# Wireless Worla <br> <br> RLECTRONICS <br> <br> RLECTRONICS <br> Radio . Television 




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ELECTRONICS, RÅIO, TELEVISION

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Transistor circuits are usually designed to operate from one battery. They should be stabilised as described in Transistors for the Experimenter. A two-battery system can be used instead (see figure). It is especially suitable for transformer coupling. The collector supply voltage $\mathrm{V}_{\mathrm{cc}}$ is provided by two batteries in series, and the base supply voltage $\mathrm{V}_{\mathrm{bb}}$ is obtained by connecting one resistor $\mathrm{R}_{\mathrm{b}}$ from the base to the common point of the two batteries.

In transformer coupled two-battery circuits $\mathrm{R}_{\mathrm{b}}$ can be very small and even zero. $\mathrm{R}_{\mathrm{b}}$ draws base current only from $\mathrm{V}_{\mathrm{bb}}$, it does not bleed current from $\mathrm{V}_{\mathrm{cc}}$. Hence $\mathrm{R}_{\mathrm{b}}$ can be small to obtain good stability, and in the limiting case when $\mathrm{R}_{\mathrm{b}}=0$ the stability is the same as for a grounded base circuit.

In R-C coupled circuits a low $\mathbf{R}_{\mathrm{b}}$ shunts the input and a value of the order of $10 \mathrm{k} \Omega$ must be used. The main disadvantage of the circuit is that since $\mathrm{V}_{\mathrm{bb}}$ is fixed, once the nominal collector current has been chosen, $R_{e}$ is automatically decided, too. For example, if $\mathrm{I}_{\mathrm{c}} \mathrm{nom}$. is 0.5 mA and $\mathrm{V}_{\mathrm{bb}}$ is 1.5 V , then $\mathrm{R}_{\mathrm{e}}$ can only be $2.7 \mathrm{k} \Omega$. There is therefore a fairly large voltage drop across $R_{e}$, hence $R_{c}$ must be lower, and the gain is reduced. A small-signal R-C coupled circuit, in fact, cannot be operated with a collector supply voltage $\mathrm{V}_{\mathrm{cc}}$ of less than 6 V .

One general advantage of the two-battery circuit, which applies to both R-C coupling and transformer coupling, is that $\mathrm{V}_{\mathrm{cc}}$ can be changed without having to re-design the circuit. If it is necessary to increase the size of the signal, one has only to increase $\mathrm{V}_{\mathrm{cc}}$ while keeping $\mathrm{V}_{\mathrm{bb}}$ the same, and the transistors operate at the same nominal collector currents. The stability is slightly worse because the junction temperature increases, but
the signal is approximately the required size.
A further advantage of the two-battery system is that the effect of resistor tolerances is much smaller. First, the circuit contains few resistors, second, the tolerance on $\mathrm{R}_{\mathrm{b}}$ exerts very little effect, a $10 \%$ spread in $\mathrm{R}_{\mathrm{b}}$ giving only about $1 \%$ spread in $I_{c}$.

Preferred circuits for an OC71 operating with a collector supply voltage $\mathrm{V}_{\mathrm{cc}}$ of 6 V and a base supply voltage $\mathrm{V}_{\mathrm{bb}}$ of 1.5 V are given in the table for R-C coupling and transformer coupling. As when working with only one battery, the design of these circuits can be very much simplified by using graphs. For two batteries the graphs are always straight lines.
 Circuit values can also be calculated, without using graphs. The procedure is as follows. (1) Choose $\mathrm{V}_{\mathrm{cc}}$, $\mathrm{V}_{\mathrm{bb}}, \mathrm{I}_{\mathrm{c}} \mathrm{nom}$., and $\mathrm{R}_{\mathrm{b}}$. (2) Then

$$
\mathrm{R}_{\mathrm{e}}=\left(\mathrm{V}_{\mathrm{bb}}-\mathrm{V}_{\mathrm{b}-\mathrm{e}}+\mathrm{I}_{\mathrm{c}\left(\rho_{0}\right)} \mathrm{R}_{\mathrm{b}}\right) / \mathrm{I}_{\mathrm{c}}-\mathrm{R}_{\mathrm{b}} / a^{\prime} .
$$

(3) Calculate $K$ and $I_{c(q)}$ max., assuming $T_{j}=T_{a m b}$, and allowing $5 \%$ for resistor tolerances if $\mathrm{R}_{\mathrm{e}}$ and $\mathrm{R}_{\mathrm{b}}$ are both $\pm 5 \%$. (4) Find $V_{c(q)} \min$. (5) Choose $R_{c}$ less than the value given by

$$
\mathrm{V}_{\mathrm{c}(q)} \min =\mathrm{V}_{\mathrm{cc}}-\mathrm{I}_{\mathrm{c}(\mathrm{q})} \max \cdot\left(\mathrm{R}_{\mathrm{c}}+\mathrm{R}_{\mathrm{e}}\right)
$$

(6) Find $T_{j}$, and if $T_{j}$ is more than $1^{\circ} \mathrm{C}$ greater than $\mathrm{T}_{\mathrm{amb}}$, recalculate $\mathbf{I}_{\mathrm{c}(\mathrm{q})} \mathrm{max}^{\text {max }}$ etc. accordingly.

PREFERRED CIRCUITS FOR OC7I

|  | Vce | $V_{b b}$ | Re | $\mathrm{R}_{\mathrm{b}}$ | Re | Ic nom. | $I_{c}(q)$$\max .$ | $V_{c(q)} \mathrm{min}$. |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  | at nom. $\mathrm{V}_{\mathrm{cc}}$ | at min. $\mathrm{V}_{\mathrm{cc}}$ |
| R-C.Coupling | 6 V | 1.5 V | $2.7 \mathrm{k} \Omega$ | $10 \mathrm{k} \Omega$ | $3.9 \mathrm{k} \Omega$ | 0.5 mA | 0.81 mA | 0.5 V | 0.3 V |
|  | 6 V | 1.5 V | $2.7 \mathrm{k} \Omega$ | $6.8 \mathrm{k} \Omega$ | $4.7 \mathrm{k} \Omega$ | 0.5 mA | 0.72 mA | 0.5 V | 0.3 V |
|  | 6 V | 1.5 V | $1.2 \mathrm{k} \Omega$ | $10 \mathrm{k} \Omega$ | $2.2 \mathrm{k} \Omega$ | 1.0 mA | 1.6 mA | 0.5 V | 0.3 V |
| Transformer Coupling | 6 V | 1.5 V | $0.47 \mathrm{k} \Omega$ | 0 | $200 \Omega$ | 2.85 mA | 3.05 mA | 3.9 V | 2.6 V |
|  | 6 V | 1.5 V | $1.2 \mathrm{k} \Omega$ | 0 | $200 \Omega$ | 1.1 mA | 1.2 mA | 4.2 V | 2.8 V |
|  | 6 V | 1.5 V | $2.7 \mathrm{k} \Omega$ | 0 | $200 \Omega$ | 0.5 mA | 0.57 mA | 4.2 V | 2.8 V |

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## F.M. Distortion

AT a time when the obvious advantages of v.h.f. broadcasting are being widely publicized it may seem churlish to draw attention to the obverse side of the picture. In the initial phases of the establishment of a nation-wide service there will be many fringe areas in which listeners who change from medium waves to v.h.f. will find that they have substituted a background of steam locomotive shunting noises (arising from aircraft reflections) for heterodyne whistles and monkey chatter. Even in areas of good signal strength there may be distortions arising from multi-path propagation and reflections from fixed objects which would not arise with amplitude modulation.

The problem of aircraft flutter in severe cases where the field strength periodically goes through zero can be solved only by the use of diversity reception with aerials spaced an odd number of quarter-wavelengths apart, but much can often be done with directional aerials to mitigate the lesser troubles by discriminating against the indirect or reflected signal.

Another approach is to improve the capture ratio of the receiver, and in this issue we publish an article which reflects the trend of thought in America and outlines the methods by which the ratio has been narrowed from the normal average of 20 dB to 1 dB or less.

It may well turn out that less drastic and expensive methods will suffice to combat the commoner forms of f.m. propagation distortion, as experienced in Britain, and we hope to deal more fully with these in a subsequent issue.

## Domestic Equipment

LOOKING back at the Radio Show, it seems hardly an exaggeration to say that technical interest has shifted from television to v.h.f. sound broadcasting. . In an appraisal of the exhibits printed elsewhere in this issue, the conclusion is reached that a condition of near-standardization has now been reached in television. For the first time it is possible to put forward a general description of a receiver which will apply with remarkable accuracy to the great majority of modern sets. There is even a measure of standardization in tube sizes. The 17 -inch is apparently the most popular with viewers, but, incidentally, it was interesting
to observe the favourable reactions of ordinary Show visitors to the picture quality offered by the 9 -inch portables. That may be because such small pictures now have a novelty value.

Provision for v.h.f. reception is now practically universal in sound broadcast receivers. Though designs have been generally tidied up, there has been little development in circuitry. The ratio detector is chosen on economic grounds for the cheaper sets and the Foster-Seeley discriminator for the more expensive tuner units. Naturally enough, continuous tuning is used on v.h.f. in the combined a.m./f.m. receivers, but it is surprising to find that push-button (or three-position switch) selection of stations has not gained more ground in v.h.f.-only sets. It seems the natural thing for a three-programme service like ours. No doubt the reluctance to use switch tuning is partly connected with the difficulty in avoiding frequency drift; to fit an external trimmer would probably be considered an admission of defeat. In several new receiver designs advantage is taken of the wider frequency range obtainable from v.h.f. broadcasting by providing better audio sections and loudspeakers.

There are more v.h.f.-only receivers than last year, but their number is still small. According to some authorities, that is because the public generally is believed to insist on provision for reception of Radio Luxembourg.

With the spread of the transistor, the "personal" portable set has at last come into its own. No longer need the user be faced with heavy batteryreplacement costs. Another solution to this problem may be provided by the "hybrid" receiver, which economizes battery current by using transistors in the output stage, with valves elsewhere. A similar technique is applied to a car radio receiver.

It will be interesting to see the long-term public reaction to two novelties in sound reproduction introduced at the Show. The first is a home recorder using a magnetic disc instead of tape; it has the advantage that, by substituting the appropriate pickup heads, it can be used for the reproduction of normal (non-magnetic) disc records. The second novelty is the "transistorized" portable record player, produced by several firms. This remarkable example of electro-mechanical ingenuity is run from a 6 -volt dry battery which supplies the motor and transistors,

# Aeronautical Radio Developments 

NEW EQUIPMENT AT THE S.B.A.C. AIR SHOW

WHILST some new equipment made an appearance at the Farnborough Air Show, organized by the Society of British Aircraft Constructors, there was very little evidence that any strikingly new developments would emerge in the immediate future. Taking a long-term view, however, it seems that some changes are envisaged in channel spacing in the present air-to-ground communications band of 118 to $136 \mathrm{Mc} / \mathrm{s}$. Evidence for this was the introduction by Marconi's of a new high-power aircraft v.h.f. transmitter designed for operation on channels $50-\mathrm{kc} / \mathrm{s}$ wide, which is just about half the current width; as a result double the number of operating channels are available, but not necessarily usable at present in this country. This set, Type AD305, gives 30 watts output, which is considerably more than customarily obtained-from aircraft v.h.f. transmitters.

This increase in power of airborne transmitters is matched by a corresponding increase in some ground sets. Pye have a v.h.f. ground transmitter rated at 1 kW r.f. output, whereas the average output of such sets is in the 50 - to 60 -watt category. It appears that the higher-powered equipments are envisaged for use when the newest types of large jet aircraft come into service on the long-distance routes, as their higher speeds will necessitate v.h.f. communication over greater distances than hitherto.

Long-distance radio telephone communication on air routes has been envisaged for some time past and all the new aircraft communication equipment introduced over the past few years for use in the 2 -to-20$\mathrm{Mc} / \mathrm{s}$ band has been designed with this possibility in view. The Marconi Type AD307 is a case in point and so is the Standard Telephones STR18 series. This year a new STR18D has appeared which provides 200 working spot frequencies selectable from either of two remote control units. Its outstanding features are, fully automatic control by motor operation of the tuning elements of circuits, 100 watts r.f. output on all bands on telephony and transmission on c.w. or i.c.w. for use when a wireless operator is included in the aircrew; otherwise it is operated entirely by remote control by the pilot on telephony.

## Airborne Teleprinter

The tendency towards pilot operation of communication equipment adds considerably to the existing heavy burdens of pilots, but some relief from radio watch-keeping can now be provided by a new and sompact selective calling attachment for aircraft receivers introduced by Marconi's. Known as "Secal" it operates on a two-pulse four-tone signal which is coded by combinations of the tones and only the aircraft decoder adjusted for a particular combination will respond and un-mute the receiver. A warning tone is generated and fed into the pilot's


Decca "Wind-Finding" radar; the operators sit in the enclosed cabin behind the aerial "dish."
headphones, but visual warning can be employed instead if preferred.

An interesting development, which at present is only in the experimental stage, is an airborne teleprinter system. Examples of the equipment were shown by Marconi's and by Standard Telephones; the system operates on the comparatively low radio frequencies of 90 to $130 \mathrm{kc} / \mathrm{s}$. It has been tried out quite successfully on the trans-atlantic air routes for passing meteorological and routine flight information.

Point-to-point communications along the principal air routes is usually effected now by teleprinter and high-speed radio telegraphy using the FSK (frequency shift keying) system. While no basic changes seem imminent a few new items of equipment have made an appearance. One is an electronic receiver muting unit (Type PV97A), developed by Plessey, which operates from the i.f. amplifier and mutes the receiver in the absence of a usable signal. However, in addition it provides a "cleaned up" signal as filters attenuate the noise while an amplifier enhances the signal. Plessey exhibited also a new high-stability version of their dual-diversity receiver (Type PV102B) which provides the correct form of output signal for direct operation of teleprinters.

A few new communications receivers were shown this year, principally by Amalgamated Wireless of Australia, by Cossor and by Redifon, the lastmentioned showing one designed for very accurate resetting to any previously logged signal, or to a desired frequency. It is known as the R145 and embodies a very massive coil turret allowing for fourteen frequency ranges in the $2-$ to $-30-\mathrm{Mc} / \mathrm{s}$ band. Provision is made for double- or single-sideband reception with switch selection of either the upper or the lower sideband.
Improvements in radio navigational aids were exemplified by a sub-miniature version of the Standard Telephones f.m. radio altimeter and by a new automatic radio-direction finder for aircraft (AD712) introduced by Marconi's. This set is based
on the Bellini-Tosi crossed-loop aerial system using flat, dust-iron cored, "dragless" loops in a fibreglass housing; it covers the bands 100 to $415 \mathrm{kc} / \mathrm{s}$ and 490 to $1,779.5 \mathrm{kc} / \mathrm{s}$ with remote selection of any frequency in this range in steps of $0.5 \mathrm{kc} / \mathrm{s}$.

Ekco have produced a new lightweight scanner for their TE120 airborne search radar which cuts 33 lb off the total weight. The addition of a small unit to this equipment now gives precise information on aircraft drift by utilizing the Doppler effect of modulation.

A new 3-cm aircraft radar was shown by Elliott. Based on a Bendix design, it gives warning of heavily charged storm clouds ahead and it can be used also for mapping the ground along the flight path. The working range is 150 nautical miles.

The Doppler effect is utilized in the design of a new Marconi aircraft navigational aid, Type AD2000. The airborne equipment radiates signals towards the ground and these are modulated by the movement of the aircraft over the ground at frequencies bearing a direct relationship to the aircraft's ground speed and drift angle. The reflected signals containing this data are fed to a computer, together with gyro-compass readings, and direct information is provided in the aircraft as continuous readings of latitude and longitude. Drift and ground speed are also available and all this information is obtained without the co-operation of ground stations.

Among the improvements to ground radar equipment is a fixed-coil assembly for use in place of the customary rotating coil assemblies employed for p.p.i. displays. This system enables much greater flexibility in the type of display to be provided solely by switching. It is a Decca development and the same firm have introduced a new $3-\mathrm{cm}$ radar for plotting the tracks of meteorological balloons. They call it a "Wind-Finding" radar and it has an effective working height of $60,000 \mathrm{ft}$.

Standard Telephones have introduced a number of modifications into their precision approach radar, which in its present form (Type SLA3) is considerably smaller, has only four-fifths the former number of valves and is simpler to maintain. A new display console is employed which is fitted with two 17 -in c.r. tubes arranged one above the other for easier
interpretation of the azimuth, elevation and range information shown on the tubes.

Pye exhibited their latest ILS equipment, which is now in production form, and Cossor demonstrated a new method, using a special form of radio "lens," for cancelling the clutter on p.p.i. display tubes produced by heavy rain and storm clouds. The lens imparts a circular polarization to linearly polarized waves and is interposed between the horn and the aerial reflector. It was demonstrated in conjunction with their Type 21 airfield surveillance radar which operates on 10 cm . With the Decca Type 424 radar the operator can now select any type of polarization from linear through elliptical to circular as a means of cancelling rain and snow clutter without affecting the echoes from the aircraft, other than rendering them visible whereas they would otherwise be lost in the clutter.

The two rescue beacons, Ultra "Sarah" and Burndept "Talbe," for aircrews forced down in the sea, were shown with various modifications. The most interesting of these is a transistor-type h.t. supply unit for "Sarah" which functions on the ringingchoke principle and derives its primary power from a 12 -volt Kalium battery. Germanium diodes serve as rectifiers. It is for beacon operation only (not radio telephony) and the efficiency is said to be over $80 \%$.

Display console of the Standard Telephones SLA3 precision approach radar.

Left : Redifon 14-range communications receiver Type R145. Note

Trabsistor h.t. supply unit for Ultra "Sarah" rescue beacon.
the massive coil turret.


Marconi AD712 receiver withdrawn from its case.

# WORTLD OF WIIREIESS 

## Amateur Emergency Operation

THE LICENCE clause prohibiting British amateurs sending messages for a third party has been amended. Although limited to communication "during disaster relief operations or during any exercise relating to such operations conducted by the British Red Cross Society", it is a noteworthy move. It is, however, interesting to note that the lifting of the ban applies to all licensed radio amateurs and not only those linked with the Radio Amateur Emergency Network organized by the R.S.G.B., although the amendment has been made as a result of the efforts of that organization.

It is stressed that the primary duty of R.A.E.N. in an emergency is to transmit information to the nearest centre linked with the G.P.O. telephone-not to usurp the G.P.O.'s function.

## Interference : More Recommendations

ABOUT 80 delegates, representing 17 countries and five international organizations, attended the fifth plenary session of the International Special Committee on Radio Interference (C.I.S.P.R.) held in Brussels a few weeks ago under the chairmanship of O. W. Humphreys (director, G.E.C. Research Laboratories). The United Kingdom delegation was led by E. L. E. Pawley (B.B.C.) and consisted of 14 members drawn from the G.P.O., R.I.C., B.E.A.M.A., I.E.E., Electrical Research Association, Society of Motor Manufacturers and Traders and the Radio Research Laboratory. The main considerations were (a) limits of interference and methods of control; (b) measurement of radio interference; and (c) safety aspects of suppression techniques.
It was recommended that the countries represented should set (within the frequency bands quoted) the following limits to the interfering voltage at the terminal of appliances.

| $150-200 \mathrm{kc} / \mathrm{s}$ | $500-1,500 \mu \mathrm{~V}$ |
| :--- | :--- |
| $200-285 \mathrm{kc} / \mathrm{s}$ | $500-1,000 \mu \mathrm{~V}$ |
| $525-1,605 \mathrm{kc} / \mathrm{s}$ | $500-1,000 \mu \mathrm{~V}$ |


C. J. V. LAWSON

E. ALLARD
(See "Personalities")

## Organizational, Personal and

Industrial Notes and News

## Mobile Radio Changes

THE MAIN proposals in the Second Report of the Mobile Radio Committee», which has been approved by the P.M.G., are narrower channelling and a revised sub-allocation plan. It is proposed that channels in the $165-173-\mathrm{Mc} / \mathrm{s}$ band should be reduced from $100 \mathrm{kc} / \mathrm{s}$ to $50 \mathrm{kc} / \mathrm{s}$, as in the $71.5-$ $88-\mathrm{Mc} / \mathrm{s}$ band. It is further recommended that the possibility of introducing $25-\mathrm{kc} / \mathrm{s}$ channelling in both bands should be examined.
The factors, both technical and operational, affecting channel spacing for mobile radio users are examined in an article on page 481.

* H.M.S.O., price is 6 d .


## PERSONALITIES

J. A. Smale, C.B.E., A.F.C., B.Sc., M.I.E.E., engineer-in-chief of Cable and Wireless, Limited, since 1948, will be retiring in March. As, however, the deputy engineer-in-chief, W. J. Knight, M.B.E., will be retiring at the end of this year it has been necessary to appoint another deputy who will succeed Mr. Smale in April. The new deputy e.-in-c. is C. J. V. Lawson, M.I.E.E., aged 50, who since 1954 has been manager of the cable station and head of the school of telecommunication engineering at Porthcurno, near Land's End. He has had many years' service overseas since joining the Eastern Telegraph Company in 1923, and has been in charge of installation work at a number of wireless stations. Mr. Smale has been with C. \& W. since 1929, having previously been on the staff of the Marconi W.T. Company for ten years. Last year he became parttime chairman of the Cyprus Inland Telecommunications Authority.
Five appointments are announced by English Electric Valve Company: E. Allard, assistant to the general manager; J. Dain, chief of microwave research; and R. G. Roach, Dr. F. C. Thompson and W. E. Turk as managers respectively of the valve, radar tube and photoelectric divisions. E. Allard, B.Sc., A.M.I.E.E., who joined the company a year ago, was with Edison Swan during the war as a valve development engineer and later engineer in charge of a shadow factory. In 1946 he joined Plessey and was with them as chief of physics research until 1951 when he went to the Ministry of Supply as engineer in charge of valve production. J. Dain, M.A., a Wrangler of St. John's College, Cambridge, joined the Telecommunications Research Establishment in 1942 working on pulse modulators. In 1946 he transferred to the Atomic Energy Research Establishment. Mr. Dain joined the English Electric Valve Company in 1954 where he has been engaged in research on, and design of, special types of microwave tube. R. G. Roach, B.Sc., joined the company in February, 1951, and has worked on the design and production of large transmitting valves. He was formerly in the valve division of Standard Telephones and Cables, which he joined in 1937. F. C. Thompson, Ph.D., A.M.I.E.E., has been with the company for seventeen years. During the war he was seconded to T.R.E. where he worked on airborne radar. W. E. Turk, B.Sc., A.M.I.E.E., joined the company in February, 1953, having previously been in the E.M.I. Research Laboratories where he was concerned with the design and development of photoelectric devices including television pick-up tubes.

Dr. D. Gabor, F.R.S., Mullard Reader in Electronics at Imperial College, has been elected a member of the council of the Physical Society. Dr. Gabor has been working for some time on the development of a flat, thin, neckless television tube, on which he will deliver a lecture at the meeting of the Television Society on October 25th.
F. R. W. Strafford, M.I.E.E., has resigned from Belling and Lee, where he has served for many years as technical manager, to become a consulting radio and electronics engineer on his own account. He is well known to Wireless World readers for his many contributions on aerials and interference suppression and has been intimately associated with the technical side of the industry for 33 years. His work has included domestic radio, telecommunications, radar and general applications of electronics. While at Belling-Lee, Mr. Strafford originated G9AED, the pilot transmitter for the I.T.A. stations.

F. R. W. STRAFFORD

B. H. DOUTHWAITE
B. H. Douthwaite, A.M.I.E.E., rejoins Belling and Lee on October 1st as administrative head of the research and development activities of the company. He was previously with Belling-Lee from 1945-47 and was subsequently a divisional manager at the research laboratories of Elliott Bros. (London), Limited. "During the war, as a member of the research staff of G.E.C., he was engaged on radio counter-measures.
G. A. Briggs, managing director of Wharfedale Wireless Works, Limited, is to address a New York meeting of the Audio Engineering Society on sound reproduction in concert halls on September 28th. He will also be giving a lecture-demonstration in Carnegie Hall, New York, on October 3rd for which he will be joined by H. J. Leak.
A. Whitaker, O.B.E., M.A., M.I.E.E., F.Inst.P., has been appointed a director on the board of SiemensEdiswan, Limited. He is director of engineering, covering the Edison Swan interests of the company; his headquarters are at Cosmos. Works, Brimsdown, Middlesex. Mr. Whitaker joined the Gramophone Company in 1927 to initiate a research department and subsequently took charge of product engineering for domestic appliances and cathode-ray tubes. Since leaving the Gramophone Company he has held various posts in industry.
B. G. H. Rowley, M.A.(Oxon), A.M.I.E.E., Marconi's new assistant commercial manager, joined the company in 1950 as their technical representative in the United States, returning to this country at the end of 1954 to become manager of the maritime division. Early in the war he was engaged on radar development at the Royal Naval Signal School. In 1942 he was attached to the United States Navy for radar liaison duties in the North Atlantic, and was later appointed to the staff of the British Admiralty Delegation in Washington.
J. P. Wykes, A.M.I.E.E., formerly works manager at Marconi's, Chelmsford, has been appointed manager of the company's maritime division. He joined the Marconi International Marine Communication Company as a sea-going radio officer in 1918, and subsequently served in various capacities ashore. In 1934 he transferred to the research and development staff of Marconi's W.T. Company and was engaged on the design and development of marine equipment. He became manager of the crystal department in 1942 and four years later was appointed assistant engineer-in-chief in charge of test at the Chelmsford works, of which he became manager in 1949.

## OUR AUTHORS

Harold J. Leak, who, with A. B. Sarkar, writes on electrostatic loudspeaker design in this issue, needs little introduction to audio engineers. He founded the company bearing his name in 1934. For five years prior to 1934 he worked as an installation engineer of motion picture equipment. Mr. Leak is leaving London on September 25th on a seven-week tour of countries in the Western Hemisphere in which his company has agents. Whilst in Brazil he plans to conclude an agreement with the Feigenson Company, of Sao Paulo, under which it will manufecture Leak products under licence.
A. B. Sarkar, co-author of the article on page 486 , joined H. J. Leak and Company for research on sound reproduction problems about a year ago, prior to which he was with Standard Telephones and Cables. Mr. Sarkar, who is 28 , received an M.Sc. degree from London University for a thesis on measurement of acoustic impedance which he wrote following research in the physics department of Chelsea Polytechnic.
-J. R. Humphreys, who discusses on page 481 the question of reduced band-width for mobile radio, has been with the Pye organization since 1948 . He has been with Pye Telecommunications, of which he was appointed chief engineer in January, since 1950 . For the past three years he has worked on the design and development of v.h.f. mobile equipment. Mr. Humphreys entered the radio industry in 1938 at the age of sixteen as a junior technical assistant at Marconi's, Chelmsford. From 1946 until he joined Pye he was senior design engineer with Denco, of Clacton.

Lawrence W. Johnson, author of the article on f.m. receiver design, is a development engineer in the laboratory of the Hewlett-Packard Company, Inc., of Palo Alto, California, which he joined earlier this year, and is at present concerned with the development of oscilloscopes. He obtained a B.S. degree in physics and an M.S. degree in electronic engineering from the Carnegie Institute of Technology, Pittsburgh. From 1948 until 1953 he was sesearch assistant at the Institute's Nuclear Research Centre, where he was concerned with instrumentation for nuclear research, including the design of the $\mathrm{f} . \mathrm{m}$. oscillator for a $400-\mathrm{meV}$ syncrocyclotron.

## OBITUARY

George M. Wright, C.B.E., B.Eng., M.I.E.E., engineer-in-chief of Marconi's Wireless Telegraph Company until his retirement in 1954, died on August 26th in his sixty-sixth year. After obtaining his B.Eng. degree at Sheffield University he joined the Marconi Company's research department in 1912 where he assisted C. S. Franklin, pioneer of the beam system, and Captain H. J. Round, of valve fame. After service in the first World War, when he was closely associated with the establishment of the naval d.f. network, he returned to the company's research department, of which he subsequently became head. During the last war he was seconded to the Admiralty and became chief scientist at the Admiralty Research Establishment. Mr. Wright returned to

Marconi's as engineer-in-chief in 1946. He was a member of the Radio Research Board of D.S.I.R. from 1948 to 1950.
W. A. Ferguson, B.Sc.(Eng.), A.C.G.I., Grad.I.E.E., author of the articles on the design of a 20 -watt highquality amplifier which appeared in our May and June issues last year, died early in August at the age of thirty-two. He had been in Mullard's valve measurement and application laboratory since 1949 where he was initially concerned with the measurement of power output and distortion but latterly had concentrated on amplifier design. During the latter part of the war he was in the Royal Navy, having previously been at the National Physical Laboratory's Radio Division, Slough, from 1942 to 1944.
W. MacLanachan, who was well known in the radio industry as a technical writer, died at the end of August. "Mac," who was sixty, was for some years technical adviser to The Observer on radio and television matters. His book "Fighter Pilot" described his experiences in the first World War. In the last war he served for a time on the staff of Combined Operations Headquarters.

## IN BRIEF

Broadcast receiving licences current in the United Kingdom at the end of July totalled $14,361,465$, including $5,979,510$ for television and 307,294 for car radio sets.

Television licences increased by 57,490 during the month.

Audio Fairs, Limited, has been formed as a non-profit-making company "to assume responsibility for future Audio Fairs in Great Britain." The original organizing committee, consisting of representatives of audio manufacturers, "wlll continue to serve as members of this company on a purely voluntary basis." Enquiries regarding the 1957 Audio Fair (Waldorf Hotel, April 12th-15th) should be sent to the company at 21, Old Buildings, Líncoln's Inn, London, W.C.2.

Restoring "Top."-The filters recently installed by the B.B.C. at the Wrotham v.h.f. station to restrict upper modulation frequencies to $10 \mathrm{kc} / \mathrm{s}$ have now been removed. See our Editorial, August issue.

London Television Power.-We stated last month that both the Lichfield and Croydon I.T.A. stations had increased their power, whereas, in fact, the increased e.r.p. ( 120 kW ) from Croydon was not finally introduced until September 8th. Two days later the e.r.p. of the B.B.C. station at Crystal Palace was also increased to 120 kW . This was made possible by the installation of a new aerial at a height of 400 feet on the support mast.

Scottish I.T.A. Station.-Work has begun on the transmitter building at the site for the I.T.A. station àt Blackhill, Lanarkshire. The transmitting equipment, mast and aerial array will be supplied by Marconi's.

As might be expected, a television manufacturer was the first in this country to use closed-circuit television on a large scale for a national sales conference. Existing television links between London, Birmingham, Manchester, Glasgow, Bristol and Dorking were used by Pye for their conference in August.
"Too Old At-?"-In this article on age and hearing in the September issue, page 439 , the curve in Fig. 2 labelled "9" should be " 19. ."
"Cascode A.F. Amplifier."-In the first equation under Fig. 2, page 284, of the June issue, the last term on the right-hand side should be $-E_{1}$, to agree with the second equation, which is correct. The designation of V5 in the list of parts should be 5T4 (not 574). A British equivalent of this American rectifier is the Brimar 5U4.

## PUBLICATION DATE

Owing to a temporary rearrangement of our printing schedule the publication date of the November issue of Wireless World will be advanced to October 16th.

Hearing Aid Show.-Twenty-five manufacturers of hearing aids, associated equipment, components and accessories are participating in a one-day exhibition and convention organized by the Society of Hearing Aid Audiologists. It will be held on October 6th ( 10.30 a .m. to 12.30 p.m. and 2.15 to $3.15 \mathrm{p} . \mathrm{m}$.) in the Park Lane Hotel, Piccadilly, London, W.1. Details are obtainable from W. A. Cullen, 31, Highfield Avenue, Pinner, Middlesex.
Pulse Techniques.-A course of twenty-three lectures and a 12 -week laboratory course under the general heading of pulse techniques, are being conducted by the Borough Polytechnic, Borough Road, London, S.E.1. The main course is on Monday evenings (beginning October 15th) and the laboratory course on Monday afternoons or Thursday evenings (beginning October 22nd). The fees are, respectively, £2 10s and £1.

A series of short lecture-demonstrations covering various aspects of control engineering-industrial instrumentation, automatic process control and servo-mechanisms-will be given at the Battersea Polytechnic, London, S.W.11, during the new session. The Polytechnic also provides a one-year evening course on linear servomechanisms which commences on October 3rd.
Among the advanced lectures in electrical and mechanical engineering listed in a booklet issued by the Manchester and District Advisory Council for Further Education are an eight-lecture course on transistor circuit techniques and a nine-lecture course on sound recording and reproduction. Both courses will be held at the Manchester College of Science and Technology, the first on Tuesday evenings, commencing October 16th, and the second on Mondays, commencing January 7th. Fees 30s per course.

Evening courses on transistors, servomechanisms, pulse techniques, radar and computers-varying from eight to twenty-four lectures-have been arranged for the new session by the Technical College, Beaconsfield Road, Southall, Middlesex. A course on colour television will begin in January.
Dr. G. N. Patchett, head of the department of electrical engineering at the Bradford Technical College, has sent us details of part-time day and evening courses in radio and allied subjects provided at the college. They include courses for the Higher National Certificate and professional qualifications in electrical, electronic, radio and television engineering and courses in radio and television servicing.

Those concerned with the psycho-acoustic problems involved in teaching deaf children to talk will be interested in a speech training hearing-aid amplifier introduced by Amplivox. The unit is a three-stage amplifier embodying volume compression.

## FROM ABROAD

Audio in the U.S.-The eighth annual convention of the Audio Engineering Society, which now has a membership of nearly 2,000 , will be held in New York from September 26th to 29th in conjunction with the High Fidelity Show (September 27th to 30th) sponsored by the Institute of High Fidelity Manufacturers. Both functions will be held in the New York Trade Show Building. It is understood that the Audio Fair fixed for October 4th to 7th in the Hotel New Yorker has been cancelled.

Broadcasting in India.-India's second five-year plan provides for expenditure of Rs. 90 M on broadcasting, including the installation of four new $100-\mathrm{kW}$ transmitters at Delhi and $50-\mathrm{kW}$ transmitters at Bombay, Calcutta and Madras. Provision is also made for spending Rs. 4 M on an experimental television service.

South African Broadcasting.-Later this, year the new short-wave broadcasting centre of the South African Broadcasting Corporation at Paradys, near Bloemfontein, in the Orange Free State, will be officially brought into service. Marconi's have supplied nine $20-\mathrm{kW}$ transmitters for the centre which is already in partial operation.

A sixty-foot diameter paraboloid, accurate to about 3/16ths of an inch, has been built by Bell Telephone Laboratories at Holmdel, New Jersey, for the study of scatter propagation. The $5 \frac{1}{2}$-ton paraboloid is intended for use at 460 and $4,000 \mathrm{Mc} / \mathrm{s}$ at which it has a gain of 37 and 54.6 dB respectively. It has also been tested at $9,400 \mathrm{Mc} / \mathrm{s}$ at which the gain was 61.1 dB .

## BUSINESS NOTES

The Ministry of Supply has ordered from Marconi's a considerable quantity of equipment for ionospheric scatter transmission and reception. Some of the equipment, which is designed to operate in the $35-55 \mathrm{Mc} / \mathrm{s}$ band, will be used to establish a communications system between the United Kingdom and Malta, which will eventually be extended to Cyprus and the Middle East. Marconi's are undertaking the complete installation of the first section of the system. The company has also supplied a transmitter to the Admiralty. This has been set up in Gibraltar for experimental work on ionospheric scatter between the Rock and this country.

Anglo-American Company.-Ketay Limited has been formed jointly by the Plessey Company, of Ilford, and Norden-Ketay Corporation, of New York, to manufacture data transmission units. These include synchros, servo motors, computing synchros and tachometer generators. The new company's offices are at Eddes House, Eastern Avenue, West Romford, Essex.

Elliott-Swartwout Agreement.-Elliott Brothers (London), Limited, have concluded a licence and technical agreement with the Swartwout Company, of Cleveland, Ohio, by which they become the manufacturing licensees and sole agents in the British Commonwealth (except Canada) and Europe for the complete Swartwout range of electronic control equipment.
G.E.C. Laboratories.-Three laboratories of the General Electric Company, Limited, at Stanmore (Middlesex), Allesley (Coventry) and Salisbury (South Australia) have been renamed under a group title of Applied Electronics Laboratories. This change does not affect the Research Laboratories at Wembley.

Alma Components. Limited, was recently formed with premises at 165, Ossulston Street, Euston, London, N.W. 1 (Tel.: Euston 2977) to manufacture precision wirewound fixed resistors. The maximum resistances in each of the three wattage ratings ( $\frac{1}{4}, \frac{1}{2}$ and 1 watt) are respectively $600 \mathrm{k} \Omega, 800 \mathrm{k} \Omega$ and $2 \mathrm{M} \Omega$.

A feature of the radio equipment installed by Marconi's in the new 20.527-ton troopship Nevasa is the sound amplifying system. A total power of 420 watts is available to feed 214 loudspeakers. An aerial distribution system enabling 75 receiver points to be fed from a single aerial has also been fitted.

The laboratories and production departments of Fortiphone, Limited, manufacturers of hearing aids and miniature components, are now at the comoany's new head office at 92, Middlesex Street, London, E.1. (Tel.: Bishopsgate 0871.) A West End showroom and retail sales office for hearing aids is heing maintained at 247, Regent Street, London, W.1. (Tel.: Langham 3773.)

Craven Electronic Instrument Company, of Bradford, will in future be known as Craven Electronics, Limited
G. V. Planer, Limited, consultants with laboratories for rescarch, development and experimental production of electronic components and materials, have moved to new premises at Windmill Road, Sunbury-on-Thames, Middlesex. (Tel.: Sunbury-on-Thames 2266.)

Multicore Solders (Australia) Pty., Ltd., formed just over a year ago as a sales organization, has acquired a factory at Alexandria, Sydney, where Ersin Multicore solder will be manufactured.

Ke.vin and Hughes (Marine), Limited, have opened a depot at 4, Central Road, Docks, Southampton.

The head office, sales and service departments of Runbaken Electrical Products are now at 45, Oxford Road, Manchester 1.

Standard Telephones and Cables have established a regional office at Coronation House, 69/71, Market Street, Manchester 1.

## EXPORT NEWS

A mobile display of Garrard record changers an players has been touring the Continent giving demonstrations at various centres including exhibitions in Brussels, Leipzig, Vienna and Strasbourg.

Thirty-four member firms of the Scientific Instrument Manufacturers' Association participated in a display of instruments at the St. Eriks Fair held in Stockholm from September 1st to 16th.

Television Receivers.-Seng Guan Hong Co., of 854/62 Talad Noi, New Road, Bangkok, are interested in securing the agency of a United Kingdom manufacturer of television receivers. Thailand has adopted the C.C.I.R. 625 -line standard.

Tape Recorders.-A report on trade in Cyrenaica, Libya, states that there is a market for a small British tape recorder. The cheapest British equipment on sale at present costs about $£ 70$ c.i.f., whereas Italian tape recorders can be obtained for £27 c.i.f.

Record Changers.-K. F. Moseley, sales director of Birmingham Sound Reproducers, makers of the Monarch record changer, has been on a tour of the U.S.A. and Canada, during which he visited the new Broadway, New York, premises of the Discus Corporation (the B.S.R. sales and service office in America). Shipments to the States of Monarch record changers are now considerably in excess of the 4,500 units dispatched weekly at the beginning of this year.



Trends in Vision and Sound Broadcast Receivers - and Some Highlights

## TELEVISION RECEIVERS

IT has been evident for some years that the trend of development in television receivers is towards a high degree of standardization between the products of different manufacturers. By this is meant a standardization of the kind of circuitry and not of mechanical form or external appearance. In this, television is merely following in the footsteps of sound broadcast receivers.
This uniformity of circuitry is not a result of the efforts of any standardization committee, but is a natural result of the fact that there is technically a best way of achieving a required performance. Sometimes there are equally good technical alternatives, of course, but then one is usually cheaper than another and the laws of economics dictate the choice.
Complete identity of circuitry is not, of course, obtained. Probably no two receivers by different makers have exactly the same circuit, but the differences are more in biasing and decoupling arrangements than in the main general form of the circuits. In one thing, however, there is virtually complete standardization, and that is in picture sizes. This has been brought about by the c.r. tube manufacturers, who have adopted the range of $9-, 12-, 14-$, $17-$ and 21 -inch sizes, the dimensions referring to the screen diagonal or diameter. Of these, only the three largest sizes are in common use, the 9-and 12 -inch tubes being normally reserved for portable apparatus. At the present time, the 17 -inch tube appears to be the most popular, but there are still plenty of 14 -inch and nearly all set makers are producing 21 -inch models.
The effect of tube size upon the circuitry is very small. It actually affects only the timebases and
e.h.t. supplies. Because of the large deflection angle and higher final-anode voltage a large tube needs more deflection power than a small one. In the main this is offset by the use of deflector coils and transformers of higher efficiency, and the net result is that there are negligible circuit differences.

It is now possible to attempt a general description of a television receiver which will apply with remarkable accuracy to the vast majority of modern sets. The general practice is to include a tuner covering Bands I and III which has a double-triode cascode r.f. amplifier and a triode-pentode frequency changer. There is a single-tuned input circuit coupling the aerial feeder to the grid of the r.f. stage, the first triode of which is neutralized by a capacitance bridge circuit. A coupled pair of circuits is used between the r.f. stage and the pentode mixer, and the oscillator is of the Colpitts form.

Four tuned circuits thus require alteration for station selection. There are three alternative methods of doing this. The usual one is the turret tuner, but the so-called incremental-inductance method with switch selection is still employed by a number of firms. Bush also retain the method by which a set of Band I and a set of Band III coils have their cores ganged to provide continuous tuning over the bands, but are provided with a camcontrolled mechanism and an automatic change-over switch, so that the user still has a multi-position selector control as in the other systems.
The standard intermediate frequencies of 34.65 $\mathrm{Mc} / \mathrm{s}$ for vision and $38.15 \mathrm{Mc} / \mathrm{s}$ for sound are now used by nearly everyone. Coupled pairs of circuits are usual for the intervalve couplings of both sound and vision amplifiers, with three stages on vision and two on sound. Some makers, however, manage with a stage less, but there is a tendency for these



Fig. 2. H.M.V. receivers have a pentode video stage $V_{1}$ with a triode black-spotter $V_{2}$.

Ekco) a Metrosil voltagestabilizing element is added to the e.h.t. circuit. Linearity control is generally by means of a small coil having a core which is controllably saturated by a small permanent magnet. Width control is usualiy by a coil having a movable ferrite core.

The output stage is driven by either a blocking oscillator or a multivibrator; in a few cases, the output stage itself forms a part of the multivibrator. Last year there seemed to be a definite
spotter or noise inverter. It is then usually a cathode-input triode connecting the video output to the tube grid. It is biased off over the video range of signals, but comes into operation on noise of greater amplitude than peak white and then provides on the grid a noise signal in the same phase as that applied in the normal way to the cathode. The two signals on grid and cathode thus tend to cancel in their effect on the tube.

Figs. 1 and 2 both show this type of black-spotter-in Fig. 1 fed from a cathode-follower and in Fig. 2 from the video stage itself. In the latter the complete video signal is fed through the contrast control to the cathode of the c.r. tube in more or less the usual way. It is also fed to the cathode of $V_{2}$ which is biased by $R_{4}$ so that it is nonconductive over the whole range of normal variations of the waveform. The signal is increasingly negative for increasing whiteness and a noise peak beyond peak white pulls down the cathode of $\mathrm{V}_{2}$ sufficiently for it to conduct. An amplified noise pulse then appears on the anode in the same phase and is applied to the tube grid to nullify or overcome the noise pulse applied in the same phase to the tube cathode. Although some of the details of the circuit are different, the black-spotter $\mathrm{V}_{3}$ of Fig. 1 operates in much the same manner.

The form of contrast control in Fig. 2 is an unusual one. The video-stage coupling resistor is $R_{2}$ and the required fraction of the voltage at this point is taken off the potentiometer $R_{1}$ to the c.r. tube. The "earthy" end of $R$, is returned to a point on the potential-divider chain $R_{2}, R_{4}$ to remove the standing voltage and so permit $R_{1}$ to be of low resistance without drawing excessive current.

Time-bases circuits have a somewhat less standardized appearance than elsewhere, but not very much. For the line scan there is invariably a pentode or beam tetrode output stage which is nearly always coupled to the deflector coils by an autotransformer. A few people use the so-called directdrive circuit; for example, Bush. The e.h.t. is a by-product of the flyback and is obtained from an over-wind on the transformer, a half-wave thermionic rectifier being used. In a few sets (e.g.,
trend towards the use of flywheel sync, although the simpler direct-locking method was still quite widely used. This year the position is much the same, but one could hardly now call it a trend towards flywheel sync; the proportions of sets using the two methods appear to be much the same. There is, in fact, a feeling in some quarters that direct locking is to be preferred when interference is not such as to render flywheel sync a necessity. Murphy, for instance, use direct locking in all their ordinary models and include flywheel sync only in the fringe area types. In the frame sync circuits, considerable minor variations still exist. In the main, however, the waveform generator is a blocking oscillator or multivibrator, with a pentode output stage. For linearity, the well-known feedback circuit is still a favourite.

On sound, a diode detector and a simple, but effective, diode ignition-interference limiter are always used. A pentode output stage, often preceded by a triode a.f. amplifier, is employed. A.G.C. on the sound channel is obtained in a conventional manner. On vision, however, a.g.c. is a good deal more complex and there is no great uniformity of methods. It is not always included at all, of course, but when it is the mean level system still has its adherents. In this the grid voltage of the sync separator is smoothed and applied via a delay diode to one or more i.f. stages and to the r.f. stage. This is the simplest arrangement, but it is one in which the mean brightness of the picture is necessarily affected by the control, for the a.g.c. voltage depends on the mean brightness as well as upon signal strength per se.

Gated a.g.c. systems, in which the signal is sampled during the back porch, are free from this effect. The level of signal during the back porch is independent of the mean brightness of the picture. Such systems are a good deal more complicated for they require the production of a gating pulse at some time during the back porch. This is usually developed from the pulse which occurs in the line timebase during line flyback, and quite often a simple RC network will provide the requisite delay. The methods used, however, vary a good deal. One is illustrated in Fig. 1. The diode $\mathrm{V}_{4}$ is normally
non-conductive but is rendered conductive during the back porch by a pulse which is derived from a tap on the output auto-transformer of the line timebasc. The capacitor $C$ then becomes charged to a value dependent upon the various voltages involved of which only the back-porch signal level is a variable; the total charge thus varies according to the black level of the signal and so to signal strength. The voltage across C is applied through the delay diode $\mathrm{V}_{5}$ to the a.g.c. line. The diode also prevents any possibility of the a.g.c. line becoming positive.

Power supplies are almost invariably of the a.c./ d.c. type with series heaters, while the h.t. is derived through a half-wave rectifier, either valve or metal. Auxiliary supplies are few, for the electromagnetic focus coil vanished some years ago. However, the permanent magnet focus unit is now showing signs of disappearing also. Some of the new c.r. tubes have electrostatic focus. The voltages needed are fairly low and can be obtained from the h.t. supply or, when this is too low, from an extra rectifier connected to a tapping on the line-scan transformer.

Electrostatic focus means that the focus control can once again be conveniently placed instead of being an inaccessible lever projecting through the back of the cabinet. However, this does not mean that it is always so fitted; there is a tendency to operate under fixed focus conditions or, if not this, to regard it as a pre-set and not a user control.

Remote control for television sets is coming somewhat to the fore. Ekco have a remote-control unit for some models which contains contrast and volume
controls. It is connected by a multi-core cable and the connector plugs into the set. Philco have a more elaborate arrangement in which a pecking motor is used to operate the turret tuner so that station selection can also be effected remotely.

