe that this stressing of the plural of Programme is so important. I was, I am and I believe I always will be convinced that the technical method by which broadcast programmes are distributed pays greater respect to the art it serves as, within reason, the number of different programmes it offers, simultaneously, for the individual's choice is the greater. This conviction determined me, after I had left the B.B.C. (1929), to do all I could to proselytize and develop rediffusion, i.e., schemes whereby programmes are distributed through wire networks rather than by radio. Need I stress the limitation of radio in being very spare of channels in the frequency bands available, while relatively the wire suffers no such restriction?

During the late twenties and early thirties the Post Office, the B.B.C. and the Radio Trade more or less openly opposed the development of rediffusion; in spite of so formidable a combination it grew, and when given a chance, goes on growing. This fact reaffirms an unshakable conviction that a majority want a reasonably wide choice between different kinds of clearly produced programmes.

This proposition might well have been denied when, in the early days, the passion for home-building radio receivers was at its height. Many of my readers must remember those delightful times when they would hear one of the cognoscenti boasting his home-built set and how he "received Zloik (station in Czechoslovakia, old man) on my Superwoppodyne; 'phones were lying on the kitchen table and I heard the station quite clearly, while I was upstairs changing my shirt." Well, "it was swell while it lasted"; it was that rarity a hobby that produced a full-scale manifestation. In contrast you built a model steam engine and the consummation was a smell of meths. and a jerky puffing concealed in a pale mist-how different from Zloik "clear as a bell and no fading, old man."

The hobby died, the public bought the superhet,
the programme was the thing and this gave rediffusion its opportunity.

I cannot refrain from taking this opportunity to air a grievance. Briefly it is that when there actually was a means to prevent the extension of rediffusion the vested interests made full use of it. My friend and colleague Rupert Carpenter and I devised a system whereby four to six programmes could be sent through the electric mains to householders who, by the movement of a switch, could select any one of them. There happens to be an Act of Parliament, dated 1882, which forbids the electricity authorities to use their wires "for the purpose of sending a telegram." After a demonstration of the practicability of our method it seemed to certain vested interests and their Parliamentary sympathizers that this act was hardly less important than Habeas Corpus. And so, in the "land of opportunity" (see Press) we were forbidden to prove how right our opponents were when they said, as they did, that our scheme would not work and that if it did it would introduce a "dangerous new principle."

The issue requires little elaboration, the proposals we made about it, also about the wider implications of wire-broadcasting received either contemptuous dismissal by Government committees and commissions of enquiry or combative assertions about technical method; funds were even raised to oppose the passage of a bill through Parliament revoking the ancient statute; today, with the impact of television the issue about the method itself is dead but not, I trust, the implications of the story.
I shall not say much here about the more important phases in the development of sound broadcasting, its rejuvenation by v.h.f., the introduction of the "Third" (the most notable and admirable of the B.B.C.'s innovations), automatic monitoring, the overseas service and its intricacies of switching programmes, so admirably conceived and executed and, above all, television. I excuse this unbalance by remarking that it is all recorded elsewhere whereas in like degree heterodox opinion, which becomes me better, has not. I could not, however, even begin
to excuse the dismissal of television and so, since it belongs more to Time Future than to Time Past I will have a good deal to say about in in my next and last article.
While I maintain that the most remarkable outcome of the invention of the valve is the broadcasting service, obviously parallel developments are nearly as important.

It was said in my first article that, broadly speaking, the first decade of the development of wireless proved it, as a means to link stations separated by world distances, a comparative failure. It was of course the valve that raised the status of wireless as a world communicator so that it became, under the aegis of private enterprise, a competitor with the under-sea intercontinental cable. In order that private enterprise should not become indecently enterprising the Establishment decided to synthesize thesis and antithesis and so brought the Public Utility "Cables and Wireless" into being.

The invention that brought about the merger was the Marconi beam system whereby it was proved that radio was capable of penetrating to distances of the order of $\pi$ times the radius of the globe; a globe assumedly well wrapped in an ionized blanket. It was his ability to see the wood, without confusion of trees, that, just after the conclusion of the first war, made Marconi say, "Now that we have the valve why don't we try short waves again?" The sentence implies an appreciation of the signal-to-noise ratio; the higher the frequency of the signal the less the noise. On the other hand the shorter the wave the greater the overland attenuation. Before the aerial currents could be amplified, short-wave ranges were limited to very short distances; once the more feeble but less interrupted signals could be amplified the overall gain was, to say the most, fantastic.

I have never been able to find out whether and if so in what degree Marconi was driven to follow his hunch, so neatly stated, by the mass observation of amateurs; it is surely fitting to remember that these keen experimenters, driven away from the mediumwave gamut, did prove, by their skill and patience, that, on the lowest terms, an investigation of the commercial potentialities of short waves was well worth making. We may also note that the less imaginative authorities were still tied to lower frequency and higher power and still higher aerials.

In commending the beam system we should pay tribute to the genius of C. S. Franklin who designed the transmitters, aerials and receivers, and to T. L. Eckersley whose original work on the physical properties of the ionosphere made it possible to match optimum frequency with world paths and diurnal times.

I am drawn, in these concluding paragraphs to hover on the edge of prophecy; a giddy state and therefore exciting. Boldly stated, it is that point-topoint communication over ocean distances will eventually and as to the greater part be made by cable, while overland communication, as to a considerable part, will be consummated by radio. I stress ocean distances meaning communication where oceans get in the way; there is no need to define the term overland, it means over land. I stress also that it is point-to-point communication that is in question; obviously radio is the only viable method of communication for mobile services.
To expand this thesis we see already how the telephone and telegraph service between, in effect, Europe
and North America has been improved by substituting cable for radio. We learn of plans to bridge the Pacific and link the Commonwealth by supplementing this ocean cable by another spanning the Indian Ocean. We know that when the transistor and its associated components become more reliable that great benefits will be conveyed to the ocean cables. We can foresee intercontinental television exchanges (which are almost impossibly expensive when relying upon a radio link) becoming an everyday occurrence when there are sufficient cables to carry them.

As to overland communication by radio there are in operation today many systems, the wave frequencies climbing into and above the four thousand megacycle landmark. Line-of-sight transmission, from hill-top to hill-top, invites shorter and shorter waves and as the systems develop and according to the nature of the terrain we may see them, as we do today in various parts of the world, supplanting the coaxial and multi-quad cables by means of which many hundreds of messages are sent simultaneously through the same link. There seems to be, at first blush, something of a paradox if radio, hitherto used as an ocean bridge, should give up its role in this respect and take on another where, superficially speaking, the physical conductor would seem the obvious link. For overland communication the paradoxical aspects may fade, as does radio, when it is seen that signalling by refracted and reflected waves has a hazardous aspect when compared with a conductor which guides the waves to their destinations.

It may seem as if this adumbration of possible future developments is outside the terms of reference of an article headed Time Past. But no! Time past has seen the wide use of radio for point-to-point communication over land as it has also seen the introduction of the telephone cable bridging the Atlantic; the facts are there; all that has been done, in a time future category, is to postulate the continuation of a tendency already manifest in Time Past.

But "Amarath an Amarath succeeds"; a new form of multi-electrode amplifier-the transistor-is already taking up the work begun by its forbear the thermionic valve.

I end as I began by stressing the importance of the invention of the valve without which the Time Future of telecommunication, which I hope to glance at in my next article, would not, in all probability, be worth writing about.

## Television Society's Exhibition

AS in past years, the keynote of the Television Society's Exhibition, which opens at the Royal Hotel, Woburn Place, London, W.C.1, on March 3rd will be television research rather than domestic reception. The three-day exhibition opens at 11.30 on the first day and at noon on the following two days. The respective closing times are 8,0, 8.0 and 7.0. At the time of going to press the following had taken space at the exhibition:-

British Communications \& Electronics
Chapman \& Hall
Cossor
E.M.1. Electronics Ever Ready

Hallam, Sleigh \& Cheston
Livingston Laboratories
Mullard
C. H. Nokes

Standard Insulator Co.
20th Century Electronics
N. E. B. Wolters
N. E. B. Wolters

John Ware
Rank Cintel
Rank Cintel
Wireless World
Wireless World and Radio \& Elec-
tronic Engineer
Admission is by ticket obtainable free from the Society at 166 Shaftesbury Avenue, London, W.C.2.

# LETTERS TO THE EDITOR 

The Editor does not necessarily endorse the opinions expressed by his correspondents

## Stereophonic Records

HAVING just returned from a brief visit to America, where stereo rules the roost, I read your February Editorial with much interest. (Over there it is almost impossible to sell any new higk-grade mono equipment, and the mere sight of a new mono record, however good, scares the pants off record dealers.)

Mr. Haddy, of the Decca Studios, has now convinced me by demonstration that the sound from disc stereo can be as good as the same recording heard direct from $15-\mathrm{in} / \mathrm{sec}$ tape, provided the pickup is of adequate quality. The stylus should maintain contact with the groove walls at a playing weight of 3-4 grammes, with no audible needle chatter with the ear three or four feet away. As the Decca pickup fulfils these requirements and is now available through the trade, and other good stereo models are being produced, my previous complaint about inaccessible high-quality units no longer holds.

This leaves us with the great question: How does the best stereo compare with the best single-channel recording? The idea to have direct comparison, put forward in your Editorial, is excellent and should also be applied to pickups. With the co-operation of E.M.I. and Decca we hope to do something on these lines at our Royal Festival Hall Demonstration on the 9th May. If really omni-directional speakers are used I feel sure that a worth-while comparison can be made, even in the R.F.H. The big idea is to play some original passage of music through the same amplifiers and speakers with three different types of input as follows:
(1) Single-channel record with high-quality mono pickup.
(2) Same as No. 1, but with the best available stereo pickup connected for mono output.
(3) Stereo record with same pickup as used for No. 2, but connected for two-channel output.

Even if we prove nothing, it should be good, clean fun.
Idle, Bradford.
G. A. BRIGGS,

Wharfedale Wireless Works Ltd.

## Stereophonic Sound

I SUGGEST that the mode of sound reproduction, commonly known as " stereo", will only prove technically satisfactory when some manufacturers realize that the "ting" of a cash register is not the only sound that requires reproducing.

Southam.pton.
JULIAN GARDNER

## Stereophonic Standards

ONE of the great dangers of a rapid advance in the commercial application of a scientific discovery is that of ill-conceived standards. We are now, for instance, reconsidering our television system, and have recently asked ourselves why we chose odd standards for tape speeds. Now that stereophonic sound is a commercial proposition, we ought surely to start considering very carefully the standards we are going to accept, such as loudspeaker spacing, cross-talk levels and so on. It would be interesting to hear, for example, why 6 feet has been chosen by some as a loudsoeaker spacing. One can hardly imagine the most tolerant and adoring wife accepting two reflex cabinets or column radiators placed 6 feet apart in the living room or the lounge. Could we not have 12 feet? Practically every living room in the country must be very close to this dimension in either length or breadth. In this case they could be
placed in two of the corners of the room, which should be more compatible with the domestic situation. It would be interesting, therefore, to hear the experts declare whether such an arrangement would be acceptable from à technical standpoint.

Sevenoaks.
J. R. OGILVIE

## Rigidity of L.S. Diaphragms

MR. BARLOW'S comment in the February issue on my letter intrigued me very much. I was very interested to find he had had similar results to mine-that a rigid (well, somewhat rigid) diaphragm didn't obey the rules. I have been brooding on this problem for some time and one or two thoughts occur to me.

In the matter of extended frequency response, as compared with a paper cone, there are two possible theories, but I wouldn't be prepared to say that either was a good one. Experimental proof is so very difficult. The more credible one seems to me to be this: it is common knowledge that liquids and solids conduct sound more efficiently than gases, and a hard solid is a better conductor than a soft one. One would not use a rolled-up newspaper as a car engine "stethoscope", but a metal rod behaves very well in locating the source of engine noises. I suggest that in a hard synthetic resin diaphragm there is good transmission of high frequencies from the cone apex throughout the main body of the diaphragm which results in a greater area for transferring the sound waves to the air, whereas with the customary felted paper cone this transmission does not occur owing to the (literally) absorbent nature of the material; hence the only part of the cone which does propagate the sound waves is the apex.
The alternative theory is that the hard cone material "rings" and produces spurious extreme treble which is not detected by rough-and-ready frequency response measurements but which might be detected (as was pointed out to me by Mr. Voigt) by applying the output of the measuring microphone to a sensitive oscilloscope. Unfortunately this test will not work with comparatively small inputs to the speaker since it is almost impossible to detect small departures from the sine wave on the 'scope.
This I can say: the old pre-war Hartley-Turner speaker had a phenolic resin cone, and its chief defect was excessive output in the 5 to $6 \mathrm{kc} / \mathrm{s}$. region. When I thought I had finally solved the problem of a synthesized resin cone I found it did exactly the same thing. The top sounded wonderful, but it wasn't the real thing, and the effect was only removed when I modified the formulation to reduce the hardness of the resin and give it (to use an ad. man's phrase) "built-in darnping."

This leads to a note on the matter of ordinary damp, which puzzles Mr. Barlow. The cones which were developed in the U.S. were designed to be as hard and stiff as possible. This involved using a very hard filler in the resin formulation. The filler was hygroscopic but I thought that perfect sealing off by the resin would make the cone waterproof, since the resin used was supposed to have negligible water absorption. Not a bit of it! I left the cones in a damp atmosphere for a month, and they turned limp. On heating them I could see steam coming off, and then the cones were hard again. I do not believe that any syrthetic resin in comparatively thin films can reject moisture. The figures quoted by manufacturers are based on the immersion of quite thick pieces of resin and, in the case of polystyrene for example, I feel pretty certain that the figure
of $2.8 \%$ quoted by Mr. Barlow is almost a "skin effect". If the major part of the resin used in the cone is a skin then the actual absorption would be quite a high percentage.

If so, then Mr. Barlow's thick sandwich is basically a very good idea, for he seals off the resin on both sides, and doubtless could seal the edges too. But the size/ weight figures he gives should, I think, be improved if good reble response is to be maintained. He quotes 15 gm for a $7 \frac{1}{2}$-in disc; I stuck to a conical shape as being, in my opinion, the strongest shape, and I finally managed to get down to 11 gm for an $8 \frac{1}{2}$-in cone. A felted paper cone of the same size and shape comes out at $5-6 \mathrm{gm}$, and the resin cone had more real top than the paper one. But if I increased the weight I am sure that the "roll-off" would be fairly severe. One sample had a weight of 22 gm and had the same treble response as the paper cone, but, of course, far better bass.

My work, and Mr. Barlow's, clearly indicates that, whilst we may feel pretty certain that felted paper's sole advantage is ease of production, change to another material which denies accepted design dogma involves basic research which is not very easy to carry out. There are so many possible permutations and combinations of materials.

Exton, Southampton. H. A. HARTLEY.

## Miller Sweep Circuit

THE Miller sweep circuit described by C. S. Speight in your January issue was first developed some three years ago by J. D. Julian and myself. It may be of interest that the original object was to improve by Miller feedback, the linearity of saw-tooth obtained from the multivibrator with its excellent synchronizing properties.
The multivibrator circuit employed is shown in the accompanying diagram, and is due to H. E. Anthony*.


The timing capacitor $C$ is in the cathode circuit of V2. Because of the low impedance of V2, C will charge rapidly when V2 grid is switched "on" by the multi"vibrator action. During the period when V2 grid is "off," no current will flow through V2 and C will discharge logarithmically through $R$. The essential feature of the improved circuit is the substitution for R of a Miller valve.
The circuit is capable of producing more than 200 volts of almost perfectly linear saw-tooth with 350 volts h.t. supply. This can be maintained, with careful design, up to $500 \mathrm{kc} / \mathrm{s}$ sweep recurrence. The sweep amplitude can be easily stabilized, permitting accurate time calibration; in addition all the advantages of easy synchronization are retained.

It is not clear why Mr. Speight refers to the circuit

[^3]as combining the Miller and Puckle circuits, as surely the essential ingredient of Puckle's timebase is the use of a constant-current pentode as a linear charging device. With the Miller-Multivibrator, as I prefer to designate it, it is the run-down which is linear. The constant-current anode characteristic of the pentode is not employed at all.

Bournemouth.

## L. FREEMAN,

Waveforms, Ltd.
The author replies:
A timebase is generally referred to by a name which suggests either a special method of generating a linear sweep (e.g., Miller bootstrap) or a special method of recharging the sweep capacitor (e.g., thyratron, griddiode). In my view, the Puckle circuit is essentially a recharging device, in which a hard valve trigger circuit replaces the obsolete thyratron circuit.
F. J. M. Farley* describes the Puckle timebase with a resistor in place of the more usual constant-current pentode, and the arrangement then resembles that to which Mr. Freeman refers.

It is worth noting here that the factors influencing the operation of the Puckle circuit are quite different from those of the standard multivibrator and the only real similarity between the two circuits is in the method of applying positive feedback.

In reply to the letter of Mr. J. D. Julian, published in the February issue, I should like to make the following comments.

Apparently, Mr. Julian has overlooked the main difficulty, encountered in combining the Miller and Puckle circuits, which is to ensure that the current drawn by the Miller valve during flyback does not subtract substantially from the recharging current.

The situation is aggravated by the feedback action of the sweep capacitor tending to drive the Miller valve into heavy conduction. Ideally, this valve should be cut off, during flyback, but in practice it is more convenient to limit the anode current to a fraction of the recharge current. In my circuit D1 and $R_{8}$ perform this function, and Mr. Julian is incorrect in saying: "D1, it will be appreciated, is not essential to the operation of the circuit and in the interests of economy can well be left out, together with $\mathrm{R}_{8}$." This statement appears to be based on the assumption that these components serve merely to protect the valve against overload. I would suggest that equivalent components might be inserted into the circuit proposed by Mr . Julian, for, as the circuit stands, it is difficult to see how the sweep capacitor can recharge at all.

I cannot agree that synchronization and triggering efficiency can be improved by applying the pulses concerned to V3 cathode. Synchronization is most effective when it is applied to the initiation of the flyback, rather than to the initiation of the sweep, particularly when the ratio of sweep to flyback periods is as high as that of the circuit described. Another point in favour of the original method of sync. injection was mentioned in the article, and this arises from the action of V2(b) in amplifying the sync. pulses before they become effective.

The statement of Mr . Julian concerning capacitive loading of V2 is irrelevant. The resistor $R_{1}$ is introduced to avoid this effect in the free-running circuit, whilst the diode D2 performs a similar function in the triggered version. The diode will cut off, should the rate of rise of voltage at V2(a) anode exceed that at the trigger input, thus isolating this point from stray capacity at the input. When the trigger pulse has a fast rising front, this will be transmitted through D2, contributing directly to the c.r.t. "bright-up" pulse.

Automatic "gate-out" of negative pulses arriving during the run-down is achieved by means of D2, whilst the grid-cathode diode of V2(b) reduces the amplitude of positive pulses to negligible proportions.

The term " initial fast region" was used deliberately

[^4]to suggest the effect of the Miller step on the c.r.t. display. I can see no point in trying to connect this indeterminate arithmetical error with the formal definition of non-linearity. I agree, however, that more should have been said about the Miller step, especially concerning its dependence on sweep speed.

The advantages of the circuit, which I overlooked, unfortunately do not include: "Constant amplitude of run-down, unaffected by sync." As was originally pointed out, the lower limit of sweep depends on the anode voltage of V2(b) during run-down, and this will vary with sync. amplitude.

Mr. Julian states that I overlooked mentioning "the superior triggering or synchronizing ability" of my circuit. It would, therefore, seem that the circuit performs satisfactorily in this respect and no modification is required. This is, in fact, the case.

It would be interesting to compare the commercial circuits referred to by Mr. Julian, with that which I submitted. The fact that these have some properties in common confirms my original opinion, that a timebase can be devised, which is no more complex than the standard Puckle circuit, yet compares with those presently available in commercial equipment.

It was thought that a description of such a circuit might be of interest to readers, but I must apologise to anyone who has been concerned with this type of circuit for not acknowledging the work done in this field.

Southend-on-Sea.
C. S. SPEIGHT

## Printed Circuils

IN his reply to my letter (January issue) on printed circuits Mr. W. I. Flack accurately states the case from the manufacturers' point of view, and in so doing confirms my contention that most set makers are very much out of touch with conditions prevailing in the dealers. service department. Regarding his doubt as to a $24-$ hour service, I can assure him that although I do not regard our service department as being in anyway exceptional, the majority of our repairs are completed within 24 hours. Indeed, on the day that my Wireless World arrived and I read Mr. Flack's letter I looked at our job book and found that three TVs reported faulty in the morning had been collected, repaired and returned to the owners by 6 p.m. It will no doubt surprise Mr. Flack and other manufacturers to know that this is the sort of service which the customer expects and which I think he is entitled to receive. On the rare occasions when we have had to return a receiver to the makers for repair we never see it again for at least ten days. I am afraid most manufacturers do not know the meaning of the word "service".
Contrary to popular belief the average dealer's workshop does not consist of several fully qualified and whitecoated engineers surrounded with wobbulators and signal generators, busily aligning i.f. strips and making expert repairs to printed circuit panels. If it did, Mr. Flack would be quite right. It would take just as long to repair a "steam wired" set as a printed circuit set. After 30 years' experience I have found that the most efficient dealer service should be run by a few "Old Hands" with the necessary "know-how" and experience to diagnose faults quickly and accurately, (generally aided by nothing more complicated than a valve voltmeter). The faulty set is then passed to a less experienced engineer who executes the necessary, repair. Under these conditions the "steam-wired" receiver wins hands down for accessibility and general speed of service.

Is Mr . Flack really serious when he suggests that dealers should stock a complete set of printed circuits for every receiver on the market? At the present rate of progress most units would be obsolete within a year. A good idea from the manufacturers' point of view but as a dealer I will not trust myself to say more on the subject.

I am sure Mr. Flack will agree that when we reach the heart of the matter in this controversy, it doesn't really matter what he, as a manufacturer, or I, as a dealer, prefer. It is the customet that counts, and if he wants his faulty TV back in time to see "Emergency Ward 10" or the Cup Final, there's not much doubt which sort of service he would prefer!

In my original letter I did not question the reliability of the printed circuit, but since Mr. Flack has raised the issue by stating that they are superior in this respect I should like to point out that it is rather early days for such an assumption. I may be wrong but it occurs to me that when some of these printed-circuit receivers have been in use for five or six years, possibly under conditions of extreme condensation, etc., they may not be looking to well. Perhaps we shall then find that-as in the case of push buttons, black screens, plastic c.r.t. masks, etc.,-all new ideas are not necessarily good ideas.
It is unfortunately true that more and more manufacturers are falling for printed circuits but I think that the few who are sticking to their "steam-wired" designs will reap the benefit of bigger sales in the long run. I have found generally that sets that appeal to the service department are good sellers and vice-versa, but only time will tell!
London, N.W.6.
A. G. TUCKER.

YOUR correspondent, Mr. Flack (February issue), in defence of printed wiring (quite distinct from printed circuits) advances the argument that his technicians employed on production test, like, or even prefer, to work on these panels as against conventionally constructed chassis.

I should like to enquire if Mr. Flack would prefer to fault-find in an unfamiliar conventional chassis where all sleeving is of the same colour, or in one which employed the recognized sensible variety of coloured wiring? Since printed wiring falls into the first category, he would appear to prefer the former.

The second disadvantage is, that since wiring and components are on opposite sides of the printed panel, the eyes are required to move constantly from one side to the other, resulting in both eye and mental fatigue.

Thirdly, looking at the component side, this holds a disjointed array of resistors and capacitors fixed apparently at random and one is unable to see the interconnections. This is not the case with a well-designed conventional assembly.

Fourthly, the experienced and skilled technician is used to looking at valveholders and coils frob below, and by seeing all the wiring attached to these components in a well-defined and logical sequence, he makes his measurements without hesitation, just as Mr. Flack would, no doubt, unhesitatingly spell out the word Czechoslovakia. He suggests, however, that it is just as easy, or even easier, to spell the name of that country backwards.

I am well aware that to print and to read in the fashion we, in fact, do is merely an acquired habit. But I hope Mr. Flack will agree with me that it would not be a particularly good idea to print alternate words on the front and back of pages. It would take longer to read and be more strenuous if this method were, in fact, adopted. Although sooner or later one would, of course, get used to it and live. with it.

To say, therefore, that printed wiring offers any advantage other than perhaps to reduce the manufacturing costs is grossly misleading. Its general adoption will certainly increase maintenance and repair costs. On balance, therefore, the set owners will not gain, but the engineers' span of life is certain to become shorter.
London, N.W.6.
E. KISCH

MR. W. FLACK'S remarks in last month's issue are not altogether complimentary to service engineers in general. He states that the apparent dislike engineers
have for printed circuits is because servicing of such equipment calls for a good deal more care and technical skill, whereas with conventional wired circuits, hit-andmiss methods are the order of the day. Anyone who has had experience in servicing printed circuitry is fully aware of the care necessary in handling it, and has, no doubt, discovered that the application of a soldering iron can wreak havoc if not handled very carefully. If this means technical skill as interpreted by Mr. Flack, all good and well. But he goes on to say that there is less room for hit-and-miss methods in servicing printed circuitry than in conventional wiring circuits. I can only assume that this statement applies to "pirates" and the like, who are, unfortunately, to be found in every trade or profession. With a thorough knowledge of basic theory, combined with a logical approach, servicing of both types of circuits should not present undue hazards.

A final point on the cost factor of the printed versus wiring circuits. If the former is supposed to be much cheaper to manufacture, to whom has the saving been passed?

Sevenoaks.
A. W. WESLEY-COLLINS

## What Makes Currents Flow?

"CATHODE RAY'S" article in the January issue concerning the causes of current flow led me to read the monograph by P. Hammond to which he refers, and I am prompted to make one or two general comments on the subject.
First, let me say that I support Hammond in his view that the current in a simple battery and resistance circuit is caused by charge distributions, whilst agreeing with "Cathode Ray" that the prime cause may be chemical action or even the act of assembling the battery. The term e.m.f. is one which we employ to describe one manifestation of a particularly complex process of electro-chemical action, charge distribution and charge motion, just as we use the term eleatric field to conceal our ignorance of the true laws of force between charges.
If a straight conductor is positioned between two terminals having a potential difference but without connecting to them, the charges in the conductor must redistribute themselves until the resultant electric field is everywhere zero; but one might hesitate before committing oneself to saying just what form this redistribution will take. For instance, since the applied field is apparently being set up by two equal and opposite conzentrations of charge at the terminals, one suitable charge configuration in the conductor should be with similar but opposite concentrations at its very ends. Is this the only suitable configuration or are there others; and if there are which one is correct?
The answers to these questions are by no means obvious, even in this straighforward case of a straight conductor carrying no current, but if current is permitted to flow by connecting the conductor to the terminals and the conductor is no longer assumed straight the problem is complex indeed and I do not pretend to know the carrect solution.
"Cathode Ray's" trump card, the closed ring surrounding a varying magnetic flux is, however, fairly easy to explain, at least in terms of the concepts of induced e.m.f. and lumped resistance. The effect of transformer induction, of which this is an example, is basically due, as E. G. Cullwick ${ }^{\star}$ points out, to the fact that charges in motion (namely those charges which constitute the current system setting up the magnetic flux) exert forces on stationary charges as a result of this motion. These forces, of whose true nature we are quite ignorant, may be regarded as producing an electric field in the closed ring and this, over a finite length of ring, is equivalent to an e.m.f. tending to drive current through the resistance of the ring.

However, this e.m.f. must not be regarded as concentrated at one point in the circumference of the ring but, like the resistance, as uniformly distributed around
it. The figure shows an equivalent circuit containing, say $n$ sections each comprising a small e.m.f. $\delta \mathrm{E}$ in series with a small resistance $\delta$ R. The current will be $\mathrm{I}=n \mathrm{E}$ / $n \delta \mathrm{R}=\delta \mathrm{E} / \delta \mathrm{R}$ and the voltage drop from $A$ to $B$ is $\mathrm{I} \delta \mathrm{R}-\delta \mathrm{E}=0$ and similarly between any two such points of the circumference. If one imagines $\delta \mathrm{E}$
 and $\delta \mathrm{R}$ reduced to infinitesimal proportions the reason why no voltage drop can be measured around the ring is at once apparent.
This manner of explanation should satisfy those who prefer to tackle this kind of problem in terms of lumped parameters, but I do not claim that it will bear close inspection from the viewpoint of charge distributions and electric fields. In general it is obviously futile to try to explain fundamental mechanisms in terms of those very concepts which have been adopted purposely to avoid having to explain them.

Cranfield.
G. H. STEARMAN.
*."The Fundamentals of Electromagnetism", (Cambridge Univ.
Press), page 87 . Press), page 87.
The author replies:
The essential words in Mr. Stearman's letter are presumably. "in a simple battery and resistance circuit." For having voted in favour of the view that in such a circuit the current is caused by charge redistribution, he refers to examples of (1) a charge redistribution with no current, and (2) a current with no charge redistribu-
tion. tion.

As if this were not enough to underline my doubts about regarding charge redistribution and current as cause and effect, he goes on to emphasize how difficult it is to calculate the charge redistribution in (1) and how easy it is to handle (2)* on the conventional e.m.f. basis.

So he encourages me to go farther than I did in the January issue, by saying that if charge redistribution were to be substituted for e.m.f. as the cause of electric currents it would transform electrical technology into an unteachable mystery.
"CATHODE RAY."

* We are both taking for granted that on this level of electrical engineering we don'l venture into atomic physics.


## Licence Reminders

IN his letter published in the December issue, Mr. W. R. Gregory complained that he was unable to renew his wireless licence the day before it expired on 31 st August because he could not produce a reminder notice.
I should like to assure Mr. Gregory that there is no question of the issue of a renewal licence being conditional on the production of the relative reminder notice and there is no reason why he should not have been allowed to renew his licence a day in advance as he wished. I can only assume that the counter clerk who refused to issue a licence was acting under a misunderstanding and I must apologise on behalf of the Post Office for the inconvenience caused.
London, E.C.1.
T. A. O'BRIEN,

Public Relations Officer,
General Post Office.

## Transistor Tape Pre-Amplifier

IN Mr. Ridler's article in the December, 1958, issue he chooses the inductance-resistance integrator in preference to the resistance-capacitance integrator as a means of providing the bass-lift for a tape pre-amplifier. He makes this decision on the ground that a high input impedance implies high thermal noise. This is not correct, since the input is shunted by the source which is an inductance and,
(Continued on page 125)
in so far as it is pure, no thermal noise exists in it. The argument based on electrostatic pick-up also falls down for the same reason.

This would not matter if the alternative chosen by Mr . Ridler were as good, but this unfortunately is not the case. Even with a perfect integrating amplifier, the resistance included in the source limits the bass correction severely at the frequency at which the $Q$ of the head falls to unity. Taking the $\frac{1}{2}$-henry head in the article and associating it with the probable value of 400 ohms d.c. resistance, the frequency of 3 dB loss is $127 \mathrm{c} / \mathrm{s}$ and the loss at $50 \mathrm{c} / \mathrm{s}$ is over 11 dB . It was this consideration which caused me to reject it as a practical possibility, when considering the problem in 1957. It does not help if the amplifier is inexpensive if the heads have to be of enormously high $Q$ to operate it; the high impedance alternative is more complex, but at least it will work with any head up to an inductance of say one henry, without the risk of leaking electrolytics polarizing the head and, with exact corrections independent of the $Q$.

The explanation of the failure of Mr. Ridler's curves to show the real loss is that the test tape does not have continuous gliding tone and the number of fixed points is too low to draw conclusions from. The fact that the inevitable rippling in the head's response in the bass region due to the outer gap effect is also not shown must be due to the same cause. The curves for the amplifier and head inductance can be obtained by using, as a signal source, a constant current into a mutual inductance, say one milli henry, and injecting this in series with the head. The added inductance is small compared with the inductance of the head itself and has no effect on the response curve, which should now be flat up to the chosen turnover frequency. In this way it is much easier to see the departure from the strict law, than if a constant voltage input is used.
My second point concerns the bias loop circuit to which Mr . Ridler draws attention in his reference 3. (E.R.E., May, 1957, p. 161) and the essential distinction between that circuit and Mr. Ridler's Fig. 4. In this there is no d.c. resistance included in the first emitter circuit. The omission of this resistance results in an entirely different mode of operation. In this mode the current in the second transistor depends on the baseemitter drop of the first transistor and on the base current of the first transistor, which in turn depends on the alpha gain of the first transistor. As the base-emitter drop of the first transistor is scarcely affected by the battery voltage, there is no term which relates the current in the second transistor to that voltage and the circuit of his Fig. 4 fails when the battery voltage falls since the collector of the second transistor can bottom. In the circuit described in ref. 3 the current in both transistors is practically independent of the base-emitter drops and of the base currents, and is a function of the product of some fixed ratios and the battery voltage. This is due to the swamping of the relatively small collector-emitter voltage of the first transistor, by the drop in the emitter resistance, which may be ten or fifty times as great, and thus will determine the base potential of the second transistor with great precision.

Mr. Ridler is fully justified in claiming the low noise output of his amplifier and it is a pity that such an elegant solution as his Fig. 4 would have been, cannot form the basis of a tape characteristic corrector of the highest class. London, N.W. 2 .
J. SOMERSET MURRAY

## The author replies:

I am afraid that I must still hold to my original contention that the thermal noise in the L-R integrator is lower than in the R-C circuit. The basic equations for the two types are

$$
\begin{align*}
& \mathrm{E}_{\text {out } t}=-\frac{1}{\mathrm{RC}} \int e_{i n} \mathrm{~d} t  \tag{RC}\\
& \mathrm{E}_{\text {out }}=-\frac{\mathrm{L}}{\mathrm{R}^{1}} \int e_{\text {in }} \mathrm{d} t \tag{LR}
\end{align*}
$$

and assuming that the circuits are so adjusted that each
gives the same gain at a particular frequency, they will have identical signal performances. However, considering the input circuits

the equivalent noise voltage
in the R-C case is $e_{n}=\sqrt{4 k T \Delta f(R+r)}$
and
in the L-R case $e_{n}=\sqrt{4 k T \Delta f r}$
neglecting in each case the input impedance of the amplifier, which will be very low due to feedback. As R must be larger than the reactance of L at the highest frequency, then the ratio of the signal to noise ratios from this cause will be nearly $\sqrt{\mathrm{R} / r}$.

There is also a much more serious effect. Unless the first transistor is fed from nearly the optimum source resistance, the semiconductor noise will be far in excess of the thermal noise in the source resistance. With the R-C circuit, the first transistor sees a resistance $R$ in parallel with a capacitor GC, where G is the gain of the amplifier. The resistive component of this will be much greater than optimum at low frequencies, but improve at higher frequencies. In the L-R case the transistor sees the impedance of the head in parallel with a resistance R'G. This is much lower than the optimum at low frequencies and tends to the correct value at high frequencies. Probably both circuits are equally good in this respect. I am presently engaged in a detailed analysis of the problem.
I see that I have committed a grievous error in not giving particulars of the tape head used, as this has led J. S. M. to a false conclusion. The head is one from a Collaro Tape Transcriptor and has a d.c. resistance of 50 ohms; this, of course, gives an equalization loss of -3 dB at $16 \mathrm{c} / \mathrm{s}$. Low-frequency variations in output were observed, but were neglected in the curves as they tended to obscure the main issue in an elementary treatment. I am most grateful to J. S. M. for pointing out the technique of injecting an e.m.f. via a mutual inductance, as this does simplify measurements considerably.

With regard to J. S. M.'s second point, I am afraid that I do not have a copy of his paper on hand as our copy here has gone to the binder, but I am sure that what he says is correct. However, the circuit as described has been functioning for fifteen monchs, and two versions have been tried using transistors of different characteristics but all other components the same. The range of ambient temperature here is extreme; winter minimum is about $50^{\circ} \mathrm{F}$ and summer maximum about $110^{\circ}$ in the house. Another $15^{\circ}$ could probably be added for the additional rise inside the apparatus cabinet. The same batteries have been used over this period. The drain is only 1.5 mA but due to the heat the voltage has fallen from 12 V to 8.5 V without noticeably affecting the results as judged audibly through a high-quality loudspeaker system. I think that although I have apparently misused J. S. M.'s citcuit the results are satisfactory.

Khartoum, Sudan.
PHILIP F. RIDLER.

# Physical Society's Exhibition 

NEW TECHNIQUES IN ELECTRONICS AND MEASUREMENT

ASLIGHT but nevertheless welcome reversion to its old character was noticed in this year's Physical Society's Exhibition. More of the exhibits seemed to be devoted to the results of research and fewer to the "bread-and-butter" sort of developments. Indeed some techniques were openly admitted to be quite impractical and were only shown because of their interest value. Unfortunately this trend was also accompanied by one of the old troubles of the earlier exhibitions-lack of space between the stands for visitors to circulate freely and see things in comfort (either that, or too many tickets were issued). There is a great need for improvement here. In the following pages we have made a selection of items which we think will be of particular interest to our readers.

MASERS rely for their operation on the release of energy stored in molecules elevated to a higher-thannormal energy level. In the threelevel paramagnetic maser, shown by the Royal Radar Establishment, molecules of a paramagnetic substance placed in a magnetic field are elevated from the first energy state to
the third by the absorption of a locally generated "pump" signal (which is higher in frequency than the signal input).

The energized molecules return naturally to the first level in a time known as the relaxation time. The random emission of the corresponding frequency is not wanted and, as


Above: S.E.R.L. parametric amplifier. Silicon diode is in centre of cavity (shown open) on polystyrene insulator. Pump and signal power is introduced via probes seen in side walls of cavity.

Left: Three-level paramagnetic maser assembly for insertion in cryostat. Cavity containing maser crystal is located at bottom of pump waveguide and signal cable (both enter at top): pipes at top are used for cooling. (Royal Radar Establishment).
the relaxation time falls rapidly with increasing temperature, the maser cavity and crystal must be kept very cool: this is achieved by boiling-off a liquefied gas round the cavity.

A $3-\mathrm{cm}$ signal is applied to the resonant cavity (which has modes at both signal and pump frequencies) via a coaxial cable. The incoming signal "triggers-off" the relaxation process in proportion to its strength, molecules returning to a lower state and emitting radiation at the signal frequency. The amplified signal is carried back up the coaxial cable and separated from the incoming signal by a directional coupler.

Originally, masers were operated at liquid helium temperatures-in the region of $1.5^{\circ} \mathrm{K}$. However. recent work at R.R.E. has established that the rate of decrease of relaxation time with increase of temperature is much smaller for ruby than for some of the other substances used and practical masers operating at about $60^{\circ} \mathrm{K}$ have been produced. This has the enormous advantage that these temperatures can be reached by boil-ing-off liquid oxygen which is far more readily available and much cheaper than helium. A large step along the path towards making the solid-state maser a practical device has thus been taken, and already useful gains together with usable bandwidths have been achieved at both 10 and 3 cm .

Instead of using a pump signal to elevate the energy state of molecules, the ammonia maser (built by Glass Developments, Ltd, for the Signals Research and Development Establishment, Christchurch) sorts, so to speak, the sheep from the goats by means of an electrostatic field. In this device a "jet" of ammonia gas (about 2 c.c. per day!) enters a vertical "tunnel" formed by eight wires alternately at earth potential and 20 kV . This strong non-uniform field affects the molecules having a low energy far more than those in a higher state: these latter pass down the tunnel and into a cavity resonant at the frequency corresponding to the radiation emitted when they drop back to the lower state (about $23.87013 \mathrm{kMc} / \mathrm{s}$ ). The "goats." or low energy molecules, are deflected


Klystron grids made by E.M.I. process mounted on microscope slide. Note very small area of grid presented normal to electron beam.
by the electrostatic field and are condensed on a cylinder which is cooled $b_{j}$ liquid nitrogen and encloses the wires.

This maser is inherently a narrowband device and is likely to have greater application as a frequency standard of extremely high purity and stability (possibly 1 part in $10^{10}$ ) than as an amplifier. How narrow this bandwidth is can be judged from the fact that the cavity was machined to within $1 / 10$ "thou" and then its temperature was controlled to about $\pm 2^{\circ} \mathrm{C}$, adjustment of this being used for fine tuning.
Parametric Amplifiers offer, like masers, some very attractive features; but they are likely to be of use at rather lower frequencies, probably in the region where specialized microwave valves take over from the conventional types. It can be shown that if the tuning of a circuit (i.e. one of the parameters $L$ or $C$ ) is altered at twice the frequency to which the circuit is tuned (again this is called "pumping") the circuit will exhibit negative resistance provided that the relative phase of sigmal and pump frequency is correct. This is an onerous requirement so the signal is placed deliberately off-frequency by a small amount: the device will amplify and attenuate alternately as the phase relationship varies, so that the amplified output obtained is modulated at the frequency corresponding to the beat between twice the signal frequency and the pump irequency. Sometimes it is necessary to add a third tuned circuit known as the "idler") resonant to this; but the $Q$ of the signal frequency circuit may be low enough to include this frequency within its pass-band. This type of amplifier is known as a three-frequency amplifier and it can exhibit a useful bandwidkh. As
the amplified output depends upon the power supplied by the pump circuit, increasing the pump power increases the gain-eventually oscillation results.

