

The illustration depicts a set of modern "AVO " testgear being used to measure the " Q " of the secondary winding of the second I.F. transformer on a chassis of unknown characteristics-just one of many tests which can be performed by this combination of instruments.
A signal of predetermined frequency from the "AVO" Wide Range Signal Generator is being fed into the Electronic Test Unit, where it is amplified and fed to the secondary winding of the transformer. The Electronic Testmeter is connected across the tuned circuit under test and from the readings obtained and the controls of the Electronic Test Unit, the " Q " of the circuit can be determined.
The three instruments, shown as a team, cover a very wide field in measurement and form between them a complete set of laboratory testgear, ruggedly constructed to withstand hard usage.

ELEGTRONIC TESTMETER A 56 -range instrument combining the A. 5 sitivity of a delicate gase of handsenstithe robustress and andi-range metable with of an ordinary of a highy from ling of an basically oltmeter, presenting Consists alve Millivo and prider test. D.C. Vatve variations circuit under nezigible to measure:- 5 mV to $10,000 \mathrm{~V}$.
Sivicched Volts: 5 mV to to to Amp.
D.C. Current: iv to $2,50 \mathrm{Mcls}$. M.S.
iv to ${ }_{200}^{250} \mathrm{Mc} / \mathrm{s}$.

Output:
A.C. Power ${ }_{5}^{50 \mathrm{Watts}}+20 \mathrm{db}$.

Decibels: - Zero level 50 to $50 \mu \mathrm{~F}$.
Capacitance: .0001 to to 10 Megogms
Resistance: $100-130 \mathrm{v}$. and $200-260 \mathrm{v}$



## And its SPECIFICATION

The Tape Player itself is of unit construction which enables many special requirements and applications to be met without undue modification. The following features can be provided to special order:-

Tape speeds 71" and 15"per second, or $3 \frac{8}{4}^{\circ}$ and $1 \frac{7}{4}^{\prime \prime}$ per second.
-
Synchronous drive motor. -
Remote operation or foot control.
-
Automatic back spacing and reverse drive for dictation purposes.
-
Cassette tape loading. -
Rack mounted assembly.

TAPE SPEEDS TRACKS PLAYING TIME PER TRACK SPOOLS

SENSE OF SPOOLING
REWIND TIME
HEADS

## TAPE

OPERATION

74* and $33^{* *}$ per second.

- 1 " wide. Number of tracks 2. 30 minutes at $71^{\circ}$ per second. 60 minutes at $3 \mathbf{y}^{\prime \prime}$ per second.
Standard $7^{\prime \prime}$ and $5^{\prime \prime}$ plastic or metal.
From left to right with tape coating inwards.
One minute for $1,200 \mathrm{ft}$. of tape (approx.).
R.F. erase head. Record/ playback head off-set for recording on upper track. Provision on player unit for additional monitoring head for special applications.
Single control provides:Record, Playback, Fast Forward, Cueing, Rewind.
To ensure additional safety against accidental erase, an additional record/playback adwition is provided on the amplifier assembly. Power and amplifer assembly. Power and brake operation is by means of
a relay which will enable rea relay which will enable re-
mote operation to be provided mote operation to be provided
in special applications.

FREQUENCY RESPONSE

DISTORTION
At 71 per second 60-10,000 C.P.S. plus or minus 3 db . At $33^{\circ}$ per second $70.7,000$ C.P.S. plus or minus 3 db .

Less than 24 \% total harmonic distortion at normal operating
level.

SIGNAL/NOISE Approximately 50 db . using RATIO

INPUTS
(1) $U_{p}$ to 50 ohms low level -110 db . microphone input. (2) High Z up to 100 K at 1 v . unbalanced (radio input).
OUTPUTS
(1) $2 \frac{1}{2}$ ohms at 3 watts to internal loudspeaker.
(2) 15 ohms at 3 watts for external speaker.

WOW AND FLUTTER

Total wow and flutter content less than $\cdot 2 \%$.
MAINS SUPPLY $200 / 250$ v. 50 cycles 230 V.A. Other voltages and frequencies supplied to special order.
DIMENSIONS $16^{\circ}$ wide $\times 11^{\circ}$ high $\times 18^{\circ}$ deep epprox.
WEIGHT 45 lbs .

New Standard of

## Efficiency in a V.H.F

Power Tetrode

The Mullard power-tetrode, QY3-125, is the most efficient valve of its class in the world. Its outstanding characteristics include low driving power, low power consumption, and high power gains at V.H.F. With a maximum anode dissipation of 125 Watts, this valve will give an output of 375 Watts up to 120 Megacycles, and 225 Watts up to 200 Megacycles. The QY3-125 is particularly recommended for use as an oscillator or R.F. amplifier in compact communications and allied electronic equipments, operating in the V.H.F. range.

## SPECIAL FEATURES

* Low-inductance Anode Lead - Higher Frequency Performance
* Double Helical Filament - Low Inductance and Low Hum
* Bottom Screen Shield—Low Anode/Filament Capacitance
* Zirconium Coated Carbon Anode-Continuous Gettering
* Powdcred Glass Base - Low R.F. Losses
* American Giant 5-Pin Base

Full technical details of this and other types in the Mullard range of all-glass V.H.F. valves will be supplied on request to the address below.


MULLARD LTD., COMMUNICATIONS \& INDUSTRIAL VALVE DEPT., CENTURY HOUSE, SHAFTESBURY AVE., LONDON, W.C.2.

## You must nof miss this great display



## BRITISH



## EXHIBITION AND CONVENTION

OLYMPIA•LONDON
JUNE 8-18

## MATERIALS

## OF HIGHEST QUALITY

## MACHINES

OF LATEST DESIGN

## FABRICATORS WITH <br> NEWEST TECHNIQUES

All important British producers, moulders and fabricators of plastics materials, and those who supply raw materials or equipment, will be "At Home" to the world for the 2nd British Plastics Exhibition and Convention. Exhibits will include :

Materials-synthetic resins, moulding materials, laminates, sheeting and other essential ingredients.

Products-technical and industrial components, consumer goods, etc.

Plant \& Equipment-presses, extrusion machines, moulds and accessories, etc. The Convention, held simultaneously, will promote a broader understanding of the Industry and provide a progress report on recent developments. Writc NOW for full details.

FIRM $\qquad$

MAIL THIS TODAY TO BRITISH PLASTICS Dorset House, Stamford Street, London, S.E.1.

Please send me the 1953 Exhibition Brochure, Convention details, free season ticket, etc.

. . . with safety in the hazardous enterprise of the deep sea trawler is its radio and radar equipment upon which sale navigation depends. Thousands of soldered joints contribute to the efficient functioning of this delicate apparatus. One dry or H.R. joint could mean the breakdown of a circuit, the destruction of the vital link, a perilous voyage.

FAULTLESS FLUXING PRESERVES THE VITAL LINK
Dry or H.R. joints are impossible with Superspeed for the flux is always released in exactly the correct proportion. This faultless fluxing action is ( achieved by the unique STELLATE core which gives six points of rapid solder collapse. At soldering temperature the activated rosin flux is released immediately for effective spreading and wetting. Superspeed is being used more and more in the production of radio and radar equipment where faultless joints are essential.

Cored and solid solde tings and solid solder washers supplied to customers' specifications.

# SUB-MINIATURE ELECTRONIC COMPONENTS 

New-Better-and Smaller too!

## Earpieces Type MME/G

are of electro-magnetic design and their air gap is controlled to within .00025 inch. Each unit is checked throughout the frequency range after a prolonged test at overload conditions. Four alternative frequency characteristics are available, at impedances between 30 ohms and 1000 ohms. These earpieces can be used with polarising currents.

## Plugs Type PL6

are tiny standard round pin plugs moulded to the flexible cord. They fit the earpiece or socket with a positive detent action. They are designed for low contact resistance and minimum wear.
A range of plugs is available and special plugs can be manufactured to order.

## Transformers Type T4

are miniature output or coupling transformers for valve or transistor circuits. The winding terminations consist of solder tags moulded into the robust thermosetting bobbins. Before lamination each unit is checked to ensure no short-circuited turns. During final inspection the transformer is checked for efficiency throughout the frequency band. With a 70 kilohms source and a direct current of .35 mA the variation in response is 3 db between 200 and 4000 cps. Turns ratios and impedance can be varied to customer's specification.

Write or telephone for prices, further details, and samples

# FORTIPHONE LIMITED <br> FORTIPHONE HOUSE, 247 REGENT STREET, LONDON, WI 

## ALL-POWER

## REGULATED POWER SUPPLIES

## SERIES 500

An entirely new range of units, designed for the highest possible performance and overall efficiency.

Now entering into large scale production and available for prompt delivery.

(FITTED WITH END FRAMES)

## ABRIDGED DATA

(FURTHER INFORMATION ON REQUEST)

| DATA |  | MODEL 501 | MODEL 502 | MODEL 503 | MODEL 504 | MODEL 505 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | OUTPUT | $\begin{gathered} 200-500 \mathrm{v} . \\ 250 \mathrm{~mA} \end{gathered}$ | $\begin{gathered} 200-500 \mathrm{v} . \\ 250 \mathrm{~mA} \end{gathered}$ | $\begin{aligned} & 0-500 \mathrm{v.} \\ & 250 \mathrm{~mA} \end{aligned}$ | $\begin{aligned} & 0-500 \mathrm{v} . \\ & 250 \mathrm{~mA} \end{aligned}$ | $\begin{aligned} & 0-500 \mathrm{r} . \\ & 150 \mathrm{~mA} \end{aligned}$ |
|  | NUMBER OF RANGES | 2 | 2 | 3 | 3 | 1 |
|  | VOLTAGE <br> STABILIZATION | $\pm 0.02 \%$ | $\pm 0.002 \%$ | $\pm 0 \cdot 1 \%$ | $\pm 0.002 \%$ | $\pm 0.1 \%$ |
|  | EFFECTIVE OUTPUT RESISTANCE (MAX.) | $0.2 \Omega$ | $0.002 \Omega$ | 0.58 | $0.02 \Omega$ | $0 \cdot 5 \Omega$ |
|  | OUTPUT RIPPLE (R.M.S. MAX.) | 2 mV | 1 mV | 3 mV | 1 mV | 3 mV |
|  | OUTPUTS | - | - | $\begin{gathered} 250 \mathrm{v} .5 \mathrm{~mA} \\ 0.250 \mathrm{v} .1 \mathrm{~mA} \end{gathered}$ | $\begin{aligned} & 250 \mathrm{~V} .5 \mathrm{~mA} \\ & 0.250 \mathrm{~V} .1 \mathrm{~mA} \end{aligned}$ | $\begin{gathered} 250 \mathrm{~V} .5 \mathrm{~mA} \\ 0.250 \mathrm{v} .1 \mathrm{~mA} \end{gathered}$ |
|  | VOLTAGE <br> STABILIZATION | - | - | $\pm 0.05 \%$ | $\pm 0.002 \%$ | $\pm 0.05 \%$ |
|  | OUTPUT RESISTANCE (MAX.) | - | - | $1 \Omega$ | $0.01 \Omega$ | 18 |
|  | OUTPUT RIPPLE <br> (R.M.S. MAX.) | - | - | 2 mV | 1 mV | 2 mV |
| UNSTABILIZED + VE. <br> H.T. SUPPLY 250 mA MAX. |  | $\begin{aligned} & 470 \mathrm{v} \\ & 630 \mathrm{v} \end{aligned}$ | $\begin{aligned} & 470 v \\ & 630 v \end{aligned}$ | $\begin{aligned} & 320 \mathrm{y} \\ & 470 \mathrm{y} \\ & 630 \mathrm{y} \end{aligned}$ | $\begin{aligned} & 320 \mathrm{v} \\ & 470 \mathrm{v} \\ & 630 \mathrm{v} \end{aligned}$ | 630 v |
| UNSTABILIZED A.C. SUPPLY |  | 6.3 v. 10 A | 6.3 v .10 A | 6.3 v. 10 A | 6.3 V .10 A | 6.3 v .10 A |
|  | PRICE | 670 | 691 | ¢81 | 6100 | 675 |

## STANDARD UNITS

All models are supplied as standard for mounting in 19 in . racks and are fitted with fully protective covers.
EXTRAS To convert from rack mounting to bench use the following extras are available :-
Polished hardwood reinforced end frames ... ... ... ... ... ... ... ... \&I 15 per pair.
Steel instrument case of new design ... ... ... ... ... ... ... ... ... 4410 each.
PRICES. PRICES ARE QUOTED NET EX WORKS AND ARE SUBJECT TO VARIATION WITHOUT NOTICE.
ALL - POWER TRANSFORMERS LTD.
CHERTSEY ROAD, BYFLEET, SURREY


Chosen for its brilliance, resolution, long life and high safety factor, the 'English Electric' British Made Mctal C.R. Tube T901 is the one around which most home constructors' circuits - including the Magna-View - are designed. This comprehensive booklet gives you full instructions for building the famous Magna-View 16-inch receiver, together with details of the necessary components.

## 'ENGLISH ELECTRIC'

LONG LIFE I6-inch METAL C.R. TUBE

British Made

FILL IN AND POST THIS COUPON TO THE ADDRESS BELOW WITHOUT DELAY

Please send me POST FREE full instructions for building the big screen Magna-View Television receiver.
NAME.
ADDRESS


## Microwave Test Gear

Metropolitan-Vickers Electrical Company announce a complete range of precision microwave test gear for use in $3 \mathrm{in} . \times \frac{1}{2} \mathrm{in}$. waveguide over a band of wavelengths from 10 cm . to 11 cm .


1 PRECISION ATTENUATOR Type 501
2 MATCHED LOAD Type 506
3 OSCILLATOR Type 508
4 FIXED ATTENUATOR Type 519
5 SHORT CIRCUIT Type 510
6 DIRECTIONAL COUPLER Type 504


6


Other Metrovick microwave equipment includes variable attenuator type 502, standing wave detector type 512, wave meter type 517, high power load type $515, \mathrm{~S} \& \mathrm{X}$ band spectrometer type 518.

Full technical details will be sent on request.

# GOODMANS P.M. VIBRATION GENERATORS 

 vibration pick-ups and ancillary equipmentThese vibration generators provide vibratory sinusoidal forces of frequency and amplitude by which specific vibratory conditions con be simulated. They provide the means of assessing the effects of sudden acceleration on materials, structures and components.

MODEL V47 tor the vibration tronic components, optical-cell research. hairspring torque testing etc.
 Thrust. . ....... Force factor 0.9 lbs . per amp
Mox. Continuous
Current Roting 1.5 amps; 2.0 amps up to (R.M.S.) Stroke....... 2 min. duration. impedance ..... 0.2 in toral excursion. - load berween 3 and 10 ohms.

Frequency Range
Weight of Moring
Weight of Moving
System
Stray Fields ...... 6.5 grams. Operating zone less than
Flux Density. 25 gauss.
Flux Density. . . .
Weight . . . . . 2 lbs.


MODEL 390A A medium duaty
ing an alternating force of approximately

## $\pm 25$ tbs.

Thrust. - . . . . . Force factor 4.7 lbs . per amp Mox. Continuous 2 amps uncooled; 4 amps Current Rating with air cooling of approx.
(R.M.S.)
Stbs. per sq. in.
(R.M.S.)

Stroke....... S lbs. per sa. in.
Impedance . . . . . . 0.5 in total excursion.
Frequency Range. . Uohms matching.
Weight of Moving
System......... 0.16 lbs
Stray Fields ..... Operating zone less than
Flux Density . . . . 11.000 gauss
Werght ........ 26 lbs.


MODEL 790 For Vibratinz ponents, and is capable of producing ponents, and is cap.
a force of $\pm 50 \mathrm{lbs}$.
This unit has a force factor of approxi mately 9.5 amps and total current capacity. with air cooling. of 4 amps (R.M.S.)

Stroke........ 0.5 in total excursion. impedonce..... is ohms matching Frequency Range. Up to $5.000 \mathrm{c} / \mathrm{s}$. Frequency Range
Weight of Moving Systemt of Moving 8 ozs. (approx.) System ${ }^{\text {Stay }}$...... 8 ozs. (approx.)
Stray Fields.... Operating zone less Operating zone less
than 100 gauss. than 100 gaus.
11.000 gauss. Flux Density. . . . 70 lbs. (inc. trunnion)

MODEL 8/600
For the vibration of heavy loads or complete assemblies. Has a total force of approximately $\pm 250 \mathrm{lbs}$. Stroke......... I in. total impedance. . . . . excursion. impedance. . . . . to suit driving equipFrequency Range Up so 3.000 Weight of Moving c/s. System ...... 6 lb. (approx.) tray Fields .... Operating
zoneless
chan 25
gauss.
Flux Density . . . 10.000 gaus
Flux Density . . . 4 cwt . rotal Wesghe This unit can be fitted with (a) buite in air cooling blower (b) switch to give high or low impedance armature coil and (c) pick-up unit for monitoring wave form and amplitude.

## Driving Equipment for Model 390A

(A) POWER AMPLIFIER TYPE D/I20 (120 watts continuous rating) covering a frequency range of $10 \mathrm{c} / \mathrm{s}$ to $10,000 \mathrm{c} / \mathrm{s}$ and giving full output over the entire range. A specially designed tuning unit increases the current and hence the thrust of the generator with which it is used. This arrangement, in addition to providing a greater current output, also ensures a purer wave-form.
(B) TRANSFORMER UNIT, TYPE 120 This is provided as a separate unit purely for reasons of physical accommodation.
(C) STABILISED POWER SUPPLY UNIT, TYPE DS/120 By means of this unit the. H.T. supplies for the amplifier are stabilised so that the full power of the amplifier will be maintained even if the mains voltage drops. Subsidiary unstabilised heater and biasing voltages are also derived from this unlt.
(D) OSCILLATOR, TYPE RC/D.I. This fully covers the range of the amplifier, i.e., $10 \mathrm{c} / \mathrm{s}$ to $10,000 \mathrm{c} / \mathrm{s}$. Apart from being a highly efficient oscillator, specifically designed for operation with the Type D. 120 amplifier, this is in effect a control unit, and includes (a) a valve. volt meter calibrated directly in amperes to indicate the current in the moving coil of the vibrator; (b)
a gain control for varying the current to the generator, and so to control the amplitude of its vibration: (c) facilities for incorporating a phase meter, to provide the measurement of phase angles from which motional impedence can be plotted.


Driving equipment alsa available for Models 790 and $8 / 600$.
Full technical information and Data Sheets are available on request to "Special Products" Division D

## GOODMANS INDUSTRIES LTD. <br> AXIOM WORKS • WEMBLEY • MIDDX. Telephone: Wembley 1200 (8 lines)

## PAINTON

mamme
By Sppointment to the Pipessional Engineer...

Altemuatoss
Faders
Stud Switches and Toggle Suitches
Wirewound Polentiometers
High Stability Caibon Resistors
Wircwound Resistors
Plugs and Sockels
Terminals
Tinots Dials and Pinters

PAINTON
Sorthampton Oingland


Complete assembly sealed in resin block-suitable for incorporation in every type of electronic equipment-designed to meet, at low cost, the stringent requirements of Specification RCS.214-saving in weight and volume-suitable for power, signal, pulse and high voltage applications-designed to individual specification.

## Complete Transformer Service to the Electrical Equipment Manufacturer


' $K$ ' Series-impregnated open type ' C' core construction, conforming with RCL. 216.

' H' Senes-hermetically sealed oil filled components ' C ' core construction, conforming with RCL. 215.

' $M$ ' Series-hermetically sealed miniature type for power and signal applications



## ONLY the LAB unit has all these features...


$\star$ Continuous storage $\star 700$ resistors in a space $12^{\prime \prime} \times 4^{\prime \prime} \times 4^{\prime \prime}$
$\star$ Ohmic values separately carded $\star$ Finger-tip selection

Designed to provide a complete range of resistors for research and experimental laboratories and small production units. As easy to use as a card index. Rapid selection from 700 sorted and carded resistors. Continuous storage - empty cards merely replaced with full cnes available from stock. The Lab Continuous Storage Unit is supplied FREE with initial purchase of 180 Type R Resistors (Order LSUC $\frac{1}{2}$ ) or 240 Type T $\quad$ (Order LSUC $\frac{1}{4}$ )


The Lab Continuous Storaze Units are available from your normol source of supply but more detailed information con be obtained from

THE RADIO RESISTOR COMPANY LTD. 50 ABBEY GARDENS, LONDON, N.W. 8 - Telephone: Maida Vale 5522

'Frequentite' is the most suitable insulating material for all high frequency applications. Seventeen years ago we introduced the first British-made low-loss ceramic, and consultation with us before finalising the design of new components is a wise precaution.

## STEATITE \& PORCELAIN PRODUCTS LTD. B <br> S.P. 66



## Introducing the



## LOFTEX TELEVISION AERIAL

- ADJUSTABLE FOR HEIGHT OR "STAND OFF"
- ROTATABLE FOR DIRECTION
- TAKES MINIMUM LOFT SPACE

Here is the new " Loftex" loft aerial that combines simplicity with sound design-and as a result provides an efficiency that is truly phenomenal. With facilities for easy adjustment it utilises the minimum of loft space yet can be mounted to provide maximum signal strength since it incorporates the well-tried " Antex" principle. Sheer simplicity and sound design have permitted the production of this aerial at a price that will commend itself to all.

Can be fitted into the following spaces

| Transmission | Channel* | Spare Required |  |
| :--- | :---: | :---: | :---: |
| LONDON | 1 | Height $8^{\prime} 2^{\prime \prime} \times 7^{\prime} 11^{\prime \prime}$ wide |  |
| GLENCAIRN | I/HOR | Length $8^{\prime} 2^{\prime \prime} \times 7^{\prime} 11^{\prime \prime}$ wide |  |
| HOLME MOSS | 2 | Height $7^{\prime} 5^{\prime \prime} \times 7^{\prime} 0^{\prime \prime}$ wide |  |
| KIRK'O SHOTTS | 3 | Height $6^{\prime} 9^{\prime \prime} \times 6^{\prime} 6^{\prime \prime}$ wide |  |
| BRIGHTON | 3 | Height $6^{\prime} 9^{\prime \prime} \times 6^{\prime} 6^{\prime \prime}$ wide |  |
| MIDLANDS | 4 | Height $6^{\prime} 2^{\prime \prime} \times 5^{\prime} 11^{\prime \prime}$ wide |  |
| WENVOE | 5 | Height $5^{\prime} 8^{\prime \prime} \times 5^{\prime} 6^{\prime \prime}$ wide |  |
| PONTOP PIKE | $5 / H O R$ | Length $5^{\prime} 8^{\prime \prime} \times 5^{\prime} 6^{\prime \prime}$ wide |  |



* State channel number required as shown above.

AHTIFERENCE UMIED BICESTER ROAD AYLESBURY, BUCKS.


## AIR-SPACED ARTICULATED


offer a unique combination of
$\checkmark$ FRACTIONAL CAPACITANCE HIGH IMPEDANCE

MINIMUM ATTENUATION
ALONG WITH
EXCEPTIONAL FLEXIBILITY
LIGHT WEIGHT

## 38 STOCK TYPES

## FOR ANY OF YOUR STANDARD

 OR SPECIAL APPLICATIONSA few of the very low capacitance types are:

| Type No. | Capacit. $\mu \mu$ F/ft. | Impedance ohms | O.D. |
| :--- | :---: | :---: | :---: |
| C. 44 | 4.1 | 252 | $1.03^{\prime \prime}$ |
| C. 4 | 4.6 | 229 | $1.03^{\prime \prime}$ |
| C.33 | 4.8 | 220 | $0.64^{\prime \prime}$ |
| C.3 | 5.4 | 197 | $0.64^{\prime \prime}$ |
| C. 22 | 5.5 | 184 | $0.44^{\prime \prime}$ |
| C.2 | 6.3 | 171 | $0.44^{\prime \prime}$ |
| C.1I | 6.3 | 173 | $0.36^{\prime \prime}$ |
| C.I | 7.3 | 150 | $0.36^{\prime \prime}$ |

## nin <br> TRANSRADIO <br> CONTRACTORS TO h.M. GOVERNMENT

[^0]
## PROGRESS

## in Counting ~

 through the agesMISTORICAL I_LUSTRATIONS EY COURTESY OF THE SGIENCE MUSEUM LONDON.

LEIBNITZ'S STEPPED RECKONER

Invented in 1694 by the German phlosopher and practicable calculating machines

NAPIER'S BONES

Devised in 1617 by John Napier. Scottish nobleman.
es a me thanical means of multiplication by addition.

ROMAN ABACUS

Known to have been in use at the time of Juliws Caesar. A surviving =pe:imen is preserved $n^{-3}$ Eibliothèque Nationike, Paris.

## Cold Cathode

 TubesFOB Modern

## COUNTING. CALCULATING SELECTING. SWITCHING



## THE SCIENTIFIC

 VALVEHIVAC
COLD CATHODE
TUBES
The most modern devices for use in electronis systems.
douglas and macadi

## coll wind <br> Automatic <br> DE TR

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& \text { ton, aud or na in rendering } \\
& \text { be pleased to assist in }
\end{aligned}
$$

overseas AGENTS We have Agents in all parts of the world. Names and addresses will be furnished on application.

## MARCONI communication systems



## serve mankind

Wherever men live logether fin industrial ateas or in tropical archipelagoes) prosperity is largely a matter of communications.
bland communitios seem blessedly seltcombined. Fut, for ellective govermment,
hygene, adety, hunbandry. commerce welfare, it short, and prosperity-archipelagoes are increasingly interdependent. Marconi's invention is hringing 10 island peoples a fuller and richer way of life.


PLANNED


INSTALLED

## G.E.C. RADIO COMMUNICATION EQUIPMENT FOR USE IN ALL PARTS OF THE WORLD




The Advance type H1 Audio
Signal Generator completely covers the unusually wide range of $15 \mathrm{c} / \mathrm{s}$ to $50,000 \mathrm{c} / \mathrm{s}$.

It is characterised by its extremely low distortion and level output over the entire range; provides both sine and square wave output. A robust, reliable and accurate instrument for the discriminating service engineer - Accuracy $\pm 1 \%, \pm 1 \mathrm{c} / \mathrm{s}$. Distortion less than $1 \%$ at $1,000 \mathrm{c} / \mathrm{s}$. Output from 200 microvolts to 20 volts with accuracy of
2 db . Weight 14 lb . Size $131^{\circ} \times 101^{\prime \prime} \times 8^{\prime \prime}$ $\pm 2 d b$. 200 microvolts to 20 volts with accuracy of
2 db . Weight 14 lb . Size $131^{\circ} \times 101^{\prime \prime} \times 8^{\prime \prime}$200 microvolts to 20 volts with accuracy of
2 db . Weight 14 lb . Size $131^{\circ} \times 101^{\prime \prime} \times 8^{\prime \prime}$

Full technical details available in Folder S.I6/W.

## ADVANCE

 TYPE HI
for 1450 Valves


Recognised as the Most Reliable Valveholders

## B7G Valveholders

are now available moulded in:Phenol Formaldehyde (Black).
Nylon loaded Phenol Formaldehyde
(Notural Brown).
P.T.F.E.

## On STAND 13

# MUREX LTD 



## 'SINCOMAX' <br> SINTERED COMPOSITE MAGNETS

 SINTERED PERMANENT MAGNETS • Also TUNGSTEN MOLYBDENUM, TANTALUM \& ZIRCONIUM METALS . . . and fabricated instrument \& valve components

MUREX LIMITED (Powder Metollurgy Division) RAINHAM • ESSEX - Telephone: Rainhom, Essex 3322 LONDON SALES OFFICE: CENTRAL HOUSE, UPPER WOBURN PLACE, W.C.I. Telephone: EUSton 8265

## HEITISM anal pround of it!



## EXACT EFFORTLESS TUNING



The S.L. 8 Spin wheel drive gives easy control through a ratio 24-1. Fitted with constant velocity coupling, it eliminates strain on the Condenser, providing mechanical and electrical isolation from vibration and noise.

Complete with 3 -band glass scale 9 in. $\times 4 \frac{1}{4} \mathrm{in}$. Printed short, medium and long wavebands with station names. Scale length 7in. Supplied with florentine bronze escutcheon.

PRICE 27/6.
S.L.5, similar but fitted with reverse vernier drive, gives ratios of 18-1 search and 50-1 reverse vernier.

PRICE 26/6.
Replacement Scales calibrated to Copenhagen
Plan now available for:
Airplane drive $\quad 2 / 3$ retail Full Vision Drive $\quad 2 / 9$ retail Squareplane Drive $2 / 6$ retail S.L. 8 or S.L. 5 Drive $/ / 6$ retail

## PRECISION COMPONENTS BY



## JACKSON

BROS. (LONDON) LIMITED KINGSWAY•WADDON• SURREY Telephone: Groydon 2754
Telegrams; Wclfilco. Souphone, London

## . . picture perfection with the Aerialite DU BL EX (Trade Mark)


these features, PLUS excellent mechanical design, ensure that you get the best performance from your receiver. Moreover, the Dublex is moderately priced at $£ 4.8 .6$ ( 7 ft . mast complete with brackets, etc.); $\mathbf{4 7 . 1 5 . 0}$ ( $10 \mathrm{ft} . \times 2 \mathrm{in}$. mast complete with double lashings, etc.).

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| 100 | 350400 | 550 | $4 \frac{1}{2}^{\text {m }}$ | $18^{\prime \prime}$ | L32 | K136 | 13/6 |
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6 or 12 V D.C., positive or negative earth

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| eceiver only. | 3.3 mmps .) | At 6 volts these figures |
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* Carefully designed to reduce moving parts to the very minimum. giving long trouble-free life.
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| $\left.\begin{array}{l} \text { fits on } \\ \text { GD.071 } \\ \text { LD. } 071 \end{array}\right\}$ | Bulkhead (Junction) adaptor | vD. 071 |
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FU24. 0-12-24 v. at I amp.
F5. 6.3 v at 10 amps . or 5 v . at 10 amps . or 12.6 v at 5 amps . or 10 v . at 5 amps.
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## VRTVES, "LUBES \& CIRCUTHE

## 4. "FLYWHEEL" SYNCHRONISATION OF TELEVISION LINE TIMEBASES


#### Abstract

The problem of ensuring good line synchronism in the presence of large noise pulses has received more attention in the design of receivers for the negative modulation systems used in the U.S.A. and on the Continent of Europe than in the design of receivers for the British system which employs positive modulation. Nevertheless, a low signal-to-noise ratio in fringe areas in Great Britain also causes poor synchronism when direct synchronising of the line timebase is adopted. This is due, of course, to the timebase being triggered by noise pulses which occur just before the synchronising pulse. One method of preventing this is to control the frequency of the timebase oscillator by a direct potential which varies with the tendency of the oscillator frequency to drift. The variation of the control potential is obtained by comparing the relative phase of the synchronising pulses with that of the flyback pulse from the timebase oscillator, and this is done by a so-called coincidence detector. A typical circuit is shown in Fig. 1, and its operation is as follows: The synchronising pulse input, of approximately 100 volts peak-to-peak, is differentiated and applied to the grid of the triode $V_{1}$. The grid is so biased that only the negative portions of the differentiated synchronising pulses are amplified by the triode. The pulses appearing at the triode anode then have a leading edge corresponding to the leading edge of the synchronising pulses, but are of shorter duration than the synchronising pulses. The pentode $\mathrm{V}_{2}$ is employed as the coincidence detector. The pulses appearing at the anode of $V_{1}$ are applied to the screen grid of $\mathrm{V}_{2}$, and a series of pulses which correspond to the flyback period of the line timebase oscillator are applied to the control grid. The mean anode current of $\mathrm{V}_{2}$ will thus vary with the phase difference between these two sets of pulses, that is to say with the degree of coincidence between them, as illustrated in Fig. 2. The voltage at the anode of $\mathrm{V}_{2}$ will vary accordingly.


$V_{3}$ and $V_{4}$ operate as a multivibrator in the conventional manner and constitute the line timebase oscillator, the grid resistor of $\mathrm{V}_{3}$ being returned to the anode of the coincidence detector $V_{2}$. A positive-going pulse corresponding to the line flyback pulse appears at the anode of $\mathrm{V}_{4}$, and it is this pulse which is applied to the grid of the coincidence detector as previously mentioned. The drive to the line ontput valve is taken from the anode of V3, being suitably shaped by $R_{1}$ and $C_{1}$.
The tuned circuit $\mathrm{L}_{1}, \mathrm{C}_{2}$ in the anode circuit of $\mathrm{V}_{4}$ is tuned to the line timebase frequency, and is inserted to improve the long-term stability of the oscillator.
The cycle of operation of this system is:
(1) Line oscillator frequency falls
(2) Line scanning time increases
(3) Line flyback pulse is delayed
(4) Degree of coincidence between flyback pulse and synchronising pulse at the coincidence detector is reduced
(5) Anode current of coincidence detector falls
(6) Anode voltage of coincidence detector rises
(7) Line scanning time falls
(8) Line oscillator frequency rises


Fig. I-Typical " Flywheel " Synchronising Circuit.


Fig. 2-Positive-going synchronising pulse and line flyback pulse applied simultaneously to the coincidence detector.

For an initial rise in line oscillator frequency, the converse of the above stages occurs.

## Looking Ahead in Television

THE most fervent advocate of our present television system would hardly claim that it should be retained for ever. Sooner or later, there will be a demand for something better-higher definition, colour, three-dimensional or perhaps a combination of these qualities. Depending on the plans we make now, the transition to a new system will be either easy or difficult.

In the midst of the present uncertainties, when the television picture is clouded by party-political acerbity, those responsible for development may be forgiven if their thoughts tend to be deflected from the long-term view to matters of the moment.

Of course, it may be argued that the only practicable way of making a change is to ensure that the new and improved system shall be "compatible" with the old. That is to say, receivers already installed shall be usable with the improved system, though they may derive no benefit from the improvement. If we allow all the available television channels to be filled by transmissions operating on the present standards, there will be no choice; compatibility will be a necessity, and posterity must look after itself.

The other-perhaps less realistic-way of making a change is to introduce new and improved (but incompatible) standards as a parallel de luxe service for those who are willing to buy new receivers. In the course of time, when most of the old receivers are worn out, this service would become the national standard, and the original stations would either be converted to it, or, perhaps to system No. 3, something still more refined and highly developed than system No. 2. All this implies, of course, the availability of at least one spare set of channels providing virtually national coverage at the time when the first change of standards is made.

So far as can be foreseen at present, compatibility must imply compromise and the need for tolerating something short of the best. From the technical point of view, everything is against compatibility as a principle of development, but everything is in its favour as a matter of economics.

These thoughts are provoked by the recent appointment of a technical sub-committee to the

Television Advisory Committee. The task of that body is not only to advise the Government on the highly controversial issue of competitive television services, but also on the general development of the art. So far as the long-term view is concerned, the technical advisers seem to us to have one fundamental decision to make. Will the problems of the future be solved by compatibility, or should the door be left open for the introduction of some radically new system that may be inherently incompatible with existing practice?

## Unjust Tas

AT this time of the year, when the Budget is being prepared, it is customary for those with specialist interests to beg the Chancellor of the Exchequer to lighten their own particular burdens. We will resist the temptation to plead for remission of the very heavy purchase tax carried by broadcast receivers and valves, but feel impelled to draw attention to the recently introduced impost on loudspeakers. This measure, imposed with the idea of bringing the broadcast relay companies into the taxation net in company with makers of receivers for direct radio reception, seems to have worked most unfairly and with harmful results in many directions.

When the tax was introduced last August, we stigmatized it as a drag on progress. There is no link in the chain of sound reproduction that is more susceptible to development than the loudspeaker, and the imposition of tax has lessened the incentive towards improvement. We now understand there has been a sharp fall in the sales of those types liable to tax.

The position of those installing equipment for public address seems to be particularly unfortunate and also anomalous. Some firms escape the tax altogether, while others have to pay. In particular, there is a tendency to penalize low-level sound reinforcement systems of the most up-to-date kind that employ a large number of small cone speakers. Then there is considerable difficulty in deciding precisely what type of reproducer comes within the scheme.

# Television "Booster" Stations 

Filling the Gaps in the B.B.C. Coverage

By P. J. HARVEY, ${ }^{*}$ b.sc., A.c.g.I.

IN any scheme of nation-wide television coverage based on a chain of high-power transmitters, there will inevitably be appreciable areas in which the received signal is either unsatisfactory or virtually non-existent. This state of affairs is noticeable in the British Isles, much to the annoyance of long-suffering viewers in the fringe areas, but it will undoubtedly be of much greater concern in countries with more rugged geographical features.
In Britain the television development programme of the B.B.C. still leaves some 20 per cent or more of the population outside the effective coverage of the transmitters. The areas which are not adequately catered for, and for whom no future plans have yet been dis-


Map showing in black the areas of the country not covered by the stations of the complete B.B.C. network (outside the $100 \mu \mathrm{~V} / \mathrm{m}$ contour). Masses of population in these black areas (over 500 people per square mile) are indicated by the small white houses.
closed, lie in parts of East Anglia, North Wales, Northumberland and the Lake District, in the remoter parts of Scotland, Northern Ireland and the West Country, and in the Channel Islands and Isle of Man.

Even if the B.B.C. finally decides to follow up its existing programme with additional facilities for the fringe areas, it will probably find it uneconomic by present methods to effect a complete coverage of the country. Thus, unless a somewhat different approach to the problem is considered, there will always be isolated communities without a satisfactory service.

In the circumstances the only practicable means of achieving a more complete coverage would appear to be the introduction, where necessary, of some form of television relay, to pick up, amplify and re-distribute the signal of the existing network throughout the locality.
The first essential for the successful operation of any relay system of this nature is a satisfactory site at which the network signal can be received at an acceptable level and signal-to-noise ratio. This should be not more than, say, 5 miles from the centre of the community it is to serve. It is seldom that such a site cannot be found, since the geographical obstacles which give rise to most of these black spots can be of assistance in eliminating them, as they will usually provide excellent sites for reception of the network signal. It is economically desirable that all such relay schemes should be fed by the direct reception of the main transmitter's signal, for if extensive cable or radio relays were used to bring the signal into the area the total cost of the installation would probably become prohibitive.

Whilst there must be a basic similarity between all television relay systems, there is some difference of opinion on the best method of distribution. To date the only method that has gained favour is distribution by wire. An alternative to this is distribution by reradiation. In this form of relay the receiver could be identical with that used in the wired system, but the station would include low-power vision and sound transmitters and function as a local booster to a main station.

## Designing the Station

The essential features of an experimental low-power booster station are shown in Fig. 1. The basis of the design is a high-quality sound and vision receiver, from which audio and video signals are obtained and used for the modulation of conventional sound and vision transmitters, using established circuit techniques.

In most applications it is necessary for the operating frequencies of the equipment to be so arranged that

[^9]

Fig. 1. Block diagram showing the essential features of on experimental low-power booster station.
the separation between the received and re-radiated sound and vision carriers is at least equal to the normal channel spacing. This is to prevent blocking of the receiver by the re-transmitted signals, and to obviate the possibility of loop feedback. If adjacent channel working is to be used, adequate filtering at the receiver input circuit is necessary. The exception to this is the rare instance when the natural features of the locality can be used to supplement the screening between the receiving and re-radiating aerials and to ensure that there is no overlapping of the main and booster station service areas. Re-radiation can then be done on the original sound and vision frequencies. $\dagger$

It might be thought that the obvious basis for design would be to change the sound and vision fre-

[^10]quencies at r.f. and then amplify the converted signals to the required output level. Two main obstacles were encountered in designing the vision stages of a prototype equipment on this basis. The first was the difficulty of maintaining adequate bandwidth and linearity throughout the relatively large number of linear amplifiers necessary to build up to the final level of r.f. output. The second was the multiplicity of spurious frequencies that can be generated by unwanted heterodyning, in the frequency converter stage, of various harmonics of the incoming, outgoing and local-oscillator frequencies. Serious patterning on vision can result if these spurious frequencies are radiated, even at comparatively low power. The effect is produced by the interference of such frequencies with the vision carrier, local oscillator or i.f. of the distant receiver, or with any harmonic of these which produces a difference frequency within the video band. These tendencies become increasingly marked with reduction of the spacing between the incoming
and outgoing frequencies. If the frequency relationships are arranged so as to eliminate the effect with one design of receiver, this does not necessarily hold for another using a different i.f.

Both these problems are more easily dealt with in a design in which vision and sound are recovered from the incoming signal and remodulated on the transmitted carriers.
Except in favourable conditions, the receiver is preceded by a pre-amplifier accepting both the sound and vision signals, a cascode circuit being used to realize optimum noise factor. If the level of the received signal is at all low, the pre-amplifier unit is located at the top of the aerial mast and connected directly to the receiving aerial. The increase in the overall noise factor due to the feeder losses is thus avoided.

## Automatic Gain Control

The design of the vision and sound receiver is largely conventional. The common r.f. amplifier is an earthed-grid circuit, ensuring correct termination of the input feeder. Automatic gain control is effected in the vision receiver by gating and using the frame pulses. The control voltage thus derived, which is proportional to the sync pulse amplitude, is amplified and applied to the suppressor grid of the first i.f. amplifier. The a.g.c. hold thus obtained is adequate, the variation of output being not more than 1 db for 30 db variation of input. A.G.C. is applied to the sound receiver in a similar manner and a comparable hold is obtained. A switching voltage is derived from the sound carrier for automatic stopping and starting of the transmitters as the station is normally unattended.
The carrier frequencies of both vision and sound


The complete equipment is housed in a single 4-ft rack.
transmitters are derived by frequency multiplication from crystal oscillators. It is essential that the multiplication be as small as possible to minimize the number of spurious frequencies to be reduced to an acceptable level, so overtone circuits are used. In fact, if line, or even lumped-constant, circuits operating at the final carrier frequencies can be designed to an adequate stability they are to be preferred from this point of view.

An r.f. power range of $10-250$ watts for the vision transmitter is considered suitable to cater for the differing requirements of various applications. Grid modulation of the final amplifier is used, since at these levels of r.f. power little difficulty is experienced in obtaining adequate vision signal power for the purpose. In a power amplifier stage of this type, which is rated for continuous operation at peak power, no increase of efficiency is gained by using anode modulation, and the slight improvement of linearity does not justify the extra video modulating power required. The linearity obtained with grid modulation is quite acceptable, provided that the amplifier is only driven into moderate grid current and the regulation of both the driver and modulator is sufficiently good. The regulation of the h.t. and bias supply is equally important. The driver and modulator stages, therefore, are both designed to present the lowest possible source impedance, and in fact the design principle used in the video circuits is to maintain a low impedance throughout, even at the expense of some efficiency.

The d.c. component is re-inserted at the grid of the modulator, the anode of which is then directly coupled to the grid of the power amplifier. Clamping at the black level would be preferable, but the d.c. restorer gives satisfactory results, provided the a.g.c. hold of the receiver is maintained, and is used by reason of its comparative simplicity.

The distortion of the modulation envelope, first by the detection characteristic of the receiver and secondly by the modulation characteristic of the transmitter, largely affects the sync pulses only. To compensate for this distortion a certain amount of reshaping of the pulses is done by means of a separate sync chain, the output of which is applied to the power amplifier grid in parallel with that of the modulator.

A vestigial sideband filter is used to suppress the unwanted sideband of the vision transmitter if the licensing regulations specify such a restriction.

## Coverage

The design of the sound transmitter closely follows that of the vision transmitter except that anode modulation is used with advantage since, in this case, the modulation frequency range does not have to extend down to zero. The carrier power is adjusted to be approximately one-quarter of the peak power of the associated vision transmitter. The required range of carrier power for the sound transmitters is therefore $5-50$ watts.

An indication of the r.f. power necessary for a typical booster station is provided by tests with the experimental installation referred to above, which radiated 50 watts (peak) vision and 15 watts sound. Satisfactory coverage was obtained over an area of approximately 50 square miles, involving a population of some 100,000 , in conditions that were by no means ideal. Where the local topography allows line-of-sight propagation the coverage obtainable with


On the left is shown the picture as received by the booster station, and on the right, for comparison, is the transmitted picture.
radiated powers of this order could be appreciably greater.
The design of the transmitting aerials is largely dependent on the position of the aerial relative to the area to be covered and relative to any distant transmitter operating on the same channel. In the ideal arrangement, which can frequently be achieved, the transmitting aerials are sited at the same vantage point as the receiving aerials and beamed to "look down" on their service area. The coupling between transmitting and receiving aerials is normally minimized by directivity, change of polarization, or in unfavourable circumstances by the use of special reflector screens.
The question of frequencies on which such booster stations might be permitted to radiate is a somewhat vexed one. The lower u.h.f. band ( $400-600 \mathrm{Mc} / \mathrm{s}$ ) is not yet overcrowded, and a few channels in the portions of this band allocated for television purposes might be approved by the licensing authority for relay schemes.
Consideration would have to be given to the design of commercial receivers for these frequencies. Unless the relaying of television programmes at u.h.f. gained general acceptance it would not be economic for manufacturers to market a special set for these bands, and to begin with converters would have to be used with existing v.h.f. receivers. However, it is reasonable to assume that the cost of such converters would not be greatly in excess of that of the pre-amplifier at present common in fringe areas. When set off against the additional cost of the multi-element aerial arrays which are also widely used in these areas, the u.h.f. converter should not involve any alarming increase in the total cost of a receiving installation and might even reduce it.