It was evident last year that manufacturers were not slow to realize that the television set lends itself to Band II f.m. reception and quite a few models with provision for this were shown. This year there are, if anything, more of them. In such sets, Band II coils are fitted to the turret tuner and the sound i.f. amplifier is arranged to feed a discriminator and ratio detector. Some switching is involved in the audio circuits, but mainly in the power supply, since the tube and vision circuits are put out of action on Band II. This switching is mechanically linked to the turret.

Although the combination of Band II sound with the television receiver seems fairly popular the inclusion of a medium and long-wave receiver with it has never been common and of recent years such models have become very rare. A new one this year is produced by Pye and includes also a recordplayer and four loudspeakers.

In the bigger television sets, like this one, there is certainly a tendency to improve the sound side. The H.M.V. 1847 table model with a 21 -inch tube, for example, has two loudspeakers of which one is a high-frequency tweeter. A very unusual point is that provision is made for fitting the bass speaker on either side of the cabinet so that it may be placed on the better side from the point of view of room acoustics.

1. Pye portable with a 14 -inch tube. 2. Spencer-West portable. A 9 -inch tube is used and the controls are on the side. 3. One of the ponels in the Pam printed circuit television receiver. The upper view shows the "connection "side of the panel and the lower the component side.


Last year a new. trend in television appeared; the portable set. Ekco produced a 9 -inch tube model with a vibrator h.t. supply and Murphy a 12 -inch with a.c. supply; both are continued this year. There are now several newcomers. Pye have one with a 14 -inch tube for a.c. mains only. Spencer-West have a 9 -inch tube model.
It is characteristic of these portable models that they have no true cabinet, but light cases of purely functional character. They are, however, of extremely attractive appearance.

## TELEVISION AERIALS

A FEW new television aerials made an appearance this year but the general impression given by close examination of the exhibits in this category was that most aerial makers have been busy consolidating the position created last year. Then, the approach of commercial television brought about a spate of dual-band and Band-III aerials, all of which had had to be designed before actual transmissions commenced; Band III was unmapped territory so far as television was concerned in this country.

A year's experience has shown that very few changes in the design of these aerials has been necessary. Mostly they concern the mechanical design rather than the electrical, as, for example, the practice of assembling aerials in the factory in such a way that they can be sent out with all elements correctly assembled, but with the aerial collapsed rather like an umbrella. This simplifies the work of erection as the assembly merely consists of opening out the aerial and possibly tightening a few wing nuts. In some cases even this is not necessary. Generally a locating clip or pin, as used on the Belling-Lee collapsible Band-III aerials, is used to ensure that the elements take up their correct positions and do not become displaced under the stress of vibration after erection. The collapsible technique is not restricted to the smaller Band-III aerials, but is adopted also for some of the dual-band models made by Aerialite, Antiference, Labgear and Telerection.
There is a tendency to replace seamless, drawn
tube by rolled tube for aerial elements, it being said that the rolled variety is tougher than the drawn and less liable to fracture in high winds. Wolsey have gone over to rolled stainless steel tube for their elements, but retain aluminium for cross-arms and the supporting pole. In all cases where rolled tube is used for the elements drawn tube seems to be retained for cross-arms and supporting poles.
As rolled tube is not watertight, its adoption has led to a re-design, in some cases, of the centre insulator on the aerial in order to prevent water getting into the cable compartment via the aerial rods. This might not have any serious consequences when solid dielectric cable is fitted, apart from the general undesirableness of dampness anywhere in electrical circuits, but it can be quite disastrous with semi-air spaced cables, used extensively on Band-III aerials and often on Band-I models in fringe areas.
Of the new developments an interesting one is the Telerection "phase-corrected" dual-band aerial. The one shown had two Band-I dipoles combined with the reflector and director elements of two Band-III Yagis. The Band-I dipoles function also as the driven elements (to use a transmitting term for easier explanation) on Band III and the complete aerial behaves as a broadside array on both bands. The "delta" matching system, for which this firm is noted, is employed, and, owing to the proportioning of the number of the elements, the matching holds good on both bands.
The unique feature of this aerial is that the Band-I dipoles are broken up into three collinear half-wave Band-III sections by what appear to be insulators, but which are actually housings for small LC circuits resonating in Band III. These confine the out-of-phase r.f. currents along the long dipoles when used on Band III to these non-radiating parts of the system and each thus behaves as three
(Continued on page 473)


Belling-Lee collapsible Band-Ill aerial embodying a locating device for the elements.

Left: Telerection "Double Delta Four'" combined aerial howing phasing sections inserted in the Band-1 dipoles.
collinear half-wave dipoles operated in phase. The principle of operation is similar to that of some of the American wide-band television aerials, but in those the makers use either folded or open stubs resonating in their "high" band. The principle was explained in "Wide-Band Television Aerials" in Wireless World of June, 1956.

Without these "phase correctors," as Telerection call them, the vertical polar diagram on Band III would have two principal lobes, one above and one below the horizontal, with very poor response at the normal angles of reception. The phase correctors cause the two lobes to merge into one.

The addition of reflectors and directors for Band III reinforces the forward gain and cancels all or most of the back responses. As shown, with only two Band-I phase-corrected dipoles, the aerial is bi-directional on Band I and uni-directional on Band III, but with the addition of Band-I reflectors the array would be uni-directional on both bands.

Another new aerial of unorthodox construction was shown by Labgear. It is described as the "Bi-Square" and lends itself admirably for erection in a loft, as the space occupied is much less than that of an aerial of more orthodox construction and equivalent gain. Basically it consists of two fullwave dipoles each bent to form a square and mounted about $\lambda / 4$ apart. The front one, to which is connected the feeder ( $75 \Omega$ ), is the "live," or driven, element and the other behaves as a parasitic reflector and occupies the rear position (in relation to the station being received). In order for the rear element to function as a reflector it must be made slightly longer, physically, than the driven element and this is achieved by inserting folded stubs in the centre of each half-wave section.
With the aerial rotated so that the "insulators" (the only real insulator is where the feeder is connected) are in the vertical sides, the aerial responds to vertically polarized waves and if turned through $90^{\circ}$ so that the insulators are in top and bottom sides the polarization then becomes horizontal. As implied by the mention of a reflector, the aerial is uni-directional and this is at right-angles to the plane of the "loops." The elements may be referred to as loops as the reflector, for example, is continuous electrically, while the only discontinuity in the "live" element is at the feeder connection.

The "Bi-Square" is for use on Band I and is said to give a gain of 12 dB compared with a plain dipole. For Channel I the overall size is approximately 5 ft square and 4 ft from back tc front, so the earlier statement that it is well suited for loft mounting is quite justified. For other Band-I channels the aerial will be smaller; a separate model is supplied for each. It is claimed that when erected in a loft, and despite the reduction in efficiency usually encountered under these conditions, its performance is comparable to that of a 3 - or 4 -element outdoor aerial and good reception can generally be provided out to 50 or 60 miles from a high-power television station. The aerial can be used, of course, out of doors, when it should be equivalent to a 7- or 8 -element Yagi, or to two 3- or 4 -element Yagis in broadside, so far as gain is concerned.

A novel dual-band aerial, also of unorthodox design, is the new J-Beam "Hornbeam." The Band-III portion, which is called a "skeleton horn," consists of a square frame erected vertically with a horizontal " $V$ " part behind it, the open ends of


Labgear "Bi-Square" aerial, note the tuning stubs in the reflector element.


Two new Aerialite coaxial cables: (a) with catenary suspension wire (b) with double screening.
the " $V$ " being joined to the centres of the top and bottom sides of the, square. All parts, such as the sides of the square and the arms of the " $V$," are very approximately $\lambda / 2$ long on Band III.

Two vertical elements are joined to the top and bottom sides of the square at the junctions of the ends of the "V." These vertical rods are telescopic and have to be adjusted in length for each Band-I channel. The square plus "V" comprise the Band-III aerial and the vertical rods, in conjunction with the two halves of the square in parallel, form a plain dipole on Band I. In this form the aerial is known as the "Hornbeam I."
The precise method of operation of this aerial on Band III is not very clear, but it appears to behave as two inclined vertical collinear $\lambda / 2$ dipoles (the " $V$ " portion) with the vertical sides of the square reinforcing the forward response by behaving as two vertical $\lambda / 2$ dipoles connected in phase by the horizontal sides. This assumes it is orientated for vertical polarization as illustrated in last month's Wireless World (page 419).

Now a 4-element Yagi gives a gain of about 9 dB and the 7 to 9 dB claimed for the "Hornbeam" on Band III agrees with what might be expected from an aerial having four effective $\lambda / 2$ elements. A Band-I reflector can be added thereby making the aerial uni-directional on both bands; in this form it is described as the "Hornbeam II."

Apart from its unique construction the "Hornbeam" is characterized by having a very wide bandwidth; it is said to cover 170 to $230 \mathrm{Mc} / \mathrm{s}$ giving a


Antiference one-piece coaxial plug showing method of fitting the cable.
gain of 7 dB at the lower frequency and 9 dB at the higher.

Only infrequently does anything interesting in the shape of coaxial cables appear at the Show. This year Aerialite had two new types which have been produced primarily for use in television relay installations, in blocks of flats wired for television and wherever television is "piped" to a number of receiving points from a central position. One new cable takes the form of a double-screened version of the Super Aeraxial, the additional copper-braid sheath being separated by polythene insulation from the normal outer conductor.

The other cable, which has characteristics similar to the Standard Aeraxial, includes a 0.048 -in (No. 18 s.w.g.) galvanized steel catenary wire embedded in the outer PVC covering. This cable is intended for use where long overhead runs are required, or wherever frequent anchorage points cannot be provided. It is a single-screened variety and, in common with the double-screened cable, has a nominal impedance of $73 \Omega$.

Coaxial plugs and sockets used on television receivers, v.h.f. sets, audio equipment at times and some test gear are now more or less standardized and few changes are encountered as a rule. This year, however, Antiference has introduced a new one-piece cable plug for which certain advantages are claimed. There are no loose parts to get lost; it has polystyrene insulation and is completely insulated by a Neoprene sleeve. Assembling (or loading, as it is often called) the cable in the plug merely consists of stripping back sufficient inner and outer insulation on the cable; threading the centre conductor through the hollow pin on the plug, and tip soldering it, then squeezing the skirt of the plug on to the outer braid of the cable. Ordinary flatnosed pliers can be used.

## SOUND RECEIVERS AND REPRODUCERS

UNDOUBTEDLY the predominant development in sound receivers is the establishment of the v.h.f. range as an essential alternative to the short-, medium- and long-wave bands. The success of the early transmissions from Wrotham, the now rapid expansion of the v.h.f. service to the provinces and the renaissance of public interest in sound broadcasting freed from interference and improved in quality have convinced the industry of the wisdom of the B.B.C's policy. Two years ago some, last year many and this year all domestic sound receivers are equipped for reception on Band II.

No significant developments in circuitry are
evident; the neutralized double-triode mixeroscillator and the ratio type discriminator continue to provide the economic answer for most manufacturers. The Foster-Seeley discriminator is to be found in high-quality tuner units and a fresh example was noted in the Whiteley Electrical tuner.

The improved quality of reproduction on v.h.f. is due primarily to the removal of restrictions on the bandwidth of the i.f. amplifier which must of necessity be narrow to exclude adjacent-channel interference on medium waves. To exploit the widened frequency response permissible on Band II several manufacturers have revised their designs to provide better sound reproduction. In the Murphy A272C, for example, a return has been made to large baffle mounting for a loudspeaker of better quality. The tuning scale, too, implies that the v.h.f. is the principal waveband. Ekco have gone the whole hog and produced three sets, for v.h.f. only, in which a combination of two speakers is used to improve quality.

Multiple speakers are also being more widely used for omni-directional (3D) sound distribution in the manner popularized by Continental manufacturers, and the Continental influence is to be seen in many cabinets with dark polished finish and gilt ornamentation.

Two portable receivers made by the printed-circuit technique and employing transistors in all stages were shown. The Pam Model 710 which employs eight transistors (oscillator, mixer, two i.f., detector, a.f. amplifier and push-pull output) has an intermediate frequency of $315 \mathrm{kc} / \mathrm{s}$. In the Cossor set there are six transistors, the first is an oscillatormixer and is followed by a single i.f stage ( $460 \mathrm{kc} / \mathrm{s}$ ), detector, two drivers and the customary push-pull output stage.

Transistors have virtually ousted valves in small portable record reproducers, which are now available with a battery-driven turntable for $45-\mathrm{r} . \mathrm{p} . \mathrm{m}$. records. Four U2 dry cells ( 6 volts) serve to run the motor and energize the transistor amplifier. Examples of this new trend were to be seen on the stands of Cossor, Philco, Philips, Pye, Roberts and Vidor. In the Philips AG2130 and the Philco Model 3755 three-speed battery-driven turntables are provided.

The transistor has also invaded the car radio field and is used in the latest Pye receiver not only for the output stage but also as a switching device in place of the more usual mechanical vibrator for h.t. generation.

The widely held opinion that transistors, while being all very well for hearing aids and small portables, are unlikely for a long time to come to displace valves in high-quality power amplifiers may have to be revised sooner than most people think. Last year we saw their use in a low-distortion tape preamplifier (Reflectograph) and this year Mullard demonstrated an experimental high-quality amplifier with less than $0.4 \%$ total distortion at a rated maximum output of 4 watts (at 1 watt the distortion is $0.13 \%$ a $1000 \mathrm{c} / \mathrm{s}$ ). Push-pull OC16 transistors are used in the output stage and there are two feedback loops, the inner from the OCl 6 collectors to the bases of the driver transistors and the outer from the output transformer secondary to the base of the input stage. Stability problems with large degrees of feedback are eased by the fact that direct couplings can be used between the driver and output stages, and


Right: Murphy A272C "baffle"
 type console.

Right: Connections of transformerless output stage in the Philips AG2126 "Magic Box" record player (below).

Cossor portable receiver using transistors throughout in a printed circuit.

Philco "Transistor" 3 speed portable record player.


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by the use of an OC45 to give a wide-band, highgain input stage. Heater hum problems are nonexistent and supply voltages are low.

Although an output transformer is used in the Mullard amplifier it is often dispensed with in transistor push-pull output stages, which then feed a centre-tapped high-impedance moving-coil loudspeaker directly. A move in another direction to eliminate the output transformer was noted in the Philips AG2126 "Magic Box" record reproducer. This is for a.c. operation and uses valves throughout. The interesting point is that the two output valves are operated in series with feedback applied in the upper valve. The result is a useful increase in the current gain of the stage as a whole and an equivalent output impedence low enough to feed a highimpedence moving çoil directly.

Electrostatic loudspeakers are still subsidiary to the moving coil as far as most receivers and reproducers are concerned. Two are used in conjunction with a 12 -in moving coil in the McMichael 1555 radiogramophone and Pye have added an electrostatic unit to extend the frequency range of their "Super Black Box " record reproducer. An interesting combination of the moving-coil and electrostatic principles is under development by Whiteley Electrical who showed one of their 12-in "Duplex" models in which three of the mid-frequency stabilizing patches in the main diaphragm. were replaced by small circular electrostatic "wafers" connected in parallel.

An unexpected development by Goodmans is a return to pressure units and horn loading for middle as well as high frequencies, and a three-unit system employing a large direct-radiating "woofer" for the bass showed itself capable of handling the high power outputs which are now called for by many American buyers of "hi-fi" equipment.

Tape recording activities are centred chiefly on the reproduction of "Stereosonic" twin-channel records and H.M.V. can now supply their tape deck (Model 3035), with equalized amplifiers and a con-
trol unit for incorporation in existing high-quality installations.
G.E.C. have gone into production with a complete stereophonic sound reproducing system incorporating standard G.E.C quality amplifiers and metal-cone loudspeaker units and using a Truvox Mark IV tape deck with a special dual playback head.

A disc-type magnetic recorder has been developed by Pye for home entertainment. The record/playback head is mounted in a tone arm, and the discs, which are impregnated with magnetic oxide, have an unmodulated spiral groove impressed on both sides which serves to track the head. The turntable has four speeds and playing times up to 12 minutes per side at $16 \frac{2}{3}$ r.p.m. are obtained. Recording is made with a.c. bias and erasure by separate magnetic "wiper" blade. Accessories include a two-station radio tuner unit, and normal gramophone pickup heads can be purchased separately to convert the instrument to a record player. The amplifier has an output of 4 watts and enough gain for use with a microphone.

- Normal disc reproducing techniques, as exempli-


Goodmans "Trebax" hornloaded high-frequency unit.

R.C.A. eight-pole "New Orthophonic" pickup.

fied at the Show, have not changed fundamentally, but the pickup marketed by R.C.A. (Great Britain) Ltd. is of interest. It serves to underline the dictum that method of execution is often more important than principle of operation in electro-acoustic transducers. The moving-iron (variable reluctance) principle has had more detractors than most, yet it was shown many years ago that distortion could be reduced to negligible proportions by the use of wide air gaps and proper siting of the pickup coil. In the R.C.A. design no fewer than eight poles are used and the armature has a $90^{\circ}$ twist above and below the pickup coil. This results in what might be termed double differential cancellation of any residual nonlinearities near the neutral positions of the two halves of the armature.

## VALVES AND SĖMICONDUCTORS

JUNCTION transistors, at one time only available for low-power amplification at audio frequencies, are now branching out in two directions-towards higher powers and higher frequencies. This puts us in the pleasant position of having several categories of transistors from which to choose. The groups are fairly well defined by the limiting values of the transistors concerned, and they correspond roughly to the various departments of a sound broadcast receiver.
First of all there are the low-power, low-frequency types primarily intended for audio input and driver stages. They have collector dissipations in the region of $25-50 \mathrm{~mW}$ and alpha cut-off frequencies of about $500-800 \mathrm{kc} / \mathrm{s}$. Some of them have rather higher cut-off frequencies, of the order of $1-2 \mathrm{Mc} / \mathrm{s}$, and these, particularly, can be used for i.f. amplifier stages working at $200-500 \mathrm{kc} / \mathrm{s}$ when their internal feedback is neutralized by suitable circuit design. A wide range of types by various manufacturers is already on the market.

Going up in frequency brings us to the next cate-gory-r.f. transistors. These are low-power devices (collector dissipations of $10-25 \mathrm{~mW}$ ) with cut-off frequencies in the region of $3-10 \mathrm{Mc} / \mathrm{s}$. They can be used as r.f. amplifiers, oscillators and frequency changers for the medium- and long-wave bands and also, of course, as i.f. amplifiers. At the moment they are not readily available on the general market, but should be in quantity production at the beginning of next year.
In roughly the same stage of development are the transistors in the high-power category. These are mainly intended for audio output stages and are often used in Class B push-pull circuits. The smaller types, which are easily obtainable, have collector dissipations of about 50 mW and in push-pull circuits will give outputs of approximately 200 mW . The larger ones, which are still to some extent experimental, will dissipate $2-8$ watts when suitably cooled and provide outputs (push-pull) in the range $3-10$ watts.

At the Show a complete range of transistors representing each of these categories was offered by Ediswan, who have just entered the field. Their XA101, for example, with a cut-off frequency of $4.5 \mathrm{Mc} / \mathrm{s}$, is suitable for i.f. amplification at $250-$ $500 \mathrm{kc} / \mathrm{s}$, while the XA102, with $7 \mathrm{Mc} / \mathrm{s}$ cut-off, can be used as a frequency changer or local oscillator. Types XB102 and XB103 are for a.f. amplifier or driver stages, while the $\mathrm{XC101}$ comes in the small-size power category ( 220 mW push-pull output). All of


Mullard u.h.f. transmitting double tetrode QQVO2-6.

Ediswan transistor of metal construction.
them are hermetically sealed in welded metal cases.
Pye Industrial Electronics also have a complete range, with three audio and i.f. types, prefixed V10; three r.f. types, prefixed V6, having cut-off frequencies between 3 and $10 \mathrm{Mc} / \mathrm{s}$; and six large-size power types, prefixed V15 and V30, with collector dissipations of up to 4 watts.

One of the earliest power transistors, the Mullard OC15; has now been superseded by a new experimental type, the OC16. This operates with collector voltages up to 16 V and has a maximum dissipation of 8 watts at $25^{\circ} \mathrm{C}$. Another new experimental type from the same firm is the OC45 r.f. transistor, which is of all-glass construction and has a cut-off frequency of $6 \mathrm{Mc} / \mathrm{s}$ The equivalent-circuit collector capacitance is 12.5 pF .

The range of G.E.C. junction transistors hitherto known by the prefix EW are now being changed to GET type numbers. The GET3, GET4 and GET6 are all primarily intended for low-power a.f. and i.f. applications, but the GET6 is notable for its low-noise performance, while the GET4 characteristics are such that it can be used for Class B pushpull output stages of up to 200 mW . A transistor specifically designed for output stages is the GET5. This is similar internally to the GET4 (both have $f_{\alpha}$ at about $1.4 \mathrm{Mc} / \mathrm{s}$ ) but has a nickel-plated copper mounting arrangement designed to give low thermal resistance (see sketch), which allows the collector dissipation to be as high as 200 mW at $45^{\circ} \mathrm{C}$. Moreover, the variation of current gain factor with emitter current is quite small. In a push-pull output stage


Electrode structure of Brimar 6AM4. The grid (normally earthed) is connected to five pins by flat metal sheets.
working from 20 V a pair of GET5s will give 800 mW .

As for the rest of the semi-conductor field, the more recent developments were represented as follows: avalanche transistors for high-speed pulse circuits (Mullard); silicon junction diodes for high temperature operation (Ferranti, G.E.C. and S.T.C.); silicon junction power rectifiers (Ferranti); germanium junction photocells (S.T.C.); and photo-transistors (Mullard). Two new silicon junction diodes, SX641 and SX642, for operation up to $150^{\circ} \mathrm{C}$ were shown by G.E.C. At $100^{\circ} \mathrm{C}$ they have reverse currents as low as $5 \mu \mathrm{~A}$ with peak inverse voltages of 60 and 120 respectively.

Amongst metal rectifiers the most interesting exhibit was a rectifier/stabilizer for the $1.4-\mathrm{V}$ valve filament supplies of mains/battery portable receivers. It consists of two small selenium rectifiers mounted on the same insulated spindle. The first is used for obtaining the d.c. low-tension voltage from the mains transformer and comprises two units which can be connected for either half- or full-wave rectification. The second rectifier, also in two sections but joined permanently in series, is shunted across the valve filaments, and it acts as a voltage stabilizer-with the aid of a series resistor-by virtue of the non-linear voltage/current properties of the forward characteristic.

There were no really outstanding developments in valves to be seen this year, except perhaps the introduction by G.E.C. of a "big brother" for the well-known KT66 audio output pentode. Known as KT88, it has the increased anode dissipation of 35 W (as against 25 W ), a slope of $11 \mathrm{~mA} / \mathrm{V}$ (compared with $6.3 \mathrm{~mA} / \mathrm{V}$ ), and in a push-pull stage with fixed bias will give twice the output power of an equivalent KT66 stage.
If a trend in valves is to be detected at all, it is perhaps the gradual introduction of u.h.f. types for operation in Bands IV and V. This is made necessary by the possibility of colour television and mobile radio moving into these bands. The rather surprising thing about the new valves is that they look like ordinary miniature receiving types on B7G and B9A bases, but in fact their internal electrode structures are somewhat different, being mostly constructed on
S.T.C. rectifier'stabilizer for mains battery portables.
the planar-electrode principle. In a sense they represent a half-way house between conventional electrode structures and the expensive disc-seal techniques.

Examples at the Show were the G.E.C. type A2521 low-noise receiver input valve, the Brimar 6AF4A oscillator triode, the Brimar 6AM4 r.f. amplifier or mixer triode, and the Mullard QQVO2-6 poweramplifier double tetrode. The last-mentioned, which is intended for mobile radio transmitters, uses a new type of control grid to give the high slope (for u.h.f.) of $7 \mathrm{~mA} / \mathrm{V}$, and in a power amplifier circuit working at $490 \mathrm{Mc} / \mathrm{s}$ will deliver 3.5 watts to the aerial.

The trend in television cathode-ray tubes towards bigger and bigger screens has now largely stabilized at the 21 -inch rectangular type with a $90^{\circ}$ deflection angle. Ediswan, however, were showing a 24 -inch rectangular tube. Will this be the next stage-or are we beginning to stretch our 405 lines beyond the limit?

## OTHER EXHIBITS

IN addition to ERNIE, the random number generator described in our last issue, the G.P.O. demonstrated several other unusual electronic devices. Perhaps the most impressive was an electronic letter-sorting machine, which enables a single operator to divide mail into 144 groups at one sorting as against the 48 groups in ordinary hand sorting. The letters are actually distributed by a "conveyor belt" system made up of small rollers with receiving compartments underneath, and it is the function of the electronic circuits to select and open the lids of the compartments for the travelling letters to fall into.
First, the letters are automatically passed in front of the operator, who observes the town of destination and then, in accordance with a memorized code, presses two buttons simultaneously. There are 12 buttons for each hand, so that a total of 144 combinations is available. The buttons switch on voltages, which are applied to a $12 \times 12$ matrix of 144 cold-cathode tubes, so that for each combination a particular tube is triggered. The output signal from the tube is then used to actuate electromechanical control gear, which opens the appropriate receiving compartment after a delay to allow for the travelling of the letter.
The process of selection by electronic means was also to be seen in a demonstration of electronic switching for automatic telephone exchanges. Here the method of communication between any two subscribers, over a circuit common to many subscribers, was based on the time-division multiplex principle. The connection of any two subscribers is achieved when coincident pulses are applied to germanium rectifiers in each of their circuits, and at these instants speech currents held as charges in capacitors are transferred from one end to the other. The switching pulses themselves are prevented from
coming on the line by arrangements of transformers. Ringing is achieved by transmitting a $2-\mathrm{kc} / \mathrm{s}$ tone broken at $17-\mathrm{c} / \mathrm{s}$ intervals. At the receiving end this is amplified by a transistor, which allows a capacitor to charge and so produce a $17-\mathrm{c} / \mathrm{s}$ pulsating voltage for operating the bell.

Precise timing was also the essence of another demonstration, in the "Careers in Electronics" section, showing the electronic method used for calculating the speed of the Fairey Delta II aircraft on its record-breaking $1,132-\mathrm{m}$.p.h. run. Special cameras were set up at each end of a 9 -mile course. When the aircraft came in the field of view of the first camera an operator exposed the film, and the shutter mechanism started an electronic timer count-


Display version of G.P.O. electronic random number generator for providing two digits.
ing pulses from a crystal-controlled source. At the far end of the course the second camera was operated as the aircraft came in view and here the shutter was used to stop the timer. Thus, from the number of pulses counted (representing a period of time), the known length of the course, and the position of the aircraft relative to the graticule of each camera when actually photographed, it was possible to calculate the speed. An overall accuracy of 1 in 2 or 3 thousand was said to be obtained.

An unusual type of display seen on a cathode-ray tube on the Mullard stand was the spot tracing out a word in the form of handwriting. The tube was a 12 -inch p.p.i. radar type with a long-persistence screen. The necessary $X$ and $Y$ deflection wave-


Pye transistorized loud-hailer with trigger switch.

T.C.C. printed-circuit wafer switches with flush contacts to reduce wear and prevent contoct bounce.


Left: special camera used for electronic timing of Fairey Delta II aircraft. Right: Midget receivers being soldered with Multicore ".Savbit" alloy containing copper to reduce bit wear.

forms for doing this were obtained from photocells scanning rotating masks, which were shaped to represent graphs of the waveforms in poiar coordinates. By adjusting the X and Y gain controls it was possible to alter the character of the handwriting.

As one might expect, transistors were to be seen in a good deal of apparatus, especially where small size, light weight and economical operation were the desirable features. A good example was the Pye transistorized loud-hailer, which incorporates a folded-horn loudspeaker, a 3-watt transistor amplifier, an electromagnetic microphone and batteries, all in a single hand unit weighing only 5lb. The speaking range is up to 500 yards, while the average consumption from the $12-\mathrm{V}$ battery is 120 mA , giving a battery life of about 20,000 ten-second operations, or up to about six months on normal use. The amplifier consists of a Class A driver transistor in a common-emitter circuit coupled by a phase-splitting transformer to a pair of 3 -W dissipation power transistors in a Class B push-pull stage, which drives the $15-\Omega$ loudspeaker through an output transformer. The frequency response is flat within 3 dB from $200 \mathrm{c} / \mathrm{s}$ to $4 \mathrm{kc} / \mathrm{s}$.

An example of how transistors can be used to advantage in test gear was to be seen on the Labgear stand in the shape of a portable crystal-controlled r.f. oscillator not much bigger than a matchbox. It is intended for servicing radio equip-
ment in the field when mains supplies and normal test gear are not available. An r.f. junction transistor is

Wolsey "balun" 75-300-ohm coupling unit.

Below: Circuit display boards used in Philco training eq sipment.

used in the oscillator circuit, operating in the range $2-6 \mathrm{Mc} / \mathrm{s}$ (as ordered) and giving an output (unloaded) of at least 1.5 V r.m.s. with a frequency stability of $0.01 \%$. The current drain from the built-in miniature battery is 1 mA , giving an average battery life of 40 hours.

Several experimental transistor receivers for medium and long waves were shown by Mullard, but perhaps the most interesting example of transistorization here was an unorthodox approach to the problem of v.h.f. reception. Obviously, even the latest r.f. transistors could not be used for r.f. amplification at $90 \mathrm{Mc} / \mathrm{s}$, so the aerial was connected straight to a diode mixer. The local v.h.f. oscillation was obtained by using an avalanche transistor to generate very sharp pulses and then selecting the appropriate harmonic from them by filters. Another interesting circuit was a television sync separator using two OC71 transistors. This gave line and frame sync pulses of about 10 V amplitude suitable for all normal types of timebase oscillator. Approximately 1 volt of composite video signal was required at the input.

An excellent aid to practical training in radio and electronics was demonstrated for the first time by a new department of Philco (Overseas). This consisted of a series of displayed circuit panels, 34 in all, each of which carried a circuit diagram, painted in black on a yellow background, and an actual working circuit laid out correspondingly. The components and valves are on the front of the panels and are plugged in alongside their appropriate circuit symbols. Up to eight panels can be mounted on a large display rack, which has wiring facilities for interconnecting the units. Accompanying this demonstration equipment is an equivalent set of 34 small standardized chassis for actual laboratory work. These duplicate the circuits on the demonstration panels and can be connected together in the same way in racks but are constructed on more orthodox lines.

On the same general principle of building up one's own apparatus-though not entirely for educational purposes-were the two instrument kits for home construction shown by Cossor Instruments. One of these was for a single-beam 4-inch oscilloscope with a timebase variable in frequency from $20 \mathrm{c} / \mathrm{s}$ to $250 \mathrm{kc} / \mathrm{s}$ and a $\mathbf{Y}$ amplifier having a useful response up to $10 \mathrm{Mc} / \mathrm{s}$ and giving a maximum sensitivity of 50 mV per cm . The other kit was for a valve voltmeter covering $40 \mathrm{c} / \mathrm{s}-1 \mathrm{Mc} / \mathrm{s}$ and having r.m.s. ranges up to 1.5 kV , peak-to-peak ranges up to 4 kV , and also ranges for ohms and d.c. volts.

Among the various accessories to be seen this year was a small "balun" for matching a $75-\Omega$ coaxial cable to a $300-\Omega$ balanced input circuit of the kind found in some f.m. receivers and tuners. It is made by Wolsey and has a coaxial socket at one end and a short length of $300-\Omega$ ribbon feeder at the other. An interesting aerial development is a new Belling-Lee "Skyrod" for commercial rather than "domestic" use which is fitted with a precipitation static discharger at the top. The specimen shown was about 15 ft long and made of $1 \frac{1}{2}$ in-diameter tube.

A wide range of printed circuits for various applications was to be seen on the T.C.C. stand, including small panels for mounting sub-assemblies and wafer switches with flush contacts.

# Mohile Radio Development 

V.H.F. Operation with Reduced Channel Spacing

By J. R. HUMPHREYS*

THE need to reduce channel spacing as a means of providing more channels within the frequency bands used by v.h.f. mobile radio in this country is not disputed by anyone. It is a factor of great importance to those actively concerned with finding sufficient channels to allocate, and those who wish to operate in these bands with freedom from undue interference.

The success of our efforts to reduce channel spacing, however, depends to a great extent on our ability to solve a number of important problems. Not the least of these is that of introducing new equipment to new standards into a field of operation now occupied by a variety of equipment, some of which is not always capable of providing inter-ference-free operation with the present channel spacing.

It is with this in view that the author intends to start with a brief examination of the history, method of operation, and channelling standard of equipment currently in use in the two v.h.f. bands; then pass to a more detailed examination of operational and technical factors affecting channel spacing in particular.

Present Practice and Standards in the U.K.Over the period of the last ten years a very considerable quantity of v.h.f. equipment has been installed and is operating in the bands 71.5 to $88 \mathrm{Mc} / \mathrm{s}$ (called the low band) and 156 to $184 \mathrm{Mc} / \mathrm{s}$ (called the high band).

The low band, which was the first to be used, has been allocated on a $50-\mathrm{kc} / \mathrm{s}$ channel separation basis, and schemes using this band have tended in the main to be those wishing to cover as wide an area of operation as possible.

The high band, which has at present a channel spacing of $100 \mathrm{kc} / \mathrm{s}$, is more suitable, by reason of its shorter range propagation characteristics, for use in highly industrialized areas and for schemes not requiring a large area of operation.

There is considerable need for more channels in both bands, but the problem in the high band has assumed a higher degree of priority, because of a recent decision to reduce its upper limit to $174 \mathrm{Mc} / \mathrm{s}$.

The bulk of equipment in these two bands operates on a two-frequency simplex system, with a frequency spacing of the crder of $10 \mathrm{Mc} / \mathrm{s}$, which permits a close geographical spacing between base stations.

When the existing equipment specifications were drawn up by the General Post Office, in conjunction with the radio industry in 1948, the accent lay on what it would be reasonably possible to produce rather than what was actually necessary to give a certain desirable system performance. Inevitable though this was at the time, it has produced a specification which has ihree important shortcomings from a channel separation point of view. These are:-
(a) When the maximum frequency errors occur,
two adjacent channels may be as close as $64 \mathrm{kc} / \mathrm{s}$ ai $180 \mathrm{Mc} / \mathrm{s}$, or $33 \mathrm{kc} / \mathrm{s}$ at $85 \mathrm{Mc} / \mathrm{s}$.
(b) The specified minimum selectivity of receivers is such that, even if the maximum errors in (a) do not occur, unrestricted and interference-free use of a number of adjacent channels cannot be guaranteed in the same area. This is particularly true of low band equipment where $50-\mathrm{kc} / \mathrm{s}$ channelling makes the selectivity requirements more arduous.
(c) There is no limitation placed upon the amount of energy radiated by a transmitter in channels adjacent to its own carrier frequency. This unwanted radiation becomes important for spacings of $50 \mathrm{kc} / \mathrm{s}$ or less when a high degree of adjacent channel protection is needed.

These factors have an important bearing upon the present allocation as well as on future plans.

Frequency Allocation.-This is a complicated subject upon which it is only too easy to generalize and it must be borne in mind here that the problem of allocation is often very difficult. Even so, the following two questions may fairly be asked:-

First, is the best use being made of the existing frequency bands, with the channel spacing currently employed, and the equipment at present in use? Secondly, with new equipment becoming available, what channel spacing should be adopted in the high and the low band and how can this best be introduced?

First, let us examine some of the methods which are at present in use, or could be introduced in these bands, to increase the use of the spectrum.
A well-known method, that of frequency sharing or co-channel operation, has obvious advantages particularly when applied to areas sufficiently spaced in distance to prevent interference. It also has limited use for "same area" users who may operate on a time-sharing basis. This latter type of sharing has obvious limitations, but applied intelligently is a useful example of frequency economy.

Another method often advocated, but not in use in this country, is that of allocating split channels in areas sufficiently separated in distance to enable existing equipment to be used. This would appear at first sight to offer considerable advantage and would allow $50-\mathrm{kc} / \mathrm{s}$ channels to be used in the high band and $25-\mathrm{kc} / \mathrm{s}$ in the low band. If this could be successfully applied over the country, the economy effected would enable more channels with the present spacing to be used in those areas already congested.

These advantages cannot be realized with present equipment, however, due mainly to the fact, already mentioned, that owing to frequency errors and lack of selectivity the geographical separation necessary would not be appreciably less than that for co-channel working. For example, a typical $50-\mathrm{kc} / \mathrm{s}$ channel equipment meeting existing specifications, but having maximum frequency errors, might well give only

[^0]$10-\mathrm{dB}$ rejection of a signal in an adjacent channel of only $25 \mathrm{kc} / \mathrm{s}$ separation.

Even when new and technically superior equipment is used this form of allocation does impose a greater burden of responsibiity upon those concerned with planning and frequency allocation. They might be expected to favour what at first sight would appear to be a simpler course of allocation which avoids this type of geographical and technical consideration.

Apart, therefore, from the other advantages to be gained, the introduction of equipment which will operate satisfactorily on adjacent channels in the same area, with reduced channel spacing, appears to offer planners a simpler solution to their problems.

Coming now to the second question, it seems a logical step to aim at the minimum achievable spacing that can be introduced in both bands. This has the very great advantage of avoiding the need for more than one change and one type of equipment and will ensure a very long period of operational stability.

Bearing in mind the present spacings of 50 and $100 \mathrm{kc} / \mathrm{s}$, the obvious reduction is to a spacing of $25 \mathrm{kc} / \mathrm{s}$ throughout, which will allow double the number of low-band channels, provide four times the number of high-band channels and at the same time will require no alteration of the frequencies at present in use.

There is considerable evidence to show, taking into account all foreseeable technical advances, that a separation of much less than $25 \mathrm{kc} / \mathrm{s}$ would not permit the satisfactory operation of adjacent channel schemes in the same area. There is, however, good reason to believe that equipment operating satisfactorily in the same area with a spacing of $25 \mathrm{kc} / \mathrm{s}$, could be used on split channels of $12.5 \mathrm{kc} / \mathrm{s}$ for operation in areas with suitable geographical spacing.

A very similar proposal has been made by the American F.C.C. authority who are splitting their present $60-\mathrm{kc} / \mathrm{s}$ channels into $30-\mathrm{kc} / \mathrm{s}$ and then $15-$ $\mathrm{kc} / \mathrm{s}$ channels with a geographical safeguard.

Introducing New Equipment.-The method of introduction of new equipment is obviously of great importance as it raises the problem of compatibility with, and interference to and from, old equipment. Iet us examine briefly the question of introducing two frequency simplex equipment in the high and the low bands which will: (a) operate satisfactorily in the same area on adjacent $50-\mathrm{kc} / \mathrm{s}$ channels, and (b) on adjacent $25-\mathrm{kc} / \mathrm{s}$ channels.

The introduction of equipment described in (a) into existing schemes presents a compatibility problem in the high band but not the low band.

If (a) equipment is added to existing high-band schemes, then existing base transmitters will need to have improved frequency stability. Base receivers must remain "wide band" unless all old mobile transmitters are improved in stability or until they become obsolete.

If (a) equipment is used in the high band and new $50-\mathrm{kc} / \mathrm{s}$ channels are interleaved between the old $100-\mathrm{kc} / \mathrm{s}$ channels, then it will not be possible to allocate them in the same area as existing $100-\mathrm{kc} / \mathrm{s}$ schemes, because of adjacent channel interference to old and new equipment.

The geographical separation between schemes which will allow an interleaving policy, is not easy to estimate owing to the uncertainty of the errors of existing equipment. A rough estimate would be
approximately half the co-channel spacing distance.
The introduction of equipment described in (b) into existing schemes presents a compatibility problem in the low and the high band, and existing base transmitters will need to have a much higher degree of frequency stability to work into the narrow bandwidths of mobile receivers. Similarly, base receivers will have to remain wide band until all old mobiles become obsolete or can be stabilized.
The interleaving of (b) equipment on new $50-\mathrm{kc} / \mathrm{s}$ channels in the high band will produce the same sort of result as for (a) equipment with rather less interference to the new services from the old.

The allocation of (a) equipment to new $25-\mathrm{kc} / \mathrm{s}$ channels, can be made in those areas where old equipment is not in use, or is at least $100 \mathrm{kc} / \mathrm{s}$ away in frequency.

Operational Limitations.- With the reduction of channel spacing and the consequent increase in the number of channels and users in any particular area; there will arise a number of operational factors not all of which are directly related to channel width.

The first, and perhaps the most important of these, is the effect of inter-modulation in receivers from a combination of unwanted transmitters spaced close in frequency to the wanted transmission. The subject has been examined in detail by Bullington ${ }^{1}$ and Babcock ${ }^{2}$ who, writing from their experience of American mobile v.h.f., had to consider the twin effects of congestion of operation, as well as the use of a single-frequency simplex system.

Because v.h.f. mobile systems in this country are operated mainly on a two-frequency basis, the problem is simplified considerably. This is because combinations of base transmitters, which might cause harmful interference in base receivers in the same area, tend not to occur. Also, if interference does arise, it is comparatively simple to eliminate.

The problem nevertheless exists to the same degree in respect of combinations of base transmitters causing interference in mobile receivers. It will also be caused, though less frequently, by a number of mobile transmitters interfering with base receivers.

To give some idea of the magnitude of these effects, the following example taken from recent field measurements using equipment of new design may be of interest:-

Three co-sited base transmitters, A, B and C, were operated in the $150-\mathrm{Mc} / \mathrm{s}$ band with powers of approximately 15 watts, with heights above the ground of approximately 100 ft and at a frequency spacing between each of $25 \mathrm{kc} / \mathrm{s}$.

It was observed that a mobile receiver operating on the frequency of transmitter A received interference from the combination of B and C (in the absence of a transmission from $A$ ) when it operated in an area around the site of between 200 yds and 800 yds radius. It is interesting to compare this figure with the radius of adiacent-channel interference which was confined to a distance of between 50 and 100 yds . Whilst it is true that with co-sited schemes the wanted transmission will generally "capture," this will not occur in the majority of cases. The planning, therefore, of numerous v.h.f. schemes in a small area and using closely spaced channels must take increasing note of this problem which rapidly becomes more difficult as the number of channels increases. Inter-modulation can, of course,
(Continued on page 483)
occur between transmitters but this is not a serious problem with transmitters of small power rating.
Equipment Design.-The technical factors which set limits to the reduction of channel width, and which primarily concern equipment design, are as follows:-
I. Bandwidth of transmission and reception.
II. Frequency stability of transmitters and receivers.
III. Inter-modulation effects in receivers.
IV. Susceptibility of narrow band receivers to noise interference.
I-Transmission Bandwidth.-Consideration of the "intelligence bandwidth" necessary for amplitude modulation, frequency modulation and single sideband systems, is in favour of the s.s.b. system. But it is only proposed in this article to consider the use of a.m. and f.m. systems as it is considered that sufficient technical progress has not yet been made to permit s.s.b. systems to be used for v.h.f. mobile radio.
The comparison of intelligence bandwidth of transmission between a.m. and f.m. is in favour of the a.m. system which in the ideal case only occupies a total spectrum of twice the highest modulating frequency in use. The f.m. system occupies a spectrum total of twice the highest audio frequency, plus twice the peak frequency deviation. Of more practical significance, however, is the attenuation of the unwanted sideband energy as this becomes an important and limiting factor in the ultimate reduction of channel spacing.

Measurements made by the author on a.m. and f.m. systems indicate that, from the point of view of modulation sidebands, a low-deviation f.m. transmitter is very similar to a carefully designed a.m. transmitter. The noise sidebands from f.m. transmitters, however, occupy a very much wider frequency spectrum, which becomes a limiting factor
not present with an a.m. transmission.
The modulation and noise sideband spectra of an f.m. transmitter are shown in Fig. 1 together with measurements of an a.m. transmitter designed for narrow-channel operation. This illustrates the very wide noise spectrum of a frequency modulated transmitter.

It seems probable also that in order to achieve channel spacings as low as $20 \mathrm{kc} / \mathrm{s}$, f.m. systems will need to restrict peak deviation to $5 \mathrm{kc} / \mathrm{s}$. In both a.m. and f.m. systems modulation limiting must be employed and followed by lowpass filters to prevent frequencies in excess of $3,000 \mathrm{c} / \mathrm{s}$ from entering the modulated stage.
Receiver Bandwidth.-The effective bandwidth, or "nose" bandwidth, of a receiver must allow for the modulation width, plus the frequency error of the wanted transmitter as well as its own frequency error and i.f. drift.

In the case of f.m. the minimum "nose" bandwidth ( 6 dB bandwidth) can be made equal to twice the r.m.s. frequency deviation plus twice the system frequency error.

In the case of a.m. the minimum "nose" bandwidth can be either twice the highest modulation frequency or twice the r.m.s. system frequency error, whichever is the greater.
The required slope response, or "bandwidth factor," of the main selectivity determining elements of receivers is not easy to specify accurately. This is because of the wide variations which are possible between the response of the filters used and the "true" selectivity when measured by "two signal" methods (i.e., the level needed from an interfering signal to degrade a certain level of wanted signal by a fixed amount). This is particularly true of f.m. receivers and was adequately described by Nicholson. ${ }^{3}$ The attenuation characteristics of filters must also allow for the frequency errors of the receiver itself and of the adjacent interfering transmitter.

Some conclusions drawn by the author are :-

1. A smaller bandwidth factor (i.e. a better filter) will be necessary for an f.m. system for the same end result as an a.m. system.
2. In general, up to 20 dB more attenuation is necessary from the filter characteristics than can be expected from a two-signal measurement on a.m. and probably nearer 40 dB for f.m.
3. With due allowance made for system frequency errors, there is no great operational virtue in making the true selectivity of receivers significantly better than the degree to which adjacent channel transmitter sideband energy can be attenuated.

II-Frequency Stability Considerations.-The degree of frequency stability obviously becomes greater as the channel spacing is decreased and if both fixed and mobile equipments were produced regardless of cost then even the most stringent of requirements could be met. Even if this were the case, however, it would not produce a proportional reward in terms of channel width reduction, as the present limitation (ignoring inter-modulation effects) is the


Fig. 1. Comparison of modulation and noise spectra of a.m. ana $\dagger$.m. transmitters designed for narrow-channel operation.
frequency spectrum occupied by both a.m. and f.m. transmitters.

This being so, the requirements of frequency stability are as follows:-
(a) The frequency stability of transmitters and receivers must be such that, under the worst conditions, the channel spacing is not reduced to less than a figure which will be determined by the onset of interference in receivers due to transmitter sideband radiation. For instance, in a $25-\mathrm{kc} / \mathrm{s}$ system this might well be not less than $20 \mathrm{kc} / \mathrm{s}$.
(b) The transmitter and receiver stability must be such that under conditions of maximum frequency error the degradation of "on-channel" performance is not unreasonable.

These two factors are, of course, related to the receiver "nose" bandwidth and bandwidth factor, and a compromise is necessary bearing in mind that a greater frequency error may be tolerated in an a.m. system for a given degree of distortion. In any event, the less the frequency errors the greater may be the permissible b.w.f. and the smaller the "nose" width. In this last connection it should be noted that the reduction of "nose" bandwidth is beneficial because it will reduce the degree of energy received from a transmitter's unwanted sidebands.