In the three-frequency amplifier shown by the Services Electronics Research Laboratory, Baldock, the parameter varied was "C"-the variable element comprised a small, back-biased, silicon junction diode mounted in the centre of a cavity resonant (in different modes) to pump ( $1500 \mathrm{Mc} / \mathrm{s}$ ), idler ( $972 \mathrm{Mc} / \mathrm{s}$ ) and signal ( $528 \mathrm{Mc} / \mathrm{s}$ ) frequencies. The pump power applied was about 80 mW : this realised a gain of 30 dB and a bandwidth of $2.5 \mathrm{Mc} / \mathrm{s}$, and 25 to 30 dB was quoted as being the maximum usable gain at the present stage of development, as stability is difficult to maintain at higher gains.
Microwave Valves. - Broad-band voltage-tuned O-type backward-wave/oscillators-that is oscillators in which the electrons interact with an r.f. wave travelling with a similar velosity but in the opposite direction, and in which a magnetic field is used only for focusing the electron beamare now available from several manufacturers. Two shown by Standard Telephones and Cables obtain suitable r.f. waves travelling much slower than in free space by using a set of interlinking hairpins as a slow-wave structure rather than the more usual interdigital line. The hairpin line has the higher impedance, and this, as well as the fact that the electron beam can completely interlink its open structure, makes the interaction considerably more efficient.
Klystron Grids have in the past been made from round wires formed into a mesh: this provides a poor hole-to-wire-ratio and reduces efficiency. E.M.I. Electronics, Ltd had on show some klystron grids made by a pro-


Glass Development's ammonia maser. Resonant cavity is in base, wire " tunnel " forming non-uniform electrostatic field for molecule separation is inside cylinder.
cess which gives an extremely thin section to parts of the grid at rightangles to the electron beam. Aluminium wires are given a thin coating of copper and packed into a copper tube which is then sintered to join all the copper coatings together and to the tube. The whole is "sliced" and the aluminium dissolved out, leaving a grid with very thin "inter-cell" walls.
Microwave Components.-A fourway switch shown by the G.E.C. Research Laboratories used a gaseous discharge plasma initiated by short d.c. pulses to bridge the gaps between the inners of the input and alternative output coaxial lines. R.f. propagation is sustained for several hundred microseconds until the free electron density in the discharge decays below a certain critical value dependent on the transmitted frequency. The rate of decay of the electron density can be varied by changing the gas material and pressure, and its initial value changed by altering the input pulse power. Since the electrons rapidly lose their energy to the positive ions and neutral atoms little noise is produced. This type of switch has
been found to be usable from 600 to $4,000 \mathrm{Mc} / \mathrm{s}$.
Propagation Test Equipment. Elliott Brothers (London), Ltd were showing propagation test equipment (which could also be used as a 24 -channel telephone or multichannel data link with a minimum of modification) for the $7.5 \mathrm{kMc} / \mathrm{s}$ band. This equipment comprises a selfcontained transmitter and receiver unit fitted in a cylindrical, weatherproof trunnion-mounted case with the aerial "dish" mounted at one end, the feeder and radiator (Cutlertype twin-slotted cavity) projecting through the end of the cylinder into the middle of the dish. One unusual feature is that the same klystron (1.2W output) is used for transmission and reception, the signals received being separated from the transmission by a wide-band ferrite isolator with a 46 dB " reverse" loss.' This, of course, necessitates that the incoming signals (from a similar unit) are frequency-displaced by the i.f. ( $70 \mathrm{Mc} / \mathrm{s}$ ) and automatically ensures that the remote unit is tuned to the transmission from the other. To avoid any chance of adjacent-channel interference a filter (rejection 50 dB ) consisting of three tunable cavities in the $\mathrm{H}_{011}, \gamma / 4$ and $\mathrm{H}_{012}$ modes is fitted between the balanced mixer and isolator. The whole unit can be tracked in azimuth, elevation and height from a remote point and a full complement of test equipment is built into the waveguide assembly, direct readings of received and transmitted power, etc., being obtained with a display unit fed from the balanced mixer and a thermistor fitted in the waveguide.
Filter Crystals often exhibit undesired modes outside the required response. This makes it difficult to secure satisfactory rejection over a wide band, typical responses for crystal filters having points outside the passband where rejection falls to only about 17 dB . Some results of work on this problem at the G.E.C. Research Laboratories were on show, the improved crystals having out-ofband peaks about 25 dB down on the main response and of negligibly small amplitude compared with the rest of the out-of-band response. This is achieved by restricting oscillation of the crystal to the main mode by bevelling the edges and using only small-area contacts. Satisfactory results in crystals for frequencies between 1 and $10 \mathrm{Mc} / \mathrm{s}$ have been produced by this procedure.

Micro-measurement Technique.-
The measurement with an accuracy
of $\pm 1 \%$ of a change of $1.0 \times 10^{-7} \mathrm{~cm}$ seems formidable enough without making the operation far more difficult by having to do it at temperatures near absolute zero; but. this is the problem presented by the measurement of coefficients of expansion of small samples at liquid helium temperatures. An apparatus for doing this has been developed at the Royal Radar Establishment and was shown on the Ministry of Supply stand.

The sample (about 1 cm long) and a reference are placed in a brass "pot", so that both the sample and the reference each produce a capacitance of the order of 10 pF to a common electrode mounted on the top plate. A bifilar-wound transformer secondary and these capacitances make up a bridge circuit and the energizing supply (derived from a $100 \mathrm{kc} / \mathrm{s}$ transistor oscillator) develops about 40 V across the bifilar winding. The balance detector (a transistor amplifier feeding headphones or a visual indicator) is connected to the common plate of the capacitors and the bridge is balanced by introducing a further signal, which is adjustable in amplitude and phase, to the centre tap of the transformer. By itself, the sensitivity of the bridge to a small unbalance condition is hardly sufficient-to improve this a high-Q coil tuned to $100 \mathrm{kc} / \mathrm{s}$ is connected across the output to the null detector. When the temperature has been changed by a small amount, say $1{ }^{\circ} \mathrm{C}$, and the bridge has been rebalanced, the relative coefficient of expansion of the reference and sample is obtained.

The sensitivity is such that light pressure applied from a pencil on the top plate of the " pot" causes a large unbalance indication due to distortion of the half-inch-thick pot walls; likewise the expansion caused by breathing momentarily on the pot whilst it is at room temperature gives a clear out-of-balance indication.
Bridges.-A small (1-inch) c.r.t. indicates the phase and amplitude balance in the Cossor LCR bridge Model 1446. The $2-\mathrm{kc} / \mathrm{s}$ input signal is fed to the X-plates and the error signal from the Wheatstone bridge to the Y-plates via a high-gain differential amplifier. Balance is thus indicated when the trace is a horizontal straight line. Resistance is measured using d.c. (central spot for balance) so that the resistance of an inductor can be determined.

Impedance comparators containing two resistive arms of a bridge and a high-gain amplifier feeding a detector were shown by Griffin \& George
and Dawe. In the Dawe Type 304 an internal $1,000-\mathrm{c} / \mathrm{s}$ oscillator is provided, and both phases and amplitudes can be compared on a meter. In the compact Griffin \& George "Panmetron" the a.c. input is derived from the mains, and a rectangular tuning indicator shows balance at maximum shadow.

Wayne Kerr showed that transformer bridges can be wound with sufficient accuracy to make absolute measurements accurate to within $0.01 \%$ and comparisons to within $0.001 \%$. The impedance looking into this type of bridge can be made less than $10^{-5}$ of the impedance to be measured. As the velocity of light is now known to better than 1 part in $10^{6}$ and linear dimensions can be measured to a similar order of accuracy, a standard capacitance of about 10 pF can easily be constructed to an accuracy of $0.01 \%$. Other standards of different sizes can then be obtained using such accurate transformers.
The Griffin-Raleigh Elution Bridge (Griffin and George) is notable for its extreme simplicity of operation. This bridge, used for determination of the end of a washing operation to remove ionically-dissociated solutions, consists of two conductivity cells connected in a Wheatstone bridge circuit and a transistoramplified balance detector. The influent runs through one cell and the effluent through the other, the bridge being balanced initially in the absence of the material to be washed. This procedure eliminates errors due to temperature, carbon-dioxide absorption and cell differences. When the material is introduced the conductivity of the effluent cell increases, so unbalancing the bridge which returns to balance when the concentration of ionized material in the effluent has fallen to the value that it had when the bridge was first balanced. As one of the conductivity cells is a standard the bridge can be used also for the direct calibration of other cells and for the determination of the resistivity of solutions. It is mains operated and low-voltage a.c. is used to energize the bridge circuit.
Digital Meters.-A sufficient number of these were already on show last year to illustrate most of the common general methods in use (see Wireless World, May 1958, p. 222). One simple new system was seen this year however in the Nash and Thompson prototype digital ohmmeter. Here measurements are made using an a.c. bridge whose off-balance
(Continued on page 129)
voltage is amplified and used to drive a 2-phase motor. The motor drives a helical potentiometer forming one arm of the bridge. As the bridge passes through balance, $180^{\circ}$ phase shift occurs in the off-balance output and the motor reverses and comes to rest: The time taken to travel the full 3 decades of the motor driven counter is approximately 10 seconds, and the input impedance is about 1,000 $2 / \mathrm{V}$.
Analogue-to-Digital Converters.Mechanical "digitizers" for converting shaft rotations into coded pulse output signals are usually commata-tor-type devices with wiping contacts or photocells for sensing purposes. G.E.C. have introduced an entirely new system which depends on electromagnetic induction. It is basically a transformer with a single energizing winding and a number of secondary windings corresponding to the rumber of digits. The rotating part is cup-shaped and contains a magaetic element which provides the coupling between the energizing winding and the various digit windings (which are arranged to conform to some code). A typical energizing voltage is 10 V r.m.s. at $20 \mathrm{kc} / \mathrm{s}$, giving an outpat of 1.5 V r.m.s.

An entirely electronic analogue/ digital converter shown by Mullard samples an input analogue voltage in the range $0-10 \mathrm{~V}$ and gives a binary output of 10 bits either serially or in parallel. Using transistors throughout, it is based on ten bi-stable circuits which are switched either "on" or "off" according to a comparison process which balances the current produced by the input voltage through a resistor with currents switched by the bi-stable circuits through graded resistors from a $10-$ volt source. The bi-stable circuits are triggered in sequence at a rate of $12 \mathrm{kc} / \mathrm{s}$ so the sampling period for a 10 -digit output is less than 1 millisecond.
Digital Storage Systems.-The principle of the superconductive storage element, based on a property of metals at very low temperatures, has already been explained in Wireless World (July, 1957, p. 326). A particular embodiment of this principle was described in the January, 1958, issue (p. 32), in which circulating currents are magnetically induced in a superconductive sheet by a drive conductor, the direction of circulation signifying whether a " 1 " or a " 0 " is stored. The Royal Radar Establishment demonstrated this idea, using three elements made from evaporated films of tin on a mica base
with drive and pick-up conductors, the whole being held at a temperature of about $3.7^{\circ} \mathrm{K}$ in a flask of liquid helium. Pulses of $50 \mathrm{~m} \mu \mathrm{sec}$ duration were written in, and read out when required, through special transmission lines with Constantan inners to reduce the thermal conductivity.

An unusual type of static storage system capable of giving a serial pulse output was shown by Elliott. It is based on both magnetic recording and magnetostriction principles. Information is stored permanently as a pattern of remanent magnetism along the length of a nickel-iron wire. It is read out by passing a single current pulse through the wire The magnetic field of this pulse interacts with the static magnetic pattern and causes magnetostrictive action, which creates an acoustic wave train corresponding to the stored pattern of information. This train travels along the wire and is converted into a serial pulse output by a pick-up coil, as in normal magnetostriction delay-line stores. The reading-out process does not destroy the stored information, which can be used over and over again.

Ericsson demonstrated the principle of the "twistor" magnetic storage system already described in the January, 1958, issue (p. 32) and February, 1959, issue (p. 80). Information was written into a $4 \times 4$ mat-
rix store by coincidence of currents through columns of coils $(100 \mathrm{~mA})$ and currents through the twisted magnetic wires ( 200 mA ) A p.r.f. of $500 \mathrm{kc} / \mathrm{s}$ was used, with transistor writing and reading circuuts.
A new kind of square-loop ferrite store, shown by Mullard, is intended for computer speeds up to ten times faster than are possible with the existing types of magneric matrix stores. It achieves the higher speed of operation by not switching the cores into their full states of saturation. Two ferrite cores are required for each digit stored. They are switched in opposite directions of magnetisation by a wire carrying the input digit pulse and, coincidentally, in the same direction by a read/write drive wire. Thus, one core is magnetized more than the other-which one depending on whether a " 1 " or a " 0 " is stored. The output wire threading the two cores combines their outputs in series opposition (corresponding to the input wire) When the read/write drive wire is energized to "read", the differing magnetic flux changes in the two cores induce a combined current pulse in the output wire, and the direction of this depends on whether a " 1 " or a " 0 " was stored.

## Very Low-Frequency Generators.-

 A popular circuit, which provides frequencies as low as 1 cycle in $10^{4}$ seconds in the Solartron JO 744 for
example, uses two mtegrator stages and overall feedback with $180^{\circ}$ phase shift. This can be regarded as providing the sine-wave solution of the differential equation $\mathrm{d}^{2} y / \mathrm{d} t^{2}=$ $-y$. This method conveniently provides outputs from the two integrators whose phases differ by $90^{\circ}$. In the Solartron oscillator the effective input impedance of the integrators is multiplied by 1,000 times up to $1000 \mathrm{M} \Omega$ by feeding them from a 1000 -to-1 ratio potential divider. This keeps the integrator capacitors to reasonable values, but even so leakage across these capacitors and the integrator amplifiers themselves results in damping being effectively applied in the circuit. A gradual reduction in the output level is however avoided by sampling it at its peak value and using current feedback to keep this peak value constant.

Low frequencies are obtained by a different method in the Servomex Transfer Function Analyser Type TFA46. A plane loop of wire is rotated at the required modulating low frequency ( $\omega_{m} / 2 \pi$ say) in a magnetic field which is alternating at a much higher carrier frequency ( $\omega_{c} / 2 \pi$ say). If $\mathbf{B}_{\text {max }}$ is the maximum flux linking the loop, then at any time $t$ the flux linking it can be expressed as $\mathrm{B}_{\max } \sin \omega_{c} t \sin \omega_{m} t$. The e.m.f. induced by the changing flux linkage is the differential of this quantity, i.e. $B \max \omega_{c} \cos \omega_{c} t \sin \omega_{m} t$ $+B_{\text {max }} \omega_{m} \sin \omega_{c} t \cos \omega_{m} t$. The second term causes the amplitude to vary with the modulating frequency. It represents effects due to the flux linkage changing at the modulating frequency, and is as might be expected negligible when the modulat-
ing frequency is very much lower than the carrier frequency ( $\omega_{m} \ll \omega_{c}$ ). The loop output is then equal to $\mathrm{B}_{\text {max }} \omega_{c} \cos \omega_{c} t \sin \omega_{m} t$ and the lowfrequency modulation term $\sin \omega_{m} t$ can be obtained by demodulating the carrier frequency. Simple avoidance of the use of a carrier frequency by rotating the loop in a fixed magnetic field at the required low frequency would not be nearly so effective since the output would vary with frequency and also be very low at very low frequencies. Direct modulation of a valve-generated high-frequency signal by means of a motor-driven capacitor is used in the Airmec Type 257 to provide (after demodulation) frequencies from 0.03 to $30 \mathrm{c} / \mathrm{s}$. A second set of stator vanes on the capacitor can be manually rotated to provide a second output with a continuously adjustable phase difference from the first. These last two methods have a number of advantages. First, the output frequency can be varied by changing the motor speed without any delay due to long time constants, and its level can also be set quickly and accurately simply by varying the carrier level. Secondly, the purity and level stability of the output are independent of frequency. Finally, the output frequency is readily measured from the motor frequency.

Audio-frequency Oscillators. - A Wien bridge which is balanced by the addition of two extra resistive arms (one of which is an amplitudestabilizing thermistor) and which is fed from a push-pull transistor amplifier is used in the Dawe Type 421 to avoid the low input impedance and changing output phase-

U.K. Atomic Energy transistor sampling oscilloscope displaying a pulse a few millimicroseconds wide. The sampling pulse generator and sampling circuit are housed externally to the oscilloscope in a probe attached to the avalanche input pulse source to avoid using cable to carry very short pulses.
shift difficulties normally associated with transistors. A standard Wien bridge oscillator using valves is an addition to the range of Mullard educational constructional circuits.

A parallel-T network is used to control feedback in the Muirhead D-888-A analyser-oscillator to provide either narrow-band amplification with an equivalent $Q$ of at least 70 or, with additional feedback, oscillation at frequencies beyond the audio region up to $650 \mathrm{kc} / \mathrm{s}$.
Ultrasonics.-Two new metal-testing sets were shown covering the range 1 to $10 \mathrm{Mc} / \mathrm{s}$, this latter high frequency being preferable for the examination of materials with a tight-knit grain structure, such as austenitic steels. The idea behind this is to use a shorter, higherpowered pulse (consequently the higher frequency so that a reasonable number of cycles form the pulse) to give higher definition, so making clearer the difference between the grain structure "grass" and the echoes due to faults in the material : smailer faults, too, are detectable. The Ultrasonoscope MkII flaw detector provides a non-linear amplifier facility so that this "grass" may be compressed in amplitude.

Kelvin Hughes, in their Mk. 5 set, provide two gated amplifiers whose outputs are displayed on muchbrightened portions of the timebase. This, it is said, eases the problem of locating "dangerous echoes" in production testing, by setting the gates so that only the points where faults could be the cause of failures in the object being tested are displayed at full brightness.

A rather amusing demonstration of the potentialities of barium titanate transducers was given on the stand of Sir Howard Grubb, Parsons and Co., Ltd. This took the form of a small transducer propagating a $1.6-\mathrm{Mc} / \mathrm{s}$ particle vibration in air and a receiver whose output was mixed with the transmitted frequency and fed to an a.f. amplifier and loudspeaker. Relative movement of the transmitter and receiver resulted in a doppler "burp" from the loudspeaker. This work, of course, has a serious application: it forms part of a research programme with a view to improving "sonic" gas analysers by raising the frequency used from 3 or $4 \mathrm{kc} / \mathrm{s}$ possibly into the megacycle region.

Short-Pulse Generators.-A second-ary-emission pentode is used to generate pulses with a rise time of less than $15 \mathrm{~m} \mu \mathrm{sec}$ in the Wayne Kerr Type P131. The pentode is
normally cut-off, but on triggering with a positive input pulse it conducts, and a negative pulse is sent from the anode down a shortcircuited delay line, this negative pulse being also fed back to the cathode. This positive feedback to the cathode results in a cumulative process, since owing to the secondary emission the anode current is several times the cathode current. The secondary emission also results in the secondary-emission electrode (dynode) losing electrons, and so its potential rises and provides the leading edge of the output pulse. At a time determined by the line constants the negative pulse from the anode is reflected from the shortcircuited end with its phase reversed. When this positive pulse is received back at the anode, the feedback to the cathode cuts off the pentode and the dynode potential immediately falls, terminating the output pulse. Owing to stray capacitances this fall would normally be rather slow. These strays are however discharged by using the positive reflected pulse to switch on a normally cut-off ordinary pentode connected between dynode and cathode. The resulting output pulse fall time is less than $20 \mathrm{~m} \mu \mathrm{sec}$.

Medical Electronics.-A mechanical scanning and recording machine for displaying pictorially the distribution of radioactivity in human organs (introduced deliberately by radioactive isotope techniques) was demonstrated by Hammersmith Hospital Department of Physics. The patient himself is moved on a float-ing-top couch in a scanning raster under a stationary radiation detector. The output from the detector is fed to a ratemeter and a recording meter which is fixed to the moving couch. According to the rate of count the meter deflection causes printing ribbons of different colours to move under a printing stylus, which makes dots through them on to a fixed sheet of paper. The result is an even array of dots covering an area equal to the area of patient scanned. The regions giving a count rate within certain limits come out in the same colour, rather like height contours on a map.

Bristol University Department of Anæsthetics had an electroencephalographic display system, based on a standard c.r.o., which gave inherent rhythm and phase information like the Grey Walter toposcope but by using a television type of timebase instead of the rotating p.p.i. kind. A 10 -second frame timebase is used, with a line-time-


Printing mechanism and control gear of the Hammersmith Hospital body radiation scanning machine.
base period depending on the frequency component of the brain wave under investigation. The output of the e.e.g. amplifier modulates the brightness of the tube spot, and the display (on a long-persistence screen) appears as vertical bands of light and dark.

Oscilloscopes.-A response from d.c. to $60 \mathrm{Mc} / \mathrm{s}$ ( 3 dB down) with a calibrated maximum sensitivity of $50 \mathrm{mV} / \mathrm{cm}$ can be achieved in the new Cossor Model 1076 using the 1078 plug-in Y pre-amplifier. Twenty-four calibrated timebase speeds from $0.02 \mu \mathrm{sec} / \mathrm{cm}$ to $5 \mathrm{sec} /$ cm are available, and a $200 \mathrm{Mc} / \mathrm{s}$ ( $\pm 2 \%$ ) oscillator provides intensity modulation dots for the measurement of pulse rise-times. A trigger level control is one of the facilities in the 1070 plug-in trigger unit. Other plug-in units, including, for example a differential $Y$ pre-amplifier, are being developed for this oscilloscope.

Infinite persistence storage oscilloscopes for viewing " once in a lifetime" phenomena, using a tube similar to that described in our review of last year's Physical Society Exhibition (May 1958, p. 221), have previously been shown by Cawkell and are now also made by Solartron. Facilities available in the new Solartron QD 910 include a triggered timebase with 49 alternative speeds from $1 \mu \mathrm{sec} / \mathrm{cm}$ to $10 \mathrm{sec} / \mathrm{cm}$ and two identical Y-amplifiers with responses from d.c. to $1 \mathrm{Mc} / \mathrm{s}$ ( 3 dB down) and single or differential inputs.

A sampling principle for viewing repetitive waveforms is used to give an effective response from d.c. to
$300 \mathrm{Mc} / \mathrm{s}$ ( 3 dB down), using transistors except only in the final c.r.t. plate driving stages, in an oscilloscope shown by the U.K. Atomic Energy Authority. The maximum sensitivity is $200 \mathrm{mV} / \mathrm{cm}$ and the equivalent input noise 100 mV . The input impedance is as high as 1 pF in parallel with $0.25 \mathrm{M} \Omega$. Signals with p.r.f.'s from $100 \mathrm{c} / \mathrm{s}$ to $10 \mathrm{kc} / \mathrm{s}$ can be observed, and from 40 to 320 samples taken.
The principle consists in using a very short pulse to sample a small portion of each input pulse, the samples being taken at a time delay from the beginning of the input pulse which is increased slightly between successive input pulses. Thus the sampling pulse gradually progresses through the input pulse and builds up a picture of it in the process. The sampled pulse can be amplified throughout the period between input pulses so that only a comparatively poor response is needed for this amplifier. Moreover, if each sampled pulse is referred to the same zero potential, the sampled pulse amplifier need not even respond to d.c. Since the period between input pulses in which the sampled pulse is amplified and fed to the c.r.t. is " most of the time", this period is nearly independent of the input p.r.f., and so the brightness also is unaffected by the input p.r.f.

The equivalent rise time is equal to the width of the sampling pulse. The very narrow sampling pulses thus required to obtain a short rise time are obtained by avalanche operation of a transistor. In such operation the transistor emitter current is
initially cut off by making the base positive with respect to the emitter. Its collestor is connected to a negative potential of several hundred volts through a high resistance so as to draw a current of several hundred microamperes, the collector taking up a potential of about -50 V . If now the base potential is allowed to fall to that of the emitter, cumulative electron multiplication occurs in the transistor and the collector potential very rapidly rises to that of the emitter. By differentiating this rise, a very short pulse of several hundred milliamperes into a few tens of ohms can be obtained. By selecting ordinary $10 \mathrm{Mc} / \mathrm{s}$ OC44 transistors, pulses as short as 1 to $2 \mathrm{~m} \mu \mathrm{sec}$ can be obtained in this way. Sampling is carried out in the oscilloscope by feeding the avalanche pulse to the base of a $50 \mathrm{Mc} / \mathrm{s}$ SB100 (this is the only high-frequency transistor required) and the input pulse to its emitter. The difference between these two pulses then appears on the collector. Most of the noise is produced in this sampling process and in the avalanche sampling pulse itself.
Amplitude Distribution Measure-ment.-A sampling technique was also used in an instrument shown by Plessey. The input waveform is fed to one plate of a special type of c.r.t. and the sawtooth sampling waveform (which has a much lower frequency) to the opposite plate. A thin wire parallel to these plates replaces the usual screen. For a fixed sampling voltage, only a certain voltage on the input waveform will deflect the beam so as to strike the thin wire. The current thus produced will be proportional to the length of time during which this input voltage occurs. As the sampling voltage gradually changes, the current in the wire thus gives the lengths of time during which the various voltages present in the input occur.
C.R.O. Tubes. - Development in this field is aimed at the twin requirements of higher deflection sensitivity and increased trace brightness (especially with high-speed writing). Unfortunately these two requirements often conflict. The use of post-deflection acceleration to increase the brightness, for example, tends to reduce the deflection sensitivity and distort the trace. If the deflection plates are brought very close together to increase their effect on the beam, the anode aperture has to be made proportionately smaller so that the beam electron density, and hence the brightness, is reduced. An interesting method of

tube envelope to increase the deflection sensitivity. The coils in their experimental tube (two pairs) are arranged sequentially like electrostatic deflection plates and measure about $3 \mathrm{~cm} \times 1 \mathrm{~cm}$, with inductances of about $20 \mu \mathrm{H}$. The deflection sensitivity is 1 mA per millimetre in both directions.
Scan Magnification.-A means for increasing the spot movement produced by a c.r.t. magnetic-deflection system without increasing the power input to the coils has been developed by Mullard, Ltd. The basic principle of this involves passing the electron beam through a magnetic diverging lens after deflection; but this is not so simple as it seems, for two reasons. The spot will increase in size and become distorted, due to divergence of the beam itself, and there is difficulty in producing a magnetic lens of the required form. The solution adopted is to use a system of quadrupoles, which are "lenses" having a diverging effect in one direction and a converging effect at $90^{\circ}$ to this ( $x$ and $y$ deflections, say). One quadrupole is placed between the scanning coils and the c.r.t. screen: this will cause a movement of the beam in the $x$ direction to increase and also mis-shape the spot. In the converging, or $y$, direction the deflection would be reduced; to obtain an increase in this direction the beam has to enter the lens in such a way that a focus is formed inside the lens; then divergence of the emergent beam is effected and the quadrupole magnifies (although to a lesser degree than in the $x$ axis). Thus the beam entering the magnifying quadrupole must have different $x$ and $y$ foci and be shaped in such a way that the ellipticity introduced by the magnification process exerts a correcting rather than a distorting influence. This beam requirement is met by employing two more quadrupoles, with their axes crossed at $90^{\circ}$, as the focusing system. The demonstration on the Mullard stand showed the potentialities of this system very well-the image on the scan-magnified display filled the tube
and the conventional display was rather less than half-an-inch across. Some increase in spot size (about 5 times) does occur; but with modern c.r.t.s this should not prove an insurmountable problem; the quadrupoles, too, have the advantage of being of fairly simple constructionsmall bar magnets and simple pole pieces are used.

Electroluminescence, alreačy well established for illuminated signs and notices, is now being develaped for more complex indicators. Thorn, for example, were showing a new type of digital indicator for displaying the numerals 0 to 9 on a single flat plate (which could be viewed from very oblique angles). The numerals are built up by activating various combinations of electrodes on the back of the electroluminescent layer, the front electrode being continuous over the whole surface. A coding matrix, formed by a printed circuit with pellets of non-linear resistance material, selects the right combination of back electrodes to construct a particular numeral when the activating voltage is applied to the appropriate one of ten input wires. About 500 V a.c. is required for aetivation, with a current of 0.5 mA .

Also demonstrated by Thorn was a crude but nevertheless interesting attempt at achieving scanning over a series of electroluminescent elements (with perhaps the eventual aim of television picture presentation). A row of electroluminescent and photoconductive elements were linked

Image Intensifiers. - The principle of electroluminescence (see above) was used in an X-ray image intensifier demonstrated by Thorn in association with the Physics Department of King's College Hospital. It consists of an electroluminescent layer and a photoconductive layer separated by an optically opaque but electrically conducting material. An alternating voltage of about $1,000 \mathrm{~V}, 50 \mathrm{c} / \mathrm{s}$, is applied across the whole "sandwich ". X-ray radiation falling on the photoconductor causes it to conduct in proportion to the intensity. This produces a corresponding pattern of increased voltage across the electroluminescent layer, and light is emitted to form a visual image. The picture is $20-50$ times brighter than that obtained from a conventional fluorescent screen.

Brightness amplification can also be obtained by using television techniques, and it is then possible to introduce extra facilities like variable contrast to assist in viewing. English Electric Valve Company showed a $4 \frac{1}{2}$-inch image orthicon pick-up tube intended for this purpose.

A more direct method of using the television pick-up tube for image intensification-this time for an electron microscope-was shown by Siemens Edison Swan. The fluorescent screen of the microscope is replaced by a photoconductive type of pick-up tube, the two envelopes being sealed together and evacuated as one. The sensitive screen of the pick-up tube is a layer of amorphous selenium, and the resistivity of this
is modified in a pattern when bombarded on one side by the highenergy electrons from the microscope beam. The other side of the selenium is scanned with 405 lines by the electron beam of the pick-up tube, and a signal corresponding to the electronmicroscope image is obtained from a signal plate as in television technique.
Photoelectric Devices.-One of the latest and most efficient devices for converting light into electrical energy is the silicon junction "solar cell." Ferranti were showing examples which gave open-circuit voltages of 500 mV for a light intensity of about 1,000 foot-candles, and currents of 20 mA per square inch of active area. They are available in various sizes with rectangular and circular shapes and in multiple form. One demonstration showed the ability of these cells to resolve high-frequency pulses of light (above $1 \mathrm{Mc} / \mathrm{s}$ ).

The photomultiplier tube has hitherto been rather a large and clumsy device because of the space required for its electron multiplier structure. This is becoming increasingly less true nowadays. For example, 20th Century Electronics were showing a single-stage tube measuring only $30 \mathrm{~mm} \times 6 \mathrm{~mm}$. It has a sensitivity comparable with a gas-filled photocell (about $100 \mu \mathrm{~A}$ per lumen) and the speed of response of a vacuum photocell.
Semiconductor Devices.-There was evidence of continuing development in power and high-frequency transistors. Silicon transistors are

together electrically and optically so that the light from one electroluminescent element activated a photoconductor connected in series with the next electroluminescent element. The output voltage from the photoconductor causes the second electroluminescent element to light up and illuminate the next photoconductor, and so on, causing a spot of light to travel down the row. The speed of scanning is severely limited at the moment by the slow response of the photoconductors.


Above: Mullard scan magnification: both 5 -in tubes are operated under identical conditions except that one has scan magnification, the other conventional deflection and focusing. Gain is about 12 times.

Right: Image intensifier tube mounted under the desk top of an electron microscope (Siemens Ediswan Swan).

becoming more common and B.T.H have now entered the field with several diffused-junction development types. These include an h.f. transistor with an alpha cut-off frequency of $30 \mathrm{Mc} / \mathrm{s}$ and a power type with a collector dissipation of up to 100 watts. Thorn demonstrated the switching of a 300 -watt lamp by their GT422 power transistor from a 45 -volt supply. G.E.C. had a new h.f. drift transistor, EW69, with a cut-off frequency of $30 \mathrm{Mc} / \mathrm{s}$.
Amongst the more specialized semiconductor devices was a p-n-p-n "sandwich" diode structure with a characteristic suitable for bistable switching circuits (see Technical Notebook, October 1957, p. 502, for principle). It has two states of conductivity and is made to switch from one to the other according to the value of the applied voltage. The triggering voltage is 100 V and the holding current for the fully conducting state is 50 mA at less than 1.5 V . Another semiconductor switching device giving two states of conductivity (shown by Plessey) depended on a phenomenon which occurs in germanium at very low temperatures, when the material is virtually an insulator. If a voltage is applied across the germanium the few remaining free current carriers are accelerated until, at a critical field strength, their energy is sufficient to ionize the impurity centres and cause a non-destructive breakdown. The resistance of the germanium then falls from about $10 \mathrm{M} \Omega$ to $20 \Omega$.

Transistor Tester.-Noise can be measured in the new battery operated Avo Transistor Analyser by comparison with the output from a $1,000 \mathrm{c} / \mathrm{s}$ stabilized transistor oscillator. By restricting the measurement to noise above $200 \mathrm{c} / \mathrm{s}$, spurious hum pick-up is avoided and amplifier interstage coupling made easier. Current gain is measured by comparing the output from the oscillator before and after amplification using a calibrated attenuator and high-gain transistor amplifier.

Barium Titanate Capacitors are difficult and expensive to make in high values-difficult, because barium titanate is very brittle in the form of ceramic sheets, so making it hard to "stack" them without breakage; expensive, because the sintering process is carried out in air, and palladium is the cheapest metal able to withstand the sintering and remain unaffected. In the new Plessey process the barium titanate powder is mixed with a


Avo transistor analyser allowing noise measurement.
plastic binder to form a flexible sheet and the binder is volatilized during the sintering process, leaving the barium titanate dielectric. To overcome the expense associated with palladium they have also developed a process by which a sufficiently high dielectric resistivity may be obtained by sintering in hydrogen, so that nickel can be used for the "plates".

Resistance Wires for use in extremely stable highly accurate resistors are sometimes wound in pairs, the two wires having mutually compensating temperature coefficients and changes of coefficient. Johnson, Matthey have, in their Silver Minalpha wire, carried this technique a stage further by using the compensating alloy as a sheath to Minalpha wire. Minalpha itself (copper-nickel-manganese) has a positive temperature coefficient of about 5 parts in $10^{6}$ per ${ }^{\circ} \mathrm{C}$ up to $26^{\circ} \mathrm{C}$ : this then turns negative after reaching a maximum. The silveralloy sheathing (silver-tin-manganese) exhibits almost the reverse of this characteristic, having a small negative coefficient up to $20^{\circ} \mathrm{C}$, going positive above this point, so that, by choosing the optimum proportions of Minalpha to sheath, a resistance wire with built-in compensation from $10^{\circ} \mathrm{C}$ to $100^{\circ} \mathrm{C}$ is produced.

Printed Circuits.-A recent development in this field was shown by G. V. Planer, Ltd, being a copper circuit on a glass substrate. The copper is deposited by a direct metallizing process, no "glue" of any type being used. Circuits produced by this method are said to be inert, stable and suitable for high-temperature operation. The examples shown included some inductors buitt up from double-sided " windings," the copper continuing over the edge of the glass to make contact with the
coil on the other side and the coils on other sections of the stack.
Magnetic Materials. - Comalloy (Murex Ltd. is a cobalt-aluminiummolybdenum material with an energy product of $1.0 \times 10^{6}$ gauss-oersteds, the remanence being $9.7 \times 10^{3}$ gauss. Although the energy product is not as high as that of some "magnetic-ally-hard" materials, this sintered material possesses the valuable characteristic of being machinable before its final heat treatment is applied. In its pre-treatment state the cutting characteristics resemble those of mild steel, and it is said that little distortion occurs during treatment, so that it is usually necessary only to restore the surface finish.
Melting by bombardment with an electron beami was perhaps the most direct application of electronics shown at this year's exhibition. Two demonstrations were given, one by S.E.R.L. (on the National Research Development Corporation's stand) and the other by Associated Electrical Industries. The S.E.R.L. apparatus was being used for zone refining: this was accomplished by encircling the rod of material to be refined with a heated tungsten filament and beaming the emission with two annular plates above and below, and connected to, the filament. The current is controlled by the variation of the filament temperature and e.h.t. and for zone refining either the cathode assembly or the rod can be inoved vertically (surface tension of the molten material keeps the rod whole). The temperatures achieved are sufficient to melt tantalum (m.p. $3000^{\circ} \mathrm{C}$ ) and tungsten. A $3 / 16-\mathrm{in}$ diameter ruthenium (m.p. $2450^{\circ} \mathrm{C}$ ) rod requires a bombarding current of about 100 mA at 3.75 kV .
Four electron guns consisting of a retractable tungsten filament in a slot in an air-cooled brass block (again connected to the filament) are used in the A.E.I. equipment: these are mounted above the material (e.g., silicon) to be melted and the beams are focused and deflected on to the material by electromagnets. 5 kV is applied to the guns and silicon crystal-pulling can be achieved with an input power of about 1 kW (rough comparison: 4 kW for r.f. and silica crucible method). The container is, of course, continuously evacuated, a pressure of at least $1.0 \times 10^{-5} \mathrm{~mm}$ mercury being necessary. Silicon crystals produced by this process are extremely pure as the molten material is not in contact with any contaminating substance (e.g. silica) and most of the other impurities are volatilized.

By "CATHODE RAY"

## 2. Throwing Weight Around

There was a young lady named Bright
Who travelled much faster than light.
She started one day
In a relative way,
And came back the previous night.
Traditional

THE difficulty about relativity is not so much that it is complicated as that it upsets things we had come to regard as absolutely basic, such as length and time. It is as if we were suddenly asked to accept that two and two make three. Perhaps an arithmetic of that kind would be no more complicated and difficult than the one we know, but it would mean we would have to think hard about every step instead of reeling it off from memory.
As we saw last month, Einstein's Special Theory of Relativity follows naturally from two simple facts, but the results are startling.
The facts are:
(1) The speed of light in empty space, denoted by $c$, is always the same.
(2) Nothing has been found in the whole universe which can be shown to $b \approx$ fised* and thereby entitled to a better claim than anywhere else as a reference point for measurement.

We drew a graph of time against length or distance (that is to say, space in one dimension) as reckoned from one viewpoint (ours) and then plotted on it the varying position with time of another viewpoint travelling at a uniform speed relative to us (Fig. 1). We also plotted on the same graph the progress of a beam of light, showing it travelling at the frequently and accurately measured speed of almost $3 \times 10^{8}$ metres $/ \mathrm{sec}$. Our problem was to provide the graph with time and distance scales applicable to the other viewpoint, such that according to them too the speed of light would be $3 \times 10^{8}$ metres $/ \mathrm{sec}$., as per Fact No. 1.

We were given the other viewpoint's zero-distance line, because it was our line representing their position relative to us. It then became clear that in order to comply with the requirements we had to abandon the usual assumption that what to the observers at the other viewpoint were a second and a metre must necessarily be the same to us.
Having done that, we had an infinite choice of time and distance scales. For instance, we could have made their metre measure up on our scale as two-thirds of a metre, leaving their time scale the same as ours. But that arrangement would have made our metre look like $1 \frac{1}{2}$ metres to them, and in the light of Fact No. 2 such a lack of reciprocity would be anomalous. In other words, since there is no essential distinction between the two view-

[^5]points, it would be unaccountable that one of them should find the other's lengths smaller and one should find the other's lengths larger.

We found we could avoid this anomaly by making the other people's time markings different, when seen from our viewpoint, in the same way as their distances. This, shown in Fig. 2, enabled us to arrive at a formula (part of the celebrated " Lorentz transformation ") for converting either's scales of seconds and metres to the other's, the difference being due to each viewpoint travelling steadily

Fig. I. Simple time/distance graph, showing the progress of a beam of light and of another viewpoint.

relative to the other at a velocity $v$. The conversion factor is

$$
\sqrt{1-\frac{v^{2}}{c^{2}}}
$$

and we see that according to it all lengths (in the direction of motion) and time intervals in a system moving towards or away from any observer measure less to him than to an observer travelling with that system. So a chronometer on a rocket hastening to the moon would, so far as perfect measurements on earth could tell, be running slightly slow, though it would be correct to an observer with equally perfect measuring gear in the rocket.

We also soon see, if we try plotting the conversion f2 tor against $v$, as in Fig. 3, that it doesn't begin to differ appreciably from 1 until $v$ is far greater than can be achieved (relative to earth) by the fastest
moon rocket, or even by a car on the Preston bypass. So for most practical purposes relativity doesn't make a ha'porth (or even a microfarthingsworth) of difference. But in modern research certain small particles, such as electrons and mesons, sometimes approach the speed of light quite closely, with remarkable relativistic results. In the limit it would appear that to photons-which are light itself-all distance in empty space is shrunk to nothing, and that what to us are the thousands of million of years light takes to reach us from the far nebulæ is to them no time at all! At least we can say that any speed greater than $c$ is unimaginable and presumably impossible, so we need have no fear of being faced with the situation outlined in the verse about Miss Bright.

What then, we may ask, would happen if a particle travelling at nearly the speed of light were to shoot off a sub-particle at high speed in the same direction? If $v$ was the velocity of the particle (relative, say, to us) and $u$ the velocity of the projectile relative to the


Fig. 3. Distance and time measures on one viewpoint always look less on another travelling at relative velocity $v$, the contraction factor being as plotted here.

Fig. 4. Diagram for calculating how to combine two velocities which are appreciable in comparison with that of light.

particle, then we would usually reckon that the velocity of the projectile relative to us (w) was $v+u$. So if $v$ was, say $\frac{3}{4} c$, and $u$ was also $\frac{3}{4} c$, then $w$ would be $1 \frac{1}{2} c$, which is impossible.

Any such question would reveal that we were still bound by our two-and-two-make-four habits. Fig. 2 is a reminder that we were successful in using the graphical method to arrive at the Lorentz formula by simple geometry, which seems to me much more convincing and easily visualized than the usual textbook method by algebra, so let us try following it up to discover the correct result of combining two velocities in the same direction.

To save effort let us use our previous diagram with its dotted line (OBC in Fig. 4) to represent our tracking of the particle moving at the rate $v$. Remember, in this diagram the distance $3 \times 10^{8}$ metres is represented by the same length horizontally on the
paper as 1 second vertically, so $c$ appears as the diagonal of any square. Therefore

$$
\frac{\text { Speed of particle relative to us }}{\text { Speed of light }}=\frac{v}{\mathrm{c}}=\frac{\mathrm{AB}}{\mathrm{OA}}=\frac{x_{b}}{t_{b}}=
$$

Note that the quantities distance and time are now denoted by $x$ and $t$ respectively; and $x_{b}$ means the distance from zero, and $\tau_{b}$ the time after zero, represented by point B -and so on.