Tests carried out at v.h.f. indicated that the signal which could be laid down by a booster station of this type would permit the use of an inexpensive indoor aerial in some 75 per cent of the service area. There is no reason to assume that results would be greatly inferior at u.h.f.
It might be pertinent to consider also whether the existing channels in the v.h.f. band ( $41-67 \mathrm{Mc} / \mathrm{s}$ ) could be put to further use. The primary consideration for operating booster transmitters on any of these channels would be the complete elimination of any
possible interference with existing services on the same channel. The magnitude of this problem should not be over-estimated, since with radiated powers of this low order and suitably directive transmitting aerials it should not be difficult to confine the effective service area of the transmitters within quite narrow limits. A change of polarization from that of the original service on the channel might lessen the possibility of mutual interference, although such interference would probably occur in conditions of tropospheric propagation and the possible rotation of the plane of polarization in these conditions would tend to lessen the advantage.

## Economics of the System

In comparison with distribution by wire the booster transmitter scheme has the advantage of relative simplicity, resulting in an appreciably lower capital outlay for the relay installation and lower maintenance costs. The circuitry of the complete booster station is comparable with that of the central receiving and amplifying equipment of the wired system. The receiving equipment used by the viewer with either system is also similar, although in the case of the booster station scheme there is the added advantage of using standard television receivers supplied through normal retail channels. The additional complexity of the wired system thus lies in the extensive network of feeders, amplifiers, matching pads and equalizers required for distributing the relayed signals, and this additional complexity is a measure of the difference in cost of the two systems.

The problem of providing a satisfactory service in the fringe areas must eventually resolve itself into one of economics. At present, distribution of television programmes in this country is organized almost solely on a national basis, with the exception of one or two relay systems installed by large commercial organizations. However, in the fringe areas, owing to the uneven distribution of communities it may well be that the problems of distribution can only be sclved by independent small-scale relay schemes of the type envisaged in thls article.

There is considerable doubt that the B.B.C. or any commercial concern would find it economic to operate such relay schemes. The type of organization which
suggests itself for the purpose is a syndicate representing local television interests, functioning on a parttime non-profit-making basis. The booster station equipment described above is particularly suited for such a venture by reason of the low capital outlay involved, the fact that it is designed to operate unattended, and that maintenance would be well within the scope of the local television dealers' service organizations. A company could be formed to implement these proposals, representing receiver owners, local television dealers and the Town Council, or other local authority, but care would be necessary to ensure that such a body would show no less a sense of responsibility than the national authority in guaranteeing the permanence, continuity and quality of the service.
Any such departure from established practice would of course necessitate the approval first of the B.B.C. in the matter of copyright and secondly of the G.P.O. for the grant of an operating licence. Local television relay schemes would not involve the B.B.C. in any additional expenditure, but the increased licence
revenue resulting from the extension of coverage should benefit them financially.

It seems likely that the cost of such relay schemes would have to be borne solely by the local viewers, since it is doubtful if any portion of the licensing fec could be diverted for this purpose. However, an intelligent guess at the balance sheet of a typical relay service of the booster transmitter type indicates that the additional annual cost to individual viewers need not be alarming. It is estimated that the initial outlay on equipment for a booster station would not exceed $£ 4,000$, and the annual cost of operation and maintenance $£ 500$. Assuming the average size of a community involved to be 50,000 , a rough estimate of the number of television receiving licences might be 3,000 . An extra charge of 10 s per annum per licence would realize an income of $£ 1,500$, which should be adequate to finance the scheme.

The material for much of this article was obtained from experimental investigations of the possibilities of booster television stations. The work was done under G.P.O. licence on behalf of the author's firm.

## RADID IN INDIA

WITH but one broadcast receiver per thousand of the population, India is virgin soil for the radio manufacturer. It must not be thought, however, that it is open to prospectors, for restrictions bar the import of complete receivers. There is, therefore, a growing indigenous radio industry, which, according to figures recently given, now produces over a quarter of a million broadcast receivers a year.

Surveying the general situation both industrially and culturally, an article in Industrial Review* criticizes the Indian government's general broadcasting policy and calls for a commission to investigate the working of All India Radio with a view to improving the service. Despite the fact that the licence fee has been increased by 50 per cent-to Rs. 15 p.a.-it is evident that the government cannot afford to undertake any major expansion programme. The writer considers that the only alternative is to introduce commercial broadcasting by private enterprise. The interest in, and regular reception of, the transmissions from the commercial stations in Goa and Ceylon are cited as proof of the failure of A.I.R. to meet the needs of the average listener.
Many other aspects of the Indian radio industry are discussed in the Industrial Review survey including component manufacture, receiver exports and, too, the need for breaking down the Indian's prejudice against articles marked "Made in India."

Another of the more convincing arguments for the introduction of sponsored broadcasting in India is adauced by the editor of the Radio Times of India in the issue of December 16th. With the total press circulation in the country reaching barely 1 per cent of the population, the need for a medium which will reach the millions who do not, or cannot, read the newspapers and magazines is obvious and can, it is claimed, be met by commercial broadcasting. It will,

[^11]of course, be argued that the present radio density of $1 / 1,000$ is even lower than the press circulation, but few advertisers take space in all publications. Moreover, it is claimed that better programmes from more stations will create a greater demand for sets.

Illiteracy is, of course, a tremendous drawback to the advancement of any country, and, in an endeavour to overcome the present confused multi-lingual situation in India the government has decreed that Hindi will be the national language and will, during the next 15 years, gradually replace English.

Considerable concern is felt in academic circles over this change. The question of translating textbooks into Hindi will inevitably take many years and a letter from a number of distinguished Indian educationists is published in the December 16th issue of the Radio Times of India criticizing the method of transition. "We see," they write, "no possibility of our being able to do without some good knowledge of English, for any number of translations will never completely meet the needs of all the subjects taught in our universities. . . . It follows that in the interests of higher education in India it is eminently necessary to keep up an adequate standard of English even after it ceases to be the medium of instruction. . . . It is a regrettable feature of our present-day education that the standard of English has been steadily going down, which has very seriously affected the study of all subjects."

$$
\begin{gathered}
\text { Oscillalor/Filter Līnil } \\
-a \text { Correction }
\end{gathered}
$$

IN the circuit, Fig. 3, p. 130 of the last issue the ganged switch $S_{1}, S_{2}$ is in the position giving rejection filter characteristics and not the connections for oscillation. The lettering of the switch should, therefore, be reversed.

# Electronic Film-Making 

Application of Television Principles to Motion-picture Production

ALTHOUGH WORK which has been carried out on the process to be described was primarily intended for the production of cinematograph films for theatrical release, it is becoming increasingly obvious that another and equally important application exists in the need for an economic method of "canning" television programmes.
It would be difficult to run a sound broadcasting service without constant recourse to recorded programmes. The need is even greater in a television service where a much greater rehearsal-to-transmission ratio has to be faced and where studio sets, lighting and all the paraphernalia of vision have to be erected and dismantled. In consequence, the amount of studio space to maintain a wholly "live" relevision programme becomes prohibitive and "prefilming" is considered to provide the only means of satisfying the otherwise insatiable demand.
Other arguments for pre-filming television programmes are based on the difficulty of securing the services of leading artists at the studio during the most important viewing hours, when they may be working in the theatre or cabaret. Also, in a large country like America, time differences are of extreme importance. Last, but not least, due to the high costs of television programmes, some permanent record, which can be used for a repeat performance without further encroachment on studio or artists' time, and which can be sold to other broadcasting organizations, becomes an economic necessity.
The use of an intermediate electronic process in the production of motion-picture film is an idea which has intrigued technicians for some years. The major attraction is that it would confer upon the film maker the ability to produce a motion picture in precisely the same way as a television broadcast programme is produced, by substituting a motion-picture camera (photographing a cathode-ray screen) for the radio transmitter. The camera would faithfully record on film an entire production, complete with all cuts, fades, wipes, titles, etc., in full continuity in the actual time taken to enact the performance. The resultant film would, in theory, be capable, after printing and development, of projection as a finished product without curting or further editing.

Doubtless in practice this ideal would be difficult, though not impossible, to achieve in its entirety, but the sponsors of this system of production are confident that by its use films can be produced in a much shorter time than by orthodox methods using optical cameras. The general arrangement of the apparatus used in the proposed process, where electronic cameras are substituted for optical cameras on the studio floor and the pictures emanating from them displayed at a control point simultaneously on a number of viewing units, is shown in the diagram.

Any number of cameras may be used, each with its separate viewing unit, but, in practice, the number is likely to be limited to three or four.

At the control point, the director can, by means of instantaneously acting controls, select, cut, mix or superimpose any of the available shots, and judge the finished result on a master viewing screen. In parallel with this master viewing tube, but elsewhere in the building, is a monitor c.r.t. from which the photographic image is recorded. Continuous recording can be achieved by the provision of two motionpicture cameras operated in relays with a small overlap.

The use of multiple optical cameras, in orthodox film-making, has been regarded by most film technicians as impracticable because of the difficulty of reconciling the different viewpoints; not to mention the difficulty of lighting satisfactorily for more than one camera.

It must be conceded that the use of more than one optical camera is artistically too hazardous. The value and composition of a picture and the effectiveness or otherwise of the lighting can only be judged by the operator who sees the scene through his camera's eye by looking into the viewfinder. Even so, he can only look through the viewfinder of his own camera and cannot simultaneously assess the complementary viewpoints of other cameras focused on the scene from different angles.
The director, for his part, is working blind the whole time and must rely on his skill and judgment and the advice of his cameraman to form a mental picture of the scene as it will be recorded by the camera. He, moreover, knows that he will not be able to see the result of his work until the next day when the "rushes" are developed and printed, by which time it will be difficult and expensive to retake scenes which, when viewed subsequently in the projection theatre, fall short of the effect he strove to secure.

It is natural, therefore, that both director and cameraman should feel impelled to embark on a lengthy series of retakes until each is satisfied that there is sufficient satisfactory material in the "can" to render it virtually certain that a good sequence can be extracted from it.

This and the elaborate cutting process that naturally follows are two of the main reasons why the production of motion pictures is such a protracted and costly business, often yielding only one or two minutes of finished film per day from each studio stage, because only a very small proportion of the large footage shot is actually used. If electronic cameras are used, however, an entirely different state of affairs supervenes.

First, the electronic camera can be made smaller * High-Definition Films, Lid.
and lighter than its optical counterpart. It is more sensitive and hence needs less lighting intensity, it is noiseless in operation and requires no reloading.

These, however, are incidental gains and the real advantage reposes in the fact that the director can see on individual screens before him the simultaneous picture output of one, two, three or more cameras and can mix, fade or cut them at will. By directing his camera operators through headphones he can secure exactly the effect he desires on the master screen, with the full knowledge that what he is seeing is being photographed by a motion-picture camera from the screen of another c.r.t.
His judgment can be confirmed and strengthened by his advisers looking at duplicate screens, either on the spot or elsewhere in the studio. The precise effectiveness of the lighting for all cameras can be assessed by the lighting expert and, due to the flexibility of the electronic apparatus, minor defects in lighting can be corrected while actually observing the result of the adjustment on the screen.

Because of this, the number. of retakes can be reduced to an absolute minimum and the amount of finished film produced per day should be capable of being raised to 10 or even 20 minutes, which represents a large saving in time and hence cost.

In addition, the facilities provided by the electronic process may be fully exploited during rehearsal when cameras can be used and a large part of the work normally carried out in the cutting room or laboratory subsequent to taking can be worked out and scripted during rehearsal thus ensuring that the actual shooting will proceed smoothly and rapidly. Exterior shots taken with normal optical cameras can be inserted in their proper context during studio shooting by means of film scanning apparatus, and long sequences of the production can be assembled in proper continuity and presented as a "rough cut" of the finished picture-all without recourse to cutting-room work or the introduction of opticals in the laboratory.

## Multiple Filming

Because the actual cinematography is from the screen of a cathode-ray tube several negatives may be simultaneously produced by operating a number of tubes in parallel. For example, both a $35-\mathrm{mm}$ and a $16-\mathrm{mm}$ negative can be produced simultaneously and to these can be added a $16-\mathrm{mm}$ direct positive, made by reversing the picture electrically and photographing the negative image on the c.r.t. screen. This direct positive can be "hot-processed" and, in less than a minute after production, can be televised back on to the director's master screen for his scrutiny or, alternatively, projected for the enlightment of the artists.
Apart from the savings in time and cost outlined above, there are additional, less obvious but not unimportant advantages. To mention only a few :-

First, the director himself remains in personal control and can supervise and modify camera height, angle, speed of tracking, panning, etc., all with complete assurance that comes from being able to see what he is doing.

Secondly, the use of the electronic multi-camera technique enables successive sequences to be shot with a high degree of continuity and this, with the ability to see the resulting picture throughout and not to have to wait a day to see disconnected rushes, enables the director to preserve an equally high degree of con-
tinuity of thought and grasp of the theme of the story in all its subtle nuances of presentation. Without the electronic method and faced with the limitations of a single optical camera technique, the director is in the position of a painter who is denied the ability to sketch in his outline first and must bring one by one all the small details to perfection first, like a worker in mosaic.

Thirdly, the ability to work rapidly and in logical sequence of continuity tends to be of great assistance to the artists and enables them to give a better, more consistent and personally satisfying performance.

Finally, an advanced stage has been reached in the development of artificial scenic devices including various inlay and overlay processes. By these methods all types of glass shots, back projection and travelling matte shots may be effected electronically during the actual shooting and without subsequent optical work in the laboratory. By immensely fast electronic switching from a master foreground camera depicting the artists to a slave background camera focused on a small photograph of the required scenery, a perfect illusion can be achieved. The potential saving in cost offered by this process requires no further comment.

## Alternative Electronic Aids

A number of attempts have been made from time to time to fit optical motion-picture cameras with "electronic viewfinders." These consist of a television camera coupled by a split optical system to a common taking lens. The director can thus assess the picture being received by each camera and, by remote switching, each film camera can be operated at will.

There is no doubt that this scheme offers advantages over the use of purely optical cameras and makes multi-camera working a much more practical proposition. In the view of the author and his associates, however, it represents only a half-solution of the problem because it does not eliminate the necessity for subsequent work in the cutting room or production of optical fades, etc., in the laboratory. Moreover, artificial scenic devices of the electronic type cannot be used nor is immediate "playback" available.

Added to this, the combined optical and television camera becomes extremely bulky and heavy and it is not any more silent than a normal motion-picture camera.

In consequence of these limitations, it seems fairly clear that this "half-way" process can never compare in time-saving and efficiency of production with the "all-out" electronic system advocated by the author, There is ample justification for pressing forward with the development of the system.

## Technical Considerations

It is obvious that, to be an acceptable substitute for orthodox direct photographic methods, the electronic film-making process must be indistinguishable in quality from the result obtained by normal processes. To achieve the requisite standard of quality poses severe but not insoluble problems.

When considering the problems of generating television images, it is extremely difficult to think in terms other than those of television broadcasting where many factors, both economic and practical have to be studied. Many of the limitations imposed by broadcasting, however, can be disregarded when the whole of the work is carried out on "closed circuit" within the confines of the studio.

Consider first the question of definition. It is well


Block schematic of the general arrangement of the equipment used in the system of electronic film-making described by the author. By using a number of c.-r. tubes in parallel several negatives may be produced simultaneously.
known that the transmission of a television picture involves the transmission of a very wide band of frequencies. Broadly speaking, the wider the band of frequencies, the better the definition of the picture. There are limitations to the bandwidth which can be transmitted and the broadcasting engineer has, in fact, to make a compromise between what he would like to do and what he can do.

* In the first place there is the question of ether space, which is extremely limited. Consequently a band of only five megacycles wide can be allocated in the ether for each transmitting station. Into this has to be fitted television picture and television sound (with enough spacing between them to avoid mutual interference) and guard bands on each side to avoid interference with adjacent channels. As a result of this limitation the picture is bound to be of low definition in terms of cinematographic standards.

Next, there is the cost of transmission of television signals over coaxial circuits. If the vision signal consists, as it does in this country, of a band of frequencies $3 \mathrm{Mc} / \mathrm{s}$ wide, the cable repeaters have to be spaced about 12 miles apart which, with the cost of the cable, represents a very substantial item. If the band of frequencies were extended appreciably, the number of repeater stations would go up until, in the absurd case, there would be one continuous line of valves, anode to cathode from one end of the country to the other.

Again, the cost of television receivers is greatly increased if the definition standards are raised substantially, and they become too expensive to command extensive sales.

## Standards of Definition

As a result of all this, the standards of definition used for broadcast television must of necessity be kept comparatively low.

On the other hand, the electronic process for film production described is subject to no such limitations. Working on a closed-circuit system and operated under laboratory conditions, there are no limitations to the bandwidth which may be used, there is no radio link, no long cable and no television receiver. Instead, any standard of definition that appears desirable and any signal waveform that may be convenient can be used. There is no necessity to mix in synchronizing information to form a composite signal; synchronizing signals may be kept entirely separate and sent down a separate cable to the reproducing monitor, which materially eases the operation of the apparatus.

As regards the actual standards of definition required to produce by electronic means a picture on film which is indistinguishable from one taken by normal cinematography, opinions differ over very wide limits.

The writer, however, reasoned on these lines. In the average motion-picture negative the limiting resolution on axis is about 40 lines per mm. and is degraded to about 25 lines per mm . in the release print. This appears to represent a value which has been determined empirically over many years as being acceptable to the public. No doubt it would be extremely uneconomic to work to a significantly higher standard, and the resultant picture would probably be considered too sharp and hard for artistic portrayal of many scenes.

If 40 photographic lines per mm . are taken as an
average figure for motion-picture definition, this corresponds to 80 conventional television lines per mm. or about 1,300 total for the frame. Actually there are certain corrections which can be applied. For example, the fine definition response of a television system can be maintained by "top boost," whereas under conditions used in motion film photography it tails off fairly steeply to the extinction point. The author, therefore, considers that it may not be necessary to use so many lines but that a 1,000 -line picture with depth of modulations maintained right up to cut-off frequency might be sufficient for the purpose in hand.

The production of such a picture does not present any very serious difficulties, as electronic systems capable of resolving 1,500 to 1,800 lines are a demonstrable possibility to-day.

## Contrast Range

Turning to the question of contrast; broadcasting imposes severe limitations due to such things as the high electrical interference level from motor cars, etc., in all but the areas of high field-strength, and the electrical overload capacity of the transmitter. These limit the dynamic range usefully available in the transmission, with the result that both the black and white ends of the transmitted tonal scale must be crushed, if the radiated signal is not to be continuously undermodulated, to the detriment of all but the viewers near the transmitter.

In electronic film-making no such limitations exist, and any desired range of contrast can be realized with complete freedom from the crushing of any part of the tonal characteristic. In fact, compensation can be introduced to straighten the far from linear overall transfer characteristic of the photographic processes from master negative to release print. For this reason more faithful results can, in theory, and indeed in practice, be achieved, once the immensely flexible electronic link has been introduced into the picture. It has been suggested that electronic filmmaking can do for the motion picture what electronic recording has done for the sound disc, a view with which the author confidently concurs.

Prior to the development of high-definition electronic methods of film-making, the cost of making pre-filmed programmes for television by orthodox optical camera methods has been too high for wide use even if the utmost despatch in studio production is achieved. It is firmly believed, however, that the use of electronic methods, probably in conjunction with $16-\mathrm{mm}$ film, will enable programmes to be prefilmed at a sufficiently economic cost and with a standard of technical quality to satisfy the most critical observer.

In this way, the establishment of an exchange of television programmes throughout the world becomes practicable; the introduction of sound tracks in different languages being an already known art should not be difficult.

From this it is but a short step to the application of the electronic process to films for theatrical release, and the extension of the process to colour film is not far off.

In conclusion, the author is convinced that in the development of this process, with its manifold applications, a new and tremendously flexible tool is being forged, which will have great potentialities in both the film and television industries.



Diagram " $A$ " shows the correct positioning of a Brimistor in the average circuit.
Diagram "B" shows an oscillogram of the normal "switch-on" surge which damages valves and dial lights.
Diagram " $C$ "': this oscillogram shows the gentle rise of BRIMISTOR controlled current and complete surge suppression.

## OTHER USES

I. Wired in series with the lead to the reservoir condenser, Brimistors can be used to reduce the voltage surge associated with directly heated rectifiers.
2. Connected in series with a choke or focus coil, the negative temperature co-efficient of a Brimistor may be used to compensate for the positive temperature co-efficient of the coil.

## cURRENTSURGE

BRIMISTOR RESISTORS

Wired in series with valve heaters . . . dial lights . . . BRIMISTORS hold the starting current at a low value, permitting it gradually to rise until working temperature is reached, thus ensuring maximum life for both valves and lamps.
When cold the resistance of a BRIMISTOR is high. As it warms up with the passage of current, however, this resistance falls gradually to a very low level which usually can be ignored. At the very most a small reduction in the line cord or other voltage dropping resistance will provide the necessary compensation.

## CHARACTERISTICS OF BRIMISTORS

| Type | Cold resistance$20^{a t} \mathrm{C} .$ | Resistance with the following currents flowing |  |  |  | Max Current | Price List |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 0.1 amp . $0.15 \mathrm{amp} .0 .2 \mathrm{amp} \cdot 0.3 \mathrm{amp}$. |  |  |  |  |  |
| CZ1 | ohms 3,000 | ohms 180 | ohms 100 | ohms $75$ | ohms 44 | $\mathrm{amp}_{0 .}$ |  |
| CZ2 | 3,000 5,500 | 180 170 | 100 90 | 66 | 44 38 | 0.3 | $3 / 6 d$ $\cdot 3 / 6 d$ |
| CZ3 | 1,500 | 100 | 50 | 35 | - | 0.2 | -1/6a |
| C4 | 760 | - |  |  |  | 1.25 | +5) - d |
| CZ6 | 3.000 | 200 | 120 | 80 | 45 | 0.45 | *3/6a |

* Brimistors are not subject to Purchase Tas

Use CZI for 0.3 amp . and 0.1 amp . valves.
Use CZ2 for 0.15 amp . and 0.2 amp . valves.
Wire type CZ3 across the dial lamp (in conjunction with a CZI or CZ2 as above).

## Notes on the use of BRIMISTORS

1. Owing to the high operating temperature (up to $250^{\circ}$ C.) Brimistors must be spaced away from coils and waxed components.
2. They should be inserted in the "live " end of the heater chain-i.e. between mains resistance and rectifier valve heater.
3. At least $\frac{1}{2}$ in. of wire must be left at each end before soldering to a tag.

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# Phase-Angle Measurements 

A Note on the Accuracy of the Various Methods

By F. A. BENSON,*<br>M.Eng., Ph.D., A.M.I.E.E., M.I.R.E.

THE CATHODE-RAY tube still finds application for determining the phase angle between two quantities. In making the measurements two voltages are produced proportional to the quantities, one voltage (maximum value $V_{1}$ ) is then applied to the vertical deflector plates of the tube and the other (maximum value $\mathrm{V}_{2}$ ) to the horizontal plates. If voltage $\mathrm{V}_{2}$ leads voltage $\mathrm{V}_{1}$ by an angle $\theta$ an ellipse will result, provided the voltages are sinusoidal, which is bounded by a rectangle with sides 2 X and 2 Y (Fig. 1) and which has major and minor axes of lengths $2 a$ and $2 b$ respectively.

Two well-known methods have been given by Carter and the author ${ }^{1}$ for deter-
 mining $\theta$ from the ellipse and they have made estimations of the magnitudes of the maximum errors likely to be caused by trace thickness and any harmonics present. In the first method BC and AD are measured (Fig. 1) and $\sin \theta=\mathrm{BC} / \mathrm{CD}$. For the second method the major and minor axes of the ellipse are measured together with 2 X and 2 Y . Then $\sin \theta=(2 a)(2 b) /(2 \mathrm{X})(2 \mathrm{Y})$.

A note has been published recently ${ }^{2}$ showing that $\theta$ can be calculated from the simple expression $2 \tan ^{-1}$ ( $2 b / 2 a$ ) provided that the horizontal and vertical quantities produce equal-amplitude deflections. It follows that here the angle subtended by the minor axis at the end of the major axis is the phase angle. It is stated in the note that this method does not appear to be generally known but this is thought to be unlikely since it has been given already, along with a proof by Fleming ${ }^{3}$, as early as 1925 and by Glaser ${ }^{4}$ in 1952. Attention has also been drawn to the method by Bainbridge-Bell ${ }^{5}$.

It is interesting to calculate the maximum errors likely to be produced by trace thickness for this final method, however, and to compare the results with those previously published ${ }^{1}$ for the first two methods. In making such calculations it will be assumed that no harmonics are present and the amplitudes of the two quantities are made exactly equal although it will be appreciated that this condition may be difficult to obtain in practice and will lead to additional errors in determining the phase angle. In making any measurements on the screen of the cathode-ray tube errors are introduced due to the thickness of the

[^12]trace at either end of each dimension involved. It will be assumed, as previously ${ }^{1}$, that the position of the end of each dimension can be estimated to within 0.5 mm since a trace thickness of about 1 mm is typical of ordinary oscilloscopes. The errors in phase angle can then be written down. Thus, since $\theta$ is calculated from $2 \tan ^{-1}(2 b / 2 a)$, the maximum and minimum values of $\theta, \theta$ max and $\theta$ min respectively, are given by :-
$$
\tan \left(\theta_{\max } / 2\right)=(2 b+1) /(2 a-1)
$$
and $\tan \left(\theta_{\min } / 2\right)=(2 b-1) /(2 a+1)$
where $2 a$ and $2 b$ are expressed in mm . The curves of Fig. 2 show the variation of error with $\theta$. The errors have been calculated from the expression :-
$$
\delta \theta= \pm(1+\tan \theta / 2) /\left[a\left(1+\tan ^{2} \theta / 2\right)\right]
$$
for values of $2 a$ equal to 60,100 and 140 mm , which are suitable for screen diameters of about $2 \frac{1}{2} \mathrm{in}$., $4 \frac{1}{2} \mathrm{in}$. and 6 in . respectively.

It is seen that the maximum errors in the three cases are $\pm 0.99^{\circ}, \pm 1.38^{\circ}$ and $\pm 2.31^{\circ}$ respectively and occur at $45^{\circ}$. Thus, the present method shows considerable improvement over the other two methods for phase angles above $20^{\circ}$ providing the assumption of equal amplitudes can be justified in practice and further that the graticule used for measuring $2 a$ and $2 b$ can be positioned accurately on the face of the cathoderay tube screen. The latter condition will almost certainly present difficulties and lead to additional errors which may be large compared with the tracethickness errors calculated.


# Saving Television Bandwidth 

Must We Transmit ALL the Information

in the Picture?

LJESS than a year after R. V. L. Hartley had published his now-famous theory on the relation of information to bandwidth, the principle was invoked-perhaps unknowingly-in an interesting suggestion for reducing bandwidths in television systems. This was in 1929, when the old Nipkow scanning disc was still the rule. In the words of the Patent specification': "It has been customary in the past to transmit successive complete images of the transmitted picture. This method of picture transmission requires a band of frequencies dependent on the number of images transmitted per second. Since only a limited band of frequencies is available for picture transmission, the fineness in the detail of the transmitted picture has therefore been determined by the number of picture elements and the number of pictures transmitted per second. In accordance with the invention, this difficulty is avoided by transmitting only the difference between the successive images of the object. By this mode of operation no signal is transmitted when there is no change in the image of the picture or object and its fineness of detail is limited only by the speed of the action to be transmitted."
This proposal recognized the important fact about television that, as a result of the scanning process, a lot of the information transmitted by the signal consists of useless repetition. In the language of Information Theory, the signal contains a large amount of redundancy. The only really new information is represented by the changes in the picture -the amount by which each picture differs from the previous one (Fig. 1). By transmitting only these differences, or "error signals," the amount of information to be conveyed in a certain time-the rate of information-could be reduced considerably, and

[^13]this would make possible a corresponding reduction in bandwidth.

The method suggested for doing this was quite simple (see Fig. 2). The signal from the photo-cell was applied to one half of the primary winding of a transformer and to the other half through a delay network. The delay introduced by this was equal to the complete scanning cycle of the Nipkow disc. On the first scan the complete picture was transmitted through the top half of the transformer. But during the next scan, $S_{\text {", }}$, the delayed version of the first scan $\mathrm{S}_{-1}$ was present in the bottom half of the transformer in phase opposition; so if the next scan was exactly the same as the first the two cancelled out and no signal was transmitted. If the next scan was different, then a corresponding difference signal was induced in the transformer secondary and transmitted.
Of course, nothing was actually done about reducing bandwidths along these lines. But now that the bandwidth problem is becoming more and more acute a great deal of fresh attention is being paid to the subject. The new approach has been greatly influenced, if not actually stimulated, by the development of Hartley's ideas into modern Information Theory, for information is undoubtedly at the heart of the matter. The average television channel, in fact, regarded as a means of conveying information, proves to be grossly inefficient.

The central problem, then, is the large amount of repeated information, or redundancy, in television signals and how it can be reduced to make possible a saving in bandwidth. The 1929 proposal was concerned with the long-term redundancy caused by the similarity, or correlation, between successive pictures. This is very great because of the repetition in time necessary to achieve continuity of vision-especially in backgrounds, which do not change very often.

Fig. I. Illustrating the technique of transmitting only the changes in the picture: (a) represents the first scan, and the second, while (b) shows the amount by which (c) differs from (a) as a result of the figure moving to the left.


Nowadays, a great deal of interest is being shown in the short-term redundancies caused by the similarity of adjacent picture elements. This correlation has to do with the repetition necessary to build up the picture spatially, in two dimensions, rather than with the repetition necessary to give continuity of vision in time. It is highest, of course, in the flat unchanging areas of the picture, the areas containing the least amount of detail. For practical purposes it can be analysed into two dimensions-the redundancy between adjacent elements along a line, or horizontal redundancy, and that between adjacent elements in successive lines, or vertical redundancy. But, of course, there is no significant difference between the two, except that the vertically adjacent elements are in different line scan periods.
By working on this correlation between adjacent elements Kretzmer ${ }^{2}$ has attempted to measure the total redundancy in various pictures. He estimates from his results that if it could be eliminated a 50 per cent reduction in bandwidth should be made possible. This, however, does not seem to be the limit of possible achievement. Gouriet, for example, has adopted the opposite approach to the problem. He has measured the amount of information in various pictures and compared it with the maximum possible information that could be in them-the amount allowed by the bandwidth. The discrepancy is then a measure of the redundancy. The amount of information is measured in terms of the brightness changes in the picture, while the maximum possible information is represented by the maximum possible number of brightness changes-a $3-\mathrm{Mc} / \mathrm{s}$ chequerboard pattern. Even with a highly detailed picture the information rarely exceeds 5 per cent of the maximum possible. This suggests a redundancy of something like 90 per cent to be exploited. The great problem is how to do it. Can we really achieve a ten-to-one reduction in bandwidth?

## Statistical Approach

A fresh attack has been made on the problem recently,' more or less along the lines of the original 1929 proposal. That is, to transmit only error-signals representing the changes in the picture. In this case, however, the error-signals are not the differences between successive pictures but between adjacent picture elements. So the redundancy eliminated is of the short-term kind associated with the two-dimensional structure of the picture, as explained above. Moreover, this latest method has been influenced somewhat by the statistical approach which modern science is nowadays adopting towards so many problems. The original method predicts, in effect, that the next picture element will be exactly the same as the corresponding one in the previous picture-and any discrepancy in this prediction is transmitred as an error-signal. But the new method works on the principle that the next element will probably follow the general trend of several previous elements. Which, of course, is a more likely situation. So the error-signal transmitted is the difference between the actual next element and a predicted next element derived from the statistics of the ones before.
The apparatus which does the job (Fig. 3) is very similar to that in the early proposal. It uses a delay

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Fig. 2. Proposed system for transmitting only the differences between successive pictures (circa 1929). The scene is scanned by a Nipkow scanning disc and the light variations are picked up by a photocell.


Fig. 3. An improved modern version of the Fig. 2 system. The actual value of a picture element is compared with its predicted value and the difference is transmitted as on error signal.
network, to hold up the previous elements until the next element arrives for comparison, and a cathodecoupled differential amplifier (instead of the transformer) to produce the difference signal. The significant modification is that the delay network has now been extended to give delayed versions of several previous elements instead of one. The delays are such that the signals all combine at the grid of the valve and the resulting composite signal amounts to a prediction of what the next element, $\mathrm{S}_{\text {s }}$, will be. The attenuators merely adjust the relative contributions of the various signals to this prediction.

As this system is designed to work on the correlation between adjacent picture elements, the delays introduced by the network are made equal to the time intervals between these elements. That is, between adjacent elements along a line or between corresponding elements in successive lines, depending on whether it is the horizontal or the vertical redundancy which is to be removed. Fig. 4 shows how the error signals appear. By using a combination of two such "decorrelators" one with element-to-element delays and the other with line-to-line delays, it is possible to remove both the vertical and the horizontal redundancy.

At the receiving end, of course, some system is needed to restore the redundancy, so that the picture can be presented in its original form. In other words, a certain amount of the received information has to be repeated; and in practice this means that it has to be stored in some way. In the 1929 system the receiving screen consisted of a mosaic of little reflectors which were moved in and out of the light by the signals. The first scan set them all in their appropriate positions and they remained there permanently until moved by the error signals representing the changes in the picture. Thus the information was stored and,

(Courtesy Bell System Technical Journal.)
Fig. 4. The results of reducing redundancy. At (a) is the original picture, while (b) shows, in two-dimensional form, the error signals produced by horizontal decorrelotion and (c) the error signals produced by vertical decorrelation. Note the absence of horizontal contours in (b) and vertical contours in (c).
in effect, repeated in a succession of picture periods. Nowadays the process could be done electronically, and a picture storage tube would be used in place of the mechanical mosaic.

One very obvious advantage of transmitting only error signals, as in Fig. 4, is that their power is very much lower than that of the actual picture waveform: they only represent the brightness changes. But this does not constitute a bandwidth saving as such. The sharpness or harmonic-content of these error signals is still as great, and requires as much bandwidth, as in the original signal. But the power-saving can be translated into bandwidth saving by a suitable method of coding. In other words, the information in the brief error signals can now be spread out over a greater length of time-we can take more time in transmitting them. As a result of this " smoothing out " process we can obtain a continuous and very much lower rate of information. And with the lower rate of information our bandwidth can be made correspondingly smaller. A coding device, then, is an essential part of the television channel.

## "Smoothing Out" the Information

There are various methods of coding or "smoothing out " the flow of information, and one of the most promising was described recently in an I.E.E. paper by Cherry and Gouriet '. This system is really intended for decorrelating, or reducing the redundancy, in the original television waveform, but it could equally well be used as a means of coding the error signals produced by the "prediction" method of decorrelation. The principle is simply to vary the speed of the scanning system so that more time is taken to transmit the high-information parts of the picture and less time to transmit the low-information parts. The highinformation parts are the sharp transitions between different brightness values, which are represented by correspondingly sharp transitions on the camera mosaic and in the waveform (see Fig. 5); so the velocity of the scanning beam is made to vary according to the sharpness or slope of the waveform. The apparatus suggested for doing this is shown in Fig. 6. A differentiating circuit measures the slope of the waveform and controls the scanning velocity accordingly. Thus the high slope of a sharp edge tends to slow down the beam, and as a result the slope is itself reduced until it settles down to an equilibrium value. Similarly a

[^15]low slope causes the beam to go faster. The arrangement is really a self-adjusting or servo system in which the beam velocity is controlled by a function of itself. Fig. 7 shows the effect on the television waveform. In the receiver, the velocity of the scanning beam of the c.r. tube is varied in synchronism, and the information for doing this is derived from the modified slopes of the received waveform.

To use this system for coding the error signals produced by the "prediction" method of decorrelation it would be necessary first to store the error signals on the target of an electrostatic storage tube so that they could be scanned. But, as stated above, the system is, in itself, a decorrelator, as well as being a coding device. One can see from the waveforms in Fig. 7, for example, that very little time is wasted on transmitting the flat redundant parts of the picture-a time saving which is in fact used for sending the sharp transitions in the picture at a lower speed.


Fig. 5. Line waveform showing low-information parts and high-information parts.


Fig. 6. System for controlling the velocity of the camero sconning beam according to the amount of information in the picture.

Of course, with the Fig. 6 arrangement the decorrelation is done only along the lines-in the horizontal dimension. The vertical redundancy, between adjacent lines, remains. Gouriet and Cherry, however, have pointed out that this vertical redundancy could actually be reduced in a similar way. The modified line scan waveforms (Fig. 7(b)) would be stored in a picture storage tube in the form of a horizontallydecorrelated electrical image of the picture. This would be done by the "writing" beam of the tube. The "reading" beam would then scan the stored image in the vertical direction, and it would use the same variable-velocity system so that this time the vertical redundancy would be removed. The output waveform would look very much like Fig. 7(b) but owing to the horizontal redundancy having already been reduced the differences between successive line waveforms would be much greater. At the receiving end the same process would be done in reverse, using a storage tube with a variable-velocity " writing" beam to restore the vertical redundancy and a variablevelocity "reading" beam to restore the horizontal redundancy.

## Matching to the Eye

These bandwidth saving schemes are, of course, all based on the general argument that there is no point in transmitting more information than is needed to reassemble the picture on the receiving screen. But the matter does not really end there. One could also argue that there is no point in presenting on the screen more information than the human visual system is able to cope with. At the moment it seems that we are doing just that. In other words, it is probable that we could make further reductions in the rate of transmitting information, and hence in bandwidth, by taking more advantage of the limitations of the human visual system. The screen should be "matched" to the eye.

One way of doing this would be to reduce the scanning rate-either the number of pictures per second or the number of lines per picture. The number of lines per picture is, of course, an old problem and everybody has different ideas on the subject. In practice it has been a matter of using as many lines as possible without making the bandwidth too unreasonable. This has been based on the assumption that the more lines there are in the picture the sharper it will appear. But in fact this assumption is only partly true. The subjective sharpness increases rapidly with number of lines up to a certain point, but after this the addition of further lines makes very little difference. In other words, after this point an extra expenditure of bandwidth brings no worthwhile returns in subjective sharpness and so is largely wasted. The French 819line pictures certainly do not look twice as sharp as our 405 -line pictures. The optimum point is actually in the region of 400-500 lines, so in the British and American systems, at any rate, a reasonable matching of screen to eye has already been achieved, and little more can be done in this direction.
There remains the possibility of reducing the number of pictures per second. At the moment, in the British system, we have 25 pictures per second. But we do not really need this picture frequency to achieve continuity of vision, for the human visual system is still incapable of separating successive images when they are presented at a rate as low as 10 per second. (This may have something to do with the

Fig. 7. At (a) is a normal line waveform while (b) shows how it would appear with variablevelocity scanning as in Fig. 6.


Fig. 8. Illustraticn of the dot-interlace system of television in which the picture is built up from four sets of dots.
$10-\mathrm{c} / \mathrm{s}$ electrical alpha rhythm of the brain which is thought to be a kind of physiological scanning mechanism.) So it seems that we could quite safely halve the picture frequency to $12 \frac{1}{2}$ per second and thereby make possible a 50 per cent reduction in bandwidth.

The only difficulty in this is that although the eye may only need $12 \frac{1}{2}$ pictures per second for continuity of vision, it becomes very sensitive to flicker at anything below 50 per second. In the present system, of course, we have overcome this difficulty by the ingenious expedient of presenting the 25 pictures per second as a series of interlaced half-pictures, or frames, at 50 per second. As a result of this interlacing the flicker is transferred from between successive pictures to between adjacent lines. This is just right for the eye, since the only part of the retina which is capable of resolving the lines (the fovea in the centre) happens to be insensitive to flicker. The parts of the retina which are sensitive to flicker (the peripheral regions) are incapable of resolving the lines and so cannot see it.

The logical thing to dc, then, is to extend the interlacing principle and halve the picture frequency to $12 \frac{1}{2}$ per second, while still keeping the flicker frequency at $50 \mathrm{c} / \mathrm{s}$. In other words, each picture would now consist of four frames instead of two ${ }^{5}$. This, in fact, is the principle of the dot-interlace system of television, in which each picture is built up from four interlaced sets of dots (see Fig. 8); each set contributes a quarter of the total picture information. Dot interlacing could therefore be used in this way to halve the bandwidth by halving the picture frequency. (Alternatively, of course, with the existing bandwidth

[^16]it could be used to double the definition.) Another method of halving the picture frequency while keeping the flicker frequency at $50 \mathrm{c} / \mathrm{s}$ is to use a picture storage tube. gIn other words, each picture is stored and flashed on the screen four times before changing to the next picture-a principle which has already been utilized in certain film projectors. The main disadvantage of reducing the picture frequency is that fast-moving objects appear as multiple images when the necessary measures have been adopted in order to overcome flicker.

There is also the possibility of exploiting the limitations of visual acuity, or resolving power of the eye, as it is sometimes called. Jesty and Phelp have shown, for example, that visual acuity falls off rapidly with reduced brightness and contrast ${ }^{6}$. This suggests that it might be possible to reduce the amount of information in the shadowy parts of the picture without anybody being any the wiser-and such a reduction would certainly enable a bandwidth saving to be made. Then again, visual acuity is also very much lower at the edges and corners of the picture, since these are the parts which are projected on to the insensitive peripheral regions of the retina. It is true that we are free to focus the sensitive fovea of the retina on to these parts but, generally speaking, we keep it at the centre of the picture. So, again, a reduction of information might be made towards the outer edges of the picture. Another interesting feature of visual acuity is that it seems to fall off as the screen is brought nearer to the eye. This suggests that a small and near picture can satisfy the viewer with less information than a large and distant one subtending the same angle at the eye. Or, for the same bandwith, the small and near picture will look sharper. Unfortunately this physiological effect seems to work in the opposite way to the psychological effect described by C. Burns in the January issue, by which the bigger picture appears to be the sharper of the two.

## Colour Television

With colour television, of course, the bandwidth problem is even more acute, for there is so much more information to be transmitted. Various methods have been proposed for reducing the bandwidth by successive sampling of the three colour components, but a new idea put forward recently by Valensi seems to get much closer to the heart of the matter. In this, the components of the colour are not transmitted at all, but just a code signal representing the colour itself. The colours are analysed in the usual way by three pick-up tubes with filters in front, and the voltage outputs from these form a combination which actuates the transmission of the appropriate code signal.

The fact that a person's visual acuity is about 50 per cent lower with colour pictures than with black and white has already been exploited in the "mixed highs" system of colour television. Since there is no point in describing the high-information parts of the picture in colour when the eye cannot see the result, the bandwidth of the colour channel can be reduced by about 50 per cent without any noticeable difference in the picture. The high-information parts are made visible to the eye by being reproduced in black and white. Again, the visual acuity with blue is very much lower than with the other two colour components, so this makes possible an additional bandwidth reduction in the blue channel.

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## HOOKS IRECEIVED

Principles of Radar, by Members of the Staff of the Radar School, Massachusetts Institute of Technology. Third Edition, by J. F. Reintjes and G. T. Coate. Extensively revised test book of the basic concepts and and techniques of radar. Pp. $985+\mathrm{XV}$; Figs. 642. McGraw-Hill Publishing Company, 95, Farringdon Street, London, E.C.4. Price 55s 6d in U.K.

Sound Reproduction (Third Edition), by G. A. Briggs. Revised and enlarged survey of loudspeakers, recording and allied subjects Pp. 368; Figs. 315 . Wharfedale Wireless Works, Bradford Road, Idle, Bradford, Yorks. Price 17s 6d.
TV Manufacturers' Receiver Trouble Cures, Volume I, by Milton S. Snitzer. More than 300 troubles met with in current American television receivers, and the cures recommended by their makers. Pp. 115; Figs. 51. John F. Rider Publisher, 480, Canal Street, New York, N.Y. Price $\$ 1.80$.
Television and Education in the United States, by Charles A. Siepmann. Sociological study of the influence of television viewing on the school work of American children. (A UNESCO pamphlet.) Pp. $120+x i$; Her Majesty's Stationery Office, Kingsway, London, W.C.2. Price 6s.

## Manmiacturers: Literature

Portable Disc Recorder containing a four-stage amplifier with push-pull output of 8 watts. Illustrated brochure with specification and response curves and a leaflet describing the ribbon microphone that goes with the equipment, from E.M.I. Sales \& Service, Hayes, Middlesex.