The implications of tighter frequency tolerance are of great practical importance in relation to the design, production and maintenance of equipment. In respect of frequency stability the important difference to be expected between present equipment and equipment for considerably narrower channel operation, i.e. $30 \mathrm{kc} / \mathrm{s}$ or less, are as follows :-

1. All equipment will have to be adjusted with a high degree of accuracy to their assigned frequency.
2. The use of crystal-oven control, to maintain the required degree of stability, will be important on base station equipment, and in the case of f.m. probably on mobile equipment also.
3. Facilities must exist for periodicaliy checking the frequency of base station and mobile equipment which must be provided with sufficient adjustment to correct any frequency error.
One method, which has much to recommend it, is to periodically check the base equipment frequency and provide a simple means of checking the relative error of the mobiles. This may be accomplished by the use of a stable oscillator on the centre frequency of the receiver's i.f.' amplifier which in the presence of the wanted transmission will give a beat note equal to the prevailing system error. An illustration of a miniature portable oscillator which has been specially. designed for this purpose is shown in Fig. 2. This oscillator is crystal controlled and employs one transistor.
III-Inter-modulation.-The operational effect of inter-modulation in receivers has already been stressed and therefore all measures to reduce this factor will be of great benefit. Three approaches are possible :-
4. Provide sufficient r.f. selectivity to attenuate the unwanted adjacent channels.
5. Use r.f. amplifiers and mixers which have a high degree of linearity.
6. Keeping the voltage gain of r.f. stages to a minimum, consistent with good noise factor.
It is clearly uneconomical to provide a sufficient degree of r.f. selectivity before non-linear valves which produce these harmful distortion products, and therefore it is not of practical importance.

The last two solutions are immediately applicable but only (3) is entirely in the hands of the equipment designer.
IV-Noise Considerations.-The increase in susceptibility of receivers to impulsive noise as the passband is reduced is both an unfortunate fact and a very complex problem, which presents' difficulty for the engineer both in respect of calculation and measurement.
Analysis and measurements which have been conducted on f.m. systems ${ }^{4}$ tend to show that, dependent upon the deviation employed, the level of input and the degree of frequency error, a $25-\mathrm{kc} / \mathrm{s}$ receiver will be on the average $3-\mathrm{dB}$ (but could be $6-\mathrm{dB}$ ) worse than a $50-\mathrm{kc} / \mathrm{s}$ receiver as far as impulsive noise is concerned.
Where fluctuation noise is considered, this average figure is not measurably different, but for small input levels the narrow-band' receiver can be slightly superior.

For a.m. receivers fluctuation noise is less in direct proportion to the bandwidth, a fact which renders this narrow-band type of receiver much quieter, in the absence of any impulsive type of interference, than its wide-band counterpart.
Narrow-band a.m. receivers are more susceptible to impulse noise and this has been adequately discussed by Toth ${ }^{5}$ and many others. Measurements made by the author show that, depending upon the input level, but not to any great extent the frequency error, there is an inferiority of approximately 3 dB compared with a $50-\mathrm{kc} / \mathrm{s}$ receiver.

Trends in Equipment Design and Measurement Technique.-It is interesting to examine what has made possible the solution of many of the problems just enumerated, problems that it was not possible to solve economically as late as 1950. Taking the technical parameters one by one, we come first to the big advantage to be gained in frequency stability by the use of close tolerance AT-type crystals. These can be produced economically with as little as $\pm 0.0015 \%$ change in frequency for a temperature variation of some $80^{\circ} \mathrm{C}$. Coupled with this is the use in base or mobile equipment of the miniature type crystal oven (illustrated in Fig. 3). This type of unit has long been in use in the U.S.A. and Canada and for a relatively small cost will reduce crystal frequency drift to less than $\pm 0.0005 \%$.
With the use of ferrites, small highly stable capacitors and methods of encapsulation, it is no longer a formidable problem to produce the necessary receiver selectivity in a small size and with the required stability. As an example of what may be


Fig. 2. Miniature crystal controlled transistor oscillator for testing v.h.f. mobile equipment (Labgear).


Fig. 3. Miniature crystal oven for use in mobile v.h.f. equipment.
achieved, an i.f. amplifier using six i.f. transformers and employing ferrite coils can be made with a "nose" width of $12 \mathrm{kc} / \mathrm{s}$, a bandwidth factor of 3 , and a frequency drift not exceeding $1 \mathrm{kc} / \mathrm{s}$.

Conclusions.-In conclusion it is intended to describe and comment in brief upon a number of the most important factors affecting the introduction and operational success of narrow-channel spacing in the v.h.f. bands.

1. The technical specifications to which this type of equipment must be designed and operate, can and are being written in a manner which permits satisfactory operation on adjacent channels in the same area. In this respect it will be possible to give a considerable measure of improvement with narrow-channel spacing, compared with current apparatus complying with existing specifications at the present channel spacing.
2. Consideration should be given to the use of "same area" $25-\mathrm{kc} / \mathrm{s}$ channel-spaced equipment, and frequency separation of, say, $12.5 \mathrm{kc} / \mathrm{s}$ with geographical spacing.
3. Everything else being equal, some small degradation must be expected in the performance of narrowchannel equipment under conditions of impulsive interference but this should be more than compensated by the availability of more channels.
4. There is no reason to think that narrow-channel equipment will be less reliable, indeed it is likely that in view of the more stringent requirements in design, reliability will be improved.
5. There will inevitably be problems of compatibility between existing and new equipment and old and new frequency allocations. Careful attention to a comparatively small number of essential technical considerations, however, should ensure a smooth period during the introduction of new equipment.
6. There should be only a small increase in maintenance cost as compared with a properly maintained wide-band system. It must be borne in mind, however, that satisfactory maintenance will demand accurate test equipment and well trained personnel.
7. The maximum use of the extra channels, made possible by narrower channel spacing, will not be
obtained in areas using a large number of channels, unless there is a reasonably co-operative approach to the problems of the transmitter location, radiation pattern, and other means for reducing the effects of inter-modulation.
8. There is an obvious advantage to be gained by the immediate introduction of equipment which will operate satisfactorily on the very minimum of channel spacing. This will ensure a long period of technical and operational stability and at the same time offer the very maximum of channels to the user.

Finally, the way in which our limited frequency spectrum is to be divided between the various classes of users is a factor which will have a considerable bearing on the best use of the extra channels which technical achievement has made possible. The maximum benefit to be gained will, in the end, be dependent on the good sense of administrators, manufacturers and users.

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${ }^{3}$ "Comparison of AM and FM" by M. G. Nicholson. Wireless Engineer, July, 1947.
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## COMMERCIAL LITERATURE

Square Pulse Generator giving a main pulse, variab!e in width from $0.2 \mu \mathrm{sec}$ to 2 secs, from one output and a $0.2 \mu \mathrm{sec}$ "pre-pulse", precedi: ; it by an interval variable between $0.2 \mu \mathrm{sec}$ and 2 secs , from another output. Main pulse has rise time of $0.01 \mu \mathrm{sec}$ and can be triggered at frequencies up to $3 \mathrm{Mc} / \mathrm{s}$. Specification from Nagard, 18, Avenue Road, Belmont, Surrey.

The Narda Corporation, of Mineola, New York, ask us to state (with reference to the item Waveguide Attenuators in the August issue) that their British representatives are now B and K Laboratories, 57, Union Street, London, S.E.1.

Optical Lenses and hand magnifiers made from synthetic resins, suitable for laboratories and inspection departments. Illustrated leaflet of types available from Combined Optical Industries, Plasta Works, Bath Road, Slough, Bucks.

Anchoring Devices for circuit wiring. Technical bulletin (No. 4) with drawings of recent introductions from Harwin Engineers, 101-105. Nibthwaite Road, Harrow, Middlesex.

Aerials and Accessories for sound and television, indoor and outdoor, twin-band, v.h.f. and cars. Convertors and pre-amplifiers are included. A 1956-57 catalogue of the complete range made by Aerialite, Castle Works, Stalybridge, Cheshire.

Contact Bi-Metal consisting of an inlay or facing of contact material integrally bonded to a base metal backing. Suitable for manufacture of contact parts, and will withstand shere zing, punching and bending. Also finished bi-metal contacts. Electrical Engineering Data Sheet 1300:345 from Johnson, Matthey and Co., 73-83, Hatton Garden, London, E.C.I.

Pillar Signal Lamp, like a miniature lighthouse, suitable for illuminating baseboards, etc. Also a 3-position jack switch, signal-lamp lenses, micro-switches with toggle (lever) operation and rubber-covered crocodile clips. Leaflet of Jew products from A. F. Bulgin and Co., Bye Pass Road, Barking, Essex.
Aerial Masts for rhombic, turnstile, delta match, v.h.f., u.h.f. microwave and scatter propagation arrays. Also transmission lines and other accessories. Illustrated catalogue from the Tower Construction Company, Sioux City, Iowa, U.S.A., whose installation services include survey and layout of site, computation of stress analysis and construction of feeders.

# Full-Range Electrostatic 

By H. J. LEAK, M. Brlt. I.R.E., and A. B. SARKAR, M.Sc. (Cal.), M.Sc. (Lond.), Grad. Inst. P., Grad. Brit. I.R.E.

I

T is now well known that the push-pull electrostatic loudspeaker is capable of a quality of reproduction far in advance of that obtainable from the moving-coil system. So far as the authors are aware previous literature on the push-pull condition and practical attempts at the construction of push-pull loudspeakers ${ }^{1,2}$ have been on the assumption that construction will be as shown in Fig. 1. Hunt ${ }^{3}$ (pp. 167-212) has given a detailed analysis of this transducer system and further contributions have been made by Walker ${ }^{4}$, Cocking ${ }^{5}$ and Nuttall ${ }^{6}$.

In the course of research and development on full-range electrostatic loudspeakers the authors have discovered that an alternative method of construction has theoretical and practical advantages.

It will first be necessary to consider briefly the construction of Fig. 1. The active element consists of a very thin film of plastic material (D) stretched midway between two insulated and perforated metal plates ( $\mathrm{P}_{1}$ and $\mathrm{P}_{2}$ ). The gaps $\mathrm{P}_{1} \mathrm{D}$ and $\mathrm{P}_{2} \mathrm{D}$ allow movement of the diaphragm, which must have high resistivity in ohms per sq. centimetre. The alternative arrangement of inserting a very high resistance (such as R in Fig. 1) has been shown in previous literature, the stated purpose being to prevent the electrostatic charge on the moving diaphragm from varying during the lowest audio-frequency cycle to be reproduced. This is an impracticable solution in a full-range electrostatic loudspeaker, the reason being that the capacitances will be small enough to necessitate $R$ being of the order of thousands of megohms. If the full d.c. potential is to be applied to the diaphragm then the insulation must be in the region of many thousands of megohms, and this is impracticable with normal humidity variations.
Let us examine some aspects of the mounting of the diaphragm of Fig. 1. The diaphragm must be tensioned to give a positive stiffness sufficient to counteract the negative stiffness or force resulting from the constant charge. The acoustical loading on either side of the diaphragm is low at low frequencies. With the thin, light plastic materials at present available for use, the elastic restoring force may not be sufficiently stable with respect to time to prevent eventual collapse. Of course, greater stability can be obtained by sub-dividing the diaphragm into smaller areas bounded by supports, and better acoustic impedance matching and directivity may be obtained by making the diaphragm vibrate in the end of a long tube ${ }^{2-10}$, and in this case the radiation impedance for the loudspeaker will be approximately the same as if it were mounted in an infinite baffle. The directivity pattern can also be altered by mechanical and acoustical treatment in the long tube. At first glance it appears that by these means one can avoid the use of the type of cabinet so essential for dynamic loudspeakers, but the work involved in production is complicated and little less expensive than a cabinet.
The foregoing considerations led the authors to the new method of construction shown in Fig. 2.

The loudspeaker consists of an insulated and perforated metal plate fixed rigidly between two stretched thin and light diaphragms, with air backing. Thus, the a.c. signal will be applied on the membranes (D1, D2) and the d.c. potential on the fixed plate (P).
The transformation of electrical energy into mechanical energy involves the interactions between magnetic (in the case of the moving coil transducer) or electric (in the case of electrostatic, electrostrictive and piezoelectric transducers) fields and matter. The present theoretical concept is based on any of these physical effects since all types of electroacoustic transducer follow the reciprocity law. Table 4.8 of ref. 11 shows the general relationships for these electro-mechanical transducer principles. It must be remembered that in the case of the movingcoil transducer the force per unit current is a function of the inductance and of the negative stiffness resulting from the steady magnetic field, and the solution therefore entails the concept of "motional impedance." But, in the case of the electrostatic transducer the force per unit voltage is a function of the capacitance and of the negative stiffiness resulting from the steady electrostatic field, and hence the analysis can best be carried out using the concept of "motional admittance "3 (p. 202), particularly when considering light plastic diaphragm materials comparable to the density of air.

We know that a mechanical force $f$ produces a particle velocity $u$ when applied over a transducer surface S. If we consider plane compressional waves in a medium of low viscosity, then we have $p \mathrm{~S} / u=\rho_{\mathrm{a}} c_{\mathrm{a}} \mathrm{S}=\mathrm{Z}_{\mathrm{B}}=$ mechanical radiation resistance (for lossless medium), where $\rho_{\mathrm{a}} c_{\mathrm{a}}$ is the impedance of the medium (air), and $p$ is the pressure.

We can draw now the equivalent circuit of an airbacked transducer which may take the form of Fig. 3, in which $L=$ equivalent inductance $=M / 4 t_{\mathrm{t}}{ }^{2}$, where $\mathrm{M}=$ motional mass and $t_{r}=$ transformation factor; $\mathrm{C}=$ equivalent capacitance, where $\mathrm{K}=$ motional stiffness; $\mathrm{R}_{\mathrm{m}}=\mathrm{Z}_{\mathrm{R}} / 4 t_{r}{ }^{2}$, and $\mathrm{C}_{0}=$ clamped electrical capacitance.


Fig. 1. In the majority of push-pull electrostatic loudspeakers a central diaphragm vibrates between perforated fixed outer electrodes.


Fig. 2. Alternative arrangement oi electrodes offering many advantages both in design and manufacture.

# Loudspeakers 

A New Approach to Practical Design

We know that the mechanical $Q$ of the transducer plays an important part in controlling the frequency response of the radiated power. The calculated value of the mechanical Q is generally lowered owing to the mounting loss of the transducer. Thus mounting is very important, since this additional damping reduces the intensity of the sound radiation. We know that it is possible to represent mechanoelectric networks either by the sum of admittance or by the sum of impedance components (velocity and current as independent variables), and this so-called " duality " behaviour ${ }^{3,12}$ plays an important part in the analysis of complex electro-acoustic systems. An example is the operation of a spring ${ }^{11}$ (a simple harmonic motion device).

Let us now study the vibration of a stressed membrane which will be applicable to the motion of our diaphragms in Fig. 2. Morse ${ }^{13}$ has dealt in detail with the vibration of a stretched membrane, and he has shown that the shape of the boundary line on which the diaphragm is stretched has considerable effect on the solution of the equation, which is beyond the scope of this paper. One of his classic examples is a kettledrum, which we will consider briefly as it is applicable to our loudspeaker design. Striking one of the stretched membranes causes alternating compression and expansion of the enclosed air, which exerts force on both membranes and modifies their modes of vibration. When the velocity of transverse waves in the membranes is low compared with the velocity of sound in air, then the effect of the motion of one part of the membrane transmits very rapidly through the air to affect the other parts, and so, on the whole, the reaction of the air is uniform over the membrane's surface. When the membrane vibrates, due to the adiabatic pressurevolume variations the excess pressure is represented with a negative sign, since this pressure always opposes the displacement of the diaphragm. The load offered to the diaphragm is expressible as a resistive term (additional reactive load is added when the mass of the diaphragm is not negligible). The resistive term varies with frequency, except when the speed of sound in the medium is less than that in the membrane, in which case the resistance is constant ( $=\rho_{\mathrm{a}} c_{\mathrm{a}}, \rho_{\mathrm{a}}$ being the density of the air and $c_{\mathrm{a}}$ sound velocity in air).

The foregoing physical phenomena can be applied to the design of a modern condenser microphone and, remembering Rayleigh's and Helmholtz' reciprocity theorems and also the acoustical principle of similarity ${ }^{0}$, we know that these phenomena will be equally applicable to the construction of our fullrange electrostatic loudspeaker. Considering the microphone, we must know the driving force due to the incident wave (uniform, approximately, over the diaphragm), and the reaction force per unit area of the medium (both sides of the membrane), the latter being proportional to the average displacement. The proportionality factor contains the specific acoustic impedance term of the medium. The resistive part
of this impedance consists of (1) the radiation resistance of the air next to the outer part of the membrane, and (2) the resistance due to reaction offered to the side facing the inside of the microphone case containing small holes (viscous friction is produced with the motion of air through these holes). The reactive part of the impedance due to the outer air is masslike (i.e., positive reaction) and that due to the air inside is stiffiness controlled. The motion of the diaphragm is stiffiness controlled when the frequency of mechanical resonance lies near the upper limit of the frequency range. To achieve this a metallic diaphragm must be tensioned, limited by the tensile strength of the metal. In the case of a nonmetallic diaphragm of low tensile strength, resonance will occur too low in the frequency range unless means can be found to bolster up the stiffness of the material. This can be done ${ }^{14,15}$ by a construction similar to Fig. 2, but omitting an electrical connection to one diaphragm. Assuming that the boundaries are air-tight, the trapped air serves the acoustic purpose of providing the necessary stiffiness to the working diaphragm.

We have applied the foregoing principles to the

Fig. 3. Electrical analogue of - loudspeaker with the electrode arrangement of Fig. 2.

construction of a full-range electrostatic loudspeaker,* which takes the form of Fig. 2. Consider the static condition: when the polarizing potential is applied the charge is constant over the entire surface of the central conducting electrode, because it is effectively infinitely rigid. Both diaphragms move small equal distances towards the central electrode, and there is no unbalancing of the two capacitances; therefore the balanced push-pull condition is maintained. One important advantage of our construction is that we can dispense with $R$ (shown dotted) altogether, without altering the operation of the loudspeaker; because the charged plate is not moving, there is no necessity for a large RC constant. The problem of achieving resistance and insulation values of thousands of megohms is thereby eliminated. Of course, a low, practical value of $\mathbf{R}$ can be inserted as a protective device.

The plastic diaphragms in our design are given outside resistive coatings, the values of these being designated $\mathrm{R}_{1}$ and. $\mathrm{R}_{2}$ in Fig. 2; therefore they are not external physical resistors. This construction prevents migration of charge, and also gives us a built-in RC transmission line offering a partial solution to the variable-area requirement necessary in a fullrange loudspeaker.

On applying the polarizing potential, in the absence of a signal, the behaviour of the diaphragms can be studied under two conditions: (1) when the interior air cavity is perfectly air-tight, which (if it can be practically realized over a long period) necessitates calculations based on innumerable

[^1]atmospheric variations; and (2) when a pinhole duct is deliberately introduced to allow for atmospheric pressure equalization. We will deal here only with condition 2. The force of attraction between the perforated plate and one diaphragm can be represented by: $\mathrm{F}=\xi \mathrm{V}_{0}{ }^{2}$ where $\xi$ is a constant.

Consider now the dynamic conditions when a signal voltage is applied, comprehending that the pinhole introduced to allow for atmospheric pressure equalization will have negligible effect at any audiofrequency on the stiffness of the enclosed air, and therefore on the damping effect of the resonant modes of vibration of the diaphragm. It is to be noted here that the displacement of the diaphragms will be due only to the magnitude of the constant charge on the fixed plate and the a.c. field resulting from the signal voltage between the diaphragms. Assuming $e$ and $e^{\prime}$ to be the maximum values of the signal and $\omega$ and $\omega^{\prime}$ their respective angular velocities, we can write the Fourier series (for the force between direct and alternating voltages) as:

$$
\begin{align*}
& f=\xi\left(\mathbf{V}_{0}+e \cos \omega t+e^{\prime} \cos \omega^{\prime} t\right)^{2} \\
&=\xi\left[\left(\mathbf{V}_{0}+\frac{1}{2} e^{2}+\frac{1}{2} e^{\prime 2}\right)+\right. \\
&+e e^{\prime}\left\{\cos \left(\omega+e^{\frac{1}{2}}\left(e^{2} \cos 2 \omega t+e^{\prime 2} \cos 2 \omega^{\prime} t\right) t \cos \left(\omega-\omega^{\prime}\right) t\right\}\right. \\
&+\left(2 \mathbf{V}_{0}\left(e \cos \omega t+e^{\prime} \cos \omega^{\prime} t\right)\right] \\
& \quad \text { (on expansion) } \ldots
\end{align*}
$$

The first term of equation (1) is a steady component whose function is to displace the diaphragm from its stationary position; hence it does not contribute to the reproduction of sound. The second term will produce second harmonic distortion, for we can see that it contains terms with frequencies double those of our applied frequencies. Similar distortion in the sound output will occur with the third term containing the sum and differences of the applied frequencies, so it is only the fourth term, which contains the applied frequencies, that we desire to be reproduced. Obviously, for low acoustic distortion we must make $\mathrm{V}_{0} \gg e$ and $e^{\prime}$, so that $2 \mathrm{~V}_{0} e$ and $2 \mathrm{~V}_{0} e^{\prime} \gg e, e^{\prime}, \frac{1}{2} e^{2}$ and $\frac{1}{2} e^{\prime 2}$. The same argument regarding the force can then be applied to the other diaphragm, remembering that the displacement and the a.c. signal change sign on passing from one side to the other. Thus not only the efficiency, but also the quality of sound output is dependent on a high polarizing potential, particularly at low frequencies, since amplitude and sound energy is directly proportional to $\mathrm{V}_{0}$ and $\mathrm{V}_{0}{ }^{2}$ respectively. Note that this loudspeaker acts as a doublet. Note also that we have not taken into account, for the sake of simplicity, the non-planar shape of the vibrating diaphragms and the corresponding variation of electrostatic force on different parts of the diaphragms.
As previously stated, the air enclosed by the diaphragms acts to stiffen them, and this helps towards an extended low-frequency response because the need for highly tensioning the diaphragms is reduced. The stiffness behind the diaphragms is dependent to some extent on the dimensions of the perforations in the central electrode. Changing the dimensions, and hence the air resistance, will alter the displacement and response curves, as shown by Morse (ref. 13, p. 197).

The advantages of this system are considerable. Harmonic distortion, transient distortion and frequency deviations are all very much lower than on moving coil systems. The construction is simple, and reliable; the dimensions need not be unreasonably
large; the diaphragms need not be very tightly stretched; they are visible for inspection, and they form a dust-proof barrier protecting the highlycharged central electrode.

In the previous convention of Fig. 1 the single diaphragm can be properly tensioned in the first instance on one of the perforated plates, but when the second plate is applied it is difficult to ensure exact coincidence of the spacers, and also the diaphragm is not accessible for inspection or adjustment. There is also the serious practical difficulty of holding the spacers on the two plates in intimate contact with the diaphragm.

The authors hope that the wide application of their ideas will speed the availability of better and cheaper high-fidelity loudspeakers, for it can be stated that the cost will be considerably lower than the best moving-coil systems, whilst the overall improvement in listening quality is demonstrably beyond question.

Thanks are due to our colleagues E. H. Ashley and P. H. Biggs for their practical help in constructing many development models.

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## Solid "Electrolytic" Capacitor

A METHOD of eliminating the electrolyte in tantalum oxide capacitors is reported by Bell Telephone Laboratories. After formation of the oxide film on the surface of the sintered tantalum anode in the usual way, the electrolyte is removed and replaced by successive deposited layers of manganese dioxide, carbon and lead alloy. The barrier layer established by tantalum, tantalum oxide and manganese dioxide is said to have high stability with time and temperature and gives a capacitance of $500 \mu \mathrm{~F}$ per cubic inch for a voltage rating of 35 .
Arrangements for manufacture are in the hands of the Western Electric Company.

## LETCMERS TO THE EDITOR

The Edior does not necessarily endorse the opinions expressed by his correspondents

## Echoes of the Show

HAVING heard several "high-quality" demonstrations at this year's Radio Show it has struck me how unsatisfactory the set-up is. There are no standards of comparison. One demonstrator was frank enough to say the tone controls were set to suit himself and he could only hope it suited his audience. On another stand I was told that a speaker cabinet had been designed for another room, hence its admisted rather poor performance. The recordings were chosen either to show good points or conceal the bad ones.

I would suggest in future all the demonstration rooms should be made to a standard design and part of the demonstration should include standard records of widely varying character, all played with tone controls set level. Of course, other recordings and settings could be used for other parts of the demonstration. And dare I ask for "independent" rooms for comparing different makers' products?
London, N. 8.
C. STREATFIELD.

AT the Radio Show a great deal of time was given to demonstrating the reproduction of gramophone records of popular music played on an electronic organ with a rhythmic accompaniment.

In a type of instrument commonly used for this sort of entertainment, harmonics generated by rotating electromagnetic tone-wheels are synthesized into complex sounds; but only harmonics up to the eighth are used and the highest frequency generated is about 6,000 cycles per second. Moreover-and this applies to any electronic instrument-the final complex sound itself issues from a loudspeaker which may be incapable of reproducing the highest audio frequencies.

I do not suggest that the instruments in question are defective; on the contrary, the limitations I quote are largely the result of careful and ingenious design in a practical instrument. It is clear, however, that electronic organs, whatever their merits otherwise, are of limited value as test material for wide-range sound equipment. In the demonstration I heard this fact was concealed by the drums accompanying the organ which naturally stood out in startling fashion when an efficient tweeter was brought into use. Unfortunately, the ability of a loudspeaker to reproduce toneless percussion instruments is no guarantee of a clean upper register; and many reproducers able to recreate vividly the snare drum, castanet and tambourine fail miserably in the reproduction of sustained complex tones at the top of the musical scale. Records of electronic organ and drums may therefore be positively misleading as a guide to the capabilities of a wide-range reproducer.

Sawbridgeworth, Herts.
H. GLOVER.

## Receiver Radiation

"DIALLIST," in your August issue, is a bit belated in suggesting that U.K. manufacturers should emulate the F.C.C. and establish limits for line timebase and other radiation from television receivers. My Association has had limits (evolved in collaboration with the G.P.O.) for line timebase radiation and tentative limits for oscillator radiation for about four years. In due course these limits will be published as part of BS 905 .

We claim that, in the field of receiver measurement and interference radiation measurement and suppression techniques, the U.K. radio manufacturing industry is well ahead of its counterparts in other countries. In fact, many of the B.R.E.M.A. proposals to the B.S.I. were submitted in unabridged form as British proposals to International Standards conferences and have since
been incorporated, basically unaltered, as standards
The only criticism of British manufacturers in this respect which we feel "Diallist" could legitimately make is that the industry has been too overloaded to publicize the work it has completed in collaboration with the G.P.O. However, it is to be hoped that this will be remedied by the publication of the British Standard.
S. E. ALLCHURCH.

Secretary, The British Radio Equip-
ment Manufacturers' Association.

## Headphone Portables

WHY does no manufacturer or kit supplier provide a cheap light portable or mains transportable for headphone use? The solution of the difficulties and domestic ill-feeling that arise when different members of a family want different programmes is not going to be solved by the production of small loudspeaker receivers for alternative use. Each loudspeaker programme needs a separate room, but the average English home in winter has only one warm habitable room; also there are times when some of us prefer silence.

It is simple enough, I know, if one has the time and facilities, to modify existing sets and circuits for headphone use, but why should it be necessary? Besides, a manufacturer has the resources to design such a set so as to achieve the greatest compactness. A mains headphone set is what is wanted.

Reading, Berks.
THOMAS ASHBY.

## "Economy in Receiver Design"

I SHOULD like to comment on the use of negative feedback into a diode load, as in the circuit of the simple superheterodyne described in your August issue. In order to be negative, the feedback voltage must oppose the audio voltage on the diode due to demodulation, causing the diode to "see" an impedance lower than the load resistance. As the following amplifier does not pass d.c. the result is a deterioration of the a.c./ d.c. load ratio, bound up with the well-known clipping at high modulation.

There is also another way of looking at this. Assume at the diode a carrier of, say, 1 volt nearly fully modulated with a sine wave, and the volume control set so as to get, say, 2 volts appearing across the speaker terminals (about 1.5 watts). Then at the instant of the centre of the modulation trough the negative diode voltage due to r.f. (or i.f.) will be at its minimum (a.f positive going), but a negative peak of 2.8 volts appears at the other end of the diode load. It is clear from this that the diode is cut off, at least during an appreciable part of the modulation trough. Admittedly this is about the worst possible case, but some clipping must occur, especially with weak carriers and at high volume.

Negative feedback into a diode-triode stage seems a tricky business. The feedback signal can be féd either (a) into the grid circuit, or (b) into the cathode circuit, but in both cases the same problem arises, unless in case (b) the entire grid and diode circuit is also returned to the feedback point, having the disadvantage of losing all the gain of the triode stage, so far as the feedback loop is concerned. It would seem better to combine the diode (or diodes) with the i.f. stage (6B8G, EBF80, etc.), and feed back into the audio triode cathode. The ensuing loss of maximum gain. (feedback being constant now) is probably not so serious, as the maximum possible gain would hardly be wanted in connection with the then rather high hiss level, especially with good quality i.f. coils and a ferrite rod aerial.

London, N.W. 2 .
G. N. E. PASCH

# Flexibility in Cabinet Design 


"Knock-down " or "pack-flot" television cabinet which utilizes two identical moulded side cheeks. The loudspeaker louvers point forward, guiding the sound toward the front of the set; the controls may be easily recessed. Decorative schemes could be changed by pads in the moulding tools, and the glass courd be silk-screened around the edges to mask internal fixings. For 14 -in and 27 -in television sets, the top, base, back, mask and glass could be altered dimensionally, still leaving the mouldings intact. The top panel could be veneered ply or leathercloth-covered hardboard. With this arrangement, cabinet storage and transport difficulties are overcome. All the complex shape is token by the plastic moulding, whlle the other parts are comparatively plain.


## -

A different approach to the relevision console or the radio equivalent to the furniture monufocturer's " $G$ " plan.

Carries things a stoge further. Here are three alternatives; a pull-out drawer type gramophone makes a fourth.

THE CASE FOR PLASTICS, ALONE OR

IN COMBINATION WÍTH WOOD

IN these days of constant change in cabinet fashions radio manufacturers are often reluctant to make use of the many advantages of plastics when faced with high initial tool cost and the possibility of a "fixed " design. With a wooden cabinet, although more expensive in itself, there are no tool costs and alterations can be accomplished with relative ease to give the model a "face lift" at a later date.

In spite of this, I hope to show by the accompanying examples that plastics can give the designer much more scope than is perhaps realized. The combined use of plastics and wood can often give strength and enhance the appeal of the design as a whole. The sketches are put forward simply as basic ideas to start trends of thought and they will have served their purpose if they lead to something new or help people to view plastics in a different light.

Application has been made for registration of the designs presented in this article.


Complex curved sides, top and back in phenolic; chassis sits in virtually a plain wooden box with a foncy front. The moulding is used as a dust cover and holds tne glass and mask, the glass being detachable via the wooden trim across the front. Plain colours, stoved enamel or the patented "Oxvar" finish could be used on the phenolic cover.


By JOHN W. MOORE, M.S.I.A.
(Formerly Chief Designer, British Moulded Plastics, Ltd.)


A method of sealing the air space between the glass and the tube face and moulding the mask as an integral part of the cabinet. As an alternative, the front portion alone could be moulded and attached to a simple wrapped-ply cobinet.


A
One-piece phenolic cover. Everything is there in one moulding: loudspeaker apertures, vented back, rabbet for glass with simple removal. Chassis, tube and controls mounted on plywood base. A separate centre pad in the tool could give variety in the speaker openings which coula also be covered or pointed to give colour changes. An extruded flexible p.v.c. trim provides a non-scratch, nonskid buffer for a polished table top and allows yet another possible colour change.

The cabinet body should be of relatively simple shape with a plain frontal area. The tuning scale could be applied as a transparent disc, or moulded-in and paint filled. The frontal covering could be virtually anything with enough holes in to pass the sound. Metal or plastic trim strips could be trapped between the cabinet and front cover.


A shot ot the future. If it makes you shudder, I hope it will also make you think. In effect it is on ideal moulding form with its sweeping curves and compound radif.


The main objective is to cover the vacant stare of the cathode-ray tube when not in use. This is achieved on more expensive models by hinged doors or wooden shutters. Plastic shuttering achieves cheapness, colour and durability in one go.



An alternative treatment of the two-piece cabinet in which the front escutcheon is constant. Either a moulded or a wrapped-ply back cover with internal fittings could be used.


Possibly the best shape from the portable point of view. The long handle is designed to take the pull of the set's weight at the proper points and could be made to pivot to one side to disclose the scale and knobs. These would fit neatly into a dished form or could be attached to the chassis and pushed into the top part of the recess that takes the baffle, thereby giving a frontal instead of top scale presentation. The moulding itself would be inexpensive and strong and could be either colourful or unobtrusive.


Method of achieving maximum internal flexibility in a portable with the minimum of tool cost using "identical" back and front forms, but with a slight omendment to the top tool. For example, if we require holes in the front of the cabinet we mould, say, 5,000 off. The pad is then changed to seal off those holes and 5,000 backs are moulded.


Method of changing the "face " of a cabinet. Not new, but could be more widely known. Although the body remains basically unaltered the part that really matters may be subjected to innumerable alterations.

## Centimetre-wave Beacon



Microwave harbour beacon equipment, with portable receiver in the foreground.

A COMMERCIAL version of the microwave harbour beacon described in Wireless World, November, 1955, by A. L. P. Milwright (Admiralty Signal and Radar Establishment) has now been produced by Elliott Brothers. The equipment, though basically unchanged, has been modified in external form, particularly as regards the receiver. It works on an adaptation of the Lorenz system of overlapping signal sectors with an equi-signal path between them.

In its present form the gear provides a simple and relatively inexpensive navigational aid for small craft, such as fishing vessels, when entering harbours. The transmitter, installed on shore, costs $£ 550$ and receivers, one of which is needed for each vessel using the beacon service, cost $£ 55$ each.

The receivers, which are extremely light and easily portable, employ transistors; they can also be used as simple direction-finders as an aid to finding the beam-path: Agents for installation and maintenance are Coastal Radio, Ltd., Hope Crescent, Edinburgh, 7.

# Designing Decade Units <br> By C. D. LINDSAY, B.Sc. 

Simple Method of making<br>useful items of test gear

TRIAL and error is often the method by which the optimum value of a component for a particular circuit function is found. This is a simple matter in the case of resistors, but capacitors or inductors are more difficult to deal with. In most cases the usual way is to use a decade box. It is often thought that such a box is large and costly, but it need not be so, as it is now proposed to explain.

It is well known that if one has units of capacitance of $1,2,2$, and 5 , by suitable choice, any value from zero to 10 in steps of 1 may be formed. At first sight the switch to make this choice may appear to be somewhat complicated, but in point of fact it is possible to do this with a single wafer-type switch, actually much simpler than the wavechange switch in the average radio set. Moreover, no insulated contacts are used, as this would seriously increase the stray capacitances. In an actual decade the error caused by the switch amounted to only 13 pF .

For most practical purposes $10 \%$ tolerance components are satisfactory, but $5 \%$ selected capacitors are normally available at slightly increased cost from at least one well-known manufacturer (T.C.C.). Naturally a box using a number of decades of these components would be correct only to one significant figure.

Those who wish to construct a closer tolerance box, and have available a good quality bridge, may use silvered mica capacitors, in particular the type one eyelets together, and which are made by Johnson Matthey. These should be built up to slightly more capacitance than required (but not more than the capacitance of one plate above the value required). They should then be adjusted to the correct value by scraping and then dipped in hot wax. Remove from the wax after about 15 seconds and shake off excess wax. After cooling, adjust on the bridge again (they will have increased in value due to wax dipping). After final adjustment dip in wax for 3 to 4 seconds which will build up a thick protective coating. The value of the capacitor may, if it will show through the wax, be written in ink on the plate remote from the one on which any capacitance adjustment is to be made.

A single wafer switch intended for inductors or resistors (or anything that adds when joined in series) is shown here. Again it uses no insulated contacts. As it was not found possible to design such a switch using the more


# Transistor R.F. Amplifiers 

1.-Available Junction Types as High Frequency Circuit Elements

By D. D. JONES,* M.Sc., D.I.C.

JUNCTION transistors will probably find one of their main markets in battery-operated broadcast radio receivers such as the "personal" portable and car types. The reason for there being only a few of such receivers on the British market at present is that transistors for handling large output powers and for use in high-frequency circuits are not yet in largescale production. Transistors suitable for the highfrequency stages of receivers for the medium waveband are now in an advanced stage of development but it is unlikely that all the requirements of a v.h.f. f.m. receiver will be met for a considerable time.

Difficulties in obtaining high-frequency transistors inevitably mean that the existing types are used at higher frequencies than are intended by their designers. Since this state of affairs may last for some time, it is important to realize what parameters set a limit to the frequency at which the transistor can be used. It is equally important to consider the effect of these parameters on circuit design, in order that as good performance as possible can be obtained.

In one respect, junction transistors differ very considerably from thermionic valves. Whilst the valve is substantially free from internal feedback effects up to comparatively high frequencies, the transistor suffers from feedback at all frequencies. Because of this, the input and output circuits of a transistor amplifier are not isolated from each other as are the grid and anode circuits of a "common cathode." thermionic valve amplifier. At high frequencies this internal feedback, coupled with phase changes due to transit time effects in the transistor, can be very serious and can easily lead to an amplifier becoming unstable.

This part of the article discusses how to represent a transistor as a high-frequency circuit element, while Part 2 will show how the effects of internal feedback can be neutralized. As an example, the use of a p-n-p transistor, type GET4, in a $465-\mathrm{kc} / \mathrm{s}$ i.f. amplifier is considered. This transistor is really intended for use in amplifiers at much lower frequencies, but its use at $465 \mathrm{kc} / \mathrm{s}$ is fairly typical of the " marginal" application of transistors that is inevitable until newer types are readily available. The problem of automatic gain control will also be discussed in Part 2.
Probably the most suitable, if not the only, tool for studying the behaviour of transistors at high frequency is the a.c. equivalent network. This is a network used to represent the a.c. characteristics of a transistor under the chosen d.c. operating conditions; it is assumed that the d.c. bias circuit does not shunt or otherwise upset the a.c. circuit. Although many different networks have been suggested, it is found that the T type is as simple and convenient as any.

[^2]At low frequencies a junction transistor operated in a common-base amplifier arrangement can be represented as shown in Fig. 1. The current generator $\alpha i_{e}$ means that a voltage $\alpha r_{c} i_{e}$ is developed across the collector resistance $r_{c}$ when a current $i_{e}$ flows in the input circuit. The arrows indicate positive directions of currents and voltages.

At high frequencies it is necessary to change this network in a number of ways:-
(a) Because of transit time effects the value of the current gain factor, $\alpha$, decreases as the frequency increases. It is found that this can be expressed as
$\alpha=\frac{\alpha_{0}}{1+j f \mid f_{\alpha}}$ where $\alpha_{0}=$ value of $\alpha$ at low frequency and $f_{\alpha}=$ frequency where $|\alpha|=0.7 \alpha_{0}$. This value $f_{\alpha}$ is sometimes known as the " 3 dB down frequency ", or, more often, as the " alpha cut-off frequency." The above expression is by no means exact, but is found to be sufficiently accurate at frequencies up to about $0.5 f_{\alpha}$.
(b) It is necessary to take into account the capacitance $\left(\mathrm{C}_{e}\right)$ that appears across the collector-base $\mathrm{p}-\mathrm{n}$ junction.
(c) There is also a capacitance $\mathrm{C}_{e}$ across the emitter resistance, $r_{e}$. This is not a straightforward $\mathrm{p}-\mathrm{n}$ junction capacitance, but is due to transit time effects and is known as the " emitter diffusion capacitance." $\mathrm{C}_{e}$ may be deduced from the expression

$$
\mathrm{C}_{e}=\frac{1}{2 \pi f_{\alpha} r_{e}}
$$

$r_{e}$ being given by

$$
r_{e}(\text { ohms }) \sim \frac{25}{I_{e}}
$$

where $I_{e}$ is the emitter bias current in milliamps.
(d) The value of the base resistance is much lower at high frequency (e.g. $100 \mathrm{kc} / \mathrm{s}$ ) than at low frequency (e.g. $100 \mathrm{c} / \mathrm{s}$ ). The high-frequency resistance is known as the extrinsic base resistance, $r_{b o}$ (or, sometimes, $\left.\cdot r_{b}\right)_{\text {, }}$ and is the resistance of the section of germanium between the point where the base lead is soldered on to the germanium and the point where "the transistor proper" may be considered to start.
As an example, consider a GET4 transistor operated at a collector voltage of -12 volts and an emitter current of +1 mA .

$$
\begin{aligned}
r_{e} & =25 \Omega \\
\mathrm{C}_{e} & =0.006 \mu \mathrm{~F} \\
f_{\alpha} & =1.4 \mathrm{Mc} / \mathrm{s} \\
r_{c} & =1 \mathrm{M} \Omega \\
\mathrm{C}_{e} & =45 \mathrm{pF} \\
r_{b o} & =70 \Omega \\
\alpha_{o} & =0.975
\end{aligned}
$$

At $465 \mathrm{kc} / \mathrm{s}$ the value of $\mathrm{C}_{c}$ is such that it heavily shunts $r_{c}$; it may therefore be possible to ignore $r_{c}$ (compared with $\mathrm{C}_{c}$ ) except in circuits where $\mathrm{C}_{c}$ is part of a tuned circuit and where very high values of


Fig. 1. Low frequency equivalent circuit.


Fig. 2. High frequency equivalent circuit.
dynamic impedance are used. However, at $465 \mathrm{kc} / \mathrm{s}$, the reactance of $\mathrm{C}_{e}$ is roughly twice the magnitude of $r_{e}$ and in certain circuits it must be taken into account.

The high-frequency equivalent network thus becomes that shown in Fig. 2. This relates to the common-base arrangement but may easily be adapted for the case of the common emitter amplifier by simply rotating the arms of the T ; this is shown in Fig. 3. However, in the common-emitter arrangement, it is sometimes preferable to express the current generator that indicates amplification (i.e. the $\alpha i_{e}$ part) in terms of the current flowing in at the input terminals. The input current is now $i_{b}$ and by using the expression

$$
i_{b}+i_{c}+i_{e}=0
$$

it is possible to transform the network of Fig. 3 to that of Fig. 4.

Using these various equivalent networks it is possible to calculate input and output impedances when using various generator and load impedances, and also to determine the gain of the amplifier. They can also be used, and this is at least as important, to indicate how the internal feedback occurs and thus how to design circuits that neutralize its effects.

The importance of the various factors that limit high-frequency performance will now be considered in turn.

Alpha cut-off frequency.-The most obvious limitation is the reduction of $\alpha$ with increasing frequency. The effect of this in circuit performance depends on the values of load and generator impedances. As an example, consider a common-base amplifier working into a load $\mathrm{R}_{\mathrm{L}}$ which is very small compared with $r_{0}$. Using the equivalent network shown in Fig. 2 the input impedance at the emitter is then given approximately by the expression $Z_{i n}=Z_{e}+r_{b o}(1-\alpha)$ where $Z_{e}$ is the impedance of $\mathrm{C}_{e}$ and $r_{e}$ in parallel. Hence, if the amplifier is


Fig. 3. The circuit of Fig. 2-rototed to represent the common emitter arrangement.


Fig. 4. The circuit of Fig. 3 modified so that amplification is expressed in terms of $i_{b}$
supplied from a generator $\left(v_{g}\right)$ of resistance $r_{g}$, we find that the current flowing in the input loop is given by

$$
\mathrm{I}_{i n}=\frac{v_{a}}{r_{g}+\mathrm{Z}_{e}+r_{b o}(1-\alpha)}
$$

Since $\alpha$ decreases with increasing frequency, $Z_{\text {in }}$ will increase and hence $I_{\text {in }}$ will decrease. In order that this reduction in $\mathrm{I}_{i n}$, which results in a decrease in the gain of the amplifier, be kept as small as possible it is necessary that
(i) $r_{b o}$ be as low as possible.
(ii) $r_{g}$ be as high as possible.

The frequency $f_{\alpha}$ is determined by applying a constant-current signal into the emitter and varying the frequency until the voltage developed across a very low resistance (e.g. $300 \Omega$ ) in the collector circuit is reduced to 0.7 of its amplitude at low frequency. Thus $f_{\alpha}$ may be regarded as the bandwith of a common-base amplifier driven from a very high-impedance generator and working into a lowimpedance load. As was shown above, the effect of changes in $\alpha$ with frequency become more marked as lower values of $r_{g}$ are used. Since it is necessary, in a practical circuit, to use a value of $r_{g}$ of the same order of magnitude as $\mathbf{Z}_{\text {in }}$ in order to obtain high gain, the limitations due to the variation of $\alpha$ with frequency are•likely to occur at frequencies well below $f_{\alpha}$.

In fact, the bandwidth of an untuned amplifier working into a low value of $\mathrm{R}_{\mathrm{L}}$ is found to be given approximately by

$$
\left\{1-\frac{\alpha_{0} r_{b 0}}{r_{g}+r_{e}+r_{b 0}}\right\} f_{\alpha}
$$

The complex nature of $\alpha$ also means that $\mathbf{Z}_{\text {in }}$ has an inductive component.

Another case worth considering is the effect of
the variation of $\alpha$ with frequency on the performance of a common-emitter amplifier. The network in Fig. 4 shows that the current gain factor in this case is given by the factor $\alpha /(1-\alpha)$; this is often termed $\alpha_{e b}$. In this case the current gain cut-off frequency, $f_{a c b}$, is found to be given by

$$
f_{x c t}=\left(1-\alpha_{0}\right) f_{\alpha}
$$

Taking the GET4 example considered earlier, we obtain a value of $35 \mathrm{kc} / \mathrm{s}$. That is, the bandwidth of an untuned common emitter amplifier (measured, as before, from zero frequency) fed from a high generator impedance (constant-current source), and working into a very low load resistance, is given by $f_{\text {acb }}$. It may at first sight, therefore, be surprising that such a transistor as the GET4 should be considered at all for use in a common-emitter amplifier at $465 \mathrm{kc} / \mathrm{s}$.
However, for the common-emitter amplifier (with low $R_{L}$ ) the input impedance is given approximately by

$$
Z_{i n}=r_{b o}+\frac{r_{e}}{1-\alpha}
$$

and, hence, using a generator, $r_{g}$, the input current is given by

$$
\mathrm{I}_{i n}=\frac{v_{y}}{r_{\theta}+r_{b o}+\frac{r_{e}}{1-\alpha}}
$$

In this case, as the frequency is increased, $\alpha$ decreases as before but the effect now is to reduce the value of $Z_{i n}$ and hence to increase $I_{i n}$. As a result the bandwidth now becomes substantially greater than $f_{\alpha c b}$; the lower the value of $r_{g}$, the higher does the useful range of operating frequency extend. In practical amplifiers $r_{g}$ is comparable with $Z_{i n}$ and this results in a substantial increase in bandwidth.
Base resistance and collector capacitance.-In both the common-base and common emitter cases considered above, the effect of frequency variations of $\alpha$ on the frequency range of the amplifiers becomes more pronounced as $r_{b o}$ is increased.
Consider also the feedback due to $r_{b o}$ and $\mathrm{C}_{c}$. This may be seen from the network shown in Fig. 2. If
an a.c. voltage $\mathrm{V}_{1}$ is applied to the collector terminal, and the emitter terminal is a.c. opez circuited, a voltage $\mathrm{V}_{f}$ will appear at the emitter terminal. At a frequency $f, \mathrm{~V}_{f}$ will be given by

$$
\mathrm{V}_{g}=\frac{j 2 \pi f \mathrm{C}_{0} r_{b o}}{1+j 2 \pi f \mathrm{C}_{b} r_{b o}} \mathrm{~V}_{1}
$$

This expression shows that the product $\mathrm{C}_{\mathrm{e}} r_{b o}$ must be kept as low as possible if the resulting internal feedback is to be small. It also shows that it is the product of $\mathrm{C}_{c}$ and $r_{b o}$ that is important.

The parameter $r_{b o}$ depends largely on the resistivity of the germanium used and on the geometrical construction of the transistor; it is found to be substantially independent of the d.c. operating conditions.