Next we measure off along the particle's time and distance scales-the sloping dotted ones-1 second (represented by anywhere on the upper dotted line) and $C D$, so that
$\underline{\text { Speed of projectile relative to particle }}=\frac{u}{C D}$
Speed of light

It follows that

$$
\begin{aligned}
& \frac{\text { Speed of projectile relative to us }}{\text { Speed of light }}=\frac{w}{c}=\frac{x_{d}}{t_{d}} \\
& =\frac{O C \sin \theta+C D \cos \theta}{O C \cos \theta+C D \sin \theta}
\end{aligned}
$$

$$
(\text { from }(2))=\frac{\mathrm{OC} \sin \theta+\mathrm{OC} \frac{u}{c} \cos \theta}{\mathrm{OC} \cos \theta+\mathrm{OC} \frac{u}{c} \sin \theta}
$$

$$
=\frac{\sin \theta+\frac{u}{c} \cos \theta}{\cos \theta+\frac{u}{c} \sin \theta}
$$

$$
=\frac{\cos \theta \tan \theta+\frac{u}{c} \cos \theta}{\cos \theta+\frac{u}{c} \cos \theta \tan \theta}
$$

$$
=\frac{\tan \theta+\frac{u}{c}}{1+\frac{u}{c} \tan \theta}
$$

$$
\begin{aligned}
\mathrm{n}(1)) & =\frac{\frac{v}{c}+\frac{u}{c}}{1+\frac{v u}{c^{2}}} \\
\therefore w & =\frac{v+u}{1+\frac{v u}{c^{2}}}
\end{aligned}
$$

So if (to go back to the example that started this off)

$$
\begin{aligned}
& v=u=\frac{3}{4} c, \text { we find they add up to } \\
& w=\frac{1 \frac{1}{2} c}{1+\left(\frac{3}{4}\right)^{2}}=0.96 c
\end{aligned}
$$

In fact, even if $v$ and $u$ both go to the limit - $c-w$ is still no more than $c$. Sounds crazy, of course, but can you find any flaw in the argument?

One can hardly tamper so drastically with all we knew about length, time, and speed, and expect the consequences to go no farther. As the schoolmaster said, you will hear more of this. To go straight to the heart of the matter, consider energy and mass-the basic ingredients of the universe.

When a mass is moving, it has kinetic energy, which we are told at school is equal to $\frac{1}{2} m v^{2}$. We are also told that energy is conserved; that is, it
can't just disappear without trace-it can only change into an equal amount of energy of another kind. Well, even without bringing Einstein into it, that ought to make one think. For we used also to be told that mass is another indestructible quantity. So $m$ in the energy formula was just a constant. But nobody can give an authoritative ruling on how much, in any particular case, $v$ is. Take the earth, for instance. Galileo was threatened with torture if he didn't agree that its $v$ was definitely zero. And certainly, from the point of view of his inquisitors, who were sitting on it, that was true, and its kinetic energy therefore also zero. An astronomer on Mars would observe that its $v$ relative to him was at times very considerable, and he would be glad that the planetary orbits were such that there was no risk of collision, which would demonstrate the earth's kinetic energy in no uncertain fashion. (We would of course take a different view of the catastrophe, blaming the k.e.-and the k.o.!on Mars.)

So it appears that a body can have a lot of kinetic energy and at the same time none at all, depending on what point one happers to measure its velocity from. Which makes the law of conservation a little less simple than we may have thought. But let that pass.

Consider again the projectile-firing situation, but at such a low speed that Einstein can safely be ignored. Suppose you are cruising along a street in your car and, not liking the face of a man standing in your way, you hurl at it a custard pie of mass $m$, with a velocity (relative to yourself) $v$, which, by a curious coincidence, is equal to that of the car relative to the man in the street (Fig. 5). The latter, who happens to be interested in such problems, notes that your action has increased the k.e. that the pie had when it was travelling along with the car $\left(\frac{1}{2} m v^{2}\right)$ to $\frac{1}{2} m(2 v)^{2}$; that is to say, you have increased its energy by $1 \frac{1}{2} m v^{2}$. You, on the contrary, are firmly of the opinion that you have imparted a velocity $v$ to a previously stationary and unaggressive pie, thereby increasing its energy by only $\frac{1}{2} m v^{2}$. So here is another potential cause of friction between you and the man in the street.

I, as an unbiased spectator, am chiefly disturbed by the thought that not onlly is it impossible to obtain agreement about the absolute kinetic energy of a body (which, after all our talk about relativity, was perhaps only to be expected), but that even the k.e. relative to what it was before the imparting of a given amount is disputable. How much energy did you in fact give the pie? Its impact on the man is four times as devastating as if you had thrown it with equal exertion from a standing car, or three times as much as it had while in the moving car. There is a 3-to-1 discrepancy between different but apparently faultless methods of calculation.
Much to my relief I found on looking into it more closely that although you were correct in supposing that it cost you no more effort to throw your missile from a moving car, and the target was also correct in supposing that the missile from the moving car did four times as much work on his face as one thrown equally hard from a standing car, nevertheless there is no discrepancy. It would spoil the fun to give the solution now; perhaps the Editor might offer a small reward for the best one sent in.

At least we need have no doubts about the correctness of the k.e. formula-l $m v^{2}$. So, $m$ being assumed


Fig. 5. This incident poses some interesting scientific problems.
constant, doubling the velocity quadruples the energy. At normal car and pie-throwing speeds there need be no uncertainty about what is meant by doubling the velocity (relative to a specified origin). But if our car were travelling towards the man in the street at the speed (relative to him) of $\frac{1}{2} c$, and you were to throw the pie at him with the same speed relative to yourself, we have discovered by a chain of inexorable logic that the man sees the pie coming towards him not at the speed $c$ but $0.8 c$. According to classical or school dynamics, your giving the pie the same velocity from yourself as you and it already had relative to the specified origin ought to quadruple its energy, whether the velocity in question was 10 m .p.h. or $\frac{1}{2} c$. But $\frac{1}{2} m(0.8 v)^{2}$ is obviously not four times $\frac{1}{2} m(0.5 v)^{2}$-in point of fact it is 2.56 times.

Here is another apparent discrepancy. But even if you didn't know the catch before I started, I have given so many hints that you oughtn't to have much hesitation in suggesting that it lies in wrongly assuming mass to be constant. There is some excuse for this, when we are brought up on the Law of Conservation of Mass-or at least in my day we were-and anyway one has an intuition that the mass of a thing is something pretty definite, that can't be annihilated or created, least of all just by changing its speed of movement. One would as soon expect an electron to vary its charge as it goes along. Nevertheless mass is actually one of the most difficult things to define satisfactorily. Of course it is mixed up in our minds with weight, and although we know that the weight of a thing depends on where it is (being almost nothing in a distant space ship or in a lift falling freely down its shaft) the fact that weight-other things being equal-is exactly proportional to mass is a strange coincidence, which led Einstein to his General Theory of Relativity.

## Another Conservation Law

Don't panic! We're not going to embark on that; we had just reached the point of presuming (correctly) that if the relativity law for adding high speeds was not to land us in serious discrepancies about energy we would have to abandon the idea of constant mass. Without going into the mathematical details one can see that what energy is lost by the final velocity of the pie, being only $0.8 c$ instead of $c$, could be made up if its mass were to increase with velocity.

The actual relationship of mass to velocity is most easily calculated on the basis of another conservation law-that of momentum, which is equal to mass times velocity. What you do is describe the momentum conditions of two elastic bodies before and after they have collided, from two points of view moving relative to one another. That brings in the Lorentz transformation. The whole thing is given in simple algebra by R. C. Tolman in his book
"Relativity, Thermodynamics and Cosmology," p. 43. Very conveniently, the law for relativistic increase in mass is that the mass ( $m$ ) at velocity $v$ is equal to the mass at rest ( $m_{o}$ ) divided by our old friend the Lorentz shrinkage factor:

$$
m=\frac{m_{0}}{\sqrt{1-\frac{v^{2}}{c^{2}}}}
$$

Because $m$ is divided by this factor instead of being multiplied by it, the graph of mass against velocity is Fig. 3 upside down; at speeds not comparable with $c$ it doesn't differ appreciably from $m_{o}$, but ultimately it rises to infinity as the speed reaches $c$.

Obviously that means that nothing having mass can ever reach, let alone exceed, the speed of light. Photons themselves, which are light (I am still using that word to include all electromagnetic radiation), are the only things that can reach that speed, and they have no rest mass at all-if for no other reason than that they are never at rest! Now zero divided or multiplied by any finite number is still zero, but when the velocity $v=c$ then the Lorentz factor becomes zero, and $0 / 0$ can be anything. In the particular case of a photon it can be shown by other means that it is equal to $h f / c^{2}$, $h f$ being the "quantum" of energy of the photon, made up of the frequency of its radiation ( $f$ ) and Planck's constant ( $h$ ). The photon's energy hf is therefore equal to its mass multiplied by $c^{2}-$ a fact we'll return to presently.
We left our pie approaching the face of the man in the street with a velocity $v=0.8 c$. He, poor chap, can take no comfort from the fact that this is less than $c$, which is what it would have been according to what he had been taught at school, because he (having studied Einstein) knows that so far as he is concerned its energy is just as great as he had feared, owing to its increase in mass. So he wisely takes evasive action by retreating with velocity $0.8 c$. Relative to him the pie is now at rest, so its mass is $m_{o}$ only. (He notes, however, that the street and everything attached to it, flying past him at $0.8 c$, have become $66 \frac{2}{3} \%$ more massive).
Mass, then, joins length and time as a quantity which has no absolute value, but varies according to the relative speed between it and the observer. These three quantities, you may notice, are the three usually regarded as basic-hence the m.k.s. and c.g.s. systems of units.
That is not to say that the relativistic variation in mass is a sort of hallucination, with no reality. The nuclear physicists, who play about with particles at speeds close to $c$, are obliged to make very real and practical allowance for relativistic mass in the design of their large and expensive equipment. 300 kV , which is not very much nowadays, is enough to accelerate an electron to $0.8 c$, which makes quite a substantial difference to its mass.
It is fair to mention that an increase in mass of moving electric charges, such as electrons, had been predicted before the theory of relativity-by J. J. Thomson as far back as 1881, and with the actual relativity formula by Lorentz in 1904-but Einstein showed that this was just part of a general law applying to all masses.

If one calculates the kinetic energy given to a body in accelerating it from rest to velocity $v$ as the work done on it, taking into account this depen-
dence of mass on velocity, one arrives at the result

$$
\text { kinetic energy }=\mathrm{E}_{k}=\frac{m_{0} c^{2}}{\sqrt{1-\frac{v^{2}}{c^{2}}}}-m_{0} c^{2}
$$

For comparatively low velocities we can use the approximation

$$
\begin{aligned}
\mathrm{E}_{k} & =m_{o} c^{2}\left(1+\frac{v^{2}}{2 c^{2}}\right)-m_{o} c^{2} \\
& =\frac{1}{2} m_{0} v^{2}
\end{aligned}
$$

in accordance with school dynamics.
And in general, as $m_{0} / \sqrt{ }\left(1-v^{2} / c^{2}\right)$ is the variable mass $m$,

$$
\mathrm{E}_{k}=\left(m-m_{o}\right) c^{2}
$$

so the increase in mass due to acceleration from rest to velocity $v$ is

$$
\frac{\mathrm{E}_{k}}{c^{2}}
$$

In other words, the increase in mass is equal to the increase in energy multipled by the very small constant $1 / c^{2}$.

If the masses of the colliding elastic bodies are calculated for the instant when they are both at rest, their deformed shape signifying the stored potential energy which is just going to fling them apart, they are found to be greater by an amount equal to this potential energy divided by $c^{2}$. And if alternatively they are assumed to be perfectly inelastic bodies, like lumps of dough, there is again an increase in mass corresponding to the heat energy generated. Whatever the form any quantity of energy takes, it can be shown to be accompanied by $1 / c^{2}$ times that quantity of mass, over and above the mass of the same body without that energy.
The 64,000 dollar question-no; literally the 64,000 billion dollar question-that follows from this is: If these infinitesimal increases in mass represent changes in energy, what does the comparatively enormous rest mass $m$, represent?
Einstein's answer, expressed in the simple celebrated equation $\mathrm{E}=m c^{2}$, is that it represents an enormous amount of energy. And you don't need me to tell you how practically this affects every one of us now. The sudden destruction of quite a moderate fraction of $m_{o}$ is accompanied by the spectacular release of energy in a nuclear explosion, and its more gradual destruction is at this moment running electrical appliances in homes through the activity of Calder Hall.
So it seems that mass and energy are inseparable; two manifestations of the same thing-the basic stuff of the universe. When an electron gains in mass without anything being added to it except speed, that is not really a breach of the law of conservation of mass; it comes from the energy given to it. Whatever has given it that energy must have lost an equal amount of mass. Millions of tons of the sun's mass are streaming off every second into space as radiant energy. They are not lost, for the photons emitted have exactly that amount of mass, due to their speed, $c$. Some scientists believe that all the time the radiant energy of the universe is condensing back into matter.
To say anything at all after this picture, in which magnitudes range from the great universe itself working as a whole down to the activity of the minutest particles and waves, which are responsible for that working, would be an anticlimax. So I just stop, leaving plenty for the mind to follow up.

# Evaluating Aerial Performance 

## 2. Multi-element and Long-wire Systems:

Receiving Aerials: Matching

By L. A. MOXON, B.Sc. A.M.I.E.E.

(Concluded from page 65 of the February issue)

A$S$ pointed out in the first part of this article aerial gain can be calculated with the aid of tables of mutual impedance. This is, in general, complex, but the reactive component can be got rid of by choosing an element spacing of $\lambda / 8$, and the calculations then become quire simple and give the results plotted in Fig. 12 (Ref. 2). Note that maximum gain, 5.3 dB , requires a phase shift $\varphi$ of $0.6 \phi_{0}$ and the gain curve is equally valid for driven or parasitic beam systems except that in the former case it can be applied up to somewhat wider element spacings. With a parasitic reflector, the maximum gain falls to just over 4 dB at $\lambda / 4$ spacing, the corresponding figure for a director being only about 2.5 dB . A simple method of introducing the phase shift is to arrange that the two elements, together with the connection between them (Fig. 11 (b)) are exactly resonant, and move the feed point slightly off centre. This is equivalent to lengthening one element and shortening the other by the same amount, and inserts equal positive and negative reactances, the values of these being given by the $2 \pi b Z_{0}$ formula of Part 1 . The resulting phase shift is given in degrees for $\dot{\lambda} / 8$ spacing by $120 \pi b Z_{0} / R_{r}$, where $\mathrm{R}_{\mathrm{r}}$ is the radiation resistance of the individual dipoles. A number of practical arrangements based on this principle have been described elsewhere ${ }^{2,9}$. It follows from the mode of operation outlined above, that Fig. 12 can be drawn as a universal set of curves

Fig. 12. Variation of gain, radiation resistance and front-toback ratio with phase angle for pairs of close-spaced, end-fire elements. $\phi_{0}$ is the phase angle corresponding to the spacing. The phose difference between the element currents is $180^{\circ}-\phi$.

for pairs of close-spaced equal-current elements, the use of $\lambda / 8$ spacing for the calculations being merely a subterfuge to simplify the algebra. It is further evident that it can be applied to elements of any shape so long as radiation takes place mainly in one plane, and provided that the dimensons remain sufficiently small in comparison with half a wavelength. It

Fig. 13. Single square loop as used in the "Quad" system. Arrows indicate direction of current flow. The portion efoab $=$ bcde $=\lambda / 2$. Figures indicate relative fields produced by each bit of the loop in accordance with Fig. 6(a). Main lobe of radiation in direction at right angles to plane of paper. Alternotive mounting is with
 diagonal vertical with feed point at lowest corner.
therefore holds, more or less, for arrangements such as the popular "cubical quad" aerial system. This usually takes the form of a pair of loops of the kind illustrated in Fig. 13; arrows show the direction of current flow, and also indicate, to scale, the relative magnitudes of the fields which each bit of the loop would produce on its own. It will be seen that radiation from the sides involves currents of opposite sign which tend to cancel each other. The two horizontal portions bear some resemblance to a folded dipole, but relative to this they give a $30 \%$ reduction of field strength in the vertical direction. This increases the gain slightly, but only to the extent of about 1 dB . This figure was obtained by regarding the loop as a 2 -element broadside array with rather close spacing for which handbook data is available, but it is also in reasonable accord with the radiation pattern. It is quite easy to show, with the aid of Fig. 6, that a $14-\mathrm{Mc} / \mathrm{s}$ quad will give comparable performance at frequencies up to $21 \mathrm{Mc} / \mathrm{s}$ in spite of a rather untidy current distribution which has given rise to a belief that the quad is a single-frequency system. A $21-\mathrm{Mc} / \mathrm{s}$ quad can also be used at $14 \mathrm{Mc} / \mathrm{s}$, the loops being roughly equivalent to shortened dipoles of about 28 ohms radiation resistance.

We now come to the interesting question of what happens with more than two elements. Obviously, the larger the number of elements the more variables we have at our disposal and the larger the number of directions in which the radiation can be made to cancel, to the advantage of those directions in which it does not quite cancel. It is at this point that the process acquires the label of "super-gain" and
becomes really fascinating. The gain theoretically obtainable ${ }^{3}$ is nearly equal to $\mathrm{N}^{2}$, which checks quite well with the figure of 5.3 dB which was obtained for $\mathrm{N}=2$. Beyond this point the practical difficulties increase rather rapidly, but there is nothing, in principle, to prevent the design of, say, an aerial the size of a matchbox with a gain of a million; for this purpose, however, the matchbox would have to contain 1,000 elements all with zero loss resistance, and the element currents would have to be adjusted to various different values of amplitude and phase with fantastic precision. Having achieved this the bandwidth of the array would be


Fig. 14. Current distribution in a long-wire aerial is shown at (a). Standing-wave pattern disappears if aerial is terminated by a suitable load. Average current is then same at all points, ignoring losses. In (b) long wire of (a) is represented by point sources of alternate sign. Radiation in direction $A C$ is zero when $A B-A C=\lambda / 2$, which occurs when $\gamma=\sqrt{ } 2 / 1$ where $l$ is in wavelengths and $\gamma$ is in radians. (c) is a typical radiation pattern of a long-wire aerial. Null directions indicated by arrows are those for which $\gamma=\sqrt{ } 2 /$.
so narrow that it would be impossible to use it for signalling at a useful rate. As another example, one author ${ }^{4}$ has designed on paper a 9 -element "broadside " version having a gain of 8.5 dB based on its directivity, but an efficiency of only $10^{-14}$ of onc per cent at $10 \mathrm{Mc} / \mathrm{s}$ assuming $\frac{1}{2}$-in diameter, $\lambda / 2-$ elements. To achieve this the currents have to be adjusted to better than 1 part in $10^{11}$ !
In practice there is some doubt as to how far the process can be carried usefully beyond two elements whi $: h$, as we have seen, present no problem. A 4element end-fire super-gain array has been constructed ${ }^{3}$, but although fairly wide spacing was used, making the dimensions $0.6 \lambda$ from back to front, the gain realized was only 8.7 dB out of the theoretical 10.2 dB , adjustments being critical and the bandwidth only $1 \mathrm{Mc} / \mathrm{s}$ at $75 \mathrm{Mc} / \mathrm{s}$. It seems difficult to reconcile this result with the high gains sometimes quoted for conventional parasitic arrays having three or more elements. Systems of this type are commonly known as Yagi aerials and can give large gains when their length is long in terms of wavelengths. Calculations for an idealized system of this type ${ }^{3}$ have
predicted a possible power gain of $1.8+5.6 l$ where $l$ is the length in wavelengths. The case of 3 -element parasitic arrays having equal spacing has been investigated by Walkinshaw ${ }^{6}$ who obtained a theoretical gain of just over 7 dB , but this entailed the very low figure of about 4 ohms for the radiation resistance, which would present difficulties of impedance matching, and a poor front-to-back ratio. On the other hand reference 3 contains a hint that some improvement might be achieved by the use of unequal spacing.

There is one important class of aerial which, at first sight, appears to require a different kind of treatment. This is the "long-wire aerial," the wires being usually several wavelengths long and arranged in pairs (V-beams) or groups of four (rhombics). Let us first consider the case of a very long wire as in Fig 14(a) with its far end insulated so that a wave travelling down the wire is reflected from the far end and a standing-wave pattern is set up containing numerous current reversals. Such a wire can be regarded as a row of dipoles of alternate sign, as in Fig. 14(b), and for most directions the majority of the dipoles can be paired off with others which give an opposing field so that the effective radiation tends to zero. On the other hand, for directions nearly in dine with the wire positive and negative dipoles have their centres nearly half a wavelength apart and thus produce fields which add in phase. The great length of the wire makes up for the low value of $\cos \theta$. Referring to Fig.14(c), note that the main lobes of the radiation pattern are sandwiched between the nulls given by $\gamma=0$ and $\gamma=\sqrt{2 / l}$, where $l$ is the length of the wire in wavelengths. The width of these lobes between half-power points is $0.5 \sqrt{2 / l}$ but a large proportion of the power is radiated in the minor lobes. The power gain is approximately equal to $l / 2$.
It is common practice, particularly with rhombics, to terminate the wires in their characteristic impedance, as in Fig.15, in which case there is no reflection and no standing waves, but the finite velocity of the wave travelling down the wire means that at any given instant there are current reversals and, in the direction of travel of the wave down the wire, the field is additive as before. There is, however, no radiation in the opposite direction since, considering any short section of the aerial, the radiation from it is cancelled by that from the corresponding section half a wavelength farther along, which is out of phase having originated half a cycle earlier; this has travelled one wavelength farther, i.e. half a wavelength out and back, and therefore remains out of phase, whereas for the forward direction it has a half-wavelength start


Fig. 15. Terminated rhombic aerial. It consists of four long wires arranged so that lobes in the direction of the arron reinforce each other.


Fig. 16. Illustration of aerial images: (a) due to horizontal dipole and (b) to vertical dipole. (c) represents end view of horizontal dipoles. Note that distance travelled by reflected roy is greater than that of direct ray by distance $A^{\prime} B$. Since aerial and image are of opposite sign fields reinforce each other when $A^{\prime} B=\lambda / 2$. Perfect "earth" is ossumed.
and arrives in the same phase. The energy not radiated backwards is absorbed in the terminating resistance and does not add to the forward-power gain but, conversely, on reception the termination absorbs half the noise power picked up on the aerial and may, therefore, result in a doubling of the signal-to-noise ratio.

Effect of the Ground.-The ground acts in general as a good reflector of radio waves and therefore, by optical analogy, the aerial must be considered as having an image as shown in Fig.16. This image is to be regarded as a duplicate of the aerial and modifies the radiation pattern so that, referring to Fig.16(c) the radiated field is zero along the ground or when $A^{\prime} B$ is an integral number of wavelengths, but is equal to twice the tree-space field when $A^{\prime} B$ is an odd number of half wavelengths. For intermediate angles of elevation the field must be worked out from the phase difference corresponding to the number of wavelengths in $A^{\prime} B$. Obviously the greater the height the lower the angle of the lowest lobe of radiation. Putting $2 h \sin \theta=\lambda / 2$, we find that $h$ for an angle of $5^{\circ}$ (which is a desirable angle for long-distance propagation at the higher frequencies) is just under three wavelengths. If, however, the ground slopes down at $25^{\circ}$ an angle of $30^{\circ}$ to the ground will give the required $5^{\circ}$ angle of elevation and requires an aerial height of only $\lambda / 2$. In practice a height of 20 ft only at $14 \mathrm{Mc} / \mathrm{s}$ with a ground slope of about $20^{\circ}$ was found to give near-optimum performance on the long route to Australia, an aerial near the bottom of the slope being just as good as one near the top provided both the aerial and the point of reflection were on the slope. Ground sloping in the desired direction is more or less equivalent to an increase of aerial height, and is a valuable asset for v.h.f. and u.h.f. as well as h.f. communication.

The effect of the ground is much more complicated in the case of vertical polarization. From Fig.16(b) one would expect the radiation to be concentrated at very low angles, but in general 'this is not the case in practice. The reason is that the
representation is only valid for a perfectly conducting earth, fairly high angles of radiation, or comparatively low frequencies. With normal ground it is found that as the angle of radiation is reduced more and more of the reflected wave is absorbed until an angle known from optical analogy as the pseudo-Brewster angle is reached. Below this the reflection coefficient increases again but the image is reversed in sign. This means that for low-angle radiation at the wavelengths used for long-distance communication, assuming level ground, the performance of a vertical aerial is roughly equivalent to that of a horizontal one at the same height.

With a centre-fed vertical aerial the feeder system is liable to present mechanical difficulties since it requires to be brought away from the radiator at right angles. One answer to this problem is the ground-plane aerial, Fig.17. Radiation from the horizontal wires cancels out in all directions, so that the radiatior is effectively half of a dipole. It has been shown that the radiation pattern of short elements is nearly the same as that of a half-wave dipole regardless of the precise shape of the current distribution and the ground-plane aerial is no exception to this. Being half the length, however, it requires twice the current to produce the same field and must therefore be assumed to have a radiation resistance of 18.3 ohms, although the figure usually quoted is slightly higher, about 22 ohms. It is important to realize that the so-called ground plane is merely a device for matching the aerial to an unbalanced feeder; it is not a substitute for the actual ground, and has no reflecting properties.

In certain handbooks a distinction is made between the "free space directivity gain" with which we have been concerned hitherto, and the

Fig. 17. Ground-plane aerial. $O A$ is a $\lambda / 4$ vertical radiator, $O B_{1}$. $O B_{2}$ etc. are three or four equally spaced horizontal radials, each $\lambda / 4$ long.

"practical DX signal gain." This appears to be nowhere fully explained, but seems to arise in part from the practice of comparing vertical stacks of dipoles with single dipoles at the same average height. Because of the relatively complicated geometry, the nulls in the vertical pattern may not occur at exactly the same angle in both cases, and large differences either way may be observed at certain angles of elevation. There is another effect which may account for apparent changes of gain with height, amounting to about $\pm 1.5 \mathrm{~dB}$ for aerial heights greater than $0.18 \lambda$; this is due to the mutual coupling between the aerial and its image, which causes the radiation resistance of a horizontal dipole to vary with height between limits of about 60 and 100 ohms. This causes a variation in current, and therefore of the field at a distance. Beam aerials tend to radiate less energy upwards and downwards which means in general that there is less coupling between the aerial and its image; the " 8 JK " aerial
provides a simple illustration of this because the two image elements must obviously induce nearly equal and opposite currents in each of the real elements, assuming that, as usually is the case, the spacing is small compared with the height. This almost completely removes the variation of radiation resistance with height, the apparent changes in gain being due to the reference dipole. Another possible cause of gain variation, in reception, is nonuniformity in the field surrounding the aerial. This could be due to reflections from neighbouring objects, e.g. telephone wires, and would of course affect the gain equally in the case of tranmission. This effect is one of the causes of error in gain measurements as discussed by Strafford ${ }^{1}$. These effects will usually be small and may work in either direction. There is no justification for the belief that beam aerials may give large gains due to the "lowering of the angle of radiation" in addition to their free space gain.

The Effective Keceiving Gain of an Aerial.-With modern receivers the range of reception is normally limited by the strength of the background noise, and not by lack of sensitivity. This means that aerial gain is of no value as such, and it is quite common to sacrifice gain in order, for example, to improve the back-to-front ratio.

The problem in reception is to achieve the best possible ratio of wanted signal to noise or interference. For this purpose the proper criterion of aerial performance depends on the nature of the unwanted background and may consist of the power gain, the directivity gain (which is the same as power gain if there are no losses), the depth of nulls in the radiation pattern, or more probably some rather complicated function of the directivity. These cases will be considered in turn, and it will be shown


Fig. 18. Effect on back radiation of alternative phasing. Angle over which field is reduced to $10 \%$ is greater when goin is higher although nominal back/front ratio is poor.
that, as an index of performance, the nominal back-to-front ratio is more likely to be misleading than otherwise.

At frequencies greater than $100 \mathrm{Mc} / \mathrm{s}$ or so signals are mainly interfered with by noise generated in the receiver, and the signal-to-noise ratio is then directly proportional to the power gain of the aerial. This situation may be affected in the foreseeable future by the introduction of new low-noise amplifiers such as masers and parametric devices.

A very similar situation exists if the sources of noise are external to the receiver but are uniformly distributed in space. Halving the aerial beam width
halves the number of sources which contribute to the noise level but doubles the aerial gain and therefore the strength of those noise sources which are within the beam. The received noise power is therefore independent of beam width, and signal-tonoise ratio is again directly proportional to aerial gain, but this time it is the gain as calculated from the directional pattern, losses in the aerial or feeder system being of no account since they act equally against signals and noise. The losses cannot, however, be allowed to increase indefinitely since, depending on the noisiness of the receiver, a point will eventually be reached where reception is limited by receiver noise or the thermal noise of the aerial loss-resistance ${ }^{7}$. This case is typical of normal reception against a background of galactic or atmospheric noise, although th.e spatial distribution of such noise is not strictly uniform and this may sometimes have to be taken into account.

The next case to be considered arises when reception is limited by noise or interference from one particular source. Provided the two signals are not in the same direction, the unwanted one can in principle be phased out with at worst some reduction in the level of the wanted one. This can be done either by means of two aerials, with appropriate adjustment of relative phases and amplitudes, or by making use of nulls in the radiation pattern. These nulls occur in the end-on directions and in other directions depending on the phase shift; Fig. 12 shows one such null occurring in the $180^{\circ}$ direction for a phase shift of $\phi_{o}$, which is the much sought after "infinite front-to-back ratio" condition. Referring now to Fig.18, it is interesting to observe what happens as $\phi$ is reduced; the nominal front-to-back ratio drops to about 4 to 1 but two nulls have now appeared instead of one. The gain is higher, which is useful if the aerial is also to be used for transmission, and moreover, we are free to select whichever null gives least reduction of the wanted signal relative to noise or other interference. One obvious application of this technique would be in television reception for the removal of ghost images caused by one or two indirect signal paths.

If interference is likely to come equally from all direction it might be thought reasonable to treat it as omni-directional noise, and the ratio of average signal to average interference power will of course be equal to the directivity gain. This is not, however, the correct approach if it is required, for example, to separate weak wanted from strong unwanted signals. It is obviously useless to reduce the strength of an unwanted signal merely by a factor of 10 if it is 100 times stronger than the wanted signal! The objective therefore is not to reduce the average level of interference but rather to reduce the number of occasions on which unwanted signals exceed the level of wanted ones. A deep but narrow null in the radiation pattern, unless it can be moved at will, is of very little use for this purpose; on the other hand, reduction of say 10 to 1 or more in field strength over a wide range of angles is of great value and by this criterion the higher gain pattern shown in Fig. 18 is slightly better. In one typical case, the reception of amateur signals from Australia via South America in the presence of short-skip interference from Europe, the author has found a parasitic beam to be just as good in practice as a driven arrangement, despite the in(Continued on page 143)
equality of element currents which causes partial filling in of the nulls. This is, of course, to be expected if angle is more important than depth of rejection.

Interfering signals vary in amplitude between very wide limits; they are almost as likely to exceed a weak wanted signal by 20 dB as by 10 dB , and may well be as much as 60 dB stronger, so that to make a large reduction in the number of occasions on which interference occurs the response of the aerial in unwanted directions must also be reduced by a large amount. Higher gain means that weaker signals can be received and correspondingly greater discrimination is therefore required against strong unwanted signals. These arguments underline the need to reduce the general level of side lobes of high-gain receiving aerials, even if this has a negligible effect on gain.
Feeder Losses.-Consider the case shown in Fig. 19 of a source of power having an internal impedance $R_{1}$ connected to a load $R_{2}$ through a line having a characteristic impedance $\mathrm{Z}_{o}$. If $\mathrm{R}_{1}=\mathrm{R}_{2}=\mathrm{Z}_{o}$, and there are no losses in the transmission line, all the power available will be delivered to the load. If now the load is changed to some new value, power is reflected from the load and on its way back, interacts with the wave travelling towards the load to produce a standing wave, i.e., varying amplitudes of current and voltage along the line. If the load remains resistive, the new value being $R_{2}^{\prime}$, the ratio of maximum to minimum current along the line is equal to $R_{2} / R_{2}^{\prime}$ or $R_{2}^{\prime} / R_{2}$, whichever is greater. This ratio or its reciprocal is known as the standing wave ratio or s.w.r. The alteration made to $R_{2}$ means that the generator row "sees" a less suitable value of load impedance and delivers less power, the mismatch loss being given as a power ratio ${ }^{2}$ by $(1+\sigma)^{2} / 4 \sigma$ where $\sigma$ is the s.w.r. This loss is significant in cases such as TV reception since, in the absence of matching adjnstments at the receiver, it reduces the ratio of signal to internally-generated noise. It is not, however, applicable in the case of a transmitter supplying power to an aerial, since the mismatch is usually taken care of automatically by adjusting the aerial coupling to obtain correct loading of the transmitter. When this is done, it follows that if the line has zero resistance all the power leaving the transmitter must reach the aerial because there is nowhere else where it can be dissipated. The distinction frequently made between a "matched" and a "resonant" feeder, is not basic, as sometimes represented, but related in some arbitrary manner to the degree of mismatch coupled with the absence of "marching devices." A typical example of a "resonant" feeder system would be an open-wire line of about 600 ohms impedance feeding a half-wave dipole, the s.w.r. being then given approximately by $600 / 73$, i.e., just over 8 to 1 . Sometimes a resonant feeder is used from the aerial to some point such as ground level, at which adjustments can be made, and it is there matched into a non-resonant line which can be as long as necessary; one such case is recalled in which a certain successful aerial system having an s.w.r. of possibly 100 to 1 in the "resonant" portion was publicly criticized, without contradiction, on the grounds of a 3 to 1 s.w.r. in the non-resonant part of it!
Generally speaking the intensive efforts so frequently made to reduce s.w.r. to near unity, in the case of amateur transmitting aerials, bring very
little reward in the shape of increased radiation efficiency. High s.w.r. does not in itself introduce losses, although it does accentuate any series resistance losses which may be already present in the line. The reason for this is apparent when we recollect that losses in a resistance are proportional to the square of the current. Series resistance tends to be uniformly distributed along the line, and an s.w.r. of 4 means that the current varies between twice and a half its value for the matched condition; there is thus a 4 to 1 increase of loss in parts of the line, and this is counterbalanced to only a small extent by the reduction of losses to a quarter at the current minima. For standing wave ratios up to


Fig. 19. Source of r.f. power connected to a load via a transmission line.
3 to 1 the extra loss in decibels never exceeds $40 \%$ of the matched-line loss, and reaches a maximum value of about 1.2 dB when the matched-line loss is large. At 5 to 1 s.w.r. the corresponding figures are $100 \%$ and 2.5 dB .
Other disadvantages of resonant or mismatched feeders include narrow bandwidth, as previously discussed, and in some cases the possibility of voltage breakdown or excessive heating. The reduced overall bandwidth may be unacceptable in the case of television reception, particularly as it is likely to be associated with picture defects caused by multiple reflections in the line, and in amateur transmission it can be a nuisance owing to the necessity of retuning the transmitter when small changes of frequency are made. A high s.w.r. increases the dielectric and leakage losses in a lowimpedance feeder, in much the same way as it increases the series loss since the voltage wave along the feeder varies in a similar manner to the current wave although displaced from it by $\lambda / 4$. In the case of an open-wire line, however, standing waves can sometimes be used to reduce losses of this type by placing the insulators at low-voltage points in the system.
The following example of a 700 -ohm line in use by the author provides a simple illustration of the calculation of feeder losses. The length is 100 metres and the wire is 20 s.w.g. which is much thinner, and also much cheaper, than the 14 s.w.g. normally recommended. From Fig. 5, the loss per half wavelength at $15 \mathrm{Mc} / \mathrm{s}$ ( 20 metres) for one wire is 3.4 ohms. The total loss resistance therefore is equal to this multiplied by 20 , i.e., 68 ohms. With the line matched, this loss-resistance can be regarded as more or less in series with the terminating resistance so that $1 / 11$ of the total power into the line is lost. This amounts to just over $1 / 3$ of a decibel, rising to 0.5 dB at an s.w.r. of 3 . At $28 \mathrm{Mc} / \mathrm{s}$ the loss rises to 0.5 dB , if the line is matched, or $2 \mathrm{~dB}-$ just noticeable in terms of signal strength reportsif the line is used as a resonant feeder to a half-wave dipole. There is, of course, some loss by radiation from an open-wire feeder, but this is usually very small and would amount in the above example to an additional loss resistance of only 1 or 2 ohms.

Out-of-balance currents can, however, cause a considerable loss by radiation, and can be caused by anything which disturbs the symmetry of an aerial system. When coaxial cable is used for transmitting it is important to ensure that no currents flow on the outer of the cable, in other words voltage should not be induced from the transmitter, or the radiator itself, into the aerial-to-ground path formed by the outer conductor. This is, of course, similar to the requirement, for reception, that currents induced in the outer conductor must not be allowed to flow through the receiver input circuit. To this end the feeder should be brought away from the aerial for a considerable distance at right angles, but in the case of a vertical dipole of tubular construction it can be brought down through the centre of the lower half. An approximation to this last arrangement is obtained with a ground-plane type of construction when the radials are allowed to slope downwards. The lower end of the feeder should be firmly earthed to the case of the transmitter or receiver. As a further precaution the path to earth via the outer conductor can be made non-resonant by adjustment of feeder length.

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## BOOKS RECEIVED

The Services Textbook of Radio. Volume 5 Transmission and Propagation by E. V. D. Glazier, Ph.D. (Eng.), B.Sc., A.M.I.E.E. and H. R. L. Lamont, Ph.D., M.A., A.M.I.E.E. Deals with the propagation of electromagnetic energy on transmission lines, in waveguides, and in free space. Practical as well as theoretical aspects are covered and chapters are devoted to waveguide components and couplings and to aerial systems for all useful frequency bands. The text is arranged and marked for reading at elementary, intermediate and advanced levels. Pp. 500; Figs; 425. Price 25s. H.M. Stationery Office, York House, Kingsway, London, W.C.2.

Basic Electricity by Van Valkenburgh, Nooger and Neville, Inc. Treatise in five parts designed originally as a course of training for U.S. Navy technicians with no previous knowledge of electricity. The text is profusely illustrated with simple diagrams in which the approach is often anthropomorphic. The present series has been Anglicized and adapted for British and Commonwealth use by a team of the Royal Electrical and Mechanical Engineers. Pp. 120 (approx.) per volume. Price 12s 6d per part or 55 s per complete set. The Technical Press Ltd., 1 Justice Walk, London,
S.W.3.

Basic Electricity for Communications by W. H. Timbie and F. J. Ricker. Second edition of a textbook on d.c. and a.c. circuits and conduction in gases and semiconductors. Numerous worked examples show the application of basic laws. Pp. 538; Figs 450. Chapman \& Hall, 37 Essex Street, Strand, London, W.C.2.
International Radio Tube Encyclopædia by Bernard B. Babani. Third Edition (1958-59) containing data of 27,500 types including transmitting and microwave types. Pp. 768. Price 63s. Bernards (Publishers) Ltd., The Grampians, Western Gate, London, W.6.

Tube and Semiconductor Selection Guide 1958-59. (Philips' Technical Library). Compiled by 'Th. J. Kroes. Simplifies selection of preferred types in the Philips range for any given purpose and indicates possible replacements for obsolete types. Pp. 160. Price 9s 6d. Obtainable through the Cleaver-Hume Press, Ltd., 31 Wrights Lane, London, W.8.

## SHORT-WAVE CONDITIONS



G.M.T

THE full-line curves indicate the highest frequencies likely to be usable at any time of the day or night for reliable communications over four long-distance paths
from this country during from this country during March.
Broken-line curves give the highest frequencies that will sustain a partial service throughout the same period.


GMT

# The Bifilar-T Circuir 

# An Important Filter Investigated from First Principles 

(Concluded from page 71 of February issue)

WE saw last month that the bifilar-T circuit is not really novel at all, but is, if used properly, an economical way of building a perfectly ordinary filter using a long-known equivalent circuit. The shape of the rejection peak is iust what you can get with any other equivalent of the same basic network, provided that resistance cancellation is used to push the attenuation up. Just to see what happens let us take the simplest network which might do the job and turn it into a lattice by means of Bartlett's Bisection Theorem. There it is, in Fig. 15, large
a resonance, and displace the standard curve accordingly.
I must digress for a moment. A couple of paragraphs back I made use of Bartlett's Bisection Theorem, comforting myself with the knowledge that it is in the books for you to look up if you will not trust me. Unfortunately it is only in some of the books, and anyway we do not really need it here. Look back to Fig. 15(a) and imagine, if you will, that a perfect centre-tapped coil, of infinite inductance and unity coupling between the two

(a)


Fig. 15. Applying Bartlett's Bisection Theorem to (a) we get the lattice shown in (c).
as life and twice as natural. If you compare this with Fig. 11 (last month's issue) you see that you have lost the freedom to produce the curve in Fig. 11(a) with $\mathrm{C}_{\mathrm{B}}$ less than $\mathrm{C}_{\mathrm{A}}$, because that $\mathrm{L}_{2} \mathrm{C}_{2}$ must produce a real frequency of infinite attenuation. This is a normal limitation when a ladder is used in place of a lattice. It does not worry us, though, because we do want that peak. The circuit values are all positive, of course, and it is easy to see from what we calculated above that $L_{2}$ should be about $1 / 10$ th of $\mathrm{L}_{1}$.
The reason why negative element values are not needed is that I have been quite content to neglect the sense of the output terminals. The bridge shown in Fig. 16 shows that there is not any real justification for associating one input terminal with one output terminal. I do not feel sufficiently interested in this matter of phase reversal to work out the answer. If you happen to care, the shape of the image phase characteristic is in the book and all you have to do is work out the phase shift at one convenient frequency, such as zero, or infinity, or
halves, is added in parallel with $\mathrm{L}_{2} \mathrm{C}_{2}$. Obviously this cannot alter the conditions at all; since the inductance is infinite it cannot affect the behaviour of the circuit in any way. However, we can now carry out the operation shown in Fig. 14 and transform $\mathrm{L}_{2} \mathrm{C}_{2}$ into a lattice form, in which, of course, $Z_{b}$ is infinite. Then using the transformation of Fig. 5 we bring the $\mathrm{L}_{1} \mathrm{C}_{1}$ 's into the lattice. The only difference is that this time, since I did not have

Fig. 16. Drawing the lattice as a bridge emphasizes the essential symmetry of the arms.