Diamond Wheels in various shapes and sizes for cutting and grinding ceramics and glass. Leaflet giving general information and describing improved method of manufacture, from Colton \& Company (Lapidaries), Diamond Tool Division, Walpole Road, London, S.W.19.

Electronic Measuring Instruments for laboratory and test purposes. A brochure giving short descriptions and photographs of the complete range available from Philips Electrical, Century House, Shaftesbury Avenue, London, W.C.2. Also a leaflet on the Philips range of variable transformers.

Hi-K Ceramic Capacitors; a technical bulletin (No. 27, series 2) giving specifications of wire-ended and screw-base tubulars and lead-through and miniature-bead types, from the Telegraph Condenser Company (Radio Division), North Acton, London, W. 3.

Components, accessories and construction kits; a selection of more popular items from the general stock list of City and Rural Radio, 101, High Street, Swansea, Glam.

Fractional Horse-Power Motors and others; an abridged list with prices of the types available from Higgs Motors, Witton, Birmingham, 6 .

Megohmmeter with a range of $0.3-20,000,000 \mathrm{M} \rho$, enabling changes in resistance to be measured without adiustment of controls. This and other electronic measuring instruments described briefly in an illustrated broadsheet from Electronic Instruments, Red Lion Street, Richmond, Surrey.

Stabilized Power Supply with d.c. output voltage continuously variable from 200 V to 350 V and stability of $\pm 0.5 \%$ for $\pm 10 \%$ mains variations. Specification in a leaflet from Harvey Electronics, 273, Farnborough Road, Farnborough, Hants. Also a leaflet describing their Production Facilities available for precision mechanical and electro-mechanical work.

Batteries; h.t. and l.t. for receivers and hearing aids. A brochure giving voltage tappings, weights, sizes and prices from Siemens Electric Lamps and Supplies, 38 and 39, Upper Thames Street, London, E.C.4.

Gramophone Amplifiers, 4 W and 10 W outputs, with tone controls but no output transformers. Descriptive leaflet and list of accessorics from Ian M. Ross, Electro-Acoustic Laboratories, Knockbreck Road, Tain, Ross-shire, Scotland.

Resistance and Ratio Boxes, d.c. bridges and similar instruments for the laboratory; brief specifications on an illustrated leaflet from Croydon Precision Instrument Company, 116, Windmill Road, Croydon, Surrey.

## THE "BELLING-LEE" PACE <br> Providing technical information, service and advice in relation to our products and the suppression of electrical interference

## R.E.C.M.F

Readers of this page, receiving invitations, will be visiting the R.E.C.M.F'. Exhibition from the Ifth to the roth April.

As most readers are aware, this is a private exhibition, held under the auspices of the Radio \& Electronic Component Manufacturers Federation, and as in recent years, will take place at Grosvenor Honse, I ark I ane.

The majority of those receiving invitations will be representatives of the Services and Supply Ministries. Industrial concerns and technical establishments will be represented by designers, technicians and engineers, together with a very large number of overseas buyers.

We hope any readers who may be present, will come and see us and what we have to show, on Stand No. 52.

## Ignition Suppression of Motor Cars, etc.

Lately the correspondence columus of "The Autocar" have carried it number of very interesting letters on this subject, and we have the lidlitor's and the Author's permission to reprocluce one letter which appeals to us. We have never met Mr. Goodger, but his


The oscillatory mature of the ignition capacity spark current is shown above.
letter disposes of a common misconception in a very concise manner, as in it, he refers to the alteration of waveform. but does


Introduction of a suppressor resistor effectively eliminates the oscillation: of the spark current.
not illustrate this. The two curves shown above are our contribution.
"Alteration of Current Waveform
( 64652 ) -Is your correspondent (64627) really correct in assuming that the insertion of a
suppressor-a resistor of, say, 5,000 ohms-in the h.t. lead at the distributor necessarily reduces the voltage at the spark gap?

Don't forget that, until the actual moment of the spark jumping to earth, the circuit is not completed and what the resistor cloes, surely, is to enable the voltage to build up to maximum before jumping the plug gaps.

The suppressor does not work as such by lowering the voltage by any appreciable extent, but by altering the waveform of the h.t. current.

An electrical current which rises quickly to maximum and peaks sharply will radiate much more interference than one which builds up comparatively gradually and collapses in the same way. Hence the resistor at the distributor. Ancl what about side itenms, like longer plug life and easier starting from cold ?

I own a somewhat lethargic 8 h.p. saloon and, since fitting a suppressor some while ago, I have noticed no drop in performance or increase in petrol consumption, but, contrariwise, I have noticed a very slight but definite increase in the engine's ability to pull at low speeds.
B. Goodger.'

Oswestry, Shropshire.
Electric Cooker as an Interference Suppressor
This is a hardy annual: we are often asked why it is that interference which makes radio intolerable, sometinies ceases when an


Belling-Lee" Capacitor Suppressor
L. 750
electric cooker is switched on. This was given quite a lot of publicity recently in one of our daily papers which has a very wide circulation. The answer is, that by its nature, a cooker has a very considerable capacitance between each line and earth, and this by-passes the unwanted H.F. interference to earth. Naturally this rather large "capacitor" is only effective when
in circuit, i.e., when switched on, and is therefore an "expensive" suppressor. A similar effect may be expected from a standard

" Belling-Lee " Set Lead Suppressor L.300/3
capacitor suppressor such as the " Belling-Lee" L. 750 which measures $3 \frac{1}{4} \mathrm{in} . \times 2 \frac{1}{2} \mathrm{in} . \times 3$ in. and of course passes no current. When used in this way, it is generally fitted where the mains enter, i.e. at the meter board between neutral and live, the centre point being earthed. A set lead suppressor (containing chokes) is often preferred; this is L.300/3 and is fitted, as its name implies, between the supply point and the recciver.

## Ever Hopeful

Recently, one of our engineers had occasion to visit places rather remote from the bustle and strife of London. Television was something that was likely to arise " in a few months time." He heard of a prospective viewer who would only buy a broadcast receiver that had a loudspeaker grill large enough to take a television screen when the time came!

## STOP PRESS

All efficient new cars have at least one ignition resistor fitted as standard.
Bring your car up-to-date for 2/6d. Fit a "Sparkmaster" for controlled spark, easier cold starting, reduced pinking and plug burning.
Cars so fitted do not interfere with T.V. reception.
It will pay you to fit a " BellingLee" "'Sparkmaster" to-day.

Written 28 February, 1953.


## MARCONI communication systems



Before Guglielmo Marconi's invention of wireless, a ship out of sight of land was literally cut off from the world. Today you can, in a matter of minutes, contact a ship in
mid-Atlantic. Such is the measure of his service to communications and safety at sea. Such too, is the measure of his ideals, which still inspire the Company he founded.


PLANNED


# Presenting Technical Information 

Six More Writing Awards by the R.I.C.

THE NEED for more articles making known to executives and administrators, as well as to technicians, new and interesting developments in radio, television and electronics in this country is stressed by Vice-Admiral Dorling, director of the Radio Industry Council, in announcing six further awards of premiums for technical writing. It is the writing of this type of article that the R.I.C. wants to encourage and Vice-Admiral Dorling emphasized that it is almost essential when preparing a highly technical article to start with a "simple, concise explanation of the aims and applications of the techniques described. Greater attention should also be paid to clarity."

Four awards of 25 guineas will be made to the authors of each of the following articles published during 1952:-
P. H. Parkin, B.Sc., A.M.I.E.E., who is principal scientific officer in charge of the sound insulation and acoustics section of the Physics Division at the D.S.I.R. Building Research Station, and P. H. Taylor, A.M.I.E.E., technical director of Pamphonic Reproducers, Ltd., for their article "Speech Reinforcement in St. Paul's Cathedral," published in the February and March issues of Wireless World.
T. Somerville, B.Sc., M.I.E.E., F.Inst.P., head of the Electro Acoustics Group of the Engineering Research Department, B.B.C., and C. L. S. Gilford, M.Sc., F.Inst.P., head of the acoustics section of the Group, for an article in the B.B.C. Quarterly, Spring, 1952, on " Composite Cathode Ray Oscillograph Dis-
plays of Acoustic Phenomena and their Interpretation."
J. A. Jenkins, M.A., M.Inst.P., who is in charge of the Photoelectric Division of the Mullard Vacuum Physics Research Laboratories, and his chief assistant, R. A. Chippendale, B.Sc., for "Some New Image Converter Tubes," published in the July, 1952, issue of Electronic Engineering.
W. R. Stamp, on the staff of the Admiralty Research Laboratory, Teddington, for an article in Discovery, September, 1952, on " Underwater Television."

Ex gratia awards of $£ 10$ each are made to:-
T. W. Bennington, who is engaged in ionospheric and short-wave propagation work in the Research Department of the B.B.C., for his article on "Propagation of v.h.f. via Sporadic E," (Wireless World, January, 1952), and
G. N. Patchett, Ph.D., B.Sc., A.M.I.E.E., M.I.R.E., A.M.Brit.I.R.E., head of the Electrical Engineering Department, Bradford Technical College, for his contribution on "Faulty Interlacing," (Wireless World, July and August, 1952).

The awards will be presented at a luncheon in London on April 9.
An interim award of 25 guineas was made at last year's Radio Show.
Full details of the scheme for encouraging technical writing, which provides for the awarding of up to six 25 -guinea premiums each year, are available from the Secretary, R.I.C., 59, Russell Square, London, W.C.I.

## NEWS FROM

Brighton.-"Sub-miniature Equipment" is the subject of a talk to be given by a representative of Multitone Electric, Ltd., to the Brighton \& District Radio Club on April 14th. On April 28th F. R. Canning will speak on "Designing a Simple Transmitter." Meetings are held at 7.30 at the Eagle Inn, Gloucester Road, Brighton, 1. Sec.: R. T. Parsons, 14, Carlyle Avenue, Brighton.

Chester.-The April programme of the Chester \& District Amateur Radio Society, which meets each Tuesday at 7.30 in the Tarran Hut, Y.M.C.A. Grounds, Chester, includes: a discussion on a top-band transmitter ( 7 th ); "Problems of Transmitters in North Africa" by Capt. Carpenter, MD2B ( 14 th ); and a discussion on radio interference (21st). Sec.: N. Richardson, 1, Victory Villa, Newton Lane, Upron, Chester.

Cleckheaton.-At the meeting of the Spen Valley \& District Radio \& Television Society on April 8th. A. Smith, B.Sc. (G2BOO), will speak on "Nuclear Physics." The club meets on alternate Wednesdays at 7.30 in the Temperance Hall, Cleckheaton. Sec.: N. Pride, 100, Raikes Lane, Birstall, Nr. Leeds.

Coventry.-Members of the Coventry Amateur Radio Society will be given a "Radio Reader's Digest" at the meeting on April 13th. A lecture-demonstration on "A Multi-purpose Power Supply" will be given by Ray Bastin (a member) on April 27 th . The club meets on alternate Mondays at 7.30 at the Y.W.C.A., Queen's Road, Coventry. Sec.: K. Lines, 142, Shorncliff Road, Coventry.

East Brighton.-We have been informed by F. J. Walton, secretary of the East Brighton Short Wave Club, that owing to lack of support, the affairs of the club are being wound up.

Manchester.-The new secrctary of the South Manchester Radio Club (M. Barnsley) will show members the Mullard Film Strip No. 5 on April 10th. Meetings are held on alternate Fridays at 7.30 at Ladybarn House, Mauldeth Road, Fallowfield, Manchester. 14. The club transmitter (G3FVA) is on the air on $3.5 \mathrm{Mc} / \mathrm{s}$ on alternate Fridays. Sec.: M. Barnsley (G3HZM), 17, Cross Street, Bradford, Manchester, 11.

Peterborough.-P. J. Walker, of the Acoustical Manufacturing Co., will give a lecture-demonstration on "High Quality in Small Rooms" to members of the Peterborough Radio and Scientific Society (G3DQW) on April 9th. The club meets each Thursday at 7.30 at its headquarters in St. Paul's Road, Peterbornugh. Sec.: S. Woodward, 72, Priory Road, Peterborough.

Portsmouth.-Meetings of the Portsmouth \& District Radio Society, which is affiliated to the Royal Marine Signal Club, are held each Tuesday at 7.30 at the Royal Marine Barracks, Signal Club, Portsmouth. Sec.: L. Rooms (G8BU), 51, Locksway Road, Milton, Portsmouth.

Southend.-The lecture programme of the Southend $\&$ District Radio Society includes "The Diode Detector" by K. J. Varcoe, A.M.I.E.E., on April 17th." The Principles and Use of the Cathode-Ray Oscilloscope" by S. F. Asquith, A.M.Brit.I.R.E., and "The Manufacture of Piezo-Electric Crystals" by Dr. C. P Fagan of Cathodeon, Ltd., are among the forthcoming events. Also visits are planned to the Cable and Wireless station at Brentwood, North Foreland station and the Rochford Aerodrome. Sec.: G. Chapman, Bell Hotel, 20 Leigh Hill. Leigh-on-Sea, Essex.

# Designing a Tape Recorder 

By J. M. CARTER. b.sc.*

## 2-The Amplifier and its Associated Oscillator and Level Indicator

ALL STUDIO TYPE tape recorders employ separate record and playback heads, each function occurring simultaneously, so that the playback head provides a continuous monitoring of the recorded signal. In this case, therefore, two separate amplifiers are necessary for record and playback. For almost all other non-professional applications, however, this is not strictly necessary as with a good instrument it is extremely rare for anything to go wrong with the recording process. For this reason, and also because of the saving in space and cost, the majority of commercial equipments use a combined record-playback head, and with the addition of the appropriate switching the amplifier is also made to


Fig. 2. Frequency response curve (uncompensated) of head and transformer, from a recarding made under constantcurrent conditions.
serve both purposes. What are the main requirements for such an amplifier, and how may they most easily be met?

The first essential is high overall gain. Assuming that the combined record-playback head is of the usual twin-track type and is used with a step-up matching transformer to the first valve grid, the gain at $50 \mathrm{c} / \mathrm{s}$ from this point to the output will require to be roughly 110 db . With this order of gain hum can become a major problem, especially when a metalcored inductor, i.e., the head, is connected to the input! However, a carefully planned layout of the components, and some special precautions, can reduce it to a negligible value. The head matching transformer should preferably be of the "astatic" type, to assist hum pick-up cancellation, and, in addition, be completely screened in a Mumetal case, it is also advantageous to mount it so that it can be rotated to the position of minimum hum pick-up. This rotatable feature is also desirable with the mains transformer, as often just a few degrees of movement of the main flux axis can effect a marked reduction in hum. These adjustments, will be made, of course, when the equipment is completely assembled, and with the motors running, in case a slight amount of hum "bucking" is necessary.

To exclude the possibility of internally generated valve hum in the low level stages, only valves with a " bifilar" heater should be used, and as an additional safeguard the heater winding on the mains transformer


Fig. 3. Playback equalizing circuit. Fixed resistors can all be of $\frac{1}{2}$ watt rating, but $R_{3}$ and $R_{11}$ should be of the high-stability type. A silvered-mica capacitor should be used for $C_{4}$. Transformer Type 977 and choke Type 727, made by Wright and Weaire, can be used for $T_{1}$ and $L_{1}$ respectively.

[^18]should be centre-tapped to chassis. The first and second valves must also be free from microphony, and the compact form of the modern all-glass miniature type is a decided advantage in this respect, although if a "lock-in" base is employed it is advisable to mount the valveholder assembly on rubber and use flexible leads to the valveholder tags, to avoid transmitted vibrations from the chassis. To avoid chassis currents, a single chassis " earth" for all the earth return leads of the first two valves, input transformer and playback head is essential. Another desirable feature in connection with the first two valves is the use of highstability anode load resistors to eliminate resistor noise.

The amplifier must also be equalized to accommodate the varying response from the head on playback, which occurs when recording is made under virtually constant-current conditions in the usual way (Fig. 2). This may be done in several ways, but the method shown in Fig. 3 is recommended for its simplicity, stability and possibility of adjustment to almost any unequalized tape response.

## Equalizer Design

As can be seen, it takes the form of a combined potentiometer-shunt circuit across the output from the first valve anode load resistor. When the CRL network between points A and B has a high impedance, which occurs at low audio frequencies, and also at the high audio frequencies in the region round the resonance point of $\mathrm{L}_{1} \mathrm{C}_{4}$, the maximum voltage will be tapped from the potentiometer formed by its series connection to the $100-\mathrm{k} \Omega$ resistor $\mathrm{R}_{30}$ and the coupling capacitor $\mathrm{C}_{24}$. At the same time the shunting effect of the whole network across the anode load of V1 will be a minimum. In the middle audio-frequency range, say $1,000-5,000 \mathrm{c} / \mathrm{s}$, the impedance of $\mathrm{C}_{5}$ and $\mathrm{L}_{1} \mathrm{C}_{4}$ will be low (the latter being well off resonance), and the value of the variable resistance $R_{5}$ will form the greater part of the total impedance, which will be at its lowest value. Hence a much smaller voltage will be tapped off the potentiometer network, and the shunt effect also will be at its maximum. The resonance point of $L_{1} C_{1}$ is usually fixed between 11 and $13 \mathrm{kc} / \mathrm{s}$ and forms the upper limit to the frequency response, after this point the response falling away

rapidly. To avoid hum pick-up, the inductor $\mathrm{L}_{\text {}}$ should either be screened in Mumetal or be of astatic construction.

The second valve V2 is connected as a straight RC-coupled amplifier, with the cathode resistor not bypassed, to provide some measure of current feedback and further improve the linearity.

From the anode of V2 the amplifier can follow any prescribed pattern to the output stage, providing (a) that the remaining amplification is linear and adequate, and (b) that some 13 volts or so of undistorted signal are available at some point for recording.

## Main Amplifier

Perhaps the simplest output stage fulfilling these conditions, and also providing 3 watts of output at 3 per cent distortion, is that shown in Fig. 4. The output from V2 is fed via the volume control $R_{12}$ to the last two valves V3 and V4. A "loop "embracing some 14 db of negative feedback is run from the secondary of the output transformer $\mathrm{T}_{2}$ to the cathode of V3 and, as previously stated, enables 3W of audio power to be extracted from V4 at a low percentage distortion. It also ensures a very flat response from $40-15,000 \mathrm{c} / \mathrm{s}$ from the microphone or radio jack to the output transformer secondary.

When the amplifier is used for recording, the action of plugging the microphone into the microphone jack (Fig. 3) automatically disconnects the first stage and the equalizing circuit, and V2, V3 and V4 then provide a gain of approximately 8,000 to the anode of V 4 . As the voltage required for the peak recording level will only be about 13 , this means that a signal of 1.6 mV from the microphone will fully load the tape. The "Radio" input jack should be used when a large recording input of perhaps 100 mV and upwards is available, as very large inputs would overload the grid of V2. If the source of this larger input is of a high-impedance nature, a shorting plug should be inserted in the microphone jack to prevent any first stage noise coming through; if of low impedance it will, of course, shunt the noise output from the first two stages, and this trouble will not arise.

The amplified signal to be recorded will appear at the anode of V4 and from there will pass through the isolating capacitor $C_{27}$ to the recording network which consists of a $51-\mathrm{kc} / \mathrm{s}$ trap ( $\mathrm{L}_{4}$ and $\mathrm{C}_{13}$ ) and the main series recording resistor $\mathrm{R}_{19}$ and its associated pre-emphasis capacitor $\mathrm{C}_{12}$. The $3,400-\mathrm{pF}$ capacitor $\mathrm{C}_{11}$ is merely to depress the feedback " hump" in the supersonic region.

The $51-\mathrm{kc} / \mathrm{s}$ trap is a parallel resonant LC circuit tuned to the bias frequency by means of a dust core in the coil and serves two purposes: (a) to provide an effective block to bias volts appearing at the anode

Fig. 4. Power amplifier circuit. Resistors are of $\frac{1}{2}$-watt rating. $C_{11}, C_{12}$ and $C_{13}$ are silvered mica. Components suitable for $L_{4}$ and $T_{2}$ are the choke Type 666 and transformer, Type T/428, made by Wright and Weaire.
of V4, and consequently its output network, and (b) to maintain a high impedance to $51 \mathrm{kc} / \mathrm{s}$ across the signal winding of the head and so prevent its feed circuit from imposing too great a reflected shunt load on the bias winding and supply. $\mathrm{R}_{19}$ has a value of $100 \mathrm{k} \Omega$ and $\mathrm{C}_{12}$ of 250 pF , and as the head impedance reaches a maximum of $12,000 \Omega$ at $10,000 \mathrm{c} / \mathrm{s}$, this ensures that the recording current through the head is essentially constant up to the higher audio frequencies, where $\mathrm{C}_{12}$ provides an increase which serves to neutralize the extra head losses and also provides some degree of treble boosting.

## Recording Level

To obtain the best signal-noise ratio the tape must be recorded to that intensity which gives the maximum permissible distortion on subsequent playback. Too low a recording level will result in a lower signal output from the tape requiring more gain from the amplifier and thus, inevitably, more background noise. This monitoring of the recording voltage applied to the head is done by some form of visual indicator. " Magic eyes" which deflect, or neon lamps which glow on the audio peaks are often used. For close control on professional equipment, however, a form of meter is always used and the simple valve voltmeter shown in Fig. 5 can easily be built into the amplifier at little extra expense to provide an excellent peak level indicator.
Part of the signal is taken from the potentiometer $R_{33}$ and applied through $R_{34}$ to the strapped anode and grid of one triode half of $V_{5} . \quad C_{14}$ has a value of $1,000 \mathrm{pF}$ and is intended to prevent any stray bias volts causing a "permanent" meter deflection. $\mathrm{R}_{34}$ is to prevent diode " clipping " on playback when a large voltage may appear across the $10-\mathrm{M} \Omega$ resistor, $\mathrm{R}_{20}$ and cause current to flow between the grid and cathode of the second triode half of $V_{5}$ and thus through the $12 \mathrm{k} \Omega$ resistor $\mathrm{R}_{22}$ to earth. $\mathrm{R}_{20}$ and $\mathrm{C}_{15}$ form the diode load and their values are chosen to have a fairly large time constant, so that the meter needle will "hang" on even the most transient peaks. The second triode half of $V_{5}$ is backed off to zero anode current (at the no input condition) by the potentiometer chain $\mathrm{R}_{22}, \mathrm{R}_{23}$.

By adjusting the setting of $\stackrel{R}{2}_{33}$, the peak recording

Fig. 5. Level-monitoring valve voltmeter. Fixed resistors are oll $\frac{1}{2}$-wott.


Fig. 6. Bias oscillotor circuit. $R_{26}$ should be rated at I watt. Tag morkings for $L_{2}$ relate to the Wright and Weaire Type 579 coil.
level, the setting of which will be described later may be made to correspond to any convenient deflection or marked point on the meter scale.

Turning now to the oscillator section, there are several requirements to be met. If the frequency response is to extend to 10 or $12 \mathrm{kc} / \mathrm{s}$, the bias frequency should be at least five times this, to avoid the production of whistles due to possible harmonic interference from the audio signal. The waveform of the bias should be pure, to prevent a noisy background appearing on the recording, second harmonic distortion being especially bad in this respect. The grid drive, and, if possible, the anode current waveform of the oscillator valve also should be free from distortion, to prevent the radiation of high-order harmonics which, though small, may be of comparable magnitude to radio signals and cause undesirable whistles when recording these. Finally, it must be arranged that the bias will "die" slowly when the oscillator is switched off, as otherwise a sudden break will often result in a "polarized" head and consequent noisy background to recording and playback.

When correctly set up, the oscillator circuit shown in Fig. 6 conforms to these requirements and has proved satisfactory and reliable in practice. It uses a coil with a movable dust core, so that if, in spite of all precautions, an interference whistle is produced between the oscillator and a radio signal being recorded, it may be used to tune out the offending beat-frequency note.

By means of the potentiometer $\mathrm{R}_{25}$ the grid drive volts should be made the maximum possible without distortion becoming evident when the waveform is viewed on a cathode ray tube, this being done, of course, with the normal load connected. The h.t. is supplied across $\mathrm{C}_{20}$, which is a $16-\mu \mathrm{F}$ electrolytic, and is switched beyond this so that the capacitor acts as a reservoir and ensures that a slow dying of the oscillations takes place. $C_{17}$ is intended to prevent any hum getting through from the oscillator to the record and erase heads, and $\mathrm{C}_{16}$ is to prevent any shunting action of the oscillator output circuit on the signal winding, by reflection through the head.
(To be concluded)

## Broadcast Transmitter Distortion

IN reply to a complaint regarding severe distortion manifesting itself as a transient "rattle" heard throughout a large number of broadcasts in recent months, I have been informed by the B.B.C. that "All our domestic service transmitters of 10 kW and over have had compression introduced into the audio chain in order to offset the interference from foreign stations. This means that the transmitter will peak to 100 per cent modulation more frequently than is usual in a broadcasting system, and distortion, particularly on transients, is bound to occur."

I imagine that no high-power transmitter can handle such high modulation under any conditions without considerable distortion being introduced, and when to this is added the product of a.f. compression, the result is intolerable.

In recent years a greater number of people than ever before have given their efforts to designing, and their money to purchasing, receivers capable of giving a really high quality of reproduction from broadcast transmissions; their intentions are stultified when violent rattles are radiated by the transmitter, while the result of the "average" commercial receiver trying to cope with high modulation levels needs no description!

Automatic compression has been spurned in good recording practice for some time on account of its essential accompaniment of distortion, and it would seem that the B.B.C. is compromising too much with quality in return for a but fractional gain in signal-to-noise ratio. Surely equivalent results could be obtained by manual monitoring, the quietest passages being boosted. This would not seem to require excessive trouble on the part of the producer, and would entirely avoid the severe disadvantages of the present method.

The B.B.C. should not forget that some of its listeners are able to receive its transmissions under reasonable conditions-and, anyway, I can assure them that, other things being equal, a clear and undistorted signal sounds better "through" background mush than a muzzy one. London, N. 10 .

IAN LESLIE.

## "Geared to Greenwich"

IN the March issue of Wireless World there are several references to the MSF standard frequency transmissions, including an editorial comment in which some disappointment is expressed that the 24 -hour transmissions planned for the future will not be modulated continuously by $1-\mathrm{c} / \mathrm{s}$ impulses. Although no official announcement has yet been made about the revised programme I can say that the information you give is substantially correct, and I hope this letter will explain why it is not proposed to transmit pulses continuously, and will also clear up one or two misunderstandings that appear to exist.

The transmissions form part of an internationally coordinated experiment designed to discover the best means of securing world-wide reception at a useful level of a standard frequency. Although frequency standardization is the primary object, as much useful information as possible is added in the form of various modulation frequencies. The programme of modulation could be changed to meet new requirements, and any suggested alternatives will be considered in relation to other, and possibly conflicting, requirements, and to other services of standard frequency transmissions. It will be appreciated that compromise solutions may have to be adopted.

It is stated in the editorial comment that the continuous transmission of $1-c / \mathrm{s}$ impulses would make practicable a truly radio-driven clock for anyone who cared to set up a radio receiver; but this is an over-simplification of the
problem. The carrier frequencies available are, of course, allocated by international agreement and although for this country a lower frequency would be more suitable, the difficulties of securing such a channel will be well understood. It is hoped that ultimately at least one of the available frequencies from one or other of the transmitting stations will be well received at any time of the day; but it is expected that automatic selection of the frequency giving the strongest signal would have to be used. Until the service has reached this stage and until it is decided that the rather complicated reception and control equipment is economically reasonable for application to pendulum clocks, there is no advantage in transmitting the pulses continuously and thereby making the service less suitable for other purposes.
The statements made about the service in the article "Time and Telearchics" are also somewhat misleading. The transmissions are the responsibility of, and are monitored by, the National Physical Laboratory, although the very high standard of performance that has been achieved depends on the close co-operation of the Post Office and the Royal Observatory. The $1-\mathrm{c} / \mathrm{s}$ impulses are obtained directly from the quartz standard controlling the carrier waves and the accuracy of time interval is, therefore, the same as that of frequency. The impulses are not independently phased except on rare occasions when on the first day of the month adjustments of 50 ms or 100 ms are made to bring the actual time of transmission nearer to uniform time. During the last year it has not been necessary to make any adjustments. This uniform time is not quite the same as astronomical time but has the effect of a small periodic variation in the rotation of the earth removed. The discrepancy between the time of the impulses and astronomical time may therefore approach 100 ms . In order that there should be only one value of absolute time in use throughout the country the results given for MSF in Wireless Engineer include the timedifference between the $1-\mathrm{c} / \mathrm{s}$ impulses and the Rugby, GBR, time signals.
National Physical Laboratory, R. S. J. SPILSBURY.
Teddington

## "Magnetic Powder Cores"

THE excellent article by Champion and wilkins in your February issue was very interesting although, to me, a little disappointing in one respect.
I was hoping to find some reference to a very important effect which I observed during investigation into the reason for the rejection of too high a percentage of small powdercore r.f. transformers. I refer to the change in the permeability of the core through shock. An effect which does not seem to have been treated anywhere, although it must have been noticed by others.

In the investigation referred to all the usual factors were carefully eliminated; i.e., the accuracy of test gear, testing personnel, manufacturing tolerances and materials. Then a "passed" specimen of a completed transformer was taken straight from test, re-tested and found to be well within the tolerance figures. The main parameter was the primary inductance. The transformer was then subjected to a sharp shock by the simple expedient of dropping it three feet on to a concrete floor. A re-check showed that the inductance had increased sufficient to take it outside the tolerance limits. Other specimens were then given a sharp tap on the core with a hammer with a similar result.
It was first thought the change was due to a variation taking place in the air gap (the moulded cores were E's and I's) so it was decided to measure the permeability of the E's alone, and in nearly every case there was a rise in
the permeability as a result of the E being struck with a hammer.
In these days when apparatus is being subjected to more shocks than ever before, especially in guided missiles where acceleration of the order of several times $g$ are not unusual, it might be revealing to have a more thorough investigation of the subject than was possible during the pressure of wartime production.
It is assumed that the change is brought about by re-orientation of the magnetic particles in the core material by being shock-excited in the earth's magnetic field. But it may not be.
Cambridge.
C. H. BROAD.

## Resistor Colour Code

SOME time ago "Diallist" called attention to the risk of errors in the interpretation of the colour code for resistors, now laid down in BS1852:1952. Even the fourring or "Group-Captain" display is not foolproof, as he showed; but the "tip" method, which is allowed as an alternative, seems to leave considerable scope for ambiguity.
For example, quite a likely value for a resistor is $12 \mathrm{k} \Omega$. The appropriate code is a brown body, red tip, and orange spot or ring. But what about the tolerance, indicated by the colour of the other tip? If the colour is regarded as "none" then the tolerance would be taken as $\pm 20 \%$, but if regarded as brown it would be assumed to be $\pm 1 \%$-a significant difference. Meanwhile, someone looking at the resistor from the other side would see it as brown-brown-orange-red, meaning $11 \mathrm{k} \Omega \pm 2 \%$ !

It is all very well for BS1852:1952 to say "The positions of the coloured markings shall be such that no ambiguity can arise in their interpretation"; an officially approved system ought to be inherently unambiguous.

Bromley, Kent.
M. G. SCROGGIE.

## More Interference

VERY shortly there will be installed all over the country thousands of flashing illuminated beacons on zebra crossings. In addition, the Minister of Transport is reported to be about to permit flashing direction indicators on cars.

It is understood that the beacons will be fitted with suppressors to prevent interference with radio and television reception; but when it is realized that the beacon flashers will have to operate continuously day and night at the phenomenal rate of 57,600 times per day, one can only hope that the suppressors will be effective when the flasher contacts have become pitted, as they inevitably must under such arduous conditions of service.

It would appear essential for arrangements to be made to service these devices frequently, not only to ensure that they are flashing properly, but to check that they are not causing interference due to deterioration of contacts.

The problem will be intensified if the Ministry of Transport permits flashing car indicators. These will also require efficient interference suppressors, but it is difficult to see how the requirement can be enforced. It is to be hoped therefore, that the Minister will not give his blessing to so doubtful an innovation.

Uppermill, Lancs.
JOHN BAGGS.

## "Flywheel Synchronization"

I WAS very interested to read B. T. Gilling's article in the March issue. One important advantage of the balanced circuit such as he has described is the neutralization of impulsive-type interference due to equal conduction of each diode across a common load impedance. In this way the circuit becomes almost immune to the magnitude of the interfering signal and only responsive to its phase relationship with the required synchronizing pulse. This is an advantage not shared with the many, often
complex, circuits, previously described in several American journals.

There are, however, a few points of comment and question that I should like to raise, if I may, in connection with the operation of the circuit described.

The first is one of terminology and concerns the use of the expression "frequency discriminator" used throughout the article. Surely the arrangement involved is that of a phase discriminator, its purpose being constantly to compare the phase of the incoming synchronizing pulses with that of a wave-form derived from the line scanning circuit and to provide a potential difference between two terminals whose magnitude is directly proportional to this difference in phase? This idea of phase discrimination is borne out by the fact that the control voltage generated by the discriminator can only vary to a positive or negative maximum over a phase angle, dependent on the duration of the line flyback period (as we do not wish the synchronizing blanking to occur during the scan period). This will be of the order $\pm \frac{\pi}{10}$ radians for a 405-line system.
I observe from the waveforms shown in Figs. 3 and 4 (which, incidentally, do not agree, the former showing a negative going sawtooth, whilst the latter has a positive going slope), that the waveform at the anode of the line output valve is shown as a sawtooth of potential. In a line time-base designed for magnetic deflection, such as in the Wireless World receiver, the anode load of the output valve is predominantly inductive. As a consequence, if a sawtooth of current is required in the load impedance the waveform of potential appearing across it will be a pulse wave having a mark/space ratio of roughly $1: 10$. In point of fact, due to resistance and transformer core losses, the waveform will have a very small sawtooth component, but of an amplitude considerably smaller than that of the positive pulse ${ }^{1}$. In order to compare the phase of this pulse waveform with that of the incoming synchronizing pulses it is necessary to convert the former into a sawtooth waveform. This may be achieved by coupling the line output circuit with the discriminator via a CR integrating circuit whose time constant is made large with respect to the pulse repetition frequency ${ }^{2}$. This circuit has been included in Fig. 5, and indeed is commented upon, where the need to maintain this time constant at a large enough value to ensure sufficient integration is stressed by keeping R10 at a high value.

Finally, from the operation of the discriminator shown in Fig. 3, it will be seen that when the phase discrepancy has zero value (3b) then the commencement of the line flyback period occurs slightly before the arrival of the leading edge of the synchronizing pulse. The amount of this delay is dependent on the width of the line synchronizing pulse and more particularly on the period of time required by the line time-base circuit to complete the flyback (assuming no phase shift in the synchronizing separation circuit.
Should this delay prove excessive then a portion of the picture on the right-hand side of the screen will be lost and provide brightness modulation to the cathode ray tube for the first part of the flyback period; i.e., a "foldover" effect will be observed.
A method for the correction of this phase error has been described by Wendt and Fredendall ${ }^{3}$ and involves delaying the integrated pulse waveform from the line output circuit, by an appropriate amount to ensure a portion of the blanking period appearing on the right-hand side of the picture.

Coventry
K. G. BEAUCHAMP.

[^19]
# Extending Television Service * Production <br> and Export Figures • New SOS Frequency 

## Brighton "Booster"

THE TEMPORARY low-power television booster to be installed near Brighton to improve reception in the coastal area between Brighton and Worthing will be in operation before the Coronation.

The transmitter which will operate in channel 3, eventually to be used by the permanent station at Rowridge in the Isle of Wight (56.75 $\mathrm{Mc} / \mathrm{s}$ vision and $53.25 \mathrm{Mc} / \mathrm{s}$ sound), will be set up on Truleigh Hill, $3 \frac{1}{2}$ miles north of Kingston-upon-Sea. The temporary station will employ vertical polarization and use the asymmetric side-band method of transmission, as will the permanent Isle of Wight station.

A design for a converter, which will enable sets at present tuned to Alexandra Palace to receive the booster station and eventually the Isle of Wight transmitter, will be given in our next issue. Viewers in the area who receive a satisfactory signal from Alexandra Palace need not, of course, make any change.

## News in Morse

A REVISED SCHEDULE of the London Press Service transmissions of news in Morse having been introduced at the beginning of March, we tabulate below a summary of the details. In addition to the bulletins listed, which are transmitted at 20-27 w.p.m. by the Post Office stations, there are a number of Hellschreiber transmissions.

| G.M.T. | Call | Freq. ( $\mathrm{kc} / \mathrm{s}$ ) | Area |
| :---: | :---: | :---: | :---: |
| 0030-0230 M | GIR | 5,715 | 1 |
|  | GDI | 7,780 | 2 |
| 0130-0300 w | GKU4 | 4,025 | 3 |
|  | GPE26 | 6,977.5 | 4 |
| 0130-0300 d | GPT28 | 8,192.5 | 5 |
| 1315-1415 w | GDZ | 13,910 | 8 |
| 1600-1700 M,Sa | GAY29 | 9,332.5 | 7 |
|  | GDT29 | 9,310 | 6 |
| 1700-1800 w | GAY29 | 9,332.5 | 7 |
| 1715-1815 w | GAH | 8,065 | 6 |
| 1815-1930 d | GIJ | 6,985 | 7 |
|  | GAH | 8,065 | 6 |
| 1845-1945 w | GDG | 6,912 | 3 |
|  | GAY29 | 9,332.5 | 4 |
| 1945-2215 w | GBD2 | 5,885 | 7 |
|  | GAD6 | 7,355 | 6 |
| 2000-2115 M, F | GIH | 10,650 | 2 |
| 2015-2100 M, Sa | GPT | 9,270 | 5 |
| 2045-2200 S | GBD2 | 5,885 | 7 |
|  | GAD6 | 7,355 | 6 |
| 2100-2200 w | GPT | 9,270 | 5 |
| 2115-2215 w | GIH | 10,650 | 2 |
| 2230-2300 M-F | GDI | 7,780 | 2 |
| 2245-0230 w | GIR | 5,715 | 1 |
| 2300-0030 S | GPT | 9,270 | 5 |
| 2300-0230w | GDI | 7,780 | 2 |

The numeral in the fourth column of the table indicates the area to which the transmission is beamed: 1, N. America; 2, S. America; 3, distant Europe; 4, Middle East; 5, Africa; 6, N. E. Asia, Australia and New Zealand; 7, S.E. Asia; 8, China.

These transmissions are not intended for reception in this country, but there is no restriction on the use outside the U.K. of the material broadcast.

## Two Million Viewers

THE LARGEST INCREASE in any month $(110,617)$ brought the number of television licences in force in the U.K. at the end of January to $2,003,449$. The total number of broadcast receiving licences (both sound and vision) on January 31 st was $12,868,183$.

The total includes 179,544 licences for car radio sets. Incidentally, under a new Road Vehicles Regulation, which came into force at the beginning of March, motorists are being asked when applying for vehicle licences whether car radio is fitted.

## T.A.C. Sub-Committee

THOUGH obtained from an official source, the list of members of the technical sub-committee of the Television Advisory Committee, as published in our last issue, proves to have been inaccurate. The corrected list is: Dr. W. G. Radley (engineer-in-chief, G.P.O.; chairman), H. Bishop (B.B.C.), H. Faulkner (G.P.O), R. T. B. Wynn (B.B.C.), Dr. R. L. Smith-Rose (D.S.I.R.), Professor Willis Jackson (Imperial College), G. E. Condliffe (E.M.I.), B. J. Edwards (Pye) and K. I. Jones (Cossor).

## Distress Frequency

A WORLD-WIDE radio-telephony maritime distress and general calling frequency of $2,182 \mathrm{kc} / \mathrm{s}$ will be introduced on May 1st in place of the present $1,650 \mathrm{kc} / \mathrm{s}$ used in the European region.

This is in compliance with the maritime mobile frequency plans drawn up at the Extraordinary Administrative Radio Conference in Geneva in 1951, which provides for all maritime mobile stations to operate within the band 1,605-2,850 $\mathrm{kc} / \mathrm{s}$. The introduction of the plan will also necessitate a change of frequency for coast stations, and it is hoped to publish a list of British stations next month.

## Set Production

TELEVISION RECEIVERS became a significant factor in the British radio export figures for the first time last year, when 6,000 were sold overseas. Production of television sets rose by 14 per cent during the year and the sale of receivers in the United Kingdom reached 782,000 .
The production of domestic broadcast receivers and radio gramophones fell by 41 per cent (compared with 1951) to $1,228,000$. Home sales accounted for 789,000 and export sales for 523,000 .

These figures were given at the annual general meeting of the British Radio Equipment Manufacturers' Association by P. H. Spagnoletti, who was re-elected chairman.

## Exporting Television

TRANSMITTING equipment, as well as receivers, for television is now a "significant" British export.
Marconi's, who have already supplied mobile and studio equipment for the first two television stations in Canada (Toronto and Montreal), are to install $5-\mathrm{kW}$ vision and $3-\mathrm{kW}$ sound transmitters in Ottawa. The equipment is to be installed temporarily in the building housing the existing medium-wave transmitter in time for viewers to see telefilms of the Coronation. Initially, it will use a single-stack aerial, but when erected on the permanent site it will employ a 12 -stack array, which, it is claimed, will increase the radiated power to about 55 kW .
Pye's are to supply complete equipment for two studios for the Belgian television service operated by the Institut National Belge de Radiodiffusion. The Belgian service is planned to operate on both 819 and 625 lines and an outstanding feature of the Pye equipment is that by a simple switch the necessary changeover can be made.

## Television Costs

THE CAPITAL COST of each of the five permanent low-powered tejevision stations scheduled for Plymouth, Isle of Wight, Pontop Pike, Belfast and Aberdeen, was given by the Assistant Postmaster-General as of the order of $£ 150,000-£ 200,000$. In reply to a further question in the House he stated that the booster station at Brighton, which was "sanctioned as an experimental station and
is being constructed as such," would cost under $£ 10,000$.

The G.P.O. link between London and the Isle of Wight, which will probably be by cable, is estimated to cost about $£ 125,000$, while the radio relays linking Plymouth with Wenvoe and $\Lambda$ berdeen with Kirk o'Shotts will each cost approximately $£ 200,000$.

Figures were not given for the link to Pontop Pike-it is, of course, on the existing radio chain between Manchester and Kirk o'Shotts. The temporary Belfast station will rely on direct radio reception of the Kirk o'Shotts transmitter for rebroadcasting; the permanent link between Northern Ireland and the mainland nas not yet been approved.

The Assistant P.M.G. stated it was estimated it would take approximately two years to construct each of the five permanent low-powered stations from the time resources are made fully available. Since that statement was made, the B.B.C. has ordered from Marconi's the $5-\mathrm{kW}$ vision and $2-\mathrm{kW}$ sound transmitters for three of the stations.

## Licensing Set Manufacturers

NEW PATENT LICENCES are now available to manufacturers of receivers from the Broadcast Licensing Pool (F. C. Topham, Marconi House, Strand, London, W.C.2).

The Grantors who contribute their present and future patents for the period December 1st, 1953, to December 31st, 1957, are: B.T-H., Cintel, E.M.I., G.E.C., Marconi's W.T., Murphy, Philips, Pye and Standard Telephones.
The two licences which will be issued are:-
Licence A7-T, which will relate to broadcast television receivers and all combined instruments incorporating television receivers. The basic royalty on television receivers will be 3 per cent of the manufacturer's net turnover in such apparatus, with, however, a reduction in the cases of combined instruments and export.
Licence A7-S, which will relate to broadcast sound-only receivers and radio-gramophones. The royalty proposed for the present is 0.5 per cent of the net turnover, with a reduction for export.

## PERSONALITIES

H. L. Haslegrave, M.A., Ph.D., M.Sc.(Eng.), M.I.E.E., at present principal of the Leicester College of Technology, has been appointed principal of Loughborough Institute of Technology. Dr. Haslegrave, who was from 1938 1943 head of St. Helen's Municipal Technical College, will be the first principal of the Loughborough I.T., which was the title given to the Engineering and Science Departments of Loughborough College when they were recently reorganized by the Ministry of Education as a separate establishment.

Guy R. Fountain, chairman of the Cannoy group of companies, who is on a two-months' tour of America and the

West Indies, will also visit Canada, where he plans the formation of a Canadian Tannoy company.

Robert Telford, B.A., A.M.I.E.E., has been appointed to the new post of general works manager of Marconi's Wireless Telegraph Co. He graduated at Cambridge in 1937 and the same year joined Marconi's. Two years later he became assistant to the works manager. For four years from 1946 Mr. Telford was managing director of the subsidiary Companhia Marconi Brasileira in Rio de Janeiro and in 1950 became assistant to the general manager at Chelmsford.
C. G. White, director and general manager of the Marine Division of Kelvin $\&$ Hughes, Ltd., has been reelected vice-chairman of the R.C.E.E.A.
L. W. D. Sharp, M.A. (Cantab), A.M.I.E.E., who joined the Plessey Company in 1948, has been appointed chief engineer, Components Division. For the past two years he has been chief radio engineer, Telecommunications Division, engaged in the development of mobile radio equipment. Prior to joining Plessey's, Mr. Sharp was with E. K. Cole, Ltd., where he was for some time responsible for the development of car radio equipment.