The collector capacitance $\mathrm{C}_{c}$, on the other hand, is highly dependent on the operating conditions. Thus for a p-n-p transistor, such as the GET4, made by the "alloy" process, $\mathrm{C}_{c}$ varies with the d.c. collector voltage $\mathrm{V}_{c}$ as follows: $-\mathrm{C}_{c} \propto 1 / \sqrt{ } \mathrm{V}_{c}$. Thus if $\mathrm{V}_{c}$ is reduced from -12 volts to -6 volts in the example considered earlier, the value of $\mathrm{C}_{\mathrm{c}}$ is increased by $50 \%$.
Figure of merit.-It has been shown above that, in order to obtain good high-frequency performance, it is necessary that
(i) $f_{\alpha}$ be as high as possible.
(ii) $r_{b o}$ and $\mathrm{C}_{c}$ be as low as possible.

A useful performance criterion is given by what has become accepted as a " figure of merit " for transistor amplification. This is designated by the factor $M_{q}$, where
$\mathrm{M}_{g}=\frac{f_{\alpha}}{r_{b o} \mathrm{C}_{\mathrm{c}}}$, with $\mathrm{f}_{\alpha}$ expressed in $\mathrm{kc} / \mathrm{s}, r_{b o}$ in ohms and $\mathrm{C}_{c}$ in pF .
For a transistor having the values quoted above, $\mathrm{M}_{\mathrm{g}}$ is 0.44 .
The more practical aspects of designing transistor h.f. amplifiers, with particular reference to neutralization and a.g.c., will be discussed in Part 2 of this article.
(To be continued.)

## SHOIRT-WAVE CONIITIONS Prediction for October



# F.M. Receiver Design 

Methods of Improving Capture Ratıo to Combat Multi-path

## and Co-Channel Interference

By LaWRENCE W. JOHNSON*

CIONTRARY to expectations, frequency modulation has so far failed to bring about the revolution in broadcasting which has been predicted from time to time. F.M. was announced over twenty years ago; its advantages have been widely admitted, but nowhere has it replaced a.m. as the main broadcasting medium. The reasons are many, varied, and interesting; but in this article it is proposed to discuss one single contributory factor, namely the imperfect performance of f.m. receivers of what we may call " traditional" design.

Receivers at present used for f.m. reception are variants, almost without exception, of the basic design used by Major Armstrong in his tests in the middle thirties. On the one hand this might not seem too surprising, since a.m. receivers have undergone relatively little refinement in basic design in the same 20 -year period. Yet it is perhaps just this tendency to think about new things in terms of similar subjects with which we are already familiar that leads to such " follow-the-leader" situations. Much that has been added to the literature of f.m. receiver design in the last 20 years has yet to find its way into receivers in commercial production. It cannot be denied that Armstrong's work ${ }^{1}$ represented a thrilling example of creative engineering in the face of monumental scepticism on the part of organized radio, and it is no reflection on his work that the type of receiver he originated has since been shown to suffer from shortcomings, albeit subtle ones. But first let us review what has been accomplished in spite of them.

Before an f.m. broadcast service is established, field tests are generally made. Inspection of the available reports ${ }^{2,3}, 4,5$ of these field tests reveals an interesting point. Although they were carefully conducted and meticulously recorded, it is apparent that some of the conclusions can be questioned. It must be clearly realized that such conclusions are valid concerning only the whole combination of transmitter, propagition medium, and receiver, rather than purely and simply about f.m. as a basic system of broadcasting. It appears that this distinction has been lost. Consider an analogy with a.m. broadcasting; one would not perform an a.m. field test with a crystal set, or even with a 1936 t.r.f. receiver, yet the diligent reader can find examples of recent f.m. field tests using receivers which from all accounts appear to be no more than refined versions of Armstrong's original design, to which we

[^3]In America, where more than 500 f.m. broadcasting stations are in regular service, the problems of co-channel interference have lately directed attention to the inadequacy of "traditional" receiver designs.

This article presents the arguments which have led to the adoption of wideband discriminator circuits in some of the more advanced American commercial f.m. tuners, and outlines alternative techniques for improving f.m. reception under adverse conditions either of receiver siting or anomalous propagation.
have assigned, we hope without offence, the descriptive adjective " traditional." It is of considerable importance to note that f.m. has been adopted in spite of the receivers used.

No stigma should be attached to the incorrectness of conclusions mentioned above, for the history of radio shows that many of the foremost authorities in the field, past and present, have jumped to incorrect conclusions about f.m.'s capabilities. It seems, in fact, to have been a sort of occupational hazard for radio engineers and scientists, afflicting the great and small alike. Perhaps it still is.

Many vague and excessively inclusive claims have been made at one time or another for f.m. Some have been incorrectly stated and others have been idealized versions of what f.m. can do when receivers of advanced design are used; but all these claims, of whatever validity, are based on the capture effect, by virtue of which a signal effectively "takes over" at the f.m. detector if its amplitude exceeds the sum of the amplitudes of any other signals present there. Proper consideration of the relevant vector diagrams will yield the correct answer. The literature ${ }^{6,7,8}$ explores the subject in detail and we will do no more than outline some of the procedure by which the details of performance can be deduced.

Consider the rather special situation that exists when two signals of constant power are present on the same carrier frequency in nearly the same strength. Suppose that one signal has a field strength which yields one millivolt at the aerial terminals. while the other yields nine tenths of a millivolt, and that both are now frequency-modulated. Let us examine their vector sum during a period short compared to the highest modulating frequency. Of particular interest is the angular velocity of the resultant, $R$, since it is that angular velocity which carries the intelligence we are planning eventually to recover. At the same time we must not lose sight of the amplitude behaviour of the resultant, since that too must influence our design decisions. Suppose for the time being that we have available a limiter circuit which will accept the sum signal and yield an output of constant amplitude, which will of necessity have the same angular velocity-or instantaneous frequencycharacteristics as the original sum vector. For convenience we shall choose the one-millivolt vector as our reference in time, and we shall suppose that the other vector is slightly higher in frequency during the period of our examination, and so will be rotating anti-clockwise about our reference
vector. Fig. 1(a) shows the situation when the two vectors are pointing very nearly in the same direction; then the resultant is nearly 1.9 millivolts, and is rotating with an angular velocity relative to the $1-\mathrm{mV}$ vector only slightly different from half that with which the $0.9-\mathrm{mV}$ vector is rotating.
Some time later-approximately one half cycle of the frequency difference between the two component vectors-the situation is as shown in Fig. 1(b); here the two vectors are very nearly directly subtracting. Now the resultant is very nearly 0.1 mV , and of particular interest is the fact that its angular velocity relative to the $1-\mathrm{mV}$ vector is now something like nine times that with which the $0.9-\mathrm{mV}$ vector is rotating; but note that now the resultant vector is going clockwise, and so corresponds to an instantaneous frequency considerably below the frequency of the $1-\mathrm{mV}$ signal, while in Fig. 1(a) its direction of rotation and relative angular velocity corresponded to an instantaneous velocity slightly above the frequency of the $1-\mathrm{mV}$ signal. Thus the instantaneous frequency of the resultant varies over a considerable range during one cycle of the difference frequency, but since in the long run the resultant makes the same number of total revolutions as the longer vector, their average angular velocities must be equal. Also shown in Fig. 1(b) is the locus of the tip of the sum vector for one complete difference frequency cycle; this is to emphasize the extent of the amplitude variation during the difference frequency cycle.

## Frequency Spectrum After Limiting

Thus we see that the resultant instantaneous frequency goes through periodic variations at the difference frequency, its average frequency being that of the larger vector; and the amplitude also undergoes variations at the difference frequency. A plot of the instantaneous frequency versus time would show a series of sharp spikes whose maximum frequency deviation from the nominal carrier frequency can exceed by far the nominal $75-\mathrm{kc} / \mathrm{s}$ peak deviation. Prior to limiting, that is to say for all stages which do not limit, purposely or accidentally, a $150-\mathrm{kc} / \mathrm{s}$ bandwidth will nevertheless suffice for undistorted transmission of the resultant; for since such a bandwidth would suffice for either signal separately it will by superposition, valid for linear systems, transmit them equally well simultaneously. But once limiting has taken place, having the useful effect of removing the amplitude variations, we encounter as an unavoidable consequence a broadening of the frequency spectrum which means that succeeding stages must have the new wider bandwidth if they are to transmit faithfully the limited signal. The frequency spectrum for the general case of the limited signal requires a bandwidth wide enough to accept at least the highest angular frequency deviation which the process described above may bring about; and this bandwidth must be present in all circuits after the first nonlinear circuit, thus including, in terms of conventional design, the anode circuit of the first limiter, any subsequent limiters, and the detector.

For our purposes it will suffice here to record that the expansion of required bandwidth brought about by the ideal limiting of the resultant of two signals depends in a simple fashion upon the ratio of the magnitudes of the two signals.


Fig. 1. The resultant of two signals of nearly equal strength and frequency varies in "instantaneous frequency" as well as amplitude, but the average frequency is exactly equal to that of the stronger signal.

If we let $a$ be the ratio of the weaker signal strength to the stronger, and $B$ be the expanded bandwidth, if $f_{d}$ is the maximum permitted frequency deviation at the transmitter, we have the relation:

$$
\mathrm{B}=2 f_{\mathrm{d}} \cdot \frac{1+a}{1-a}
$$

If, for example, it is desired to receive the stronger of two signals when the weaker is $95 \%$ as strong as the former, in an f.m. system using $75 \mathrm{kc} / \mathrm{s}$ as the permissible peak deviation, the relation gives $5.95 \mathrm{Mc} / \mathrm{s}$ as the bandwidth required of the limiter and detector. This is quite a startling departure from the usual $150 \mathrm{kc} / \mathrm{s}$, to say the least. That this extended bandwidth is necessary, but not sufficient, will be discussed below. Before this result has a chance to discourage the reader, let us turn to the general question of the applicability of the rather special problem we started out with, involving two signals of nearly equal strength.

It is plain that our problem applies directly to co-channel interference; that is the problem associated with the reception of one or the other (or, in unfortunate cases, both) of two signals using the same carrier frequency. That it should apply to most other interference problems as well takes a little more explaining. Adjacent-channel interference, so important in a.m., cannot be ignored in f.m. The example above covers adjacent-channel interference if we note that the difference frequency would simply be bounded differently. In the cochannel case the difference frequency can vary from zero to $150 \mathrm{kc} / \mathrm{s}$, while in the adjacent-channel case it can vary from $50 \mathrm{kc} / \mathrm{s}$, when carriers of adjacent channels are modulated the maximum amount toward one another, up to $350 \mathrm{kc} / \mathrm{s}$, when they are modulated the maximum amount away from one another. Multi-path transmission, responsible for ghosts and aircraft interference on television, is perhaps for British listeners the most important of the types of interference; it arises when two paths of appreciably different length are possible, as in the case of reflection from a mountain, building, or airplane. When one observes that two versions of the same signal, one delayed, appear to the receiver much the same as two separate and distinct signals
on the same channel, it is apparent that the same situation regarding expansion of bandwidth applies for multi-path transmission interference.

Impulse noise goes a little farther afield; what happens is that the transient shock-excites the receiver's front - end circuits at their natural resonant frequencies. Since these frequencies are within the pass-band of the receiver, they are amplified along with the desired signal, and appear to the limiter and discriminator stages as discontinuous bursts of un-frequency-modulated co-channel signal. Thus for a period of time of the order of the ringing time of the front-end circuits the situation is roughly the same as that which we chose, with almost startling foresight, as our special example. In this connection we note that the conditions of maximum selectivity and minimum ringing time are mutually incompatible, making some sort of a compromise necessary.

It should be clear, then, that the problem brought about by spectrum expansion is substantially the same for all the kinds of interference mentioned above, which with their variations constitute a fairly complete list of the things that keep radio transmission from working properly or perfectly. And we have established certain bandwidth requirements which when met allow the distortionless reproduction of the patterns of frequency spikes mentioned earlier. At this point one may sensibly ask whether or not this is worth doing, granting that the proper cir-cuitry-rather fancy circuitry-can do it. This very question must have bothered some fairly highpowered authorities for some years; the principles behind the formation of the spikes had been well understood for some time before anyone decided that building a receiver with wide-band limiters and detector to do a good job of demodulating them might in fact be worth while. Until the work in the middle and late forties done by a group at M.I.T. under L.B. Arguimbau, it had apparently been concluded that the spikes represented unavoidable distortion that sufficed to prevent successful reception of the stronger of two signals of nearly equal strength. And it may be a holdover of this same feeling which accounts for the apparently wide-spread opinion that noise suppression suffers if the bandwidth is widened beyond, say, $200 \mathrm{kc} / \mathrm{s}$.

## Capture Ratio

The work of Arguimbau's group established that the actual audible distortion introduced by the spikes can be held to a very low degree, if the customary de-emphasis time constant is used at the receiver. Arguimbau's work, described in many publications, ${ }^{6,7,9,10,11}$ was directed primarily at multi-path problems, and in particular at investigating the possibility of trans-Atlantic communication via f.m. ${ }^{12,}{ }^{13}$ It is interesting to note that his tests revealed that the principal drawback in that application lies in the fact that in trans-Atlantic work a multiplicity of signals is involved, no one of which is greater than the sum of the rest; since the best that can be done today in receiver design requires that one signal exceed the sum of all others, success has not yet been achieved.

It is probably an over-simplification, perhaps a permissible one, to say that the principal conclusion of the M.I.T. work is that an f.m. receiver should have a good capture ratio. After explaining the term "capture ratio," we will discuss the capture ratio
of the "traditional" receiver when compared with that of receivers built according to the M.I.T criteria, following which there will be a few words. about the steps one takes to embody these criteria.

We spoke earlier of the quantity $a$, which was defined as the ratio of the weaker signal strength to the stronger. The largest value of $a$-that is, the nearest to unity-for which a receiver will provide an interference-free signal is that receiver's capture ratio. (This definition glosses over the question of just exactly when is a signal interference-free; since there seems to be good precedent for this neglect, no more will be said.) One may also encounter capture ratio expressed in decibels; this is obtained by taking the negative of 20 times the common logarithm of the quantity $a$, or alternatively, 20 times the common logarithm of the reciprocal of $a$. Bearing in mind that capture ratio is a quantity of importance in reduction of all types of interference, let us consider the capture ratios of the general run of f.m. receivers. Several references point out that in general it has been observed that the desired signal must exceed the undesired by some 20 to 30 decibels for noise-free reception. Thus we may deduce directly that the receivers used in those tests had capture ratios of no better than 20 dB , or 0.1 in the fractional notation. Preliminary tests made as part of the work at M.I.T. support these observations, so that it is quite safe to say that until Arguimbau's group fabricated the first wide-band f.m. receiver in the early forties, capture ratios of twenty decibels and more were the order of the day.

## Commercial Wide-band Receivers

Here we might pause and note that, while f.m. has been adopted by several nations in spite of the handicap under which it functions when receivers of poor capture ratio are employed, as far as can be determined no official field test has yet been conducted with receivers designed to take advantage of the M.I.T. research. It is fortunate that f.m. still surpassed a.m. even when forced, so to speak, to labour under an unfair handicap.

Earlier it was indicated that a receiver whose limiter and discriminator had bandwidths of 5.95 megacycles could possibly have a capture ratio of 0.95 , or in decibels, 0.45 dB . The above phraseology is intended to suggest that other requirements must be met as well; if, for example, the intermediate frequency pass-band is $x \mathrm{~dB}$ down $\pm 75 \mathrm{kc} / \mathrm{s}$ from channel centre, then the capture ratio cannot be better than $x \mathrm{~dB}$, even if the limiter and detector bandwidths are infinite. More about this later; suffice it to say that it is possible to build f.m. receivers with capture ratios as good as $\frac{1}{2}$ decibel. It is not easy, nor is it inexpensive, but such receivers are described by Arguimbau, Granlund, Paananen, and Cross. ${ }^{9,10,11,14.15}$

A short description of what is now commercially available along these lines may be of interest. One manufacturer, Radio Engineering Labs., intimately associated with Major Armstrong during his f.m. work, makes an adaptation of the $\frac{1}{2}$-decibel M.I.T. receiver; that this receiver should be the most expensive (over $\$ 300$ ) on the American market is easily understood when the reports on its ancestor are inspected. Two other manufacturers, H. H. Scott and the National Company, make less expensive receivers ( $\$ 100$ to $\$ 200$ ) embodying many of the characteristics recommended in the same and later
M.I.T. work. The capture ratios of these two are of the order of two decibels, while it appears that other manufacturers propose that the same amount of money should be paid for a receiver or tuner with a capture ratio of the order of 20 dB . Having no positive information about the characteristics of British or European f.m. equipment, the author would be pleased to think that the situation is more hopeful outside the U.S., but as yet no indication that this is so has been seen.
The bandwidth requirement for limiters and detectors has been dwelt on in some detail, and the requirement for flatness of i.f. amplifier pass-band touched upon. Further details along that line are contained in the literature; in passing it may be noted that if a receiver of infinitely wide limiter and detector bandwidths has an i.f. down 3 dB at $\pm 75$ $\mathrm{ke} / \mathrm{s}$, then, when it receives a signal at band centre which is (at the aerial) 3 dB below a signal $75 \mathrm{kc} / \mathrm{s}$ away from band centre, the two signals will be of equal amplitude at the discriminator, and the effort devoted to widening the bandwidth of the limiter and detector will have been wasted. Thus we can see the type of reasoning behind the requirement that there be negligible ripple in the i.f. passband, where negligible means a variation in response small compared to the relative sizes of signals it is desired to separate.

The bandwidth and flatness requirements bear equally on all types of interference, as can be seen from the preceding discussion. A special problem associated with adjacent-channel interference is that of selectivity; it should be clear that its minimization dictates a maximum of selectivity as early in the receiver as possible. This same conclusion presents itself as a means for admitting a minimum amount of wide-band noise.

## Ease of Tuning

An'extra dividend is gained through the use of a wide-band detector in conjunction with a flat-top steep-skirted i.f. This dividend, having nothing to do directly with interference suppression, is that the receiver is many times easier to tune than the " traditional" design with round-topped i.f., and detector bandwidth of the same order as the i.f. bandwidth. In the "traditional" design one encounters "threepoint tuning", which is an unavoidable consequence of the S-curve detector characteristic; thus the subsidiary linear sections of the S-curve on each side of the main linear section give rise to additional responses to the same station as one tunes on either side of the main response. These responses are generally weaker than the main response, are usually noticeably distorted, and can serve to confuse the operator. In addition there is the fact that the limited width of the main linear portion of the characteristic makes tuning for minimum distortion very critical with receivers of "traditional" design. The wide-band flat-top steep-skirt design, on the other hand, gives a tuning ease comparable to if not exceeding that encountered in a good a.m. receiver. This is by virtue of the fact that the discriminator characteristic, as modified by the i.f. response, is more like a letter N than an S on its side; thus the subsidiary responses are so very narrow, being associated with slope detection on the i.f. skirts, that they are heard only as noisy spots on each side of a broad area of undistorted reception. This
fortunate situation seems generally to make unnecessary automatic frequency control.

This must not be interpreted as justification for avoiding the building of a stable local oscillator. It is simply that a.f.c.'s chief reason for beingease of tuning-is no longer existent. Just as one should not design a sloppy audio amplifier, expecting to clean up its deficiencies later on with inverse feedback, neither should one depend on slipping by with an unstable receiver with the idea of covering up those deficiencies with a.f.c.

The considerable difference between frequency allocation policy in the United States and Great Britain places differing degrees of emphasis on the various types of interference. In many U.S. locations, particularly along the Atlantic seaboard, adjacent-channel and co-channel problems are serious, while there are probably a negligible number of such problems in Great Britain. On the other hand, multi-path transmission and impulse noise are not mitigated by careful and intelligent allocation planning, so that these problems are encountered in varying degrees in all parts of both areas, and thus provide good reason for the desirability of good capture ratio in f.m. receivers everywhere.

An interesting example of the importance of good capture ratio, suggested by B. G. Cramer ${ }^{16}$, is particularly applicable to the co-channel situation common in the U.S., but is of sufficient interest and importance to be included here. It involves the rather theoretical situation of two transmitters on the same channel situated, say, some 100 miles apart on a flat earth free from mountains or other reflecting bodies. This distance of separation of co-channel stations is a realistic one for the U.S., and so is the assumption that they have the same effective radiated power, made only for convenience. Finally, if we assume, also for convenience, that the receivers to be considered use non-directional aerials we can plot contours which enclose areas for which a receiver of a given capture ratio will receive the stronger signal without interference from the weaker. Fig. 2 shows this situation plotted for stations 100 miles apart when receivers of 0.1 capture ratio (small circles) and 0.9 capture ratio (large circles). One would have predicted that the contours were circles, and naturally enough they are not centred on the transmitting sites. The usable areas inside the contours are actually reduced in accordance with whatever figure we may choose as the maximum service range for an f.m. transmitter of a given power. If we choose 100 miles as that maximum range, the shaded areas shown join the excluded cross-hatched areas. There is still quite a difference between the service areas brought about by the capture ratio difference between the receivers used.

We will now proceed with a few notes on the means now available for achieving the ends described above. Regarding broad-band detectors, it may be noted that several varieties are successfully used. The simplest are nothing more than versions of the familiar ratio detector modified for bandwidths of the order of megacycles, while considerably more complicated designs are used to achieve the $6-\mathrm{Mc} / \mathrm{s}$ bandwidth mentioned earlier. Regarding the ratio detector, it should be remarked that its inherent limiting properties are a valuable adjunct to limiters which precede it, but that these limiting properties are not sufficient to do a good job when used alone in a receiver intended to have a good capture ratio.


Fig. 2. Service areas (stippled) for capture ratios of 0.1 and 0.9 with co-channel transmitters spaced 100 miles apart.

Other things being equal, as one widens the detector bandwidth, the audio output amplitude for a given deviation decreases; as a consequence one must supply more audio amplification than is customary in conventional designs, but in these days of highquality disc and tape recorder pre-amplifiers this presents no problem.

Another point in detector design is the choice of the intermediate frequency. On the chance that some may be daring enough to break away from the wellestablished $10.7 \mathrm{Mc} / \mathrm{s}$, we may note that it is only with considerable effort that a $6-\mathrm{Mc} / \mathrm{s}$ bandwidth can be achieved at the standard i.f.; this is understandable in view of the large fraction, some $60 \%$, of the centre frequency which the bandwidth represents. On the other hand, if an i.f. in the neighbourhood of $30 \mathrm{Mc} / \mathrm{s}$ were chosen, the problems associated with achieving a $6-\mathrm{Mc} / \mathrm{s}$ bandwidth are reduced considerably. It is a slight over-simplification to say that achieving $6 \mathrm{Mc} / \mathrm{s}$ at $30 \mathrm{Mc} / \mathrm{s}$ is just like achieving $2 \mathrm{Mc} / \mathrm{s}$ at $10.7 \mathrm{Mc} / \mathrm{s}$, yet it may be that in these days of considerable experience with television's rather high i.f.'s, a higher f.m. i.f. might be a very sensible choice. Other pros and cons enter; spurious responses are probably reduced, local oscillator interference problems may be increased, etc.

The actual design decision regarding capture ratio deserves careful consideration. Here an analogy may be drawn with the sometimes-hazy subject of the source impedance of audio amplifiers used to drive loudspeakers. That certain benefits are derived from driving the loudspeaker from a source impedance low compared to the loudspeaker impedance is well known; the trend which some years ago led from un-fed-back pentodes, with their source impedances considerably higher than the load, has now brought us to source impedances of the order of one tenth the load impedance. The point here is that very little is gained in lowering further the source impedance, say, to one hundredth the load impedance, since the gain in performance is likely to be imperceptible while the effort involved in accomplishing
the reduction is considerable. We are faced with a very similar situation regarding capture ratio; proceeding from the rather poor ratio of 0.1 up to a respectable 0.7 or 0.8 represents a great improvement in performance, while the additional effort required to extend the capture ratio to $0.9,0.95$, or even 0.99 would not be proportionately reflected in performance improvement, even though it would be interesting from a technical standpoint.

## Improved Limiters

Limiters play an exceedingly important role in the overall system; they must remove the violent variations in amplitude which occur from many causes, including the mixing of two nearly-equal signals, as set forth earlier. Because of the speed with which amplitude variation may take place, the limiter must be fast-acting. Such is not the case with conventional designs depending on gridcurrent and cut-off limiting with pentodes operated at low screen and plate voltages; recent designs have unanimously adopted other means for limiting. The simplest quick-acting limiter employs biased diodes, with care taken to assure that time constants in the bias source do not lead to the same recovery time troubles which hamper pentode limiters.

A means which does not present the signalattenuation disadvantage of the diode limiter is available through use of the gated-beam tube (6BN6) as a limiter. This tube was originally designed as a combination limiter-discriminator ${ }^{17,18}$, and is so employed in many television receivers. Its use as a detector in high quality receivers has not yet been reported, but when properly used it is unsurpassed as a limiter ${ }^{19}$. It has the fortunate property of depending on electron-optical beam-switching for its limiting action, rather than on grid-current biasing. Literature on its application is extensive but apparently not widely familiar. Interested experimenters might do well to investigate its use as a combined wideband limiter and discriminator. Of course, in all limiting means mentioned, care must be taken that the bandwidth appropriate for the desired capture ratio is maintained; in general such wide-band limiters have single-tuned low-Q circuits and are only broadly tuned.
I.F. and front-end design should follow general good design practice for low noise, with special attention to selectivity and flat-top characteristics. As mentioned earlier, it is important to obtain as much selectivity as early in the receiver as possible. Special attention should also be directed to the overload characteristics of the front end, with an eye to minimizing spurious responses. Note also that the later stages of the i.f. amplifier should not limit, for if they do, the spectrum is broadened with consequent possible distortion and degradation of capture ratio due to loss of sidebands. Both of these last two points indicate the desirability of an effective, fastacting automatic-gain-control system. If the a.g.c. is fast-acting enough, it will in fact be of considerable assistance to the limiters in maintaining a constant signal amplitude at the detector.

And now, in closing, a few remarks about some alternative schemes and some new developments. A device used successfully some years ago, which does not seem to have been exploited in the design of receivers of good capture ratio is the lockedoscillator detector ${ }^{20}, 21$. In short, this scheme locks
the frequency of an oscillator in the receiver which is normally operating at the i.f. to that of the received signal; it is the resulting variations in the frequency of this oscillator which are detected. An advantage of this system is its inherently perfect limiting, since the oscillator's output amplitude depends in no way on the incoming signal amplitude. For satisfactory locking it is obvious that the incoming signal would have to exceed some threshold, as is always the case with any detector. With suitable design the frequency excursions encountered under interference conditions can be handled by making the oscillator such that it cannot quite follow the extreme variations; it will thus perform, in effect, to limit bandwidth.

## Pulse-counting Discriminators

Counter-type detectors are frequently proposed for f.m. receivers ${ }^{22,}{ }^{23,}{ }^{24}$. Their advantages are considerable, the principal ones being excellent linearity over the design range, and in most cases, admirable simplicity. Since they customarily use a low i.f., of the order of $150 \mathrm{kc} / \mathrm{s}$, it is easily predictable that strange things must happen under interference conditions. If the instantaneous frequency were to head towards $150 \mathrm{kc} / \mathrm{s}$ below channel centre, the frequency into the counter would approach zero, which it cannot be expected to detect satisfactorily. And if the instantaneous frequency heads for a frequency more than $150 \mathrm{kc} / \mathrm{s}$ below channel centre, it is plain that the i.f. output into the counter will reverse phase at zero frequency and start back up again, giving rise to considerable distortion, since if the instantaneous frequency went $300 \mathrm{kc} / \mathrm{s}$ low, the counter would think that it was seeing the same $150 \mathrm{kc} / \mathrm{s}$ that corresponds to an unmodulated carrier. These results are somewhat analogous to overmodulation in an a.m. transmitter, or to partial carrier suppression in an a.m. receiver; the consequence is that the obtainable capture ratio is severely restricted by the use of the low i.f. that is dictated by practical considerations in counter-detector design. Use of higher i.f.'s in counter-detector circuits brings with it considerable complication; if money and size were no object, one might employ a digital frequency divider to proceed from the customary i.f. range down to the neighbourhood where a counter discriminator can operate conveniently. The deviation would then have been reduced by the dividing ratio, making good capture ratio possible with a second i.f. of the order of $150 \mathrm{kc} / \mathrm{s}$; other things held constant, the output voltage would be reduced by the same division factor.

A recent paper ${ }^{25}$ describes the theory behind a plan to accomplish interference rejection without recourse to the wide-band limiter-detector system discussed above. This suggested system, results of the experimental confirmation of which have not yet been reported, is an outgrowth of the M.I.T. work mentioned earlier. Its conclusion is that the wide-band scheme is, in the mathematical sense, sufficient but not necessary, the newer plan involves alternate stages of amplitude limiting and bandwidth limiting. Thus every time a limiter broadens out the spectrum by removing amplitude modulation, a steep-skirted bandpass filter reduces the bandwidth at least part way back to its original value. Cascading a succession of such ideal limiters and bandpass filters is shown to be capable of yielding a good capture ratio without the necessity of including
broad-band circuits. Pending further experimentation with this idea, one can be reasonably certain that the broad-band technique will do the job. Perhaps in a few years the broad-band techniques will have become the traditional techniques, with which the newer narrow-band system will be competing for recognition.

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## PIE-TEA

Regrettably Theoretical Refreshment

By "CATHODE RAY"

IY apologies for the title. The idea was to attract at least the momentary attention of the Billy Bunters of the electronic world. But perhaps there is small hope of expecting that attention to continue once I have confessed that the title really ought to have been " $\AA 1-\mathrm{T}$ " and that I am offering no facilities for raising $\pi$ to the $t$ th in any grossly materialistic sense.

This is not the first time it has been appropriate to recall the celebrated radio journalist who passed off a fairly conventional receiver circuit as something startlingly novel by the simple expedient of drawing It in an unfamiliar pattern. (One of the features that seemed to put it in a class by itself was the 1.t. battery in series with the aerial.) He was taking advantage of the very large extent to which our minds are guided by the way in which a circuit diagram is set out. In more recent times the late Mr . Bainbridge-Bell, who also realized this fact, made commendable use of it in his efforts to standardize a clear and logical style. But there are still differences between ourselves and the old heavy electrical engineers, who spend a lot of their time contemplating 3-phase power systems. Three impedances (or generators, or what have you), represented in Fig. 1 by oblongs, can be arranged symmetrically between three terminals in the two ways shown; the first is for obvious reasons called "star" or " $Y$ " or "wye," and the second "delta" or " $\Delta$ " or " mesh." We people, on the other hand, are more interested in filters and attenuators, which are examples of 4-terminal networks or "quadripoles." There are two basic forms of these, shown in Fig. 2 and known for equally obvious reasons as " $T$ " and " $\Pi$ " respectively.

Students who have had to become acquainted with both the two main divisions of electrical engineering, and therefore with both Figs. 1 and 2, may (unless it was pointed out to them at the time) have experienced an appreciable time lag in seeing that electrically these two pairs of circuit formation are identical, the differences being confined to the way in which they are drawn. Billy Bunter, for example, or Pilot Officer Prune (if either of them got as far as a telecommunications course) would probably need a soldering iron for changing a delta system into a $\Pi$, and thereby raise a cackle from the superior mortals who had seen the light five minutes earlier. We, being more sophisticated still, are supposed to have reached the stage where we are no longer misled by any apparent differences between Figs. 1 and 2 but instead are interested in (a) and (b) in either Fig. being electrically identical.

For purpose of discussion it is desirable to settle on one particular nomenclature, and although our
professional inclinations will naturally be towards Fig. 2, I am sufficiently conscious of the value of visual arrangement to admit that in this case the heavies have something. The triangular symmetry of Fig. 1 best matches the cyclic symmetry of its mathematics, as we shall see.

And so, although " $\Pi$-T" may have a more pleasurable sound for some, we are going to call our subject (embracing both Figs. equally) the Star-Delta Theorem. It states that any three impedances connected in one of these ways can be exactly replaced by three other impedances connected the other way, provided that their values are correctly chosen. It is only fair to add that the substitute impedances are liable to be awkward from a practical point of view-they may have to vary with frequency in a rather impossible manner, and resistances may come out negative. But at any one frequency the most that is needed to make each of the three "arms" is one resistance in series or parallel with one reactance.

The next thing is to find the formulx for the substitute impedances. To do this we must have symbols for the three impedances in each formation. The general symbol for impedance being $Z$, the usual practice is to distinguish the six concerned here by subscripts- $Z_{1}, Z_{a b}$, or whatnot. But seeing that " $Z$ " is common to all six, the information conveyed by that symbol is nil*, so it is redundant. It is much quicker and simpler and less liable to error (as I am sure the printer will agree) to label them simply $a, b$ and $c$ in the star and $\mathrm{A}, \mathrm{B}, \mathrm{C}$ in the delta, and away with anyone who grumbles that C stands for capacitance.

To preserve the symmetry, the $a$ in the star should

* If this proposition is new to you, consult any book on Information Theory, or "Cathode Ray " in the Sept. 1952 issue.

(a)

(b)

Fig. 1. Diagrams like this (or drawn with inductances in place of the oblongs) are common in books on 3-phase electric supply systems. They are called the star $(a)$ and delta (b).


Fig. 2. These diagrams of the $T(a)$ and $n(b)$ networks are common in books on communication circuits, and of course are electrically the same as those in Fig. I.
come opposite the $A$ in the delta, and so on, as in Fig. 3.

One procedure is to reckon the impedance between each of the three pairs of terminals in turn in both arrangements. Then, if the two arrangements are electrically the same, their corresponding impedances must be the same, which we can express by equating them. The impedance between terminals 1 and 2 is $b+c$ for the star and $A$ in parallel with $B+C$ for the delta. We are going to make much use of the rule that to find the result of two impedances in parallel you divide their product by their sum. So

$$
b+c=\frac{\mathrm{A}(\mathrm{~B}+\mathrm{C})}{\mathrm{A}+\mathrm{B}+\mathrm{C}}
$$

The same method applies to the other two pairs of terminals, with the results

$$
\begin{aligned}
& c+a=\frac{B(C+A)}{B+C+A} \\
& a+b=\frac{C(A+B)}{C+A+B}
\end{aligned}
$$

This is an example of the mathematical cyclic symmetry I mentioned; having one equation, we can arrive at the others by moving all the letters one place round each time, as in the Mad Tea Party

Now if we add the last two lots together (to give $c+a+a+b$ ) and then deduct the first, we eliminate $b$ and $c$ and leave $2 a$, thus:

$$
\begin{aligned}
2 a & =\frac{\mathrm{B}(\mathrm{C}+\mathrm{A})+\mathrm{C}(\mathrm{~A}+\mathrm{B})-\mathrm{A}(\mathrm{~B}+\mathrm{C})}{\mathrm{A}+\mathrm{B}+\mathrm{C}} \\
& =\frac{\mathrm{BC}+\mathrm{BA}+\mathrm{CA}+\mathrm{CB}-\mathrm{AB}-\mathrm{AC}}{\mathrm{~A}+\mathrm{B}+\mathrm{C}} \\
& =\frac{2 \mathrm{BC}}{\mathrm{~A}+\mathrm{B}+\mathrm{C}} \\
\text { So } a & =\frac{\mathrm{BC}}{\mathrm{~A}+\mathrm{B}+\mathrm{C}}
\end{aligned}
$$

In exactly the same way (or by using cyclic symmetry as a short cut) we find that

$$
b=\frac{C A}{A+B+C}
$$

$$
\text { and } c=\frac{\mathrm{AB}}{\mathrm{~A}+\mathrm{B}+\mathrm{C}}
$$

So if we know the three impedances A, B and C in a delta we can calculate the correct values for the three impedances $a, b$ and $c$ to make a star that is electrically the same.

Although the foregoing, or its equivalent, is the treatment given in some books, it is really more by luck than judgment that it happens to be right. The only thing we have established is that the two arrangements related by the formulæ are equivalent as viewed between two of their terminals, the third being unconnected. But that is not how stars or deltas are normally used. The usual thing is to use one pair of terminals as the input and another as the output, and the impedance measured at one pair depends on what is connected between the other pair. It is not obvious that the formulæ we have derived for infinite output impedance would hold good for any impedance. When output impedances are brought into the problem, simple algebra is a clumsy tool for solving it, and I will refer you to the more advanced books for the full
treatment, which by good fortune does give the same result.

To be fully armed, we need formulæ to perform the reverse transformation; i.e., given $a, b$ and $c$, to find the equivalent $A, B$ and $C$. One way of arriving at this second set of formulæ is to follow the same procedure as for the first, but shortcircuiting one of the unused pairs of terminals in each arrangement. Alternatively, the second set can be derived from the first by algebra, but a little less easily than you might think. In any case, I consider it neater to call in the duality principle.

This was explained in the April 1952 issue, but as that is rather far back I had better just mention that equations which are true of one electrical circuit also hold if they and the circuit are systematically turned upside down. Among other things, series is replaced by parallel, open-circuit by short circuit, and impedance by admittance.

Now the delta is the dual of the star, and if the same procedure as before is followed, short-circuiting the output instead of open-circuiting it, and working in admittances instead of impedances, the form is exactly the same. I suggest you try it for yourself to make sure; what I am going to do is assume that duality works and use it to transform a sample equation in our existing set. It is

$$
a=\frac{\mathrm{BC}}{\mathrm{~A}+\mathrm{B}+\mathrm{C}}
$$

Interchanging large and small letters and standing them all on their heads, we get

$$
\frac{1}{\mathrm{~A}}=\frac{\frac{1}{b c}}{\frac{1}{a}+\frac{1}{b}+\frac{1}{c}}
$$

From which $\mathrm{A}=b c\left(\frac{1}{a}+\frac{1}{b}+\frac{1}{c}\right)=\frac{b c}{a}+c+b$


Fig. 3. The impedances in the two formations will be denoted by these letters.

(a)

(b)

Fig. 4. These two attenuators are completely interchongeable.


Fig. 5. By substituting a delto (b) for the stor in (a), we find that the same job can be done with one resistor fewer.

This doesn't look very symmetrical, but it can be made to do so by writing it a little more fully:

$$
\mathrm{A}=\frac{a b+b c+c a}{a}
$$

Encouraged by the fact that either of these is correct by the reference books, we write down the other two simply by cyclic symmetry:

$$
\begin{aligned}
& \mathbf{B}=\frac{c a}{b}+a+c \\
& \mathbf{C}=\frac{a b}{c}+b+a
\end{aligned}
$$

A still neater use of the duality principle would be simply to say that as the two arrangements are duals of one another the second set of formulæ is exactly the same as the first except that the letters represent admittances instead of impedances.

It would be a pity if, having equipped ourselves with these two sets of formulæ, we were to find no use for them. Even in Mr. Squeers' system of education, theory was followed by practice. So the rest of the session will be devoted to this aspect of the matter.

Suppose an increase in power of the local television signal has made it necessary to cut it down at the receiving end, say by 10 dB in a 75 -ohm aerial line. We may have been informed that this can be done by inserting between the line and the receiver a $T$ attenuator (Fig. 4(a)) in which the horizontal arms are $39-\Omega$ resistors and the upright one $53 \Omega$. But if we had no resistors as low as this we might try the $\Pi$ form (b) instead. The uprights are what we have been denoting by C and B in Fig. 3, and substituting $b=c=39$ and $a=53$ we get

## Similarly <br> Silarly

$$
B=C=53+53+39=145
$$

$$
A=\frac{39^{2}}{53}+39+39=107
$$

The "preferred values" $150 \Omega$ and $100 \Omega$ would do quite nicely, giving an attenuation of 9.6 dB at $75 \Omega$.

In this case it would be just as easy to find the values directly from the formula for matched $\Pi$ attenuators, provided of course we happened to have it. However; it makes a nice simple example

Fig. 6. The ordinary series and parallel additions can be used to find the single resistance equivalent to networks

like (o), but

(a)

(b)

Fig. 7 . . . they fail when applied to a bridge circuit (a). However, the star-delta transformation succeeds (b).
of the use of our Star-Delta Transformation. In so far as all the arms of the system can be regarded as pure resistance, the two arrangements are truly interchangeable at all frequencies. At Band-III frequencies, however, we have to be careful about the type of resistor we use; as little as 2 pF stray capacitance is about $400 \Omega$ reactance, so is not a negligible path in comparison with 100 or $150 \Omega$ Not that it would be likely to cause trouble in such a non-critical application as a TV attenuator.

It is interesting to note in passing that the equivalent of a star in which all the arms are equal is a delta in which all the arms have three times the im-pedance-as can easily be discovered by using the transformation formulæ.

Another very simple purely resistive example was given by $W$. Tusting in the November 1954 issue, p. 552, where he showed that if one wished to run a resistance-coupled valve from a tapped-off point on the h.t. supply, as in Fig. 5(a), it was really unnecessary to use three resistors, because if the star system they form is transformed into its equivalent delta (b) it turns out that one of the resistors $\left(\mathrm{R}_{6}\right)$ serves no useful purpose and can be omitted.

Most of the elementary books on electricity, after they have dealt with Ohm's law and series and parallel arrangements, show how the currents, etc., in complicated networks can be calculated by successive reduction. For instance, $R_{1}$ and $R_{2}$ in Fig. 6(a) can be replaced by $\mathbf{R}_{12}\left(=\mathbf{R}_{1} \mathbf{R}_{2} /\left(\mathbf{R}_{1}+\right.\right.$ $\mathbf{R}_{2}$ ) at (b), and this in turn can be added to $\mathrm{R}_{3}$ to give simple $R$ at (c). But with some circuits this doesn't work. Perhaps the most important of these exceptions is the bridge network, Fig. 7(a). Suppose we wanted to find the resistance of this bridge as a whole, between the terminals 1 and 2. The difficulty is $R_{5}$, because it is not directly in


Left:-Fig. 8. This lattice circult is just a different way of drawing the bridge-Fig. 7(a). Right:-Fig. 9. The bridged $T$ is onother network thot can be simplified by the star-delta transformation.


Left:-Fig. 10. And so is the parallel $T$ or twin $T$, of which this is an example. Right:-Fig. II. The equivalent path between the top two (" live ") terminals of Fig. 10 is shown here; it consists of a tuned circuit with infinite impedance at resonance.
paraliel with any of the other arms. But if we spot that with $R_{3}$ and $R_{4}$ (or with $R_{1}$ and $R_{2}$ ) it makes a delta, we have only to transform it into its equivalent star, as in Fig. 7(b), which (after adding two of the new arms to $R_{1}$ and $R_{2}$ ) is the form we handled successfully in Fig. 6.

Incidentally, this is another example of how the same circuits can be made to look different, according to who is drawing the diagrams; or even the same person can draw them differently and not realize that they are the same. In books on electrical measurements one could hardly fail to find Fig. 7(a) -the familiar diamond-shaped Wheatstone bridge. In books on filters and attenuators one could hardly fail to find the lattice network (Fig. 8). Yet these are identical. Personally I consider Fig. (7)a much clearer than Fig. 8-it is easier to see that the attenuation can be made infinite, and what are the conditions for this-but perhaps that is because I studied Wheatstone bridges before attenuators.

Another commonly used arrangement that defies the series-parallel treatment is the bridged-T (Fig. 9). I will not go into this in detail, because Mr. Tusting did so in the article just referred to, but will just point out that the upper part of it forms a delta. When transformed into the equivalent star or T, it combines with its original "stalk" to form a simple T. Alternatively, the lower part of Fig. 9 is a T which can be transformed into a $\Pi$ into which the original "roof" can be merged by paralleling.

In practice this sort of network nearly always has reactive arms, as in the Tusting example. This makes no difference to the formulx, but in place of simple arithmetic we have to use the so-called "complex" algebra whose mysteries I attempted to dispel in the February 1953 issue. $\dagger$

[^4]As an example of this let us take a slightly more complicated network-the parallel-T, èmbodied in a distortion meter described by V. f. Tyler in the September 1953 issue. In this particular case, Fig. 10, half its arms were resistors and half capacitors, and their ratios were as shown. The object of the thing was to remove entirely the fundamental of the signal being investigated and allow the harmonics to go on and be measured by a suitable valve voltmeter. So it must have infinite attenuation at some selected frequency and as little as possible at all others. Hence the need for reactances, to discriminate between one frequency and others. Like a bridge, which also can be adjusted to balance out the signal but has the disadvantage that there is no earthable terminal common to input and output, it is called a " null" network.

As the name "parallel T" suggests, it consists of two T filters in parallel between the usual input and output terminals. But since no arm is in simple series or parallel with any other, it completely defies straightforward calculation. However, by transforming both Ts into $\Pi \mathrm{s}$, each of the three arms in one $\Pi$ comes directly in parallel with the corresponding arm of the other so can be merged with it to form a single $I$. And if you are wondering why Mr. Tyler didn't call for this single $\Pi$ (or its equivalent single T ) in the first place, instead of his complicated double T , wait and see.
Suppose what we want to find is the correct value of C for rejecting a given frequency completely. That means the C that makes the impedance of arm A in Fig. 3 infinitely large. This arm will consist of two $\Pi$ arms in parallel, say $A_{1}$ and $A_{2}$. Then

$$
A=\frac{A_{1} A_{2}}{A_{1}+A_{2}}
$$

and this is infinite if $\mathrm{A}_{1}+\mathrm{A}_{2}=0$ (assuming $A_{1} A_{2}$ is not zero). Now as regards the capacitors in Fig. 10 it is the reactance of


Fig. 12. The " bolted lattice " networkanother that can be simplified by the stardelta transformation. C that concerns us, which is $1 / j \omega C$. Using our transformation and remembering that $j^{2}=-1$ we have

$$
\begin{aligned}
& \mathrm{A}_{1}+\mathrm{A}_{2}=\frac{b_{1} c_{1}}{a_{1}}+b_{1}+c_{1}+\frac{b_{2} c_{2}}{a_{2}}+b_{2}+c_{8} \\
&= 2 \mathrm{R}^{2} j \omega \mathrm{C}+\mathrm{R}+\mathrm{R}+\frac{2}{\mathrm{R}(j \omega \mathrm{C})^{2}}+\frac{1}{j \omega \mathrm{C}}+\frac{1}{j \omega \mathrm{C}} \\
&=2 \mathrm{R}^{2} j \omega \mathrm{C}+2 \mathrm{R}-\frac{2}{\mathrm{R} \omega^{2} \mathrm{C}^{2}}-j \frac{2}{\omega \mathrm{C}}
\end{aligned}
$$

To make this add up to nothing, the resistive and reactive parts must separately cancel out, i.e.,

$$
\begin{aligned}
& 2 \mathbf{R}^{2} j \omega \mathrm{C}-j \frac{2}{\omega \mathrm{C}}=0 \\
& \text { and } \quad 2 \mathrm{R}-\frac{2}{\mathrm{R} \omega^{2} \mathrm{C}^{2}}=0
\end{aligned}
$$

In both cases this happens when $R=1 / \omega C$; in other words, when the frequency is such as to make the reactance of $C$ equal to $R$. In the distortion meter
the middle value of R was $15 \mathrm{k} \Omega$, so for rejecting $1 \mathrm{kc} / \mathrm{s}$ (say) the reactance of C must be $15 \mathrm{k} \Omega$ at $1 \mathrm{kc} / \mathrm{s}$. That works out at about $0.01 \mu \mathrm{~F} . \ddagger$

Fig. 11 shows the arms $A_{1}$ and $A_{2}$ in parallel, and answers the question about the point of using a double T. As our calculations show, the reactance in $A_{1}$ is $2 R^{2} j \omega C$, which varies directly with frequency, so is equivalent to an inductance equal to $2 R^{2} \mathrm{C}$. With $\mathrm{R}=15 \mathrm{k} \Omega$ and $\mathrm{C}=0.01 \mu \mathrm{~F}$, that is 4.5 henries; and such an inductor alone would cost and weigh more than the whole double $T$. And what about $A_{2}$, which contains a negative resistance! To give us the equivalent of an easily tunable audio-frequency resonant circuit with infinitely high $Q$, this double $T$ is a masterpiece of economy.