Fig. 17. Adding $C_{3}$, here split to p.eserve the symmetry, to the network of Fig. 15 will give a worth-while improvement.


Fig. 18. This configuration is an equivalent for the series orm in Fig. 17.
to try to remember Bartlett's Theorem, the output terminals are crossed over.

The digression in the last paragraph was intended to show how these simple equivalences can be used to establish circuit conditions very easily. I do not think it is really necessary for me to show, by exactly the same reasoning as that used in the last paragraph, that a resistance $R$ connected from the centrepoint of $L_{2}$ can be called $Z_{b} / 2$ in Fig. 14 and thus introduced as 2 R across each of the lattice arms made up of $L_{1}$ and $C_{1}$ only. This resistance is used to balance the loss in $\mathrm{L}_{2}$. You can do this calculation by a T-T transformation, but although I actually did this in these columns a good many years ago, the lattice treatment is much simpler and much more elegant.
From the discussion above it seems fairly clear that the bifilar-T, to do the job claimed for it, needs these end circuits and that it is, in fact, nothing more or less than a rather complicated way of making a simple fult section of an unsymmetrical band-pass filter. I do not imagine the phase reversal is important, but an extra winding on the inductance at either end would do this equally well and it would be possible to save one element. If the phase reversal is not needed the element saved is a double-wound coil, which, although not expensive, must still cost something.

It is thus rather a problem to find the advantage
of the bifilar-T network. I would expect the values to be more convenient in the $\pi$-network, because the inductance ratio is halved. Tuning in the $\pi$-network is more direct, with the top arm actually tuned to the notch frequency. The near edge of the pass-band is then fixed by the inductance ratio, but the tuning of the end circuits does not affect the notch. With the bifilar-T all the elements seem to control the position of the notch. The actual shape of the characteristic is the same for both circuits if they are designed to the same rules. Even the limitations on resistance cancellation seem to be the same. I must confess that I did not expect to reach this conclusion when I began to examine the bifilar-T, but I am now convinced that it is just a dreary old filter circuit dressed up in a new package. The chief feature of the new package is, indeed, the complete lack of design data.

The effect of leakage inductance in the bifilar coil, especially if one were to choose to work with less tight coupling, and of an appropriate capacitance in the top arm to provide both d.c. blocking and an additional impedance element, might be considered. After sketching out the appropriate variants on Fig. 11 I decided that the results were not of sufficient interest to discuss in any detail. The chief reason for this is that these extra elements appear in both arms of the lattice and thus lead us to rather restricted structures. It is for the reader to follow this up if he will, but I fear he will find that his principal satisfaction came in travelling hopefully.

Two topics seem to follow naturally on from the discussion of the bifilar-T and its equivalent ladder network. The first is the question of slightly more elaborate notch systems, elaborate in the sense of having more complex reactance diagrams. The obvious step is to add a single capacitance element in the series arm of the filter. To get the symmetry needed for ideal resistance cancellation, if we really want this, the capacitance must be split, and in drawing Fig. 17, and its footnote Fig. 18, I have assumed that we shall want this symmetry. I am not showing the equivalent lattice network diagrams as a figure because here again we have one of the standard filters of network theory. This structure, indeed, provides what is called a confluent band-pass characteristic, which means that in general it has two pass bands but that in practice you always arrange them side by side, flowing together. Although it is not a symmetrical filter, it is more symmetrical than the one we have been discussing up to now. Whether the top arm of Fig. 17 is used or whether you use Fig. 18 is something you have to calculate for each application. Sometimes it is convenient to take the coil shunt capacitance directly into the network, as in Fig. 17: sometimes Fig. 18


Fig. 19. Two variants of a circuit sometimes used. In practice the compensation resistonce must not be allowed to leak anode
voltage to the following grid.


Fig. 20. The circuit of Fig. 15 (a) may give awkwardly high values of $L_{2}$. This is one way of getting a lower value.


Fig. 21. These two equivalences may often be used to provide more easily realized elements.
gives more convenient values. The arrangement shown in Fig. 18 is slightly more convenient for adjustment, since $L^{\prime}{ }_{2}$ can be tuned with $\mathrm{C}_{3}^{\prime}$ to the centre of the pass-band, which is also the antiresonant frequency of $L_{1} \mathrm{C}_{1}$, and then $\mathrm{C}_{2}^{\prime}$ is tuned to give the correct peak frequency. In Fig. 17 we must either alter $\mathrm{C}_{3}$ or dodge backwards and forwards between $\mathrm{L}_{2}$ and $\mathrm{C}_{2}$, each of these affecting both bandcentre and peak frequencies. All the design formule are in the standard reference books.

There is a considerable temptation at this point to discuss those natty little shunt absorber circuits which some designers like. The two variants shown in Fig. 19 are electrically equivalent. Although they look fairly simple they do take slightly longer in computation, still using the standard books, than the network of Fig. 17, though I do not think there is much to choose in performance. The reason for the extra work is that you must first consider $L_{1} C_{1}$ and part of C as a half-section of a filter, then treat
the middle bit with the rest of each C and the shunt arm as a full section of a different kind of filter, and then put the whole lot together. Mind you, it is nothing to the time you could spend messing about measuring characteristics with lots of different coils if you did no calculations at all. Time spent in reconnaissance, the manual says, is seldom wasted, and nowhere is that more true than in circuit design.
There are, no doubt, other ingenious notch circuits which may be considered, but I cannot offhand think what they are. In ordinary i.f. use they will all be connected between tuned anode and tuned grid circuits and I shall be very surprised if they do not all turn out to be quite conventional filter networks. In audio-frequency work we sometimes wish to put a notch into a system of such a large bandwidth that this approach is not appropriate. Frequently, too, we are actually working between resistive terminations. Here we have the case of a first order filter designed to provide a stop band and I have already covered the theory of this, the simple first order filters and the frequency transformations, in "Filters Without Tears" (August, September, November and December issues, 1954).
Sometimes one hears the objection that the design of these interstage networks by conventional filter theory leads to impossible values of components. The objector is usually a man who has muddled up a circuit by trial and error and has then declared himself happy with a much poorer performance than the one he specified for his calculations. If we take the circuit of Fig. 15(a) as an example, there is rarely any trouble about $L_{1}$ because this is just the usual anode or grid inductance we have in any ordinary amplifier. It is always $L_{2}$ which causes the trouble. One way in which the vaiue of $L_{2}$ may be reduced is shown in Fig. 20. For example we might centre-tap $\mathrm{L}_{1}$ : the impedance level would be reduced to one quarter, so that $\mathrm{L}^{\prime}{ }_{2}=\mathrm{L}_{2} / 4$ and $\mathrm{C}_{2}^{\prime}=4 \mathrm{C}_{2}$.
The other way which can sometimes be used to make component values more convenient is to use a capacitance transformation. It was obvious in Fig. 20 that what we really did was introduce an ideal transformer in parallel at a different impedance level from the ends. If we consider the first pair of networks in Fig. 21 we can write

$$
\begin{aligned}
& \mathrm{V}_{1}=j \omega\left(\mathrm{C}_{3} / \mathrm{C}_{2}\right) \mathrm{V}_{2}+1 / j \omega \mathrm{C}_{2} \mathrm{I}_{2} \\
& \mathrm{I}_{1}=j \omega\left(1+\mathrm{C}_{1}\left(1+\mathrm{C}_{3} / \mathrm{C}_{2}\right)+\mathrm{C}_{8}\right) \mathrm{V}_{2}+\left(1+\mathrm{C}_{2}\right. \\
& \text { and } \\
& \mathrm{V}_{1}=n \mathrm{~V}_{2}+1 / j \omega n \mathrm{C}_{5} \mathrm{I}_{2} \\
& \mathrm{I}_{1}=1 / n \mathrm{I}_{2}
\end{aligned}
$$

There are several ways of arriving at these equa-

Fig. 22. An ideal transformer is introduced into the middle of a conventional half-section bandpass filter.

tions and I do not want to use up space on such a straightforward operation. If now we put $\mathrm{C}_{2} / \mathrm{C}_{3}=$ $\kappa_{3}$ and $\mathrm{C}_{2} / \mathrm{C}_{1}=\kappa_{1}$ we can go on to demand that both pairs of equations should be identical. Again I skip the algebra to tell you that the answer in first stage of boiling down is:

$$
\begin{gathered}
n \mathrm{C}_{5}=\mathrm{C}_{2} \\
\frac{\kappa_{1}}{\kappa_{3}} \frac{1+\kappa_{3}}{1+\kappa_{1}}=n^{2} \\
\text { and } 1+\kappa_{1}+\kappa_{3}=0
\end{gathered}
$$

This last equation implies that either $C_{1}$ or $C_{3}$ must be negative. In fact

$$
\begin{aligned}
& \kappa_{1}=n /(1-n)=-n /(n-1) \\
& \kappa_{3}=-1 /(1-n)=1 /(n-1)
\end{aligned}
$$

Thus if $n$ is greater than unity $\mathrm{C}_{1}$ is negative, and if $n$ is less than unity $\mathrm{C}_{3}$ is negative. The equations for the other transformation are obtained in the same way and you will find them in Shea's book. The negative capacitance is rather a nuisance though it should not cause any more alarm than does the $-M$ in a conventional transformer. But you do have to find a positive capacitance to marry it in with. In the sequence shown in Fig. 22 an ideal transformer is put into the middle of a constant $-\kappa$ half-section band-pass filter. Ideal transformer plus $\mathrm{C}_{5}$ is equivalent to the $\pi$. of capacitors $\mathrm{C}_{1} \mathrm{C}_{2} \mathrm{C}_{3}$
shown in the third diagram. If we have a step-up in the transformer $\mathrm{C}_{3}$ will be negative, because $n$ will be less than unity. We can choose the value of $n$ to make $\mathrm{C}_{3}=-\mathrm{C}_{6}$, so that the parallel end capacitance vanishes and we are left with the very simple network you see in Fig. 22. In one particular design this gave a ten-to-one step-up and was terminated in the losses of the end inductance.

The other topic which arises from consideration of the bifilar-T is the general use of coil-pairs in filter networks. I propose to treat this as the subject for a separate article because I have to meet a man in a glacier to morrow, and anyway I do want my notes.
In this study of the bifilar-T trap and matters which arise from it I hope I have conveyed to some readers at least the importance of making full use of the elementary filter theory which is in every book. The transformations used for twisting the circuits around are not difficult and once you have twisted the circuit into a standard form you can always find equations for the values in the standard references. If you want a superior design there is a great deal of superior filter theory available. Certainly you will not do any better by just using coils you have handy and hoping for the best, and if you cannot put numbers into a simple formula you just ought to give up design work.

# An Interim Statement on the International Geophysical Year 

By T. W. BENNINGTON*

IN a booklet published by the Royal Society $\dagger$ an interim statement is made concerning some of the I.G.Y. achievements. The main impression one gathers from this is that, during a period of record high sunspot activity which provided a unique opportunity for the study of solar-terrestrial relationships, a vast programme of observational work was satisfactorily accomplished, but that it is much too early yet to attempt to discern even the outstanding results of this work. Nevertheless matters of interest have arisen :n each of the 15 scientific subjects studied in this country, and among them are some closely connected with radio.
The arrival in the earth's atmosphere of ultraviolet rays, X-rays and charged particles from the sun is of importance in radio because of the effects of these radiations upon the ionosphere, both in maintaining the structure of the ionized layers and in disrupting it. The latter aspect was studied by means of special observations made when it seemed likely that exceptional outbursts of solar radiation would occur. To ensure this 38 sequences of "alerts" were issued on a world-wide basis, and these culminated in 43 "special world intervals", during which the observations were made. Not all of these did, in fact, coincide with unusual solar activity, though it seems likely that a few exceptional outbursts were fully observed.
A vast measurement programme has been carried out upon the ionosphere, which included vertical sounding by pulsed radio signals, studies of absorption using radio waves from radio stars and reflec-

[^6]tions from meteor trails, and measurements of atmospheric noise produced by lightning flashes. At Halley Bay, Antartica it was observed that in winter, despite the fact that at the layer height the sun never rises, the $\mathrm{F}_{2}$ electron densities at noon are ten times greater than those at midnight, whilst in summer, when the sun never sets, the maximum density at noon is less than that at midnight. The maximum density at noon is, in fact, greater in winter than in summer, whilst at the equinoxes there is a sudden change from winter to summer conditions. It is concluded that, since the direct ionizing action of the sun is small, these variations in ionization are mainly due to movements in the ionosphere, and that in winter the layer is replenished by horizontal movements of ionization.

Studies of solar activity made in the U.S.A. and U.S.S.R. appear to indicate that when a solar flare occurs-such as may give rise to a sudden ionospheric disturbance-there is a sudden conversion of magnetic energy in the sun into wave and particle energy, for the magnetic lines of force undergo a sudden redistribution.

Many data on the earth's magnetic field have been obtained, the main use of which will begin when they can be correlated with simultaneous observations of aurora, the icnosphere and activity on the sun, and some of which may help to define the position of the electric currents in the high atmosphere which are responsible for geomagnetic-and ionos-pheric--disturbances. The auroral observations appear to indicate that displays of aurora australis and aurora borealis progress very similarly, and that the aurora penerates farthest towards the equator in the "summer" hemisphere.

# News from the Industry 

Thorn-Philco-Under an agreement with Philco International Corp., of New York, all its "radio and monochrome television engineering knowledge, designs and developments will become available through a licence to Thorn Electrical for manufacture and sale of these products in the United Kingdom." Thorn also acquires all the issued capital stock of Philco (Overseas), Ltd., Philco's manufacturing unit in this country, which also owns the U.K. sales company Philco (Great Britain), Ltd. Thorn Electrical will manufacture export receivers under the Philco trademark for Fhilco International's overseas distribution. Thorn will also manufacture and sell sound and television receivers and radiogramophones, under the Philco trademark in the U.K. Philco is now added to the names of Ferguson, H.M.V., Marconiphone, Champion and Avantic, already in the Thorn group.

Cossor Radar and Electronics, Ltd., have moved from Highbury, London, to new premises in Harlow New Town, Essex. The company has for some time occupied two small factories in the New Town, and to these has been added a large third factory. The site covers some nine acres and the factory, which has a floor-space of 95,000 square feet, houses the principal research and development laboratories and workshops, the main production unit, and the administrative and sales offices.

International Computèrs \& Tabulators, Ltd., is the title under which the recently merged British Tabulating Machine Co. and Powers-Samas Accounting Machines will trade. In the field of electronic calculators and computers both of the original companies are associated with concerns specializing in electronic developments. B.T.M. are associated with the G.E.C. (they jointly own Computer Developments, Ltd.) and Powers-Samas with Ferranti. B.T.M. also have a link with Laboratory for Electronics, of Boston, U.S.A. I.C.T.'s main manufacturing establishments are at Letchworth, Croydon and Castlereagh (N. Ireland).

Ferranti-designed klystrons are to be manufactured in the U.S.A. under an agreement concluded between Raytheon Corporation, Ltd., of Boston, Mass, and Ferranti, Ltd. A sum of $\$ 250,000$ is involved in this sale of British" know-how" to the United States of America. The agreement involves two tubes used in Doppler radar systems. They are intended for use in military equipment and production in the U.S.A. is expected to begin early in 1960.

Marconi's have received a contract from the Ministry of Supply for v.h.f. direction finders for a number of civil airports and airfields in the U.K. The type ordered is AD 210 C , the first of a new series of automatic d.f. equipment developed by Marconi's.


Sperry Gyroscope Co., of Brentford, Middlesex, announce a substantial re-organization. The main objective is a de-centralization of management and the establishment of three separate operating divisions each with its own sales, design and manufacturing organization. These three divisions will be known as the Brentford, Bracknell and Industrial Divisions. The Brentford Division will be responsible for the company's aeronautical, marine and naval activities, the Bracknell Division will concentrate on Government contract work for guided weapons and inertial navigation, and the company's interests in industrial control engineering will be concentrated in the Industrial Division. M. L. Jofeh, formerly the company's chief enginecr, is manager of the Industrial Division, Wing Commander J. C. G. Bell, manager of the Brentford Division, and H. B. Sedgfield, manager of the Bracknell Division.

Radio Rentals.- The net profit of the Radio Rentals Group for the year ended last August, after deducting £911,947 for taxation, totalled £824,940-an increase of $£ 219,221$ on the previous year. The group's manufacturing subsidiary is Mains Radio Gramophones, Ltd., of Bradford, Yorks.

Burnhope.-All the main transmitting equipment, including the aerial system, for the recently opened I.T.A. station at Burnhope, near Durham, was supplied by Marconi's. The Burnhope transmitters (two $4-\mathrm{kW}$ vision and two $1-\mathrm{kW}$ sound) are identical to those used at Chilterton Down.
E.M.I. Vidicon film-scanning equipment to the value of over $£ 30,000$ has been supplied to Tyne Tees Television, the programme contrestors for the Burnhope station.

20th Century Electronics, Ltd., of King Henry's Drive, New Addington, Surrey, have appointed Peter Holton to take charge of their Photomultiplier Applications Advisory Service. Mr. Holton was until recently in charge of testing and application of photomultiplier tubes in the E.M.I. Photomultiplier Production Group at Ruislip.

Griffin \& George (Research and Development), Ltd., has been formed by the Griffin \& George Group to conduct research into, and the development of, new and improved scientific instruments and apparatus for laboratory use and process control.

Redifon have received an order from the G.P.O. for a number of i.s.b. drive units valued at $£ 30,000$.

Printed Circuits, Ltd., of Borehamwood, Herts., which recently became associated with the London Electric Wire Co. \& Smiths, Ltd., are establishing an information service in the form of a series of technical bulletins. Those wishing to receive these brochures are asked to write stating their particular interest.

Consoles. - R. H. Minns has resigned his technical directorship of Hatfield Instruments, and, with J. S. Jordan, has formed Consoles. The company, which has premises at Hersham Trading Estate, Walton-on-Thames, will produce metal cabinets and control consoles, and a range of measuring instruments, transformers and aerial equipment.

A new plastics factory to make equipment for the radio industry has been opened at Reading by Resinoid and Mica Products, Ltd., which was formerly a subsidiary of Southern Areas Electricity Corp., Ltd.

Stella.- The head office and showroom of Stella Radio and Television Co. is now at Astra House, $121 / 3$, Shaftesbury Avenue, London, W.C.2. The telephone number (Gerrard 7086) is unchanged.

Amos of Exeter are moving on March 2nd to larger premises at Weircliffe Court, Exwick, Exeter. The telephone number (Exeter 72132) is unchanged.

Kelvin Hughes Survey Department offers a world-wide service for hydrographic surveys, in the sounding and sampling of river, coastal and sea areas, tide and tide stream studies, together with the necessary land surveys. All preliminary marine survey work for the proposed sixth nuclear power station at Sizewell, Suffolk, was undertaken by the department.

CQ Audio, Ltd., have arranged with Technical Suppliers, Ltd., of 63, Goldhawk Road, London, W.12, to act as sole distributors for their complete range of equipment.

## EXPORTS

Trunk Radio-telephones.-Recent orders for the Murphy MR851 trunk radio-telephone equipment include 42 terminal installations-each comprising output unit, modulator, demodulator, r.f. unit and supply units -for British Guiana's first trunk telephone network. A second order for MR851 equipment for Australia's Snowy Mountains Hydro Electric Authority has also been received. The equipment provides 24 traffic channels.

India. - Solartron Electronic Group, who had a stand at the exhibition held in association with the recent Indian Science Congress, are in the course of forming, with Indian interests, a company in India.
V.H.F. communications equipment valued at more than $£ 600,000$ has been ordered from Plessey for the Australian Army. The equipment comprises the latest frequencymodulated v.h.f. multi-channel radio-telephone transmitter/receivers to have been proved in armoured fighting vehicles under field conditions. Three sets are involved, these being known to the British Army as the C42, the B47 and the B48.

Hanover Trade Fair.-Electronic Components Centre (Great Britain), Ltd., which represents a number of component manufacturers, have concluded a 10 -year agreement with the Hanover Fair Authorities to exhibit as the British Electronic Centre. Among the firms associated with the organization, which has offices at 31 Morden Road, London, S.E.3, are Ardente, Geo. Bray, British Electric Resistance, Cosmocord, E.M.I. Electronics, Hunt, London Electric Wire Co., Painton, and Telcon.

Norwegian Navy has ordered Decca true-motion radar, type TM909, for its new "Nasty" class of motor torpedo-boats.

Denmark.-The display of British products, including domestic sound and television equipment, announced in our last issue (page 98) as taking place in Copenhagen in May, has been postponed until November.

## MARCH MEETINGS

Tickets are required for some meetings; readers are advised therefore to

## LONDON

2nd. British Computer Society.-"Pseudo-random elements for com-puters-a survey of methods" by Dr. E. s. Page at 2.30 at Northampton College of Advanced Technology, St. Juhn's Street, E.C.I.
Srd. Association of Supervising Electrical Engineers.-" Wired television systems-installation and maintenance problems" by L. A. Isaacson (E.M.I.) at 7.30 at Windsor Castle Hotel, 134 King Street, Hammersmith, W.6.
4 th . Association of Supervising Electrical Engineers. -"High - frequency heating" by a representative of Philips Electrical at 7.45 at Wood Gieen Town Hall, N. 22.

5th. I.E.E.--" The reliability and life of impregnated paper capacitors" by J. P. Pitts at 5.30 at Savoy Place, W.C. 2 . 10th. I.E.E.-Discussion on "The Laplace transform-a tool for the electrical engineer" opened by A. C. Sim at 6.0 at Savoy Place, W.C.2.
10th. Association of Supervising Electrical Engineers. " Radio - frequency heating" by P. W. Ainscow (G.E.C.) at 7.45 at Eltham Green School, Queenscroft Road, S.E.9.
12th. Armed Forces Communications and Electronics Association.Visit to M.O. Valve Co.'s works at Hammersmith.

13th. Television Society.-"Training
television servicing" be in television servicing " by G. C. Barker (Murphy) at 7.0 at the Cinematograph

Exhibitors' Association, 164 Shaftesbury Avenue, W.C. 2.

13th. Radar and Electronics Association, Student Section.-"Radar data handling" by Dr. L. C. Payne (Decca Radar), at 7.0 at Norwood Technical College, Knight's Fiill, S.E.27.
18th. I.E.E.-" New amplifying techniques" by C. W. Oatley at 5.30 at Savoy Place, W.C.2.
18th. British Computer Society."An approach to learning and teaching machines" by C. E. G. Bailey at 6.15 at the Northampton College of Advanced Technology, St. John's Street, E.C 1.

18th. British Kinematograph Society. - Film and dem nstration of stereophinly at 7.30 at Mullards Theatre, Wuliard House, Torrington Place, WCI.
19th. Brit.I.R.E. Medical Electronics G1oup.-"Instrumentation in field physiology" by Dr. H. Wolff at 6.30 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1.

19th-20th-I.E.E.-Radio and Telecommunication Section convention on sterf ophonic sound recording, reproducLinn and broadcasting.
20th. Institute of Navigation. - "The imract of radar on the rules of the road at sea"" by Capt. F. J. Wylie, R.N. (president), at 5.15 at the Royal Geographical Society, 1 Kensington Gore, S.W. 7

20th. R.S.G.B.-" Single sideband techniques" by B. J. Rogers, G3ILI, (Hush) at 6.30 at the I.E.E. Savoy Mlace, W.C.2.
20th. B.S.R.A.-"F.M. feeder units" by R. S. Roberts at 7.15 at the Royal Society of Arts, John Adam Sureet, W.C. 2.
23rd." I.E.E.-" High-quality microphones" by M. L. Gayford at 5.30 in the Lecture Theatre at Savoy Place, W.C. 2 .

23rd. I.E.E.-" Effects of argon content on the characteristics of neonargnn glow-discharge reference tubes" by Dr. F. A. Benson and P. M. Chaliners a: 5.30 in the Tea Room at Savoy Place, W.C.2.
25th. Brit.I.R.E.-Papers on radio telemetry including "Engineering aspects of a 24 -channel f.m.-a.m. telemetry svesem" by W. M. Rac and "A sixcharmel high-frequency telemetry system" by T. C. R. S. Fowler at 6.30 at the London School of Hvgiene and Tropical Medicine, Keppel Street, w.C.I.

## ABFRDEEN

13th I.E.E.-" The relation between picture size, viewing distance and picture Guality with special reference to colour television and 10 spot-wobble technicues" by L. C. Jesty at 7.30 at Robert Gordon's Technical College.

## BELFAST

، 3ra. I.E.E.--Faraday lecture on "Automation" by Dr. H. A. Thomas at 7.30 at the Sir William Whitla Hall, Queen's University, Stranmillis Road.

10th. I.E.E.-" The B.B.C. sound broadcasting service on very-high frequencies" by E. W. Hayes and H. Page ai 6.30 at the David Keir Building, Queen's University.

## BIRMINGHAM

23rd. I.E.E - "The history of B.B.C. television" by R. T. B. Wynn at 6.0 at the James Watt Memorial Institute.

## BRISTOL

10th. Television Society.-" Industrial television" by J. G. M. Downs (Pye) at 7.30 at the Hawthornes Hotel, Clifton.

24th. Brit.I.R.E.-" Recent advances in travelling-wave tubes" by P. F. C. Burke at 7.0 at the School of Management Studies, Unity Street.

## CAMBRIDGE

2nd. I.E.E.-"The recognation of moving vehicles by electronic means" by T. S. Pick and A. Readman at 8.0 at the Cavendish Laboratory, Free School Larfe.

17th. I.E.E.-" Reliability of electronic components" by G. W. A. Dummer at 8.0 at the Cavendish Laboratory, Free School Lane.

## CARDIFF

10th. Association of Supervising Electricai Engineers. - " High-quality sound renroduction" by a representative of G.E.C. at 7.30 at the Angel Hotel.
11th. Brit.I.R.E.-"Applications of photo-electric cells" by Dr. F. A. Benson at 6.30 at the College of Advanced Technology.

## CHATHAM

2nd. I E.E.- " Domestic high-fidelity reproduction" by J. Moir at 7.0 at the Medway College of Technology.

## CHESTER

25 th . Society of Instrument Tech-nology.-" Industrial applications of transistors" by D. G. Holloway at 7.0 at the Grosvenor Museum, Grosvenor Street.

## CHRISTCHURCH

25th. I.E.E.-" The application of transistors to line communication equipment" by H. T. Prior, D. J. R. Chapman and A. A. M. Whitehead at 6.30 at the Kings Arms Hotel.

## DUNDEE

12th. I.E.E.-"The relation between picture size, viewing distance and picture quality with special reference to colour television and to spot-wobble techniques" at 7.0 at the Electrical Engineering Department, Qucen's College.

## EDINBURGH

17th. I.E.E.-" The application of transistors to line communication equipment" by H. T. Prior, D. J. R. Chapman and A. A. M. Whitehead at 7.0 at the Carlton Hotel, North Bridge.
20th. Brit.I.R.E.-" Application of magnetic amplifiers to electrical switching" by J. A. Purdie at 7.0 at the Department of Natural Philosophy, The University, Drummond Street.

24th. I.E.E.-Faraday lecture on "Automation" by Dr. H. A. Thomas at 7.0 at the Usher Hall.

## EXETER

121h. I.E.E.-" Gcrmanium and silicon power rectifiers" by T. H. Kinman, G. A. Carrick, R. G. Hibberd and A. J. Blundell at 3.0 at S.W.E.B. Showrooms, Bedford Street.

## FARNBOROUGH

24th. I.E.E.-" Space rescarch" by Dr. R. L. F. Boyd at 6.0 at Farnborough Technical Coliege, Boundary Road.

## FAWLEY

6th. Society of Instrument Tech-nology.-"Swartwout electronic instrunient system" by M. V. Needham at 5.30 at Copthorne House.

## GLASGOW

19th. Brit.1.R.E.-"Application of magnetic amplifiers to electrical switching" by J. A. Purdie at 7.0 at the Institution of Engineers and Shipbuilders, 39 Elmbank Crescent.

## LEEDS

16 th .
I.E.E-Faraday lecture on "Automation" by Dr. H. A. Thomas at 7.0 at the Town Hall.

## LIVERPOOL

11th. Institution of Production En-ginecrs.-" Electronic control of machine tools" by J. A. Stokes at 7.30 at the Exchange Hotel, Tithebarn Street.

## MANCHESTER

5th. Brit.I.R.E.-" Closed circuit television equipment" by R. E. Blythe at 6.30 at Reynolds Hall, College of Technology, Sackville Street.
18th. I.E.E.-" Bridging the Atlantic" by A. H. Mumford at 6.15 at the Engineers' Club, Albert Square.

## NEWCASTLE UPON TYNE

2nd. ,"I.E.E.-" High-quality microphones" by M. L. Gayford at 6.15 at King's College.
11th. Brit.I.R.E.-" Microwave stripline circuits for radar equipment" by K. Foster at 6.0 at the Institution of Mining and Mechanical Engineers, Neville Hall, Westgate Road.
19th. I.E.E.-Faraday lecture on "Automation" by Dr. H. A. Thomas at 7.0 at the City Hall.

## PRESTON

11th. I.E.E.-" Computers" by Dr. R. L. Grimsdale at 7.15 at the NorthWestern Electricity Board Demonstration Theatre, Friargate.

## READING

10th. I.E.E. Graduate \& Student Scction.-"Stereophonic sound" by E. W. Berth-Jones at 7.0 at Reading Technical College.

## SHEFFIELD

6th. I.E.E.-Faraday lecture on "Automation" by Dr. H. A. Thomas at 7.0 at the City Hall.

## WOLVERHAMPTON

11th. Brit.I.R.E.-" The develop,ment of high-frequency tape recording" by P. J. Guy at 7.15 at the Wolverhampton and Staffordshire College of Technology, Wulíruna Street.

## WORCESTER

25th. Institution of Production En-gineers.-Annual general meeting at 7.0 followed by "Numerical control of machine tools" by O. S. Puckle at the Star Hotel.

## IATE-FEBRUARY MEETINGS

23rd. B.S.R.A.-"The application of transistors to low roise pre-amplifiers" by J. Somerset Murray at 7.0 at the Ruyal Society of Arts, john Adam Stieet, W.C.2.

27 th. R.S.G.B.-"Recent developaneris in the microwave field" by K. W. Dremmond (Muilard) at 6.30 at the I.E.E.W Savoy Place, W.C.2.

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## By "DIALLIST"

## Long-lived C.R.T.s

MY request for information about old television c.r.t.s which are still going strong after years of use produced some quite amazing figures not only for the tubes themselves, but also for valves and other components working in the same sets. The oldest c.r.t. reported still at work belongs to a Horley reader. The set was bought in 1948 and it has averaged not less than three hours' use an evening ever since. During all that time the only bits and pieces replaced have been one EF50, five EB41s and one autotransformer. Though still unboosted, the 9 -in tube is claimed to give an excellent picture. From Washington, D.C., comes an account of an R.C.A. 16AP4 tube which has been in use from late afternoon till bedtime every day since 1949. About four years ago the emission fell off so the heater voltage was raised permanently to 7.5 V and it has needed no further attention. My correspondent, who is an electrical engineer, puts its service life so far at 17,000 hours plus, for it gets some very hard work at the week-ends and during the children's school holidays.

## Can You Say?

Several readers ask whether I can suggest reasons for the longevity of some c.r.t.s. I only wish I could! If you take a batch of well-made and
well-pumped c.r.t.s of the same type, a few (a very few) will fall by the way in the first six months of use; the great majority will have the about average service life for their classsome a little above and some a little below the mean figure. But just the odd one here and there will turn out to be a long-lived prodigy. I suppose, really, that that's only to be expected, for the same sort of thing seems to be true not only of most electrical and mechanical appliances, but also of human beings and other animals. Incidentally, my American reader asks why in this country we use c.r.t. and not c.r.v. as the abbreviation for the cathode-ray tube; in other words, why do we call it a tube and not a valve? The answer to that one is that the c.r.t. was developed (wasn't it by Crookes?) years before the electronic valve was even thought of.

## The Sun Does Its Best

DURING the International Geophysical Year, which recently ended its eighteen-months span, the sun played its part in a way which exceeded all expectations. It's some 200 years since records of sunspot activity began to be kept and never in the whole of that time has solar activity been so marked as in the present maximum period. Many of the I.G.Y. programmes were concerned with such activity and some remarkable discoveries will no doubt come

to light when all the results have been fully digested. Two have already been disclosed. The first is that of the Van Allen belt of intense radiation, which surrounds the whole of the Earth, except for those regions which are near the magnetic poles. It was data from Explorer satellites which led to the knowledge of this belt's existence. Then, we seem now to have found the reason-or at any rate one of the reasons-why large "flares" in the sun are followed by wireless blackouts. Whenever such a flare occurred the U.S. authorities sent up special rockets and from the data received from these it was found that these activities are accompanied by a tremendous emission of X-rays, one of whose effects is to cause the wireless blackouts.

## Some Bed!

ON the whole, I don't think I'll go in for one of the super-beds on show at the recent Furniture Exhibition in London. I use my bed for sleeping purposes. There are doubtless those who like to go abed and watch the "tele" or listen to the "wireless" and I've no doubt that there'll be quite a run on Slumberland's masterpieces, even of the variety with mink coverlets, which is to sell for a mere £2,500. I might be tempted to do something about it if the TV set were equipped to deal with tape recordings of single-file processions of sheep jumping one after another over a gate. That would indeed be a luxury beyond price to those who like myself sometimes have bouts of insomnia. But I don't think I need push-button curtain opening or shutting gadgets and I'm sure I don't want a bed fitted with an intercom. telephone-unless this is so arranged that it can be used only for outgoing calls. No, I think I'll remain content with my present simple, but very comfortable couch.

## Non-detachable Backs

WHAT a blessing it would be if the television sets sold to the ordinary viewer could be made with backs that couldn't be taken off except by the use of a key or a special tool. Like me, I expect you've met with instances of the damage that can be done when someone who knows nothing about the works gets poking about in them
with a screwdriver. And then there's the question of the risk of accidents. By the law of averages about half the TV sets used with 2 -pin mains plugs are at any time so connected that their chassis are live. Yet people will get fiddling about with adjustments which can be made only if the set is switched on and working. One way of making backs non-detachable except by the serviceman would be to mount one side of the back on pintles like those used for rudders and to have a lock or a sealed fastening between the other side and the cabinet. Probably the seal would be the simplest. The dealer would fix it when he had installed the set and would impress his own mark on it.

## CLUB NEWS

Barnet.-F. J. H. Charman (G6CJ) will give a lecture-demonstration on aerials to members of the Barnet \& District Radio Club on March 31st. The club's lecture-meetings are held at 7.30 on the last Tuesday of each month at the Red Lion Hotel, High Barnet. Morse and instructional meetings are held on the second Tuesday.

Battersea.-The London Short-wave Club meets every Friday at 7.30 at the L.C.C. Men's Institute, Latchmere Road, London, S.W.11. Lecturedemonstration meetings are arranged for alternate Friday evenings and the club station (G2CLR) is "on the air" on the intermediate Fridays between 7.30 and 9.30 .

Birmingham.-The March programme of the Slade Radio Society includes a Mullard film show on the 6th at the Y.M.C.A., Snow Hill; a talk on the 13 th by J. F. Moseley, of Pye, on v.h.f. business radio, illustrated by a film; and on the 27 th two members, T. J. Hayward (G3HHD) and G. Nicholson (G3HKC) will deal with the construction and use of test equipment. Except where otherwise stated, meetings are held at the Church House, High Street, Erdington, at 7.45.

Brighton.-Meetings of the Brighton and District Radio Club (G3EVE) are held at 8.0 on Tuesdays at the Eagle Inn, Gloucester Road, 1.

Bury.-B. P. Clear, of the Jodrell Bank Research Station, will talk on some aspects of the station's work at the March 10th meeting of the Bury Radio Society. The club meets at 8.0 at the George Hotel, Kay Gardens. On the 24th members are visiting the Mullard works at Simonstone.

Halifax.-High-quality recordings will be demonstrated by Fane Electronics to members of the Halifax \& District Amateur Radio Society on March 3rd. The club meets on the first Tuesday of each month at the Sportsman Inn, Bradshaw.

South Kensington.-D. E. A. Harvey, of Siemens, will speak on the manufacture and use of transistors at the March 10th meeting of the Civil Service Radio Society. Meetings are held at 6.0 in the Science Museum, South Kensington, London, S.W.7.

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## Extra-spatial Electrons

NOT long after the Russians launched their circumsolar satellite, I attended a lecture at the London Planetarium in which this man-made planet was shown, together with the other planets in their respective orbits, travelling around the sun. The lecturer must have been an electronics fan-which is not quite the same thing as an electric fan-as he was at pains to explain what a great part electronics played in space navigation.

In my opinion the next step will obviously be to launch a satellite at a high enough velocity for it to escape from the sun's gravitational field, and go into orbit around the nearest star, Proxima Centauri, which is only $4 \frac{1}{3}$ light years away. Then will come the day when a spaceship escapes out of our own Milky Way, and journeys towards another galactic system such as the giant nebula in Andromeda which is one of the nearer ones, only a matter of a million light years distant.

When that day dawns will there be no other fields to conquer? To the unthinking materialist it might seem so, but to the more intelligent members of the community who read $W . W$. it will be clear that the ultimate triumph will not come until a "spaceship" breaks out of this universe of space and time into the timeless and non-spatial one which seems to be inhabited by poltergeists and other clammy entities who, if the ghost hunters are to be believed, not only pass freely through brick walls, without let or hindrance, but have the ability to be in two or more places at once.

Now I have never actually seen a ghost, although I once mistakenly thought I did, as you will see by the accompanying 20 -year-old sketch. It may surprise you, therefore, when I say that I do believe in them, and I
think I can see the way in which we may eventually be able to launch a psychic sputnik.
Actually it was "Cathode Ray" who gave me the idea when he told us in the November issue that an electron seemed to consist of an uncancelled $\psi$ wave, which, as he explained, is a wave in nobody knows what. Now obviously these $\psi$ waves must possess the various attributes which we associate with waves such as length, amplitude and what-haveyou.
We all know that radio waves of various lengths and amplitude are constantly passing through our receivers and through each other without difficulty. Working on this analogy I have the idea that ghosts are built of atoms made up of $\psi$ waves having a different length or other dimension from those which we call electrons.

It therefore follows that if we can find means to change the length or other dimension of the electronic $\psi$ (psi) waves, we shall turn them into what I will call psychic electrons of which ghosts and ghostly walls are built. In other words if we change the wavelength of electrons or $\psi$ wave in, say, a lump of sugar, it will disappear from our world into the fourth dimension, like the Time Machine in H. G. Wells's famous sciencefiction novel.

## Challenge Accepted

IN the January issue Mr. James M. Hoy in a letter to the Editor invites me to visit the Institution of Civil Engineers for the purpose of reading a certain paper in the Proceedings of this learned society, after which I am further invited to give a definition of the phrase "common usage ". All this, because I stated in $W . W$. that the word "valve" had, by common usage, come to mean a one-way device.