J. P. Wykes, A.M.I.E.E., who joined the Marconi Marine Company as a seagoing wireless operator in 1918, will continue as works manager of the Marconi Works at Chelmsford. During the war he was manager of the crystal manufacturing department and in 1946 became chief of the Test Division.
E. B. Greenwood, A.M.I.E.E., has been appointed works manager of Marconi's factory now being built at Basildon New Town, Essex. A graduate of Leeds University, he held appointments with Murphy Radio, Ltd., and E. K. Cole, Ltd., before joining the Marconi Company in 1951.

Dr. Vladimir K. Zworykin's "outstanding contribution to the concept and design of electronic components and systems" was cited by the American I.E.E. in awarding him the 1952 Edison Medal. Now vice-president and technical consultant of the R.C.A. Laboratories at Princeton, N.J., he has been associated with R.C.A. since 1929 and was a pioneer in the development of "all-electronic". television, inventing the iconoscope in 1923. Dr. Zworykin, who was born in Moscow in 1899, has been domiciled in the U.S.A. since 1920.

William Davies, M.B.E., who was the first official sea-going wireless operator having joined the Marconi Marine Company in 1902, has been awarded the Marconi Memorial Medal of Service by the Veteran Wireless Operators' Association of America.
T. E. Goldup, M.I.E.E., has been elected chairman of the Radio Communication \& Electronic Engineering Association for 1953 in succession to K. S. Davies. Mr. Goldup, who is a director of Mullard, Ltd., which he oined in 1923, is a vice-president of the Institution of Electrical Engineers and chairman of the governors of the Ministry of Supply School of Electronics, Malvern.

Leslie Cooper, G5LC, the new president of the Radio Society of Great Britain (see last issue), is works manager of the Phoenix Telephone and Electric Works, Ltd., manufacturers of telecommunications equipment, and of its associate companies which include Correx Communication (1949), Ltd.

## OUR AUTHORS

T. C. Macnamara, contributor of the article in this issue on the use of electronics in film making, has been technical director of High-Definition Films, Ltd., since it was formed in 1951 to develop the use of electronic apparatus in the film industry. From 1923 to 1950 he was with the B.B.C. and was largely concerned with the establishing of such transmitting stations as Droitwich, Ottringham and Alexandra Palace. During his last three years with the Corporation he was head of the Planning and Installation Department. On leaving the B.B.C. he joined ScophonyBaird, from which the resigned a year ago to devote himself exclusively to development work with High-Definition Films, Ltd.
P. J. Harvey, who contributes the article on television booster stations in this issue, graduated B.Sc., A.C.G.I., in 1941 and from 1942-47 served as signals officer in the R.A.F. He subsequently joined E. K. Cole, Ltd., specializing in the development of communication equipment and was largely responsible for the design of the prototvpe television booster referred to in the article.
A. Poliakoff, managing director of Multitone, Ltd. which he joined in 1931 soon after the formation of the company, contributes an article on hearing aids in this issue (p. 182). He studied physics at University College, London, and is a member of the B.S.I. Committee on hearing aids and audiometers.

## OBITUARY

It is with regret that we record the death of John Satchell Smith on February 21 st at the age of 64 . Until his retirement in 1950 he was manager, North West Area, for the Marconi International Marine Communication Co., which he joined in 1910. He was at one time Marconi Marine representative in New York.

We record with regret the death of James H. Webb, principal of the British School of Telegraphy, where wireless operators have been trained since 1906. During the two world wars radio instruction was given under Mr. Webb's supervision to large numbers of men for the fighting Services. Training of students for the Postmaster-General's certificates is to continue under the present staff of instructors.

## IN BRIEF

Dry Batteries have been reduced in price by about 10 per cent. In announcing this in the middle of March, the Association of Radio Battery Manufacturers stated that reductions in the cost of raw materials, notably zinc, have made this possible. A 120 -volt battery is now 17 s 1d (inc. p.t.) instead of 19s 1 d .

Medium-wave Coverage of Germany by the B.B.C. has been improved by the use of the high-power transmitter at Osterloog, near Norden, operated by the Nordwestdeutsche Rundfunk. This transmitter, which is now being used for the B.B.C. European Service instead of the station at Ottringham, Yorks, is radiating on $1,295 \mathrm{k}=/ \mathrm{s}(231.7 \mathrm{~m})$. The programmes are fed by line from London.

Television Test Transmissions from B.B.C. stations have been extended by an hour each week-day until June lst. The transmissions during the third hour (1200-1300) consist of Test Card $C$ with $440 \mathrm{c} / \mathrm{s}$ tone on the sound channel.

Fortieth Anniversary of the founding of the Radio Society of Great Britain on July 5th, 1913 (then known as the London Wireless Club), is to be celebrated by holding a dinner in London later in the year.

Retailers' Conference.-The first national conference of retailers, organized by the Radio and Television Retailers' Association, will be held at Eastbourne from April 27th to 29th. Particulars of the conference are obtainable from the Secretary, R.T.R.A., 26, Fitzroy Square, London, W.l.

Ferguson.-A series of technical discussions on Ferguson television receivers has been arranged by Thorn Electrical Industries, Ltd. The fourth in the series will be held at the Royal Hotel, Dundee, on April 7th and 8th, and the fifth at the Kensington Hotel, Belfast, on April 8th and 9th. Sessions are from 2.0-5.0 and 7.0-10.0 each day. Admission is by invitation, and those wishing to receive invitations should apply to Ferguson Sales Division, Thorn Electrical Industries, Ltd., 233, Shaftesbury Avenue, London, W.C.2.

Decca Radar was used to check the series of timed runs over the Newbiggin measured mile when heavy mist prevented a visual check of the markers being made during the acceptance trials of the m.s. Camellia.
B.R.E.M.A. - The following 12 mem-ber-firms forming the executive council of the British Radio Equipment Manufacturers' Association were re-elected at the recent annual general meeting (the representative's name is in brackets): A. J. Balcombe (E. K. Balcombe); Bush Radio (G. Darnley-Smith); E. K. Cole (G. W. Godfrey); A. C. Cossor (H. Roberts); G.E.C. (M. M. Macqueen); Gramophone Co. (F. W. Perks); Kolster-Brandes (P. H. Spagnoletti); McMichael Radio (C. G. Allen); Philips (A. L. Sutherland); Pilot Radio (H. L. Levy); Pye (C. O. Stanley) and Ultra Electric (E. E. Rosen). P. H. Spagnoletti and E. K. Balcombe were re-elected chairman and vice-chairman, respectively.

Communication Theory.-A week's course of lectures on the theory of communication has been organized by D. A. Bell (Reader in Electromagnetism at Birmingham University) from July 20th25 th at the Centre for Continued Studies, Primrose Hill, Selly Oak, Birmingham, 29. A syllabus of the course, for which the fee is £3, excluding meals and accommodation, is obtainable from the director of Extra-Mural Studies, The University, Birmingham, 3.

## EXHIBITION NEWS

American Buyers are chartering a plane to bring them to this country for the R.E.C.M.F. exhibition which opens at Grosvenor House, London, W.1, on April 14th. The party is being organized by Leonard Carduner, president of the British Industries Corporation, of New York, in co-operation with W. T. Ash, secretary of R.E.C.M.F. Visits to a number of radio factories, including Garrard, Multicore and G.E.C., are


WHALING BUOY. Lightweight version of Venner's standard radio-transmitting dahn buoy (see Nov. 1952 issue, p. 468) for marking whales after the kill to facilitate recovery by the parent ship. It has a range of 30 miles and radiales for 22 hours.
being arranged for overseas visitors to the exhibition.
B.I.F.-For the first time since the war all three sections of the British Industries Fair (normally restricted to trade buyers) will be open to the public every afternoon and all day on Saturday. The fair opens at Castle Bromwich (Birmingham) and at Earls Court and Olympia (London) on April 27th. Each section will be open daily from 9.30 to 6.0 until May 8th.

Electronics Exhibition.-The eighth annual electronics show organized by the North-Western Branch of the Institution of Electronics will be held at the College of Technology, Sackville Street, Manchester, from July 15 th to 21 st. In addition to the usual commercial section there will be a section devoted to exhibits from research establishments. One portion of the research section will cover the electro-medical applications of electronics. The organizing secretary is W. Birtwistle, 17 , Blackwater Street, Rochdale, Lancs.

Navigation,-An exhibition, "Navigation Today," is being held at the Science Museum, South Kensington, London, S.W.7, from March 31st until the middle of September. The navigator's problems and some of the methods and equipment used by mariners and airmen to solve them are being shown.

Radio and Models Show.-The London Radio Group of the Institution of Post Office Electrical Engineers will be holding its second radio and models exhibition in the Metropole Hall, Northumberland Avenue, London, W.C.2, on May 7 th, 8 th and 9 th, from 11.0 to 8.0. Although there will be some commercial exhibits, the emphasis will be on home construction.

## PUBLICATIONS

New Audio Journal.-Thirty papers presented at the last annual meeting are printed in the first issue of the fournal of the Audio Engineering Society, which appeared in March. There will be four issues in a year, and the subscription rate to non-members is $\$ 8.00$. The editor is Lewis S. Goodfriend, P.O. Box 12, Old Chelsea Station, New York 11 , U.S.A.
"Television Oscilloscope."-A reprint is now available of the articles by $W$. Tusting on the design of a television oscilloscope using a 5 -in cathode-ray tube, which appeared in the June and July, 1952, issues of Wireless World. "Television Oscilloscope" is obtainable from our Publishers price 9d (postage 1 $\frac{1}{2}$ d).

Interference Suppression.-A " Code of Practice" in draft form on the general aspects of radio interference suppression has been issued by a joint committee of the I.E.E. and the British Standards Institution. Copies of the code, CO(ELE)6125, which is mainly for the guidance of those installing and using interference-producing devices, may be had from B.S.I., 24, Victoria Street, London, S.W.I, at 2 s .

British Standard for synthetic-resinbonded paper tubes of circular crosssection for use at r.f. has been issued as B.S.1951:1953. The electrical and physical properties covered by the specification include the power factor and permittivity; insulation resistance along laminæ; density; water absorption; cohesive strength between layers and
dimensional tolerances. Copies of the Standard are obtainable from the B.S.I., Sales Branch, 24, Victoria Street, London, S.W.1, price 2 s 6 d .

Our Raw Material.-Each of the 70 domestic, commercial, industrial, farm and miscellaneous tariffs for electricity supplied by the various Area Boards of the British Electricity Authority are given in the "Electricity Tariff Handbook" published by Electrical Review Publications, Ltd., price 7/6.

## BUSINESS NOTES

African Broadcasting.-Transmitting and studio equipment for the extension of the broadcasting services in the Colonies has been ordered by the Crown Agents for the Colonies from Marconi's W.T. Co. A 20-kW h.f. transmitter, to supplement the two existing low-power stations, together with four complete studios, is to be installed at Accra, Gold Coast. H.F, transmitters have also been ordered for Lagos ( 20 kW ), Enugu ( $2 \frac{1}{2} \mathrm{~kW}$ ) and Kaduna ( $7 \frac{1}{2} \mathrm{~kW}$ ) in Nigeria; Dar-es-Salaam (20kW) in Tanganyika; and a $7 \frac{1}{2}-\mathrm{kW}$ station for Uganda. Studio and o.b. equipment is to be installed in Nairobi, Kenya.
Large-screen television projectors have been supplied by CinemaTelevision, Ltd., to the German cincequipment manufacturers U.F.A. for installation experimentally in a cinema in Düsseldorf. Standard equipment, which has been or is being installed in eight cinemas in this country, will be used. In this country, of course, it operates on 405 lines and the change to 625 lines (the standard adopted by Germany) can be made "in less than a minute."

Ten Tankers-16,000-ton motor ves-sels-under construction on the Clyde, Wear and Tyne for the Lowland Tanker Co., are to be equipped with radio communication and navigational gear by the Marconi International Marine Communication Co. Eight new tankers being built for the British Tanker Co., the largest of which is 32,000 tons, are also to be fitted with Marconi radio equipment. The company is also installing communication and navigational gear on the 7,000-ton Norwegian passenger steamer Leda on behalf of its associated company Norsk Marconikompani.

Ekco Electronics, Ltd., has been registered as a subsidiary of E. K. Cole, Ltd., for the marketing of the company's electronic equipment.

Telephone number of Marris \& Cartin, Ltd., 42, Brook Street, London, W.1, was misquoted in the advertisement on page 139 of our February issue. It should be Grosvenor 5571.

Lancs., not Yorks.- The address of Holiday and Hemmerdinger, Ltd., was incorrectly given as Leeds, instead of Manchester, on p. 118 of our March issue.

## MEETINGS

## Institution of Electrical Engineers

Radio Section-"Special Effects for Television Studio Productions" by A. M. Spooner, B.Sc.(Eng.), and T. Worswick, M.Sc., on April 9th.

Discussion on "The Relative Merits of Broad-Band Transmission by Beam, Cable and Waveguide," opener: E. C. H. Organ, O.B.E., on April 13 th .
"An Investigation of the Characteristics of Cylindrical Surface Waves" by Professor H. M. Barlow, Ph.D., B.Sc. (Eng.), and A. E. Karbowiak; and "Surface Waves" by Professor H. M. Barlow, Ph.D., B.Sc.(Eng.), and A. L. Cullen, Ph.D., B.Sc.(Eng.), on April 22nd. Measurements Section. " Digital Computers at Manchester University" by T. Kilburn, M.A., Ph.D., G. C. Tootill, M.A., M.Sc., D. B. G. Edwards, M.Sc., and B. W. Pollard, M.A.; "The Construction and Operation of the Manchester University Computer" by B. W. Pollard, M.A., and K, Lonsdale, B.Sc. Tech.; "Universal High-Speed Digital Computers: A Decimal Storage System " by T. Kilburn, M.A., Ph.D., and G. Ord, M.Sc.; and "Recent Advances in Cathode-Ray Tube Storage" by Professor F. C. Williams, O.B.E., D.Sc., D.Phil., F.R.S., T. Kilburn, M.A, Ph.D., G. N. W. Litting, B.Sc., D. B. G. Edwards, M.Sc., and G. R. Hoffman, B.Sc., on April 14th.

All the above meetings will be held at 5.30 at Savoy Place, London, W.C.2.

North-Western Centre.-"The Lon-don-Birmingham Television Cable System " by T. Kilvington, B.Sc.(Eng.), F. J. M. Laver and H. Stanesby, at 6.15 on April 14th at the Engincers' Club, Albert Square, Manchester.
South Midland Centre.-." The Devising of Examination Questions" by Professor G. W. Carter, M.A., at 7.0 on April 9th at the College of Technology, Electrical Engineering Dept., Birmingham.
"Electronic Telephone Exchanges" by T. H. Fiowers, M.B.E., B.Sc., at 7.15 on April 21st at the Winter Gardens Restaurant, Great Malvern.

South Midland Radio Group.-The Transmission of Pictures by Radio," by A. W. Cole and I. A. Smale, B.Sc, at 6.0 on April 27 th at the James Watt Memorial Institute, Great Charles Street, Birmingham.

Southern Centre.-"Television Broadcasting Stations," by P. A. T. Bevan, B.Sc., at 6.30 on April 22 nd at the South Dorset Technical College, Weymouth.

London Students' Section.-Visit to the B.B.C. Monitoring Station, Tatsfield, Surrey, at 2.30 on April 25 th.

## British Institution of Radio Engineers

London Section.-" Lens Aerials for Centimetric Wavelengths" by Lt.-Col. J. P. A. Martindale, B.A., at 6.30 on April 8th at the London School of Hygiene and Tropical Medicine, Keppel Street, London, W.C.I.

Scottish Section.-" The Principle and Applications of the Telescribe " by C. A. Gilbert at 7.0 on April 2nd at the Department of Natural Philosophy, Edinburgh University.
"Remote Control Devices and Servo Mechanisms" by A. E. W. Hibbitt at 7.0 on April 9 th at the Institute of Engineers and Shipbuilders, Glasgow.

North-Eastern Section. -. Annual General Meeting followed by a demonstration of stereophonic reproduction at 6.0 on April 8th at the Institute of Mining and Mechancial Engineers, Neville Hall, Westgate Road, Newcastle-onTyne.

## British Sound Recording Association

London.-"Sound Reproduction" by D. T. N. Williamson at 7.0 on April 17th at the Royal Society of Arts, John Adam Street, London, W.C.2.

Manchester Centre.-"Practicalities of Recording" by G. F. Budden (B.B.C.) at 7.30 on April 27th at the Engineers' Club, Albert Square, Manchester.

## Institution of Electronics

North-Western Branch.-"Present Technique of Colour Television" by J. A. Darbyshire, M.Sc., Ph.D., F.Inst.P., A.M.I.E.E. (Ferranti, Ltd.), at 7.0 on April 24 th in the Reynolds Hall, College of Technology, Manchester.

## Institute of Physics

Symposium on "Practical and Theoretical Aspects of Ultrasonic Testing" at 6.30 on April 17 th at 47, Belgrave Square, London, S.W.1.

## Institute of Navigation

"The Use of Radar for Preventing Collisions at Sea " by Capt. F. J. Wylie, R.N. (Ret.), at 5.0 on April 17 th at the Royal Geographical Society, 1, Kensington Gore, London, S.W.7.

## hadio dnit for hearing aids

MENTION WAS MADE in our March issue of a home-made detector unit for converting a hearing aid into a broadcast receiver. We now learn that a unit for this purpose is manufactured commercially by Acousticon, and that the hearing aids made by that firm are now fitted with sockets for connection of the adaptor.

The "Radion" unit, as it is called, comprises a B.T.H. germanium diode and a coil with dust-iron core, adjustment of which provides tuning control over the range $550-1,500 \mathrm{kc} / \mathrm{s}$. A short length of flexible wire plugged into the unit serves as an aerial at very short ranges. For greater signal pick-up, either a "random" or normal full-length aerial may be connected.

Though the tuner is only singlecircuit, selectivity is greater than that of the old-fashioned crystal set, and is sufficient for all except the least favourable conditions.


## Vision A.G.C.

Stabilizing the Picture Against<br>Signal Fading<br>and Aeroplane "Flutter"



HE problem of applying automatic gain control to the vision channel of a television receiver is not quite as simple as it may seem at first. Obviously the broad principle is the same as in a sound broadcast receiver -just negative feedback-but the difficulty comes in presenting the a.g.c. circuit with the right kind of information on which to work.

The main trouble lies in the peculiar nature of the television signal. In sound broadcasting the mean value of the carrier wave remains the same whatever the modulation. It only varies with the signal strength, so the a.g.c. circuit can use it for deriving its control voltage. But in television the mean value of the carrier wave is changing continually with the nature of the picture, so it never gives a really true indication of the signal strength. It can be used to give some measure of a.g.c., which is a good deal better than none at all, and in the October, 1952, issue G.

The Pye model $V 4$ receiver incorporating vision a.g.c. It has flywheel sync and an inverter for suppressing interference pulses. Both the 14 -in c.r. tube screen and the implosion guard are tinted, and the screen is tilted to face downwards away from the room lighting. The set has continuous tuning covering all the B.B.C. television stations.

Johnson described a simple type of circuit working on this principle. As this was, in effect, a negative feedback system stabilizing the mean value of the signal, it tended to keep the average brightness of the picture constant; consequently the contrast and black level were not always maintained as they should be.
There is, however, one part of the television signal which does give a true indication of the signal strength, and that is the black level, represented in the waveform by the so-called "back porch" following each line sync pulse. To make use of this it would

Fig. I. Simplified circuit diagram showing the a.g.c. chain and how it is connected into the receiver circuit.

be necessary for the a.g.c. circuit to sample the signal at the correct instants and use these black-level samples to produce the control voltage. In adjusting the gain of the receiver to give correct contrast, its effect would be to maintain constant not the mean brightness of the picture, but the black level, which, of course, is just what is wanted.

This scheme has now actually been introduced into a commercial receiver, the Pye model V4, under the name of "automatic picture control." It does not, as one might suppose from this title, endow the set with any powers of æsthetic discrimination, but it will maintain constant the brilliance and contrast of the pictures, whatever they may be, against signal variations of up to 40 db . This 40 db represents the total range, so how the system behaves depends on where the mean signal voltage lies between the absolute upper and lower limits. If it comes in the middle, the picture will be stabilized with signal variations of up to $\pm 20 \mathrm{db}$; if it is a lower voltage the signal can increase more than 20 db ; if it is higher the signal can decrease inore than 20 db .

No doubt the system will be of most value in fringe areas, where fading is more troublesome, but it is also quite effective against the "flutter" caused by passing aircraft, which can be bad in any area. It is sufficiently quick-acting to counteract this flutter up to a frequency of about $50 \mathrm{c} / \mathrm{s}$, and anything above this is not perceptible to the eye anyway.

For the purpose of sampling the black level of the signal, a large pulse is derived from the flyback of the line scan waveform. This is done by means of a small pulse transformer, as shown in Fig. 1, fed from a winding on the line scan output transformer. Since the pulse comes from the flyback it occurs at the same time as the line sync pulse in the received signal, so it has to be delayed a few microseconds to make it coincide with the back porch following the sync pulse. The delaying is done by the RC network connected to the transformer secondary.

This sampling pulse, which is negative-going, passes through $\mathrm{C}_{1}$ te the cathode of the diode $\mathrm{V}_{3}$ and causes it to conduct. As a result the diode acts as an electronic switch and connects the video output of the cathode follower, $\mathrm{V}_{2}$, to the cathode of $\mathrm{V}_{4}$ during the black level period. The video output signal is inverted (the sync pulses being positive with respect to the picture) so the situation that occurs can be depicted as in Fig. 2. The sampling pulses coming through $\mathrm{C}_{1}$ are d.c. restored by $\mathrm{V}_{3}$ to the d.c. value of the black level in the video output waveform. In other words, what would normally be negative peaks in an a.c. waveform are now held at an absolute positive potential of about 100 V . And, of course, when


Fig. 2. Showing how the sampling pulses are d.c. restored to the black level patential of the inverted video waveform from the cathode follower.
the black level potential fluctuates with changes in signal strength the sampling pulses ride up and down with it.

Thus the absolute positive potential of the sampling pulse peaks now becomes a true indication of the signal strength, and the pulses can be used to produce a proportional negative bias voltage for controlling the gain of the receiver. They are actually rectified by the diode $\mathrm{V}_{5}$ and smoothed by the associated RC network, the resultant bias voltage being applied to the grid of the first i.f. amplifier, $V_{1}$, as shown.

Before this happens, however, the pulses are amplified in $\mathrm{V}_{4}$. The cathode of this valve therefore receives a series of negative pulses d.c. restored to a positive potential representing the black level-in other words it receives a positive potential with troughs in it. The grid of $V_{4}$ also has a positive voltage applied to it, as bias, by the potentiometer $\mathrm{R}_{2}$, but this is somewhat lower than the cathode potential so the valve is normally cut off until a pulse comes along and drives the cathode negatively; then current flows in the valve and a corresponding negative pulse appears at the anode. Thus the setting of $\mathrm{R}_{2}$-represented by the chain line in Fig. 2-determines the input voltage below which the pulse is amplified. In other words it determines the amplitude of the pulse at the anode, hence the value of the negative bias voltage and hence the gain of the i.f. amplifier for a given signal strength. So $\mathrm{R}_{2}$ acts as a manual gain (or contrast) control quite independently of the automatic gain control.

Thus when the signal strength falls the black level in the video output waveform goes more positive, the sampling pulse is d.c. restored to a higher positive potential, a smaller amount of pulse is amplified by $V_{4}$ and the resultant negative bias control voltage is reduced. The gain of the first i.f. stage is thereby increased and the video output restored to its original amplitude. Similarly, if the signal strength rises; the black level positive potential falls, more of the sampling pulse is amplified and the negative control voltage increases so that the gain is reduced.

One valuable feature of the system is that it is not affected by interference pulses which may occur during the black level period. Since the interference pulse will be negative-going on the video output waveform, the diode $V_{3}$ acts in its capacity as an electronic switch and cuts off, the time constant of $C_{1} R_{1}$ being too long to allow the cathode to fall to a comparable potential during this brief period. Thus the sampling pulse applied to $\mathrm{V}_{3}$ is not d.c. restored to the lower positive potential produced by the interference pulse-it remains more or less at the true black level. In other words the system simply "misses a beat" and the resultant negative control voltage is not affected.

Of course, since the object of the system is to keep the picture stable, one is not aware from the screen that anything is happening when the signal strength varies. The effect of the system is very obvious, however, when the V4 receiver is compared with another set which is fed from the same aerial but has no a.g.c. If changes in signal strength are introduced by a variable attenuator in the common aerial lead, the picture on the ordinary set can be varied from a blacked-out condition to a grossly overloaded condition, but the picture on the V4 receiver is not noticeably affected. One is only aware of what is happening by watching the noise level, which, of course, becomes rather more noticeable at the lower signal levels.

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3-Earthed-emitter and Earthed-collector<br>Circuits as Amplifiers and Oscillators

ANY HISTORICALLY MINDED young engineer might find it interesting to search through the literature to see how long it was before the users of valves tried unusual connections. Terman gives a 1928 reference for what I prefer to call a diminisher circuit, with anode input and grid output, no reference for the cathode follower (would this be about 1936?) and 1940 for the earthed grid. All the early progress, in fact, seems to have been made by shouting for better valves. In the transistor circuit field there has been no such delay. The circuit designers have applied themselves from the beginning to a study of almost every conceivable way of connection and interconnection of the transistor electrodes. In the previous article we considered the equivalent circuit for the fundamental earthed-base connection of the transistor. We can, of course, equally well connect either emitter or collector to earth, and in particular we shall often find it useful to apply the input to the base and take the output from the collector in the earthed-emitter circuit. It is necessary, therefore, to make a fairly close examination of the equivalent network which can be used.

The circuit shown in Fig 1(a) is the one discussed in last month's article and is for the earthed-base connection, but the transistor part of it, shown in Fig. 1(b) has no earth connection inherent in the equivalent circuit, so there is no reason why it should not be twisted round. This is quite an important matter, and really deserves to be reached by a much longer and more ponderous discussion. The transistor equivalent network made up of the four resistance elements $r_{e} r_{c} r_{b}$ and $r_{m}$ is a true equivalent for threeterminal working, not an approximation for special conditions of use like the ordinary valve equivalent.

Of course, it is still an approximation in another sense, because the $r$ 's are not constants, but are

dependent on the actual currents in the elements of the transistor. We know how to deal with this, though, because we have the same problem with ordinary valve circuits. I suspect that we sometimes exaggerate the importance of the non-linearities when we are being theoretical. Our amplifiers are usually worked in regions where the non-linearity is not more than 5 per cent, which should not be enough to complicate the theory, and circuits in which the non-linearity is used, like multivibrators, are designed on quite a different principle. For sine-wave operation we can say that the network of Fig. 1(b) is a complete and accurate equivalent for the point transistor.

We can immediately apply this equivalent to study the earthed-emitter amplifier shown in Fig. 2.* The basic equations are

$$
\begin{aligned}
& i_{1}\left(\mathbf{R}_{a}+r_{e}+r_{b}\right)+i_{2} r_{e}=v_{\theta} \\
& i_{1}\left(r_{e}-r_{m}\right)+i_{2}\left(\mathbf{R}_{\mathrm{L}}+r_{e}+r_{c}-r_{m}\right)=v_{\mathrm{L}}
\end{aligned}
$$

In deriving this it must be remembered that the current which appears in the generator $r_{m} i_{e}$ is here $-\left(i_{1}+i_{2}\right)$. Just as we did at the end of the previous article, we can take $v_{\mathrm{L}}=0$ or $v_{\mathrm{G}}=0$ and solve to find the input and output impedances. The results are:

Input impedance

$$
\mathbf{R}_{11}=r_{b}+r_{e}+\frac{r_{e}\left(r_{m}-r_{e}\right)}{\mathbf{R}_{\mathrm{L}}+r_{e}+r_{e}-r_{m}}
$$

Output impedance

$$
\mathbf{R}_{22}=r_{c}+r_{e}-r_{m b}+\frac{r_{e}\left(r_{m}-r_{e}\right)}{\mathbf{R}_{0}+r_{b}+r_{e}}
$$

As before, just as you would expect, the input impedance includes a term in the load, $\mathrm{R}_{\mathrm{L}}$ and the

* Based on Fig. 10, p. 376 of "Some Circuit Aspects of the Transistor "by R. M. Ryder and R. J. Kircher, B.S.T.7. Vol. XXVIM, July 1949.

Fig. 2. Earthed-emitter amplifier (a), and its equivalent circuit (b).

output impedance includes a term containing $\mathbf{R}_{0}$.
The condition for stability is that the total impedance in either the input loop or the output loop should be positive. It does not matter which we take, because, as you can check for yourself, the same answer is obtained. Choosing $R_{22}+R_{L}>0$ we have

$$
\begin{aligned}
& \quad\left(\mathrm{R}_{\mathrm{L}}+r_{c}+r_{e}-r_{m}\right)+\frac{r_{e}\left(r_{m}-r_{e}\right)}{\mathrm{R}_{\mathrm{G}}+r_{b}+r_{e}}>0 \\
& \text { or, }\left(\mathrm{R}_{\mathrm{L}}+r_{c}+r_{e}-r_{m}\right)\left(\mathrm{R}_{\mathrm{G}}+r_{b}+r_{e}\right)+r_{e}\left(r_{m}-r_{e}\right) \\
& =\Delta \xrightarrow{=}>0
\end{aligned}
$$

If this way of using the transistor is unstable it must be because ( $\mathrm{R}_{\mathrm{L}}+r_{c}+r_{e}-r_{m}$ ) is negative, in the first place, and because the second bracket, the one containing $\mathrm{R}_{6}$, is too large. This, of course, depends on the known smallness of $r_{e}$ compared with $r_{m}$. Forgetting $r_{e}$ and $r_{b}$ wherever possible, the stability depends on

$$
\left(\mathrm{R}_{\mathrm{L}}+r_{c}-r_{m}\right) \mathrm{R}_{0}+r_{e} r_{m}
$$

This very approximate expression shows how a sufficiently large $\mathrm{R}_{\mathrm{L}}$ will make the circuit quite safe. Very often we shall find that we are trying to work with $\mathrm{R}_{\mathrm{L}}=r_{c}$ and that $r_{m}$ is approximately equal to $2 r_{c}$, so that ( $\mathrm{R}_{\mathrm{L}}+r_{c}-r_{m}$ ) is zero. Everything is then satisfactory at low levels, but with a large input signal the transistor swings into a region where $r_{0}$ is reduced and $r_{m}$ increased. If this happens we may get momentary instability.
We can manipulate our two equations to give the forward power gain $\mathrm{G}_{\mathrm{F}}=4 \mathrm{R}_{6} \mathrm{R}_{\mathrm{L}}\left(\frac{r_{m}-r_{e}}{\Delta}\right)^{2}$
where $\Delta$ is the long expression given earlier, which must be positive for stability.
In this arrangement it is sometimes of interest to use the circuit backwards, with the input applied to



Fig. 3. Earthed-collector amplifier (a) and its equivalent circuit (b).

Fig. 4. This circuit oscillates because the anti-resonant impedance in the base lead supplies positive feedback. The theoretical circuit (a) requires supply connections as shown
in (b).

(a)

(b)
the collector and the output taken from the base. The backward power gain can be obtained by manipulating the two basic equations. It is

$$
\mathrm{G}_{\mathrm{R}}=4 \mathrm{R}_{\mathrm{a}} \mathrm{R}_{\mathrm{L}} \cdot\left(\frac{r_{e}}{\Delta}\right)^{2}
$$

Not surprisingly, this way of using the transistor gives more gain than the earthed-base connection. The reason is that, like it or not, you have positive feedback provided by $R_{G}$, which is in the base lead. An advantage of this connection, however, is that local negative feedback can be provided by adding resistance in the emitter lead, just like the cathode resistor in an ordinary valve circuit. In many ways, indeed, the problems resemble those of a highfrequency tuned-grid tuned-anode amplifier, but this analogy must not be pushed too far.

The third way of operating the point transistor is with the collector connected to earth. The circuit and its equivalent are shown in Fig. $3 \dagger$ and the equations are

$$
\begin{aligned}
& i_{1}\left(\mathrm{R}_{\mathrm{G}}+r_{b}+r_{c}\right)+i_{2}\left(r_{c}-r_{m}\right)=v_{\mathrm{a}} \\
& \quad i_{1} r_{c}+i_{2}\left(\mathrm{R}_{\mathrm{L}}+r_{e}+r_{c}-r_{m}\right)=v_{\mathrm{L}}
\end{aligned}
$$

Performing all the algebraic operations as before, we find that for stability we need:

$$
\left(\mathrm{R}_{\mathrm{o}}+r_{b}+r_{c}\right)\left(\mathrm{R}_{\mathrm{L}}+r_{e}+r_{c}-r_{m}\right)+r_{c}\left(r_{m}-r_{c}\right)
$$

The input impedance is

$$
\mathrm{R}_{11}=r_{b}+r_{c}+\frac{r_{c}\left(r_{m}-r_{c}\right)}{\mathrm{R}_{\mathrm{L}}+r_{e}+r_{c}-r_{m}}
$$

and the output impedance

$$
\mathrm{R}_{22}=r_{c}+r_{c}-r_{m}+\frac{r_{c}\left(r_{m}-r_{c}\right)}{\mathrm{R}_{a}+r_{b}+r_{c}}
$$

The power gain using the base as input and the emitter as output electrode is

$$
\mathrm{G}_{\mathrm{F}}=4 \mathrm{R}_{\mathrm{G}} \mathrm{R}_{\mathrm{L}}\left(\frac{-r_{c}}{\Delta}\right)^{2}
$$

and in the opposite direction

$$
\mathrm{G}_{\mathrm{R}}=4 \mathrm{R}_{\mathrm{a}} \mathrm{R}_{\mathrm{L}}\left(\frac{r_{m}-r_{c}}{\Delta}\right)^{2}
$$

The power loading at the emitter is usually quite a lot lower than the permitted collector input, and without having made any tests I would expect to find that reasonably low distortion would give a limit around 1 mW for this connection. For an average point transistor there is one very interesting feature of this arrangement: $r_{m}$ is about twice $r_{e}$ so that the two gains are roughly equal. This means that it is possible to make an amplifier which can be put in a telephone line and will amplify the signals in both directions. The gain of 6-12 decibels which can be obtained is considered well worth while by telephone engineers, especially as the microphone current is available to supply the amplifier. I expect to see amplifiers of this kind fitted in my telephone some day. The economic value lies in the saving of copper by the use of thinner wires from the telephone subscribers to the exchange.

With this classical background to the transistor, we can turn for the remainder of the article to some oscillator circuits. It is an interesting thing that oscillators represent in some ways the simplest and in other ways the most complicated of the circuit problems with which communication engineers are concerned. To make an oscillator work you need only produce an amplifier with a gain of rather more than

[^20]unity, put in a tuned circuit somewhere, and connect input to output. Apart from the necessity to watch that there isn't a 180 -degree phase shift, the circuit will oscillate. On the other hand, if you want an oscillator which is not affected by the supply voltages, valves, weather and the cost of living, the design problem is much more complicated, and even "Cathode Ray" has been known to express doubts. Transistors do not make good oscillator circuits, because they are rather lacking in gain, but it is certainly possible to get a stability of one part in a thousand for a wide range of supply voltage.
The special feature of the point transistor which makes some of the oscillator circuits so simple is the possibility of getting a negative resistance by choosing the external resistances properly. Let us take some simple numbers which we can use for calculation purposes. A hypothetical transistor, operating at low currents, might have :
\[

$$
\begin{array}{ll}
r_{e}=200 \text { ohms } & r_{b}=200 \text { ohms } \\
r_{c}=10,000 \text { ohms } & r_{m}=30,000 \mathrm{ohms}
\end{array}
$$
\]

We might choose to match the output at the collector, and make $\mathrm{R}_{\mathrm{L}}=10,000$ ohms. This should give the maximum power from our oscillator. Then if we use the carthed-emitter circuit, the input impedance will be, from the equation given earlier

$$
\begin{aligned}
\mathrm{R}_{11} & =200+200+\frac{200(30,000-200)}{20,200-30,000} \\
& \bumpeq 200+200-600=-200 \mathrm{ohms}
\end{aligned}
$$

This means that if we connect a circuit in the base lead which has an impedance of 200 ohms at one frequency, the system will be unstable. A normal parallel-tuned circuit could be used, with an impedance at the anti-resonance of, say 250 ohms. But if we have a $Q$ of 50 , which is not unreasonable for an oscillator frequency determining circuit, the reactance of the coil and the capacitor would have to be only 5 ohms at the working frequency. This means a very small inductance and a relatively enormous capacitance. Fortunately there is an casy way out of the difficulty. Let us put, say, 5,000 ohms in the lead between emitter and earth. This increases the effective value of $r_{G}$ to $5,200 \mathrm{ohms}$, and makes the input impedance at the base about - 25,000 ohms. This is just about the right sort of impedance magnitude for conventional tuned circuits of good Q . The circuit shown in Fig. 4(a), using a $100-\mathrm{mH}$ inductance with a Q of 50 at $800 \mathrm{c} / \mathrm{s}$ will just oscillate nicely with a good transistor and suitable supplies. The resistance in the emitter lead can be used as a regencration control. The modifications needed to connect the supplies are shown in Fig. 4(b).

There is an alternative form of this circuit shown in Fig. 5(a), which avoids this use of the emitter resistance by tapping across a small portion of a conventional tank circuit to get the 250 -ohm antiresonant impedance. It would, of course, be at $1 / 10$ th of the turns of our $100-\mathrm{mH}$ coil in the example chosen. I prefer, however, to get the impedancestabilizing effect of the emitter resistance. Another variant is obtained by moving the earth connection, as shown by the two circuits in Figs 5(b) and 5(c). These circuits seem to offer some difficulties in applying the proper bias, and I have not yet tried them.

A modified circuit of the type shown in Fig. 4(a), has been made to oscillate at frequencies up to $300 \mathrm{Mc} / \mathrm{s}$. I have used the word " modified," but the actual modification consisted of the use of a special


Fig. 5. Variants of the base-controlled oscillator circuit.


Fig. 6. Oscillator circuit (a) using negative emitter impedance when the base load resistance is high enough, and (b) method of connecting suppliss.
transistor, and the base-emitter capacitance plays an important part in the functioning of the circuit. This circuit does not, therefore, fall into the simple negative-resistance class we are discussing at the moment.
A second oscillator type is obtained by operating with an earthed-base connection, using the negative input resistance at the cmitter. As we saw in the previous article, the input resistance is

$$
r_{e}+r_{b}-\frac{r_{b}\left(r_{b}+r_{m}\right)}{\mathbf{R}_{\mathrm{L}}+r_{e}+r_{b}}
$$

Using the same numbers as before, this is

$$
\mathrm{R}_{11}=400-\frac{200(30,200)}{20,200} \text { or about } 100 \text { ohms. }
$$

This will not do, but if we put an 800 -ohm resistor in series with the base, we make the effective value of $r_{b}=1,000 \mathrm{ohms}$, and then

$$
\mathrm{R}_{11}=1200-\frac{1,000(31,000)}{21,000} \text { or about }-300 \text { ohms. }
$$

It is very easy to design a circuit which will be 300 ohms at one frequency, and a higher impedance at other frequencies. A coil having a total loss resistance of 300 ohms and a series capacitor is all we need. For a coil giving a Q of about 30 , we have $\omega \mathrm{L}=\mathrm{QR}$, or about 10,000 ohms. Choosing $798 \mathrm{c} / \mathrm{s}$ as the frequency, we find that a $2-\mathrm{H}$ inductor is needed: a thoroughly practical value is obtained. The circuit is basically that shown in Fig. 6(a), and needs only


(a)

Fig. 7. The use of collector negative impedance, as in this circuit is not satisfoctory if oppreciable output power is needed.

(b)

Fig. 8. (o) Shows a resonant circuit in the output (collector) loop of an earihed-emitter circuit. Re-drawn as in (b), with resistors added, it becomes on earthed-base amplifier with feedback from collector to emitter.
the additional resistors shown in Fig. 6(b) to provide the necessary supplies. This circuit can be controlled in amplitude by varying the resistor in the base lead. Theoretically it can use a thermistor in the base lead, so that increasing oscillation level reduces the regeneration. I say theoretically, because the immediate difficulty is obtaining a suitable thermistor. The base current will be only about 1 mA , so that the thermistor must work on powers of the order of 1 mW .

The two circuits described have the advantage that they are power oscillators: they leave the collector free to work with a normal load resistance, so that the full available output can be taken from the transistor. This output is enough for many jobs, and will always be enough to drive a push-pull transistor amplifier stage. There are two more interesting circuits which can be designed in the same way. One of these is shown in Fig. 7. It starts off from the equation for the output impedance of the earthed base circuit

$$
\mathbf{R}_{2 \underline{2}}=r_{c}+r_{b}-\frac{r_{b}\left(r_{b}+r_{m}\right)}{\mathbf{R}_{\mathrm{f}}+r_{e}+r_{b}}
$$

with $\mathbf{R}_{6}$ taken'as zero, and $r_{b}$ for convenience, increased to 1,000 ohms by adding an 800 -ohm resistor in the base lead. Then

$$
R_{22}=11,000-\frac{1,000(31,000)}{1,200}=-14,000 \mathrm{ohms}
$$

This sort of negative resistance is sometimes useful in designing crystal oscillator circuits: it is not very good for conventional LC working. Another disadvantage is that there is nowhere to put in a load circuit, unless it is in series with the controlling LC circuit, and this is not desirable as it degrades the $Q$ of the system and makes the frequency stability worse.

The final circuit in this class is obtained by considering the output impedance of the earthed-emitter circuit. We shall want to look at this in a rather different way later, but for the moment we can observe that the output impedance is

$$
\mathrm{R}_{22}=r_{c}+r_{e}-r_{m}+\frac{r_{e}\left(r_{m}-r_{e}\right)}{\mathrm{R}_{\mathrm{a}}+r_{b}+r_{b}}
$$

With no external resistance at all this reduces to

$$
\begin{aligned}
\mathrm{R}_{22} & =10,000+200-30,000+\frac{200(29,800)}{400} \\
& \text { or about }-5,000 \mathrm{ohms} .
\end{aligned}
$$

A circuit of the type, shown in Fig. 8, will therefore oscillate, provided that the resonant circuit drops to a resistance of less than 5,000 ohms at resonance. This is crystal magnitude again, of course, but as before we have nowhere to take out the load. When we use this circuit, which turns out to be a very satisfactory one, we must rearrange it in the way shown in Fig. 8(b). Then the negative impedance calculation is not so simple, and we adopt a more normal amplifier approach.

This treatment of the negative resistance oscillators may seem rather scanty, but we shall be returning to the question of negative resistance circuits in a later article. For the moment, I want just to show the sort of circuits which can be produced, and the lines along which the design process must operate. Conventional oscillators, in which we feed back from an output electrode to an input electrode, and amplifiers must be considered before we look again at the very interesting negative-resistance circuits which can be used for pulse production, standardization and manipulation.

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## Multi-Layer R.F. Coil-Winder

## Constructional Details of

 a Machine for MakingWave-zwound Coils

By B. V. NORTIIALL



WAVE-WOUND COILS are so extensively used that it is thought an account of the author's winding machine may be of interest. It is capable of winding coils up to $\frac{1}{2}$ in wide and can accommodate formers 2 in long. The machine is of simple construction, does not require any specially machined parts and is made largely from pieces of scrap. Few detailed measurements only are given in the following description as the majority can be varied according to the materials available.

The basic design is shown in Fig. 1. A variable friction drive is used in place of a gear wheel. Rotation of the cam slides the wire-feed finger to and fro, the amount of travel being varied by adjustment of the cam-follower mounting in the direction $a-a$. Gear changes are made by adjusting the drive roller in direction $b-b$. The wire is fed to the finger over pulleys which may be used to operate a brake acting on the wire bobbin.

Typical paths traced out by the finger on the coil former are shown in Fig. 2. The gear ratios are actually adjusted to slightly more or less than the


The various parts shown separately in the drawings can be identified in this view of the finished winder. No special parts or tools are required for its construction.


Fig. I. Basic design of the winder for making wave-wound coils

Left: The ratio between driving and driven friction discs determines the path traced out by the wire on the coil.

Fig. 2. Some typical types of wave winding possible with this machine. Coils up to $\frac{1}{2}$ in wide can be wound.
$R=\frac{\text { FORMER TURNS }}{\text { CAM TURNS }}$
$R=\frac{1}{1}$

(a)
$R=\frac{2}{1}$

(b)

(c)


Fig. 3. Method of calculating the gear ratio.