A closely related null network is shown in Fig. 12§. You may like to work it out in detail. The procedure is exactly the opposite of that followed with Fig. 10: the two deltas in series are transformed into stars, and the null occurs when the two vertical arms in series with one another, but shunted across the transmission path, add up to zero impedance. The answer (to check your working) is when (X/R) ${ }^{2}$ $=\left(1-n^{2}\right) / n^{2}$, which comes to much the same as with Fig. 10 (i.e., $|\mathrm{X}|=\mathrm{R}$ ) when $n \gg 1$.

The logical conclusion of all this is the theorem that any three or four terminal network of linear impedances can be reduced to an equivalent star or delta, and that its component values can be found by measuring any three of the four following: input impedance with output open and short-circuited; output impedance with input open and shortcircuited. It is too late now to start on these aspects of the matter, but details can be found in the books on communication circuits.

```
\ddagger If 1/(2\pi\times1,000\timesC)=15,000,C=1/(2\pi\times1,000\times15,000)=1.06\times
10-8F=0.0106 \mu\textrm{F}.
    § Due to E. M Reid, Electronic Engineering, Oct. 1954, p. }445
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## Radio Officers' Examinations

A CHANGE in the procedure and standards in the examinations for the P.M.G.'s certificate of competence in radio-telegraphy (1st and 2nd Class) is being introduced for the next exam which will be held in December. In the past, examiners have toured the country holding examinations at all the wireless schools
in turn. Under the new arrangement exams will be held three times a year-March, July and Decembersimultaneously at all schools where suitable equipment, approved by the P.M.G., is installed. A list of these schools is given below.

Under the new scheme the examinations for the P.M.G.'s certificate, which has to be obtained by all radio officers serving in British merchant ships, will be more comprehensive and the standard in general has been raised.

The syllabus of instruction for both the first and second class certificates is set out in two parts: the first, headed "technical electricity," and the second "radio communication."

The examinations for both certificates will be conducted in two parts: the first, theoretical; and the second, practical and manipulative.

The revised details for the examinations are being incorporated in the "Handbook for Wireless Operators" issued by the Postmaster General (H.M.S.O. 3s). Further information on the syllabuses is obtainable from the principals of the colleges listed below.

## ABERDEEN

Marine Radio College, 56 Union Street
BELFAST
Marine Radio College, 2 Eglantine Avenue, Lisburn Road
BRIDLINGTON
North Eastern School of Wireless Telegraphy, Radio House, Shaftes. bury Road
CARDIFF
College of Technology and Commerce, Cathays Park
COLWYN BAY
Wireless College, East Pariade
GLASGOW
Glasgow Wireless College, 26 Newton Place, C. 3
GREENOCK
Watt Memorial School, Dalrymple Street
GRIMSBY
Grimsby College of Further Education, Nautical Department, Orwel! Stree
HULL
College of Technology, Park Street
LEITH
Leith Nautical College, Edinburgh, 6
LIVERPOOL
Riversdale Technical College, Riversdale Road
LONDON
British School of Telegraphy, 179 Clapham Road, S.W. 9
London Telegraph Training College, 20 Penywern Road, S.W. 5 Norwood Technical College, Knight's Hill. S.E. 27
MANCHESTER
Wireless Telegraph College, 25 John Dalton Street
College of International Marine Radiotelegraphic Communication, Overseas House, Brook's Bar
PLYMOUTH
Plymouth and Devonport Technical College, Tavistock Road
PRESTON
Northern Counties Wireless School, 91 Lancaster Road SOUTHAMPTON

The University, Dept. of Electronics, Telecomms. and Radio The School of Marine Radio and Radar, Hamble
SOUTH SHIELDS
South Shields Marine and Technical College, Ocean Road


Typical marine radio equipment installed in the new examination room at G. P. O. headquarters. Radio officers employed at sea wishing to obtain a higher grade of certificate than that already held will be examined here. On the bench are, left to right, Marconi direction finder; Slemens emergency transmitter, receiver and automatic keying device; Marconi "Mercury " receiver; "Electra" receiver; Marconi transmitter; I. M. R.'"Solas" life-boat equipment; and Marconi automatic alarm.

# Wide-Range Audio Oscillator 

Simple Design Covering a Frequency<br>Range of $3 \mathrm{c} / \mathrm{s}$ to $300 \mathrm{kc} / \mathrm{s}$

By R. WILLIAMSON

0N a number of occasions recently the writer has been unhappily reminded of the inadequacies of his present audio oscillator (vintage 1950). A few years ago a range from $25 \mathrm{c} / \mathrm{s}$ to $25 \mathrm{kc} / \mathrm{s}$ was considered quite adequate for the vast majority of audio testing applications. Nowadays, frequencies used in audio equipment and for the testing of such equipment may extend over a very wide range. Magnetic bias frequences are often as high as $120-150 \mathrm{kc} / \mathrm{s}$ and many wide-band low-frequency amplifiers have a measurable response from perhaps a couple of cycles to over $200 \mathrm{kc} / \mathrm{s}$.

For these reasons it was decided to construct a new oscillator fully capable of use in testing all audio equipment of contemporary design. The functions desirable in the new design were carefully considered and are as follows:-
(a) The range should be from $3 \mathrm{c} / \mathrm{s}$ to $300 \mathrm{kc} / \mathrm{s}$ in five decade steps with overlap between ranges.
(b) Output to be pure sinusoidal with total harmonic content no greater than $1 \%$.
(c) Amplitude linearity should be no worse than $\pm 0.5 \mathrm{~dB}$ over the entire range and no preset controls should be necessary.
(d) Finally, the unit should be fully self-contained, portable and on economic grounds of simple circuitry using standard components.

In the prototype model the above specification

was satisfied completely, and in some instances improved upon. The unit is small, compact and of simple design. The total harmonic distortion on range 3, measured at $2 \mathrm{kc} / \mathrm{s}$ was considerably less than $1 \%$. Linearity was $\pm 0.25 \mathrm{~dB}$ from $330 \mathrm{kc} / \mathrm{s}$ to $15 \mathrm{c} / \mathrm{s}$ and 0.75 dB down at $3 \mathrm{c} / \mathrm{s}$. In fact, it proved possible to calibrate the output control directly in volts, thus obviating the necessity for a metered output stage. The maximum output voltage was 25 volts R.M.S. and the frequency drift checked after continuous operation for 3 hours was negligible.

## Maintained Accuracy

It was decided to use the Wein bridge type of oscillator. For optimum performance in accuracy and frequency stability the variable element should be a ganged capacitor. But, unfortunately, to cover the specified range would mean either an abnormally low bridge input impedance on the highest ranges or unwieldly values of resistance on the lowest ranges. In one model constructed a carbon track type of ganged potentiometer was employed, and it is still accurate to the original calibration after three months' use, with the additional advantage of $300^{\circ}$ scale coverage. However, if the constructor considers the extra expenditure worth while a wirewound ganged potentiometer gives frequency stability comparable with variable capacitors.

## LIST OF MAIN COMPONENTS WITH VALUES

## Capacitors

| $\mathrm{C}_{1}$ | $0.47 \mu \mathrm{~F}$ |
| :--- | :--- |
| $\mathrm{C}_{2}$ | $0.047 \mu \mathrm{~F}$ |
| $\mathrm{C}_{3}$ | $0.0047 \mu \mathrm{~F}$ |
| $\mathrm{C}_{4}$ | 470 pF |
| $\mathrm{C}_{5}$ | 47 pF |
| $\mathrm{C}_{6}$ | $0.47 \mu \mathrm{~F}$ |
| $\mathrm{C}_{7}$ | $0.047 \mu \mathrm{~F}$ |
| $\mathrm{C}_{8}$ | $0.0047 \mu \mathrm{~F}$ |
| $\mathrm{C}_{9}$ | 470 pF |
| $\mathrm{C}_{10}$ | 50 pF compression trimmer |
| $\mathrm{C}_{11}$ | $0.25 \mu \mathrm{~F}, 250 \mathrm{~V}$ |
| $\mathrm{C}_{12}$ | $32 \mu \mathrm{~F}, 250 \mathrm{~V}$ |
| $\mathrm{C}_{13}$ | $100 \mu \mathrm{~F}, 6 \mathrm{~V}$ | Tolerances-


| $\mathrm{C}_{14}$ | $0.25 \mu \mathrm{~F}, 250 \mathrm{~V}$ |
| :---: | :---: |
| $\mathrm{C}_{15}$ | $0.25 \mu \mathrm{~F}, 250 \mathrm{~V}$ |
| $\mathrm{C}_{16}$ | $60 \mu \mathrm{~F}\} 350 \mathrm{~V}$ |
| $\mathrm{C}_{17}$ | $60 \mu \mathrm{~F}$ |
| $\mathrm{C}_{18}$ | $50 \mu \mathrm{~F}, 50 \mathrm{~V}$ |
| Res | istors |
| $\mathbf{R}_{1}$ $\mathbf{R}_{2}$ | $100 \mathrm{k} \Omega+100 \mathrm{k} \Omega^{*}$ |
| $\mathrm{R}_{3}$ | $10 \mathrm{k} \Omega, 2 \%$ |
| $\mathrm{R}_{4}$ | $10 \mathrm{k} \Omega, 2 \%$ |
| $\mathrm{R}_{5}$ | $4.7 \mathrm{k} \Omega$ |
| $\mathrm{R}_{6}$ | $100 \mathrm{k} \Omega$, $\frac{1}{2} \mathrm{~W}$ |
| $\mathrm{R}_{7}$ | $560 \mathrm{k} \Omega$ |

*Ganged; log law (Reliance Mfg. Co.).
$\mathrm{R}_{8} \quad 15 \mathrm{k} \Omega, 1 \mathrm{~W}$
$\mathbf{R}_{9} \quad 330 \Omega, \frac{1}{2} W$
$\mathbf{R}_{10} 100 \mathrm{k} \Omega$
$\mathbf{R}_{11} 10 \mathrm{k} \Omega$, wire-wound
$\mathrm{R}_{12} 560 \mathrm{k} \Omega$
$\mathrm{R}_{13} 22 \mathrm{k} \Omega$
$\mathrm{R}_{14} 2 \mathrm{k} \Omega$, wire-wound, 3 W
$\mathrm{TH}_{1}$ S.T.C. Thermistor A2552/100
$\mathrm{T}_{1} 250 \mathrm{~V}, 30 \mathrm{~mA} ; 6.3 \mathrm{~V}, 1 \mathrm{amp}$.
$\mathrm{MR}_{1} 250 \mathrm{~V}, 50 \mathrm{~mA}$. contact-cooled type.
$\mathrm{S}_{14,}$, в 5-way, 2-pole Yaxley.


Circuit diagram of the oscillator. Frequency ranges are: 1, $3 \mathrm{c} / \mathrm{s}-33 \mathrm{c} / \mathrm{s} ; 2,30 \mathrm{c} / \mathrm{s}-330 \mathrm{c} / \mathrm{s} ; 3,300 \mathrm{c} / \mathrm{s}-3.3 \mathrm{kc} / \mathrm{s}$; $4,3 \mathrm{kc} / \mathrm{s}-33 \mathrm{kc} / \mathrm{s}$; $5,33 \mathrm{kc} / \mathrm{s}-330 \mathrm{kc} / \mathrm{s}$.

The operation of this type of RC oscillator is adequately dealt with in detail in the majority of text books, but a brief recapitulation would not be out of place. The circuit elements to determine the frequency of oscillation are in the form of a potential divider, the upper arm consisting of capacitance and resistance in series and in the lower arm identical values of resistance and capacitance in parallel. At a certain frequency attenuation is at a minimum of 9.54 dB and if the potential divider is connected in the regenerative loop of an amplifier stage with a gain slightly higher than this, oscillation will take place. Provided overloading of the amplifier does not occur, the harmonic content of the generated waveform can be of a very low order. To satisfy the stringent requirements of the amplifier stage, i.e., wide bandwidth, low distortion and minimum phase shift, a high degree of negative feedback is employed, the feedback attenuator completing the familiar Wein bridge. To compensate for minor tracking errors and maintain oscillation it is usual practice to include a non-linear element in the degenerative loop, and in this design an S.T.C. Type A thermistor is used. The oscillator has a buffer output stage in the form of a cathode follower and is partially d.c. coupled to maintain the amplitude of the lowest range. With this method of output control used the source impedance varies somewhat from about $400 \Omega$ at maximum to $2500 \Omega$ at the halfway setting In practice this is no serious disadvantage since there is no question of matching to the load, usually considerably higher.

Wiring should be rigid and as short as possible to keep stray capacitance to a minimum. Earth wiring should be via a common heavy gauge copper wire to one point on the chassis only. The accuracy of the generator will depend on the tolerance of the capacitors in the bridge, which should be as close as practicable. Of the high values in particular, padding may be necessary to bring the capacitors to within at most $2 \%$ of the specified value.

## Calibration

Although calibration can be carried out by other means it is strongly recommended that an oscilloscope be used. By injecting the reference signal to the horizontal input and the generated signal into the vertical input, calibration by Lissajous figures on the fundamental and multiples of the reference signal is easily effected. At the same time a constant visual check can be maintained on the generated waveform for distortion. The reference signal should, of course, be of known accuracy; e.g., a pre-transmission B.B.C. tone, in which case calibration should be initially -made on Range 3. Providing closetolerance capacitors have been used, calibration on other ranges other than Range 5 should be accurate and in multiples of ten. On Range 5 the scale should be set to $200 \mathrm{kc} / \mathrm{s}$ and a radio receiver in close proximity tuned to the Droitwich Light Programmes transmission. C10 should then be adjusted until a zero beat note is heard. Similar checks can be made at other frequencies on the radio scale.


Oscillograms showing oscillotor output at $3 \mathrm{c} / \mathrm{s}, 2 \mathrm{kc} / \mathrm{s}$ and $200 \mathrm{kc} / \mathrm{s}$.

# High-Temperature Components 

New Materials and Methods to Meet Service Demands

By G. W. A. DUMMER, M.B.E., M.I.E.E.

IN 1939, Service components were tested to W.T. Board Specification K.110, in which the dry heat test was exposure of the components for six hours at $70^{\circ} \mathrm{C}$. The $70^{\circ}$ figure was made up by allowing $15^{\circ}$ temperature rise in the equipment above the highest ambient temperature then measured, i.e., $55^{\circ} \mathrm{C}$. Certain components, notably selenium rectifiers, could not be operated at temperature in excess of this in any case. With the increase in complexity of equipments and in the number of components used, it became obvious that this $15^{\circ}$ rise was insufficient and therefore a new rating of $85^{\circ} \mathrm{C}$ was introduced, i.e., $55^{\circ}$ ambient $+30^{\circ}$ rise. The introduction of sealing and miniaturization techniques showed a need for a range of components rated at $100^{\circ} \mathrm{C}$ and in 1945 Specification RCS. 11 required three categories of temperatures $-40^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C},-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ and $-40^{\circ} \mathrm{C}$ to $+100^{\circ} \mathrm{C}$. Recently a $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ category has been introduced and it is interesting to compare the increase in Service category temperature of components over the years-see Fig. 1.

In recent years considerable advances have been made in miniaturizing components. The use of these extremely small components in sealed assemblies aggravates the cooling problem, since there are smaller surface areas available to trarisfer waste heat to the surroundings. Various methods of cooling miniature equipments have been adopted, notably on airborne boxes, e.g., forced convection cooling; but it is obvious that if components were available which could operate at the maximum temperature ambients of valves, etc. in a confined space without cooling arrangements, considerable gains would be made.

In airborne radar equipments the following constructions are in use : -

Open Construction (Fig. 2). In this construction, heat losses occur mainly by natural convection and some radiation. The arrangement consists of a wellventilated box with possibly an internal blower for a particular heat-producing component, for example, a magnetron or a large transmitter valve.

Pressurized Construction (Fig. 3). Heat losses in this construction are by radiation and convection; an internal fan is used to reduce excessive hot spots and an external fan, arranged to blow over the external surface, transfers heat to local air. It is possible to use ram air in place of the external fan in some cases, but either system becomes inefficient at altitudes over 30,000 feet because of the low density of the air available.

Pressurized Construction with Annular Heat Exchanger (Fig. 4). In this construction, heat losses are by radiation and forced convection. The annular space between the internal cover and outer shell is fitted with cooling fins or wires over which several pounds of cooling air can pass per minute. The source of mass flow can be air from the cabin spill-
over, a ram scoop or, in some cases, a specially designed high-altitude blower.

The latter two methods are thermally inefficient because of poor heat transfer from the components to the pressurized container and it is neceessary to expend much horse-power in cold air units or blowers to cool equipment dissipating a few hundred watts at high altitudes.
Aerodynamic Heating Effects. In low-speed flight, equipment cooling has generally depended on the losses due to natural convection and radiation, the heat being conveyed to the aircraft structure and thence to the outside air. Where forced convection cooling is used, outside air is collected and forced through ducts to the equipment or equipment bay, where it takes up heat, and is expelled as hot air farther down the aircraft. At high speeds, the temperature of the "cooling" air may be higher than the ambient temperature of the equipment itself.

Similarly, the aircraft skin temperature rises with speed until heat flows into equipment instead of away from it and even components which do not generate heat are heated above their maximum tolerable temperature. The rise in temperature in the Fairey D. 2 aircraft was recently given in a television broadcast as $100^{\circ} \mathrm{C}$. The main problem of cooling large electronic equipments in high-speed aircraft lies not so much in its practicability as in the formidable weight penalties involved in the provision of cooling apparatus.

Other Design Requirements.-With the continuing trend towards extreme miniaturization such as


Above: Fig. I. Increase in range of temperature specified for tests of Service equipment since 1938.

Right: Fig. 2. Conventional cabinet with ventilation at atmospheric pressure (low
 at high altitudes).

Army pack transmitter/receivers, computors, etc., there is little doubt that as components are made smaller, heat dissipation problems become increasingly difficult. The development of high temperature components is, therefore, useful in all fields of electronic equipments.

## Component Developments

In general, the development of components for use at high ambient temperatures precludes the use of organic materials such as paper and many of the plastics, and requires the use of inorganic materials such as glass or ceramics. Certain plastics, such as polytetrafluorethylene (p.t.f.e.) and materials such as silicones can be used with advantage.

The following is a general review of recent developments in laboratories in the United Kingdom and the United States of America and will be of interest to the electronics industry in this country.
N.B.-In all cases, temperatures on the surface of components are referred to.

Fixed Resistors ${ }^{1}$.-British laboratories have worked on metal film resistors for some years and the platinum-gold types are now becoming commercially available. The platinum-gold fixed resistor is fired on a glass base at a temperature of the order of $600^{\circ} \mathrm{C}$. The limiting factor is not in the film but in the solder which is used to attach the end connections. Temperatures of $180^{\circ} \mathrm{C}$ to $200^{\circ} \mathrm{C}$ have been reached with good stability and reasonably long life, but pressure connections in place of solder connections might enable $300^{\circ} \mathrm{C}$ to be reached. Metal oxide film resistors have also been developed, which are formed by spraying tin-antimony chlorides on glass rods at a high temperature. Upon impinging on the glass, these chlorides become oxides on the surface, forming a very hard adherent resistive film. It has been found experimentally that tin-antimony oxide powders mixed with alumina, pressed in the normal powder resistor manufacturing techniques and then fired at temperatures of the order of $900^{\circ} \mathrm{C}$, can produce Grade 2 resistors which can be operated up to at least $400^{\circ} \mathrm{C}$ with pressed end connections. The temperature coefficient is, however, rather high and work is being done on other diluting materials with the aim of improving this characteristic. Existing types of pyrolytic or cracked carbon resistors, when de-rated, can operate at $120^{\circ} \mathrm{C}$ for long periods with good stability and at even higher temperatures for short periods. The normal carbon composition resistor is limited to about $115^{\circ} \mathrm{C}$ and no power can be dissipated at this temperature. Vitreous wire-wound resistors, when de-rated, can operate at $320^{\circ} \mathrm{C}$ or, in some cases, $450^{\circ} \mathrm{C}$, depending upon the conditions.

It would seem, therefore, that already we have available a wide range of resistors, although some are in the experimental stage, capable of operating in the $150^{\circ} \mathrm{C}$ region with reasonable reliability and stability.

Variable Resistors.-Wire-wound resistors on ceramic formers and of all-metal construction are available, which can operate in excess of $150^{\circ} \mathrm{C}$, but carbon composition resistors are unable to exceed $115^{\circ} \mathrm{C}$. Some work is being done on the use of silicone mixes with carbon blacks to increase the operating temperature of the track, and it is hoped

[^5]

Fig. 3. Pressurized construction with external forced convectlon cooling by ram air or fan.


Fig. 4. Pressurized construction with onnular heat exchanger.
to produce a track operating up to $200^{\circ} \mathrm{C}$. Recently developed metal film (platinum-gold) variable resistors may operate up to $200^{\circ} \mathrm{C}$.

Fixed Capacitors.-P.T.F.E. dielectric capacitors in America have operated for 20,000 hours at $200^{\circ} \mathrm{C}$, but they are expensive and it has been found difficult to produce thin films of the material for winding. Glass dielectric capacitors, also available in America, are capable of operating continuously at $200^{\circ} \mathrm{C}$. Similar work is being done by several British firms.

Mica as a dielectric can withstand temperatures up to about $400^{\circ} \mathrm{C}$ before dehydration occurs, but mica capacitors are limited by the sealing material. Silvered mica capacitors in "Mycalex" cases will operate at about $130^{\circ} \mathrm{C}$. Vitreous glaze capacitors should operate satisfactorily at $150^{\circ} \mathrm{C}$ in sizes comparable to the mica capacitors.

High-permittivity ceramic dielectric capacitors are not, in general, capable of being operated over $100^{\circ} \mathrm{C}$ because of a degradation effect known as "creep", which becomes apparent as a change in capacitance with temperature; the mechanism of the change is not fully understood but is being studied.

Some low-permittivity ceramic capacitors can operate at temperatures up to $125^{\circ} \mathrm{C}$. Polyethylene terephthalate, known as "Melinex," is capable of operating up to $130^{\circ} \mathrm{C}$, but is temperature and frequency sensitive. Paper dielectric capacitors can be operated up to $100^{\circ} \mathrm{C}$ without too great a deterioration in insulation resistance, and metallized paper capacitors up to $125^{\circ} \mathrm{C}$. The recently introduced tantalum pellet electrolytic capacitors can operate at $150^{\circ} \mathrm{C}$ for long periods without deterioration, but

TABLE
$\left.\begin{array}{|l|l|c|c|}\hline \text { Type of Component } & & \text { Approximate } \\ \text { Maximum } \\ \text { Operating } \\ \text { Temperature }\end{array}\right]$
tantalum foil capacitors are limited to about $85^{\circ} \mathrm{C}$.
Transformers and Chokes.-It is possible to operate transformers up to $250^{\circ} \mathrm{C}$ by the use of sili-cone-impregnated glass-covered wire with siliconeimpregnated glass cloth interleaving, bobbins, etc., with normal grain-oriented silicon steel cores. Anodized aluminium wire is also being used experimentally. In America, fluoro-compound liquids with glass silicone wicks to convey the liquid to the hot spots, which act as vapour phase heat exchangers, are being used, there being a mixture of gases in the void space to give optimum heat transfer. Heat "sinks" consisting of copper plates are part of the design. In America, also, ceramic-covered wire with an external covering of p.t.f.e. is available which can operate at temperatures of up to $250^{\circ} \mathrm{C}$.

Experimental transformers for operation at temperatures up to $500^{\circ} \mathrm{C}$ are being made in Britain by the use in their construction of such substances as asbestos, mica, glass and silica and by the use of core materials with high Curie points, of the order of $700^{\circ} \mathrm{C}$. Tungsten or molybdenum wires may have to be used for conductors.

Relays.-Relays are available in the United States which operate at temperatures of $200^{\circ} \mathrm{C}$, using ceramic and glass insulation and metal covers.

Switches.-Ceramic base switches are available which can operate to $150^{\circ} \mathrm{C}$, but considerable work is being done on the development of switches for temperatures in excess of this.

Plugs and Sockets.-General purpose multi-way plugs and sockets with phenolic insulation are generally limited to approximately $85^{\circ} \mathrm{C}$, but recent developments of loaded silicone insulating materials have led to new ranges being developed which can operate up to $200-250^{\circ} \mathrm{C}$. In the United States of America, glass-sealed types of plugs and sockets are being used for operation at $200^{\circ} \mathrm{C}$. Ceramic or sintered alumina plugs and sockets are available operating up to $300^{\circ} \mathrm{C}$.

Cables and Sleeving.-Cables made with glass braid impregnated with silicone varnish and with
p.t.f.e. insulation can operate at $200^{\circ} \mathrm{C}$; non-flexible cables with ceramic insulation and copper external sleeving can run continuously at $250^{\circ} \mathrm{C}$ and, for short periods, up to $1000^{\circ} \mathrm{C}$. Silicone sleeving can be used up to $200^{\circ} \mathrm{C}$, but abrasion resistance is not yet too good.

Power Rectifiers.-The recently introduced silicon rectifiers can operate at temperatures up to $180^{\circ} \mathrm{C}$, have a much higher efficiency than the selenium or copper oxide types and are considerably smaller. Titanium dioxide rectifiers can operate for long periods at $100-150^{\circ} \mathrm{C}$, but they are larger than selenium or copper oxide rectifiers and are moisture sensitive. Tellurium rectifiers have operated at much higher temperatures but are still experimental.

Spindle Seals.-A new type of spindle seal which has recently been developed, using spring-loaded p.t.f.e. bushes, is capable of operating in excess of $150^{\circ} \mathrm{C}$ for 20,000 operations. It can be used for high-temperature potentiometers, rotary switches, etc.

The accompanying table summarizes the surface temperatures of the above components. It should be pointed out that the figures relate to reasonably long life, but many components will work at much higher temperatures if a shorter life is accepted.

It thus appears that considerable progress is being made in the development of high temperature components, but valves must also be considered. Most valves will work at temperatures between $160^{\circ} \mathrm{C}$ and $200^{\circ} \mathrm{C}$, although life may be reduced; some can operate up to $240^{\circ} \mathrm{C}$. Valves are the main source of heat production in electronic equipments and it may be that in the future valves which will work at even higher temperatures may be required.

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PRODUCTION and home and export sales curves of scund receivers-including radiograms and car radio setsplotted from figures issued by the Radio Industry Council at the Radio Show. The 1956 figures are calculated from the totals for the first six months of the year.

# IMPORTING AN INSTRUMENT 

A Guide Through the Official Maze

By A. J. REYNOLDS*

0VER the past twelve months or so the inflow of measuring instruments from abroad, and in particular from the U.S.A., has increased from a very slow drip to a thin trickle.

By virtue of their unfamiliar appearance and because of their temporary novelty value these instruments are shown to all visitors to laboratories owning them, the effect thus resembling that of the stage army-an impression of numbers out of proportion to reality.

Questions regularly put to the writer suggest that there is a great deal of ignorance concerning the legislation, practice and even ethics of importing these instruments, so perhaps an attempt at clarification might not be out of place.
To the telecommunications industry, instruments are the tools of trade. We are fortunate in this country in possessing a vigorous instrument industry which can fulfil most of our needs and at the same time make a useful contribution to exports. Only two other countries in the world are in anything like the same position-America and Germany. Even Russia, as far as can be judged, relies to a considerable extent on imported instruments.

Viewed as part of the general picture of the radio and allied industries, the turnover of the whole of the instrument industry is but a few per cent of the total turnover. That few per cent, however, has a virtually incalculable effect on the performance and abilities of the industry as a whole, the finished product being only as good as the tools with which it was made.

Just as incalculable but equally important is the effect on price of inefficient or time-consuming methods brought about by inadequate tools.

## Divergent Industries

The enormous anti-import ditch dug between the U.S.A. and ourselves just after the war was most successful in preventing apparatus crossing the Atlantic and even more undesirably prevented American technical journals from being seen by all but a favoured few. It is not surprising, then, that, kept forcibly apart from each other, two rapidly expanding industries should progress along slightly divergent paths.
Different techniques were evolving in the U.S.A., some superior, some inferior, to our own solutions to the same problems. These techniques gave rise to their own individual instrument requirements and produced many instruments of types not even attempted by our own manufacturers. The converse also applies, but instances are fewer.
An excellent example of this is in the range of instruments manufactured by the Kay Electric Company in America. Here is a whole catalogue of thirty-odd instruments, hardly one of which has a British counterpart.

[^6]The range consists for the most part of sweep generators and noise generators. Sweep generators are available giving a sweep of $\pm 30 \mathrm{Mc} / \mathrm{s}$ with a centre frequency continuously variable from $50 \mathrm{Mc} / \mathrm{s}$ to $1,000 \mathrm{Mc} / \mathrm{s}$. There is a video sweep generator having ranges of $0-5 \mathrm{Mc} / \mathrm{s}, 0-10 \mathrm{Mc} / \mathrm{s}$ and $0-20 \mathrm{Mc} / \mathrm{s}$ with several crystal markers. Noise generators are avaiłable at all common impedances and covering the frequency spectrum from the low radio frequencies well into the microwave region.

## Importance of Markets

In some cases component development plus the larger market has bred an instrument better than anything here by several orders. The availability of a valve that will work in a counter decade at $10 \mathrm{Mc} / \mathrm{s}$ with complete reliability has enabled Hewlett Packard to offer a frequency measuring counter which, with accessories, covers the fantastic frequency range of $0.006 \mathrm{c} / \mathrm{s}$ to $12,000 \mathrm{Mc} / \mathrm{s}$ with an accuracy of 1 part in $10,000,000$.
The importance of the large domestic market in the U.S.A. is perhaps not realized sufficiently. To develop, engineer and manufacture an instrument such as the above counter is an enormously expensive business and the final instrument can only be sold at an economic price if the development and tooling charges can be spread over a very large number of instruments.
There is a factor in the order of tens between the potential annual demand for a given instrument in Britain and the U.S.A. In spite of this handicap there is a sprinkling of British instruments that have no peer in the States. To quote the one example that will not be invidious, there is no American multi-range meter with the performance and versatility of the Avometer.
There are, then, occasions when the only sane engineering answer to a particular problem involves importing an instrument. What is the procedure and how does one obtain the dollars?

## Obtaining a Licence

The first step is to obtain a Board of Trade licence to import the item in question. This is achieved by filling in the inevitable form, which asks in effect: "What do you want? Why do you need it? Why won't the local product do?" The Board of Trade call on the Ministry of Supply for assistance in making their adjudication. In the case of instruments, at least, this involves the production officer responsible for the Ministry purchases of that type of equipment. These officers are engineers in their own right, with the result that, with very few exceptions, a good case is rewarded with a licence and any attempt at "pulling a fast one" reaps the raspberry it richly deserves.

Having got your licence you now have a title to the necessary dollars. Before sending off your pur-
chase order to the manufacturer, however, you must apply to the Board of Trade again, this time for a document known as an import certificate, which must accompany your order to allow the manufacturer to obtain an export licence.

The vast majority of instruments as we know them are liable to a customs duty of $33 \frac{1}{3} \%$ on arrival, being subject to the Safeguarding of Industries Act, 1921, this duty being popularly known as "key industries duty." The Act was passed just after the first world war to protect the then infant British optical and chemical industries against German competition. It has been amended from time to time, one of the most important occasions being in the Finance Act introduced in 1936, when provision was made for the duty to be waived in certain exceptional cases.

## Duty-free Import

Before the goods are imported a Treasury licence may be applied for. This document, if issued, enables goods subject to "key industries duty" to be imported duty free. The conditions required for duty-free import are very similar to those required for the original import licence. The adjudication is made on a basis of use, and available British alternatives.

Here, where the Government is concerned with money destined for its own pocket, the going becomes much tougher.

A lot turns on the question of what constitutes a "similar" instrument-and at times we appear to have the Gilbertian situation where the arbiter is the competing British manufacturer! By the time the argument has subsided the instrument has arrived and it is only necessary to wheedle it out of the customs. It may be costing $£ 800$ and be covered by a duty-free licence but do not be surprised to be charged 1s 9d duty on the mains lead if that was not listed separately on the licence.
If you see an instrument in Electronics that would halve your testing time or do a particular job three times as accurately in half the time, don't despair of getting it. There are only thirty-seven separate steps of paper work between needing it and having it on your bench.

The comparatively few instruments that have come in during the past couple of years have had a beneficial effect on the industry at large, out of all proportion to their cost. They have speeded development, cut production costs and acted as a gentle spur to many instrument manufacturers. The cost per year in dollars? Just about one token import of Californian dried fruit.

## CLUB NEWS

Chelmsford.-" $70-\mathrm{cm}$ techniques" is the title of the lecture to be given by $\mathbf{F}$. Turner at the meeting of the Chelmsford Group of the British Amateur Television Club on October 11th, at 7.30 at 10 Baddow Place Avenue, Great Baddow, near Chelmsford. Sec.: D. W. Wheele (G3AKJ), 56 Burlington Gardens, Chadwell Heath, Essex.

Crystal Palace.-At the meeting of the Crystal Palace and District Radio Club on October 20th, G. A. Bird (G4ZU) will speak on the design and construction of a three-band " minibeam" aerial. Meetings are held at 7.30 at Windermere House, Westow Street, London, S.E.19. Sec.: A. J. Worrall, 169, Kent House Road, Beckenham, Kent.

Plymouth.-Meetings of the Plymouth Radio Club are held on alternate Tuesdays at 7.30 at the Virginia House Settlement, Barbican. The next meeting is on October 2nd. Regular morse classes are held and occasional lectures given. Sec.: C. Teale (G3JYB), 3 Berrow Park Road, Peverell, Plymouth, Devon.
Sidcup.-The September meeting of the Cray Valley Radio Club will be held at the Station Hotel, Sidcup on September 25 th at 8.0 . R. G. Shears ( G 8 K W) will speak on v.h.f. mobile radio communication. Sec.: S. W. Coursey (G3JJC), 49 Dulverton Road, New Eltham, London, S.E.9.

Warrington.-It is proposed to include in the winter programme of the Warrington and District Amateur Radio Society a series of lectures on radio theory up to the standard for the Radio Amateur Examination. Meetings are held on the first and third Thursdays of each month at 7.30 at the Royal Oak Hotel, Bridge Street. Sec.: R. Dyke, 22 Stetchworth Road, Walton, Warrington, Lancs.

## OVERSEAS SALES

THE value of exports for the whole of the radio industry for the first six months of this year was nearly $£ 19.6 \mathrm{M}$ which, if continued for the rest of the year, will mean a record figure of nearly $£ 40 \mathrm{M}$. As will be seen from the figures below giving the exports for the various sections of the industry since 1950, the largest percentage increase during the past seven years has been in the value of sound reproduction equipment. But by far the largest contribution to the industry's export figures continues to be made by the manufacturers of capital goods-communications equipment, transmitters, navigational aids, industrial equipment, etc. It is estimated by the Radio Industry Council that indirect exports of capital goods (e.g., fitted in ships and aircraft) represent a further $25 \%$ increase on the figures given for this type of equipment. It is worth recording that the number of car radio sets exported during the first six months of this year was over $6,500-\mathrm{a} 50 \%$ increase on the same period last year.

| Class of Product | 1950 |  | 1951 |  | 1952 |  | 1953 |  | 1954 |  | 1955 |  | 1956 § |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | ¢M | \% | CM | \% | CM | \% | EM | \% | \&M | \% | EM | \% | EM | \% |
| Domestic Receivers | 2.740 | 15 | 4.793 | 21 | 4.636 | 19 | 3.945 | 15 | 3.628 | 12 | 3.968 | 12 | 3.919 | 10 |
| ment | 1.133 | 6 | 1.711 | 8 | 2.123 | 8 | 3.056 | 12 | 3.760 | 13 | 5.728 | 17 | 7.262 | 19 |
| Componenes $\dagger$................. | 4.607 | 25 | 6.014 | 26 | 5.931 | 24 | 5.386 | 20 | 6.543 | 22 | 7.484 | 22 | 8.294 | 21 |
| Communications \& Industrial | 7.089 | 39 | 6.440 | 28 | 8.630 | 34 | 11.655 | 44 | 13.182 | 44 | 13.628 | 41 |  | 42 |
| Valves \& Cathode-Ray Tubes* | 2.700 | 15 | 3.790 | 17 | 3.672 | 15 | 2.252 | 9 | 2.602 | 9 | 2.812 | 8 | 3.104. | 8 |
| Total | 18.269 |  | 22.748 |  | 24.992 |  | 26.294 |  | 29.715 |  | 33.620 |  | 39.176 |  |

[^7]
## DCTDBER

LONDON
4th. I.E.E.-Address by Sir Gordon Radley (president) at 5.30 at Savoy Place, W.C. 2
5th. Television Society.- "Impressions of commercial television" by Leslie Mitchell at 7.0 at 164 Shaftesbury Avenue, W.C.2.
17th. I.E.E.-"The electron:c age" by Dr. R. C. G. Williams (chairman, Radio and Telecommunication Section) at 5.30 at Savoy Place, W.C.2.

19th. I.E.E.-Discussion on "Experiments for the electronics laboratory" opened by V. Attree at 6.0 at Savoy Place, W.C. 2.

19th. B.S.R.A.-" A distortionless loudspeaker" by P. J. Walker at 7.15 at Royal Society of Arts, John Adam Street, W.C.2.
22nd. I.E.E-" Use of transistors in radio and television" by Dr. A. J. Biggs and E. Wolfendale at 5.30 at Savoy Place, W.C. 2.
23rd. I.E.E.-Discussion on "The use of electronic computers in nuclear reactor design studies" opened by R. W. Sutton at 5.30 at Savoy Place, W.C.2.

26th. R.S.G.B.- " More about the antenna match " by Frank Hicks-Arnold (G6MB) at 6.30 at the I.E.E., Savoy Place, W.C.2.

29th. I.E.E.-Five-day convention on ferrites opens at Savoy Place, W.C.2.

31st. Brit.I.R.E.-Address by George A. Marriott (president) at 7.15 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C. 1.

31st. British Kinematograph Society. -"Colour television" by B. J. Edwards (Pye) at 7.15 at the Royal Society of Arts, John Adam Street, W.C.2.

## BIRMINGHAM

8th. I.E.E.-" Germanium and silicon power rectifiers" by T. H. Kinman, G. A. Carrick, R. G. Hibberd, and A. J. Blundell at 6.0 at Regent House.

29th. I.E.E.-"Stereosonic recording and reproduction" by H. A. M. Clark at 6.0 at James Watt Memorial Institute, Great Charles Street.

## CAMBRIDGE

9th. I.E.E.-Address by J. G. Yates (chairman, Cambridge Radio and Telecommunication Group) at 6.0 at Cambridge Technical College.

23rd. I.E.E--" Generation and synthesis of music by electrical means" by A. Douglas at 8.0 at the Cavendish Laboratory.

## CARDIFF

31st. Brit. I.R.E.-" Applications of transistors to radio reception" by L. E. Jansson at 6.30 at Cardiff College of Technology, Cathays Park.

## CHELTENHAM

5th. Brit. I.R.E.-" Television lighting effects" by A. E. Robertson at 7.0 at the North Gloucestershire Technical College.

## GLASGOW

3rd. I.E.E.-Address by Prof. F. M. Bruce (chairman, S.W. Scotland Centre) at 7.0 at the Institution of Engineers and Shipbuilders, 37 Elmbank Crescent, C.2.

11th. Brit. I.R.E.-" The digita! computer and its applications" by Dr. Barnett at 7.0 at the Institution of Engineers and Shipbuilders, 37 Elmbank Crescent, C.2.

## MEETINGS

25th. I.E.E.-"An introduction to computers " by P. A. V. Thomas at 7.0 at the Institution of Engineers and Shipbuilders, 37 Elmbank Crescent, C. 2.

26th. Society of Instrument Tech-nology.-"A flexible electronic recorder controller" by S. A. Bergen (Cambridge Instrument Co.) at 7.15 at the Building Centre, 425 Sauchiehall Street.

## IPSWICH

8th. I.E.E.-Address by J. A. Sumner (chairman, East Anglian SubCentre) at 6.30 at the Crown and Anchor Hotel.

## LOUGHBOROUGH

9th. I.E.E.-Address by Dr. H. L Haslegrave (chairman, East Midland Centre) at 7.0 at Loughborough College.

## MANCHESTER

4th. Brit. I.R.E.-" Education for electronics " by R. H. Garner at 6.30 in the Reynolds Hall, College of Technology, Sackville Street.
29th. Institution of Production En-gineers.-" Computer-controlled machine tools" by D. T. N. Williamson (Ferranti) at 7.15 in the Reynolds Hall, College of Technology, Sackville Street.

## NEWCASTLE-ON-TYNE

15th. I.E.E.-Address by A. E. Twycross (chairman, North-East Radio and Measurement Group) at 6.15 at King's College.

## OXFORD

10th. I.E.E.-"A transatlantic telephone cable" by Dr. M. J. Kelly, Sir Gordon Radley, G. W. Gilman and R. J. Halsey at 7.0 at Southern Electricity Board, 37 George Street.

## PORTSMOUTH

12th. B.S.R.A.-" A distortionless loudspeaker " by P. J. Walker at 7.30 in the Lecture Hall of the Central Library

## SHEFFIELD

17th. I.E.E.-Address by G. G Nicholson (chairman, Sheffield Subcentre) at 6.30 at the Grand Hotel.

## TORQUAY

11th. B.S.R.A.-" Towards perfection : more news of the quest " by D. M. Chave (Lowther Manufacturing) at $7: 45$ at Callard's Café.

## WOLVERHAMPTON

10th. Brit. I.R.E.-Lecture on colour television at 7.15 at the Wolverhampton and Staffordshire Technical College, Wulfruna Street.
17th. Institution of Production En-gineers.- "Electronic control in industry" by E. Heys at 7.30 at the Wolverhampton and Staffordshire Technical College, Wulfruna Street.

## LATE-SEPTEMRER MEETINGS

## LONDON

26th. Brit. I.R.E.-" Some aspects of transistor progress" by Dr. H. W. Loeb at 6.0 at the London School of Hygiene and Tropical Medicine, Keppel Street, W.C.1.

28th. Physical "Society.-" Speech, music and hearing" by Dr. Winston E. Kock (Bell Telephone Labs.) at 5.30 in the Physics Department, Imperial College, Prince Consort Road, S.W.7.


The new TRIX Sound Columns introduce a system of outstanding efficiency for sound diffusion over wide areas both indoors and outdoors. The main advantages of this type are:-

* Striking improvement in acoustic efficiency.
* Marked directional flat beam effect, assisting in extending average range with minimum reverberation and feedback.
$\star$ Wiring and installation costs cut due to climination of large numbers of single units.


COLUMN LOUDSPEAKER the trix electrigal go. lid. MAPLE PLACE, TOTTENHAM COURT ROAD, Tel.: MUS 5817. Grams: Trixadio, Wesdo, London

# RANIDM RADIATIONS 

## By "DIALLIST"

## Faint, But Persuing

MY search for the v.h.f.-only, three-station, press-button receiver continues. I hear rumours that it is in course of production, or even that it has been produced by one or more manufacturers, but that's as far as it seems to get. I'm by no means alone in wanting such a receiver. Like innumerable folk who live on our east or south-east coasts, I find the medium-wave and longwave bands completely useless owing to constant interference from foreign stations and a variety of devices which make for the greater safety of those who go down to the sea in ships. Therefore I see no point in paying for tuning circuits that will never be used, for a wavechange switch and so on. On Band II only three B.B.C. programmes, Home, Third and Light, are available. Hence there's no point in buying a receiver with continuous tuning throughout the band. A set pre-tuned to the three frequencies of the local transmitting stationand possibly provided with a trimmer for drift or for small changes in valve characteristics-would be ideal for the job and I'm sure there would be a good demand for it.

## "Sound" Renaissance

INTEREST in v.h.f. broadcasting is steadily growing but the a.m. receiver with an added f.m. range
won't in most cases do justice to the high quality of the transmitted signals. Those who get their first taste of v.h.f. broadcasting, with such receivers can't possibly realize what splendid reproduction of music is obtainable from the service. People whose introduction to the new service comes by way of a good television console provided with a v.h.f. range may fare better, for many of these have audio departments designed to deal pretty well with a wider range of modulation frequencies. But there's one big improvement due to f.m. which users of both sound and television sets with the extra range are bound to observe: the completely quiet background. The spread of f.m. transmissions will, I believe, do a great deal towards restoring the popularity of sound broadcasting.

## A New Menace?

THE ever go-ahead Bell Telephone Laboratories of America have, I observe, developed an up-to-date version of a device which enables those at each end of a line to see one another while they converse. It's certainly an ingenious contrivance, for, by working at the rate of one picture every two seconds, it has been found possible to transmit and receive over ordinary telephone lines. I see, of course, the attractions of this "picture-phone" affair; but it seems

also to have rather awful possibilities unless you're provided with a switch enabling you to cut out at will the image transmitter at your end of the line. It would, for instance, no longer be seemly to answer calls clad only in a bath towel. Nor, again, could you wriggle out of an invitation on the grounds of being up to the eyes in work if a mid-morning caller could plainly see for himself that you had just tumbled out of bed. No, I think that the old-fashioned nopicture telephone will do well enough for me.

## More Power to their Elbows !

BY the time that this is in print the Crystal Palace e.r.p. will have gone up to 120 kilowatts-and no doubt there will have been a heavy demand for attenuators in Sydenham and parts adjacent! Dwellers in erstwhile fringe areas will be getting the television reception for which they have been longing, or at any rate something like it, and those in the coastal parts of Kent will no longer be feeling quite so much left out in the cold. I have a fellow feeling for all those whose homes are in places where only a poor TV signal is available, since for more than a year now I've had to make do with a mingy $25 \mu \mathrm{~V} / \mathrm{m}$, which means a "snowy" picture and other inconveniences. However, it shouldn't be long now before the permanent station at Tacolneston, near Norwich, comes into action and then, one hopes, all will be well. This will have been a big year for television development. Before it comes to an end it will have seen the move from Alexandra Palace to Crystal Palace, the "hotting up" of the latter station, of North Hessary Tor, Rowridge and Tacolneston, besides the coming into action of the I.T.A. stations at Lichfield, Winter Hill and Emley Moor. Not a bad record!