I once mistakenly thought I saw a ghost

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| Sarylces No: | - | Mod. Version of CT 316 | CT 386A | - | Mod Version of CT 316 | CT 380 | CT 414 | Admiralty <br> Mk. 29 TU | - |
| Bandwidth (3db): | Max: D.C.$10 \mathrm{Mc} / \mathrm{s}$. | Max: D.C.$5 \mathrm{Mc} / \mathrm{s}$. | Max:D.C.- $10 \mathrm{Mc} / \mathrm{s} .$ | Max:D.C.$1 \mathrm{Mc} / \mathrm{s}$. | Max: D.C.$5 \mathrm{Mc} / \mathrm{s}$. | Constant $\text { D.C. }-12 \mathrm{Mc} / \mathrm{s} \text {. }$ | Max:D.C. $7 \mathrm{Mc} / \mathrm{s}$. | $\begin{gathered} \text { Max:D.C. } \\ 20 \mathrm{Kc} / \mathrm{s} . \end{gathered}$ | Constant: $0.9 \mathrm{c} / \mathrm{s}-9 \mathrm{Mc} / \mathrm{s}$ |
| Sensitivity: | $1 \mathrm{mv} / \mathrm{cm}-$ $10 \mathrm{~V} / \mathrm{cm}$ | $\begin{aligned} & 0.4 \mathrm{~V} / \mathrm{cm} .- \\ & 10 \mathrm{~V} / \mathrm{cm} . \end{aligned}$ | $\begin{aligned} & 1 \mathrm{mV} / \mathrm{cm} .- \\ & 10 \mathrm{mV} / \mathrm{cm} . \end{aligned}$ | $\begin{gathered} 3 \mathrm{mV} / \mathrm{cm}- \\ 100 \mathrm{~V} / \mathrm{cm} . \end{gathered}$ | $\begin{aligned} & 0.4 \mathrm{~V} / \mathrm{cm} .- \\ & 10 \mathrm{~V} / \mathrm{cm} . \end{aligned}$ | $\begin{gathered} 100 \mathrm{mV} / \mathrm{cm} .- \\ 60 \mathrm{~V} / \mathrm{cm} . \end{gathered}$ | $\begin{aligned} & 3 \mathrm{mv} / \mathrm{cm} .- \\ & 100 \mathrm{~V} / \mathrm{cm} . \end{aligned}$ | $\begin{gathered} 10 \mathrm{mV} / \mathrm{cm} .- \\ 10 \mathrm{~V} / \mathrm{cm} . \end{gathered}$ | $\begin{gathered} 30 \mathrm{mV} / \mathrm{cm}_{3 .}- \\ 30 \mathrm{~V} / \mathrm{cm} . \end{gathered}$ |
| 'Y' Calibration: | Cal Sensitivity <br> Acc: 10\% | Shin Meter Acc: 3\% | Cal. Sensitivity <br> Acc: 10\% | Cal. Shift <br> Acc: 5\% | Shift Meter Acc: 3\% | Cal. Shift <br> Acc: 2\% | Cal. Shift <br> Acc. 5\% | Special Facilities | Comparison <br> A.C. Acc: 5\% |
| 'X' Calibration: | Cal. Adjustment Acc: 10\% | 'Pips' and Sinewave Acc: 2\% \& $1 \%$ | Cal. Adjustment Acc: 10\% | Cal: Adjustment Acc: 10\% | Sincwave Acc: 1\% | 'Pips' and <br> Bright-up <br> Acc: 2\% | Cal. Adjustment Acc: 5\% | Special Facilities | Brilliance Mod. Acc: 5\% |
| fweep Velocity : | $\begin{aligned} & 10 \mathrm{~cm} . / \mathrm{Sec}- \\ & 10 \mathrm{~cm} . / \mu \mathrm{Sec} . \end{aligned}$ | $0.1 \mathrm{~cm} / \mathrm{m} \mathrm{Sec}$ $10 \mathrm{~cm} / \mu \mathrm{Sec}$. | $1 \mathrm{~cm} / \mathrm{Scc}$. $10 \mathrm{~cm} . / \mu \mathrm{Sec}$. | $1 \mathrm{~cm} . / \mathrm{Sec} .-$ $1 \mathrm{~cm} . / \mu \mathrm{Sec}$. | $0.1 \mathrm{~cm} . / \mathrm{m} \mathrm{Sec}-$ $10 \mathrm{~cm} / / \mu \mathrm{Sec}$. | $\begin{aligned} & 10 \mathrm{~cm} . / \mathrm{Sec} .- \\ & 10 \mathrm{~cm} / \mu \mathrm{Sec} . \end{aligned}$ | 0.33 cm / $/ \mathrm{Sec}$. $3.3 \mathrm{~cm} . / \mu \mathrm{Sec}$. | $1 \mathrm{~cm} . / \mathrm{Sec} .-$ $100 \mathrm{~cm} . / \mathrm{Sec}$. | $0.1 \mathrm{~cm} . / \mathrm{m}$ Sec. $-2 \mathrm{~cm} . / \mu \mathrm{Sec}$. |
| ' $X$ ' Expansion ; | $\begin{array}{r} \times 0.5, \times 1.0, \\ \times 2.0, \times 5.0 . \end{array}$ | - | $\begin{array}{r} \times 0.5, \times 1.0 \\ \times 2.0, \times 5.0 \end{array}$ | $\begin{aligned} & \text { Variable } \\ & \text { up to } \times 10 \end{aligned}$ | - | $\begin{aligned} & \text { Variable } \\ & \text { up } 10 \times 100 \end{aligned}$ | $\begin{aligned} & \text { Variable } \\ & \text { up to } \times 10 \end{aligned}$ | $\left\|\begin{array}{lll} \times & 0.05, & \times \\ \times 0.1 & 0.1 \\ \times 0.2,0.5, \times 1.0 . \end{array}\right\|$ | Variable up to $\times 10$ |
| Dimensions : | $\begin{array}{r} 161^{\prime \prime} \times 10^{\prime \prime} \\ \times 22^{\prime \prime} \text { deep. } \end{array}$ | $\begin{aligned} & 12^{\prime \prime \prime} \times 9^{\prime \prime} \\ & \times 18^{\prime \prime} \text { deep. } . \end{aligned}$ | $\begin{aligned} & 161^{\prime \prime} \times 10^{\prime \prime} \\ & \times 22^{\prime \prime} \text { deep. } \end{aligned}$ | $\begin{aligned} & 16 \frac{1}{} 1 \times 10^{n} \\ & \times 22^{\prime \prime} \text { deep. } \end{aligned}$ | $\begin{gathered} 12^{\prime \prime} \times 9^{\prime \prime} \\ \times 18^{\prime \prime} \text { deep. } \end{gathered}$ | $\begin{array}{r} 20^{\prime \prime} \times 143^{\prime \prime} \\ \times 277^{\prime \prime} \text { deep. } \end{array}$ | $\begin{array}{r} 16 i_{1 \prime \prime}^{\prime \prime} \times 13^{\prime \prime} \\ \times 272^{\prime \prime} \text { deep. } \end{array}$ | $\begin{aligned} & 14^{\prime \prime} \times 10^{\prime \prime} \\ & \times 20^{\prime \prime} \text { deep. } \end{aligned}$ | $\begin{aligned} & 14 \frac{1}{" \prime}^{\times} 101^{\prime \prime} \\ & \times 19 z^{\prime \prime} \text { deep. } \end{aligned}$ |
| Weight: | 70 lb . | 40 lb . | 70 tb . | 70 lb. | 40 lb . | 140 lb . | 116 lb . | 47 lb . | 43 lb . |
| Pries: | ¢235 | ¢235 | E27 | c288 | $\leftarrow 220$ | ¢ 490 | 8390 | ¢460 | ¢145 |

Also available shortly — the new Solartron Infinite Persistence Oscilloscope Type QD 910
Why not write or call us now for a demonstration of any of the nine models listed above?
Specialist instrument engineers are immediately available to assist you, whatever your problem or field of application.

# These photocells give you the simplest photo-electric <br> <br> control possible 

 <br> <br> control possible}

Photo-electric control with the Mullard ORP11 and ORP90 cadmium sulphide cells is the simplest possible because a photocell and relay form the complete circuit.
The unusual combination of high current capacity and extreme sensitivity of these Mullard cells enables robust relays to be operated direct-amplifiers are unnecessary.
Both cells can be operated from either a.c. or d.c. supplies, they are inherently rugged and have a wide range of applications in industry.
The usable response extends through the entire visible spectrum to the near infra-red.
The ORPIl differs from the ORP90 chiefly in being "end-viewing" and having a somewhat smaller photocathode area. This type of photocell is made available to simplify mounting problems encountered in certain applications-particularly in flame failure detectors in oil fired furnaces.
Data sheets giving further information are readily available from the address below.


## abridged data

|  | ORP II | ORP 90 |
| :--- | :---: | :---: |
| Required direction of incident light | End-on | Side-on |
| Area of photo-element | $1.25 \mathrm{sq} . \mathrm{cm}$. | 2.9 sq. cm. |
| Average cell current at 10V d.c., <br> 5 foot candles and lamp colour <br> temperature 2 700 | 6 KnA | 6 mA |
| Maximum ultimate dark current <br> at 100 V d.c. | $5 \mu \mathrm{~A}$ | $<2.5 \mu \mathrm{~A}$ |
| Maximum cell dissipation at $25^{\circ} \mathrm{C}$. | 200 mW | 600 mW |
| Spectral response | Same for both cells- |  |
| see curve. |  |  |



Mullard Limited
Mullard House, Torrington Place, London, W.C.I
Telephone: Langham 6633


## Mullard

GOVERNMENT AND
INDUSTRIAL VAIVE DIVISION


## Printed Circuit Counter Panels



A complete range of transistorized counter panels of common size, fixing method and electrical connexion, designed to provide a flexible unit system whereby any special requirements in the counting or data processing fields can be quickly built up.

A fully illustrated brochure giving complete performance and specification figures for every panel in the range is available on request.

50kc/s Scaler
$1 \mathrm{Mc} / \mathrm{s}$ Scaler
Input Amplifier
Gate Unit
10ke/s Oscillator
$1 \mathrm{Mc} / \mathrm{s}$ Oscillator
Power Unit
50kc/s Read-out Scaler
$1 \mathrm{Mc} / \mathrm{s}$ Read-out Scaler
4 Channel Output Unit
Read-out Unit
Meter Display Unit
Lamp Display Unit
Numerical Indicator Tube
Shift Register Stage
Shift Register Driver

RANK CINTEL LIMITED
Worsley Bridge Road London-SE 26 HITher Green $\mathbf{4 6 0 0}$

[^7]

## When it's hot enough to fry an egg "Terecaps"are in theirelement

SIZES AND RATINGS

| Capacitance | Working Voltage d.c. at $71^{\circ} \mathrm{C}$ at $125^{\circ} \mathrm{C}$ |  | ```Test Volts d.c.``` |  | ons <br> Dia. | Catalogue Ref. |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0.1 | 150 | 125 | 300 | 11 | $\frac{1}{4}$ | 8801.C |
| 0.25 | 150 | 125 | 300 | 13 | $\frac{1}{2}$ | S-8803.C |
| 0.5 | 150 | 125 | 300 | 17 | \% | S-8800.C |
| 1.0 | 150 | 125 | 300 | 17 | ? | S-8804.C |
| 0.1 | 250 | 180 | 500 | 11 | $\frac{1}{2}$ | $8801 . C$ |
| 0.25 | 250 | 180 | 500 | 21 | 1 | 8803.C |
| 1.0 | 250 | 180 | 500 | 21 | 1 | 8804.C |
| 0.1 | 350 | 250 | 700 | 18 | $\frac{1}{2}$ | 8802.C |
| 0.25 | 350 | 250 | 700 | 13 | : | S-8804.C |
| 1.0 | 350 | 250 | 700 | 21 | 1 | 8806.C |

## "Terecap"* Capacitors

Dubílier "Terecap" Capacitors are of rubular form with extended foil metal electrodes fitted with wire tail terminations and incorporate a non-hygroscopic film dielectric. Being designed to meet abnormal atmos pheric conditons such as obtain in tropical zones the capacitors are supplied hermetically sealed in metal containers with ceramic end-seals.

* A Registered Dubiliep Trade Mark.

Dubilier "Terecap" Capacitors have these outstanding features:-

1. Can be used up to $125^{\circ} \mathrm{C}$. with voltage de-rating above $70^{\circ} \mathrm{C}$.
2. High insulation resistance, more than twenty times that of paper dielectric capacitors. $\left(10,000 \Omega \mathrm{~F}\right.$ at $20^{\circ} \mathrm{C}$.)
3. Compactness.
4. Excellent capacitance stability over a wide temperature range. (Normal capacitance tolerance $\pm 20 \%$ )
Other capacitance values can be supplied to order. We invite your enquiries. 5. Power Factor $0.5 \%$ at $20^{\circ} \mathrm{C}$. for $1 \mathrm{kc} / \mathrm{s}$.

## CAWKELL INSTRUMENTS IN ACTION ... 3

## Time Calibrator type CU3....



We had intended, this month, to show a typical user of our time marker generator. Unfortunately all the customers we approached are working on secret projects and wouldn't let our camera near the equipment.

The CU3 Time Calibrator provides crystal calibrated time markers for use with an oscilloscope. The markers can be either free running or keyed from a square wave source. The keying source can be the internal generator or external to the calibrator.

## BRIEF SPECIFICATION

Marker intervals :
0.5 - 1000 microseconds

Accuracy :
$\pm 0.05 \%$
Amplitude :
$\pm 50 \mathrm{~V}$
Rise-time :
0.1 microseconds.

A typical display of a keyed train of markers at intervals of 1 and 5 microseconds
 For full details contact:

## CAWKELL RESEARCH \& ELEC

Telephone: SOUthall 3702/588I

## Low Resonance Speakers



## SPECIFICATION

Size $34 \mathrm{in} . \times 31 \mathrm{in} . \times 12 \mathrm{in}$. Weight 641b. Impedance $8 / 15 \mathrm{ohms}$. Bass resonance $30 / 35 \mathrm{c} / \mathrm{s}$. Max. input 15 watts.

## UNITS

W12/SFB, 10 in . Bronze/SFB, Super 3. The 12in. and 10 in . units are in parallel. This arrangement gives very smooth results over the full range with a 3 dB gain at low frequencies. The Super 3 is again in parallel via a 4 Mfd . capacitor and is mounted on a small baffle facing upwards.

## Model PST/8

A new, compact low-resonance cabinet, fitted with
EXPANDED POLYSTYRENE panels.

> SIZE: $24^{\prime \prime} \times 12^{\prime \prime} \times 12^{\prime \prime}$
> Weight of cabinet 171 b .

The PST/8 has been designed to give optimum results at minimum cost in money and room space. Placed on a shelf or table, it is an ideal second speaker for STEREO with the SFB/3 or COLUMN EIGHT in the other channel, and may be laid on its back with the cone facing upwards to avoid directional effects if preferred.
The PST/8 also functions well as a second speaker to reduce room resonance on single channel input.

Made and guaranteed by

## by Whartedale

registered trade mark
Model SFB/3 (Regd. design 881,557)

## £39-10-0 Complite

- Attractive appearance
- Free-standing and easily moved
- Resonance free Sandfilled Baffle
- Omni-directional
- Frequency range: $30 \mathrm{c} / \mathrm{s}$ to $20,000 \mathrm{c} / \mathrm{s}$
- Moderate price


IMPEDANCE CURVE. Note the unusually level impedance which typifies the wide frequency response.


PRICES: (including tax where applicable) Cabinet only
£7/10/- in white wood.
£10/10/- polished and veneered walnut, oak or mahogany.
Fitted with $8^{\prime \prime}$ Bronze/FS/AL
£13/19/- in white wood.
£16/19/- polished and vencered as above.

> Fitted with Super 8/FS/AL
£16/15/- in white wood.
£19/15/- polished and veneered as above.
Leaflets giving full technical description and operational notes by G. A. BRIGGS Free on Request.
Telephone: Idle 1235/6. Grams: 'Wharfdel,' Idle Bradford.

# big Imlok developments 

TO SAVE YOU MONEY ON CASE CONSTRUCTION!


1MLOK - the unique cabinet construction system-now offers designers and engineers even more possibilities than before. Fifty-six new parts have been added to the already wide range of precision-made corner connectors, extrusions and accessories, and you can now buy Imlok in six different stages of manufacture. It means more versatility ... more scope in case and cabinet design. The Imlok system is ideal for both prototype and production work



NEW 36-PAGE IMLOK MANUAL
now available, includes illustrations and specifications of all Imlok parts, colour photographis of many cabinet and console
designs built in Imlok, as well as new Imlok uses. Send now for free copy


## 56 new parts

The addition of new connectors and extrusions means you can now have external or internal angles from $22 \frac{1}{2}^{\circ}$ to $90^{\circ}$. The new range also includes strengthening extrusions for extra heavy-duty structures, fashion trims, doorpulls, castors, and many more. Full list on request Uxbridge 6231

## 6 ways to buy

1. As individual connectors and with extrusions in 12 ft . lengths. 2. With connectors pre-drilled ready for assembly with selftapping screws. 3. With pre-drilled connectors and with extrusions custom-cut by Imhofs to your lengths and mitred. 4. As a complete frame-work, ready for fitting with your own panels. 5. As a complete structure with panels, but unpainted. 6. As a complete case, rack or console, fully finished

Alfred Imhof Ltd., Dept. M.3, Ashley Works, Cowley Mill Road, Uxbridge, Middx,

Export \& London Showrooms : 112-116 New Oxford St WCI: Museum 7878

IMHOFS AGENTS OVERSEAS
AUSTRALIA Aladdin Industries (Pty) Led, Stanmore NSW
BELGIUM Rogelec, Ghent
CANADA Measurement Engineering Led, Arnprior

NORWAY Birger Christensen, Oslo SWEDEN Elektronlund AB, Malmo C SWITZERLAND Walter Blum, Zurich 2/39
U.S.A. Bud Radio Inc, Cleveland 3, Ohio

BRIT. GUIANA British Caribbean Agencies Led,

DENMARK Tage Schouboe, Copenhagen $N$ FINLAND Oy Scienta Ab, Helsinki HOLLAND J.Th. van Reijsen, Delft ITALY Prodel SPA, Milan
NEW ZEALAND Imarex Ltd, Auckland C3

## Pye MICROWAVS Portable TV Links

## Type PTC M1000

This transportable long-range television link is suitable for use with the N.T.S.C. colour or monochrome systems, the C.C.I.R. system or the British 405-line system. A sub-carrier f.m. music link circuit is incorporated. The normal frequency range is 6875 to $7425 \mathrm{Mc} / \mathrm{s}$ but models can be supplied to cover the range of 5925 to $6425 \mathrm{Mc} / \mathrm{s}$. The r.f. power output is one watt.
The equipment can be operated back-to-back as a demodulator repeater for multi-stage transmission links. Dependent upon siting, each link is capable of transmitting a distance of 50 miles or more.
Transmitter and receiver, as well as an r.f. wavemeter and intercommunication circuits are all contained in four lightweight luggage-type cases. Spun aluminium parabolic reflectors are available in diameters up to 10 ft ., and all ancillary equipment can also be supplied.
Please write for details.

## As supplied to.

ASSOCIATED TELEVISION LTD.
scottish television ltd.
TYNE-TEES TELEVISION LTD. CENTRAL REDIFFUSION SERVIGES LTD. PORTUGUESE TELEVISION SERVICE ATOMIC WEAPONS RESEARCH ESTABLISHMENT
and many other users

# now in quantity production 



This latest ELAC deflection unit incorporates the new MULLARD Ferroxcube core Type FX 1981, enabling a 'spull back" of 4 mm to be achieved without loss of sensitivity. Line inductances of 5 to 30 mH with ${ }_{\mathrm{K}}^{\mathrm{L}}$ RATIO OF .8 and frame impedances of 2 to 70 ohms are readily available. The standard model is supplied complete with TUNGSTEN steel picture centring plates, positive tube neck clamping device and a terminal panel well removed from adjustment points.


## AN ENTHUSIAST

Results now have justified my faith in buying a Brenell, and I shall not hesitate to recommend your products and excellent service to anyone... F.A.M.
again I would say that we are more than pleased with the instrument itself and it is nice to know thot one is dealing with manufocturers who have the interest of their customers after sale so much at heart.
F.G.C.

## A MEMBER OF U.S.A.F.

having investigated tape recorders manufactured in the United States, Germany, the Netherlands and Great Britain I was greatly impressed by the performance and quality of your product: it is, in my opinion, one of the finest instruments of its type available in this wide market . .
D.B.

TAPE CLUB FOUNDER
... You can quote me at ony time as saying that pound for pound the Brenell recorder is the finest value in tape equipment on the market today . . .
C.W.A.

## A PROUD OWNER

my recorder has given me excellent service and I am really delighted with it . .

## A WARDEN

a member of the management committee of our Centre was at the Show last week, and came back full of praise for your equipment, Brenell Mk. 5 Stereophonic Record Playback. He has interested us to the extent that we would like to add this item to the Centre's equipment . . .

## BRENELL 3 STAR $\star \star \rightarrow$ PORTABLE

* Three recording speeds, $1 \frac{7}{3}, 3 \frac{3}{4}, 7 \frac{1}{2}$ i.p.s.
* Frequency compensation at all speeds
* Push button operation (interlocked)
$\star$ Printed eircuit amplifier
* Separate bass and treble controls
$\star$ High-quality loudspeaker ( 8 in . by 5 in .)
* Spool sizes up to 7 in . to take standard pre-recorded tapes (all E.M.I. pre-recorded tapes are on 7in. reels)
- Pause control
* Digital revolution counter
* Modern style wooden cabinet designed for improved acoustic performance


## 58 gns.

Price includes Mlerophone, 7in. syool and 1.200 ff . Tape
Send for full details and information on Stereophonic/dual track play-back equipment, to Sole Monufacturers:
BRENELL ENGINEERING CO.: LTD.
la DOUGHTY STREET, LONDON, W.C.I
Tel : CHAncery 5809 and HOLborn 7358


* Four recording speeds $1 \frac{7}{6}, 3 \frac{3}{\frac{1}{2}}, 7 \frac{1}{2}$ and 15 i.p.s.
$\star$ Permits use of $8 \frac{1}{2} \mathrm{in}$. reels $(2,400 \mathrm{ft}$. of tape for
long play. $1 \frac{7}{6}$ i.p.s. over 8 hours)
* Three independent motors (B.T.H.)
* Special foolproof interlocking controls

Ł Instant stop without spillage
$\star$ Pause control

* Digital rev. counter
* High quality amplifier
* Recording level indicator
* Monitoring facilities
* Azimuth head adjustment


## LIST PRICE <br>  <br> 64 <br> GNS.

including 1200 ft . of tope

* Provision for extra sound heads
* Fast rewind ( $1,200 \mathrm{ft}$. in 45 jecs.)
* Coloured signal lights

Because the Mark 5 is of unir construction the following can be supplied as separate items for incorporation in your owr equipment.
Tape deck with provision for extra heads ..................... 28 gns.
Tape Pre-amplier Type T.P.2. ..................................... 17 gns.
Power Unit T.U 2.
64.180
rtereo/Rec. Playback (including mounting rack) ............ 893 . 16.0


Yes, we receive many letters congratulating us on the quality and fine performance of our products also the excellent service we maintain. Being the sole manufacturers of Brenell recorders, we are naturally pleased to receive such praises which fully justify our claim that the equipment we manufacture and sell is the finest value money can buy.

## STEREO SOUND SUPREME/ P by <br> THE RESULT OF 4 YEARS' PROGRESSIVE DEVELOPMENT

THE STEREO PICKUP
for playing $45 / 45$ records. Miniature ceramic type with replaceable diamond stylus. Constant velocity output approximately 20 mV from each channel. Frequency range 20 tol 6,000 cycles. Channel separation $20 / 25 \mathrm{dbs}$.
(Complete as illus.) $£ 9$ plus £3 . 16. 11 P.T.
Head only $£ 5 \cdot 10,0$ plus £ $2: 7$. 0 P.T.
Arm only $£ 3$ : 10.0 plus $£ 1$ : 9 : $11 \mathrm{P} \cdot 1$.

VARIABLE 3 SPEED MOTOR TYPE B
Operates at $33 \frac{1}{2}, 45$ and 78 r.p.m. Non-ferrous turntable. Built-in large stroboscope with internal light source. Precision ground and lapped spindles. Adjustable nylon graphite bearings. Synchronous motor.

$$
\begin{aligned}
& £ 20 \cdot 10 \cdot 0 \\
& \text { plus } £ 8 \cdot 15 \cdot 3 \text { P.T. }
\end{aligned}
$$

STEREOPHONIC AMPLIFIER AND PRE-AMPLIFIER


Twin channel amplifier and pre-amplifier for reproducing monaural and stereophonic sound from disc, radio and compensated tape.
Ultra linear push/pull output giving 7.5 watts peak from each channel.

Amplifier £24. 10 . 0
Pre-amp. £16. 10 . 0

A. R. SUGDEN \& Co. (Engineers) Ltd. market street, BRIGHOUSE, YORKS.
Telephone 2142

# Magnetic Attraction Applications-1 

The earliest known effect of permanent magnets is their ability to attract ferrous objects.
The attraction or holding power of a magnet under ideal conditions can be calculated from the basic formula:-
Force in dynes $=\frac{\mathrm{B}^{2} \mathrm{~A}}{8 \pi}$
or
Force in $\mathrm{lb} .=\frac{\mathrm{B}^{2} \mathrm{~A}}{11,263,000}$
where $A=$ area of magnetic pole faces in $\mathrm{cm}^{2}$
and $B=$ corresponding fux density in gauss.
Under normal conditions joints, tolerances, misalignments and leakage flux rapidly reduce the theoretical pull; therefore the calculated value should only be used as a guide. One factor which must al ways be observed is that the magnet under the most severe conditions of open circuit should be of sufficient length to prevent self-demagnetisation-i.e. the maximum value of H should not exceed $\mathrm{H}_{\mathrm{d}}$.
This advertisement deals briefly with some of the many industrial applications using magnetic attraction, such as Machine Tool Chucks, Relays, Industrial Filtration, Door Catches, etc.


## Machine Tool Chucks

One of the most useful applications of permanent magnets is in machine tool chucks where steel articles of a form extremely difficult to clamp, are held firmly in position for machining operations.

There are many types of magnetic chuck designed for various purposes but generally with relatively wide pole pitch spacing. The chuck illustrated bas the advantage of small pole spacing and is particularly suitable for small or thin articles. It consists of thin 'Magnadur'
blocks with mild steel pole plates assembled in sandwich form giving alternate poles $1 / 8^{\prime \prime}$ apart.

The attractive force to iron and steel objects of not less than $1 / 16^{\prime \prime}$ in thickness is approximately $130 \mathrm{lb} . / \mathrm{sq}$. in. The objects are released by moving the lower section of the chuck one pole pitch to short circuit or cancel the flux in the whole chuck.

## Industrial Filtratlon Equipment

It is well known that one of the principal causes of wear in machinery is the presence of abrasive matter in the lubricating oil. A certain amount of the contamination to the oil can be prevented by careful design and dust covers but minute particles of steel can only be removed by the use of permanent magnets.

## Relays and Thermostats

For current carrying relays and thermostats, contacts should open and close with a snap action and contact pressure must be sufficient to prevent chatter and arcing. It is in applications such as these that the a ttractive force of a magnet can be used to supply the necessary minimum contact pressure and also the desired degree of snap action.

## Magnetic Fishing Tool

The tool shown is used for recovering broken rock drills or bits of iron or steel which accidentally get into deep boreholes. The one illustrated is of $14^{\prime \prime}$ diameter, and uses a magnet capable of lifting over two tons.
Photograph by courtesy of D.F.J. Burns Co. Ltd.

## Magnetic Door Catches

Magnetic door catches can be designed to be extremely small, efficient and inexpensive. As an example, a 'Magnadur' magnet $0.89^{\prime \prime} \times 0.59^{\prime \prime}$ $\mathrm{x} 0.18^{\circ}$, when fitted between mild steel pole plates, is capable of holding an armature with a force of between four and five pounds.



Fixed Station. SSBL-1 60 watt ( 500 double sideband equivalent) four channels $3-15 \mathrm{mc} / \mathrm{s}$. Over 4000 RCA single sideband equipments are in use the world over as

## FIXED AND MOBILE STATIONS

- Remote aerial tuning facility
- Rugged construcion for naval and military use
- Upper and lower sideband selection
- Corpatibility with double sideband systems


Noise limiter-clipper -filer for heavy interference conditions.


Mobile Station. SSRL-30M 30 vatt ( 250 w double sideband equivalent) four channels 3-15 me/s.

## Don't spoil <br> your <br> ship!

In terms of cost the solder content of your products may be negligible. In terms of quality, dependability, your reputation, it is all-important. Why not act today? You've only to tell your buyer - "Switch to Superspeed"- and you can relax in the knowledge that all your risks are underwritten by the greatest name in soldering history.

## ENTHOVEN

 incorporating Enthoven's unique 6channel stellate core, is unchallenged as the most efficient cored solder wire for general assembly work on radio, television, electronia and telecommunication equipment. But remember, too, that there is an Enthoven solder product for every other engineering and manufacturing application. And Enthoven always means the best ! If you use solder, please write today for the new edition of our brochure "Enthoven Solder Products"-or consult us, quite freely, on your particular problems.
Whenever soldering is discussed, ENTHOVEN enters the argument. It's a name that represents 150 years' experience in non-ferrous metals, and an incomparable record in research and development.

# ENTHOVE soloer PRODUCTS 

ENTHOVEN SOLDERS LIMITED SALES OFFICE \& WORKS:
UPPER ORDNANCE WHARF, ROTHERHITHE STREET, LONDON, S.E. 16 .
Telephone: BERmondsey 2014
HEAD OFFICE: DOMINION BUILDINGS, SOUTH PLACE, LONDON, E.C.2. Telephona: MONareh O39I

# DYNATRON 

SOME EXAMPLES OF THE COMPREHENSIVE RANGE OF

## HI-FI, STEREOPHONIC AND SINGLE CHANNEL EQUIPMENT



AMPLIFIER LF.15.CS


AMPLIFIER LF.16.CS


AMPLIFIER LF. 20


TUNER T.10A

## AMPLIFIERS

LF.15.CS 4 valve Push-Pull LF Ampllfler. 10 watts output. 1 voit R.M.S. input for full output. Auxillary power supply for tuners, etc. Operates with Tone Control TC.15.CS or TC.16.CS, or TC. 20.

E20.5.0.
LF.16.0S 4 valve Push-Pull LF Amplifier. 10 watts output. 1 volt R.M.S. input for full output. 2nd channel stereo use. Operates with tone control unit TC.16. \&17.10.0.
LF. 205 valve Push-Pull LF Amplifier. 20 watts output. 8 volt R.M.S. Input for full output. Auxiliary power
supplies. Operates with tone control unit TC.20. £29.15.0.


TUNER FM.2.LV

## TUNERS

T. 10 A The 'Ether Pathfinder' 9 valive AM/FM Tuner. Four wavebands AM 13-2000 metres, one waveband FM $88-108 \mathrm{Mc} / \mathrm{s}$. Variable selectivity 4 positions. R.F. Stage. Tuning Indicator. Large fully llluminated tuning dial. High sensltivity, and signal to noise ratio. £66.0.0. FM.2.LV V.H.F. Tuner Pre-tuned. 7 valves. 4 Pre-tuned FM Stations. Range $88-108 \mathrm{Mc} / \mathrm{s}$. Single shaft operation, Unlversal mounting. Ideal for FM conversion purposes. As used by BBC. FM.2.HY (8 valves).
25.0.6.

E29.3.9.


TONE CONTROL UNIT TC.16.CS


TONE CONTROL UNIT TC.15.CS
$\angle$ TONE CONTROL UNIT TC. 20

## TONE CONTROL UNITS

## (PRE-AMPLIFIERS)

TC.16.0S Stereo Tone Control Unit. The most advanced stereo-amplifier yet avallable. 4 Valve, Twin Channel. 6 Controls. Selectors for Volume, Balance, Treble and Bass. Operates with Amplifiers LF.15., LF.16., LF.20. Unique records, Tape or Radio Input features. £27.0.0. TC.15.6S Tone Control Unit. 2 valves, 6 controls. Selectors for Mixers, Volume Filter, Treble and Bass. Inputs for pick-ups, tape and radio. Operates with LF Amplifier LF. 15 or LF. 20.

ع17.10.0.
TC. 204 valve Controller Mixer. 6 Controls. (Level, Filter, Treble, Bass, Mix Level). Equaliser. Push-button switch and Channel Selector. Fully comprehensive input and output facilities. Operates with Amplifier LF.20, LF. 15.
£29.0.0
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Panorama Console Loudspeakers. Bass reflex chamber with acoustic resistance unit.
CLS.10. 10 watts max. $12^{\prime \prime}+$ two $5^{\prime \prime}$ Treble speakers $\quad$ £30. 9.0. CLS.15. 15 watts max. $12^{*}+$ one $5^{7 \prime}$ High Flux speakers £35. 5.0. CLS.20. 20 watts max. $12^{\prime \prime}+$ Trebax Treble Unit 846.10 .0 .


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## The TYPE DI/D v.h.f. SIGNAL GENERATOR



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Crystal Modulator, eliminating spurious frequency modulation.

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## the tuner

 with the outstanding specification

## Valves:

Gircuit Description:

Tuning Range:
Input and Output:

Performance:
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Power Supply: Consumption: Dimensions: Outstanding Features:
$3 \times$ EF80, ECF80, EF89, 2 crystal diodes.
R.F. stage; frequency changer and reactance valve; I.F.; 1st limiter; 2nd limiter; Foster Seeley discriminator. Automatic Frequency Control from discriminator to reactance valve; Automatic Gain Control from 1st limiter to I.F.
88-108 Mc/s., FM/VHF.
Standard 75 ohms co-axial aerial socket. Standard domestic co-axial audio output plug. Pre-wired on 3 feet of cable. Mains cable pre-wired with miniature 22 pin motor plug on 3 feet of cable. Motor socket for unswitched mains supply.
Frequency response: 20 to 20 K c.p.s. with standard de-emphasis applied.
$14 \mu \mathrm{v}$ for 40 Db quieting. $8 \mu \mathrm{v}$, for 20 Db quieting Automatic Frequency Control giving pull in from 300 K c.p.s. Output from $\pm 22.5 \mathrm{~K}$ c.p.s. deviation, 100 mv . across $25,000 \mathrm{ohms}$ impedance.
$210-250$ volts $\mathrm{AC}, 50$ cycles and $105-120$ volts AC 60 cycles.
30 V.A. Separate push button On-off switch, illuminated.
Height 31 ins. Width: $10 \frac{1}{2}$ ins. Depth: 5 ins.
$\star$ Incorporates a push-button on-off control and a built-in power pack $\star$ High sensitivity gives good reception up to 30 miles on indoor aerlal, and up to 80 miles on outdoor aerial $\star$ Automatic Frequency Coretrol provides ease of tuning without the need for a 'magic eye' $\star$ Twin Limiters cut out every vestige of background noise $\star$ Output is more than adequate fut out every vestige of background amplifier of comparable quality and design.


The Pye Mozart 10 -watt amplifier, identical in finish and size to the FM tuner.

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Stereophonic TWINSET
This equipment can be built up from an existing. Truvox R2 (monaural) Recorder. We fit the recorder with a stereo head and supply a ' $B$ ' Unit (comprising correctly matched amplifier and loudspeaker together with additional microphone and leads). Send for full details.
Available from all leading stores and radio dealers or full details from:-
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## R2 TAPE RECORDER

From the very first replay-pre-recorded or your own home-recorded programme-you'll be thrilled with its sureness of tonal quality and ease of control. Just as you recognise the voice of a friend on the 'phone or your favourite songster on radio or record, you'll know that this is the instrument you've always wanted. Designed and built by pioneers in the development and manufacture of Tape Decks and Tape Recording Ampli-fiers-TRUVOX are justly proud of an instrument that lives up to "all that the name implies." Increased production at our new, modern factory now enables us to offer this famous instrument at reduced prices-Models now available from 56 gns.
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The D31R is a high pertormance, double beam, measuring oscilloscope, designed to be built in as a permanent monitor for the operation, servicing and maintenance of all types of complex electronic equipment. It utilises an entirely new $3 \frac{1}{2}$-inch flat-faced, double gun cathode ray tube, designed to our specification and now manufactured exclusively for Telequipment. The two beams have independent brightness and focussing controls and are provided with identical Y amplifiers and a common X-sweep.

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## loudspeaker system that cancels distortion

At 40 cycles per second with 10 watts input, the Periphonic system reduces harmonic distortion to the incredibly low figure of $2.7 \%$. At 1000 C.P.S. the distortion is $0.3 \%$. Here is a triumph in sound reproduction achieved with two "aircoupled" G.E.C. metal cone loudspeakers and four G.E.C. presence units, mounted in a cabinet embodying entirely now principles. Exclusive to the G.E.c. and available as individual units, the metal cone loudspeakers give low inter-modulation distortion, extremely smooth high frequency response and excellent transient response. Designed as accessories to the loudspeakers, the presence units overcome point sound source and add "presence" to the 9 octave realism.

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## HIGH QUALITY SOUND EQUIPMENT

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at $3-0$ p.m.
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Introduced by G. A. Briggs with the collaboration of P. J. Walker.
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# H.F. Exponential-Line Transformers 

An article in the February issue of Electronic $\mathcal{E}$ Radio Engineer describes the design and constructional details of the four-wire exponential transmission lines, used at the high-frequency transmitting station at Rugby, to match balanced impedances of 195 and 565 ohms over the frequency range $4-27 \mathrm{Mc} / \mathrm{s}$.

## ARTICLES

IN THE MARCH ISSUE INCLUDE

## Transistor Equivalent Circuit

An equivalent circuit is described which accurately represents the transistor at any frequency where the device gives useful gain. It is shown that calculations based on these formulae agree favourably with practical measurements made with a representative selection of alloyjunction germanium transistors.
Saturable-Transformer Switches
The characteristics and design of switches which operate by virtue of magnetic saturation of transformer cores are discussed. The switches are applied in groups to select magnetic reading and writing heads for a magnetic-drum store, as in a computer.

ALSO
The unique monthly Abstracts and References feature compiled by the Radio Research Organization of the Department of Scientific and Industrial-Research.

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## बANVIEP SIGNAL GENERATORS

## HF SIGNAL GENERATOR Type 201

For accurate, stable sinusoidal signals of pure waveform from $30 \mathrm{kc} / \mathrm{s}$ to $30 \mathrm{Mc} / \mathrm{s}$. Output levels, which are stabilised by an amplified A.G.C. system, can be varied from 1 microvolt to 1.1 volts R.M.S. (or 2.2 volts R.M.S. unmodulated). A high output of 5 volts ( 10 volts unmodulated) is also provided from a 300 ohm source impedance. The attenuators are very accurately calibrated and have a constant 75 ohm output impedance, regardless of their setting and the frequency in use.

## SPECIFICATION

- Frequency range $30 \mathrm{kc} / \mathrm{s}-30 \mathrm{Mc} / \mathrm{s}$ - Crystal calibration - Harmonic components less than 1\% - Film scale giving actual scale length of $4 f t$. on each band 90:1 slow motion drive with logging scale - High output of 5 volts - Output variable from 1 microvolt to 1.1 volts at 75 ohm impedance - Output level constant over entire frequency range - Modulation depth variable from $0-100 \%$ - F.M. and A.M. less than $300 \mathrm{c} / \mathrm{s}$.



## SPECIFICATION

- Frequency range $30 \mathrm{c} / \mathrm{s}-30 \mathrm{kc} / \mathrm{s}$ - Frequency stability $+0.25 \%+0.5 \mathrm{c} / \mathrm{s}$ - A screened and balanced transformer enables balanced, unbalanced and fully floating outputs to be obtained - A 600 ohm constant impedance attenuator provides steps of $0,20,40$ and 60 db of attenuation under all output conditions.



## DELIVERY - EX STOCK

## LF SIGNAL GENERATOR Type 702

A resistance-capacitance type of oscillator is employed to obtain a stable output waveform with very low harmonic distortion.
The output circuit of the instrument uses a screened and balanced transformer to enable unbalanced, balanced and floating conditions to be obtained, and a 600 ohm constant impedance attenuator provides steps of $0,20,40$ and 60 db of attenuation under all output conditions. A single level control is incorporated for setting the output to any desired value, and the amplitude is monitored by a diode valve-voltmeter calibrated both in open circuit volts and in db relative to 1 milliwatt in 600 ohms.

# SIGNAL GENERATORS 

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## CONTINUOUS MESSAGE TRANSMISSION

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## with the A.T.E. Series TAA

TELEGRAPH TERMINALEQUIPMENT

This compact unit, comprising an electronic distributor and an operating head, is equipped with two tape readers arranged to operate alternately for the automatic transmission of messages. The space taken is kept to a minimum because
the operating head measures only $6 \frac{1}{2}{ }^{\prime \prime}$ in width. An operator can supervise two or more units without difficulty. Breaks in transmission time which often occur in normal operating practice can be eliminated and maximum circuit availability is achieved.

Special features are:
(A) Automatic transmission of:1. Station call slgn and serial number before any message. 2. A 100 -character test message. 3. Routine message.
4. Station call sign and serial number at predetermined intervals number at predetermined intervais
(B) Speeds $45.5,50,56.8$ and 75 bauds.
(C) $7,7 \frac{1}{2}$ or 8 -unit code.
(D) Stopping a transmission and restarting at the point when a transmission was interrupted, or at any point without starting a new preamble. In the case of a tape, this facility may be used to repeat, or omit a particular portion of the message.
(E) Transmitting the test message direct from the distributor.

## Operating Head

Dimensions $6 \frac{1}{2}^{\prime \prime} \times 22^{\prime \prime} \times 9^{\prime \prime}(16.5 \mathrm{~cm} . \times$ $56 \mathrm{~cm} . \times 23 \mathrm{~cm}$.)
Welght 31 lb . ( 19.1 Kg .)

## Electronic Distributor

Dimenslons $19^{\prime \prime} \times 8 \frac{1_{4}^{\prime \prime}}{} \times 3 \frac{1^{\prime \prime}}{2}$
$(48 \mathrm{~cm} . \times 21 \mathrm{~cm}, \times 9 \mathrm{~cm}$.) Weight $23 \mathrm{lb} .(10.7 \mathrm{Kg}$.


that elusive works manager ...



## in 10 seconnds

NO LOUDSPEAKERS, BELLS or FLASHING LIGHTS
only the man who's wanted knows and replies.

Selective Induction is saving time, money and worry in Offices, Factories, Hospitals, Hotels, Departmental Stores etc., all over the Country. All key personnel carry small transistorised receivers bearing a number. When they are wanted their numbered key is pressed on a small transmitter. Immediately they must respond to the URGENT 'PEEP PEEP' in their pockets which summons them and them alone to Action! A verbal message can be transmitted if desired.

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- Designed for the man who cannot afford to be tied to his office.
- Equally suitable for large or small concerns.
- Low purchase price-virtually no indoor wiring-low rental terms.

Write or 'phone for further particulars - WE CAN BE FOUND IN TEN SECONDS



During the ten years that have elapsed since we first blazed the To us it is a matter of pride that so many features of that first Ferrograph have since become standard practice and embodied in the designs of other manufacturers．

To－day we again look to the future and initiate a policy to ensure that－no matter how Tape Recording develops or for what purpose it is used－every Series 4 Ferrograph can be readily adapted at any future time for a variety of applications．To achieve this，space has been provided under the Head Cover to permit Head changes and additions for monitoring，stereo recording，stereo playback，dual track stereo to the new American standard or for lower track use．Such Heads are designed to be plugged in and are fitted with rocking facilities for azimuth correction．

This Ferrograph development，for example，permits any Series 4 A to be instantly converted into a Series ${ }_{4} \mathrm{~S}$ merely by plugging in the additional stereo Head costing seven guineas．

Simultaneously the following important design improvements have been incorporated：
＊The cap．stan motor is resiliently mounted for quiet operation．
＊Function Switch Knob is re－fashioned for greater ease of operation．
＊Brief Stop（or pause control）is now a standard fitting on all Ferrographs．
＊The Head Cover－a one－piece moulding－is hinged for easy tape loading．
＊The Turns Counter－now gear driven－is accurate to a turn．
Finally，to conform to our policy of rationalisation，the Ferrograph will be supplied only in one standard colour finish－a handsome two－tone grey．It is available in two forms，either as a portable or as a unit for installation into your own cabinet，in thes following models：－

## Series 4A

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| Input voltage | $190-260$ voles, $50-60 \mathrm{c} / \mathrm{s}$. <br> ( 115 v . model available) |
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| Output voltage | $4.5-14$ volts (Variable). |
| Output current | 500 mA . |
| Voltage regulation accuracy | $\pm 0.15 \%$ |
| Output impedance | <0.1 ohm. |
| Ripple and noise | 5 mV . p-p. |
| Ambient working | $25^{\circ} \mathrm{C}$. |
| Overall cabinet dimensions | $10 \frac{1}{2 \prime \prime}^{\prime \prime} \times 7 \frac{1^{\prime \prime}}{} \times 8 \frac{1}{1 / 2}$ deep. |
| Weight | $7 \frac{1}{1} \mathrm{lb}$. |

Models LTE-12.5-0.75 \& - 1.5

| Input voltage | 205-250 volts, $50 / 60 \mathrm{c} / \mathrm{s}$. <br> (115v. model available) |
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| Ripple and noise | 10 mV . p-p. |
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## ABRIDGED DATA

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| :---: | :---: | :---: | :---: | :---: | :---: |
| SINGLE PHASE FULL WAVE C.T. (2 Diodes) | $\begin{aligned} & 19 \mathrm{E} 2-1.1 \\ & 19 F 2-1-1 \\ & 19 \mathrm{G} 2-1-1 \end{aligned}$ | $\begin{aligned} & 28+28 \\ & 21+21 \\ & 14+14 \end{aligned}$ | $\begin{aligned} & 25 \\ & 18 \\ & 12 \end{aligned}$ | $\begin{aligned} & 40 \\ & 40 \\ & 40 \end{aligned}$ | £10. 0.0 d <br> £ 7.15.0d <br> £ 5. 7.6d |
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# " BELLING-LEE" NOTES CONTACTS PLUGS and SOCKETS 

## 3 rd of a series

In the last issue we dealt with single plugs and sockets and we have a few more general notes applicable to both single and multi-plug assemblies. A resilient socket must always be closed below the diameter of the bottom limit of the appropriate pin. A pin on the top limit must not, after repeated insertions, distort a socket on either the top or bottom limit to impair its contact resistance with a pin on the bottom limit. The force required to insert a pin will be greatest at the point of entry, before the socket has accommodated itself to the pin diameter.