Right: Some of the coils wound on this machine. The wide variety of shopes exemplifies its versatility

even figures quoted, so that successive wires lie side by side. The choice of gear ratio is influenced by the angle $\phi$ (Fig. 2a), which should not be too large or the initial turns may slip out of position on the former.

The gear ratio of the friction drive is given by $R_{a} / R_{b}$ (Fig. 3). The machine is designed to have a maximum ratio of slightly over $2: 1$ and a minimum ratio of just below $1: 2$. For a gear ratio of $2: 1$ the radius of the circle described by the drive roller on the main drive disc $\left(\mathbf{R}_{1}\right)$ is half that described on the cam dise. To reduce the ratio to unity the drive roller is withdrawn a distance equal to half $R$, so that the two radii are equal. A further equal withdrawal reduces the gear ratio to $1: 2$. Fineness of control of gear ratio depends on the size of the disc, ease of adjustment increasing with disc size.

Details of the finished machine are given in Fig. 4. The various assemblies are mounted on a 8 in $\times 9$ in wooden bascboard. Two 4 -in diameter discs are
mounted at 3 in centres on silver steel shafts. The discs should be keyed or threaded to the shafts to prevent slip, and 6-BA studs tapped into the shafts are used as keys on the machine. The driving sides of the discs are scored with fine lines to increase the grip. The main bearings are made from brass nuts soldered between two suitably cut and bent pieces of mild steel sheet. A further bearing made from an angle bracket supports the drive shaft. Jockey rollers are fitted to reduce side thrust on the bearings. The drive roller consists of a rubber washer mounted on a small ball-race which is a push fit on a $\frac{1}{4}$-in silver steel shaft. A BSF thread is used on the shaft to give fineness of adjustment. Driving friction is adjusted with the threaded brass collar (A; Fig. 4) mounted on the end of the shaft. A mechanical counter cam ( $\mathbf{B}$; Fig. 4) is used also to lock the friction adjustment. A bolt-head cut and filed to rose-bit shape forms the driving cone.

The best "lock over" of succeeding wires is pro-

Left : Fig. 4. Complete layout of the machine and (Right) Fig. 5. The wire-feed finger and traversing carriage.


Right : Fig. 6. Details of the tail stock.


duced when the path traced out by the wire-feed finger on the coil former is linear and ideally the cam should be shaped to produce such a track. A cam with a plane surface produces a sine-wave track and a face-hardened cam of this type is used on this machine. The cam angle is 30 deg. The camfollower is made from silver steel rod and ground to a point at one end, but not hardened, and is held against the cam by a coil spring and retaining collar. The mounting is made from mild steel sheet. The wire-feed carriage is mounted directly on the camfollower shaft and is locked in position with a $6-\mathrm{BA}$ screw. A pulley for the wire is loosely mounted on a length of silver steel rod just above and slightly in front of the cam-follower.

The wire-feed finger and carriage are perhaps the most important units and are illustrated in Fig. 5. The different gauges of wire require fingers having different sizes of "eye." Piano wire, bent to form an "eye," is bound with 30 s.w.g. tinned copper wire for approximately $1 \frac{1}{2}$ in and soldered to give rigidity. The stud soldered to the carriage is drilled $\frac{1}{16}$ in to accommodate the finger which is locked in position with a nut and washer. A corresponding slot is filed in the mounting collar. The pulley guides the wire to the finger and the tension of the wire passing over it keeps the finger pressed against the coil.

Details of the tail stock are shown in Fig. 6. The centre cone is a push fit in the sleeve and the end of the silver steel shaft is rounded to reduce friction. The tail stock bracket is made from mild steel sheet. A lock nut is fitted.

The tension of the wire is important. Slackness produces spongy coils and over-tightness can produce electrical faults as well as increasing the risk of the wire breaking. At hand speed a simple friction brake on the wire bobbin with a sprung pulley compensating for variations in tension produces a satisfactory wind. A wire-operated brake, however, is used on this machine and the essentials of its design are shown in Fig. 7. A $\frac{3}{8}$-in Whitworth bolt is used so that only a small downward movement of the lever is required to release the brake. A spring consisting of a few thin strands of elastic holds the lever up (brake on) and is adjusted to give the required tension on the wire. Normally the bobbin tends to apply the
brake when in motion. The method of feeding the wire is shown in Fig. 8.

Before using the machine, the cam-follower mounting is set to give the required width of coil, and the appropriate wire-feed is adjusted to approximately the centre of the former. The coil tension is checked and the required gear ratio selected. The critical angle ( $p$; Fig. 2a) at which the initial turns slip out of place varies with the size of the wire and covering, and a single turn of cellulose tape, adhesive side outward, is an assistance in keeping the first few turns in position on a smooth coil former. A few trial runs to make final adjustments should be made before a coil is wound. The gear may be adjusted so that the pattern builds up progressively or retrogressively, the latter pattern being more stable. Coil data is available in various publications, some that have been found useful being given in the references.

Mild steel sheet $\frac{1}{16}$ in thick makes mountings of adequate strength. The tools required for the construction of this machine are few and consist of :-a hand or breast drill, drills, taps, dies, hacksaw, files, vice and soldering iron. The hand drill, clamped in the vice, can be used as an improvised lathe for the turning of small parts, such as the wire guide pullevs and to true up the two large friction-drive discs.

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## Hadio-Controlled <br> Hoats

NOW that the Philips radio-controlled boats seem to have established themselves as almost a permanent feature of European radio exhibitions, many people will be interested in the appearance of a book which reveals all the technical mysteries of how they work. Under the title of "Remote Control by Radio," A. H. Bruinsma has written a very full description of his two boats in 94 pages, including four complete circuit diagrams, 28 pages of data on the valves used and various photographs and diagrams to illustrate the text.

The first part of the book deals with the smaller and simpler boat, which has a two-channel frequency-division radio system with amplitude modulation. Considerably more space is devoted to the pièce de resistance "Teleservilips VIII" which has so impressed everybody with its remarkable versatility. This has an eight-channel radio systen working on the time division principle; the channels being provided by eight interlaced sets of pulses which are modulated in amplitude by the control signals.

Since the boats have been designed mainly for demonstration purposes, regardless of complexity or cost, the systems described in the book are not likely to be of much direct practical value to the amateur radio control enthusiast. The receiver of the larger boat, for example. uses 44 valves! Nevertheless, some of the individual techniques and devices within the systems are worth knowing about. and the book as a whole is a welcome addition to the somewhat scant literature on telearchics. It is published by the Philips organization at Eindhoven and is available in this country through the Cleaver-Hume Press at 8 s 6 d .

# The Future of Hearing Aids 

Battery Consumption and Other Problems: The Case

Against Standardization

By A. POLIAKOFF*

IN THE AUGUST, 1950, issue of Wireless World the writer mentioned the complexity of the problem of amplifying sound for the hard of hearing and gave results of a survey undertaken by his company into the optimum volume requirements of the deaf. In the time that has elapsed since, an increasing number of health authorities in many countries have started laying down minimum requirements for performance of hearing aids, to which instruments must conform before they qualify for a grant-in-aid, free issue or even import permits. These requirements refer almost entirely to the response curve, the limits of which are given with varying precision and severity. Needless to say, the variations of requirements between country and country are so considerable that in one case at least, an instrument having a response curve recommended by the Medical Research Council in their Report No. 261, "Hearing Aids and Audiometers "-would not pass the test. These variations of requirements are not, however, fundamental. The fact remains that the majority of authorities now hold the view that an instrument made to a particular recipe is the best hearing aid for all cases.

- This view is a technical heresy and its widespread adoption can only result in complete stagnation of hearing aid design and development.
In case we should be suspected of an anarchical bias, whether against health or any other authority, it is as well to stress that private industry, although more flexible than public bodies, has so far not gone nearly far enough into dividing the hard of hearing into a number of groups, according to their acoustic requirements, and providing each group with an instrument giving the optimum results for the members of that group.

The wide variety of requirements of the hard of hearing present a particularly fascinating problem, and their satisfaction in the most logical manner is a real challenge to all those working in the field.

It must be made clear at the outset that the audiometric curve-that is the hearing loss of the subject measured at threshold at different frequencies-is absolutely no indication of his instrumental requirements. The knowledge that a particular case has a loss at $1,000 \mathrm{c} / \mathrm{s}$ of 40 db does not permit us to say that he needs an instrument with an amplification 40 db . He may actually be unable to tolerate more than 15 db . Cases with strong "recruitment," that is, non-linear deafness, need small amplification and small power output. Other cases need low power
output but are able to make use of quite high amplification in certain conditions. Some deaf people who are able to hear when the words are shouted straight into their ear either directly or through a trumpet, need low amplification and high output. The very deaf need high output and the highest amplification they can have without "howlback."

An extremely interesting point is the variation in the amplification that can be usefully employed between cases having the same optimum volume requirement. We have come across people with moderate conductive deafness who had very modest optimum volume requirements combined with the ability, when provided with sufficient amplification, of hearing better in difficult conditions than a person of normal hearing. Their type of deafness is such that they can interpret speech quite easily in spite of the background noise and room reverberation. The perceptive case, on the other hand, has often a very limited range of useful amplification owing to the masking effect of background noise and room reverberation. Individuals vary, therefore, not only in their minimum, optimum and maximum power requirements, but also in what may be termed their

[^21]Multitone "Universal" hearing aid which is obtainable in three versions to meet individual power requirements. The type " $Q$ " chassis of the 2 -valve model is on the right.

resolving power, which sets a limit to the amplification that they can use.

It may be useful at this stage to list the main variables that the designer has to deal with :

1. Maximum output.
2. Mean amplification.
3. Shape of the response curve.
4. Automatic compression.
5. Conversion efficiency, the ratio of acoustic output to battery power consumed.

The problem, then, is to provide each individual with an instrument which fits most closely his requirements under the first four headings, and with the highest possible conversion efficiency. In addition, the individual preference between small bulk and higher battery replacement costs and large bulk and lower battery replacement costs must be met. The range of individual requirements is, as already indicated, very large indeed.

Taking the first variable, the power output, and expressing it in terms of electric energy delivered to the earpiece, we find that nearly half of all the hard of hearing cases tested by us are satisfied with the volume resulting from a power input to the earpiece of only 0.005 milliwatt or less. This has to be contrasted with maximum power output of the average commercial instrument of 5 mW , a difference of 30 db . Further, within the $0.005-\mathrm{mW}$ group there is a proportion of cases unable to tolerate anything like the output resulting from 0.005 mW , which is the power required only by the deafest members of this group. A number of sensitive cases have a maximum of only 0.0005 mW . At the other end of the scale there are the very deaf needing as much as 50 mW or even more. The electric power output range is, therefore, at least from 0.0005 to 50 mW , a range of 50 db .

To attempt to deal with this range of outputs with the same fundamental circuit of a crystal microphone followed by two voltage amplifier valves, and an output valve does not appear, to put it in the mildest way, to be good design.

It means that the vast majority of cases have far more power than they need, while a small minority have insufficient. Over-provision of potential power on such a scale would undoubtedly be considered utterly irrational in any other field. Consider a car manufacturer who insisted on putting the same engine he used in a 10 -ton lorry into the smallest passenger saloon and telling the unfortunate owner of the saloon that, after all, he has a throttle-control and there is no need to generate more power than he requires. The case of the hearing aid is actually worse, for in the vast majority of cases the power consumption of the aid is independent of the power delivered to the earpiece at any given moment. It follows, therefore, that the most of the deafened are consuming batteries at a rate far in excess of their own requirements.

The example of the throttle-control in a car brings us to a most important point, that of the purpose of the volume control in a hearing aid. The purpose of such a control should be to take up variations between different conditions of use by the user, and not to take up variations between requirements of different users. It is surely bad design to provide a control with 260 degrees of rotation and tell an individual user that the first 30 degrees are for his adjustment and the other 230 degrees for somebody else.

We have now seen that there is a very large variation in power output as well as amplification. What of the response curve? Its shape is not at all critical. Most


Performance of automatic volume control for the four settings of the level control in the Multitone "Selector" aid (0db $=0.0002$ dyne $/ \mathrm{cm}^{2}$ ).
cases are helped by absence of sharp peaks and a general rise towards the high end. This does not mean, however, that standardizing the response curve at, say, 6 db rise per octave up to 3,500 or $4,000 \mathrm{c} / \mathrm{s}$ is the right solution. There are a number of exceptions of which the most important is that of the very deaf. It is obvious that they should be provided with instruments having the highest conversion efficiency possible. The better the response curve in the higher frequencies the lower is the efficiency of the earpiece in the part of the response curve that a very deaf subject can hear. It is very rare for these cases to respond to $4,000 \mathrm{c} / \mathrm{s}$ at all. It is wrong, therefore, to saddle them with an "academic" response, the advantages of which they themselves cannot hear. Not only does this result in the very deaf being no longer able to hear with instruments using a $22 \frac{1}{2}-V$ hightension battery, and needing as much as 45 V , which cannot be conveniently incorporated in a reasonably sized " monopack" instrument, but it also makes it difficult for them to hear at all. It is extremely hard to obtain an effective seal between the earpiece and the ear at the higher frequencies. If there is a good high-note response the aid starts whistling as soon as the amplification is sufficient for the user to hear at all, and, not infrequently, he himself is quite incapable of hearing the whistle. One of our users before the war, an M.P., was asked to leave the Chamber on one occasion on account of a continuous whistle set up by his instrument of which he was completely unaware. The "howlback" does not merely cause a disturbance, but almost always overloads the output valve and spoils the amplification of sounds the user can actuaily hear.

## Volume Compression

We now come to automatic volume compression. In most of the instruments manufactured by my company, and now a number of other manufacturers, t'ie power output can be limited to a desired figure by automatic compression, while the amplification of weak signals is controlled by a manual volume control. The reason for the manual control is to take un
large variations between noise conditions such as exist, for instance, between the cabin of an aeroplane in flight and a quiet room in the country.

Too much compression can be very objectionable, for not only will speech be flattened out, but pauses between words will be filled by the rush-in of the background noise. There is a considerable variation between the amount of compression preferred by different cases, which provides yet another variable for the designer, together with the selection of the most suitable " attack" and "decay" times.

Here is the problem, and it is difficult to see how it can be solved satisfactorily without designing special equipment for each of a number of groups.

I believe that the future of hearing aids lies in detailed study of these requirements coupled with the production of valves (later transistors) and batteries designed specifically for use for a particular group.

Having studied the volume requirements of the deafened, my company has now adopted a design policy for instruments to which we have given the name of "Fitted Power." Our new "Universal" instrument is available in three versions: the first version has only two valves, both voltage amplifiers, with a high-tension consumption of approximately a quarter of that of the average instrument with the same high-tension voltage ( $22 \frac{1}{2} \mathrm{~V}$ ) and with a lowtension drain of half of that of a normal instrument. Its maximum amplification at $1,000 \mathrm{c} / \mathrm{s}$ is in the region of 40 db . Aatomatic volume control is provided, while the maximum output with the automatic volume control switched off is in the region of 119 db referred to 0.0002 dyne $/ \mathrm{cm}^{2}$ at $1,000 \mathrm{c} / \mathrm{s}$. In cases for which the amplification is not enough a 3 -valve version is available with the low high-tension drain of 0.2 mA . Where the power output is not sufficient, which, by the way, is extremely rare, a third instrument is available with a drain of 0.4 mA .

In our "Selector" instrument we have introduced "Fitted Power" by making the high-tension drain vary with the setting of the automatic volume control level selector. This instrument employs a 15 -volt battery and the high-tension drain changes in four steps from 0.1 to 0.4 mA as the compression is varied through a range of 20 db . The accompanying graph shows the compression curves of the four positions of the level control.

The lower amplification obtained with the reduced high-tension drain is a great help with difficult perceptive cases. In the past we spent a great deal of time "calming" instruments by reducing amplification and maximum output from their standard values to deal with cases with pronounced recruitment.

Nothing has been said so far of the effect of these ideas on further miniaturization. It is obvious that instruments designed expressly for the less deaf group could be much smaller than those produced for the deafer groups. A far more radical approach, however, is required to produce really small instruments. The man who makes use of his instrument for special occasions only, such as theatres, and so on, does not need a battery that will last him longer than the duration of the play, with a little to spare, and it would be possible, with the help of the battery manufacturers, to produce a really minute instrument for him.

When it becomes possible to use transistors only in an instrument, thereby eliminating one of the batteries, miniaturization will receive a further impetus.

The transistor hearing aid is not here yet, although at least one manufacturer in the States has announced
the production of the first transistor instrument. This, however, employs two valves and uses the transistor only in the output.

The principles set out above for the design of hearing aids are quite independent of the means by which amplification is obtained. Valves, transistors, or any other devices that may appear in the future all use energy and this should not be recklessly wasted, nor should individuals be supplied with instruments potentially far too strong for them, in order to enabie somebody else to hear with the same device.

## MANGANESE-HISMUTH PEIRMANENT MMGNETS

WHEN manganese and bismuth are melted together an inter-metallic compound ( MnBi ) is formed which exhibits strong magnetic properties when the particle size is small enough. ${ }^{\text {i }}$ Recent investigations have shown $n^{2}$ that magnets made from this compound can offer a practical alternative to conventional alloys of cobalt and nickel, and that, in particular, they have a coercivity higher than that of any other known magnet. This property is of particular value where the demagnetising force is high, as in magnets of short length. Coercivities higher than 3,000 oersteds have been measured, compared with ,600-700 for alloys such as "Alnico" and "Alcomax," and the best MnBi magnets have a maximum energy product (BH) of $4.3 \times 10^{6}$ gauss-oersteds, which is of the same order as that of Alcomax. The remanence reaches 4,300 gauss compared with 7,500 for 6 per cent cobalt steel and 12,500 for Alcomax III.
Magnets are prepared by a powder-metallurgy technique, the Mn -Bi melt being first pulverized to an average diameter of 8 microns and then separated magnetically before being compacted (at a temperature of 300 deg C ) in dies at pressures up to $3,000 \mathrm{lb} / \mathrm{in}^{2}$ in a pulsating axial magnetic field with a maximum of 20,000 oersteds.

The processes are carried out in an inert atmosphere to guard against oxidation, and protective coatings for the finished magnet are necessary, particularly when they are to be used under conditions of high humidity.

[^22]
## Standardizing Laindspeaker Sizes

TO facilitate the replacement or exchange of loudspeaker units, standardization of the size and distribution of fixing holes has long been advocated.

Now a British Standard Specification (BS1927؛1953) has been issued for nominal loudspeaker sizes of $2 \frac{1}{2}, 3 \frac{1}{2}$, $5,6 \frac{1}{2}, 8,10,12,15$ and 18 in diameter and gives mounting circle diameters and the number, size and distribution of fixing holes.

It is also required that loudspeakers conforming to this specification should state the nominal fundamental resonant frequency, the nominal impedance (preferred values 3 and 15 ohms), and that one terminal should be marked red to indicate that when this terminal is positive the corresponding movement of the diaphragm is forward.

Details of the conditions of measurement of resonant frequency and nominal impedance are given and the specification, which is obtainable from the British Standards Institution, 24, Victoria Street, London, S.W.1, costs 2 s .

## 

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# " CATHODE RAY" Clears Up Some Common Misconceptions About Fields 

WHEN interference with broadcasting caused by electrical equipment in the home is being discussed, one of the ways by which such interference reaches the receiver is commonly referred to as radiation. Some of the people who talk about radiation in this connection probably realize-or would do, if they thought about it for a moment-that they are misusing the term.

What exactly is radiation? BS.204:1943 defines it quite briefly as "the emission of energy in the form of electromagnetic waves." Now the process we are considering is certainly the emission of energy. It is even true to say that part of the energy is in the form of electromagnetic waves. But it is usually an insignificant part, so "radiation" hardly seems to be the right word to describe the whole. However, if it were merely a matter of relative size it might perhaps be rather fussy to object. The reason I am objecting is that if the people who know better call this thing " radiation" others may be led to believe that it really is radiation and thereby get their ideas wrong. For this is not just some little by-way of science; it is an arterial road, and if the entrance is obscured by a fog of misunderstanding a lot of people are going to lose their way. The road leads to two fascinating fieldsthe electric field and the magnetic field. To the student these may seem vague and intangible things, compared with his clear-cut circuits carrying easily measurable currents. But in fact the energy of lowfrequency or even d.c. circuits is in the surrounding fields rather than within the wires themselves. Circuit electricity is just a very special case of the more general electromagnetism, and although quite a lot of useful work can be done without bothering much about fields, this restricted knowledge soon leads to difficulty in radio-particularly, for example, with aerials. The circuit-minded man is hopelessly at sea-or should we say in the air?-when he comes to aerials.

## "Lines of Force"

Of course, everybody who studies electricity more than a very little learns something about fields. Mathematicians take to them like ducks to water. But I am not concerned with the minority of born mathematicians but the majority, who are not really happy about symbols unless they are accompanied by a clear mental picture of what they represent. And this is where fields are a difficulty. Of course, one has only to handle a pair of magnets or look at an electrostatic voltmeter in action to be convinced that very potent forces do exist in space, but so far nobody has been able to discover how they come to be there. If one is satisfied with bare mathematics, all right; but if not, the best that can be done is a rather artificial mechanical analogy, and unfortunately this is only too liable
to give wrong impressions unless it is kept strictly in its place.

For instance, nearly all the instruction on the subject brings in " lines of force." One book, published only a few years ago, says "Attraction or repulsion (between electric charges) can be explained by studying the electric lines of force. There are electric lines of force emanating from the proton or positive charge, and if no other charges are present, these lines of force extend out to infinity in all directions." When a definite and unqualified statement like this is made on a fundamental principle by an author with eight lines of qualifications after his name on the title page, how can readers be expected to realize that it is quite un-true-a mere baseless figment of the imagination? When we are informed in song that "You're the cream in my coffee " there is no serious risk of the statement being accepted as a literal fact, because metaphor is an understood convention in that kind of thing, and it is interpreted accordingly; but when a technical instructor uses poetic imagery he is under an obligation to give warning of it. I am not suggesting that because lines of force are non-existent there is no justification at all for introducing them-personally I find them very useful-but if they are introduced it should be made quite clear that they are purely a mental or visual representation, having no closer relationship to actual fields than an arrow on a weather chart has to a wind blowing in from the Atlantic. Just as a wind fills a whole region of space, not just where the arrow is drawn, so a field is continuous, not confined to separate lines.

Some while ago I saw an instructional film in cartoon form, in which the inflammable vapours given off by natural oil were represented as imprisoned demons seeking to escape and wreak vengeance. Now if that presentation succeeded in conveying an impression of the risk due to oil vapour and the need to guard against it, well and good. And if lines of force, flux-cutting, and all the rest of the field ideas help us to make use of electromagnetism, well and good; but to suppose that they are real is like believing that each barrel of oil contains a malevolent demon with forked tail. As a matter of fact, it is possible to substitute quite different ideas for magnetic fields and still get the same results; but I am not going to be as unconventional as all that.

Beginning, then, with well-known principles, there is the electric field, associated with a difference of potential-voltage-and the magnetic field, associated with an electric current. Theoretically it is possible to have either kind of field without the other. If you charge any well-insulated capacitor, there will be a difference of potential-and hence an electric field-between the two plates, though there is no perceptible current to set up a magnetic field. It is
not so easy to demonstrate a magnetic ficld without voltage, because suitable circuits such as coils usually have resistance that needs an e.m.f. to keep the current going, but by cooling a lead ring sufficiently its resistance can be reduced to so nearly zero ("superconductivity") that a current once started flows for hours and maintains a magnetic field without any e.m.f. or electric field. I am leaving out of account permanent magnets, because although it is now believed that their fields, too, are due to electric currents (within the atoms), their circuits cannot be seen and handled.

Note that a separate electric or magnetic field can only exist at zero frequency. The moment one starts to vary an electric field it is necessary to increase or reduce an electric charge somewhere, which means moving a charge, and that is an electric current, which sets up a magnetic field. And any variation of a magnetic field generates an e.m.f. in the circuit producing the field, so an opposite e.m.f. is needed to compel the variation, and the resulting difference of potential causes an electric field.

That is a most important thing to grasp. Important enough to look into more closely. Let us look into it with particular reference to the production of the varying magnetic field. (The whole statement can be converted to apply to the varying electric field by using the duality principle explained in the April 1952 issue.) All the apparatus we need is a coil connected to an a.c. generator, as in Fig. 1. To simplify matters let us assume that the a.c. has a pure sine waveform, and that the coil is surrounded by air. Then the field strength or flux is exactly proportional to the current, so its waveform is the same and is in phase. In fact, by using suitable scales the same curve can be used for both, labelled I and $\Phi$ in Fig. 2. E.M.F. is generated not by a constant field but by variation of field and is proportional to the rate of variation, represented by the gradient of the $\Phi$ wave. The steepest gradient is where the $\Phi$ wave cuts the horizontal axis, first upward (generating a peak of positive e.m.f.) and then downward (generating a peak of negative e.m.f.). The rest of the e.m.f. curve can be filled in by plotting its amplitude at every point proportional to the gradient of the $\Phi$ wave; the result (dotted in Fig. 3) is another sine wave, or rather a cosine wave, for although the shape is the same it is quarter of a cycle ( $90^{\circ}$ ) earlier in phase.

## Direction of Flow

A little question of sign arises here. We have shown $E$ as positive when the current is increasing, but that means nothing unless the direction around the circuit is specified. By Lenz's law, the field-induced e.m.f. acts in such a direction as to oppose the cause of the current change; so when the current is increasing it is positive against the direction of current increase. The only other voltage in the circuit is the generator e.m.f., which therefore must be exactly equal and opposite. So if the curve E in Fig. 3 is understood to refer to the same direction in the circuit as the current, it represents the generator e.m.f.

The reason for being so particular about getting the directions right is that they indicate the direction of power flow, which will be the next thing to be considered. In case you are wondering why we started with the current and worked back to the e.m.f. driving it, I can only say it is easier that way. Having done what we have done, we can say that if a sinusoidal
c.m.f. is applied, say E in Fig. 3, the resulting current will be as shown by I in the same diagram, lagging $90^{\circ}$ behind the e.m.f., as the books say. Of course, it must not be imagined that this lag is a time delay, each postive peak of current being caused in some way by the positive voltage peak that occurred quarter of a cycle earlier; the current peak is actually an accumulation over half a cycle of the continuous increase needed to generate an e.m.f. to oppose the continuously positive generator e.m.f. during that halfcycle, and has now flattened out at a summit because at that same instant there is no e.m.f. to be opposed.

## An Eye on the Energy

One other side issue may be disposed of at this stage-the resistance of the circuit. Either it can be assumed to be zero, which simplifies the situation to the one described and illustrated, or alternatively, if that seems too unreal (super-conductivity demonstrations notwithstanding) the e.m.f. needed to overcome the resistance of the circuit can be assumed to be provided by an auxiliary suitably-phased series generator, which, with its e.m.f., we shall forthwith completely ignore-quoting as authority for this conduct the Superposition Theorem.
As I emphasized in January, 1952, voltage and current by themselves are less significant than energy, which is the result of the exercise of power in some direction or another for a period of time. In a purely resistive circuit, the current is in phase with the applied e.m.f., and as power is equal to voltage $\times$ current it is in this case positive in the direction of the current-meaning that power is going out of the generator and being dissipated in the resistance. But if we now draw a curve representing the power in our purely inductive circuit, finding its amplitude at every moment by multiplying together the E and I amplitudes, there are some moments when (as with resistance) both are positive, indicating positive power, and some when both are negative, indicating the same $(-1 \times-1=+1)$, and some when one is positive and the other is negative, indicating negative


Fig. 2.1 The magnetic flux ( $\Phi$ ) set up by the current (I) is in phase with the current, so can be represented by the same curve.


Fig. 3. The generator e.m.f. (E) needed to overcome the e.m.f. generated in the coil by the variation of $\Phi$ is $90^{\circ}$ ahead of the current.


Fig. 4. Power is represented by the curve $P$, plotted by multiplying I by $E$; and energy ( $=$ power $\times$ time) by the shaded areas, which are equally positive and negative in each cycle.


Fig. 5. If account is taken of the time lag needed to convey changes in field strength to a distance from the coil, the positive energy (leaving the circuit) is greater than the negative (returning to it).
power. If positive power means power going out, negative power must mean power coming back. The power curve, P in Fig. 4, shows equal positive and negative periods. The shaded areas represent power $\times$ time, or energy. So we see that in each current cycle there is first a wad of energy going out and then an equal wad of energy coming back. The circuit takes no energy as a gift; it borrows and then immediately repays in full. This, of course, is no virtue-merely lack of opportunity. The circuit has no means (such as resistance) for permanently disposing of any energy. To the question, where does it go temporarily, the answer is, into the magnetic field. A field-magnetic or electric-is stored energy. While the current is flowing with the generator e.m.f. it is building up a magnetic field-either positive or negative, according to the direction of the current through the circuit-and while it is flowing against the gencrator e.m.f. it is being driven by the energy of the now collapsing field.

## Spread of the Field

Having been so critical of other people's inaccuracies, I must confess than the foregoing is not perfectly accurate, for it is based on an untrue assumption. The assumption is that the ficld is everywhere exactly in phase with the current, as shown in Fig. 2. But, theoretically, the field fills the whole of space, and the more outlying parts of space cannot know what the current is doing immediately it does it. At the very fastest, the news cannot travel quicker than light, for, according to Einstein, the universe is so constructed that no greater speed is possible. That is, in fact, the speed at which the growth of the field does spread out through empty space. Being 186,288 miles per second, it doesn't take long to spread the field to the greatest distance likely to be of practical interest. So the current in the circuit has to grow very rapidly indeed to catch any part of the neighbouring space by surprise, as it were.

Now this is the point where things tend to become rather difficult to grasp. Let us first consider how the magnetic field would be distributed if the current had been raised to a certain amount and kept there long enough for the momentary spreading effect to be past. Much the strongest part of the field is inside
or close around the coil. Beyond that it falls off at an increasing rate, until when it is well outsidesay five or six times the diameter of the coil-it is approximately inversely proportional to the cube of the distance. So at a hundred times the diameter it would be only about one-thousandth as strong as at ten times; at a thousand times, one-millionth; and so on. But we need hardly bother about the "and so on," for any of this steady field that lies much beyond, say, one thousand times the dimensions of the circuit maintaining it is for most purposes negligible. For example, the field from a 1 -inch coil could be regarded as being all within a radius of 1,000 inches, or, say, 100 feet.

## Exit Some Energy

Now consider what happens if the current is varied, by using a.c. instead of d.c. At a distance of 100 feet, the variation does not become effective until $1 /(186,288 \times 52.8) \mathrm{sec}$, or about 0.1 microsecond, later. So when the current has reached its maximum-and likewise the field close to the coil-the field at 100 ft has still $0.1 \mu \mathrm{sec}$ in which to continue growing. This effect can be represented by a delay of the $\Phi$ curve, with a corresponding delay of the E curve. Whether or not this delay would be worth considering would depend on the frequency of the a.c. At $50 \mathrm{c} / \mathrm{s}$ it would be only one-hundred-thousandth of a cycle. But at $1 \mathrm{Mc} / \mathrm{s}$ it would be one-fifth of a cycle, or 72 deg. That is not to say, of course, that the induced e.m.f. as a whole would be delayed by anything like this amount; the strongest components, due to the close-in field, would be delayed much less. But it seems inevitable that there would be some delay, increasing with frequency. In Fig. 5 it is shown as amounting to 45 deg, to make it clearly visible; and when the power curve is plotted it can be seen that the balance between outgoing and incoming energy, noted in Fig. 4, no longer applies. More is going out than coming back. Where is it going?
Maxwell gave the answer when he showed mathematically that it was being radiated off as electromagnetic waves. What this signifies we shall see more closely in a moment, but meanwhile let us take stock of the position so far. So long as current is unvarying (d.c.) no energy can be radiated, for there is no e.m.f. (except what may be necessary to overcome the resistance of the circuit). When the current is varying slowly (low-frequency a.c.) there seems to be a possibility of radiation, but to such a small extent as to be negligible. But as the frequency is raised, it looks as if sooner or later the current would vary rapidly enough to cause an appreciable phase delay, with corresponding loss of energy from the circuit.

Well, of course, this is far too over-simplified to be a " proof" of anything; for one thing we have paid no attention to what has been happening to the electric field. Not only is there the field lagging $90^{\circ}$ behind the current (since caused by its rate of change), which field corresponds to the Ecurve in Fig. 4; there is also an electric component caused by the outward movement of the magnetic field, and therefore in phase with it and the current. This electric field plays an essential part in radiation. Without it, there would be nothing to keep the magnetic field in existence during its journey through space. But since a varying magnetic field generates an electric field, and a varying electric field generates a magnetic field, the two keep one another contimually in existence. Half


Fig. 6. Relotive magnetic field strength ot distances measured along the axis of a coil or loop Ift in diameter energized by current at $/ \mathrm{Mc} / \mathrm{s}$. The distance of the crossover point (here 156 ft .) is proportional to wavelength.
the radiated energy in any piece of space is necessarily $i_{r}$ the form of magnetic field and half in electric field. If one kind of field were in some way got rid of, the other would have to disappear, too. This is an important difference between radiated fields and induced fields (as the non-radiated variety is called) which as we saw can exist separately or in any proportion.

We have noted already that induced fields fall off very steeply with increasing distance from their source. Some books say inversely as the square, but that calculation does not apply to ordinary circuits, from which the fall-off is inversely as the cube of the distance. The energy locked up in a single radiated wave, if it is assumed to flow out equally in all directions like the surface of an expanding soap bubble, remains constant, but is distributed over an area proportional to the square of the distance from the source. Just as electrical energy is proportional to current $\times$ voltage, field energy is proportional to magnetic field $\times$ electric field, so either of these separately falls off in direct proportion to distance. Although close up to the circuit the induction part of the fields may be vastly stronger than the radiating parts, this situation becomes reversed at great distances. Suppose, for example, the source is a circuit lft in diameter, through which current is alternating at $1 \mathrm{Mc} / \mathrm{s}$. Then Fig. 6 shows the relative strengths of radiation and induction at ranges up to 300 ft along the axis of the loop. At 5 ft the radiation is negligible, being only one-thousandth of the induction field strength. But at about $5,000 \mathrm{ft}$ or rather under a mile the proportions are reversed. The distance at which thee cross-over takes place-where the induction and radiation fields are equal-is $\lambda / 2 \pi$, or a little less than one-sixth of the wavelength. In this case the wavelength is 300 metres, so the zone within which the induction predominates extends up to 156 feet from the source. On 1500 metres (Light Programme) it is 780 feet.- Now it is quite exceptional for noisemaking appliances to interfere seriously with broad-
casting at greater distances than these, especially if it is remembered that the distance is not necessarily from the appliance itself but from the nearest wires, etc., connected to it. So radiation is seldom sufficient to cause trouble. The interference comes by induction.
"So what," you say, " as long as it comes?"Well, the difference between radiation and induction is not purely theoretical. The fields themselves are the same, of course. The difference is that whereas radiation must consist of equal electric and magnetic fields, induction can be both or either. Some kinds of screening protect against magnetic fields only or electric fields only; either will do against radiation, for radiated fields cannot exist separately. But this is not so for induction. A source of interference surrounded by an iron box has most of its outer magnetic field removed, to the great benefit of portable sets in the vicinity. But if the box is not earthed an appreciable electric field may remain, to the disadvantage of sets with "picture-rail" aerials. On the other hand, earthed graphite or tinfoil screening around a television set will greatly reduce the electric field from the line-scanning components, but the magnetic field from the deflecting coils will be practically unaffected, as the portable set with its frame aerial can show.

Motor ignition is rather a special case of interference source. It is most manifest on the television wavelengths, and $1 / 2 \pi$ of a television wavelength is only about 3 feet. Viewers have no need to be told that the effective interference range of motor vehicles is not restricted to 3 feet. Therefore one can use the term "radiation" in this connection with complete confidence.

## IRMW MATEIEALS OF RAIDIO

THE interplay of supply and demand is strikingly illustrated in the relations between the chemist and the radio technician, and the history of radio progress is punctuated by the advent of new materials, developed often to meet a specific demand, but not infrequently springing fortuitously from pure research.

Polyethylene, which had a profound influence on the wartime development of radar, was but one of a whole series of strange new products, synthesized by a very high pressure technique developed by I.C.I. for pure research. It owes it early recognition as a dielectric of outstanding performance to the fact that its mechanical properties resembled those of gutta-percha, for which it was first investigated as a possible substitute in submarine cables.

Nickel-iron alloys of high permeability and polycrystalline ceramic aggregates of various permittivities and temperature coefficients may be cited as examples of materials deliberately sought for specific requirements.

To facilitate the more rapid development of new radio materials, the Radio Research Board of the Department of Scientific and Industrial Research set up, in 1948, a committee on raw materials, to co-ordinate the efforts of academic, industrial and Government research laboratories in tackling the problems posed by the demand for materials having characteristics so far unattained.

Radio Research Special Report No. 25, "Selected Problems in the Preparation, Properties and Application of Materials for Radio Purposes " (H.M. Stationery Office, price is 6 d ), presents a survey of the position up to 1951, and serves the dual purpose of acquainting readers with existing special materials and of stimulating thought on possible new lines of development. It is divided broadly under the following headings: 1, Ceramics; 2, Organic polymeric dielectrics; 3, Magnetic materials; 4, Semiconductors. An appendix gives over 60 references to recent publications on these subjects.


AT THE NEW HOUSE OF COMMONS the Research Department of Messrs. Watford Electric \& Manufacturing Co. Ltd. is using the Cossor Model 1049 Double Beam Oscillograph and Camera together with D.C. Pre-Amplifier Model 1430 to record the dynamic temperature range, the humidity level and the performance of the servo-mechanisms in the automatic controller of one of the air-conditioning plants.

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# Ignition Interiference 

## Tolerable Limits at Television

## Frequencies; Effect of Suppressors

## on Engine Performance

1PAPER* read recently in London to a joint meeting of the Institutions of Mechanical and Electrical Engineers gave a fairly comprehensive survey of ignition interference as it affects television reception. Its causes have been extensively investigated and the customary means of suppression are discussed and from the motorist's angle the effect of fitting suppressors to motor car engines has been quite thoroughly investigated.

The exact mechanism of its generation is not always so well explained as it is in this paper. The energy stored in the self-capacitance of the plug, the high-tension cables and the ignition coil is very rapidly discharged and gives rise to an oscillatory current whose peak value may exceed 100 A , but lasting for less than a microsecond.

The interference radiated has the redeeming feature that, being a pulse type of signal at a very high frequency, its worst effects can sometimes be curbed by means of noise limiters in the receiver, but as there is a limit to what can be achieved in this way, suppression at the source is by far the most satisfactory cure. Thus a large part of the paper is devoted to this aspect of the problem.

## Ignition System Layout

In the course of investigation it was revealed that interference from different types of vehicles varied between maximum and minimum levels of 6,000 and $20 \mu \mathrm{~V}$ per metre at $45-50 \mathrm{Mc} / \mathrm{s}$. The curves reproduced show that of the tests made with a number of pre-1939 vehicles, those with the more compact ignition layout, exemplified by the coil mounted on the engine block (curve 2), produced appreciably lower interference fields than those vehicles with more spread-out ignition components. The corresponding curves for the post-1945 vehicles do not show the effect of different positioning of the parts, so that direct comparisons are not possible. However, there is only a slight general improvement in the curves for the newer vehicles and without suppressors of some kind the radiated interference exceeds the $50 \mu \mathrm{~V}$ per metre level at a distance of 10 metres from the source which is the maximum radiation now permitted from all motor vehicles manufactured after July 1st, 1953. It is defined in the latest regulations dealing with the suppression of interference, "The Wireless Telegraphy (Control of Interference from Ignition Apparatus) Regulations, $1952, "$ and covers interference in the band $40-70 \mathrm{Mc} / \mathrm{s}$.

It is mentioned as a passing comment that occasionally electrical equipment apart from the ignition system can be responsible for appreciable interference of a slightly different kind, but equally annoying. Screen-wiper motors, dynamos and such-like can all add their quota and this accounts for the complaints,

[^23]often discredited, that diesel-engined vehicles sometimes interfere with television reception.
The effect of fitting resistance suppressors in the h.t. wiring of the ignition system is interesting and is stated to change ... "the high-frequency discharge into a unidirectional one of much lower duration, with a reduction of peak current to less than one ampere. The amplitude of the current in the inductive or "flame" component (of the spark), on the other hand, is substantially unaltered since the normal parameters of the circuit limit it to milliamperes."
When discussing the question of the resistance values for ignition suppressors, it is stated that suitable values lie between 5,000 and 15,000 ohms, but that these are by no means critical.
It is significant that ignition component manufacturers are now considering embodying suppressors in the design of the component. For example, in one well-known make the high-tension brush consists of three parts, the two ends being of normal brush carbon and the centre of high-resistance material which forms the actual suppressor. This gives about the same degree of suppression as a single external resistor in the coil-distributor h.t. lead and satisfies most requirements.

With regard to the effect of suppressors on engine performance, it is stated that . .. "the evidence accumulated from carefully controlled tests made by engine and car manufacturers is that, for normat engines under normal operating conditions, suppression resistors have no effect on performance. Probably some of the prejudice against resistors has arisen from the effect they have in reducing the light in-

Field strength in db above $I_{\mu} V$ per metre of interference generated by pre-1939 motor vehicles on a percentage basis of vehicles tested. Curve l, unsuppressed private vehicles ; curve 2, unsuppressed vehicles with ignition coil on engine block; curve 3, with suppressor in coil-distributor lead and coil on engine block ; curve 4, with coil-distributor and plug suppressors. The horizontal broken line is the $50_{\mu} \mathrm{V}$ metre at 10 metres level.

tensity from the spark, but it cannot be emphasized too strongly, however, that the brightness of a spark is not a criterion of its incendivity."
Typical results on the effect of resistors on engine output are given in the table, which relates to a fairly high-performance four-cylinder overhead-valve engine of $1 \frac{1}{2}$ litres capacity.

The general conclusions that emerge are that the effect of ignition suppressors on engine performance is negligible and that a single suppressor in the coildistributor h.t. lead generally gives adequate suppression.
As in the great majority of cases fitting suppressors is still voluntary, an important inducement, if one should be needed, is that suppressors lead to a reduction in plug erosion and so give a longer useful life.

TABI.E

| Engine Speed <br> r.p.m. | B.h.p. output |  |
| :---: | :---: | :---: |
|  | Without resistors | With $10-\mathrm{k} \Omega$ to 15-k $\Omega$ resistor at distributor and each plug |
| 1,000 | 13.5 | 13.5 |
| 1,500 | 22.1 | 21.9 |
| 2,000 | 30.0 | 29.9 |
| 2,500 | 36.8 | 37.1 |
| 3,000 | 42.5 | 42.9 |
| 3,500 | 45.9 | 45.9 |
| 4,000 | 46.7 | 47.2 |

## R.E.C.M.F. SHOW

WE give below a list of the 120 stand-holders at the exhibition of components, test gear, valves, materials and accessories to be held in the Great Hall, Grosvenor House, Park Lane, London, W.1, from April 14th to 16th. This is the 10th annual show organized by the Radio and Electronic Component Manufacturers’ Federation, and it will follow more or less the pattern of its predecessors. It is essentially a private exhibition, admission being limited to
those who have a professional, industrial or trade interest in the equipment exhibited. Tickets, which are gratis, are obtainable from the R.E.C.M.F., 22, Surrey Street, London, W.C.2.

The catalogue of the exhibition, which will be available at the show or from the R.E.C.M.F. (price 1s) includes a classified buyers' guide to the equipment manufactured by the members of the Federation.


## Manufacturers' Products

NEW EQUIPMENT AND ACCESSORIES FOR RADIO AND ELECTRONICS

## Television Aerial Developments

A MODIFICATION which, although apparently trivial, may have quite a marked effect on the general reliability of the aerial has recently been adopted by Wolsey Television, Ltd., 75, Gresham Road, Brixton, London, S.W.9. This consists in replacing the usual round-section "Tee" fittings on the ends of the crossarm and the pole-cap fitting by square sections.

The change is said to have the following advantage: with an " $H$ "-type aerial the vertical elements cannot slip round the crossarm and become badly misaligned; the whole aerial cannot rotate on its pole and change its orientation and the two quarterwave reflector rods can be assembled before leaving the factory. This last change means that the erection on the site is simplified since these elements have only to be turned to the


New square-section crossarm fittings used on Wolsey television aerials keep all elements properly aligned and the whole correctly orientated.
vertical and locked in position by two wing nuts. The supporting pole is secured by a bolt and wing nut.
Single element aerials will also benefit, but perhaps to a lesser degree except in the case of horizontal aerials for the low-power stations The square ends may prevent the elements sagging. The same type of fittings are to be used on all Wolsey aerials.