## Misleading?

IT seems to me rather a pity that the name ferroelectrics has been given to a range of materials which have, in fact, nothing ferrous in their make-up. They are actually mainly (if not exclusively) ceramics and none
of those in present use contains any iron at all. The term ferroelectrics was adopted because the electrical polarization of these materials is in many ways similar to the magnetization of ferromagnetic substances. Ferroelectrics are likely to play an increasingly important role in electronic apparatus. I, for one, regret that any explanation to the uninitiated of what they are and what they do must begin with some sort of statement that "ferro-" doesn't mean what it says.

## Names Again

WHY, I wonder, can't TV set makers come to an agreement to standardize the names of the "outside-the-cabinet" controls? It must often cause a good deal of puzzlement to the man in the street to find, when he buys a new set, that it appears to have controls quite different from those he had grown accustomed to in the old one. There's no line-hold, he finds, though there's a mystifying knob labelled line-speed, or perhaps hori-zontal-lock. Is frame-amplitude the same thing as height, he wonders; and can horizontal-form mean the same as line-linearity? I'm sure it would be worth everybody's while to adopt one set of names for the controls of all TV sets.

## Moving-coil Microphone

DESIGNED specially for the requirements of magnetic tape recorders, whether of the domestic or office dictating type, the new "Lustrette" Model LD61, introduced by Lustraphone, Limited, Regent's Park Road, London, N.W.1, is of the moving coil type. It is available in a choice of impedances which satisfy the input conditions of all the principal makes of recorder, and is housed in a case which is inherently stable for desk use; there is provision for stand mounting if required.

"Lustrette" moving-coil microphone for tape recorders.
 grip, is available with plain prods, with or without fuses and with or without retractable points; retractable items have slotted side-grip prod-points. The Neon Test Prod is moulded in black rubber with semi-stiff sleeved leads, totally shockproof and waterproof. The slender handle " Twistgrip " model is designed for use in inaccessible places, and prods or grips connections.



Full technical details and prices on these and all other Bulgin components are available in:

Catalogue No. 195/WW (price I/-post free).

[^8]
## UNIBIASEID

## By FREE GRID

## Freud in the Bathroom

I AM surprised there are not more lightweight TV sets on the market fitted with a handle so that they can easily be carried from room to room. I don't mean the type which runs off mains or car battery because this is produced for a special purpose. The type I have in mind is the sort which is conventional in every way except in size and weight.

You may well ask why I or anybody else should want to lug a TV set from room to room. The answer is that I don't always want to be tied to the same room when I want to do some viewing. My ideal is to have an aerial connection in every room, including the bathroom, and I don't mind telling you that I have made a start at home by equipping the latter room with one.
Nothing delights me more than to relax in a steaming hot bath and

watch television, although I must confess that so great is the power of suggestion that at first, whenever a lady announcer came on and gazed coldly at me in my bath, I used to go hot all over. I now invariably wear my bowler on these occasions. It adds a touch of sartorial formality which has an almost unbelievable psychological effect. No doubt Freud had a word for it

## Heretical Ideas

RECENTLY I took the opportunity to slip across to France to have a look at French 819-line television. I avoided Paris, as it is so full of other distractions which the Editor won't let me discuss here. I must confess that I was much impressed by the greater detail given and the absence of lininess. I know all the technical arguments against a larger number of lines but surely the need to transmit more information in a restricted bandwidth is a challenge to our engineers.

There would undoubtedly be a tremendous outcry if we all had to
scrap our present TV sets, for it would be quite uneconomic to attempt to alter them. I cannot, however, see why the B.B.C. and I.T.A second programmes, when they arrive, should not be radiated with a greater number of lines than 405 Those who wanted the second programmes would have to get a new set, but they couldn't grumble as the existing programmes would still be available on 405 lines. These could continue for some years until the owners of existing sets had received a fair run for their money.

Personally, I consider that a picture of greater detail would be preferable to colour. That brings me to yet another point. Why is so much effort being wasted in trying to produce a compatible colour system that is receivable in black and white on ordinary TV sets? Why cannot colour broadcasting-not necessarily by a compatible system-be confined for the first few years. to the proposed two second programmes? Owners of existing sets would then have no grumble, as they would still be able to receive existing programmes and could change over to colour when they felt they could afford it.

As for the cost at the transmitting end, surely money could be saved on the programme side in a very simple manner. Don't have any second programmes, but let the B.B.C. and the I.T.A. radiate "second transmissions" of the same programme in higher lineage and subsequently in colour. Owners of existing sets would then have no leg to stand on as they could not complain that they were being cheated out of a second programme.

## Radio Show Ramblings

WHY is it that the man in the street invariably wants to use the minimum in the way of an aerial? How often one heard enquirers at the Radio Show asking if this or that television set needed an outdoor aerial or whether an elaborate aerial was needed for " this 'ere v.h.f." I really wonder whether set manufacturers are to blame for this trend. Whilst they continue to produce sets with built-in or built-on aerials the wetnosed viewer and listener are naturally going to expect a receiver to work with these, irrespective of the field strength available in , their locality. I wonder if it was to set a good example that the I.T.A. adorned some of their receptionists with miniature arrays as head-gear?

At the B.B.C. enquiry bureau, where, incidentally, I noticed two well-known Wireless World contributors answering queries, the majority of the questions asked during the few minutes I was eaves-

dropping were concerned with v.h.f. That was to be expected in view of the excellent demonstration which the Corporation provided. It was certainly very convincing and I was glad to hear the demonstrator mention the need for a good aerial and refer listeners to the B.B.C. leaflet giving practical advice on v.h.f. aerials.

I was not very impressed by the "hi-fi" demonstrations at the Show. Perhaps it is that I have had a surfeit of these hyper-super-quality demonstrations. However, I was glad to find that in some of the demonstration rooms manufacturers had provided specialists to answer technical enquiries.

## Apology Impending?

IN the September issue the Editor admitted to having had his knuckles rapped by the Post Office. His offence was that he had allowed to be published a description of a telephone answering machine with an implication that it satisfied P.O. rules and regulations. As everybody knows, each breath that we in radio draw is by permission of the Post-master-General, and to offend him is a serious matter.
My sympathy went out to the Editor, but, now, in the light of something that has since happened, I doubt if there is any risk that publication of Wireless World will be suspended as a result of his indiscretion. In the latest number of the Post Office Telecommunications fournal there appears a summary of the offending Wireless World article, with no implication whatever that the device described fails to satisfy P.O. requirements. Indeed, the writer of the summary is at pains to quote the statement of the original author that the device, in "its respect for the sanctity of Post Office equipment, is virtually superhuman." Coming in a strictly official journal, that seems to me to amount to something very near official acceptance.

## (10) <br> Regd. Trade Mark <br> WIDE BAND <br> SIGNAL GENERATOR Type T.P.M.

The design of this new Wide Band Signal Generator, operating throughout on fundamentals, is the outcome of considerable research and development work to meet the stringent requirements imposed by new frequency modulation and commercial television stations.


## COVERAGE

$5-220 \mathrm{Mc} / \mathrm{s}$ in 8 ranges, CW or $400 \mathrm{c} / \mathrm{s}$ sine/square wave modulacion. Accuracy $\pm 1 \%$. Provision for spot frequency callbration.


## coverage

$65.120 \mathrm{Me} / \mathrm{s}$. Accuracy $\pm 1 \%$. Maximum devlation $\pm 150 \mathrm{Kc} / \mathrm{s}$.

This compact and versatile instrument not only meets all U.K. requirements, but will also prove useful in many other parts of the world.

## OUTPUT:

Minimum (about $2 \mu \mathrm{~V}$ ) to 100 mV continuously variable with decade multiplier. Forse output 250 mV .

## OUTPUT IMPEDANCE:

$80 \Omega, 200 \Omega$. balanced $80 \Omega$ and $300 \Omega$, isolated unbalanced $80 \Omega$

The frequency bands have been chosen in such a manner as to ensure maximum convenience when servicing and aligning T.V. and F.M. receivers.

Provision has been made for spot R.F. frequency calibration.

Facilities are provided :o ensure adequate discrimination throughout the very wide frequency band covered by the instrument.

Sine and square wave audia frequency modulation pro* vided.

The instrument is fitted with an R.F. carrier level meter
A double-ratio slow-motion mechanism, together with interpolation dial, enables the instrument to be set with a high degree of accuracy. On the F.M. range an internal phasing control enables the modulating signal to be applied to the X-plates of an oscillograph to produce a picture of a discriminator response curve.

## OPERATING VOLTAGES:

100-120, $200-260 \mathrm{~V}, 50-60 \mathrm{c} / \mathrm{s}$ A.C. malns.

DIMENSIONS:
$15 \frac{1}{4} \times 10 \frac{1}{2} \times 10$ ins. approx. with lid closed.

WEIGHT: 16 lbs. approx
LIST PRICE: 889


## A PROVED OSCILLOSCOPE

The Solarscope CD 513 is no longer a promising newcomer, with an impressive performance at a moderate price. It has proved itself the ideal general-purpose oscilloscope for a wide range of applications, including continuous wave and pulse phenomena, voltage and time measurement in radar, V.H.F. radio, telecommunications and servo analysis, and for general use in industry. Large numbers are now in service in this country and abroad, and the volume of repeat orders proves beyond doubt that Solartron have added another winner to their list.

Call Solartron for an engineer togive you a demonstration.

## BRIEF SPECIFICATION

| Max. Bandwidth: | D.C. to $10 \mathrm{Mc} / \mathrm{s}$ <br> (- $3 \cdot \mathrm{db}$ ) at 10 |
| :---: | :---: |
|  | $\mathrm{V} / \mathrm{cm}$. $/ \mathrm{cm}$ - to 60 |
|  | $\mathrm{V} / \mathrm{cm} .$ |
| Time-Base: | $0.1 \mu \mathrm{sec} / \mathrm{cm}$. to $100 \mathrm{~m} \mathrm{sec} / \mathrm{cm}$. plus expansion |
| Internal/Extern | up to $\times 5$. Sync.orTrigg |
| Dimensions: | $16^{\prime \prime} \times 10^{\prime \prime} \times 23^{\prime \prime}$ |
| Weight: | long. |

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For any equipment and for any application where low voltage stabilisers are normally employed, this 75 volt Mullard tube will give improved. performance.
Mechanically strong, the 75 Cl combines zirconium electrodes and the sputtered envelope technique pioneered by Mullard. This design gives a combination of high stability and good regulation which has never before been achieved in one tube.


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## ADVANCED FEATURES INCLUDE:

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Burning voltage of all tubes at 20 mA confined within the extremely close range of 73 to 79 volts.

Inherently Rugged
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Special uranium oxide coating ensures that the maximum striking voltage is 110 volts in BOTH DAYLIGHT AND DARKNESS.

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COMMUNICATIONS AND INDUSTRIAL VALVE DEPARTMENT

## MICROSECOND CHRONOMETER



This equipment is designed to measure small intervals of time to a high order of accuracy and two ranges are provided :
(1) $1 \mu \mathrm{sec}$. to 1 sec . in steps of $1 \mu \mathrm{sec}$.
(2) $10 \mu \mathrm{sec}$. to 10 sec . in steps of $10 \mu \mathrm{sec}$.

Accuracy of each range is better than $\pm 0.005 \% \pm$ the step interval.
Full details on this and other 'Cintel' Chronometers are avallable on request.

## CINEMA



What greater tribute to the supreme reliability of the Ferrograph than for it to be exclusively chosen by the G.P.O. for its new automatic telephone Weather Service. Opened only last Easter as a round-the-clock service for the London region, with as many as 600 lines available during peak periods, WEA 22 II has already recorded many millions of calls.

With a similar installation set up to provide Londoners with the latest Test Match scores, telephone enquiries reached the astronomical total of $2 \frac{1}{2}$ millions - big business for the G.P.O. and added lustre to
the Ferrograph reputation for dependability.
Bear these simple facts in mind when choosing your new Tape Recorder. There are many on the market costing less-some, indeed, much less. But Ferrograph standards of fidelity, performance and dependability have only been achieved by making cost considerations a secondary matter. The really important question we continually ask ourselves is "Does the Ferrograph still set the standard by which Tape Recorders are judged?" We believe that it does. We believe, too, that a lot of people are proud to say that they own Ferrographs.


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## HIGH GRADE INSTRUMENTS


$2 \frac{1}{2}$ in. scale moving coil meter. Square flush mounting. Type S25.

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"Fulscale" meter 4in. dia. scale moving coil having $270^{\circ}$ arc with a 9 in . scale length.


High torque moving coil portable meter. Precision grade to BS.89.

Multi purpose test set for simultaneous measurement of current and voltage.

Moving coil Microammeter 5in. scale. Flush mounting rectangular case. Type $S 50$.


Ûniversal multi range test set for electrical and radio engineers.


Ohmmeter for the rapid and direct measurement of very low values of resistance. Model RM.155.

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These represent just a few of our wide range of high quality instruments which are used by the electrical and

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# Distortion GorrectedTransmission Perfected 

The Regenerative Repeater TRR 1 is a start-stop, five unit code equipment, designed to correct distortion on long line or radio telegraph circuits. It covers the speed ranges 45,50 or 75 bauds, and accepts signals with up to $49 \%$ distortion. Noteworthy features for use on radio circuits are the rejection of short duration spurious start signals, the automatic insertion of correct length stop signals under deteriorating conditions, and the retransmission of long space signals during setting up.



Superbly engineered and of advanced design, the two models offered possess excellent electrical characteristics and are robustly constructed for service in : any climate. The "770 R" has contin" uous AM/FM coveragefrom $19 \mathrm{Mc} / \mathrm{s}$ to $165 \mathrm{Mc} / \mathrm{s}$; the " 770 U " from 150 $\mathrm{Mc} / \mathrm{s}$ to $500 \mathrm{Mc} / \mathrm{s}$. Both incorporate six-position turret tuning assemblies of unique design and giving high reliability. Fully illustrated literm ature with performance curves on request.

# AUDIO FREQUENCY MEASURING EQUIPMENT POWER OUTPUT METER TYPE 708 

Output powers from 1 milliwatt to 20 watts may be measured by this instrument, in the frequency range $30 \mathrm{c} / \mathrm{s}$ to $30 \mathrm{kc} / \mathrm{s}$. No tapped matching transformer is used, and the input impedance remains resistive and constant over the whole frequency range.

Power Range:

Frequency Range:
Input impedance:
I milliwatt to 20 watts in 4 ranges.
$30 \mathrm{c} / \mathrm{s}-30 \mathrm{kc} / \mathrm{s}$.
2, 4, 7, 10, 15, 20, 40, 75, $100,140,250,400,600$, 1,000, 2,000, 3,000 and 5,000 ohms.

Accuracy: $\quad+10 \%$ of full scale power from $30 \mathrm{c} / \mathrm{s}$ to $15 \mathrm{kc} / \mathrm{s}$. The meter reading is approximately 1 db low at $30 \mathrm{kc} / \mathrm{s}$.


This Signal Generator utilises a resistance-capacitance type of oscillator to provide stable, accurate and easily controlled voltages in the frequency range $30 \mathrm{c} / \mathrm{s}$ to $30 \mathrm{kc} / \mathrm{s}$. A screened and balanced output transformer and attenuator enable unbalanced, balanced and floating outputs to be obtained from a source of a constant 600 ohm impedance.
Frequency Range: $30 \mathrm{c} / \mathrm{s}-300 \mathrm{c} / \mathrm{s}, 300 \mathrm{c} / \mathrm{s}-$ $3 \mathrm{ke} / \mathrm{s}, 3 \mathrm{kc} / \mathrm{s}-30 \mathrm{kc} / \mathrm{s}$.

Stability:
Attenuator:
$\pm 0.05 \%, \pm 0.5 \mathrm{c} / \mathrm{s}$.
A 600 ohm constant impedance attenuator provides steps of 30,40 and 60 db of attenuation under all output conditions.

Full details of these instruments, which are available for immediate delivery, will be forwarded gladly on reques?.

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Our experience in the industrial field has indicated that there is a definite need for this type of outfit offering facilities for making prototype flexible remote controls as required. The two gauges of Remote Control fexible shafts in these outfits cover the range of torque loadings required for - volums controls - all types of wave change switches - condensers - all controls likely to be mat in electronic, radio and television equipment.

The outfits are reasonably priced and comprise:
No. 130 (.130 in. dia.) for remote controls up to 4 in . length............. 0.0
No. 150 (.150 in. dia.) for remote controls up to 6 in . length........£.10.0
(For use without flexible casing)

The S. S. White Company wil. be pleased to advise which Outfit is most suitoble for specific applications.

A detalled Parts List is availabie upon request.


The latest and most reliable of a famous range of Automatic Record Changers, the 456 is a 4 -speed instrument for $16 \frac{2}{3}, 33 \frac{1}{3}, 45$ and 78 r.p.m. operation. It will play 7in., IOin., and 12 in . records intermixed in any order-provided that all are of the same speed range-and is suitable for A.C. voltages of $100 / 125$ and 200/250

This new High-Fidelity Tape Deck
has been designed on Transcription quality principles for live recording, recording from F.M. Broadcasts, etc., and reproducing pre-recorded tapes. A twintrack model fitted with four heads, it runs at speeds of $3 \frac{3}{4}, 7 \frac{1}{2}$ and 15 inches per second.

4-SPEED GRAMOPHONE TRANSCRIPTION UNIT (MODEL 4T200) for truly faithful reproduction Entirely new type 4 -speed mechanism ensures absolutely uniform speed, with reproduction free from frequency modulation.


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In buying solder for manufacturing purposes there is only one sound principle . . . buy the best.
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Incorporating Enthoven's
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By relying on ENTHOVEN for all your soldering requirements you are banking on the best known name in the industry - a name that represents nearly 150 years experience in non-ferrous metals and an incomparable record in research and development.

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Specialists in HIGH STABILITY CRYSTAL UNITS in the frequency range $2,000-60,000 \mathrm{Kc} / \mathrm{s}$.

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Integral terminations at both ends ensure a wide field of application for these cores in the construction of self-resonant chokes for TV interference suppression. Indeed, this is an outstanding design feature for many types of $R / F$ coils and chokes required in small lightweight equipment. High internal resistance means that the shunting effect on the coil is negligible. The range of materials and terminations is such that most frequencies, particularly Band III, are adequately covered.


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# MINIATURE <br> CERAMIC CAPACITORS 

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Telegraphic Address: "STEATITE-BIRMINGHAM, 15 "

## A NEW-PRINCIPLE A-C AUTOMATIC VOLTAGE STABILISER



This is our Model ASR-1150 Automatic Voltage Step Regulator. It ideally fills the need for a cheap, small and light Stabiliser. Although it measures only $8 \frac{1}{2} \mathrm{in}$. $\times 4 \frac{1}{2} \mathrm{in}$. $\times 5 \mathrm{in}$., weighs as little as 11 ib ., and costs only $£ 24$ net, it has a performance fully equal to any similarly rated Automatic Stabiliser of the resonated, saturated core type, without any of the disadvantages.
ASR-1150 has a pure output waveform, is unaffected by changes in mains frequency, and works equally well from no-load to full-load, which is 1150 VA . It has a stabilised output at 230 V unless otherwise ordered.
Many other Automatic Voltage Stabilisers are now manufactured by us, and all are available for immediate delivery. In some cases the constancy of output is as high as $0.15 \%$. Models are available from 200 VA to 30 k VA , single phase. 3-Phase Stabilisers are also available. Prices are extremely competitive.

We can supply' from stock all types of American tuber, condensers, valves, potentiometers, etc.

# The NEW "ASR-I | 50" costs only £24 net 

## Outstanding range of Thorn miniature lampholders



## MINIATURE SEALED PANEL LAMPHOLDER - INDICATOR TYPE

Completely waterproof and will withstand conditions of constant vibration and shock, these lampholders are intended for installation on aircraft, armoured fighting vehicles, and marine equipment. They are sealed and insulated from the panel, the thickness of which can vary from 20 S.W.G. (.036") to 10 S.W.G. (.128\%). Thicker panels can be counterbored. Rotation is prevented by flats on the body. Mounting is by a single hole. Access to the lamp, for replacement, is from the front of the unit by unscrewing the dome. Lamps may be renewed, without breaking the seal to the equipment.
Weight : . 420 oz . ( 11.6 grammes ) with bulb.
Electrical connections: Two solder tags.
Catalogue No. MPL. 20 Red: MPL, 21 Green.
Catalogue No. MPL. 22 Amber: MPL. 23 Opalescent Ivory.
miniature sealed panel lampholder - dimmer type
Identical to the Indicator type, except for the interchangeable cap. This is ribbed for grip, continuously rotatable and contains a light output control from bright to 'blackout'. Weight: .530 oz . ( 14.8 grammes) with bulb.
Electrical connections: Two solder tags.
Catalogue No. MPL. 10 Red (Translucent). Catalogue No. MPL. 11 Green (Transparent). Catalogue No. MPL. 12 Amber (Transparent). Catalogue No. MPL. 13 Clear (Transparent).

## THORN MIDGET PANEL LAMPHOLDER

This is the simplest and most economical lampholder designed to accommodate the Atlas Midget Panel lamp. It is extremely effective and easily installed. Available with its transparent top in a variety of colours. Weight : 8.4 gr . ( 0.3 ozs .)
Can be supplied with insulated washers and connecting tags where non-earth return is desirable.

## Miniature lampholders in the Thorn range

have been made possible by the development of the Atlas Midget Panel bulb.

5 types of
midget panel bulbs are available 28 volts 0.04 amps .28 volts 0.08 amps
12 volts 0.1 amps
6 volts 0.1 amps
1.5 volts 0.75 amps


## FLUSH OR RECESSED LIGHTING UNIT

This lampholder is used as a standard unit in the Plasteck Console panel. The body of the lampholder may be retained in a countersunk hole in the panel by a hexagonal backnut and lock-washer. A small projection under the collar prevents the fitting turning in the panel. The special coloured filter is contained in a moulded screw cap and a soft rubber sealing washer prevents any light from escaping round the edge. Filters in red, green, amber and clear. Weight: .31 oz . with bulb.
Terminals: Solder tag and earth return.
Catalogue No. PPL90.
Catalogue No. PPL120 (with 6BA terminal screw and earth return, weight : .35 oz . with bulb).
Interservice ref: Type A, No. 1.
Flush type-Solder connections. Ref. No. 5C/X. 5143. Type A, No. 2.
Flush type - Screw terminals. Ref. No. 5C/X. 5144.
Can be.supplied with insulated washers and connecting tags where non-earth return is desirable.

## SURFACE TYPE LIGHTING UNIT

An alternative design to PPL90 for Plasteck and other control panels where no room exists immediately behind the metal panel. The bulk of the component projects above the face of the panel. A soft rubber sealing washer under the cap prevents the escape of light from the front of the panel. The lamp is inseued with the cap up.
Weight : . 49 oz . with bulb.
Terminals: Solder tag and earth return.
Catalogue No. PPL. 100.
Interservice Ref: Type B,
Surface type - Ref. No. 5C/X. 5145.


## heart of the

## matter...

The RCA New Orthophonic High Fidelity Pick-ups are built around an entirely new 8 -pole balanced variable reluctance cartridge.
A special feature of the design, which represents a completely new departure in gramophone disc reproduction, is the change-over mechanism. The single cantilever construction of the dual stylus model, completely eliminates the mechanical resonance previously experienced with normal type turnover Pick-ups. The dual styli are both mounted on the one cantilever and the change-over from one stylus to the other is effected by a positive toggle action.


RCA Pickups are available with single or dual stylus cartridges with diamonds and sapphires, and a choice of two arm lengths. The long arm models will track records up to $16^{\prime \prime}$ in diameter, and the normal length arm tracks any record up to $12^{\prime \prime}$ diameter. The Pick-ups will fully load any normal High Fidelity amplifier without the use of a step-up transformer. A tracking pressure selector is builtin on all models, with an adjustable pedestal to suit any height of turntable.


Thew Onthophonec dheqh Fidelity

## EDISWAN <br> 

## FLUOROGARBONS FOR ELEGTRONIGS

## P. T. F. E.

P. G. T. F. E.

EDISWAN are now the largest fabricators of fluorocarbons, P.T.F.E. and P.C.T.F.E. and were, in fact, the first fabricators of these materials in Europe and the British Commonwealth.

The wide range of products that can be supplied includes Electronic Components as well as blanks for user fabrication. Prices, because of the rapid growth in production, are substantially down. A fully Technical Advisory Service is available without charge or obligation for the benefit of Engineering and Purchasing Departments. Write for informative details, descriptive literature and our latest comprehensive Catalogue.

## EDISWAN




## (by De La Rue)

## does a doulble act for you

You already know DELARON as a laminated insulant of precisely known dielectric values and mechanical properties which fully meet all BritishStandard and Ministry specifications at every grade. It is used as a controller of electricity, from loudspeaker spiders to insulating washers.
But do you know about the second half of the double act by DELARON laminates?

In its copper-clad form it is the basis of most of the printed circuits in radio, television, radar and electronic computors - the new, simpler and cheaper method of circuit manufacture.
You should use DELARON laminates. For information, advice and service get in touch with Dept. D8C, Thomas De La Rue \& Co Lid 84-86 Regent Street, London, W1.

*
An extremely wide range of Dubilier resistors is available both for development and production purposes This range covers insulated wire-wound, power wire-wound, precision wire-wound, ultra high range, high stability, high voltage and high frequency resistors.

Type BT insulated resistors are completely protected by a phenolic resin housing which is sealed at the ends. Type BT8 is rated at $\frac{1}{2}$ watt and Type BTB at 1 watt at $70^{\circ} \mathrm{C}$. Resistance range is $100 \Omega$ to $10 \mathrm{M} \Omega$ (BTS) and $390 \Omega$ to $22 \mathrm{M} \Omega$ (BTB).

Type BTS resistors are also available in attractive handy cartons of twelve. These cartons protect the resistors from dirt and permit easy storage and selection in the laboratory or workshop. Save time and trouble by ordering all your $\frac{1}{2}$ watt resistors in the handy carton.


# TESTEQUIPMENT 

in the

## AUDIO RANGE

$15 \mathrm{C} / \mathrm{S}$ to $50,000 \mathrm{C} / \mathrm{S}$

These two most modestly priced models, in common with the comprebensive "Advance" range (which completely covers from Audio to U.H.F.) bave earned a reputation second-to-none the world over for accuracy, simplicity in use, and consistent reliability.

II signal generator
for the Audio Engineer
Covers from $15-50,000 \mathrm{c} / \mathrm{s}$ in three ranges. This model is characterised by its extremely low distortion and level output over the entire range. Accuracy $\pm ェ \% \pm \mathrm{c} / \mathrm{s}$. Output from $200 \mathrm{micro-volts}$ to 20 volts with an accuracy of $\pm 2 \mathrm{db}$.

STABLE OUTPUT OVER FULL RANGE SINE OR SQUARE WAVE OUTPUT DISTORTION LESS THAN $1 \%$ AT $1000 \mathrm{e} / \mathrm{s}$

LIST PRICE IN U.K. $\mathbf{£ 2 8}$

Full technical details in Folder Wi6.

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Covers from $15-50,000 \mathrm{c} / \mathrm{s}$ in three ranges. Accuracy $\pm$ ( $2 \%+\mathrm{Ic} / \mathrm{s}$ ). Output (continuously variable) into 600 ohms. o.imW 1 W ( $0.25-25 \mathrm{~V}$ ) $\pm 2 \mathrm{db}$, output impedance approximating to 600 ohms over the whole range. Maximum output into $s$ ohms is greater than $\frac{1}{2} W$.

LIst price in u.k £35.12s.
TYPE J 2 similor to || but with output voltage meter.
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## ADVANCE COMPONENTS LIMITED ROEBUCK ROAD • HAINAULT • ESSEX



Have you a transformer problem? If so, we can help you. We can undertake to develop and manufacture rotary transformers to your specification.
The illustration shows a typical transformer which we are manufacturing for a specific requirement. Made for 6,12 or 24 volts D.C. input, it can supply a continuous D.C. output of 350 volts at 30 mA . or an intermittent output of 310 volts at 60 mA . The no-load current consumption is 2.2 amps . at 11.5 volts and the ripple voltage is less than 6 volts r.m.s. on 60 mA . load. The size is only $4-9 / 16^{\prime \prime}$ long by $2-21 / 3^{\prime \prime}$ across the brush terminals.

## Trean NEW!!!

## T.V. Waveform \&Alignment Generator


for all T.V. systems
(1) T.V. Pattern Generator.
(2) A.M. Signal Generator with variable audio output.

7 expanded bands cover frequencies :

## 4 basic instruments

## INCORPORATED IN ONE

(1) 4 to $7 \mathrm{Mc} / \mathrm{s}$.
(4) 30 to $45 \mathrm{Mc} / \mathrm{s}$.
(2) 8 to 14 ,
(5) 45 to 80 ,
(3) 15 to 22

ALL ON FUNDAMENTALS
(3) F.M. Signal Generator.
(4) T.V. Sweep Oscillator.
(6) 85 to $145 \mathrm{Mc} / \mathrm{s}$.
(7) 150 to 220 "


#### Abstract

T.V. PATTERN GENERATOR modulation. All patterns FULLY INTERLACED with accompanying fine and frame blanking, synchronizing (and equalizing on 625, 525 ines) signals. Threc gradation patterns providing relatively high, medium and low frequency transients. Cross grid pattern for linearity checking. Cross bar pattern on which $1.5,2,2.5,3,3.5,4,4.5 \mathrm{Mc} / \mathrm{s}$ definition bars can be displayed. Horizontal bars. Vertical bars with switched Mc. bars. White raster. Bleck raster.

Output-R.F. Three position attenuator, $0,-20$ and -40 db relative to 100 mV . Output-Video. From cathode follower 2 Kohms direct, positive or negative 3 V poak to peak. output-Synchronizing. Additional sync. output from 2 Kohrms through $8 \mu \mathrm{~F}$ 10V peak to peak positive going wavetorm comprising line and frame syachronizing pulses, interlacing signals (and equallzing pulses on k 25 and 525 tines). Audio. From cathode follower 2 Kohms direct. Variable A.F. voltage at $900 \mathrm{c} / \mathrm{s}$ approx. 2 V peak-to-peak maximum Optional mains lock facility.


Main features of A.M. Signal Generator F.M. Signal Generator \& T.V. Sweep Oscillator:
Frequency: $4-220 \mathrm{Mc} / \mathrm{s}$ in 7 expanded bands. Calibration accuracy: $\pm 1 \%$. Output: A.M. 100 mV, F.M. and sweep 3 mV . Attenuation: $\frac{1}{0},-20,-40 \mathrm{db}$. Oulput Impedance: 75 ohms unbalanced. Sweep frequency: $50 / 60$ cycles. A.M. \& F.M. audio modulation: $900 \mathrm{c} / \mathrm{s}$ approximrequency: $50 / 60$ cycles A.m. band-width all controllable.
Ṕower supply: $105 / 125$ V. or $200 / 250$ V. A.C., at $40 / 100 \mathrm{c} / \mathrm{s}$. Consumption 70 watts.
Woight: 201b. Dimensions: 17in. $\times 9$ in. $\times 8$ in. $(43 \times 23 \times 20 \mathrm{~cm}$. $)$. Finish: Steel case in grey hammer finisb with Perspex escutcheon.

LIST PRICE £75.0.0.<br>CREDIT TERMS AVAILABLE WRITE FOR LITERATURE




We have always prided ourselves on service and it is no secret that our business has been built around this word.

To illustrate what this really means, we recently received cabled orders totalling 143,500 radio tubes. These were packed and shipped on one boat within six days.

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1900 types of both receiving and transmitting tubes always in stock.



* STANDARD CONNECTORS
in 8, 12, 18 and 25 way
Low priced Unitor
Natural colour Nylon-PF moulding
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Breakdown voltage 3.5 KV



## * MICRONECTOR

in 9, 18, 26 and 34 way
Precision miniature connector
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Gold plated contacts
Breakdown voltage 2.4 KV

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 Covers can be supplied for either plugs or sockets or both in all these types.$\star$ STRIP CONNECTORS
in $2,4,6,8$ and 10 way Inexpensive flat connector for Printed Circuits Woodflour-filled PF moulding Cadmium plated contacts Breakdown voltage 3.5 KV

$\star$ RED-RANGE CONNECTORS
in 8, 16, 24 and 32 way
Medium priced
Low insertion force - self-aligning
Red Nylon-PF moulding
Gold plated contacts
Breakdown voltage 3.0 KV


ISPIMMC. 1


The criterion, as always, is that the reproduced sound shall be the closest
approach to the orlginal - that the enfoyment and apprectation of music may be unimpeded. This is reflected throughout the design of the QUAD II. It is reflected. too, in the straightforward and logical system of contro
achieved without the sacrifice of a single refinement or adjustment capable of

Send for further details ana booklet




## The Marconi TF 801B/l Signal Generator

Marconi's new " 1 " version of the TF 801B is now in production. It is a premium-grade instrument meeting all signal generator requirements for the design, manufacture and maintenance of a.m. equipment in the range 12 to $470 \mathrm{Mc} / \mathrm{s}$.
Painstaking attention to detail in the design of the master oscillator and power amplifier circuits has yielded an instrument with an r.f. output of unusually high quality : f.m. on a.m. is less than $2 \mathrm{kc} / \mathrm{s}$ at all carrier frequencies up to $425 \mathrm{Mc} / \mathrm{s}$; harmonic content is less than $2 \%$; the calibration of the frequency dial is accurate to within $0.5 \%$; drift is less than $0.02 \%$ over a $10-$ minute interval.

ABRIDGED SPECIFICATION

Frequency Range : 12 to $470 \mathrm{Mc} / \mathrm{s}$ in five bands.

Output: $0.1 \mu \mathrm{~V}$ to 1 volt modulated; $0.1 \mu \mathrm{~V}$ to 2 volts unmodulated.
Source Impedance: 50 ohms.
Internal Modulation: $1000 \mathrm{c} / \mathrm{s}$ sine a.m. monitored and variable to depths up to $90 \%$.

External Modulation : Sine up to 20 $\mathrm{kc} / \mathrm{s}$ or pulse at p.r.f's. up to
$50 \mathrm{kc} / \mathrm{s}$.


AM \& FM SIGNAL GENERATORS . AUDIO \& VIDEO OSCILLATORS • VALVE VOLTMETERS • POWER METERS Q METERS • BRIDGES • WAVE ANALYSERS • FREQUENCY STANDARDS • WAVEMETERS • TELEVISION AND RADAR TEST EQUIPMENT • AND SPECIAL TYPES FOR THE ARMED FORCES


For those to whom acoustics is a mystery, the mounting of such a fine high fidelity instrument as the Tannoy Dual Concentric Speaker is inevitably something of a
hit-and-miss procedure. And it is difficult subsequently to subdue the nagging thought that perhaps a more knowledgeable approach might have produced even finer results.

It is for such people . . . people who cannot be happy with anything short of the best.. . .
that the GRF Enclosure has been produced.
Based on the larger and more costly 'Autograph ' enclosure, it has an elaborate interior system of sound source expansion, incorporating all the advantages of horn loading for
bass reproduction. Mid-range frequencies are covered by direct frontal radiation

## DIMENSIONS

Max. front to rear 29*
Max. width $38^{\prime \prime}$
Overall length $48^{\prime \prime}$
Floor clearance, on
standard legs $6 \frac{1}{2}$
via an entirely new acoustic coupling device, and the high frequency end of the spectrum
is amply catered for by the normal non-directional horn loaded source of the $15^{\circ}$
Tannoy Dual Concentric unit.
. . . . And if technicalities are of little more than passing interest, the point that really counts is that, obviously, Tannoy would never release an enclosure that did not extract the last possible degree of realism from their own Speaker unit.


Practitloners in Sound

# heats up from cold in 6 seconds! 

Manufactured for Enthoven Solders Ltd., by Scope Laboratories, Melbourne, Australia.

Designed on an entirely new principle, this light-weight, versatile iron is eminently suitable for soldering operations in the radio, television, electronic and telecommunication industries. For test bench and maintenance work it is by

## TIME/TEMPERATURE CURVE CHART from the SUPERSPEED SOLDERING IRON TIP/TEMPERATURE TIME GHEGK

The effect of different voltages on initial heating-up time is shown. Whilst 4 V is the standard voltage normally employed, 6 V will cause no harm, and accumulators are a useful source of current supply.

* Activated by light thumb pressure on the switch ring. When pressure is released, current is automatically switched off-thus greatly reducing electricity consumption, wear on copper bit and carbon element.
* Length, $10^{\prime \prime}$; weight, $3 \frac{1}{2}$ ozs.; can be used on 2.5 to 6.3 volt supply ( 4 volt transformer normally supplied) or from a car battery.
* More powerful than conventional 150-watt irons; equally suitable for light wiring work or heavy soldering on chassis.
* Simple to operate ; ideal for precision work.
* Requires minimum maintenance - at negliglble cost; shows lowest operating costs over a period.

For full particulars, including guarantee terms and free trial facilities, please write to the sole concessionaires in this country :ENTHOVEN SOLDERS LTD. (Industrial Equipment Division) 89 Upper Thames Street, London, E.C.4. MANsion House 4533

## Switch to the

 Superasperall ar the most efficient and economical soldering iron ever designed. Ideally suitable for use with Enthoven Aluminium Cored Solder (melting point $260^{\circ} \mathrm{C} .500^{\circ} \mathrm{F}$.).Soldering Iron as being used by the<br>Royal Society Antartic Expeditior<br>for the International Geophysical Year.



## Goodmans



In the new 'viscount' enclosure provision is made for the trebax high frequency pressure unit to be mounted in the front baffle. The aperture for this purpose is blanked-off by a detachable panel. A similar arrangement in the rear wall facilitates the fitting of a variable attenuator for the irbrax.

The trebax unit may be used either with the AxIOM 150 Mk. II (15 wasts) or with the AXIOM 22 Mk. II '(20 vatrs) in the 'viscount' enclosure. The crossover frequency should be 5,000 c.p.S., for which purpose crossovir Unit sype XO. 5,000 -another new Goodmans product-should be used.

## Goodrmans

GOODMANS INDUSTRIES, LTD., AXIOM WORKS, WEMBLEY, MIDDX. Telephone: WEMbley 1200 Cables: Goodoxiom, Wembley, Englond. Scottish Distributors : LAND, SPEIGHT \& COMPANY, LIMITED,

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$\star$ The 595 HS can be controlled by ultra sensitive contacts handling 0.4 mA . at 2 V . Contacts will handle 5 A , at 230 V. A.C. $\star$ The 595 HS is made to withstand exceptionally heavy shock and vibration. $\star$ The 595 HS is made to withstand dirt and humidity indefinitely.
$\star$ The 595 HS can be obtained with various contact assemblies.
$\star$ The 595 HS is low in price because of its novel design.

Specially designed for use in domestic Radio \& Television receivers, these miniature rectifier stacks have an established position with manufacturers to whom reliability, small dimensions and low costs are important.

| TYPE | RMO | RMI | RM2 | RM3 | RM4 | *RM5 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Maximum amblent temperature Maximum output current (mean) Maximum Input voltage (r.m.s.) Maximum peak lnverse voltage Max. instantaneous peak current Weight $\qquad$ | $\begin{gathered} 35^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C} \\ 30 \mathrm{~mA} \quad 15 \mathrm{~mA} \\ 125 \mathrm{~V} \\ 350 \mathrm{~V} \\ \text { Unlimited } \\ 0.82 \mathrm{oz} . \end{gathered}$ | $\begin{gathered} 35^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C} \\ 60 \mathrm{~mA} 30 \mathrm{~mA} \\ 125 \mathrm{~V} \\ 350 \mathrm{~V} \\ \text { Unllited } \\ \mathrm{I} \mathrm{OZ.} \end{gathered}$ | $35^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C}$ $100 \mathrm{~mA} \quad 60 \mathrm{~mA}$ 125 V 350 V Unilimited 1.4 oz | $\begin{gathered} 350^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C} \\ 120 \mathrm{~mA} 90 \mathrm{~mA} \\ 125 \mathrm{~V} \\ 350 \mathrm{~V} \\ \text { Unilimited } \\ 2 \text { oz. } \end{gathered}$ | $\begin{gathered} 40^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C} \\ 250 \mathrm{~mA} \quad 125 \mathrm{~mA} \\ 250 \mathrm{~V} \\ 700 \mathrm{~V} \\ \text { Unlimited } \\ 4.5 \mathrm{oz} . \end{gathered}$ | $\begin{gathered} 40^{\circ} \mathrm{C} \quad 55^{\circ} \mathrm{C} \\ 300 \mathrm{~mA} \quad 150 \mathrm{~mA} \\ 250 \mathrm{~V} \\ 700 \mathrm{~V} \\ \text { Unilmited } \\ 4.75 \mathrm{oz} \end{gathered}$ |

- For use in voltage doubler circults the peak Inverse and maximum input voltages are halved, current output belng as for half wave operatlon
- Instant starting-no warming-up period

Unlimited instantaneous overload

- No limit to size of reservolr capacitor

Simple mounting-no valve holder

- Withstand overloads such as charging current of de-formed electrolytic capacitors

Low heat dissipation
Practically indestructible In service
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## 間 <br> EIGGEE FIDEIITY STSTEMMS <br> The New Sound in Home Entertainment

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## The FIRST and ONLV Aenials writh ELIECTRONIC COUPLING

 Driven Element is "Electronically coupled" to a Band III Resonator (Prov. Pat. No. 21958/55). This eliminates the use of coupling bats, thus avoiding any impedance mis-match and at the same time has the advantage of utilising the Band I Elements on both Bands I and III. For example the HILO Model HL402/4K (Band I Dipole and Director electronically coupled to 4 Band III elements) has an equivalent Band III performance to a 6-element Acrial. This same principle is a feature of the complete HILO range.


## OFFER ANGE THE ADVANTAGES

- 1 AERIAL ONLY FOR BOTH BANDS.
- 1 JUNCTION UNIT ONLY.
- 1 DOWN LEAD ONLY.
- NO FILTER UNIT REQUIRED.
- ACCURATE MATCHING.
- ALL.CHANNEL COMBINATIONS.
- HIGHLY COMPETITIVE PRICE. STOCK UP NOW!


Band I "H" (Dipole and Director) Electronically coupled to 4 Band III elements for CHIMNEY MOUNTING. 17 HILO Models are available in all popular types of mounting, suitable for most areas, in any Band I/Band III channel combinations.

All Antiference Aerials incorporate features which are fully protected by Patents or are the subject of Patents applied for.

## television monitor comprehensive

## every tube has these advanced features

- Metal backed screen.
- Straight gun (no ion trap) permitting highest spot quality.
- Resolution to highest monitor standards, satisfying 625 line system even at high brightness of 50 ft . lamberts in peak whites.
- Screen blemishes reduced to the absolute minimum.
- Quality control throughout manufacture to ensure highest picture standards and maximum life.
- Magnetic and electrostatic focused equivalent available to suit individual design requirements.


AW13-36 Electrostatic Focus
Max. final anode voltage 14 kV Typical focus voltage range -200 V to +200 V about cathode potential
Deflection angle $53^{\circ}$
Overall length 12 inches


MW13-35
Magnetic Focus

Maximum final anode voltage 11 kV Deflection angle $53^{\circ}$ Overall length 11 inches

## ( 5 -inch TUBES

These tubes are designed for use as electronic view-finders in television cameras, but they also satisfy the requirements for compact monitor equipment in broadcast and industrial television.

## 14-inch TUBES

These tubes are designed for use in television studio monitors but they also satisfy the requirements for large screen displays in industrial television systems.

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## tubes - Britain's most

## range



## AW22-10 Electrostatic Focus

Max. final anode voltage 14 kV
Typical focus voltage range -200 V to +200 V about cathode potential
Deflection angle $58^{\circ}$
Overall length 16 inches


MW22-22
Magnetic Focus
Max. final anode voltage 14 kV
Deflection angle $64^{\circ}$
Overall length 15 inches

## 9-inch TUBES

These tubes are designed to a size convenient for use in mobile outside broadcast television equipment. They are also employed in studio floor monitors.


This tube is primarily intended for use as a televjsion studio monitor tube.

## MW43-67 <br> Magnetic Focus

Max. final anode voltage 15kV
Deflection angle $70^{\circ}$
Overall length 20 inches


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The full range consists of $2,4,6,8,10,12,18,24$ and 33 -pole sizes, and there is
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3. Breakdown or Flashover Voltage :

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CONTACT RESISTANCE
voltage rating (a) 1,000 Volts D.C. or A.C. (peak).
This applies to use in temperate climates under normal conditions.
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Alt plugs and sockets will withstand a voltage proof test of 2.5 KV between contacts, and 3.0 KV between contacts and the mounting flange or cover.
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I. Average Contact Resistance: Below 0.002 Ohm.
2. Maximum Contact Resistance : 0.0025 Ohm.
(1)

## DESIGN FEATURES

The terminal numbering is moulded into both the plug and socket bodies, and appears not only in proximity to the appropriate soldering-tag. but also on the mating face.
This not only facilitates wiring; but enables complete cable forms to be tested prior to inclusion in equipments without removing the plug or socket covers.

## (2)

Four small distance pips are moulded on to the plug body and they keep the mating faces slightly a part even when the plugs and sockets are fully engaged.
This eliminates the possibility of free moisture remaining between the plug moisture remaining between the plug and socket face, and is instrumental in "Multicon" plugs and sockets.

## (3)

The single-piece body mouldings are nylon-filled to provide o high insulation ord tracking resistance.
(4)

All socket clips and plug blades are located in recessed cavities in the mouldings.
This also provides o high tracking resistance This also provides o high tracking resistance
between contacts, on, in the socket between contacts, and, in the socket
version, the enclosed contacts enable the version, the enclosed contacts enable the
maximum voltage to be safely utilised maximum voltage to be safely utilised
(provided the -direction of voltage supply feed is from socket to plug).

## (5)

Each socket clip has split limbs, so that there are four individual areas in contact with each' plug blade.
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This enables rows of plugs or sockets to be mounted either end-to-end or side-by-side with a maximum saving in panel or chassis space.
(7)

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This is provided to focilkate the engagement of these larger sizes, especially in unitor applications.
(8)

Covers are provided with either a top cable-entry hole and clamp or a side cable-entry hole and clamp, to suit the needs of particular equipments and designs.