Now we propose to refresh our

memories regarding multi-way plugs and sockets. The difficulties in manufacture and gauging are considerable, but are generally eased by designing the component with the plugs and/or sockets floating, but the float must be controlled, so for practical purposes a locating dowel is essential, and dimensionally must be carefully related to the layout of the plugs and/or sockets, so that at no time may a pin find itself down the outside of a socket. Where the component takes the form of Unitor, i.e., a plug and socket fixed to two chassis units, the fixing hole must be very accurately dimensioned. The contact pressure between any pin and its socket need only be in the region of a few ounces, but the accumulated pressures set up in say a 24 -way Unitor, with the inevitable disposition of plugs and sockets, although within manufacturing limits, may result in an extraction force in the region of a pound per pin, which in the case of a free member really calls for some mechanical aid for disconnection. This is often associated with the cover.
Every effort is taken by manufacturers to reduce these pressures consistent with the retention of low contact resistance, because if thev
are too high there is great danger in the soft silver or gold plating being removed, leaving the part liable to corrosion and high resistance.
Some systems of printed circuit connectors use the actual printed board as the plug, in such cases great care must be taken that the pressures involved will not damage the deposited copper contact bars.
A number of factors come in here, e.g., is the component likely to be disconnected many thousands of times or only occasionally?
Many types of multi-way plugs and sockets provide two pins of larger diameter to carry heater currents, and there must be adequate spacing between pins, their associated sockets and the chassis for the voltages likely to be involved.

The material used as the insulant must be considered in relation to possible applications, anti-tracking, moisture resistant, high softening point, etc. Capacity or screening between certain pins, or pins and chassis may have to be considered.
The question of sealing will be dealt with generally later on in this series.
The materials from which plugs and sockets are manufactured are most important. When plugs are solid, brass is adequate, suitably protected. If sockets are turned, again brass is suitable, but care must be taken to ensure that the protective plating finish "throws" to the bottom of the socket, otherwise corrosion will commence there and spread upwards. There is an increasing tendency to favour sockets pressed from beryllium copper suitably treated, this can be hard and springy, and consideration must be given to the presence of burrs which can quickly remove protective surfaces in the plug. Leads will have to be soldered to the back end of both plugs and sockets, and it is good practise to present these suitably finished so that they will easily accept solder. When a bucket spill is provided, this should be tin dipped or silver plated to the bottom of the bucket, so that the conductor will "wet" with solder all the way. Spills left silver plated may be further protected by petroleum jelly dip. With pressed sockets, care must be taken to prevent solder from running, and damaging the contact surfaces.

In the next issue we will deal with coaxial plugs and sockets.

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It is a 12 -way block, flexible, to fit uneven surfaces or shaped contours and is unbreakable. It can be sub-divided with a knife. Screw terminals are secured against accidental withdrawal and loss, even when upside down, and are vibration resistant.
Conservatively rated at 2 amperes, 250 v . A.C. (350v. D.C.) terminals are voltage proof at 2000 v . between adjacent terminals and chassis. Breakdown voltage is 3500 v . and 3000 v . D.C. (min.) between terminals and chassis. Permissable voltage based on creepage distance (B.S.415) between terminals 1135 v . peak; between terminals and chassis 320 v . peak. Capacitance 0.4 pF between adjacent ways and 0.7 pF to chassis. The material is tracking resistant.

See also p. 101.

Most "Belling-Lee" products are covered by patents or registered designs, or applications therefor.

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104 Waveguide Transmission
106 World of Wireless
109 European Television Stations
117 Time Past-Beam and Broadcast By P. P. Eckersley
121 Letters to the Editor
126 Physical Society's Exhibition
135 Relativity-2
139 Evaluating Aerial Performance-2
144 Short-wave Conditions
145 The Bifilar-T Circuit-2
149 News from the Industry
150 March Meetings
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[^10]
## Transistors

## A Simple Temperature-control System

This system involves the use of transistors to keep a block of copper at constant temperature and thus to provide thermostatic conditions for the transistors of any experimental or industrial circuit. The collector leakage current of a transistor varies with ambient temperature. The drift which thus arises affects the stability of d.c. amplifiers and test equipment.


COPPER BLOCK (approx. full size)
A copper block was constructed to the design shown in the drawing above. The block is made circular for ease of machining and for uniform heating effect. A layer of aluminium foil covering the heater coil reflects heat into the block.

A closed-loop servoméchanism controls a power transistor, which supplies the current for heating a coil wound uniformly around the block. The sensing element, used to provide the error signal, is an a.f. transistor mounted in the block. The leakage current $I^{\prime}{ }_{c o}$ of this transistor changes markedly with temperature, so that suitably amplified changes in I' ${ }_{c o s}$ are used to control the current through the heating coil. It was found that an initial current through the coil of 1.5A gave a good heating rate without much overshoot. For a 12 V supply the resistance of the coil becomes $8 \Omega$. To provide sufficient length of wire to wind uniformly, five strands of thin enamelled constantan wire are wound in parallel.
The control circuit is shown here. The connection of Tr 3 and Tr 4 allows the OC 72 to drive the OCl 6 without danger of excessive dissipation. The gain of the combination is then the product $\alpha_{1}^{\prime} . \alpha_{2}^{\prime}$. The combined base-emitter voltage provides sufficient voltage for the collectors of Tr 1 and Tr 2 ; at
the same time it limits the maximum possible dissipation of these transistors. The variable resistance provides a path for some of the leakage current, allowing adjustment of the temperature to which the block is set.

When the circuit is switched on there is very little leakage current flowing in the sensing transistor. Accordingly the OC16 output transistor 'bottoms' and the current flowing in the heating coil is very nearly $\mathrm{V}_{\mathrm{cc}} / \mathrm{R}_{\mathrm{H}}$ ( $\mathrm{V}_{\mathrm{cc}}$ is the supply voltage and $\mathbf{R}_{\mathbf{H}}$ the heating coil resistance). The block heats rapidly to nearly $40^{\circ} \mathrm{C}$, at which temperature the leakage current of Tr 1 rises rapidly and the OCl 6 is cut-off, and remains cut-off till the block temperature has dropped to the set vaiue. There is a slight overshoot of temperature because the junction temperature of Tr 1 does not react to changes instantaneously.

Trl is the temperature-sensing transistor. Its collector voltage is limited to about 1 V , whilst the working voltage is about 0.6 V . To limit the maximum change of temperature inside the block to $0 \cdot 1^{\circ} \mathrm{C}$, the maximum variation of $\mathrm{I}^{\prime}$ o must be $0.012 \%$. The actual stability achieved over at least 24 hours satisfies this condition.

A long-tailed-pair amplifier using germanium transistors was tested in the block. A variation of ambient temperature from 20 to $35^{\circ} \mathrm{C}$ resulted in a change of $0.75^{\circ} \mathrm{C}$ in the block. The drift, referred to the input, was about $75 \mu \mathrm{~V}$, or $5 \mu \mathrm{~V} /{ }^{\circ} \mathrm{C}$.

[5518]

CONTROL CIRCUIT


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It will not come as a surprise to "Wireless World" readers to be told that we-Belling \& Lee Limited-do not make cookers or electric fires. Neither are television aerials our main interest, although we are considered specialists and consultants in all matters relating to the reception of V.H.F. signals, be they television, F.M. sound or professional communication.

Since the New Year we have opened a new aerial research laboratory divorced entirely from the main research facilities which are, of course, fully equipped for the most stringent testing of electronic components. The new laboratories have been specially equipped for the research and development of aerials, aerial distribution systems including amplifiers, interference filters and last but by no. means least, screened compartments.
Whilst domestic television aerials take up most of the time, there are always "special purpose" professional type aerials required, for example we have always developed and made the aerials for the T.V. pilot transmitter used by I.T.A. We also make aerials for the B.B.C. and "all round looking" aerials as used on lightships.

Aerials can rarely be finalised in a laboratory. There is inevitably field testing to be carried out, for which purpose we have a specially built


A general view across one side of the laboratory showing a screened comportment in the for corner. There are microwave facilities ovailable and a very comprehensive range of some of the finest measuring equipment ovailable on either side of the Atlantic and a lot more to come.


The Mobile Research Unit is complete with work bench and seats. It has self-contained 50 cycle 250 volt supply taken to a separate distribution system. The mast is winch-operated and may be raised to 32 feet. When on location the vehicle is fitted with a Television receiver and necessary measuring equipment.


One of the most modern and interesting pieces of equipment available in the laboratory is the impedance measuring set which plots the results directly on to a Smith chart thereby cutting out much tedious work.
mobile research unit, with its own power supply. We have also a number of permanent field sites that are in use in all but the very worst of weather. Delightful in spring and summer, when it comes, not so good in the winter, but the show goes on.

We have some very interesting equipment enabling us to work up to 2,000 M/c. Even Smiths charts are no longer laboriously plotted by hand, they are taken off a Diagraph.

We are looking for some extra staff, one senior engineer preferably of B.Sc. standard or equivalent, and a few junior engineers. If any readers would like to hear more about these jobs please write to the Secretary.


For work on any of our three field sites, we use 4 wheel drive vehicles as illustrated which carry a small petrol electric set as the supply unit.

[^11]
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Two-segment solid flared core
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To those concerned with driving electron beams around almost impossible angles, these new Ferramic* Scan Coil Cores bring a large measure of relief. Desirable electrical characteristics have been combined with sound mechanical design, facilitating the rapid production of coils.

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* FERRAMIC is a Registered Trade Mark.


## Aspects of design

This is the ninth of a series of special features dealing with advanced problems in television and radio circuit design to be published by Siemens Edison Swan. The Ediswan Mazda

Applications Laboratory will be pleased to deal with any questions arising from this or other articles, the tenth of which will appear in the April 1959 issue.

## REQUIREMENTS OF AN R.F. STAGE

In a television front end tuner consisting of a cascode double triode R.F. stage followed by a triode-pentode mixer the R.F. stage has many functions to perform.

One of its requirements is to reduce the effect of the mixer noise which in this case would have a noise factor of the order of 40 at $200 \mathrm{Mc} / \mathrm{s}$. There must therefore be a stage of signal frequency amplification ahead of the mixer to render the mixer noise negligible in comparison with the noise of the input circuit and this can be done by providing a sufficiently high gain between the mixer grid and the grid of the R.F. valve.

The R.F. stage in providing the necessary gain must itselt make the least possible contribution to the receiver noise as well as meet the other necessary requirements of the input stage. These can be summarised as:

To attenuate the local oscillator voltage and prevent radiation via the receiving aerial.

To provide a reasonable match for the aerial feeder by means of an input transformer of the required bandwidth.

To handle large signals with the minimum cross-modulation effects between sound and vision carriers.

To maintain these characteristics at reduced gain with A.G.C. bias.

In designing an R.F. stage to give the minimum noise figure it is necessary to reduce as far as possible noise produced in the valve itself arising from normal shot noise and induced grid noise. In general a valve with the highest mutual conductance per milliamp of anode current working at a high cathode current density will provide the highest gain and the lowest noise figure. With this end in view, valves Kave been designed having a high slope per milliamp of anode current and a corresponding improvement in gain and noise performance. Unfortunately
their comparatively high cost prevents wide adoption, except .n receivers intended for real fringe area reception, but an improved version of the 30 L 1 presents an attractive compromise.

As an alternative to using a more costly valve in fringe sets only, it is possible to provide the improved valve costing not much more than the 30L1. Such a valve could be fitted in every set and would have a noise performance suitable for use in most fringe areas and at the same time would provide more gain on both Band I and III-a useful feature where a recelver is required to operate from a simple indoor aerial.

The Ediswan Mazda 30 L15 has been designed to meet these requirements and it gives a gain increase of 4 to 5 dB over the 30 L 1 with an average noise factor of 5.5 dB on Band III compared to 7.5 dB obtainable with the 30 L 1 .

If the 30 L 15 is plugged into an existing design in place of the 30 L 1 and only circuit re-trimming is carried out a gain improvement of up to 3 dB and a noise improvement of ${ }^{-1}$ to 1.5 dB may be realised. To take full advantage of the valve's capabilities some changes will be required to the bias resistor and slight readjustment of the coupling to the aerial feeder may be necessary.

## NOISE FACTOR AND A.G.C.

The mistuning of the input circuit when A.G.C. bias is applied to the R.F. valve produces a change in the response curve of the tuner as shown in Fig. 1.

The amount of mistuning, or "tilt", with a given valve is largely controlled by the working $Q$ of the aerial circuit. If a higher working $Q$ aerial circuit giving more than about 1.5 dB of tilt could be accepted, lower noise figures than those given above could be obtained. It follows then that a full assessment of the performance of the input stage should consider noise factor and. gain in relation to the amount of tilt, or A.G.C. mustunung, present in any particular tuner. The practice in these Laboratories is to design and quote performance figures on the basis of a tilt of up to 2 dB for a gain reduction of ten times.


## IMPROVED MIXER VALVE

A companion valve to the 30 L 15 has been designed to give higher conversion gain in the mixer stage of a television tuner. It has the same base connections as the 30 C 13 which it has superseded and is therefore specially suitable for printed circuit use but can also be used in a wired tuner to provide approximately 3.5 dB more gain than the 30 Cl .

The new valve will be described more fully in a later article on Aspects of Design.

FIG. 1.

Effect of A.G.C. bias on overall tuner response curve

## NEW HIGH PERFORMANCE CASCODE AMPLIFIER

## EDISWAN MAZDA 30L15.

The 30L15 is a V.H.F. twin triode specially designed for use as a cascode amplifier in television tuners where a high gain and good noise performance are required. The 30 L 15 will provide a gain improvernent of 4 to 5 dB and a noise façtor improvement of 2 dB as compared with the 30 L 1 .
Heater Current (amps)
Ih
Heater Voltage (volts)
$V_{h}$
0.3

MAXIMUM DESIGN CENTRE RATINGS
Anode Dissipation, either section (watts)
$\mathrm{P}_{\mathrm{a} \text { (max) }}$
Cathode Current, per section (mA) .. $I_{k(m a x)}$
Anode Voltage (volts) .. .. .. Va(max) 2.0

Negarive Grid Voltage (vults) $\quad . \quad \mathrm{Vg}_{\mathrm{max}}$ ) 250

Grid to Cathode Resistance, section
1 (k $\Omega$ ) Cathode Resistance, section
Grid to Cathode Resistance, section $2(\mathrm{k} \Omega)$
$\mathrm{Rg}^{\prime}$ - $\mathrm{k}^{\prime}$ (max) 500

Effective Grid to Earth Resistance, section $2(\mathrm{k} \Omega$ )

22*
$\mathrm{R}_{8}-\mathrm{E}$ (max)

* Grid current bias.
** With potentiometer bias from anode supply.


## INTER-ELECTRODE CAPACITANCES ( $\mathbf{p} \mathbf{F}$ ) $\dagger$

Grid 1 to Anode 1
Anode 1 to Cathode 1, Heater, Shield Grid 1 to Cathode 1, Heater, Shield. Anode 2 to Cathode 1, Heater, Shield Grid 1 to Anode 2

| $\mathrm{Cg}^{\prime}-\mathrm{a}^{\prime}$ | 1.5, |
| :---: | :---: |
| $\mathrm{Ca}_{\mathrm{a}^{\prime}-\mathrm{k}} \mathrm{k}^{\prime} \mathrm{h}, \mathrm{s}$ | 1.9 |
| $\mathrm{cg}^{\prime}-\mathrm{k}^{\prime}, \mathrm{b}, \mathrm{s}$ | 3.1 |
| $\mathrm{Ca}_{3^{\circ}-\mathrm{k}} \mathrm{k}^{\prime}, \mathrm{h}, \mathrm{B}^{\prime}$ | 3.6 |
| $\mathrm{Cg}^{\prime}=\mathrm{a}^{\prime \prime}$ | 0.005 | $\dagger$ Inter-electrode capacity with holder capacity balanced out but with cylindrical screen can.

BASE:
NOVAL-B9A



MAXIMUM DIMENSIONS
Overall Length mm ) 56
Seated Height (mm) 49 Diameter
(mm) 22.2

## TYPICAL CASCODE OPERATION

Conditions given for circuit with the second section using potentiometer bias from the anode supply, making $\mathrm{V}_{\mathrm{g}^{*}-\mathrm{E}}=99 \mathrm{~V}$. Anode supply voltage (volts) ..

| Anode decoupling resistor, section $2(\mathrm{k} \ddot{\Omega})$ | $\mathbf{R}_{a^{\prime}}$ | 2.2 |  |  |
| :--- | :--- | :--- | :--- | ---: |
| Anode current (mA) | $\cdots$ | $\cdots$ | $\cdots$ | $\mathbf{I}_{n}$ |
| Self bias resistor, section $1(\Omega)$ | $\cdots$ | $\mathbf{R}_{g^{\prime}}$ | 100 |  |

Self bias resistor, section $1(\Omega) \quad \cdots \quad \mathbf{R}_{\boldsymbol{g}}$
Mutual conductance, section 1 (mA/V)
gm
Combined mutual conductance (mA/V)
A.G.C. Voltage (volts) to give $\triangle I_{B}$ $\Delta V_{g}=0.1 \mathrm{~mA} / \mathrm{V}$
$V_{g^{\prime}}$
Input capacity working ( pF )
Change in input capacity by biasing to
cut-off ( pF )

Characteristic Curves of Ediswan Mazda Valve Type 30L15-each section


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and low thermal resistance allows High Output Powers

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| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MAXIMUM CURRENT 250 ma | MAXIMUM CURRENT 250 mA |  |  | MAXIMUM CURRENT 350 ma |  |  |
| GET 106 | GET 114 | GET 103 | GET 104 | GET 115 | GET 116 | GET 105 |
| $V_{\text {ce }}$ max (volts) | 15 | 30* | 30 | 15 | 30* | 40* |

* $\mathrm{R}_{\mathrm{e}} / \mathrm{R}_{\mathrm{b}}>0.03$ † Also supplied in matched pairs

Further details of these and other semiconductor devices manufactured by the G.E.C. are obtainable from the
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Read "Transistor Audio Frequency Amplifiers", written by the staff of G.E.C. Research Laboratories, published by Iliffe \& Sons Ltd. Price 21/-
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## RADIO TELEPRINTER OPERATION

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A comprehensive brochure is available on request. Please ask for Plessey Publication No. 168.

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$\star$ The meter fitted for reading signal level will also read bias voltage to enable a level response to be obtained under all circumstances. A control is provided for bias adjustment to compensate low mains or ageing valves.
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$\star$ The 0.5 megohm input is fully loaded by 18 millivolts and is suitable for crystal P.U.s, microphone or radio inputs.
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| 6J5 |  | $12976 T$ | 7/10 | ECF80 12/m | E241 10/- | PZ30 | 12/m | 266 | $9 / 6$ |
| 6.J6 | $4 / 6$ | 20.D1 | 6 | 6 | E280 8/6 | 41 | $2 / 6$ | Z719 | $7 /$ |

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## 1959

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#### Abstract

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## WESTON MODEL 772 TESTMETER



| A.C. VOLTS | D.C. | A.C. CUR- |
| :--- | :--- | :--- |
| 2.5 v. | CURRENT | RENT |
| 10 v. | $100 \mathrm{mic} \cdot / \mathrm{a}$. | 500 ma. |
| 50 v. | 1 ma. | 1 amp. |
| 250 v. | 10 ma. | 5 amp. |
| $1,000 \mathrm{v}$. | 50 ma. | RESIS. |
| D.C. VOLTS | 100 ma. | ANCE |
| 2.5 v. | 500 ma. | 100 ohms |
| 10 v. | OUTPUT | 1,000 ohms |
| 50 v. | METER | 100 k. ohms |
| 250 v. |  | 10 megohm |
| $1,000 \mathrm{v}$. |  |  |

1,000 v.
Supplied in perfect working order complete with rexine carrying case, internal batteries and instructions, E8/19/6 each. P/P. 4/-.

COSSOR DOUBLE BEAM OSCILLOSCOPE


TYPE 339

Operation $110 / 200 / 250$ volts A.C. 120 watts. Time Base 10 positions. 6 cps . to $250,000 \mathrm{cps}$. Amplifier 10 eps. to $2,000,000$ cps. Sensitivity, YI.Y2.3.1 $v$. D.C. I.I. v. rms. X. 2.25 v. D.C. .8 v. rms.
Supplied in good working order complete with handbook and circuit. £27/10/- each. P/P. £1.
 $3 / 6$ per reel. $P / P$. $1 /$.
LEACH AERIAL CHANGEOVER RELAYS. 12 v. D.C. double pole transmitter type. New, boxed, $7 / 6$ each. P/P. 9d.

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 TF517. Frequency coverage $10-18 \mathrm{mc} / \mathrm{s}$. $33-58 \mathrm{mc} / \mathrm{s}$ and $150-300 \mathrm{mc} / \mathrm{s}$. Operation $200 / 250$ volt A.C. Supplied in good working order, $\mathrm{f} 12 / 10 / \mathrm{F}$ each. P/P. $10 /=$750.WATT AUTO TRANSFORMERS. EX Admiralty, fine jobs. Tapped from 110 to 230 volts. Brand new, $69 / 6$ each. P/P, 5/-.
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MUIRHEAD PRECISION STUD SWITCHES
 4 banks, 1 pole 24 positions each bank. Self cleaning heavy duty contacts. Brand new, $17 / 6$ each. P/P. I/.,
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Complete set of new valves $2 \times 66,2$ U50, 2 DH63, 2 KT63, 6 KTW6!. Also set of resistors, condensers, pors, toggle switch and output transformer. Supplied new and boxed, 59/6 each. P/P. 4/6.

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12 v. D.C. input, watrs, 50 cycles ourput. Housed in wooden case and fitted with voltage control slider resistance, switch, plugs and A.C. mains voltage output check perfect condition, individually tested. $\mathbf{6 9 / 1 9 / 6}$ each. P/P. 10/-

PARMEKO MAINS TRANSFORMERS, Input 230 volts. Output $350 / 0 / 350$ volts 150 mA 6.3 v. 4 amp., 5 v. 4 amp . Brand new, $32 / 6$ each. P/P $2 / 6$.
R.II55 " N "TYPE SUPER SLOW MOTION DRIVES. Brand new, $12 / 6$ each. P/P. I/-.

$100 \mathrm{KC} / \mathrm{S}$ CRYSTALS. 3in. spacing, $15 /$ each CV967 lin. C.R.T. 4 v. HEATER. Suitable for oscilloscopes, etc., 25/- each. P/P. 1/-.CRYSTAL MICROPHONE INSERTS. Only 4/6 each. P/P. 6d
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MIDGET NIFE ACCUMULATORS. Single units, ideal for models, etc., $2 / 3$ each. P/P. 9d. I2-VOLT MOBILE AMPLIFIERS. Ex Admiralty. Mic. or gram. inputs, 10 watts output to 3 or 15 ohm speakers. Not new but in good working order, $€ 8 / 19 / 6$ each. P/P. 5/-.

## RCA ET 4336 PLATE TRANSFORMERS.

 Special release, brand new in original transit cases. Primary tapped $200 / 250$ v. 50 cycless. Secondary, 2,0000/0/2,000 v. 400 ma ., tapped $1,500 / 0 / 1,500 \mathrm{v}$. Price $£ 12 / 10 /$ each. P/P. $£ 1$.

## AMERICAN MULTI-RANGE TESTMETERS



1,000 ohms per voit, 400 microamp basic movement. A.C. VOLTS D.C. VOLTS $\begin{array}{ll}2.5 \mathrm{v} & 2.5 \mathrm{v} \\ 10 \mathrm{v} & \\ 50 \mathrm{v} & 10 \mathrm{v} .\end{array}$ $\begin{array}{ll}50 \mathrm{v} & 50 \mathrm{v} . \\ 250 \mathrm{v} . & 250 \mathrm{v} .\end{array}$ $1,000 \mathrm{v}$ 1,000 v 5,000 v. 5,000 v. D.C. CURRENT RESIST'CE $1 \mathrm{ma} \quad 500$ ohms 100 ma . $\quad 100 \mathrm{k}$, ohms I amp. DECIBELS -10 to +69
ALL BRAND NEW. COMPLETE WITH INTERNAL BATTERY TEST PRODS AND INSTRUCTIONS. E5/I9/6 EACH. P/P 3/-


## LORAN indicators APN4

Another release, all brand new These units contain a 5CP1 C.R.T. 14 6SN7 valves, 8 6H6 3 6SL7, I $6 S J 7$ and a $100 \mathrm{kc} / \mathrm{s}$ crystal, also many thousands of useful components. Ideal for conversion to an oscilloscope 65/19/6 each. Carriage 10/-.

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200/250 volt A.C. input. Output 250 volts 150 mA . and 6.3 volts 6 amps . Fully smoothed, double choke and paper condensers, fused and fitted with input and output plugs. Sockets are provided on the front panel for meter check. Housed in grey metal case for standard 19in. rack mounting. Supplied brand new, 59/6 each. P/P 7/6.

## FERRANTI TESTMETERS TYPE 0.

| D.C. | A.C. | D.C. | Oh |
| :---: | :---: | :---: | :---: |
| VOLTS | VOLTS | Current |  |
| 3 v . | 15 v . | 7.5 ma . | 25,000 |
| 30 v . | 30 v . | 30 ma . |  |
| 150 r . | 150 v . | 150 ma . |  |
| 600 v . | 600 v . | 750 m |  |

500 ohms per volt on all ranges. B.S.S. first grade aceuracy on all self contained ranges. Supplied in perfect working order complete with leads, battery, instructions and rexine covered carrying case.
Price $72 / 6$ each. P/P 2/6.


6-VOLT VIBRATOR POWER PACKS, Output 120 volts 30 mA . Fully smoothed, uses standard 4 -pin mallory vibrator. New and boxed, 12/6 each. P/P 2/-.

## METER BARGAINS

50 M croarup. D.C. M/C., fush so. Ln. 50 Microamp. D.C. M/G., prof. rd. $2 \mathrm{tin}$.
100 Mieroamp. D.C. M/C., fueh rd. 2 iti . 500 Microamp. D.C. M/C., luubh rd., 2in.
 1 Milliamp. D.O. M/C., flush rd. 2/in. 50 Milliamp D.C. M/C., Aush sq. 2in. 200 Milliamp. D.C. M/C., fush rd. 2 if ${ }_{30}$ Amp. D.C. M/C., flush re.., $\downarrow$ inin. I5 Volt D.C. M/C., flueb rd., 1 ini. ${ }_{120}{ }^{2}$ Volt D.C. M/C., proj. rd. 2in. 120 Volt D.C. M/C., fuush rd. 3in. 300 Yoit A.C. M/L., flush rd. 24 ln .
300 Volt A. ${ }_{500}^{300}$ Volt AC. M/T, Aush rid fush rd., 2 |in. 500 Volt ALL MRAND NEW AND TESTED

## $119 / 6$ $49 / 8$ <br> $15 / 6$ $19 / 6$ | 195 |
| :--- |
| 25 |
| 716 | 916 1016 916 18 1016 1016 $32 / 6$ $32 / 6$ $25 /-$ $25 /-$

MARCONI TF. 373 UNIVERSAL IMPED ANCE BRIDGES. Recond. as new, 655 each. P/P 10/-.
A.C. MAINS VOLTAGE REGULATOR TRANSFORMERS. Input 230 volts. Variable output 185 to 250 volts at 24 amps. $\mathrm{C} / 2 / 10 /=$ each. P/P. 10/-.


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R.II55 COMMUNICATION RECEIVERS MODELS L AND N. Incorporate the TRAWLER BAND. Coverage 200 to 500 $\mathrm{kc} / \mathrm{s}$, and $600 \mathrm{kc} / \mathrm{s}$. to $18 \mathrm{mc} / \mathrm{s}$. Supplied in perfect working order, aerial tested, £/2/19/6 each, P/P. 7/6. Standard model B also available fitted with improved N type drive, periect order, $£ 7 / 19 / 6$ each. Combined A.C. mains power pack and audio output stage suitable for either of above, 85/Instruction booklet supplied with each receiver.


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BRAND NEW instruments by famous manufacturer. Housed in polished teak case. Moving iron movement reading A.C. or D.C. volts on 2 ranges, $0-160 \mathrm{v}$. or $0-320 \mathrm{v}$., 8 in . mirror scale. Accuracy within $2 \%$. Supplied at a fraction of original cost, $\mathbf{E 5 / 1 9 / 6}$ each. P/P $3 / 6$.

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Small, compact, accurate instrument. Resistance measurements from 0 to 20 k . ohms, D.C. volts from 0 to 500 v ., A.C. volts from 0 to 500 v., D.C. current from 0 to 500 mA . Supplied in perfect working order, complete with leather case and leads. $\mathbf{6 5 / 1 0 / -}$ each. P/P $2 / 6$.



SHERRILL U.S.A. COMPASS, magnetic type, directional indicator. Complete with buile-in variable horizon corrector, induced error corrector, variable course indicator, deviation calculator, ete., with provisions for internal lighting. Brand new. € $3 / 10 /-$ ea. P. \& P. 2/-


SOUND POWER TELEPHONE UNIT, no batteries required. Fitted with neon indicator lamp and high pitched buzzer, operated by built-in generator. Entirely self-contained, ex Admiralty. Rebuilt and guaranteed working. Effective up to half a mile, waterproof.
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NEW
NROL IOWat TROL. 25 ohms, plus 25 ohms. $7 / 6$ each. P. \& P. $1 / 6$.

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AUTO TRANSFORMERS, step up, step down. 110-200-220-240 v. Fully shrouded. New.
300 watt type $£ 2 / 2 /$ - each. P. \& P. $4 / 6$. 500 watt type
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9 amp. D.C. Hot Wire W.R.

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Milliammeters
5 mA . M.C. 2 in . fi. sq.
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$50 \mathrm{~mA} . \mathrm{M.C.}^{2} \mathrm{in}$. fl. sq . 200 mA. M.C. 2 tin. fl. rnd 500 mA . M.C. $2 \frac{1}{2} \mathrm{in}$. fl. rnd. 1 mA. M.C. 2 in . sq. 11 .. 500 Microamp latest type Ernest Turner 2in. fl. rnd. with mounting ring and scaled 0-5, moving coil
500 Microamp ex. equip. M.C.
2 in. rnd. scaled $0-15$ and
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A.C. and D.C. volts, ohms and milliA.C. and D.C. volts, ohms and milliamps., basic movermeni amps., in wooden carrying case, complete with test prods, new batteries,
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P. \& P. 2/6.


MIDGET ROTARY TRANSFORMERS. $2 \frac{1}{4} \mathrm{in}$. dia. $\times 4 \frac{1}{2} \mathrm{in}$. input 11.5 volr. Output $310 / 365$ volts at 30 mA . Brand new. 17/6 each. P. \& P. 1/6.
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H.S. 30 L.R. U.S.A., min. ear pes. 19/P. \& P. on above $1 / 6$ each.


MINIATURE UNISELECTOR SWITCH, two banks of ten plus home contacts, one bank continuous of normal. 30 ohms coil for 24 volt operation. Brand new, manulacturer's
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MINIATURE P.M. MOTOR $12 / 24$ volt, reversible, $1 \frac{1+i n}{4}$. dia. New. Price 9/6 each. P. \& P. 1/-.


TWELVE PLATE F.W. BRIDGE CON NECTED RECTIFIER mounted on 200/250 volt A.C. input transformer. Output $36 / 40$ volt D.C. at 1.2 amps. New, perfect. Price 16/6. P. \& P. 3/6.


200/250 v. A.C. MOTORS. New $1 / 80$ h.p., 2 drives, direct 6000 r.p.m., reduc-


No. 100 RM VARIABLE VOLTAGE TRANSFORM ERS, as illustrated above. Brand new in manulacturers original cases. inpur 230 volt A.C. output variable from 0 to 270 volt at 9 amperes. Price $\boldsymbol{E}$ i' each, plus carr. 12/6.

No. 200 CUH YARIAC ex unit but new and unused. Input 230 volt A.C., output variable from 0 to 270 volt at $2 \frac{1}{2}$ ampere. Price $\in 7 / 10 /-$, plus carr. $10 /$.
No. 200 CU VARIAC, Tandem model. Input 230 volt A.C., output variable from 0 to 270 volt at $4 \frac{1}{2}$ ampere. New. Price 69 each, plus carr. 10/-


NEW UNCHARGED UNFILLED 12 VOLT AC. CUMULATOR 9 ampere in unspillable plastic cases. Comprises $6 \times 2$ volt separate cells connected by terminal 2/9. Wooden carrying case for same with lid and strap price $3 / 6$.


12 v D.C. AMPLIFIER, as new, for operation on 12 v car battery, 10 watts undistorted output, with $6 L 6$ valves $62,100,250$ or 500 ohms. E12/100 each. Carr. 15/-
L.T. TRANSFORMER, real heavy duty job, extremely well made for continuous duty. New in original manufacturers" cases. Input 110 v.-260 v. multi-tapped 50 cycles, single phase. Output 28 -29-30-31 y at 21 amperes. Price 69/6. Carr. 9/-


NEW CARPENTER'S PENTER'S PRISEP PE LAYS. $2 \times$ 9500 LAYS. $2 \times$ 1,685 ohms.

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NEW MOVING EADSETS Hemplete with Tomplete with hand carbon hand microplug suitable plug suitable Price: 12,6 each plu

RCA AR88-LF, the "Rolls-Royce" of receivers. Covers $73-550 \mathrm{Kc} / \mathrm{s}$. and 1.48 $30.5 \mathrm{Mc} / \mathrm{s}$ in 6 bands. Controls: Variable Selectivity (crystal filter). $16 \mathrm{Kc} / \mathrm{s}-500 \mathrm{c} / \mathrm{s}$., Noise Limiter, Tone, BFO, etc. Has 2 RF Noise Limiter, 1 Ifiene: "Bro, etc. Has 2 in HT. For $110-230 \mathrm{v}$. A.C. mains. Supplied complete with 14 valves and handbook, but less case. in first class condition and perfect working order. Price 645.


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The smallest pocket radio with
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The product of a well known
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* 6-TRANSISTOR SUPERHET.
* Medium and long wave (Light programme)
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* Attractive cabinet with gold
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$\star$ OC72 push-Pull output.
$\star$ Not a do-it-yourself.


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# THE "TRANSISTOR-8" COMBINED CAR-RADIO/PORTABLE PUSH-PULL SUPERHET 

> This Portable 8 Transistor 8uperhet in tunable for both Medium and Long Waves and is comparable in performance to any equivalent Comunercial Transistor Set.
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star features

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$\star$ Medium and Long Waves.
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* $7 \times 4$ Elliptical HIgh Effictency Speaker. $\star$ Drilled Paxolin Chassls $81 \times 2$ inin. $\star$ Transistor Holders.
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GALL AND HEAR DEMONSTRATION MODEL

SIX TRANSISTOR POCKET SUPERHET

## STAR FEATURES

* Medium and Long Wave.
* 6 Selected Transistors. * Printed Circuit.
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* 30 ohms Speaker. * Instruction Booklet. \& Low consumption. $\star$ Attractive Plastic Cabinet. $\star$ (Red, Blue and White colours) t 9 v. P.P. 4 Battery. * Easy to Build.


## "THE MINOR"

The smallest transistor radio offered on the market. Case size only $3 \times 2 \times$ in. Variable tuning over medium waves. Home, Light and Third programmes without an aerial in areas of reasonable reception. Uses a three-stage reflex circuit of high efficiency, Total cost including Personal phone; transistor; long life miniature battery, eircuit and complete layout diagrams and all components: * Internal ferrite aerial. $\star$ Weight less than 2 ozs. All components sold separately. Circuit, layout diagrams and shopping list free.


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* No aerial or earth. * Variable tuning ove medium waveband.
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K Economical ( $1 \frac{1}{2}$ mA, consumption). Size $4 \frac{1}{2} \times 3 \times 1 \frac{3}{2} \mathrm{in}$. Total A May be assembled within an hour. weight less than 4 oz . \$ Complete layout diagrams.
We can supply all items including EDISWAN 7246
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COLLARO 4 -speed single player with crystal turnover head, E6/19/6, P.P. 3/6.
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 ACH OF THE ABOVE RECORD PLAYERS INCORPORATE THE LATEST FEATURES: \& LIGHT WEIGHT ASCTION. $A$ EIFE. $\star$ EASE OF FITING AND USE. NNG LIFE.
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RCA $6 \frac{1}{2}$-inch P.M. SPEAKER in Cabinet. With vol. control and 600 ohm Line Trans.

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27 / 6
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P.P.2/6.

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Tuning, 30 to $40 \mathrm{Mc} / \mathrm{s}$. Includes
TYPE 26: Variable zuning, 50 to
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## £7/12/6 ${ }_{\text {Post }}{ }^{3 / 6}$ 2/12/6 (8OTH MODELS)筑

 INET, $27 / 6$ EXTRA.SUPER REXINE PORTABLE CABINET (Illustrated) 37/6 EXTRA.

The product of a well-known radio manufacturer:
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Models available:
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Medium Wave, 180 to 550 metres.
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PYE $45 \mathrm{Mc} / \mathrm{s}$. STRIP TYPE 3583.

Complete with 12 valves. 10-EF50; EB34; EA50, with modification data. ABSOLUTE BARGAIN $39 / 6 \begin{gathered}\text { Carriage } \\ 5 / 6 .\end{gathered}$

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The ideal F.M. conversion unit as described in May, 1957. ComMay, 1957. com plete with 6 valves, EF92's and one EB91.I.F.T's etc., in absolutely new condition With circuit and conversion data
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COMPLETE set of instruments including calibrator and PYE Scalelamp galvo.

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Complete with the following valves:
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For No. 19 Set.
$10 \mathrm{Kc} / \mathrm{s} ; 100 \mathrm{Kc} / \mathrm{s}$.; $1 \mathrm{Mc} / \mathrm{s}$; ; spot frequencies; Crystal controlled oscillators; includes 52SC7 valves, neon modulator handbook, etc. BRAND NEW E4/19/6 $\begin{gathered}\text { Post } \\ \text { free }\end{gathered}$

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Complete with Valves, High Resistance Head. phones, Handmike and instruction Book and circuit. Frequency Range 41.0 to $61 \mathrm{Mc} / \mathrm{s}$. Range approximately 3 to 8 miles.
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A quality 4 -speed single -record unit complete with TPA12 tranacription arm and OC8 crystal pick-up. Size (apace reqd.) $174 \times 18 \downarrow \times 3$ iin.
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(22) TELETRON "OOMPANIOON "3-Transistor Printed Circuit
23) TELETRON "TRANNSIDYNE "Transistor Portable
(24) "DE-LUXE " Printed Circult Superher
25) JASON J.T.V.
26) RADIO JACK MULIARD TYPE : C : $i$ Tape pre-amp
(28) TAPE RECORDER (wlth Coltaro Mk. iv deck)
(29) JASON J. 3-3 Stereo pre-amp.

30 JASON PER Stereo pre-amp.......
32) New Jason F.M. Tuner with built-in power supplies \& cabinet
33) Now Jason Fringe Area F.M. Tuner as above


MULLARD 510 HIGH-FIDELITY AMPLIFIER
 Our printed circuit version
of this exellent ampliter,
with ULTRA-IINEAR with ULTRA-IINEAR
PUSH-PULL output stage Piving an exceptionally high quallity output of 10 watt (max.). Built-in Controls are provided for independ ent bass and treble Tone correction to suit all typee
of signal input. WIII match all erystal or high isapedance misgaetic pick up haads, P.M., A.M. or A.M./F.M. tuners or tape recorder output. All re
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AN ELECTRONIC ORGAN. Many readers will be Interested to hear that we are in the process of constructing the excellent Eectronic
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METERS. We carry Large ptocks of Meters from 50 microamps to 1.500 v. A few of the most popalar types are:- $\mathbf{1 0 0}$ microanaps 24 th
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 be plessed to quote for spectal meters to your own spacification.
RADIO JACK. The range of equlpment to butld yoursell. Covers local Medlum Wave variably tuned. Compart selfsell. Covers local Medfum Wave Btation to aerlal (no pormpact selif-contained onit requiring only connection in confunction with your tape recorver or high galn amplifer. Al necessary compo
plua $1 / 6$. \& $P$.