## Polystyrene Capacitors

ALTHOUGH polystyrene-insulated capacitors have been in existence for some time certain restrictions have been imposed on their use


Suflex polystyrene capacitotors.
by the comparatively low operating temperature of this material.

By a new process of manufacture Suflex, Ltd., 35, Baker Street, London, W. 1 have produced a polystyrene capacitor claimed to be stable over the range of temperatures -40 deg C to +85 deg C .
Other advantages claimed for this type are:-insulation resistance 10 million megohms; power factor 0.00025 and stability of capacitance 0.1 per cent. They are enclosed in moulded polystyrene and have end connecting wires.

The new capacitors are available in values ranging from 5 pF to $5,000 \mathrm{pF}$ and for working voltages of 125, 250 and 500 d.c. A $500-\mathrm{V} 1,000-\mathrm{pF}$ type measures $\frac{5}{5}$ in long and $\frac{3}{8}$ in in ciameter while one of 400 pF for 250 V is $\frac{15}{32}$ in long and $\frac{7}{32}$ in in diameter.

## Conductive Ceramic

A NEW material for use as a dummy load, or as an attenuator, for absorbing power at centimetric and millimetric wavelengths in waveguides has been introduced by The Plessey Co., Ltd., Ilford, Essex.

It is a form of conducting ceramic called Caslode and is a homogenous substance of high intrinsic resistivity which does not contain dispersed conductive particles. At a wavelength of 3.2 cm its dielectric constant is about 20 and the loss angle 10 to 20 deg .
During manufacture the material is fired at a high temperature to form a structure which is stable under extreme conditions of temperature and humidity.
High-power dummy loads of this material have been used as matched terminations for waveguides without recourse to water cooling.

A Caslode attenuator pad is still in the experimental stage, but units have been produced having an attenuation of 28 db with a match of 0.95 .


Wedge-shaped dummy loads of Caslode for use in waveguides.

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# RaNDOM RADIATIONS 

By "DIALLIST"

## Cathodic Protection

My best thanks to the several experts in cathodic protection who have written to supplement my note in the February issue of $W . W$. My information came from a friend who, though not a technical man, has a better-than-average working knowledge of electricity. We discussed the subject at some length after his recent return from the Persian Gulf, but I was pretty sure that I hadn't got quite the whole story. Readers now tell me that the metal generally used for the negative elements is not aluminium, but a special magnesium alloy. I ought, anyhow, to have spotted that the action of sea-water on steel piers is electro-chemical. It's bound to be, when you come to think of it. Sea-water contains many salts in solution, the original molecules being dissociated into their component atoms and radicals in the form of positive and negative ions. When one considers that the negative ions must consist to an important extent of halogen atoms as eager to be up and doing as chlorine and bromine, and of such active radicals as $\mathrm{SO}_{4}$, it's no wonder that unprotected steel piers suffer!

## Electronic Fish-finding

Some readers have written asking me to say a bit more about electronic and electrical fishing. Well, here's a summary. Electronic gear is regularly and successfully used by our modern fishing vessels to enable them to locate shoals of fish. The echo-sounder gives quite good results, but the best gadget yet is the German Fischlupe (or "Fishscope"). In the "search" position this gives a small-scale picture on a c.r. tube of the water, from surface to seabed, in the neighbourhood of the vessel using it. A shoal of fish then appears as a luminous patch. Turning a switch changes the picture into a large-scale one of a small area and a knob enables the shoal to be centred on the screen. Its depth can be read direct from a scale and the kind of fish of which it is composed can be determined with fair certainty; for the shoal now appears as a collection of luminous points, each representing the echo from one fish. The depth of the shoal, its
extent and shape, the formation in which the fish swim and their size as shown by the individual echoes, give the expert fisherman all the indications that he needs.

## Another Fish Story

The actual capture of fish by the use of an electric field has not got beyond the small-scale experimental stage at sea. It has been used, however, in trout streams for some little time as the most effective method known of ridding such waters of cannibals, predatory fish, diseased fish and other undesirables. The principle is this: in a direct field fish are attracted towards the positive pole and repelled from the negative; if the field exceeds a certain intensity, they are temporarily stunned, recovering quickly when the field ceases to exist. One method is to connect the negative terminal of a generator on the bank to a vertical metal plate in the water by means of a cable; the positive terminal is similarly connected to a kind of large landing net, made of wire mesh and handled by a man in a punt. Fish swim into the net and the unwanted are easily disposed of. The other method uses a field strong enough to stun fish. The undesirables having been picked out, the generator is switched off and a move made to
another stretch of the river. Neither method should be used before preliminary experiments on a very small scale have been made to determine the effects produced by fields of different intensity. Otherwise, there is a real risk of causing serious injury to immature fish-and even to the fisherman himself. It is interesting to note that the application of low-intensity fields has proved of great value in some of the salmon rivers dammed under the Scottish hydro-electric scheme. Salmon running up from the sea are kept out of the dangerous tail-races of power stations by negatively charged gratings, whilst positively charged gratings guide them irresistibly to the fish-passes provided for them. Similar arrangements prevent the young salmon smolts on their way down to the sea from being drawn into the intakes of power stations.

## F.M./A.M. Reception

At the moment I am busy arranging for the installation of a receiver for the B.B.C.'s transmissions from Wrotham. (I wonder, by the way, how many of $W . W$.'s far-flung army of readers are puzzled about the pronunciation of that queerlooking name! It's Root-um, the accent being on the first syllable.) I live inside the $3 \mathrm{mV} / \mathrm{m}$ contour, which means that nothing very special will be needed in the way of an aerial; in fact, what I'm putting up is almost too simple to be truebut I think it will work all right. What I'm doing is just this. To a convenient chimney stack a piece of well-painted two-by-two wood, just

over five feet in length, is being firmly fixed. From the frame of the window near which the set will stand a length of ordinary p.v.c.-covered flex runs up to the wooden crosspiece on the chimney. There, the two leads are separated. One runs out on small insulating cleats to the right; the other to the left. And there, provided the dimensions are right, I should have an effective weather-proof and cheap horizontal half-wave dipole. Useful formulx for calculating the dimensions of dipoles are given in the Wireless World Diary.

## Widecombe Fair?

B.B.C. television engineers tell me that a new O.B. unit will enable viewers to see events in the West Country and in Wales. That's firstrate news, for one of the most worthwhile jobs that television can do is to show each half of our world how the other half lives. The West Country should be a rich hunting ground for O.B. producers. Right away I can think of Hungerford's Tutty Men, of the Floral Dance at Helston, of the Devon and Somerset Staghounds, of pilchard fishing, of Cornish wrestling, of Army and Air Force happenings on Salisbury Plain, of Widecombe Fair . . . I could go on and on. And then Wales, with its Eisteddfod, its industries, its mountain climbing, its hill farming and the marvellous bird-life of its off-shore islands. I've always maintained that broadcasts of interesting current events are one of the most attractive possibilities of television. Every use should be made of O.B.s and of mobile film units and their coverage of the whole country should be extended as rapidly as possible.

## NUNCTIONALISM



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## Home Recording

IT WOULD appear from recent correspondence in the columns of Wireless World that for a long time I have been guilty of robbing authors, composers and others, whose works the B.B.C. broadcasts, of the few crusts that their toil brings. For a very long time I have used a tape recorder, operated by a time switch, for the sole purpose of enabling myself to enjoy the breathless excitment of Mrs. Dale's Diary in the evening when I got home. I have always regarded this as, in the words of the Editor, " a delayed method of radio reception."

Even if I am quite wrong in my views I intend to continue doing it. I make no pretence that I record daytime programmes for any other purpose than that of entertaining myself, Mrs. Free Grid and any friends that may be sharing my evening bread and margarine. If I am legally in the wrong I have little doubt that the injured parties will invoke the law to redress their grievances. If Mrs. Dale or her creator would like to make a test case of me I am perfectly willing that they should do so. There is no question of my recording items for private study, research, criticism or any other kind of hypocritical under-the-counter equivalent. I do it for entertainment and for that alone.

The time is coming when every radio receiver will have a built-in tape recorder and time switch so that pre-selected items can be recorded for later consumption by the set owner. The sooner the makers of laws and moral codes recognize this fact the better. I have often wondered why such all-in "receivercorders" are not manufactured as there is no technical reason why they should not be. It is, I suppose, either patent rights or copyright which stands in the way. Probably it is a bit of both. It may be, therefore,

An additional offence.

that I am infringing patents as well as copyright and I must remember to ask the court to take this additional offence into consideration when assessing damages.

## S.O.S. on Land

NATIONAL DISASTERS involving great loss of life such as the Harrow train smash six months ago forcibly bring home to us what very poor use we make of modern radio communication technique. We have heard a lot of talk of costly automatic signalling systems, which will take a long time to install even if eventually decided upon, but nobody in authority seems to have pointed out that by the use of very simple radio methods a repetition of this type of disaster could be avoided.

Every train is already provided with its emergency communication cord whereby anybody on the train can quickly, and simply, bring it to a standstill. There seems to be no reason why this system could not be operated by a short-wave radio signal from any neighbouring signal box by applying well-known and well-tried telcarchic techniques. The simplest of transmitters, set in action by merely pushing a button, is all that would be needed in the signal box.

It would also be possible for each platform of every station to be provided with two or three operating posts of the same type as the familiar strect fire alarm. Operation of one of these by any railway employee who saw the approach of danger would set the signal-box transmitter at work. Had this been available at Harrow, the station master, who heard the third train approaching. would at least have had a chance of preventing the second collision.
Of course, the effect would be to bring to a standstill every train within range of the transmitter and a false call made by some unnecessarily anx:ous official could waste a lot of time and money. Such cases would be rare, however, and the wastage small compared with the benefits.
There are plenty of objections to the idea, both technical and otherwise, and it is up to those who see the snags to put forward improved ideas. One obvious objection is that if the receiver on a train failed through a minor fault the system would become valueless. There is, however, no technical difficulty in arranging that the radio relay on the train normally held the emergency brake in the "off" position so that the development of a fault would result in the stopping of the train.

## Ar Hyd y Nos

WE HAVE read a lot in the newspapers about what is being done to enable the coronation pageantry to be scen by the privileged ones inside the Abbey and by those who can afford the price of a seat on the processional route, complete with champagne lunch. I see that some of the plutocrats of Pall Mall are to have jarm on it, as a television set, complete with hooded screen to shut out the daylight, is to be fitted in front of each seat so that the ceremony in the Abbey can be seen as well as the procession.

But nothing whatever seems to


Home comforts.
have been done for impecunious proletarians like myself who will be compelled to spend a chilly night on the kerb daydreaming like an out-ward-bound sailor of the home comforts we are missing. It is true that the streets are to be liberally supplied with loudspeakers so that we bankrupt bacteria can be kept acquainted with the progress of the day's pageantry. Nobody seems to have suggested that these loudspeakers should be kept going throughout the preceding night to revive our flagging spirits and our flag-wag-weary bodies with music and song.

I cannot doubt, however, that commercial interests will take advantage of such a splendid opportunity and I am taking my personal portable along to pick up Radio Luxemburg which, I feel sure, will be on the air. The programme sponsors will certainly have the good sense to start the night's entertainment with a choir of Welsh singers, and what more appropriate signature tune could they use than Ar Hyd y Nos ("All through the Night").

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# of SOUND REPRODUCTION 




Frequency amplitude curves for the "TREBLE-3" position ( $5 \mathrm{kc} / \mathrm{s}$ turnover). Curves of the some slopes are obtained on the other two positions turning over ot $7 \mathrm{kc} / \mathrm{s}$ and $9 \mathrm{kc} / \mathrm{s}$ (" -2 "' and " -1 "positions).

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The filters consist essentially of twin-T resistor-capacity networks inserted in the return circuit of a single-loop feedback amplifier. The more obvious advantages of this electronic feedback method over conventional choke filters include : (a) Improved transient response characteristics (due to absence of chokes having self-capacitance) and the consequent reduction of "ringing."
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(c) No discontinuities in the rates of slope when the slope control is operated, and no change in signa level at frequencies below turnover. (Both these faults occur in variable-slope choke filters due to the slope control, altering the terminating impedance and the insertion loss.)
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- Normally supplied with seven crystal controlled frequencies between 46-52 cycles.
- Variable oscillator which may be standardised against crystal frequencies by means of Synchroscope comparator.
- Output $100-120$ volts or $\mathbf{2 2 0 - 2 6 0}$ volts at 15 watts with negligible distortion at 300 ohms source impedance. - Auxiliary low voltage output at 5 amps for checking Wattmeters, etc. (Brought out at rear of instrument.) - Eight crystal controlled frequencies between 100 cycles and 12,800 cycles in binary steps. (Brought out at rear of instrument.)
- Provision for measuring the frequency of an external supply by means of Synchroscope comparator with same degree of accuracy as crystal controlled frequencies.
- Selection by single switch. Accuracy better than . 01 per cent.
- Built in voltmeter.
- Operation from A.C. mains 200-250 volts, 40-100 cycles.
- Warming-up time approximately $\frac{1}{2}$ minute.
- Consumption 170 watts. Continuously rated.
- Cased for portability in teak, with adequate ventilation for operating in different positions.

is to be found in its Transformer, which, like the human heart must keep the lifeblood flowing evenly and constantly under both normal and abnormal conditions.


## POTTED COMPOUND FILLED TRANSFORMERS AND CHOKES

manufactured by WODEN have been designed to fulfil such a purpose, and for this reason they have been standardized by many leading Radio, TelevisionandElectronicmianufacturers and also Government Research Departments. Their choice is only made after exhaustive tests for accuracy and reliability. To merit this confidence, there is a constant need to provide components of the highest quality and our ample research and testing facilities ensure continued progress in this direction.

TONE CONTROL NETWORK
Can be fitted into any circuit and will give independent treble and bass lift. Kit comprises 2 Condensers, 2 Resistors, and 2 potentiometers. Complete with instructions only $8 / 6$. Instructions available separately $1 / 6$.


BUILD AN OUTPUT METER Switch, Transformer, meter and rectifier to build a dual range O.P Meter with instructions. £1/15/-.

## TWO ITEMS F OR V.C.R. 97 USERS

(I) RF. E.H.T. UNIT

To take the place of the mains E.H.T. transformer, has the following advantages.
(a) Is more reliable.
(b) Is cheaper
(c) Can be repaired.

Complete kit comprises 2 valves, smoothing condenser, filament transformer and all necessary parts. Price $20 /$ - plus $1 / 6$ post. Constructional and operational data ree with kit or available separately price 2/6.
(2) INTERNAL MAGNIFIER KIT
The kit comprises a veneered and polished wooden surround, special mask, oil filled enlarger, and four chrome-head fixing screws.
Has these advantages :-
(a) It gives the impression of being a standard 9in. tube.
(b) Saves the cost of a 6 in . mask. (c) Protects magnifier from accidental damage.
(d) Is equally suitable for use with a 9in. tube.
Price of kit $39 / 6$ plus $2 / 6$ post and insurance.

SPRING
LOADED
BLOCK
Fully insula -

for mains,
terminal point
fitted on bench of workshop or laboratory. Also suitable for temporary hook ups when testing components etc., will save its cost the first week of use. Price $3 / 6$.

## COIL PACK SNIP



Manufactured by quite a famous company, this 3 wave Coil Pack incorporates a gram position and Long, Medium and Short wave band, designed for $465 \mathrm{kc} / \mathrm{s} \mathrm{IF}$. Brand new and fully guaranteed, Complete with circuit, only 19/6, limited quantity so act quickly.

## ELECTRONIC PRECISION EQUIPMENT LTD

## ELPREQ LEAD AGAIN!



Not a kit but assembled, tested and ready for use, soon as tape is fitted. The new " Elpreq " Tape Recorder is a 4 stage unit of advanced design employing a 12 AT7 valve for 1 st and 2 nd Amplifying stages followed by a ECL80, the triode section of which is used for further amplification and correction, the pentode section acting as R.F. Bias oscillator in the Erase and Record position and as output valve in the play-back position.
In spite of the use of double yalves and high gain circuitry, exceptional freedom hum and microphony has been achieved.
The signal is fed into double triode 1st section, RC coupled to 2nd section. Particulat care has been taken with the dimensioning of the H.T. supply circuit in this stage to ensure absolute stability and minimum hum pick-up. The 2nd stage which is gain controlled is fed R/C. coupled to the 3rd stage in the diode of which a variable tone control network accommodates various tape peculiarities. In the Record and Erase position, the pentode section of the ECL 80 acts as a R.F. Bias Oscitlator (Hartley). In the play-back position this section acts as an output pentode with a degree of bass boosting applied in the grid circuitry to ensure high quality reproduction. A small degree of negative feed-back is applied in this stage.

R1155 COMMUNICATION RECEIVER FOR ONLY $£ 2 / 14 /-$
 DEPOSIT This set, as most will know, is conof the to be one of the finest
communications communications able to-day, The frequency range is $75 \mathrm{kc} / \mathrm{s}$ to 18 $\mathrm{Mc} / \mathrm{s}$. It is complete with 10 valves and is fitted in a black metal case. Made for the R.A.F., so obviously a robust receiver which will give years of service. Slightly used but completely overhauled and guaranteed in perfect working order. PRICE $£ 7 / 19 / 6$ or will be sent against a deposit of $£ 2 / 14 / \mathrm{m}$, balance of 12 monthly payments of $11 / 6$ If you cannot call to collect, please include an additional 10/- to cover cost of transit and carriage. This partly returnable to you if and when you return the transit case.

MAINS POWER PACK FOR R. 1155
With Pentode output stage. Plugs into socket on receiver so no internal modifications are required. Price $£ 5 / 10 /$ - complete with 5 in. speaker ready to work, carriage $3 / 6$.
A few picked out at random from our

| Position | Poles | Banks | Spindle Length | Price |
| :---: | :---: | :---: | :---: | :---: |
| 2 | 2 | 1 | 星 | 2/2 |
| 2 | 2 | 1 | $1 \frac{3}{3} \mathrm{in}$. | 2/9 |
| 3 | 3 \& S. Link | 1 | 1 l in. | 2/2 |
| 3 3 | 3 | 1 |  | 2/2 |
| 3 | 4 | 1 |  | $2 / 6$ |
| 4 | 4 | 2 | $2 \%$ | 2/9 |
| 4 5 |  | 3 | $1 \frac{1}{2}$ in. | 3/3 |
| 5 5 | $2 \& 2^{1}$ Links | 1 | lin. | 1/6 |
| 5 | $2 \& 2{ }_{7}$ Links | 4 |  | $2 / 2$ $3 / 9$ |

EXPERIMENTERS SWITCH ASSORTMENT No. 1.
1 each of the above types, Price 19/.
EXPERIMENTERS SWITCH ASSORTMENT No. 2.
A useful assortment of 10 new switches from which most types of switch can be constructed, by breaking down and re-building. Bargain price $7 / 6$ for 10 .


LAST FEW. E3/19/6. LAST FEW The Lectross Warms room as it dries clothes. Size 3 ft . Wide, 3 ft . high and 5in. deep. It has four stove cnamelled rails and works off AC or DC mains, consuming 650 watts. Fully guaranteed. Price £3/19/6 plus 7/6 carriage.


EXTENSION SPEAKER IN METAL CABINET
This has a $6 \frac{1}{2}$ in. P.M. Goodmans Speaker, heavy magnet type, complete with an output transformer. It is fitted in grey steel case with perforated front and back, ideal for P.A. work in canteens, clubs, etc., $27 / 6$ each

P.M.

SPEAKERS
2tin. .. 15/6 31in. .. 13/6 5in. .. 12/6 $\begin{array}{lll}6 \frac{1}{i n} . & 12 / 6 \\ 8 \mathrm{in} . & \text {.. } & 15 /-\end{array}$ 10in. . . 25/-

RELAYS


EYING GEAR BOX Ratio per tourn.
Suitable for adapcoil counter, rev counter and many other interesting uses. Price $\mathbf{8 / 6}$ post free.

## MAINS TRANSFORMER

 SNIPHeavy duty mains transformer, with 5 secondaries, suitable for big amplifier T.V. etc., Primary, 200,220 \%. Secondary 1. 350-0350 v. at $200-250 \mathrm{~mA}$. S2. 6.3 v . at 5 amp . S3. 4 v . at 3 amp . S4. 4 v . at 3 amp . S5, 5 v . at 3 amp . Half shrouded drop through chassis mounting. Price 29/6 plus I/6 postage.

## ELECTRONIC PRECISION EQUIPMENT LTD



SERVICE DATA
100 service sheets, covering British receivers which have been sold in big quantities, and which every service engineer is ultimately bound tc meet. The follow ing makers are included: Aero dyne, Alba, Bush, Cossor, Ekco, Ever-Ready, Ferguson, Ferranti, G.E.C., H.M.V., Kolster Brandes, Lissen, McMichael, Marconi, Mullard, Murphy, Philco, Philips, Pye, Ultra. Ưndoubtedly a mine of information invaluable to all who earn their living from to all who earn their living from
radio servicing. Price $£ 1$ for the radio servicing.
complete folder.
complete folder. 2 consists of 100 data sheets covering most of the popular American T.R.F., and superhet receivers " all dry," etc. which have been imported into this country. Names include Sparton, Emmerson, Admiral, Crossley, R.C.A. Victor, etc. Each sheet gives circuit diagrams and component values, alignment procedure, etc., etc. Price for the procedure, etc., etc. Price free.


AUTO-
MATIC STARTER
For remote control of D.C. motor between 1 and 3 kw ., adjust ment for 100 v or 230 v .
Unused andin first class con-
dition, complete with metal and wired glass cover. request.


MOTE
MOTE
COT
$\mathrm{T}_{\mathrm{W}} \mathrm{R}_{\mathrm{i}}^{\mathrm{O}} \mathrm{L} \frac{\mathrm{L}}{\mathrm{h}}$ only one pair of $\begin{array}{llll}\mathrm{w} & \mathrm{i} & \mathrm{r} & \mathrm{e} \\ \mathrm{a} \\ \mathrm{n} & \mathrm{d} & \\ \mathrm{a}\end{array}$ simple push button you any one of four stations without leaving your armchair. This is iust one of the many applications of our impulse relay. There are many other purposes to which it can be put. Note they are somewhat soiled due to storage but mechanically O.K. Price 2/6.

## TWO-VOLT

ACCUMULATORS
Made for the Forces by one of the most 15 amp .-hour size approx. 15 amp.-hour size approx. bin. $\times$ lit. ebonite case, pre-charged, only need filling with acid, $4 / 9$ each, plus 9 d . post and insurance.


PLUGS FOR MODERN VALVE HOLDERS Each is
fitted with a rubber shroud.
For $B 7 G$ button base. Price $1 / 4$ each, discounts for quantities.

WHITEWOOD CABINETS
Available by post or from our Ruislip Depot only. (Carriage 4/6 per cabinet).

Hundreds of uses in the home, office or workshop; for instance two of these and $3 /$ - worth of hardboard (obtained locally) make an excellent desk or workbench. Sturdily constructed of well seasoned softwood frames and fronts, birch ply top, sides and doors, hardboard backs and bottoms. External dimensions $29 \mathrm{in} . \times 14$ in $\times 131 \mathrm{in}$.


Model 314
MODEL 313B. Sturdy cupboard and drawer. Fitted full shelf inside the cupboard. 42/6. MODEL 313C. Drawer space and cupboard, half shelf inside cupboard. Price 39/6. MODEL 314, Five drawer chest, suitable for personal papers, etc. Price 54/-.


Model 313C
MODEL 312. Has platform drawer on runners in top space, cupboard underneath. Ideal record player cabinet. Price $54 /=$. MODEL 313A. Well built cabinet with shelf and cupboard. Fitted half shelf inside cupboard. Price 37/6.


Model 314 \& 313 as desk


HANCE OFFER a large Last year we purchased a Collaro Auto quantity of the Collaro Auto 3 speed suitable for all types 3 speed suitable for all types of records with the latest crystal pick-up but these have been selling very rapidly and it may well be that unless you buy one this month you will not be able o again, at this special price. We urge you therefore to order right away, the price is 11 gns., plus $7 / 6$ carriage and insurance.

## 29 GNS.

## AUTORADIOGRAM

29 GNS.

## CONSOLETYPE CABINET

With full grained walnut finish, will take standard type auto change gram unit. Price £11/10/\% H.P. terms £3/17/deposit, and 12 monthly payments of $16 / 9$, plus $15 /$ - carr. RADIO CHASSIS TO SUIT. Long, Medium, Short wavebands, three colour illuminated sands, three colour Model. Price scale. AC/DC Model. Price
£8/19/6. H.P. terms $£ 3$ deposit and 12 monthly payments of $13 / 2$ plus $7 / 6$ carr. A.C. only, model. Price £9/19/6. H.P. terms £3/7/deposit and 12 monthly payments of $14 / 6$ plus $7 / 6$ carriage. AUTOCHANGE UNITS. Collaro 3 speed for long playing and standard records, comolete with dual purpose head. Jrice 11 gns.
SPECIAL OFFER. Cabinet, A.C./D.C. model Radio Chassis; and Auto Changer. Price 29 gns H.P. terms $£ 10 / 14 /-$ deposit and 12 monthly payments of $£ 2 / 3 /=$ plus $£ 1$ carriage and insurance. Booklet of photos circuit diagrams, etc., $2 / 6$ (returnable)


NEW 12in. T.V. TUBES A special purchase enables us to offer some brand new, latest type 12 in . T.V. tubes at the special price of $£ 15$ each.
Each tube is in original carton and complete with makers guarantee.


12 CELL ACCUMULATOR This accumulator can be coupled up to give $24 \quad v$. with all cells connected in series or 12,6 or 2 volts in series parallel arrangements They were originally mens, They pareinally made for Admiralty by a leading manuacturer, have never been filled, and are in excellent condition. Each is contained in a wooden crate as illustrated. Post and insurance 2/6. Price 27/6.

FLEXIBLE COUPLINGS
(\%)7T1)Th? These are sometimes known as bellows couplings because they will extend as well as bend. They are ideal for joining shafts which are out of alignment and for slug tuning controls where the core has to come in and out. Price $1 / 9$ each.


ALL MAINS CHASSIS
This is the equivalent of a 4-valve receiver for it uses three valves and a metal rectifier. It is all wired up ready to work off A.C. mains, complete with valves, ganged tuning, dust cored coils, on metal chassis. Tunes long and medium wavebands. Large clear dial. Receives Home Services, Light programme, LuxServices, Light programme, Lux-
emburg, etc., Chassis size approxemburg, etc., Chassis size approx-
imately $9 \mathrm{in} . \times 4 \mathrm{in}$. $\times 5 \mathrm{in}$. imately 9in. $x$ in. $x$ in.
Complete with valves, but less Complete with valves, but less
speaker. Obviously not the speaker. Obviously not the last word in receivers, but useful standby or for workshop, bedroom or even the greenhouse where a radio helps to make the job more pleasant. Few only left to clear at $49 / 6$ plus $3 / 6$ packing and insurance. Suitable speaker with matching transformer, 16/6. Nothing else needed.

PYREX
AERIAL
INSULATORS


Ideal for aerial connections through cabin walls or through panels. Consists panels. Consist of glass come with threaded rod and terminal ends, and metal fixing flange
Price $2 /-$ each.

## ELECTRONIC PRECISION EQUIPMENT LTD

## ELECTRICAL BARGAINS

In addition to our large range of radio accessories we also carry a good stock of electrical wiring accessories details of a few of these can be found below:WOOD BLOCKS
Varnished Walnut.
Size New Scratched $2 \frac{1}{2}$ in. $\times 2 \frac{1}{3}$ in. $\times 1$ in 3in. $\times 3$ in. $\times \operatorname{lin}$. 6in. $\times 3$ in. $\times 1$ in. $9 \mathrm{in} . \times 3 \mathrm{in} . \times 1 \mathrm{in}$. 3 in. $\times 1$ in.

| 7 d. | 5 d. |
| :--- | :--- |
| 8 d. | 5 d. |
| $1 / 1$ | 8 d. |
| $1 / 4$ | 11 d. |
| $8 \frac{\mathrm{~d} .}{}$ | 6 d. |

## CARBON BRUSHES

Pre-bedded with springs.辛in. $\times$ tin. $\times \frac{3}{3}$ in. 8 d . pr. $\frac{1}{4}$ in. $\times \frac{1}{8}$ in. $\times \frac{\text { sin. } 6 \text { d. pr. }}{}$

## RUBBER TAPE

8 oz. reels 6d. each.

## ARROW ROTARY SWITCHES

4 position on/off hot/cold. Suitable for hair dryers etc., Price 6/6.

## ROSS COURTENEY TAGS

Packet of 100 assorted Price 3/6. rev counter and many other interesting uses. Price $8 / 6$ post free.


SURFACE SWITCHES SWITCHES
-HICRAFT OblongBrown Plastic 1-way, $1 / 3$ each. Oblong White Plastic 1-way,
Oblong Brown 2 way Oblong White 2 way Round Brown 1 -way Round Brown 1-way
Round White 1-way Round White 1-way
Round Brown 2-way Round Brown 2-way
Round White 2-way

## SOCKETS-

Flush type for skirtings 5 amp. 3-pin shuttered, 1/3 each ; ditto with switch. $2 / 3$ each.


CEILING SWITCHES HICRAFT With cord and acorn. Brown or acorn.
White,
1-way,
$3 / 9$ $\begin{array}{ll}\text { White, } & 1 \text {-way, } 3 / 9 \\ \text { each; } & 2 \text {-way, } 4 / 3\end{array}$ each; each.

## LAMP HOLDERS

Bakelite 1/- each or $10 / 6$ doz. Bakelite skirted Batten holder 1/6 or 15/- doz
Bakelite type threaded for ${ }_{8} \mathrm{in}$. with HO skirt, $1 / 6$.
10 per cent. discount if bought in dozens.

## ADJUSTABLE <br> ADJUSTABLE

250 v. heavy
silver contacts
can be adjust-
between $70^{\circ}$ -
$300^{\circ} \mathrm{F}$. These are suitable for aquarium heaters, electric blankets, etc.
1 Amp. Model 3/6.
2 Amp. Model 5/6.
2 Amp. Modedel $5 / 6$. $14 / 6$, post etc.,
6d. each.


Special this month is the Portable illustrated alongside. We offer a bakelite cabinet with carrying handle, metal chassis, battery housing and two waveband dial, all for $27 / 6$ plus 5/-carr. and insurance. This cabinet and set of parts is ideal for making up either an all dry battery receiver for holidays, pienics etc., or a battery mains set for everyday use. Constructional details of two suitable circuits using 1.5 V. valves 1 R5, etc., will be given free with cabinet assembly, or is available separately price $1 / 6$ post frec.

SPECIAL RADIOGRAM OFFER
To those who want an auto radiogram at a low price, we offer the cabinet illustrated alongside complete with Collaro three speed record changer with dual purpose crystal pick up, at a special bargain price of $£ 17 / 16 / 8$ plus $12 / 6$ carriage and insurance or plus i2/6e c6ill 3 colour scale, scale pan, chassis, pulley, driving head, springs, etc., etc., to radio cabinets are available as a parcel at $15 /-$, plus $1 / 6$ post.

Cabinct separately $£ 7 / 10$ /- (or £2/10/- deposit) plus 10/- carriage and insurance


TV OFFER-LAST CHANCE
The 12 in . T.V. with front flap for controls as illustrated alongside is still available at $£ 7 / 10 /-$ or $£ 2 / 10 /-$ deposit plus 10/- carriage and insurance, but stocks are rapidly going and this in all probability is your last chance to secure one of these really superior cabinets.


The table model also illustrated is still available in fair quantity at £ $3 / 17 / 6$ plus $7 / 6$ carriage and insurance, which price includes the armour plate glass and surround.
Mechanical details for the Console or Table Model are available as a parcel: Punched and prepared metal chassis, punched outrigger valve plate with spacers, 12 in . Tube Clamping ring, tube rear support brackets, etc. Price $25 /$ - plus $2 / 6$ post.

15in. MAGNETIC TELEVISION TUBE.
By famous maker, as used in many popular Television receivers (list on request). Specification Blue/White screen 9 Kv . ion trap triode, heater 6.3 v . at .55 amp ., 50 deg. deflection. New, with written guarantee offered at approximately half price, $£ 12 / 10 /$ - each, plus 10 carriage and insurance. H.P. monthly payments of $18 / 3$.
Limited quantity so order
 immediately

FREE. With this tube we will give free a circuit diagram of a complete T.V. designed around the tube.

## STOP PRESS !

Terrific Bargain through special purchase from manufacturer. Famous Hale Electric "Midgetronic" radio cabinet, bakelite, size $7^{\prime \prime}$ wide, $4^{\prime \prime}$ deep and $8 \frac{1}{2}{ }^{\prime \prime}$ high, with scale, pointer, punched metal four-valve chassis and hardboard back. Price only $15 /$, plus $2 / 6$ post and insurance.

ONLY A LIMITED QUANTITY-SO ORDER TO-DAY!
To ensure receiving prompt reply, please enclose stamped addressed envelope, when writing for

IDEAL GENERAL PURPOSE RECEIVER


The Elpreq "Wolsey" 5 valve A.C./D.C. superhet has a built-in aerial and is of convenient size and weight to carry from room to room. Powerful reception on long, medium and short waveshandsome wooden cabinetilluminated glass dial, with stailluminated glas Wial, with ion names, A.V.C. and usual refinements. Size 11 in . $\times 5 \frac{1}{2} \mathrm{in}$. $\times$ 7 in . with B.V.A. valves, 12 months' guarantee. Limited quantity only. $59 / 5 /-$ or $£ 3 / 2 /-$ deposit and balance over 12 months, carriage and insurance, 5/-


EX-ROYAL NAVY SOUND POWERED TELEPHONE
These require no batteries, and will go for long periods without attention. Complete with generator and sounder which gives a high pitched note, easily heard above any other noise. Also fitted with an indicator lamp which in quiet situations can be used instead of the sounder, or where several 'phones are used together will indicate which one is being called.
Size 73 in . $\times 9 \mathrm{in}$. $\times 7 f \mathrm{in}$., wall Size 7 in. $\times 9$ in. $\times 7$ in., wall
mounting, designed for ships use, but equally suitable for home, office, warehouse, factory, garage, etc. Price $57 / 6$ each, plus $4 / 6$ carriage.

## -GREATLY REDUCED- <br> CATHODE RAY TUBES

VCR97. Brand new and unused, ideal for 'scope, etc. Price $12 / 6$. Carriage and insurance $5 /$ - extra.


VCR5I7. Blue and White $6 \frac{1}{2}$ in. guaranteed full picture. 29/6 plus 5/- carriage and

VCR139A. 2 in. $32 / 6$ plus $2 / 6$ carriage, etc. VCR138. $3 \frac{1}{2} \mathrm{in}$. electrostatic short persistence, suitable for T. $V$. and ideal for scope work, $37 / 6$ plus $3 / 6$ carriage, etc.

VCR112. 5in. electrostatic, persistence not known, 15/- each plus 5/-carriage, etc.
CV996. 6in. electrostatic, persistence not known, 15/- each plus 5/-carriage, etc.
CV1140, CV1590, CV1546. All 12 in . magnetic long persistence, $£ 2 / 10 /=$ plus $10 /$ - carriage.


CHASSIS ASSEMBLY
3 colour, 3 waveband scale covering standard Long, Medium, and short wavebands, scale pan, chassis punched for standard 5 valve superhet, pulley driving head, springs etc., to suit Scale size $14 \frac{1}{2}$ in. $\times 3 \frac{1}{2}$ in. Chassis size, $15 \mathrm{in} . \times 5 \mathrm{in} . \times 2 \mathrm{in}$. deep. Price 15/- plus $1 / 6$ post. Note this is the one that fits ou C7/10/- Radiogram cabinet.


DEMOBBED VALVES Gives the commercial equivalents of many thousands of Service Valves, an invaluable publication recently revised. Price $2 / 3$.


TOGGLE SWITCHES
Metal body standard size, made by a leading maker. Available with round dolly or with special $V$ cut dolly. State which type when ordering. Price while stocks last only $2 / 3$.


RRASS CASED PLUG Seven-way brass cased plug, ideal for portable apparatus. Price $2 / 3$ each half. D33BR and D3313L.

GERMANIUM DIODES.
Wire Ended
Several alternative types available at bargain prices.
Types Equivalents Price
Red CG4GEX55, WG7A 10/-
Green CG5, GEX45,
WG7B, GEX33, WG4A
Orange CG6, GEX45, $\quad 7 / 6$
Green For super detectors 5/6
Yellow For Crystal Re4/6
Blue For General Experimental purposes $2 / 3$
Parcel containing one each of the above. Price 30/-.

## THIS MONTH'S SNIP

## ELECTRONIC PRECISION EQUIPMENT LTD

Combined 15 in . Television and 2 wave Radio chassis. Brand NewReady to work-Fully guaranteed. Complete with Tube, Mask, knobs, and speaker, Adjustable to any channel. Price £55, carriage and ins. $f 2$.


## TECHNICAL DESCRIPTION

TELEVISION. This is an A.C. chassis of Superhet Design employing 14 valves viz. 7, 6F1 Mazda; 1, CP25 Mazda; 2 6D2 Mazda; 16 L 18 Mazda; 1, EY51 Mullard; 1 EL38 Mullard; 16 SN7 Brimar ; Tube is Cossor 85 K
Special' Features : Focus, PM with pre-set centring and tilting facilities, and continuously adjustable focus control, accessible to the user.
Picture Tone Control: 3 position Gradation Control which adjusts the level of vision interference suppression.
Aerial Attenuation: A screen Aerial Attenuator Box is provided on the chassis for use in the vicinity of transmitters.
Sensitivity: 50 micro volts for peak white on the screen giving extremely good results in fringe areas.
extremely good results in iringe areas. Picture Centring: 2 slider controls accessible from the
the picture to be moved horizontally or vertically to the centre of the screen.
RADIO. This is a 4 Valve A.C./D.C. chassis employing superher principle. A two wave band tunable dial for the long and medium waves.


#### Abstract

PROFESSIONAL RADIOS YOU CAN MAKE You will find that the building of our all-mains radio receivers $s$ simplicity itself, and the more you make the less time each takes, every bolt is supplied last nut and bolt is supplied. and everything fits together in a professional manner. When finished the receiver looks and plays as well as those being offered in radio shops at anyhing between 10 and 14 . The one illustrated above we call the "Occasional," in a choice of colours, Ivory or Walnut and the T.R.F "Occasional" in a choice of colours, Iv $£ 2 / 1 / 6$ deposit and 10 monthly payments of $10 / 6$.




EX-GOVT. BATTERY SET Medium waveband superhet complete with 4 Mazda valves, speakers, etc. Storage soiled, complete but may need attention, few only to clear. Price 55/0 plus 5/- carr. Ditto less speaker, Price 45/- plus 5/- carr.

LAMP HOLDERS
We carry a large assortment of dial lamp holders. Price 6d. each. Order by list numbers.


FUSE BOARD
In metal case in metal conduit knockouts, 12:6.


TELEPHONE JACK PLUGS As illustrated 7d. each. Sockets to suit 10 d . each.

## ELECTRONIC PRECISION EQUIPMENT lid.

Post orders should be addressed to:-
ELPREQ HOUSE (Ref. 2), HIGH STREET, WEALDSTONE, MIDDX.

Personal shoppers however must coninue to call at:-

## 42-46, WINDMILL HILL, RUISLIP, MIDDX.

Phone: RUISLIP 5780
(Half-day, Wednesday),
152-153, FLEET STREET, E.C.4.
Phone: CENTRAL 2833
(Half-day, Saturday)


Ia

Bargains in Ex-Service Radio and Electronic Equipment

## RECEIVER UNIT TYPE 25

Ref. IOP/IL part of the TRII96 equipment. Frequency range, $4.3-6.7 \mathrm{Mc} / \mathrm{s}$
Valves: 2 VR53 (EF39), 2 VR56 (EF35), VR55 (EBC33), VR57 (EK32), 2 IFT $460 \mathrm{Kc} / \mathrm{s}$, plus (EBC33), VR57 (EK32), 2 IFT $460 \mathrm{Kc} / \mathrm{s}$, plus
various microdensers, mic. and output transvarious microdensers, mic. and output trans-
formers, pots., condensers, resistors, etc., connecformers, pots., condensers, resistors, etc., connec-
tions brought to 10 pin Jones type chassis plug. tions brought to 10 pin Jones type chassis plug. enclosed in case, $8 \frac{1}{2} \mathrm{in} . \times 6 \frac{3}{4} \mathrm{in}, \times 6 \frac{1}{4} \mathrm{in}$. Would make the basis of an All Wave Receiver, required tuning pack and power supply.
Ask for Price $35 /=$
Post
No. X/H29 Paid
(Conversion Data ("R.Con." Reprint). I/6) SPARES KIT
For R.C.A. Type TEI49 Crystal Wavemeter A wood case, dim.; $11 \frac{1}{2}$ in $\times 10 \frac{1}{4}$ in. $\times 4$ in. containing A wood case, dim.; $11 \frac{1 i n}{2} \times 10 \frac{1}{4} \mathrm{in}$. $\times 4$ in. containing
Crystal (in holder) $1,000 \mathrm{Kc} / \mathrm{s}$ plus and minus Crystal (in holder) $1,000 \mathrm{Kc} / \mathrm{s}$ plus and minus
.005 per cent. ( $(\mathrm{V}$. cut low temp. co-efficient), 2 . 005 per cent. (V. cut low temp. co-efficient), 2 switches, 2 pole, 2 way, 5 fixed resistors. A.F. transformer. A.F. choke, variable inductance, 2 fixed inductances, S.C.C. jack, valveholders I.O., 4 knobs, terminal, 4 pin plug, etc. Ask for No. X/H917

49/6
Carriage
Paid
V.H.F. R/TI32 Rack Mtg.
V.H.F. R/T RECEIVER UNIT

Range $100-124 \mathrm{Mc} / \mathrm{s}$
A 10 valve superhet with 4 VR53 (E539), VR54 (EF34), VR57 (EK32), 2 VR65 (SP61), VR66 (P61), VR67 ( 655 G ), plus stabiliser VS70' (7475), " S ", meter, screened R.F. section B.F.O., etc. etc., in enclosed chassis, size 19 in. $\times 19 \frac{1}{2}$ in. $\times 11$ in., finish dark grey. Circuit supplied.
Ask for
No. X/H915
2.19.6 $\qquad$ Carriage MILLIAMETER RANGE 0-5 Ask for 2 in. square drilled flange
No. X/H53
12/6
Each
Post
Paid

MFG. BANKRUPT STOCK BARGAINS TUBULAR ELECTROLYTIC CONDENSER (All. Can)
Capacity 8 -16 mfd., max. wkg. 450 volts, Clip mtg. Ht. $2 \frac{5}{8} \mathrm{in}^{2}$., dia. I I in. Aluminium, case.
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Heavy Duty 24 Volts Starter Motor, as used in Aircraft, with recessed splined spindle and mounting by 6 Bolts, Dim. $: 8 \frac{1}{2} \mathrm{in}$. long, $4 \frac{1}{2} \mathrm{in}$. dia. Ask for $10 /=\quad$ Carriage $\begin{array}{ll}\text { No. } \mathrm{X} / \mathrm{H} 870 & 10 /=\text { Each } \quad \text { Paid }\end{array}$

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AS A COMMUNICATIONS RECEIVER with 9 valves for $\mathbf{2 0 0 - 2 5 0}$ v. A.C. Mains. Comprises RECEIVER UNIT RII 55 Plus

| Plus $\qquad$ | OWER PACK |  |
| :---: | :---: | :---: |
| Appearance as new <br> Ask for No. X/E6 AC | Borh Units | $\begin{aligned} & \text { Carr. } \\ & \text { Paid } \end{aligned}$ |
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| RII55 RECEIVER | R UNIT ONLY |  |
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£7.19.6
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R.F. UNIT TYPE 24
 Switched tuning, 5 pre-tuned spot. freq., 3/VR65s (SP61), Cutput approx. $7-8 \mathrm{Mc} / \mathrm{s}$. in metal case, $9 \frac{1}{2} \times 7 \frac{1}{2} \times 4 \frac{3}{4} \mathrm{in}$. In Orizinal Carton $\quad 30 /=\quad$ Post Ask for No. X/H850 $\quad 30 /$ Each $1 / 6$ R.F. UNIT TYPE 25

Frequencies covered $40-50 \mathrm{Mc} / \mathrm{s}$. (6-7.5 metres). Otherwise as Type 24 above.
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Unit with all valves. As new... 63000
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Superhet. Mediwave bands. $200-$ 500 M . and 19-50 M. Uses 1 each 1R5, 1T4, 1S5 and $3 \mathrm{Kc} / \mathrm{s}$. valves. ${ }^{\text {I.F. }} 465$ availabie.
Fully assembled Fully assembled and wired, Ready to fit into a vabinet. H.T. 90 volts D.C., L.T. Size of chassis :9in. wide, $5 \frac{1}{2}$ in. deep, 2in. high, Gins overall height.
LASKY'S PRICE 85/-
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YOUR BATTERY PORTABLE CAN OPERATE FROM THE MAINS.
A.C. MAINS POWER BRAND NEW. IN ORIGINAL CARTONS
Size: 5 in. wide, 2 is in . high, 7 票in. long. Input 200/250 volts A.C Output: H.T. 90 volts, D.C.; L.T. 1.4 volts D.C.