(9)

Two alternative facilities can be proviced for earthing the plug or socket (gA.)
In one version an earthing tag attached to the moulding connects the inside of the cover to the highest numbered contact so that an earth lead in the cableform. connected to the highest numbered contact, automatically earths the cover.
(9B.)
In the alternative form, an earth tag is riveted directly to the outside of the cover and is suitable for the direct connection of an earth lead.
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## (10)

All sizes of plug or socket covers can be fitted with retaining blades to secure the unit to the panel or chassis.
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(1)

(2)

(7)


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\text { C.V. number } \\
26 \\
2666 \\
788
\end{gathered}
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## POWER AMPLIFIER

Push-pull distributed load output stage producing an output of 27 watts at $\pm 0.1 \%$ total distortion
Frequency response: $\pm 1 \mathrm{~dB} \mid \mathrm{c} / \mathrm{s}$. to $100 \mathrm{Kc} / \mathrm{s}$.
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# Wireless Woild 

ELECTRONICS, RÅDIO, TELEVISION

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## OCTOBER 1956

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Transistor circuits are usually designed to operate from one battery. They should be stabilised as described in Transistors for the Experimenter. A two-battery system can be used instead (see figure). It is especially suitable for transformer coupling. The collector supply voltage $\mathrm{V}_{\mathrm{cc}}$ is provided by two batteries in series, and the base supply voltage $\mathrm{V}_{\mathrm{bb}}$ is obtained by connecting one resistor $\mathrm{R}_{\mathrm{b}}$ from the base to the common point of the two batteries.
In transformer coupled two-battery circuits $\mathrm{R}_{\mathrm{b}}$ can be very small and even zero. $\mathrm{R}_{\mathrm{b}}$ draws base current only from $\mathrm{V}_{\mathrm{bb}}$, it does not bleed current from $\mathrm{V}_{\mathrm{cc}}$. Hence $\mathbf{R}_{\mathrm{b}}$ can be small to obtain good stability, and in the limiting case when $\mathrm{R}_{\mathrm{b}}=0$ the stability is the same as for a grounded base circuit.
In $R-C$ coupled circuits a low $R_{b}$ shunts the input and a value of the order of $10 \mathrm{k} \Omega$ must be used. The main disadvantage of the circuit is that since $\mathrm{V}_{\mathrm{bb}}$ is fixed, once the nominal collector current has been chosen, $\mathrm{R}_{\mathrm{e}}$ is automatically decided, too. For example, if $\mathrm{I}_{\mathrm{c}}$ nom. is 0.5 mA and $\mathrm{V}_{\mathrm{bb}}$ is 1.5 V , then $\mathrm{R}_{\mathrm{e}}$ can only be $2.7 \mathrm{k} \Omega$. There is therefore a fairly large voltage drop across $R_{e}$, hence $R_{c}$ must be lower, and the gain is reduced. A small-signal R-C coupled circuit, in fact, cannot be operated with a collector supply voltage $\mathrm{V}_{\mathrm{cc}}$ of less than 6 V .

One general advantage of the two-battery circuit, which applies to both R-C coupling and transformer coupling, is that $\mathrm{V}_{\mathrm{cc}}$ can be changed without having to re-design the circuit. If it is necessary to increase the size of the signal, one has only to increase $\mathrm{V}_{\mathrm{cc}}$ while keeping $\mathrm{V}_{\mathrm{bb}}$ the same, and the transistors operate at the same nominal collector currents. The stability is slightly worse because the junction temperature increases, but
the signal is approximately the required size.
A further advantage of the two-battery system is that the effect of resistor tolerances is much smaller. First, the circuit contains few resistors, second, the tolerance on $\mathbf{R}_{\mathrm{b}}$ exerts very little effect, a $10 \%$ spread in $\mathrm{R}_{\mathrm{b}}$ giving only about $1 \%$ spread in $I_{c}$.

Preferred circuits for an OC71 operating with a collector supply voltage $\mathrm{V}_{\mathrm{cc}}$ of 6 V and a base supply voltage $\mathrm{V}_{\mathrm{bb}}$ of 1.5 V are given in the table for R-C coupling and transformer coupling. As when working with only one battery, the design of these circuits can be very much simplified by using graphs. For two batteries the graphs are always straight lines.
 Circuit values can also be calculated, without using graphs. The procedure is as follows. (1) Choose $\mathrm{V}_{\mathrm{cc}}$, $\mathrm{V}_{\mathrm{bb}}$, $\mathrm{I}_{\mathrm{c}} \mathrm{nom}$., and $\mathrm{R}_{\mathrm{b}}$. (2) Then

$$
\mathrm{R}_{\mathrm{e}}=\left(\mathrm{V}_{\mathrm{bb}}-\mathrm{V}_{\mathrm{b}-\mathrm{e}}+\mathrm{I}_{\mathrm{c}(o)} \mathrm{R}_{\mathrm{b}}\right) / \mathrm{I}_{\mathrm{c}}-\mathrm{R}_{\mathrm{b}} / a^{\prime} .
$$

(3) Calculate $K$ and $I_{c(q)}$ max., assuming $T_{j}=T_{\text {amb }}$, and allowing $5 \%$ for resistor tolerances if $\mathrm{R}_{\mathrm{e}}$ and $\mathrm{R}_{\mathrm{b}}$ are both $\pm 5 \%$. (4) Find $V_{c(q)}$ min. (5) Choose $R_{c}$ less than the value given by

$$
V_{c(q)} \min =V_{c c}-I_{c(q)} \max \cdot\left(R_{c}+R_{e}\right)
$$

(6) Find $T_{j}$, and if $T_{j}$ is more than $1^{\circ} \mathrm{C}$ greater than $\mathrm{T}_{\mathrm{amb}}$, recalculate $\mathbf{I}_{\mathrm{c}(\mathrm{q})}$ max. etc. accordingly.

PREFERRED CIRCUITS FOR OC7I

|  | $V_{\text {cc }}$ | Vbb | Re | $\mathrm{R}_{\mathrm{b}}$ | Re | Ic nom. | $l_{c}(q)$$\max .$ | $\mathrm{V}_{\mathrm{c}}(\mathrm{q}) \mathrm{min}$. |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |  |  | at nom. $\mathrm{V}_{\text {cc }}$ | at min. $\mathrm{V}_{\text {ce }}$ |
| R-C.Coupling | 6 V | 1.5 V | $2.7 \mathrm{k} \Omega$ | $10 \mathrm{k} \Omega$ | $3.9 \mathrm{k} \Omega$ | 0.5 mA | 0.81 mA | 0.5 V | 0.3 V |
|  | 6 V | 1.5 V | $2.7 \mathrm{k} \Omega$ | $6.8 \mathrm{k} \Omega$ | $4.7 \mathrm{k} \Omega$ | 0.5 mA | 0.72 mA | 0.5 V | 0.3V |
|  | 6 V | 1.5 V | $1.2 \mathrm{k} \Omega$ | $10 \mathrm{k} \Omega$ | $2.2 \mathrm{k} \Omega$ | 1.0 mA | 1.6 mA | 0.5 V | 0.3 V |
| Transformer Coupling | 6 V | 1.5 V | $0.47 \mathrm{k} \Omega$ | 0 | $200 \Omega$ | 2.85 mA | 3.05 mA | 3.9 V | 2.6 V |
|  | 6 V | 1.5 V | $1.2 \mathrm{k} \Omega$ | 0 | $200 \Omega$ | 1.1 mA | 1.2 mA | 4.2 V | 2.8 V |
|  | 6 V | 1.5 V | $2.7 \mathrm{k} \Omega$ | 0 | $200 \Omega$ | 0.5 mA | 0.57 mA | 4.2 V | 2.8 V * |



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Most types are available within a reasonable period. In many instances additional types can be made available for special applications.


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## About these eacos.g heads ...

11 Carr Street, Gouge; Sydney, Australia. Dear Sirs,
" - will track with ease all present day records". So reads your ad. for the new Hi-g Heads. "We will soon see about that", I said. The first test for this new L.P. Head was Decca's Brahms Fourth - always very difficult to handle I found on the old head. The shock I received was enough to put me to bed for a month -where was all that distortion? Where was all that groove jumping? Having recovered my strength and secretly suspecting it was. just a fluke, I tried the Swan Lake-also another jumper - and then in a determined effort to prove you wrong, on went the Symphonic Fantastique and Rite of Spring.

At this stage the neighbours and family were seriously alarmed at sundry cries issuing from my room-they need not have worriedthey were cries of pure joy. I had seriously considered installing expensive magnetic Pick-ups - of which I knew very little - but this will obviously be quite pointless now.
"-will track with ease all present day records" to which I say, "blessed be the name of Aces Hi-g".

Yours with relief,
(Sgd.) Cliff Davidson.


FHEE The subject of Hi-g cannot be adequately explained in an advertisement, so we have produced an interesting booklet -" The ABC of Hi-g '. May we send you a copy?


## "BELLING-LEE" NOTES

## New Band III Broadsides from Enfield



We can announce several new band III designs which we were able to finalise after the last issue of the "Wireless World " went to press, and before the Radio Show opened its doors to the public. By the time you are reading this, we will be delivering all these new types-they are in production.
The great success of the "Belling-Lee" "double-six" broadside array has resulted in our producing a "double-three" and a "double-nine." These new designs are particularly useful for the reduction of "ghost" images in "difficult" towns such as Sheffield, where the town is surrounded by hills.

In localities not far enough away to warrant a "double-six" the "double-three" will be found a boon. Both of these new designs are available for channels 8,9 or 10. The "double-three" has either a cranked mast and single lashings, or a nine foot mast and double lashings. But the " doublenine" is provided with a 14 foot mast and double lashings only, and is probably the most efficient commercial aerial available from production lines at the moment.

## | Band I "H" with

5 Band III elements
This aerial is the logical development from our band III adaptor kit, which consists of two add-on elements alongside a band I dipole (L. 924 Patent Application No. 17696/54). The patent will be seen to be two years old. This combination results in an aerial which will give good results in the secondary areas of both B.B.C. and I.T.A. services.

It was originally designed for channels 1 and 9 only, but is now available in channels 2 and 10, so they will soon be seen
on the skyline in Yorkshire. It is supplied with a 9 foot mast and lashings, and has an array only suitable for fixing to customers own tubular mast with $1 \frac{1}{2}$ in. bore.

The Announcement of a LongLife Collapsible Band I " $X$ " Aerial
Ever since "Belling-Lee" invented the " $V$ " type array first produced in September 1946,
 we have toyed with the idea of setting up two of them " back to back" to make a band I "X."

Such an aerial is obviously cheaper to manufacture than an "H" as there is no crossarm. But no " X " is as efficient as a good "H" although there are many sites where maximum efficiency is not required, and it must be admitted that the "X " has "caught on." So, as a result of repeated requests from the public and trade alike here is a "Belling-Lee" " X."

The mechanical design is very interesting, it almost " springs to attention." It is definitely a long life array, easily erected, with particular care taken to ensure easy and quick connection of the feeder.


We have been able to improve both the appearance and performance of the "Double V" array for use with Lichfield and Sutton Coldfield. The new list number is L. 920 and it should be noted that L. 927 ceases to be shown in our catalogue and lists.

Advertisement of
BELLING \& LEE LTD.
Great Cambridge Rd., Enfield, Middx. Written 20th August. 1956

## "BELLING-LEE" INTERFERENCE FILTER


L. 1314 For 2-core cables, 2 amp. 250 v. A.C./D.C. This new small flex lead filter is designed for the suppression of interference at band I television frequencies only, and is for insertion in the flex lead within 6 in. of the motor of an appliance. This is the most convenient form of filter which can be readily installed and is complete with terminals, cord erips, etc

L.199. (2 or 3 -core) 2 amp. 250 v. A.C.ID.C. Inductor and capacitor filter effective at band I television frequencies and short and medium wavebands. This is an inductor and capacitor filter designed for connection in the lead of domestic electrical equipment such as hair dryers, vacuum cleaners, sewing machine motors, electric fans, etc., and to be truly effective must be fitted within 6 in . of the connections of the moror in the appliance.

L.1334.2 Amp. This very small inductor is essential for the filtering of interference on band III, and is individually tuned for use on band I. It must be fitted inside the casing of the appliance. When dealing with these very high trequencies, it is generally quite useless to attempt filtering in the flex lead, as the odd 6 in. of lead together with the overall dimensions of the appliance is an appreciable factor of the wavelengths and the whole acts as a radiator of interference.

# COMBINED COUPLING 

IS THE SUBJECT OF A PATENT APPLICATION FOR
(Patent Application 37/89/1954)

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## WOLSEY TELEVISION LIMITED

## PACEMAKERS TO THE AERIAL INDUSTRY

## MARBONN LOW FREOUENGY GRYYSTAIS

The illustration shows
a 1.6 and $3 \mathrm{Kc} / \mathrm{s}$ Carrier Oscillator Unit for transmitting information to a remote bearing indicator as used in the Marconi V.H.F. Direction Finder Type ADzoo.

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[^11]

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# - 



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OCTOBER/56

## Model 1058 Single Beam Osililograph

Designed for laboratory use, this new oscillograph provides a $Y$ amplifier with a very useful
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Frame or Line sync. pulse in a I volt D.A.P. (positive) composite video signal. Five calibrated time base ranges are provided giving spot velocities from $30 \mathrm{~cm} / \mathrm{sec}$ to
$1.5 \mathrm{~cm} / \mathrm{mic}$ crosec. An $X$ amplifier with a maximum sensitivity of $0.5 \mathrm{~V} / \mathrm{cm}$ and bandwidth $20 \mathrm{c} / \mathrm{s}-250 \mathrm{kc} / \mathrm{s}(-50 \%)$ is included and allows time base expansion, continuously variable, of up to five times. Time measurement is by calibrated shift control. The instrument operates from $100-130$ or 200-250 volt mains supplies.

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* The total hum and noise at $7 \frac{1}{2}$ inches per second $50-12,000$ c.p.s. unweighted is better than 50 dbs .
$\star$ The meter fitted for reading signal level will also read bias voltage to enable a level response to be obtained under all circumstances. A control is provided for bias adjustment to compensate low mains or ageing valves.
$\star$ A lower bias lifts the treble response and increases distortion. A high bias attenuates the treble and reduces distortion. The normal setting is inscribed for each instrument.
$\star$ The distortion of the recording amplifier under recording conditlons is too low to be accurately measured and is negligible.
* A heavy mu-metal shielded microphone transformer is built in for $15-30$ ohms balanced and screened line, and requires only 7 micro-volts approximately to fully load. This is equivalent to 20 ft . from a ribbon microphone and the cable may be extended 440 yds . without appreciable loss.
$\star$ The .5 megohm input is fully loaded by 18 millivolts and is suitable for crystal P.U.s, microphone or radio inputs.

> The amplifier, speaker and case, with detachable lid, measures $8 \frac{1}{4} \mathrm{in} . \times 22 \frac{1}{2} \mathrm{in} . \times 15 \frac{3}{4} \mathrm{in}$. and weighs 30 lb .
> PRICE, complete with WEARITE TAPE DECK £84 0
$\star$ A power plug is provided for a radio feeder unit, etc. Variable bass and treble controls are fitted for control of the play back signal.
$\star$ The power output is 3.5 watts heavily damped by negative feedback and an oval internal speaker is built in for monitoring purposes.

* The play back amplifier may be used as a microphone or gramophone amplifier separately or whilst recording is being made * The unit may be left running on record or play back, even with $1,750 \mathrm{ft}$. reels, with the lid closed.

POWER SUPPLY UNIT to work from 12 volt Battery with an output of 230 volts, 120 watts, 50 cycles, within 1\%. Suppressed for use with Tape Recorder. PRICE E18.

## FOUR CHANNEE BLECHRONIC MIXPR

is almost essential for the professional or semiprofessional where a number of different items have to be inixed on one tape recording.
It is recommended by a number of tape-recorder manufacturers for this purpose.
Any normal input impedance can be supplied to order, balanced or unbalanced, the standard being 15-30 ohms balanced.
The normal output is 0.5 volt on 20,000 ohms or less, but 600 ohms is available as an alternative.
The steel stove-enamelled case is polished and fitted with an engraved white panel suitable for making temporary pencil notes.
An internal screened power pack and selenium rectifier feed the five low-noise non-microphonic
 valves.
Used in many hundreds of large public address installations, recording studios and Broadcasting Stations throughout the world.

# Train for a wonderful future in ELECTRONICS 

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## OLYMPIC HONOUR FOR BRITAIN

We are proud to announce that our TL/'12 amplifiers have been chosen for use at the 1956 Olympic Games to be held in Australia.

It was in 1945 that H. J. Leak revolutionised the performance standards for audio amplifiers by designing the original "Point One" series, and we became the first firm in the world to market amplifiers having a total distortion content of 0.1 per cent. This claim was received with incredulity, but it was subsequently confirmed by the National Physical Laboratory and since then hundreds of TL/12 amplifiers have been used by the B.B.C., and Commonwealth and forelgn broadcasting authorities, and thousands have been used by recording studios, leading musicians and musiclovers throughout the world. We were the only British exhibitor at the world's first Audio Fair which was held in New York in 1949 and the volume of our exports to the United States of America has grown steadily since then.

Further development work resulted in our producing, at a much lower price but with the same high performance standards, the TL/10 amplifier. The TL/10 amplifier and "Point One" pre-amplifier received such an excellent reception when they were first exhibited at the Audio Fair in New York in October, 1953, that we received an initial order for 1,000 sets. Since then several thousand sets have been sold throughout the world. The output of the TL/10 is ample for highfidelity home music systems, and the quality of reproduction obtained is equal in every respect to that of the TL/12. We always use the TL/10 amplifier and "Point One" pre-amplifier for our public demonstrations of high-fidelity reproduction of gramophone records and radio. The TL/10 amplifier, when used with the best available complementary equipment, gives to the music-lover a quality of reproduction unsurpassed by any equipment at any price. Even when the complementary equipment falls below that of the best obtainable, the use of these amplifiers will enable one to obtain very marked improvements in reproduction.


## HIGH FIDELITY EQUIPMENT

## ELECTROSTATIC LOUDSPEAKERS

Reprints of "The Gramophone" article (May, 1955) by H. J. Leak, summarising his work and findings on Electrostatic and Dynamic Loudspeakers, are available on request, free of charge.

Write for leaflet $W \nmid$


## Make LEAK equipment the heart of your Hi-Fi system ... .

Illustrated above is the LEAK TL/IO AMPLIFIER $£ 17.17 .0$ and "PO:NTONE" PRE-AMPLIFIER, $\{10.10 .0$. These prices are made possible only by world-wide sales.

## SPECIFICATION :

## Circuitry

A triple loop feedback circuit based on the famous TL/I2. The output transformer is the same size as in the TL/12.
Maximum power output: 10 watts.
Frequency Response: $\pm 1 \mathrm{db} 20 \mathrm{c} / \mathrm{s}$ to $20,000 \mathrm{c} / \mathrm{s}$.
Harmonic Distortion: $0.1 \%, 1,000 \mathrm{c} / \mathrm{s}, 7.5$ watts output.
Feedback Magnitudes 26 db , main loop.
Damping Factor: 25.
Hum: -80 db referred to 10 watts.
Loudspeaker Impedances: 16 ohms, 8 ohms, and 4 ohms.

## "POINT ONE " PRE-AMPLIFIER

The handsome gold escutcheon plate contributes to the elegant appearance and blends with all woods.

## $\star$ Pickup

The preamplifier will operate from any pickup generally available in the world, A continuously variable input attenuator A continuously variable input attenuator at the rear of the pre-amplifier permits
the instantaneous use of crystal, movingthe instantaneous use of crystal, moving iron and moving-coil pickups.

## $\star$ Radio

The radio input sockets at the rear permit the connection of the LEAK V.S. tuner unit. An input attenuator is fitted. H.T. and filament supplies are available" from the pre-amplifier.
$\star$ Distortion
Of the order of $0.1 \%$.

## * Hum

Negligible, due to the use of recently developed valves and special techniques. * Input selector * Input selector

Radio, tape, records: any and all records can be accurately equalised. * Treble

Continuously variable +9 db to -15 db at $10,000 \mathrm{c} / \mathrm{s}$.

* Bass

Continuously variable +12 db to -13 db at $40 \mathrm{c} / \mathrm{s}$.
t Volume Control and Switch
The switch controls the power supply The switch controls the power
to the TL/IO power amplifiers. to the TL/10 power amplifier
An exclusive feating Jacks
An exclusive feature. Readily accessible jacks are provided on the front panel for instantaneous use with Tape Recorders which have built-in (low level) amplifiers.

OTHER LEAK PRODUCTS -

| Varislope II pre-amplifier | $£ 1616$ | 0 |  |  |
| :---: | :---: | :---: | :---: | :---: |
| TL/12 power amplifier | 8287 | 0 |  |  |
| TL/25A power amplifier | 8347 | 0 |  |  |
| Leak dynamic pick-up:Arm | \&2 15 | 0 p.t. ¢1 | 3 | 1 |
| LP head with diamond stylus | ¢5 15 | 0 p.t. 82 | 8 | 4 |
| 78 head with diamond stylus | £5 15 | 0 p.t. 82 | 8 | 4 |
| Mu-metal cased transformer | \&1 15 | 0 |  |  |
| Trough-Line FM tuner unit w supply | ¢25 | 0 p.t. $£ 10$ |  | 0 |
| Those seeking to obtain the highest quality of gramophone and radio reproduction are invited to ask their dealer for a demonstration of Leak products which, with their tradition of excellence, represent the bes that can be obtained. |  |  |  |  |

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## PRODUCTION IN. CREASED - CIRCUIT IMPROVED - PRICE REDUCED <br> 

To-day's beat value in Band III converters, authable for your TV or money refunded Complete ready to operate, $49 / 6$ nonmains, or $69 / 6$ mains. Post and insurance $3 / 6$.

## BAND III PRE-AMP



## BAND III AERIALS



THE • READY
This is a $\ddagger$ wave-length, 3 element array. Of all alloy construction, and ready for instant mounting in loft, bedroom cupboard, window frame, etc. Price 12/6, plus 2/-.
3 element array with swayneck mast with "U " bole clamp for fitting to existing masts from $\frac{1}{2}$. to 2 in . dia. $41 / 6$
3 element array with cranked mast and wall mounting bracket
3 element array with cranked mast and chimney lashing equipment
5 element array with swanneck mast and " U " bolt clamp for fitting existing mast from din. to 2 in. dia.. .

52/6
5 element array with cranked mast and chimney lashing equipment
8 element array with swanneck mast and "U" bolt clamp for fitting to tin. to $2 i n$. dia. mast

## T.R.F. CONVERTER

New this month is a converter for T.R.F. Set Viewmaster-Electronic Engineering, etc. Small mods. to the TV are necessary as this must be rumed into a superhet to stop re-radiating. Price comto stop re-radiating. Price com-
plete with two valves $E 6 / 10 \%$ assembled-ready Oct./W.W.

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- Ferrite Rod Aerial. (6) Low consumption valves (DK96 range). - Superhet circuit with A.V.C.
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| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CV1199 | 5/6 | ML4 | 816 | 5016 |  |  | 9/6 |
| DAP91 |  |  | $7 / 6$ | $5{ }^{5} 49$. | $8 / 6$ | 7 C 5 | $8 / 6$ |
| DDT4 | $8 / 6$ | RK34 |  | ${ }^{5} 3$ | 718 | 786 | $8 / 6$ |
| DDT13 | 816 | 8217 | $4 / 6$ | $6 \mathrm{AC7}$ | 716 | 787 | $8 / 6$ |
| DF91 | 81- | ${ }^{\text {sp61 }}$ | 41. | ${ }^{68 \mathrm{~B}} 8$ | $12 / 6$ | 787 | 18 |
| DH7sm | 101- | TDD 4 | $8 / 6$ | 687 | $7 / 6$ | 787 | 916 |
| DL74 | 9/6 | TDD13 | 8/6 | 688 | 8/8 | 724 | $8 / 6$ |
| EB34 | 816 | U71 | $8 / 6$ | ${ }^{6 C B 6}$ | $12 / 6$ | 7193 | $6 / 6$ |
| Esc33 | $12 / 6$ | U74 | $8 / 6$ | 6 C 4 | 716 | 72 | 1016 |
| HCH35 | 13/6 | VP23 |  | ${ }^{606}$ | 6/6 | 78 | 716 |
| EF36 | 6/- | VRs5 | 716 | ${ }^{\text {6D6 }}$ | 73 | 807 | 6/6 |
| EF39 | 8/6 | VR65 | 216 | ${ }_{6}^{655}$ |  | 84 | $8 / 6$ |
| EL32 | 1016 | Vr52 | 81- | ${ }^{676}$ | 716 | $866{ }^{8}$ | $8 / 6$ |
| ${ }_{\text {ELL5 }}$ | ${ }_{1}^{12 / 6}$ | V872 | $4 /-$ | ${ }_{\text {BFG6a }}^{\text {BF8 }}$ | 9/- | 9D2 | 4/- |
| H63 | $7 / 6$ | Xhis.s | $4 / 8$ | 6H8 | 3/6 | 954 | 4/6 |
| HL23DD | $8 /-$ | XPL.s | $4 / 8$ | 635 | 8/6 | 9004 | 7/6 |
| HLA | $8 / 6$ | ${ }^{173}$ | $9 / 6$ | 6.57 | $8 / 6$ | 12sK7c | 8/- |
| HL13 | 10/6 | ${ }^{63}$ | 91 | ${ }^{6 \mathrm{~K}} 6$ | $6 / 6$ | 12 SH 7 | 6/- |
| ${ }_{\text {KT63 }}$ | ${ }_{9 / 6}^{7 / 6}$ | ${ }_{183}$ ILD5 | 9/- | 6K7 | 8 | 12SE7 | 8\% |
| KT72 | $12 / 6$ | IT4 | 8 8/- | 6L69 | 9/- | 12 Y 4 | 9/- |
| KT81 | 8/6 | IV | 10/6 | 6N7 | \% | 1625 | $8 / 6$ |
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| KTW73 | $8 / 8$ $8 / 6$ |  | 718 | 6857 | ${ }_{71}^{6 / 1}$ | 230 XP | 6/6 |
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It is an ideal unit for a quality radio-
gram.
Special features include magic eye tuning indicator, extra long scale and pointer travel-latest circuitry employing full A.V.C. feedback, etc., etc.
Undoubredly one of the finest AM/FM chassis available today.
Chassis size $17 \mathrm{fin} . \times 6 \frac{1}{\mathrm{i}} \mathrm{i} . \times 7 \frac{\mathrm{tin}}{} \mathrm{in}$. Price $£ 23 / \mathbf{1 7} / 6$, carriage, packing and insurance $20 /$ extra.

THIS MONTH'S SNIP
THE WOLSEY 4-VALVE SUPERHET


This exeellent little receiver employs atandard circuitry and is ideal as a necond recelver for bedroom, kitchen, etc. It is a broadesst band aet and wiil recelve with ouly a few feet of nerial alt atations can be recelved. Complete, ready to work in modern: looking oak cabinet-limited quantity orfered thas month at $£ 6 / 15 /$ - plus
Orriage and insurance.
Overall size approximately 11 |lin $\times 71 \mathrm{n}$. $\times 81 \mathrm{u}$.

RII5 YOURS FOR 12


The R1155 is constdered to be one of the Hinest communication recelvers avallable to-day. Its frequency range is $75 \mathrm{kc} / \mathrm{s}$ s and is fitted ln a black metal case. Made for the R.A.P. so obviously a robust receiver which will give years of service. Completely overhauled and guaranteed in perfect working order. Price $£ 9 / 19 / 6$ or $\delta$ payments. of $£ 2$ each. Carriage and Traunit case 151- extra. Mains Powd


MINIATURE MOTOR

size only 2 ifin. long by $1 \mathfrak{i n}$. diameter American made-iaminated poles and armature-intended for 28 volt. Price 10/6. post $2 / \mathrm{F}$

F.M. TUNER

This tuner is based npion the very suocessful circuit in the booklet pablished by Dats
Publications. We hevo made up models at all branches and will be glad to demonstrate. Cost of all parts including valves, prepared metal chassia, wound colls and stove enamelled scale, slow motion drive, pointer, tunlng knob, tn fact everything
needed to make the complete unit, is $£ 6 / 12 / 6$. Data is Included iree with the parts or is avallable separately price $2 /$-.

THIS IS ON OFFER AT APPROX. HALF COST TO MAKE


An mpressive, costly jooking cabinet originally designed for T.V. but simple modification tuskes the cabinet sultable reflox speaker-size gisin. wide. wita deep and $37 \downarrow 1 \mathrm{n}$. high. Limited quantity at $28 / 15 /-8 a c h$, cartiage $12 / 8$.


This overhead heater warms only the area Fithla ite radiant rays, and su effects a considerable saving of fuel. Its benefts are perlod. It is essentlally a personal type of heater, having controls within easy reach of the operative. The controle give four variations of heat and "Off," At maximum heat the unit consumes 1 kW .
The Infray Major is of particular use:-
(8) In large rooms, warchouses, lofte, machine shops, etc., where the cost of level would be too greato
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Price is cy/10/-, carriage pald. ELECTRIC BLANKET WiRE Waterproot P.V.C. covered, so blanket Washable. 164 ohms per foot- $1 / 6$ per yard. It yarde, ideal for a verage blanket, El post free.

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This is a complete flourescent fitting, stove enamelled white, with starter and ballast all ready to install. Price 29/6, plus $4 / 6$ carriage and packing. 40 -watt tube $10 /$-, no extra for packing, if ordered with fitting.


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$£ 12 / 10 /=$, plus $10 \%$ cerriazs E12/10/-, plus 10/- carriage and naurance. Alternatively, if you wish to make up the unit yoursel re shall be glad to supply the
components separately. Send for the Mullard amplifer shopplag list.

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All parts to build 6 . and 12 -volt charyer Which can be connerted to a "Hat," tarted instantly. Kit comprikiog the Mailowing transformer 5 -amp rectifer
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Uses high-effeiency colls-covers long and medium wavebands and fits into the and white or brown bakelite cabinetlimited quantity only. All the parts, including cablnet. valves, in fact, everything, es $/ 10 /$-, pliss $3 / 8$ post. Constructlonal data free with the parts, or avail able separately, $1 / 8$.
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New G.P.O. telephone sets with internal bell and push button switch easily conaected together to form office intercom.

## CIRCUIT DETAILS

Dlagrams and other information extracted
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Delivery of all the above is from stock. We can also supply the LEAK F.M. TUNER. Wharfedale, Goodmans and Tannoy loudspeakers, etc., Connoisseur Variable 3 -speed Motors and all other Quality Equipment on EASY TERMS. Please send us your requirements. Quotation by return of post.
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### 237.10.0

THE
T.R. "Master-Link" FOR USE BETWEEN ANY TAPE-DECK AND ANY AMPLIFIER

The T.R. "Master-Link" is a completely versatile
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of tape-deck and any amplifier system. The advantages
are immediately apparent. Not only are considerable
expense and trouble avoided through not having to
installamplifying stages exclusively for the deck select-
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sound equipment so that much higher standards of
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Featuresinclude instant bias matching; input matching
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eurrent supply for solenoid operation where needed;
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3ł. $7 \frac{1}{2}$ and 15 i.p.s. Its fine circuitry has been
produced to ensure the highest possible standards
of recording, and it connects between tape-deck
andamplifieraseasilyas plugging in a
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Strongly recommended for the T.R. £19.10 Williamson
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Kit complete with valves, pre-amp
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GARRARD BC. 1101 ! 3 -speed mixer auto-changer unit with G.C. 2 to crysta head. Greamand brown fandah. Brand Dew In sealed manufacturer's cartons wita fiting and operating instructions

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Easily converted to 2 metres or 70 cm . In Copper-Plated metal case $3 \frac{1}{2} \mathrm{in} . \times 4 \frac{1}{2} \mathrm{in} . \times 5 \frac{1}{2} \mathrm{in}$, with dial calibrated $0-100$ and 80 V Neon Tube. Coverage approx. $190-210 \mathrm{Mc} / \mathrm{s}$. New, $6 / 6$ each, post paid.

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$10.72 \mathrm{Mc} / \mathrm{s}$. I.F.s. Frequency $100-120$ $\mathrm{Mc} / \mathrm{s}$, suitable for conversion to 2 metres and Wrotham.
Owing to a large purchase we can offer these units fully valved with circuit diagram at $25 /$ - each plus 3/- post/packing. Valve line-up: (4) EF50, (1) EL32, (2) EF39, (1) EBC33, (1) EA50.


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12 V Input 150 V 40 mA Size $5 \mathrm{ilin} . \times 5 \mathrm{tin} . \times 3 \mathrm{in}$. 12 V Input 275 V 80 mA Size $5 \frac{1}{\mathrm{in}} . \times 5 \frac{1}{2} \mathrm{in} . \times 3 \mathrm{in}$. 6 V Input 275 V 80 mA Size $4 \frac{\mathrm{in}}{\mathrm{in} .} \times 6 \mathrm{in} . \times 4 \frac{1}{2} \mathrm{in}$.

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With one $0-30 \mathrm{~mA}$ and one $0-100 \mathrm{~V} 2 \frac{1}{2} \mathrm{in}$. Panel Mtg. Meters, P.O. Switch, 3 Panel Mounting. Fuse Holders in strong polished wooden case with lined lid and handle. Size 8 in . wide $\times 5 \frac{3}{4} \mathrm{in}$. $\times 5 \frac{3}{2} \mathrm{in}$. Price $15 /-$ each. Plus $2 / 6$ p.p.

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BENDIX I.F. Transformers $1.63 \mathrm{Mc} / \mathrm{s}$. complete in cans, set of two new and boxed. Size $2 \times 1 \frac{1}{6} \times 3 \frac{s}{4} i n ., 5 /-$, p.p. $1 / 6$.
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R.F. $24 \quad 20-30 \mathrm{Mc} / \mathrm{s}$. Switched Tuning.

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2-3 mA movements in metal case measuring $4 \frac{1}{i n}$. high, 38 in . wide and 5 in . deep. $7 / 6$ plus 2/- p.p.

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This Unit consists of Magnet, and Coil which is attached to an aluminium diaphragm suspended freely and perforated to prevent air damping. Mounted on a Ceramic cover which sits over the diaphragm is a form of 2-Gang capacitor which has a swing from $10-50 \mathrm{pF}$.
The above unit is used as part of Wobbulator described on page 252 of the June "Wireless World." PRICE 7/6 p.p.


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A sub-chassis $31 \times 6 \underset{6}{2} \times 2 \frac{1}{i n}$. houses a Receiver tuned to the transmitting frequency. Contains TWO 9004 valves. For use in 70 cm . band.

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20 watt 15 ohm with internal Multitapped transformer in strong metal case measuring 10 in . dia. $\times 5 \mathrm{in}$. deep. Brand new in original packing, $£ 2 / 10 /=$ plus $7 / 6$ carr.

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Comprising two small counters. Two Desyn type follower motors (ideal for antenna direction indicators). Size of motors $1 \frac{1}{2}$ in. long, lin. dia. Terminal block 6-way, toggle sw. Housed in metal outer case, fitted with plastic 360 -degree dial. Price $8 / 6$, post paid.

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Size $10 \frac{1}{2} \times 2 \hbar \times 3 \mathrm{in}$. Frequency $9.72 \mathrm{Mc} / \mathrm{s} .2 \mathrm{EF} .92 \mathrm{~s}$ and 1 EF 91 I.F. amps. EB.91. DET/ AGC. EF. 91 AGC. Amp. and EF. 91 Limiter. Circuit supplied.

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(For use with Receiver R1392) Double Smoothed 200-250 v. 50 c Input. 240 V .100 mA .6 .3 at 6 amps , with Volt Meter reading input and output voltages. Size: $19 \mathrm{in} . \times 10 \mathrm{in} . \times 6 \frac{1}{2} \mathrm{in}$. Standard Rack Mounting. Price \&4/10/each, plus $7 / 6$ carriage.


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 Incorporating high " $Q$ " technique using the New Ferrite rod. Made possible by simple conversion of an ex-Govt. Hearing Aid.Technical Details. A Germanium Diode Detector circuit followed by the existing 3-valve Amplifier, giving adequate amplification throughout the medium wave band.

This conversion can be carried out in approximately 30 minutes.

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RECEIVER TYPE R1132
Frequency $101-126 \mathrm{Mc} / \mathrm{s}$. I1-valve Superhet.


Valve line-up: R.F. Amplifier VR. 65 (SP.61); Frequency changer VR. 65 (SP.61); Local Oscillator VR. 66 (P.61); Stabilizer VS. 70 (7455).
$3 \times$ I.F. Amplifiers VR. 53 (EF.39); B.F.O. VR. 53 (EF.39); Detector VR. 54 (EB.34); A.F. Amplifier VR. 57 (EK.32); Output VR. 67 (6J5).
Switchable A.G.C. and A.V.C. Variable B.F.O.
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* PRINTED CIRCUIT, size $7 \frac{1}{\mathrm{in}} \mathrm{n} . \times 2 \frac{1}{\mathrm{i}} \mathrm{in}$.
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Very special offer of Record Player Carrying Cases with space for almost any type of auto-changer and room for amplifier. Latest design, handsome maroon/cream finish. Length $16 \frac{\mathrm{in}}{}$., width 141 in ., depth 8 in . Size of record player baseboard, $14 \times 12$ tin. Fitted strong plastic carrying handle. Limited number only, cannot be repeated.

LARGE SELECTION OF CABINETS for radiograms, carrying cases for record players, tape recorders, TV, etc. Your enquiries invited.


BARGAIN VALUE IN SINGLE RECORD PLAYERS
Limited number only B.S.R. type TU. 8 3-speed motor and pick-up with HGP. 59 t.o. crystal, complete with two styli.

LASKY'S PRICE 92/6 Post $3 / 6$.
3 -speed motor and turntable without pick-up, 57/6. Post $2 / 6$

SPECIAL OFFER! GOODMANS 12in. AUDIOM 50 P.M. SPEAKERS

LASKY'S PRICE
10 watts. Limited quantity only
97/6

## Laskss. RADIO



## 6-VALVE RADIOGRAM

 CHASSIS COMPLETE WITH VALVESFamous Manufacturer's surplus. 6 valve 3 wave Superhet. $13-50 \mathrm{~m}$. short. 200550 m . medium, $1,000-2,000 \mathrm{~m}$. long. Brand new Mullard valves: ECH42, EF4I, L63,
EB41, 6 V 6 g.t., EZ 40 tund finest quality components. Gram. switch, $465 \mathrm{kc} / \mathrm{e}$ I.F., $13 i \times 5$ in., height $12 \nmid \mathrm{in}$. Aperture required for diule and controls $11 \times 3 \mathrm{H} \mathrm{I}$. Complete with valves, output trans., knobe, etc. LASKY'S PRICE $810 / 19 / 6$ Curriage and packing 7/6 extra.

## 5-VALVE

RADIOGRAM CHASSIS
A.C. mains, 3-wave superhet. Large full vision dial, $11 \frac{1}{2} \times 4 \frac{1}{2}$ in. Overall dimensions $14 \times 6 \times 7$ in. Valve line-up: 12AH8, 6BA6, 6AT6, 6BW6,.6X4.
LASKY'S PRICE, complete with valves,
£9/19/6
Table Cabinet for above, complete with $6 \frac{1}{2} \mathrm{in}$. P.M. speaker, $49 / 4$.


## COMPLETE 5-VALVE RADIO CHASSIS

Brand new and unused. A.C. 1 D.C. 200/250 volts. I.F. 465 kc/s. A.V.C., 4 watts output, -station pre-set, frame aerial, funy aligned, chassis $10 \times 5$ zin., wred and ready for use with Wired and ready for use, with the addition of a speaker and output transformer. Two controls, volume and station switch. Valves used: $10 \mathrm{Cl}, 10 \mathrm{~F} 9$ or UF41, 10LD11, 10P14, U404 or UY41.
REDUCED To $52 / 6 \begin{gathered}\text { less valves. Poat } \\ 3 / 6 \text { extra. }\end{gathered}$ spool:

25/-

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All leading makes and types in stock. A few examples :-
THE NEW JASON "ARGONAUT" AM/FM TUNER of brand new surplus and imported valves:-

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LASKY'S PORTABLE TAPE RECORDER KIT

- Fully fitted portable case as illustrated
- Truvox Deck, latest type,
- Tape Recorder Amplifier,

12 gns

- $7 \times 4$ Elliptical Speaker, 19/6
- E.M.I. Tape, 1,200 feet, 35/-
(7in. Plastic Spool
4/6
LASKY'S PRICE FOR $\mathbf{T} 35$ Carr. \& Pkg. 25/-.



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SPECIAL OFFER. Magnetic Recording Tape, kraft base. On Cyldon metal spools:

1,200ft., 11/6; 600ft., 7/6.
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PURETONE TAPE on plastic
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All above, 1/- extra by post.
An maker of Tape including the new thin long-playing in stock

Also all types of Spools.

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TAPE DECK MOTORS
Anti-clockwise, shaded pole. Collaro Garrard B.T.H.

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26 / 6
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super-sensitive Tune and Receiver for F.M. and medium waves. Complete parcel containing everything to build the "Argonaut," with power supplies,

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All components available separately; send for itemised price list. Chassis Assembly, complete, 57/9. I.F. and Coil Set, complete, 78/-. Post $2 / 6$.

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Special parcel containing data book, chassis, front panel, dial, drive tuning condenser, full set of coils, I.F.'s, ratio detector, etc
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Incorporates its own power supply and provides complete F.M and provides Operates with most coverage. Oper with mos Amplifier. Valve line-up EABC80, Amplifier. Valve line-up EABC80, ECC85, two EF89, 6x4 (Rect.), EM80 Indicator. Incorporate GORLER Inductance Tuning Heart, and magic eye tuning indicator.
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Dial

O Overall size $9 \times 6 \times 5$ tin. high. Complete
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2 valves and metal rectifiers, metal case. Contains power pack for 200-250 v. A.C. List £8/10/-.

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KK I Complete rit to buld thin verter, drilled chassis, condensers, resistances, coils, 2 -EF80 valves, ctc., with circuit diagrain and instructions. Post $1 / 6$

Dritied chasels only, 3/9. 48/6
MK. I. Uses latest type valves. Cancode
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 TAPE EQUIPMENTDeck. 3-speed, 3 motors, record and play back. 18 Gns.
Amplifier Mk. II. 5 watts, for use with 3 ohms speaker. Magic eye. $18 \frac{1}{2}$ Gns.
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Complete equipment with mike and tape, in carrying case, ready for use. 51 Gns.

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FILAMENT TRANSFORMERS All 200-250 v. 50 c.p.s. primary, finest quality, fully guaranteed. 6.3 v. 1.5 amp 6.3 v .2 amp .

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$4 / 6$
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Now avallable on easy terms In original wood transit cases Brand New Secondhand, Grade I Ditto, Grade 2 Carr. $12 / 6$.
Power Pack and Output Stage with $6 \frac{1}{2}$ in. Speaker. . 555


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Push pull, very high gain. 4 valves: two ULA1 (p.p.), UCH42, UAF42. Input voltage $100 /$ 100 A.C./D.C. Easily converted to 230 v . Ideal for record players, tape recorders, baby alarms, etc. Supplied fully assembled with valves, circuit diagram and details.
REDUCED TO 50/-

REMPLOY LNSTRUMENT SOLDERING IRONS. 200-250 $\begin{array}{ll}\text { Neon indicator } & 19 / 3\end{array}$

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GEX. 34
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## MAGNIFICENT VALUE IN TV CABINETS

THE DE LUXE. Complete with mask, glass, castors, shelf, bearers, C.R.T. néck end protector, back, speaker, fret and baffle board. Finished in beautiful figured medium light or dark walnut veneer, with high polish. Receivers, including the "Viewmaster," "Practical Television," "Tele-King," "Magnaview" "Wireless World," etc Supplied with cut-out for 14 in ., 16 in . and 17 in CR allowance of $4 / 6$ will be made if the mask is not required.
Inside dim.: Depth $16 \frac{1}{2} \mathrm{in}$., width 17 3in. Height 2 sin . Overall height 32 in . Width $18 \frac{1}{2} \mathrm{in}$. Adaptor frames for fitting 9 in . or 18 i in. Adaptor frames for fitting 9 in. ${ }^{\circ} \mathrm{i}$
loin. C.

I.ASKY'S PRICE $\mathbf{C 8 / 1 0 / =}$
Carriage $12 / 6$.

Carriage $12 / 6$.


THE ROTHESAY. Outstanding contemporary design. Absolutely rigid construction throughout with the finest-laminated woods, veneered in walnut, polished light, medium or dark shade. Fitted with gold anodised speaker grille. C.R.T. aperture frame is detachable, supplied to suit any size tube o order. NOTE SIZES:
Outside dim.: 341 in . high, 21 kin . wide, 21 in . deep. Inside dim.: $18 \frac{1 \mathrm{in}}{}$, wide, 19 kin . deep. Size of top: $221 \times 21$ gin. Thickness $\frac{1}{2}$ in

LASKY'S $89 / 19 / 6$ Carriage 15/WITH FULL-LENGTH DOORS veneered both sides, polished to match the cabinet and mounted with full-length piano hinges. $£ 14 / 9 / 6$. H.P. Terms arnanged for any of above Cabirets.

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All specified components used, with your choice of transformers and chokes. Fully assembled and ready for use.
The MULLARD 510. Price, according to transformers used, from 15 GNS. The Book, 3/6. OSRAM 912. From 19 GNS. The Book, $4 /$-, post free. All components available separately, also printed Circuits.


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light in handle
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## PORTABLE GRAM AMPLIFIER

Uses 3 latest miniature valves, U78, N78, DH77. Volume, bass and treble controls; extension L.S. socket and internal L.S. switch, indicator lamp. Mounted on wood baffle, overall size $14 \times 4$ \}in. with speaker centralised. All top quality new components. For A.C. mains, 200-250 v. Ideal for portable record players, input will match Monarch, RC54, 3/554, etc.
Price, complete with 3 new Osram valves, $7 \times 4 \mathrm{in}$. Goodmans elliptical speaker, métal speaker grille, mains lead, and knobs. $\& 5 / 9 / 6$

Carr. 5/-

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All 200-250 v. 50 c.p.s. primary, finest quality, fully guaranteed. MBA/3. $350-0-350$ v. 80 mA . 6.3 v. 4 a., 5 v. 2 a. Both fila6.3 V. 4 a., 5 v. 2 a. Both filaMBA/7. 250-0-250 v. 80 mA 6.3 v. 3 a., 5 v. 2 a. Both fila6.3 V. 3 a., 5 v. 2 a. Both fia-
ments tapped at 4 volts. 19/6. ments tapped at 4 volts. $19 / 6$.
AT/3. Auto trans. $0-10-120$, AT/3. Auto trans. 0-10-120,
$200-230-240$ v. 100 watts. $19 / 6$. $\begin{array}{ll}\text { 200-230-240 } \\ \text { MT/340. } 100 \text { watts. } & \text { Tapped input } 200 \text {. }\end{array}$ MT/340. Tapped input 200
$250 \mathrm{v} .300-0-300,100 \mathrm{~mA}, 5$ $3 \mathrm{amp}, 6.3 \mathrm{v} .1 .5 \mathrm{amp} .{ }^{2} 16 / 6$. MT/341. Tapped input 250-0$250,120 \mathrm{~mA} ., 6.3$ v. 5 amps , fully shrouded. $27 / 6$.

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 WIDE ANGLE 38 mm . Line E.H.T. trans., $\mathcal{\text { Serroxecube }}$ core, 9-16 kV. Iow imp. line and Scauning Coils, low imp. line andframe ...........................
frame
Ferrox-cube cored Scanning Coils
and Line Output Trans., $10-15$ kV, EYporates width and linearity control. Complete with elrcuit dlagram the pair
rame Output Traneformer
scanning Colls low imp. Hine and Frame or
rame or line blocking osc. (rane ocus Magneta Ferrox-dure P.M. Focus Magnets, Iron Cored Duomag Focallsers $300 \mathrm{~m} / \mathrm{a}$. Smoothing chokes Enstromagnetio cocals Line Output Transformers. No
E.H.T.
Line Output Transformers 6.9 kV .
E.H.T. and 6.3 \%. winding. Ferrox-cule . v. winding Scanning coils. Low mmp . Hne and frame..
Ditto by Igranic,
Frame or line blocking oscillator
Framesormer biocking cecilator
Frame output transformer.
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Focus Colls Flectro-magnetic
$200 \mathrm{~m} / \mathrm{s}$. Smoothing Chokes

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A 7 -valve 3 -waveband superhet chassis having a Push-Pull stage for approximately 2 watts output. PRICE E12-19-6 H.P. TERMS: Deposit Plus $7 / 6$ carr, \& lan. (E6/9/9 and 8 monthly payments of $18 / 9$

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THESE CHASSIS HAVE "GRAM " POSITION and are IDEAL REPLACEMENT CHASSIS FOR THAT "OLD RADIOGRAM "-Send S.A.E. for complete details.

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The following items form complete Recorder. Plus attractive Carrying Case.

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3 Shated-Pole motors. Drop-in Tape Loading. Push Button Control. Separate Push Button Brake. Fast forward and fast reverse. Silent drive eliminating Wow and Flutter. Half Track working and 2 speeds, $3 \frac{1}{8}$ in. and $7 \frac{1}{2} \mathrm{in}$. per sec. Positive Azimuth Adjustment. Overall size only 14$\} \times 12 \frac{1}{4} i n$. Available for $£ 23 / 2 /$.
THE MODEL T.R.I./F. AMPLIFIER Has been expressly designed to meet the requirements of enthuslasts for fidelity reproduction, and in particular to CORRECTLY operate the above TRUVOX DECK, it is supplied complete with a matched Elliptical 3 ohm P.M. Speaker, it incorporates an efficient Tone Control arrangement and has a Magic Eye Indicator (Operative on Record). A Co-axial Socket is also incorporated for MONITORING on amplifier. The Amplifier can also be used for high qualizy reproduction of gramophone records direct from a gram unit. Available for $£ 14 / 14 / \%$.

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NEW : \& A COMBINED AM and FM RADIOGRAM CHASSIS of EXCEPTIONAL HIGH QUALITY and VERY PLEASING APPEARANCE. PRICE IS ONLY $£ 21 / 10 \%$. SEND S.A.E. for full details.
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Comprises the MAIN AMPLIFIEB of very popular
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press purpose of ite use
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side. TWO PGE-AMY LacriER illustrated along. resultant reproduction is genuinely in the BuGH fidelity oat la ridiculously low
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The Portable Case (with room to
house Autochanger,