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(2) "RAMBLER" Mains Unit, £3/5/-; (7) T.S.L. F.M. Tuner, $£ 13 / 15 /-$; (4) F.M. Powor Pack
$52 / 6$; (15) R.C. $3 / 4$ watt Amplifier, $55 /-j$ (16) 2-amp. Battery charger,
BABY" Alarm, 89/6; (21) Mullard 510, ci2/12/-; Instruction Books which contain fulk des-
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A practical cabinet nicely designed, cloth covered twotone (brown and coffee). Size $15 x$ $17 \times 8 \mathrm{sin}$. deep. Takes B.S.R. 4speed Autochanger and $6 \frac{1}{2} \mathrm{in}$. round or elliprical speaker Carr. and Ins. 4/6.

## 69/6

'A beautifully styled cabinet. Made by a famous manufacturer. In polka dot cloth with clipped lid and carrying handle. Size $16 \times 14 \frac{1}{2} \times 8 \frac{1}{2}$ B.S.R. Monareh 4 speed Autochanger tical speaker, and mostof the modern portable amplifiers Carr. and Ins. 4/6.


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Polished wood 19/9
cabinet of attrac-
tive appearance. Fitted with 8in. P.M. Speaker W.B. or Goodmans of the
highest quality. Standard highest quality. Standard
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(2-5 ohms). Switch and flex included. Ins. carr. 3/6.
Ideal for stereophonic sound.
8 in. P.M. Speakers $8 / 9$. With O.P. transformer fitted $10 /$.
To suit most Record Players.
$6 \frac{1}{4} \mathrm{in}$. Round P.M. Speakers, $12 / 6$.
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STEREOPHONIC AMPLIFIER ET/19/6 Beautifully made for portable stereophonic record players. Latest design with printed circuit. Dimensions $3 \times 5 t \times 9$ lin. A.C. only. Mains isolated. Twin amplifiers each side ECL82 triode pentode valve. Full tone, volume and balance controls. Complete and ready to fit. Knobs $3 / 6$ per set extra. P.P. and ins. 4/6.

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Rexine covered. Size Takes 150 slides. Neep. Takes partides. Numbered partitions. Plated elip. Strong carryin
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MOTOR BOARDS $2 / 6$
For 4 -speed Autochangers. P. \& P. $1 / 3$. TELEVOX TELEPHONE ( AMPLIFIER 89/6


Invaluable in a noisy office or workshop. suction upe urin in suction type vibration microphone A.C./D.C.
Size of amplifier $7 \times 11 \times 3 \mathrm{in}$. Fits any type of size of amplifier $7 \times 11 \times 3 \mathrm{in}$. Fits ${ }^{2}$
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## 

## B.S.R. MONARCH 4-SPEED AUTOCHANGER <br> £6.19.6

Incoroorating aveo and manual control com. =plete with zurnover crrstal p.u. and sapphire str) lus. P.P. and Ins. $5 / 6$.
COLLARO 4 -SPEED AUTOCHANGERS
£7.19.6 $\begin{gathered}\text { Incorporating auto and } \\ \text { manual } \\ \text { mintrol }\end{gathered}$ plete with studio crystal p.u. and sapphire Bstyus. P.P. and Ins 5/6.


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- 12 MONTHS GUARANTEE
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- MARK D1.
- Brand new. Latest

- design with printed 59/6
- Mains isolated ${ }^{\text {cirens }} 7 \times 2 t \times 5$ in. A.C. onily. - porating EL84 as high gain outpur valve. - Volume and tone controls. Knobs $2 / 6$ exera. - $P$. and $P 3 / 6$.
- PORTABLE AMPLIFIER Printed circuit.
- MARK D. 2 79/6 Latest desisn.
- only. Mains isolated $3+2+\times$ Sin. A.C.
- Inly. Mains isolated 3-4 watts outpur.
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AMPLIFIER

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- manufacturer. Especi-
ally built for portable

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- B.S.R. FUL-FI CRYSTAL TURNOVER -
- CARTRIDGES $19 / 6$

Brand new. Including sapphire needles for -- L.P. and Standard, giving fullest range and -- finest tone obtainable for any player. Can be - fitted to all standard pick-up arms $P$. and $P$. - 9d.

79/6
Seylish cabinet by famous manufacturer. Cloth covered in contrasting colours (red and grey). Grilled front controls panel. Size $15 \times 19 \times 8_{\frac{3}{4} \mathrm{in} \text {. }}$ deep. Reautifully made-a cabinet of which you can be
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R.C. $121 / \mathrm{Mk}$. II. Automatic selection for any size record. Separate heads for monaural and stereo. Carr. and Ins. 5/6.

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STEREO AUTOCHANGERS 11 gMS. with zurnover cartridge for stereo L.P. and standard. Carr. and Ins. 5/6.

A "MUST" for


Less speaker fret. Size $133 \times 15 \times 8 \frac{1}{2}$ in. deep. Detachable lid with compartment for spare rape. Covered in 8 r
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STURDY CASE 12/6
 18 Tin . long playing records. P. and P. $2 / 6$

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RECORD PLAYER CABINETS
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## 0 <br> REGETTERED IMPROVED VACUUM T/V TUBES <br> 12 MONTHS GUARANTEE

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## 17" T.V. CHASSIS, TUBE AND SPEAKER. 16 GNS.

I7in. Rectangular Tube on modlfied chassis supplied as single channel chassis covering .B.C. Channels 1.5 or, incorporating Turre Tuner, which can be added as an extra, at our special price to chassis purchasers of $50 /$ giving choice of any two channels (B.B.C. and ITT.A.). Extra channels can be supplied at $7 / 6$ each. Chassis size $12 \times 14 \frac{1}{2} \times 11$ in. less valves. Similar chassis are used by well known companies because of their stability and eliability. With Tube and Speaker (less valves) 16 guineas. Complete and working with valves and Turrer Tuner, 24 guineas. 12 months' guarantee on the Tubes. 3 months' guarantee on the valves and chassis. Ins. carr. (incl. Tube), 25/-.


## 

## * TRANSFORMERS

Drop Through Type $12 / 9$.
$350-0-350$ volts at $250 \mathrm{~mA} ., 6.3$ volt at 4 amp . 6.3 volt at 4 amp ., 4 volt at 3 amp ., 22 volt at $.3 \mathrm{amp} ., 4$ volt centre tapped at 1.5 amp Primary $200-250$ volt. 50 cycles. P. and P. 3/9.
Drop through type 12/9.
$350-0-350$ volts at 250 mA . 6.3 volt at 5 amp . 4 volt at 4 amp. 4 volt at 7 amp. ${ }^{4}$ volt centre tapped at
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## * HEATER

$12 / 9$ volt at $\frac{1}{2}$ amp. $0-200-25$

## Upright type $3 / 9$.

$350-0-350$ volt at 80 mA ., 12 volt at 1.5 amp
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MAINS AUTO $12 / 6$
0-205-225-245 volts at 300 mA . Isolated windings of 6.3 volt at $2-6 \mathrm{amp}$. 6.3 volt at 3 Windings of 2 volt at $1-4 \mathrm{amp}$. P. and P. $3 / 9$

## TRANSFORMERS *

2-1 ratio or $1-2$ ratio auto trans-
former 2 volt at 1.4 amp. primary,
$2-1$
former 2 volt at 1.4 amp. primary, $3 / 9$
4 volt secondary. ${ }^{\text {P. and }}$ P. $1 / 9$.


POWER PACK AND AMPLIFIER $19 / 6$ R.F. E.H.T. Not tested. Amplifier stage 6 V6 with O.P. trans. 3 ohms matching. Smoothed H.T. 350 volt at 250 mA .6 .3 volt at 5 amp . 22 v . at $3 \mathrm{amp} ., 6.3 \mathrm{v}$. at 4 amp . and 4 v . centre tapped. Less valves. Drawings free. Size $14 \frac{1}{2} \times 8 \times 7 \mathrm{in}$. ns. Carr. 5/6.

## POWER PACK AND AMPLIFIER $9 / 9$

Output stage PEN45. O.P. trans. choke. Smoothed H.T. 325 volt at 250 mA .4 v . at 5 amp .6 .3 v . at 5 mp . 4 v . at 5 amp. centre tapped. a . standard plugs. Less valves. Ins. Carr. 5/6.

or car wiring. Revolutionary in design. Instantly ready for use and cannot burn. In light metal cas with full instructions for use. Post $2 / 9$
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Containing scanning coils, focus unit line transformer, etc. Less valves. Drawings FOCUS MAGNET $9 / 9$. Brand new 38 mm . 9 Brand new. 38 mm . Incorporating picture SCANNING COils SCANNING COILS 10/6.
Low impedance. 38 mm . Brand new. COLVERN PRESET POTENTIO. METERS $2 / 9$
Brand new. 200 ohms. 10 K . and 20 K . P. and P. 6 d . T.V. MÁSKS-WHITE, $9 / 9$
(I7in.) GREY, $10 / 9$.
Brand new. Good quality plastic. Post T. $2 / \dot{\mathrm{V}}$. MASKS—12im., $1 / 9$. Soiled-needs washing. Flat face. Used. Post $1 / 6$.
Finest quality. 75 ft . $\times \frac{1}{2}$ in. in sealed metal Finest quality. 75 ft . $\times$ tin. in sealed metal
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 switched for 1 or 2 kW . Illuminated grille Size $26 \times 18 \times 7$ tin. deep. Ins, carr $10 / 6$

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 CATALOGUE14in. T.V. CHASSIS, TUBE AND SPEAKER, 11 guineas
As above, with 14 in . Rectangular Tube. I2 months' guarantee on Tube; 3 months' guaran tee on chassis and valves. Chassis with Tube and Speaker (Less valves) II guineas. Complete and working with valves and Turres Tune

## $\rightarrow$ SUPER CHASSIS 99/6

5 valve superhet chas sis including Bin. P.M. speaker and valves.
(tone, volume, tuning w/change switch). Four w/bands with position for gram. p.u. and extension speaker. A.C. Ins. Carr. 5/6.

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6 or 8 valve. Latest type, midget valve design for A.M. or F.M. Brand new. Cadnium. Size $12 \frac{1}{6} \times 7 \frac{1}{4} \times 2 \frac{7}{6} \mathrm{in}$. P. and P. $1 / 9$.
12in. T.V. CHASSIS 29/6
Complete chassis by famous manufacturer R F. E.H.T. unlt included. Easily fited to table or console model owing to this chassis being in three separate units-(power; s/vision strip; t/base interconnected). This chassis is less valves and tube. Speaker FREE. I.F's 16-19.5 $\mathrm{Mc} / \mathrm{s}$ vision. Channels $1-5$ easily converted to I.T.A. by use of a Turret Tuner. Drawings available at 2/6 or FREE with order. Ins. Carr. 10/6. 12 in . Tube available at $£ 6$. plus Ins. Carr. 15/6.

- popular 12in. PLESSEY T.V.


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This is a real bargain for anyone wanting to make up their own T.V. at a very low cost. I.F's $10.5-14 \mathrm{Mc} / \mathrm{s}$. Simply adapted for a 12 Channel Turret Tuner and can be modified to take a larger Tube. A chassis in one unit. Untested. Less valves, tube, speaker and scanning coils. (All can be supplied as extras.) Circuit diagram available at $3 / 6$ or FREE with order. Carr. and Ins. 10/6.

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For Cathode Ray Tnbes having Heater/Cathode short cireuit and for C.R. Tabes with lalling emission. and $50 \%$. Tapped mails primaries.


100 ohms to 10 meg. Ditco $5 \%, 9 a ., 10 \%, 6 \mathrm{a}$
$\left.\begin{array}{l}5 \text { watt } \\ 10 \\ \text { watt }\end{array}\right\}$ WIRE-WOUND RESISTORS
, 50,000 bmas- 10,000 ohms WLRE WOUND POTS. 3 WATT LAB. COLVERN, ETC.
 All values 25 ohme to 25 K . $3 L^{\circ}$ ea., 30 K. 60 K , $4 /$. Ditto Carbon
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$0 / P$ TRANSFORMERS. Heavy Daty $50 \mathrm{~mA} .4 / 6 / \mathrm{malti}$ Push-pall 10 watts, $15 / 6$. MULLARD 6510,46 . $4 / 61$ L.F. CKOKES 15/1018 60/65 ma., 5/-. 10 B 85 ma . $10 / 6$. 10H 150 ma . $14 / \mathrm{t}, 5 \mathrm{H} 850 \mathrm{ma} .16 / 6$.

MAINS TRANSFORMERS $200 / 250$ v, A.C.
STANDARD $250-0-250,80 \mathrm{~mA}$., 8.3 \%. 3.5 a .

MIDGET, 220 ₹. $45 \mathrm{~mA}_{\text {, }}, 6.3 \mathrm{v}$.2 a .
SMALL, $220-0-220,50 \mathrm{~mA}, 6.3$ ซ. 3 a
HEATER TRANS. $6.3 \mathrm{v} .1+\mathrm{A} .7 / 6$. 3 amp.

ALADDIN FORMERS and cores, in. 8d. ifin, 10 d, nd in. $x$.
SLOW MOTION DRIVES. Epicyclic ratio $8: 1_{1} 2 / 3$. TYANA. MIdget Soldering Iron. 200;2LO F. or $230 / 200$. $17 / \beta_{\text {, }}$
 LINE CORD. 3 amp., 80 ohms, per $1000,4 \mathrm{amp}$.
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CRYSTAL MIKE INSERT by Acos 6/6
Precision engineered. Size only $\frac{7}{8} \times \frac{3}{6}$ in. Bargain.
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 12in, Baker 15 wt. 3 ohm and 15 ohm models $105 / \mathrm{m}$
18 in , 15 ohm . Plessey 10 wt . with Twester, $87 / 6$
1.F. TRANSFORMERS $7 / 6$ pair

485 kofs , slug tunlag miniature can $21 \times 1 \times 1 \mathrm{in}$. High Wearite M800 L.F. Miniature $465 \mathrm{kc} / \mathrm{s}, 12 / 6$ pair,

CRYSTAL DIODE G.E.C. $2 /$-. GEX34, 4/-0 40 Cireaits, $3 /-$ A.R. HEADPLONES. 4,000 ohms, brand new $16 / 6 \mathrm{pai}$ WITCH CLEA CONDENSERS 365 . $4 / 3$ Mia. $1 / 4$ $\times 1 \frac{1}{2} \mathrm{in}_{0} \times 1 \frac{1}{5} \mathrm{in} ., 10 /$ - $\quad .0005$ Standard with trimmers $9 /-$ i less trimmers, $8 /-$ Midgat, 7/6; 8ingle 50 p.1., 2/8
 B12A, CRT, $1 / 3$. Eng. and Amar. $4,5,6,7$ and 9 pin, $1 /$ d. ByG with can, 1/6: B12A. $1 / 8$. B9A with oan, $2 / 6$ CERAMIC, EF50, B7G, B9A, OCt., 1/n, B7G with can, 1/9

 WAVECHANGE SWITCHES
D. 2way, 3 p. 2-way, short spindle

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GOLTOP TRANSISTORS
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Complete Dato Supplied
AUDIO V.10/15s suitable R.F. V.6/R2 sultable for for high gain and low ire- osclllators, frequency
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1959 RADIOGRAM CHASSIS FINEST VALUE


THREE WAVEBANDS
S.W. $16 \mathrm{~m} .-50 \mathrm{~m}$.
M.W. $200 \mathrm{~m} .-550 \mathrm{~m}$. ECE 42, EF41, EBC41 12 month Guaranteo. A.C. $200 / 250$ Enti, EZ 4 -way awiteh Short-Medimmont-Gram. A.Y.C. and Negative Feedrack, 4.2 watts. Chassis $18 \%$ in, $\times 5$ hin $\times 2 \nmid i n$ 2 Pilot Lamps, Four Knobs, Walunt or Ivory, aligne and calibratod, Chassis isolated from mains.
BRAND NEW £9.10.0 carr. a/a.
TERMS: Deposit $£ 5 / 5 /$ and 5 monthly payments of $£ 1$.
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CHANGERS RC121/D MKII MODELS Brand now and fally guaranteed 12 months.

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Designed to play 18, 33, 45, 78 r.p.m. Records 7in., 10 mm . 2in. With plug-m NORMAL
OUR PRICE $\frac{1}{1} 0.150$ each. Post Free. Optional Plag-in STEREO HEAD 22 extra.

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4-SPEEDS-10 RECORDS
With Studio "O" pick-Up
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OUR PRICE $£ 6-19-6$ post free STEREO MODEL JA12, £11/17/6.

AUTOCHANGER ACCESSORIES Suitable player cabinets (uncut boards) ... 49/6 Amplifier player cabinets with cur boards $63 /-$ 2 valve amplifier and $6 \frac{1}{2} \mathrm{in}$. speaker for above $72 / 6$ 3 valve amplifier and $6 \frac{1}{\frac{1}{2}} \mathrm{in}$. speaker for above 95/Wired and zested ready for use.

## EXCEPTIONAL OPPORTUNITY

COLLARO 4-speed MODEL $4 / 564$ Single Player, heavyweight turntable and lightweight Player, heavyweight turntable and lightweight
Studio O Piek-up with turnover Xtal mounted on baseplate. Autostop fitted. $6 \mathrm{~g} \mathrm{~g}_{\text {I }}$ Carr.4/-

Amplifier Player Cabinets, 45

## * GARRARD 4-SPEED,SINGLE

 £8 AUDIO PERFECTION POST De luxe Cabinet with amplifier and $6 \frac{1}{2}$ in. speaker $\mathbf{\epsilon} 6 / 15 /=$ or complete kit $£ 14 / 10 /=$.Teletron Transistor Pocket Radios Designers Speeffied Kits
COMPANION PRINTED CIRCUIT 3 Local station receiver kie, $4 \frac{1}{2} \times 3 \times 1 \frac{1}{2}$ in. 84.19 .6 with 3 genuine Pye Goltop transistors; plans 6d.

Transidye Superhet Six $6 \times 4 \times 1$ Izin.
T.C.C. Printed Circuit, internal Ferrite aerial cabiner and 6 genuine Pye Goltop cransistors.
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No surplus reject transistors supplied

THE HI-GAIN BAND 3 PRE-AMP. Cascode circuit using Valve ECCE4. I7db gain. Kit $29 / 6$ less power; or 49/6 with Alser pack kit Plans Also Band I version same Prices.

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 Long $\begin{gathered}\text { Midget sizz } \\ \text { spindles. Guaran- }\end{gathered}$ 5 K. ohms to 2 Meg.
No Switch D.P.SW.
Linear or Log Tracks
Linear or Log Tracks Semi-air spaced Polythene core. Ideal Band III $9_{b y}^{d}$
Losses cut $50 \%$, FRINGE QUALITY ATRSPACED .... 1/6 yd. $\begin{array}{lllll}\text { COAXIAL PLUGS } & 1 /- & \text { DOUBLE } & \text { SOCKET } & 1 / 3 \\ \text { SOCKETS } & 1 / \mathrm{OUTLET} & \text { BOXES } & 4 / 6\end{array}$ BALANCED TWIN FEEDER pER yd. 6d $80 \Omega$ or $300 \Omega$. TWIN SCREENED BALANCED FEEDER $1 / 6 \mathrm{yd}, 80 \mathrm{ohm}$. $1 / 3$. $250 \mathrm{pf.} 1 /$,6 . $600 \mathrm{pf}, 7,750 \mathrm{pt}$., $1 / 9$. Phillips, $1 / \mathrm{eca}$. ALUMINIUM CHASSIS. $18 \mathrm{~s}, \mathrm{w}-\mathrm{g}$. Platn, undrilled with 4 sides, riveted corners and lastice eixing holes,
with 2 in. sides. $7 \times 4$ in. $4 / 6 ; 9 \times 7$ in $5 / 9 ; 11 \times 7$ in. with 21 in sides, $7 \times 4 \mathrm{in}, 416 ; 9 \times 7 \mathrm{in}, 5 / 9 ; 11 \times 7 \mathrm{in}$,
$6 / 9 ; 13 \times$ oin., $8 / 6 ; 14 \times 11 \mathrm{in} ., 10 / 6 ; 15 \times 14 \mathrm{in} ., 12 / 6 ;$

BLAOK CRACKLE PAINT. Air drying, $3 / \mathrm{m} \mathrm{im}$. P.V.C. CONN. WTRE, 8 colours, single or stranded, 2d. Fd, CORED SOLDER RADIOGRADE $3 /$ yds, $9 \mathrm{~d}, \frac{1}{2} \mathrm{lb}, 2 / 6$.
PAXOLIN SHEET, 8in $\times 10 \mathrm{in}, * 1 / 6$. ION TRAPS $5 /-$

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$1,200 \mathrm{ft}$. Paper tape on 7in. metal reels
Spare Reels $5 \frac{3}{4} i n$. plastic, $3 /-; 7 \ln$. plastic, $4 /$
SUPERIOR 1,200tt. Plastic Tape 21/-
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"INSTANT" Bulk Tape Eraser, 200/250v. A.C. For any make and size of tape.
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 6/-; RM3, 120 mA ., $8 /-\mathrm{F}$ RM4, 250 N. $275 \mathrm{~mA}, 16 /=$
MINIATURE CONTACT COOLED RECTIFIERS. $250=50 \mathrm{~mA}$ 2/6. 60 mA 8/6: $85 \mathrm{~mA} 9 / 6$
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With Jason Calibrated dial and 4 valves, c6/15/,
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CLOSE TOLERANCE ( $\pm 1 \mathrm{pt}$.) 1.5 pf, to 47 pt ., 1/6. DITTO
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COMPLETE KITS OF PARTS FOR THE "HI-FI" ENTHUSIAST

Designed by MULLARD-Presented by U8 strictly to their specification

THE VERY POPULAR MULLARD" 5-10 MAIN AMPLIFIER

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Based entirely on the present very popular " $3-3$ " model and designed to operate in conjunction with the new 2-stage PRE-AMPIIFIER (shown bere) thus providing alt the faciltiles associated with the more expensive "Hi-Fi" equipment. We
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IT INCORPORATES:

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- HIGH QUALITY 7in. $\times 4 \mathrm{in}$. P.M. Speaker.

1,200ft. reel EMI tape.
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Incorporating 3-SPEEDTREBLE EQUALISATION by means of t'ie lotest FERROX CUBE POT CORE INDUCTOR. PRICE for COMPLETE $\mathrm{E} 12 / 15 /-$
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A very high quality amplifier based on the very successful Type " $A$ ", design completed in the MULLARD LABORATORIES. ONLY NEW HIGH-GRADE COMPONENTS are incorporated including MULLARD VALVES and a GILSON OUTPUT TRANSFORMER . . . other features are: Magic Eye Recording Hand Indicator-Effective Tone Control-Monitoring and Extension Speaker Sockets-has own Power Supply and can be used as independent Amplifier for direct reproduction of Gram Records or from Radio Tuner. Overall size $11 \times 6 \times 6 \mathrm{in}$.-Truvox-Collaro-Lane-Brenell or Morek Decks-please specify which.
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 CRYETAL ${ }^{\text {TAPE }}$ MLKE (E1)15/-). ROLA A 10n. ( $21 / 10 / 0$ ) ALL FOR
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Assembled and Teated
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 Two are perrectig sultable to operate with ONE main Ampllier , and the second main Amplifer FULLy descriptive leaflets are available. enclose an ent the stereo,
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The latest COLLARO "CONTINENTAL" Autochanger, Studto "C " Piok-up The COLLARO 4 -speed Single Record THE NEW B.S.R. model UA12 Is in Stock MIXER AUTOCHANGER.
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The latest GARRABD TRANBCRIPTION Dator " 301 " with Stroboscopically murked 828.0 .11 turntable The new GARRARD Model 4HF High Quality SIngle Record Player fitted with the latest T.P.A. 12 plek-up arm and
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 COMBINED AM/FM TUNING UNIT INCORPORATING OWN POWER SUPPLY is in Stock. Epecifcally designed for operation with High on Long, Medium and short Wavebands plus the complete VHF/FM Transmissions. Thoroughly reoommended where a complete self-powered All Wave Tuner is wanted. PRICE 224. 19.0 (Plus 7/6 carr. \& ins.) GURE PURCRASE: Deposit 25 and 12 monthu at $£ 1 / 16 / 7$.A fully Illustrated leafiet is avallable, please enciose S.A.E.

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Manufactured for the Admiralty in 1952 by Burndept, this utilises 4 valves, 1 each 5 Z4G, 6V6G, 677G, 6J5G. and high quality components such as "C" Core Transformers and Block Paper Smoothing Condensers. Has A.C. Mains Pack for nominal $110 / 230$ volts. Provision for 600 ohms or High Impedance Input, and has Output to 600 ohm Line. For normal use only requires changing Output Transformer. Can be used for Speech or Music, giving High Quality Reproduction. Output approximately 4 watts. Enclosed in metal case, and designed for Standard 19in. Rack Mounting, having grey front panel size 19in. x 7in. with Chromium Handles. All connections to rear panel, front having "On/Off" Switch, Gain Control, Indicator Light, Fuses and Valves Inspection Panel. BRAND NEW IN MAKERS' PACKING. ONLY $£ 4 / 9 / 6$ (carriage 10/6).

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A.C. MAINS POWER PAOK OUTPUT STAGE, in black metal case to match recelver, enabling it to be operated immediately, by just plugging in withont any modification. Fitted with Sin. P.M. Speaker \&6/10/. DEDUCT 101. IF PURCEASING RECEIVER AND POWER PACK TOGETHER.
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OSCILLOSCOPE No. II by Cossor A First Grade L.F. Oscilloscope incorporating a Hard Valve Time Base with speeds of $1-5-40$ milliseconds, but easily converted for a few shillings to produce $3 \mathrm{c} . \mathrm{p} . \mathrm{s}$. to $30 \mathrm{kc} / \mathrm{s}$. Has High Class Amplifier with Fine and Coarse Gain controls, Brightness and Focus controls, $X$ and $Y$ shifts. A.C. mains pack for 115 v. 230 v. nominal, fully fuse protected. Employs $2 \frac{3}{4}$ in. Tube ACR 10. Front panel I9in. x 7in., for rack mounting, depth 12 in ., or can be used in Steel Transit Case on bench. Complete with suggested Modification data, BRAND NEW AND UNUSED. ONLY E $12 / 10 /$ - (carriage $15 / \%$ ). Rll55 SUPER SLOW MOTION TUNING ASSEMBLY. As used on all late models 1155 s . Easily fitted to "A" sets etc. ONLY $12 / 6$.
ROLA $6 \frac{1}{2}$ in. P.M. SPEAKER. Mounted in grey crackled metal cabinet $9 \times 9 \times 4 \frac{2}{4}$ in., and with volume control. BRAND NEW AND UNUSED. ONLY $27 / 6$.
MAINS ISOLATING TRANS. FORMER. Manufactured by Vortexion. Fully shrouded. Will provide true $1: 1$ Ratio from nominal 230 v. Primary. Rated at 100 watts. BRAND NEW. ONLY $22 / 6$ (post $2 / 6$ ). 6 v. VIBRATOR PACKS. Output vprox. 130 v . at 30 mA . fully filtered and smoothed. Complete ONLY $12 / 6$.

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[^13]
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AO50 AUDIO OSCILLATOR covers 20 cps to $200,000 \mathrm{cps}$ in four ranges. Output variable up to 10 volts.
Details of above sent by return on receipt of stamped addressed envelope. HIRE PURCHASE available. TRADE supplied direct.

## GRAYSHAW INSTRUMENTS

126 Sandgate High Street, Folkestone, Kent Phone: folkestone 78618

## MIDLAND INSTRUMENT CO.

HUGEES MOTORS, ahant wound 12 v . 1 -amp. speed 5,000 r.p.m.., reversing, size 3 in. long. 11 in . dia., tin. shaft, weight 20 oz ., a very superior motor designed for anti-radar equipment.
new. unused, $10 /-$, post $1 / 6$. Ditto fitted renew. unused, $10 /$, post
duction zear, giving a final drive ( f in. ahanft) duction gear, giving a final drive (tin. anatuied
either 320 or 160 r.p.lu., state which required. $\begin{array}{ll}\text { either } & 320 \text { or } \\ 12 / 6 \text {, poat } 1 / 9 .\end{array}$
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BARR \& STROUD RANGEFINDERS. 1 -metre base, coincidence type, a hand-bel, 1 instrument that gives range in yards or any distant object from 500 to 20,000 ysurds
( 12 miles). The variable focus is $\times$ rigbt eyepiece provides two innages of the object ( 12 milies). The variable focus $14 \times$ right eyepiece provides two innges of the object
viewedt one from the night objective the other trom the ieft. When hese two images
 eyepiece. Fitted two fiters and other reânements. A very high-quality instrument, eriginal cost $£ 180$, our price new or near new condition, supplied In stout albre con N.I. $201-$

CEASSIS. U.S. manufacture all aluminium, $12 \pm \times 8 \times 5$ fin. complete with top cover
gome items have been removed, remalning are: -25 Amphenol midzet ceramic B7G some items have been removed, remaining are:- 25 Amphenol midget ceramic B7G
type v /holders, complete with cans, over 70 colour coded $5 \%$ resistors, alao many type vhoiders. complete what, trimmers., padders, fixed and variable inductances, trans. former, v/engtrol, etc., new, unused, bargain, $10 /-$, post $3 / 6$.
OPTICAL UNITS, consiats of a brass mount holding two high-grade $40-\mathrm{mm}$, dia. aehromats, each 31n, focal length, forming a Petzral aystem lenses easily removed by unscrowing retaining ring, new and periect.
are 3 in. focal length, $12 / 6$, port $1 / 6$.
R.C.A. ROTARY RELAYS, $12 \cdot v$. D.C., powerfuil 18 -deg. movement, actuating 2 heary and 1 light duty 3 -pole changeover contacta, slao fitted contact mutomatically openlng coil resistance from 8 to 118 -ohms holding, size 3in. $\times 3 i n . \times 2 H 1 \mathrm{n}$. , welght 13 oz., new boxed, well worth $50 /$. Our price 7/6, poat 1/6.
JOEN OSTER MOTORS. 12 v. 1.4 amp. , inflal drive 5.600 r.p.m., , atted enclosed gearbox with two kin dia, shaft drives, one 8 and the other 24 r.p.m., also a hinear Tociprocating shaft in guides, with win. movement opera ing at 24 unoesa minute, easily YACUUM GADGES, panel mounting 24 in. square front dial graduated in units up to $B$ FACUUM GAVGES, panel mourdive boxed $5 /-$, post $1 / 6$.
CEARCING SETS, ughtweight, 46 lbs. easily carried, 4 -stroke aircooled, runsi or 18 hours from 1 gall. petrod, D.C. output 12 to 18 voits at 80 watts, complete with exhaust and silencer, canvas cover, completely reconditioned by makers and now as new,
supplied in stout canee, our price $£ 8 / 10 /$,- carriage $100 \mathrm{~m} .12 / 6,200 \mathrm{~m} .16 / 6,300 \mathrm{~m}$. $20 /$ gupplied in sto
(inland only). (thland only).
TELEPPHONE
TELEPPHONE SETS, consist of two comblned milcrophones and recelvers, which when wired ny by ordloary twin flex, provide perfect 2 -way communication, excollent results new unused, 7/6, post $1 / 3$. snitable twin 14/3 sapplied, postage each 2oft. flex 3d. extra.
SELENIOM nECTIFIERS. These sre latest brand new G.E.C. aupply, not ex-Govt. or assembled from bits, full-wave bridge, 12-v. 11 amp. cont., 2 smp. Int., 10/a, poat $1 /$.
 $230 / 250$ v., with $5-11-17$ v. output to charge a 2,6 or $12 v$. batiery, brand new, 1$\}$ an

Many other Bargains ; send stamped addressed envelope for lists.
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RCP MULTMMETERS. A.O. and D.C. volts, to $2.6_{1}, 100 \mathrm{~mA}, 0-1$ Ampe.; Resistance, 0 to 500 , $100 \mathrm{~K}, 1 \mathrm{Meg} ;$; DeciBels, -10 to +69 . 1,000 ohms per volt ( 400 microamps basic). In light oak case leade and prods, Internal battery, and instruction manual. All BEAND NEW and tested. $25 / 19 / 6$.

SELENIUM BRIDGE RECTIPIERS. Funnel cooled. A.C. input 45 . RMg. D.C. ontput
$90 \nabla$. 10 amps. BRAND NEW. Boxed. $45 /=$. Poot $3 / 6$.

DUAL PURPOSE TRANSFORMERS (Gresham). Prt 230/250 v. Secc. 240-0-240 v. 1.5 amps., $5 \nabla^{\nabla}$. 12.5 amps. fr.


ADMIRALTY ET TRANSFORMERS. PTH 230 ₹. $50 \mathrm{c} / \mathrm{s}$. Secs. $620-550-375-0-375-550-620$ ₹. ( 620 and 650 v. $200 \mathrm{~m} / \mathrm{ampse}, 375 \mathrm{v} .250 \mathrm{~mm}$ amps.), plus two 5 V. 3 Amp, rectifler Windings. Total

Carr. 5/:.
TRANSFORMER BARGAIN. Tnput 0 -200/250 tapped. Outputs $250.0-250 \mathrm{~V} .80 \mathrm{~m} / \mathrm{amps} .5 \mathrm{v}$. 2 amps: 6.3 v. 4.5 amps. Upright mote. BRAND NEW. Boxed. Ex-Admiralty mado 1952. A Ane $50 \mathrm{c} / \mathrm{s}$. mains tranny for ONLT 16/6, post FREE.

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| :--- |
| Lavest type potted, oil filled, Pri. 230 v. 50 c/s. |
| Sec. $0.70-76-80$ v. 4 amps. gize $5 \frac{5}{2} \times 4 \frac{1}{2} \times 6$ inn. |
| high. Wt. 19 lb. BEAND NEW. $42 / 6, ~ c a r r, ~$ |

AR88D MAINS TRATSFORMERS. Input 110-240 v.
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MODULATION TRANSFORMERS. COlling type 20 watts 807 to $807,8 / 6$ each. Post $1 / 6$. FERRANTII TYPE, tor Tx 36 etc., puch-pull 807's | FERRAN |
| :--- |
| to plate and screen modulato push-push $807 \mathrm{~s}_{4}$ |
| Wh. 61 lb . $17 / 6$. | ratio 2:1.

Post 2/6.

## RESISTORS

Margan "T" (1 watt) and " R"' (1 watt). Lateat types, al
Post $1 /-$

RCA TE-149 HETERODYNE WATEMETERS, Employ V-cut $1 \mathrm{Mc} / \mathrm{s}$. crystal ( $0.005 \%$ ). Overal accuracy better than $0.02 \%$. $105 \mathrm{Mc} / \mathrm{\beta}$. Useful calibrated every to $20 \mathrm{Mc} / \mathrm{s}$. Provision for ftting internal dry batteriea. As new, In original transit case. f9/10/6. Carr. 5/6.


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## RANGE TYPE SIZE PRICE

 25 Microamp. D.C. M/C atln. Flust Circ. Scale "Rontgens" 25 Microsmp. D.C. M/C 2\$l. Prol. Circ. Bcale " Rontgens ${ }^{\text {² }}$ 100 Microamp. D.C. M/C 3 1 ld . Flush Circ. Scale $0-50 / 0-1000 \mathrm{v}$. 100 Mlcroarap, D.C. M/C 2 In. Flush Circular. 1 Milliamg. D. M. M/C 3/in. FIush Circular
1 Miliamp. D.C. M/C 21n Flush Square. Fe/NFe. 1954 1 Miliamp. D.C. MC 6in. Flush Clre. \& 200 Milliamp . D.C. M/C 21 ln . Flush Creular
1 Amp. Thermoconple 2 in. Frin. Flush Circular
4 Amp. Thermocouple $2 / \mathrm{n}$. Flush Square
300 Voits A.C. M/I 6in. Flush Clircular. Made 1955 300 Volts A.C. M/I 2 2in. Flush Circular 500 Volts $\quad$ A.C. M1I $2 / \mathrm{in}$. Flush Circular
10 Amperes METAL RECTIFIERS. Full wave bridge.
$1 \mathrm{~mA} ., 8 / 6$. $\quad \mathrm{mA} ., 8 / 6 . \quad 8 T C 2 \mathrm{~mA}, 5 / 6$.

CQSSOR 343 GANGMG OSCILLATORS. A.M./F.M. signal generator of $70 \mathrm{Kc} / \mathrm{B}$. to $21 \mathrm{Mc} / \mathrm{s}$. im 3 ranges. A.M. modulated 400 c/s., F.M. sweep cloan condition, but may have mNNOR faults. floan condig/6. Cart. 7/6.

HICKOCK 1-177 VALVE TESTERS. Checks EICKOCK 1-177 dynamic mutual condnctance, shorts, emiseion, gas, and nolse. For UX4, UX5, UX8, UX7, Octal, Loctal B7G, and Acorn types. Portable in wooden carrying case $161 \times 8 \times 5$ in. Wt.
1921b. BRAND NEW. Complete with Instru1921b. BRAND NEW. Complete with instru-
tion book and valve testing charts. For $117 \mathrm{\nabla}$. tion book and valve
A.C. 10 gns. Carr. $7 / 0$. Matching auto. tranafurmer tor 230 v. A.C., $12 / 6$.

MULLARD C. \& R. BRIDGES. 0.1 ohm to 10 Megohms in 4 ranges; 10 prd. to 10 minges: Callbrate. Open Bridge, and $\%$ range 3 ranges: Calibrate. Malnen Brige, and guaranteed.
For $100-550 \mathrm{~F}$ A.C. maled For 100-5 $\mathrm{F} / 10 \mathrm{~F}$. Pot $3 / 6$.

CRYSTAG CALIBRATORS. Give 1 Mc/s. CRYSTAE CALIBRATORS.
$100 \mathrm{Kc} / \mathrm{e}$, and $10 \mathrm{Kc} / \mathrm{s}$. "pips" with or without
Employ
dual
$1,000 / 100 \mathrm{Kc} / \mathrm{s}$. ruodulation. Employ dual $1,000 / 100 \mathrm{KC/s}$. crystal ralvos. Made by Canadian Marconl Co.
 Operate from 12 v. 0.45 ampa. L.T. and 250 v.
D.C. H.T. In neat metal casc $2 \times 44 \times 94 \mathrm{in}$.
 deep. BR connectors, ete. $84 / 19 / 6$.

R1155 RECEIVFRRS. With latest type super slow mothon drive. In good condition and perfect working order, re.aligned and air tested. Model (covers trawler and shipping bands) \&12/1916. Carr, (either) 10/6. Send 8.A.E. for detalls of sets and power units, or $1 / 3$ for Illustrated booklet.

INVICTA LOUDSPEAKERS. Good quality 101n. unit (Impodance 3 ohmas). In wooden cabinet $17 \times 17 \times 6 \mathrm{in}$ Complete with 50it. lead and
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MOVING COL PHONES. Finest quallty Canadian, with chamois ear-muffe and leather-covered headband With lead and jack plug. Noise excluding
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TX-36 MODULATOR/POWER UNIT. Supplies HT 500 ₹. $200 \mathrm{~m} / \mathrm{Amps}$ twice, I.T. 6.6 v. 8 Ampw and blas. Also modulator \$ 6CJG7, 2807 , and 3 FW $4 / 600$ Recta In oak case $24 \times 161 \times 1501$
Wht. 120 lb . Circult supplled. TREMENDOUS BARGAIN. EA/19/6, cart. $15 / 6$ Eng. \& Wales
D.C./A.C. ROTARY CONVERTERS Input 12 v D.C. Output 230 จ. $50 \mathrm{e} / \mathrm{s}$. A.C. at 135 w'atts. Fitted with $0-300$ v. A.C. 2 in. meter and slider resistor for voltage adjusiment,. In stout wooden carrying case with lid. Perlect working order.
c9/19/6. Carr. 10/6.

## FERRANTI

 VOLTMETERS N5. 0-300 volts, 25. $100 \mathrm{c} / \mathrm{s}$. Moving iron 6 ln scale. Fl. mtg. Hermetically sealed, grade IN. Made 1955. BRAND NEW. B o $\quad$.$79 / 6$, post $3 / 6$.


## AUTO-CHANGER <br> COLLARO RC54 3-8PEED MIXER Fitted studfo pick-up with turnover head. Brand new, cartoned, but for 110 v. 50 c.p.s. A.C. mains. So that the unlt can be operate 1 from normal mans. So that madns we are supplying free with every changer a suitable  W.B. "STENTORIAN" HIGH FIDELITY P.M, TF1012 SPEAKERS <br> really good quality speaker at a low spesch coil. Where gighy recommend this unit with an ambazing performance \$4/10/9. Please state whether 3 ohm or 15 ohm required.



AM/FM RADIOGRAM CHASSIS, HIGH QUALITY, PUSH-PULL. 6-8 WATTS OUTPUT. Current manufacture. 12 months ${ }^{\text {O }}$ guarantee. For 200-250 v. mains. Covers
L. and M. Wavebands plus F.M. L. and M. Wavebands plus F.M. Includes 8 latest type miniature B.V.A. valves. Only 22 gns. plus $7 / 6$ carr. Or deposit $£ 2 / 12 /-$ and 9 monthly payments of $£ 2 / 12 /=$


BRAND NEW LUCAS MOTOR CYCLE BATTER TES. 6 v. 22 A.H (in sealed cartons). Limited number available at only $29 / 9$ each. Carr. 3/6. Normal price 71/- each.