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TRANCFORMERS
6.3 v. $1.5 \mathrm{a} ., 7 / 11$.
6.3 v. 3 a., $12 / 6$.

Special Transformer. 2 amps, with the following tappings: $3,4,5,6,8,9,10,12,15,18,20$, 24 and 30 volts. Price $17 / 6$.

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 Midget. $465 \mathrm{Kc} / \mathrm{s}$. Iron dust cores in can, size :-1in. $\times 1 \mathrm{in} . \times$ 2tin. high. $12 / 6$ per pair WEARITE TYPE 500. Range $450-470 \mathrm{Kc} / \mathrm{s}$. Dust cored compression trimmer tuned. In can, size: $-3 \frac{1}{2} \mathrm{in}$. high, $1 \frac{1}{2} \mathrm{in}$. square. $12 / 6$ per pair.WEARITE TYPE 550. Range 445-520 Kc/s. Permeability tuned. In can, size as for 500 series. 12/6 per pair.

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MBA/6. $\quad 350-0-350$ v. 100 mA , 6.3 v. 3 a., 5 v. 2 a. With mains tapping board. Price 22/6. MBA/7. $\quad 250-0-250$ v. 80 mA , 6.3 v. 3 a., 5 v. 2 a. Both filaments tapped at 4 volts. Price
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chrome plated.
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Fused, with fully retractable points. Price 4/11 per pair.

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CONDENSERS, . 0005 mfd . No. 1. Miniature. With Perspex dust cover and trimmers. Size : $1 \frac{1}{2} \times 2 \times 1 \frac{1}{2}$ in., $\frac{1}{1} \mathrm{in}$, spindle. LASKY'S PRICE

86
No. 2. Midget. With trimmers. Size: $24 \times 1 \frac{1}{4} \times 1$ in., $\frac{1}{3} \mathrm{in}$. spindle. LASKY'S PRICE

86
No. 3. Midget. Less trimmers. Size: $2 \times 1$ 㝵 $\times 1$ in., din. spindle. LASKY'S PRICE

66

No. 4. Standard type. Size: $2 \frac{1}{2} \times 2 \frac{9}{4} \times 1$ in., tin. spindle. LASKY'S PRICE

66


## R. 1155 RECEIVERS. BRAND NEW. AERIAL TESTED BEFORE DESPATCH

These well-known ex-Air Ministry Receivers need no further introduction.
Supplied complete with 10 valves.

| LASKY'S | P11. 19.6 |
| :--- | :--- | :--- |
| PRICE | 87 |
| USED | 19.6 |

Carriage $12 / 6$ per unit extra.

Fully Assembled Power Pack and OutPower Pack and
put Stage, for R. 1155 put Stage, for use on 200- 250 volts AC mains.
LASKY'S $\mathbf{7 9 / 6}$ LASKY'S $\quad 79 / 6$
PRICE
Carriage $5 /$ - extra.

INTERCOM UNITS
4-station operation. For use on A.C. D.C. mains $200-250$ volts. Supplied complete, with 3 new valves, ready for immediate installation. Fitted in attractive plastic cabinet.
Suitable for use as baby alarm.
MASTER UNIT $£ 7 / 15 /-$. Carr. Extension Units. Price 21/each complete. Carriage 2/each extra.

AMPLIFIERS. Fully assembled and wired.
25 Watt Model. By Romac, with radio tuner, long and medium wave. 7 valves, including 2 6L6 in push-pull. Provision for high and low impedance microphone. Absolutely perfect, new and unused. For $200-250$ volts 50 c.p.s. mains. LASKY'S PRICE £25 Carriage extra. complete. MODEL J/RA/2. 10-12 watt. A LASKY'S PRICE $£ 7196$ Carriage 10/- extra.
4-Watt Model. Ex-Government. Complete with 10 valves: 2 25L6, $16 \mathrm{H} 6,125 \mathrm{Z} 6,66 \mathrm{SK} 7$. For operation on 110 volts A.C./D.C. Balance and pushpull. High, medium and low impedance inputs, A.G.C., etc. LASKY'S PRICE £7/19/6 complete. No circuits available.

Carriage 10/- per unit extra.
P.A. SPEAKERS. Re-entrant horn, 10 watts, beam proiector, weatherproof speakers. 7.5 ohm impedance. Horn diameter 14 in . LASKY'S PRICE 99/6
Spigot mounting 10/- extra.
Carriage and packing 15/- each.
TANNOY PRESSURE UNITS 10 watts. 7.5 ohms impedance. LASKY'S PRICE

79/6

10,000 VALVES IN STOCK.
R.V.A., Special Purpose, Trans-
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Raytheon sub, miniature
CK510AX, pentode, 3/11. 954,
956, VU120, 32,57, E1148. All
at $2 / 6$ each. RK $34,2 /-$, MS/PEN;
$3 / 6 ;$ VUi $11,3 / 6$.
(red) EF50 $12 / 6$.

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We now have in stock 3 types, including T.R.F. and 5 channel superhet models. Prices range from £6/19/6 complete. For full description write for our latest list.

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All less o'trans. new and unused. First quality $\begin{array}{lll}3 \text { in. } & 12 / 11 \\ 5 \text { in. } & 12 / 6\end{array}$ $\begin{array}{lll}\text { 5in. } & . . . & 12 / 6 \\ 6 \text { in. } \\ 13 / 6\end{array}$ $\begin{array}{lll}6 \operatorname{lin} . & 13 / 6 \\ 8 & 15 / 6\end{array}$ $\begin{array}{lll}8 \mathrm{in} & \ldots . & 15 /- \\ 10 \mathrm{in} & & 32 / 6\end{array}$

## OUTPUT

TRANSFORMERS

| 40 mA . Multi ratio | 6/11 |
| :---: | :---: |
| 80 mA . Multi ratio | 14/11 |
| 80 mA . Pentode | $12 / 6$ |
| 60 mA . Plessey, 6,000 |  |
| ohms | 5/11 |
| Standard pentode | 4/11 |
| Pentode | 3/6 |
| Midget pentode | 4/3 |
| Miniature pentode. |  |
| PX4 Intervalv |  |
| 5:1 Intervalve | $5 / 1$ | 80 mA . Multi ratio. . 14/11 60 mA . Plessey, 6,000 Standard pentode . Pentode Miniature pentode. 3S4, 1S4 5:1 Intervalve

$8 / 6$
5/11

SPEAKER FRET
Expanded Meral. Silver Finish. $12 \times 12$ in. $3 / 11$. $12 \times 18$ in. 5/11. Plastic. White. $12 \times 5$ in., $2 /-$. Wire. Bronze. $11 \times 8 \mathrm{in}$.

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G.E.C. Type, glass, wire ends $3 / 3$. B.T.H. Type, plastic ...... 3/-.

ION TRAPS. All types available. PRICE 5/- each. State type number of c.r.t.

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> C.R.T. OFFER

Brand new and unused 12 in . ion trap cathode ray tubes. 6.3 volt heater, $7-9 \mathrm{Kv}$. E.H.T. 35 mm . neck. Black and white picture. £11/19/6.
Screen has very slight blemishes. PERFECT
Carriage and insurance 15/- per tube extra.
R1132 RECEIVERS. New, boxed, with all valves. $£ 3 / 19 / 6$. Carriage 10 i-.

Used C.R. TUBES. For Callers only, With heater-cathode and/or ion burns. Suitable for testing or experimental purposes. 9 in.-, $12 \mathrm{in} .55 / \mathrm{l}$.
ION TRAP AND ALUMIN. ISED TUBES. These have only heater-cathode shorts, no ion burns. 9 in., $45 /-; 10 \mathrm{in} ., 55 /-$; 12in., $75 /-$; $15 \mathrm{in} ., 95 /-$.

RF25 UNITS. New, with valves. 19/11. Carriage 2/6.
EX. GOVERNMENT TEST METERS. Brand new and No. 1. AC/DC AVO MINOR 500 ohms per volt. Ranges: $0-5,25,100,250,500$ volts AC/DC. $0-2.5,5,25,100$, $500 \mathrm{~m} / \mathrm{a}$. AC/DC. $0-20,000$ ohms : Up to 10 meg. with external battery. Size: $4: \frac{3}{1} \times 3 \frac{3}{3} \times 1 \frac{3}{2}$ in. Full instructions supplied. £5/19/6 complete. Postage 5/extra. Leather carrying case
$7 / 6$ extra.
soxa

S9/6 complete. Postage 5/- extra.

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 transformer10/6
Scanning coils. High impedance frame, low line. By Plessey
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$8 \mathrm{mfd} .450 \mathrm{v} . \mathrm{w}^{2}$
$8 \mathrm{mfd} .450 \mathrm{v.w}$. 8 mfd .500 v.w.
$16 \mathrm{mfd} . ~$
50
v.w. 16 mfd .350 v.w.
16 mfd .500 v.w. 16 mfd .500 v.w. 32 mfd .500 v.w.
50 mfd.
350 v.w. 50 mfd .350 v.w.
60 mfd .350 v.w. 60 mfd .350 v.w.
$250 \mathrm{mfd} . ~$
22
v.w. 250 mfd .350 v.w $8+8 \mathrm{mfd} .450$ v.w. $8+16 \mathrm{mfd} .500 \mathrm{v} . \mathrm{w}$
$2 / 3112 \times 12 \mathrm{mfd} .350$ v.w. .. $\quad 3 / 6$ $2 / 11 \quad 16+16$ mfd. 500 v.w. ... $\quad 4 / 6$ $\begin{array}{lll}2 / 6 & 16+24 \text { mfd. } 450 \text { v.w. } & \text { m } \\ 3 / 6 & 3 / 11\end{array}$ $\begin{array}{llll}3 / 6 & 32+100 \mathrm{mfd} . & 450 \text { v.w. } & 7 / 6 \\ 4 / 11 & 60+100 \mathrm{mfd} . & 350 \text { v.w. . } & 9 / 6\end{array}$ 4/11 $60 \div 100 \mathrm{mfd} .350$ v.w... $9 / 6$ 3/11 3/11 2/- 25 mfd. 25 BIAS $\quad . . .1 / 6$ 41150 mfd .12 v.w. $4 / 650 \mathrm{mfd} .50 \mathrm{v} . \mathrm{w}$ $4 / 11 \mid 75$ mfd. 12 v.w. ALL BRAND NEW AND GUARANTEED ALL BRAND NEW AND GUARANTEED (NOT EX-
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12 ソ. 1 а. 6/9 12 v. $\frac{1}{2}$ a. 3/6 6 v. 1 a. 3/11 6 v. 4 a. $17 / 6$ 12 v. 4 a. $17 / 6$
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All models available including filter chokes. Price (except Alexandra Palace), 28/per set. Alexandra Palace, 20/- per set L9 RF choke, $2 /$ -

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Sound-Vision, 18/6. Power-Pack/Time base, $18 / 6$.
Support for S.V. chassis, 6/-.

W/B \& PLESSEY
Line EHT trans.
32/6
$3 \mathrm{Mc} / \mathrm{s}$ boor $25 / 6$ choke Width cont... $5 / 9$ Scanning cool. . 10/Main choke .. 33/3 Focus ring Heater trans. WB/ 103 WB/103A Front and rear ront and rear Corts $\begin{aligned} & \text { C.R. }{ }^{2} \text {..... 21/6 }\end{aligned}$

## G.B. T.V. COMPONENTS

Line Trans. .. 29/6 Frame Trans. 22/6 Main choke .. 12/6 Width Coil .. 6/11 Boost Choke. . 3/9
All suitable for Home Construction T/V.

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 SELENIUMThe very latest " Sentercell" S.T.C. range. $\mathrm{K} 3 / 40,3.2 \mathrm{kV}$. $7 / 6$ $\mathrm{K} 3 / 45,3.6 \mathrm{kV}$. $8 / \mathbf{2}$ K3/50, 4.0 kV . $\quad 8 / 8$
K3/100, $8.0 \mathrm{kV} \mathrm{14/8}$
K3/160, $12.8 \mathrm{kV} \mathrm{21/6}$

METAL RECTI-

## FIERS

WX3 and 6. Ea. 14D36
14A86
36EHT45 ..... 20/4
$\begin{array}{llll}36 \text { EHT45 } & \cdots . & 23 / 8 \\ & 26 / 1\end{array}$
36EHT100.... $29 / 6$

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Width per pair 25/Width Control 6/6
NOW AVAILABLE. LARGE SCREEN, WIDE ANGLE CONVERSIONDETAILS FOR THE TAILS FOR TH
VIEWMASTER. VIEWMASTER. for
Send 3d. stamp for full data. Fully itemised price list of all Viewmaster components now available.

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The television set you can build at home from stanbuild at home from standard parts.
A MODEL FOR EACH FREQUENCY. State staFREQ required. State sta-
Brilliant high definition black and white picture. Superb reproduction.
Uses 9in. or 12 in . Cathode Ray Tube. Table or Console Model.
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Television for the home constructor at its finest. Send to-day for the CON STRUCTION ENVELOPE 32-page booklet crammed with top-rate information and all the necessary data, also 8 fullsize working drawings and stage by stage wiring instructions.
Alexandra Palace, Sutton Coldfield, Holme Moss, Kirk o'Shotts, Wenvoe, Pontop Pike, Belfast. State model required.

PRICE $7 / 6$

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PLATE GLASS
15 in . Actual size
$18 \mathrm{gin} . \times 19 \mathrm{in} . \times \begin{aligned} & \text { Actual } \mathrm{in} .\end{aligned}$
12in. Actual size
$13 \mathrm{in} . \times 10 \mathrm{i} \mathrm{in} . \times$ in. $4 /-$
9 in . Actual size 9 in.
$\times 8 \mathrm{in} . \times$ in.
. 3/-

Co-Axial Cable. $\quad 70-80$ ohms impedance.
Single core, 9/- doz. yards.
Iwin core, $12 /$ - doz. yards.
Twin feeder, 6/-doz. yards.
Co-Axial Connectors. For standard $\frac{1}{1} \mathrm{in}$. cable, $1 / 11$.
PERSPEX. $13 \frac{1}{3} \times 101 \times 1 \mathrm{in}$. Neutral shade, slightly marked. 5/11 per piece.
JUST ARRIVED. 15 in . C.R.T. MASKS. New aspect ratio. MASKS. New aspect ratio. Unused. Overall size: 17 in . wide,
13in. high. 17/6. Post $2 /$-extra.

LATEST 12in. PLASTIC MASK Incorporates gold finish tube escutcheon and dark escutcheon and dark screen filter. O all dimensions: high. $17 / 6$.

Post 2/6 extra.

BRAND NEW 12 in . MASKS. LATEST ASPECT RATIO. Round face, with fitted dark screen filter. Dustproof. SPECIAL OFFER 29/11. COMPLETE. POST FREE.
T.C.C. VISCONOL HIGH VOLC. (Cathodray).
$.001 \mathrm{mfd} .15 \mathrm{kV} \quad . . . .$. .001 mfd .25 kV
0005 mfd .25 k V
$0005 \mathrm{mfd} 12.5 \mathrm{kV} \cdot \cdots \cdot{ }^{18 /-}$
Plastic case, single bolt fixing.
Other high voltage condensers.
.1 mfd .7 kV
0.04 mfd .12 .5 kV
.001 mfd .12 .5 kV

## DE LUXE TELEVISION CABINETS

For 12 in . cathode ray tubes Beautiful figured medium walnut finish, with high polish. Fitted with shelf for receiver, glass speaker baffle and fret, and castors for easy and fret, and castors for easy movement. Undrilled. Suitable for use with the Tewmaster, "Practical Television," "Practical World " televisors.
LASKY'S
£8.10.0
Carriage $12 / 6$ extra.
Outside dimensions of cabinet $17 \frac{1}{2} \mathrm{in} . \times 16 \frac{1}{2} \mathrm{in} . \times 32 \mathrm{in}$. Why not convert your table receiver to a console? Adaptor frames for fitting 9 in . or 10 in . C.R. tubes available if required.

This cabinet can also be supplied cut out for a 16 in . C.R. tube.


CRYSTAL DIODES. Germanium Vacuum sealed glass type with wire ends, $2 / 8$ each or 30/- per dozen.
ELECTROLYTIC CONDENSERS. 32.32 mfd . $350 \mathrm{v} ., 3 / 3$ each or $36 /-$ per doz.
WHANDA WIREAND CABLE STRIPPERS, to take allsize flexes and cables up to $\frac{3}{8}$ in. diameter, with 3 alternative heads and triple screw adjustment. These are brand new and boxed, and the meriginal price was 15/- each. Our Price $4 / 3$ original price was
each or $48 /-$ per doz.
CARBON RESISTORS. $\frac{1}{4}-\frac{1}{2}$ watt, 3d, each. Virtually all standard values in stock. Nearest value supplied, unless otherwise stated. Special quantity prices, $f 1$ per 100,68 per 1,000 of any one value.
SILVER MICA OR CERAMIC CONDENSERS ( pFs ). 2, 4, $5,10,15,20,22,25,30,33$, $47,50,75,100,160,200,220,300,330,470,500$ $1,000,2,000,4,000,4,500,4,700$. All at 5 d . each, or 4/- per doz.
WAX TUBULAR CONDENSERS. . 1 mfd 350 v. 01 mfd .450 v., 4 d . each or $3 / \mathrm{P}$ per doz. SILVER MICA CONDENSERS. . 01 mfd 500 v., 6d. each or 5/- per doz.
EHT CONDENSERS. . $001 \mathrm{mfd}, 5 \mathrm{kV}$. A.C. Test, $1 / 6$ each or $15 /-$ per doz. $.02 \mathrm{mfd} ., 5 \mathrm{kV}$, D.C. working, $1 / 6$ each or $15 /-$ per doz.

SPEAKERFRET. Expanded Metal, finished Silver $6 \times 6,1 / 3 ; 9 \times 9,2 / 6 ; 12 \times 12,3 / 9 ; 18 \times 18,8 / 6$. SPEAKER FRET. Expanded metal, finished gold $6 \times 6,1 / 6 ; 12 \times 12,4 /-$
SPEAKER FABRIC. Fawn or brown, $12 \times 12$, 2/-; $18 \times 18,4 / 6$.
HEADPHONES. 4,000 ohms, per pair, $11 /$-.
METAL RECTIFIERS. RMI, 125 v @ 80 mA $3 / 11$. RM2, 125 v.@ $100 \mathrm{~mA}, 4 / 3$. $14 \mathrm{D} / 972$, $4 / 6.12 \mathrm{v} .2 \mathrm{~A}, 12 / 6.12 \mathrm{v} .2 \frac{1}{2} \mathrm{~A} ., 16 / 6.12 \mathrm{v} .4 \mathrm{~A} ., 21 /=$

FLEX CONNECTORS. $2 \frac{1}{\alpha} \times \frac{1}{2}$ in., for 250 v ., 1/- complete.
TWO GANG TUNING CONDENSERS. 0.0005 mfd ., with fixing feet. Price $7 / 9$ each.

THYRATRONS. Type NGTI (CVI|4I), 4 volt heater. Price $6 / 6$ each.
TRIMMERS. $50+50 \mathrm{pF}, 100+100 \mathrm{pF}, 100+$ $500 \mathrm{pF}, 500+500 \mathrm{pF}$, ceramic mica, 9d. each: $250 \mathrm{pF}, \mathrm{I}, 000 \mathrm{pF}$, 9 d , each; $50 \mathrm{pF}, 75 \mathrm{pF}$, airspaced pre-set, $1 / 3$ each; 75 pF , air-spaced, 2 in . spindle, 2/- each.
SWITCH SOCKETS. Flush mounting 250 volt, 3 pin, 5 amp., bakelite. Price 3/6.
LINE CORD. 3-way, 0.3 amp., 60 ohms per foot, 1/9 per yard.
ELECTROLYTIC CONDENSERS. 32 mfd ., 450 valts, 250 mA ripple, can., $4 / 6 ; 8 \mathrm{mfo},. 450 \mathrm{v} .$, $1 / 9 ; 8+8 \mathrm{mfd} ., 450 \mathrm{v} ., 3 / 3 ; 8+16 \mathrm{mfd} ., 450 \mathrm{v} .$, , $4 /-; 8+32 \mathrm{mfd} ., 450 \mathrm{v} .14 / 6 ; 16+16 \mathrm{mfd} .$, 450 v., $5 /-; 32+32 \mathrm{mfd} ., 350 \mathrm{v} ., 3 / 6 ; 25 \mathrm{mfd}$., $25 \mathrm{v}, \mathrm{I} / 9$; $50 \mathrm{mfd} .12 \mathrm{v},. \mathrm{I} / 9$.
MAINS DROPPERS. Standard 0.2 and 0.3 amp. Price $3 / 9$ each.
AERIAL AND OSCILLATOR COILS. For medium and short waves. Price 5/- per set of 4 coils.
HEATER TRANSFORMERS. 230 v . input. $6.3 \mathrm{v} ., 1.5 \mathrm{amp}$. output, $5 / 6$ each.
MOULDED BAKELITE ESCUTCHEONS. $8 \frac{3}{6} \mathrm{in} . \times 2 \frac{3}{3} \mathrm{in}$. with opening $6 \frac{3}{3} \mathrm{in} . \times 1 \frac{1}{8} \mathrm{in}$., $1 /-$ each. GENERAL PURPOSE TRIODES. Type 7193, 6.3 v . heater. Similar to 6 J 5 G . Price $2 / 6$ each. ROTARY TOGGLE SWITCHES. 4-pole bunching. Price $1 / 6$ each.

## 

 PLUGS AND SOCKETS. 5 way, $2 /-, 7$ way, $2 / 3$ complete. JONES PLUGS AND SOCKETS. 6 way, I/9; 8 way $\mathbf{2}$-complete.
E.H.T. PLUGSANDSOCKETS. 1- complete PYE $\frac{1}{\text { I COAXIAL PLUGS AND SOCKETS. }}$ 1/- per pair, complete.
4-WAYMOULDED PLUGS AND SOCKETS 2/6 per pair.
ZINC PLATED CHASSIS. $13 \frac{1}{2} \times 6 \times 2 \frac{1}{2} \mathrm{in}$. drilled for five valves, 2/6.
DUMMY AERIAL LOADS. Tapped at 20 , 10 and 5 ohms, 100 wates. British 5 -pin base. Callers only, $1 / 6$ each.
EPICYCLIC DRIVES. Reduction 5 : I. Price 1/3 each. With 1 Iin. drum, price $1 /$-.
METAL CUTTING SHEARS. With one inch blades. Price $1 / 3$ per pair.
PLASTIC SPEAKER CABINETS. Louvred PLASTIC SPEAKER CABINETS. Louvred
for Sin speaker. Callers only. Price $10 /$ each. IMPULSE MOTORS. Double circuit control, two solenoids, less contacts. Price 2/ each. Ex-equipment.
IGNITION SWITCHES. LOw voicage, high current, in bakelite case. Price 9 d . each.
S.W. TUNING CONDENSERS. 160 pF , with fixing feet. Price $2 / 3$ each.
3-GANG TUNING CONDENSERS. $75+$ 75 pF , split stator each section. Price 2/6.
CATHODERAY TUBES. Type 5CPI. sin. green screen, electrostatic focussing and deflection. Callers only. Price $22 / 6$ each.
PAXOLIN COIL FORMERS. 5/16th in. diameter, lizin. long. Price $2 /$ per doz.

NEON LAMPS. 85 volt striking, S.B.C. sentre contact base. Price $2 / 6$.
SLEEVING. Approx. 2 mm . Price $1 / 6$ per doz. yard lengths.

CARTRIDGE FUSES. All lisual values in $1 \frac{1}{4}$ in., 5d. each. Other sizes 6d. each.
SWITCHBOARD METERS. $6 \frac{1}{2} \mathrm{in}$. scale, $0-5$ v. A.C./D.C. full scale. Price $30 / \mathrm{h}$.
HIGH FREQUENCY PENTODES. Type 6SH7, high slope, octal based. 6.3 v . heaters. Price 6/-.
GROUP BOARDS. Paxolin 7in. x 2 in., 45 contacts. Price $1 / 3$.
GARLAND TAPE RECORDER OSCILLATOR UNITS. With valve, for use with high impedance heads only. Price 35/*.
TAPE RECORDER OSCILLATOR COILS. $6.3 \mathrm{mH}, 45 \mathrm{kc} / \mathrm{s}$, for high impedance heads only. Price $6 / 9$ each.
TELEVISION MAGNIFYING LENSES. 6 in . clear, 19/6; 9in. clear or filter, $50 /-$; 12 in ., clear or filter, 70/-. Please state which and add 5/- for carriage and packing.
DIAL BULBS. $6-8$ v. 0.3 amp., M.E.S. fitting Price 3 for $1 / 9$.


PAPER CONDENSERS. Ex-equipment. 0.01 mfd. $500 \mathrm{v}_{\mathrm{L}}, 6 \mathrm{~d} . ; 0.01 \mathrm{mfd} ., 750 \mathrm{v}, 6 \mathrm{~d} . ; 0.25 \mathrm{mfd}$. $500 \mathrm{v} ., 6 \mathrm{~d} . ; 1 \mathrm{mfd} .400 \mathrm{v} ., 6 \mathrm{~d} . ; 1 \mathrm{mfd} .500 \mathrm{v},. 9 \mathrm{~d} . ;$ $1 \mathrm{mfd} .600 \mathrm{v}_{\mathrm{H}}, 1 / \mathrm{F} ; 1 \mathrm{mfd} .800$ v., $1 / 3$; $1 \mathrm{mfd} 1,500 \mathrm{v}$. $1 /=3 \mathrm{mfd} .750 \mathrm{v},. 1 / 9 ; 10 \mathrm{mfd} .450 \mathrm{v},. 3 / 6$.
OIL FILLED CONDENSERS. Ex-equipment Imfd. 600 v., 3/-; 1 mfd. 1,500 v., $2 / 9$.
AERIAL FILTER UNIT. No. 112. 10P/13089 2-each.

ALADDIN $\frac{3}{3}$ in. COIL FORMERS. Ex-Govt., wound. 2/- per doz.
ADMIRALTY TRANSFORMERS. 36 vA . 500/1,000 c. Pri. 0-80-180 v. Sec. 0-900-2,400 v. Can be used in reverse on $50 \mathrm{c}, 3 / 9$.
MINIATURE MUMETALTRANS. FORMERS. Auto-wound, giving approx. 2/l ratio. Ideal rewind as head lift transformer, or lams can be used for recording heads. 2/6 each.

DIMMER SWITCHES. 5C/725. Wire-wound, approx. 300 ohms, with off position. 1/3 each.

ELECTRON COPPER AERIALS. 50ft., 2/-; 100ft., 3/9.

DECALS, 500 tin. high white transfer letters and words for marking electronic equipment. Price 4/9 per book. The new Decals book for the amateur now available. 29 words per page, 4 pages radio and audio, 4 pages $T / V$ and Scope, 2 pages misc. incl. Tx. and Tape Recording. 3/6 per book.

METERS. 2 in . square M/C. $0-300$ Y., with separate resistor, thus usable as $0-5 \mathrm{~mA}$ meter. 8/6.

ALL GOODS NEW AND UNUSED (except where otherwise stated). ITEMS (C) REQUIRE CRATING FOR SAFE DESPATCH. CRATES CRATING FOR SAFE DESPATCH. CRATES
ARE NOT CHARGED PROVIDED YOU ARE NOT CHARGED PROVIDED YOU
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PLEASE ADD POST OR CARRIAGE ON ALL ITEMS. KINDLY PRINT NAME AND ADDRESS. POST ORDERS TO OUR DEPTFORD ADDRESS. EARLY CLOSING THURSDAY, OPEN ALL DAY SATURDAY.

THELATEST LANETAPETABLE Incorporating three heavy duty Lane motors fast rewind and wind-on without tape handling automatic braking: high impedance half-track heads: hub locking device. Tape speed 7 thin per second. Price $\mathfrak{f} 17 / 10 / \%$, Carriage $10 /$.
" MOTEK " K3 TAPE UNIT. A high impeddance Tape Unit embodying 3 motor drive unit with high impedance Record-Playback and erase heads. Electronic braking system, push button controls, rewind and fast forward wind without tape handling. $7 \frac{1}{2}$ in. per second, twin track thus giving approx. one hour playing time with 7in. spool. Standard size unit: $16 \frac{1}{4} \mathrm{in} . \times 11 \frac{1}{2} \mathrm{in} . \times$ $4 \frac{1}{8} \mathrm{in}$. Price 16 gns , plus $10 /$ - carriage.

HIGH QUALITY LOUDSPEAKERS
GOODMANS Audiom 60, $12 \mathrm{in} ., \mathrm{E} 11 / 5 /-$.
GOODMANS Axiom 150 Mk . II, 12 in ., £14/3/4.
Wharfedale Bronze 10 in ., $\mathbf{6 5 / 0 / 4}$.
WHARFEDALE Golden
10in., $£ 8 / 6 /$.
WHARFEDALE Golden
10in. CSB, $49 / 5 /$.

## (6)



CERAMIC SWITCHES. Single pole, eight way, $3 / 6$ each.
HIGHCAPACITY PRECISION MICA CONDENSERS. We have still a few thousand of these left, but as we can no longer offer our original comprehensive range, we are clearing the balance. in the values shown hereunder, at $1 /$ - each, or 9/- per dozen. Accuracy in all cases is plus or minus point five ( 0.5 ) per cent. Ruby mica and copper foil or silver. (a) $0.017970 \mu \mathrm{~F}$, (bl) $0.027400 \mu \mathrm{~F}$. (d) $0.040710 \mu \mathrm{~F}$. (e) $0.055820 \mu \mathrm{~F}$, (g) $0.087460 \mu \mathrm{~F}$, (h) $0.108435 \mu \mathrm{~F}$, (i) $0.123750 \mu \mathrm{~F}$. MULTI-PURPOSE TOOL. Bends, shears, punches and threads sheet-strip and rod. For all the little workshop jobs that waste your time. Complete with jig, gauge and protractor, enabling repetition work to be carried out with precision. 17/6.

UNDRILLED CHASSIS. in 20 s.w.g., bright mild steel : four-sided size $13 \mathrm{in} . \times 7 \mathrm{in} . \times 2 \frac{1}{2} \mathrm{in}$., 5/- each; two-sided with two straps, 12 in. $x$ 4 in. $x 2 \frac{1}{2}$ in., price $3 /-$ each. Two-sided with two straps, size $6 \times 5 \times 2$ in., price 2/- each. Add onethird to above prices if chassis required in aluminium.

## BOOKS FOR RADIO ENGINEERS

Mullard Valve and Service Reference Manual Mullard Amateurs Guide to Valve Selection Osram Valve Manual, Part 1
Brimar Radio Valve and Teletube Manual.. Wireless World Radio Valve Data, 3rd edition
Radio Valve Guide. By W. J. May
The Williamson Amplifier Manual, latest edition
Wireless World High Quality Amplifier Manual
T.V. Fault Finding

Television Faults
Television Explained (Miller)
Viewmaster Envelope (state transmitter for which required)
Tele-King Envelope
The Oscilloscope Book
Magnetic Recording (Quartermaine)
(Add 3d. to price in all cases for postage.)
TYANA SOLDERING IRONS. Lightweight, 40 watt irons with easily interchangeable elements and $3 / 16$ in. diameter bits. Voltage ranges, $100 / 110 \mathrm{v} ., 200 / 220 \mathrm{v}$. and $230 / 250 \mathrm{v}$. Price $16 / 9$. "The iron that makes soldering a pleasure."

WIRE WOUND RESISTORS. Open, cement coated or vitreous enamelled. 4 watt, 50,90 , 1 k . Price $1 /$ - each. 6 watt, $30,145,27010 \mathrm{k}$., 15 k . Price $1 / 6$ each. $10-15$ watt, $5,90,100$, $120,170,175,200,400,700,950$, I k.. 3.5 k ., $4.5 \mathrm{k} ., 4.7 \mathrm{k}$. , I $\mathrm{k}_{\mathrm{i}}$, $15 \mathrm{k} ., 25 \mathrm{k} ., 1 / 9$ each. 15 watt, 650 ohm. Price $2 /$.

## HEAVY DUTY MAINS TRANSFORMERS

137: Heavy-duty shrouded; 250-0-250 $120 \mathrm{~mA} ., 6.3$ v. 7 a., 5 v. 5 a., $37 / 6$.
190: Heavy-ducy shrouded; 350-0-350 120 mA., 6.3 v. 5 a. C.T., 5 v. 5 a., 39/6.

5/$5 /-$
$1 / 6$
$5 /-$

All new and unused. Spindle diam. $\frac{1}{6} \mathrm{in}$. in all cases. * are earthed slider.

TOROIDAL CERAMIC POTENTIOMETERS. $260 \Omega 50 \mathrm{w} ., 6 / 6.17 \mathrm{k} .100 \mathrm{watt}, 8 / 6$.

VARLEY MAINS TRANSFORMERS. Primary $10-0-200-220-240$ volts. Secondary $300-0-300$ volt at 150 mA ., 5 volt at $3 \mathrm{amps} ., 6.3$ volt at 4 amps . 6.3 volt at I amp. Open type construction. 6.3 volt a

ENGRAVED KNOBS. I $\frac{1}{\mathrm{i}}$. diameter, fluted in Walnut or lvory, with the following markings: Volume. Vol-On-Off, Treble, Bass, Tone, Tuning, Wavechange, S-M-L-On-Of,'Brilliance, Brilliance-On-Off. Contrast, Focus, RI-R2-PB. Price 1/6 each. Plain knobs to match. $1 / 3$ each.
RECTANGULAR KNOBS. Size $1 \frac{1}{4} \mathrm{in} . \times \frac{3}{4} \mathrm{in}$. with gold indicating spot; to fit standard $\frac{1}{4}$ in. spindies. Price 9d. each.

FULLERPHONE KNOBS. Black flanged type for $\ddagger \mathrm{in}$. spindle, engraved $0-8$. Special price, 2/- per dozen.
BRIMISTORS. Non-linear resistors to protect valves from current surges: CZI, 0.3A, 3/6; CZ2, 0.3A, $2 / 6 ; C Z 3,0.2 \mathrm{~A}, \mathrm{I} / 6 ; \mathrm{CZ} 4,1.25 \mathrm{~A}$, 5/-; CZ6, 0.45A, 3/6.
MÓULDEDBROWNBAKELITE CABINETS. Suitable for fitting Decca 3 -speed gram. motor, amplifier or loudspeaker. Outside gram. motor, amplifier or loudspeaker.
 Mess of wall, 1/Gin. Price, 22/6, to callers only. at 80 mA ., $8 / 6$ each.


AMPLION TESTMETER, 10 ranges A.C. and D.C. up to 500 v. Resistance up to 200,000 ohms, 1,800 ohms per volt A.C. and D.C. Price $\mathrm{f5}$, HIGH WATTAGE WIRE WOUND RESISTORS. Capped end type, porcelain covered, at the following prices: 20 watt, $1 /-; 40$ watt, $1 / 3 ; 80 \mathrm{watt}, 1 / 6 ; 100 \mathrm{watt}, 1 / 9 ; 200 \mathrm{watt}$, $2 / 6 ; 3$ ohm, semi-variable, 20 wate ; 4 ohm , 40 watt: $13.852+1.352+8.352$ ohm, 40 watt; $80 \mathrm{ohm}, 40$ watt; $350 \mathrm{ohm}, 40$ watt; 430 ohm , $80 \mathrm{ohm}, 40$ watt ; $350 \mathrm{ohm}, 40$ watt ; 430 ohm, 200 watt ; $5 \mathrm{k}, 40 \mathrm{watt} ; 7.5 \mathrm{k}, 40 \mathrm{watt}$;
$12 \mathrm{k}+2 \mathrm{k}, 80$ watt ; $20 \mathrm{k}, 80$ watt; 50 k . 100 watt ; $72 \mathrm{k}+2 \mathrm{k}, 80$ watt ; $20 \mathrm{k}, 80$ watt; 50 k . 100 watt ; carriage extra TWIN SCREENED CABLE. Suitable for carrying currents of up to 5 Amps. Cotton covered, 9d. per yard. Ditto uncovered, 8d. per yard.
RUBBERTUBING. External diameter, 0.25 in , internal diameter, $0 . l i n$. Price $2 d$ d. per yard. WALNUT VENEERED CABINETS. Size $12 \mathrm{in} . \times 7 \mathrm{in} . \times 5 \frac{1}{2} \mathrm{in}$. Suitable for housing a T.R.F. receiver or inter-com. Complete with back, chassis, dial and clips. Price, to callers only, 22/6. LARGE DIAMETER SLEEVING. Transparent plastic, $\frac{1}{3} \mathrm{in}$. diameter, 6 d . per yard length
LITZ WOUND INDUCTORS. 199 micro Henry, wound on Aladdin $\frac{3}{3} \mathrm{in}$. coil former (no iron dust core). Price, $2 / 6$ per half-dozen. ELECTRO-MAGNETIC CONTACTORS. Energised at $9-14$ volt, $\frac{1}{2}$ Amp; maximum switched current 40 Amps. In bakelite case. Price 2/9. PAXOLIN PANELS. $3 \frac{1}{2} i n . \times 1 \frac{1}{2} \mathrm{in} . \times 3 / 32 \mathrm{in} .$, $1 /=$ per doz., $5 /-$ per $100.2 \frac{1}{2} \mathrm{in} . \times 2 \mathrm{in} . \times 1 / 16 \mathrm{in}$., 1/3 per doz., 6/= per 100. Many thousands available.
CONNECTION BLOCKS. Moulded plastic, with brass inserts. 3-way, IOd.; 4-way, 1/-. WESTECTORS: Type W6 half-wave, 9d. each ; Type WMX2201 full-wave, $1 /$ - each

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WE HAVE BEEN APPOINTED SOLE DISTRIB UTORS FOR GREAT BRITAIN FOR THE "BRENETTE" RANGE OF CELL MICROPHONES, WHICH WILL SET A NEW STANDARD IN QUALITY AND VALUE. THE FIRST OF THESE, WHICH WE ARE INTRODUCING WHICH WE ARE INTRODUCING 9ND. BALL MICROPHONE. THE 9ND. BALL MICROPHONE. THE
PRICE OF THIS ITEM, COMPLETE PRICE OF THIS ITEM, COMPLETE
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A MIDGET, 4-STATION "PRE-SET" REGEIVER A complete Kit to build a

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Receiver ior A.C. mains The set is ny threestatigned to receire waveband and one on lium wave, each station being received by the turn of a rotary suitch- no tuning
being necessiry
 expensive reaty-made set, but can be more
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A Widget f-ralve superhet Portaile
net covering medium and long wave
Designed to operate on A.C. mains The ent volta, or by an "Alldry". battery. The set is so designed that the mains section is aupplied as a separate unit which may
be atded at any time. The Kit thereiore II be sup, lied (a) ar an ". Alldry ", Baittery Superhet Personal Set which catu thery Le accommodated in the attache cispe as illustrated (size 9 jin. $\times 44 \mathrm{in}, \times$
7 in ). this is attractively finished 7in.). this is attractively finished
in izzard, maroon, dark green or blue rexine, or (b) as a combinied
 wood catinet is available to acconmorlate woth Mains Utit and Batterice toge lier.
Crecuitincorporates delayed A. V.C.


 Practica! Larout and a

## TWO BATTERY PORTABLES

(a) THE "MIN! TWO-THREE
 metres, with use of uhort trailer acrial.
 a 3-valve zet or a- a-ralve (afterwards easily converted to the 3 -valve) can be made
Consists of a T. R. F. ciruvit using a regenerattve detector With H.F, statre gud a h.
line up IT4-1T4-DL94.
The 2 -valve set ean be completely built ior $94 / 3 / 6$ (less case), and the 3 -walve iur $25 / 3 /$ - (less case
includesvalven, spakerind drilledchassis.
includes valven, speakerand drilled chassis.
Send $1 / \theta$.
Send 1/9 ior the asserably instructions; thes include simple and complote practical component layouts and dia-
grams, which enable fre most inexperienced construcgrams, which enable tre most inexlerienced construc-

(b) THE *MINI-FOUR

A 4 -valve Baitery Superhet Receiver degigned to recejve long wave to suitlocal cosditions. Each station is obtained on the set by thm turn of a rotary switch. No tuning is necessary.
It is of midget size, being only $4 \frac{1}{2} \mathrm{in}, \times 6 \frac{1}{\mathrm{in}} \times \mathbf{4} 1 \mathrm{in}$. when completely built and is wery easily assembled irom diagruns supplied.
Cost of all components t.e build thls get. in accordance with
 tive carrying ease finished io blue leatherette, $16 / 9$. Conpleteconstructional data with a blue print, which shows the practical eomponent lafout and wiring diagram, together with an individjas component price list, is available
separntely. $1 / 8$. Gur battery eliminators (illustrated above) separately. $1 / 8$. Gur bathery elfminators ( illustrated abov

## THE FAMOUS "SHAFTESBURY

 RIBBON MIKEIncorporating internal line transformer having transformation ratio from ribbon impedance up to $500-600$
ohms. . reduced from 10 gas. to fe. A special line to grid, $\mathbf{5} 00-600$ ohms transformer aleo available for 25 /.

## DENCO I.F. LINER

Foraceurately giving 465 Kc and 1.6 Mc . I. E. channels
and associate circuits. Battery $\left\{\begin{array}{l}\text { completely seli-contained, 59/6 (plus battery 2/2). }\end{array}\right.$

## THE "WIRELESS WORLD MIDGET A.C. MAINS 2-VALVE REGEIVER

We can supply all the components to build this set, including valves and moving coil speaker, for $£ 3 / 10 /=$ including designers' complete building instructions
(these are availableseparately for od). (these are available separately for 9 d .).

## THE VIEWMASTER TELEVISOR

customers to build this T/V and can supply a SPECIFIED COMPONENTS EX-STOCK. The assembly available ior $7 / 6$ for London, Sutton Coldfleld, Holme Moss, Kirk-o'-Shotts and Wenvoe. Completeteleviaion price list is contained in our general STOCK LIST at 9d.,including Haynes, etc., components.

THE DENGO ULTRA MIDGET SUPERHET COIL TURRETS WITH A ROTARY TURRET ACTION
Type CT8 consists of a four station " pre-set" unit irsm long wave cha be received by a turu of the turret one on Price 48/10. Type CT10. 18 a 3 wave band coil pack incorporating a fourt gwitch position ior 'Gram. Comple te coverage 18 , long wavebatid $700-2.000$ metres, medium waveband $190-570$ ind short ware 15.50 metres. Price $£ 2 / 12 / \%$
Acoment with each turret. These can be aupplied separately for 6 d

## COMPLETE BATTERY CHARGER KIT

 For charging elther $2,4,6$ or 12 volis at 1 ampComplete in metal case, and incorporiting a limiter Complete in metal case, and incorporating a limiter
resistor. Size 54 in. $\times 54 \mathrm{in}, \times 4$ in. Conplete and resistor. Size 54 in
ready ior use, $39 / 6$.

## ELECTRIC TRAIN CONTROL UNIT

 Provides 4, 6, 8, 10 or 12 volts 1 amp., and has provision for forward and reverse motion. Corntlete in metal box. Size $5 \frac{1}{2} \mathrm{in} . \times \overline{5}$ in $\times 4$ in. $49 / 6$.
## PICK.UPS

Cosmocord " G.P.20," forstandard records, e3/11/5. interchangeable(G.F.19) headfor L.P.records, $\mathbf{E 2 / 3 / 4}$. Deccalightweight' ' turnover head" type, for L.P. and standard records, £3/19/2.
Goldring, Standard, Itghtweight Magnetic, $35 / 10$. For Standard or L.P. Records. £3/11/5.

## BATTERY CHARGER KITS

All kits incorporate metal rectifiers and are for use on
A.C. maing 220.200 volts. All kits include an easily A.C. mains $220-250$ Folts. All kits include an easily followed wring dagram. All prices inclindera TAPPED
RESISTOR and o tive-position SELACTOR SWITCH RESISTOR and is tive-position SELECTOR SWITCH
to enable the charging rate to he valiod. For ti or 14 volt batteries at max. 1 im (excluding Resistor ard Switch, \&1/3/6) excluding Resistor and Switch. £1/8:8) For 6 or 12 volt batterie at max. $2 \frac{1}{2}$ amp excluding Resistor and Switch. $\mathbb{L 1}^{1 / 19 / 6}$ ), $2 / 14 / 6$ For ${ }^{6}$ or 19 volt batteries at max. 4 , imps., $£ 3 / 2 / 6$

Send gd. for our STOCK LIST it shows hundreds of RADIO AND TELEVISION COMPONENTS and many KITS OF PARTS for both Sets and Battery Chargers. When ordering please include approx. cost of Post and Packing.