§3,17/6 $6^{(\text {thuas }}$ (earr) Amplifier with Speaker, has separa $44 / 12 / 6$ (olun 4/ ate Bass and Treble Coatrolis, Send S.A.E. for ILLUSTRATED and DESCRIPTIVE leaflet.
EXCEPTIONAL OFFER FOR CASH The B.S.R. ONLY MONARCH 3-SPEED AUTOCRANGER
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## MIXED 7 in

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- They have aeparate amphires for L.P. and 78 r.p.an. which are moved Into positton by a single awitch.
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NEW 1956 DESIGN. HIGH FIDELITY PUSH-PULL UNIT EMPLOYING SIX VALVES. Tone Control Pre-amp stages are ncorporated. Sensitivity is extremely high. for full output. THIS ENSURES THE SUITABILITY OF ANY TYPE THE MAKE OF MICROPHONE OR PICKUP. Separate Bass and Treble controls give both "lift" and "cut" with ample tone correction for long playing wecords. AN correction for long playing records. AN INCLUDED FOR SUPPLY OF 300 v . 20 mA , and 6.3 v .1 .5 a . FOR A RADIO FEEDER UNIT. Price in kit form with easy-th-follow wiring diagrams. Only GNS. Or Factory built with 12 months' guarantee, 50/- extra. TERMS ON ASSEMBLED UNITS with extra input. DEPOSIT 28/9 and 9 monthly payments of $28 / 9$. If required an extra input with associated vol. control can be provided so that two separate inputs such as "mike and gram., etc., etc., can be simultaneously applied for mixing purposes.


Type 807 output valves are used with High former specialiy designed for output transoperation. Negative feedback of $17 \mathrm{D} B$ in main loop. CERTIFIED PERFORmain loop. CIGURES ARE EQUAL TO MANCE FIGURES ARE EQUAL TO MOST EXPENSIVE UNITS AVAIL-$30-20,000 \mathrm{c} / \mathrm{cs} ., 12 \mathrm{D} . \mathrm{B}$. "lift "at $50 \mathrm{c} / \mathrm{cs}$ $30-20,000 \mathrm{c} / \mathrm{cs}$. ," 12 D.B. "lift" at $50 \mathrm{c} / \mathrm{cs}$.,
12 D.B. "lift" at $12,000 \mathrm{c} / \mathrm{cs}$, Hum and noise 70 D.B. down. Good quality reliable components used. Chassis finish blue crackle. Overall size $12 \times 9 \times 9$ in. approx. Power consumption 150 watts. For A.C. mains $200-230-250 \mathrm{v} .50 \mathrm{c} / \mathrm{cs}$. Outputs for 3 and 15 ohm speakers. EQUALLY SUITABLE FOR THE CONNOISSEUR OR FOR LARGE HALLS, CLUBS, or OUTSIDE FUNCTIONS. IDEAL FOR USE WITH MUSICAL INSTRUMENTS SUCH AS STRING BASS, ELECTRONIC ORGAN, GUITAR, etc. FOR DANCE
 R.S.C. TAI HIGH QUALITY TAPE DECK AMPLIFIER
FOR ALL DECKS WITH HIGGH IMPEDANCE RECRDDPLAYBACK AND
ERASE MEADS. Snch as Lane, Truvox, etc., or matched to low impedance erase heads as fitted latest COLLARO TAPL TRANSCRIPTOR. Chassis
 Output for standard $2-3$ ohm speaker. Only 15 millivolts laput required for
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 can be used ant gram. ampliffer with input of 0.75. . . R.M.B. Negative fee

11Ready for use Facilitles for recordings at 15in., 7ina. or 33 lin . Ready for ure per secono. Authmatic equalination at ecord tirn heads Is ansured. PERFORMANCE IS IS COMPARABLE WITH UNTTE AT OVER TWICE THE COBT. LEAFLET 6 d .

GARRARD 3-SpEED AUTOMATIC RECORD CHANGERS. Latest Model. Type RC110. Fitted high fldelity turnover crystal plok-up head. For $200-250$ v. A.C. mains. Limited
cumber. Brand new cartoned. Only
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Unapproachable value at $57 / 15 /-$ or factory built 45/- extra. Carriage 10/-. If required louvred metal cover With 2 carrying handles cas be supplied for 17/6. TERMS ON ASSEMBLED UNITS with exira"input as mentioned below. DEPOSIT £2/19/6 and nine monthly payments of £1. control is requited for mixing purposes thls can be provided ior 13/-extra.

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Controls are Tuning, w., Ch, and Vol. Output will load Controis are Tuning, W.. Ch, and Vol, Output will load A.e. location. Only 250 v. 18 mA . Y.T., and L.T. off 6 . A v 1 amp requircd from ampliffer. Bizo of unit approx. $9-6.7 \mathrm{in}$. high Diagrama and Instructions, $2 / 6 \mathrm{~d}$. Complete set of parts £ 2.15.0.

## What the reviewers say about the RD JUNIOR <br> This month we have devoted our advertisement to extracts from two reviews which have

 appeared recently dealing with the RD JUNIOR Amplifier and Control Unit. Written by acknowledged authorities in the high fidelity field they speak for themselves..."This amplifier has deservedly established a high reputation for itself. Wherever I have wandered during the past year, I have found universal praise for it from dealers and others who have had an opportunity of putting it through its paces."
"The amplifier is inherently stable under quite adverse conditions of load, and this means not only that high output power can be maintained with low distortion over the whole frequency range, but that the condition of stability will persist even as the valves and other components begin to age. No wonder the makers can afford to give " an unconditional two year guarantee against faulty materials and workmanship."
"The control unit has also been designed to give the utmost that can be expected from an economical unit. It only uses one double triode valve and yet gives not only half a dozen different input arrangements and independent treble and bass controls (the former continuous, the latter in steps) but also a variable steep-cut treble filter without the use of chokes."
"In performance the amplifier has given no surprises: just the clear, crisp, effortless power that one expected from its specification. Which means that it takes a place as one of the best three or four 10 -watt amplifiers that are available on the British market at present."
(The above extracts are taken from TECHNICAL REPORT by P. Wilson, M.A., in the July 1956 issue of THE GRAMOPHONE).
"It can be truthfully stated that the RD JUNIOR is an established product and is widely used as the basis of many
average domestic high fidelity systems. For this reason it was chosen by us as one of the equipments installed for record playback purposes in the Gramophone Record Review demonstration room at the London Audio Fair.'
"A convenient method is adopted for altering the output marching by means of an appropriate "impedance plug" which is plugged in and the correct feedback resistor (for the three impedances) is automatically placed in circuit."
"The units are constructed on steel chassis heavily primed and stove enamelled in an attractive dark maroon. The components are of first grade with the valve and associated components in the control unit mounted on a floating sub-chassis to avoid microphony. The workmanship is superb, and the neat wiring layout can be seen in our under-chassis view."
" Listening quality is of a high order and I found the controls, particularly the low pass filter used in conjunction with the treble tone control, flexible enough to cope with modern Lp's, old or worn 78 's, and poor radio transmissions. Apart from the electrical and mechanical features of this design, the purchaser will be given confidence by the fact that the manufacturer is willing to supply complete circuit diagrams and components lists and guarantee the apparatus (excluding valves) for a period of two years."

This RD JUNIOR Amplifier/Control Unit exemplifies a much sought after, but not so often achieved, combination of high quality with moderate price."
(The above extracts are taken from HOME TEST No. 27 by Donald W. Aldous in the July 1956 issue of the Gramophone Record Review).

A 12-page Illustrated Booklet devoted solely to the RD JUNIOR Amplifier and Control Unit, together with
reprints of the full reviews appearing in the Gramophone and the Gramophone Record Review, post frec on request. Send also for details of companion equipment in the RD JUNIOR Home High Fidelity System-the RD JUNIOR FM Unit and RD JUNIOR Corner Horn enclosure.

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| UF41 | 11/- |  | 81- | EBF80 | 11/6 | ECC84 | 12/6 |
| UCE 42 | $12 / 6$ | PY88 | 101- | EFB6 | 10/6 | 6AQ5 | $101-$ |
| UBC41 | 101 | PCCs4 | $12 / 6$ | EFP99 | 10/6 | ${ }^{\text {PCLE } 22}$ | 12/6 |
| DK40 | 10\% | PCF82 $12 A U 7$ | 12/6 | EF88 | 12/6 | 3.45 | $12 / 6$ $12 / 6$ |

[^13]
## TRANSISTORS <br> JUNCTION TYPE (Red.Spoet) OEFERED AT LESS THAN HALF-PRICE.

Designed for A.F. application up to $800 \mathrm{Kc} / \mathrm{s}$ and is suitable for use in Radio Control, Signal Tracers, Local Station Receivers, Oscillators, Transistor Voltmeters, Microphone Pre-Amplifiers etc.
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(Testef and complete with Dita and Circuits)
N.B. These Transistors may be used in place of Mullard OC7I or similar Transistors.
R.F. TRANSISTORS (BLUE SPOT) $1.6 \mathrm{Mc} / \mathrm{s}$ 15/- each.

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Complete KIt of Parta with 4. Tranalstors and 3in. Speaker. L.P.T.'s. 2-gang minuature cond. V/C. Ferrite Rod, Cond. snd Res. £6/10/-
" TELETRON " Tranaiator Superhet with Push-Pall Output, 6 Transistors, $6 \times 4$ Euliptical Speaker, I.F.T.'s, 2-gang miniature coud. V/C. Ferrite Rod, Cond. and Rea. car/-
(Call and hear Demonstration Modely worklng.)

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High Speed Vmcuum tester,
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## U.S.A. INDICATOR UNIT

Complete with 3BPI C/B tube and ncreen. 7 valves-2-68N7GT, $2-6 \mathrm{HBGT} .6 \mathrm{~GB}, 2 \times 2$.
$6 \times 5 \mathrm{a}$, volume controls condensers, etc. $6 \times 50$, volume controls, condensers, etc.
Ideal for portable iscope. In black crackle case stze 15 in $\times 91 \mathrm{~m} \times 9$ in. BRAND NEW. 65/-, catr. FREE.

$$
\begin{aligned}
& \text { TRANSMITTER RECEIVER } \\
& \text { " } 38 \text { " WALKIE TALKIE SETS } \\
& \text { Speclal ofier of above set, complete with } 5 \\
& \text { valves, 4-ARP12 and ATP4, with circult. } \\
& \text { Range } 7.4 \text { to } 9 \text { Mc/s. These sete are not } \\
& \text { guarateed but are serviceable. } \\
& 25 /- \\
& \text { Junction bux } 2 / 6 \text { extra. }
\end{aligned}
$$

62A INDICATOR UNIT
Containing VOR97 with Ma-Metal Screen. 21V I lves: 12 -EFBO, i-RP61, 3-EA50, 2-FB34. Plus Pots., Switches, H. $\overline{\text { P. Cond., Resistors }}$
Muirhead S/X Dlal: Double Deck Chasals Muirhead s/M Dlal: Double Deck Chassis
and Crystal. BRAND NFW ORIGINAL and Crystal. BRAND NFW
CASES, 6\%/6. CARR, FREE.


## INDICATOR UNIT TYPE I82A

 Uall contaius VCR5 17 Cathode Ray fin. tubc, complete with Mu-Metal screen. 3EF50, $48 P 61$ and 1 SU4G valves, 9 wire. wound volume controls and quantity of respators and condensere. Offered BRAND NEW (less relas) at 67/6. Plus 7/6 carr "Radio-Constructor" scope circuit included.


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Minlature w/W IV V/Controlo
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Teletron minfature oscillator
colls . . . . . . . . . . . .........
616 each
Teletron malaiature I.F.T.'s
colls
$6 / 6$ each

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Complete wlth 11 valves 8-SPBI 5 U 40 , VU120, V R92. As apecified tor inexpenslve T..

In absolute new condition. 27/6, eart $5 / \mathrm{F}$
R.F. 24 10/- R.F. 25 12/6. R.F. 26 25/-

BRAND NEW WITH VALVES. Carr. 2/6.

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VCRsl7\%. Guaranteed full T/V
MU-METAL SCREENS for
6in. ENLARGER for VCR97 or
6 in . ENLARGER for VCR97 or
j 17 . P.P. $1 / 6$
VCR97. Slight cut-on. Carr. of
3BP1. Brand new
£1 150
£1 150
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£1 100

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Brand new and guaran'eed Lleted at $£ 16 / 10$. e7/19/6. carr. paid.

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 STRIP "TYPE 81"Size $7!\mathrm{in} . \times 6 \mathrm{in} . \times 3 \mathrm{in}$. Complete with Valver Type OV415, CV309, 2-6AM6, 2-7D9 snd Quartz Grystal, $4,860 \mathrm{kc} / \mathrm{s}$. Fully wired with oircuit.

84/10/- complete.

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"373" 9.72 MEG.
Brand new miplatare 1.F. Strip aize 101 fn . $x$ $2\} \mathrm{in} \times 3 \mathrm{in}$. bigh. Vatsn liue-up.
3 -EP91 and E891. W'th circoit.
Price (lesa valven) 7/8.p. \& p. $1 / 8$, This I.F. Strip is part of above equipment.


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Drod thro' $350-0.350$ v. $70 \mathrm{~mA} .6 \quad \vee$. 2.5 smp. 5 จ. 2 amp., $14 / 6$.

Chassis mounting or drop-thro*. Pri. $110-150$ ₹. Bec. $350-0-350,250 \mathrm{~mA}$.
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$3 / 6 . \quad 32 / 6$.

Chassis mounted and fully ahrouded, $80 \mathrm{~mA} .6 \mathrm{v} .3 \mathrm{mmp} ., 5 \mathrm{~F} .2$ amp., $14 / 6$. Drop, thro' $270 \cdot 0-27060 \mathrm{~mA} ., 6 \mathrm{~V}$. 3 amp., 11/6.
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$5 \%$
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R. \& T.V. Energised 6 gin. Spesker.
with o.P. itrane. field coll.
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T.V. Coils, moulded former, froncored wound ior re-winding purposes
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Used Metal Rectiffer, 250 v. 150 mA .,
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Standard type, 8,000 ohms imp., 4/9, $4 / 3$. Miniature $4-2,13 / 3$. Multiratio $3,500,7,000$ and $14,000,5 / 6$.
10 watt push-pull, 6 V 6 matchtig, $10-$ watt push-pull, 6V6, matching,
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3-SPEED AUTOMATIC CHANGER
WHitake 10 ree ords, 7 in ., 101 n . or 12in. Mixed, tarnover crybtal head, brand

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COMMERCIAL TELEVISION CONVERTER
SUITABLE ANY T.V. using lower side band
NO ALTERATIONS TO SET
Complete with built-in power supply, 230-250 r. A.C. malns. Crackle finish case 611 u , long, 31 in . wide, $4 / \mathrm{in}$ high. Incorporating galn control and band switch.


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comprising 5 chamel tuner, mains transformer, metal rectifer, electrolytio, 2 valves 12AT7 data, $£ 2 / 5 /-$. P. \& $P .2 / 6$.

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Comprising :in. moving coil meter, scale callibrated in A.C./D.C. volts, ohms and malliamps. Voltage range A.C./D.C, $0-10$, 0.100 and 0.500 . Milliamps $0-10,0-100$; ohins $0-10,000$. Front panel, range switch, wire wound pot (for ohms zero setting), two toggle switches, resistors and meter rectifier.
Campare tinawe grey $19 / 6$
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Coverage $120 \mathrm{Kc} / \mathrm{s} .230 \mathrm{Kc} / \mathrm{s}$,
$300 \mathrm{Kc} / \mathrm{s} .900 \mathrm{Kc}, ~$ $300 \mathrm{Kc} / 4 .+900 \mathrm{Kc} / \mathrm{s.0} .900 \mathrm{Kc} / \mathrm{B} .2$.
 $\mathrm{Mc} / \mathrm{s}, 24 \mathrm{Mc} / \mathrm{M} \cdot 84 \mathrm{Mc} / \mathrm{B}$. Metal ouse 20 in . $\times 6 \mathrm{fln} \times 4 \mathrm{in}$ in. Size of scale, 641 n . $\times 3$ 3.ln. 2 valves 250 v . Internal modulation of 400 c.p.s. to a depth of 30 per cent., modulated or unroopulated
R.F., output continuoualy rarR.F., outpur continuoualy variable 100 milli-volts. C.W. and mod, ewitch, variable A.F. output Mad moving coll output
meter. Rlack crackle finlohed meter. and white panel. Acouracy case or nilnus 2\%. $£ 419 / 8$
or 34i- deposit and 3 monthly or 34/- deposit and 3 monthly
payments $25 /-$. $P$. it $P$.
$4 / 6$ extra.

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 IN PLASTIC CABINET3 vaive plus metal rectifer, A.O. maius 2000250 r . Medium and long wave日. In pastel blue or brown. Valve line-up:
VR65S and VT5a, Size 15 and VR65s and VT52, Size lyitn. long by 9in. high by 7hn deep. $£ 3 / 18 / 6$. P. \& P. $4 / 6$.
A point-to-point wiring diagram. 1/6. Free 4 point-to-point wi

All parts supplied separately.


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Coverage $7.6 \mathrm{Mc} / \mathrm{s} .-210 \mathrm{Mc} / \mathrm{s}$ in five bands, all on fundamentals, slow-motion tuning, audio outpur, 8 vertical and horizontal bars, logging scalc.
In grey hammer finished case with carrying handle. Accuracy $\pm 1 \%$.
A.C. mains 200-250 v.
£6-19-6 P. \& P. 5/6.

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GOODS NOT DESPATCHED WHERE CUSTOMS DECLARATION IS APPUCABLE
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As above, but complete with line and 25 mA . choke transiormers. 5 Henry $250 \mathrm{wkg}, 380 \mathrm{~mA}$. AC. ripple. $£ 2 / 9 / 6$. P. \& P. 3/-

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Combined 12in. mask and escutoheon In lightly tinted Perspex. New aspect edged in brown. $12 / 6$. As above for 15 In. tube, cabine
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BY FAMOUS MANUFACTURER


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We can supply meters with NON-STANDARD CURRENT and VOLTAGE RANGES to any specification. DELIVERY $7-14$ days. MCV. ING IRON THERMO AND ELECTROSTATIC INSTRUMENTS. 2t in SQUARE FLUSH, $3 \frac{1}{2}$ in. ROUND FLUSH and INDUSTRIAL SWITCHBOARD INSTRUMENTS ALSO AVAILABLE FOR PROMPT DELIVERY.

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Range 7.5 to $330 \mathrm{mc} / \mathrm{s}$. in 5 bands. Calibration Accuracy: 2\%.
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22-A SIGNAL GENERAT
FREQUENCY METËR
Crystal controlled, range $8-15 \mathrm{me} / \mathrm{s}$ and 150 . $230 \mathrm{mc} / \mathrm{s}$. PRICE, with accessories $£ 13 \quad 50$

[^14]

This is a portable Precision Heterodyne Type Frequency Meter.

* High Stability.
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(Aluminium body), with built-in Decade Box. PRICE, completely overhated ... $£ 35$ 0 0
G.E.C. TYPE BW232 E.H.F. SIGNAL GENERATOR, range $500-1,000 \mathrm{mc} / \mathrm{s}$. PRICE

EDDYSTONE TYPE 358 X (Admiralty
Type B-34) COMMUNICATION Type B-34) COMMUNICATION
RECEIVERS (OMM


Range $90 \mathrm{ke} / \mathrm{s}$ to $31 \mathrm{Me} / \mathrm{s}$. obtainable with 9 Pjug


Selectivity: $2 \mathrm{kc} / \mathrm{s}$ at 2.5 db down; $5 \mathrm{kc} / \mathrm{s}$ at 35 db down; $150 \mathrm{c} / \mathrm{s}$ at 40 db down with crystal.
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Image ratio: from $33 / 1$ at $20 \mathrm{mc} / \mathrm{s}$ to $500 / 1$ at $3 \mathrm{mc} / \mathrm{s}$.
Output: 1.5 watts max. into 120 or 2,000 ohms.
Supply requirements : 6 v. 1.4 amp. ; $175 / 180 \mathrm{v}$. Supply
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Circuit: variable mu pentode HF amplifier. triode-hexode frequency changer, two If triode-hexode frequency changer, two IF
amplifiers ( $450 \mathrm{kc} / \mathrm{s}$ ); crystal filter, AVC/ amplifiers ( $450 \mathrm{kc} / \mathrm{s}$ ); crystal filter, AVC/
detector $/ \mathrm{AF}$ amplifier, output stage, BFO, valve check meter.
Price, complete with set of 9 plug-in coils (not aligned), aerial checked ...... $\leqslant 12100$ Packing and carriage ..................... \&I 10 . * These Receivers can also be supplied completely overhauled to manufacturers specification, aligned with Test Certificate, suitable for use as RF. bridges, etc., at a price of

MARCONJ TYPE TF-428 (or equivalent ex-service type) valve voltmeters, range 0.150 voles, 817 .

RECORD BONDING TESTERS (Jow range ohmmeters), range 0 to 0.10 ohms, complete. with Alkaline Cell and matched pair of leads, approximately 60 and 6 ft . long respecrively........................

HALLICRAFTER S-27C COMMUNICATION RECEIVER, range $130-210 \mathrm{mc} / \mathrm{s}$ AM-FM reception ...................... $£ 90.00$

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Compiled by "Wireless World"

All European long and medium-wave broadcasting stations and over 1,900 short-wave transmitters in 125 countries are listed both geographically and in order of frequency and wavelength in this revised edition. The information, checked against measurements made at the BBC receiving station at Tatsfield, has been corrected to mid-July. Frequencies of over 500 v.h.f. broadcasting stations in Europe and operating details of European television stations are also included.
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500 microamps D.C.
1 ma. D.C.
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$10 \mathrm{~m} . \mathrm{a}$, D.C.
$150 \mathrm{~m} . \mathrm{a}$. D.C.
$200 \mathrm{m.a}$. D.C.
$500 \mathrm{~m} . \mathrm{a}$. thermo 4 amp . D.C.
20 amp . D.C.
40 amp . D.C.
30-0-30 amp. D.C.
15 volts A.C.
300 voles A.C.
2 Kilovoles A.C...
300 volts D.C.

SIZE AND TYPE
2 in. Flush circular
$2 i n$. Flush square
$2 i n$. Flush square
2in. Flush square (scaled $0-100$ m.a.).

2 in. Flush circular (blank scale) 2 in. Flush square
$2 \frac{1}{2} \mathrm{in}$. Flush circular
2in. Proj. circular
$2 \frac{1}{2}$ in. Plush circular
2in. Proj, circular
2in. Prój. circular
Car type moving iron $2 \frac{1}{2}$ in. Flush circular moving $2 \frac{1}{2}$ in. Flush circular moving $2 t \mathrm{in}$. Proj. circular electro$2 i n$. Flush square

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Price $£ 21$, plus $K 2$ carr, and pkg. ( $K 1$ refund Price $£ 21$, plus $\& 2$ carr, and
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## R.C.A

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range
50 Microamp.
100 Microsimp. 500 Microasup
amp.
1 Milliamp.
100 Miliamp.
100 Milliamp.
150 Milliamp.
150 Milliamp.

| 200 Mml |
| :--- |
| $\begin{array}{l}1 \\ 4 \\ 4 \\ 4 \mathrm{Amp} \text {. } \\ 20\end{array}$ |

${ }_{20}{ }^{2} \mathrm{Amp}$ Ap.
$30-0-30$ Amp
15 Volta
15 Volts
300 Volts
300 Volte
METER RECTIFIERS. Salford Instruments, I mA, 8/6; 5 mA , $6 / 9$ : S.T.C., 2 mA ., as used in E.M.I. Output Meter, $5 / 6$. All are full wave bridge and brand new.
 T.in. diam. spindle, we. $50 z$. Will work from 6 v . dry battery and are reversible. Has ball bearings, $8 / 6$.

INDICATING UNIT 277. Case size $5 \frac{1}{2} \mathrm{in}, \times 7 \mathrm{in} . \times 12 \mathrm{in}$, deep. Contains lin. C.R. tube type VCR 522 (same as that used in G.E.C. "Miniscope"), 4 VR91, 2 VR92, and a host of useful modern components. Fitted with "Focus" and "Brightness" controls, etc. Should convert to useful miniature ascillostope. All tubes tested. 59/6.
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amp., and 5 v .2 amp . BRAND NEW exU.S.A. $29 / 6$.

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An elegant cabinet in richly figured walnut veneer, internal panels in polished sycamore. A drop front lid covers a sloping, uncut control panel ( 16 in . long $x$ $10_{\frac{3}{3}}^{3} \mathrm{n}$. high) alongside which is an uncut base-board ( $15 \frac{3}{2} \mathrm{in}$. long $\times 13 \frac{1}{2} \mathrm{in}$, back to front). The inside of the drop front lid is panelled in beige leatherette. In the lower part of the cabinet are two large torage cupboards ( 13 t in . high, 7 itin . wide, $16 \frac{1}{2}$ in. deep). The lid and cupboard handles are in chased Florentine bronze. Overall dimenslons (33in. high, 34in. long, $16 \frac{1}{2} \mathrm{in}$. deep.)
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200 Assorted Moulded Mica Condensers, popular values......... £2 10 200 Assorted Silver Mica Condensers, popułar values................ $£ 210$. 200 Assorted Carbon Resistors, $\frac{1}{4}$, $\frac{1}{2}$ and I watt. Good selection €| 10 PAXOLIN SHEET. $18 \times 4 \frac{1}{2} \times \frac{1}{1} \mathrm{in} ., 1 / 6 ; 10 \times 10 \times \frac{1}{3} \mathrm{in}, 1 / 6 ; 20 \times 10 \times$ 直 in $3 /-; 10 \times 10 \times \frac{1}{16}$ in: $2 /-; 20 \times 10 \times \frac{1}{16}$ in, 4/-. Minimum $P$ \& Pkg. $1 / 6$.

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B.T H. CRYSTAL DIODES, $1 / 3$. Very special price for large quantities. RESISTORS. $\ddagger$ watt, $2 / 6$ doz ; $\frac{1}{2}$ w=tt, $3 /$-doz ; I watt, $4 /$ - doz ; 2 watt, 6/- doz.
W.W. RESISTORS. 5 watt, $1 / 6 ; 10$ watt, $2 / 6 ; 15$ watt, $3 /-; 20$ watt, 3/6. We carry stocks of resistors from 2 watt to 150 watt W.W. Your enquiries invited.
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VALVE HOLDER FITTED WITH LOWER CAN, $1 / 6$ per doz
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Double wound Transformers, Pri. 250 v., Sec 130 v. 18 amps. £6/10/-, carr. 7/6. We have a large selection of heavy duty LT Transformers. Let us know your requirements. Alkaline Batteries. Crates of five cells giving 6 v , at 58 AH . Size of wood crate 15 x $5 \frac{1}{4} \times 1, \frac{1}{2}$ in. $65 / 19 / 6$. Carr 7/6. 6 v. 75 AH E6/19/6, carr. 7/6. 24 v. 125 AH, $£ 3 / 10 /$, carr. 5/-. Willard Aircraft Batteries, 24 v. If AH. Size $8 \times 7 \frac{1}{2} \times 7 \frac{1}{2}$ in. New in maker's cartons, $49 / 6$, carr. 7/6. Exide 10 v. 5 AH Glass Accumulators. Size $7 \times 2 \frac{12}{3} \times 1 \frac{1}{8} \mathrm{in}$. Suitable for HT construction and models, ete. New in maker's cartons, 8/6, P.P 2/6.
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There are numerous advantages to be gained in joining one of the enthusiastic teams engaged in advanced super priority fields of research, design and development.

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ELECTRONIC ENGINEER to head à section being formed to cover technical liaison and servicing of Guided Weapon test equipment. Training will be given but industrial or Service experience is needed. Ref. 41 J.

ELECTRONIC ENGINEER for a newly formed Reliability Section concerned with breakdown of defect information assessing from all stages of manufacture and use, recording, auditing and final assessment. Ref. 42 J.

## TELEMETRY DEVELOPMENT

 ENGINEER for design and development of equipment. Preferably with degree in engineering or physičs, but at least H.N.C. and four years' experience in this or related fields. Ref. 44J.TELEMETRY ENGINEER for practical work with minimum of direction. An opportunity for recently graduated engineer to gain valuable experience of techniques prior to promotion. Experienced applicants without academic qualifcations also invited. Ref. 45J.

TELEMETRY APPLICATIONS ENGINEER for selection and testing of units for specified functions. At least two years' experience, practical interest in electro-mechanical apparatus and considerable initiative. Ref. 46 J.

TECHNICAL ILLUSTRATOR for Technical Publications Department, able to read engineering drawings and produce line illustrations of all descriptions. Ref. 53J.

TECHNICAL AUTHOR - Electronic, for preparation of descriptive and instructional literature. Ability to read engineering drawings essential, knowledge of methods of print production and associated processes an advantage. Ref. 55J.

ELECTRONIC ENGINEER for research design and some development work on part of a Guided Weapon system. A responsible post for a man of initiative, preferably of degree standard and with at least two years' experience. Ref. 57J.

ASSISTANT ENGINEER to work with well qualified engineers engaged on Guided Weapon system design. Knowledge of electronics or mechanical engineering required. Good promotion prospects. Ref. 58J.

LABORATORY ASSISTANT for Guided Weapon systems research and development. Some electronic experience needed. Improvement of theoretical and practical knowledge will be encouraged. Ref. 59J.

ENGINEER fully conversant with construction and wiring of subminiature electronic equipment. Machine tool operating experience an advantage. Interesting and progressive work. Ref. 61J.

ELECTRONIC ENGINEER to lead a small team engaged in design and prototype construction of low frequency, low noise, high gain amplifiers and associated measuring equipment. Will also be concerned with problems related to instrumentation of high altitude aircraft and other projects. Applicants must be physically fit and willing to take part in flight and ground trials. Ref. 63J.

WIREMAN to produce finished model from sketches and verbal instructions, requiring some design and testing knowledge. Excellent prospects. Ref. 60 J .

## LABORATORY TECHNICIAN for

 wiring and testing electronic equipment in close co-operation with the designer. An excellent background for a young man studying for H.N.C. electrical or equivalent with skill in wiring and laying out miniature assemblies. Must be fit for field trials and interested in problems of installation and operation in aircraft. Ref. 64J.
## METALLURGICAL LABORA-

 TORY ASSISTANT with some industrial experience. School Certificate standard in science subjects. Ref. 65J.RADIOGRAPHER for Metallurgical Laboratory with some experience in this field and suitable academic qualification. Ref. 66J.

TECHNICAL ENGINEER to be responsible for technical education of trials team and organisation of information service within operation group. Ref. 69J.

TECHNICAL ASSISTANT for development of missile and aircraft power supplies. H.N.C. and ability to conduct tests with minimum supervision. Ref. 73J.

SERVO ENGINEER able to offer degree and/or good practical experience of hydraulic, pneumatic and electrical servo systems-informal interviews in London available if preferred. Ref. 75J.

Applications, quoting reference of post in which interested and giving brief details of qualifications, experience and age should be addressed to:-

Personnel Manager, De Havilland Propellers Limited, HATFIELD, Herts.

## H O L L ERITH

ELECTRONIC COMPUTING

The continuing expansion of the British Tabulating Machine Co. Ltd., in the electronic computing field, in collaboration with the General Electric Co., Ltd., and the International Computer Corporation of America, offers excellent opportunities for junior or senior engineers interested in the design and construction of computing and calculating machines. We undertake to train at full pay, by practical work on development projects, applicants with suitable qualifications but with no previous experience in this field.
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Applicants must possess a University Degree, or equivalent, in Engineering or Physićs and be capable of accepting full Technical and Administrative responsibility for a Project involving a Team of Development Engineers.
Experience in one or more of the following is essential: Servo-mechanisms, Computing Techniques, Missile Guidance and Control Systems.
This is a Senior Appointment with great scope, and carrying an attractive Salary for the right man.
The Post is Pensionable and Permanent, and Assistance is available with Housing and Removal.
An Interview can be arranged either in London or Belfast.
Write, giving full details to:-
Staff Appointments Officer, SHORT BROTHERS \& HARLAND, LIMITED, P.O. Box 241, BELFAST, quoting S.A. 169.

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"DANGER-HIGE VOLTAGE," slide-off transters, $1 \frac{1}{4} \mathrm{in} . \times 3 \mathrm{in}$, red on white background, $4 / 8$ doz., $27 / 6$ gross.
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Applications are invited for the undermentioned posts which are vacant with the Telecommunications Laboratory staff of British Insulated Callender's Cables Ltd., at Kirkby, Near Liverpool.
GRADUATE ENGINEER for the design and development of Electrical Measuring Equipment for Telephone Cables. Candidates should be capable of liaising with Drawing Office and Model Shop for the production of prototype equipment. Previous experience in this type of work would be an advantage.
JUNIOR ENGINEER to assist in the design and development of electronic test gear necessary for the testing and manufacture of Telephone Cables. Minimum qualification is National Certificate standard and candidates should possess a knowledge of valve applications.
Applications quoting reference $\mathrm{P} / 43 / 56$ should be addressed to the Staff Officer, B.I.C.C. Ltd., Prescot, Lancs.

# THE <br> MULLARD RADIO VALVE CO. LTID. MITCHAM JUNCTION SURREY. <br> PROGRESS IN ELECTRONICS <br> <br> RECEIVING VALVES <br> <br> RECEIVING VALVES <br> TRANSMITTING VALVES <br> CATHODE RAY TUBES GAS-FILLED TUBES <br> <br> SEMI-CONDUCTOR DEVICES 

 <br> <br> SEMI-CONDUCTOR DEVICES}

In all the above fields far-reaching and rapid developments are taking place and absorbing work for suitably qualified persons is available in the Production and Development Departments in each of these fields.

## PRODUCTION

The appointments in the Production Departments require a high level of organising ability to maintain successful relations with all levels of Staff concerned. Production Engineers are responsible for the overall efficiency in the Section, including a complete range of duties that characterise a junior management post with the Company.

## DEVELOPMENT

Posts in the Development Departments involve the scientific investigation and basic engineering developments of both Company Products and the processes involved. Persons interested in this field should have some inclination for work which requires some inventive flair, tenacity of purpose, and a sensible practical approach to the problems that are likely to be met. Development Engineers are also expected to keep abreast with current scientific progress in their field of work.

For all vacancies the technical nature of the work demands a high level of academic training and it is felt that the posts would particularly appeal to Graduates in Physics, Mathematics, Electronic Engineering or, in a few cases, Chemistry or equivalent qualifications, and have had some period of industrial experience in ValveMaking or other industry and who now feel they would like to further their career in these fields.

A feature of employment within the Company for Graduates is the encouragement given to the younger Manager or Engineer to accept full responsibility for his particular project in the Development or Production field.

The Company salaries can be considered as progressive and arrangements for holidays, sickness and pension scheme have long been established.
Applications in writing, which will be treated in confidence, should be addressed to

The Personnel Officer, The Mullard Radio Valve Co. Ltd., New Road, Mitcham Junction, Surrey, quoting reference $\mathrm{JFG} /$ Tech Gen/W1.

## ENGLISH BLECTRIC COMPANY HIGH POWER ENGINEERING

THE WORLD IS DEMANDING larger and larger power plant for the rapidly expanding needs of industry and commerce.

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NEW AND ADVANCED LABORATORIES have been set up at our works at Stafford to carry out mechanical and electrical research and development on a large number of high voltage and high power engineering projects.

MODERN TEST PLANT is available, such as $2 \frac{1}{2}$ million K.V.A. short circuit apparatus and a 3.2 million volt generator for the development of equipment under the most severe conditions.

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WE ARE LOOKING FOR MEN AND WOMEN with good degrees in mechanical or electrical engineering, physics, and mathematics, who desire to build up a rewarding career in these laboratories.

THE WORK IS OF OUTSTANDING NATIONAL IMPORTANCE and covers a vast field of engineering research giving exceptional opportunities to engineers, physicists and mathematicians. Previous research experience is not essential-an enquiring mind and an interest in solving unusual problems are the chief requirements. WE ARE PREPARED TO PAY GENEROUS SALARIES to those with the enthusiasm and curiosity needed to concentrate on this kind of development. These positions are permanent and every assistance will be given in securing housing accommodation.

IF YOU WOULD LIKE TO HEAR more about these interesting prospects, we would be glad to hear from you. Reply to Dept. C.P.S., 336/7, Strand, W.C.2, quoting Ref. 1216.

## THE MULLARD RABIO VALVE CO. LTD.

This Company, increasing its contribution in the field of Electronics, is widening its activities in the field of VALVE DEVELOPMENT. There are a number of vacancies for younger men and women in the Valve Development and Quality Departments of the Company. The work is concerned with the experimental manufacture and/or the specialised testing of sample quantities of Valves and provides the opportunity for both interesting and varied work. The technical nature of this employment makes it necessary that applicants possess the G.C.E. at "A " level, particularly in Physics or equivalent qualifications. There are facilities for further study leading to higher academic qualifications and subsequently for posts of higher responsibility within the concern.

The commencing salaries will be according to qualifications and experience, and the Company's salary policy can be considered to be competitive. The Company operates a liberal holiday plan and a pension scheme, and a feature of employment is that an endeavour is made to grant technical responsibility to employees at as early a stage in their careers as possible.

Applications, in writing, should be made to The Personnel Officer, The Mullard Radio Valve Co. Ltd., New Road, Mitcham Junction, Surrey, quoting reference JFG/RVDD/W1.

## IMPERIAL CHEMICAL INDUSTRIES

 LIMITED, NOBEL DIVISIQN, has vacancies for Electronic Engineers, Physicists, or Electrical Engineers to work on the application of electronic techniques to a wide range of industrial problems.Positions in both senior and junior grades are available For the senior grade a good Honours Degree or an equivalent qualification is required, while for the junior grade, applicants should have at least Higher National Certificate, C \& G Final Certificate in Tele-communications or equivalent. Experience of industrial electronics is desirable but not essential.
Starting salaries would be in accordance with qualifications, age and experience.
The vacancies are in the Research Department which is situated in the Division's main factory at Ardeer, Stevenston, on the Firth of Clyde Working conditions are good and the factory works a five-day week. The Company operates a profit-sharing scheme and a contributory pension fund. Selected candidates will be interviewed and expenses will be refunded.

After joining the staff, married men who have to move their home to the Stevenston area will receive a reasonable refund of removal (including travel) expenses. To assist in house purchase, facilities are available in approved cases for loans; in addition legal fees may be advanced.
Application should be made in writing to the Staff Manager, 460 Sauchiehall Street, Glasgow, C.2.

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DESIGN, DEVELOPMENT AND LAYOUT OF RADAR SYSTEMS AND ALL TYPES OF SPECIALISED ELECTRONIC EQUIPMENT.
PERMANENT AND PROGRESSIVE EMPLOYMENT WITH A STEADILY EXPANDING ORGANISATION, IN A FRIENDLY AND INFORMAL ATMOSPHERE.

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[^16]CANADIAN AVIATION ELECTRONICS LTD., Montreal, Canada, require several Senior Development Engineers having some years experience in the development and maintenance of Fire Control Radar, Computers, Servo-Mechanisms and allied equipments.

Two year employment contract and excellent promotion prospects in rapidly growing organization occupying a leading position in its field in Canada.

The plant is modern, being situated on the outskirts of Montreal where accommodation is readily obtainable.

The company operates hospital and sickness insurance and pension schemes.
Salaries, fully commensurate with qualifications and experience range from "\$450 per month upwards.

QUALIFICATIONS DESIRED
B.Sc. (Maths., Physics or Eng.)
A.M.I.E.E.
A.M.Brit.I.R.E.
C.G.L.I. Full Technological Certificate in Telecommunications H.N.C.

Interviews will take place at various centres in England. Application should be sent to Canadian Aviation Electronics Representative, Care of Canadian Department of Labour, 61 Green Street, London, W.I.

CANADIAN AVIATION ELECTRONICS LTD., of Montreal, Canada, have openIngs for Technical Representatives and Technicians for the maintenance of electronic equipment in the field, at locations within the temperate regions of Canada. These appointments carry commencing salaries in the range $\$ 400$ to $\$ 475$ per month for Technlcal Representatives and $\$ 1.78$ to $\$ 2.10$ per hour for Technicians. Living and travelling expenses and Field Bonus will be paid where applicable.

Appointments will be made on a firm contract for a period of two years in the first instance, with an option of renewal at the end of that time. The Company operates a Penslon Plan, Group Insurance and Hospitalization plans. Assistance will be given with passages if required.

These posts carry excellent prospects of promotion and provide an opportunity for men with drive and initiative to succeed in the rapidly expanding economy of this country.

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 A.R.T.C., M.I.E.E., M.I.MECH.E Offers a first course of twenty Monday evening lectures in:-SERVO-MECHANISMS AND CONTROL SYSTEMS commencing on 8th October at 7 p.m.
This course is intended for engineers holding a science or engineering degree or H.N.C. who wish to study the principles of control systems. The course will include appropriate demonstration and laboratory work.
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Applications are invited from young men (preferably ex-National Servicemen) under 25 years of age who possess Inter B.Sc. or 3 "A" Level G.C.E.|Science Passes (preferably Physics, Pure and Applied Maths.) for admission to a newly instituted NDUSTRY-BASED SANDWICH SCHEME.
This is a four year course and involves alternating periods o 6 months' fulltime studying and interesting varied practica work at these Laboratories. It is designed primarily for Physicists and Electrical Engineers and will lead to the London B.Sc. (Special Physics) or B.Sc.(Eng.). Applications are also invited from potential chemists, metailurgists and mathematicians for similar courses, as well as from those interested in a career in patents work. Apart from a generous salary all College and examination fees will be gener
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- Applications should be made within 3 weeks of the appearance of this advertisement and should be addressed to :-
THE EDUCATION AND TRAINING OFFICER, RESEARCH LABORATORIES, G.E.G. LTD., EAST LANE, N. WEMBLEY, MIDDX.

QUOTING REFERENCE RLL/107.

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## ASSISTANT CHIEF ENGINEERS REQUIRED WITH EXPERIENCE IN ANY OF THE FOLLOWING FIELDS:

1. Design and development of transformers of types used in radio equipment and electrical appliances.
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The vacancies present excellent opportunities to engineers capable of taking full control of the project in hand and a very attractive salary is offered, together with good future prospects. The Company's extensive laboratory and production facilities are situated in the London area.
Please reply, in utmost confidence, giving details of qualifications and experience, to Box No. 3282.

TECHNICAL ASSISTANTS-There will be a number of openings for young TECHNICAL ASSISTANTS of both sexes in the new MULLARD SOUTHAMPTON WORKS. The minimum educational standard required is the General Certificate of Education in science subjects, and the posts will be concerned with research, development and applications work in the field of semi-conductors. There are also openings for more highly qualified personnel, interested in making careers in this expanding branch of electronics.

The commencing salaries paid will be based on age, qualifications and experience and the Company's general conditions of service will be found to be attractive. Interested condidates are invited to apply in writing to: The Personnel Officer;' THE MULLARD RADIO VALVE CO., LIMITED, New Road, Mitcham Junction, Surrey, quoting reference-GBK/S.13.

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An OPPORTUNITY is available for a first-class COMPUTER ENGINEER to join the group responsible for the design of the World's most advanced machine tool control system. A sound knowledge of computer logical design and a familiarity with high-speed digital techniques is required. The Laboratory is situated in delightful surroundings near the sea shore on the outskirts of Edinburgh, and provides ideal working conditions. The project may involve travel in the United States and other countries. The post offers excellent prospects and a full staff superannuation scheme is in operation.
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Applicants should possess a Degree, preferably with Honours, in Mathematics, Physics or Electrical Engineering, and should have had some experience of the theory of feedback systems and servo-mechanisms.
The work is concerned with guided missile and navigational systems, automatic pilots, analogue computors, flight simulators, servos and automatic control developments. The Appointment is Permanent and Pensionable.
The Organisation is expanding and its Laboratories and Equipment are all new. Salaries and prospects are good.
Assistance with Housing and with removal expenses.

Apply with full details to:
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## SENIOR ENGINEER

## Felgate Radio Ltd.

Degree or similar qualification necessary. Candidates who have been associated with production will receive preference.

Duties include the design of domestic radio and television, high fidelity audio equipment and miniature transmitter receivers.

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(Senior) required by an engineering company situated on the south coast for the development of small electro-mechanical radio components. Design experience of variable capacitors an advantage but not essential. Attractive salary to selected applicant. Please reply giving full details to Box No. 3792 c/o "W.W."

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Automatic Controls for Vertical Take Off and other aircraft, guided missiles, computing systems, etc.

QUALIFICATIONS: University Degree, capacity to control a development team.

EXPERIENCE: Several years of one of the following:
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This is a Permanent and Pensionable Appointment in an expanding organisation with new and well equipped Laboratories.
Salaries are commensurate with responsibility and prospects are good. Assistance with Housing and with Removal Expenses.

Apply with full details to:
Staff Appointments Officer,
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COMMERCIAL TELEVISION AND F.M. BROADCASTING HAVE RESULTED IN VACANCIES BECOMING AVAILABLE FOR MEN INTERESTED IN THE DEVELOPMENT OF V.H.F. AND U.H.F. TUNERS INVOLVING NEW TECHNIQUES OF DESIGN AND MANUFACTURE. SALARIES IN THE RANGE OF £650 TO £1,200 ARE OFFERED TO ENGINEERS WITH THE REQUIRED EXPERIENCE, AND PROSPECTS OF FUTURE ADVANCEMENTS ARE GOOD. WRITE, IN CONFIDENCE, GIVING FULL PARTICULARS OF EXPERIENCE AND QUALIFICATIONS TO BOX NO. 3880 c/o. "W.W."

APPLICATIONS ARE INVITED FOR OPENINGS which occur in the COMPUTOR AND MACHINE TOOL ELECTRONIC CONTROL GROUPS of a large engineering company situated in the eastern suburbs of London. Vacancies exist for the following staff:-

1. DESIGN ENGINEERS with experience of counting and pulse circuits, magnetic storage and amplifiers.
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3. MECHANICAL DESIGN ENGINEER (small mechanies).
B.Sc. an advantage but not essential. Excellent commencing salaries to the selected candidates. Superannuation scheme in operation. Please reply, giving full details of qualifications and experience to Box No. 3791.

CONSULTING GEOPHYSICISTS have vacancies in seismic oil exploration field parcies for young men, preferably single, with electronic or radar engineering experience, with Higher National or C. \& G. Certificates in Electrical Engineering or Telecommunications. £600 x $£ 72$.

We offer a permanent career to men of initiative prepared to accept responsibility and work in camp conditions in all parts of the world. Liberal home leave and allowances.
Box 3741, c/o "Wireless World."

## SOLARTRON

## ASSISTANT CHIEF INSPECTOR

Solartron Laboratory Instruments Ltd., a member of the rapidly expanding Solartron Group of Companies, is seeking a man with several years' test and supervisory experience to fill the newly created position of Assistant Chief Inspector.

Applicants should be conversant with A.I.D. procedure and be abreast of modern techniques of test and inspection.

The Group operates a generous non-contributory Pension and Life Assurance Scheme.
Apply in writing to the Group Personnel Officer, The Solartron Electronic Group Lid., Thames Ditton, Surrey.

## ELECTRONIC APPLICATION ENGINEER

Senior Development Engineer required for Circuit Application work on Electronic Components. Experience in one or more of the following fields is necessary.

1. DOMESTIC RECEIVER DESIGN
2. TELEVISION SET DESIGN
3. COMMUNICATIONS EQUIPMENT
4. TRANSISTOR CIRCUITS.

The situation is in the outskirts of London and a salary in the range $£ 700$ to $£ 1,200$ is offered, dependent upon age, qualifications and experience. Please write, giving full details, to Box No. 3881 c/o."W.W."

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Quoting S.A.175.

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[^5]:    'See "Radio and Electronic Components, Vol. 1-Fixed Resistors," published by Sir Isaac Pitman \& Sons.

[^6]:    * Livingston Laboratories.

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