ELECTROLYTICS (current production) Not Ex Govt
Tubular Types
$8 \mathrm{mfd} 450 \mathrm{v} \quad 1 \mathrm{~g} \quad 16 \mathrm{~F} 450$ Types
$8 \mathrm{mfd} .500 \mathrm{v} \quad 2 / 6 \quad 32 \mu \mathrm{~F} 350 \mathrm{v}$
$16 \mu$ F 350 v
$16 \mu \mathrm{~F} 350 \mathrm{v}$. $\quad 1 / 11$
$16 \mu \mathrm{~F} 450 \mathrm{v} . \quad 2 / 9$
$16 \mu \mathrm{~F} 500$ v. $\quad 3 / 9$
$25 \mu \mathrm{~F} 25 \mathrm{v}$. $\quad 1 / 3$
$50 \mu \mathrm{~F} 12 \mathrm{v}$, $1 / 3$
50 mfd .25 v. $\quad 1 / 9$
$50 \mu \mathrm{~F} 50 \mathrm{v}$. $\quad 1 / 9$
$100 \mathrm{mfd} .12 \mathrm{v} . \quad 1 / 9$
100 mfd. $25 \mathrm{v}, \quad 2 / 3$
$3,000 \mathrm{fmd} .6 \mathrm{v}$. $3 / 9$ 32 mfd 450 100 mfd .450 v . $\begin{array}{ll}100 \mathrm{mfd} .450 \text { v. } & 4 / 9\end{array}$ $8-8 \mu \mathrm{~F} 450$ v. $16-16 \mu \mathrm{~F} 450$ v. $\quad 3 / 11$ $16-32 \mathrm{mfd} .350$ v. $4 / 6$ $16-32$ mfd. 350 v. $4 / 6$ $32-32 \mu \mathrm{~F} 350 \mathrm{v}$ $100-100 \mathrm{mfd} .350$ $6,000 \mathrm{mfd} .6 \mathrm{v} . \quad 3 / 11$

150 mfd .450 v. $5 / 9$

desyan or

 T.R.F. receiver thaer. For inclusion in cabinet iflustrated or walnut
vencered type. It vencered type. It
employs
valves $6 \mathrm{K7}, \mathrm{SP61}, \mathrm{BF6G}$, denigned for simplicity in wiring. Sonsitivity and quatity outu up to standard. Foint-to-point wiring diagram, Dor a maximum parts list, 1/9. Thls receiver can be built on brown or cream bakelite including cabinet. Available

## SELENIUM RECTIFIERS



## BATTERY C

ASSEMBLED CHARGERS 6 v. 1 а. . . . . . . . . . . . 19/9 6 v .2 a.
$6 / 12$
$6 / 12$
v.
1 $\frac{1}{}$ a. $6 / 12$ v. 2 a.
$6 / 12$
v. 4 a. Above 4 a......... Above ready for use with mains and output leads. Cases well ventilated and finished in stoved blue hammer. Carr. \& Pkg. 3/6.

CHARGER
TRANSFORMERS 200-230-250 v. $50 \mathrm{cf} / \mathrm{s}$, 0-9-15 v. $1 \frac{1}{\text { a }}$ a., 11/9; 0-9-15 v. 3 a., 16/9; $\begin{array}{lllll}0-9-15 & \text { v. } & 5 & \text { a., } & 19 / 9 ; \\ 0-9-15 & \text { v. } & 6 & \text { a., } & 23 / 9 .\end{array}$

## LINEAR L3/3 STEREOPHONIC AMPLIFIER

Sensitivity 150 m.v. for 3 watts output on each channel. Ganged Vol, and Tone Controls. Pre-set balance control. Outputs for two matched 2-3 ohm speakers. (Can be used as straight 6 -watt amplifier.) Provides remarkably realistic

output when connected to 200-250 v. A.C. mains point. Stereophonic pick-up head and good quality speakers. Instructions and guarantee included.

Carr. free
Send S.A.E. for leafler.
Or deposit $27 / 9$ and 5 monthly payments $27 / 9$.

7 gns.

12in. 10 WATT HIGH QUALITY LOUD


## EXTENSION SPEAKERS

Limited number in handsome Walnut veneered cabinets. 2-3 ohm speech coils, $6 \frac{1}{2}$ in. 29/9، 8in. 35/9. 10in. 56/9.
DRY SHAVERS. Brand new in carrying case. Operation from 3 U2 batteries, fitted in case. Just the thing for travel. Only 59/6 (approx. half price)
RECORDING TAPE. 600ft. reels, 9/9 1,200ft. reels $14 / 9$.
ELECTRIC SOLDERING RONS. Light weight type for radio work, 19/9.
Cabinet can be supplied separately at $47 / 6$,
GARRARD 4-SPEED AUTO-CHANGER. at $£ 10 / 19 / 6$. Carr. 5/6.

COLLARO 4-SPEED AUTO-CHANGERS With studio pick-up with turnover head. 200-250 N A. Cartoned latest model. For Conguest $57 /$. . mains. Very limited number Conquest $£ 7 / 19 / 6$. Continental 9 gns. Carr. $5 / 6$.

Model RC120/4H. Limited number only
ACOS Crystal Microphone Inserts. Brand new. Only $5 / 11$ ea. Ex. Equip. 4/11 ea. ACOS HGP59 Hi Fi Crystal Cartridges. (Turnover type with sapphire stylus.) Standard
replacement for Garrard and B.S.R. Only

## R.S.C. BATTERY TO MAINS CONVERSION UNITS

Type BM1. An all dry battery eliminator. Size $5 \frac{1}{2} \times 4 \frac{1}{2}$
2in. approx. Completely replaces batteries supply 1.4 v
and 90 v . where A.C. mains $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$ is available.
1.4 v. and 90 v . This includes latest low consumption types Complete kit with diagram. 39/9 or ready for use $46 / 9$.
Type BM2. Size $8 \times 5 \frac{1}{2} \times 2$ itin. Supplies 120 V .90 v ., and 60 V. 40 mA . and 2 v. 0.4 a. to 1 amp. fully smoothed THEREBY COMPLETELY REPLACING BOTH H.T. BATTERIES AND L.T. 2 V. ACCUMULATORS When FOR ALL BATTERY RECEIVERS normally using $2 v$ accumulator. Complete kit with diagrams and instructions, 49/9, or ready for use, 59/6.

## R.S.C. TRANSFORMERS

## FULLY GUARANTEED

MAINS TRANSFORMERS
FULLY SHROUDED UPRIGHT MOUNTING $250.0-250 \mathrm{v} .60 \mathrm{~mA}, 6.3 \mathrm{v}, 2 \mathrm{a}, \mathrm{E} v .2 \mathrm{a}$.
$250-0-250 \mathrm{v}, ~$
$200 \mathrm{~mA}, 6.3 \mathrm{v}, 4 \mathrm{a}, 5 \mathrm{v}, ~$

$300-0-300$ ₹. 130 mA ., 6.3 マ. 4 a., 5 จ. $\qquad$ kuitable for Mullard 510 Amplifier
 8. 3 a.

TOP SHROUDED DROP-THROUGE TYPE
 $250-0 \cdot 250$ v. $100 \mathrm{~mA}, 6.3$ v. $4 \mathrm{a}, 5 \mathrm{~F} .2 \mathrm{a}$. $300-0-300$ จ. $100 \mathrm{~mA}, 6.3$ マ. 4 a., 5 v. 3 a. $350=0-360$ v. $100 \mathrm{mAA}, 6.3$ v. $4 \mathrm{~B} ., 5$ v. 3 a. $350-0-350$ จ 150 mA ., 6.3 จ. 4 a ,
ELIMINATOR TRANSFORMERS Primaries 200-250 v. $50 \mathrm{c} / \mathrm{s}$.


Primarlea $200-250 \vee 50 \mathrm{cta}$



OUTPUT $7 / 9$$8 / 11$OUTPUT TRANSFORMERS
man Pentode $5,000 \Omega$ to 30Standard Peutode $5,000 \Omega$ to $3 \ddot{n}$
Giaudars Pentode $8,000 \Omega$ to $3 \Omega$
Push-pull 8 watte 876 toPuah-pull 8 watts $8 \mathbf{V} 6$ to 5 ohmsPush-pull 10-12 watts 8 V 6 to $3 \Omega$ or $15 \Omega$
Puah-pull 10-12 watts to match 6V6 to $3-5.8$ or $15 \Omega$Push-pull 15-18 watte, sectlonally wound, 6L6, KTG6etc., to 3 or 15 ohms
Push-pull
20
watt high-quality6L6, KT66, etc., to 3 or 15SMOOTHING CHOKES$250 \mathrm{~mA}, 5 \mathrm{H} ., 100$ ohm$150 \mathrm{mA},. 7-10 \mathrm{H} ., 250 \mathrm{ohms}$0 mA ., 10 E. 200 ohmsA0 mA, $10 \mathrm{H} ., 360$ ohm

## HARGING EQUIPMENT

 BATTERY CHARGERKITS ASSEMBLED Consisting of Mains Trans- CHARGER former F. Weifil vendge. Metal Rectified, well ventilated steel case. Fuses, Fuse-holders, Grommets, panels and circuit. Carr. 2/6 extra.6 v . or 12 v .1 amp .
As above, with ammeter 6 v. 2 amps. 6 v . or 12 v .2 amps. 6 v. or 12 v. 2 amps (inclusive of ammeter) 6 Y. or 12 v. 4 amps. $\begin{array}{lll}\text { (inclusive of ammeter) } & 41 / 6 & \text { tractive hammer } \\ \text { Y O. } 12 \text { V. } 4 \text { amps. } \\ \text { BATTERY CHARGE } & 53 / 9 & \text { blue. Ready for }\end{array}$ $6 / 12$ v., 6 amp . consisting of F.W. Bridge Rectifier Mains Trans. and ammeter. 49/9. Post 4/6.
.22/9
6 v . or 12 v .
2 amps .

## Fitted Ammeter

 $32 / 9$ and selector plug $25 / 9$ for 6 v . or 12 v . 31/6 Louvred metal case, finished atuse with mains Double Fused Only Carr. 3/9.All for A.C. Mains $200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s}$ Guaranteed 12 months

## ASSEMBLED 6 v . or 12 v .



## 4 amps.

Fitted Ammeter and variable charge selector Also selector plus for 6 v or 12 v , charging. Double fused. Well ventilated steel case with blue hammer finish. 75 / hammer finish. 75/= weady for use with malns and output leads. Carr. 4/6. Or monthly payments $14 / 11$. 1411 and five As above but nts 14/11
5 GNS. Carr. 5 amp. charging. 19/9 and Carr. 5/- Or Deposit 19/9 and five monthly payments of
19/9.

## R.S.C. A. 10 ULTRA LINEAR 30 WATT AMPLIFIER

FIGH FIDELITY PUSH-PULL UNIT EMPLOYING SIX VALVES. EF86 EF86, ECC83, 807, 807, G234. Tone Sensitivity is extremely high. Only 12 millivoits minimum input is required for full ABMITYOFANY TYPEOR MAKEOF MICROPHONE OR PICK-UP. Separate Bass and Treble controls give both "lift" and "cut" with ample tone correction for long playing records. An extra input with associated vol. control is provided so that gram, gram, etc., etc., can be simultaneously applied for mixing purposes. AN OUTPUT FOR SUPPLY OF 300 v. 20 mA . and FOR SUPPLY OF
6.3 v. 1.5 a. FOR A RADIO 20 mA FEDER 6.3 V. 1.5 a. FOR A RADIO FEEEDER
UNIT. Price in kit form with easy-to-follow wiring diagrams.
Cover as illustrated 11 GnS. Cover as illustrated

11 Gn
 TERMS ON ASSEMBLED UN
and 12 monthly payments of $24 / 9$.


Type 807 output valves are used with High Quality Sectionally wound output trans former specially designed for Ultra Linear operation. Negative feedback of 20 D.B. in main loop. CERTIFIED PERFORMANCE FIGURES ARE EQUAL TO MOST EXPENSIVE UNITS AVAII. ABLE. Frequency response $\pm 3$ D.B. $30-20,000 \mathrm{c} / \mathrm{cs}$. , Tone Controls $\pm 12$ D.B. at $50 \mathrm{c} / \mathrm{cs}$, , $+12 \mathrm{D} . \mathrm{B}$. to $-6 \mathrm{D} . \mathrm{B}$. at 12,000 c/cs. Hum and noise 70 D.B. down. Good quality reliable components used. Chassis finish blue hammer. Overall size $12 \times 9 \times$ 9 in . approx. Power consumption 150 watts. For A.C. mains $200-250-250$ v. $50 \mathrm{c} / \mathrm{cs}$. Outputs for 3 and 15 ohm speakers. CONNOISSEUR OR FOR LARGE HPLLS, CLUBS or OUTSIDE FUNCTIONS. IDEAL FOR USE WITH MUSICAL INSTRUMENTS SUCH AS STRINGBASS,ELECTRONICORGAN, GUITAR, etc. FOR DANCE BANDS,


LINEAR LT/45 HIGH QUALLITY TAPE DECK AMPLIFIER COMPLETE WITG POWER PACK and OSC. STAGE. Sultable for Collaro,
 maing Output for standard $2-3$ ohm speaker. Only 15 millivolts input required for full recording. Only 2 millivolts minimum output required from recording head. Magic Eye recording level indicator. Pronetion for feeding P.A. amplifier. Negative feed back equalisation. Linear frequency repponse $\pm 3$ D.B. $50-11,000$ ofcs. Facillties for recordlogs at 18 in ., 7 ilin . or 33 in , 12 Ready for use tuy from record to playback position automatic Or Dep. 22/3 and gain and output controls. Valves type ECC83,
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COLLARO JUNIOR 4 SPEED REGORD PLAYER with separate pick-up having dual point sapphire stylus. Brand new, cartoned. For 200-250 v. A.C. maine only. Only £4/10/-, Post $3 / 6$

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12 VOLTS 10 WATT AMPLIFIER Complete with internal dynamotor 2-6L6s push-pull, 2-6N7. Incorporating mike and gram inpurs, speaker outpurs, tone and volume controls. Size $12 \frac{1}{2} \times$ $64 \times 8 \mathrm{in}$. Sprung mounted. ONLY 68/10/-. Carr. 10/-
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Worth 520 . Comcabinets $16 \times 7 \times 6 i n$. Worth 22 . ComFREQUENCY METER. LM14 120 Ke/s. to $20 \mathrm{Me} / \mathrm{s}$. accuracy $0.01 \%$. Brand new. Only $£ 27$ each, carriage 10/-. WESTINGHOUSE J.50 PENCIL RECTIFIERS, $500 \mathrm{v}$. mA., 5/-. P. \& P. 1/-

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Contains Hand Generator giving 6 V output with overload cut-out, idea to boost a flat car battery; 6 volt Vibrator Pack giving 3 voles and 150 volts with output leads, plugs, etc., for Army 38 and 18 sets; Battery box for three 2 V . accumulators. This is a fully portable and multipurpose power supply unit. BRAND NEW.

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TRANSMITTER: Is 4-channel erystal controlled, usin 6G6 ose., $12 A 6 \mathrm{harm}$. ampl. 832 har . ampl and driver, and $6 G 6$ ose., 12 A6 harm. ampl. 832 harm. ampl. and driver, and
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Covering $1.5-12 \mathrm{Mc} / \mathrm{s}$. De signeation, consists of V.F.O Buffer, Doubler, P.A. with internal push/pull modulator and provision for V.F.O. or crystal control on 4 channels Output 40 watts phone 100 watts C.W. Complete with 7 valves, aeria current meter and R.F current meter.
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$15 /-$
Senior Model. Total frequency coverage is $50 \mathrm{Kc} / \mathrm{s}-30 \mathrm{Mc} / \mathrm{s}$, obtained by inserting coil sets of which 4 sets are supplled with every receiver. Incorporates 9 valves, 2 R.F., 2 I.F. and 2 audio amplifiers. Crystal gate with crystal phasing control. Variable selectivity. Effective tuning scale of $\mathbf{1 2}$ feet. R.F. and L.F. gain controls. Signal strength meter. B.F.O. on/off and pitch control. H.T. and "S" meter on/off switch. Phone jack and loudspeaker terminals. Black crackle cabinet, $17 \frac{1}{2} \mathrm{in} . \times 9 \mathrm{in} . \times 12 \mathrm{in}$ deep. Power requirements 240 voles D.C. at 70 M.A. and 6.3 volts at 3.4 amps. Fully tested and guaranteed. Each member of our staff has purchased one of these receivers-need we say more? We are proud to offer these magnificent receivers complete with 110 volt or 6 volt (state which preferred) power pack

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A truly magnificent first grade L.F. oscilloseope incorporating a hard valve time base, speeds 1-5-40 milliseconds, easily extended for a few shillings to 3 c.p.s. to $30 \mathrm{kc} / \mathrm{s}$. speeds. Has high-class amplifier with controls. Brightness and focus controls, $X$ \& $Y$ shifts. 1.15 V . and 230 v . mains power pack fully in eded. Employs $2 \frac{1}{2}$ in.

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Covenng 5 to $13 \mathrm{Mc} / \mathrm{s}$. The most compact ocelver ever produced for the U.S. forces. Only gin. $\times 2$ zin. $\times 1$ in. Frighing 3000 oz . vaive permeabis. convertor (oscilistor olxer), (1)-I.8.5. diode detector, AVC ad first audlo ampliher, (1)-I.L.4. 2nd audlo, power amplleter. Lncorporater ou/oli gain and frequency controls. Sapplied com pleve with aesdphones, aerial, aratohing attery container, canvas carrying case with rom and siandard dry batteries, iconsumption 6.5 mA.). Manufactured by Emerson Radio Brand new th original cartons. Fully guaran bed and ready for ONLY

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0-300 v. A.C. $2 \frac{1}{2} \mathrm{in}$. F/M., 251-
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A.H. less handie, $5 /=$ P. \& P. $2 /-6$ for A.H. less handle,
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 26 ohms at 6.5 amps, very liberally ratedBrand new and unused, 251, \& $P .3 / 6$. TRANSMITTER RECEIVER No 19, Mk. I6. Complete station comprising Transmitter/receiver, power supply unit, Transmitter/receiver, power supply unit, aerial, variometer, control box, head-
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Output 12 watts fitted for mike and gram inputs. A sound and practical unit in inputs. A sound and practical unit in
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 2.9 kVA. Pri. 230 V., $50 / 60$, Carr, paid.
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E.H.T. TRANSFORMERS. $3,850 \mathrm{v}$, at 50 mA with two additional 4 V . L.T. windings for 230 v .50 cycles primary. New and boxed. 63/15/-. Carr. $5 /$-. E.H.T. TRANSFORMER. 1,800-0-1,800 at I kVA. 230 v. 50 eycles primary. Fully tropicalised. New and boxed. E8/15/-. Carr. 10/.
HEAVY DUTY LT TRANSFORMER. 230 v .50 cycles pri. 11.12 .6 v . at 70 amps . sec. 230 v . 130 cycles pri. 11.12 .6 v . at 70 amps. sec-
Ditto 13.15 v . sec. at 60 amps . Both capable o earrying $25 \%$ over actual rating. Perfect condition. earrying $25 \%$ over actual rating.
ONLY $115 /$ each. Carr. $5 /-$.

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12 in. P.M. 15 ohms 15 warts, $30-14,000$ c.p.s. Our price $£ 4 / 10 /-$
H1-FI MASTER" 12 in. 15 ohms. 12 wates, 20-16,000 e.p.s. Flux density approx. $14-15,000$. OUR PRICE E7/10/-.
S5PER MI-F $25,12 \mathrm{in} ., 15$ ohms, 25 watts 25-20,000 c.p.s. Flux density 17,600 . OUR PRICE 9/9/- All the above speakers are Brand New and full deseriptive specification is availablo.


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amps., $£ 7 / 10 / \mathrm{m}$. Carr. $10 /$. amps. $£ 7 / 10 / \ldots$ Carr. $10 /$.
CONSTANT VOLTAGE
190-260 V primary, sec. 115 190-260 v. primary, sec. I 15 v . at I $\frac{1}{2} \mathrm{kVA}$. (listed at $2 \mathrm{kVA})$. Brand new and unused. $£ 25$ or $£ 45$ per pair. Carr. 20/- each.

## CEAVY DUTY-ALL STEER TRIPOD STANDS

Adjustable every 6 in. tó approx. 9ft. Gin. when fully extended. (Fo'ds up to only 4 ft . 6in. for storage). Suitable for outdoor speakers, public address systems. floodlighting, etc., etc.



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TEST SET TS-26/TSM. This volt ohmmeter is the correct tester for EE8 tolephones and all standard telephone equipment. Brand new and boxed, with full technical data and calibration charts. $£ 7 / 10 /$-.
PRECISION SERIES 834-S (U.S.A.) Multi range tester for A.C./D.C. volts, ohms and milliamps. Basic movement 400 mieroamps. Housed in wooden box with lid and carrying strap. Over-
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EVERSHED \& VIGNOLES 100 v. MEGGER. Good working order. Limited quantity. ONLY E4/15/0.
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RECORD MEGGERS. 500 v . insulation tester $0-20$ megohms. In leather case, good condition, $t 8$. EVERSHED \& VIGNOLES WEE MEGGER 500 v. New and unused. Only $\mathrm{f} 12 / 10 / \mathrm{-}$. Ditto $250 \mathrm{v}, \mathrm{E} 10 / 10 / \mathrm{F}$. P, \& P, 3/- on each.
AVO TEST BRIDGE. A.C. mains operated from $\mathbf{2 0 0 - 2 5 0} \mathrm{v}$. Will test resistance from 5 ohms to 50 megohms and capacity from 00001 to 50 mids. A most useful instrument for everyday uses. Our price ONLY $£ 7 / 19 / 6$. P. \& P $3 / 6$.


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Variable tre ble cut and bass boost controls sensitivity 100 MV for 3 -watt output
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+or
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Complete amplifier wired and tested with quality sectionalised output translormer t mumara specifieation.
(less speaker)
Carr. and ins.
88/8/-
Wired Power O/Put Socket with Additional Smoothing for F.M. Tuner $10 / 6$ extra.
Stereo version now under development.

## JASON F.M. TUNER UNIT $87-105 \mathrm{mc} / \mathrm{s}$

Designer-Approved Kit of parta to build this modern bighly successful unit drilled chassis and auperior type dial. Colls, caris and all qualty components, etc., for oniy 5 gns. .post free. Set of 4 spec. EF91 or equiv. valves $30 /$-post free. Illusirated
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1001250 v.
$8+8 / 450$
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$16+16 / 450 \quad 3 / 6$ $38 / 350 \mathrm{\nabla}$ $\begin{array}{lll}32 / 500 \\ 32+32 / 450 & 5 / \%\end{array}$ Comprehensiv MIDGET TRANSISTRange in stook.
 CONDENSERS-Silver Mica. All pres. values, 2 pl to 1,000 pl., 6 d . each. Ditto ceramios 9 d . each. Tubulars 450 o. T.c.C. 02, 1/500 $\quad$ V., 1j- each. .25 \#uats 1/6. 5 T.C.C. 1/O. $0018 \mathrm{kv}, 5 / 6.0120 \mathrm{kF}$, $8 / 6$. RESISTORS-FULL RANGE 10 ohms10 megrohms $20 \%$ \&w. $3 \mathrm{~d} . \mathrm{i}+\mathrm{w} .5 \mathrm{~d} ., 1$ w. 6 d. $5 \%$. $9 \mathrm{d} .10 \$.
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are built in heavy metal cabinets approx. Weight are built in heavy metal cabinets approx. Weight
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 $\frac{\text { ABBREVIATIONS: }}{\text { SINGLES }}$|  | SINGLES |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Capacity <br> (Mids.) | Wkg. |  | Price | Capacity | Wkg. |  |  |
|  | 275 | W/8 | Price | (\%ms.) | Volts Siz | Type | Prioe |
| ${ }_{2}^{2}$ | ${ }_{6}{ }^{12} \times 1$ |  | 116 | ${ }_{84}^{80}$ | $35011 \times 2$ $2751 \times 3$ | ${ }_{\text {T/8 }}$ | 216 218 |
| 8 | $12 \times 1$ |  | 116 | 84 | $3501 \times 3$ | 0 | 31. |
| 2 | $50 \times 14$ | w | $1 /$ | 70 | $3{ }^{3} 5 \times$ | -8 | $1 / 6$ |
| ${ }_{2}$ | ${ }_{200}^{70}$ of ${ }^{\text {of }}$ | W | 1/6 | ${ }_{75}^{75(R e v)}$ | $121 \times 3$ | c | 10 d . |
| 4 | 150 | T/8 | $10 \mathrm{d}$. | ${ }_{80}^{75}$ |  | $\mathrm{T}_{\mathrm{W} / \mathrm{s}}$ | 1/6 |
| 4 | $150 \times 1 \frac{1}{6}$ | w | 1/- | 90 | $35011 \times 3$ | T/8 | 2/6 |
| 5 | $531 \times 1$ |  | $1 / 8$ | 100 | $12 \times 18$ | W/8 | 1/. |
| 5 | $100 \times 1$ |  | 1/6 | 100 | $25 \times 1$ | T | 1. |
| 8 | $1211 \times 1$ |  | 1/6 | 100 | $23 \times 18$ | T | $1 \%$ |
| 8 | 15 \% $\frac{5}{16} \times$ |  | 1/6 | 100 | $25 \times 14$ | T// | $1 / 3$ |
| 8 | $251 \times 1 t$ | T-8 | 1/- | 100 | $25.1 \times 1{ }^{1}$ | W/8 | 10 d . |
| 8 | 150 | w | 1\% | 100 | $27014 \times 2$ |  | $2 / 3$ |
| 8 | 150 | T | 9d. | 100 | $2751 \times 3$ | 0 | $2 / 6$ |
| 8 | ${ }_{20}^{200}$ | w | 1/2 | 100 | $27511 \times 2$ | C/S | 216 |
| 8 |  | ${ }_{\text {c }}$ | ${ }_{21}^{1 / 0}$ | 100 | $\bigcirc{ }^{275} 1 \times 3$ | P18 | $2 / 6$ |
| 8 | $450 \times 1$ | w/s | $1 / 11$ | 100 | $3001 \times 3$ <br> 350 <br> 1 |  | $2 / 9$ |
| 10 | 15 1 $\times 1$ | w/s | 116 | 100 | $5001 \% \times 3$ |  | 4. |
| 10 | $25 \times 1 \frac{1}{2}$ | w | 1/- | 150 | $251 \times 2$ | T |  |
| 10 | 480 | W/8 | $1 / 8$ | 150 | $1501 \times 3$ | W/s | $1 / 8$ |
| 18 | ${ }_{150}^{25}$ |  | $1 / 6$ | 200 | ${ }_{8}{ }_{1} \times 1$ |  | 1/. |
| 18 18 |  | T/8 | 1. | 200 200 | ${ }_{35}^{25} 1 \times 17$ | T/8 | 10 d . |
| 16 | 273 |  | $1 / 8$ | 200 | ${ }^{350} 1 \times 1{ }^{1} \times 13$ | C/8 | 10 2- |
| 16 | $300 \times 1{ }^{1}$ | W-s | 1/3 | 200 | $27511 \times 3$ |  | $2 / 6$ |
| 18 | 350 | P | $1 / 6$ | 250 | $61 \times 2$ | T/8 | 1. |
| 16 | $3501 \times 2$ | T | 19 | 250 | $12 \times 1$ | W | $1 /$ |
| 18 |  | $\stackrel{8}{8}$ | $1 / 9$ | 250 | 12 | W/8 | 1/3 |
| 18 | 450 |  | 2/\% | 250 | ${ }_{25}^{28} \times 1 \times 1$ | - | $1 / 3$ |
| 18 | $4501 \times 3$ | 0 | 2/- | 250 | $25 \times 1$ | W/T | $1 / 3$ |
| 20 | ${ }^{6}$ f $\times 1$ ¢ |  | $1 / 6$ | 250 | $25 \times 1$ | W/8 |  |
| 20 | $1501 \times 1$ | T/W | 1/- | 250 | $501 \times 2$ |  | 1/6 |
| 20 | $4501 \times 2$ | W/s | $2 \%$ | 250 | $10011 \times 3$ | P | 1/3 |
| 20 | $4501 \times 3$ | T/88 | 2- | 250 | $20011 \times 2$ | T | 1/6 |
| 25 | $251 \times 14$ | W/8 | 10 d | $400(\mathrm{Rev})$ | $61 \times 2$ | $\stackrel{\text { P }}{ }$ | 1.- |
| ${ }_{2} 5$ | ${ }_{50} 5$ | W/8 | $1 /$. | 500 | ${ }^{6}{ }^{1} \times 18$ | T | 10 d. |
| 25 | 350 | w | $8 \%$ | 500 | $121 \times 17$ | T | 1 10d. |
| 30 <br> 38 <br> 8 | $12{ }^{12} \times 1 \times$ | -8 | $1 / 6$ | 500 | $121 \times 2$ | ${ }^{\text {T }}$ | 10 d |
| 32 | $2751 \times 2$ | 0 | $1 / 6$ | 500 |  |  | 1/3 |
| 32 | $2751 \times 2$ | ${ }_{\text {P }}$ | $1 / 6$ | 1000 | $3351 \times 3$ | c | 1/6 |
| ${ }_{40}^{38}$ | (150 | W/8/8 | $2 /$ | 3000 | $2811 \times 4{ }^{6}$ | 0 | 4/- |
| 40 | $3501 \times 2$ | ${ }^{\text {P }}$ | 21. | 5000 5000 |  | -8 | 3/- |
| 40 | $4501 \times 3$ | W/8 | 216 | 6000 | $811 \times 3$ | -8 | 4/- |
| 50 | $50 \times 1$ | T | 113 |  |  |  |  |
| 5 | 50 |  | 1/6 |  | OUBLE |  |  |
| 50 50 |  | W/8 | $1 / 6$ | 8+8 | ${ }_{350}^{275} 1 \times 2$ |  | 3 |
| 50 | $3301 \times 2$ | W/8 | $2 / 3$ | $8+8$ | ${ }_{450}^{3501}$ | W/8 | 2/6 |
| 60 | $2751 \times 3$ | W | $2 /$. | $8+8$ | $4501 \times 2$ | W/8 | 2/9 |


| Capacity (Mids.) | Whg. volts Size" | Type | Price | Cajacity (MIds.) $\begin{gathered}\text { Wke. } \\ \text { Volts. } \\ \text { Size }\end{gathered}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $8+16$ | 450, $1 \times 1$ | W/8 | 3/8 | $100+100$ | Type | ${ }_{1 / 6}$ |
| $10+10$ | $4501 \times 2$ | w/B | 2/8 | $100+100 \quad 27511 \times 3$ |  | 6 |
| $12+12$ | $2761 \times 2$ |  | $2 /-$ | $100+200 \quad 251 \times 2$ | ${ }^{\text {IP }}$ |  |
| $12+18$ | $3601 \times 2$ |  | 2/6 | 100+200 | - | 16 |
| $12+24$ | $2751 \times 2$ |  | 2/- | $100+2000^{2750} 10 \times 44$ | ${ }_{0}$ | \% |
| $12+88$ | $2751 \times 2$ |  | 2/- |  |  |  |
| $18+16$ | 150 \% $\times 1$ | . 8 | 1/- | $100+300$ 2751104 |  | - |
| $16+16$ | $3501 \times 2$ | C | 3/- | $100+400 \quad 2752 \times 44$ |  |  |
| $16+32$ | $27514 \times 4$ |  | $2 / 6$ | $600+1400 \quad 2762\} \times 4{ }^{2}$ |  |  |
| $16+32$ | $2751 \times 2$ | P | $2 / 6$ |  |  | 9/6 |
| $20+10$ | $4501 \times 3$ | - | 3/- | TRIPLES | tc. |  |
| $20+20$ | $1501 \times 2$ | W/S | 1/- |  | 1 |  |
| $20+20$ $24+24$ | $4501 \times 3$ $2751 \times 3$ | W/8 | $3 / 9$ | ${ }_{12+12+18}^{18}$ | 1 | - |
| $24+24$ $24+24$ | $\begin{aligned} & 2751 \\ & 350 \\ & 1\end{aligned} \times 8$ | c | ${ }_{3 / 6}^{2 / 6}$ | $\begin{array}{lll}19+24+24 & 275 \\ 16+8+4 & 275 & \times 3 \\ 1 & \times 3\end{array}$ | P | - |
| $25+25$ | $3001 \times 2$ | T | $2 /$ | $16+16+16$ | ${ }_{0}$ | - |
| $30+30$ | 150 1 $\times 17$ | w/8 | 1/- | $19+39+39 \quad 2751 \times 3$ |  | 6 |
| $32+18$ | $2001 \times 2$ | P | 1/6 | $20+10+10 \quad 35014 \times 3$ |  |  |
| $32+16$ $38+18$ | $35017 \times 2$ 8501 | T | $3 / 6$ | $\begin{array}{ll}20+15+15 & 450 \\ 20+20+80 & \text { P3 } \\ 200\end{array}$ | -8 | 6 |
| $32+3{ }^{2}$ | $1501 \times 2$ | T | 316 | $25+25+25 \quad 251 \times 2$ | C/8 | 8 |
| $32+32$ | $2501 \times 24$ |  | 1/6 | $\begin{array}{ll}30+30+30 & 275 \\ 32+8+8 & 185 \\ 188 \\ 182\end{array}$ |  |  |
| 32 233 | $27514 \times 2$ | c | $2 / 9$ | $32+32+8{ }^{275}$ 11 $\times 2$ | o | $2 / 6$ |
| $32+38$ | $27514 \times 3$ |  | $2 / 9$ | $32+32+8 \quad 27511 \times 2$ |  | 2- |
| 俍 $\begin{aligned} & 32+32 \\ & 32+32\end{aligned}$ | - $3501{ }^{1} \times 2 \times 2$ | ${ }_{-8 / 8}$ | 4/- | $32+3{ }^{32}+8 \quad 2751 \times 3$ | P/C | - |
| $32+32$ | $45011 \times 3$ | W/g | 4/6 | $3{ }^{32}+32+168176$ |  |  |
| $32+324 / 10$ | ${ }^{350} 11 \times 2$ | T | 36 | $22+32+255275 / 251 \times 3$ | T |  |
| $40+18$ $40+80$ | $2501 \times 3$ $1501 \times 2$ | T | $1 / 6$ |  |  |  |
| $40+40$ | $150.1 \times 3$ | ${ }^{\text {P }}$ | 1\% | $\left.{ }_{32}+300+70 \quad 27511 \times 4\right\}$ |  |  |
| $40+40$ | $1501 \times 2$ | W/8 | 1/- | $40+20+10 \quad 3501 \times 3$ | P | 6 |
| $40+40$ | ${ }^{275} 11 \times 2$ | C/s | $2 / 9$ | $40+30+20 \quad 30011 \times 2$ |  | $2 / 9$ |
| $40+40$ $40+40$ | ${ }_{4}^{300} 1 \times{ }^{1} \times 1 \times 3$ |  | $2 / 8$ | $40+30+20 \quad 1501 \times 2$ |  |  |
| $50+3$ | $1501 \times 2$ | W/8 | 1/. | $40+40+12 \quad 27511 \times 2$ | P | - |
| $50+50$ | ${ }^{25} 1 \times 2$ | C | 1/6 | $40+40+20 \quad 27511 \times 2$ |  | 8 |
|  | $1501 \times 2$ | w/8 | $1 / 6$ | $40+40+20 \quad 2751 \times 3$ | P |  |
| $50+30$ | $2001 \times 3$ |  | 1/6 | $40+40+32 \quad 27812 \times 24$ |  | $2 / 8$ |
| 50+50 | ${ }_{275}^{2511 \times 2}$ | P | 1/9 | $\begin{array}{llll}40+120+70 & 278 \\ 44 & 11 \times 4\end{array}$ |  | 8 |
| $50+50$ | $27511 \times 3$ | c | $2 / 9$ | $50+24+24 \quad 27511 \times 3$ |  |  |
| $50+50$ | $27511 \times 2$ | T | $2 / 8$ | $50+50+6 \quad 27511 \times 3$ |  |  |
| $50+30$ | $30011 \times 2$ | P | $3 /-$ | $50+50+10 \quad 15011 \times 2$ | T |  |
| $50+50$ | $3001 \times 2$ | T | 3/- | $50+50+50 \quad 35011 \times 3$ | P | 316 |
| 80+500 | 3005 | P | $3 /$ | $80+300+30 \quad 27519 \times 4\}$ |  | 18 |
| $65+100$ | $25011 \times 3$ | P | 3/8 | (100+100+50300 $11 \times 3$ |  | 16 |
| $60+200$ | $27511 \times 4$ | c | 41 | $100+250+250275{ }_{2} \times 4$ | C |  |
| ${ }^{60}+200$ | 27814 | O | $41-$ | $100+400+1627512 \times 4$ |  |  |
| $80+250$ 80 80 | ${ }_{275}^{275} 11 \times 4$ | - | 518 | $100+400+3227517 \times 4$ |  | - |
| $100+35$ | $25011 \times 3$ | P | $\begin{aligned} & 3 / 6 \\ & 4 /- \end{aligned}$ | $200+250+2502752+48$ $40+20+10+1035011$ | 0 | $8 / 6$ $3 / 6$ |

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This is an excellent position with very good prospects with expanding CompanyA group Pension Scheme is in operation. Previous experience in calibration up to $250 \mathrm{Mc} / \mathrm{s}$ desirable. Write stating age, full details of previous experience, academic qualifications and salary envisaged to:

## Farnell Instruments Limited, Wetherby Industrial Estate, York Road, Wetherby,

quoting reference: "Calibration" on top of envelope.

## TELEVISION DEVELOPMENT ENGINEER

Due to expansion in our Development Laboratory, applications are invited for the following posts:-

1. DEVELOPMENT ENGINEER with previous experience on Time Base and Scanning Circuits.
2. DEVELOPMENT ENGINEER with previous experience on Tuner Design, preferably with Printed Circuit knowledge.
Both these positions are permanent and pensionable. The factory is situated on the South coast, with many attractive local amenities.
Written applications, with full particulars including age, qualifications, experience, and salary required, should be made to:-

The Chief Engineer
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Applications to:
The Works Labour Manager (at the appropriate Works address)

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to work in a production organisation on work connected with the testing of missile systems or engaged on the investigation and preparation of test facilities and procedures.
For the senior vacancies, applicants must have had some years' experience working on electronic, particularly radar, equipment. Attendance on Services Radar training courses and/or some knowledge or experience of vibrational testing would be an advantage.

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Grayshaw Instruments
Griffths Marten, Lta. .i.................... 116

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Modern Electrics. Ltd.
Modern Techniques
M. R. Supplies. Ltd
M.S.S. Recording Co.. Ltd. .............. 89

Mullard, Ltd,
Multimustc. Lid....... 3. 15. 44, 79. 91. 98, 164
Cover IV Multitone Electric Co........................................... IV

Nash \& Thompson. Ltd. Newnes, George. Ltd. ................................. 41
 \(\begin{array}{lll}\text { Northern Radio Services, Ltd. . . . . . . . . . . . } & 76 \\ \text { Northworks, Ltd. . . . . . . . . . . . . . . . . . } & 117\end{array}\)

Oddie Bradbury \& Cull. Ltd.



Quartz Crystal Co.. Ltd.
174

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Samsons Surplus Stores ............. 118, 15
Savage Transformers, Lto
.... 175
Semiconductors, Ltd.
- 126, 127

Sifam Electrical Instruments,
Stmmonds, L. E. Ltd.
Smith, G. W. (Radio), Litd. ...........i24, i25. 148

Solartron Electronic Group, Ltd.
Sound News Productions
Southern Radio Supply, Ltd. Spencer-West. Ltd.
Stamford, A. L. ............................. 162
Standard Telephones \& Cables, "Ltd, \(135,139,1\)

Sugden, A. R., \& Co. (Engineers); Litd
Stratton \& Co, Ltd.
Superior Radio Supplies, Lta
Swan Electronics


Vacwell Engineering Co., Ltd. ................................ 78
Valradio, Ltd.
Venner Electronics, Litd.
Vitality Bulbs, Ltd.
Vitavox, Lid.
Electrical Service ............................ 109
Walmore Electronics, Ltc
Watts, Cecil E, Ltd. Order), Lid.
Wayne 5 err Laboratorles, Ltd. The .... 154
Webber. R. A., Ltd. .....................
ii9. 1146
Webb's Radio
West Insulating Co.
Westool, Ltd. \(1 . . . . . . . . . . .\).
Weymouth Radio Mfg. Co., Ltd., The \(9,{ }^{\prime} 33,74\)
Wharfedale WIreless Works. Ltd. \(79 . .9 .{ }_{14}\)
Wi'kinson, L. (Croydon), Ltd.
Wright \& Wealre, Ltd.
z. \& I. Aero Services, Ltd. .............. 170

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\section*{OPINIONS OF TECHNICAL EXPERTS}
P. Wilson, M.A., "The Gramophone"

This is without doubt the most versatile domestic tape recorder that I have had the pleasure of trying out, and the quality, both of its recording and its playback, is of exceptionally high standard. For quality of performance, then, I give the instrumen full marks: I know of no better. For the construction I have nothing but praise. There is nothing fimsy about it either as a piece of mechanism or on the electronic side. It is a fine piece of engineering up to the highest British standards.
D. W. Aldous, M.Inst.E., M.B.K.S.,
"The Gramophone Record Review"
The separate record and replay amplifiers make possible the direct monitoring from this facility is certainly a boon. There is no doubt whatever that when one has used this type of recorder one never wishes to return to the combined record/playback type of instrument. ! have never heard better quality at \(7 \frac{1}{2}\) in.p.s. from any tape recorder that has passed through my hands. The "Reflectograph" is a pedigree tape recorder of immaculate construction and impecof immaculate performance.
cable performance.
Separate motors are used for capstan drive and both spools, all three motors being of Garrard manufacture. The overall im pression after some months of use is that the machine is convenient and pleasant to
handle, while the exira facilities make it handle, while the exira facilities
very suitable for professional use.```


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[^1]:    * These channels are outside the imits of Band III (174-216 Mc/s).

[^2]:    * J. M. Furnival a pioneer of aircraft wireless, and now Consultant to Marconi Instruments Ltd.

[^3]:    * Electronics, Vol. 29, No. 2, p. 186 (Feb. 1956).

[^4]:    * "Elements of Pulse Circuits," by F. J. M. Fariey (Methuen).

[^5]:    *I omitted to mention last month that following the MichelsonMorley experiment, the null result of which could have been explained by the Lorentz contraction without necessarily abandoning the aether, there was a Kennedy-Thorndike experiment which would have showed the existence of an aether if there had been one-in spite of the Lorentz contraction-and it didn't.

[^6]:    * Research Department, British Broadcasting Corporation.

    Royal Some International Geophysical Year Achievements", The Royal Society, December, 1958.

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[^11]:    "Belling-Lee Notes " appear this month on page 69.

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