TWO COMPLETELY ASSEMBLED "ALL-WAVE" SUPERHET CHASSIS


## A GENUINE SPECIAL OFFER!

PLESSEY 3-SPEED AUTO CHANGE UNITS

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$1 \mathrm{~m} / \mathrm{a}_{\mathrm{a}}, \mathrm{M} / \mathrm{c}$., $2 \frac{1}{2} \mathrm{in}$., flush panel mounting ................................................................ 22/6
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$5 \mathrm{~m} / \mathrm{a}$., $\mathrm{M} / \mathrm{c}$, , 2 in. . square panel mounting $10 \mathrm{~m} / \mathrm{a}$., M/c., $2 \frac{1}{2}$ in., flush panel mounting $30 \mathrm{~m} / \mathrm{a}$., M/c., $2 \frac{1}{2} \mathrm{in}$. round panel mounting $30 \mathrm{~m} / \mathrm{a} ., \mathrm{M} / \mathrm{c} ., 2 \frac{1}{2} \mathrm{in}$., flush panel mounting $50 \mathrm{~m} / \mathrm{a} ., \mathrm{M} / \mathrm{c} ., 2 \mathrm{in} .$, square panel mounting $150 \mathrm{~m} / \mathrm{a} ., \mathrm{M} / \mathrm{c}$., 2 in ., square panel mounting $200 \mathrm{~m} / \mathrm{a} ., \mathrm{M} / \mathrm{c}$., $2 \frac{1}{2} \mathrm{in}$., flush panel mounting $500 \mathrm{~m} / \mathrm{a}$. . M/c., $2 \frac{1}{2} \mathrm{in}$., flush panel mounting

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6
18
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| Tubuiar Types |  | Can Types |  |
| :---: | :---: | :---: | :---: |
| $8 \mu \mathrm{~F} 450 \mathrm{v}$. | 1/11 | $16 \mu \mathrm{~F} 450 \mathrm{v}$. | 219 |
| $8 \mu \mathrm{~F} 500 \mathrm{v}$. | 2/9 | $24 \mu \mathrm{~F} 350 \mathrm{v}$. | $2 / 11$ |
| $16 \mu \mathrm{~F} 350 \mathrm{v}$. | $2 \cdot 3$ | $3 \hat{2}_{2 / 2} \mathrm{~F} 350 \mathrm{~V}$. | 211 |
| $16 \mu \mathrm{~F} 450 \mathrm{v}$. | 2/9 | 32 mfd .450 v . | 4/9 |
| $16{ }_{\mu} \mathrm{F} 500 \mathrm{v}$. | 3/9 | $40 \mu \mathrm{~F} 450 \mathrm{v}$. | 4/9 |
| $24 \mu F 350$ v. | 3/3 | $50 \mu \mathrm{~F} 350 \mathrm{v}$. | 4/9 |
| $32 \mu F 350$ v. | 3/9 | $8-8 \mu \mathrm{~F}$ \$50 v. | 3/8 |
| 32 mfd .500 v . | 5/9 | $8-8 \mu \mathrm{~F} 450 \mathrm{v}$. | $3 / 11$ |
| $8-16 \mu \mathrm{~F} 500 \mathrm{v}$. | 4/11 | $8.16 \mu \mathrm{~F} 450 \mathrm{v}$. | 4/6 |
| $25 \mu \mathrm{~F} 25 \mathrm{v}$. | 1/3 | $16.16 \mu \mathrm{~F} 450 \mathrm{v}$. | 4.11 |
| $50 \mu \mathrm{~F} 12 \mathrm{v}$. | 1/3 | $16-16 \mathrm{mfd} .500$ | - 5/3 |
| $50 \mu \mathrm{~F} 50 \mathrm{v}$. | 2/3 | $16-32 \mu \mathrm{~F} 350 \mathrm{v}$. | 4/9 |
| Gan Types |  | $32-32 \mu \mathrm{~F} 350 \mathrm{v}$. | 4/9 |
| 8 mfd .450 v . | 2/3 | $32-32 \mu F 450 \mathrm{v}$. | $5 / 11$ |
| 8 mifi. 500 v . | 2/11 | 60-100 mfd. 350 | v.7/6 |
| 16 mifd .350 v . | 1/11 | $3,000 \mathrm{mid} .6 \mathrm{v}$. | 6/9 |

$32-32-8 \mu \mathrm{~F} 350$ CAN TYPES
6-16 $\mu \mathrm{F} 450$ V $5 / 1$
$50 \mu \mathrm{~F} 350$ v. plus $250 \mu \mathrm{~F} 12 \mathrm{v}$. ............................ 411
CHASSIS. (16 s.w.g. Undrilled Aliminiums) Re ceiver type. $6 \times 3 \frac{8}{8} \times 1 \frac{1}{2}$ in., $1 / 11 ; 71 \times 4 \frac{9}{4} \times 2 \mathrm{in}$. $2 / 9 ; 16$ s.w.g. $10 \times 5\} \times 2$ in., $3 / 9 ; 11 \times 0 \times 21$ im. $4 / 3 ; 12 \times 8 \times 2 \frac{1}{2}$ in., $5 / 3 ; 16 \times 8 \times 2 \frac{1}{2}$ in., $7 / 6$; $20 \times 8 \times 2$ in., $8 / 11$; Amplifier Type, $12 \times 8 \times$ $2 \operatorname{lin} ., 7 / 11 ; 16 \times 8 \times 2 \frac{1}{2} \mathrm{in} ., 10 / 11 ; 14 \times 10 \times 3 \mathrm{in}$. $13 / 6 ; 20 \times 8 \times 2 \frac{1}{2} \mathrm{in}$., $13 / 6$.

[^25]\[

$$
\begin{aligned}
& \text { A PUSH-PULL 3-4 watt HIGH-GAIN } \\
& \text { AMPLIFIER FOR } \mathbf{3} \mathbf{3} 12 / 6 \text {. For Mains iuput } \\
& 200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s} \text {. Complete kit of parts } \\
& \text { including circuit diagram and instruc ions. } \\
& \text { (Point-to-point wirime diagrams available for } \\
& 1 / 6 \text { extra.) Anplifier can be used with any } \\
& \text { type of Feeder (Tnit or l'ick-up. Ontput is for } \\
& \text {-3 ohm speaker. (We can supply a very } \\
& \text { suitable l(0in. mit by Goodmans at } 31 /- \text {.) } \\
& \text { The amplifier can be supplied ready for use } \\
& \text { for } 84 / 17 / 6 \text {. liull descriptive leaflet } 1 /=\text {. }
\end{aligned}
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VOLUME CONTROLS with long spindles, all values ess switch 2/9, with S.P. switch 3/11. WIRE WOUND POTS.: $5 \mathrm{~K}, 10 \mathrm{~K}, 20 \mathrm{~K}, 25 \mathrm{~K} .50 \mathrm{~K}$ (mediun length spindles), 2 /9.

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Potted Types

| 330 mA . 5 H. 50 ohns. | 12/9 |
| :---: | :---: |
| $22^{2} 1$ mA. 5 H. 50 ohms. | 10/3 |
| 150 mA .10 H .200 hmms | 10/11 |
| $50 \mathrm{~mA} .50 \mathrm{H} .1,250$ ahms | 8/11 |

EX-GOUT. BLOCK PAPER MANSBRIDGE TYPE CONDENSERS


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3 a., $9 / 11 ; 6.3$ v. 6 a., $17 / 6 ; 0-2-4-5-6.3$ v. 4 a., 3 a., $9 / 11 ; 6.3$ v. 6 a., $17 / 6 ; 0-2-4-5-6.3$ v. 4 a.,
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## SMOOTHING CHOKES

|  | $16 / 9$ |
| :---: | :---: |
| 250 M | $11 / 9$ |
| $100 \mathrm{mit}, 10 \mathrm{H} .100$ ohms. | $7 / 6$ |
| $80 \mathrm{~mA}, 10 \mathrm{H} .350$ ohms. | 5/6 |
| $60 \mathrm{~mA} ., 10 \mathrm{H}$. | 4/11 |
| $50 \mathrm{~mA} ., 50 \mathrm{H} .1,000$ ohms. | 9/11 |
| ELIMINATOR TRANSFORMERS |  |
| $1{ }^{\text {'rimaries }} 200-250 \mathrm{v} .50 \mathrm{c} / \mathrm{s} ., 120 \mathrm{v} .40 \mathrm{~mA}$. | 7/11 |
| $90 \mathrm{v} 10 \mathrm{~mA}, 9-0-9 \mathrm{v}, 250 \mathrm{~mA}$. | 106 |
| $20 \mathrm{v} 40 \mathrm{~mA} .,$.6 v. 1.5 a. | 14/9 |
|  | 15/9 |

## FULLY SHROUDED UPRIGHT

$250-0$ - 250 v. $60 \mathrm{~m} .4 ., 6.3$ v. 2 a., 5 v. 2 a.
Midget type 2f-3-3in. ….......................... 17/6
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25/9
for R1355 conversion............................... 29/9
300-0-300 v. 100 nıA., 6.3 v 4 v. 4 a. c.t.
0-4-5 v. 3 a. ..............................................
350-0-350 v. $100 \mathrm{mA}. ., 6.3$ v.-4 v. 4 a. $\mathrm{t}_{\mathrm{t}}$,,
 $350-0-350$ v. 150 mA ., 6.3 v. „ a., 6.3 v, 2 a. 5 v. 3 а.
5 v. 3 a. .................................................... 45/9
$350-0.350$ v. 250 mA., 6.3 v. 6 a., 4 v. 8 a. 0-2-6 v. 2 a., 4 v. 3 a. for Electronic Eng Televisor
67/6
425-0-425 v. $200 \mathrm{~mA} ., 6.3$ v. -4 v. 4 a. c.t.,
Williamson Amplifier, etc. ....................
$425-0-425$ v. 250 mA .6 .8 v. 6 a., 6.3 v. 6 a.

325-0-325 v. $20 \mathrm{~mA} ., 6.3 \mathrm{v} .0 .5 \mathrm{a}$., 6.3 v
1.5 a. for Williamson Preamplifier .......... 17

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65/6
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300
mA.
6 $3 \mathrm{amp} ., 5$ volt $2 \mathrm{amp} ., 25$ /.
Drop-thro' $350-0-350 \mathrm{v} .70 \mathrm{~mA}$. Drop-thro $350-0-350$ v. 70 mA.
6 v. 2.5 amp., 5 v. 2 amp., $14 / 6$. 6 v. $2.5 \mathrm{amp},{ }^{5} \mathrm{~s} \mathrm{v} .2 \mathrm{amp}, 14 / 6$.
Drop-thro $250-0.250 \mathrm{v} .80 \mathrm{~mA}$ $6 \mathrm{v} .3 \mathrm{amp},{ }^{5} \mathrm{v} .2 \mathrm{amp} ., 16 / 6$ Drop-thro' $110-0-11060 \mathrm{~mA}$ $6 \mathrm{v} .0 .5 \mathrm{amp} ., 8 / 6$. 280-0-280, drop-through, 80 mA . $6 \mathrm{v} .3 \mathrm{amp} ., 5 \mathrm{v} .2 \mathrm{amp} ., 16 / 6$. Auto-wound, H.T. 280 volts at $360 \mathrm{~mA}, 4 \mathrm{v} .3 \mathrm{amp} ., 2 \mathrm{v} .3 \mathrm{amp}$. or 6 v. 3 amp. Separates 4 v . 3 amp., re-tifier winding (upright or drop-through), 10/6.
Auto-transformer, various combinations of voltages includ. ing 110 v .70 wates and $3 / 4$ volt windings at I amp., 2 vult I amp. drop-through or upright mountn5, 10/6.
$\mathbf{2 5 0 - 0} \mathbf{- 2 5 0}, 60 \mathrm{~mA} .6$ v. 3 amp . 2,
250-0-250, 80 mA .6 v .4 amp , $14 / \mathrm{-}$ Pri. 230 v . Sec. 200-0-200 35 mA . 6 v .1 amp., $8 / 6$.
Pri. 200/250 v. secondary 3, 4,5 $6,8,9,10,12,15,18,20,24$ and 30 volt at 2 amps., 13 /-
5 emi-shrouded drop-through 200 250 V. primary; sec. 280 $0-280250 \mathrm{~mA}$. 6 v .6 amps . 5 v . 3 amps., 29/6. ${ }^{\text {P. }} \dot{\&}$ P. 8 . Heater Transformer, Pir. 230. $\begin{array}{ll}\text { Heater Transformer, Pir. } & 230 \text { - } \\ 250 \text { v. } 6 \text { v. } & \frac{1}{2} \text { amp., } 6 \%: \\ 2\end{array}$
 $2 \frac{1}{2}$ amps., $5 / 0: 2^{2,4} 4$ or 6 v. at $2 \mathrm{amps}, 7 / 6 ; 2$ Y. $2 \frac{1}{2}$ amp. and 8/6. P. \& P. each I/-

## P.M. SPEAKERS <br> $\underset{\substack{\text { (Closed } \\ \text { ficld) }}}{\text { ( }}$

 with less P. \& P. on the above $1 /$ - each.
loin. ess trans., 25/-. P. \& P. $1 / 6$. R. \& A. 8in. M.E. Speaker field R. \& A. 8in. M.E. Speaker field
coil, 1,600 ohms O.P. trans. coll, 1,600 ohms O.P. trans.
5,000 ohms, impedance, $18 / 6$. Post and packing 2/
bitin. Energised Television Speaker by PIESSEY. Field resistance 68 ohms with humbucking coil. Will pass up to 300 mA ., requires minimum 200 mA . to energise, $9 / 6$ each, 2 for $18 \%$.
Volume Controls, by famous manufacturer. Long spindle less switch, $50 \mathrm{~K}, 500 \mathrm{~K}$, I meg. 2/6 each. P.'\& P. 3d. each. ${ }^{\text {meg }}$ Expanded aluminium Expanded aluminium
speaker fret, $13 \frac{1}{4} \times 9 \mathrm{in}$., $2 /=$. speaker fret,
Volume Controls by famous Volume Controls by famous manufacturer. Long spindle and
switch, $\frac{1}{4}, \frac{1}{2}, 1$ and 2 meg., $4 /-$ switch, $\frac{1}{4}, \frac{1}{2}, 1$ and 2 meg., 4/-
each; $50 \mathrm{~K}, 3 / 6$ each. $\frac{1}{2}$ and 1 each: $50 \mathrm{K.}, \mathrm{3/6} \mathrm{each}. \frac{1}{2}$ and I
meg., long spindle double pole meg., long spindle double pole
switch, miniature, 5/-. P. \& P. 3d. each.
Trimmers, $5-40$ pf., 5 d . ; $10-110$ $10-250,10-450 \mathrm{pf} .10 \mathrm{~d}$.
Twin-Gang. 0005 Tuning Condenser, 5/-. With trimmers, 7/6. P. \& P. $1 /=$

Line Cord. 3-way 0.3 amp. 180 ohms, per yard, 1/3 per yard. Twin-gang .0005 with feer, size $3 \frac{1}{4} \times 3 \times 1 \frac{5}{8} \mathrm{in} ., 6 / 6$.
3-gang .0005 , with feet, size $4 \frac{3}{4} \times 3 \times 1 \frac{3}{6} i n ., 7 / 6$.
Television Coils wound in Television Coils wound in
alican, size $2 \frac{3}{6} \times \frac{7}{8} i n$, with former alican, size $2 \frac{3}{6} \times$ zin. with
and iron core, $1 /$ each.

# D. COHEN RADIO \& TELEVISION COMPONENTS 



KIT OFPARTS FOR SIGNAL GEN ERATOR. Coverage $110 \mathrm{Kc} / \mathrm{s} .-320 \mathrm{Kc} / \mathrm{s}$., $320 \mathrm{Kc} / \mathrm{s} .990 \mathrm{Kc} / \mathrm{s} ., 900 \mathrm{Ke} / \mathrm{s} .-2.75 \mathrm{Mc}$ s. ., $2.75 \mathrm{Mc} / \mathrm{s}-8.5 \mathrm{Mc} / \mathrm{s}$., $8.5 \mathrm{Mc} / \mathrm{s} .20 \mathrm{Mc} / \mathrm{s}$. Metal case $10 \mathrm{in}, \times 6 \frac{3}{8} \mathrm{in}, \times 4 \mathrm{din}$, size of scale $6 \frac{1}{2} \mathrm{in} . \times 3$ tin. 2 valves and 1 rectifier valve, A.C. mains 230/250. Internal modulation 400 c.p.s. to a depth 30 per cent. Frequency calibration accuracy plus or minus I per cent. Modulated or unmodulated R.F. output continuously variable 100 millivolts, $£ 3 / 10 / \mathrm{F}$, P. \& P. 4/-. This includes the return to us for checking and calibration. We will build same for 15 /- extra. Circuit and point-to-point wiring diagram, $3 / 6$. Kit of parts for above, less checking and calibration, £3, plus $2 / 6 \mathrm{P}$. \& P .

Constructor's parcel, comprising chassis $12 \frac{1}{2} x$ $8 \times 2$ mining chassis ${ }^{2 \frac{1}{2}} \times$ $8 \times 2 \mathrm{in} .$, cad. plated 18 gauge, $v / h$. IF and trans cut-outs, back-plate, 2 sup-
porting brackets, 3 wave porting brackets, 3 wave band scale, new wavelength station names. Size of scale $11 \frac{1}{2} \times 4 \frac{3}{2} i n$., drive spindle, drum, 2 pulleys, pointer, 2 bulb holders, 5 paxolin international octal valve holders, 4 knobs, and pair of 465 IF's., 16/6. P. \& P. I/9.


CRYSTAL PICK-UP by famous manufacturer complete with sapphire trailer needle and volume control, 23/-. Less volume control, $\mathbf{2 1 / = \text { , post }}$ and packing on each $1 /$-.


EX-GOVT. RECEIVER TYPE B28. Complete coil unit, 6 bands, $60 \mathrm{ke} / \mathrm{s}$. $420 \mathrm{ke} / \mathrm{s}$-, $500 \mathrm{kc} / \mathrm{s} .-30 \mathrm{Mc} / \mathrm{s} ., 21 /$. Plus 2/- P. \& P. Circuit for above, 4/-Variable Selectivity IF Switch to


CABINET as illustrated, $11 \frac{1}{2} \times 6 \frac{1}{2} \times 5 \frac{1}{2}$, in walnut or cream, complete with T.R.F chassis, 2 waveband scale, station names, new wareband, back-plate, drum, pointer, spring, drive spindle, 3 knobs and back, 22/6. P. \& P. $2 / 6$. knobs and back, 22/6. P. \& P. 2/6. $7 / 6$. Medium and long T.R.F. coils, 5/6 per pair.
Circuit and point-to-point wiring diagram for $\mathbf{3 - v a l v e}$ plus rectifier T.R.F. A.C. mains Receiver which could be built for approx. 44 (using the above cabinet), 1/6.
WATERHOUSE 5in. EXTENSION SPEAKER, complete with vol. control, in gold and green, 22/6. P. \& P. 1/
MAINS OR BATTERY SUPERHET PORTABLE COILS. Mediumwaved frame aerial and long-wave loading coil, used as aerial coils. Midget iron-core screened L/M ose. coils, with circuit I.F. 465 Kc ., 9/6.
465 KC. MIDGET I.F.s. Q 120 , size $1 \frac{1}{\frac{1}{2}} \mathrm{in}$. long, lin. wide, $\frac{7}{1} \mathrm{in}$. deep by very famous manufacturer. Pre-aligned adjustable iron-dust cores, per pair, 12/6. Both these items fl , post paid.
CONSTRUCTOR'S PARCEL comprising chassis 8 in. $\times 4$ in. $\times 1 \frac{1}{2}$ in., with speaker and valveholder cut-outs, 5 in . P.M. speaker with transformer, twin gang with trimmers, pair T.R.F. coils long and medium, iron-cored, four valveholders, 20 K . volume control and wave-change switch, $23 /$.. $P$. \& $\mathbf{P} .1 / 6$.
OUTPUT TRANSFORMERS. Standard type 5,000 ohms imp., 2 ohms speech coil, 4/9; 42-I speech coil 2 -ohm with extra feed-back winding, 4/3; Miniature $42-12-0 h m$ speech coil, 3/3. Multi-ratio 3,500, 7,000 and 14,000 2 -ohm speceh coil, 5/6. To watt push-pull, 6 V 6 matching, 2 ohm speech coil, $7 /-$ Television Chassis: Size $9 \frac{3}{4} \times 9 \frac{1}{2} \times$ $3 \frac{1}{4}$ in., 18 gauge steel cadmium plated complete with 5 -coil cans size $1 \frac{1}{2} \times$ lin. with iron cored former. These are wound for television frequency, 6/6. P. \& P. I/6.
Push-back connecting wire. Doz yds $1 / 6$ post paid.

Standard Wave-change Switches. 6-pole 3-way, 2/-; 4-pole 3-way, 1/9; 5-pole 3-way, 1/9 ; 3-pole 3-way, 1/9; 9-pole 3-way, $3 / 6$; Miniature type, 9-pole 3 -way, $3 / 6$; Miniature type,
long spindle, 3 -pole 4 -way, 2 -pole long spinde,
5 -way.
P. \& P.

Valve Holders, moulded, octal Mazda, and loctal, 7d. each, Paxolin, octal, Mazda and locral, 4d. each. Moulded B7G, B8A and B9A, 7d. each. B7G moulded with screening can, $1 / 6$ each $32 \mathrm{mid}, 350 \mathrm{wkg}$
$16 \times 24350 \mathrm{wkg}$.
$4 \mathrm{mfd} ., 200 \mathrm{wkg}$.
$40 \mathrm{mfd} ., 450 \mathrm{wkg}$.
$16 \times 8 \mathrm{mfd} ., 500 \mathrm{wk}$

$16 \times 8 \mathrm{mfd} ., 500 \mathrm{wkg}$. $16 \times 16 \mathrm{mfd}, 500 \mathrm{wkg}$ $8 \times 8 \mathrm{mfd}$., 450 wkg . | $32 \times 32 \mathrm{mfd}, 350 \mathrm{wkg}$. |
| :--- |
| $32 \times 32 \mathrm{mfd}, 350 \mathrm{wkg}$. | $25 \mathrm{mfd} ., 25 \mathrm{wkg}$.

$25 \mathrm{mfd}, 25 \mathrm{wkg}$.
$1 / d$.
$250 \mathrm{mfd} ., 12 \mathrm{v}$ wk
11 d
16 mfd .500 wkg , wire ends $8 \mathrm{mfd} .500 \mathrm{v} . \mathrm{wkg}$., wire ends $2 / 6$ $8 \mathrm{mfd} ., 350 \mathrm{v}$. wkg., tag ends $1 / 6$ 50 mfd .25 v . wkg., wire ends $1 / 9$ Ex-Govt. 8 mid., 500 v . wkg.,
Ex-Govt. $8 \mathrm{mid}, 500 \mathrm{v}$. wkg., $60+100 \mathrm{mfd}$. 280 v . wkg. $16 \times 32 \mathrm{mfd}$, 350 wkg . 50 mfd .180 wkg . $65 \mathrm{mfd} ., 220 \mathrm{wkg}$ 8 mfd ., 150 wkg.
$60+100 \mathrm{mfd}$., 350 wkg . 50 mfd ., 12 wkg . $50 \mathrm{mfd}, 50 \mathrm{wkg}$., 275 wkg . $4 / 6$ Miniature wire ends moulded 100 pf ., 500 pf , and. 001 ea. 7 d . Combined $12 i n$. mask and escutcheon New lightly tinted perspex. New aspect, edged in brown. Fits on front of cabinet, 17/6. P. \& P. 2/-.
Frame Oscillator Blocking Transformer, 4/6.
Frame O.P. Transformer. Inductance 10 hy ratio $10: 1,9 / 6$. Tube Mounting Bracket, size $9 \frac{1}{2} \times 4 \frac{2}{4}$ in., with $4 \frac{3}{4}$ in. dia. cut-out. In 18 gauge cadmium plated steel and 12 in. tube clamps, 2
Smoothing Choke, 2 henry $150 \mathrm{~mA}, 3 / 6$.
Smoothing Choke, 250 mA . 4 henry, $5 /-; 250 \mathrm{~mA} .5$ henry, 6 -. P.M. Focus Unit for Mazda tube, $15 /-$ P. \& P. $1 / 6$. Similar to above with front adjustment, $2 / 6$ extra.
P.M. Focus Unit for any Sin. or 12 in . tube, 35 mm . neck except Mazda 12 in ., state tube $12 / 6$.
Similar to above, but with front adjustment, 2/6 each extra. Ion Traps for Mullard or English Electric Tubes, $5 /=$, post paid. 465 Kc. J.F.s, size $2 \frac{1}{2} \times 1 \frac{1}{8} \mathrm{in}$. Q. 110 removed from American equipment, 5,- per pair Stan $4 \times 11 \times 1$, iron-cored F $4 \times 1 \frac{1}{2} \times 1 \frac{1}{2}$ in., per pr. 7/6. Wear
 Ir'S. $3 \frac{1}{2} \times 1 \frac{3}{8} \times 18 \mathrm{lin}$, per pr. $9 / 6$.
Iron-Cored 465 Kc . Whistle Iron-Cored
Filter, $2 / 6$.
Television Masks. White Rubber, 9 in. with glass, $7 / 6$. Cream rubber, 12 in . with armourplate glass, $15 /-$, plus $1 / 6$ P. \& P. T.V. Width Control, 3/6.

Two-piece Octal Screening Can, 9d. P. \& P. 3d.
Three-bank, $50 \mathrm{pf}, \mathrm{i} / 3$. Fourbank, 50 pf., I/B.
Mains Droppers, 0.3 amp., 460 ohms, tapped 280 and $410,1 / 6$. 0.2 amp., 717 ohms, tapped at 100 ohms, vitreous, $1 / 6$; 0.3 amps . 950 ohms, tapped 700 and 825 , 2/6; 0.2 amp., 1,000 ohms, vitreous, tapped, $2 / 6$. Vitreous 3 amp. 700 tapped $680,640,600,3 / 6$. P. \& P. on each 3d.
E.M.I. potted low resistance Pick-up Matching Transformer, $7 / 6$.
Speaker Material, 12 in . $\times 10 \mathrm{in}$. 1/6, post paid.
Germanium Crystal Diode 2/3 post paid.

Terms of Business: Cash with order. Despatch of gaods within 3 days fram receipt of order. Where post and packing charge is not stated please

## SPECIAL NOTE: NO GOODS SENT WHERE CUSTOMS DECLARATION IS APPLICABLE.

23 HIG

## 23

## "MUST HAVE" BARGAINS

UNIVERSAL AVOMETERS. Ex-W.D. items that have been thoroughly reconditioned and are in perfect order. We have a limited number only of the 36 range Model at 10 gns . each, and the Model 40 at $\mathrm{f} 12 / 10 /$ each
COMMUNICATIONS RECEIVER R.II55. The famous exBomber Command Receiver known the world over to be supremc in its class. Covers 5 wave ranges $18.5-7.5 \mathrm{Mc} / \mathrm{s}, 7.5-3.0 \mathrm{Mc} / \mathrm{s} 1,500$ $600 \mathrm{kc} / \mathrm{s} 500-200 \mathrm{kc} / \mathrm{s}, 200-75 \mathrm{kc} / \mathrm{s}$, and is easily and simply adapted for normal mains use, full details being supplied. Aerial tested before despatch these are BRAND NEW AND UNUSED IN MAKER'S ORIGINAL TRANSIT CASES, ONLY £II 196.
A few used receivers, also tested working before desparch, are available at $£ 7 / 19 / 6$.
A factory made Power Pack, Output Stage and Speaker, contained in a black crackled cabinet to match the receiver, can be supplied at ONLY $£ 5 / 10$-. Operates receiver immediately. PACK TOGETHER.
Please add carriage costs of $10 / 6$ for receiver, and $5 /$ - for power pack.
RF UNITS TYPE 26 AND 27. The very popular variable tuning units, which use 2 valves EF 54 and I EC 52. Type 26 covers 65$50 \mathrm{Mc} / \mathrm{s}$ ( $5-6$ metres), and Type 27 covers $8 \mathrm{~s}-6 \mathrm{~s}$ Mc/s ( $3.5-5$ metres). BRAND NEW IN MAKER'S CARTONS. ONLY $59 / 6$ each.
VIBRATOR UNITS. 2 volt type, American made, delivers 67 volts at $4.7 \mathrm{~mA}, 130$ volts at 20 mA , and 1.4 v . L.T. Easily adapted for use with any battery receiver, full details being supplied. ONLY 50/- (postage $2 /-$ ).
6 volt type, made by The National Co. of America for use with HRO Communication Receivers, supplying 165 volts at 85 mA . fully smoothed D.C. Complete with vibrator and $6 \times 5$ rectifier in black crackle cabinet, size 7in. $\times 7 \frac{1}{2} i n . \times 6 i n$. Slightly used. ONLY 39/6.
12 volt type, made by Masteradiof or the Admiralty, Delivers 300 volts at 100 mA , and are complete ready for use. BRAND NEW IN MAKER'S CARTONS, ONLY 35/- (postage, etc., 2/6). INDICATOR UNIT TYPE 62A. Contains VCR97 tube with mu metal screen, 12 valves EF50, 4 of SP6I, 3 of EA50, and 2 of EB34. Built on a two deck chassis containing hundreds of condensers and resistors, potentiometers, etc. In BRAND NEW CONDITION IN MAKER'S TRANSIT CASES. ONLY ET/10/(carriage 9/6)
6 VOLT BATTERIES. By famous American makers, these have genuine hard rubber cases, and are BRAND NEW AND UNUSED IN MAKER'S PACKING. Size $8 \frac{1}{2} i n$. long $\times 6 \frac{1}{4} \mathrm{in}$. wide $\times 7 \frac{1}{2} \mathrm{in}$. high. ONLY 59/6 (carriage, etc. 7/6).
10 VALVE I $\frac{1}{2}$ METRE SUPERHET ZC 8931. For long distance TV results. Valve line up is 6 of VR 65, 2 of VR 92 , and I each VR 136 and VR 137, and the $12 \mathrm{Mc} / \mathrm{s} 6$ stage I.F. Strip gives tremendous amplification with ample bandwidth of $4 \mathrm{Mc} / \mathrm{s}$. Easily modified. Full details supplied. ONLY $59 / 6$ (carriage, etc., 5/-).
TELESCOPIC AERIAL. Pulls out of metal tube ISin. Iong to extend to 73 in . BRAND NEW. ONLY $7 / 6$ (post IOd.).
194 I.F. STRIP. An easily modified I.F. Strip recommended for TV constructors who want good results at moderate cost, or for those who have built televisors but are having trouble in the vision or sound receivers. Can also be modified for 2 channel working as per details in "Practical Television"" October issue. This 6 stage strip measures 18 in. $\times 5 \mathrm{in} . \times 5 \mathrm{in}$., and contains 6 valves VR 65 , I VR 92 and 1 of VR 56 or VR 53 . Mod. data supplied. ONLY 45/- (postage, etc., 2/6).
208 AMPLIFIER. Ideal for conversion into a high gain TV preamp. Complete with 2 valves EF 50. ONLY 15/- (postage, etc., 1/6).
CERAMIC 2 WAY 3 BANK SWITCHES, $7 / 6$ each.
CHOKES. $10 \mathrm{H} .60 \mathrm{~mA}, 3 / 9 ; 30 \mathrm{H} .100 \mathrm{~mA}, 12 / 6 ; 5 \mathrm{H} .200 \mathrm{~mA}$. 6/-(postage 1/- per croke).
CHASSIS OF POWER UNIT 529. An ideal unit for seripping, or for using to build an amplifier, etc. Contains valveholders. resistors, potentiometer, chokes, and block and tubular condensers. Brand new, and housed in grey metal case size 12 in . $\times 8 \frac{1}{2} \mathrm{in}$. $\times 7 \frac{1}{2} \mathrm{in}$. ONLY 10/-(carriage, etc., 3/6).
TRANSFORMERS. Manufactured to our specification and fully guaranteed. Upright mounting, fully shrouded normal primaries $425-0.425$ v. $200 \mathrm{~mA}, 6.3$ v. 6 a., 6.3 v. 6 a., 5 v. 3 a., $0-2-4-6.3$ v. 3 a., $72 / 6$.
$425-0-425$ v. $200 \mathrm{~mA}, 6.3$ v. 4 a., 6.3 v. 4 a., 5 v. 3 a., $50 / \mathrm{m}$
$350-0-350$ v. $160 \mathrm{~mA}, 6.3$ v. 6 a., 6.3 v. 3 a., 5 v. 3 a., $42 / 6$.
$250-0-250$ v. $100 \mathrm{~mA}, 6.3 \mathrm{v} .6$ a., 5 v. 3 a., $32 / 6$.
Please add $1 / 6$ per transformer postage.
TRANSFORMERS FILAMENT. 6.3
TRANSFORMERS, FILAMENT. $6.3 \mathrm{v}$.2 a., $7 / 6 ; 6.3 \mathrm{v}$.3 a., 10/6. (Postage 1/-).
TRANSFORMERS EHT, Upright mounting. EHT for VCR 97 tube, 2,500 v. $5 \mathrm{~mA}, 2-0-2$ v. I. 1 a., 2-0-2 v. 2 a., $37 / 6$.
EHT 5,500 v. $5 \mathrm{~mA}, 2$ v. 1 a., 2 v. 1 a., $72 / 6$.
EHT 7,000 v. $5 \mathrm{~mA}, 4$ v. $1 \mathbf{a}, 82 / 6$. Please add $1 / 6$ per transformer postage.

Cash with order please, and print name and address clearly. Amounts given for carriage refer to inland only.

## U.E.I. CORPORATION

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Phone: TERMINUS 7937.
(Open until I p.m. Saturdays We are 2 min. from High Holborn (Chancery Lane Station) and 5 min . by bus from King's Cross.)
U.S. NAVY OSCILLOSCOPE UNITS-Containing 5 BPI. 5in. Tube with fully screened mu shield isolating heater trans Dozens of H.V. Cond., Resistors, Pots, etc. The finest value offered to date in "Scope" units. Price'57/6.
LABORATORY TEST EQUIPMENT. For aligning and checking Trans./Receivers covering 150 to 234 Mcs. comprising. Type BC906. Frequency Dip Grid Meter 145-235 Mcs.
Type BC906. Frequency Dip Grid Meter.
Type I-196-B. Signal Generator. $150-234$ Nics.
Type BCIO66-B. Radio Receiver. $150-234$ Mcs. Price $\mathbf{f} 12$ the Set. Carriage extra.
EX-W.D. TRA NSFORMERS, input 200/250 v., output $0-460 \mathrm{v}$ $200 \mathrm{~m} / \mathrm{A} .210 \mathrm{v} .15 \mathrm{~m} / \mathrm{a} .6 .3 \mathrm{v} .5 \mathrm{a} ., 22 / 6$.
VALVES. IS4, 10/6; 6AG5, $10 / 6$; II7Z6, $12 / 6$ : 65 H 7 , 6/6; EF50, 8/6;955, 954, 6/-; SG215, 6/6; Pen 220A, $6 / 6$; TTII, $8 / 6$ VR $150,10 / 6 ; 42,10 / 6$; CK5I2AX, 9 ;- 6 GAG5, $10 / 6 ; 9001,9002$ 9003, 7/6;954-955, 6/6.
RF24 UNITS. Converted to $28 \mathrm{Mc} / \mathrm{s}$ Band. Variable tuned with $100-1$ geared S.M. Dial. Complete with plug and leads for imme diate use $£ 3 / 5$.-
MAINS TRANSFORMERS. Input 200/240 v. Output 350-0-350 or $250-0-250$ volt 80 mA . and 4 and $6.3 \mathrm{v}, 4 \mathrm{a}$, and 4 and 5 v .2 a Price 21,6 . Input $200 / 240 \mathrm{v}$. Output tapped $3,4,5,6,8,9,10,12$, $15,18,20,24,30$ volts, 2 amps., $21 / 6$. All with one year's guarantee 15, $18,20,24,30$ volts, 2 amps., $21 / 6$. All with one year's guarantee D.P.D.T. RELAYS. Operate at 200/300 volts Dake and break, 8/6. We can supply any type of voltage and make and break, $8 / 6$. W
contacts at varying prices.
NEW SELENIUM RECTIFIERS. F.W. $12 / 6$ volt 3 amps., $14 / 6$
$4 \mathrm{amp} ., 26 /-; 6 \mathrm{amp} ., 30 /-; 1 \mathrm{amp} . \mathrm{B} / 6 ; 12 \mathrm{v} .100 \mathrm{~mA} ., 3 /-$ 250 v .100 mA . H.W., $9 / \mathrm{H}: 80 \mathrm{~mA} ., 6 / 6$.
TYPE IN34-GERMANIUM CRYSTAL DIODES, 5/6.
B.C. 603 RECEIVER AND B.C. 604 TRANSMITTER. 20 to
$28 \mathrm{Mc} / \mathrm{s}$. Including 80 crystals, e 35 , carriage extra.
U.S. MANUALS for Test Equipment IE-56-A, SCR-729-A and SCR729AZ. Contains all the gen on Freq. Meter B.C. 906. Receiver B.C. 1066 and Sig. Gen. I-196-B, $21 /$-.

VCR97 CRTs. New and crated. Picture tested, 45/6. Bases, 3.6 NEW MAG SLIP TRANSMITTER MOTORS. Made by G.E.C $50 \mathrm{v} ., 50$ cycle. Mk. I-2in., Mk. 11 and IV-3in., $15 /$ e each.
NEW P.M. SPEAKERS. 5in., $14 / 6$; 6 in., $16 / 6$; $8 \mathrm{in} ., 20 /-$ NEW P.M.
$10 \mathrm{in} ., 29.6$.
CARBON MICROPHONES with Matched Trans,, $10 / 6$
$0-500$ MICROAMMETERS $2 \mathrm{in}$. 16/6.
M/C MICROPHONES with Trans. 15 /
4ft. ROD AERIALS, set of three $5 / 6$. Base $3 /$.
All Carriage paid in the U.K. from Dept. W.W.,
The RADIO \& ELECTRICAL MART
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[^26]For spare parts for American transmitters see our advertisement on page 58.

## DE HAVILLAND

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Technicians are needed for key positions in an expanding team engaged on research, design and development work. Permanent and progressive posts, both senior and junior, are available for men of ability and enthusiasm who can offer sound knowledge and experience. This is a unique opportunity of acquiring advanced techniques and promotion prospects are good.

## Reference:

17 DESIGNERS experienced in servo mechanisms, gyros, hydraulics, pneumatics, materials and structures related to aircraft or armament work.

17A DRAUGHTSMEN-SENIOR, intermediate and sunior for varied aspects of design.
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21 ELECTRONIC ENGINEERS - SENIOR - to be responsible for section engaged in work as reference 22 below.
22 electronic engineers experienced in one or more of following-circuit development, miniature and sub-miniature techniques, magnetic amplifers and equipment.

23 development engineers, H.N.C. standard, experienced structural testing (preferably aircraft), knowledge of strain-gauge equipment an advantage. Capable of working on own initiative planning test equipment and analysing test results.

24 development engineers, H.N.C. standard, familiar small electro-mechanical equipment, pneumatic and hydraulic scrvo systems for testing and developing. Knowledge electronic test equipment a distinct advantage. Should have experience in undertaking planning, carrying out and analysing of complete tasks.

26 INSTRUMENT ENGINEERS-experience of electro-mechanical instruments (gyros, small specialised motors. relays, etc.).

27 SERVO engineers-practical experience of servo mechanisms.
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A IRGRAFT Radio Mechanlc required, prefergineer's licence but not essential; to be based at Gineer's licence but not essential; to be based at Writing giving full particulars to-The Personnel Manager. The Fairey Aviation Co., Ltd., Hayes. Middlx. E land manufacturers of radio equipment, permanent posts located in the Midlands are offered to men with experience of radar, radio control, ence and salary required, to Personnel Manager,
\([9792\) Box 5744 . 1 Services electronic equipment, substantial exp. on transformer design essen., must be fully alive to modern techniques and methods, sai. to 42 years; London area.-Write fullest particulars to Box 5735.
[9786
MATHEMATICIAN required by a large enginsuburbs of London; applicants should have experience of electronic digital computing machines.-please write. in confidence. quoting reference \(W W / \mathrm{M}\), giving detalls of qualificaions. to Box 5698
DECA RADAR, Ltd., require a Senior Installation Engineer to take charge of a new section of the company responsible for erection and commissioning of high power centimetric systems. The appointment of ers an excelent panding oompıny
APPLICANTS must have sound engineering trraining and extensive experience in commissioning of high power radar or radio systems. An adequate salary will be offered to a man with the right qualities APPLY in confidence to Managing Director, S.W 9 dar, Lu.. 1-3. Brixton Rd., London. R ADIO and radar testers, first-class men tegear and Government contracts for radio and radar equipment by Midland manufacturers; men with wide experience of fault finding in any o the flelds mentioned should write. giving full
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tails apply O.C. \\
REME. Shrapnel Baryacks. S.E.18. & Group \\
\hline 9760
\end{tabular}

GEEFifiDUM METAL RECTIFIERS
\begin{tabular}{|c|c|c|c|c|}
\hline Volts & Amps & & & Price \\
\hline 230 & 11/2 & F/bridge & E4 & 00 \\
\hline 165 & 4 & " & ¢8 & 0 \\
\hline 110 & \(\frac{1}{2}\) & " & ¢1 & 76 \\
\hline 50 & \(\frac{1}{2}\) & " & & 196 \\
\hline 24 & 10 & " & ¢4 & 00 \\
\hline 24 & 6 & " & fl & 150 \\
\hline 24 & \(2 \frac{1}{2}\) & " & E1 & 50 \\
\hline 24 & 1 & " & & 136 \\
\hline 12 & 10 & " & £2 & 100 \\
\hline 12 & 6 & " & \(\pm 1\) & 50 \\
\hline 12 & 4 & , & & 186 \\
\hline 12 & \(2 \frac{1}{2}\) & " & & 166 \\
\hline 12 & 2 & " & & 120 \\
\hline 12 & 1 & " & & 76 \\
\hline 6 & 1 & & & 76 \\
\hline RM1-125v. & 60 mA . & H/wave & & 46 \\
\hline RM2-125\%, & 80 mA . & " & & 50 \\
\hline RM3-125v. & 100 mA . & . \(\cdot\) & & 60 \\
\hline RM4-250v. & 250 mA . & " & & 186 \\
\hline K3/200-5.2kV. & 1 mA . & " & \&1 & 6 8 \\
\hline K3/180-4.660kV. & 1 mA . & , & El & 43 \\
\hline K3/160-4.JkV. & \(I m A\). & ", & El & 16 \\
\hline K \(3 / 140-3.600 \mathrm{kV}\). & 1 mA . & * & & 193 \\
\hline K 3/120-3.08kV. & 1 mA . & " & & 168 \\
\hline K3/100-2.6kV. & 1 mA . & ", & & 148 \\
\hline K \(3 / 90-2.280 \mathrm{kV}\). & \(\operatorname{lm} A\). & , & & 136 \\
\hline K3/80-2.030kV. & \(\operatorname{Im} A\). & " & & 124 \\
\hline K3/70-1.708kV. & 1 mA . & , & & 110 \\
\hline K \(3160-1.500 \mathrm{kV}\). & \(\operatorname{ImA}\). & " & & 98 \\
\hline K3/50-1.260kV. & 1 mA . & , & & 88 \\
\hline \(\mathrm{K} 3 / 45-1.140 \mathrm{kV}\). & 1 mA . & " & & 82 \\
\hline K \(3 / 40\)-1 kV. & 1 mA . & , & & 76 \\
\hline K3/35-885v. & \(\operatorname{ImA}\). & " & & 610 \\
\hline K/30-755v. & 1 man . & , & & 60 \\
\hline K3/25-655v. & \(\operatorname{ImA}\). & " & & 58 \\
\hline K3/20-500v. & \(\operatorname{Im} A\). & " & & 51 \\
\hline J50-400\%. & 2 mA . & , & & 69 \\
\hline J25-250v. & 2 mA . & " & & 36 \\
\hline PULSE CIRCUIT & & & & \\
\hline 36 EHT 25 W'house & 2 mA . & " & & 13 \\
\hline 36 EHT 35 ", & 2 mA . & , & & 17 \\
\hline 3 EHT 40 " & 2 mA . & , & El & 00 \\
\hline 36 EHT 45 " & 2 mA . & ", & £I & 26 \\
\hline 36 EHT 50 & 2 mA . & " & El & 6 \\
\hline 36 EHT ICO & 2 mA . & & \(\pm 1\) & 8 \\
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